## Renesns



## SH7040，SH7041，SH7042， SH7043，SH7044，SH7055 Group <br> Hardware Manual

（CPU Core SH－2）

Renesas 32-Bit Single-Chip RISC Microprocessor
SuperH RISC engine Family/ SH7040 Series (CPU Core SH-2)

SH7040, SH7041, SH7042, SH7043, SH7044, SH7055 Group

Hardware Manual

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## Preface

The SH7040 Series (SH7040, SH7041, SH7042, SH7043, SH7044, SH7045) single-chip RISC (Reduced Instruction Set Computer) microprocessors integrate a Renesas Technology-original RISC CPU core with peripheral functions required for system configuration.

The CPU has a RISC-type instruction set. Most instructions can be executed in one clock cycle, which greatly improves instruction execution speed. In addition, the 32-bit internal-bus architecture enhances data processing power. With this CPU, it has become possible to assemble low cost, high performance/high-functioning systems, even for applications that were previously impossible with microprocessors, such as real-time control, which demands high speeds. In particular, the SH7040 series has a 1-kbyte on-chip cache, which allows an improvement in CPU performance during external memory access.

In addition, the SH7040 series includes on-chip peripheral functions necessary for system configuration, such as large-capacity ROM and RAM, timers, a serial communication interface (SCI), an A/D converter, an interrupt controller, and I/O ports. Memory or peripheral LSIs can be connected efficiently with an external memory access support function. This greatly reduces system cost.

There are versions of on-chip ROM: mask ROM, PROM, and flash memory. The flash memory can be programmed with a programmer that supports SH7040 series programming, and can also be programmed and erased by software.

This hardware manual describes the SH7040 series hardware. Refer to the programming manual for a detailed description of the instruction set.

## Related Manual

SH7040 series instructions
SH-1/SH-2/SH-DSP Programming Manual
Please consult your Renesas Technology sales representative for details for development environment system.

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## List of Items Revised or Added for This Version



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| Section | Page | Description |
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| 11.2.3 DMA <br> Transfer Count Registers 0-3 (DMATCRODMATCR3) | 220 | Description amended <br> The data for the upper 8 bits of a DMATCR is 0 $\square$ when read. |
| 11.2.4 DMA Channel Control Registers 0-3 (CHCRO-CHCR3) | 221 | Description amended <br> - Bits 31-21—Reserved bits: Data are 0 $\square$ when read. The write value always be 0 . |
|  | 224 | Description amended <br> - Bit 7—Reserved bits: Data is 0 $\square$ when read. The write value always be 0 . |
| 11.2.5 DMAC Operation Register (DMAOR) | 226 | Description amended <br> - Bits 15-10—Reserved bits: Data are 0 $\square$ when read. The write value always be 0 . |
|  | 227 | Description amended <br> - Bits 7-3—Reserved bits: Data are 0 $\square$ when read. The write value always be 0 . |
| 11.3.3 Channel Priority | 233 | Figure amended |
| Figure 11.3 Round Robin Mode |  | Channel 0 is given the lowest priority. |
| 12.4.5 Cascade Connection Mode | 337 | Figure amended |
| Figure 12.23 Cascade Connection Operation Example (Phase Counting Mode) |  | TCLKC <br> TCLKD |
| 12.4.9 <br> Complementary <br> PWM Mode <br> Figure 12.55 <br> Example of Output <br> Phase Switching by <br> External Input (1) | 373 | Figure amended <br> When $B D C=1, N=0, P=0, F B=0$, output active level $=$ high |
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| 16.7.2 Handling of <br> Analog Input Pins <br> Figure 16.8 Example <br> of Analog Input Pin <br> Protection Circuit |  | Note amended |
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| Notes: |  |  |

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| 21.2 .2 Socket |  |
| Adapter Pin |  |
| Correspondence and |  |
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| Figure 21.3 SH7042 |  |
| Pin and HN27C 101 |  |
| Pin Correspondence |  |
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| 22.7.4 Erase-Verify <br> Mode <br> Figure 22.14 <br> Erase/Erase-Verify <br> Flowchart |  | Figure amended <br> Notes: *1 Preprogramming (setting erase block data to all " 0 ") is not necessary <br> *2 Verify data is read in 32-bit (longword) units. <br> *4 Erasing is performed in block units. To erase a number of blocks, each block must be erased in turn <br> *5 Make sure to set the wait times and repetitions as specified. Erasing may not complete correctly if values other than the specified ones are used. |
| 24.4.2 Canceling the Standby Mode |  | Cancellation by a Manual Reset deleted |
| 25. Electrical Characteristics (5V, 33.3 MHz Version) | - | Deleted |


| Section | Page | Description |
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| $25.2 \mathrm{DC}$ <br> Characteristics | 751 | Note amended |
|  |  | *2 5 mA in the A mask version, except for F-ZTAT products. |
| Table 25.2 DC Characteristics |  |  |
| 25.3.2 Control | 754 | Note amended |
| Signal Timing |  | Note: * The $\overline{\mathrm{RES}}, \overline{\mathrm{MRES}}, \mathrm{NMI}, \overline{\mathrm{BREQ}}$, and $\overline{\mathrm{RQR7}}-\overline{\mathrm{RQO}}$ signals are |
| Table 25.5 Control Signal Timing |  | asynchronous inputs, but when thesetup times shown here are provided, the signals are considered to have produced changes at clock rise (for $\overline{\mathrm{RES}}, \mathrm{MRES}, \overline{\mathrm{BREQ}}$ ) or clock fall (for NMI and $\overline{\mathrm{IRQ}}-\overline{\mathrm{IRQO}}$ ). If the setup times are not provided, recognition is delayed until the next clock rise or fall. |

25.3.3 Bus Timing 763 Figure amended

Figure 25.12 DRAM
Cycle (Normal Mode,
1 Wait, TPC=0,
RCD=0)


Figure 25.13 DRAM 764 Figure amended
Cycle (Normal Mode,
2 Waits, TPC=1, RCD=1)


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| 25.4 A/D Converter Characteristics | 779 | Table amended |  |  |  |  |  |
| Table 25.16 A/D Converter Timing (A mask) |  | Non-linearity error* |  |  |  |  |  |
|  |  | Offset error* |  |  |  |  |  |
|  |  | Full scale error* |  |  |  |  |  |
|  |  | Quantize error** |  |  |  |  |  |
| 26.2 DC Characteristics | 782 | Table amended |  |  |  |  |  |
| Table 26.2 DC Characteristics |  |  | $\mathrm{V}_{\mathrm{T}}^{+}-\mathrm{V}_{\mathrm{T}}^{-}$ | $\begin{gathered} v_{c \mathrm{cx}} x \\ 0.00 \end{gathered}$ | - - | v | $\frac{\mathrm{VT}^{+} \geq \mathrm{V}_{\mathrm{cc}} \times 0.9 \mathrm{~V}(\text { min })}{\mathrm{VT}^{-} \leq \mathrm{V}_{\mathrm{cc}} \times 0.2 \mathrm{~V}(\text { max })}$ |
|  | 783 | Table amended |  |  |  |  |  |
|  |  | Analog supply current |  | - | $48$ | mA | $\mathrm{f}=16.7 \mathrm{MHz}$ |
|  |  |  | $\mathrm{Al}_{\text {er }}$ | - | $0.5 \quad 1^{* 3}$ | mA | QFP144 version only |

## *3 2 mA in the A mask version of MASK products.

26.3.2 Control 786 Note amended

Signal Timing
Table 26.5 Control
Signal Timing

Notes: *1 SH7042/43 ZTAT (excluding A mask) are 3.2V.
*2 The $\overline{\mathrm{RES}}, \overline{\mathrm{MRES}}, \mathrm{NMI}, \overline{\mathrm{BREQ}}$, and $\overline{\mathrm{RQ} 7}-\overline{\mathrm{RQO}}$ signals are asynchronous inputs, but when the setup times shown here are provided, the signals are considered to have produced changes at clock rise (for RES, $\overline{\text { MRES }}$, $\overline{\mathrm{BREQ}}$ ) or clock fall (for NMI and $\overline{\mathrm{RQ} 7}-\overline{\mathrm{RQO}}$ ). If the setup times are not provided, recognition is delayed until the next clock rise or fall.
26.3.3 Bus Timing 795 Figure amended

Figure 26.12 DRAM
Cycle (Normal Mode, 1 Wait, TPC = 0,
RCD = 0)



Figure 26.14 DRAM 796 Figure amended Cycle (Normal Mode, 3 Waits, TPC $=1$, $R C D=1$ )

26.3.5 Multifunction 802 Figure amended

Timer Pulse Unit Timing
Figure 26.23 MTU I/O Timing


| 26.3.11 | 810 | Title amended |
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| Measurement |  | Output Load Circuit |
| Conditions for AC |  |  |

Conditions for AC
Characteristics
Figure 26.33 Output
Load Circuit

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| Appendix B Block Diagrams | 844 | Note added |  |  |  |  |  |  |  |
| Figure B. 19 |  | On-chip flash memory* |  |  |  |  |  |  |  |
| $\begin{aligned} & \frac{\mathrm{PB} 4 / \overline{\mathrm{RQ} 2} / \overline{\mathrm{POE} 2 /}}{\overline{\mathrm{CASH}}, \overline{\mathrm{PB3} / \mathrm{IRQ1}} /} \\ & \hline \mathrm{POE} 1 / \mathrm{CASL} \\ & \text { Block Diagram } \\ & \text { (F-ZTAT Version) } \end{aligned}$ |  | A17 |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |
|  |  | Note: * Only when $\mathrm{n}=4$. |  |  |  |  |  |  |  |
| Appendix C Pin States <br> Table C. 1 Pin Modes During Reset, Power-Down, and Bus Right Release Modes (144 Pin) | 865 | Table amended |  |  |  |  |  |  |  |
|  |  | Pin Function |  | Pin modes |  |  |  |  | Standby in Bus Right Release |
|  |  |  |  | $\frac{\text { Reset }}{\text { Power-OnManual }}$ |  | $\frac{\text { Power-Down }}{\text { Sus Right }}$ |  |  |  |
|  |  | Class | Pin Name |  |  |  |  |  |  |
|  |  | Clock | CK | $\bigcirc$ | 0 | ${ }^{* *}$ | $\bigcirc$ | $\bigcirc$ | - |
|  |  | System | RES | 1 | 1 | 1 | 1 | 1 | 1 |
|  |  | control | MRES | $\mathrm{z}^{\text {ma }}$ | 1 | z | 1 | 1 | z |
|  |  |  | WDTOVF | O*3 | O*3 | 0 | $\bigcirc$ | $\bigcirc$ | - |
|  |  |  | BREQ | $\mathrm{z}^{\text {44 }}$ | 1 | z | 1 | 1 | 1 |
|  |  |  | $\overline{\text { BACK }}$ | $\mathrm{z}^{\text {*4 }}$ | 0 | z | 0 | L | L |
|  |  | Interrupt | NMI | 1 | 1 | 1 | 1 | 1 | 1 |
|  |  |  | $\overline{\mathrm{RQQ0}-1 \mathrm{RQ7}}$ | $\mathrm{Z}^{\text {**4 }}$ | 1 | z | I | 1 | z |
|  |  |  | $\overline{\text { RQOUT (PD30) }}$ | $\mathrm{z}^{\text {*4 }}$ | 0 | ${ }^{*}{ }^{*}$ | H | $\bigcirc$ | ${ }^{*}{ }^{\text {+ }}$ |
|  |  |  | $\overline{\text { IROOUT (PE15) }}$ | $\mathrm{z}^{\text {ma }}$ | $\bigcirc$ | z | H | 0 | z |
|  |  | Address bus | A0-A21 | $0^{* 2}$ | $\bigcirc$ | z | $\bigcirc$ | z | z |
|  |  | Data bus | D0-D31 | $\mathrm{Z}^{\text {** }}$ | I/ | z | $1 / 0$ | z | z |
|  |  | Bus | WAIT | $\mathrm{Z}^{\text {** }}$ | । | z | 1 | z | z |
|  |  | control | RD/WR, $\overline{\text { RAS }}$ | $\mathrm{z}^{\text {** }}$ | $\bigcirc$ | 0 | $\bigcirc$ | z | z |
|  |  |  | $\frac{\overline{\overline{\mathrm{CASH}}, \overline{\mathrm{CASL}},}}{\overline{\mathrm{CASLH}}, \overline{\mathrm{CASLL}}}$ | $z^{\text {** }}$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | z | z |
|  |  |  | $\overline{\text { RD }}$ | H | 0 | z | - | z | z |
|  |  |  | $\overline{\text { CSO, }}$ CS 1 | H | $\bigcirc$ | z | $\bigcirc$ | z | z |
|  |  |  | CS2, Cs3 | $\mathrm{Z}^{* 4}$ | 0 | z | O | z | z |
|  |  |  | $\overline{\overline{W R H H}, \mathrm{WRHL}}$, WRH, WRL | H | $\bigcirc$ | z | - | z | z |
|  |  |  | $\overline{\overline{A H}}$ | $\mathrm{z}^{\text {*4 }}$ | 0 | z | 0 | z | z |
|  |  | DMAC | DACK0, DACK1 (PD26, PD27) | $\mathrm{z}^{\text {ta }}$ | 0 | 0*1 | $\bigcirc$ | 0 | 0*1 |
|  |  |  | DACK0, DACK1 (PE14, PE15) | $\mathrm{Z}^{* 4}$ | 0 | z | O | 0 | z |
|  |  |  | DRAKO, DRAK1 | $\mathrm{z}^{* 4}$ | $\bigcirc$ | O*1 | - | $\bigcirc$ | 0*1 |
|  |  |  | $\overline{\text { DREQ0, }{ }^{\text {DREQ1 }}}$ | $\mathrm{z}^{\text {*4 }}$ | 1 | z | 1 | 1 | z |

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| Section | Page | Description |  |  |  |  |  |  |
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| States |  |  |  |  |  |  |  |  |
| Table C. 2 Pin Modes During Reset, Power-Down, and Bus Right Release Modes (112 Pin, 120 Pin) (cont) |  | Pin Function |  | Reset |  | -Down | Bus Right | Standby in Bus |
|  |  | Class | Pin Name | Power-On Manual | Stan | Sleep | Release | Right Release |
|  |  | Port control | POE0-POE3 | $\mathrm{Z}^{* 4}$ | Z | 1 | 1 | Z |
|  |  | SCI | SCK0-SCK1 | $\mathrm{z}^{* 4} \quad 1 / 0$ | Z | I/O | 1/0 | Z |
|  |  |  | TXD0-TCD1 | $\mathrm{Z}^{* 4} 0$ | O*1 | 0 | 0 | O*1 |
|  |  |  | RXD0-RXD1 | $\mathrm{Z}^{* 4}$ | Z | 1 | 1 | Z |
|  |  | A/D converte | $\overline{\text { ADTRG }}$ | $\mathrm{Z}^{* 4}$ | z | 1 | 1 | Z |
|  |  | control | AN0-AN7 | z | Z | 1 | 1 | Z |
|  |  | I/O Port | PA0-PA15 | $\mathrm{Z}^{* 4} \quad 1 / 0$ | $\mathrm{K}^{* 1}$ | K | 1/0 | $\mathrm{K}^{* 1}$ |
|  |  |  | PB0-PB9 |  |  |  |  |  |
|  |  |  | PC0-PC15 |  |  |  |  |  |
|  |  |  | PD0-PD15 |  |  |  |  |  |
|  |  |  | PE0-PE8-PE10 |  |  |  |  |  |
|  |  |  | PE9,PE11-PE15 | $\mathrm{Z}^{* 4} \quad 1 / 0$ | Z | K | I/O | Z |
|  |  |  | PF0-PF7 | Z | Z | 1 | 1 | Z |
|  |  | Notes: 1. There are instances where bus right release and transition to software standby mode occur simultaneously due to the timing between $\overline{\mathrm{BREQ}}$ and internal operations. In such cases, standby mode results, but the standby state may be different. <br> The initial pin states depend on the mode. See section 18, Pin Function Controller (PFC), for details. <br> 2. I: Input, O: Output, H: High-level output, L: Low-level output, Z: High impedance, K: Input pin with high impedance, output pin mode maintained. <br> * 1 If the standby control register port high-impedance bits are set to 1 , output pins become high impedance. <br> *2 A21-A18 will become input ports after power-on reset. <br> *3 Input in the SH7044/SH7045 F-ZTAT version. <br> *4 General use I/O ports PAn, PBn, PCn, PDn, and PEn, as well as pins multiplexed with them, are unstable during the $\overline{\operatorname{RES}}$ setup time (tress) immediately after the $\overline{\operatorname{RES}}$ pin goes to low level. |  |  |  |  |  |  |

Appendix E Product 876, Table amended Code Lineup 877
Table E. 1 SH7040, SH7041, SH7042, SH7043, SH7044, and SH7045 Product Lineup

| Product Type |  | Mask Version | Product Code | Mark Code | Package | Order Model No.*2 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SH7040A | Mask ROM verion | A MASK | HD6437040AF28 HD6437040AVF16 HD6437040AVX16 | HD6437040A (**) F28 HD6437040A ${ }^{(* *)}$ )VF16 HD6437040A(**) VX16 | QFP2020-112 QFP2020-112 TQFP1414-120 | HD6437040A*** HD6437040A*** HD6437040A**X |
|  |  |  | HD6437040ACF28 HD6437040AVCF16 | $\begin{aligned} & \text { HD6437040A }{ }^{(* * *)} \text { CF28 } \\ & \text { HD6437040A }(* *) \text { VCF16 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { QFP2020-112Cu*1 } \\ & \text { QFP2020-112Cu* } \\ & \hline \end{aligned}$ | $\begin{aligned} & 1 \\ & 1 \\ & 1 \\ & 1 \\ & H D 643737040 A^{* * * * *} \mathrm{CF} \\ & \hline \end{aligned}$ |
|  | ROM less verion | A MASK | HD6417040AF28 <br> HD6417040AVF16 <br> HD6417040AVX16 <br> HD6417040ACF28 <br> HD6417040AVCF16 | HD6417040AF28 <br> HD6417040AVF16 <br> HD6417040AVX16 <br> HD6417040ACF28 <br> HD6417040AVCF16 | QFP2020-112 <br> QFP2020-112 <br> TQFP1414-120 <br> QFP2020-112Cu*1 <br> QFP2020-112Cu* ${ }^{* 1}$ | HD6417040AF28 HD6417040AVF16 HD6417040AVX16 HD6417040ACF28 HD6417040AVCF16 |
| SH7041A | Mask ROM verion | A MASK | HD6437041AF28 HD6437041AVF16 | $\begin{aligned} & \hline \text { HD6437041A }\left({ }^{* * *}\right) \text { F28 } \\ & \text { HD6437041A }\left({ }^{* * *}\right. \text { VF16 } \end{aligned}$ | QFP2020-144 QFP2020-144 | $\begin{aligned} & \text { HD6437041A***F } \\ & \text { HD6437041A } A^{* * *} \text { F } \end{aligned}$ |
|  |  |  | HD6437041ACF28 HD6437041AVCF16 | $\begin{aligned} & \text { HD6437041A (***)CF28 } \\ & \text { HD6437041A(***)VCF16 } \end{aligned}$ | $\begin{aligned} & \text { QFP2020-144Cu*1 } \\ & \text { QFP2020-144Cu }{ }^{\star 1} \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { HD6437041A***CF } \\ & \text { HD6437041 }{ }^{* * * *} \mathrm{CF} \end{aligned}$ |
|  | ROM less verion | A MASK | HD6417041AF28 <br> HD6417041AVF16 | HD6417041AF28 HD6417041AVF16 | QFP2020-144 QFP2020-144 | HD6417041AF28 HD6417041AVF16 |
|  |  |  | HD6417041ACF28 HD6417041AVCF16 | HD6417041ACF28 HD6417041AVCF16 | $\begin{aligned} & \text { QFP2020-144Cu*1 } \\ & \text { QFP2020-144Cu }{ }^{\star 1} \\ & \hline \end{aligned}$ | HD6417041ACF28 HD6417041AVCF16 |
| SH7042 | Mask ROM verion | - | HD6437042F28 HD6437042VF16 | $\begin{aligned} & \text { HD6437042 (***)F28 } \\ & \text { HD6437042 (**)VF16 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { QFP2020-112 } \\ & \text { QFP2020-112 } \end{aligned}$ | $\begin{aligned} & \text { HD6437042**F } \\ & \text { HD6437042*** } \end{aligned}$ |
|  | $\begin{aligned} & \hline \text { Z-TAT } \\ & \text { version } \\ & \hline \end{aligned}$ | - | HD6477042F28 HD6477042VF16 | HD6477042F28 HD6477042VF16 | $\begin{aligned} & \text { QFP2020-112 } \\ & \text { QFP2020-112 } \end{aligned}$ | HD6477042F28 HD6477042VF16 |
| SH7042A | Mask ROM verion | A MASK | HD6437042AF28 HD6437042AVF16 HD6437042AVX16 | $\begin{aligned} & \hline \text { HD6437042A }\left({ }^{(* *)}\right. \text { F28 } \\ & \text { HD6437042A }\left({ }^{* * *)}\right. \text { VF16 } \\ & \text { HD6437042A( }{ }^{* * *)} \text { VX16 } \end{aligned}$ | QFP2020-112 <br> QFP2020-112 <br> TQFP1414-120 | HD6437042A***F HD6437042A*** HD6437042A**X |
|  |  |  | HD6437042ACF28 HD6437042AVCF16 | $\begin{aligned} & \text { HD6437042AA(**)CF28 } \\ & \text { HD6437042A(**)VCF16 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { QFP2020-112Cu*1 } \\ & \text { QFP2020-112Cu } \\ & \hline \end{aligned}$ | $\begin{aligned} & 1 \mathrm{HD}^{1} 437042 \mathrm{~A}^{* * *} \mathrm{CF} \\ & 1 \\ & \hline \end{aligned}$ |
| Product Type |  | Mask Version | Product Code | Mark Code | Package | Order Model No. ${ }^{\text {² }}$ |
| SH7042A | $\begin{aligned} & \hline \text { Z-TAT } \\ & \text { version } \end{aligned}$ | A MASK | HD6477042AF28 HD6477042AVF16 HD6477042AVX16 | HD6477042AF28 HD6477042AVF16 HD6477042AVX16 | QFP2020-112 <br> QFP2020-112 <br> TQFP1414-120 | HD6477042AF28 HD6477042AVF16 HD6477042AVX16 |
|  |  |  | HD6477042ACF28 HD6477042AVCF16 | HD6477042ACF28 HD6477042AVCF16 | $\begin{aligned} & \text { QFP2020-112Cu*1 } \\ & \text { QFP2020-112Cu* } \\ & \hline \end{aligned}$ | HD6477042ACF28 HD6477042AVCF16 |
| SH7043 | Mask ROM version | - | HD6437043F28 HD6437043VF16 | $\begin{aligned} & \text { HD6437043(**)F28 } \\ & \text { HD6437043(**)VF16 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { QFP2020-144 } \\ & \text { QFP2020-144 } \end{aligned}$ | $\begin{aligned} & \text { HD6437043**F } \\ & \text { HD6437043**F } \\ & \hline \end{aligned}$ |
|  | Z-TAT version | - | HD6477043F28 HD6477043VF16 | HD6477043F28 HD6477043VF16 | $\begin{aligned} & \hline \text { QFP2020-144 } \\ & \text { QFP2020-144 } \end{aligned}$ | HD6477043F28 HD6477043VF16 |
| SH7043A | Mask ROM version | A MASK | HD6437043AF28 HD6437043AVF16 | $\begin{aligned} & \text { HD6437043A(**)F28 } \\ & \text { HD6437043A(**)VF16 } \end{aligned}$ | $\begin{aligned} & \text { QFP2020-144 } \\ & \text { QFP2020-144 } \end{aligned}$ | $\begin{aligned} & \text { HD6437043A***F } \\ & \text { HD6437043A**F } \end{aligned}$ |
|  |  |  | HD6437043ACF28 HD6437043AVCF16 | $\begin{aligned} & \text { HD6437043AA (**)CF28 } \\ & \text { HD6437043A }{ }^{(* *)} \text { VCF16 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { QFP2020-144Cu*1 } \\ & \text { QFP2020-144Cu } \end{aligned}$ | $\begin{aligned} & \text { HD6437043A** CF } \\ & \text { HD6437043 }{ }^{* * *} \text { an } \end{aligned}$ |
|  | $\begin{aligned} & Z-\mathrm{TAT} \\ & \text { version } \end{aligned}$ | A MASK | HD6477043AF28 HD6477043AVF16 | HD6477043AF28 HD6477043AVF16 | $\begin{aligned} & \hline \text { QFP2020-144 } \\ & \text { QFP2020-144 } \end{aligned}$ | HD6477043AF28 HD6477043AVF16 |
|  |  |  | HD6477043ACF28 HD6477043AVCF16 | HD6477043ACF28 HD6477043AVCF16 | $\begin{aligned} & \text { QFP2020-144Cu*1 } \\ & \text { QFP2020-144Cu* } \end{aligned}$ | HD6477043ACF28 HD6477043AVCF16 |
| SH7044 | Mask ROM version | A MASK | HD6437044F28 | HD6437044(**)F28 | QFP2020-112 | HD6437044**F |
|  | F-ZTAT version |  | HD64F7044F28 | HD64F7044F28 | QFP2020-112 | HD64F7044F28 |
| SH7045 | Mask ROM version | A MASK | HD6437045F28 | HD6437045(**)F28 | QFP2020-144 | HD6437045**F |
|  | F-ZTAT version |  | HD64F7045F28 | HD64F7045F28 | QFP2020-144 | HD64F7045F28 |

$\left.{ }^{* * *}\right)$ is the ROM code.
NoteS: 1. Package with Copper used as the lead material.
2. ${ }^{* * *}$ in the Order Model No. is the ROM code, consisting of a letter and a two-digit number (ex. E00). The letter indicates the voltage and frequency, as shown below.

- E, F, G, H: 5.0 V, 28 MHz
- P, Q, R: $3.3 \mathrm{~V}, 16 \mathrm{MHz}$


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## Section 1 SH7040 Series Overview

### 1.1 SH7040 Series Overview

The SH7040 Series (SH7040/41/42/43/44/45) CMOS single-chip microprocessors integrate a Renesas-original architecture, high-speed CPU with peripheral functions required for system configuration.

The CPU has a RISC-type instruction set. Most instructions can be executed in one clock cycle, which greatly improves instruction execution speed. In addition, the 32-bit internal-bus architecture enhances data processing power. With this CPU, it has become possible to assemble low cost, high performance/high-functioning systems, even for applications that were previously impossible with microprocessors, such as real-time control, which demands high speeds. In particular, the SH7040 series has a 1-kbyte on-chip cache, which allows an improvement in CPU performance during external memory access.

In addition, the SH7040 Series includes on-chip peripheral functions necessary for system configuration, such as large-capacity ROM and RAM, timers, a serial communication interface (SCI), an A/D converter, an interrupt controller, and I/O ports. Memory or peripheral LSIs can be connected efficiently with an external memory access support function. This greatly reduces system cost.

In addition to the masked-ROM versions of the SH7040 series, the SH7042 and SH7043 have a ZTAT ${ }^{\text {TM }}{ }^{* 1}$ version with user-programmable on-chip PROM and the SH7044 and SH7045 have an F-ZTAT ${ }^{\mathrm{TM} * 2}$ version with on-chip flash memory. These versions enable users to respond quickly and flexibly to changing application specifications, growing production volumes, and other conditions.

Notes: *1 ZTAT (Zero Turn-Around Time) is a registered trademark of Renesas Technology Corp.
*2 F-ZTAT (Flexible ZTAT) is a trademark of Renesas Technology Corp.

### 1.1.1 SH7040 Series Features

## CPU:

- Original Renesas architecture
- 32-bit internal data bus
- General-register machine
- Sixteen 32-bit general registers
— Three 32-bit control registers
— Four 32-bit system registers
- RISC-type instruction set
- Instruction length: 16-bit fixed length for improved code efficiency
- Load-store architecture (basic operations are executed between registers)
- Delayed branch instructions reduce pipeline disruption during branch
- Instruction set based on C language
- Instruction execution time: one instruction/cycle ( $35 \mathrm{~ns} /$ instruction at $28.7-\mathrm{MHz}$ operation)
- Address space: Architecture supports 4 Gbytes
- On-chip multiplier: multiplication operations ( 32 bits $\times 32$ bits $\rightarrow 64$ bits) and multiplication/accumulation operations ( 32 bits $\times 32$ bits +64 bits $\rightarrow 64$ bits) executed in two to four cycles
- Five-stage pipeline


## Cache Memory:

- 1-kbyte instruction cache
- Caching of instruction codes and PC relative read data
- 4-byte line length (1 longword: 2 instruction lengths)
- 256 entry cache tags
- Direct map method
- On-chip ROM/RAM, and on-chip I/O areas not objects of cache
- Used in common with on-chip RAM; 2 kbytes of on-chip RAM used as address array/data array when cache is enabled


## Interrupt Controller (INTC):

- Nine external interrupt pins (NMI, $\overline{\text { IRQ0 }}-\overline{\text { IRQ7 }}$ )
- Forty-three internal interrupt sources (forty-four for A mask)
- Sixteen programmable priority levels


## User Break Controller (UBC):

- Generates an interrupt when the CPU or DMAC generates a bus cycle with specified conditions
- Simplifies configuration of an on-chip debugger


## Bus State Controller (BSC):

- Supports external extended memory access
— 16-bit (QFP-112, TQFP-120), or 32-bit (QFP-144) external data bus
- Memory address space divided into five areas (four areas of SRAM space, one area of DRAM space) with the following settable features:
- Bus size ( 8,16 , or 32 bits)
- Number of wait cycles

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## Renesns

- Outputs chip-select signals for each area
- During DRAM space access:
- Outputs $\overline{\text { RAS }}$ and $\overline{\text { CAS }}$ signals for DRAM
- Can generate a RAS precharge time assurance Tp cycle
- DRAM burst access function
- Supports high-speed access mode for DRAM
- DRAM refresh function
- Programmable refresh interval
- Supports CAS-before-RAS refresh and self-refresh modes
- Wait cycles can be inserted using an external $\overline{\text { WAIT }}$ signal
- Address data multiplex I/O devices can be accessed


## Direct Memory Access Controller (DMAC) (4 Channels):

- Supports cycle-steal transfers
- Supports dual address transfer mode
- Can be switched between direct and indirect transfer modes (channel 3 only)
- Direct transfer mode: transfers the data at the transfer source address to the transfer destination address
- Indirect transfer mode: regards the data at the transfer source address as an address and transfers the data at that address to the transfer destination address


## Data Transfer Controller (DTC):

- Data transfer independent of the CPU possible through peripheral I/O interrupt requests
- Transfer mode can be set for each interrupt factor (transfer mode set in memory)
- Multiple data transfers possible for one activating factor
- Abundant transfer modes
- Normal mode/repeat mode/block transfer mode selectable
- Transfer unit can be set to byte/word/longword
- Interrupts activating the DTC requested of the CPU
- Interrupts can be generated to the CPU after completion of one data transfer
- Interrupts can be generated to the CPU after completing all designated data transfers
- Transfer can be activated by software


## Multifunction Timer/Pulse Unit (MTU):

- Maximum 16 types of waveform output or maximum 16 types of pulse I/O processing possible based on 16-bit timer, 5 channels
- 16 dual-use output compare/input capture registers
- 16 independent comparators
- 8 types of counter input clock
- Input capture function
- Pulse output mode
— One shot, toggle, PWM, phase-compensated PWM, reset-synchronized PWM
- Multiple counter synchronization function
- Phase-compensated PWM output mode
- Non-overlapping waveform output for 6-phase inverter control
- Automatic setting for dead time
- PWM duty cycle can be set from 0 to $100 \%$
- Output off function
- Reset-synchronized PWM mode
- 3-phase output of any duty cycle positive phase/reverse phase PWM waveforms
- Phase calculation mode
- 2-phase encoder calculation processing


## Compare Match Timer (CMT) (Two Channels):

- 16-bit free-running counter
- One compare register
- Generates an interrupt request upon compare match


## Watchdog Timer (WDT) (One Channel):

- Watchdog timer or interval timer
- Count overflow can generate an internal reset, external signal, or interrupt


## Serial Communication Interface (SCI) (Two Channels):

## (Per Channel):

- Asynchronous or clock-synchronous mode is selectable
- Can transmit and receive simultaneously (full duplex)
- On-chip dedicated baud rate generator
- Multiprocessor communication function


## I/O Ports:

- QFP 112 (SH7040, SH7042, SH7044), TQFP-120 (SH7040, SH7042)
- Input/output: 74
— Input: 8

4
— Total: 82

- QFP 144 (SH7041, SH7043, SH7045)
- Input/output: 98
— Input: 8
— Total: 106


## A/D Converter:

- 10 bits $\times 8$ channels
- Conversion upon external trigger possible
- Sample and hold function: two on-chip units (two channels can be sampled simultaneously)
- Depending on the product, there is a high speed, mid-accuracy A/D on-chip type and a midspeed, high accuracy A/D on-chip type. For details, see the product lineup.


## Large Capacity On-Chip Memory:

- ROM (128 kbytes PROM, 256 kbytes/128 kbytes/64 kbytes mask ROM, 256 kbytes flash ROM)
- SH7044, SH7045: 256 kbytes (flash ROM, mask ROM)
- SH7042, SH7043: 128 kbytes (ZTAT, mask ROM)
- SH7040, SH7041: 64 kbytes (mask ROM)
- RAM: 4 kbytes ( 2 kbytes when cache is used)


## Operating Modes:

- Operating modes
- Expanded mode with ROM disabled
— Expanded mode with ROM enabled
- Single-chip mode
- Processing states
- Program execution state
- Exception processing state
- Bus-released state
- Power-down modes
- Sleep mode
- Software standby mode


## Clock Pulse Generator (CPG):

- On-chip clock pulse generator
— On-chip clock-doubling PLL circuit

|  | «әрәлиоэ a／甘 pəәds －P！W，，əәS | słsənbəય <br> sә！oN дәృsueд uo spoчłəш <br>  |  | дәнәлиоо <br>  Іdnидәц！әц！әбиецว | 9トココヘ $\forall$ をも0LEt90H 8乙コつVとャ0LEt9OH 9トコオ 8ZコVEt0LEt9OH | $\begin{array}{r} \wedge \varepsilon \in \varepsilon \\ \wedge G \\ \wedge \varepsilon \cdot \varepsilon \\ \wedge G \end{array}$ | $\begin{aligned} & \text { zHW 91 } \\ & \text { zHW 8Z } \\ & \text { zHW 91 } \\ & \text { zHW 8Z Э.GL Ol Э.OZ- } \end{aligned}$ | ＊nづカレ－OZOZd」O <br> カート－OZOZdコO | $\begin{array}{r} \text { (pəədS-p!w) } \\ \text { gS7ヵ干 } \end{array}$ | S＋19 乙८ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | «әиәлиоэ <br> a／$\forall$ pəәds <br> －4б！！ |  |  |  | 9トコノ＾と†0LEt90H 8てヨとャ0んとち90H | $\wedge \varepsilon \cdot \varepsilon$ $\wedge$ | zHW 91 <br> zHW 8己 〇．s 0 Oł 〇．OZ－ | カカレ－0ZOZd」O | （pəədS－46！！$)$ GS7SI耳 | S！！ |
|  | «әрәлиоう <br> a／$\forall$ pəods <br> －P！${ }_{1}$, әәS | słsənbəı <br> sәłоN дәృsueд uo spoцəәш <br>  |  | ләцәлиоэ <br>  <br>  |  <br>  91Xへ $\forall$ Zち0LEt9OH 9トコイオZヤロ <br>  | $\begin{array}{r} \wedge \varepsilon \varepsilon \\ \wedge G \\ \wedge \varepsilon \varepsilon \\ \wedge \varepsilon \cdot \varepsilon \\ \wedge G \end{array}$ | zHW 91 <br> zHW 82 <br> zHW 91 <br> zHW 91 <br>  |  |  | St！ 91 |
|  | ＂диәлиоэ <br> a／$\forall$ pəəds <br> －46！！${ }^{\text {，}}$ ，әәS |  |  |  | 9トコへટち0LEt90H 8Zヨで0LEち90H | $\begin{array}{r} \wedge \varepsilon \in \varepsilon \\ \wedge \mathrm{G} \end{array}$ | zHW 91 <br> zHW 8Z 〇．SL Ol O．OZ－ | ZLL-OZOZdJO | $\begin{array}{r} \text { (pəəds-46!! }) \\ \text { gSTSI耳 } \end{array}$ | St！ 91 |
|  | «әнәлиоэ <br> a／$\forall$ pəәds －P！${ }^{\prime \prime}$ ，әəS | słsənbə』 <br> sәłon дәృsueł uo spoчəәш <br>  |  | дәдәлиоэ <br>  <br>  | 9トココベートナOL\＆t90H 8己ココナレャ0LEt9OH 9トコイシレャOLEt9OH 8Zコトレレ0LEt90H | $\begin{array}{r} \wedge \varepsilon \varepsilon \\ \wedge G \\ \wedge \varepsilon \varepsilon \\ \wedge G \end{array}$ | $\begin{aligned} & \text { zHW 91 } \\ & \text { zHW 8Z } \\ & \text { zHW 91 } \\ & \text { zHW 8Z ЭஃGL Ol Э.OZ- } \end{aligned}$ |  <br> カセレ－OZOZdコO | $\begin{array}{r} \text { (pəəds-p!w) } \\ \text { gS7t } \mp \end{array}$ | S＋！9 乙८ |
|  | «әрәлиоう a／$\forall$ pəods －P！${ }^{\prime \prime}$, әəS |  | sıołəə＾OıQ pue sрочłәu ssəээ厄 <br>  | дәәлиоэ <br>  łdnયəəృu！əપł əదиец૭ | 9トココヘ $\forall 0+0<\varepsilon t 90 H$ 8乙コЭVOtOLEt9OH 91Xへ甘OtOLEt9OH 9ト－－$\wedge \forall 0 ャ 0 \angle E t 90 H$ <br>  | $\begin{array}{r} \wedge \varepsilon \varepsilon \varepsilon \\ \wedge G \\ \wedge \varepsilon \varepsilon \varepsilon \\ \wedge \varepsilon \cdot \varepsilon \\ \wedge G \end{array}$ | zHW 91 <br> zHW 82 <br> zHW 91 <br> zHW 91 <br> zHW 8Z Э．SL Oł О．OZ－ |  | $\begin{array}{r} \text { (pəədS-p!w) } \\ \text { gS7t } \mp \end{array}$ | St！ 91 |
|  | «әиәлиоэ a／甘 pəəds －p！$N_{n}$ әәS | słsənbəィ <br>  <br>  |  | дәцәлиоэ <br>  <br>  |  | $\wedge G$ |  | カセレ－0Z0ZdコO | $\begin{array}{r} \text { (pəəds-p!w) } \\ \text { gS7t干 } \end{array}$ | S＋！9 乙८ |
|  | «әнәлиој a／$\forall$ pəәds －P！${ }^{\prime}$, әәS | słsənbəィ <br> sәłon dəృsueд uo spoyłəu <br>  |  | дәиәлиоэ <br>  <br>  |  | $\wedge G$ | zHW 8Z О．GL Ol O．OZ－ | ZLL-OZOZdJO | $\begin{array}{r} \text { (pəəds-p!w) } \\ \text { gS7t } \mp \end{array}$ | St！ 91 |
|  | «әрәлиоо a／$\forall$ pəәds －P！${ }^{\prime \prime}$, әәS | słsənbəィ <br> sә！oN дәృsueд uo spoчəәш <br>  |  sрочəәu ssәээе પヨコロ әЧા әбиடчэ | дәрәлиоэ <br>  ¡dnגəəu！әцł әбиецつ | 9トヨコノ $\forall$ をも $0 \angle \angle \triangleright 90 H$ 8乙コつV\＆ャ0LLt9OH <br>  8ZコV\＆ャ0LLャ90H | $\begin{array}{r} \wedge \varepsilon \in \varepsilon \\ \wedge G \\ \wedge \varepsilon \varepsilon \varepsilon \\ \wedge G \end{array}$ | zHW 91 <br> zHW 8己 <br> zHW 91 <br> zHW 8Z Э．GL Ol 〇．OZ－ | ＊ n 〇カャレー0ZOZd」O <br>  | $\begin{array}{r} \text { (pəəds-p!w) } \\ \text { GS7t干 } \end{array}$ | S＋19 乙८ |
|  | «әиәлиој a／$\forall$ pəods －46ิ！Н，әәS |  |  |  | 9トヨএ＾をt0LLt90H 8てヨど0LLt90H | $\begin{array}{r} \wedge \varepsilon \\ \wedge \\ \wedge \end{array}$ | zHW 91 <br> zHW 8Z 〇．SL Ol O．OZ－ |  | （pəəds－46！！$)$ gS7s！ | S＋！9 乙८ |
|  | «әрәлиој <br> a／$\forall$ pəәds <br> －P！${ }^{\prime}$, әәS | słsənbəય <br> sәృоN дәృsueд uo spoчłәш әб匕ел әцІ әбиечว би！џәs әцІ әбиечว | sıołəə＾О」 pue sроцłәu ssəээ厄 பヨコロ әЧъ әбиечว | дәдәлиоэ <br> a／甘 рәдеןәд sıоㅟәл <br>  |  8Zコづ 91Xへ $\forall$ ZっOLLt9OH <br>  8ZコVZち0LLャ9のH | $\begin{array}{r} \wedge \varepsilon \varepsilon \varepsilon \\ \wedge G \\ \wedge \varepsilon \varepsilon \varepsilon \\ \wedge \varepsilon \cdot \varepsilon \\ \wedge G \end{array}$ | zHW 91 <br> zHW 8Z <br> zHW 91 <br> zHW 91 <br> zHW 8Z Э．SL Ol 〇．OZ－ |  | $\begin{array}{r} \text { (pəədS-p!W) } \\ \text { GS } 7 \uparrow \mp \end{array}$ | St！ 91 |



|  <br>  | ＂әрәлиој <br> a／$\forall$ pəәds －P！${ }^{N}$, әəS | słsənbəય <br>  <br>  |  sрочəәш ssәээе <br>  |  | 9トヨコヘシレレ0くレカ9aH <br>  9トコイシトゥ0んเ⿱9OH 8てヨVトレ0くเナ9のH | $\begin{array}{r} \wedge \varepsilon \varepsilon \\ \wedge G \\ \wedge \varepsilon \cdot \varepsilon \\ \wedge G \end{array}$ | zHW 91 <br> zHW 8Z <br> zHW 91 <br> ZHW 8Z | O.GL Ol O.OZ- | ＊ n <br> 什レ－OZOZd」O | $\begin{array}{r} \text { (pəəds-p!w) } \\ \text { gS7ヶ } \end{array}$ | S4！$¢$ 乙८ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  ןอэ！џәઇヨ，әәS | ＂əрәлиој <br> a／$\forall$ pəəds －p！${ }^{\prime}$, әәS | słsənbəィ sәło дəృsueג uo spoyłəш <br>  |  sрочңәш ssəәэ Чヨコロ әЧł әбиечว | ґәдәлиоэ <br> a／甘 рәృe｜ər sıoloəィ <br>  |  <br>  91Xへ $\forall$ OtOLレカ9OH 9トヨイン <br>  | $\begin{array}{r} \wedge \varepsilon \varepsilon \\ \wedge G \\ \wedge \varepsilon \varepsilon \\ \wedge \varepsilon \varepsilon \\ \wedge G \end{array}$ | zHW 91 <br> zHW 8Z <br> ZHW 91 <br> zHW 91 <br> zHW 8Z | O．S Ol О．OZ－ | ＊ nJてLL－0ZOZdコO OZl－カレカーdコO1 てLI－OZOZd」O | $\begin{array}{r} \text { (pəods-p!w) } \\ \text { gS7ヶ } \end{array}$ | Stı9 91 |
|  | ＂дрәлиој <br> a／$\forall$ pəәds －P！${ }^{\prime}$, әәS | słsənbə» <br> səloN dəృsueł uo spoyłəu <br>  |  sрочғәш ssәээе <br>  | дәиәлиог <br>  łdnયəıu！әчł әбиечว | 8Z」St0LEt90H | $\wedge \mathrm{G}$ | zHW 8Z | O．GL Ol О．0Z－ | 什し－0ZOZdコO | （pəəds－p！w） GS $7 \downarrow \mp$ | S＋！ 9 |
|  | ＂диәлиој <br> a／$\quad$ pəәds | słsənbəม <br> sәlon дәృsueд uo spoцłəu |  sрочңәш ssəәэе | дәцәлиоэ <br>  |  |  |  |  |  | （pəəds－p！W） |  |

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### 1.2 Block Diagram

Figure 1.1 is a block diagram of the SH7040 Series QFP-112 pin and TQFP-120 pin. Figure 1.2 is a block diagram of the SH7040 Series QFP-144 pin.


Figure 1.1 Block Diagram of the SH7040, SH7042, SH7044 (QFP-112 Pin), SH7040, SH7042 (TQFP-120 pin)


Notes: *1 $\mathrm{V}_{\mathrm{CC}}$ in the mask and ZTAT versions; FWP in the F-ZTAT version (however, FWE in writer mode) *2 $\mathrm{V}_{\mathrm{pp}}$ : ZTAT version only

Figure 1.2 Block Diagram of the SH7041, SH7043, SH7045 (QFP-144 Pin)

### 1.3 Pin Arrangement and Pin Functions

### 1.3.1 Pin Arrangment

Figure 1.3 shows the pin arrangement for the QFP-112 (top view).


Figure 1.3 SH7040, SH7042, SH7044 Pin Arrangement (QFP-112 Top View)

Figure 1.4 shows the pin arrangement for the TFP-120 (top view).


Figure 1.4 SH7040, SH7042 Pin Arrangement (TQFP-120 Top View)

Figure 1.5 shows the pin arrangement for the QFP-144 (top view).


