19-1246; Rev 0; 7/97

Low-Cost, High-Speed, SOT23, Single-Supply Op Amps with Rail-to-Rail Outputs

General Description

The MAX4012 single, MAX4016 dual, MAX4018 triple, and MAX4020 quad op amps are unity-gain-stable devices that combine high-speed performance with Rail-to-Rail® outputs. The MAX4018 has a disable feature that reduces power-supply current to 400µA and places its outputs into a high-impedance state. These devices operate from a +3.3V to +10V single supply or from ±1.65V to ±5V dual supplies. The common-mode input voltage range extends beyond the negative power-supply rail (ground in single-supply applications).

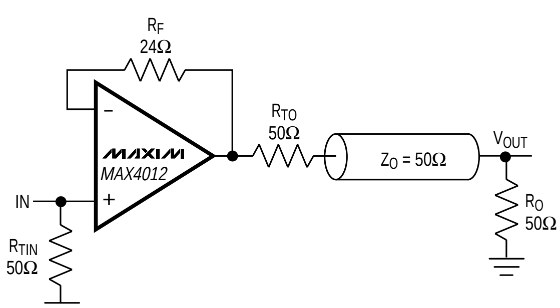
These devices require only 5.5mA of quiescent supply current while achieving a 200MHz -3dB bandwidth and a 600V/µs slew rate. These parts are an excellent solution in low-power/low-voltage systems that require wide bandwidth, such as video, communications, and instrumentation. In addition, when disabled, their high output impedance makes them ideal for multiplexing applications.

The MAX4012 comes in a miniature 5-pin SOT23 package, while the MAX4016 comes in 8-pin µMAX and SO packages. The MAX4018/MAX4020 are available in a space-saving 16-pin QSOP, as well as a 14-pin SO.

Applications

- Set-Top Boxes
- Surveillance Video Systems
- Battery-Powered Instruments
- Video Line Driver
- Analog-to-Digital Converter Interface
- CCD Imaging Systems
- Video Routing and Switching Systems

Typical Operating Circuit



UNITY-GAIN LINE DRIVER ($R_L = R_O + R_{TO}$)

Features

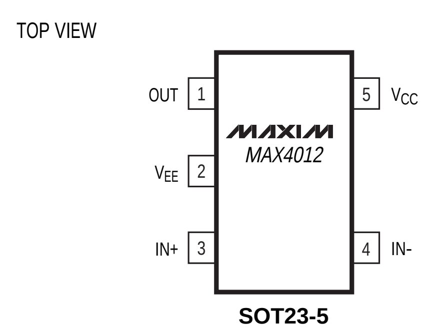
- **Low-Cost**
- **High Speed:**
  200MHz -3dB Bandwidth (MAX4012)
  150MHz -3dB Bandwidth (MAX4016/18/20)
  30MHz 0.1dB Gain Flatness
  600V/µs Slew Rate
- **Single 3.3V/5.0V Operation**
- **Rail-to-Rail Outputs**
- **Input Common-Mode Range Extends Beyond V_{EE}**
- **Low Differential Gain/Phase: 0.02%/0.02°**
- **Low Distortion at 5MHz:**
  -78dBc SFDR
  -75dB Total Harmonic Distortion
- **High Output Drive: ±120mA**
- **400µA Shutdown Capability (MAX4018)**
- **High Output Impedance in Off State (MAX4018)**
- **Space-Saving SOT23-5, µMAX, or QSOP Packages**

Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE	SOT TOP MARK
MAX4012EUK	-40°C to +85°C	5 SOT23-5	ABZP
MAX4016ESA	-40°C to +85°C	8 SO	—
MAX4016EUA	-40°C to +85°C	8 µMAX	—

Ordering Information continued at end of data sheet.

Pin Configurations



SOT23-5

Pin Configurations continued at end of data sheet.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V_{CC} to V_{EE})..+12V
IN_-, IN_+, OUT_, EN_.....................(V_{EE} - 0.3V) to (V_{CC} + 0.3V)
Output Short-Circuit Duration to V_{CC} or V_{EE}............ Continuous
Continuous Power Dissipation (T_A = +70°C)
5-Pin SOT23 (derate 7.1mW/°C above +70°C)...........571mW
8-Pin SO (derate 5.9mW/°C above +70°C).................471mW
8-Pin µMAX (derate 4.1mW/°C above +70°C) 330mW
14-Pin SO (derate 8.3mW/°C above +70°C)...............667mW
16-Pin QSOP (derate 8.3mW/°C above +70°C)..........667mW
Operating Temperature Range ....-40°C to +85°C
Storage Temperature Range.............................-65°C to +150°C
Lead Temperature (soldering, 10sec)............................+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or at any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

(V_{CC} = +5V, V_{EE} = 0V, EN_ = +5V, R_L = ∞ to V_{CC} / 2, V_{OUT} = V_{CC} / 2, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Input Common-Mode Voltage Range	V_{CM}	Guaranteed by CMRR test		V_{EE} - 0.20		V_{CC} - 2.25	V
Input Offset Voltage (Note 2)	V_{OS}				4	20	mV
Input Offset Voltage Temperature Coefficient	TC_{VOS}				8		µV/°C
Input Offset Voltage Matching		Any channels for MAX4016/MAX4018/ MAX4020			±1		mV
Input Bias Current	I_B	(Note 2)			5.4	20	µA
Input Offset Current	I_{OS}	(Note 2)			0.1	20	µA
Input Resistance	R_{IN}	Differential mode (-1V ≤ V_{IN} ≤ +1V)			70		kΩ
		Common mode (-0.2V ≤ V_{CM} ≤ +2.75V)			3		MΩ
Common-Mode Rejection Ratio	CMRR	(V_{EE} - 0.2V) ≤ V_{CM} ≤ (V_{CC} - 2.25V)		70	100		dB
Open-Loop Gain (Note 2)	A_{VOL}	0.25V ≤ V_{OUT} ≤ 4.75V, R_L = 2kΩ			61		dB
		0.5V ≤ V_{OUT} ≤ 4.5V, R_L = 150Ω		52	59		
		1.0V ≤ V_{OUT} ≤ 4V, R_L = 50Ω			57		
Output Voltage Swing (Note 2)	V_{OUT}	R_L = 2kΩ	V_{CC} - V_{OH}		0.06		V
			V_{OL} - V_{EE}		0.06		
		R_L = 150Ω	V_{CC} - V_{OH}		0.30		
			V_{OL} - V_{EE}		0.30		
		R_L = 75Ω	V_{CC} - V_{OH}		0.6	1.5	
			V_{OL} - V_{EE}		0.6	1.5	
		R_L = 75Ω to ground	V_{CC} - V_{OH}		1.1	2.0	
			V_{OL} - V_{EE}		0.05	0.50	
Output Current	I_{OUT}	R_L = 20Ω to V_{CC} or V_{EE}		±100	±120		mA
Output Short-Circuit Current	I_{SC}	Sinking or sourcing			±150		mA
Open-Loop Output Resistance	R_{OUT}				8		Ω

DC ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = +5V, V_{EE} = 0V, EN_ = +5V, R_L = ∞ to V_{CC} / 2, V_{OUT} = V_{CC} / 2, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Power-Supply Rejection Ratio (Note 3)	PSRR	V_{CC} = 5V, V_{EE} = 0V, V_{CM} = +2.0V	46	57		dB
		V_{CC} = 5V, V_{EE} = -5V, V_{CM} = 0V	54	66		
		V_{CC} = 3.3V, V_{EE} = 0V, V_{CM} = +0.90V		45		
Operating Supply-Voltage Range	V_S	V_{CC} to V_{EE}	3.15		11.0	V
Disabled Output Resistance	$R_{OUT\ (OFF)}$	EN_ = 0V, 0V ≤ V_{OUT} ≤ 5V (Note 4)	28	35		kΩ
EN_ Logic-Low Threshold	V_{IL}				V_{CC} - 2.6	V
EN_ Logic-High Threshold	V_{IH}		V_{CC} - 1.6			V
EN_ Logic Input Low Current	I_{IL}	(V_{EE} + 0.2V) ≤ EN_ ≤ V_{CC}		0.5		µA
		EN_ = 0V		200	300	
EN_ Logic Input High Current	I_{IH}	EN_ = 5V		0.5	10	µA
Quiescent Supply Current (per Amplifier)	I_S	Enabled		5.5	7.0	mA
		MAX4018, disabled (EN_ = 0V)		0.40	0.55	

