Am2909A/Am2911A

Microprogram Sequencers

DISTINCTIVE CHARACTERISTICS

- 4-bit slice cascadable to form longer word width
- Branch input for N-way branches
- 4 x 4 file with stack pointer and push/pop control for nesting microsubroutines
- Zero input for returning to the zero microcode word

- Individual OR input for each bit for branching to higher microinstructions (Am2909A only)
- Am2909A in 28-pin package & Am2911A in 20-pin package

MICROPROGRAM SEQUENCER BLOCK DIAGRAM
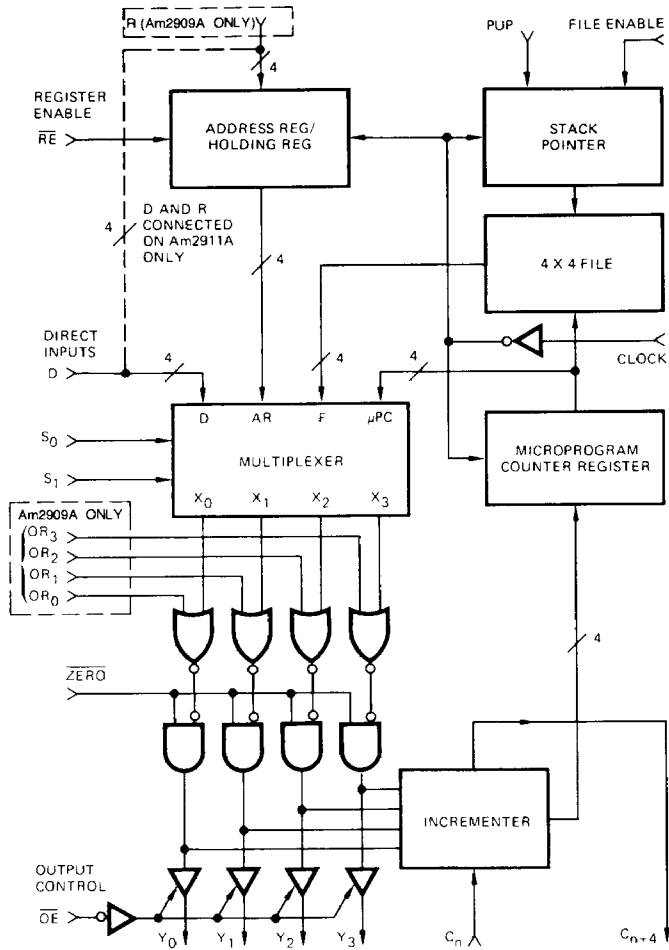
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GENERAL DESCRIPTION

The Am2909A is a 4-bit wide sequencer intended for sequencing through a series of microinstructions contained in memory. Two Am2909As may be interconnected to generate an 8-bit address (256 words), and three may be used to generate a 12-bit address (4K words).

The Am2909A can select an address from any of four sources: 1) a set of external direct inputs (D); 2) external data from the R inputs, stored in an internal register; 3) a 4-word deep push/pop stack; or 4) a program counter register (which usually contains the last address plus one). The push/pop stack includes certain control lines so that it can efficiently execute nested subroutine linkages. Each of the four outputs can be OR'ed with an external input for conditional skip or branch instructions, and a separate line forces the outputs to all zeros; the outputs are three-state.

The Am2911A is an identical circuit to the Am2909A, except the four OR inputs and the R inputs are removed and the D inputs connect to the ADDRESS REG/HOLDING REG. The Am2911A is in a 20-pin, 0.3" centers package.

RELATED AMD PRODUCTS

Part No.	Description
Am2918	Pipeline Register
Am2922	Condition Code MUX
Am29803A	16-Way Branch Control Unit
Am29811A	Next Address Control
Am25LS163	4-Bit Counter
Am27S35	Registered PROM

For applications information, see Chapter 11 of **Bit Slice Microprocessor Design,**
Mick & Brick, McGraw Hill Publications.

CONNECTION DIAGRAMS

Top View

Am2909A

DIPs

LCC*

CD004333

CD004312
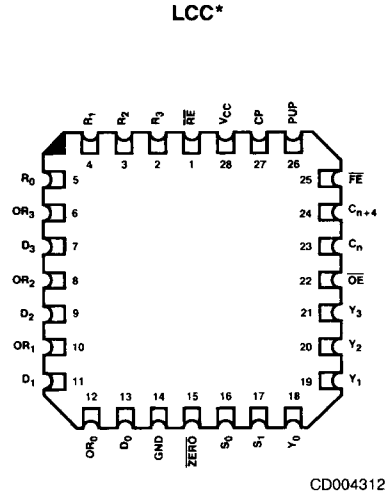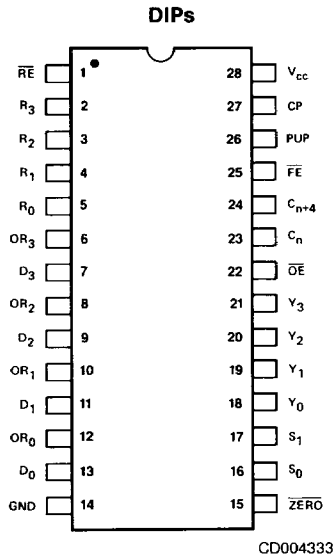
*Am2909A is also available in a 28-Pin Ceramic Flatpack; pinout is identical to the LCC.

Am2911A

DIPs

LCC

CD004322

CD004302
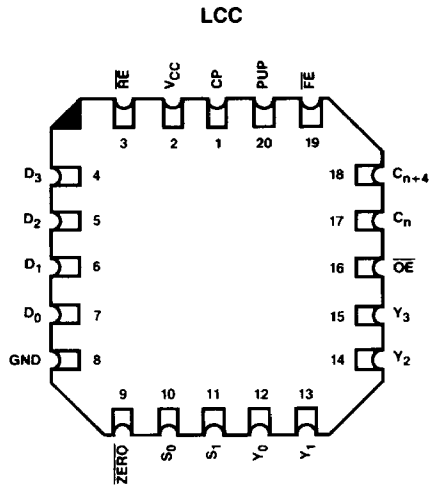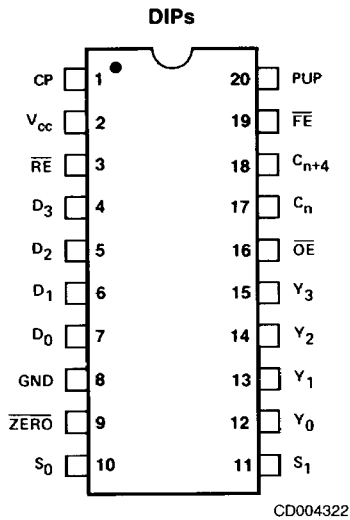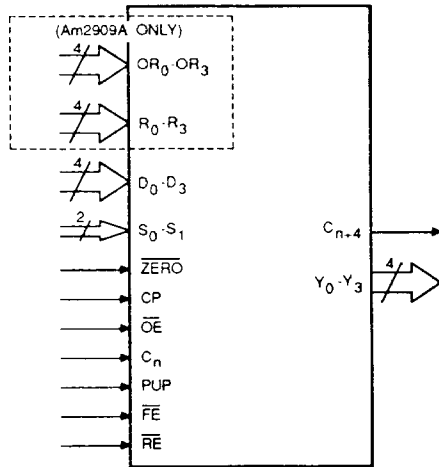
Note: Pin 1 is marked for orientation.

LOGIC SYMBOL

Am2909A/Am2911A

V_{CC} = Power Supply
GND = Ground

LS003110

Am2909A ORDERING INFORMATION

Standard Products

AMD standard products are available in several packages and operating ranges. The order number (Valid Combination) is formed by a combination of:

a. Device Number
b. Speed Option (if applicable)
c. Package Type
d. Temperature Range
e. Optional Processing

AM2909A   D   C   B

e. OPTIONAL PROCESSING
Blank = Standard processing
B = Burn-in

d. TEMPERATURE RANGE
C = Commercial (0 to +70°C)
M = Military* (−55 to +125°C)

c. PACKAGE TYPE
P = 28-Pin Plastic DIP (PD 028)
D = 28-Pin Ceramic DIP (CD 028)
X = Dice

b. SPEED OPTION
Not Applicable

a. DEVICE NUMBER/DESCRIPTION
Am2909A
Microprogram Sequencer

Valid Combinations	
AM2909A	PC, DC, DCB, XC, XM

*Military or Limited Military temperature range products are "NPL" (Non-Compliant Products List) or Non-MIL-STD-883C Compliant products only.

Valid Combinations

Valid Combinations list configurations planned to be supported in volume for this device. Consult the local AMD sales office to confirm availability of specific valid combinations, to check on newly released combinations, and to obtain additional data on AMD's standard military grade products.