Figure 1.5 SH7041, SH7043, SH7045 Pin Arrangement (QFP-144 Top View)

### 1.3.2 Pin Arrangement by Mode

Table 1.2 Pin Arrangement by Mode for SH7040, SH7042 (QFP-112 Pin)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 1 | PE14/TIOC4C/DACK0/AH | $\mathrm{V}_{\mathrm{cc}}$ |
| 2 | PE15/TIOC4D/DACK1//RQOUT | $\overline{C E}$ |
| 3 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 4 | PCO/AO | A0 |
| 5 | PC1/A1 | A1 |
| 6 | PC2/A2 | A2 |
| 7 | PC3/A3 | A3 |
| 8 | PC4/A4 | A4 |
| 9 | PC5/A5 | A5 |
| 10 | PC6/A6 | A6 |
| 11 | PC7/A7 | A7 |
| 12 | PC8/A8 | A8 |
| 13 | PC9/A9 | NC |
| 14 | PC10/A10 | A10 |
| 15 | PC11/A11 | A11 |
| 16 | PC12/A12 | A12 |
| 17 | PC13/A13 | A13 |
| 18 | PC14/A14 | A14 |
| 19 | PC15/A15 | A15 |
| 20 | PB0/A16 | A16 |
| 21 | $V_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 22 | PB1/A17 | NC |
| 23 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 24 | PB2// $\overline{\mathrm{RQO}} / \overline{\mathrm{POE} 0} / \overline{\mathrm{RAS}}$ | NC |
| 25 |  | $\overline{\mathrm{OE}}$ |
| 26 |  | $\overline{\text { PGM }}$ |
| 27 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 28 | PB5/IRQ3/ $\overline{\mathrm{POE} 3 / R D W R ~}$ | $\mathrm{V}_{\mathrm{cc}}$ |

Table 1.2 Pin Arrangement by Mode for SH7040, SH7042 (QFP-112 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 29 | PB6//ІQ4/A18/BACK | NC |
| 30 | PB7//]RQ5/A19/BREQ | NC |
| 31 | PB8/IRQ6/A20/WAIT | NC |
| 32 |  | NC |
| 33 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\mathrm{ss}}$ |
| 34 | PA14/ $\overline{\mathrm{RD}}$ | NC |
| 35 | $\overline{\text { WDTOVF }}$ | NC |
| 36 | PA13/ $\overline{\mathrm{WRH}}$ | NC |
| 37 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 38 | PA12/VRL | NC |
| 39 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 40 | PA11/CS1 | NC |
| 41 | PA10/CS0 | NC |
| 42 | PA9/TCLKD/ $\overline{\text { RQ3 }}$ | NC |
| 43 | PA8/TCLKC/ $\overline{\text { RQ2 }}$ | NC |
| 44 | PA7/TCLKB/ $\overline{C S 3}$ | NC |
| 45 | PA6/TCLKA/ $\overline{\mathrm{CS} 2}$ | NC |
| 46 | PA5/SCK1//̄REQ1/]/RQ1 | NC |
| 47 | PA4/TXD1 | NC |
| 48 | PA3 /RXD1 | NC |
| 49 | PA2/SCK0/ $\overline{\mathrm{DREQ0}} / \overline{\mathrm{RQQ}}$ | NC |
| 50 | PA1/TXD0 | NC |
| 51 | PA0/RXD0 | NC |
| 52 | PD15/D15 | NC |
| 53 | PD14/D14 | NC |
| 54 | PD13/D13 | NC |
| 55 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 56 | PD12/D12 | NC |
| 57 | PD11/D11 | NC |
| 58 | PD10/D10 | NC |

Table 1.2 Pin Arrangement by Mode for SH7040, SH7042 (QFP-112 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 59 | PD9/D9 | NC |
| 60 | PD8/D8 | NC |
| 61 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\mathrm{ss}}$ |
| 62 | PD7/D7 | D7 |
| 63 | PD6/D6 | D6 |
| 64 | PD5/D5 | D5 |
| 65 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 66 | PD4/D4 | D4 |
| 67 | PD3/D3 | D3 |
| 68 | PD2/D2 | D2 |
| 69 | PD1/D1 | D1 |
| 70 | PD0/D0 | D0 |
| 71 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 72 | XTAL | NC |
| 73 | MD3 | $\mathrm{V}_{\mathrm{cc}}$ |
| 74 | EXTAL | $\mathrm{V}_{\text {ss }}$ |
| 75 | MD2 | $\mathrm{V}_{\mathrm{cc}}$ |
| 76 | NMI | A9 |
| 77 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 78 | MD1 | $\mathrm{V}_{\mathrm{cc}}$ |
| 79 | MD0 | $\mathrm{V}_{\mathrm{cc}}$ |
| 80 | PLLVCC | $\mathrm{V}_{\mathrm{cc}}$ |
| 81 | PLLCAP | $\mathrm{V}_{\text {ss }}$ |
| 82 | PLLVSS | $\mathrm{V}_{\text {ss }}$ |
| 83 | PA15/CK | NC |
| 84 | $\overline{\mathrm{RES}}$ | $\mathrm{V}_{\text {PP }}$ |
| 85 | PE0/TIOCOA/DREQ0 | NC |
| 86 | PE1/TIOCOB/DRAK0 | NC |
| 87 | PE2/TIOC0C/DREQ1 | NC |
| 88 | PE3/TIOC0D/DRAK1 | NC |

Table 1.2 Pin Arrangement by Mode for SH7040, SH7042 (QFP-112 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 89 | PE4/TIOC1A | NC |
| 90 | $\mathrm{V}_{\mathrm{ss}}$ | $\mathrm{V}_{\text {ss }}$ |
| 91 | PFO/AN0 | $\mathrm{V}_{\text {ss }}$ |
| 92 | PF1/AN1 | $\mathrm{V}_{\text {ss }}$ |
| 93 | PF2/AN2 | $\mathrm{V}_{\text {ss }}$ |
| 94 | PF3/AN3 | $\mathrm{V}_{\text {ss }}$ |
| 95 | PF4/AN4 | $\mathrm{V}_{\text {ss }}$ |
| 96 | PF5/AN5 | $\mathrm{V}_{\text {ss }}$ |
| 97 | $\mathrm{AV}_{\text {Ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 98 | PF6/AN6 | $\mathrm{V}_{\text {ss }}$ |
| 99 | PF7/AN7 | $\mathrm{V}_{\text {ss }}$ |
| 100 | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 101 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 102 | PE5/TIOC1B | NC |
| 103 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 104 | PE6/TIOC2A | NC |
| 105 | PE7/TIOC2B | NC |
| 106 | PE8/TIOC3A | NC |
| 107 | PE9/TIOC3B | NC |
| 108 | PE10/TIOC3C | NC |
| 109 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 110 | PE11/TIOC3D | NC |
| 111 | PE12/TIOC4A | NC |
| 112 | PE13/TIOC4B/MRES | NC |

Table 1.3 Pin Arrangement by Mode for SH7040, SH7042 (TQFP-120 Pin)

| TQFP120 Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 1 | NC | NC |
| 2 | PE14/TIOC4C/DACK0/AH | $\mathrm{V}_{\mathrm{cc}}$ |
| 3 | PE15/TIOC4D/DACK1/IRQOUT | CE |
| 4 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 5 | PCO/A0 | A0 |
| 6 | PC1/A1 | A1 |
| 7 | PC2/A2 | A2 |
| 8 | PC3/A3 | A3 |
| 9 | PC4/A4 | A4 |
| 10 | PC5/A5 | A5 |
| 11 | PC6/A6 | A6 |
| 12 | PC7/A7 | A7 |
| 13 | PC8/A8 | A8 |
| 14 | PC9/A9 | NC |
| 15 | PC10/A10 | A10 |
| 16 | PC11/A11 | A11 |
| 17 | PC12/A12 | A12 |
| 18 | PC13/A13 | A13 |
| 19 | PC14/A14 | A14 |
| 20 | PC15/A15 | A15 |
| 21 | PB0/A16 | A16 |
| 22 | $\mathrm{V}_{\text {c }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 23 | PB1/A17 | NC |
| 24 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 25 | PB2/ $/ \overline{\mathrm{RQ0}} / \overline{\mathrm{POE} 0} / \overline{\mathrm{RAS}}$ | NC |
| 26 | PB3/ $\overline{\mathrm{RQ}} 1 / \overline{\mathrm{POE}} / \mathrm{CASL}$ | OE |
| 27 | PB4/ $\overline{\mathrm{RQ} 2} / \overline{\mathrm{POE} 2} / \overline{\mathrm{CASH}}$ | PGM |
| 28 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 29 | PB5/IRQ3/ $\overline{\mathrm{POE3}} / \mathrm{RDWR}$ | $V_{c c}$ |
| 30 | NC | NC |
| 31 | NC | NC |

Table 1.3 Pin Arrangement by Mode for SH7040, SH7042 (TQFP-120 Pin) (cont)

| TQFP120 Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 32 | PB6/IRQ4/A18/BACK | NC |
| 33 | PB7/IRQ5/A19/ $\overline{\mathrm{BREQ}}$ | NC |
| 34 | PB8/IRQ6/A20/WAIT | NC |
| 35 | PB9/\} \overline {  RQ7  /  A21/ADTRG  } | NC |
| 36 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 37 | PA14/ $\overline{\mathrm{RD}}$ | NC |
| 38 | WDTOVF | NC |
| 39 | PA13/WRH | NC |
| 40 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 41 | PA12/WRL | NC |
| 42 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\mathrm{ss}}$ |
| 43 | PA11/ES1 | NC |
| 44 | PA10/CS0 | NC |
| 45 | PA9/TCLKD/\} \overline {  RQ3  } | NC |
| 46 | PA8/TCLKC/\} \overline {  RQ2  } | NC |
| 47 | PA7/TCLKB/ $\overline{\mathrm{CS3}}$ | NC |
| 48 | PA6/TCLKA/ $\overline{C S 2}$ | NC |
| 49 | PA5/SCK1//̄REQ1//]R1 | NC |
| 50 | PA4/TXD1 | NC |
| 51 | PA3/RXD2 | NC |
| 52 | PA2/SCK0/ $\overline{\text { DREQ0 }} / \overline{\mathrm{RQO}}$ | NC |
| 53 | PA1/TXD0 | NC |
| 54 | PA0/RXD0 | NC |
| 55 | PD15/D15 | NC |
| 56 | PD14/D14 | NC |
| 57 | PD13/D13 | NC |
| 58 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 59 | PD12/D12 | NC |
| 60 | NC | NC |
| 61 | NC | NC |
| 62 | PD11/D11 | NC |

Table 1.3 Pin Arrangement by Mode for SH7040, SH7042 (TQFP-120 Pin) (cont)

| TQFP120 Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 63 | PD10/D10 | NC |
| 64 | PD9/D9 | NC |
| 65 | PD8/D8 | NC |
| 66 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 67 | PD7/D7 | D7 |
| 68 | PD6/D6 | D6 |
| 69 | PD5/D5 | D5 |
| 70 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\text {cc }}$ |
| 71 | PD4/D4 | D4 |
| 72 | PD3/D3 | D3 |
| 73 | PD2/D2 | D2 |
| 74 | PD1/D1 | D1 |
| 75 | PD0/D0 | D0 |
| 76 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 77 | XTAL | NC |
| 78 | MD3 | $\mathrm{V}_{\mathrm{cc}}$ |
| 79 | EXTAL | $\mathrm{V}_{\text {ss }}$ |
| 80 | MD2 | $\mathrm{V}_{\text {cc }}$ |
| 81 | NMI | A9 |
| 82 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\text {cc }}$ |
| 83 | MD1 | $\mathrm{V}_{\text {cc }}$ |
| 84 | MD0 | $\mathrm{V}_{\mathrm{cc}}$ |
| 85 | $\mathrm{PLLV}_{\text {cc }}$ | $\mathrm{V}_{\text {cc }}$ |
| 86 | PLLCAP | $\mathrm{V}_{\text {ss }}$ |
| 87 | PLLV ${ }_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 88 | PA15/CK | NC |
| 89 | $\overline{R E S}$ | $\mathrm{V}_{\text {PP }}$ |
| 90 | NC | NC |
| 91 | NC | NC |
| 92 | PE0/TIOCOA/ $\overline{\text { DREQ0 }}$ | NC |
| 93 | PE1/TIOC0B/馬RAK0 | NC |

Table 1.3 Pin Arrangement by Mode for SH7040, SH7042 (TQFP-120 Pin) (cont)

| TQFP120 Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 94 | PE2/TIOC0C/DREQ1 | NC |
| 95 | PE3/TIOC0D/DRAK1 | NC |
| 96 | PE4/TIOC1A | NC |
| 97 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 98 | PFO/AN0 | $\mathrm{V}_{\text {ss }}$ |
| 99 | PF1/AN1 | $\mathrm{V}_{\text {ss }}$ |
| 100 | PF2/AN2 | $\mathrm{V}_{\text {ss }}$ |
| 101 | PF3/AN3 | $\mathrm{V}_{\text {ss }}$ |
| 102 | PF4/AN4 | $V_{\text {ss }}$ |
| 103 | PF5/AN5 | $V_{\text {ss }}$ |
| 104 | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 105 | PF6/AN6 | $\mathrm{V}_{\text {ss }}$ |
| 106 | PF7/AN7 | $\mathrm{V}_{\text {ss }}$ |
| 107 | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{V}_{\text {cc }}$ |
| 108 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 109 | PE5/TIOC1B | NC |
| 110 | NC | NC |
| 111 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 112 | PE6/TIOC2A | NC |
| 113 | PE7/TIOC2B | NC |
| 114 | PE8/TIOC3A | NC |
| 115 | PE9/TIOC3B | NC |
| 116 | PE10/TIOC3C | NC |
| 117 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 118 | PE11/TIOC3D | NC |
| 119 | PE12/TIOC4A | NC |
| 120 | PE13/TIOC4B/MRES | NC |

Table 1.4 Pin Arrangement by Mode for SH7041, SH7043 (QFP-144 Pin)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 1 | PA23/WRHH | NC |
| 2 | PE14/TIOC4C/DACK0/ $\overline{\text { AH }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 3 | PA22/WRHL | NC |
| 4 | PA21/CASHH | NC |
| 5 | PE15/TIOC4D/DACK1/\RQOUT | $\overline{C E}$ |
| 6 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 7 | PCO/AO | A0 |
| 8 | PC1/A1 | A1 |
| 9 | PC2/A2 | A2 |
| 10 | PC3/A3 | A3 |
| 11 | PC4/A4 | A4 |
| 12 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 13 | PC5/A5 | A5 |
| 14 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 15 | PC6/A6 | A6 |
| 16 | PC7/A7 | A7 |
| 17 | PC8/A8 | A8 |
| 18 | PC9/A9 | NC |
| 19 | PC10/A10 | A10 |
| 20 | PC11/A11 | A11 |
| 21 | PC12/A12 | A12 |
| 22 | PC13/A13 | A13 |
| 23 | PC14/A14 | A14 |
| 24 | PC15/A15 | A15 |
| 25 | PB0/A16 | A16 |
| 26 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 27 | PB1/A17 | NC |
| 28 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 29 | PA20/CASHL | NC |
| 30 | PA19/䃌CK/DRAK1 | NC |

Table 1.4 Pin Arrangement by Mode for SH7041, SH7043 (QFP-144 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 31 | PB2/ $\overline{\mathrm{RQQ}} / \overline{\mathrm{POE} 0} / \overline{\mathrm{RAS}}$ | NC |
| 32 |  | $\overline{\mathrm{OE}}$ |
| 33 |  | NC |
| 34 | PB4/ $\overline{\mathrm{RQ2}} / \overline{\mathrm{POE} 2} / \overline{\mathrm{CASH}}$ | $\overline{\text { PGM }}$ |
| 35 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 36 | PB5/IRQ3/ $\overline{\mathrm{POE3}} / \mathrm{RDWR}$ | $\mathrm{V}_{\text {cc }}$ |
| 37 | PB6/IRQ4/A18/ $\overline{\mathrm{BACK}}$ | NC |
| 38 | PB7/IRQ5/A19/ $\overline{\mathrm{BREQ}}$ | NC |
| 39 | PB8/\} \overline {  RQ6/A20/WAIT  } | NC |
| 40 | $V_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 41 | PB9/ $\overline{\text { RQ7 } / \text { A21/ADTRG }}$ | NC |
| 42 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 43 | PA14/ $\overline{\mathrm{RD}}$ | NC |
| 44 | WDTOVF | NC |
| 45 | PD31/D31//̈DTRG | NC |
| 46 | PD30/D30/IRQOUT | NC |
| 47 | PA13/WRH | NC |
| 48 | PA12/WRL | NC |
| 49 | PA11/ES1 | NC |
| 50 | PA10/ES0 | NC |
| 51 | PA9/TCLKD/ $\overline{\text { RQ3 }}$ | NC |
| 52 | PA8/TCLKC/ $\overline{\mathrm{RQ2}}$ | NC |
| 53 | PA7/TCLKB/ $\overline{C S 3}$ | NC |
| 54 | PA6/TCLKA/ $\overline{C S 2}$ | NC |
| 55 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 56 | PD29/D29/CS3 | NC |
| 57 | PD28/D28/CS2 | NC |
| 58 | PD27/D27/DACK1 | NC |
| 59 | PD26/D26/DACK0 | NC |
| 60 | PD25/D25/DREQ1 | NC |

Table 1.4 Pin Arrangement by Mode for SH7041, SH7043 (QFP-144 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 61 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 62 | PD24/D24/DREQ0 | NC |
| 63 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 64 | PD23/D23/IRQ7 | NC |
| 65 | PD22/D22/IRQ6 | NC |
| 66 | PD21/D21/]/RQ5 | NC |
| 67 | PD20/D20//7RQ4 | NC |
| 68 | PD19/D19/㡽Q3 | NC |
| 69 | PD18/D18/IRQ2 | NC |
| 70 | PD17/D17/IRQ1 | NC |
| 71 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 72 | PD16/D16/IRQ0 | NC |
| 73 | PD15/D15 | NC |
| 74 | PD14/D14 | NC |
| 75 | PD13/D13 | NC |
| 76 | PD12/D12 | NC |
| 77 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 78 | PD11/D11 | NC |
| 79 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 80 | PD10/D10 | NC |
| 81 | PD9/D9 | NC |
| 82 | PD8/D8 | NC |
| 83 | PD7/D7 | D7 |
| 84 | PD6/D6 | D6 |
| 85 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\text {cc }}$ |
| 86 | PD5 /D5 | D5 |
| 87 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 88 | PD4/D4 | D4 |
| 89 | PD3/D3 | D3 |
| 90 | PD2/D2 | D2 |

Table 1.4 Pin Arrangement by Mode for SH7041, SH7043 (QFP-144 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 91 | PD1/D1 | D1 |
| 92 | PDO/D0 | D0 |
| 93 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 94 | XTAL | NC |
| 95 | MD3 | $\mathrm{V}_{\text {cc }}$ |
| 96 | EXTAL | $\mathrm{V}_{\text {ss }}$ |
| 97 | MD2 | $\mathrm{V}_{\text {cc }}$ |
| 98 | NMI | A9 |
| 99 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 100 | PA16/ $\overline{\text { AH }}$ | NC |
| 101 | PA17/WAIT | NC |
| 102 | MD1 | $\mathrm{V}_{\mathrm{cc}}$ |
| 103 | MD0 | $\mathrm{V}_{\text {cc }}$ |
| 104 | PLLVCC | $\mathrm{V}_{\text {cc }}$ |
| 105 | PLLCAP | $\mathrm{V}_{\text {ss }}$ |
| 106 | PLLVSS | $\mathrm{V}_{\text {ss }}$ |
| 107 | PA15/CK | NC |
| 108 | $\overline{\text { RES }}$ | $\mathrm{V}_{\text {PP }}$ |
| 109 | PE0/TIOCOA/気REQ0 | NC |
| 110 | PE1/TIOCOB/DRAK0 | NC |
| 111 | PE2/TIOC0C/DREQ1 | NC |
| 112 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 113 | PE3/TIOCOD/DRAK1 | NC |
| 114 | PE4/TIOC1A | NC |
| 115 | PE5/TIOC1B | NC |
| 116 | PE6/TIOC2A | NC |
| 117 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 118 | PFO/AN0 | $\mathrm{V}_{\text {ss }}$ |
| 119 | PF1/AN1 | $\mathrm{V}_{\text {ss }}$ |
| 120 | PF2/AN2 | $\mathrm{V}_{\mathrm{ss}}$ |

Table 1.4 Pin Arrangement by Mode for SH7041, SH7043 (QFP-144 Pin) (cont)

| Pin No. | MCU Mode | PROM Mode |
| :---: | :---: | :---: |
| 121 | PF3/AN3 | $\mathrm{V}_{\text {ss }}$ |
| 122 | PF4/AN4 | $\mathrm{V}_{\text {ss }}$ |
| 123 | PF5/AN5 | $\mathrm{V}_{\text {ss }}$ |
| 124 | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 125 | PF6/AN6 | $\mathrm{V}_{\text {ss }}$ |
| 126 | PF7/AN7 | $\mathrm{V}_{\text {ss }}$ |
| 127 | AVref | $\mathrm{V}_{\mathrm{cc}}$ |
| 128 | $\mathrm{AV}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 129 | $V_{\text {Ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 130 | PA0/RXD0 | NC |
| 131 | PA1/TXD0 | NC |
| 132 | PA2/SCK0/DREQ0 $/ \overline{\text { REQ }}$ | NC |
| 133 | PA3/RXD1 | NC |
| 134 | PA4/TXD1 | NC |
| 135 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 136 | PA5 /SCK1/(DREQ1//̄REQ1 | NC |
| 137 | PE7/TIOC2B | NC |
| 138 | PE8/TIOC3A | NC |
| 139 | PE9/TIOC3B | NC |
| 140 | PE10/TIOC3C | NC |
| 141 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 142 | PE11/TIOC3D | NC |
| 143 | PE12/TIOC4A | NC |
| 144 | PE13/TIOC4B /MRES | NC |

Table 1.5 Pin Arrangement by Mode for SH7044 (QFP-112 Pin)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 1 | PE14/TIOC4C/DACK0/AH | NC |
| 2 | PE15/TIOC4D/DACK1/\RQOUT | NC |
| 3 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 4 | PC0/A0 | A0 |
| 5 | PC1/A1 | A1 |
| 6 | PC2/A2 | A2 |
| 7 | PC3/A3 | A3 |
| 8 | PC4/A4 | A4 |
| 9 | PC5/A5 | A5 |
| 10 | PC6/A6 | A6 |
| 11 | PC7/A7 | A7 |
| 12 | PC8/A8 | A8 |
| 13 | PC9/A9 | A9 |
| 14 | PC10/A10 | A10 |
| 15 | PC11/A11 | A11 |
| 16 | PC12/A12 | A12 |
| 17 | PC13/A13 | A13 |
| 18 | PC14/A14 | A14 |
| 19 | PC15/A15 | A15 |
| 20 | PB0/A16 | A16 |
| 21 | $V_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 22 | PB1/A17 | NC |
| 23 | $V_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 24 | PB2/ $\overline{\mathrm{TRQ} 0} / \overline{\mathrm{POE} 0} / \overline{\mathrm{RAS}}$ | NC |
| 25 |  | NC |
| 26 | PB4/ $\overline{\mathrm{IRQ2}} / \overline{\mathrm{POE} 2} / \overline{\mathrm{CASH}}$ | A17 |
| 27 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 28 | PB5//]RQ3/ $\overline{\mathrm{POE} 3} / \mathrm{RDWR}$ | NC |
| 29 | PB6//]RQ4/A18/ $\overline{\mathrm{BACK}}$ | NC |
| 30 |  | NC |
| 31 | PB8/I/RQ6/A20/WAIT | NC |
| 32 | PB9//̄RQ7/A21/ADTRG | NC |

Table 1.5 Pin Arrangement by Mode for SH7044 (QFP-112 Pin) (cont)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 33 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 34 | PA14/ $\overline{\mathrm{RD}}$ | NC |
| 35 | WDTOVF | NC |
| 36 | PA13/WRH | NC |
| 37 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 38 | PA12/WRL | NC |
| 39 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 40 | PA11/ट्CS1 | NC |
| 41 | PA10/CS0 | NC |
| 42 | PA9/TCLKD/\} \overline {  RQ3  } | $\overline{C E}$ |
| 43 | PA8/TCLKC/ $\overline{\text { RQ2 }}$ | $\overline{\mathrm{OE}}$ |
| 44 | PA7/TCLKB/CS3 | WE |
| 45 | PA6/TCLKA/CS2 | NC |
| 46 | PA5/SCK1/DREQ1/IRQ1 | $\mathrm{V}_{\mathrm{cc}}$ |
| 47 | PA4/TXD1 | NC |
| 48 | PA3/RXD1 | NC |
| 49 | PA2/SCK0/DREQ0/IRQ0 | $\mathrm{V}_{\mathrm{cc}}$ |
| 50 | PA1/TXD0 | $\mathrm{V}_{\mathrm{cc}}$ |
| 51 | PA0/RXD0 | NC |
| 52 | PD15/D15 | NC |
| 53 | PD14/D14 | NC |
| 54 | PD13/D13 | NC |
| 55 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 56 | PD12/D12 | NC |
| 57 | PD11/D11 | NC |
| 58 | PD10/D10 | NC |
| 59 | PD9/D9 | NC |
| 60 | PD8/D8 | NC |
| 61 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 62 | PD7/D7 | D7 |
| 63 | PD6/D6 | D6 |
| 64 | PD5/D5 | D5 |

Table 1.5 Pin Arrangement by Mode for SH7044 (QFP-112 Pin) (cont)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 65 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 66 | PD4/D4 | D4 |
| 67 | PD3/D3 | D3 |
| 68 | PD2/D2 | D2 |
| 69 | PD1/D1 | D1 |
| 70 | PDO/D0 | D0 |
| 71 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 72 | XTAL | XTAL |
| 73 | MD3 | MD3 |
| 74 | EXTAL | EXTAL |
| 75 | MD2 | MD2 |
| 76 | NMI | $\mathrm{V}_{\mathrm{cc}}$ |
| 77 | $\mathrm{V}_{\text {cc }}$ (FWP)* | FWE |
| 78 | MD1 | MD1 |
| 79 | MD0 | MDO |
| 80 | $\mathrm{PLLV}_{\mathrm{cc}}$ | PLLV ${ }_{\text {cc }}$ |
| 81 | PLLCAP | PLLCAP |
| 82 | PLLV ${ }_{\text {ss }}$ | PLLV ${ }_{\text {ss }}$ |
| 83 | PA15/CK | NC |
| 84 | $\overline{\mathrm{RES}}$ | $\overline{\mathrm{RES}}$ |
| 85 | PE0/TIOCA/馬REQ0 | NC |
| 86 | PE1/TIOCB/DRAK0 | NC |
| 87 | PE2/TIOCC/DREQ1 | NC |
| 88 | PE3/TIOCD/DRAK1 | NC |
| 89 | PE4/TIOC1A | NC |
| 90 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 91 | PFO/AN0 | $\mathrm{V}_{\text {ss }}$ |
| 92 | PF1/AN1 | $\mathrm{V}_{\text {ss }}$ |
| 93 | PF2/AN2 | $V_{\text {Ss }}$ |
| 94 | PF3/AN3 | $\mathrm{V}_{\text {ss }}$ |
| 95 | PF4/AN4 | $\mathrm{V}_{\text {ss }}$ |
| 96 | PF5/AN5 | $\mathrm{V}_{\text {ss }}$ |

Note: * $\mathrm{V}_{\mathrm{cc}}$ in the mask version; FWP in the F-ZTAT version (however, FWE in the writer mode)

Table 1.5 Pin Arrangement by Mode for SH7044 (QFP-112 Pin) (cont)

| PinNo. | MCU | Writer mode |
| :--- | :--- | :--- |
| 97 | AV $_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 98 | PF6/AN6 | $\mathrm{V}_{\text {ss }}$ |
| 99 | PF7/AN7 | $\mathrm{V}_{\text {ss }}$ |
| 100 | AV $_{\text {cC }}$ | $\mathrm{V}_{\text {cc }}$ |
| 101 | $\mathrm{~V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 102 | PE5/TIOC1B | NC |
| 103 | $\mathrm{~V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 104 | PE6/TIOC2A | NC |
| 105 | PE7/TIOC2B | NC |
| 106 | PE8/TIOC3A | NC |
| 107 | PE9/TIOC3B | NC |
| 108 | PE10/TIOC3C | NC |
| 109 | $\mathrm{~V}_{\text {SS }}$ | $\mathrm{V}_{\text {ss }}$ |
| 110 | PE11/TIOC3D | NC |
| 111 | PE12/TIOC4A | NC |
| 112 | PE13/TIOC4B/MRES | NC |

Table 1.6 Pin Arrangement by Mode for SH7045 (QFP-144 Pin)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 1 | PA23/WRHH | NC |
| 2 | PE14/TIOC4C/DACK0/AH | NC |
| 3 | PA22/WRHL | NC |
| 4 | PA21/CASHH | NC |
| 5 | PE15/TIOC4D/DACK1/IRQOUT | NC |
| 6 | $\mathrm{V}_{\text {SS }}$ | $\mathrm{V}_{\text {ss }}$ |
| 7 | PCO/AO | A0 |
| 8 | PC1/A1 | A1 |
| 9 | PC2/A2 | A2 |
| 10 | PC3/A3 | A3 |
| 11 | PC4/A4 | A4 |
| 12 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 13 | PC5/A5 | A5 |
| 14 | $\mathrm{V}_{\text {SS }}$ | $\mathrm{V}_{\text {ss }}$ |
| 15 | PC6/A6 | A6 |
| 16 | PC7/A7 | A7 |
| 17 | PC8/A8 | A8 |
| 18 | PC9/A9 | A9 |
| 19 | PC10/A10 | A10 |
| 20 | PC11/A11 | A11 |
| 21 | PC12/A12 | A12 |
| 22 | PC13/A13 | A13 |
| 23 | PC14/A14 | A14 |
| 24 | PC15/A15 | A15 |
| 25 | PB0/A16 | A16 |
| 26 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 27 | PB1/A17 | NC |
| 28 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 29 | PA20/CASHL | NC |
| 30 | PA19/BACK/DRAK1 | NC |
| 31 | PB2//]R0/ $\overline{\mathrm{POE} 0} / \overline{\mathrm{RAS}}$ | NC |
| 32 |  | NC |
| 33 | PA18/䃌EQ/DRAK0 | NC |
| 34 | PB4//̄R2//®OE2/CASH | A17 |
| 35 | $\mathrm{V}_{\text {SS }}$ | $\mathrm{V}_{\text {ss }}$ |
| 36 | PB5/IRQ3/ $\overline{\mathrm{POE} 3} / \mathrm{RDWR}$ | NC |

Table 1.6 Pin Arrangement by Mode for SH7045 (QFP-144 Pin) (cont)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 37 | PB6/IRQ4/A18/BACK | NC |
| 38 | PB7/IRQ5/A19/BREQ | NC |
| 39 | PB8/IRQ6/A20/WAIT | NC |
| 40 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 41 | PB9//̄RQ7/A21/商TRG | NC |
| 42 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 43 | PA14/ $\overline{\mathrm{RD}}$ | NC |
| 44 | WDTOVF | NC |
| 45 | PD31/D31//̈DTRG | NC |
| 46 | PD30/D30/IRQOUT | NC |
| 47 | PA13/WRH | NC |
| 48 | PA12/WRL | NC |
| 49 | PA11/CS1 | NC |
| 50 | PA10/CS0 | NC |
| 51 | PA9/TCLKD/\} \overline {  RQ3  } | $\overline{C E}$ |
| 52 | PA8/TCLKC/\} \overline {  RQ2  } | $\overline{\mathrm{OE}}$ |
| 53 | PA7/TCLKB/CS3 | WE |
| 54 | PA6/TCLKA/CS2 | NC |
| 55 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 56 | PD29/D29/CS3 | NC |
| 57 | PD28/D28/CS2 | NC |
| 58 | PD27/D27/DACK1 | NC |
| 59 | PD26/D26/DACK0 | NC |
| 60 | PD25/D25/DREQ1 | NC |
| 61 | $\mathrm{V}_{\text {SS }}$ | $\mathrm{V}_{\text {ss }}$ |
| 62 | PD24/D24/\REQ0 | NC |
| 63 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 64 | PD23/D23//TRQ7 | NC |
| 65 | PD22/D22/]RQ6 | NC |
| 66 | PD21/D21/厄/RQ5 | NC |
| 67 | PD20/D20/\RQ4 | NC |
| 68 | PD19/D19/IRQ3 | NC |
| 69 | PD18/D18/\} \overline { R 2 } | NC |
| 70 | PD17/D17/\} \overline {  RQ1  } | NC |
| 71 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 72 | PD16/D16/IRQ0 | NC |

Table 1.6 Pin Arrangement by Mode for SH7045 (QFP-144 Pin) (cont)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 73 | PD15/D15 | NC |
| 74 | PD14/D14 | NC |
| 75 | PD13/D13 | NC |
| 76 | PD12/D12 | NC |
| 77 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 78 | PD11/D11 | NC |
| 79 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {SS }}$ |
| 80 | PD10/D10 | NC |
| 81 | PD9/D9 | NC |
| 82 | PD8/D8 | NC |
| 83 | PD7/D7 | D7 |
| 84 | PD6/D6 | D6 |
| 85 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 86 | PD5/D5 | D5 |
| 87 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 88 | PD4/D4 | D4 |
| 89 | PD3/D3 | D3 |
| 90 | PD2/D2 | D2 |
| 91 | PD1/D1 | D1 |
| 92 | PD0/D0 | D0 |
| 93 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 94 | XTAL | XTAL |
| 95 | MD3 | MD3 |
| 96 | EXTAL | EXTAL |
| 97 | MD2 | MD2 |
| 98 | NMI | $\mathrm{V}_{\mathrm{cc}}$ |
| 99 | $\mathrm{V}_{\text {CC }}$ (FWP)* | FWE |
| 100 | PA16/ $\overline{\mathrm{AH}}$ | NC |
| 101 | PA17/WAIT | NC |
| 102 | MD1 | MD1 |
| 103 | MD0 | MD0 |
| 104 | $\mathrm{PLLV}_{\text {cc }}$ | $\mathrm{PLLV}_{\text {cc }}$ |
| 105 | PLLCAP | PLLCAP |
| 106 | $\mathrm{PLLV}_{\text {ss }}$ | $\mathrm{PLLV}_{\text {ss }}$ |
| 107 | PA15/CK | NC |

Note: * $\mathrm{V}_{\mathrm{cc}}$ in the mask version; FWP in the F-ZTAT version (however, FWE in the writer mode)

Table 1.6 Pin Arrangement by Mode for SH7045 (QFP-144 Pin) (cont)

| PinNo. | MCU | Writer mode |
| :---: | :---: | :---: |
| 108 | $\overline{\text { RES }}$ | $\overline{\mathrm{RES}}$ |
| 109 | PE0/TIOCOA/ $\overline{\text { DREQ0 }}$ | NC |
| 110 | PE1/TIOC0B/DRAK0 | NC |
| 111 | PE2/TIOC0C/DREQ1 | NC |
| 112 | $\mathrm{V}_{\mathrm{cc}}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 113 | PE3/TIOC0D/DRAK1 | NC |
| 114 | PE4/TIOC1A | NC |
| 115 | PE5/TIOC1B | NC |
| 116 | PE6/TIOC2A | NC |
| 117 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 118 | PFO/AN0 | $\mathrm{V}_{\text {ss }}$ |
| 119 | PF1/AN1 | $\mathrm{V}_{\text {SS }}$ |
| 120 | PF2/AN2 | $\mathrm{V}_{\text {SS }}$ |
| 121 | PF3/AN3 | $\mathrm{V}_{\text {SS }}$ |
| 122 | PF4/AN4 | $\mathrm{V}_{\text {SS }}$ |
| 123 | PF5/AN5 | $\mathrm{V}_{\text {SS }}$ |
| 124 | $\mathrm{AV}_{\text {ss }}$ | $\mathrm{V}_{\text {ss }}$ |
| 125 | PF6/AN6 | $\mathrm{V}_{\text {SS }}$ |
| 126 | PF7/AN7 | $\mathrm{V}_{\text {ss }}$ |
| 127 | AVref | $\mathrm{V}_{\text {cc }}$ |
| 128 | $\mathrm{AV}_{\text {cc }}$ | $\mathrm{V}_{\mathrm{cc}}$ |
| 129 | $\mathrm{V}_{\text {SS }}$ | $\mathrm{V}_{\text {ss }}$ |
| 130 | PA0/RXD0 | NC |
| 131 | PA1/TXD0 | $\mathrm{V}_{\text {cc }}$ |
| 132 | PA2/SCK0/DREQ0/IRQ0 | $\mathrm{V}_{\text {cc }}$ |
| 133 | PA3/RXD1 | NC |
| 134 | PA4/TXD1 | NC |
| 135 | $\mathrm{V}_{\text {cc }}$ | $\mathrm{V}_{\text {cc }}$ |
| 136 | PA5/SCK1/DREQ1/IRQ1 | $\mathrm{V}_{\mathrm{cc}}$ |
| 137 | PE7/TIOC2B | NC |
| 138 | PE8/TIOC3A | NC |
| 139 | PE9/TIOC3B | NC |
| 140 | PE10/TIOC3C | NC |
| 141 | $\mathrm{V}_{\text {ss }}$ | $\mathrm{V}_{\text {SS }}$ |
| 142 | PE11/TIOC3D | NC |
| 143 | PE12/TIOC4A | NC |
| 144 | PE13/TIOC4B/MRES | NC |

### 1.3.3 Pin Functions

Table 1.7 lists the pin functions.
Table 1.7 Pin Functions

| Classification | Symbol | I/O | Name | Function |
| :---: | :---: | :---: | :---: | :---: |
| Power supply | $\mathrm{V}_{\text {cc }}$ | 1 | Supply | Connects to power supply. <br> Connect all $\mathrm{V}_{\mathrm{cc}}$ pins to the system supply. No operation will occur if there are any open pins. |
|  | $\mathrm{V}_{\text {ss }}$ | I | Ground | Connects to ground. <br> Connect all $\mathrm{V}_{\text {ss }}$ pins to the system ground. No operation will occur if there are any open pins. |
|  | $\mathrm{V}_{\text {PP }}$ | I | Program supply | Connects to the power supply ( $\mathrm{V}_{\mathrm{cc}}$ ) during normal operation. <br> When in PROM mode, apply 12.5 V . |
| Clock | PLLVCC | I | PLL supply | On-chip PLL oscillator supply. |
|  | PLLVSS | I | PLL ground | On-chip PLL oscillator ground. |
|  | PLLCAP | I | PLL capacitance | On-chip PLL oscillator external capacitance connection pin. |
|  | EXTAL | I | External clock | Connect a crystal oscillator. Also, an external clock can be input to the EXTAL pin. |
|  | XTAL | I | Crystal | Connect a crystal oscillator. |
|  | CK | O | System clock | Supplies the system clock to peripheral devices. |
| System control | $\overline{\mathrm{RES}}$ | I | Power-on reset | Power-on reset when low |
|  | MRES | I | Manual reset | Manual reset when low |
|  | $\overline{\text { WDTOVF }}$ | O | Watchdog timer overflow | Overflow output signal from WDT |
|  | $\overline{\text { BREQ }}$ | 1 | Bus request | Goes low when external device requests bus right release |
|  | $\overline{\text { BACK }}$ | O | Bus request acknowledge | Indicates that bus right has been released to external device. The device that output the $\overline{\mathrm{BREQ}}$ signal receives the $\overline{B A C K}$ signal, notifying the device that it has obtained the bus right. |

## Renesns

Table 1.7 Pin Functions (cont)

| Classification | Symbol | I/O | Name | Function |
| :---: | :---: | :---: | :---: | :---: |
| Operating mode control | MD0-MD3 | I | Mode set | Determines the operating mode. Do not change input value during operation. |
|  | FWP | I | Flash memory write protect | Protects flash memory from being written or deleted. |
| Interrupts | NMI | I | Non-maskable interrupt | Non-maskable interrupt request pin. Enables selection of whether to accept on the rising or falling edge. |
|  | $\frac{\overline{\mathrm{IRQ0}}}{\overline{\mathrm{IRQ7}}}$ | I | Interrupt requests 0-7 | Maskable interrupt request pins. Allows selection of level input and edge input. |
|  | $\overline{\text { IRQOUT }}$ | 0 | Interrupt request output | Indicates that interrupt cause has occurred. Enables notification of interrupt generation also during bus release. |
| Address bus | A0-A21 | 0 | Address bus | Outputs addresses. |
| Data bus | $\begin{aligned} & \text { D0-D15 } \\ & \text { (QFP-112) } \end{aligned}$ | I/O | Data bus | 16-bit (QFP-112 pin and TQFP-120 pin versions) or 32-bit (QFP-144 pin |
|  | $\begin{aligned} & \text { D0-D31 } \\ & \text { (QFP-144) } \end{aligned}$ |  |  | version) bidirectional data bus. |
| Bus control | $\overline{\mathrm{CSO}}-\overline{\mathrm{CS3}}$ | O | Chip selects 0-3 | Chip select signals for external memory or devices. |
|  | $\overline{\mathrm{RD}}$ | 0 | Read | Indicates reading from an external device. |
|  | $\overline{\text { WRH }}$ | 0 | Upper write | Indicates writing the upper 8 bits (15-8) of external data. |
|  | $\overline{\text { WRL }}$ | 0 | Lower write | Indicates writing the lower 8 bits (7-0) of external data. |
|  | WAIT | I | Wait | Input causes insertion of wait cycles into the bus cycle during external space access. |
|  | $\overline{\text { RAS }}$ | 0 | Row address strobe | Timing signal for DRAM row address strobe. |
|  | $\overline{\text { CASH }}$ | 0 | Upper column address strobe | Timing signal for DRAM column address strobe. |
|  |  |  |  | Output when the upper 8 bits of data are accessed. |

Table 1.7 Pin Functions (cont)

| Classification | Symbol | I/O | Name | Function |
| :---: | :---: | :---: | :---: | :---: |
| Bus control (cont) | $\overline{\text { CASL }}$ | O | Lower column address strobe | Timing signal for DRAM column address strobe. |
|  |  |  |  | Output when the lower 8 bits of data are accessed. |
|  | RDWR | O | DRAM read/write | DRAM write strobe signal. |
|  | $\overline{\mathrm{AH}}$ | O | Address hold | Address hold timing signal for devices using an address/data multiplex bus. |
|  | $\overline{\text { WRHH }}$ <br> (QFP-144) | O | HH write | Indicates the writing of bits 31 to 24 of external data. |
|  | WRHL <br> (QFP-144) | O | HL write | Indicates the writing of bits 23 to 16 of external data. |
|  | $\overline{\text { CASHH }}$ <br> (QFP-144) | O | HH column address strobe | Timing signal for DRAM column address strobe. Output when bits 31 to 24 of data are accessed. |
|  | $\overline{\text { CASHL }}$ <br> (QFP-144) | O | HL column address strobe | Timing signal for DRAM column address strobe. Output when bits 23 to 16 of data are accessed. |
| Bus control multifunction timer/pulse unit | TCLKA | I | MTU timer clock input | Input pins for external clocks to the MTU counter. |
|  | TCLKB |  |  |  |
|  | TCLKC |  |  |  |
|  | TCLKD |  |  |  |
|  | TIOCOA | I/O | MTU input capture/ output compare (channel 0) | Channel 0 input capture input/output compare output/PWM output pins. |
|  | TIOCOB |  |  |  |
|  | TIOCOC |  |  |  |
|  | TIOCOD |  |  |  |
|  | TIOC1A | I/O | MTU input capture/output compare (channel 1) | Channel 1 input capture input/output compare output/PWM output pins. |
|  | TIOC1B |  |  |  |
|  | TIOC2A | I/O | MTU input capture/output compare (channel 2) | Channel 2 input capture input/output compare output/PWM output pins. |
|  | TIOC2B |  |  |  |

## Renesns

Table 1.7 Pin Functions (cont)
$\left.\begin{array}{lllll}\text { Classification } & \text { Symbol } & \text { I/O } & \text { Name } & \text { Function } \\ \hline \begin{array}{l}\text { Bus control } \\ \text { multifunction } \\ \text { timer/pulse unit } \\ \text { (cont) }\end{array} & \text { TIOC3A } & \text { I/O } & \text { MTU input } \\ \text { capture/output } \\ \text { TIOC3C }\end{array} \quad \begin{array}{l}\text { Channel 3 input capture input/output } \\ \text { compare output/PWM output pins. } \\ \text { (channel 3) }\end{array}\right]$.

Table 1.7 Pin Functions (cont)

| Classification | Symbol | I/O | Name | Function |
| :---: | :---: | :---: | :---: | :---: |
| I/O ports | $\overline{\overline{\mathrm{POEO}}}-$ | 1 | Port output enable | Input pin for port pin drive control when general use ports are established as output. |
|  | PA0- <br> PA15 <br> (QFP-112) | I/O | General purpose port | General purpose input/output port pins. <br> Each bit can be designated for |
|  | PAO- <br> PA23 <br> (QFP-144) |  |  | input/output. |
|  | PB0-PB9 | I/O | General purpose port | General purpose input/output port pins. |
|  |  |  |  | Each bit can be designated for input/output. |
|  | $\begin{aligned} & \text { PC0- } \\ & \text { PC15 } \end{aligned}$ | I/O | General purpose port | General purpose input/output port pins. |
|  |  |  |  | Each bit can be designated for input/output. |
|  | PD0- <br> PD15 <br> (QFP-112) | I/O | General purpose port | General purpose input/output port pins. <br> Each bit can be designated for |
|  | $\begin{aligned} & \text { PD0- } \\ & \text { PD31 } \\ & \text { (QFP-144) } \end{aligned}$ |  |  | input/output. |
|  | $\begin{aligned} & \text { PE0- } \\ & \text { PE15 } \end{aligned}$ | I/O | General purpose port | General purpose input/output port pins. |
|  |  |  |  | Each bit can be designated for input/output. |
|  | PF0-PF7 | I | General purpose port | General purpose input port pins. |

## Usage Notes

1. Unused input pins should be pulled up or pulled down.
2. The $\overline{\text { WDTOVF }}$ pin should not be pulled down in the SH7044/SH7045 F-ZTAT version. However, if it is necessary to pull this pin down, a resistance of $100 \mathrm{k} \Omega$ or higher should be used.