AC ELECTRICAL CHARACTERISTICS

(V_{CC} = +5V, V_{EE} = 0V, V_{CM} = 2.5V, EN_ = +5V, R_F = 24Ω, R_L = 100Ω to V_{CC} / 2, V_{OUT} = V_{CC} / 2, A_{VCL} = +1, T_A = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Small-Signal -3dB Bandwidth	BW_{SS}	V_{OUT} = 20mVp-p, MAX4012		200		MHz
		V_{OUT} = 20mVp-p, MAX4016/MAX4018/MAX4020		150		MHz
Large-Signal -3dB Bandwidth	BW_{LS}	V_{OUT} = 2Vp-p		140		MHz
Bandwidth for 0.1dB Gain Flatness	$BW_{0.1dB}$	V_{OUT} = 20mVp-p (Note 5)	6	30		MHz
Slew Rate	SR	V_{OUT} = 2V step		600		V/µs
Settling Time to 0.1%	t_S	V_{OUT} = 2V step		45		ns
Rise/Fall Time	t_R, t_F	V_{OUT} = 100mVp-p		1		ns
Spurious-Free Dynamic Range	SFDR	f_C = 5MHz, V_{OUT} = 2Vp-p		-78		dBc
Harmonic Distortion	HD	f_C = 5MHz, V_{OUT} = 2Vp-p, 2nd harmonic		-78		dBc
		f_C = 5MHz, V_{OUT} = 2Vp-p, 3rd harmonic		-82		dBc
		f_C = 5MHz, V_{OUT} = 2Vp-p, Total harmonic distortion		-75		dB
Two-Tone, Third-Order Intermodulation Distortion	IP3	f1 = 10.0MHz, f2 = 10.1MHz, V_{OUT} = 1Vp-p		35		dBc
Input 1dB Compression Point		f_C = 10MHz, A_{VCL} = +2		11		dBm
Differential Phase Error	DP	NTSC, R_L = 150Ω		0.02		degrees
Differential Gain Error	DG	NTSC, R_L = 150Ω		0.02		%
Input Noise-Voltage Density	e_n	f = 10kHz		10		$nV/\sqrt{Hz}$
Input Noise-Current Density	i_n	f = 10kHz		6		$pA/\sqrt{Hz}$
Input Capacitance	C_{IN}			1		pF
Disabled Output Capacitance	$C_{OUT (OFF)}$	MAX4018, EN_ = 0V		2		pF
Output Impedance	Z_{OUT}	f = 10MHz		6		Ω
Amplifier Enable Time	t_{ON}	MAX4018		100		ns
Amplifier Disable Time	t_{OFF}	MAX4018		1		µs
Amplifier Gain Matching		MAX4016/MAX4018/MAX4020, f = 10MHz, V_{OUT} = 20mVp-p		0.1		dB
Amplifier Crosstalk	X_{TALK}	MAX4016/MAX4018/MAX4020, f = 10MHz, V_{OUT} = 2Vp-p, R_S = 50Ω to ground		-95		dB

Note 1: The MAX4012EUT is 100% production tested at T_A = +25°C. Specifications over temperature limits are guaranteed by design.

Note 2: Tested with V_{CM} = +2.5V.

Note 3: PSR for single +5V supply tested with V_{EE} = 0V, V_{CC} = +4.5V to +5.5V; for dual ±5V supply with V_{EE} = -4.5V to -5.5V, V_{CC} = +4.5V to +5.5V; and for single +3.3V supply with V_{EE} = 0V, V_{CC} = +3.15V to +3.45V.

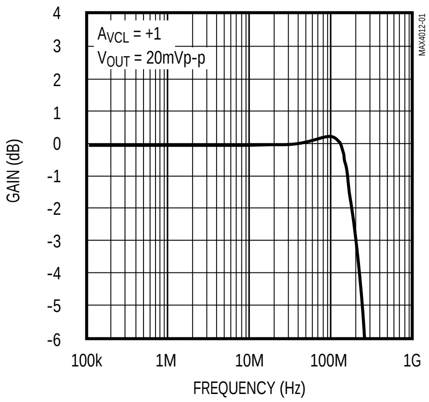
Note 4: Does not include the external feedback network's impedance.

Note 5: Guaranteed by design.

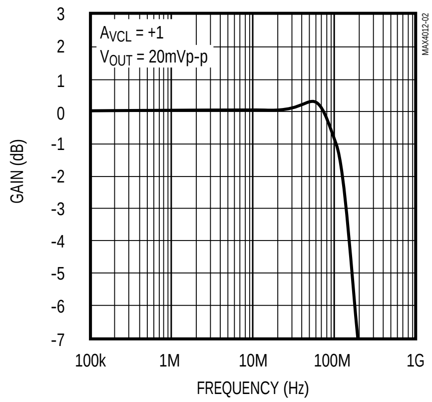
Typical Operating Characteristics

(V_{CC} = +5V, V_{EE} = 0V, A_{VCL} = +1, R_F = 24Ω, R_L = 100Ω to V_{CC} / 2, T_A = +25°C, unless otherwise noted.)

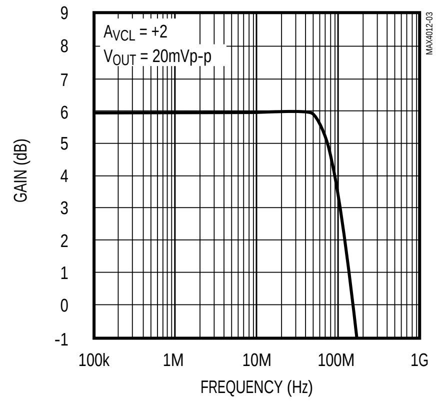
MAX4012
SMALL-SIGNAL GAIN vs. FREQUENCY
(A_{VCL} = +1)



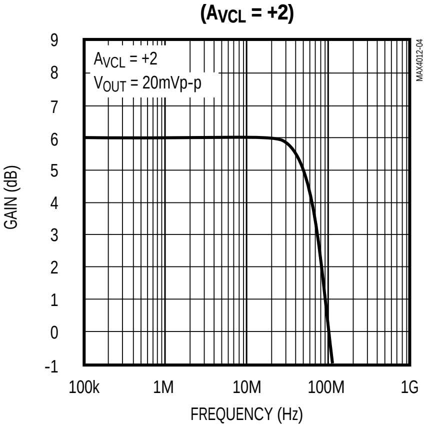
MAX4016/18/20
SMALL-SIGNAL GAIN vs. FREQUENCY
(A_{VCL} = +1)



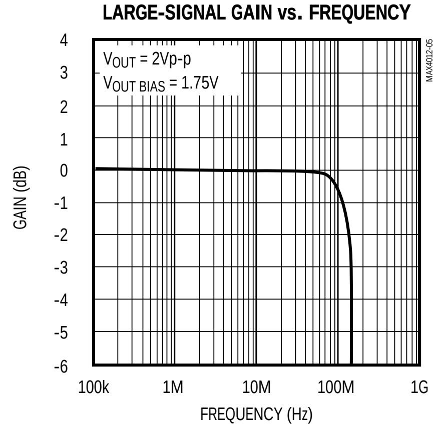
MAX4012
SMALL-SIGNAL GAIN vs. FREQUENCY
(A_{VCL} = +2)



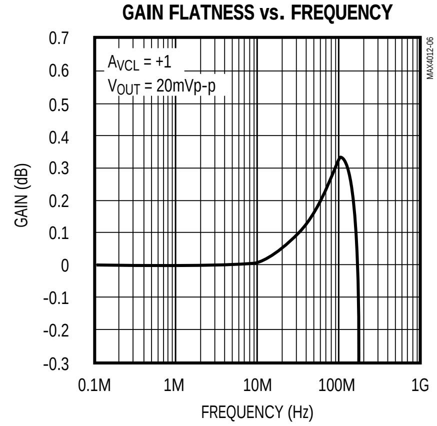
MAX4016/18/20
SMALL-SIGNAL GAIN vs. FREQUENCY
(A_{VCL} = +2)



