Final Electrical Specifications

LT1790-2.5

# 2.5V Micropower SOT-23 Low Dropout Reference

March 2000

#### **FEATURES**

High Accuracy:

A Grade—0.05% Max

B Grade—0.1% Max

■ Low Drift:

A Grade—10ppm/°C Max

B Grade—25ppm/°C Max

■ Low Supply Current: 60µA Max

Sinks and Sources: 5mA Min

Low Dropout Voltage

■ Guaranteed Operational –40°C to 125°C

Wide Supply Range: 2.6V to 18V

#### **APPLICATIONS**

- Handheld Instruments
- Negative Voltage References
- Industrial Control Systems
- Data Acquisition Systems
- Battery-Operated Equipment

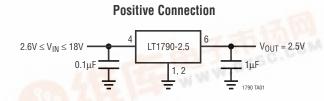
# DESCRIPTION

The LT®1790-2.5 is a SOT-23 micropower low dropout series reference that combines high accuracy and low drift with low power dissipation and small package size. This micropower reference uses curvature compensation to obtain a low temperature coefficient and trimmed precision thin-film resistors to achieve high output accuracy. In addition, the LT1790-2.5 uses post-package trimming to greatly reduce the temperature coefficient and increase the output accuracy. Output accuracy is further assured by excellent line and load regulation. Special care has been taken to minimize thermally induced hysteresis.

The LT1790-2.5 is ideally suited for battery-operated systems because of its small size, low supply current and reduced dropout voltage. This reference provides supply current and power dissipation advantages over shunt references that must idle the entire load current to operate. However, since the LT1790-2.5 can also sink current, it can operate as a micropower negative voltage reference with the same performance as a positive reference.

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# TYPICAL APPLICATION



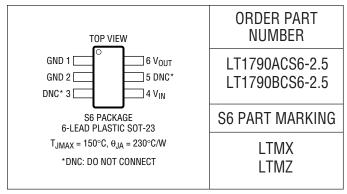
#### Typical V<sub>OUT</sub> Distribution 167 UNITS 45 40 LT1790BC LIMITS 35 NUMBER OF UNITS LT1790AC LIMITS 30 25 20 15 10 5 0 2.498 2.499 2.501 2.502 2.500 OUTPUT VOLTAGE (V)



# **ABSOLUTE MAXIMUM RATINGS**

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# PACKAGE/ORDER INFORMATION



Consult factory for Industrial and Military grade parts.

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes specifications that apply over the specified temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . $V_{IN} = 3V$ , $C_L = 1\mu F$ unless otherwise specified.

PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
Output Voltage (Notes 3, 4)	LT1790ACS6-2.5		2.49875 -0.05	2.50	2.50125 0.05	V %
	LT1790BCS6-2.5		2.4975 -0.1	2.50	2.5025 0.1	V %
Output Voltage Temperature Coefficient (Note 5)	LT1790ACS6-2.5 LT1790BCS6-2.5	•		5 12	10 25	ppm/°C ppm/°C
Line Regulation	$3V \le V_{IN} \le 18V$	•		50	170 220	ppm/V ppm/V
Load Regulation (Note 6)	I <sub>OUT</sub> Source = 5mA I <sub>OUT</sub> Source = 5mA I <sub>OUT</sub> Sink = 5mA	•		80 70	160 250 110	ppm/mA ppm/mA ppm/mA
Dropout Voltage (Note 7)	$V_{\text{IN}} - V_{\text{OUT}}, \Delta V_{\text{OUT}} \le 0.1\%$ $I_{\text{OUT}} = 0\text{mA}$ $I_{\text{OUT}} \text{ Source} = 5\text{mA}$ $I_{\text{OUT}} \text{ Sink} = 5\text{mA}$	•		60 300 40	100 400 250	mV mV mV
Supply Current	V <sub>OUT</sub> = 2.5V	•		35	60 75	μA μA
Minimum Current	V <sub>OUT</sub> = -2.5V			100	125	μА
Turn-On Time	$C_{LOAD} = 1\mu F$			700		μS
Output Noise (Note 8)	$0.1$ Hz $\leq$ f $\leq$ 10Hz 10Hz $\leq$ f $\leq$ 1kHz			12 33		μV <sub>P-P</sub> μV <sub>RMS</sub>
Long-Term Drift of Output Voltage (Note 9)				50		ppm/√kHr
Hysteresis (Note 10)	$\Delta T = 0$ °C to $70$ °C $\Delta T = -40$ °C to $85$ °C	•		40 60		ppm ppm

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

**Note 2:** The LT1790S6 is guaranteed functional over the operating temperature range of  $-40^{\circ}$ C to 125°C.

**Note 3:** If the part is stored outside of the specified temperature range, the output voltage may shift due to hysteresis.