### 1.4 The F-ZTAT Version Onboard Programming

There are 2 modes on the F-ZTAT version: a mode that writes and overwrites programs using the special writer and a mode that writes and overwrites programs onboard the application system.

When rebooting after setting each mode pin and FWP pin during the reset condition, the microcomputer will transfer to one of the modes indicated in figure 1.6. In the user mode, data can be read from the flash memory but cannot be written or deleted. Use the boot mode and the user program mode to write to the flash memory or delete data.

In the boot mode, SCI1 (TXD1, RXD1) is used for data transfer. It is possible to automatically adjust the transfer bit rate to the transfer bit rate of the host.

Table 1.8 Pins during the Onboard Programming Mode

| Notation | I/O | Function |
| :--- | :--- | :--- |
| FWP | Input | Hardware protected flash memory write/delete |
| MD1 | Input | User programming mode/boot mode setting |
| MD2 | Input | Clock mode (PLL) setting |
| MD3 | Input | Clock mode (PLL) setting |
| TxD1 | Output | Serial sent data output |
| RxD1 | Input | Serial receive data input |



Figure 1.6 Condition Transfer for Flash Memory


Figure. 1.7 Data Transfer during Boot Mode

## Section 2 CPU

### 2.1 Register Configuration

The register set consists of sixteen 32-bit general registers, three 32-bit control registers and four 32-bit system registers.

### 2.1.1 General Registers (Rn)

The sixteen 32-bit general registers ( Rn ) are numbered R0-R15. General registers are used for data processing and address calculation. R0 is also used as an index register. Several instructions have R0 fixed as their only usable register. R15 is used as the hardware stack pointer (SP). Saving and recovering the status register (SR) and program counter (PC) in exception processing is accomplished by referencing the stack using R15. Figure 2.1 shows the general registers.

| RO *1 |
| :---: |
| R 1 |
| R 2 |
| R 3 |
| R 4 |
| R5 |
| R6 |
| R7 |
| R8 |
| R9 |
| R10 |
| R11 |
| R12 |
| R13 |
| R14 |
| R15, SP (hardware stack pointer) |

Notes: *1 R0 functions as an index register in the indirect indexed register addressing mode and indirect indexed GBR addressing mode. In some instructions, R0 functions as a fixed source register or destination register.
*2 R15 functions as a hardware stack pointer (SP) during exception processing.

Figure 2.1 General Registers

### 2.1.2 Control Registers

The 32-bit control registers consist of the 32-bit status register (SR), global base register (GBR), and vector base register (VBR). The status register indicates processing states. The global base register functions as a base address for the indirect GBR addressing mode to transfer data to the registers of on-chip peripheral modules. The vector base register functions as the base address of the exception processing vector area (including interrupts). Figure 2.2 shows a control register.


Figure 2.2 Control Registers

### 2.1.3 System Registers

System registers consist of four 32-bit registers: high and low multiply and accumulate registers (MACH and MACL), the procedure register (PR), and the program counter (PC). The multiply and accumulate registers store the results of multiply and accumulate operations. The procedure register stores the return address from the subroutine procedure. The program counter stores program addresses to control the flow of the processing. Figure 2.3 shows a system register.


Figure 2.3 System Registers

### 2.1.4 Initial Values of Registers

Table 2.1 lists the values of the registers after reset.
Table 2.1 Initial Values of Registers

| Classification | Register | Initial Value |
| :--- | :--- | :--- |
| General registers | R0-R14 | Undefined |
|  | R15 (SP) | Value of the stack pointer in the vector address table |
| Control registers | SR | Bits I3-I0 are 1111 (H'F), reserved bits are 0, and other <br> bits are undefined |
|  | GBR | Undefined |
|  | VBR | H'00000000 |
| System registers | MACH, MACL, PR | Undefined |
|  | PC | Value of the program counter in the vector address table |

### 2.2 Data Formats

### 2.2.1 Data Format in Registers

Register operands are always longwords ( 32 bits). When the memory operand is only a byte ( 8 bits) or a word ( 16 bits), it is sign-extended into a longword when loaded into a register (figure 2.4).

| 31 | Longword |
| :---: | :---: |

Figure 2.4 Longword Operand

### 2.2.2 Data Format in Memory

Memory data formats are classified into bytes, words, and longwords. Byte data can be accessed from any address, but an address error will occur if you try to access word data starting from an address other than 2 n or longword data starting from an address other than 4 n . In such cases, the data accessed cannot be guaranteed. The hardware stack area, referred to by the hardware stack pointer (SP, R15), uses only longword data starting from address 4 n because this area holds the program counter and status register (figure 2.5).


Figure 2.5 Byte, Word, and Longword Alignment

### 2.2.3 Immediate Data Format

Byte (8-bit) immediate data resides in an instruction code. Immediate data accessed by the MOV, ADD, and CMP/EQ instructions is sign-extended and handled in registers as longword data. Immediate data accessed by the TST, AND, OR, and XOR instructions is zero-extended and handled as longword data. Consequently, AND instructions with immediate data always clear the upper 24-bits of the destination register.

Word or longword immediate data is not located in the instruction code, but instead is stored in a memory table. An immediate data transfer instruction (MOV) accesses the memory table using the PC relative addressing mode with displacement.

### 2.3 Instruction Features

### 2.3.1 RISC-Type Instruction Set

All instructions are RISC type. This section details their functions.
16-Bit Fixed Length: All instructions are 16 bits long, increasing program code efficiency.
One Instruction per Cycle: The microprocessor can execute basic instructions in one cycle using the pipeline system. Instructions are executed in 35 ns at 28.7 MHz .

Data Length: Longword is the standard data length for all operations. Memory can be accessed in bytes, words, or longwords. Byte or word data accessed from memory is sign-extended and handled as longword data. Immediate data is sign-extended for arithmetic operations or zeroextended for logic operations. It also is handled as longword data (table 2.2).

Table 2.2 Sign Extension of Word Data

| SH7040 Series CPU |  | Description | Example of Conventional CPU |  |
| :---: | :---: | :---: | :---: | :---: |
| MOV.W | @ (disp, PC) , R1 | Data is sign-extended to 32 | ADD.W | \#H'1234,R0 |
| ADD | R1, R0 | bits, and R1 becomes |  |  |
| ADD | R1,R0 | H'00001234. It is next |  |  |
|  |  | operated upon by an ADD |  |  |
| . DATA.W | H'1234 | instruction. |  |  |

Note: @(disp, PC) accesses the immediate data.

Load-Store Architecture: Basic operations are executed between registers. For operations that involve memory access, data is loaded to the registers and executed (load-store architecture). Instructions such as AND that manipulate bits, however, are executed directly in memory.

Delayed Branch Instructions: Unconditional branch instructions are delayed. Executing the instruction that follows the branch instruction and then branching reduces pipeline disruption during branching (table 2.3). There are two types of conditional branch instructions: delayed branch instructions and ordinary branch instructions.

Table 2.3 Delayed Branch Instructions

| SH7040 Series CPU | Description | Example of Conventional CPU |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| BRA | TRGET | Executes an ADD before | ADD.W | R1,R0 |  |
| ADD | R1,R0 | branching to TRGET | BRA | TRGET |  |

Multiplication/Accumulation Operation: 16-bit $\times 16$-bit $\rightarrow$ 32-bit multiplication operations are executed in one to two cycles. 16-bit $\times 16$-bit +64 -bit $\rightarrow 64$-bit multiplication/accumulation operations are executed in two to three cycles. 32 -bit $\times 32$-bit $\rightarrow 64$-bit and 32 -bit $\times 32$-bit +64 bit $\rightarrow$ 64-bit multiplication/accumulation operations are executed in two to four cycles.

T Bit: The T bit in the status register changes according to the result of the comparison, and in turn is the condition (true/false) that determines if the program will branch. The number of instructions that change the T bit is kept to a minimum to improve the processing speed (table 2.4).

Table 2.4 T Bit

| SH7040 Series CPU |  | Description | Example of Conventional CPU |  |
| :---: | :---: | :---: | :---: | :---: |
| CMP / GE | R1, R0 | $T$ bit is set when $R 0 \geq R 1$. The | CMP.W | R1, R0 |
| BT | TRGET0 | program branches to TRGET0 when R0 $\geq$ R1 and to TRGET1 | BGE | TRGET0 |
| BF | TRGET1 | when R0<R1. | BLT | TRGET1 |
| ADD | \#1, R0 | T bit is not changed by ADD. T bit is | SUB.W | \#1, R0 |
| CMP / EQ | \#0,R0 | set when $\mathrm{RO}=0$. The program branches if $R 0=0$. | BEQ | TRGET |
| BT | TRGET |  |  |  |

Immediate Data: Byte (8-bit) immediate data resides in instruction code. Word or longword immediate data is not input via instruction codes but is stored in a memory table. An immediate data transfer instruction (MOV) accesses the memory table using the PC relative addressing mode with displacement (table 2.5).

Table 2.5 Immediate Data Accessing

| Classification | SH7040 Series CPU |  | Example of Conventional CPU |  |
| :---: | :---: | :---: | :---: | :---: |
| 8-bit immediate | MOV | \# H' 12,R0 | MOV.B | \# ${ }^{\prime}$ '12, R0 |
| 16-bit immediate | MOV.W | @ (disp, PC) , R0 | MOV.W | \#H'1234, R0 |
|  |  | . . . . . . . . . |  |  |
|  | . DATA.W | H'1234 |  |  |
| 32-bit immediate | MOV.L | @ (disp, PC) , R0 | MOV.L | \#H'12345678, R0 |
|  |  | .............. |  |  |
|  | . DATA.L | H'12345678 |  |  |

Note: @(disp, PC) accesses the immediate data.

Absolute Address: When data is accessed by absolute address, the value already in the absolute address is placed in the memory table. Loading the immediate data when the instruction is executed transfers that value to the register and the data is accessed in the indirect register addressing mode (table 2.6).

Table 2.6 Absolute Address Accessing


Note: @(disp,PC) accesses the immediate data.

16-Bit/32-Bit Displacement: When data is accessed by 16-bit or 32-bit displacement, the preexisting displacement value is placed in the memory table. Loading the immediate data when the instruction is executed transfers that value to the register and the data is accessed in the indirect indexed register addressing mode (table 2.7).

Table 2.7 Displacement Accessing

| Classification | SH7040 Series CPU |  | Example of Conventional CPU |  |
| :---: | :---: | :---: | :---: | :---: |
| 16-bit displacement | MOV.W | @ (disp, PC) , R0 | MOV.W | @ (H'1234, R1) , R2 |
|  | MOV.W | @ (R0, R1) , R2 |  |  |
|  |  | .......... |  |  |
|  | . DATA.W | H'1234 |  |  |

Note: @(disp,PC) accesses the immediate data.

### 2.3.2 Addressing Modes

Table 2.8 describes addressing modes and effective address calculation.
Table 2.8 Addressing Modes and Effective Addresses

| Addressing <br> Mode | Instruction <br> Format | Effective Addresses Calculation |
| :--- | :--- | :--- | :--- |

Table 2.8 Addressing Modes and Effective Addresses (cont)

| Addressing Mode | Instruction Format | Effective Addresses Calculation | Equation |
| :---: | :---: | :---: | :---: |
| Indirect register addressing with displacement | ```@(disp:4, Rn)``` | The effective address is Rn plus a 4-bit displacement (disp). The value of disp is zeroextended, and remains the same for a byte operation, is doubled for a word operation, and is quadrupled for a longword operation. | Byte: Rn + disp <br> Word: Rn + disp $\times 2$ <br> Longword: Rn $+\operatorname{disp} \times 4$ |
| Indirect indexed register addressing | @ (R0, Rn) | The effective address is the Rn value plus RO. | $\mathrm{Rn}+\mathrm{Ro}$ |
| Indirect GBR addressing with displacement | $\begin{aligned} & \text { @ (disp: 8, } \\ & \text { GBR) } \end{aligned}$ | The effective address is the GBR value plus an 8 -bit displacement (disp). The value of disp is zeroextended, and remains the same for a byte operation, is doubled for a word operation, and is quadrupled for a longword operation. | Byte: GBR + disp <br> Word: GBR + disp $\times 2$ <br> Longword: <br> GBR + disp $\times$ <br> 4 |

Table 2.8 Addressing Modes and Effective Addresses (cont)


Table 2.8 Addressing Modes and Effective Addresses (cont)

| Addressing Mode | Instruction Format | Effective Addresses Calculation | Equation |
| :---: | :---: | :---: | :---: |
| PC relative addressing | disp:8 | The effective address is the PC value sign-extended with an 8-bit displacement (disp), doubled, and added to the PC value. | PC + disp $\times 2$ |
|  |  | PC |  |
|  |  |  |  |
|  | disp:12 | The effective address is the PC value sign-extended with a 12-bit displacement (disp), doubled, and added to the PC value. | PC + disp $\times 2$ |



### 2.3.3 Instruction Format

Table 2.9 lists the instruction formats for the source operand and the destination operand. The meaning of the operand depends on the instruction code. The symbols are used as follows:

- xxxx: Instruction code
- mmmm: Source register
- nnnn: Destination register
- iiii: Immediate data
- dddd: Displacement

Table 2.9 Instruction Formats

| Instruction Formats |  |  |  | Source Operand | Destination Operand | Example |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 format |  |  |  | - | - | NOP |  |
| 15 0 |  |  |  |  |  |  |  |
| XXXX | XXXX | XXXX | XXXX |  |  |  |  |
| n format |  |  |  | - | nnnn: Direct register | MOVT | Rn |
| 15 |  |  |  | Control register | nnnn: Direct | STS | MACH, Rn |
| XXXX | nnnn | XXXX | xxxx | or system register | register |  |  |
|  |  |  |  | Control register or system register | nnnn: Indirect predecrement register | STC.L | SR, @-Rn |
| m format |  |  |  | mmmm: Direct register | Control register or system register | LDC | Rm, SR |
| 15 |  |  |  | mmmm: Indirect | Control register or | LDC.L | $@ R m+$, SR |
| Xxx | mmmm | xxxx | xxxx | post-increment register | system register |  |  |
|  |  |  |  | mmmm: Direct register | - | JMP | @Rm |
|  |  |  |  | mmmm: PC relative using Rm | - | BRAF | Rm |

Table 2.9 Instruction Formats (cont)

| Instruction Formats |  |  |  | Source Operand | Destination Operand | Example |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| nm format |  |  |  | mmmm: Direct register | nnnn: Direct register | ADD $\mathrm{Rm}, \mathrm{Rn}$ |
|  |  |  |  | mmmm: Direct | nnnn: Indirect | MOV.L Rm, @Rn |
| xxxx | nnnn | mmmm | xxxx | register | register |  |
|  |  |  |  | mmmm: Indirect post-increment register (multiply/ accumulate) | MACH, MACL | MAC.W @Rm+, @Rn+ |
|  |  |  |  | nnnn*: Indirect post-increment register (multiply/ accumulate) |  |  |
|  |  |  |  | mmmm: Indirect post-increment register | nnnn: Direct register | MOV.L @Rm+, Rn |
|  |  |  |  | mmmm: Direct register | nnnn: Indirect predecrement register | MOV.L Rm, @-Rn |
|  |  |  |  | mmmm: Direct register | nnnn: Indirect indexed register | $\begin{aligned} & \text { MOV.L } \\ & \text { Rm, @ (RO, Rn) } \end{aligned}$ |
| md format 15 |  |  |  | mmmmdddd: indirect register with displacement | R0 (Direct register) | $\begin{aligned} & \text { MOV.B } \\ & \text { @ (disp, Rm) , R0 } \end{aligned}$ |
|  | xxxx | mmmm |  |  |  |  |
| nd4 format$15$ |  |  |  | R0 (Direct register) | nnnndddd: Indirect register with displacement | $\begin{aligned} & \text { MOV.B } \\ & \text { RO, © (disp, Rn) } \end{aligned}$ |
| xxxx | xxxx | nnnn | dddd |  |  |  |
| nmd format$15$ |  |  |  | mmmm: Direct register | nnnndddd: Indirect register with displacement | MOV.L <br> Rm, @ (disp, Rn) |
|  | nnnn | mmmm |  |  |  |  |
|  |  |  |  | mmmmdddd: <br> Indirect register <br> with <br> displacement | nnnn: Direct register | $\begin{aligned} & \text { MOV.L } \\ & \text { @ (disp, Rm), Rn } \end{aligned}$ |

Note: * In multiply/accumulate instructions, nnnn is the source register.

Table 2.9 Instruction Formats (cont)

| Instruction Formats |  |  |  | Source Operand | Destination Operand | Example |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| d format <br> 15 <br> xxxx | xxxx | dddd | $\overbrace{\text { dddd }}{ }^{0}$ | dddddddd: <br> Indirect GBR <br> with <br> displacement | R0 (Direct register) | $\begin{aligned} & \text { MOV.L } \\ & \text { @ (disp, GBR), RO } \end{aligned}$ |  |
|  |  |  |  | R0(Direct register) | dddddddd: Indirect GBR with displacement | $\begin{aligned} & \text { MOV.L } \\ & \text { RO, @ (di } \end{aligned}$ | (sp, GBR) |
|  |  |  |  | dddddddd: PC relative with displacement | R0 (Direct register) | MOVA <br> @ (disp | $, P C), R 0$ |
|  |  |  |  | dddddddd: PC relative | - | BF | label |
| d12 format $15$ <br> 0 |  |  |  | dddddddddddd: PC relative | - | BRA label <br> (label = disp + PC) |  |
| nd8 format$15$ |  |  |  | dddddddd: PC relative with displacement | nnnn: Direct register | $\begin{aligned} & \text { MOV.L } \\ & \text { @ (disp, PC) ,Rn } \end{aligned}$ |  |
| i format |  |  |  | iiiiiiiii: Immediate | Indirect indexed GBR | AND. B <br> \#imm, @ (RO, GBR) |  |
| xxxx | xxxx | iiii | $\mathrm{iiii}^{0}$ | iiiiiiiii: Immediate | R0 (Direct register) | AND | \#imm, R0 |
|  |  |  |  | iiiiiiiii: Immediate | - | TRAPA | \#imm |
| ni format$15$ |  |  |  | iiiiiiii: Immediate | nnnn: Direct register | ADD | \#imm, Rn |
| xxxx | nnnn | iiii | iiii |  |  |  |  |

### 2.4 Instruction Set by Classification

Table 2.10 Classification of Instructions

| Classification | Types | Operation Code | Function | No. of Instructions |
| :---: | :---: | :---: | :---: | :---: |
| Data transfer | 5 | MOV | Data transfer, immediate data transfer, peripheral module data transfer, structure data transfer | 39 |
|  |  | MOVA | Effective address transfer |  |
|  |  | MOVT | T bit transfer |  |
|  |  | SWAP | Swap of upper and lower bytes |  |
|  |  | XTRCT | Extraction of the middle of registers connected |  |
| Arithmetic operations | 21 | ADD | Binary addition | 33 |
|  |  | ADDC | Binary addition with carry |  |
|  |  | ADDV | Binary addition with overflow check |  |
|  |  | CMP/cond | Comparison |  |
|  |  | DIV1 | Division |  |
|  |  | DIV0S | Initialization of signed division |  |
|  |  | DIVOU | Initialization of unsigned division |  |
|  |  | DMULS | Signed double-length multiplication |  |
|  |  | DMULU | Unsigned double-length multiplication |  |
|  |  | DT | Decrement and test |  |
|  |  | EXTS | Sign extension |  |
|  |  | EXTU | Zero extension |  |
|  |  | MAC | Multiply/accumulate, double-length multiply/accumulate operation |  |
|  |  | MUL | Double-length multiply operation |  |
|  |  | MULS | Signed multiplication |  |
|  |  | MULU | Unsigned multiplication |  |
|  |  | NEG | Negation |  |
|  |  | NEGC | Negation with borrow |  |
|  |  | SUB | Binary subtraction |  |
|  |  | SUBC | Binary subtraction with borrow |  |
|  |  | SUBV | Binary subtraction with underflow |  |

Table 2.10 Classification of Instructions (cont)

| Classification | Types | Operation Code | Function | No. of Instructions |
| :---: | :---: | :---: | :---: | :---: |
| Logic operations | 6 | AND | Logical AND | 14 |
|  |  | NOT | Bit inversion |  |
|  |  | OR | Logical OR |  |
|  |  | TAS | Memory test and bit set |  |
|  |  | TST | Logical AND and T bit set |  |
|  |  | XOR | Exclusive OR |  |
| Shift | 10 | ROTL | One-bit left rotation | 14 |
|  |  | ROTR | One-bit right rotation |  |
|  |  | ROTCL | One-bit left rotation with T bit |  |
|  |  | ROTCR | One-bit right rotation with T bit |  |
|  |  | SHAL | One-bit arithmetic left shift |  |
|  |  | SHAR | One-bit arithmetic right shift |  |
|  |  | SHLL | One-bit logical left shift |  |
|  |  | SHLLn | n-bit logical left shift |  |
|  |  | SHLR | One-bit logical right shift |  |
|  |  | SHLRn | n-bit logical right shift |  |
| Branch | 9 | BF | Conditional branch, conditional branch with delay (Branch when $\mathrm{T}=0$ ) | 11 |
|  |  | BT | Conditional branch, conditional branch with delay (Branch when $\mathrm{T}=1$ ) |  |
|  |  | BRA | Unconditional branch |  |
|  |  | BRAF | Unconditional branch |  |
|  |  | BSR | Branch to subroutine procedure |  |
|  |  | BSRF | Branch to subroutine procedure |  |
|  |  | JMP | Unconditional branch |  |
|  |  | JSR | Branch to subroutine procedure |  |
|  |  | RTS | Return from subroutine procedure |  |

Table 2.10 Classification of Instructions (cont)

|  | Operation <br> Classification <br> Code |  | Function |
| :--- | :--- | :--- | :--- | | No. of <br> Instructions |
| :--- |
| System <br> control |

Table 2.11 shows the format used in tables 2.12 to 2.17 , which list instruction codes, operation, and execution states in order by classification.

Table 2.11 Instruction Code Format

| Item | Format | Explanation |
| :--- | :--- | :--- |
| Instruction | OP.Sz SRC, DEST | OP: Operation code |
|  |  | Sz: Size (B: byte, W: word, or L: longword) |
|  |  | SRC: Source |
|  |  | DEST: Destination |
|  |  | Rm: Source register |
|  |  | Rn: Destination register |
|  |  | imm: Immediate data |
|  |  | disp: Displacement ${ }^{* 1}$ |
|  |  | mmmm: Source register |
|  |  | nnnn: Destination register |
| Instruction | MSB $\leftrightarrow$ LSB | $0000:$ R0 |
| code |  |  |
|  |  |  |

1111: R15
iiii: Immediate data
dddd: Displacement

| Operation | $\rightarrow, \leftarrow$ | Direction of transfer |
| :---: | :---: | :---: |
|  | (xx) | Memory operand |
|  | M/Q/T | Flag bits in the SR |
|  | \& | Logical AND of each bit |
|  | \| | Logical OR of each bit |
|  | $\wedge$ | Exclusive OR of each bit |
|  | $\sim$ | Logical NOT of each bit |
|  | <<n | n -bit left shift |
|  | >>n | n -bit right shift |
| Execution cycles | - | Value when no wait states are inserted*2 |
| T bit | - | Value of T bit after instruction is executed. An em-dash (-) in the column means no change. |

Notes: *1 Depending on the operand size, displacement is scaled $\times 1, \times 2$, or $\times 4$. For details, see the SH-1/SH-2/SH-DSP Programming Manual.
*2 Instruction execution cycles: The execution cycles shown in the table are minimums. The actual number of cycles may be increased when (1) contention occurs between instruction fetches and data access, or (2) when the destination register of the load instruction (memory $\rightarrow$ register) and the register used by the next instruction are the same.

Table 2.12 Data Transfer Instructions

| Instruction |  | Instruction Code | Operation | Execu- <br> tion <br> Cycles | $\begin{aligned} & \text { T } \\ & \text { Bit } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| MOV | \#imm, Rn | 1110nnnniiiiiiii | $\begin{aligned} & \# \text { imm } \rightarrow \text { Sign extension } \rightarrow \\ & \text { Rn } \end{aligned}$ | 1 | - |
| MOV.W | @ (disp, PC) , Rn | 1001 nnnndddddddd | $(\text { disp } \times 2+\mathrm{PC}) \rightarrow \text { Sign }$ <br> extension $\rightarrow$ Rn | 1 | - |
| MOV.L | @ (disp, PC) , Rn | 1101 nnnndddddddd | $(\operatorname{disp} \times 4+\mathrm{PC}) \rightarrow \mathrm{Rn}$ | 1 | - |
| MOV | $\mathrm{Rm}, \mathrm{Rn}$ | $0110 \mathrm{nnnnmmmm0011}$ | $\mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| MOV.B | Rm, @Rn | 0010 nnnnmmmm0000 | $\mathrm{Rm} \rightarrow(\mathrm{Rn})$ | 1 | - |
| MOV.W | $\mathrm{Rm}, @ \mathrm{Rn}$ | 0010 nnnnmmmm0001 | $\mathrm{Rm} \rightarrow(\mathrm{Rn})$ | 1 | - |
| MOV.L | $\mathrm{Rm}, @ \mathrm{Rn}$ | 0010 nnnnmmmm0010 | $\mathrm{Rm} \rightarrow(\mathrm{Rn})$ | 1 | - |
| MOV.B | $@ R m, R n$ | 0110nnnnmmmm0000 | $\begin{aligned} & (\text { Rm }) \rightarrow \text { Sign extension } \rightarrow \\ & \text { Rn } \end{aligned}$ | 1 | - |
| MOV.W | $@ R m, R n$ | 0110nnnnmmmm0001 | $\begin{aligned} & (\text { Rm }) \rightarrow \text { Sign extension } \rightarrow \\ & \text { Rn } \end{aligned}$ | 1 | - |
| MOV.L | @Rm, Rn | 0110nnnnmmmm0010 | $(\mathrm{Rm}) \rightarrow \mathrm{Rn}$ | 1 | - |
| MOV.B | Rm, @-Rn | $0010 \mathrm{nnnnmmmm0100}$ | $\mathrm{Rn}-1 \rightarrow \mathrm{Rn}, \mathrm{Rm} \rightarrow(\mathrm{Rn})$ | 1 | - |
| MOV.W | Rm, @-Rn | 0010 nnnnmmmm0101 | $\mathrm{Rn}-2 \rightarrow \mathrm{Rn}, \mathrm{Rm} \rightarrow(\mathrm{Rn})$ | 1 | - |
| MOV.L | Rm, @-Rn | 0010 nnnnmmmm0110 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{Rm} \rightarrow$ (Rn) | 1 | - |
| MOV.B | $@ R m+$ Rn | 0110nnnnmmmm0100 | $($ Rm) $\rightarrow$ Sign extension $\rightarrow$ $\mathrm{Rn}, \mathrm{Rm}+1 \rightarrow \mathrm{Rm}$ | 1 | - |
| MOV.W | $@ R m+$ Rn | $0110 \mathrm{nnnnmmmm0101}$ | $($ Rm) $\rightarrow$ Sign extension $\rightarrow$ $\mathrm{Rn}, \mathrm{Rm}+2 \rightarrow \mathrm{Rm}$ | 1 | - |
| MOV.L | $@ R m+$ Rn | 0110nnnnmmmm0110 | $(\mathrm{Rm}) \rightarrow \mathrm{Rn}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 1 | - |
| MOV.B | R0, @ (disp, Rn) | 10000000 nnnndddd | $\mathrm{RO} \rightarrow(\mathrm{disp}+\mathrm{Rn})$ | 1 | - |
| MOV.W | R0, © (disp, Rn) | 10000001 nnnndddd | $\mathrm{R0} \rightarrow(\mathrm{disp} \times 2+\mathrm{Rn})$ | 1 | - |
| MOV.L | Rm, @ (disp, Rn) | 0001 nnnnmmmmdddd | $R \mathrm{~m} \rightarrow(\mathrm{disp} \times 4+\mathrm{Rn})$ | 1 | - |
| MOV.B | © (disp, Rm) , R0 | 10000100 mmmmdddd | $\begin{aligned} & (\mathrm{disp}+\mathrm{Rm}) \rightarrow \text { Sign } \\ & \text { extension } \rightarrow \mathrm{RO} \end{aligned}$ | 1 | - |
| MOV.W | @ (disp, Rm) , R0 | 10000101 mmmmdddd | $\begin{aligned} & (\text { disp } \times 2+R m) \rightarrow \text { Sign } \\ & \text { extension } \rightarrow R 0 \end{aligned}$ | 1 | - |
| MOV.L | @ (disp, Rm) , Rn | 0101 nnnnmmmmdddd | $(\mathrm{disp} \times 4+\mathrm{Rm}) \rightarrow \mathrm{Rn}$ | 1 | - |
| MOV.B | Rm, @ (R0, Rn) | $0000 \mathrm{nnnnmmmm0100}$ | $\mathrm{Rm} \rightarrow(\mathrm{R0}+\mathrm{Rn})$ | 1 | - |

Table 2.12 Data Transfer Instructions (cont)

|  |  |  | Execu- <br> tion <br> Cles | T <br> Instruction | Operation |
| :--- | :--- | :--- | :--- | :--- | :--- |

Table 2.13 Arithmetic Operation Instructions

| Instruction |  | Instruction Code | Operation | Execution Cycles | T Bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ADD | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm1100 | $\mathrm{Rn}+\mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| ADD | \#imm, Rn | 0111 nnnniiiiiiii | $\mathrm{Rn}+\mathrm{imm} \rightarrow \mathrm{Rn}$ | 1 | - |
| ADDC | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm1110 | $\begin{aligned} & \mathrm{Rn}+\mathrm{Rm}+\mathrm{T} \rightarrow \mathrm{Rn}, \\ & \text { Carry } \rightarrow \mathrm{T} \end{aligned}$ | 1 | Carry |
| ADDV | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm1111 | $\begin{aligned} & \text { Rn }+R m \rightarrow R n, \\ & \text { Overflow } \rightarrow T \end{aligned}$ | 1 | Overflow |
| CMP/EQ | \#imm, R0 | 10001000iiiiiiii | If $\mathrm{RO}=\mathrm{imm}, 1 \rightarrow \mathrm{~T}$ | 1 | Comparison result |
| CMP/EQ | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0000 | If $\mathrm{Rn}=\mathrm{Rm}, 1 \rightarrow \mathrm{~T}$ | 1 | Comparison result |
| CMP/HS | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0010 | If $R n \geq R m$ with unsigned data, $1 \rightarrow T$ | 1 | Comparison result |
| CMP/GE | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0011 | If $R n \geq R m$ with signed data, $1 \rightarrow T$ | 1 | Comparison result |
| CMP/HI | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0110 | If $R n>R m$ with unsigned data, $1 \rightarrow T$ | 1 | Comparison result |
| CMP/GT | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0111 | If $R n>R m$ with signed data, $1 \rightarrow \mathrm{~T}$ | 1 | Comparison result |
| CMP /PL | Rn | 0100nnnn00010101 | If $\mathrm{Rn}>0,1 \rightarrow \mathrm{~T}$ | 1 | Comparison result |
| CMP /PZ | Rn | 0100nnnn00010001 | If $\mathrm{Rn} \geq 0,1 \rightarrow \mathrm{~T}$ | 1 | Comparison result |
| CMP/STR | $\mathrm{Rm}, \mathrm{Rn}$ | 0010nnnnmmmm1100 | If Rn and Rm have an equivalent byte, $1 \rightarrow T$ | 1 | Comparison result |
| DIV1 | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0100 | Single-step division (Rn/Rm) | 1 | Calculation result |
| DIV0S | $\mathrm{Rm}, \mathrm{Rn}$ | 0010nnnnmmmm0111 | MSB of $R n \rightarrow Q$, MSB of $\mathrm{Rm} \rightarrow \mathrm{M}, \mathrm{M}^{\wedge} \mathrm{Q} \rightarrow \mathrm{T}$ | 1 | Calculation result |
| DIVOU |  | 0000000000011001 | $0 \rightarrow \mathrm{M} / \mathrm{Q} / \mathrm{T}$ | 1 | 0 |

Table 2.13 Arithmetic Operation Instructions (cont)

| Instruction |  | Instruction Code | Operation | Execution Cycles | T Bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DMULS.L | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm1101 | Signed operation of Rn $\times \mathrm{Rm} \rightarrow \mathrm{MACH}$, MACL $32 \times 32 \rightarrow 64$ bit | 2 to 4* | - |
| DMULU.L | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm0101 | Unsigned operation of $\mathrm{Rn} \times \mathrm{Rm} \rightarrow \mathrm{MACH}$, MACL $32 \times 32 \rightarrow 64$ bit | 2 to 4* | - |
| DT | Rn | 0100nnnn00010000 | $\mathrm{Rn}-1 \rightarrow \mathrm{Rn}$, when Rn is $0,1 \rightarrow \mathrm{~T}$. When Rn is nonzero, $0 \rightarrow \mathrm{~T}$ | 1 | Comparison result |
| EXTS.B | $\mathrm{Rm}, \mathrm{Rn}$ | 0110 nnnnmmmm1110 | A byte in Rm is signextended $\rightarrow \mathrm{Rn}$ | 1 | - |
| EXTS.W | $\mathrm{Rm}, \mathrm{Rn}$ | 0110 nnnnmmmm1111 | A word in Rm is signextended $\rightarrow \mathrm{Rn}$ | 1 | - |
| EXTU.B | $\mathrm{Rm}, \mathrm{Rn}$ | 0110 nnnnmmmm1100 | A byte in Rm is zeroextended $\rightarrow \mathrm{Rn}$ | 1 | - |
| EXTU.W | $\mathrm{Rm}, \mathrm{Rn}$ | 0110 nnnnmmmm1101 | A word in Rm is zeroextended $\rightarrow$ Rn | 1 | - |
| MAC.L | $@ R m+$, @Rn+ | 0000nnnnmmmm1111 | Signed operation of $(R n) \times(R m)+M A C \rightarrow$ MAC $32 \times 32 \rightarrow 64$ bit | 3/(2 to <br> 4)* | - |
| MAC.W | @Rm+, @Rn+ | $0100 n n n n m m m 1111$ | Signed operation of $(R n) \times(R m)+M A C \rightarrow$ MAC $16 \times 16+64 \rightarrow$ 64 bit | 3/(2)* | - |
| MUL.L | $\mathrm{Rm}, \mathrm{Rn}$ | 0000nnnnmmmm0111 | $\begin{aligned} & \mathrm{Rn} \times \mathrm{Rm} \rightarrow \mathrm{MACL}, 32 \\ & \times 32 \rightarrow 32 \text { bit } \end{aligned}$ | 2 to 4* | - |
| MULS.W | $\mathrm{Rm}, \mathrm{Rn}$ | 0010nnnnmmmm1111 | Signed operation of Rn $\begin{aligned} & \times \operatorname{Rm} \rightarrow \mathrm{MAC} 16 \times 16 \\ & \rightarrow 32 \text { bit } \end{aligned}$ | 1 to 3* | - |
| MULU.W | $\mathrm{Rm}, \mathrm{Rn}$ | 0010nnnnmmmm1110 | Unsigned operation of $\mathrm{Rn} \times \mathrm{Rm} \rightarrow \mathrm{MAC} 16 \times$ $16 \rightarrow 32 \mathrm{bit}$ | 1 to 3* | - |
| NEG | $\mathrm{Rm}, \mathrm{Rn}$ | $0110 \mathrm{nnnnmmmm1011}$ | $0-\mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| NEGC | $\mathrm{Rm}, \mathrm{Rn}$ | 0110nnnnmmmm1010 | $\begin{aligned} & 0-\mathrm{Rm}-\mathrm{T} \rightarrow \text { Rn, Borrow } \\ & \rightarrow \mathrm{T} \end{aligned}$ | 1 | Borrow |

Table 2.13 Arithmetic Operation Instructions (cont)

|  |  |  | Execu- <br> tion <br> Instruction |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Instruction Code | Operation | T Bit |  |  |  |
| SUBC | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm1000 | $\mathrm{Rn}-\mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| SUBV | $\mathrm{Rm}, \mathrm{Rn}$ | 0011 nnnnmmmm1010 | $\mathrm{Rn}-\mathrm{Rm}-\mathrm{T} \rightarrow \mathrm{Rn}$, <br> Borrow $\rightarrow \mathrm{T}$ | 1 | Borrow |

Note: * The normal minimum number of execution cycles. (The number in parentheses is the number of cycles when there is contention with following instructions.)

Table 2.14 Logic Operation Instructions

| Instruction |  | Instruction Code | Operation | Execution Cycles | T Bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AND | $\mathrm{Rm}, \mathrm{Rn}$ | 0010 nnnnmmmm1001 | $\mathrm{Rn} \& \mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| AND | \#imm, R0 | 11001001iiiiiiii | R0 \& imm $\rightarrow$ R0 | 1 | - |
| AND. B | \#imm, @ (R0, GBR) | 11001101iiiiiiii | $\begin{aligned} & (\mathrm{R} 0+\mathrm{GBR}) \& \mathrm{imm} \rightarrow \\ & (\mathrm{RO}+\mathrm{GBR}) \end{aligned}$ | 3 | - |
| NOT | $\mathrm{Rm}, \mathrm{Rn}$ | 0110 nnnnmmmm0111 | $\sim \mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| OR | $\mathrm{Rm}, \mathrm{Rn}$ | 0010 nnnnmmmm1011 | $\mathrm{Rn} \mid \mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| OR | \#imm, R0 | 11001011iiiiiiii | $\mathrm{RO} \\| \mathrm{imm} \rightarrow \mathrm{RO}$ | 1 | - |
| OR.B | \#imm, © (R0, GBR) | 11001111iiiiiiii | $\begin{aligned} & (\mathrm{R} 0+\mathrm{GBR}) \mid \mathrm{imm} \rightarrow \\ & (\mathrm{RO}+\mathrm{GBR}) \end{aligned}$ | 3 | - |
| TAS.B | @Rn | 0100 nnnn 00011011 | $\begin{aligned} & \text { If }(\mathrm{Rn}) \text { is } 0,1 \rightarrow \mathrm{~T} ; 1 \rightarrow \\ & \text { MSB of }(\mathrm{Rn})^{*} \end{aligned}$ | 4 | Test result |
| TST | Rm, Rn | 0010nnnnmmmm1000 | Rn \& Rm; if the result is $0,1 \rightarrow T$ | 1 | Test result |
| TST | \#imm, R0 | 11001000iiiiiiii | R0 \& imm; if the result is $0,1 \rightarrow T$ | 1 | Test result |
| TST.B | \#imm, @ (R0, GBR) | 11001100iiiiiiii | ( RO 0 + GBR) \& imm; if the result is $0,1 \rightarrow T$ | 3 | Test result |
| XOR | $\mathrm{Rm}, \mathrm{Rn}$ | 0010 nnnnmmmm1010 | $\mathrm{Rn}{ }^{\wedge} \mathrm{Rm} \rightarrow \mathrm{Rn}$ | 1 | - |
| XOR | \#imm, R0 | 11001010iiiiiiii | $\mathrm{RO}{ }^{\wedge} \mathrm{imm} \rightarrow \mathrm{R} 0$ | 1 | - |
| XOR.B | \#imm, @ (R0, GBR) | 11001110iiiiiiii | $\begin{aligned} & (\mathrm{RO}+\mathrm{GBR})^{\wedge} \mathrm{imm} \rightarrow \\ & (\mathrm{RO}+\mathrm{GBR}) \end{aligned}$ | 3 | - |

Note: * The on-chip DMAC/DTC bus cycles are not inserted between the read and write cycles of TAS instruction execution. However, bus release due to BREQ is carried out.

Table 2.15 Shift Instructions

| Instruction |  | Instruction Code | Operation | Execu- <br> tion <br> Cycles | T Bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ROTL | Rn | 0100nnnn00000100 | $\mathrm{T} \leftarrow \mathrm{Rn} \leftarrow \mathrm{MSB}$ | 1 | MSB |
| ROTR | Rn | 0100nnnn00000101 | LSB $\rightarrow \mathrm{Rn} \rightarrow \mathrm{T}$ | 1 | LSB |
| ROTCL | Rn | 0100nnnn00100100 | $\mathrm{T} \leftarrow \mathrm{Rn} \leftarrow \mathrm{T}$ | 1 | MSB |
| ROTCR | Rn | $0100 \mathrm{nnnn00100101}$ | $\mathrm{T} \rightarrow \mathrm{Rn} \rightarrow \mathrm{T}$ | 1 | LSB |
| SHAL | Rn | 0100nnnn00100000 | $\mathrm{T} \leftarrow \mathrm{Rn} \leftarrow 0$ | 1 | MSB |
| SHAR | Rn | 0100nnnn00100001 | $\mathrm{MSB} \rightarrow \mathrm{Rn} \rightarrow \mathrm{T}$ | 1 | LSB |
| SHLL | Rn | 0100nnnn00000000 | $\mathrm{T} \leftarrow \mathrm{Rn} \leftarrow 0$ | 1 | MSB |
| SHLR | Rn | 0100nnnn00000001 | $0 \rightarrow \mathrm{Rn} \rightarrow \mathrm{T}$ | 1 | LSB |
| SHLL2 | Rn | 0100nnnn00001000 | $\mathrm{Rn} \ll 2 \rightarrow \mathrm{Rn}$ | 1 | - |
| SHLR2 | Rn | 0100nnnn00001001 | $R n \gg 2 \rightarrow R n$ | 1 | - |
| SHLL8 | Rn | 0100nnnn00011000 | $\mathrm{Rn} \ll 8 \rightarrow \mathrm{Rn}$ | 1 | - |
| SHLR8 | Rn | 0100nnnn00011001 | $R n \gg 8 \rightarrow R n$ | 1 | - |
| SHLL16 | Rn | 0100nnnn00101000 | $\mathrm{Rn} \ll 16 \rightarrow \mathrm{Rn}$ | 1 | - |
| SHLR16 | Rn | 0100nnnn00101001 | $R n \gg 16 \rightarrow R n$ | 1 | - |

Table 2.16 Branch Instructions

| Instruction |  | Instruction Code <br> 10001011dddddddd | Operation <br> If $\mathrm{T}=0, \operatorname{disp} \times 2+\mathrm{PC} \rightarrow \mathrm{PC}$; if $\mathrm{T}=$ 1, nop | Exec. <br> Cycles <br> 3/1* | T <br> Bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| BF | label |  |  |  |  |
| BF/S | label | 10001111 dddddddd | Delayed branch, if T = $0, \operatorname{disp} \times 2+$ $P C \rightarrow P C$; if $T=1$, nop | 2/1* | - |
| BT | label | 10001001 dddddddd | If $\mathrm{T}=1, \operatorname{disp} \times 2+\mathrm{PC} \rightarrow \mathrm{PC}$; if $\mathrm{T}=$ 0 , nop | 3/1* | - |
| BT/S | label | 10001101 dddddddd | Delayed branch, if T = 1, disp $\times 2+$ $P C \rightarrow P C$; if $T=0$, nop | 2/1* | - |
| BRA | label | 1010ddddddddddddd | Delayed branch, disp $\times 2+\mathrm{PC} \rightarrow$ PC | 2 | - |
| BRAF | Rm | 0000mmmm00100011 | Delayed branch, Rm + PC $\rightarrow$ PC | 2 | - |
| BSR | label | 1011dddddddddddd | Delayed branch, PC $\rightarrow \mathrm{PR}$, disp $\times 2$ $+\mathrm{PC} \rightarrow \mathrm{PC}$ | 2 | - |
| BSRF | Rm | 0000mmmm00000011 | Delayed branch, PC $\rightarrow P R$, $\mathrm{Rm}+\mathrm{PC} \rightarrow \mathrm{PC}$ | 2 | - |
| JMP | @Rm | 0100 mmmm 00101011 | Delayed branch, Rm $\rightarrow$ PC | 2 | - |
| JSR | @Rm | 0100mmmm00001011 | Delayed branch, PC $\rightarrow \mathrm{PR}$, $R m \rightarrow P C$ | 2 | - |
| RTS |  | 0000000000001011 | Delayed branch, PR $\rightarrow$ PC | 2 | - |

Note: * One state when it does not branch.