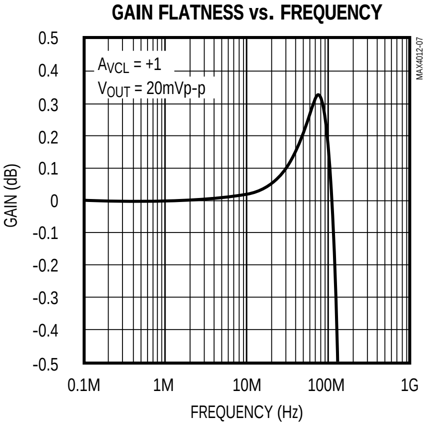
LARGE-SIGNAL GAIN vs. FREQUENCY



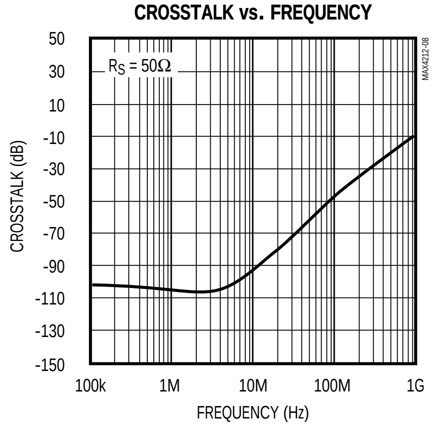
MAX4012
GAIN FLATNESS vs. FREQUENCY



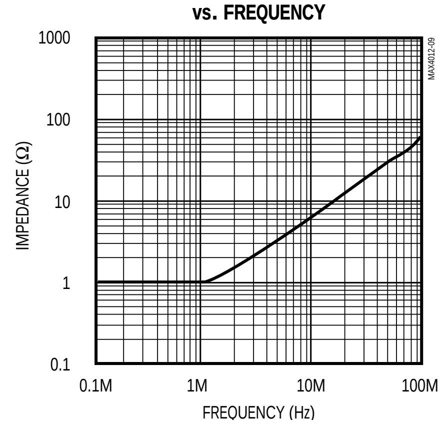
MAX4016/18/20
GAIN FLATNESS vs. FREQUENCY



MAX4016/18/20
CROSSTALK vs. FREQUENCY



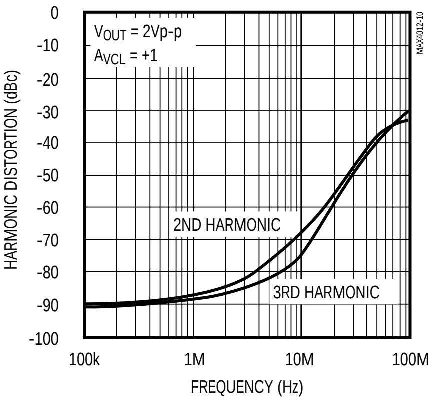
CLOSED-LOOP OUTPUT IMPEDANCE
vs. FREQUENCY



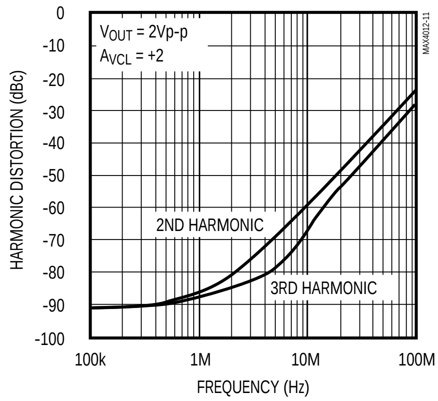
Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = 0V, A_{VCL} = +1, R_F = 24Ω, R_L = 100Ω to V_{CC} / 2, T_A = +25°C, unless otherwise noted.)

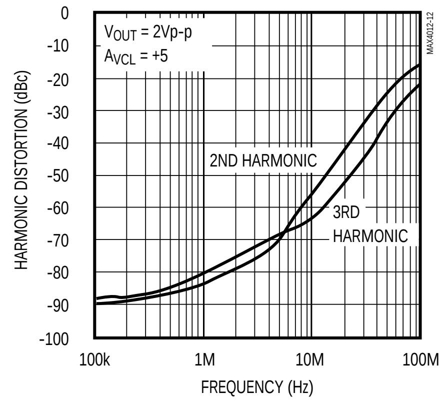
HARMONIC DISTORTION vs. FREQUENCY (A_{VCL} = +1)



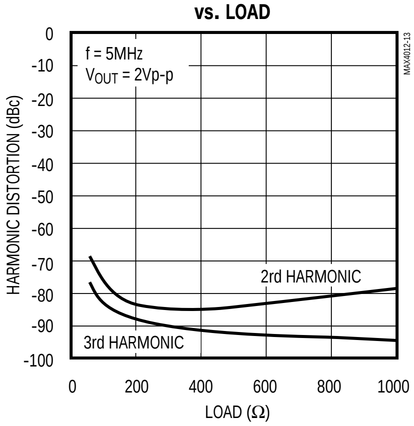
HARMONIC DISTORTION vs. FREQUENCY (A_{VCL} = +2)



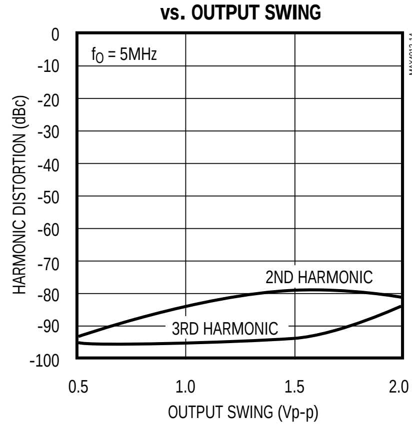
HARMONIC DISTORTION vs. FREQUENCY (A_{VCL} = +5)



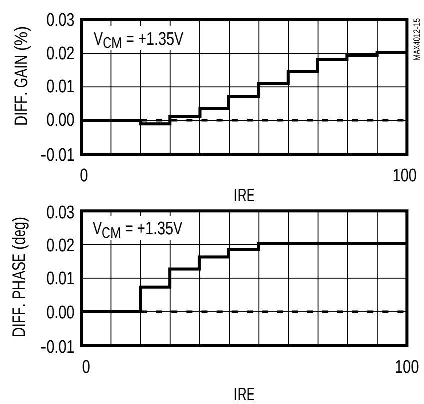
HARMONIC DISTORTION vs. LOAD



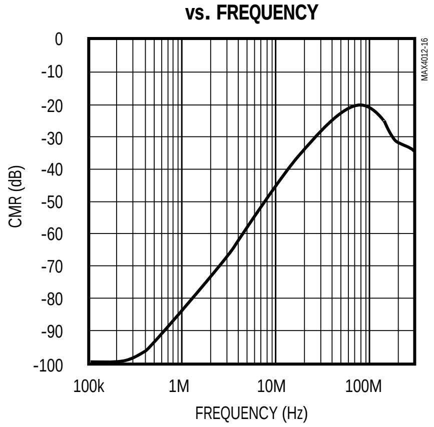
HARMONIC DISTORTION vs. OUTPUT SWING



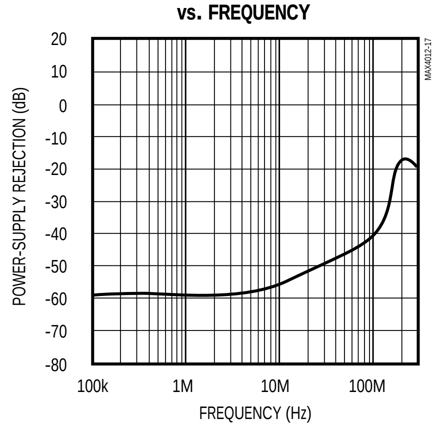
DIFFERENTIAL GAIN AND PHASE



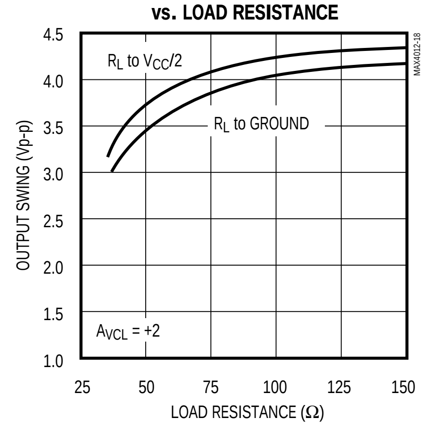
COMMON-MODE REJECTION vs. FREQUENCY



POWER-SUPPLY REJECTION vs. FREQUENCY



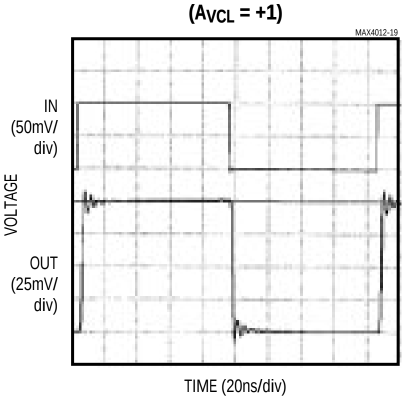
OUTPUT SWING vs. LOAD RESISTANCE



Typical Operating Characteristics (continued)

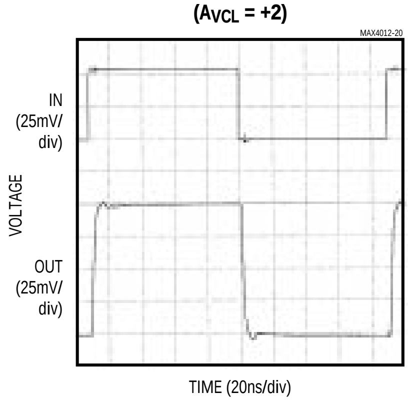
(V_{CC} = +5V, V_{EE} = 0V, A_{VCL} = +1, R_F = 24Ω, R_L = 100Ω to V_{CC} / 2, T_A = +25°C, unless otherwise noted.)