Am2909A MILITARY ORDERING INFORMATION

APL Products

AMD products for Aerospace and Defense applications are available in several packages and operating ranges. APL (Approved Products List) products are fully compliant with MIL-STD-883C requirements. The order number (Valid Combination) for APL products is formed by a combination of: **a. Device Number**
b. Speed Option (if applicable)
c. Device Class
d. Package Type
e. Lead Finish

```
AM2909A              /B    X     A
```

e. LEAD FINISH
A = Hot Solder Dip

d. PACKAGE TYPE
X = 28-Pin Ceramic DIP (CD 028)
Y = 28-Pin Ceramic Flatpack (CFM028)
3 = 28-Pin Ceramic Leadless Chip
    Carrier (CLT028)

c. DEVICE CLASS
/B = Class B

b. SPEED OPTION
Not Applicable

a. DEVICE NUMBER/DESCRIPTION
Am2909A
Microprogram Sequencer

Valid Combinations	
AM2909A	/BXA, /BYA, /B3A

Valid Combinations

Valid Combinations list configurations planned to be supported in volume for this device. Consult the local AMD sales office to confirm availability of specific valid combinations or to check for newly released valid combinations.

Group A Tests
Group A tests consist of Subgroups
1, 2, 3, 7, 8, 9, 10, 11.

Am2911A ORDERING INFORMATION

Standard Products

AMD standard products are available in several packages and operating ranges. The order number (Valid Combination) is formed by a combination of: **a. Device Number**
b. Speed Option (if applicable)
c. Package Type
d. Temperature Range
e. Optional Processing

<u>AM2911A</u>   <u>D</u>   <u>C</u>   <u>B</u>

e. OPTIONAL PROCESSING
Blank = Standard processing
B = Burn-in

d. TEMPERATURE RANGE
C = Commercial (0 to +70°C)
M = Military* (−55 to +125°C)

c. PACKAGE TYPE
P = 20-Pin Plastic DIP (PD 020)
D = 20-Pin Ceramic DIP (CD 020)
X = Dice

b. SPEED OPTION
Not Applicable

a. DEVICE NUMBER/DESCRIPTION
Am2911A
Microprogram Sequencer

Valid Combinations	
AM2911A	PC, DC, DCB, XC, XM

*Military or Limited Military temperature range products are "NPL" (Non-Compliant Products List) or Non-MIL-STD-883C Compliant products only.

Valid Combinations

Valid Combinations list configurations planned to be supported in volume for this device. Consult the local AMD sales office to confirm availability of specific valid combinations, to check on newly released combinations, and to obtain additional data on AMD's standard military grade products.
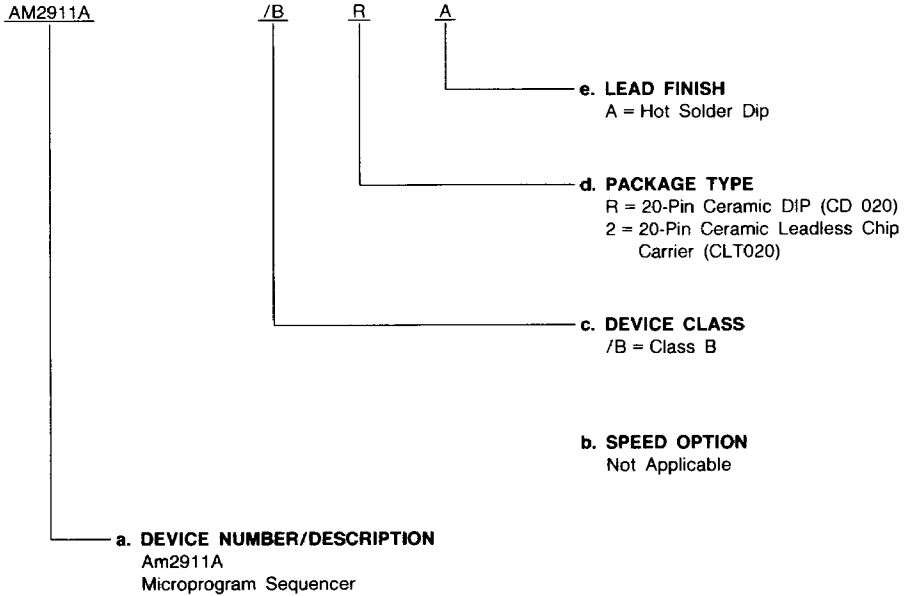
Am2911A MILITARY ORDERING INFORMATION

APL Products

AMD products for Aerospace and Defense applications are available in several packages and operating ranges. APL (Approved Products List) products are fully compliant with MIL-STD-883C requirements. The order number (Valid Combination) for APL products is formed by a combination of:
- **a. Device Number**
- **b. Speed Option** (if applicable)
- **c. Device Class**
- **d. Package Type**
- **e. Lead Finish**

```
AM2911A          /B    R    A
                            |
                            └──────  e. LEAD FINISH
                                         A = Hot Solder Dip

                       └──────────  d. PACKAGE TYPE
                                         R = 20-Pin Ceramic DIP (CD 020)
                                         2 = 20-Pin Ceramic Leadless Chip
                                             Carrier (CLT020)

                 └────────────────  c. DEVICE CLASS
                                         /B = Class B


                                     b. SPEED OPTION
                                         Not Applicable


    └──────  a. DEVICE NUMBER/DESCRIPTION
                 Am2911A
                 Microprogram Sequencer
```

Valid Combinations	
AM2911A	/BRA, /B2A

Valid Combinations

Valid Combinations list configurations planned to be supported in volume for this device. Consult the local AMD sales office to confirm availability of specific valid combinations or to check for newly released valid combinations.

Group A Tests
Group A tests consist of Subgroups
1, 2, 3, 7, 8, 9, 10, 11.

PIN DESCRIPTION

Am2909A/Am2911A

C_n Carry In (Input)
Carry input to the incrementer.

C_{n+4} Carry Out (Output)
Carry out from the incrementer.

CP   Clock Pulse (Input)
Clock input to the AR and μPC register and push/pop stack.

D_{0-3} Direct Input (Input)
Direct input to the multiplexer.

$\overline{FE}$ File Enable (Input; Active LOW)
Used along with PUP to control the push/pop stack.

$\overline{OE}$ Output Enable (Input; Three State, Active LOW)
Three-state control of Y_i outputs.

PUP   Push/Pop (Input)
Used along with $\overline{FE}$ to control the push/pop stack.

$\overline{RE}$ Register Load Enable (Input; Active LOW)
Enable input for internal address register.

S_0, S_1 Source (Input)
Control inputs to the mulitplexer that select the source of the next microinstruction address.

Y_{0-3} Address (Output; Three State)
Address outputs.

$\overline{ZERO}$ Zero (Input; Active LOW)
Forces the four Y outputs to binary zero (except when $\overline{OE}$ is HIGH).

Am2909A Only

OR_{0-3} Logic OR (Input)
Logic-OR inputs on each address output line.

R_{0-3} Register Input (Input)
Inputs to the internal address register.

FUNCTIONAL DESCRIPTION

Architecture of the Am2909A/Am2911A

The Am2909A/Am2911A are bipolar microprogram sequencers intended for use in high-speed microprocessor applications. The device is a cascadable 4-bit slice such that two devices allow addressing of up to 256 words of microprogram and three devices allow addressing of up to 4K words of microprogram. A detailed logic diagram is shown in Figure 1.

The device contains a 4-input multiplexer that is used to select either the address register, direct inputs, microprogram counter, or file as the source of the next microinstruction address. This multiplexer is controlled by the S_0 and S_1 inputs.

The address register consists of four D-type, edge-triggered flip-flops with a common clock enable. When the address register enable is LOW, new data is entered into the register on the clock LOW-to-HIGH transition. The address register is available at the multiplexer as a source for the next microinstruction address. The direct input is a 4-bit field of inputs to the multiplexer and can be selected as the next microinstruction address. On the Am2911A, the direct inputs are also used as inputs to the register. This allows an N-way branch where N is any word in the microcode.

The Am2909A/Am2911A contains a microprogram counter (μPC) that is composed of a 4-bit incrementer followed by a 4-bit register. The incrementer has carry-in (C_n) and carry-out (C_{n+4}) such that cascading to larger word lengths is straightforward. The μPC can be used in either of two ways. When the least significant carry-in to the incrementer is HIGH, the microprogram register is loaded on the next clock cycle with the current Y output word, plus one $(Y+1 \rightarrow \mu PC)$; thus sequential microinstructions can be executed. If this least significant C_n is LOW, the incrementer passes the Y output word unmodified and the microprogram register is loaded with the same Y word on the next clock cycle $(Y \rightarrow \mu PC)$. Thus, the same microinstruction can be executed any number of times by using the least significant C_n as the control.

The last source available at the multiplexer input is the 4 x 4 file (stack). The file is used to provide return address linkage when executing microsubroutines. The file contains a built-in stack pointer (SP) which always points to the last file word written. This allows stack reference operations (looping) to be performed without a push or pop.