Table 2.17 System Control Instructions

| Instruction |  | Instruction Code | Operation | Exec. Cycles | T Bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CLRT |  | 0000000000001000 | $0 \rightarrow$ T | 1 | 0 |
| CLRMAC |  | 0000000000101000 | $0 \rightarrow \mathrm{MACH}, \mathrm{MACL}$ | 1 | - |
| LDC | Rm, SR | $0100 \mathrm{mmmm0} 0001110$ | $\mathrm{Rm} \rightarrow \mathrm{SR}$ | 1 | LSB |
| LDC | Rm, GBR | $0100 \mathrm{mmmm0} 0011110$ | $\mathrm{Rm} \rightarrow \mathrm{GBR}$ | 1 | - |
| LDC | Rm, VBR | $0100 \mathrm{mmmm0} 0101110$ | $\mathrm{Rm} \rightarrow$ VBR | 1 | - |
| LDC.L | @Rm+, SR | $0100 \mathrm{mmmm00000111}$ | $(\mathrm{Rm}) \rightarrow \mathrm{SR}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 3 | LSB |
| LDC.L | @Rm+, GBR | $0100 \mathrm{mmmm0} 0010111$ | $(\mathrm{Rm}) \rightarrow \mathrm{GBR}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 3 | - |
| LDC.L | @Rm+, VBR | $0100 \mathrm{mmmm00100111}$ | $(\mathrm{Rm}) \rightarrow \mathrm{VBR}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 3 | - |
| LDS | Rm, MACH | $0100 \mathrm{mmmm00001010}$ | $\mathrm{Rm} \rightarrow \mathrm{MACH}$ | 1 | - |
| LDS | Rm, MACL | $0100 \mathrm{mmmm00011010}$ | $\mathrm{Rm} \rightarrow \mathrm{MACL}$ | 1 | - |
| LDS | Rm, PR | $0100 \mathrm{mmmm0} 0101010$ | $\mathrm{Rm} \rightarrow \mathrm{PR}$ | 1 | - |
| LDS.L | @Rm+, MACH | $0100 \mathrm{mmmm00000110}$ | $(\mathrm{Rm}) \rightarrow \mathrm{MACH}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 1 | - |
| LDS.L | @Rm+, MACL | $0100 \mathrm{mmmm00010110}$ | $(\mathrm{Rm}) \rightarrow \mathrm{MACL}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 1 | - |
| LDS.L | @Rm+, PR | $0100 \mathrm{mmmm00100110}$ | $(\mathrm{Rm}) \rightarrow \mathrm{PR}, \mathrm{Rm}+4 \rightarrow \mathrm{Rm}$ | 1 | - |
| NOP |  | 0000000000001001 | No operation | 1 | - |
| RTE |  | 0000000000101011 | Delayed branch, stack area $\rightarrow \mathrm{PC} / \mathrm{SR}$ | 4 | - |
| SETT |  | 0000000000011000 | $1 \rightarrow \mathrm{~T}$ | 1 | 1 |
| SLEEP |  | 0000000000011011 | Sleep | 3* | - |
| STC | SR, Rn | 0000nnnn00000010 | $\mathrm{SR} \rightarrow \mathrm{Rn}$ | 1 | - |
| STC | GBR, Rn | 0000nnnn00010010 | $\mathrm{GBR} \rightarrow \mathrm{Rn}$ | 1 | - |
| STC | $\mathrm{VBR}, \mathrm{Rn}$ | 0000nnnn00100010 | $\mathrm{VBR} \rightarrow \mathrm{Rn}$ | 1 | - |
| STC.L | SR, @-Rn | 0100nnnn00000011 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{SR} \rightarrow(\mathrm{Rn})$ | 2 | - |
| STC.L | GBR, @-Rn | 0100nnnn00010011 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{GBR} \rightarrow(\mathrm{Rn})$ | 2 | - |
| STC.L | VBR, @-Rn | 0100nnnn00100011 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{BR} \rightarrow(\mathrm{Rn})$ | 2 | - |
| STS | MACH, Rn | 0000nnnn00001010 | $\mathrm{MACH} \rightarrow \mathrm{Rn}$ | 1 | - |
| STS | MACL, Rn | $0000 n n n n 00011010$ | $\mathrm{MACL} \rightarrow \mathrm{Rn}$ | 1 | - |
| STS | PR, Rn | 0000 nnnn 00101010 | $\mathrm{PR} \rightarrow \mathrm{Rn}$ | 1 | - |

Table 2.17 System Control Instructions (cont)

| Instruction | Instruction Code | Operation | Exec. Cycles | T Bit |
| :---: | :---: | :---: | :---: | :---: |
| STS.L MACH, @-Rn | 0100nnnn00000010 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{MACH} \rightarrow(\mathrm{Rn})$ | 1 | - |
| STS.L MACL, @-Rn | 0100nnnn00010010 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{MACL} \rightarrow(\mathrm{Rn})$ | 1 | - |
| STS.L PR, @-Rn | 0100nnnn00100010 | $\mathrm{Rn}-4 \rightarrow \mathrm{Rn}, \mathrm{PR} \rightarrow(\mathrm{Rn})$ | 1 | - |
| TRAPA \#imm | $11000011 i i i i i i i i$ | $\begin{aligned} & \mathrm{PC} / \mathrm{SR} \rightarrow \text { stack area, } \\ & (\mathrm{imm} \times 4+\mathrm{VBR}) \rightarrow \mathrm{PC} \end{aligned}$ | 8 | - |

Note: * The number of execution cycles before the chip enters sleep mode: The execution cycles shown in the table are minimums. The actual number of cycles may be increased when (1) contention occurs between instruction fetches and data access, or (2) when the destination register of the load instruction (memory $\rightarrow$ register) and the register used by the next instruction are the same.

### 2.5 Processing States

### 2.5.1 State Transitions

The CPU has five processing states: reset, exception processing, bus release, program execution and power-down. Figure 2.6 shows the transitions between the states.


Figure 2.6 Transitions between Processing States
Reset State: The CPU resets in the reset state. When the $\overline{\mathrm{RES}}$ pin level goes low, a power-on reset results. When the $\overline{\mathrm{RES}}$ pin is high and MRES is low, a manual reset will occur.

Exception Processing State: The exception processing state is a transient state that occurs when exception processing sources such as resets or interrupts alter the CPU's processing state flow.

For a reset, the initial values of the program counter (PC) (execution start address) and stack pointer (SP) are fetched from the exception processing vector table and stored; the CPU then branches to the execution start address and execution of the program begins.

For an interrupt, the stack pointer (SP) is accessed and the program counter (PC) and status register (SR) are saved to the stack area. The exception service routine start address is fetched from the exception processing vector table; the CPU then branches to that address and the program starts executing, thereby entering the program execution state.

Program Execution State: In the program execution state, the CPU sequentially executes the program.

Power-Down State: In the power-down state, the CPU operation halts and power consumption declines. The SLEEP instruction places the CPU in the power-down state. This state has two modes: sleep mode and standby mode.

Bus Release State: In the bus release state, the CPU releases access rights to the bus to the device that has requested them.

### 2.5.2 Power-Down State

Besides the ordinary program execution states, the CPU also has a power-down state in which CPU operation halts, lowering power consumption. There are two power-down state modes: sleep mode and standby mode.

Sleep Mode: When standby bit SBY (in the standby control register SBYCR) is cleared to 0 and a SLEEP instruction executed, the CPU moves from program execution state to sleep mode. In the sleep mode, the CPU halts and the contents of its internal registers and the data in on-chip cache (or on-chip RAM) is maintained. The on-chip peripheral modules other than the CPU do not halt in the sleep mode.

To return from sleep mode, use a reset (power-on or manual), any interrupt, or a DMA address error; the CPU returns to the ordinary program execution state through the exception processing state.

Standby Mode: To enter the standby mode, set the standby bit SBY (in the standby control register SBYCR) to 1 and execute a SLEEP instruction. In standby mode, all CPU, on-chip peripheral module, and oscillator functions are halted. However, when entering standby mode, the DMA master enable bit of the DMAC should be set to 0 . If multiplication-related instructions are being executed at the time of entry into standby mode, the values of MACH and MACL will become undefined.

To return from standby mode, use a reset (power-on or manual) or an NMI interrupt. For resets, the CPU returns to ordinary program execution state through the exception processing state when placed in a reset state for the duration of the oscillator stabilization time. For NMI interrupts, the 74

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CPU returns to ordinary program execution state through the exception processing state after the oscillator stabilization time has elapsed. In this mode, power consumption drops markedly, since the oscillator stops (table 2.18).

Table 2.18 Power-Down State

|  |  | State |  |  |  |  |  | Canceling |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mode | Transition Conditions | Clock | CPU | On-Chip Peripheral Modules | CPU <br> Registers | On-Chip Cache or On-Chip RAM | I/O Port Pins |  |
| Sleep | Execute SLEEP instruction with SBY bit cleared to 0 in SBYCR | Run | Halt | Run | Held | Held | Held | - Interrupt <br> - DMA address error <br> - Power-on reset <br> - Manual reset |
| Standby | Execute SLEEP instruction with SBY bit set to 1 in SBYCR | Halt | Halt | Halt and initialize* | Held | Held | Held or Hi-Z (selectable) | - NMI interrupt <br> - Power-on reset <br> - Manual reset |

Note: * Differs depending on the peripheral module and pin.

## Section 3 Operating Modes

### 3.1 Operating Modes, Types, and Selection

This LSI has five operating modes and three clock modes, determined by the setting of the mode pins (MD3-MD0). Do not change the mode pin settings during LSI operation (while power is on). (In the F-ZTAT version, however, MD1 can be changed in the power-on reset state.)

Table 3.1 indicates the setting method for the operating mode.
Table 3.1 Operating Mode Setting

| Mode <br> No. | Pin Setting |  |  |  |  | Mode <br> Name | On-Chip ROM | CSO Area |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FWP | MD3*1 | MD2*1 | MD1 | MD0 |  |  | 112 Pin | 144 Pin |
| 0 | 1 | x | x | 0 | 0 | MCU mode 0 | Not Active | 8-bit space | $\begin{aligned} & \text { 16-bit } \\ & \text { space } \end{aligned}$ |
| 1 | 1 | x | x | 0 | 1 | MCU mode 1 | Not Active | 16-bit space | $\begin{aligned} & 32 \text {-bit } \\ & \text { space } \end{aligned}$ |
| 2 | 1 | x | x | 1 | 0 | MCU mode 2 | Active | 8/16-bit space*2 | $\begin{aligned} & \text { 8/16/32-bit } \\ & \text { space*2 } \end{aligned}$ |
| 3 | 1 | x | x | 1 | 1 | Single chip mode | Active | - | - |
| 4 | 1 | 1 | 1 | 1 | 1 | PROM mode*3 | Active | - | - |
| - | 0 | x | x | 0 | 0 | Boot mode*4 | Active | 8/16-bit space*2 | 8/16/32-bit space*2 |
| - | 0 | $x$ | x | 0 | 1 |  |  | - | - |
| - | 0 | x | x | 1 | 0 | User programming mode*4 | Active | 8/16-bit space*2 | $\begin{aligned} & \text { 8/16/32-bit } \\ & \text { space*2 }^{2} \end{aligned}$ |
| - | 0 | x | x | 1 | 1 |  |  | - | - |
| - | 1 | 1 | 1 | 0 | 1 | Flash programmer mode*4 | Active | - | - |

Notes: *1 MD2 and MD3 pins select the clock mode in modes 0-3 (table 3.2).
*2 Set by BCR2 of BSC.
*3 Only ZTAT.
*4 Only F-ZTAT.

Table 3.2 indicates the setting method for the clock mode.
Table 3.2 Clock Mode Setting

| MD3 | MD2 | Clock Mode |
| :--- | :--- | :--- |
| 0 | 0 | PLL ON $\times 1$ |
| 0 | 1 | PLL ON $\times 2$ |
| 1 | 0 | PLL ON $\times 4$ |
| 1 | 1 | Reserved (PROM mode only) |

### 3.2 Explanation of Operating Modes

Table 3.3 describes the operating modes.
Table 3.3 Operating Modes

| Mode | Description |
| :--- | :--- |
| (MCU) Mode 0 | CS0 area becomes an external memory space with 8-bit bus width for <br> the 112-pin version, and 16-bit for the 144-pin version. |
| (MCU) Mode 1 | CS0 area becomes an external memory space with 16-bit bus width for <br> the 112-pin version, and 32-bit for the 144-pin version |
| (MCU) Mode 2 | The on-chip ROM becomes effective. The bus width for the on-chip ROM <br> space is 32 bit. |
| Mode 3 (single chip <br> mode) | Any port can be used, but external addresses can not be employed. |
| Mode 4 (PROM mode) | On-chip ROM can be programmed using a general PROM writer. |
| Clock mode | The input waveform frequency can be used as is, doubled or quadrupled <br> as an internal clock in modes 0 to 3. |

### 3.3 Pin Configuration

Table 3.4 describes the function of each operating mode related pin.
Table 3.4 Operating Mode Pin Function

| Pin <br> Name | Input/Output | Function |
| :--- | :--- | :--- |
| XTAL | Input | Connects to a crystal oscillator |
| EXTAL | Input | Connects to a crystal oscillator, or used for external clock input pin |
| PLLCAP | Input | Connects to a capacitor for PLL circuit operation |
| MD0 | Input | Designates operating mode through the level applied to this pin |
| MD1 | Input | Designates operating mode through the level applied to this pin |
| MD2 | Input | Designates clock mode through the level applied to this pin |
| MD3 | Input | Designates clock mode through the level applied to this pin |

## Section 4 Clock Pulse Generator (CPG)

### 4.1 Overview

The SH7040 Series has an on-chip clock pulse generator (CPG) that generates the system clock $(\phi)$, as well as the internal clock ( $\phi / 2$ to $\phi / 8192$ ). The CPG consists of an oscillator, a PLL, and a prescaler.

### 4.1.1 Block Diagram

A block diagram of the clock pulse generator is shown in figure 4.1.


Figure 4.1 Block Diagram of the Clock Pulse Generator

### 4.2 Oscillator

Clock pulses can be supplied from a connected crystal resonator or an external clock.

### 4.2.1 Connecting a Crystal Oscillator

Circuit Configuration: A crystal oscillator can be connected as shown in figure 4.2. Use the damping resistance (Rd) listed in table 4.1. Use a $4-10 \mathrm{MHz}$ crystal oscillator (consult your dealer concerning the compatibility of the crystal oscillator and the LSI).
$\square$
Figure 4.2 Connection of the Crystal Oscillator (Example)
Table 4.1 Damping Resistance Values (Recommended Values)

|  | Frequency (MHz) |  |  |
| :--- | :--- | :--- | :--- |
| Parameter | $\mathbf{4}$ | $\mathbf{8}$ | $\mathbf{1 0}$ |
| $\operatorname{Rd}(\Omega)$ | 500 | 200 | 0 |

Crystal Oscillator: Figure 4.3 shows an equivalent circuit of the crystal oscillator. Use a crystal oscillator with the characteristics listed in table 4.2.


Figure 4.3 Crystal Oscillator Equivalent Circuit
Table 4.2 Crystal Oscillator Parameters

|  | Frequency (MHz) |  |  |
| :--- | :--- | :--- | :--- |
| Parameter | $\mathbf{4}$ | $\mathbf{8}$ | $\mathbf{1 0}$ |
| Rs $\max (\Omega)$ | 120 | 80 | 60 |
| Co $\max (\mathrm{pF})$ | 7 | 7 | 7 |

### 4.2.2 External Clock Input Method

Figure 4.4 shows an example of an external clock input connection. In this case, make the external clock high level to stop it when in standby mode. During operation, make the external input clock frequency $4-10 \mathrm{MHz}$.

When leaving the XTAL pin open, make sure the parasitic capacitance is less than 10 pF .
Even when inputting an external clock, be sure to delay until after the oscillation stabilization time (upon power-on) or after release from standby, in order to ensure the PLL stabilization time.


Figure 4.4 Example of External Clock Connection

### 4.3 Prescaler

The prescaler divides the system clock $(\phi)$ to generate an internal clock ( $\phi / 2$ to $\phi / 8192$ ) for supply to peripheral modules.

### 4.4 Oscillator Halt Function

This CPG can detect a clock halt and automatically cause the timer pins to become highimpedance when any system abnormality causes the oscillator to halt. That is, when a change of EXTAL has not been detected, the high-current six pins (PE9/TIOC3B, PE11/TIOC3D, PE12/TIOC4A, PE13/TIOC4B/MRES, PE14/TIOC4C/DACK0/AH, PE15/TIOC4D/DACK1/ $\overline{\overline{\text { IRQOUT }}) \text { are set to high-impedance regardless of PFC setting. }}$

Even in standby mode, these six pins become high-impedance regardless of PFC setting. These pins enter the normal state after standby mode is cancelled. When abnormalities that halt the oscillator occur except in standby mode, other LSI operations become undefined. In this case, LSI operations, including these six pins, become undefined even when the oscillator operation starts again.

### 4.5 Usage Notes

### 4.5.1 Oscillator Usage Notes

Since the characteristics of the oscillator are closely related to the user-defined board settings, the user should refer to the connection examples in this section and perform a careful evaluation. The oscillator circuit ratings will differ depending on factors such as the oscillator used and the stray capacitance of the mounted circuitry. Therefore, the oscillator manufacturer should be consulted before a decision is made. Make sure that the voltage applied to the oscillator does not exceed the maximum rating.

### 4.5.2 Notes on Board Design

When connecting a crystal oscillator, observe the following precautions:

- To prevent induction from interfering with correct oscillation, do not route any signal lines near the oscillator circuitry.
- When designing the board, place the crystal oscillator and its load capacitors as close as possible to the XTAL and EXTAL pins.

Figure 4.5 shows the precautions regarding oscillator block board settings.


Figure 4.5 Cautions for Oscillator Circuit System Board Design

External circuitry such as that shown in figure 4.6 is recommended around the PLL.


Figure 4.6 Cautions for Use of PLL Oscillator Circuit
Place oscillation stabilization capacitor C1 and resistor R1 near the PLLCAP pin, and ensure that these lines do not cross any other signal lines. Supply the C 1 ground from $\mathrm{PLLV}_{\mathrm{SS}}$.

Also, separate $\operatorname{PLLV}_{\mathrm{CC}}$ and $\operatorname{PLLV}_{\mathrm{SS}}$, and the other $\mathrm{V}_{\mathrm{CC}}$ and $\mathrm{V}_{\mathrm{SS}}$ pins, from the board power supply source, and be sure to insert bypass capacitors CPB and CB close to the pins.

If $\mathrm{V}_{\mathrm{CC}}$ and $\mathrm{PLLV}_{\mathrm{CC}}$ are both $3.3 \mathrm{~V} \pm 0.3 \mathrm{~V}$, it is recommended that Rp be set to $0 \Omega$.

### 4.5.3 Spread Spectrum Clock Generator Usage Notes

The following points should be borne in mind when using a spread spectrum clock generator as an external oscillator in order to reduce radiation noise.

- Set the center frequency and the spread amplitude such that the internal clock does not exceed the maximum frequency during spread spectrum operation.
- Using a spread spectrum clock generator may trigger the oscillator halt function described in section 4.4. If the system configuration is such that this function will cause problems, a spread spectrum clock generator should not be used.


## Section 5 Exception Processing

### 5.1 Overview

### 5.1.1 Types of Exception Processing and Priority

Exception processing is started by four sources: resets, address errors, interrupts and instructions and have the priority shown in table 5.1. When several exception processing sources occur at once, they are processed according to the priority shown.

Table 5.1 Types of Exception Processing and Priority Order

| Exception | Source |  | Priority |
| :---: | :---: | :---: | :---: |
| Reset | Power-on reset |  | High |
|  | Manual reset |  |  |
| Address error | CPU address error |  |  |
|  | DMAC/DTC address error |  |  |
| Interrupt | NMI |  |  |
|  | User break |  |  |
|  | IRQ |  |  |
|  | On-chip peripheral modules: | - Direct memory access controller (DMAC) <br> - Multifunction timer/pulse unit (MTU) <br> - Serial communications interface (SCI) <br> - $A / D$ converter (A/D)*3 <br> - Data transfer controller (DTC) <br> - Compare match timer (CMT) <br> - Watchdog timer (WDT) <br> - Bus state controller (BSC) <br> - Port output enable control section |  |


| Instructions | Trap instruction (TRAPA instruction) |
| :---: | :---: |
|  | General illegal instructions (undefined code) |
|  | Illegal slot instructions (undefined code placed directly after a delay branch Low instruction*1 or instructions that rewrite the PC*2) |
| Notes: $\begin{aligned} & * 1 \\ & * 2 \\ & * 3\end{aligned}$ | elayed branch instructions: JMP, JSR, BRA, BSR, RTS, RTE, BF/S, BT/S, BSRF, RAF. |
|  | tructions that rewrite the PC: JMP, JSR, BRA, BSR, RTS, RTE, BT, BF, TRAPA, /S, BT/S, BSRF, BRAF. |
|  | mask products: A/D0, A/D1. |

### 5.1.2 Exception Processing Operations

The exception processing sources are detected and begin processing according to the timing shown in table 5.2.

Table 5.2 Timing of Exception Source Detection and the Start of Exception Processing

| Exception | Source | Timing of Source Detection and Start of Processing |
| :--- | :--- | :--- |
| Reset | Power-on reset | Starts when the $\overline{\text { RES pin changes from low to high. }}$ |
|  | Manual reset | Starts when the $\overline{\text { RES }}$ pin is high and the $\overline{\text { MRES pin changes }}$ <br> from low to high. |
| Address error | Detected when instruction is decoded and starts when the <br> previous executing instruction finishes executing. |  |
| Interrupts | Detected when instruction is decoded and starts when the <br> previous executing instruction finishes executing. |  |
| Instructions | Trap instruction Starts from the execution of a TRAPA instruction. <br>  General illegal <br> instructions <br>  Starts from the decoding of undefined code anytime except after <br> a delayed branch instruction (delay slot). <br> instructions Starts from the decoding of undefined code placed in a delayed <br> branch instruction (delay slot) or of instructions that rewrite the <br> PC. |  |

When exception processing starts, the CPU operates as follows:

1. Exception processing triggered by reset:

The initial values of the program counter (PC) and stack pointer (SP) are fetched from the exception processing vector table ( PC and SP are respectively the $\mathrm{H}^{\prime} 00000000$ and $H^{\prime} 00000004$ addresses for power-on resets and the H'00000008 and H'00000000C addresses for manual resets). See section 5.1.3, Exception Processing Vector Table, for more information. 0 is then written to the vector base register (VBR) and 1111 is written to the interrupt mask bits (I3-I0) of the status register (SR). The program begins running from the PC address fetched from the exception processing vector table.
2. Exception processing triggered by address errors, interrupts and instructions:

SR and PC are saved to the stack indicated by R15. For interrupt exception processing, the interrupt priority level is written to the SR's interrupt mask bits (I3-I0). For address error and instruction exception processing, the I3-I0 bits are not affected. The start address is then fetched from the exception processing vector table and the program begins running from that address.

### 5.1.3 Exception Processing Vector Table

Before exception processing begins running, the exception processing vector table must be set in memory. The exception processing vector table stores the start addresses of exception service routines. (The reset exception processing table holds the initial values of PC and SP.)

All exception sources are given different vector numbers and vector table address offsets, from which the vector table addresses are calculated. During exception processing, the start addresses of the exception service routines are fetched from the exception processing vector table, which indicated by this vector table address.

Table 5.3 shows the vector numbers and vector table address offsets. Table 5.4 shows how vector table addresses are calculated.

Table 5.3 Exception Processing Vector Table

| Exception Sources | Vector Numbers | Vector Table Address Offset |
| :---: | :---: | :---: |
| Power-on reset PC | 0 | H'00000000-H'00000003 |
| SP | 1 | H'00000004-H'00000007 |
| Manual reset PC | 2 | H'00000008-H'0000000B |
| SP | 3 | H'0000000C-H'0000000F |
| General illegal instruction | 4 | H'00000010-H'00000013 |
| (Reserved by system) | 5 | H'00000014-H'00000017 |
| Slot illegal instruction | 6 | H'00000018-H'0000001B |
| (Reserved by system) | 7 | H'0000001C-H'0000001F |
| (Reserved by system) | 8 | H'00000020-H'00000023 |
| CPU address error | 9 | H'00000024-H'00000027 |
| DMAC/DTC address error | 10 | H'00000028-H'0000002B |
| Interrupts | 11 | H'0000002C-H'0000002F |
|  | 12 | H'00000030-H'00000033 |
| (Reserved by system) | 13 | H'00000034-H'00000037 |
|  | : | : |
|  | 31 | H'0000007C-H'0000007F |
| Trap instruction (user vector) | 32 | H'00000080-H'00000083 |
|  | : | : |
|  | 63 | H'000000FC-H'000000FF |

## Renesns

Table 5.3 Exception Processing Vector Table (cont)

| Exception Sources |  | Vector Numbers | Vector Table Address Offset |
| :---: | :---: | :---: | :---: |
| Interrupts | IRQ0 | 64 | H'00000100-H'00000103 |
|  | IRQ1 | 65 | H'00000104-H'00000107 |
|  | IRQ2 | 66 | H'00000108-H'0000010B |
|  | IRQ3 | 67 | H'0000010C-H'0000010F |
|  | IRQ4 | 68 | H'00000110-H'00000113 |
|  | IRQ5 | 69 | H'00000114-H'00000117 |
|  | IRQ6 | 70 | H'00000118-H'0000011B |
|  | IRQ7 | 71 | H'0000011C-H'0000011F |
| On-chip peripheral module* |  | 72 | H'00000120-H'00000124 |
|  |  | 255 | H'000003FC-H'000003FF |

Note: * The vector numbers and vector table address offsets for each on-chip peripheral module interrupt are given in section 6, Interrupt Controller (INTC), and table 6.3, Interrupt Exception Processing Vectors and Priorities.

Table 5.4 Calculating Exception Processing Vector Table Addresses

| Exception Source | Vector Table Address Calculation |
| :---: | :---: |
| Resets | $\begin{aligned} \text { Vector table address } & =(\text { vector table address offset }) \\ & =(\text { vector number }) \times 4 \end{aligned}$ |
| Address errors, interrupts, instructions | $\begin{aligned} \hline \text { Vector table address } & =\text { VBR }+(\text { vector table address offset }) \\ & =\text { VBR }+(\text { vector number }) \times 4 \end{aligned}$ |

Notes: 1. VBR: Vector base register
2. Vector table address offset: See table 5.3.
3. Vector number: See table 5.3.

### 5.2 Resets

Resets have the highest priority of any exception source. There are two types of resets: manual resets and power-on resets. As table 5.5 shows, both types of resets initialize the internal status of the CPU. In power-on resets, all registers of the on-chip peripheral modules are initialized; in manual resets, they are not.

Table 5.5 Types of Resets

|  | Conditions for Transition <br> to Reset Status |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Type | $\overline{\text { RES }}$ | $\overline{\text { MRES }}$ |  | CPU | On-Chip Peripheral Module |
| Power-on reset | Low | - | Initialized | Initialized |  |
| Manual reset | High | Low | Initialized | Not initialized |  |

### 5.2.1 Power-On Reset

When the $\overline{\operatorname{RES}}$ pin is driven low, the LSI does a power-on reset. To reliably reset the LSI, the $\overline{\operatorname{RES}}$ pin should be kept at low for at least the duration of the oscillation settling time when applying power or when in standby mode (when the clock circuit is halted) or at least $20 t_{\text {cyc }}$ (when the clock circuit is running). During power-on reset, CPU internal status and all registers of on-chip peripheral modules are initialized. See Appendix C, Pin States, for the status of individual pins during the power-on reset status.

In the power-on reset status, power-on reset exception processing starts when the $\overline{\operatorname{RES}}$ pin is first driven low for a set period of time and then returned to high. The CPU will then operate as follows:

1. The initial value (execution start address) of the program counter (PC) is fetched from the exception processing vector table.
2. The initial value of the stack pointer (SP) is fetched from the exception processing vector table.
3. The vector base register (VBR) is cleared to $\mathrm{H}^{\prime} 00000000$ and the interrupt mask bits (I3-I0) of the status register (SR) are set to H'F (1111).
4. The values fetched from the exception processing vector table are set in the program counter (PC) and SP and the program begins executing.

Be certain to always perform power-on reset processing when turning the system power on.

### 5.2.2 Manual Reset

When the $\overline{\operatorname{RES}}$ pin is high and the $\overline{\text { MRES }}$ pin is driven low, the LSI does a manual reset. To reliably reset the LSI, the $\overline{\text { MRES }}$ pin should be kept at low for at least the duration of the oscillation settling time when in standby mode (when the clock is halted) or at least $20 \mathrm{t}_{\mathrm{cyc}}$ when the clock is operating. During manual reset, the CPU internal status is initialized. Registers of onchip peripheral modules are not initialized. Since the BSC is not affected, the DRAM refresh control functions remain operational even when the manual reset status continues for a long period of time. When the LSI enters manual reset status in the middle of a bus cycle, manual reset exception processing does not start until the bus cycle has ended. Thus, manual resets do not abort bus cycles. However, the bus cycle ends once $\overline{\text { MRES }}$ is driven low. Hold at low level until manual
reset mode. (Keep at low level for at least the longest bus cycle.) See Appendix C, Pin States, for the status of individual pins during manual reset mode.

In the manual reset status, manual reset exception processing starts when the $\overline{\text { MRES }}$ pin is first kept low for a set period of time and then returned to high. The CPU will then operate the same as described for power-on resets.

### 5.3 Address Errors

Address errors occur when instructions are fetched or data read or written, as shown in table 5.6.
Table 5.6 Bus Cycles and Address Errors

| Bus Cycle |  | Bus Cycle Description |  |
| :---: | :---: | :---: | :---: |
| Type | Bus <br> Master |  | Address Errors |
| Instruction fetch | CPU | Instruction fetched from even address | None (normal) |
|  |  | Instruction fetched from odd address | Address error occurs |
|  |  | Instruction fetched from other than on-chip peripheral module space* | None (normal) |
|  |  | Instruction fetched from on-chip peripheral module space* | Address error occurs |
|  |  | Instruction fetched from external memory space when in single chip mode | Address error occurs |
| Data read/write | CPU or <br> DMAC <br> or DTC | Word data accessed from even address | None (normal) |
|  |  | Word data accessed from odd address | Address error occurs |
|  |  | Longword data accessed from a longword boundary | None (normal) |
|  |  | Longword data accessed from other than a longword boundary | Address error occurs |
|  |  | Byte or word data accessed in on-chip peripheral module space* | None (normal) |
|  |  | Longword data accessed in 16-bit on-chip peripheral module space* | None (normal) |
|  |  | Longword data accessed in 8-bit on-chip peripheral module space* | Address error occurs |
|  |  | External memory space accessed when in single chip mode | Address error occurs |

Note: * See section 10, Bus State Controller (BSC).

### 5.3.1 Address Error Exception Processing

When an address error occurs, the bus cycle in which the address error occurred ends. When the executing instruction then finishes, address error exception processing starts up. The CPU operates as follows:

1. The status register (SR) is saved to the stack.
2. The program counter ( PC ) is saved to the stack. The PC value saved is the start address of the instruction to be executed after the last executed instruction.
3. The exception service routine start address is fetched from the exception processing vector table that corresponds to the address error that occurred and the program starts executing from that address. The jump that occurs is not a delayed branch.

### 5.4 Interrupts

Table 5.7 shows the sources that start up interrupt exception processing. These are divided into NMI, user breaks, IRQ, and on-chip peripheral modules.

Table 5.7 Interrupt Sources

| Type | Request Source | Number of <br> Sources |
| :--- | :--- | :--- |
| NMI | NMI pin (external input) | 1 |
| User break | User break controller | 1 |
| IRQ | $\overline{\text { IRQ0-IRQ7 (external input) }}$ | 8 |
| On-chip peripheral module | Direct memory access controller (DMAC) | 4 |
|  | Multifunction timer/pulse unit (MTU) | 24 |
|  | Serial communications interface (SCI) | 8 |
|  | A/D converter | $1^{*}$ |
|  | Data transfer controller (DTC) | 1 |
|  | Compare match timer (CMT) | 2 |
|  | Watchdog timer (WDT) | 1 |
|  | Bus state controller (BSC) | 1 |
|  | Port | 1 |

Note: * For A mask products, (A/D0, A/D1) is 2

Each interrupt source is allocated a different vector number and vector table offset. See section 6, Interrupt Controller (INTC), and table 6.3, Interrupt Exception Processing Vectors and Priorities, for more information on vector numbers and vector table address offsets.

### 5.4.1 Interrupt Priority Level

The interrupt priority order is predetermined. When multiple interrupts occur simultaneously (overlap), the interrupt controller (INTC) determines their relative priorities and starts up processing according to the results.

The priority order of interrupts is expressed as priority levels $0-16$, with priority 0 the lowest and priority 16 the highest. The NMI interrupt has priority 16 and cannot be masked, so it is always accepted. The user break interrupt priority level is 15 . IRQ interrupts and on-chip peripheral module interrupt priority levels can be set freely using the INTC's interrupt priority level setting registers A through H (IPRA-IPRH) as shown in table 5.8. The priority levels that can be set are 0-15. Level 16 cannot be set. See section 6.3.1, Interrupt Priority Registers A-H (IPRA-IPRH), for more information on IPRA to IPRH.

Table 5.8 Interrupt Priority Order

| Type | Priority Level | Comment |
| :--- | :--- | :--- |
| NMI | 16 | Fixed priority level. Cannot be masked. |
| User break | 15 | Fixed priority level. |
| IRQ | $0-15$ | Set with interrupt priority level setting registers A <br> through H (IPRA-IPRH). |
| On-chip peripheral module | $0-15$ | Set with interrupt priority level setting registers A <br> through H (IPRA-IPRH). |

### 5.4.2 Interrupt Exception Processing

When an interrupt occurs, its priority level is ascertained by the interrupt controller (INTC). NMI is always accepted, but other interrupts are only accepted if they have a priority level higher than the priority level set in the interrupt mask bits (I3-I0) of the status register (SR).

When an interrupt is accepted, exception processing begins. In interrupt exception processing, the CPU saves SR and the program counter (PC) to the stack. The priority level value of the accepted interrupt is written to SR bits I3-IO. For NMI, however, the priority level is 16 , but the value set in I3-I0 is H'F (level 15). Next, the start address of the exception service routine is fetched from the exception processing vector table for the accepted interrupt, that address is jumped to and execution begins. See section 6.4, Interrupt Operation, for more information on the interrupt exception processing.

### 5.5 Exceptions Triggered by Instructions

Exception processing can be triggered by trap instructions, general illegal instructions, and illegal slot instructions, as shown in table 5.9.

Table 5.9 Types of Exceptions Triggered by Instructions

| Type | Source Instruction | Comment |
| :--- | :--- | :--- |
| Trap instructions | TRAPA | - |
| Illegal slot | Undefined code placed | Delayed branch instructions: JMP, JSR, |
| instructions | immediately after a delayed <br> branch instruction (delay slot) | BRA, BSR, RTS, RTE, BF/S, BT/S, BSRF, <br>  <br>  <br>  <br>  <br>  <br>  <br> and instructions that rewrite the |
|  |  | Instructions that rewrite the PC: JMP, JSR, |
|  | BRA, BSR, RTS, RTE, BT, BF, TRAPA, |  |
|  | BF/S, BT/S, BSRF, BRAF |  |

### 5.5.1 Trap Instructions

When a TRAPA instruction is executed, trap instruction exception processing starts up. The CPU operates as follows:

1. The status register (SR) is saved to the stack.
2. The program counter $(\mathrm{PC})$ is saved to the stack. The PC value saved is the start address of the instruction to be executed after the TRAPA instruction.
3. The exception service routine start address is fetched from the exception processing vector table that corresponds to the vector number specified in the TRAPA instruction. That address is jumped to and the program starts executing. The jump that occurs is not a delayed branch.

### 5.5.2 Illegal Slot Instructions

An instruction placed immediately after a delayed branch instruction is said to be placed in a delay slot. When the instruction placed in the delay slot is undefined code, illegal slot exception processing starts up when that undefined code is decoded. Illegal slot exception processing also starts up when an instruction that rewrites the program counter ( PC ) is placed in a delay slot. The processing starts when the instruction is decoded. The CPU handles an illegal slot instruction as follows:

1. The status register (SR) is saved to the stack.
2. The program counter ( PC ) is saved to the stack. The PC value saved is the jump address of the delayed branch instruction immediately before the undefined code or the instruction that rewrites the PC.
3. The exception service routine start address is fetched from the exception processing vector table that corresponds to the exception that occurred. That address is jumped to and the program starts executing. The jump that occurs is not a delayed branch.

### 5.5.3 General Illegal Instructions

When undefined code placed anywhere other than immediately after a delayed branch instruction (i.e., in a delay slot) is decoded, general illegal instruction exception processing starts up. The CPU handles general illegal instructions the same as illegal slot instructions. Unlike processing of illegal slot instructions, however, the program counter value stored is the start address of the undefined code.

### 5.6 When Exception Sources Are Not Accepted

When an address error or interrupt is generated after a delayed branch instruction or interruptdisabled instruction, it is sometimes not accepted immediately but stored instead, as shown in table 5.10. When this happens, it will be accepted when an instruction that can accept the exception is decoded.

## Table 5.10 Generation of Exception Sources Immediately after a Delayed Branch

 Instruction or Interrupt-Disabled Instruction|  | Exception Source |  |
| :--- | :--- | :--- |
| Point of Occurrence | Address Error | Interrupt |
| Immediately after a delayed branch instruction ${ }^{* 1}$ | Not accepted | Not accepted |
| Immediately after an interrupt-disabled instruction*2 | Accepted | Not accepted |
| Notes: $* 1$Delayed branch instructions: JMP, JSR, BRA, BSR, RTS, RTE, BF/S, BT/S, BSRF, <br> BRAF <br> $* 2$ |  |  |
| Interrupt-disabled instructions: LDC, LDC.L, STC, STC.L, LDS, LDS.L, STS, STS.L |  |  |

### 5.6.1 Immediately after a Delayed Branch Instruction

When an instruction placed immediately after a delayed branch instruction (delay slot) is decoded, neither address errors nor interrupts are accepted. The delayed branch instruction and the instruction located immediately after it (delay slot) are always executed consecutively, so no exception processing occurs during this period.

### 5.6.2 Immediately after an Interrupt-Disabled Instruction

When an instruction immediately following an interrupt-disabled instruction is decoded, interrupts are not accepted. Address errors are accepted.

### 5.7 Stack Status after Exception Processing Ends

The status of the stack after exception processing ends is as shown in table 5.11.
Table 5.11 Types of Stack Status after Exception Processing Ends

| Types | Stack Status |
| :---: | :---: |
| Address error | $\tau$ |
|  | $\mathrm{SP} \rightarrow \begin{aligned} & \text { Address of instruction } \\ & \text { after executed instruction }\end{aligned} 32$ bits |
|  | SR 32 bits |
|  | $\sim \sim$ |
| Trap instruction | $\tau$ |
|  | $\mathrm{SP} \rightarrow \begin{aligned} & \text { Address of instruction } \\ & \text { after TRAPA instruction }\end{aligned} 32$ bits |
|  | SR 32 bits |
|  | $\sim$ |
| General illegal instruction | - |
|  | $S P \rightarrow \begin{aligned} & \text { Start address of illegal } \\ & \text { instruction }\end{aligned} \quad 32$ bits |
|  | SR 32 bits |
|  | $\mathcal{L}$ |
| Interrupt | der |
|  | $\mathrm{SP} \rightarrow \begin{aligned} & \text { Address of instruction } \\ & \text { after executed instruction }\end{aligned} 32$ bits |
|  | SR 32 bits |
|  | $\mathcal{\sim}$ |
| Illegal slot instruction | F Jump destination address $\sim^{\sim}$ |
|  | SP $\rightarrow$ of delay branch instruction 32 bits |
|  | SR 32 bits |
|  | $\sim \sim$ |

### 5.8 Notes on Use

### 5.8.1 Value of Stack Pointer (SP)

The value of the stack pointer must always be a multiple of four. If it is not, an address error will occur when the stack is accessed during exception processing.

### 5.8.2 Value of Vector Base Register (VBR)

The value of the vector base register must always be a multiple of four. If it is not, an address error will occur when the stack is accessed during exception processing.

### 5.8.3 Address Errors Caused by Stacking of Address Error Exception Processing

When the stack pointer is not a multiple of four, an address error will occur during stacking of the exception processing (interrupts, etc.) and address error exception processing will start up as soon as the first exception processing is ended. Address errors will then also occur in the stacking for this address error exception processing. To ensure that address error exception processing does not go into an endless loop, no address errors are accepted at that point. This allows program control to be shifted to the address error exception service routine and enables error processing.

When an address error occurs during exception processing stacking, the stacking bus cycle (write) is executed. During stacking of the status register (SR) and program counter (PC), the SP is -4 for both, so the value of SP will not be a multiple of four after the stacking either. The address value output during stacking is the SP value, so the address where the error occurred is itself output. This means the write data stacked will be undefined.

## Section 6 Interrupt Controller (INTC)

### 6.1 Overview

The interrupt controller (INTC) ascertains the priority of interrupt sources and controls interrupt requests to the CPU. The INTC has registers for setting the priority of each interrupt which can be used by the user to order the priorities in which the interrupt requests are processed.

### 6.1.1 Features

The INTC has the following features:

- 16 levels of interrupt priority: By setting the eight interrupt-priority level registers, the priorities of IRQ interrupts and on-chip peripheral module interrupts can be set in 16 levels for different request sources.
- NMI noise canceler function: NMI input level bits indicate the NMI pin status. By reading these bits with the interrupt exception service routine, the pin status can be confirmed, enabling it to be used as a noise canceler.
- Notification of interrupt occurrence can be reported externally ( $\overline{\text { IRQOUT }}$ pin). For example, it is possible to request bus rights if an external bus master is informed that a peripheral module interrupt has occurred when the LSI has released the bus rights.


### 6.1.2 Block Diagram

Figure 6.1 is a block diagram of the INTC.


Figure 6.1 INTC Block Diagram

### 6.1.3 Pin Configuration

Table 6.1 shows the INTC pin configuration.
Table 6.1 Pin Configuration

| Name | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Non-maskable interrupt input pin | NMI | I | Input of non-maskable interrupt <br> request signal |
| Interrupt request input pins | $\overline{\text { IRQ0 }} \overline{\overline{\mathrm{RQ7}}}$ | I | Input of maskable interrupt request <br> signals |
| Interrupt request output pin | $\overline{\text { IRQOUT }}$ | O | Output of notification signal when an <br> interrupt has occurred |

### 6.1.4 Register Configuration

The INTC has the 10 registers shown in table 6.2 . These registers set the priority of the interrupts and control external interrupt input signal detection.

Table 6.2 Register Configuration

| Name | Abbr. | R/W | Initial Value | Address | Access Sizes |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Interrupt priority register A | IPRA | R/W | H'0000 | H'FFFF8348 | 8, 16, 32 |
| Interrupt priority register B | IPRB | R/W | H'0000 | H'FFFF834A | 8, 16, 32 |
| Interrupt priority register C | IPRC | R/W | H'0000 | H'FFFF834C | 8, 16, 32 |
| Interrupt priority register D | IPRD | R/W | H'0000 | H'FFFF834E | 8, 16, 32 |
| Interrupt priority register E | IPRE | R/W | H'0000 | H'FFFF8350 | 8, 16, 32 |
| Interrupt priority register F | IPRF | R/W | H'0000 | H'FFFF8352 | 8, 16, 32 |
| Interrupt priority register G | IPRG | R/W | H'0000 | H'FFFF8354 | 8, 16, 32 |
| Interrupt priority register H | IPRH | R/W | H'0000 | H'FFFF8356 | 8, 16, 32 |
| Interrupt control register | ICR | R/W | *1 | H'FFFF8358 | 8,16, 32 |
| IRQ status register | ISR | $\mathrm{R}(\mathrm{W})^{* 2}$ | H'0000 | H'FFFF835A | 8, 16, 32 |

Notes: *1 The value when the NMI pin is high is H'8000; when the NMI pin is low, it is H'0000.
*2 Only 0 can be written, in order to clear flags.

### 6.2 Interrupt Sources

There are four types of interrupt sources: NMI, user breaks, IRQ, and on-chip peripheral modules. Each interrupt has a priority expressed as a priority level ( 0 to 16 , with 0 the lowest and 16 the highest). Giving an interrupt a priority level of 0 masks it.

### 6.2.1 NMI Interrupts

The NMI interrupt has priority 16 and is always accepted. Input at the NMI pin is detected by edge. Use the NMI edge select bit (NMIE) in the interrupt control register (ICR) to select either the rising or falling edge. NMI interrupt exception processing sets the interrupt mask level bits (I3-I0) in the status register (SR) to level 15.

### 6.2.2 User Break Interrupt

A user break interrupt has a priority of level 15 , and occurs when the break condition set in the user break controller (UBC) is satisfied. User break interrupt requests are detected by edge and are held until accepted. User break interrupt exception processing sets the interrupt mask level bits (I3-I0) in the status register (SR) to level 15. For more information about the user break interrupt, see section 7, User Break Controller (UBC).

### 6.2.3 IRQ Interrupts

IRQ interrupts are requested by input from pins $\overline{\text { IRQ } 0}-\overline{\mathrm{IRQ}} 7$. Set the IRQ sense select bits (IRQ0S-IRQ7S) of the interrupt control register (ICR) to select low level detection or falling edge detection for each pin. The priority level can be set from 0 to 15 for each pin using the interrupt priority registers A and B (IPRA-IPRB).

When IRQ interrupts are set to low level detection, an interrupt request signal is sent to the INTC during the period the IRQ pin is low level. Interrupt request signals are not sent to the INTC when the IRQ pin becomes high level. Interrupt request levels can be confirmed by reading the IRQ flags (IRQ0F-IRQ7F) of the IRQ status register (ISR).

When IRQ interrupts are set to falling edge detection, interrupt request signals are sent to the INTC upon detecting a change on the IRQ pin from high to low level. IRQ interrupt request detection results are maintained until the interrupt request is accepted. Confirmation that IRQ interrupt requests have been detected is possible by reading the IRQ flags (IRQ0F-IRQ7F) of the IRQ status register (ISR), and by writing a 0 after reading a 1 , IRQ interrupt request detection results can be withdrawn.

In IRQ interrupt exception processing, the interrupt mask bits (I3-I0) of the status register (SR) are set to the priority level value of the accepted IRQ interrupt.

### 6.2.4 On-Chip Peripheral Module Interrupts

On-chip peripheral module interrupts are interrupts generated by the following on-chip peripheral modules:

- Direct memory access controller (DMAC)
- Multifunction timer/pulse unit (MTU)
- Compare match timer (CMT)
- Serial communications interface (SCI)
- A/D converter (A/D)
- Data transfer controller (DTC)
- Watchdog timer (WDT)
- Bus state controller (BSC)
- I/O port (I/O)

A different interrupt vector is assigned to each interrupt source, so the exception service routine does not have to decide which interrupt has occurred. Priority levels between 0 and 15 can be assigned to individual on-chip peripheral modules in interrupt priority registers $\mathrm{C}-\mathrm{H}$ (IPRCIPRH).