SMALL-SIGNAL PULSE RESPONSE (A_{VCL} = +1)



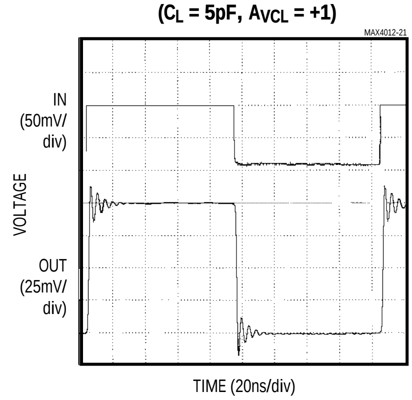
V_{CM} = +2.5V, R_L = 100Ω to GROUND

SMALL-SIGNAL PULSE RESPONSE (A_{VCL} = +2)



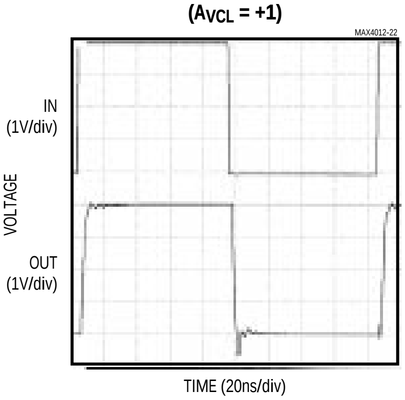
V_{CM} = +1.25V, R_L = 100Ω to GROUND

SMALL-SIGNAL PULSE RESPONSE (C_L = 5pF, A_{VCL} = +1)



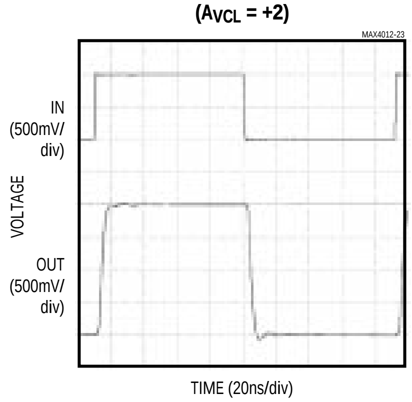
V_{CM} = +1.75V, R_L = 100Ω to GROUND

LARGE-SIGNAL PULSE RESPONSE (A_{VCL} = +1)



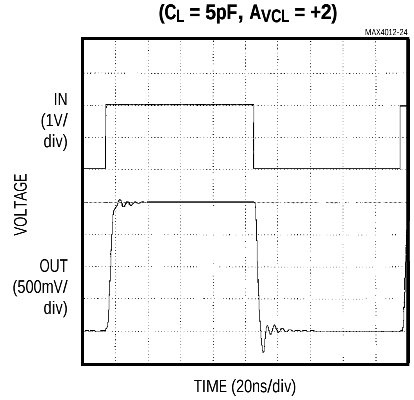
V_{CM} = +1.75V, R_L = 100Ω to GROUND

LARGE-SIGNAL PULSE RESPONSE (A_{VCL} = +2)



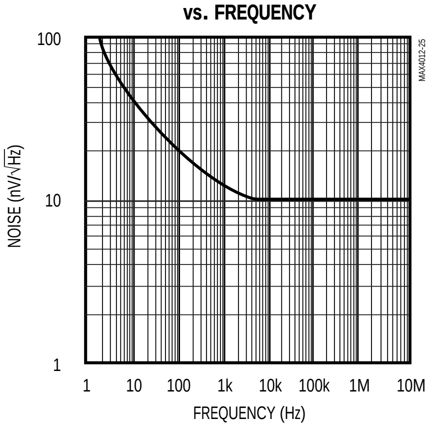
V_{CM} = 0.9V, R_L = 100Ω to GROUND

LARGE-SIGNAL PULSE RESPONSE (C_L = 5pF, A_{VCL} = +2)

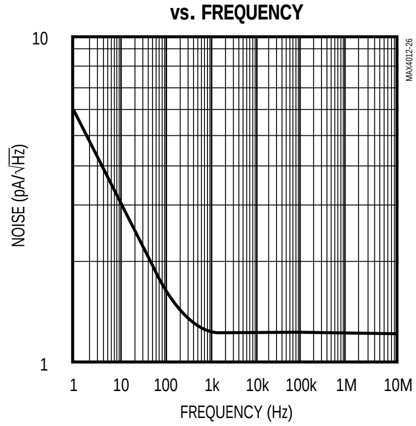


V_{CM} = +1.75V, R_L = 100Ω to GROUND

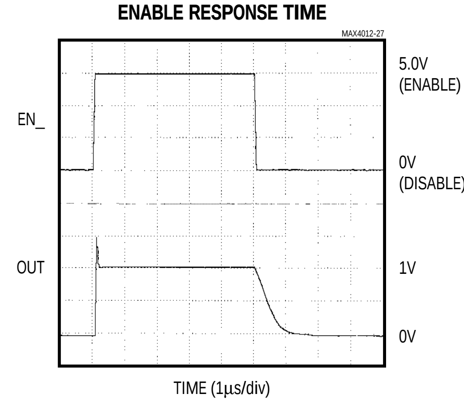
VOLTAGE-NOISE DENSITY vs. FREQUENCY



CURRENT-NOISE DENSITY vs. FREQUENCY



ENABLE RESPONSE TIME

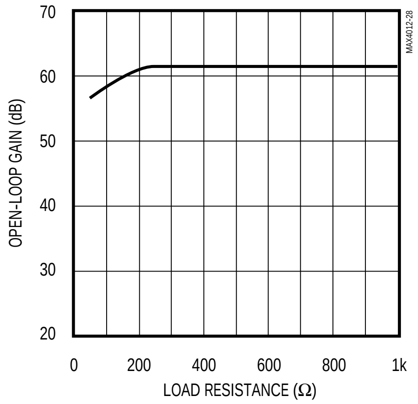


V_{IN} = +1.0V

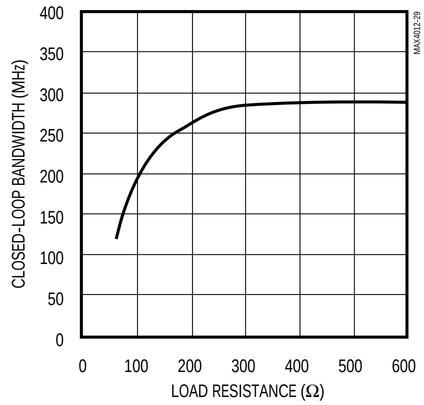
Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = 0V, A_{VCL} = +1, R_F = 24Ω, R_L = 100Ω to V_{CC} / 2, T_A = +25°C, unless otherwise noted.)