The stack pointer operates as an up/down counter with separate push/pop and file enable inputs. When the file enable input is LOW and the push/pop input is HIGH, the push operation is enabled. This causes the stack pointer to increment and the file to be written with the required return linkage — the next microinstruction address following the subroutine jump which initiated the push.

If the file enable input is LOW and the push/pop control is LOW, a pop operation occurs. This implies the usage of the return linkage during this cycle and thus a return from subroutine. The next LOW-to-HIGH clock transition causes the stack pointer to decrement. If the file enable is HIGH, no action is taken by the stack pointer regardless of any other input.

The stack pointer linkage is such that any combination of pushes, pops, or stack references can be achieved. One microinstruction subroutines can be performed. Since the stack is four words deep, up to four microsubroutines can be nested.

The $\overline{ZERO}$ input is used to force the four outputs to the binary zero state. When the $\overline{ZERO}$ input is LOW, all Y outputs are LOW regardless of any other inputs except $\overline{OE}$. In the Am2909A, each Y output bit also has a separate OR input such that a conditional logic one can be forced at each Y output. This allows jumping to different microinstructions on programmed conditions.

The Am2909A/Am2911A feature three-state Y outputs. These can be particularly useful in designs requiring external equipment to provide automatic checkout of the microprocessor. The internal control can be placed in the high-impedance state, and preprogrammed sequences of microinstructions can be executed via external access to the control ROM/PROM.

Operation of the Am2909A/Am2911A

Figure 2 lists the select codes for the multiplexer. The two bits applied from the microword register (and additional combinational logic for branching) determine which data source contains the address for the next microinstruction. The contents of the selected source will appear on the Y outputs. Figure 1 also shows the truth table for the output control and for the control of the push/pop stack. Figure 2 shows in detail the effect of S_0, S_1, $\overline{FE}$ and PUP on the Am2909A. These four signals define what address appears on the Y outputs and what the state of all the internal registers will be following the clock LOW-to-HIGH edge. In this illustration, the microprogram counter is assumed to contain initially some word J, the address register some word K, and the four words in the push/pop stack contain R_a through R_d.

Address Selection

S_1	S_0	Source for Y Outputs	Symbol
L	L	Microprogram Counter	μPC
L	H	Address/Holding Register	AR
H	L	Push-Pop stack	STK0
H	H	Direct inputs	D_i

Output Control

OR_i*	ZERO	$\overline{OE}$	Y_i
X	X	H	Z
X	L	L	L
H	H	L	H
L	H	L	Source selected by S_0 S_1

Z = High Impedance
*Am2909A only.

Synchronous Stack Control

$\overline{FE}$	PUP	Push/Pop Stack Change
H	X	No change
L	H	Increment stack pointer, then push current PC onto STK0
L	L	Pop stack (decrement stack pointer)

H = High
L = Low
X = Don't Care

Figure 1. Multiplexer Select Codes

Cycle	S_1, S_0, $\overline{FE}$, PUP	μPC	REG	STK0	STK1	STK2	STK3	Y_{OUT}	Comment	Principle Use
N	L L L L	J	K	Ra	Rb	Rc	Rd	J	Pop Stack	End
N + 1	—	J + 1	K	Rb	Rc	Rd	Ra	—		Loop
N	L L L H	J	K	Ra	Rb	Rc	Rd	J	Push μPC	Set-up
N + 1	—	J + 1	K	J	Ra	Rb	Rc	—		Loop
N	L L H X	J	K	Ra	Rb	Rc	Rd	J	Continue	Continue
N + 1	—	J + 1	K	Ra	Rb	Rc	Rd	—		
N	L H L L	J	K	Ra	Rb	Rc	Rd	K	Pop Stack;	End
N + 1	—	K + 1	K	Rb	Rc	Rd	Ra	—	Use AR for Address	Loop
N	L H L H	J	K	Ra	Rb	Rc	Rd	K	Push μPC;	JSR AR
N + 1	—	K + 1	K	J	Ra	Rb	Rc	—	Jump to Address in AR	
N	L H H X	J	K	Ra	Rb	Rc	Rd	K	Jump to Address in AR	JMP AR
N + 1	—	K + 1	K	Ra	Rb	Rc	Rd	—		
N	H L L L	J	K	Ra	Rb	Rc	Rd	Ra	Jump to Address in STK0;	RTS
N + 1	—	Ra + 1	K	Rb	Rc	Rd	Ra	—	Pop Stack	
N	H L L H	J	K	Ra	Rb	Rc	Rd	Ra	Jump to Address in STK0;	
N + 1	—	Ra + 1	K	J	Ra	Rb	Rc	—	Push μPC	
N	H L H X	J	K	Ra	Rb	Rc	Rd	Ra	Jump to Address in STK0	Stack Ref
N + 1	—	Ra + 1	K	Ra	Rb	Rc	Rd	—		(Loop)
N	H H L L	J	K	Ra	Rb	Rc	Rd	D	Pop Stack;	End
N + 1	—	D + 1	K	Rb	Rc	Rd	Ra	—	Jump to Address on D	Loop
N	H H L H	J	K	Ra	Rb	Rc	Rd	D	Jump to Address on D;	JSR D
N + 1	—	D + 1	K	J	Ra	Rb	Rc	—	Push μPC	
N	H H H X	J	K	Ra	Rb	Rc	Rd	D	Jump to Address on D	JMP D
N + 1	—	D + 1	K	Ra	Rb	Rc	Rd	—		

X = Don't care, 0 = LOW, 1 = HIGH, Assume C_n = HIGH
Note: STK0 is the location addressed by the stack pointer.

Figure 2. Output and Internal Next-Cycle Register States for Am2909A/Am2911A

Figure 3 illustrates the execution of a subroutine using the Am2909A. The configuration of Figure 6 is assumed. The instruction being executed at any given time is the one contained in the microword register (μWR). The contents of the μWR also control (indirectly, perhaps) the four signals S_0, S_1, $\overline{FE}$, and PUP. The starting address of the subroutine is applied to the D inputs of the Am2909A at the appropriate time.

In the columns on the left is the sequence of microinstructions to be executed. At address J + 2, the sequence control portion of the microinstruction contains the command "Jump to subroutine at A". At the time T_2, this instruction is in the μWR,

and the Am2909A inputs are set-up to execute the jump and save the return address. The subroutine address A is applied to the D inputs from the μWR and appears on the Y outputs. The first instruction of the subroutine, I(A), is accessed and is at the inputs of the μWR. On the next clock transition, I(A) is loaded into the μWR for execution, and the return address J + 3 is pushed onto the stack. The return instruction is executed at T_5. Figure 4 is a similar timing chart showing one subroutine linking to a second, the latter consisting of only one microinstruction.

Control Memory

Microprogram

Execute Cycle	Address	Sequencer Instruction
	J – 1	–
T_0	J	–
T_1	J + 1	–
T_2	J + 2	JSR A
T_6	J + 3	–
T_7	J + 4	–
	–	–
	–	–
	–	–
T_3	A	I(A)
T_4	A + 1	–
T_5	A + 2	RTS
	–	–
	–	–
	–	–
	–	–
	–	–

Execute Cycle		T_0	T_1	T_2	T_3	T_4	T_5	T_6	T_7	T_8	T_9
Clock Signals											
Am2909A Inputs (from μWR)	S_1, S_0	0	0	3	0	0	2	0	0		
	$\overline{FE}$	H	H	L	H	H	L	H	H		
	PUP	X	X	H	X	X	L	X	X		
	D	X	X	A	X	X	X	X	X		
Internal Registers	μPC	J + 1	J + 2	J + 3	A + 1	A + 2	A + 3	J + 4	J + 5		
	STK0	–	–	–	J + 3	J + 3	J + 3	–	–		
	STK1	–	–	–	–	–	–	–	–		
	STK2	–	–	–	–	–	–	–	–		
	STK3	–	–	–	–	–	–	–	–		
Am2909A Output	Y	J + 1	J + 2	A	A + 1	A + 2	J + 3	J + 4	J + 5		
ROM Output	(Y)	I(J + 1)	JSR A	I(A)	I(A + 1)	RTS	I(J + 3)	I(J + 4)	I(J + 5)		
Contents of μWR (Instruction being executed)	μWR	I(J)	I(J + 1)	JSR A	I(A)	I(A + 1)	RTS	I(J + 3)	I(J + 4)		

C_N = HIGH

Figure 3. Subroutine Execution

Control Memory

Microprogram

Execute Cycle	Address	Sequencer Instruction
	J – 1	–
T_0	J	–
T_1	J + 1	–
T_2	J + 2	JSR A
T_9	J + 3	–
	–	–
	–	–
	–	–
T_3	A	–
T_4	A + 1	–
T_5	A + 2	JSR B
T_7	A + 3	–
T_8	A + 4	RTS
	–	–
	–	–
	–	–
T_6	B	RTS
	–	–
	–	–