On-chip peripheral module interrupt exception processing sets the interrupt mask level bits (I3-I0) in the status register (SR) to the priority level value of the on-chip peripheral module interrupt that was accepted.

### 6.2.5 Interrupt Exception Vectors and Priority Rankings

Table 6.3 lists interrupt sources and their vector numbers, vector table address offsets and interrupt priorities.

Each interrupt source is allocated a different vector number and vector table address offset. Vector table addresses are calculated from vector numbers and address offsets. In interrupt exception processing, the exception service routine start address is fetched from the vector table indicated by the vector table address. See table 5.4, Calculating Exception Processing Vector Table Addresses.

IRQ interrupts and on-chip peripheral module interrupt priorities can be set freely between 0 and 15 for each pin or module by setting interrupt priority registers A-H (IPRA-IPRH). The ranking of interrupt sources for IPRC-IPRH, however, must be the order listed under Priority Order Within IPR Setting Range in table 6.3 and cannot be changed. A power-on reset assigns priority level 0 to IRQ interrupts and on-chip peripheral module interrupts. If the same priority level is assigned to two or more interrupt sources and interrupts from those sources occur simultaneously, their priority order is the default priority order indicated at the right in table 6.3.

Table 6.3 Interrupt Exception Processing Vectors and Priorities

| Interrupt Source | Interrupt Vector |  | Interrupt |  | Priority within IPR Setting Range |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Vector No. | Vector Table <br> Address <br> Offset | Priority <br> (Initial <br> Value) | Corresponding IPR (Bits) |  | Default Priority |
| NMI | 11 | $\begin{aligned} & \text { H'0000002C- } \\ & \text { H'0000002F } \end{aligned}$ | 16 | - | - | High <br> - |
| User break | 12 | $\begin{aligned} & \text { H'00000030- } \\ & \text { H'00000033 } \end{aligned}$ | 15 | - | - |  |
| IRQ0 | 64 | $\begin{aligned} & \text { H'00000100- } \\ & \text { H'00000103 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRA } \\ & (15-12) \end{aligned}$ | - |  |
| IRQ1 | 65 | $\begin{aligned} & \text { H'00000104- } \\ & \text { H'00000107 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRA } \\ & (11-8) \end{aligned}$ | - |  |
| IRQ2 | 66 | $\begin{aligned} & \text { H'00000108- } \\ & \text { H'0000010B } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRA } \\ & (7-4) \end{aligned}$ | - |  |
| IRQ3 | 67 | $\begin{aligned} & \text { H'0000010C- } \\ & \text { H'0000010F } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRA } \\ & (3-0) \end{aligned}$ | - |  |
| IRQ4 | 68 | $\begin{aligned} & \text { H'00000110- } \\ & \text { H'00000113 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRB } \\ & (15-12) \end{aligned}$ | - |  |
| IRQ5 | 69 | $\begin{aligned} & \text { H'00000114- } \\ & \text { H'00000117 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRB } \\ & (11-8) \end{aligned}$ | - |  |
| IRQ6 | 70 | $\begin{aligned} & \text { H'00000118- } \\ & \text { H'0000011B }^{\prime} \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRB } \\ & (7-4) \end{aligned}$ | - |  |
| IRQ7 | 71 | $\begin{aligned} & \text { H'0000011C- } \\ & \text { H'0000011F } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRB } \\ & (3-0) \end{aligned}$ | - |  |
| DMACO DEIO | 72 | $\begin{aligned} & \text { H'00000120- } \\ & \text { H'00000123 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRC } \\ & (15-12) \end{aligned}$ | - |  |
| DMAC1 DEI1 | 76 | $\begin{aligned} & \text { H'00000130- } \\ & \text { H'00000133 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRC } \\ & (11-8) \end{aligned}$ | - |  |
| DMAC2 DEI2 | 80 | $\begin{aligned} & \text { H'00000140- } \\ & \text { H'00000143 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRC } \\ & (7-4) \end{aligned}$ | - |  |
| DMAC3 DEI3 | 84 | $\begin{aligned} & \text { H'00000150- } \\ & \text { H'00000153 }^{\prime} \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRC } \\ & (3-0) \end{aligned}$ | - | Low |

Table 6.3 Interrupt Exception Processing Vectors and Priorities (cont)

| Interrupt Source |  | Interrupt Vector |  | Interrupt <br> Priority <br> (Initial <br> Value) | Corresponding IPR (Bits) | Priority within IPR <br> Setting Range | Default Priority |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Vector No. | Vector Table <br> Address <br> Offset |  |  |  |  |
| MTU0 | TGIOA | 88 | $\begin{aligned} & \hline \mathrm{H}^{\prime} 00000160- \\ & \mathrm{H}^{\prime} 00000163 \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \hline \text { IPRD } \\ & (15-12) \end{aligned}$ | High | High |
|  | TGIOB | 89 | $\begin{aligned} & \text { H'00000164- } \\ & \text { H'00000167 } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TGIOC | 90 | $\begin{aligned} & \text { H'00000168- } \\ & \text { H'0000016B } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TGIOD | 91 | $\begin{aligned} & \text { H'0000016C- } \\ & \text { H'0000016F } \end{aligned}$ | 0-15 (0) |  | $\underset{\text { Low }}{\nabla}$ |  |
|  | TCIOV | 92 | $\begin{aligned} & \text { H'00000170- } \\ & \text { H'00000173 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRD } \\ & (11-8) \end{aligned}$ | - |  |
| MTU1 | TGI1A | 96 | $\begin{aligned} & \text { H'00000180- } \\ & \text { H'00000183 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRD } \\ & (7-4) \end{aligned}$ | High |  |
|  | TGI1B | 97 | $\begin{aligned} & \text { H'00000184- } \\ & \text { H'00000187 } \end{aligned}$ | 0-15 (0) |  | Low |  |
|  | TCI1V | 100 | $\begin{aligned} & \text { H'00000190- } \\ & \text { H'00000193 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRD } \\ & (3-0) \end{aligned}$ | High $\Delta$ |  |
|  | TCI1U | 101 | $\begin{aligned} & \text { H'00000194- } \\ & \text { H'00000197 } \end{aligned}$ | 0-15 (0) |  | Low |  |
| MTU2 | TGI2A | 104 | $\begin{aligned} & \text { H'000001A0- } \\ & \text { H'000001A3 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRE } \\ & (15-12) \end{aligned}$ | High $\Delta$ |  |
|  | TGI2B | 105 | $\begin{aligned} & \text { H'000001A4- } \\ & \text { H'000001A7 } \end{aligned}$ | 0-15 (0) |  | Low |  |
|  | TCI2V | 108 | $\begin{aligned} & \text { H'000001B0- } \\ & \text { H'000001B3 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRE } \\ & (11-8) \end{aligned}$ | High |  |
|  | TCI2U | 109 | $\begin{aligned} & \text { H'000001B4- } \\ & \text { H'000001B7 } \end{aligned}$ | 0-15 (0) |  | Low | $\underset{\text { Low }}{\downarrow}$ |

Table 6.3 Interrupt Exception Processing Vectors and Priorities (cont)

| Interrupt Source |  | Interrupt Vector |  | Interrupt <br> Priority <br> (Initial <br> Value) | Corresponding IPR (Bits) | Priority within IPR Setting Range | Default <br> Priority |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Vector <br> No. | Vector Table <br> Address <br> Offset |  |  |  |  |
| MTU3 | TGI3A | 112 | $\begin{aligned} & \text { H'000001C0- } \\ & \text { H'000001C3 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \hline \text { IPRE } \\ & (7-4) \end{aligned}$ | High | High |
|  | TGI3B | 113 | $\begin{aligned} & \text { H'000001C4- } \\ & \text { H'000001C7 } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TGI3C | 114 | $\begin{aligned} & \text { H'000001C8- } \\ & \text { H'000001CB } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TGI3D | 115 | $\begin{aligned} & \text { H'000001CC- } \\ & \text { H'000001CF } \end{aligned}$ | 0-15 (0) |  | $\begin{aligned} & \text { Vow } \\ & \text { Lo } \end{aligned}$ |  |
|  | TCI3V | 116 | $\begin{aligned} & \text { H'000001D0- } \\ & \text { H'000001D3 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRE } \\ & (3-0) \end{aligned}$ | - |  |
| MTU4 | TGI4A | 120 | $\begin{aligned} & \text { H'000001E0- } \\ & \text { H'000001E3 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRF } \\ & (15-12) \end{aligned}$ | High |  |
|  | TGI4B | 121 | $\begin{aligned} & \text { H'000001E4- } \\ & \text { H'000001E7 } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TGI4C | 122 | $\begin{aligned} & \text { H'000001E8- } \\ & \text { H'000001EB } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TGI4D | 123 | $\begin{aligned} & \text { H'000001EC- } \\ & \text { H'000001EF } \end{aligned}$ | 0-15 (0) |  | $\begin{gathered} \text { Low } \end{gathered}$ |  |
|  | TCI4V | 124 | $\begin{aligned} & \text { H'000001F0- } \\ & \text { H'000001F3 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRF } \\ & (11-8) \end{aligned}$ | High |  |
|  | Reserved | 125 | $\begin{aligned} & \text { H'000001F4- } \\ & \text { H'000001F7 } \end{aligned}$ | 0-15 (0) |  | Low |  |
| SCIO | ERIO | 128 | $\begin{aligned} & \text { H'00000200- } \\ & \text { H'00000203 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRF } \\ & (7-4) \end{aligned}$ | High |  |
|  | RXIO | 129 | $\begin{aligned} & \text { H'00000204- } \\ & \text { H'00000207 } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TXIO | 130 | $\begin{aligned} & \text { H'00000208- } \\ & \text { H'0000020B } \end{aligned}$ | 0-15 (0) |  |  |  |
|  | TEIO | 131 | $\begin{aligned} & \text { H'0000020C- } \\ & \text { H'0000020F } \end{aligned}$ | 0-15 (0) |  | $\begin{gathered} \text { Low } \end{gathered}$ | $\begin{gathered} \nabla \\ \text { Low } \end{gathered}$ |

Table 6.3 Interrupt Exception Processing Vectors and Priorities (cont)

| Interrupt Source |  | Interrupt Vector |  | Interrupt <br> Priority <br> (Initial <br> Value) | Corresponding IPR (Bits) | Priority within IPR Setting Range |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Vector <br> No. | Vector Table Address Offset |  |  |  |
| SCl1 | ERI1 | 132 | $\begin{aligned} & \hline H^{\prime} 00000210- \\ & \text { H'00000213 }^{\prime} \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \hline \text { IPRF } \\ & (3-0) \end{aligned}$ | High |
|  | RXI1 | 133 | $\begin{aligned} & \text { H'00000214- } \\ & \text { H'00000217 } \end{aligned}$ | 0-15 (0) |  |  |
|  | TXI1 | 134 | $\begin{aligned} & \text { H'00000218- } \\ & \text { H'0000021B } \end{aligned}$ | 0-15 (0) |  |  |
|  | TEI1 | 135 | $\begin{aligned} & \text { H'0000021C- } \\ & \text { H'0000021F } \end{aligned}$ | 0-15 (0) |  | $\stackrel{\downarrow}{\text { Low }}$ |
| A/D* | ADI | 136 | $\begin{aligned} & \text { H'00000220- } \\ & \text { H'00000223 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRG } \\ & (15-12) \end{aligned}$ | - |
| DTC | SWDTCE |  | $\begin{aligned} & \hline \text { H'00000230- }_{\text {H'00000233 }} \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRG } \\ & (11-8) \end{aligned}$ | - |
| CMTO | CMIO | 144 | $\begin{aligned} & \text { H'00000240- } \\ & \text { H'00000243 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRG } \\ & (7-4) \end{aligned}$ | - |
| CMT1 | CMI1 | 148 | $\begin{aligned} & \text { H'00000250- } \\ & \text { H'00000253 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRG } \\ & (3-0) \end{aligned}$ | - |
| WDT | ITI | 152 | $\begin{aligned} & \text { H'00000260- } \\ & \text { H'00000263 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRH } \\ & (15-12) \end{aligned}$ | High |
| BSC | CMI | 153 | $\begin{aligned} & \text { H'00000264- } \\ & \text { H'00000267 } \end{aligned}$ | 0-15 (0) |  | Low |
| I/O | OEI | 156 | $\begin{aligned} & \text { H'00000270- } \\ & \text { H'00000273 }^{\prime} \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \hline \text { IPRH } \\ & (11-8) \end{aligned}$ | - |
| Note: * For A mask products, A/D is as follows |  |  |  |  |  |  |
| A/D | ADIO | 136 | $\begin{aligned} & \text { H'00000220- } \\ & \text { H'00000223 } \end{aligned}$ | 0-15 (0) | $\begin{aligned} & \text { IPRG } \\ & (15-12) \end{aligned}$ | High |
|  | ADI1 | 137 | $\begin{aligned} & \text { H'00000224- } \\ & \text { H'00000227 } \end{aligned}$ | 0-15 (0) |  | Low |

### 6.3 Description of Registers

### 6.3.1 Interrupt Priority Registers A-H (IPRA-IPRH)

Interrupt priority registers A-H (IPRA-IPRH) are 16-bit readable/writable registers that set priority levels from 0 to 15 for IRQ interrupts and on-chip peripheral module interrupts.
Correspondence between interrupt request sources and each of the IPRA-IPRH bits is shown in table 6.4.


Table 6.4 Interrupt Request Sources and IPRA-IPRH

|  | Bits |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
| Register | $\mathbf{1 5 - 1 2}$ | $\mathbf{1 1 - 8}$ | $\mathbf{7 - 4}$ | 3-0 |
| Interrupt priority register A | IRQ0 | IRQ1 | IRQ2 | IRQ3 |
| Interrupt priority register B | IRQ4 | IRQ5 | IRQ6 | IRQ7 |
| Interrupt priority register C | DMAC0 | DMAC1 | DMAC2 | DMAC3 |
| Interrupt priority register D | MTU0 | MTU0 | MTU1 | MTU1 |
| Interrupt priority register E | MTU2 | MTTU2 | MTU3 | MTU3 |
| Interrupt priority register F | MTU4 | MTU4 | SCI0 | SCI1 |
| Interrupt priority register G | A/D(A/D0, | DTC | CMT0 | CMT1 |
| Interrupt priority register H | A/D1)* |  |  |  |
| Note: * Excluding A mask products are A/D, A mask products are A/D0 and A/D1. |  |  |  |  |

Note: * Excluding A mask products are A/D, A mask products are A/D0 and A/D1.

As indicated in table 6.4 , four $\overline{\text { IRQ }}$ pins or groups of 4 on-chip peripheral modules are allocated to each register. Each of the corresponding interrupt priority ranks are established by setting a value from H'0 (0000) to H'F (1111) in each of the four-bit groups $15-12,11-8,7-4$, and $3-0$. Interrupt priority rank becomes level 0 (lowest) by setting $\mathrm{H}^{\prime} 0$, and level 15 (highest) by setting H'F. If multiple on-chip peripheral modules are assigned to WDT and BSC, those multiple modules are set to the same priority rank.

IPRA-IPRH are initialized to $\mathrm{H}^{\prime} 0000$ by a power-on reset or a manual reset. They are not initialized in standby mode.

### 6.3.2 Interrupt Control Register (ICR)

The ICR is a 16-bit register that sets the input signal detection mode of the external interrupt input pin NMI and $\overline{\mathrm{IRQ} 0}-\overline{\mathrm{IRQ} 7}$ and indicates the input signal level to the NMI pin. A power-on reset initializes ICR but the standby mode does not.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | NMIL | - | - | - | - | - | - | NMIE |
|  | $*$ | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Initial value: | $*$ | 0 | $R$ | $R$ | $R$ | $R$ | $R$ | $R / W$ |



Note: * When NMI input is high: 1 ; when NMI input is low: 0

- Bit 15—NMI Input Level (NMIL): Sets the level of the signal input at the NMI pin. This bit can be read to determine the NMI pin level. This bit cannot be modified.

Bit 15: NMIL Description

| 0 | NMI input level is low |
| :--- | :--- |
| 1 | NMI input level is high |

- Bits 14-9—Reserved: These bits always read as 0 . The write value should always be 0 .
- Bit 8—NMI Edge Select (NMIE)

Bit 8: NMIE Description

| 0 | Interrupt request is detected on falling edge of NMI input (initial value) |
| :--- | :--- |
| 1 | Interrupt request is detected on rising edge of NMI input |

- Bits 7-0—IRQ0-IRQ7 Sense Select (IRQ0S-IRQ7S): These bits set the IRQ0-IRQ7 interrupt request detection mode.

Bits 7-0: IRQ0S-IRQ7S Description

| 0 | Interrupt request is detected on low level of IRQ input (initial value) |
| :--- | :--- |
| 1 | Interrupt request is detected on falling edge of IRQ input |

### 6.3.3 IRQ Status Register (ISR)

The ISR is a 16-bit register that indicates the interrupt request status of the external interrupt input pins $\overline{\mathrm{IRQ}} 0-\overline{\mathrm{IRQ} 7}$. When IRQ interrupts are set to edge detection, held interrupt requests can be withdrawn by writing a 0 to IRQnF after reading an $\mathrm{IRQnF}=1$.

A power-on reset initializes ISR but the standby mode does not.


- Bits $15-8$-Reserved: These bits always read as 0 . The write value should always be 0 .
- Bits 7-0—IRQ0-IRQ7 Flags (IRQ0F-IRQ7F): These bits display the IRQ0-IRQ7 interrupt request status.

Bits 7-0:
IRQ0F-IRQ7F Detection Setting Description

| 0 | Level detection | No IRQn interrupt request exists. <br> Clear conditions: When $\overline{\mathrm{IRQn}}$ input is high level <br> No IRQn interrupt request was detected. (initial value) <br> Clear conditions: <br> 1. When a 0 is written after reading $\operatorname{IRQnF}=1$ status <br> 2. When IRQn interrupt exception processing has been executed <br> 3. When a DTC transfer due to IRQn interrupt has been executed |
| :---: | :---: | :---: |
| 1 | Level detection | An IRQn interrupt request exists. <br> Set conditions: When $\overline{\mathrm{RQR}}$ input is low level |
|  | Edge detection | An IRQn interrupt request was detected. <br> Set conditions: When a falling edge occurs at an $\overline{\mathrm{RQn}}$ in |



Figure 6.2 External Interrupt Process

### 6.4 Interrupt Operation

### 6.4.1 Interrupt Sequence

The sequence of interrupt operations is explained below. Figure 6.3 is a flowchart of the operations.

1. The interrupt request sources send interrupt request signals to the interrupt controller.
2. The interrupt controller selects the highest priority interrupt in the interrupt requests sent, following the priority levels set in interrupt priority level setting registers A-H (IPRA-IPRH). Lower-priority interrupts are ignored. They are held pending until interrupt requests designated as edge-detect type are accepted. For IRQ interrupts, however, withdrawal is possible by accessing the IRQ status register (ISR). See section 6.2.3, IRQ Interrupts, for details. Interrupts held pending due to edge detection are cleared by a power-on reset or a manual reset. If two of these interrupts have the same priority level or if multiple interrupts occur within a single module, the interrupt with the highest default priority or the highest priority within its IPR setting range (as indicated in table 6.3) is selected.
3. The interrupt controller compares the priority level of the selected interrupt request with the interrupt mask bits (I3-I0) in the CPU's status register (SR). If the request priority level is equal to or less than the level set in I3-I0, the request is ignored. If the request priority level is higher than the level in bits I3-I0, the interrupt controller accepts the interrupt and sends an interrupt request signal to the CPU.
4. When the interrupt controller accepts an interrupt, a low level is output from the $\overline{\text { IRQOUT }}$ pin.
5. The CPU detects the interrupt request sent from the interrupt controller when it decodes the next instruction to be executed. Instead of executing the decoded instruction, the CPU starts interrupt exception processing (figure 6.4).
6. SR and PC are saved onto the stack.
7. The priority level of the accepted interrupt is copied to the interrupt mask level bits (I3-I0) in the status register (SR).
8. When the accepted interrupt is sensed by level or is from an on-chip peripheral module, a high level is output from the $\overline{\text { IRQOUT }}$ pin. When the accepted interrupt is sensed by edge, a high level is output from the $\overline{\text { IRQOUT }}$ pin at the point when the CPU starts interrupt exception processing instead of instruction execution as noted in (5) above. However, if the interrupt controller accepts an interrupt with a higher priority than one it is in the midst of accepting, the $\overline{\text { IRQOUT }}$ pin will remain low level.
9. The CPU reads the start address of the exception service routine from the exception vector table for the accepted interrupt, jumps to that address, and starts executing the program there. This jump is not a delay branch.


I3 to IO: Interrupt mask bits of status register
Notes: *1 IRQOUT is the same signal as the interrupt request signal to the CPU (see figure 6.1). Thus, it is output when there is a higher priority interrupt request than the one in the I3 to IO bits of the SR.
*2 When the accepted interrupt is sensed by edge, the IRQOUT pin becomes high level at the point when the CPU starts interrupt exception processing instead of instruction execution (before SR is saved to the stack). If the interrupt controller has accepted another interrupt with a higher priority and has output an interrupt request to the CPU, the IRQOUT pin will remain low level.

Figure 6.3 Interrupt Sequence Flowchart

### 6.4.2 Stack after Interrupt Exception Processing

Figure 6.4 shows the stack after interrupt exception processing.


Notes: *1 PC: Start address of the next instruction (return destination instruction) after the executing instruction
*2 Always be certain that SP is a multiple of 4

Figure 6.4 Stack after Interrupt Exception Processing

### 6.5 Interrupt Response Time

Table 6.5 indicates the interrupt response time, which is the time from the occurrence of an interrupt request until the interrupt exception processing starts and fetching of the first instruction of the interrupt service routine begins. Figure 6.5 shows the pipeline when an IRQ interrupt is accepted.

## Table 6.5 Interrupt Response Time

|  | Number of States |  |
| :--- | :--- | :--- |
| NMI, Peripheral <br> Module | IRQ | Notes |



F: Instruction fetch (instruction fetched from memory where program is stored).
D: Instruction decoding (fetched instruction is decoded).
E: Instruction execution (data operation and address calculation is performed according to the results of decoding).
M : Memory access (data in memory is accessed).

Figure 6.5 Pipeline when an IRQ Interrupt is Accepted

### 6.6 Data Transfer with Interrupt Request Signals

The following data transfers can be done using interrupt request signals:

- Activate DMAC only, without generating CPU interrupt
- Activate DTC only, CPU interrupts according to DTC settings

Among interrupt sources, those designated as DMAC activating sources are masked and not input to the INTC. The masking condition is listed below:

Mask condition = DME • (DE0 • source selection $0+$ DE1 $\times$ source selection $1+$ DE2 • source selection $2+$ DE3 • source selection 3)

The INTC masks CPU interrupts when the corresponding DTE bit is a 1 . The DTE clear condition and interrupt source flag clear condition are listed below.

DTE clear condition = DTC transfer end $\cdot$ DTECLR
Interrupt source flag clear condition $=$ DTC transfer end $\cdot \overline{\text { DTECLR }}+$ DMAC transfer end
Where: DTECLR = DISEL + counter 0 .
Figure 6.6 shows a control block diagram.


Figure 6.6 Interrupt Control Block Diagram

### 6.6.1 Handling DTC Activating and CPU Interrupt Sources, but Not DMAC Activating Sources

1. Either do not select the DMAC as a source, or clear the DME bit to 0 .
2. For DTC, set the corresponding DTE bits and DISEL bits to 1 .
3. Activating sources are applied to the DTC when interrupts occur.
4. When the DTC performs a data transfer, it clears the DTE bit to 0 and sends an interrupt request to the CPU. The activating source does not clear.
5. The CPU clears interrupt sources with its interrupt processing routine. It then confirms the transfer counter value. When the transfer counter value $\neq 0$, it sets the DTE bit to 1 and allows the next data transfer. If the transfer counter value $=0$, it performs the necessary end processing in the interrupt processing routine.

### 6.6.2 Handling DMAC Activating Sources but Not CPU Interrupt or DTC Activating Sources

1. Select the DMAC as a source and set the DME bit to 1 . CPU interrupt sources and DTC activating sources are masked regardless of the interrupt priority level register settings or DTC register settings.
2. Activating sources are applied to the DMAC when interrupts occur.
3. The DMAC clears activating sources at the time of data transfer.

### 6.6.3 Handling DTC Activating Sources but Not CPU Interrupt or DMAC Activating Sources

1. Either do not select the DMAC as a source, or clear the DME bit to 0 .
2. For DTC, set the corresponding DTE bits to 1 and clear the DISEL bits to 0 .
3. Activating sources are applied to the DTC when interrupts occur.
4. When the DTC performs a data transfer, it clears the activating source. An interrupt request is not sent to the CPU, because the DTE bit is maintained as a 1 .
5. However, when the transfer counter value $=0$ the DTE bit is cleared to 0 and an interrupt request is sent to the CPU.
6. The CPU performs the necessary end processing in the interrupt processing routine.

### 6.6.4 Treating CPU Interrupt Sources but Not DTC or DMAC Activating Sources

1. Either do not select the DMAC as a source, or clear the DME bit to 0 .
2. For DTC, clear the corresponding DTE bits to 0 .
3. When interrupts occur, interrupt requests are sent to the CPU.
4. The CPU clears the interrupt source and performs the necessary processing in the interrupt processing routine.

## Section 7 User Break Controller (UBC)

### 7.1 Overview

The user break controller (UBC) provides functions that simplify program debugging. Break conditions are set in the UBC and a user break interrupt is generated according to the conditions of the bus cycle generated by the CPU, DMAC, or DTC. This function makes it easy to design an effective self-monitoring debugger, enabling the chip to easily debug programs without using a large in-circuit emulator.

### 7.1.1 Features

The features of the user break controller are:

- Break compare conditions can be set:
- Address
- CPU cycle or DMA/DTC cycle
- Instruction fetch or data access
- Read or write
— Operand size: byte/word/longword
- User break interrupt generated upon satisfying break conditions. A user-designed user break interrupt exception processing routine can be run.
- Select either to break in the CPU instruction fetch cycle before the instruction is executed or after.


### 7.1.2 Block Diagram

Figure 7.1 shows a block diagram of the UBC.


UBARH, UBARL: User break address registers H, L
UBAMRH, UBAMRL: User break address mask registers H, L
UBBR: User break bus cycle register

Figure 7.1 User Break Controller Block Diagram

### 7.1.3 Register Configuration

The UBC has the five registers shown in table 7.1. Break conditions are established using these registers.

Table 7.1 Register Configuration

| Name | Abbr. | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| User break address register H | UBARH | R/W | H'0000 | H'FFFF8600 | 8, 16, 32 |
| User break address register L | UBARL | R/W | H'0000 | H'FFFF8602 | 8, 16, 32 |
| User break address mask register H | UBAMRH | R/W | H'0000 | H'FFFF8604 | 8, 16, 32 |
| User break address mask register L | UBAMRL | R/W | H'0000 | H'FFFF8606 | 8, 16, 32 |
| User break bus cycle register | UBBR | R/W | H'0000 | H'FFFF8608 | 8, 16, 32 |

### 7.2 Register Descriptions

### 7.2.1 User Break Address Register (UBAR)

The user break address register (UBAR) consists of user break address register H (UBARH) and user break address register L (UBARL). Both are 16-bit readable/writable registers. UBARH stores the upper bits (bits 31-16) of the address of the break condition, while UBARL stores the lower bits (bits 15-0). Resets and hardware standbys initialize both UBARH and UBARL to H'0000. They are not initialized in manual reset or software standby mode.

## UBARH:

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| UBARH | UBA31 | UBA30 | UBA29 | UBA28 | UBA27 | UBA26 | UBA25 | UBA24 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| UBARH | UBA23 | UBA22 | UBA21 | UBA20 | UBA19 | UBA18 | UBA17 | UBA16 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

## UBARL:

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| UBARL | UBA15 | UBA14 | UBA13 | UBA12 | UBA11 | UBA10 | UBA9 | UBA8 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| UBARL | UBA7 | UBA6 | UBA5 | UBA4 | UBA3 | UBA2 | UBA1 | UBAO |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- UBARH Bits 15-0—User Break Address 31-16 (UBA31-UBA16): These bits store the upper bit values (bits 31-16) of the address of the break condition.
- UBARL Bits 15-0—User Break Address 15-0 (UBA15-UBA0): These bits store the lower bit values (bits 15-0) of the address of the break condition.


### 7.2.2 User Break Address Mask Register (UBAMR)

The user break address mask register (UBAMR) consists of user break address mask register H (UBAMRH) and user break address mask register L (UBAMRL). Both are 16-bit readable/writable registers. UBAMRH designates whether to mask any of the break address bits established in the UBARH, and UBAMRL designates whether to mask any of the break address bits established in the UBARL. Resets and hardware standbys initialize both UBAMRH and UBAMRL to H'0000. They are not initialized in manual reset or software standby mode.

## UBAMRH:

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| UBAMRH | UBM31 | UBM30 | UBM29 | UBM28 | UBM27 | UBM26 | UBM25 | UBM24 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| UBAMRH | UBM23 | UBM22 | UBM21 | UBM20 | UBM19 | UBM18 | UBM17 | UBM16 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

## UBAMRL:

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| UBAMRL | UBM15 | UBM14 | UBM13 | UBM12 | UBM11 | UBM10 | UBM9 | UBM8 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| UBAMRL | UBM7 | UBM6 | UBM5 | UBM4 | UBM3 | UBM2 | UBM1 | UBMO |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- UBAMRH Bits 15-0—User Break Address Mask 31-16 (UBM31-UBM16): These bits designate whether to mask any of the break address 31-16 bits (UBA31-UBA16) established in the UBARH.
- UBAMRL Bits 15-0—User Break Address Mask 15-0 (UBM15-UBM0): These bits designate whether to mask any of the break address $15-0$ bits (UBA15-UBA0) established in the UBARL.

Bits 15-0: UBMn Description

| 0 | Break address UBAn is included in the break conditions (initial value) |
| :--- | :--- |
| 1 | Break address UBAn is not included in the break conditions |
| Note: $\mathrm{n}=31-0$ |  |

### 7.2.3 User Break Bus Cycle Register (UBBR)

User break bus cycle register (UBBR) is a 16-bit readable/writable register that selects from among the following four break conditions:

1. CPU cycle/ DMAC/DTC cycle
2. Instruction fetch/data access
3. Read/write
4. Operand size (byte, word, longword)

Resets and hardware standbys initialize the UBBR to H'0000. It is not initialized in software standby mode.


Bits 15-8—Reserved: These bits always read as 0 . The write value should always be 0 .

- Bits 7 and 6-CPU Cycle/Peripheral Cycle Select (CP1, CP0): These bits designate break conditions for CPU cycles or peripheral cycles (DMA/DTC cycles).

| Bit 7: CP1 | Bit 6: CP0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | No user break interrupt occurs (initial value) |
| 1 | Break on CPU cycles |  |
| 1 | 0 | Break on peripheral cycles |
| 1 | Break on both CPU and peripheral cycles |  |

- Bits 5 and 4—Instruction Fetch/Data Access Select (ID1, ID0): These bits select whether to break on instruction fetch and/or data access cycles.

| Bit 5: ID1 | Bit 4: ID0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | No user break interrupt occurs (initial value) |
| 1 | 0 | Break on instruction fetch cycles |
| 1 | 1 | Break on data access cycles |

- Bits 3 and 2—Read/Write Select (RW1, RW0): These bits select whether to break on read and/or write cycles.

| Bit 3: RW1 | Bit 2: RW0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | No user break interrupt occurs (initial value) |
| 1 | Break on read cycles |  |
| 1 | 0 | Break on write cycles |
| 1 | Break on both read and write cycles |  |

- Bits 1 and 0—Operand Size Select (SZ1, SZO): These bits select operand size as a break condition.

| Bit 1: SZ1 | Bit 0: SZO | Description |
| :--- | :--- | :--- |
| 0 | 0 | Operand size is not a break condition (initial value) |
| 1 | Break on byte access |  |
| 1 | 0 | Break on word access |
| 1 | Break on longword access |  |

Note: When breaking on an instruction fetch, set the SZO bit to 0 . All instructions are considered to be word-size accesses (even when there are instructions in on-chip memory and 2 instruction fetches are done simultaneously in 1 bus cycle).
Operand size is word for instructions or determined by the operand size specified for the CPU/DMAC data access. It is not determined by the bus width of the space being accessed.

### 7.3 Operation

### 7.3.1 Flow of the User Break Operation

The flow from setting of break conditions to user break interrupt exception processing is described below:

1. The user break addresses are set in the user break address register (UBAR), the desired masked bits in the addresses are set in the user break address mask register (UBAMR) and the breaking bus cycle type is set in the user break bus cycle register (UBBR). If even one of the three groups of the UBBR's CPU cycle/peripheral cycle select bits (CP1, CP0), instruction fetch/data access select bits (ID1, ID0), and read/write select bits (RW1, RW0) is set to 00 (no user break interrupt is generated), no user break interrupt will be generated even if all other conditions are in agreement. When using user break interrupts, always be certain to establish bit conditions for all of these three groups.
2. The UBC uses the method shown in figure 7.2 to judge whether set conditions have been fulfilled. When the set conditions are satisfied, the UBC sends a user break interrupt request signal to the interrupt controller (INTC).
3. The interrupt controller checks the accepted user break interrupt request signal's priority level. The user break interrupt has priority level 15 , so it is accepted only if the interrupt mask level in bits I3-I0 in the status register (SR) is 14 or lower. When the I3-I0 bit level is 15 , the user break interrupt cannot be accepted but it is held pending until user break interrupt exception processing can be carried out. Consequently, user break interrupts within NMI exception service routines cannot be accepted, since the I3-I0 bit level is 15 . However, if the I3-I0 bit level is changed to 14 or lower at the start of the NMI exception service routine, user break interrupts become acceptable thereafter. Section 6, Interrupt Controller (INTC), describes the handling of priority levels in greater detail.
4. The INTC sends the user break interrupt request signal to the CPU, which begins user break interrupt exception processing upon receipt. See Section 6.4, Interrupt Operation, for details on interrupt exception processing.


Figure 7.2 Break Condition Judgment Method

### 7.3.2 Break on On-Chip Memory Instruction Fetch Cycle

On-chip memory (on-chip ROM and/or RAM) is always accessed as 32 bits in 1 bus cycle. Therefore, 2 instructions can be retrieved in 1 bus cycle when fetching instructions from on-chip memory. At such times, only 1 bus cycle is generated, but by setting the start addresses of both instructions in the user break address register (UBAR) it is possible to cause independent breaks. In other words, when wanting to effect a break using the latter of two addresses retrieved in 1 bus cycle, set the start address of that instruction in UBAR. The break will occur after execution of the former instruction.

### 7.3.3 Program Counter (PC) Values Saved

Break on Instruction Fetch (Before Execution): The program counter (PC) value saved to the stack in user break interrupt exception processing is the address that matches the break condition. The user break interrupt is generated before the fetched instruction is executed. If a break condition is set in an instruction fetch cycle placed immediately after a delayed branch instruction (delay slot), or on an instruction that follows an interrupt-disabled instruction, however, the user break interrupt is not accepted immediately, but the break condition establishing instruction is executed. The user break interrupt is accepted after execution of the instruction that has accepted the interrupt. In this case, the PC value saved is the start address of the instruction that will be executed after the instruction that has accepted the interrupt.

Break on Data Access (CPU/Peripheral): The program counter (PC) value is the top address of the next instruction after the last instruction executed before the user break exception processing started. When data access (CPU/peripheral) is set as a break condition, the place where the break will occur cannot be specified exactly. The break will occur at the instruction fetched close to where the data access that is to receive the break occurs.

### 7.4 Use Examples

### 7.4.1 Break on CPU Instruction Fetch Cycle

1. Register settings:

> UBARH $=\mathrm{H}^{\prime} 0000$
> UBARL $=\mathrm{H}^{\prime} 0404$
> UBBR $=\mathrm{H}^{\prime} 0054$

Conditions set: Address: H'00000404
Bus cycle: CPU, instruction fetch, read
(operand size not included in conditions)
A user break interrupt will occur before the instruction at address $\mathrm{H}^{\prime} 00000404$. If it is possible for the instruction at $\mathrm{H}^{\prime} 00000402$ to accept an interrupt, the user break exception processing will be executed after execution of that instruction. The instruction at $\mathrm{H}^{\prime} 00000404$ is not executed. The PC value saved is $\mathrm{H}^{\prime} 00000404$.

| 2. Register settings: | UBARH $=\mathrm{H}^{\prime} 0015$ |
| :---: | :--- |
|  | UBARL $=\mathrm{H}^{\prime} 389 \mathrm{C}$ |
|  | UBBR $=\mathrm{H}^{\prime} 0058$ |
| Conditions set: | Address: $\mathrm{H}^{\prime} 0015389 \mathrm{C}$ |
|  | Bus cycle: CPU, instruction fetch, write <br> (operand size not included in conditions) |
|  |  |

A user break interrupt does not occur because the instruction fetch cycle is not a write cycle.
3. Register settings: $\quad \mathrm{UBARH}=\mathrm{H}^{\prime} 0003$

UBARL $=$ H'0147
UBBR $=\mathrm{H}^{\prime} 0054$
Conditions set: Address: H'00030147
Bus cycle: CPU, instruction fetch, read (operand size not included in conditions)

A user break interrupt does not occur because the instruction fetch was performed for an even address. However, if the first instruction fetch address after the branch is an odd address set by these conditions, user break interrupt exception processing will be done after address error exception processing.

### 7.4.2 Break on CPU Data Access Cycle

| 1. Register settings: | UBARH $=\mathrm{H}^{\prime} 0012$ |
| :---: | :--- |
|  | UBARL $=\mathrm{H}^{\prime} 3456$ |
|  | UBBR $=\mathrm{H}^{\prime} 006 \mathrm{~A}$ |
| Conditions set: | Address: $\mathrm{H}^{\prime} 00123456$ |
|  | Bus cycle: CPU , data access, write, word |

A user break interrupt occurs when word data is written into address H'00123456.
2. Register settings: $\quad$ UBARH $=\mathrm{H}^{\prime} 00 \mathrm{~A} 8$

UBARL $=$ H'0391
UBBR $=\mathrm{H}^{\prime} 0066$
Conditions set: Address: H'00A80391
Bus cycle: CPU, data access, read, word
A user break interrupt does not occur because the word access was performed on an even address.

### 7.4.3 Break on DMA/DTC Cycle

1. Register settings: $\mathrm{UBARH}=\mathrm{H}^{\prime} 0076$

UBARL $=$ H'BCDC
UBBR $=$ H'00A7
Conditions set: Address: H'0076BCDC
Bus cycle: DMA/DTC, data access, read, longword
A user break interrupt occurs when longword data is read from address H'0076BCDC.
2. Register settings: $\quad \mathrm{UBARH}=\mathrm{H}^{\prime} 0023$

UBARL $=\mathrm{H}^{\prime} 45 \mathrm{C} 8$
UBBR $=$ H'000 $^{\prime}$
Conditions set: Address: H'002345C8
Bus cycle: DMA/DTC, instruction fetch, read
(operand size not included in conditions)
A user break interrupt does not occur because no instruction fetch is performed in the DMA/DTC cycle.

### 7.5 Cautions on Use

### 7.5.1 On-Chip Memory Instruction Fetch

Two instructions are simultaneously fetched from on-chip memory. If a break condition is set on the second of these two instructions but the contents of the UBC break condition registers are changed so as to alter the break condition immediately after the first of the two instructions is fetched, a user break interrupt will still occur when the second instruction is fetched.

### 7.5.2 Instruction Fetch at Branches

When a conditional branch instruction or TRAPA instruction causes a branch, instructions are fetched and executed as follows:

1. Conditional branch instruction, branch taken: BT, BF

Instruction fetch cycles: Conditional branch instruction fetch $\rightarrow$ Next-instruction overrun fetch $\rightarrow$ Next-instruction overrun fetch $\rightarrow$ Branch destination instruction fetch Instruction execution: Conditional branch instruction execution $\rightarrow$ Branch destination instruction execution
2. TRAPA instruction, branch taken: TRAPA

Instruction fetch cycles: TRAPA instruction fetch $\rightarrow$ Next-instruction overrun fetch $\rightarrow$ Next-instruction overrun fetch $\rightarrow$ Branch destination instruction fetch

Instruction execution: TRAPA instruction execution $\rightarrow$ Branch destination instruction execution
3. Conditional delay branch instruction, branch taken: BT/S, BF/S

Instruction fetch cycles: Conditional delay branch instruction fetch $\rightarrow$ Next-instruction fetch (delay slot) $\rightarrow$ Next-instruction overrun fetch $\rightarrow$ Branch destination instruction fetch
Instruction execution: Conditional delay branch instruction execution $\rightarrow$ Delay slot instruction execution $\rightarrow$ Branch destination instruction execution

When a conditional branch instruction or TRAPA instruction causes a branch, the branch destination will be fetched after the next instruction or the one after that does an overrun fetch. However, because the instruction that is the object of the break first breaks after a definite instruction fetch and execution, the kind of overrun fetch instructions noted above do not become objects of a break. If data access breaks are also included with instruction fetch breaks as break conditions, a break occurs because the instruction overrun fetch is also regarded as becoming a data break.

### 7.5.3 Contention between User Break and Exception Handling

If a user break is set for the fetch of a particular instruction, and exception handling with higher priority than a user break is in contention and is accepted in the decode stage for that instruction (or the next instruction), user break exception handling may not be performed after completion of the higher-priority exception handling routine (on return by RTE).

Thus, if a user break condition has been set for the fetch of the branch destination instruction following a branch (BRA, BRAF, BT, BF, BT/S, BF/S, BSR, BSRF, JMP, JSR, RTS, RTE, exception handling), and exception handling for this branch destination instruction with a higher priority than a user break interrupt is accepted, user break exception handling will not be performed after completion of that exception handling routine.

Therefore, a user break condition must not be set for the fetch of the branch destination instruction following a branch.

### 7.5.4 Break at Non-Delay Branch Instruction Jump Destination

When a branch instruction with no delay slot (including exception handling) jumps to the jump destination instruction on execution of the branch, a user break will not be generated even if a user break condition has been set for the first jump destination instruction fetch.

## Section 8 Data Transfer Controller (DTC)

### 8.1 Overview

The SH7040 Series has an on-chip data transfer controller (DTC), which is activated either by interrupts or software and can perform data transfers.

### 8.1.1 Features

- Arbitrary channel number transfer setting possible
- Transfer information can be established for each interrupt source
- Transfer information stored in memory
- Multiple data transfers possible (chain transfers) for one activating source
- Address space: 32-bit addresses can be designated for both transfer source and destination
- Transfer devices
- Memory: On-chip ROM, on-chip RAM, external ROM, external RAM
- On-chip peripheral modules (excluding DMAC/DTC)
- Memory-mapped external devices
- Abundant transfer modes
— Can select between normal mode/repeat mode/block transfer mode
- Can select between increment/decrement/fixed for source/destination address
- Transfer units can be set as byte/word/longword
- Interrupts activating the DTC can be requested of the CPU
- Interrupt requests can be generated to the CPU after completion of a data transfer
- Interrupt requests generated to the CPU after completion of all designated data transfers
- Transfers can be activated by software


### 8.1.2 Block Diagram

Figure 8.1 shows the DTC block diagram. DTC transfer information is located in memory.


Figure 8.1 DTC Block Diagram

### 8.1.3 Register Configuration

The DTC has five registers in memory used for storing transfer data: DTMR, DTCR, DTSAR, DTDAR, and DTIAR. It is controlled by the three registers DTER (DTEA-DTEE), DTCSR, and DTBR. The register configurations are listed in table 8.1.

Table 8.1 Register Configuration ${ }^{* 1}$

| Name | Abbr. | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DTC mode register | DTMR | -*2 | Undefined | -*2 | -*2 |
| DTC source address register | DTSAR | -*2 | Undefined | -*2 | -*2 |
| DTC destination address register | DTDAR | -*2 | Undefined | -*2 | -*2 |
| DTC initial address register | DTIAR | -*2 | Undefined | -*2 | -*2 |
| DTC transfer count register A | DTCRA | -*2 | Undefined | -*2 | -*2 |
| DTC transfer count register B | DTCRB | -*2 | Undefined | -*2 | -*2 |
| DTC enable register A | DTEA | R/W | H'00 | H'FFFF8700 | 8, 16, 32 |
| DTC enable register B | DTEB | R/W | H'00 | H'FFFF8701 | 8, 16, 32 |
| DTC enable register C | DTEC | R/W | H'00 | H'FFFF8702 | 8, 16, 32 |
| DTC enable register D | DTED | R/W | H'00 | H'FFFF8703 | 8, 16, 32 |
| DTC enable register E | DTEE | R/W | H'00 | H'FFFF8704 | 8,16, 32 |
| DTC control/status register | DTCSR | R/(W) ${ }^{* 3}$ | H'0000 | H'FFFF8706 | 8, 16, 32 |
| DTC information base register | DTBR | R/W | Undefined | H'FFFF8708 | 16, 32 |

Notes: *1 DTC registers cannot be accessed by DMAC/DTC.
*2 DTC internal registers cannot be directly accessed.
*3 Only a 0 write after a 1 read is possible for the NMIF, AE bits of the DTCSR.