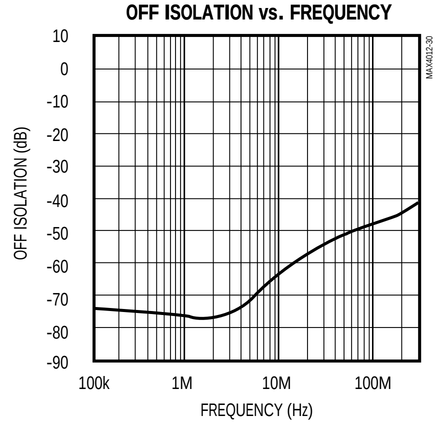
OPEN-LOOP GAIN vs. LOAD RESISTANCE



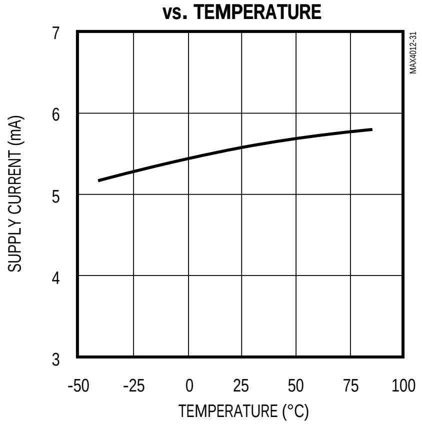
CLOSED-LOOP BANDWIDTH vs. LOAD RESISTANCE



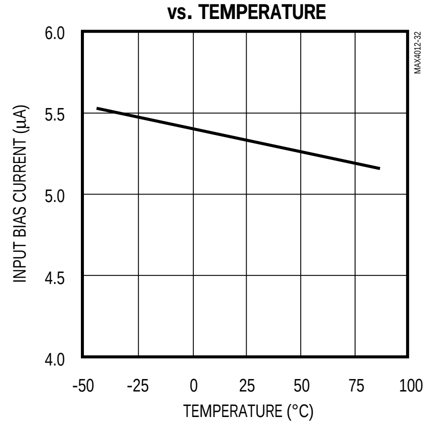
OFF ISOLATION vs. FREQUENCY



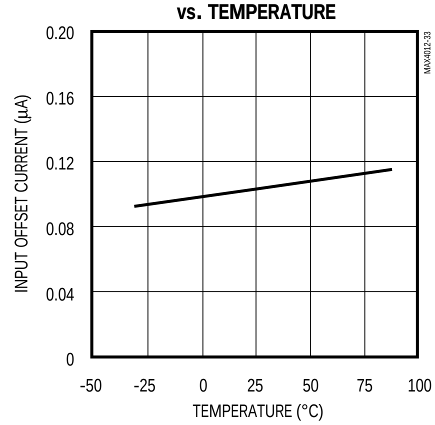
SUPPLY CURRENT vs. TEMPERATURE



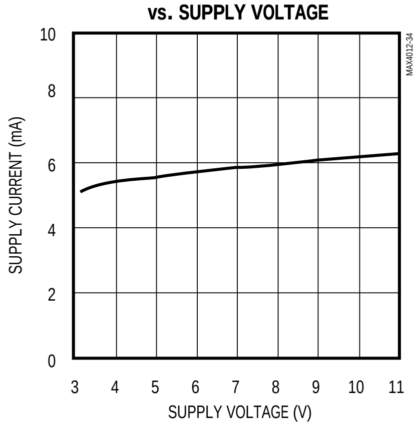
INPUT BIAS CURRENT vs. TEMPERATURE



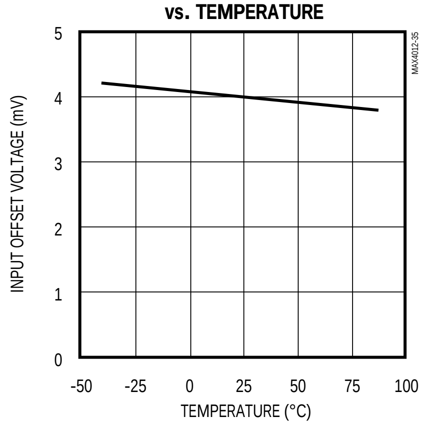
INPUT OFFSET CURRENT vs. TEMPERATURE



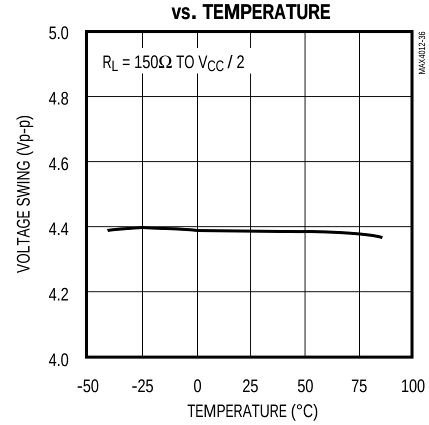
SUPPLY CURRENT vs. SUPPLY VOLTAGE



INPUT OFFSET VOLTAGE vs. TEMPERATURE



OUTPUT VOLTAGE SWING vs. TEMPERATURE



Pin Description

PIN						NAME	FUNCTION
MAX4012 SOT23-5	MAX4016 SO/μMAX	MAX4018		MAX4020			
		SO	QSOP	SO	QSOP		
—	—	—	8, 9	—	8, 9	N.C.	No Connect. Not internally connected. Tie to ground or leave open.
1	—	—	—	—	—	OUT	Amplifier Output
2	4	11	13	11	13	V_{EE}	Negative Power Supply or Ground (in single-supply operation)
3	—	—	—	—	—	IN+	Noninverting Input
4	—	—	—	—	—	IN-	Inverting Input
5	8	4	4	4	4	V_{CC}	Positive Power Supply
—	1	7	7	1	1	OUTA	Amplifier A Output
—	2	6	6	2	2	INA-	Amplifier A Inverting Input
—	3	5	5	3	3	INA+	Amplifier A Noninverting Input
—	7	8	10	7	7	OUTB	Amplifier B Output
—	6	9	11	6	6	INB-	Amplifier B Inverting Input
—	5	10	12	5	5	INB+	Amplifier B Noninverting Input
—	—	14	16	8	10	OUTC	Amplifier C Output
—	—	13	15	9	11	INC-	Amplifier C Inverting Input
—	—	12	14	10	12	INC+	Amplifier C Noninverting Input
—	—	—	—	14	16	OUTD	Amplifier D Output
—	—	—	—	13	15	IND-	Amplifier D Inverting Input
—	—	—	—	12	14	IND+	Amplifier D Noninverting Input
—	—	—	—	—	—	EN	Enable Amplifier
—	—	1	1	—	—	ENA	Enable Amplifier A
—	—	3	3	—	—	ENB	Enable Amplifier B
—	—	2	2	—	—	ENC	Enable Amplifier C

Detailed Description

The MAX4012/MAX4016/MAX4018/MAX4020 are single-supply, rail-to-rail, voltage-feedback amplifiers that employ current-feedback techniques to achieve 600V/µs slew rates and 200MHz bandwidths. Excellent harmonic distortion and differential gain/phase performance make these amplifiers an ideal choice for a wide variety of video and RF signal-processing applications.

The output voltage swing comes to within 50mV of each supply rail. Local feedback around the output stage assures low open-loop output impedance to reduce gain sensitivity to load variations. This feedback also produces demand-driven current bias to the output transistors for ±120mA drive capability, while constraining total supply current to less than 7mA. The input stage permits common-mode voltages beyond the negative supply and to within 2.25V of the positive supply rail.

Applications Information

Choosing Resistor Values

Unity-Gain Configuration

The MAX4012/MAX4016/MAX4018/MAX4020 are internally compensated for unity gain. When configured for unity gain, the devices require a 24Ω resistor (R_F) in series with the feedback path. This resistor improves AC response by reducing the Q of the parallel LC circuit formed by the parasitic feedback capacitance and inductance.

Inverting and Noninverting Configurations

Select the gain-setting feedback (R_F) and input (R_G) resistor values to fit your application. Large resistor values increase voltage noise and interact with the amplifier's input and PC board capacitance. This can generate undesirable poles and zeros and decrease bandwidth or cause oscillations. For example, a noninverting gain-of-two configuration ($R_F = R_G$) using 1kΩ resistors, combined with 1pF of amplifier input capacitance and 1pF of PC board capacitance, causes a pole at 159MHz. Since this pole is within the amplifier bandwidth, it jeopardizes stability. Reducing the 1kΩ resistors to 100Ω extends the pole frequency to 1.59GHz, but could limit output swing by adding 200Ω in parallel with the amplifier's load resistor. Table 1 shows suggested feedback, gain resistors, and bandwidth for several gain values in the configurations shown in Figures 1a and 1b.

Layout and Power-Supply Bypassing

These amplifiers operate from a single +3.3V to +11V power supply or from dual supplies to ±5.5V. For single-supply operation, bypass V_{CC} to ground with a 0.1µF capacitor as close to the pin as possible. If operating with dual supplies, bypass each supply with a 0.1µF capacitor.