Execute Cycle		T_0	T_1	T_2	T_3	T_4	T_5	T_6	T_7	T_8	T_9
Clock Signals											
Am2909A Inputs (from μWR)	S_1, S_0	0	0	3	0	0	3	2	0	2	0
	$\overline{FE}$	H	H	L	H	H	L	L	H	L	H
	PUP	X	X	H	X	X	H	L	X	L	X
	D	X	X	A	X	X	B	X	X	X	X
Internal Registers	μPC	J + 1	J + 2	J + 3	A + 1	A + 2	A + 3	B + 1	A + 4	A + 5	J + 4
	STK0	–	–	–	J + 3	J + 3	J + 3	A + 3	J + 3	J + 3	–
	STK1	–	–	–	–	–	–	J + 3	–	–	–
	STK2	–	–	–	–	–	–	–	–	–	–
	STK3	–	–	–	–	–	–	–	–	–	–
Am2909A Output	Y	J + 1	J + 2	A	A + 1	A + 2	B	A + 3	A + 4	J + 3	J + 4
ROM Output	(Y)	I(J + 1)	JSR A	I(A)	I(A + 1)	JSR B	RTS	I(A + 3)	RTS	I(J + 3)	I(J + 4)
Contents of μWR (Instruction being executed)	μWR	I(J)	I(J + 1)	JSR A	I(A)	I(A + 1)	JSR B	RTS	I(A + 3)	RTS	I(J + 3)

C_N = HIGH

Figure 4. Two Nested Subroutines (Routine B is Only One Instruction)
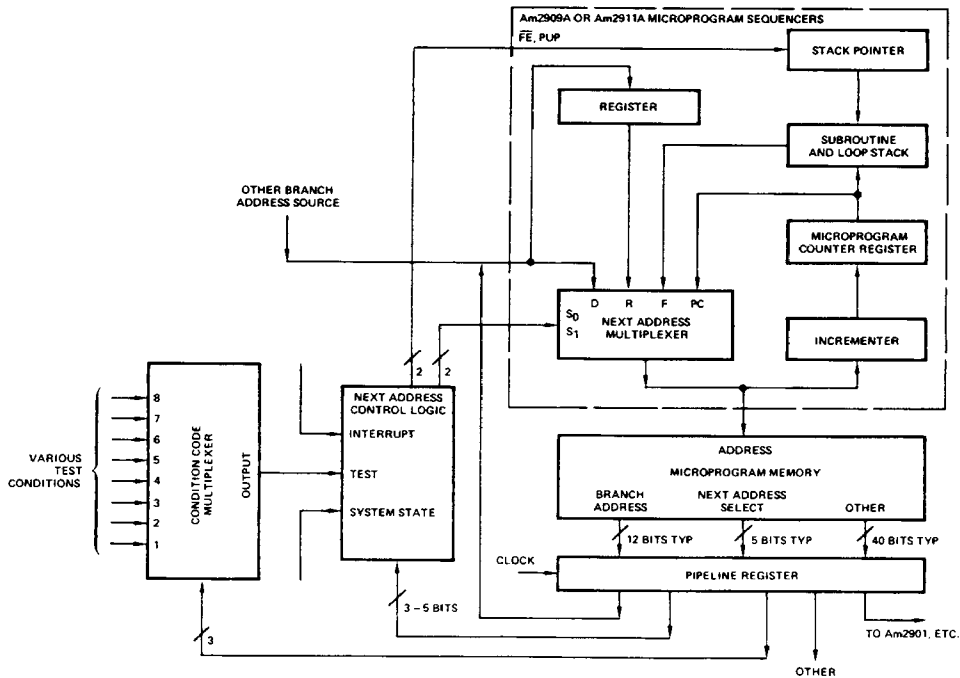
Using the Am2909A and Am2911A

The Am2909A and Am2911A are 4-bit slice sequencers which are cascaded to form a microprogram memory address generator. Both products make available to the user several lines which are used to directly control the internal holding register, multiplexer and stack. By appropriate control of these lines, the user can implement any desired set of sequence control functions; by cascading parts he can generate any desired address length. These two qualities set the Am2909A and Am2911A apart from the Am2910, which is architecturally similar, but is fixed at 12 bits in length and has a fixed set of 16 sequence control instructions. The Am2909A or Am2911A should be selected instead of the Am2910 under the following conditions:

● Address less than 8 bits and not likely to be expanded

● Address longer than 12 bits

● More complex instruction set needed than is available on Am2910

Architecture of the Control Unit

The recommended architecture using the Am2909A or Am2911A is shown in Figure 5. Note that the path from the pipeline register output through the next address logic, multiplexer, and microprogram memory is all combinational. The pipeline register contains the current microinstruction being executed. A portion of that microinstruction consists of a sequence control command such as "continue", "loop", "return-from-subroutine", etc. The bits representing this sequence command are logically combined with bits representing such things as test conditions and system state to generate the required control signals to the Am2909A or Am2911A. The block labeled "next address logic" may consist of simple gates, a PROM or a PLA, but it should be all combinational.
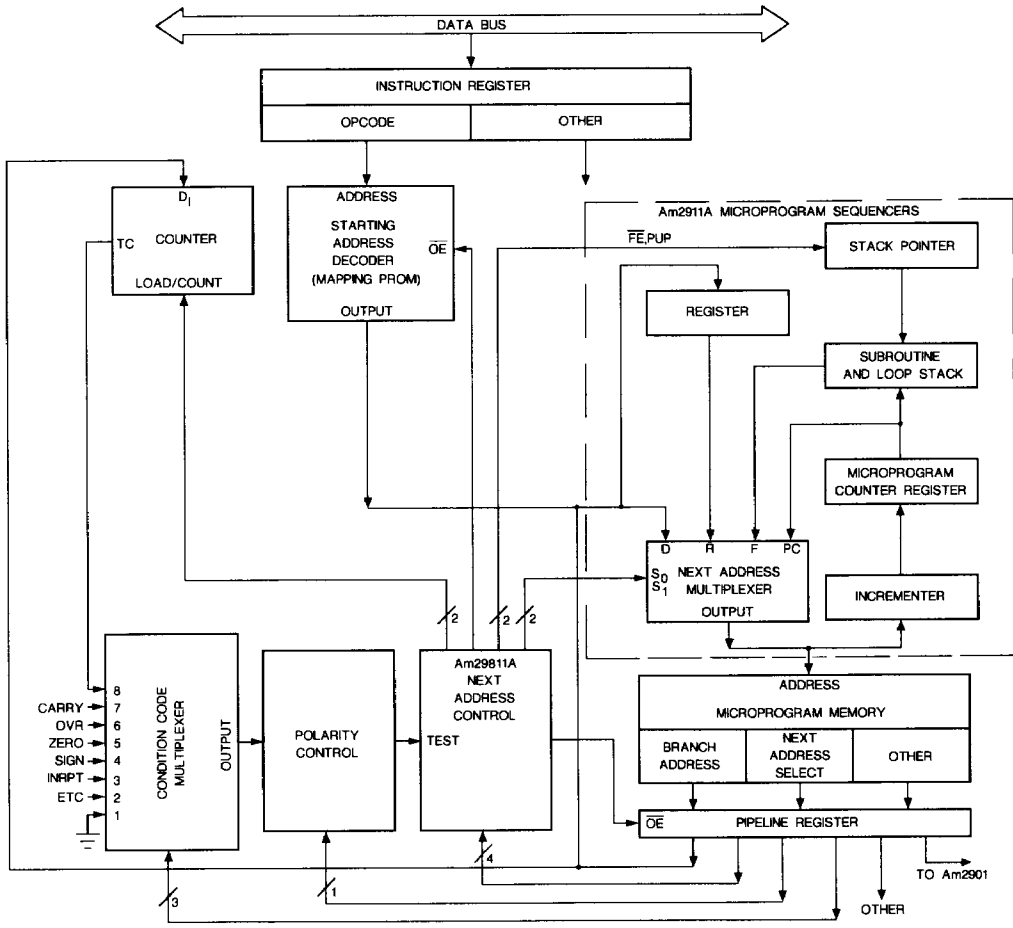

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Figure 5. Recommended Computer Control Unit Architecture Using the Am2911A or Am2909A

The Am29811A is a combinational circuit which implements 16 sequence control instructions; it may be used with either an Am2909A or an Am2911A. The set of instructions is nearly identical to that implemented internally in the Am2910.

Figure 6 shows the CCU of Figure 5 with the Am29811A in place. The Am29811A, in addition to controlling the Am2911A, also controls a loop counter and several branch address sources. The instructions which are implemented by the Am29811A are shown in Figure 7, along with the Am29811A outputs for each instruction. Generating any instruction set consists simply of writing a truth table and designing combinational logic to implement it. For more detailed information refer to "The Microprogramming Handbook".