**Note 4:** ESD (Electrostatic Discharge) sensitive device. Extensive use of ESD protection devices are used internal to the LT1790, however, high electrostatic discharge can damage or degrade the device. Use proper ESD handling precautions.

#### **ELECTRICAL CHARACTERISTICS**

**Note 5:** Temperature coefficient is measured by dividing the change in output voltage by the specified temperature range. Incremental slope is also measured at 25°C.

**Note 6:** Load regulation is measured on a pulse basis from no load to the specified load current. Output changes due to die temperature change must be taken into account separately.

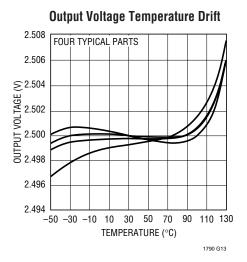
Note 7: Excludes load regulation errors.

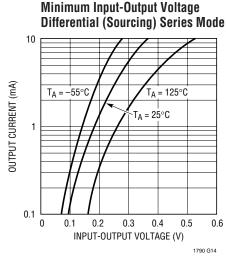
**Note 8:** Peak-to-peak noise is measured with a single pole highpass filter at 0.1Hz and a 2-pole lowpass filter at 10Hz. The unit is enclosed in a still air environment to eliminate thermocouple effects on the leads. The test time is 10 seconds. RMS noise is measured with a single pole highpass filter at 10Hz and a 2-pole lowpass filter at 1kHz. The resulting output is full-wave rectified and then integrated for a fixed period, making the final reading an average as opposed to RMS. A correction factor of 1.1 is used to convert from average to RMS and a second correction of 0.88 is used to correct for the nonideal bandpass of the filters.

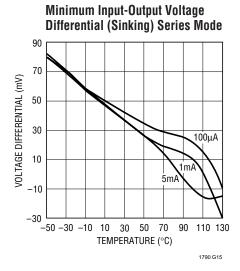
**Note 9:** Long-term drift typically has a logarithmic characteristic and therefore changes after 1000 hours tend to be smaller than before that time. Total drift in the second thousand hours is normally less than one third that to the first thousand hours with a continuing trend toward reduced drift with time. Long-term drift is affected by differential stress between the IC and the board material created during board assembly. See Applications Information.

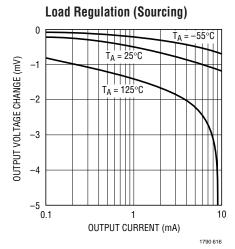
**Note 10:** Hysteresis in the output voltage is created by package stress that differs depending on whether the IC was previously at a higher or lower temperature. Output voltage is always measured at  $25^{\circ}$ C, but the IC is cycled to  $85^{\circ}$ C or  $-40^{\circ}$ C before a successive measurements. Hysteresis is roughly proportional to the square of the temperature change. Hysteresis is not normally a problem for operational temperature excursions where the instrument might be stored at high or low temperature. See Applications Information.