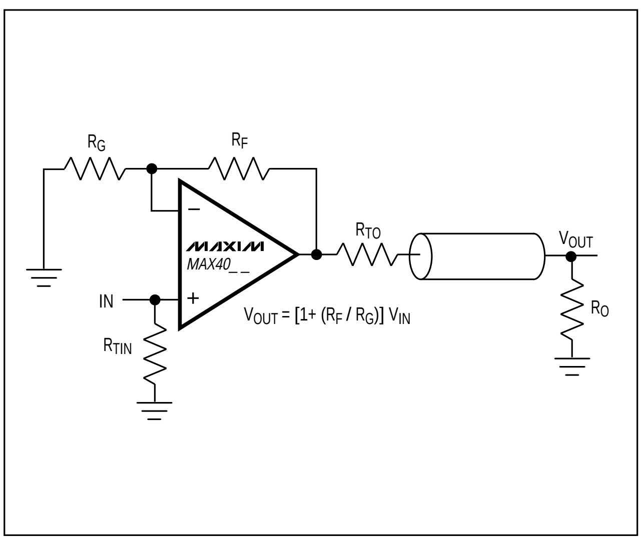


Figure 1a. Noninverting Gain Configuration

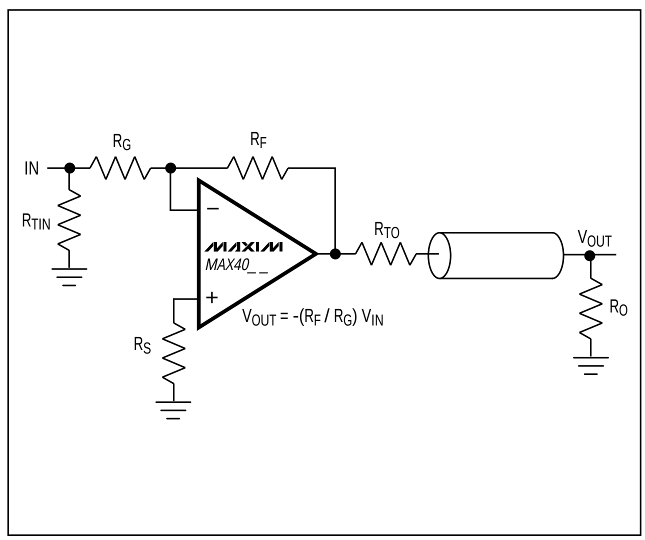


Figure 1b. Inverting Gain Configuration

Maxim recommends using microstrip and stripline techniques to obtain full bandwidth. To ensure that the PC board does not degrade the amplifier's performance, design it for a frequency greater than 1GHz. Pay careful attention to inputs and outputs to avoid large parasitic capacitance. Whether or not you use a constant-impedance board, observe the following guidelines when designing the board:

- Don't use wire-wrap boards because they are too inductive.
- Don't use IC sockets because they increase parasitic capacitance and inductance.
- Use surface-mount instead of through-hole components for better high-frequency performance.
- Use a PC board with at least two layers; it should be as free from voids as possible.
- Keep signal lines as short and as straight as possible. Do not make 90° turns; round all corners.

Rail-to-Rail Outputs, Ground-Sensing Input

The input common-mode range extends from (V_{EE} - 200mV) to (V_{CC} - 2.25V) with excellent common-mode rejection. Beyond this range, the amplifier output is a nonlinear function of the input, but does not undergo phase reversal or latchup.

The output swings to within 60mV of either power-supply rail with a 2kΩ load. The input ground-sensing and the rail-to-rail output substantially increase the dynamic range. With a symmetric input in a single +5V application, the input can swing 2.95Vp-p, and the output can swing 4.9Vp-p with minimal distortion.

Enable Input and Disabled Output

The enable feature (EN_) allows the amplifier to be placed in a low-power, high-output-impedance state. Typically, the EN_ logic low input current (I_{IL}) is small. However, as the EN voltage (V_{IL}) approaches the negative supply rail, I_{IL} increases (Figure 2). A single resistor connected as shown in Figure 3 prevents the rise in the logic-low input current. This resistor provides a feedback mechanism that increases V_{IL} as the logic input is brought to V_{EE}. Figure 4 shows the resulting input current (I_{IL}).

When the MAX4018 is disabled, the amplifier's output impedance is 35kΩ. This high resistance and the low 2pF output capacitance make this part ideal in RF/video multiplexer or switch applications. For larger arrays, pay careful attention to capacitive loading. See the *Output Capacitive Loading and Stability* section for more information.

Table 1. Recommended Component Values

COMPONENT	GAIN (V/V)									
	+1	-1	+2	-2	+5	-5	+10	-10	+25	-25
R_F (Ω)	24	500	500	500	500	500	500	500	500	1200
R_G (Ω)	∞	500	500	250	124	100	56	50	20	50
R_S (Ω)	—	0	—	0	—	0	—	0	—	0
R_{TIN} (Ω)	49.9	56	49.9	62	49.9	100	49.9	∞	49.9	∞
R_{TO} (Ω)	49.9	49.9	49.9	49.9	49.9	49.9	49.9	49.9	49.9	49.9
Small-Signal -3dB Bandwidth (MHz)	200	90	105	60	25	33	11	25	6	10

Note: $R_L = R_O + R_{TO}$; R_{TIN} and R_{TO} are calculated for 50Ω applications. For 75Ω systems, R_{TO} = 75Ω; calculate R_{TIN} from the following equation:

$$R_{TIN} = \frac{75}{1 - \frac{75}{R_G}} \Omega$$

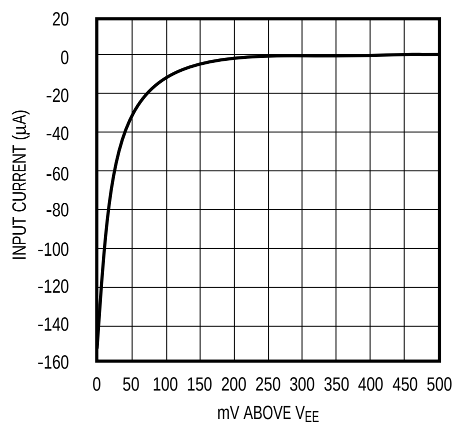


Figure 2. Enable Logic-Low Input Current vs. V_{IL}

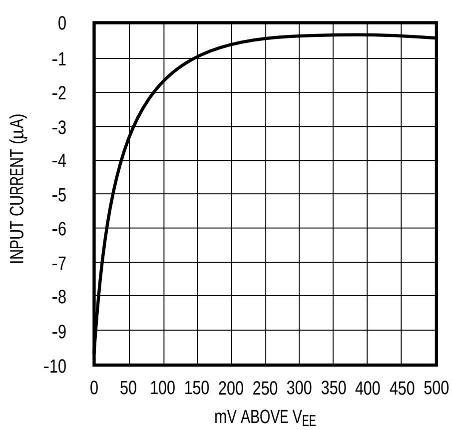


Figure 4. Enable Logic-Low Input Current vs. V_{IL} with 10kΩ Series Resistor

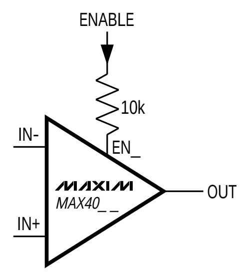


Figure 3. Circuit to Reduce Enable Logic-Low Input Current

To implement the mux function, the outputs of multiple amplifiers can be tied together, and only the amplifier with the selected input will be enabled. All of the other amplifiers will be placed in the low-power shutdown mode, with their high output impedance presenting very little load to the active amplifier output. For gains of +2 or greater, the feedback network impedance of all the amplifiers used in a mux application must be considered when calculating the total load on the active amplifier output

Output Capacitive Loading and Stability

The MAX4012/MAX4016/MAX4018/MAX4020 are optimized for AC performance. They are not designed to drive highly reactive loads, which decreases phase margin and may produce excessive ringing and oscillation. Figure 5 shows a circuit that eliminates this problem. Figure 6 is a graph of the optimal isolation resistor (R_S) vs. capacitive load. Figure 7 shows how a capacitive load causes excessive peaking of the amplifier's frequency response if the capacitor is not isolated from the amplifier by a resistor. A small isolation resistor (usually 20Ω to 30Ω) placed before the reactive load prevents ringing and oscillation. At higher capacitive loads, AC performance is controlled by the interaction of the load capacitance and the isolation resistor. Figure 8 shows the effect of a 27Ω isolation resistor on closed-loop response.