BD007591

Figure 6. A Typical Computer Control Unit Using the Am2911A and Am29811A

Expansion of the Am2909A or Am2911A

Figure 8 shows the interconnection of three Am2911As to form a 12-bit sequencer. Note that the only interconnection between packages, other than the common clock and control lines, is the ripple carry between μPC incrementors. This carry path is not in the critical speed path if the Am2911A Y outputs drive the microprogram memory, because the ripple carry occurs in parallel with the memory access time. If, on the other hand, a micro-address register is placed at the Am2911A output, then the carry may lie in the critical speed path, since the last carry-in must be stable for a setup time prior to the clock.

Selecting Between the Am2909A and Am2911A

The difference between the Am2909A and the Am2911A involves two signals: the data inputs to the holding register and the "OR" inputs. In the Am2909A, separate four-bit fields are provided for the holding register and the direct branch inputs to the multiplexer. In the Am2911A, these fields are internally tied together. This may affect the design of the branch address system, as shown in Figure 9. Using the Am2909A, the register inputs may be connected directly to the microprogram memory; the internal register replaces part of the pipeline register. The direct (D) inputs may be tied to the mapping logic which translates instruction op codes into microprogram addresses. While the same technique might be used with the Am2911A, it is more common to connect the Am2911A's D inputs to a branch address bus onto which various sources may be enabled. Shown in Figure 9 is a pipeline register and a mapping ROM. Other sources might also be applied to the same bus. The internal register is used only for temporary storage of some previous branch address.

Mnemonic	Pin No.	I_3	I_2	I_1	I_0	TEST	S_1	S_0	$\overline{FE}$	PUP	$\overline{LOAD}$	$\overline{EN}$	MAP $\overline{E}$	PL $\overline{E}$
		14	13	12	11	10	4	5	3	2	6	7	1	9
JZ	JUMP ZERO	L	L	L	L	L	H	H	H	H	L	L	H	L
		L	L	L	L	H	H	H	H	H	L	L	H	L
CJS	CON JSB PL	L	L	L	H	L	L	L	H	H	H	H	H	L
		L	L	L	H	H	H	H	L	H	H	H	H	L
JMAP	JUMP MAP	L	L	H	L	L	H	H	H	H	H	H	L	H
		L	L	H	L	H	H	H	H	H	H	H	L	H
CJP	COND JUMP PL	L	L	H	H	L	L	L	H	H	H	H	H	L
		L	L	H	H	H	H	H	H	H	H	H	H	L
PUSH	PUSH/COND LD CNTR	L	H	L	L	L	L	L	L	H	H	H	H	L
		L	H	L	L	H	L	L	L	H	L	H	H	L
JSRP	COND JSB R/PL	L	H	L	H	L	L	H	L	H	H	H	H	L
		L	H	L	H	H	H	H	L	H	H	H	H	L
CVJ	COND JUMP VECTOR	L	H	H	L	L	L	L	H	H	H	H	H	H
		L	H	H	L	H	H	H	H	H	H	H	H	H
JRP	COND JUMP R/PL	L	H	H	H	L	L	L	H	H	H	H	H	L
		L	H	H	H	H	H	H	H	H	H	H	H	L
RFCT	REPEAT LOOP, CTR # 0	H	L	L	L	H	L	L	L	L	H	H	H	L
RPCT	REPEAT PL, CTR # 0	H	L	L	H	L	H	H	H	H	H	L	H	L
		H	L	L	H	H	L	L	H	H	H	H	H	L
CRTN	COND RTN	H	L	H	L	L	L	L	H	L	H	H	H	L
		H	L	H	L	H	H	L	L	L	H	H	H	L
CJPP	COND JUMP PL & POP	H	L	H	H	L	L	L	H	L	H	H	H	L
		H	L	H	H	H	H	H	L	L	H	H	H	L
LDCT	LD CNTR & CONTINUE	H	H	L	L	L	L	L	H	H	L	H	H	L
		H	H	L	L	H	L	L	H	H	L	H	H	L
LOOP	TEST END LOOP	H	H	L	H	L	H	L	H	L	H	H	H	L
		H	H	L	H	H	L	L	L	L	H	H	H	L
CONT	CONTINUE	H	H	H	L	L	L	L	H	H	H	H	H	L
		H	H	H	L	H	L	L	H	H	H	H	H	L
JP	JUMP PL	H	H	H	H	L	H	H	H	H	H	H	H	L
		H	H	H	H	H	H	H	H	H	H	H	H	L

L = LOW
H = HIGH

Figure 7. Am29811A Truth Table
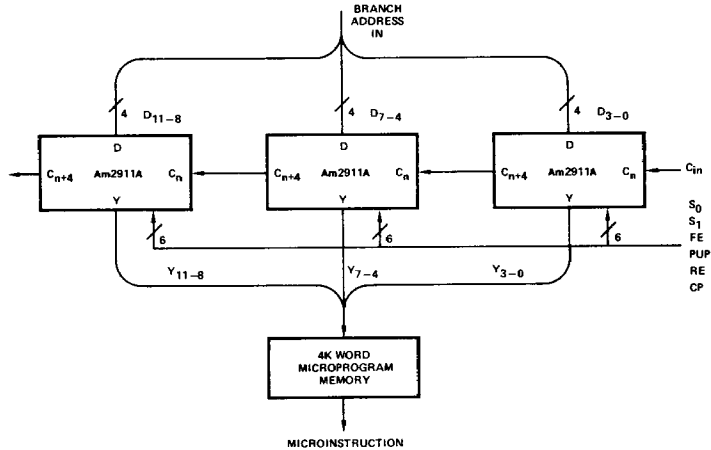
Figure 8. 12-Bit Sequencer
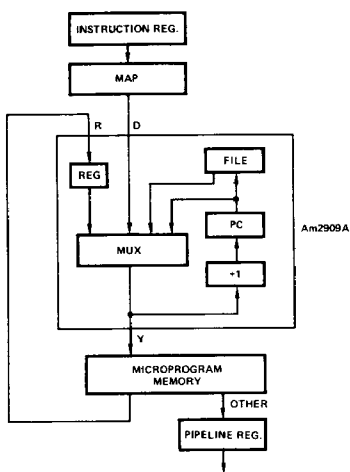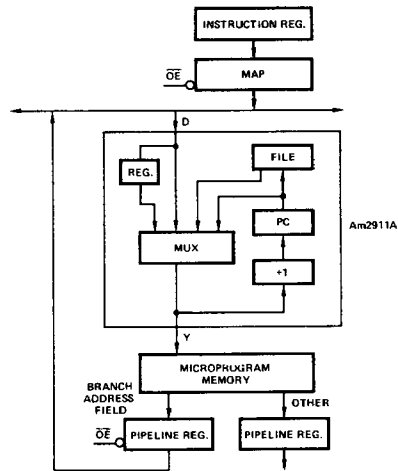
Figure 9. Branch Address Structures

The second difference between the Am2909A and Am2911A is that the Am2909A has OR inputs available on each address output line. These pins can be used to generate multi-way single-cycle branches by simply tying several test conditions into the OR lines (see Figure 10). Typically, a branch is taken to an address with zeros in the least significant bits. These bits are replaced with 1s or 0s by test conditions applied to the OR lines. In Figure 10, the states of the two test conditions X and Y result in a branch to 1100, 1101, 1110, or 1111.
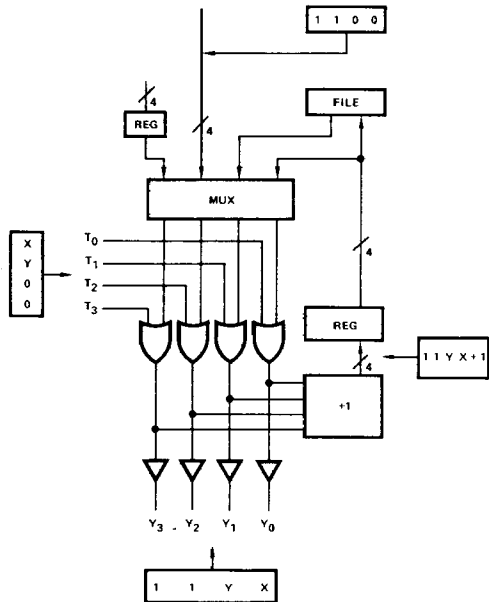
AF001610

Figure 10. Use of OR Inputs to Obtain Four-Way Branch

The Am29803A has been designed to selectively apply any or all of four different test conditions to an Am2909A. Figures 11-1 and 11-2 show the truth and function tables for this device. A nice trade-off between flexibility and board space is achieved by using a single 28-pin Am2909A for the least-significant 4 bits of a sequencer, and using the space-saving 20-pin Am2911As for the remainder of the bits. A detailed logic design for such a system is contained in "The Microprogramming Handbook."

AF001601

Figure 11-1. Am29803A Truth Table

How to Perform Some Common Functions with the Am2909A or Am2911A

1. CONTINUE

MUX/Y_{OUT}	STACK	C_n	S_1 S_0	$\overline{FE}$	PUP
PC	HOLD	H	L L	H	X

Contents of PC placed on Y outputs; PC incremented.