### 8.2 Register Description

### 8.2.1 DTC Mode Register (DTMR)

The DTC mode register (DTMR) is a 16-bit register that controls the DTC operation mode. The contents of this register is located in memory.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SM1 | SM0 | DM1 | DM0 | MD1 | MDO | SZ1 | SZO |
| Initial value: | * | * | * | * | * | * | * | * |
| R/W: | - | - | - | - | - | - | - | - |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Bit name: | DTS | CHNE | DISEL | NMIM | - | - | - | - |
| Initial value: | * | * | * | * | * | * | * | * |
| R/W: | - | - | - | - | - | - | - | - |

- Bits 15-14—Source Address Mode 1, 0 (SM1, SM0): These bits designate whether to hold, increment, or decrement the DTSAR after a data transfer.

| Bit $\mathbf{1 5}$ (SM1) | Bit $\mathbf{1 4}$ (SM0) | Description |
| :--- | :--- | :--- |
| 0 | - | DTSAR remains fixed |
| 1 | 0 | DTSAR is incremented after transfer <br> $(+1$ for byte unit transfer, +2 for word, +4 for longword) $)$ <br> 1 |

- Bits 13-12—Destination Address Mode 1, 0 (DM1, DM0): These bits designate whether to hold, increment or decrement the DTDAR after a data transfer.

| Bit $\mathbf{1 3}$ (DM1) | Bit $\mathbf{1 2}$ (DM0) | Description |
| :--- | :--- | :--- |
| 0 | - | DTDAR remains fixed |
| 1 | 0 | DTDAR is incremented after transfer <br> $(+1$ for byte unit transfer, +2 for word, +4 for longword) $)$ <br> 1 |

- Bits 11-10—DTC Mode 1, 0 (MD1, MD0): These bits designate the DTC transfer mode.

| Bit 11 (MD1) | Bit $\mathbf{1 0}$ (MDO) | Description |
| :--- | :--- | :--- |
| 0 | 0 | Normal mode |
| 0 | 1 | Repeat mode |
| 1 | 0 | Block transfer mode |
| 1 | 1 | Reserved (setting prohibited) |

- Bits 9-8—DTC Data Transfer Size 1, 0 (SZ1, SZ0): These bits designate the data size for data transfers.

| Bit $\mathbf{9}$ (SZ1) | Bit $\mathbf{8}$ (SZO) | Description |
| :--- | :--- | :--- |
| 0 | 0 | Byte (8 bits) |
| 0 | 1 | Word (16 bits) |
| 1 | 0 | Longword (32 bits) |
| 1 | 1 | Reserved (setting prohibited) |

- Bit 7—DTC Transfer Mode Select (DTS): When in repeat mode or block transfer mode, this bit designates whether the source side or destination side will be the repeat area or block area.


## Bit 7 (DTS)

## Description

| 0 | Destination side is the repeat area or block area |
| :--- | :--- |
| 1 | Source side is the repeat area or block area |

- Bit 6—DTC Chain Enable (CHNE): This bit designates whether to perform continuous DTC data transfers with the same activating source. Continued transfer information is read after the 16th byte from the start address of the previous transfer information.

| Bit 6 (CHNE) | Description |
| :--- | :--- |
| 0 | DTC data transfer end (activation wait state ensues) |
| 1 | DTC data transfer continue (read continue register information, execute <br> transfer) |

- Bit 5—DTC Interrupt Select (DISEL): This bit designates whether to prohibit or allow interrupt requests to the CPU after one-time DTC transfers.

| Bit 5 (DISEL) | Description |
| :--- | :--- |
| 0 | Prohibit interrupts to the CPU after DTC data transfer completion if the <br> transfer counter is not 0 (DTC clears the interrupt source flag of the <br> activating source to 0) |
| 1 | Allow interrupts to the CPU after DTC data transfer completion (DTC <br> clears the DTER bit for the interrupt of the activating source to 0) |

- Bit 4—DTC NMI Mode (NMIM): This bit designates whether to terminate transfers when an NMI is input during DTC transfers.

Bit 4 (NMIM)

## Description

| 0 | Terminate DTC transfer upon an NMI |
| :--- | :--- |
| 1 | Continue DTC transfer until end of transfer being executed |

- Bits 3-0—Reserved: They have no effect on DTC operation.


### 8.2.2 DTC Source Address Register (DTSAR)

The DTC source address register (DTSAR) is a 32-bit register that specifies the DTC transfer source address. An even address indicates that the transfer size is word; a multiple-of-four address means it is longword. The contents of this register is located in memory.

> Initial value:


Note: * Initial value is undefined.

### 8.2.3 DTC Destination Address Register (DTDAR)

The DTC destination address register (DTDAR) is a 32-bit register that specifies the DTC transfer destination address. An even address indicates that the transfer size is word; a multiple-of-four address means it is longword. The contents of this register are located in memory.

| Bit: | 31 | 30 | 29 | 28 | 27 |  | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  | . |  |  |  |  |  |
| Initial value: | * | * | * | * | * |  | * | * | * | * | * |
| R/W: | - | - | - | - | - |  | - | - | - | - | - |

Note: * Initial value is undefined.

### 8.2.4 DTC Initial Address Register (DTIAR)

The DTC initial address register (DTIAR) specifies the initial transfer source/transfer destination address in repeat mode. In repeat mode, when the DTS bit is set to 1 , specify the initial transfer source address in the repeat area, and when the DTS bit is cleared to 0 , specify the initial transfer destination address in the repeat area.

The contents of this register are located in memory.


Note: * Initial value is undefined.

### 8.2.5 DTC Transfer Count Register A (DTCRA)

DTCRA is a 16-bit register that specifies the number of DTC transfers. The contents of this register are located in memory.

In normal mode it functions as a 16 -bit transfer counter. The number of transfers is 1 when the set value is $\mathrm{H}^{\prime} 0001,65535$ when it is $\mathrm{H}^{\prime} \mathrm{FFFF}$, and 65536 when it is $\mathrm{H}^{\prime} 0000$.

In repeat mode, DTCRAH maintains the transfer count and DTCRAL functions as an 8-bit transfer counter. The number of transfers is 1 when the set value is DTCRAH $=$ DTCRAL $=H^{\prime} 01$, 255 when they are H'FF, and 256 when it is $\mathrm{H}^{\prime} 00$.

In block transfer mode it functions as a 16-bit transfer counter. The number of transfers is 1 when the set value is $\mathrm{H}^{\prime} 0001,65535$ when it is $\mathrm{H}^{\prime} \mathrm{FFFF}$, and 65536 when it is $\mathrm{H}^{\prime} 0000$.


### 8.2.6 DTC Transfer Count Register B (DTCRB)

The DTCRB is a 16-bit register that designates the block length in block transfer mode. The contents of this register is located in memory. The block length is 1 when the set value is $\mathrm{H}^{\prime} 0001$, 65535 when it is H'FFFF, and 65536 when it is $\mathrm{H}^{\prime} 0000$.


Note: * Initial value is undefined.

### 8.2.7 DTC Enable Registers (DTER)

The DTER (DTEA-DTEE) are five 8-bit readable/writable registers with bits allocated to each interrupt source that activates the DTC. They set disable/enable for DTC activation for each interrupt source. When a bit is 1, DTC activation by the corresponding interrupt source is enabled. Interrupt sources for each of the DTEA-DTEE registers are indicated in table 8.2.

The DTER are initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset or in standby mode. Manual reset does not initialize DTER.

For the A mask, overwrite this register as follows:
When clearing bit to 0 : read the 1 bit to clear and write 0 .
When setting bit to 1 : read the 0 bit to set and write 1 .

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DTE7 | DTE6 | DTE5 | DTE4 | DTE3 | DTE2 | DTE1 | DTE0 |
|  | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Initial value: | 0 | 0 | 0 |  |  |  |  |  |
| R/W: | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* | R/W* |

Note: * DTER bits can only be modified by writing 1 after reading 0 , or writing 0 after reading 1 .

### 8.2.8 DTC Control/Status Register (DTCSR)

The DTCSR is a 16-bit readable/writable register that sets disable/enable for DTC activation by software, as well as the DTC vector addresses for software activation. It also indicates the DTC transfer status.

The DTCSR is initialized to $\mathrm{H}^{\prime} 0000$ by power-on resets and in standby mode. Manual reset does not initialize DTCSR.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | NMIF | AE | SWDTE |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | R | R | R | $\mathrm{R} / \mathrm{W}^{* 1}$ | $\mathrm{R} / \mathrm{W}^{* 1}$ | $\mathrm{R} / \mathrm{W}^{* 2}$ |


| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bit name: | DTVEC7 | DTVEC6 | DTVEC5 | DTVEC4 | DTVEC3 | DTVEC2 | DTVEC1 | DTVECO |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W*3 | R/W*3 | R/W*3 | R/W*3 | R/W*3 | R/W*3 | R/W ${ }^{* 3}$ | $\mathrm{R} / \mathrm{W}^{* 3 * 4}$ |

Notes: *1 For the NMIF and AE bits, only a 0 write after a 1 read is possible.
*2 For the SWDTE bit, a 1 write is always possible, but a 0 write is possible only after a 1 is read.
*3 For the DTVEC7-DTVEC0 bits, writes are possible only when SWDTE $=0$.

* 4 Be sure to write 0 to the DTVEC0 bit.

Bits $15-11$-Reserved: These bits always read as 0 . The write value should always be 0 .

- Bit 10—NMI Flag Bit (NMIF): Indicates that an NMI interrupt has occurred. When the NMIF bit is set, DTC transfers are not allowed even if the DTER bit is set to 1 . If, however, a transfer has already started with the NMIM bit of the DTMR set to 1 , execution will continue until that transfer ends. To clear the NMIF bit, read the 1 from it, then write a 0 .

The NMIF bit is initialized to 0 by power-on resets and in standby mode.

| Bit $\mathbf{1 0}$ (NMIF) | Description |
| :--- | :--- |
| 0 | No NMI interrupts (initial value) |
|  | (Clear condition) Write a 0 after reading the NMIF bit |
| 1 | NMI interrupt has been generated |

- Bit 9—Address Error Flag (AE): Indicates that an address error by the DTC has occurred. When the AE bit is set, DTC transfers are not allowed even if the DTER bit is set to 1 . To clear the AE bit, read the 1 from it, then write a 0 .

The AE bit is initialized to 0 by power-on resets and in standby mode.

| Bit 9 (AE) | Description |
| :--- | :--- |
| 0 | No address error by the DTC (initial value) <br> (Clear condition) Write a 0 after reading the AE bit |
| 1 | An address error by the DTC occurred |

- Bit 8—DTC Software Activation Enable Bit (SWDTE): This bit enables/disables DTC activation by software.

The AE bit is initialized to 0 by resets and standby mode. For details, see section 8.3.2, Activating Sources.

| Bit 8 (SWDTE) | Description |
| :--- | :--- |
| 0 | DTC activation by software disabled (initial value) |
| 1 | DTC activation by software enabled |

- Bits 7-0—Software Activation Vectors 7-0 (DTVEC7-DTVEC0): These bits set the DTC vector addresses for DTC activation by software. A vector address is calculated as $\mathrm{H}^{\prime} 0400$ + DTVEC[7:0]. Always specify 0 for DTVEC0. 8 bits are available, so you can specify values H'00 (0)-H'FE (254).


### 8.2.9 DTC Information Base Register (DTBR)

The DTBR is a 16-bit readable/writable register that specifies the upper 16 bits of the memory address containing DTC transfer information. Always access the DTBR in word or longword units. If it is accessed in byte units the register contents will become undefined at the time of a write, and undefined values will be read out upon reads.

The DTBR is not initialized either by resets or in standby mode.


Note: * Initial value is undefined.

### 8.3 Operation

The DTC stores transfer information in memory. When there are DTC transfer requests, it reads that transfer information and performs data transfers based on it. It rewrites the transfer information to memory after data transfers. Storing transfer information in memory makes it possible to perform data transfers for an arbitrary number of channels. Further, setting the CHNE bit to 1 makes it possible to perform multiple transfers continuously through one DTC transfer request.

There are three DTC transfer modes: normal mode, repeat mode, and block transfer mode. After a DTC transfer, the transfer source address and transfer destination address are incremented, decremented, or kept the same, according to the respective setting.

### 8.3.1 Overview of Operation

Figure 8.2 shows a flowchart of DTC operation.


Figure 8.2 DTC Operation Flowchart

### 8.3.2 Activating Sources

The DTC performs write operations to the DTCSR with either interrupt sources or software as its activating sources. Each interrupt source is designated by specific DTER bits to determine whether it becomes an interrupt request to the CPU or a DTC activating source.

When the DISEL bit is 1 , an interrupt, established as the DTC activating source, is requested of the CPU after each data transfer in DRC. When the DISEL bit is a 0 , a request is made only after the completion of a designated number of data transfers. When the activating source interrupt is requested of the CPU, the corresponding DTER bit is automatically cleared.

In the case of software activation also, when the DISEL bit is a 1 , a software DTC activation interrupt (SWDTCE) is requested of the CPU after each data transfer. When the DISEL bit is a 0 , a request is made only after the completion of a designated number of data transfers. When no SWDTCE interrupt is requested of the CPU, the SWDTE bit of the DTCSR is automatically cleared. When a request is made of the CPU, the SWDTE bit is maintained as a 1.

When multiple DTC activating sources occur simultaneously, they are accepted and the DTC is activated in accordance with the default priority rankings shown in table 8.2.

Figure 8.3 shows a block diagram of activating source control.


Figure 8.3 Activating Source Control Block Diagram

### 8.3.3 DTC Vector Table

Figure 8.4 shows the correspondence between DTC vector addresses and register information placement. For each DTC activating source there are 2 bytes in the DTC vector table, which contain the register information start address.

Table 8.2 shows the correspondence between activating sources and vector addresses. When activating with software, the vector address is calculated as $\mathrm{H}^{\prime} 0400+$ DTVEC[7:0].

Through DTC activation, a register information start address is read from the vector table, then register information placed in memory space is read from that register information start address. Always designate register information start addresses in multiples of four.


Figure 8.4 Correspondence between DTC Vector Address and Register Information

Table 8.2 Interrupt Sources and DTC Vector Addresses

| Source <br> Generator | Activating <br> Source | DTC Vector Address | DTE | Transfer <br> Bit | Transfer <br> Source |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Destination |  |  |  |  |  | Priority.

Table 8.2 Interrupt Sources and DTC Vector Addresses (cont)

| Source Generator | Activating Source | DTC Vector Address | DTE <br> Bit | Transfer Source | Transfer Destination | Priority |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SCIO | RXIO | H'00000438-H'00000439 | DTED3 | RDR0 | Arbitrary*1 | High |
|  | TXIO | H'0000043A-H'0000043B | DTED2 | Arbitrary*1 | TDR0 |  |
| SCI1 | RXI1 | $H^{\prime} 0000043 \mathrm{C}-\mathrm{H}^{\prime} 0000043 \mathrm{D}$ | DTED1 | RDR1 | Arbitrary*1 |  |
|  | TXI1 | H'0000043E-H'0000043F | DTED0 | Arbitrary*1 | TDR1 |  |
| BSC | CMI | H'00000440-H'00000441 | DTEE7 | Arbitrary*1 | Arbitrary*1 |  |
| Software | Write to DTCSR | $\begin{aligned} & \text { H'00000400 + DTVEC[7:0] } \\ & \text { H'00000401 + DTVEC[7:0] } \end{aligned}$ | - | Arbitrary* | Arbitrary*1 | Low |

Notes: *1 External memory, memory-mapped external devices, on-chip memory, on-chip peripheral modules (excluding DMAC and DTC)
*2 Excluding A mask products are ADI, A mask products are ADIO.

### 8.3.4 Register Information Placement

Figure 8.5 shows the placement of register information in memory space. The register information start addresses are designated by DTBR for the upper 16 bits, and the DTC vector table for the lower 16 bits.

The placement in order from the register information start address in normal mode is DTMR, DTCRA, 4 bytes empty (no effect on DTC operation), DTSAR, then DTDAR. In repeat mode it is DTMR, DTCRA, DTIAR, DTSAR, and DTDAR. In block transfer mode, it is DTMR, DTCRA, 2 bytes empty (no effect on DTC operation), DTCRB, DTSAR, then DTDAR.

Fundamentally, certain RAM areas are designated for addresses storing register information.


Figure 8.5 DTC Register Information Placement in Memory Space

### 8.3.5 Normal Mode

Performs the transfer of one byte, one word, or one longword for each activation. The total transfer count is 1 to 65536. An interrupt request is generated to the CPU when the transfer with DTCRA $=1$ ends. Transfers of a number of bytes specified by the SCI are possible.

Table 8.3 shows the register functions for normal mode.
Table 8.3 Normal Mode Register Functions

|  |  | Values Written Back upon a <br> Transfer Information Write |  |
| :--- | :--- | :--- | :--- |
| Register | Function | When DTCRA <br> is other than 1 | When DTCRA is 1 |
| DTMR | Operation mode <br> control | DTMR | DTMR |
| DTCRA | Transfer count | DTCRA - 1 | DTCRA - 1 (= H'0000) |
| DTSAR | Transfer source <br> address | Increment/decrement// <br> fixed | Increment/decrement/ <br> fixed |
| DTDAR | Transfer destination <br> address | Increment/decrement// <br> fixed | Increment/decrement/ <br> fixed |

### 8.3.6 Repeat Mode

Performs the transfer of one byte, one word, or one longword for each activation. Either the transfer source or transfer destination is designated as the repeat area.

## Renesns

The total transfer count is specified between 1 and 256 . When the specified number of transfers ends, the address register of the designated repeat area is returned to its initial state and the transfer is repeated. Other address registers are consecutively incremented, decremented, or remain fixed. While DISEL $=0$, no interrupt request is made to the CPU, even if the transfer with DTCRAL $=1$ ends.

Pulses for driving the stepping motor can be output. Table 8.4 shows the register functions for repeat mode.

Table 8.4 Repeat Mode Register Functions

|  |  | Values Written Back upon a <br> Transfer Information Write |  |
| :--- | :--- | :--- | :--- |
| Register | Function | When DTCRA is <br> other than $\mathbf{1}$ | When DTCRA is 1 |
| DTMR | Operation mode <br> control | DTMR | DTMR |
| DTCRAH | Transfer count <br> maintenance | DTCRAH | DTCRAH |
| DTCRAL | Transfer count | DTCRAL -1 | DTCRAH |
| DTIAR | Initial address | (Not written back) | (Not written back) |
| DTSAR | Transfer source <br> address | Increment/decrement/ <br> fixed | (DTS = 0) Increment/ <br> decrement/ fixed |
| DTDAR | Transfer destination <br> address | Increment/decrement/ <br> fixed | (DTS = 1) DTIAR |

### 8.3.7 Block Transfer Mode

Performs the transfer of one block for each one activation. Either the transfer source or transfer destination is designated as the block area.

The block length is specified between 1 and 65536. When a 1-block transfers ends, the address register of the designated block area is returned to its initial state. Other address registers are consecutively incremented, decremented, or remain fixed. The block transfer count is 1 to 65536. An interrupt request is generated to the CPU when the transfer with DTCRA $=1$ ends.

A/D converter group mode transfers and phase compensation PWM data transfers are possible.
Table 8.5 shows the register functions for block transfer mode.

Table 8.5 Block Transfer Mode Register Functions

| Register | Function | Values Written Back upon a <br> Transfer Information Write |
| :--- | :--- | :--- |
| DTMR | Operation mode <br> control | DTMR |
| DTCRA | Transfer count | DTCRA -1 |
| DTCRB | Block length | $($ Not written back $)$ |
| DTSAR | Transfer source <br> address | $($ DTS $=0)$ Increment/ decrement/ fixed <br>  <br>  <br> DTDAR $=1)$ DTSAR initial value |

### 8.3.8 Operation Timing

Figure 8.6 shows a DTC operation timing example.


Figure 8.6 DTC Operation Timing Example (Normal Mode)
When register information is located in on-chip RAM, each mode requires 4 cycles for transfer information reads, and 3 cycles for writes.

### 8.3.9 DTC Execution State Counts

Table 8.6 shows the execution state for one DTC data transfer. Furthermore, table 8.7 shows the state counts needed for execution state.

Table 8.6 Execution State of DTC

| Mode | Vector Read I | Register <br> Information <br> Read/Write J | Data Read K | Data Write L | Internal <br> Operation M |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Normal | 1 | 7 | 1 | 1 | 1 |
| Repeat | 1 | 7 | 1 | 1 | 1 |
| Block transfer | 1 | 7 | N | N | 1 |

Note: N: block size (default set values of DTCRB)

Table 8.7 State Counts Needed for Execution State

| Access Objective |  |  | Onchip RAM | Onchip ROM | Internal I/O <br> Register |  | External Device |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Bus width |  |  | 32 | 32 | 32 |  | 8 | 16 | 32 |
| Access state |  |  | 1 | 1 | $2^{* 1}$ | 3*2 | 2 | 2 | 2 |
| Execution state | Vector read | S | - | 1 | - |  | 4 | 2 | 2 |
|  | Register information read/write | S | 1 | 1 | - |  | 8 | 4 | 2 |
|  | Byte data read | $\mathrm{S}_{\mathrm{K}}$ | 1 | 1 | 2 | 3 | 2 | 2 | 2 |
|  | Word data read | $\mathrm{S}_{\mathrm{K}}$ | 1 | 1 | 2 | 3 | 4 | 2 | 2 |
|  | Long word data read | $\mathrm{S}_{\mathrm{K}}$ | 1 | 1 | 4 | 6 | 8 | 4 | 2 |
|  | Byte data write | $\mathrm{S}_{\llcorner }$ | 1 | 1 | 2 | 3 | 2 | 2 | 2 |
|  | Word data write | $\mathrm{S}_{\llcorner }$ | 1 | 1 | 2 | 3 | 4 | 2 | 2 |
|  | Long word write | S | 1 | 1 | 4 | 6 | 8 | 4 | 2 |
|  | Internal operation | $\mathrm{S}_{\mathrm{M}}$ | 1 |  |  |  |  |  |  |

Notes: *1 Two state access module : port, INT, CMT, SCI, etc.
*2 Three state access module : WDT, CACHE, UBC, etc.

The execution state count is calculated using the following formula. $\sum$ indicates the number of transfers by one activating source (count +1 when CHNE bit is 1 ).

Execution state count $=\mathrm{I} \cdot \mathrm{S}_{\mathrm{I}}+\sum\left(\mathrm{J} \cdot \mathrm{S}_{\mathrm{J}}+\mathrm{K} \cdot \mathrm{S}_{\mathrm{K}}+\mathrm{L} \cdot \mathrm{S}_{\mathrm{L}}\right)+\mathrm{M} \cdot \mathrm{S}_{\mathrm{M}}$

### 8.3.10 DTC Usage Procedure

The procedure for DTC interrupt activation is as follows:

1. Transfer data (DTMR, DTCRA, DTSAR, DTDAR, DTCRB, and DTIAR) is located in memory space.
2. Establish the register information start address with DTBR and the DTC vector table.
3. Set the corresponding DTER bit to 1 .
4. The DTC is activated when an interrupt source occurs.
5. When interrupt requests are not made to the CPU, the interrupt source is cleared, but the DTER is not. When interrupts are requested, the interrupt source is not cleared, but the DTER is.
6. Interrupt sources are cleared within the CPU interrupt routine. When doing continuous DTC data transfers, set the DTER to 1 .

The procedure for DTC software activation is as follows:

1. Transfer data (DTMR, DTCRA, DTSAR, DTDAR, DTCRB, and DTIAR) is located in memory space.
2. Establish the register information start address with DTBR and the DTC vector table.
3. Confirm that the SWDTE bit of the DTCSR is 0 . When the SWDTE bit is 1 , the DTC is already being driven by software.
4. Write a 1 to the SWDTE bit and a vector number to the DTVEC (byte data).
5. When SWDTCE interrupt requests are not made to the CPU, the SWDTE bit is cleared. When interrupts are requested, the SWDTE bit is maintained as a 1.
6. The SWDTE bit is cleared to 0 within the CPU interrupt routine. For continuous DTC data transfers, set the SWDTE to 1.

### 8.3.11 DTC Use Example

The following is a DTC use example of a 128 -byte data reception by the SCI:

1. The settings are: DTMR source address fixed $(\mathrm{SM} 1=0)$, destination address incremented ( $\mathrm{DM} 1=1, \mathrm{DM} 0=0$ ), normal mode $(\mathrm{MD} 1=\mathrm{MD} 0=0)$, byte size $(\mathrm{SZ1}=\mathrm{SZ} 0=0)$, one transfer per activating source $(\mathrm{CHNE}=0)$, and a CPU interrupt request after the designated number of data transfers (DISEL = 0). 128 (H'0080) is set in DTCRA, the RDR address of the SCI is set in DTSAR, and the start address of the RAM storing the receive data is set in DTDAR.
2. Establish the register information start address with DTBR and the DTC vector table.
3. Set the corresponding DTER bit to 1 .
4. Set the SCI to a specific receive mode and enable RxI interrupts.
5. The RDRF flag of the SSR is set to 1 by each completion of a 1-byte data reception by the SCI, an RxI interrupt is generated, and the DTC is activated. The received data is transferred from RDR to RAM by the DTC, and the RDRF flag is 0 cleared.
6. After completion of 128 data transfers (DTCRA $=0$ ), the DTER is cleared while the RDRF is maintained as 1 , and an RxI interrupt request is made to the CPU. The interrupt processing routine clears the RDRF, and performs the other completion processing.

### 8.4 Cautions on Use

- DMAC and DTC register access by the DTC is prohibited.
- DTC register access by the DMAC is prohibited.
- When setting a bit in DTER, first ensure that all transfers on the DTC channel corresponding to that DTER have ended, or disable the transfer source for each channel so that DTC transfer corresponding to that DTER will not occur. The above restrictions do not apply for A mask due to change in the access method of DTER. However, take caution when changing LSI to A mask, since modification of the program is required.


## Section 9 Cache Memory (CAC)

### 9.1 Overview

The LSI has an on-chip cache memory (CAC) with 1 kbyte of cache data and a 256 -entry cache tag. The cache data and cache tag space can be used as on-chip RAM space when the cache is not being used.

### 9.1.1 Features

The CAC has the following features. The cache tag and cache data configuration is shown in figure 9.1.

- 1-kbyte capacity
- External memory (CS space and DRAM space) instruction code and PC relative data caching
- 256 entry cache tag (tag address 15 bits)
- 4-byte line length
- Direct map replacement algorithm
- Valid flag (1 bit) included for purges


Figure 9.1 Cache Tag and Cache Data Configuration

### 9.1.2 Block Diagram

Figure 9.2 shows a block diagram of the cache.


Figure 9.2 Cache Block Diagram

### 9.1.3 Register Configuration

The cache has one register, which can be used to control the enabling or disabling of each cache space. The register configuration is shown in table 9.1.

Table 9.1 Register Configuration

| Name | Abbreviation | R/W | Initial <br> Value | Address | Access Size <br> (Bits) |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Cache control register | CCR | R/W | H'0000* $^{*}$ | H'FFFF8740 $^{\prime}$ | 8, 16, 32 |
| Note: $*$ Bits 15-5 are undefined. |  |  |  |  |  |

Note: * Bits 15-5 are undefined.

### 9.2 Register Explanation

### 9.2.1 Cache Control Register (CCR)

The cache control register (CCR) selects the cache enable/disable of each space.
The CCR is a 16-bit readable/writable register. It is initialized to H'0000 by power on resets, but is not initialized by manual resets or standby mode.


Note: * Bits 15-5 are undefined.

- Bits 15-5—Reserved: Reading these bits gives undefined values. The write value should always be 0 .
- Bit 4—DRAM Space Cache Enable (CEDRAM): Selects whether to use DRAM space as a cache object (enable) or to exclude it (disable). A 0 disables, and a 1 enables such use.


## Bit 4 (CEDRAM) Description

| 0 | DRAM space cache disabled (initial value) |
| :--- | :--- |
| 1 | DRAM space cache enabled |

- Bit 3-CS3 Space Cache Enable (CECS3): Selects whether to use CS3 space as a cache object (enable) or to exclude it (disable). A 0 disables, and a 1 enables such use.

| Bit 3 (CECS3) | Description |
| :--- | :--- |
| 0 | CS3 space cache disabled (initial value) |
| 1 | CS3 space cache enabled |

- Bit 2—CS2 Space Cache Enable (CECS2): Selects whether to use CS2 space as a cache object (enable) or to exclude it (disable). A 0 disables, and a 1 enables such use.

| Bit 2 (CECS2) | Description |
| :--- | :--- |
| 0 | CS2 space cache disabled (initial value) |
| 1 | CS2 space cache enabled |

- Bit 1—CS1 Space Cache Enable (CECS1): Selects whether to use CS1 space as a cache object (enable) or to exclude it (disable). A 0 disables, and a 1 enables such use.


## Bit 1 (CECS1) Description

| 0 | CS1 space cache disabled (initial value) |
| :--- | :--- |
| 1 | CS1 space cache enabled |

- Bit 0-CS0 Space Cache Enable (CECS0): Selects whether to use CS0 space as a cache object (enable) or to exclude it (disable). A 0 disables, and a 1 enables such use.

Bit 0 (CECSO) Description

| 0 | CSO space cache disabled (initial value) |
| :--- | :--- |
| 1 | CSO space cache enabled |

### 9.3 Address Array and Data Array

There is a special cache space for controlling the cache. This space is divided into an address array and a data array, where addresses (tag address, including valid bit) and data (4-byte line length) for cache control are recorded. The special cache space is shown in table 9.2. It can be used as on-chip RAM space when the cache is not being used.

Table 9.2 Special Cache Space

| Space Classification | Address | Size | Bus Width |
| :--- | :--- | :--- | :--- |
| Address array | H'FFFFF000-H'FFFFF3FF | 1 kbyte | 32 bit |
| Data array | H'FFFFF400-H'FFFFF7FF | 1 kbyte | 32 bit |

### 9.3.1 Cache Address Array Read/Write Space

The cache address array has a compulsory read/write (figure 9.3).


Figure 9.3 Cache Address Array
Address Array Read: Designates entry address and reads out the corresponding tag address value/valid bit value.

Address Array Write: Designates entry address and writes the designated tag address value/valid bit value.

### 9.3.2 Cache Data Array Read/Write Space

The cache data array has a compulsory read/write (figure 9.4).


Figure 9.4 Cache Data Array
Data Array Read: Designates entry address and reads out the corresponding line of data.
Data Array Write: Designates entry address and writes designated data to the corresponding line.

### 9.4 Cautions on Use

### 9.4.1 Cache Initialization

Always initialize the cache before enabling it. Specifically, use an address array write to write 0 to all valid bits for all entries ( 256 times), that is,those in the address range H'FFFFF000H'FFFFF3FF.

Writes to the address array or data array by the CPU, DMAC, or DTC are not possible while the cache is enabled. For reads, undefined values will be read out.

### 9.4.2 Forced Access to Address Array and Data Array

While the cache is enabled, it is not possible to write to the address array or data array via the CPU, DMAC, or DTC, and a read will return an undefined value. The cache must be disabled before making a forced access to the address array or data array.

### 9.4.3 Cache Miss Penalty and Cache Fill Timing

When a cache miss occurs, a single idle cycle is generated as a penalty immediately before the cache fill (access from external memory in the event of a cache miss), as shown in figure 9.5. However, in the case of consecutive cache misses, idle cycles are not generated for the second and subsequent cache misses, as shown in figure 9.6.

As the timing for a cache fill from normal space, the CS assert period immediately before the end of the bus cycle (or the last bus cycle when two or four bus cycles are generated, such as in a word access to 8 -bit space) is extended by an additional cycle, as shown in figures 9.5 and 9.6.

Similarly, as the timing for a cache fill from DRAM space, the RAS assert period immediately before the end of the bus cycle is extended by an additional cycle. In RAS down mode, the next bus cycle is delayed by one cycle as shown in figure 9.8.


Figure 9.5 Cache Fill Timing in Case of Non-Consecutive Cache Miss from Normal Space (No Wait, No CS Assert Extension)


Figure 9.6 Cache Fill Timing in Case of Consecutive Cache Misses from Normal Space (No Wait, CS Assert Extension)


Figure 9.7 Cache Fill Timing in Case of Non-Consecutive Cache Miss from DRAM Space (Normal Mode, TPC = 0, RCD = 0, No Wait)


Figure 9.8 Cache Fill Timing in Case of Consecutive Cache Misses from DRAM Space (RAS Down Mode, $\mathbf{T P C}=\mathbf{0}, \mathbf{R C D}=\mathbf{0}$, No Wait)

### 9.4.4 Cache Hit after Cache Miss

The first cache hit after a cache miss is regarded as a cache miss, and a cache fill without idle cycle generation is performed. The next hit operates as a cache hit.

## Section 10 Bus State Controller (BSC)

### 10.1 Overview

The bus state controller (BSC) divides up the address spaces and outputs control for various types of memory. This enables memories like DRAM, SRAM, and ROM to be linked directly to the LSI without external circuitry.

### 10.1.1 Features

The BSC has the following features:

- Address space is divided into five spaces
- A maximum linear 2 Mbytes for on-chip ROM effective mode, and a maximum linear 4-Mbyte for on-chip ROM ineffective mode for address space CS0
- A maximum linear 4 Mbytes for each of the address spaces CS1-CS3
- A maximum linear 16 Mbytes for DRAM dedicated space
- Bus width can be selected for each space ( 8,16 , or 32 bits)
- Wait states can be inserted by software for each space
- Wait states can be inserted via the $\overline{\text { WAIT }}$ pin in external memory spce accesses.
- Outputs control signals for each space according to the type of memory connected
- On-chip ROM and RAM interfaces
- On-chip RAM access of 32 bits in 1 state
- On-chip ROM access of 32 bits in 1 state
- Direct interface to DRAM
— Multiplexes row/column addresses according to DRAM capacity
- Supports high-speed page mode and RAS down mode
- Access control for each type of memory, peripheral LSI
- Address/data multiplex function
- Refresh
- Supports CAS-before-RAS refresh (auto-refresh) and self-refresh
- Refresh counter can be used as an interval timer
- Interrupt request generated upon compare match (CMI interrupt request signal)


### 10.1.2 Block Diagram

Figure 10.1 shows the BSC block diagram.


Figure 10.1 BSC Block Diagram

### 10.1.3 Pin Configuration

Table 10.1 shows the bus state controller pin configuration.
Table 10.1 Pin Configuration

| Signal | I/O | Description |
| :--- | :--- | :--- |
| A21-A0 | O | Address output (A21-A18 will become input ports with power-on reset) |
| D31-D0 | I/O | 32-bit data bus. D15-D0 are address output and data I/O during address/data <br> multiplex I/O. |
| $\overline{\overline{\text { CSO }}-}$ | O | Chip select |
| $\overline{\text { CS3 }}$ | O | Strobe that indicates the read cycle for ordinary space/multiplex I/O. Also <br> output during DRAM access. |
| $\overline{\text { RD }}$ | O | Strobe that indicates a write cycle to the most significant byte (D31-D24) for <br> ordinary space/multiplex I/O. Also output during DRAM access. |
| $\overline{\text { WRHH }}$ | Strobe that indicates a write cycle to the 2nd byte (D23-D16) for ordinary <br> space/multiplex I/O. Also output during DRAM access. |  |
| $\overline{\text { WRHL }}$ | OStrobe that indicates a write cycle to the 3rd byte (D15-D8) for ordinary <br> space/multiplex I/O. Also output during DRAM access. |  |
| $\overline{\text { WRH }}$ | Strobe that indicates a write cycle to the least significant byte (D7-D0) for <br> ordinary space/multiplex I/O. Also output during DRAM access. |  |
| $\overline{\text { WRL }}$ | O | Strobe indicating a write cycle to DRAM (used for DRAM space) |
| $\overline{\text { RDWR }}$ | O | RAS signal for DRAM (used for DRAM space) |
| $\overline{\text { RAS }}$ | CAS signal when accessing the most significant byte (D31-D24) of DRAM <br> (used for DRAM space) |  |
| $\overline{\text { CASHH }}$ | O |  |
| $\overline{\text { CASHL }}$ | O | CAS signal when accessing the 2nd byte (D23-D16) of DRAM (used for DRAM <br> space) |
| $\overline{\text { CASH }}$ | O | CAS signal when accessing the 3rd byte (D15-D8) of DRAM (used for DRAM <br> space) |
| $\overline{\text { CASL }}$ | O | CAS signal when accessing the least significant byte (D7-D0) of DRAM (used <br> for DRAM space) |
| $\overline{\text { AH }}$ | O | Signal to hold the address during address/data multiplex |
| $\overline{\text { WAIT }}$ | I | Wait state request signal |
| $\overline{\text { BREQ }}$ | I | Bus release request input |
| Bus use enable output |  |  |

## Renesns

### 10.1.4 Register Configuration

The BSC has eight registers. These registers are used to control wait states, bus width, and interfaces with memories like DRAM, ROM, and SRAM, as well as refresh control. The register configurations are listed in table 10.2.

All registers are 16 bits. Do not access DRAM space before completing the memory interface settings. All BSC registers are all initialized by a power-on reset, but are not by a manual reset. Values are maintained in standby mode

Table 10.2 Register Configuration

| Name | Abbr. | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Bus control register 1 | BCR1 | R/W | H'200F | H'FFFF8620 | 8, 16, 32 |
| Bus control register 2 | BCR2 | R/W | H'FFFF | H'FFFF8622 | 8, 16, 32 |
| Wait state control register 1 | WCR1 | R/W | H'FFFF | H'FFFF8624 | 8, 16, 32 |
| Wait state control register 2 | WCR2 | R/W | H'000F | H'FFFF8626 | 8, 16, 32 |
| DRAM area control register | DCR | R/W | H'0000 | H'FFFF862A | 8, 16, 32 |
| Refresh timer control/status register | RTCSR | R/W | H'0000 | H'FFFF862C | 8, 16, 32 |
| Refresh timer counter | RTCNT | R/W | H'0000 | H'FFFF862E | 8, 16, 32 |
| Refresh time constant register | RTCOR | R/W | H'0000 | H'FFFF8630 | 8, 16, 32 |

### 10.1.5 Address Map

Figure 10.2 shows the address format used by the SH7040 Series.


Figure 10.2 Address Format
This LSI uses 32-bit addresses:

- A31-A24 are used to select the type of space and are not output externally.
- Bits A23 and A22 are decoded and output as chip select signals ( $\overline{\mathrm{CS} 0}-\overline{\mathrm{CS} 3})$ for the corresponding areas when bits A31-A24 are 00000000.
- A21-A0 are output externally.

Table 10.3 shows an address map for on-chip ROM effective mode. Table 10.4 shows an address map for on-chip ROM ineffective mode.

Table 10.3 Address Map for On-Chip ROM Effective Mode

| Address | Space | Memory | Size | Bus Width |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{H}^{\prime} 00000000-\mathrm{H}^{\prime} 0003 F F F F^{* 1}$ | On-chip ROM | On-chip ROM memory | 256 kbytes | 32 bits |
| H'00040000-H'001FFFFF | Reserved | Reserved |  |  |
| H'00200000-H'003FFFFF | CSO space | Ordinary space | 2 Mbytes | 8/16/32 bits $^{* 2}$ |
| H'00400000-H'007FFFFF | CS1 space | Ordinary space | 4 Mbytes | 8/16/32 bits $^{* 2}$ |
| H'00800000-H'00BFFFFF | CS2 space | Ordinary space | 4 Mbytes | 8/16/32 bits $^{* 2}$ |
| H'00C00000-H'00FFFFFF | CS3 space | Ordinary space or multiplex I/O space | 4 Mbytes | 8/16/32 bits*3 |
| H'01000000-H'01FFFFFF | DRAM space | DRAM | 16 Mbytes | 8/16/32 bits $^{* 2}$ |
| H'02000000-H'FFFF7FFF | Reserved | Reserved |  |  |
| H'FFFF8000-H'FFFF87FF | On-chip peripheral module | On-chip peripheral module | 2 kbytes | 8/16 bits |
| H'FFFF8800-H'FFFFEFFF | Reserved | Reserved |  |  |
| H'FFFFF000-H'FFFFFFFF | On-chip RAM | On-chip RAM | 4 kbytes | 32 bits |

Notes: Do not access reserved spaces. Operation cannot be guaranteed if they are accessed.
*1 With the 64-kbyte version of on-chip ROM, the ROM address is H'0000000$H^{\prime} 0000 F F F F$, and address H'00010000-H'0003FFFF is reserved space. With the 128 -kbyte version of on-chip ROM, the ROM address is H'00000000H'0001FFFF, and address H'00020000-H'0003FFFF is reserved space.
*2 Selected by on-chip register settings.
*3 Ordinary space: selected by on-chip register settings.
Multiplex I/O space: 8/16 bit selected by the A14 bit.

Table 10.4 Address Map for On-Chip ROM Ineffective Mode

| Address | Space | Memory | Size | Bus Width |
| :---: | :---: | :---: | :---: | :---: |
| H'00000000-H'003FFFFF | CS0 space | Ordinary space | 4 Mbytes | 8/16/32 bist $^{* 1}$ |
| H'00400000-H'007FFFFF | CS1 space | Ordinary space | 4 Mbytes | 8/16/32 bits*2 |
| H'00800000-H'00BFFFFFF | CS2 space | Ordinary space | 4 Mbytes | 8/16/32 bits*2 |
| H'00C00000-H'00FFFFFF | CS3 space | Ordinary space or multiplex I/O space | 4 Mbytes | 8/16/32 bits* $^{* 3}$ |
| H'01000000-H'01FFFFFF | DRAM space | DRAM | 16 Mby | 8/16/32 bits $^{* 2}$ |
| H'02000000-H'FFFF7FFF | Reserved | Reserved |  |  |
| H'FFFF8000-H'FFFF87FF | On-chip peripheral module | On-chip peripheral module | 2 kbytes | 8/16 bits |
| H'FFFF8800-H'FFFFEEFFF | Reserved | Reserved |  |  |
| H'FFFFF000-H'FFFFFFFFF | On-chip RAM | On-chip RAM | 4 kbytes | 32 bits |

Notes: 1. Do not access reserved spaces. Operation cannot be guaranteed if they are accessed.
2. In the single-chip mode, spaces other than on-chip ROM, on-chip RAM and on-chip peripheral modules are unavailable.
*1 Selected by the mode pin:
8/16 bit when 112 pin and 120 pin.
$16 / 32$ bit when 144 pin.
*2 Selected by on-chip register settings.
*3 Ordinary space: selected by on-chip register settings.
Multiplex I/O space: $8 / 16$ bit selected by the A14 bit.

### 10.2 Description of Registers

### 10.2.1 Bus Control Register 1 (BCR1)

BCR1 is a 16-bit read/write register that enables access to the MTU control register, selects multiplex I/O, and specifies the bus size of the CS spaces. With the 112-pin version (SH7040/SH7042/SH7044), and the 120-pin version (SH7040/SH7042), specify the bus width as word (16 bits) or less.

Write bits 8-0 of BCR1 during the initialization stage after a power-on reset, and do not change the values thereafter. In on-chip ROM effective mode, do not access any of the CS spaces until after completion of register initialization. In on-chip ROM ineffective mode, do not access any CS space other than CS0 until after completion of register initialization.

BCR1 is initialized by power-on resets to $\mathrm{H}^{\prime} 200 \mathrm{~F}$, but is not initialized by manual resets or software standbys.

## Renesns



- Bits $15,14,12-9 —$ Reserved: These bits always read as 0 . The write value should always be 0 .
- Bit 13—MTU Read/Write Enable (MTURWE): When this bit is 1, MTU control register access is enabled. See section 12, Multifunction Timer Pulse Unit (MTU), for details.


## Bit 13 (MTURWE) Description

| 0 | MTU control register access is disabled |
| :--- | :--- |
| 1 | MTU control register access is enabled (initial value) |

- Bit 8—Multiplex I/O Enable (IOE): Selects the use of CS3 space as ordinary space or address/data multiplex I/O space. A 0 selects ordinary space and a 1 selects address/data multiplex I/O space. When address/data multiplex I/O space is selected, the address and data are multiplexed and output from the data bus. When CS3 space is an address/data multiplex I/O space, bus size is decided by the A14 bit ( $\mathrm{A} 14=0: 8 \mathrm{bit}, \mathrm{A} 14=1: 16 \mathrm{bit}$ ).

| Bit 8 (IOE) | Description |
| :--- | :--- |
| 0 | CS3 space is ordinary space (initial value) |
| 1 | CS3 space is address/data multiplex I/O space |

- Bit 7—CS3 Space Long Size Specification (A3LG): Specifies the CS3 space bus size. This is effective only when CS3 space is ordinary space. When CS3 space is an address/data multiplex I/O space, bus size is decided by the A14 bit.


## Bit 7 (A3LG) Description

$0 \quad$ According to the A3SZ bit specified value (initial value)
1 Longword (32 bit) size

- Bit 6-CS2 Space Long Size Specification (A2LG): Specifies the CS2 space bus size.

| Bit 6 (A2LG) | Description |
| :--- | :--- |
| 0 | According to the A2SZ bit value (initial value) |
| 1 | Longword (32 bit) size |

- Bit 5-CS1 Space Long Size Specification (A1LG): Specifies the CS1 space bus size.

| Bit 5 (A1LG) | Description |
| :--- | :--- |
| 0 | According to the A1SZ bit value (initial value) |
| 1 | Longword (32 bit) size |

- Bit 4—CS0 Space Long Size Specification (A0LG): Specifies the CS0 space bus size.

| Bit 4 (AOLG) | Description |
| :--- | :--- |
| 0 | According to the A0SZ bit value (initial value) |
| 1 | Longword (32 bit) size |
| Note: | AOLG is effective only in on-chip ROM effective mode. When in on-chip ROM ineffective <br> mode, the CSO space bus size is specified by the mode pin. |

- Bit 3-CS3 Space Size Specification (A3SZ): Specifies the CS3 space bus size when A3LG = 0 . This is effective only when CS3 space is ordinary space. When CS3 space is an address/data multiplex I/O space, bus size is decided by the A 14 bit.
Bit 3 (A3SZ) Description

| 0 | Byte (8 bit) size |
| :--- | :--- |
| 1 | Word (16 bit) size (initial value) |
| Note:This bit is ignored when A3LG $=1 ;$ CS3 space bus size becomes longword (32 bit) (for <br> ordinary space). |  |

- Bit 2—CS2 Space Size Specification (A2SZ): Specifies the CS2 space bus size when A2LG = 0.