Coaxial cable and other transmission lines are easily driven when properly terminated at both ends with their characteristic impedance. Driving back-terminated transmission lines essentially eliminates the line's capacitance.

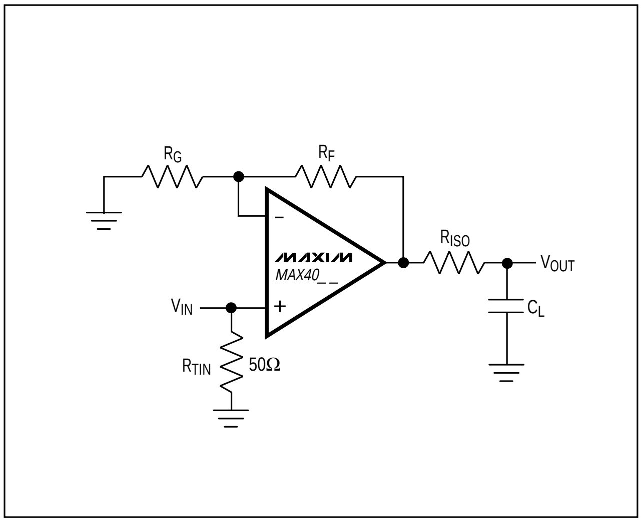


Figure 5. Driving a Capacitive Load through an Isolation Resistor

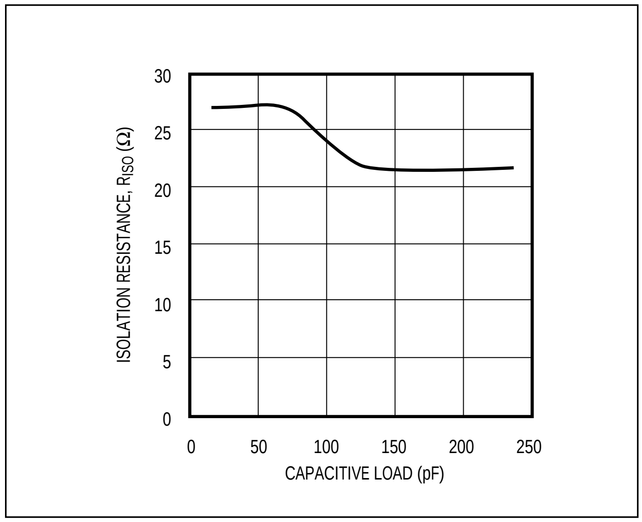


Figure 6. Capacitive Load vs. Isolation Resistance

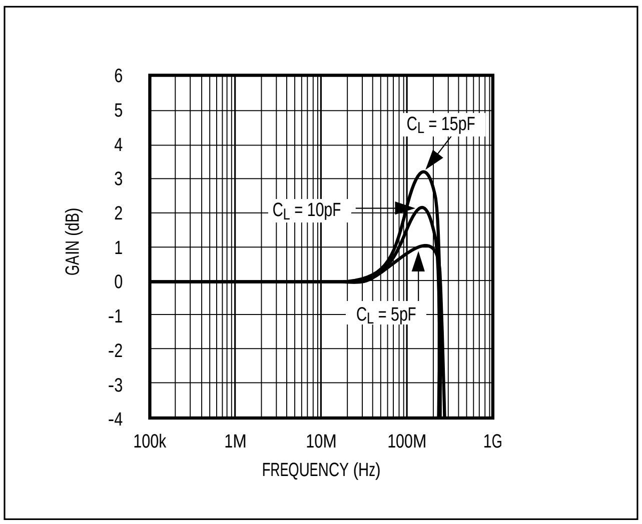


Figure 7. Small-Signal Gain vs. Frequency with Load Capacitance and No Isolation Resistor

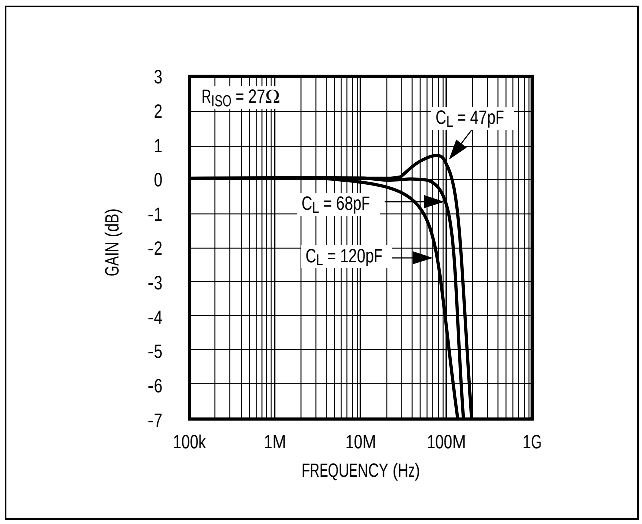
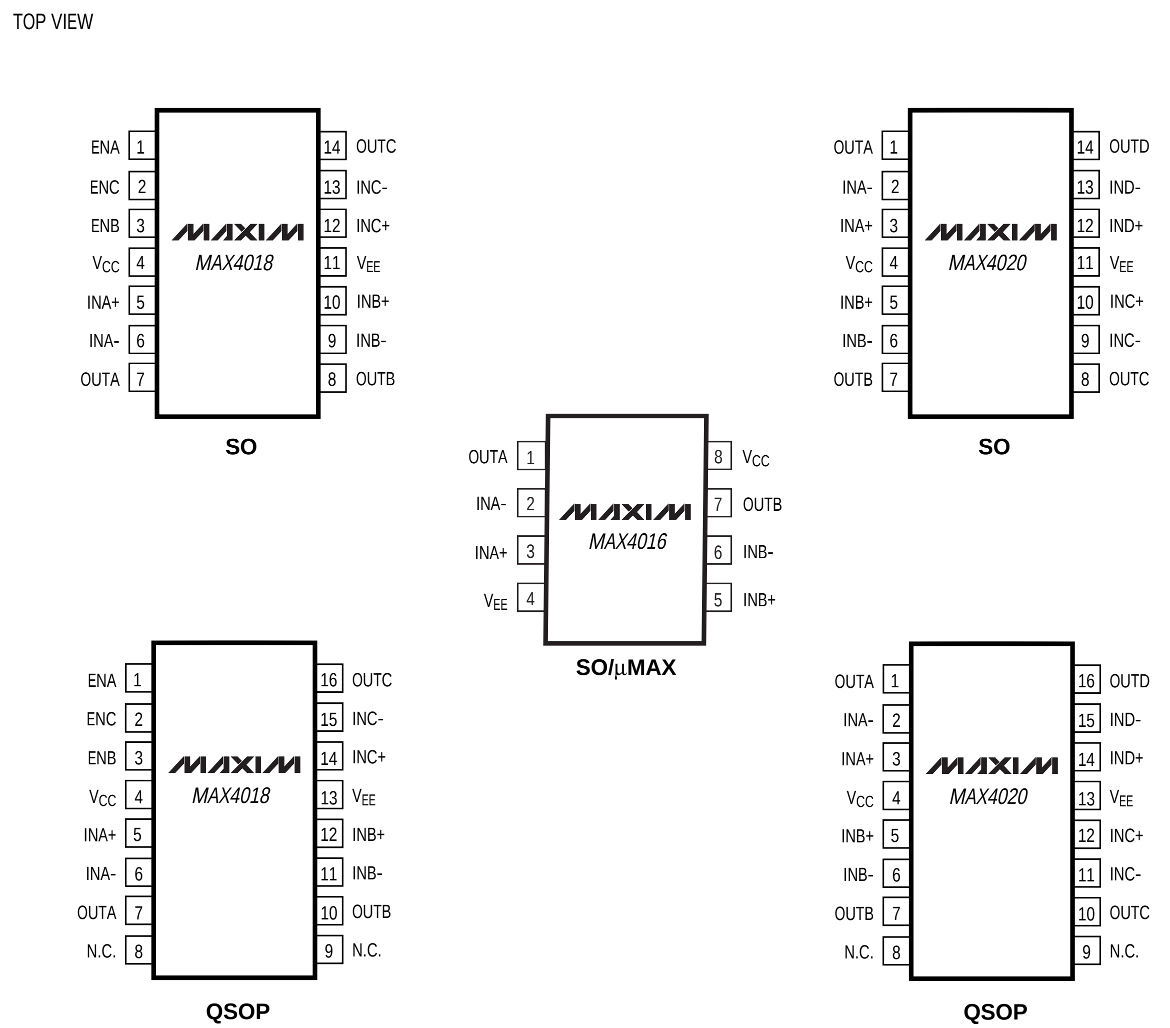