2. BRANCH

MUX/Y_{OUT}	STACK	C_n	S_1 S_0	$\overline{FE}$	PUP
D	HOLD	H	H H	H	X

Feed data on D inputs straight through to memory address lines. Increment address and place in PC.

3. JUMP-TO-SUBROUTINE

MUX/Y_{OUT}	STACK	C_n	S_1 S_0	$\overline{FE}$	PUP
D	PUSH	H	H H	L	H

Subroutine address fed from D inputs to memory address. Current PC is pushed onto stack, where it is saved for the return.

4. RETURN-FROM-SUBROUTINE

MUX/Y_{OUT}	STACK	C_n	S_1 S_0	$\overline{FE}$	PUP
STACK	POP	H	H L	L	L

The address at the top of the stack is applied to the microprogram memory, and is incremented for PC on the next cycle. The stack is popped to remove the return address.

	Branch On	I_3	I_2	I_1	I_0	OR_3	OR_2	OR_1	OR_0
NONE	NONE	L	L	L	L	L	L	L	L
Two-way Branches	T_0	L	L	L	H	L	L	L	T_0
	T_1	L	L	H	L	L	L	L	T_1
	T_2	L	H	L	L	L	L	L	T_2
	T_3	H	L	L	L	L	L	L	T_3
Four-Way Branches	T_1 & T_0	L	L	H	H	L	L	T_1	T_0
	T_2 & T_0	L	H	L	H	L	L	T_2	T_0
	T_3 & T_0	H	L	L	H	L	L	T_3	T_0
	T_2 & T_1	L	H	H	L	L	L	T_2	T_1
	T_3 & T_1	H	L	H	L	L	L	T_3	T_1
	T_3 & T_2	H	H	L	L	L	L	T_3	T_2
Eight-Way Branches	T_2, T_1, T_0	L	H	H	H	L	T_2	T_1	T_0
	T_3, T_1, T_0	H	L	H	H	L	T_3	T_1	T_0
	T_3, T_2, T_0	H	H	L	H	L	T_3	T_2	T_0
	T_3, T_2, T_1	H	H	H	L	L	T_3	T_2	T_1
Sixteen-Way Branch	T_3, T_2, T_1, T_0	H	H	H	H	T_3	T_2	T_1	T_0

Figure 11-2. Am29803A Function Table

ABSOLUTE MAXIMUM RATINGS

Storage Temperature .... −65 to +150°C
(Ambient) Temperature Under Bias .... −55 to +125°C
Supply Voltage to Ground Potential
  Continuous .... −0.5 V to +7.0 V
DC Voltage Applied to Outputs For
  High Output State .... −0.5 V to +V_{CC} max
DC Input Voltage .... −0.5 V to +7.0 V
DC Output Current, Into Outputs .... 30 mA
DC Input Current .... −30 mA to +5.0 mA

Stresses above those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent device failure. Functionality at or above these limits is not implied. Exposure to absolute maximum ratings for extended periods may affect device reliability.

OPERATING RANGES

Commercial (C) Devices
  Ambient Temperature (T_A) .... 0 to +70°C
  Supply Voltage (V_{CC}) .... +4.75 V to +5.25 V
Military (M) Devices
  Case Temperature (T_C) .... −55 to +125°C
  Supply Voltage .... +4.5 V to +5.5 V

Operating ranges define those limits between which the functionality of the device is guaranteed.

Thermal Resistance (Typical)
Am2909A

Symbol	PD 028	CD 028	CLT028	CFM028	Unit
θ_{JA}	62	55	76	82	°C/W
θ_{JC}	NA	5	7	8	°C/W

Am2911A

Symbol	PD 020	CD 020	CLT020	Unit
θ_{JA}	76	74	84	°C/W
θ_{JC}	NA	11	16	°C/W

*Military product tested at T_C = +25°C, +125°C, and −55°C.

DC CHARACTERISTICS over operating range unless otherwise specified (for APL products, Group A, Subgroups 1, 2, 3 are tested unless otherwise noted)

Parameter Symbol	Parameter Description	Test Conditions (Note 1)		Min.	Max.	Unit
V_{OH}	Output HIGH Voltage	V_{CC} = MIN, V_{IN} = V_{IH} or V_{IL}	MIL I_{OH} = −1.0 mA	2.4		V
			COM'L I_{OH} = −2.6 mA	2.4		
V_{OL}	Output LOW Voltage	V_{CC} = MIN, V_{IN} = V_{IH} or V_{IL}	I_{OL} = 16 mA		0.5	V
V_{IH}	Input HIGH Level	Guaranteed Input Logical-HIGH Voltage for All Inputs		2.0		V
V_{IL}	Input LOW-Level	Guaranteed Input Logical LOW Voltage for All Inputs			0.8	V
V_I	Input Clamp Voltage	V_{CC} = Min., I_{IN} = −18 mA			−1.5	V
I_{IL}	Input LOW Current	V_{CC} = Max., V_{IN} = 0.4 V	C_n		−1.08	mA
			Push/Pop, $\overline{OE}$		−0.72	
			Others (Note 4)		−0.36	
I_{IH}	Input HIGH Current	V_{CC} = Max., V_{IN} = 2.7 V	C_n		40	μA
			Push/Pop, $\overline{OE}$		40	
			Others (Note 4)		20	
I_I	Input HIGH Current	V_{CC} = Max., V_{IN} = 5.5 V	C_n, Push/Pop		0.2	mA
			Others (Note 4)		0.1	
I_{OS}	Output Short Circuit Current (Note 2)	V_{CC} = Max.+0.5 V V_{OUT} = 0.5V	Y_0-Y_3	−30	−100	mA
			C_{n+4}	−30	−85	
I_{CC}	Power Supply Current	V_{CC} = Max. (Note 3)	COM'L Only T_A = 0 to +70°C		130	mA
			MIL Only T_C = −55 to +125°C		140	
			T_C = +125°C		110	
I_{OZL}	Output OFF Current	V_{CC} = Max, $\overline{OE}$ = 2.7	Y_{0-3} V_{OUT} = 0.4 V		−20	μA
I_{OZH}			V_{OUT} = 2.7 V		20	

Notes: 1. For conditions shown as Min. or Max., use the appropriate value specified under Operating Ranges for the applicable device type.
    2. Not more than one output should be shorted at a time. Duration of the short-circuit test should not exceed one second.
    3. Apply GND to C_n, R_0, R_1, R_2, R_3, OR_0, OR_1, OR_2, OR_3, D_0, D_1, D_2, and D_3. Other inputs high. All outputs open. Measured after a LOW-to-HIGH clock transition.
    4. For the Am2911A, D_i and R_i are internally connected. Loading is doubled (to same values as Push/Pop).

SWITCHING CHARACTERISTICS over

operating ranges unless otherwise specified (for APL Products, Group A, Subgroups 9, 10, 11 are tested unless otherwise noted)

Tables A, B, and C below define the timing characteristics of the Am2909A/Am2911A over the operating voltage and temperature range. The tables are divided into three types of parameters: clock characteristics, combinational delays from inputs to outputs, and setup and hold-time requirements. The latter table defines the time prior to the end of the cycle (i.e., clock LOW-to-HIGH transition) that each input must be stable to guarantee that the correct data is written into one of the internal registers.

Measurements are made at 1.5 V with $V_{IL} = 0$ V and $V_{IH} = 3.0$ V. For three-state disable tests, $C_L = 5.0$ pF and measurement is to 0.5 V change on output voltage level. All outputs have maximum DC loading.

A. Cycle Time and Clock Characteristics

Time	COMMERCIAL	MILITARY
Minimum Clock LOW Time	20	20
Minimum Clock HIGH Time	20	20

B. Maximum Combinational Propagation Delays

(all in ns, $C_L = 50$ pF (except output disable tests))

From Input	COMMERCIAL		MILITARY	
	Y	C_{n+4}	Y	C_{n+4}
D_i	17	22	20	25
S_0, S_1	29	34	29	34
OR_i**	17	22	20	25
C_n	–	14	–	16
$\overline{ZERO}$	29	34	30	35
OE LOW (enable)	25	–	25	–
OE HIGH (disable)*	25	–	25	–
Clock ↑ $S_1 S_0$ = LH	39	44	45	50
Clock ↑ $S_1 S_0$ = LL	39	44	45	50
Clock ↑ $S_1 S_0$ = HL	44	49	53	58

*$C_L = 5$ pF

C. Minimum Setup and Hold Times (all in ns) (Note 1)

From Input	Notes	COMMERCIAL		MILITARY	
		Setup Time	Hold Time	Setup Time	Hold Time
$\overline{RE}$		19	4	19	5
R_i**	2	10	4	12	5
PUP		25	4	27	5
$\overline{FE}$		25	4	27	5
C_n		18	4	18	5
D_i		25	0	25	0
OR_i**		25	0	25	0
S_0, S_1		25	0	29	0
$\overline{ZERO}$		25	0	29	0

Notes: 1. All times relative to clock LOW-to-HIGH transition.
    2. On Am2911A, R_i and D_i are internally connected together and labeled D_i. Use R_i set-up and hold times when D inputs are used to load register.