Bit 2 (A2SZ)

## Description

| 0 | Byte (8 bit) size |
| :--- | :--- |
| 1 | Word (16 bit) size (initial value) |
| Note: | This bit is ignored when A2LG $=1 ; \mathrm{CS} 2$ space bus size becomes longword ( 32 bit). |

Note: This bit is ignored when A2LG = 1; CS2 space bus size becomes longword (32 bit).

- Bit 1—CS1 Space Size Specification (A1SZ): Specifies the CS1 space bus size when A1LG = 0 .

Bit 1 (A1SZ) Description

| 0 | Byte (8 bit) size |
| :--- | :--- |
| 1 | Word (16 bit) size (initial value) |

Note: This bit is ignored when A1LG = 1; CS1 space bus size becomes longword (32 bit).

- Bit 0—CS0 Space Size Specification (A0SZ): Specifies the CS0 space bus size when A0LG = 0 .

Bit 0 (AOSZ) Description

| 0 | Byte (8 bit) size |
| :--- | :--- |
| 1 | Word (16 bit) size (initial value) |

Note: A0SZ is effective only in on-chip ROM effective mode. In on-chip ROM ineffective mode, the CSO space bus size is specified by the mode pin. However, even in on-chip ROM effective mode, this bit is ignored when AOLG $=1$; CSO space bus size becomes longword (32 bit).

### 10.2.2 Bus Control Register 2 (BCR2)

BCR2 is a 16-bit read/write register that specifies the number of idle cycles and CS signal assert extension of each CS space.

BCR2 is initialized by power-on resets to H'FFFF, but is not initialized by manual resets or software standbys.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | IW31 | IW30 | IW21 | IW20 | IW11 | IW10 | IW01 | IW00 |
| Initial value: | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |


| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | CW3 | CW2 | CW1 | CW0 | SW3 | SW2 | SW1 | SW0 |
| Initial value: | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bits 15-8—Idles between Cycles (IW31, IW30, IW21, IW20, IW11, IW10, IW01, IW00):

These bits specify idle cycles inserted between consecutive accesses when the second one is to a different CS area after a read. Idles are used to prevent data conflict between ROM (and other memories, which are slow to turn the read data buffer off), fast memories, and I/O interfaces. Even when access is to the same area, idle cycles must be inserted when a read access is followed immediately by a write access. The idle cycles to be inserted comply with the area specification of the previous access. Refer to section 10.6, Waits between Access Cycles, for details.
IW31, IW30 specify the idle between cycles for CS3 space; IW21, IW20 specify the idle between cycles for CS2 space; IW11, IW10 specify the idle between cycles for CS1 space and IW01, IW00 specify the idle between cycles for CS0 space.

| Bit 15 (IW31) | Bit 14 (IW30) | Description |
| :---: | :---: | :---: |
| 0 | 0 | No idle cycle after accessing CS3 space |
|  | 1 | Inserts one idle cycle after accessing CS3 space |
| 1 | 0 | Inserts two idle cycles after accessing CS3 space |
|  | 1 | Inserts three idle cycles after accessing CS3 space (initial value) |
| Bit 13 (IW21) | Bit 12 (IW20) | Description |
| 0 | 0 | No idle cycle after accessing CS2 space |
|  | 1 | Inserts one idle cycle |
| 1 | 0 | Inserts two idle cycles |
|  | 1 | Inserts three idle cycles (initial value) |
| Bit 11 (IW11) | Bit 10 (IW10) | Description |
| 0 | 0 | No idle cycle after accessing CS1 space |
|  | 1 | Inserts one idle cycle |
| 1 | 0 | Inserts two idle cycles |
|  | 1 | Inserts three idle cycles (initial value) |


| Bit 9 (IW01) | Bit 8 (IW00) | Description |
| :--- | :--- | :--- |
| 0 | 0 | No idle cycle after accessing CSO space |
|  | 1 | Inserts one idle cycle |
| 1 | 0 | Inserts two idle cycles |
| 1 | Inserts three idle cycles (initial value) |  |

- Bits 7-4—Idle Specification for Continuous Access (CW3, CW2, CW1, CW0): The continuous access idle specification makes insertions to clearly delineate the bus intervals by once negating the $\overline{\mathrm{CSn}}$ signal when doing consecutive accesses of the same CS space. When a write immediately follows a read, the number of idle cycles inserted is the larger of the two values specified by IW and CW. Refer to section 10.6, Waits between Access Cycles, for details.
CW3 specifies the continuous access idles for CS3 space; CW2 specifies the continuous access idles for CS2 space; CW1 specifies the continuous access idles for CS1 space and CW0 specifies the continuous access idles for CS0 space.

| Bit 7 (CW3) | Description |
| :--- | :--- |
| 0 | No CS3 space continuous access idle cycles |
| 1 | One CS3 space continuous access idle cycle (initial value) |
| Bit 6 (CW2) | Description |
| 0 | No CS2 space continuous access idle cycles |
| 1 | One CS2 space continuous access idle cycle (initial value) |
| Bit 5 (CW1) | Description |
| 0 | No CS1 space continuous access idle cycles |
| 1 | One CS1 space continuous access idle cycle (initial value) |
| Bit $\mathbf{4}$ (CW0) | Description |
| 0 | No CS0 space continuous access idle cycles |
| 1 | One CS0 space continuous access idle cycle (initial value) |

- Bits 3-0- $\overline{\mathrm{CS}}$ Assert Extension Specification (SW3, SW2, SW1, SW0): The $\overline{\mathrm{CS}}$ assert cycle extension specification is for making insertions to prevent extension of the $\overline{\mathrm{RD}}$ signal or $\overline{\mathrm{WRx}}$ signal assert period beyond the length of the $\overline{\mathrm{CSn}}$ signal assert period. Extended cycles insert one cycle before and after each bus cycle, which simplifies interfaces with external devices and also has the effect of extending write data hold time. Refer to section 10.3.3, $\overline{\mathrm{CS}}$ Assert Period Extension, for details.
SW3 specifies the $\overline{\mathrm{CS}}$ assert extension for CS3 space access; SW2 specifies the $\overline{\mathrm{CS}}$ assert extension for CS2 space access; SW1 specifies the $\overline{\mathrm{CS}}$ assert extension for CS1 space access and SW0 specifies the $\overline{\mathrm{CS}}$ assert extension for CS0 space access.

| Bit 3 (SW3) | Description |
| :--- | :--- |
| 0 | No CS3 space $\overline{\mathrm{CS}}$ assert extension |
| 1 | CS3 space $\overline{\mathrm{CS}}$ assert extension (initial value) |
| Bit $\mathbf{2}$ (SW2) | Description |
| 0 | No CS2 space $\overline{\mathrm{CS}}$ assert extension |
| 1 | CS2 space $\overline{\mathrm{CS}}$ assert extension (initial value) |
| Bit $\mathbf{1}$ (SW1) | Description |
| 0 | No CS1 space $\overline{\mathrm{CS}}$ assert extension |
| 1 | CS1 space $\overline{\mathrm{CS}}$ assert extension (initial value) |
| Bit $\mathbf{0}$ (SW0) | Description |
| 0 | No CS0 space $\overline{\mathrm{CS}}$ assert extension |
| 1 | CS0 space $\overline{\mathrm{CS}}$ assert extension (initial value) |

### 10.2.3 Wait Control Register 1 (WCR1)

WCR1 is a 16-bit read/write register that specifies the number of wait cycles ( $0-15$ ) for each CS space.

WCR1 is initialized by power-on resets to H'FFFF, but is not initialized by manual resets or software standbys.


- Bits 15-12—CS3 Space Wait Specification (W33, W32, W31, W30): Specifies the number of waits for CS3 space access.

| Bit 15 <br> (W33) | Bit 14 <br> (W32) | Bit 13 <br> (W31) | Bit 12 <br> (W30) | Description |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | No wait (external wait input disabled) |
| 0 | 0 | 0 | 1 | 1 wait external wait input enabled |
|  | $\ldots$ |  |  |  |
| 1 | 1 | 1 | 1 | 15 wait external wait input enabled (initial value) |

- Bits 11-8—CS2 Space Wait Specification (W23, W22, W21, W20): Specifies the number of waits for CS2 space access.

| Bit 11 <br> (W23) | Bit 10 <br> (W22) | Bit 9 <br> (W21) | Bit 8 <br> (W20) | Description |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | No wait (external wait input disabled) |
| 0 | 0 | 0 | 1 | 1 wait external wait input enabled |
|  | $\ldots$ |  |  |  |
| 1 | 1 | 1 | 1 | 15 wait external wait input enabled (initial value) |

- Bits 7-4—CS1 Space Wait Specification (W13, W12, W11, W10): Specifies the number of waits for CS1 space access.

| Bit 7 <br> (W13) | Bit 6 <br> (W12) | Bit 5 <br> (W11) | Bit 4 <br> (W10) | Description |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | No wait (external wait input disabled) |
| 0 | 0 | 0 | 1 | 1 wait external wait input enabled |
|  | $\ldots$ |  |  |  |
| 1 | 1 | 1 | 1 | 15 wait external wait input enabled (initial value) |

- Bits 3-0—CS0 Space Wait Specification (W03, W02, W01, W00): Specifies the number of waits for CS0 space access.

| Bit 3 <br> (W03) | Bit 2 <br> (W02) | Bit 1 <br> (W01) | Bit 0 <br> (W00) | Description |
| :--- | :--- | :--- | :--- | :--- |

### 10.2.4 Wait Control Register 2 (WCR2)

WCR2 is a 16-bit read/write register that specifies the number of access cycles for DRAM space and CS space for DMA single address mode transfers.

Do not perform any DMA single address transfers before WCR2 is set.
WCR2 is initialized by power-on resets to $\mathrm{H}^{\prime} 000 \mathrm{~F}$, but is not initialized by manual resets or software standbys.


- Bits 15-6—Reserved: These bits always read as 0 . The write value should always be 0 .
- Bits 5-4—DRAM Space DMA Single Address Mode Access Wait Specification (DDW1, DDW0): Specifies the number of waits for DRAM space access during DMA single address mode accesses. These bits are independent of the DWW and DWR bits of the DCR.

| Bit 5 (DDW1) | Bit 4 (DDW0) | Description |
| :--- | :--- | :--- |
| 0 | 0 | 2-cycle (no wait) external wait disabled <br> (initial value) |
|  | 1 | 3-cycle (1 wait) external wait disabled |
| 1 | 0 | 4-cycle (2 wait) external wait enabled |
|  | 5-cycle (3 wait) external wait enabled |  |

- Bits 3-0—CS Space DMA Single Address Mode Access Wait Specification (DSW3, DSW2, DSW1, DSW0): Specifies the number of waits for CS space access (0-15) during DMA single address mode accesses. These bits are independent of the W bits of the WCR1.

| Bit 3 <br> (DSW3) | Bit 2 <br> (DSW2) | Bit 1 <br> (DSW1) | Bit 0 <br> (DSW0) | Description |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | No wait (external wait input disabled) |
| 0 | 0 | 0 | 1 | 1 wait (external wait input enabled) |
|  | $\ldots$ |  |  |  |
| 1 | 1 | 1 | 1 | 15 wait (external wait input enabled) (initial value) |

### 10.2.5 DRAM Area Control Register (DCR)

DCR is a 16-bit read/write register that selects the number of waits, operation mode, number of address multiplex shifts and the like for DRAM control.

Do not perform any DRAM space accesses before DCR initial settings are completed.
DCR is initialized by power-on resets to $\mathrm{H}^{\prime} 0000$, but is not initialized by manual resets or software standbys.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TPC | RCD | TRAS1 | TRASO | DWW1 | DWW0 | DWR1 | DWR0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | DIW | - | BE | RASD | SZ1 | SZO | AMX1 | AMX0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 15—RAS Precharge Cycle Count (TPC): Specifies the minimum number of cycles after $\overline{\mathrm{RAS}}$ is negated before next assert.

Bit 15 (TPC) Description

| 0 | 1.5 cycles (initial value) |
| :--- | :--- |
| 1 | 2.5 cycles |

- Bit $14 —$ RAS-CAS Delay Cycle Count (RCD): Specifies the number of row address output cycles.

Bit 14 (RCD) Description

| 0 | 1 cycle (initial value) |
| :--- | :--- |
| 1 | 2 cycles |

- Bits 13-12-CAS-Before-RAS Refresh RAS Assert Cycle Count (TRAS1-TRAS0): Specify the number of RAS assert cycles for CAS before RAS refreshes.

| Bit $\mathbf{1 3}$ (TRAS1) | Bit $\mathbf{1 2}$ (TRASO) | Description |
| :--- | :--- | :--- |
| 0 | 0 | 2.5 cycles (initial value) |
|  | 1 | 3.5 cycles |
| 1 | 0 | 4.5 cycles |
|  | 1 | 5.5 cycles |

- Bits 11-10—DRAM Write Cycle Wait Count (DWW1-DWW0): Specifies the number of DRAM write cycle column address output cycles.

| Bit 11 (DWW1) | Bit $\mathbf{1 0}$ (DWW0) | Description |
| :--- | :--- | :--- |
| 0 | 0 | 2-cycle (no wait) external wait disabled (initial value) |
|  | 1 | 3-cycle (1 wait) external wait disabled |
| 1 | 0 | 4-cycle (2 wait) external wait enabled |
|  | 5-cycle (3 wait) external wait enabled |  |

- Bits 9-8—DRAM Read Cycle Wait Count (DWR1-DWR0): Specifies the number of DRAM read cycle column address output cycles.

| Bit 9 (DWR1) | Bit 8 (DWRO) | Description |
| :--- | :--- | :--- |
| 0 | 0 | 2-cycle (no wait) external wait disabled (initial value) |
|  | 1 | 3-cycle (1 wait) external wait disabled |
| 1 | 0 | 4-cycle (2 wait) external wait enabled |
|  | 5-cycle (3 wait) external wait enabled |  |

- Bit 7—DRAM Idle Cycle Count (DIW): Specifies whether to insert idle cycles, either when accessing a different external space (CS space) or when doing a DRAM write, after DRAM reads.

| Bit 7 (DIW) | Description |
| :--- | :--- |
| 0 | No idle cycles (initial value) |
| 1 | 1 idle cycle |

- Bit 6-Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 5—Burst Enable (BE): Specifies the DRAM operation mode.

| Bit 5 (BE) | Description |
| :--- | :--- |
| 0 | Burst disabled (initial value) |
| 1 | DRAM high-speed page mode enabled. |

- Bit 4—RAS Down Mode (RASD): Specifies the DRAM operation mode.

| Bit 4 (RASD) | Description |
| :--- | :--- |
| 0 | Access DRAM by RAS up mode (initial value) |
| 1 | Access DRAM by RAS down mode |

- Bits 3-2—DRAM Bus Width Specification (SZ1, SZ0): Specifies the DRAM space bus width.

| Bit $\mathbf{3}$ (SZ1) | Bit $\mathbf{2}$ (SZ0) | Description |
| :--- | :--- | :--- |
| 0 | 0 | Byte (8 bits) (initial value) |
|  | 1 | Word (16 bits) |
| 1 | Don't care | Longword (32 bits) |

- Bits 1-0—DRAM Address Multiplex (AMX1-AMX0): Specifies the DRAM address multiplex count.

| Bit $\mathbf{1}$ (AMX1) | Bit $\mathbf{0}$ (AMXO) | Description |
| :--- | :--- | :--- |
| 0 | 0 | 9 bit (initial value) |
|  | 1 | 10 bit |
| 1 | 0 | 11 bit |
|  | 12 bit |  |

### 10.2.6 Refresh Timer Control/Status Register (RTCSR)

RTCSR is a 16-bit read/write register that selects the refresh mode and the clock input to the refresh timer counter (RTCNT), and controls compare match interrupts (CMI).

RTCSR is initialized by power-on resets and hardware standbys to $\mathrm{H}^{\prime} 0000$, but is not initialized by manual resets or software standbys.


- Bits $15-7$-Reserved: These bits always read as 0 . The write value should always be 0 .
- Bit 6-Compare Match Flag (CMF): This status flag, which indicates that the values of RTCNT and RTCOR match, is set/cleared under the following conditions:

| Bit 6 (CMF) | Description |
| :--- | :--- |
| 0 | Clear condition: After RTCSR is read when CMF is 1, 0 is written in <br> CMF. In some cases it will clear when DTC is activated by a compare <br> match interrupt; refer to section 8, Data Transfer Controller (DTC), for <br> details. (initial value) |
| 1 | Set condition: RTCNT = RTCOR. When both RTCNT and RTCOR are <br> in an initialized state (when values have not been rewritten since <br> initialization, and RTCNT has not had its value changed due to a count- <br> up), RTCNT and RTCOR match, as both are H'0000, but in this case <br> CMF is not set. |

- Bit 5-Compare Match Interrupt Enable (CMIE): Enables or disables an interrupt request caused by the CMF bit of the RTCSR when CMF is set to 1 .

| Bit 5 (CMIE) | Description |
| :--- | :--- |
| 0 | Disables an interrupt request caused by CMF (initial value) |
| 1 | Enables an interrupt request caused by CMF |

- Bits 4-2—Clock Select (CKS2-CKS0): Select the clock to input to RTCNT from among the seven types of internal clock obtained from dividing the system clock $(\phi)$.

| Bit 4 (CKS2) | Bit 3 (CKS1) | Bit 2 (CKS0) | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Stops count-up (initial value) |
|  | 1 | 0 | $\phi / 2$ |
|  | 1 | $\phi / 8$ |  |
| 1 | 0 | 0 | $\phi / 32$ |
|  | 1 | $\phi / 128$ |  |
|  | 1 | 0 | $\phi / 512$ |

- Bit 1—Refresh Control (RFSH): Selects whether to use refresh control for DRAM.

| Bit 1 (RFSH) | Description |
| :--- | :--- |
| 0 | Do not refresh DRAM (initial value) |
| 1 | Refresh DRAM |

- Bit 0—Refresh Mode (RMD): When the RFSH bit is 1, this bit selects normal refresh or selfrefresh. When the RFSH bit is 1 , self-refresh mode is entered immediately after the RMD bit is set to 1 . When RMD is cleared to 0 , a CAS-before-RAS refresh is performed at the interval set in the refresh time constant register (RTCNT).
When set for self-refresh, the SH7040 Series enters self-refresh mode immediately unless it is in the middle of a DRAM access. If it is, it enters self-refresh mode when the access ends. Refresh requests from the interval timer are ignored in self-refresh mode.


## Bit 0 (RMD) Description

| 0 | CAS-before-RAS refresh (initial value) |
| :--- | :--- |
| 1 | Self-refresh |

### 10.2.7 Refresh Timer Counter (RTCNT)

RTCNT is a 16-bit read/write register that is used as an 8-bit up counter for refreshes or generating interrupt requests.

RTCNT counts up with the clock selected by the CKS2-CKS0 bits of the RTCSR. RTCNT values can always be read/written by the CPU. When RTCNT matches RTCOR, RTCNT is cleared to H'0000 and the CMF flag of the RTCSR is set to 1 . If the RFSH bit of RTCSR is 1 and the RMD bit is 0 at this time, a CAS-before-RAS refresh is performed. Additionally, if the CMIE bit of RTCSR is a 1 , a compare match interrupt (CMI) is generated.

Bits $15-8$ are reserved and play no part in counter operation. They are always read as 0 .
RTCNT is initialized by power-on resets $\mathrm{H}^{\prime} 0000$, but is not initialized by manual resets or software standbys.


### 10.2.8 Refresh Time Constant Register (RTCOR)

RTCOR is a 16-bit read/write register that establishes the compare match period with RTCNT. The values of RTCOR and RTCNT are constantly compared. When the values correspond, the compare match flag of RTCSR is set and RTCNT is cleared to 0 .

When the refresh bit (RFSH) of the RTCSR is set to 1 and the RMD bit is 0 , a refresh request signal is produced by this match. The refresh request signal is held until a refresh operation is performed. If the refresh request is not processed before the next match, the previous request becomes ineffective.

When the CMIE bit of the RTSCR is set to 1 , an interrupt request is sent to the interrupt controller by this match signal. The interrupt request is output continuously until the CMF bit of the RTSCR is cleared.

Bits $15-8$ are reserved and cannot be used in setting the period. They always read 0 .
RTCOR is initialized by power-on resets to $\mathrm{H}^{\prime} 0000$, but is not initialized by manual resets or software standbys.


## Renesns

### 10.3 Accessing Ordinary Space

A strobe signal is output by ordinary space accesses to provide primarily for SRAM or ROM direct connections.

### 10.3.1 Basic Timing

Figure 10.3 shows the basic timing of ordinary space accesses. Ordinary access bus cycles are performed in 2 states.


Figure 10.3 Basic Timing of Ordinary Space Access
During a read, irrespective of operand size, all bits in the data bus width for the access space (address) are fetched by the LSI on $\overline{\mathrm{RD}}$, using the required byte locations.

During a write, the following signals are associated with transfer of these actual byte locations:
$\overline{\mathrm{WRHH}}$ (bits 31-24), $\overline{\mathrm{WRHL}}$ (bits 23-16), $\overline{\mathrm{WRH}}$ (bits 15-8), and $\overline{\mathrm{WRL}}$ (bits 7-0).

### 10.3.2 Wait State Control

The number of wait states inserted into ordinary space access states can be controlled using the WCR settings (figure 10.4).


Figure 10.4 Wait Timing of Ordinary Space Access (Software Wait Only)

When the wait is specified by software using WCR, the wait input $\overline{\mathrm{WAIT}}$ signal from outside is sampled. Figure 10.5 shows the $\overline{\mathrm{WAIT}}$ signal sampling. The $\overline{\mathrm{WAIT}}$ signal is sampled at the clock rise one cycle before the clock rise when $\mathrm{T}_{\mathrm{w}}$ state shifts to $\mathrm{T}_{2}$ state.


Figure 10.5 Wait State Timing of Ordinary Space Access (Wait States from Software Wait 2 State + $\overline{\text { WAIT }}$ Signal)

### 10.3.3 $\overline{\mathrm{CS}}$ Assert Period Extension

Idle cycles can be inserted to prevent extension of the $\overline{\mathrm{RD}}$ signal or $\overline{\mathrm{WRx}}$ signal assert period beyond the length of the $\overline{\mathrm{CSn}}$ signal assert period by setting the SW3-SW0 bits of BCR2. This allows for flexible interfaces with external circuitry. The timing is shown in figure 10.6. $\mathrm{T}_{\mathrm{h}}$ and $\mathrm{T}_{\mathrm{f}}$ cycles are added respectively before and after the ordinary cycle. Only $\overline{\mathrm{CSn}}$ is asserted in these cycles; $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WRx}}$ signals are not. Further, data is extended up to the $\mathrm{T}_{\mathrm{f}}$ cycle, which is effective for gate arrays and the like, which have slower write operations.


Figure 10.6 $\overline{\mathbf{C S}}$ Assert Period Extension Function

### 10.4 DRAM Access

### 10.4.1 DRAM Direct Connection

When address space A31-A24 = H'01 has been accessed, the corresponding space becomes a 16Mbyte DRAM space, and the DRAM interface function can be used to directly connect the SH7040 Series to DRAM.

Row address and column address are always multiplexed for DRAM space. The amount of row address multiplexing can be selected as from 9 to 12 bits by setting the AMX1 and AMX0 bits of the DCR .

Table 10.5 AMX Bits and Address Multiplex Output

|  |  | Row Address |  |  | Column Address |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AMX1 | AMXO | Shift <br> Amount | Output Pins | Output Address | Output Address | Output Pins |
| 0 | 0 | 9 bit | A21-A15 | A21-A15 | A21-A0 | A21-A0 |
|  |  |  | A14-A0 | A23-A9 |  |  |
| 0 | 1 | 10 bit | A21-A14 | A21-A14 | A21-A0 | A21-A0 |
|  |  |  | A13-A0 | A23-A10 |  |  |
| 1 | 0 | 11 bit | A21-A13 | A21-A13 | A21-A0 | A21-A0 |
|  |  |  | A12-A0 | A23-A11 |  |  |
| 1 | 1 | 12 bit | A21-A12 | A21-A12 | A21-A0 | A21-A0 |
|  |  |  | A11-A0 | A23-A12 |  |  |

In addition to ordinary read and write accesses, burst mode access using high speed page mode is supported.

### 10.4.2 Basic Timing

The SH7040 Series supports 2 CAS format DRAM access. The DRAM access basic timing is a minimum of 3 cycles for normal mode. Figure 10.7 shows the basic DRAM access timing. DRAM space access is controlled by $\overline{\mathrm{RAS}}, \overline{\mathrm{CASx}}$, and RDWR signals. The following signals are associated with transfer of these actual byte locations: $\overline{\text { CASHH }}$ (bits 31-24), $\overline{\text { CASHL }}$ (bits 23-16), $\overline{\mathrm{CASH}}$ (bits 15-8), and $\overline{\mathrm{CASL}}$ (bits 7-0). However, the signals for ordinary space, $\overline{\mathrm{WRx}}$ and $\overline{\mathrm{RD}}$, are also output during the DMAC single transfer column address cycle period. $T_{p}$ is the precharge cycle, $T_{r}$ is the RAS assert cycle, $T_{c}$ is the CAS assert cycle and $T_{c 2}$ is the read data fetch cycle.


Figure 10.7 DRAM Bus Cycle (Normal Mode, $\mathbf{T P C}=\mathbf{0}, \mathbf{R C D}=\mathbf{0}$, No Waits)

### 10.4.3 Wait State Control

Wait state insertion during DRAM space access is controlled by setting the TPC, RCD, DWW1, DWW0, DWR1, and DWR0 bits of the DCR. TPC and RCD are common to both reads and writes. The timing with waits inserted is shown in figures 10.8 through 10.11 . External waits can be inserted at the time of software waits 2,3 . The sampling location is the same as that of ordinary space: at one cycle before the $\mathrm{T}_{\mathrm{c} 2}$ cycle clock rise. Wait cycles are extended by external waits.


Figure 10.8 DRAM Bus Cycle (Normal Mode, $\mathbf{T P C}=0, R C D=0$, One Wait)


Figure 10.9 DRAM Bus Cycle (Normal Mode, $\mathbf{T P C}=1, R C D=1$, Two Waits)


Figure 10.10 DRAM Bus Cycle (Normal Mode, $\mathbf{T P C}=\mathbf{0}, \mathbf{R C D}=\mathbf{0}$, Three Waits)


Figure 10.11 DRAM Bus Cycle (Normal Mode, $\mathbf{T P C}=\mathbf{0}, \mathbf{R C D}=\mathbf{0}$, Two Waits + Wait Due to $\overline{\text { WAIT }}$ Signal)

### 10.4.4 Burst Operation

High-Speed Page Mode: When the burst enable bit (BE) of the DCR is set, burst accesses can be performed using high speed page mode. The timing is shown in figure 10.12. Wait cycles can be inserted during burst accesses by using the DCR.


Figure 10.12 DRAM Bus Cycle (High-Speed Page Mode)
RAS Down Mode: There are some instances where even if burst operation is selected, continuous accesses to DRAM will not occur, but another space will be accessed instead part way through the access. In such cases, if the $\overline{\mathrm{RAS}}$ signal is maintained at low level during the time the other space is accessed, it is possible to continue burst operation at the time the next DRAM same row address is accessed. This is called RAS down mode.

To use RAS down mode, set both the BE and RASD bits of the DCR to 1 .

Figures 10.13 and 10.14 show operation in RAS up and down modes.


Figure 10.13 DRAM Access Normal Operation (RAS Up Mode)


Figure 10.14 RAS Down Mode

### 10.4.5 Refresh Timing

The bus state controller is equipped with a function to control refreshes of DRAM. CAS-beforeRAS (CBR) refresh or self-refresh can be selected by setting the RTCSR's RMD bit.

CAS-before-RAS Refresh: For CBR refreshes, set the RCR's RMD bit to 0 and the RFSH bit to 1. Also write the values in RTCNT and RTCOR necessary to fulfill the refresh interval prescribed for the DRAM being used. When a clock is selected with the CKS2-CKS0 bits of the RSTCR, RTCNT starts counting up from the value at that time. The RTCNT value is constantly compared to the RTCOR value and a CBR refresh is performed when the two match. RTCNT is cleared at that time and the count starts again. Figure 10.15 shows the timing for the CBR refresh operation.

The number of RAS assert cycles in the refresh cycle is set by the TRAS1, TRAS0 bits of the DCR.


Figure 10.15 CAS-Before-RAS Refresh Timing (TRAS1, TRAS0 $=0,0$ )

Self-Refresh: When both the RMD and RFSH bits of the RTCSR are set to 1 , the $\overline{\mathrm{CAS}}$ signal and $\overline{\mathrm{RAS}}$ signal are output and the DRAM enters self-refresh mode, as shown in figure 10.16. Do not access DRAM during self-refreshes, in order to preserve DRAM data. When performing DRAM accesses, first cancel the self-refresh, then access only after doing individual refreshes for all row addresses within the time prescribed for the particular DRAM.

For external bus right requests during self-refreshes, to preserve DRAM data at the time of releasing the bus rights, only $\overline{\text { CASx }}, \overline{\mathrm{RAS}}$, and RDWR are output and the bus rights are released to the external device with the self-refresh maintained. Consequently, do not perform DRAM accesses from external devices at such a time.


Figure 10.16 Self-Refresh Timing

### 10.5 Address/Data Multiplex I/O Space Access

When the BCR1 register IOE bit is set to 1 , the D15-D0 pins can be used for multiplexed address/data I/O for the CS3 space. Consequently, peripheral LSIs requiring address/data multiplexing can be directly connected to this LSI.

Address/data multiplex I/O space bus width is selected by the A 14 bit, and is 8 bit when $\mathrm{A} 14=0$ and 16 bit when A14 $=1$.

### 10.5.1 Basic Timing

When the IOE bit of the BCR1 is set to 1, CS3 space becomes address/data multiplex I/O space. When this space is accessed, addresses and data are multiplexed. When the A14 address bit is 0 , the bus size becomes 8 bit and addresses and data are input and output through the D7-D0 pins. When the A14 address bit is 1, the bus size becomes 16 bit and address output and data I/O occur through the D15-D0 pins. Access for the address/data multiplex I/O space is controlled by the $\overline{\mathrm{AH}}, \overline{\mathrm{RD}}$, and $\overline{\mathrm{WRx}}$ signals.

Address/data multiplex I/O space accesses are done after a 3-cycle (fixed) address output, as an ordinary space type access (figure 10.17).


Figure 10.17 Address/Data Multiplex I/O Space Access Timing (No Waits)

### 10.5.2 Wait State Control

Setting the WCR controls waits during address/data multiplex I/O space accesses. Software wait and external wait insertion timing is the same as during ordinary space accesses. The timing for one software wait + one external wait inserted is shown in figure 10.18.


Figure 10.18 Address/Data Multiplex I/O Space Access Wait State Timing (One Software Wait + One External Wait)

### 10.5.3 CS Assertion Extension

The timing diagram when setting CS assertion extension during address/data multiplex I/O space access is shown in figure 10.19.


Figure 10.19 Wait Timing in Address/Data Multiplex I/O Space when CS Assertion Extension is Set

### 10.6 Waits between Access Cycles

When a read from a slow device is completed, data buffers may not go off in time to prevent data conflicts with the next access. If there is a data conflict during memory access, the problem can be solved by inserting a wait in the access cycle.

To enable detection of bus cycle starts, waits can be inserted between access cycles during continuous accesses of the same CS space by negating the $\overline{\mathrm{CSn}}$ signal once.

### 10.6.1 Prevention of Data Bus Conflicts

For the two cases of write cycles after read cycles, and read cycles for a different area after read cycles, waits are inserted so that the number of idle cycles specified by the IW31-IW00 bits of the

BCR2 and the DIW of the DCR occur. When idle cycles already exist between access cycles, only the number of empty cycles remaining beyond the specified number of idle cycles are inserted.

Figure 10.20 shows an example of idles between cycles. In this example, 1 idle between CSn space cycles has been specified, so when a CSm space write immediately follows a CSn space read cycle, 1 idle cycle is inserted.


Figure 10.20 Idle Cycle Insertion Example
IW31 and IW30 specify the number of idle cycles required after a CS3 space read either to read other external spaces, or for this LSI, to do write accesses. In the same manner, IW21 and IW20 specify the number of idle cycles after a CS2 space read, IW11 and IW10, the number after a CS1 space read, and IW01 and IW00, the number after a CS0 space read.

DIW specifies the number of idle cycles required, after a DRAM space read either to read other external spaces (CS space), or for this LSI, to do write accesses.

0 to 3 cycles can be specified for CS space, and 0 to 1 cycle for DRAM space.

### 10.6.2 Simplification of Bus Cycle Start Detection

For consecutive accesses of the same CS space, waits are inserted so that the number of idle cycles designated by the CW3-CW0 bits of the BCR2 occur. However, for write cycles after reads, the number of idle cycles inserted will be the larger of the two values defined by the IW and CW bits. When idle cycles already exist between access cycles, waits are not inserted. Figure 10.21 shows an example. A continuous access idle is specified for CSn space, and CSn space is consecutively write accessed.


Figure 10.21 Same Space Consecutive Access Idle Cycle Insertion Example

### 10.7 Bus Arbitration

The SH7040 series has a bus arbitration function that, when a bus release request is received from an external device, releases the bus to that device. It also has two internal bus masters, the CPU and the DMAC, DTC. The priority ranking for determining bus right transfer between these bus masters is:

$$
\text { Bus right request from external device }>\text { refresh }>\text { DTC }>\text { DMAC }>C P U
$$

However, during a read or write in DMAC dual address mode, a burst transfer, or indirect address transfer mode operation, the DMAC continues operating even if a DTC request is received.

Through port register settings, $\overline{\text { IRQOUT }}$ is asserted to indicate that a CAS-before-RAS refresh request for DRAM has been generated during release of bus rights to an external device. Use this
to cause the external device to negate the $\overline{\mathrm{BREQ}}$ and return the bus rights to the SH7040 Series. Please note that if the external device does not return the bus rights within the time prescribed for the DRAM refresh interval, this LSI will not be able to perform the refresh operation and the DRAM contents cannot be guaranteed.

Figure 10.22 shows the bus right release procedure.


Figure 10.22 Bus Right Release Procedure

### 10.8 Memory Connection Examples

Figures 10.23-10.31 show examples of the memory connections.
As A21-A18 become input ports in power-on reset, they should be handled (e.g. pulled down) as necessary.


Figure 10.23 8-Bit Data Bus Width ROM Connection


Figure 10.24 16-Bit Data Bus Width ROM Connection


Figure 10.25 32-Bit Data Bus Width ROM Connection

| SH704x | $23 k \times 8$ bits SRAM |
| :---: | :---: |
| $\overline{\mathrm{CSn}}$ | $\overline{\mathrm{CS}}$ |
| $\overline{\mathrm{RD}}$ | $\overline{\mathrm{OE}}$ |
| A0-A16 | A0-A16 |
| $\overline{\text { WRL }}$ | $\overline{W E}$ |
| D0-D7 | I/O0-I/O7 |
|  |  |

Figure 10.26 8-Bit Data Bus Width SRAM Connection


Figure 10.27 16-Bit Data Bus Width SRAM Connection


Figure 10.28 32-Bit Data Bus Width SRAM Connection


Figure 10.29 8-Bit Data Bus Width DRAM Connection


Figure 10.30 16-Bit Data Bus Width DRAM Connection


Figure 10.31 32-Bit Data Bus Width DRAM Connection

### 10.9 On-Chip Peripheral I/O Register Access

On-chip peripheral I/O registers are accessed from the bus state controller, as shown in table 10.6.
Table 10.6 On-Chip Peripheral I/O Register Access

| On-chip <br> Peripheral Module | SCI | $\begin{aligned} & \text { MTU, } \\ & \text { POE } \end{aligned}$ | INTC | PFC, PORT | CMT | A/D* | UBC | WDT | DMAC | DTC | CACHE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Connected bus width | 8bit | 16bit | 16bit | 16bit | 16bit | 16bit | 16bit | 16bit | 16bit | 16bit | 16bit |
| Access cycle | 2 cyc | 2 cyc | 2cyc | 2 cyc | 2cyc | 2cyc | 3cyc | 3cyc | 3 cyc | 3cyc | 3cyc |

Note: * A/D of A mask products are accessed in 8-bit width, 3 cyc.

## Cycles in which Bus is not Released

(a) One bus cycle:

The bus is never released during a single bus cycle. For example, in the case of a longword read (or write) in 8-bit normal space, the four memory accesses to the 8 -bit normal space constitute a single bus cycle, and the bus is never released during this period. Assuming that one memory access requires two states, the bus is not released during an 8 -state period.


Figure 10.32 One Bus Cycle

### 10.10 CPU Operation when Program is in External Memory

In the SH7040 Series, two words (equivalent to two instructions) are normally fetched in a single instruction fetch. This is also true when the program is located in external memory, irrespective of whether the external memory bus width is 8 or 16 bits.

If the program counter value immediately after the program branches is an odd-word $(2 n+1)$ address, or if the program counter value immediately before the program branches is an even-word $(2 n)$ address, the CPU will always fetch 32 bits (equivalent to two instructions) that include the respective word instruction.

## Section 11 Direct Memory Access Controller (DMAC)

### 11.1 Overview

The SH7040 Series includes an on-chip four-channel direct memory access controller (DMAC). The DMAC can be used in place of the CPU to perform high-speed data transfers among external devices equipped with DACK (transfer request acknowledge signal), external memories, memorymapped external devices, and on-chip peripheral modules (except for the DMAC, DTC, BSC, and UBC). Using the DMAC reduces the burden on the CPU and increases operating efficiency of the LSI as a whole.

### 11.1.1 Features

The DMAC has the following features:

- Four channels
- Four Gbytes of address space in the architecture
- Byte, word, or longword selectable data transfer unit
- 16 Mbytes ( $16,777,216$ transfers, maximum)
- Single or dual address mode. Dual address mode can be direct or indirect address transfer.
- Single address mode: Either the transfer source or transfer destination (peripheral device) is accessed by a DACK signal while the other is accessed by address. One transfer unit of data is transferred in each bus cycle.
- Dual address mode: Both the transfer source and transfer destination are accessed by address. Dual address mode can be direct or indirect address transfer.
- Direct access: Values set in a DMAC internal register indicate the accessed address for both the transfer source and transfer destination. Two bus cycles are required for one data transfer.
- Indirect access: The value stored at the location pointed to by the address set in the DMAC internal transfer source register is used as the address. Operation is otherwise the same as direct access. This function can only be set for channel 3. Four bus cycles are required for one data transfer.
- Channel function: Transfer modes that can be set are different for each channel. (Dual address mode indirect access can only be set for channel 1 . Only direct access is possible for the other channels.)
- Channel 0: Single or dual address mode. External requests are accepted.
- Channel 1: Single or dual address mode. External requests are accepted.
- Channel 2: Dual address mode only. Source address reload function operates every fourth transfer.
- Channel 3: Dual address mode only. Direct address transfer mode and indirect address transfer mode selectable.
- Reload function: Enables automatic reloading of the value set in the first source address register every fourth DMA transfer. This function can be executed on channel 2 only.
- Transfer requests: There are three DMAC transfer activation requests, as indicated below.
- External request: From two DREQ pins. DREQ can be detected either by falling edge or by low level. External requests can only be received on channels 0 or 1 .
- Requests from on-chip peripheral modules: Transfer requests from on-chip modules such as SCI or A/D. These can be received by all channels.
- Auto-request: The transfer request is generated automatically within the DMAC.
- Selectable bus modes: Cycle-steal mode or burst mode
- Two types of DMAC channel priority ranking:
- Fixed priority mode: Always fixed
- Round robin mode: Sets the lowest priority level for the channel that received the execution request last
- CPU can be interrupted when the specified number of data transfers are complete.


### 11.1.2 Block Diagram

Figure 11.1 is a block diagram of the DMAC.


Figure 11.1 DMAC Block Diagram

### 11.1.3 Pin Configuration

Table 11.1 shows the DMAC pins.
Table 11.1 DMAC Pin Configuration

| Channel | Name | Symbol | I/O | Function |
| :--- | :--- | :--- | :--- | :--- |
| 0 | DMA transfer request | $\overline{\text { DREQ0 }}$ | I | DMA transfer request input from <br> external device to channel 0 |
|  | DMA transfer request <br> acknowledge | DACK0 | O | DMA transfer strobe output from <br> channel 0 to external device |
| DREQ0 acceptance <br> confirmation | DRAK0 | O | Sampling receive acknowledge output <br> for DMA transfer request input from <br> external source |  |
| 1 | DMA transfer request | $\overline{\text { DREQ1 }}$ | I | DMA transfer request input from <br> external device to channel 1 |

### 11.1.4 Register Configuration

Table 11.2 summarizes the DMAC registers. DMAC has a total of 17 registers. Each channel has four control registers. One other control register is shared by all channels

Table 11.2 DMAC Registers

| Channel | Name | Abbreviation | R/W | Initial Value | Address | Register Size | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | DMA source address register 0 | SAR0 | R/W | Undefined | H'FFFF86C0 | 32 bit | 16, 32*2 |
|  | DMA destination address register 0 | DAR0 | R/W | Undefined | H'FFFF86C4 | 32 bit | $16,32^{* 2}$ |
|  | DMA transfer count register 0 | DMATCR0 | R/W | Undefined | H'FFFF86C8 | 32 bit | $16,32^{* 3}$ |
|  | DMA channel control register 0 | CHCRO | R/W*1 | H'00000000 | H'FFFF86CC | 32 bit | $16,32^{* 2}$ |
| 1 | DMA source address register 1 | SAR1 | R/W | Undefined | H'FFFF86D0 | 32 bit | $16,32^{* 2}$ |
|  | DMA destination address register 1 | DAR1 | R/W | Undefined | H'FFFF86D4 | 32 bit | $16,32^{* 2}$ |
|  | DMA transfer count register 1 | DMATCR1 | R/W | Undefined | H'FFFF86D8 | 32 bit | $16,32^{* 3}$ |
|  | DMA channel control register 1 | CHCR1 | R/W*1 | H'00000000 | H'FFFF86DC | 32 bit | $16,32^{* 2}$ |
| 2 | DMA source address register 2 | SAR2 | R/W | Undefined | H'FFFF86E0 | 32 bit | $16,32^{* 2}$ |
|  | DMA destination address register 2 | DAR2 | R/W | Undefined | H'FFFF86E4 | 32 bit | $16,32^{* 2}$ |

Table 11.2 DMAC Registers (cont)

| Channel | Name | Abbreviation | R/W | Initial Value | Address | Register Size | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \hline 2 \\ & \text { (cont) } \end{aligned}$ | DMA transfer count register 2 | DMATCR2 | R/W | Undefined | H'FFFF86E8 | 32 bit | 16, 32 *3 |
|  | DMA channel control register 2 | CHCR2 | R/W*1 | H'00000000 | H'FFFF86EC | 32 bit | $16,32^{* 2}$ |
| 3 | DMA source address register 3 | SAR3 | R/W | Undefined | H'FFFF86F0 | 32 bit | 16, 32*2 |
|  | DMA destination address register 3 | DAR3 | R/W | Undefined | H'FFFF86F4 | 32 bit | $16,32^{* 2}$ |
|  | DMA transfer count register 3 | DMATCR3 | R/W | Undefined | H'FFFF86F8 | 32 bit | 16, 32*3 |
|  | DMA channel control register 3 | CHCR3 | R/W*1 | H'00000000 | H'FFFF86FC | 32 bit | $16,32^{* 2}$ |
| Shared | dDMA operation register | DMAOR | R/W*1 | H'0000 | H'FFFF86B0 | 16 bit | 16, 32*4 |

Notes: Do not attempt to access an empty address. If an access is attemped, the system operation is not guarenteed.

* 1 Write 0 after reading 1 in bit 1 of CHCRO-CHCR3 and in bits 1 and 2 of the DMAOR to clear flags. No other writes are allowed.
*2 For 16-bit access of SAR0-SAR3, DAR0-DAR3, and CHCR0-CHCR3, the 16-bit value on the side not accessed is held.
*3 DMATCR has a 24-bit configuration: bits 0-23. Writing to the upper 8 bits (bits 24-31) is invalid, and these bits always read 0 .
*4 Do not make 32-bit access for DMAOR.


### 11.2 Register Descriptions

### 11.2.1 DMA Source Address Registers 0-3 (SAR0-SAR3)

DMA source address registers 0-3 (SAR0-SAR3) are 32-bit read/write registers that specify the source address of a DMA transfer. These registers have a count function, and during a DMA transfer, they indicate the next source address. In single-address mode, SAR values are ignored when a device with DACK has been specified as the transfer source.

Specify a 16-bit or 32-bit boundary address when doing 16-bit or 32-bit data transfers. Operation cannot be guaranteed on any other addresses.

The initial value after power-on resets or in software standby mode is undefined. These registers are not initialized with manual reset.


### 11.2.2 DMA Destination Address Registers 0-3 (DAR0-DAR3)

DMA destination address registers 0-3 (DAR0-DAR3) are 32-bit read/write registers that specify the destination address of a DMA transfer. These registers have a count function, and during a DMA transfer, they indicate the next destination address. In single-address mode, DAR values are ignored when a device with DACK has been specified as the transfer destination.

Specify a 16-bit or 32-bit boundary address when doing 16-bit or 32-bit data transfers. Operation cannot be guaranteed on any other address. The initial value after power-on resets or in software standby mode, is undefined. These registers are not initialized with manual reset.


### 11.2.3 DMA Transfer Count Registers 0-3 (DMATCR0-DMATCR3)

DMA transfer count registers 0-3 (DMATCR0-DMATCR3) are 24-bit read/write registers that specify the transfer count for the channel (byte count, word count, or longword count). Specifying a H'000001 gives a transfer count of 1 , while $\mathrm{H}^{\prime} 000000$ gives the maximum setting, $16,777,216$ transfers. The data for the upper 8 bits of a DMATCR is 0 when read. Always write 0 . The initial value after power-on resets or in software standby mode is undefined. These registers are not initialized with manual reset.