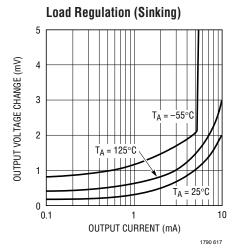
#### TYPICAL PERFORMANCE CHARACTERISTICS

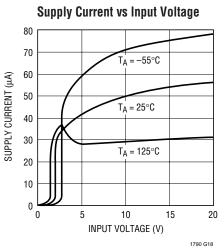




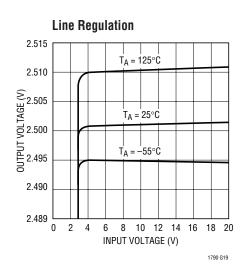


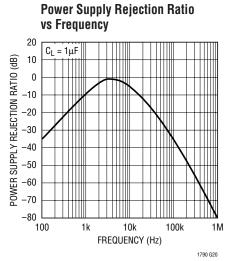


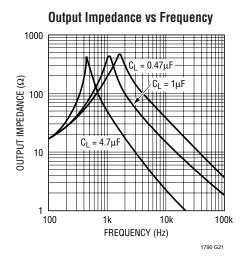




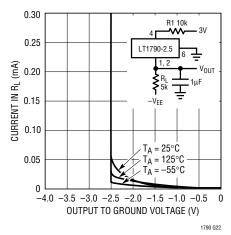
# TYPICAL PERFORMANCE CHARACTERISTICS



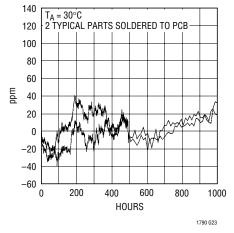




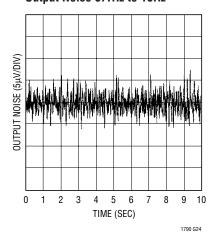




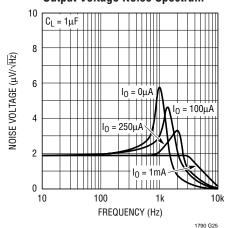




#### Output Noise 0.1Hz to 10Hz



#### Output Voltage Noise Spectrum



#### APPLICATIONS INFORMATION

#### **Bypass and Load Capacitors**

The LT1790-2.5 voltage reference should have an input bypass capacitor of  $0.1\mu F$  or larger, however the bypassing of other local devices may serve as the required component. This reference also requires an output capacitor for stability. The optimum output capacitance for most applications is  $1\mu F$ , although larger values work as well. This capacitor affects the turn-on and settling time for the output to reach its final value.

Figure 1 shows the turn-on time for the LT1790-2.5 with a  $1\mu F$  input bypass and  $1\mu F$  load capacitor. Figure 2 shows the output response to a 0.5V transient on  $V_{IN}$  with the same capacitors.

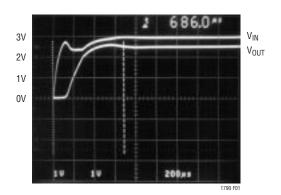


Figure 1. Turn-On Characteristics of LT1790-2.5

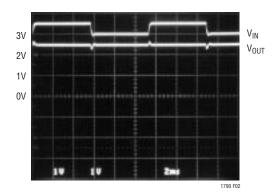


Figure 2. Output Response to 0.5V Ripple on VIN

The test circuit of Figure 3 is used to measure the stability of various load currents. With  $R_L = 1k$ , the 1V step produces a current step of 1mA. Figure 4 shows the response to a  $\pm 0.5$ mA load. Figure 5 is the output response to a sourcing step from 4mA to 5mA, and Figure 6 is the output response of a sinking step from -4mA to -5mA.

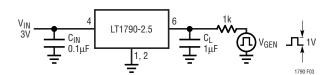


Figure 3. Response Time Test Circuit

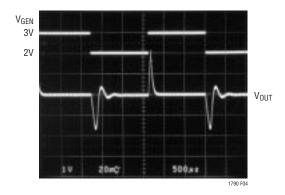


Figure 4. LT1790-2.5 Sourcing and Sinking 0.5mA

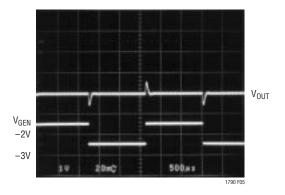


Figure 5. LT1790-2.5 Sourcing 4mA to 5mA

# **APPLICATIONS INFORMATION**

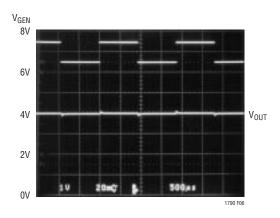


Figure 6. LT1790-2.5 Sinking -4mA to -5mA

#### **Positive or Negative Operation**

Series operation is ideal for extending battery life. If the LT1790-2.5 is operated in series mode it does not require an external current setting resistor. The specifications guarantee the LT1790-2.5 operates from 2.6V to 18V. When the circuitry being regulated does not demand current, the series connected LT1790-2.5 consumes only a few hundred  $\mu W$ , yet the same connection can sink or source 5mA of load current when demanded. A typical series connection is shown on the front page of this data sheet.

The circuit in Figure 7 shows the connection for a -2.5V reference. The LT1790-2.5 can be used as a very stable negative reference, however, it requires a positive voltage applied to Pin 4 to bias internal circuitry. This voltage must be current limited with R1 to keep the output PNP

transistor from turning on and driving the grounded output. C1 provides stability during load transients. This connection maintains the accuracy and temperature coefficient of the positive connected LT1790-2.5.