**Am2909A only.

SWITCHING WAVEFORMS

Key to Switching Waveforms

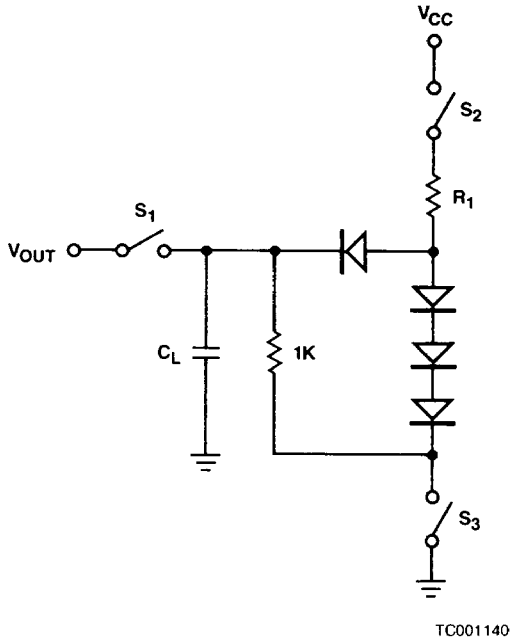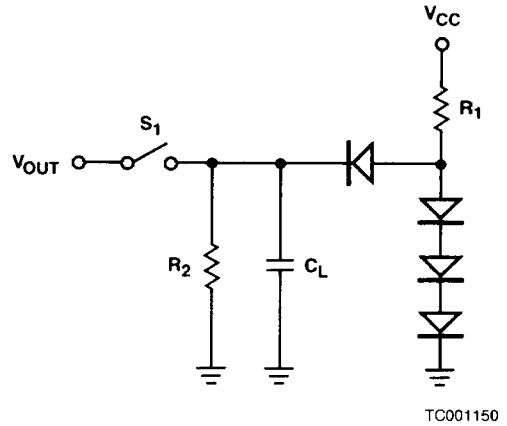| WAVEFORM	INPUTS	OUTPUTS
———	MUST BE STEADY	WILL BE STEADY
⬛	MAY CHANGE FROM H TO L	WILL BE CHANGING FROM H TO L
⬛	MAY CHANGE FROM L TO H	WILL BE CHANGING FROM L TO H
⬛	DON'T CARE, ANY CHANGE PERMITTED	CHANGING: STATE UNKNOWN
⬛	DOES NOT APPLY	CENTER LINE IS HIGH IMPEDANCE "OFF" STATE

KS000010
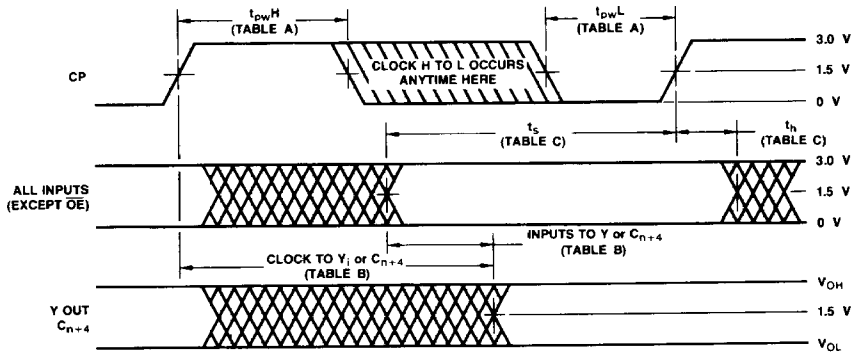
WF002601

WFR02992

19
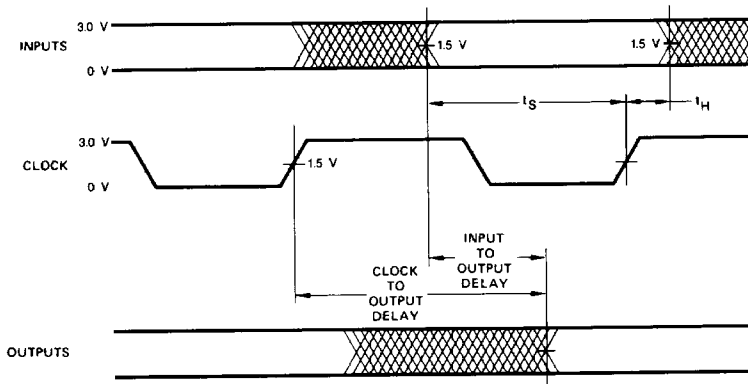
SWITCHING TEST CIRCUITS

TC001140

TC001150

$$R_1 = \frac{5.0 - V_{BE} - V_{OL}}{I_{OL} + V_{OL}/1K}$$

$$R_2 = \frac{2.4V}{I_{OH}}$$

$$R_1 = \frac{5.0 - V_{BE} - V_{OL}}{I_{OL} + V_{OL}/R_2}$$

A. Three-State Outputs

B. Normal Outputs

Notes: 1. C_L = 50 pF includes scope probe, wiring and stray capacitances without device in text fixture.
2. S_1, S_2, S_3 are closed during function tests all and AC tests except output enable tests.
3. S_1 and 3_3 are closed while S_2 is open for t_{PZH} test.
   S_1 and S_2 are closed while S_3 is open for t_{PZL} test.
4. C_L = 5.0 pF for output disable tests.

TEST OUTPUT LOADS

Device	Pin #	Pin Label	Test Circuit	R_1	R_2
Am2909A	18-21				
Am2911A	12-15	Y_{0-3}	A	220	1K
Am2909A	24				
AM2911A	18	C_{n+4}	B	220	2.4K

20

TEST PHILOSOPHY AND METHODS

The following nine points describe AMD's philosophy for high volume, high speed automatic testing.

1. Ensure that the part is adequately decoupled at the test head. Large changes in V_{CC} current as the device switches may cause erroneous function failures due to V_{CC} changes.

2. Do not leave inputs floating during any tests, as they may start to oscillate at high frequency.

3. Do not attempt to perform threshold tests at high speed. Following an output transition, ground current may change by as much as 400 mA in 5-8 ns. Inductance in the ground cable may allow the ground pin at the device to rise by hundreds of millivolts momentarily.

4. Use extreme care in defining point input levels for AC tests. Many inputs may be changed at once, so there will be significant noise at the device pins and they may not actually reach V_{IL} or V_{IH} until the noise has settled. AMD recommends using $V_{IL} \leqslant 0$ V and $V_{IH} \geqslant 3.0$ V for AC tests.

5. To simplify failure analysis, programs should be designed to perform DC, Function, and AC tests as three distinct groups of tests.

6. Capacitive Loading for AC Testing

   Automatic testers and their associated hardware have stray capacitance that varies from one type of tester to another, but is generally around 50 pF. This, of course, makes it impossible to make direct measurements of parameters which call for smaller capacitive load than the associated stray capacitance. Typical examples of this are the so-called "float delays," which measure the propagation delays into the high-impedance state and are usually specified at a load capacitance of 5.0 pF. In these cases, the test is performed at the higher load capacitance (typically 50 pF), and engineering correlations based on data taken with a bench setup are used to predict the result at the lower capacitance.

   Similarly, a product may be specified at more than one capacitive load. Since the typical automatic tester is not capable of switching loads in mid-test, it is impossible to make measurements at both capacitances even though they may both be greater than the stray capacitance. In these cases, a measurement is made at one of the two capacitances. The result at the other capacitance is predicted from engineering correlations based on data taken with a bench setup and the knowledge that certain DC measurements (I_{OH}, I_{OL} for example) have already been taken and are within spec. In some cases, special DC tests are performed in order to facilitate this correlation.

7. Threshold Testing

   The noise associated with automatic testing (due to the long, inductive cables) and the high gain of the tested device when in the vicinity of the actual device threshold, frequently give rise to oscillations when testing high speed circuits. These oscillations are not indicative of a reject device, but instead of an overtaxed test system. To minimize this problem, thresholds are tested at least once for each input pin. Thereafter, "hard" high and low levels are used for other tests. Generally this means that function and AC testing are performed at "hard" input levels rather than at V_{IL} Max. and V_{IH} Min.

8. AC Testing

   Occasionally, parameters are specified that cannot be measured directly on automatic testers because of tester limitations. Data input hold times often fall into this category. In these cases, the parameter in question is guaranteed by correlating these tests with other AC tests that have been performed. These correlations are arrived at by the cognizant engineer by using precise bench measurements in conjunction with the knowledge that certain DC parameters have already been measured and are within spec.
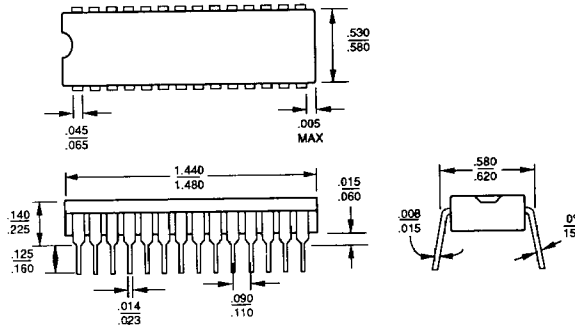
   In some cases, certain AC tests are redundant, since they can be shown to be predicted by some other tests which have already been performed. In these cases, the redundant tests are not performed.