Figure 8. Small-Signal Gain vs. Frequency with Load Capacitance and 27Ω Isolation Resistor

Pin Configurations (continued)



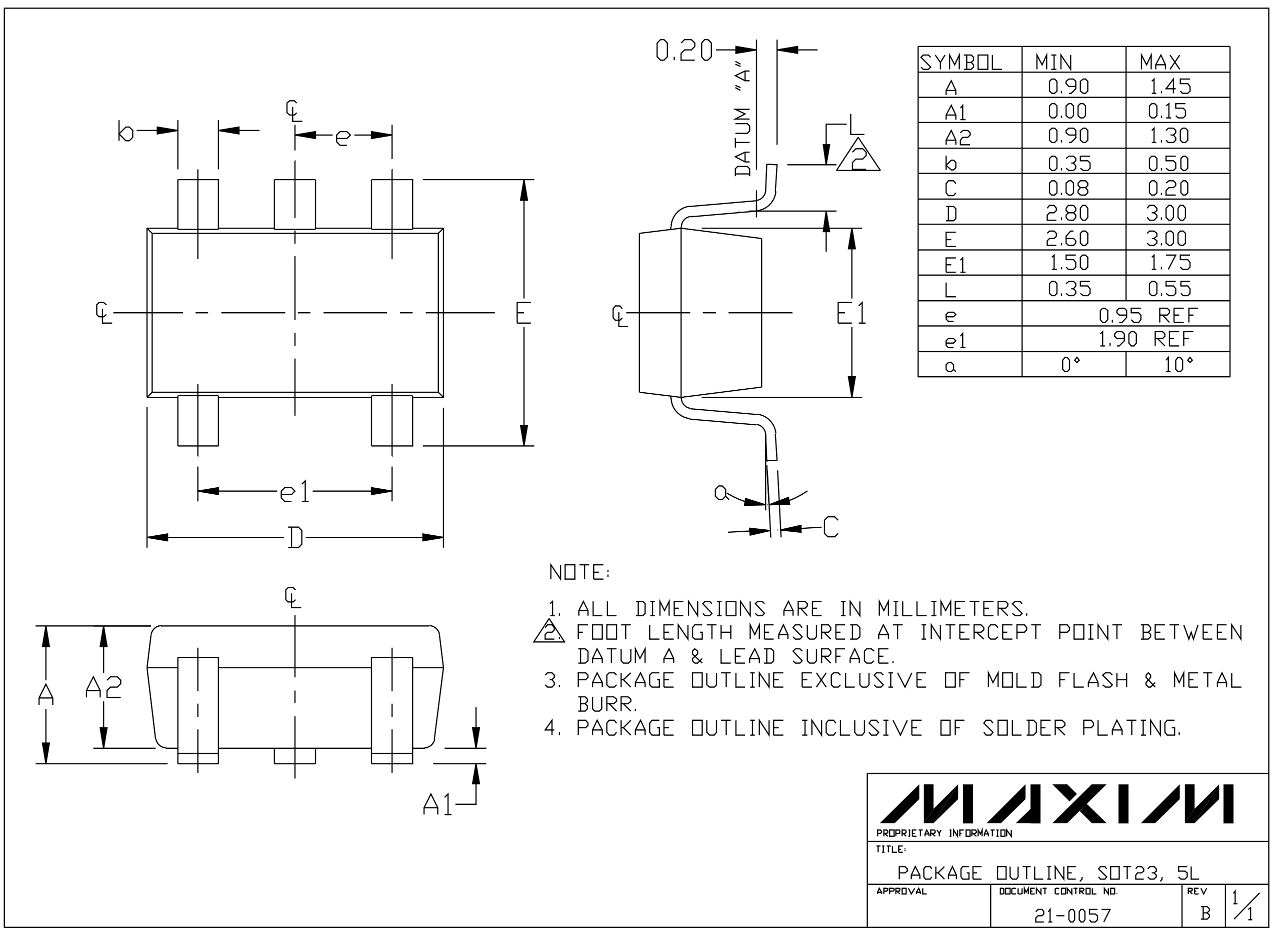
Ordering Information (continued)

PART	TEMP. RANGE	PIN-PACKAGE	SOT TOP MARK
MAX4018ESD	-40°C to +85°C	14 SO	—
MAX4018EEE	-40°C to +85°C	16 QSOP	—
MAX4020ESD	-40°C to +85°C	14 SO	—
MAX4020EEE	-40°C to +85°C	16 QSOP	—

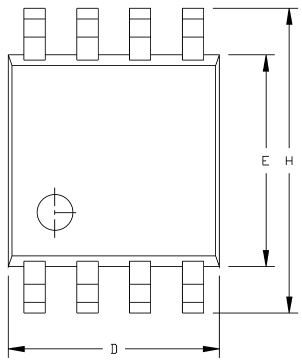
Chip Information

PART	TRANSISTOR COUNT
MAX4012	95
MAX4016	190
MAX4018	299
MAX4020	362

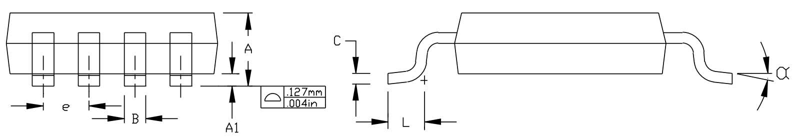
Package Information



Package Information (continued)



	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.036	0.044	0.91	1.11
A1	0.004	0.008	0.10	0.20
B	0.010	0.014	0.25	0.36
C	0.005	0.007	0.13	0.18
D	0.116	0.120	2.95	3.05
e	0.0256		0.65	
E	0.116	0.120	2.95	3.05
H	0.188	0.198	4.78	5.03
L	0.016	0.026	0.41	0.66
α	0°	6°	0°	6°

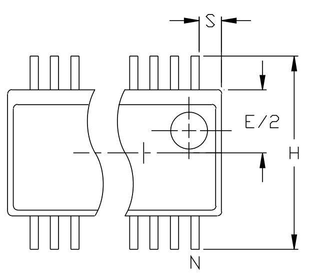


NOTES:
1. D&E DO NOT INCLUDE MOLD FLASH.
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED .15mm(.006").
3. CONTROLLING DIMENSION: INCHES

PROPRIETARY INFORMATION

TITLE: 8LD uMAX PACKAGE OUTLINE DWG.

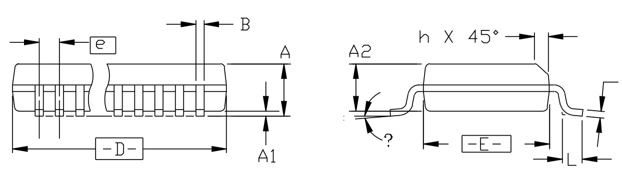
APPROVAL	DOCUMENT CONTROL NO.	REV	
	21-0036	D	1/1



	INCHES		MILLIMETERS	
DIM	MIN	MAX	MIN	MAX
A	.061	.068	1.55	1.73
A1	.004	.0098	0.127	0.25
A2	.055	.061	1.40	1.55
B	.008	.012	0.20	0.31
C	.0075	.0098	0.19	0.25
D	SEE VARIATIONS			
E	.150	.157	3.81	3.99
e	.025 BSC		0.635 BSC	
H	.230	.244	5.84	6.20
h	.010	.016	0.25	0.41
L	.016	.035	0.41	0.89
N	SEE VARIATIONS			
S	SEE VARIATIONS			
?	0°	8°	0°	8°

VARIATIONS:

	INCHES		MILLIMETERS			
	MIN.	MAX.	MIN.	MAX.	N	
D	.189	.196	4.80	4.98	16	AA
S	.0020	.0070	0.05	0.18		
D	.337	.344	8.56	8.74	20	AB
S	.0500	.0550	1.27	1.40		
D	.337	.344	8.56	8.74	24	AC
S	.0250	.0300	0.64	0.76		
D	.386	.393	9.80	9.98	28	AD
S	.0250	.0300	0.64	0.76		



NOTES:
1. D & E DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED .006"
3. CONTROLLING DIMENSIONS: INCHES

PROPRIETARY INFORMATION

TITLE: PACKAGE OUTLINE, QSOP, .150 INCH, .025" LEAD PITCH

APPROVAL	DOCUMENT CONTROL NO.	REV	
	21-0055	A	1/1

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