#### **Long-Term Drift**

Long-term drift cannot be extrapolated from accelerated high temperature testing. This erroneous technique gives drift numbers that are widely optimistic. The only way long-term drift can be determined is to measure it over the time interval of interest. The LT1790S6 drift data was taken on over 100 parts that were soldered into PC boards similar to a "real world" application. The boards were then placed into a constant temperature oven with  $T_A = 30^{\circ}\text{C}$ , their outputs scanned regularly and measured with an 8.5 digit DVM. Long-term drift curves are shown in the Typical Performance Characteristics.

#### **Hysteresis**

Hysteresis data shown in Figures 8 and 9 represent the worst-case data taken on parts from  $0^{\circ}$ C to  $70^{\circ}$ C and from  $-40^{\circ}$ C to  $85^{\circ}$ C. Units were cycled several times over these temperature ranges and the largest change is shown. As expected, the parts cycled over the higher temperature range have higher hysteresis than those cycled over the lower range.

When the LT1790-2.5 is IR reflow soldered onto a PC board, the output shift is typically just 150ppm (0.015%).

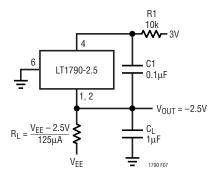
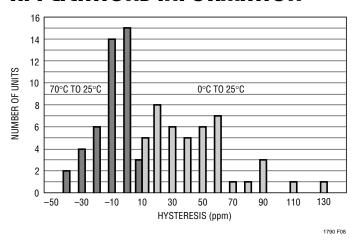


Figure 7. Using the LT1790-2.5 to Build a -2.5V Reference

# **APPLICATIONS INFORMATION**

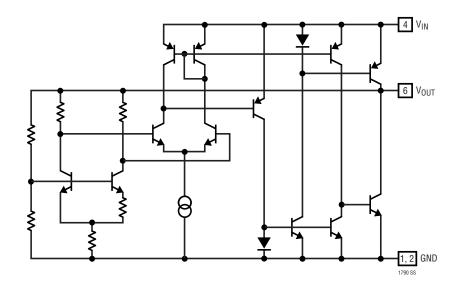


SUND 9 8 7 6 5 4 3 2 1 0 -100 -80 -60 -40 -20 0 20 40 60 80 100 120 HYSTERESIS (ppm)

Figure 8. Worst-Case 0°C to 70°C Hysteresis on 44 Units

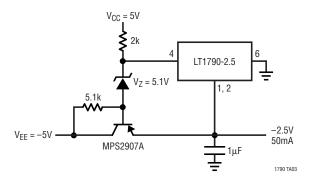
Figure 9. Worst-Case -40°C to 85°C Hysteresis on 44 Units

# SIMPLIFIED SCHEMATIC



# TYPICAL APPLICATION

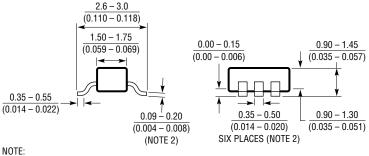
# -2.5V Negative 50mA Series Reference No Load Supply Current $I_{CC}=1.6mA$ $I_{EE}=440\mu A$

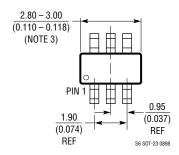


# PACKAGE DESCRIPTION

Dimensions in inches (millimeters) unless otherwise noted.

#### S6 Package 6-Lead Plastic SOT-23 (LTC DWG # 05-08-1634)





- 1. DIMENSIONS ARE IN MILLIMETERS
- 2. DIMENSIONS ARE INCLUSIVE OF PLATING
- 3. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR
- 4. MOLD FLASH SHALL NOT EXCEED 0.254mm
- 5. PACKAGE EIAJ REFERENCE IS SC-74A (EIAJ)

# **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS			
LT1019	Precision Reference	Bandgap, 0.05%, 5ppm/°C			
LT1027	Precision 5V Reference	Lowest TC, High Accuracy, Low Noise, Zener Based			
LT1236	Precision Reference	5V and 10V Zener-Based 5ppm/°C, SO-8 Package			
LTC®1798	Micropower Low Dropout Reference	0.15% Max, 6.5μA Supply Current			
LT1460	Micropower Precison Series Reference	Bandgap, 130μA Supply Current, 10ppm/°C, Available in SOT-23			
LT1461 Micropower Precision Low Dropout Reference		Bandgap 0.04%, 3ppm/°C, 50μA Max Supply Current			
LT1634	Micropower Precision Shunt Voltage Reference	Bandgap, 0.05%, 10ppm/°C, 10μA Supply Current			