9. Output Short-Circuit Current Testing

   When performing I_{OS} tests on devices containing RAM or registers, great care must be taken that undershoot caused by grounding the high-state output does not trigger parasitic elements which in turn cause the device to change state. In order to avoid this effect, it is common to make the measurement at a voltage (V_{OUTPUT}) that is slightly above ground. The V_{CC} is raised by the same amount so that the result (as confirmed by Ohm's law and precise bench testing) is identical to the $V_{OUT} = 0$, $V_{CC} =$ Max. case.

PHYSICAL DIMENSIONS

PD 028

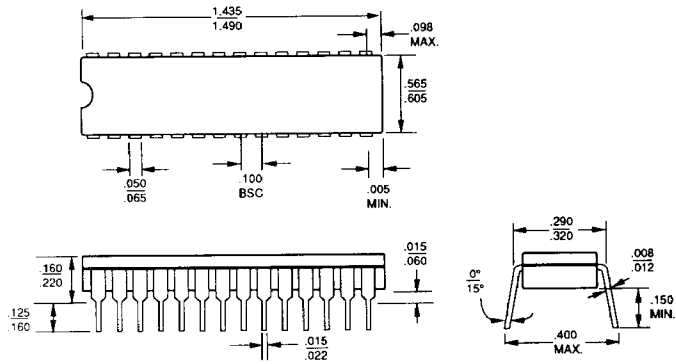.530/.580

.005 MAX

.045/.065

1.440/1.480

.015/.060

.580/.620

.140/.225

.008/.015

.125/.160

0°/15°

.014/.023

.090/.110

06842B

*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

CD 028

.290
.320

.008
.012

0°
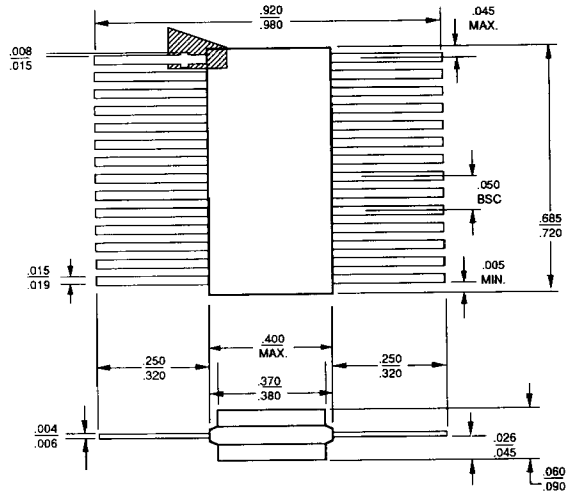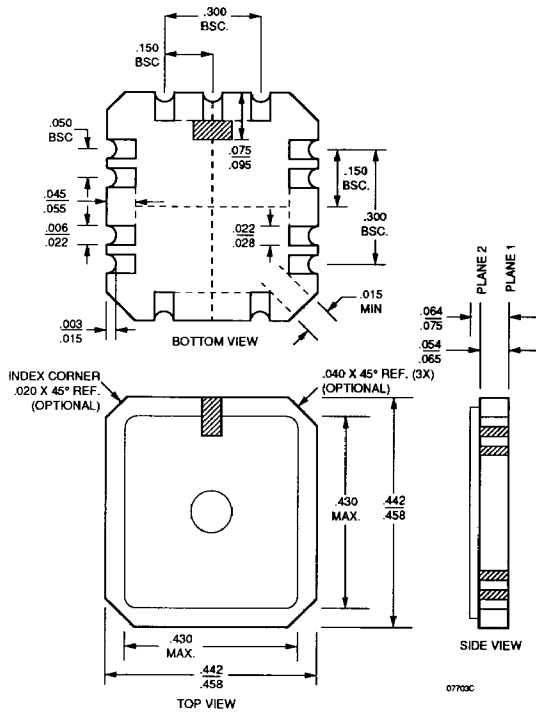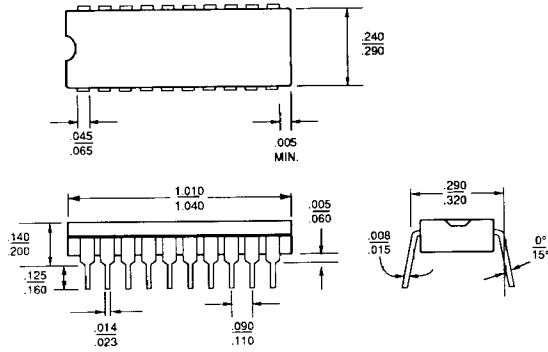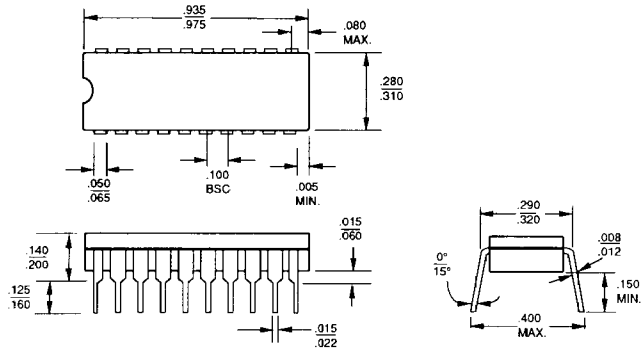15°

.150
MIN.

.400
MAX.

1.435
1.490

.098
MAX.

.565
.605

.050
.065

.100
BSC

.005
MIN.

.015
.060

160
220

.125
.160

.015
.022

06837C
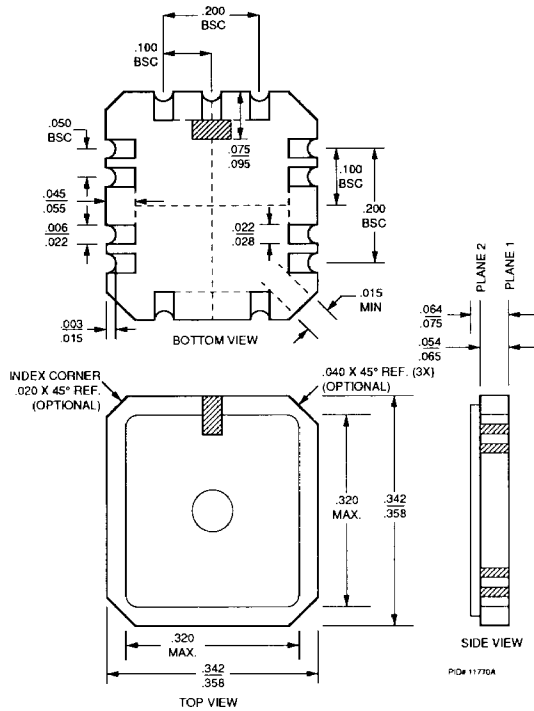
*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

CFM028

.920 / .980

.045 MAX.

.008 / .015

.050 BSC

.685 / .720

.015 / .019

.005 MIN.

.400 MAX.

.250 / .320

.250 / .320

.370 / .380

.004 / .006

.026 / .045

.060 / .090

097028

*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

24

CLT028

.300
BSC.

.150
BSC

.050
BSC

.075
.095

.150
BSC.

.300
BSC.

.045
.055

.006
.022

.022
.028

PLANE 2

PLANE 1

.003
.015

BOTTOM VIEW

.015
MIN

.064
.075

.054
.065

INDEX CORNER
.020 X 45° REF.
(OPTIONAL)

.040 X 45° REF. (3X)
(OPTIONAL)

.430
MAX.

.442
.458

.430
MAX.

.442
.458

TOP VIEW

SIDE VIEW

07703C

*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

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PD 020

.240 / .290

.045 / .065

.005 MIN.

1.010 / 1.040

.005 / .060

.290 / .320

.140 / .200

.008 / .015

0° / 15°

.125 / .160

.014 / .023

.090 / .110

07552A

*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

CD 020

PiD# 07553B

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*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

PHYSICAL DIMENSIONS (Cont'd.)

CLT020

TOP VIEW

BOTTOM VIEW

SIDE VIEW

INDEX CORNER
.020 X 45° REF.
(OPTIONAL)

.040 X 45° REF. (3X)
(OPTIONAL)

PLANE 2   PLANE 1

PID# 11770A

1/89
AN 28
CD

*For reference only. All dimensions are measured in inches. BSC is an ANSI standard for Basic Space Centering.

026702

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