Always write 0 to the upper 8 bits of a DMATCR.


## Renesns

### 11.2.4 DMA Channel Control Registers 0-3 (CHCR0-CHCR3)

DMA channel control registers $0-3$ (CHCR0-CHCR3) is a 32 -bit read/write register where the operation and transmission of each channel is designated. They are initialized by a power-on reset and in software standby mode. There is no initializing with manual reset.


Notes: *1 TE bit: Allows only 0 write after reading 1.
*2 The DI, RO, RL, AM, AL, or DS bit may be absent, depending on the channel.

- Bits 31-21—Reserved bits: Data are 0 when read. The write value always be 0 .
- Bit 20—Direct/Indirect (DI): Specifies either direct address mode operation or indirect address mode operation for channel 3 source address. This bit is valid only in CHCR3. It always reads 0 for CHCR0-CHCR2, and cannot be modified.


## Bit 20: DI Description

| 0 | Direct access mode operation for channel 3 (initial value) |
| :--- | :--- |
| 1 | Indirect access mode operation for channel 3 |

- Bit 19—Source Address Reload (RO): Selects whether to reload the source address initial value during channel 2 transfer. This bit is valid only for channel 2. It always reads 0 for CHCR0, CHCR1, and CHCR3, and cannot be modified.

Bit 19: RO Description

| 0 | Does not reload source address (initial value) |
| :--- | :--- |
| 1 | Reloads source address |

- Bit 18—Request Check Level (RL): Selects whether to output DRAK notifying external device of $\overline{\mathrm{DREQ}}$ received, with active high or active low. This bit is valid only for CHCR0 and CHCR1. It always reads 0 for CHCR2 and CHCR3, and cannot be modified.


## Bit 18: RL Description

| 0 | Output DRAK with active high (initial value) |
| :--- | :--- |
| 1 | Output DRAK with active low |

- Bit 17—Acknowledge Mode (AM): In dual address mode, selects whether to output DACK in the data write cycle or data read cycle. In single address mode, DACK is always output irrespective of the setting of this bit. This bit is valid only for CHCR0 and CHCR1. It always reads as 0 for CHCR2 and CHCR3, and cannot be modified.

Bit 17: AM Description

| 0 | Outputs DACK during read cycle (initial value) |
| :--- | :--- |
| 1 | Outputs DACK during write cycle |

- Bit 16—Acknowledge Level (AL): Specifies whether to set DACK (acknowledge) signal output to active high or active low. This bit is valid only with CHCR0 and CHCR1. It always reads as 0 for CHCR2 and CHCR3, and cannot be modified.

Bit 16: AL Description

| 0 | Active high output (initial value) |
| :--- | :--- |
| 1 | Active low output |

- Bits 15 and 14—Destination Address Mode 1, 0 (DM1 and DM0): These bits specify increment/decrement of the DMA transfer destination address. These bit specifications are ignored when transferring data from an external device to address space in single address mode.


## Bit 15: DM1 Bit 14: DMO Description

| 0 | 0 | Destination address fixed (initial value) |
| :--- | :--- | :--- |
| 0 | 1 | Destination address incremented (+1 during 8-bit transfer, +2 <br> during 16-bit transfer, +4 during 32-bit transfer) |
| 1 | 0 | Destination address decremented (-1 during 8-bit transfer, -2 <br> during 16-bit transfer, -4 during 32-bit transfer) |
| 1 | 1 | Setting prohibited |

- Bits 13 and 12—Source Address Mode 1, 0 (SM1 and SM0): These bits specify increment/decrement of the DMA transfer source address. These bit specifications are ignored when transferring data from an external device to address space in single address mode.


## Bit 13: SM1 Bit 12: SM0 Description

| 0 | 0 | Source address fixed (initial value) |
| :--- | :--- | :--- |
| 0 | 1 | Source address incremented (+1 during 8-bit transfer, +2 <br> during 16-bit transfer, +4 during 32-bit transfer) |
| 1 | 0 | Source address decremented (-1 during 8-bit transfer, -2 <br> during 16-bit transfer, -4 during 32-bit transfer) |
| 1 | 1 | Setting prohibited |

When the transfer source is specified at an indirect address, specify in source address register 3 (SAR3) the actual storage address of the data you want to transfer as the data storage address (indirect address).

During indirect address mode, SAR3 obeys the SM1/SM0 setting for increment/decrement. In this case, SAR3's increment/decrement is fixed at $+4 /-4$ or 0 , irrespective of the transfer data size specified by TS1 and TS0.

- Bits 11-8—Resource Select 3-0 (RS3-RS0): These bits specify the transfer request source.

| Bit 11: <br> RS3 | Bit 10: <br> RS2 | Rit 9: <br> RS1 | Bit 8: <br> RS0 | Description |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | External request, dual address mode (initial value) |
| 0 | 0 | 0 | 1 | Prohibited |
| 0 | 0 | 1 | 0 | External request, single address mode. External address <br> space $\rightarrow$ external device. |
| 0 | 0 | 1 | 1 | External request, single address mode. External device $\rightarrow$ <br> external address space. |
| 0 | 1 | 0 | 0 | Auto-request |
| 0 | 1 | 0 | 1 | Prohibited |
| 0 | 1 | 1 | 0 | MTU TGI0A |
| 0 | 1 | 1 | 1 | MTU TGI1A |
| 1 | 0 | 0 | 0 | MTU TGI2A |
| 1 | 0 | 0 | 1 | MTU TGI3A |
| 1 | 0 | 1 | 0 | MTU TGI4A |
| 1 | 0 | 1 | 1 | A/D ADI* |
| 1 | 1 | 0 | 0 | SCI0 TXI0 |
| 1 | 1 | 0 | 1 | SCI0 RXI0 |
| 1 | 1 | 1 | 0 | SCI1 TXI1 |
| 1 | 1 | 1 | 1 | SCI1 RXI1 |

Notes: External request designations are valid only for channels 0 and 1. No transfer request sources can be set for channels 2 or 3.

* ADI1 for A mask.
- Bit 7—Reserved bits: Data is 0 when read. The write value always be 0 .
- Bit 6- $\overline{\mathrm{DREQ}}$ Select (DS): Sets the sampling method for the $\overline{\mathrm{DREQ}}$ pin in external request mode to either low-level detection or falling-edge detection. This bit is valid only with CHCR0 and CHCR1. For CHCR2 and CHCR3, this bit always reads as 0 and cannot be modified.
Even with channels 0 and 1 , when specifying an on-chip peripheral module or auto-request as the transfer request source, this bit setting is ignored. The sampling method is fixed at fallingedge detection in cases other than auto-request.

Bit 6: DS Description

| 0 | Low-level detection (initial value) |
| :--- | :--- |
| 1 | Falling-edge detection |

- Bit 5—Transfer Mode (TM): Specifies the bus mode for data transfer.

| Bit 5: TM | Description |
| :--- | :--- |
| 0 | Cycle steal mode (initial value) |
| 1 | Burst mode |

- Bits 4 and 3—Transfer Size 1, 0 (TS1, TS0): Specifies size of data for transfer.

Bit 4: TS1 Bit 3: TSO Description

| 0 | 0 | Specifies byte size (8 bits) (initial value) |
| :--- | :--- | :--- |
| 0 | 1 | Specifies word size (16 bits) |
| 1 | 0 | Specifies longword size (32 bits) |
| 1 | 1 | Prohibited |

- Bit 2—Interrupt Enable (IE): When this bit is set to 1, interrupt requests are generated after the number of data transfers specified in the DMATCR (when TE = 1).

| Bit 2: IE | Description |
| :--- | :--- |
| 0 | Interrupt request not generated after DMATCR-specified transfer count <br> (initial value) |
| 1 | Interrupt request enabled on completion of DMATCR specified number <br> of transfers |

- Bit 1—Transfer End Flag (TE): This bit is set to 1 after the number of data transfers specified by the DMATCR. At this time, if the IE bit is set to 1 , an interrupt request is generated.
If data transfer ends before TE is set to 1 (for example, due to an NMI or address error, or clearing of the DE bit or DME bit of the DMAOR) the TE is not set to 1 . With this bit set to 1 , data transfer is disabled even if the DE bit is set to 1 .

| Bit 1: TE | Description |
| :--- | :--- |
| 0 | DMATCR-specified transfer count not ended (initial value) |
| Clear condition: 0 write after TE $=1$ read, Power-on reset, standby <br> mode |  |
| 1 | DMATCR specified number of transfers completed |

- Bit 0—DMAC Enable (DE): DE enables operation in the corresponding channel.

| Bit 0: DE | Description |
| :--- | :--- |
| 0 | Operation of the corresponding channel disabled (initial value) |
| 1 | Operation of the corresponding channel enabled |

Transfer mode is entered if this bit is set to 1 when auto-request is specified (RS3-RS0 settings). With an external request or on-chip module request, when a transfer request occurs after this bit is set to 1 , transfer is enabled. If this bit is cleared during a data transfer, transfer is suspended.

If the DE bit has been set, but $\mathrm{TE}=1$, then if the DME bit of the DMAOR is 0 , and the NMI or AE bit of the DMAOR is 1 , transfer enable mode is not entered.

### 11.2.5 DMAC Operation Register (DMAOR)

The DMAOR is a 16 -bit read/write register that specifies the transfer mode of the DMAC
Register values are initialized to 0 during power-on reset or in software standby mode. Manual reset does not initialize DMAOR.


Note: * 0 write only is valid after 1 is read at the AE and NMIF bits.

- Bits 15-10—Reserved bits: Data are 0 when read. The write value always be 0 .
- Bits 9-8—Priority Mode 1 and 0 (PR1 and PR0): These bits determine the priority level of channels for execution when transfer requests are made for several channels simultaneously.

| Bit 9: PR1 | Bit 8: PR0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | $\mathrm{CH} 0>\mathrm{CH} 1>\mathrm{CH} 2>\mathrm{CH} 3$ (initial value) |
| 0 | 1 | $\mathrm{CH} 0>\mathrm{CH} 2>\mathrm{CH} 3>\mathrm{CH} 1$ |
| 1 | 0 | $\mathrm{CH} 2>\mathrm{CH} 0>\mathrm{CH} 1>\mathrm{CH} 3$ |
| 1 | 1 | Round robin mode |

- Bits 7-3-Reserved bits: Data are 0 when read. The write value always be 0 .
- Bit 2—Address Error Flag (AE): Indicates that an address error has occurred during DMA transfer. If this bit is set during a data transfer, transfers on all channels are suspended. The CPU cannot write a 1 to the AE bit. Clearing is effected by 0 write after 1 read.

| Bit 2: $\mathbf{A E}$ | Description |
| :--- | :--- |
| 0 | No address error, DMA transfer enabled (initial value) |
|  | Clearing condition: Write $A E=0$ after reading AE $=1$ |
| 1 | Address error, DMA transfer disabled |
|  | Setting condition: Address error due to DMAC |

- Bit 1—NMI Flag (NMIF): Indicates input of an NMI. This bit is set irrespective of whether the DMAC is operating or suspended. If this bit is set during a data transfer, transfers on all channels are suspended. The CPU is unable to write a 1 to the NMIF. Clearing is effected by a 0 write after 1 read.

| Bit 1: NMIF | Description |
| :--- | :--- |
| 0 | No NMI interrupt, DMA transfer enabled (initial value) |
|  | Clearing condition: Write NMIF $=0$ after reading NMIF $=1$ |
| 1 | NMI has occurred, DMC transfer prohibited |
|  | Set condition: NMI interrupt occurrence |

- Bit 0—DMAC Master Enable (DME): This bit enables activation of the entire DMAC. When the DME bit and DE bit of the CHCR for the corresponding channel are set to 1 , that channel is transfer-enabled. If this bit is cleared during a data transfer, transfers on all channels are suspended.
Even when the DME bit is set, when the TE bit of the CHCR is 1 , or its DE bit is 0 , transfer is disabled in the case of an NMI of the DMAOR or when $\mathrm{AE}=1$.

Bit 0: DME Description

| 0 | Disable operation on all channels (initial value) |
| :--- | :--- |
| 1 | Enable operation on all channels |

### 11.3 Operation

When there is a DMA transfer request, the DMAC starts the transfer according to the predetermined channel priority order; when the transfer end conditions are satisfied, it ends the transfer. Transfers can be requested in three modes: auto-request, external request, and on-chip peripheral module request. Transfer can be in either the single address mode or the dual address mode, and dual address mode can be either direct or indirect address transfer mode. The bus mode can be either burst or cycle steal.

### 11.3.1 DMA Transfer Flow

After the DMA source address registers (SAR), DMA destination address registers (DAR), DMA transfer count register (DMATCR), DMA channel control registers (CHCR), and DMA operation register (DMAOR) are set to the desired transfer conditions, the DMAC transfers data according to the following procedure:

1. The DMAC checks to see if transfer is enabled $(\mathrm{DE}=1, \mathrm{DME}=1, \mathrm{TE}=0, \mathrm{NMIF}=0$, $\mathrm{AE}=0$ ).
2. When a transfer request comes and transfer has been enabled, the DMAC transfers 1 transfer unit of data (determined by TS0 and TS1 setting). For an auto-request, the transfer begins automatically when the DE bit and DME bit are set to 1 . The DMATCR value will be decremented by 1 upon each transfer. The actual transfer flows vary by address mode and bus mode.
3. When the specified number of transfers have been completed (when DMATCR reaches 0 ), the transfer ends normally. If the IE bit of the CHCR is set to 1 at this time, a DEI interrupt is sent to the CPU.
4. When an address error occurs in the DMAC or an NMI interrupt is generated, the transfer is aborted. Transfers are also aborted when the DE bit of the CHCR or the DME bit of the DMAOR are changed to 0 .

Figure 11.2 is a flowchart of this procedure.


Figure 11.2 DMAC Transfer Flowchart

### 11.3.2 DMA Transfer Requests

DMA transfer requests are usually generated in either the data transfer source or destination, but they can also be generated by devices and on-chip peripheral modules that are neither the source nor the destination. Transfers can be requested in three modes: auto-request, external request, and on-chip peripheral module request. The request mode is selected in the RS3-RS0 bits of the DMA channel control registers $0-3$ (CHCR0-CHCR3).

Auto-Request Mode: When there is no transfer request signal from an external source, as in a memory-to-memory transfer or a transfer between memory and an on-chip peripheral module unable to request a transfer, the auto-request mode allows the DMAC to automatically generate a transfer request signal internally. When the DE bits of CHCR0-CHCR3 and the DME bit of the DMAOR are set to 1 , the transfer begins (so long as the TE bits of CHCR0-CHCR3 and the NMIF and AE bits of DMAOR are all 0 ).

External Request Mode: In this mode a transfer is performed at the request signal ( $\overline{\mathrm{DREQ}})$ of an external device. Choose one of the modes shown in table 11.3 according to the application system. When this mode is selected, if the DMA transfer is enabled ( $\mathrm{DE}=1, \mathrm{DME}=1, \mathrm{TE}=0$, NMIF $=0$, $\mathrm{AE}=0$ ), a transfer is performed upon a request at the $\overline{\mathrm{DREQ}}$ input. Choose to detect $\overline{\mathrm{DREQ}}$ by either the falling edge or low level of the signal input with the DS bit of CHCR0-CHCR3 (DS = 0 is level detection, $\mathrm{DS}=1$ is edge detection). The source of the transfer request does not have to be the data transfer source or destination.

Table 11.3 Selecting External Request Modes with the RS Bits

| RS3 | RS2 | RS1 | RS0 | Address Mode | Source | Destination |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | Dual address <br> mode | Any* | Any* |
| 0 | 0 | 1 | 0 | Single address <br> mode | External memory or <br> memory-mapped <br> external device | External device with <br> DACK |
| 0 | 0 | 1 | 1 | Single address <br> mode | External device with <br> DACK | External memory or <br> memory-mapped <br> external device |

Note: * External memory, memory-mapped external device, on-chip memory, on-chip peripheral module (excluding DMAC, DTC, BSC, UBC).

On-Chip Peripheral Module Request Mode: In this mode a transfer is performed at the transfer request signal (interrupt request signal) of an on-chip peripheral module. As indicated in table 11.4, there are ten transfer request signals: five from the multifunction timer pulse unit (MTU), which are compare match or input capture interrupts; the receive data full interrupts (RxI) and transmit data empty interrupts (TxI) of the two serial communication interfaces (SCI); and the A/D conversion end interrupt (ADI1 for A mask, ADI for others) of the A/D converter. When DMA
transfers are enabled $(\mathrm{DE}=1, \mathrm{DME}=1, \mathrm{TE}=0, \mathrm{NMIF}=0, \mathrm{AE}=0)$, a transfer is performed upon the input of a transfer request signal.

The transfer request source need not be the data transfer source or transfer destination. However, when the transfer request is set by RxI (transfer request because SCI's receive data is full), the transfer source must be the SCI's receive data register (RDR). When the transfer request is set by TxI (transfer request because SCI's transmit data is empty), the transfer destination must be the SCI's transmit data register (TDR). Also, if the transfer request is set to the A/D converter, the data transfer destination must be the $\mathrm{A} / \mathrm{D}$ converter register.

Table 11.4 Selecting On-Chip Peripheral Module Request Modes with the RS Bits

| RS3 | S2 | RS | RSO | DMAC Transfer Request Source | DMA Transfer Request Signal | Source | Destination | Bus Mode |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 1 | 1 | 0 | MTU*2 | TGIOA | Any*1 | Any*1 | Burst/cycle steal |
| 0 | 1 | 1 | 1 | MTU*2 | TGI1A | Any*1 | Any*1 | Burst/cycle steal |
| 1 | 0 | 0 | 0 | MTU*2 | TGI2A | Any*1 | Any*1 | Burst/cycle steal |
| 1 | 0 | 0 | 1 | MTU*2 | TGI3A | Any*1 | Any*1 | Burst/cycle steal |
| 1 | 0 | 1 | 0 | MTU*2 | TGI4A | Any*1 | Any*1 | Burst/cycle steal |
| 1 | 0 | 1 | 1 | A/D | ADI*5 | ADDR** | Any*1 | Burst/cycle steal |
| 1 | 1 | 0 | 0 | SCIO*3 transmit block | TxIO | Any*1 | TDR0 | Burst/cycle steal |
| 1 | 1 | 0 | 1 | SCIO*3 transmit block | Rxi0 | RDR0 | Any*1 | Burst/cycle steal |
| 1 | 1 | 1 | 0 | SCl1*3 transmit block | TxI1 | Any*1 | TDR1 | Burst/cycle steal |
| 1 | 1 | 1 | 1 | SCl1 ${ }^{* 3}$ transmit block | RxI1 | RDR1 | Any*1 | Burst/cycle steal |

Notes: *1 External memory, memory-mapped external device, on-chip memory, on-chip peripheral module (excluding DMAC, DTC, BSC, UBC).
*2 MTU: Multifunction timer pulse unit.
*3 SCI0, SCl1: Serial communications interface.
*4 ADDR0, ADDR1: A/D converter's A/D register.
*5 ADI1 for A mask.

In order to output a transfer request from an on-chip peripheral module, set the relevant interrupt enable bit for each module, and output an interrupt signal.

When an on-chip peripheral module's interrupt request signal is used as a DMA transfer request signal, interrupts for the CPU are not generated.

When a DMA transfer is conducted corresponding with one of the transfer request signals in table 11.4 , it is automatically discontinued. In cycle steal mode this occurs in the first transfer, and in burst mode with the last transfer.

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### 11.3.3 Channel Priority

When the DMAC receives simultaneous transfer requests on two or more channels, it selects a channel according to a predetermined priority order, either in a fixed mode or in round robin mode. These modes are selected by priority bits PR1 and PR0 in the DMA operation register (DMAOR).

Fixed Mode: In these modes, the priority levels among the channels remain fixed.
The following priority orders are available for fixed mode:

- $\mathrm{CH} 0>\mathrm{CH} 1>\mathrm{CH} 2>\mathrm{CH} 3$
- $\mathrm{CH} 0>\mathrm{CH} 2>\mathrm{CH} 3>\mathrm{CH} 1$
- $\mathrm{CH} 2>\mathrm{CH} 0>\mathrm{CH} 1>\mathrm{CH} 3$

These are selected by settings of the PR1 and PR0 bits of the DMA operation register (DMAOR).
Round Robin Mode: In round robin mode, each time the transfer of one transfer unit (byte, word or long word) ends on a given channel, that channel receives the lowest priority level (figure 11.3). The priority level in round robin mode immediately after a reset is $\mathrm{CH} 0>\mathrm{CH} 1>\mathrm{CH} 2>\mathrm{CH} 3$.

## Transfer on channel 0



Channel 0 is given the lowest priority.

## Transfer on channel 1



When channel 1 is given the lowest priority, the priority of channel 0 , which was above channel 1 , is also shifted simultaneously.

## Transfer on channel 2

 request for channel 1 only.

## Transfer on channel 3



When channel 2 receives the lowest priority, the priorities of channel 0 and 1 , which were above channel 2, are also shifted simultaneously. Immediately thereafter, if there is a transfer request for channel 1 only, channel 1 is given the lowest priority, and the priorities of channels 3 and 0 are simultaneously shifted down.

No change in priority.

Figure 11.3 Round Robin Mode

Figure 11.4 shows the changes in priority levels when transfer requests are issued simultaneously for channels 0 and 3 , and channel 1 receives a transfer request during a transfer on channel 0 . The DMAC operates in the following manner under these circumstances:

1. Transfer requests are issued simultaneously for channels 0 and 3 .
2. Since channel 0 has a higher priority level than channel 3 , the channel 0 transfer is conducted first (channel 3 is on transfer standby).
3. A transfer request is issued for channel 1 during a transfer on channel 0 (channels 1 and 3 are on transfer standby).
4. At the end of the channel 0 transfer, channel 0 shifts to the lowest priority level.
5. At this point, channel 1 has a higher priority level than channel 3 , so the channel 1 transfer comes first (channel 3 is on transfer standby).
6. When the channel 1 transfer ends, channel 1 shifts to the lowest priority level.
7. Channel 3 transfer begins.
8. When the channel 3 transfer ends, channel 3 and channel 2 priority levels are lowered, giving channel 3 the lowest priority.


Figure 11.4 Example of Changes in Priority in Round Robin Mode

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### 11.3.4 DMA Transfer Types

The DMAC supports the transfers shown in table 11.5. It can operate in the single address mode, in which either the transfer source or destination is accessed using an acknowledge signal, or dual access mode, in which both the transfer source and destination addresses are output. The dual access mode consists of a direct address mode, in which the output address value is the object of a direct data transfer, and an indirect address mode, in which the output address value is not the object of the data transfer, but the value stored at the output address becomes the transfer object address. The actual transfer operation timing varies with the bus mode. The DMAC has two bus modes: cycle-steal mode and burst mode.

Table 11.5 Supported DMA Transfers

| Source | Destination |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | External Device with DACK | External <br> Memory | Memory- <br> Mapped <br> External <br> Device | On-Chip Memory | On-Chip Peripheral Module |
| External device with DACK | Not available | Single | Single | Not available | Not available |
| External memory | Single | Dual | Dual | Dual | Dual |
| Memory-mapped external device | Single | Dual | Dual | Dual | Dual |
| On-chip memory | Not available | Dual | Dual | Dual | Dual |
| On-chip peripheral module | Not available | Dual | Dual | Dual | Dual |

Notes: 1. Single: Single address mode
2. Dual: Dual address mode; includes both direct address mode and indirect address mode.

### 11.3.5 Address Modes

Single Address Mode: In the single address mode, both the transfer source and destination are external; one (selectable) is accessed by a DACK signal while the other is accessed by an address. In this mode, the DMAC performs the DMA transfer in 1 bus cycle by simultaneously outputting a transfer request acknowledge DACK signal to one external device to access it while outputting an address to the other end of the transfer. Figure 11.5 shows an example of a transfer between an external memory and an external device with DACK in which the external device outputs data to the data bus while that data is written in external memory in the same bus cycle.


Figure 11.5 Data Flow in Single Address Mode
Two types of transfers are possible in the single address mode: (a) transfers between external devices with DACK and memory-mapped external devices, and (b) transfers between external devices with DACK and external memory. The only transfer requests for either of these is the external request $(\overline{\mathrm{DREQ}})$. Figure 11.6 shows the DMA transfer timing for the single address mode.


Figure 11.6 Example of DMA Transfer Timing in the Single Address Mode

### 11.3.6 Dual Address Mode

Dual address mode is used for access of both the transfer source and destination by address. Transfer source and destination can be accessed either internally or externally. Dual address mode is subdivided into two other modes: direct address transfer mode and indirect address transfer mode.

Direct Address Transfer Mode: Data is read from the transfer source during the data read cycle, and written to the transfer destination during the write cycle, so transfer is conducted in two bus cycles. At this time, the transfer data is temporarily stored in the DMAC. With the kind of external memory transfer shown in figure 11.7, data is read from one of the memories by the DMAC during a read cycle, then written to the other external memory during the subsequent write cycle. Figure 11.8 shows the timing for this operation.

1st bus cycle


The SAR value is taken as the address, and data is read from the transfer source module and stored temporarily in the DMAC.

2nd bus cycle


The DAR value is taken as the address, and data stored in the DMAC's data buffer is written to the transfer destination module.

Figure 11.7 Direct Address Operation during Dual Address Mode


Figure 11.8 Example of Direct Address Transfer Timing in Dual Address Mode

Indirect Address Transfer Mode: In this mode the memory address storing the data you actually want to transfer is specified in DMAC internal transfer source address register (SAR3). Therefore, in indirect address transfer mode, the DMAC internal transfer source address register value is read first. This value is stored once in the DMAC. Next, the read value is output as the address, and the value stored at that address is again stored in the DMAC. Finally, the subsequent read value is written to the address specified by the transfer destination address register, ending one cycle of DMA transfer.

In indirect address mode (figure 11.9), transfer destination, transfer source, and indirect address storage destination are all 16 -bit external memory locations, and transfer in this example is conducted in 16 -bit or 8 -bit units. Timing for this transfer example is shown in figure 11.10.

In indirect address mode, one NOP cycle (figure 11.10) is required until the data read as the indirect address is output to the address bus. When transfer data is 32 -bit, the third and fourth bus cycles each need to be doubled, giving a required total of six bus cycles and one NOP cycle for the whole operation.

1st, 2nd bus cycles


The SAR3 value is taken as the address, memory data is read, and the value is stored in the temporary buffer. Since the value read at this time is used as the address, it must be 32 bits. When external connection data bus is 16 bits, two bus cycles are required.

3rd bus cycle


The value in the temporary buffer is taken as the address, and data is read from the transfer source module to the data buffer.

4th bus cycle


The DAR3 value is taken as the address, and the value in the data buffer is written to the transfer destination module.

Note: Memory, transfer source, and transfer destination modules are shown here. In practice, connection can be made anywhere there is address space.

Figure 11.9 Dual Address Mode and Indirect Address Operation (When External Memory Space is $\mathbf{1 6}$ bits)


Figure 11.10 Dual Address Mode and Indirect Address Transfer Timing Example 1 (External Memory Space to External Memory Space)

Figure 11.11 shows an example of timing in indirect address mode when transfer source and indirect address storage locations are in internal memory, the transfer destination is an on-chip peripheral module with 2 -cycle access space, and transfer data is 8 -bit.

Since the indirect address storage destination and the transfer source are in internal memory, these can be accessed in one cycle. The transfer destination is 2-cycle access space, so two data write cycles are required. One NOP cycle is required until the data read as the indirect address is output to the address bus.


Figure 11.11 Dual Address Mode and Indirect Address Transfer Timing Example 2 (On-chip Memory Space to On-chip Memory Space)

### 11.3.7 Bus Modes

Select the appropriate bus mode in the TM bits of CHCR0-CHCR3. There are two bus modes: cycle steal and burst.

Cycle-Steal Mode: In the cycle steal mode, the bus right is given to another bus master after each one-transfer-unit (byte, word, or longword) DMAC transfer. When the next transfer request occurs, the bus rights are obtained from the other bus master and a transfer is performed for one transfer unit. When that transfer ends, the bus right is passed to the other bus master. This is repeated until the transfer end conditions are satisfied.

The cycle steal mode can be used with all categories of transfer destination, transfer source and transfer request. Figure 11.12 shows an example of DMA transfer timing in the cycle steal mode. Transfer conditions are dual address mode and $\overline{\mathrm{DREQ}}$ level detection.


Figure 11.12 DMA Transfer Example in the Cycle-Steal Mode
Burst Mode: Once the bus right is obtained, the transfer is performed continuously until the transfer end condition is satisfied. In the external request mode with low level detection of the $\overline{\mathrm{DREQ}}$ pin, however, when the $\overline{\mathrm{DREQ}}$ pin is driven high, the bus passes to the other bus master after the bus cycle of the DMAC that currently has an acknowledged request ends, even if the transfer end conditions have not been satisfied.

Figure 11.13 shows an example of DMA transfer timing in the burst mode. Transfer conditions are single address mode and $\overline{\mathrm{DREQ}}$ level detection.


Figure 11.13 DMA Transfer Example in the Burst Mode

### 11.3.8 Relationship between Request Modes and Bus Modes by DMA Transfer Category

Table 11.6 shows the relationship between request modes and bus modes by DMA transfer category.

Table 11.6 Relationship of Request Modes and Bus Modes by DMA Transfer Category

| Address Mode | Transfer Category | Request Mode | Bus* ${ }^{*}$ <br> Mode | Transfer Size (Bits) | Usable Channels |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Single | External device with DACK and external memory | External | B/C | 8/16/32 | 0, 1 |
|  | External device with DACK and memory-mapped external device | External | B/C | 8/16/32 | 0, 1 |
| Dual | External memory and external memory | Any*1 | B/C | 8/16/32 | $0-3 * 5$ |
|  | External memory and memory-mapped external device | Any*1 | B/C | 8/16/32 | $0-3 * 5$ |
|  | Memory-mapped external device and memory-mapped external device | Any*1 | B/C | 8/16/32 | $0-3 * 5$ |
|  | External memory and on-chip memory | Any*1 | B/C | 8/16/32 | $0-3$ *5 |
|  | External memory and on-chip peripheral module | Any*2 | B/C*3 | 8/16/32*4 | $0-3$ *5 |
|  | Memory-mapped external device and on-chip memory | Any*1 | B/C | 8/16/32 | $0-3 * 5$ |
|  | Memory-mapped external device and on-chip peripheral module | Any*2 | B/C*3 | 8/16/32*4 | $0-3 * 5$ |
|  | On-chip memory and on-chip memory | Any*1 | B/C | 8/16/32 | $0-3$ *5 |
|  | On-chip memory and on-chip peripheral module | Any*2 | B/C*3 | 8/16/32*4 | $0-3 * 5$ |
|  | On-chip peripheral module and onchip peripheral module | Any*2 | B/C*3 | 8/16/32*4 | $0-3 * 5$ |

Notes: *1 External request, auto-request or on-chip peripheral module request enabled. However, in the case of on-chip peripheral module request, it is not possible to specify the SCl or A/D converter for the transfer request source.
*2 External request, auto-request or on-chip peripheral module request possible. However, if transfer request source is also the SCl or A/D converter (A/D1 for A mask), the transfer source or transfer destination must be the SCl or A/D converter (A/D1 for A mask). For A mask, setting A/D0 as the transfer request source is not permitted.
*3 When the transfer request source is the SCI, only cycle steal mode is possible.
*4 Access size permitted by register of on-chip peripheral module that is the transfer source or transfer destination.
*5 When the transfer request is an external request, channels 0 and 1 only can be used.
*6 B: Burst, C: Cycle steal

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### 11.3.9 Bus Mode and Channel Priority Order

When a given channel is transferring in burst mode, and a transfer request is issued to channel 0 , which has a higher priority ranking, transfer on channel 0 begins immediately. If the priority level setting is fixed mode $(\mathrm{CH} 0>\mathrm{CH} 1)$, channel 1 transfer is continued after transfer on channel 0 are completely ended, whether the channel 0 setting is cycle steal mode or burst mode.

When the priority level setting is for round robin mode, transfer on channel 1 begins after transfer of one transfer unit on channel 0 , whether channel 0 is set to cycle steal mode or burst mode. Thereafter, bus right alternates in the order: channel $1>$ channel $0>$ channel $1>$ channel 0 . Whether the priority level setting is for fixed mode or round robin mode, since channel 1 is set to burst mode, the bus right is not given to the CPU. An example of round robin mode is shown in figure 11.14.


Priority: Round-robin mode
ch0: Cycle-steal mode
ch1: Burst mode
Figure 11.14 Bus Handling when Multiple Channels Are Operating

### 11.3.10 Number of Bus Cycle States and $\overline{\text { DREQ }}$ Pin Sample Timing

Number of States in Bus Cycle: The number of states in the bus cycle when the DMAC is the bus master is controlled by the bus state controller (BSC) just as it is when the CPU is the bus master. For details, see section 10, Bus State Controller (BSC).
$\overline{\text { DREQ }}$ Pin Sampling Timing and DRAK Signal: In external request mode, the $\overline{\mathrm{DREQ}} \mathrm{pin}$ is sampled by either falling edge or low-level detection. When a $\overline{\text { DREQ }}$ input is detected, a DMAC bus cycle is issued and DMA transfer effected, at the earliest, after three states. However, in burst mode when single address operation is specified, a dummy cycle is inserted for the first bus cycle. In this case, the actual data transfer starts from the second bus cycle. Data is transferred continuously from the second bus cycle. The dummy cycle is not counted in the number of transfer cycles, so there is no need to recognize the dummy cycle when setting the TCR.
$\overline{\text { DREQ }}$ sampling from the second time begins from the start of the transfer one bus cycle prior to the DMAC transfer generated by the previous sampling.

DRAK is output once for the first $\overline{\mathrm{DREQ}}$ sampling, irrespective of transfer mode or $\overline{\mathrm{DREQ}}$ detection method. In burst mode, using edge detection, $\overline{\mathrm{DREQ}}$ is sampled for the first cycle only, so DRAK is also output for the first cycle only. Therefore, the $\overline{\mathrm{DREQ}}$ signal negate timing can be ascertained, and this facilitates handshake operations of transfer requests with the DMAC.

Cycle Steal Mode Operations: In cycle steal mode, $\overline{\text { DREQ }}$ sampling timing is the same irrespective of dual or single address mode, or whether edge or low-level $\overline{\mathrm{DREQ}}$ detection is used.

For example, DMAC transfer begins (figure 11.15), at the earliest, three cycles from the first sampling timing. The second sampling begins at the start of the transfer one bus cycle prior to the start of the DMAC transfer initiated by the first sampling (i.e., from the start of the CPU(3) transfer). At this point, if DREQ detection has not occurred, sampling is executed every cycle thereafter.

As in figure 11.16, whatever cycle the CPU transfer cycle is, the next sampling begins from the start of the transfer one bus cycle before the DMAC transfer begins.

Figure 11.15 shows an example of output during DACK read and figure 11.16 an example of output during DACK write.


Figure 11.15 Cycle Steal, Dual Address, and Level Detection (Fastest Operation)


Figure 11.16 Cycle Steal, Dual Address, and Level Detection (Normal Operation)

Figures 11.17 and 11.18 show cycle steal mode and single address mode. In this case, transfer begins at earliest three cycles after the first $\overline{\mathrm{DREQ}}$ sampling. The second sampling begins from the start of the transfer one bus cycle before the start of the first DMAC transfer. In single address mode, the DACK signal is output during the DMAC transfer period.


Figure 11.17 Cycle Steal, Single Address, and Level Detection (Fastest Operation)


Figure 11.18 Cycle Steal, Single Address, and Level Detection (Normal Operation)

Burst Mode, Dual Address, and Level Detection: $\overline{\text { DREQ }}$ sampling timing in burst mode with dual address and level detection is virtually the same as that of cycle steal mode.

For example, DMAC transfer begins (figure 11.19), at the earliest, three cycles after the timing of the first sampling. The second sampling also begins from the start of the transfer one bus cycle before the start of the first DMAC transfer. In burst mode, as long as transfer requests are issued, DMAC transfer continues. Therefore, the "transfer one bus cycle before the start of the DMAC transfer" may be a DMAC transfer.

In burst mode, the DACK output period is the same as that of cycle steal mode. Figure 11.20 shows the normal operation of this burst mode.


Figure 11.19 Burst Mode, Dual Address, and Level Detection (Fastest Operation)


Figure 11.20 Burst Mode, Dual Address, and Level Detection (Normal Operation)

Burst Mode, Single Address, and Level Detection: $\overline{\text { DREQ }}$ sampling timing in burst mode with single address and level detection is shown in figures 11.21 and 11.22.

In burst mode with single address and level detection, a dummy cycle is inserted as one bus cycle, at the earliest, three cycles after timing of the first sampling. Data during this period is undefined, and the DACK signal is not output. Nor is the number of DMAC transfers counted. The actual DMAC transfer begins after one dummy bus cycle output.

The dummy cycle is not counted either at the start of the second sampling (transfer one bus cycle before the start of the first DMAC transfer). Therefore, the second sampling is not conducted from the bus cycle starting the dummy cycle, but from the start of the $\mathrm{CPU}(3)$ bus cycle.

Thereafter, as long the $\overline{\mathrm{DREQ}}$ is continuously sampled, no dummy cycle is inserted. $\overline{\mathrm{DREQ}}$ sampling timing during this period begins from the start of the transfer one bus cycle before the start of DMAC transfer, in the same way as with cycle steal mode.

As with the fourth sampling in figure 11.21 , once DMAC transfer is interrupted, a dummy cycle is again inserted at the start as soon as DMAC transfer is resumed.

The DACK output period in burst mode is the same as in cycle steal mode.


Figure 11.21 Burst Mode, Single Address, and Level Detection (Fastest Operation)


Figure 11.22 Burst Mode, Single Address, and Level Detection (Normal Operation)

Burst Mode, Dual Address, and Edge Detection: In burst mode with dual address and edge detection, $\overline{\mathrm{DREQ}}$ sampling is conducted only on the first cycle.

In figure 11.23, DMAC transfer begins, at the earliest, three cycles after the timing of the first sampling. Thereafter, DMAC transfer continues until the end of the data transfer count set in the TCR. $\overline{\text { DREQ }}$ sampling is not conducted during this period. Therefore, DRAK is output on the first cycle only.

When DMAC transfer is resumed after being halted by a NMI or address error, be sure to reinput an edge request. The remaining transfer restarts after the first DRAK output.

The DACK output period in burst mode is the same as in cycle steal mode.


Figure 11.23 Burst Mode, Dual Address, and Edge Detection

Burst Mode, Single Address, and Edge Detection: In burst mode with single address and edge detection, $\overline{\mathrm{DREQ}}$ sampling is conducted only on the first cycle. In figure 11.24, a dummy cycle is inserted, at the earliest, three cycles after the timing for the first sampling. During this period, data is undefined, and DACK is not output. Nor is the number of DMAC transfers counted. Thereafter, DMAC transfer continues until the data transfer count set in the DMATCR has ended. $\overline{\mathrm{DREQ}}$ sampling is not conducted during this period. Therefore, DRAK is output on the first cycle only.

When DMAC transfer is resumed after being halted by a NMI or address error, be sure to reinput an edge request. DRAK is output once, and the remaining transfer restarts after output of one dummy cycle.

The DACK output period in burst mode is the same as in cycle steal mode.


Figure 11.24 Burst Mode, Single Address and Edge Detection

### 11.3.11 Source Address Reload Function

Channel 2 has a source address reload function. This returns to the first value set in the source address register (SAR2) every four transfers by setting the RO bit of CHCR2 to 1 . Figure 11.25 illustrates this operation. Figure 11.26 is a timing chart for reload ON mode, with burst mode, autorequest, 16-bit transfer data size, SAR2 increment, and DAR2 fixed mode.


Figure 11.25 Source Address Reload Function


Figure 11.26 Source Address Reload Function Timing Chart

The reload function can be executed whether the transfer data size is 8,16 , or 32 bits.
DMATCR2, which specifies the number of transfers, is decremented by 1 at the end of every single-transfer-unit transfer, regardless of whether the reload function is on or off. Therefore, when using the reload function in the on state, a multiple of 4 must be specified in DMATCR2. Operation will not be guaranteed if any other value is set. Also, the counter which counts the occurrence of four transfers for address reloading is reset by clearing of the DME bit in DMAOR or the DE bit in CHCR2, setting of the transfer end flag (the TE bit in CHCR2), NMI input, and setting of the AE flag (address error generation in DMAC transfer), as well as by a reset and in software standby mode, but SAR2, DAR2, DMATCR2, and other registers are not reset. Consequently, when one of these sources occurs, there is a mixture of initialized counters and uninitialized registers in the DMAC, and incorrect operation may result if a restart is executed in this state. Therefore, when one of the above sources, other than TE setting, occurs during use of the address reload function, SAR, DAR2, and DMATCR2 settings must be carried out before reexecution.

### 11.3.12 DMA Transfer Ending Conditions

The DMA transfer ending conditions vary for individual channels ending and for all channels ending together.

Individual Channel Ending Conditions: There are two ending conditions. A transfer ends when the value of the channel's DMA transfer count register (DMATCR) is 0 , or when the DE bit of the channel's CHCR is cleared to 0 .

- When DMATCR is 0 : When the DMATCR value becomes 0 and the corresponding channel's DMA transfer ends, the transfer end flag bit (TE) is set in the CHCR. If the IE (interrupt enable) bit has been set, a DMAC interrupt (DEI) is requested of the CPU.
- When DE of CHCR is 0: Software can halt a DMA transfer by clearing the DE bit in the channel's CHCR. The TE bit is not set when this happens.

Conditions for Ending All Channels Simultaneously: Transfers on all channels end when the NMIF (NMI flag) bit or AE (address error flag) bit is set to 1 in the DMAOR, or when the DME bit in the DMAOR is cleared to 0 .

- When the NMIF or AE bit is set to 1 in DMAOR: When an NMI interrupt or DMAC address error occurs, the NMIF or AE bit is set to 1 in the DMAOR and all channels stop their transfers. The DMAC obtains the bus rights, and if these flags are set to 1 during execution of a transfer, DMAC halts operation when the transfer processing currently being executed ends, and transfers the bus right to the other bus master. Consequently, even if the NMIF or AE bits are set to 1 during a transfer, the DMA source address register (SAR), designation address register (DAR), and transfer count register (TCR) are all updated. The TE bit is not set. To resume the transfers after NMI interrupt or address error processing, clear the appropriate flag bit to 0 . To avoid restarting a transfer on a particular channel, clear its DE bit to 0 .

When the processing of a one unit transfer is complete. In a dual address mode direct address transfer, even if an address error occurs or the NMI flag is set during read processing, the transfer will not be halted until after completion of the following write processing. In such a case, SAR, DAR, and TCR values are updated. In the same manner, the transfer is not halted in dual address mode indirect address transfers until after the final write processing has ended.

- When DME is cleared to 0 in DMAOR: Clearing the DME bit to 0 in the DMAOR aborts the transfers on all channels. The TE bit is not set.


### 11.3.13 DMAC Access from CPU

The space addressed by the DMAC is 3-cycle space. Therefore, when the CPU becomes the bus master and accesses the DMAC, a minimum of three basic clock (CLK) cycles are required for one bus cycle. Also, since the DMAC is located in word space, while a word-size access to the DMAC is completed in one bus cycle, a longword-size access is automatically divided into two word accesses, requiring two bus cycles (six basic clock cycles). These two bus cycles are executed consecutively; a different bus cycle is never inserted between the two word accesses. This applies to both write accesses and read accesses.

### 11.4 Examples of Use

### 11.4.1 Example of DMA Transfer between On-Chip SCI and External Memory

In this example, on-chip serial communication interface channel 0 (SCI0) received data is transferred to external memory using the DMAC channel 3.

Table 11.7 indicates the transfer conditions and the setting values of each of the registers.
Table 11.7 Transfer Conditions and Register Set Values for Transfer between On-Chip SCI and External Memory

| Transfer Conditions | Register | Value |
| :---: | :---: | :---: |
| Transfer source: RDR0 of on-chip SCIO | SAR3 | H'FFFF81A5 |
| Transfer destination: external memory | DAR3 | H'00400000 |
| Transfer count: 64 times | DMATCR3 | H'00000040 |
| Transfer source address: fixed | CHCR3 | H'00004D05 |
| Transfer destination address: incremented |  |  |
| Transfer request source: SCIO (RDRO) |  |  |
| Bus mode: cycle steal |  |  |
| Transfer unit: byte |  |  |
| Interrupt request generation at end of transfer |  |  |
| Channel priority ranking: $0>1>2>3$ | DMAOR | H'0001 |

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### 11.4.2 Example of DMA Transfer between External RAM and External Device with DACK

In this example, an external request, serial address mode transfer with external memory as the transfer source and an external device with DACK as the transfer destination is executed using DMAC channel 1.

Table 11.8 indicates the transfer conditions and the setting values of each of the registers.
Table 11.8 Transfer Conditions and Register Set Values for Transfer between External RAM and External Device with DACK

| Transfer Conditions | Register | Value |
| :---: | :---: | :---: |
| Transfer source: external RAM | SAR1 | H'00400000 |
| Transfer destination: external device with DACK | DAR1 | (access by DACK) |
| Transfer count: 32 times | DMATCR1 | H'00000020 |
| Transfer source address: decremented | CHCR1 | H'00002269 |
| Transfer destination address: (setting ineffective) |  |  |
| Transfer request source: external pin (㽬EQ1) edge detection |  |  |
| Bus mode: burst |  |  |
| Transfer unit: word |  |  |
| No interrupt request generation at end of transfer |  |  |
| Channel priority ranking: $2>0>1>3$ | DMAOR | H'0201 |

### 11.4.3 Example of DMA Transfer between A/D Converter and On-Chip Memory (Address Reload On) (Excluding A Mask)

In this example, the on-chip $A / D$ converter channel 0 is the transfer source and on-chip memory is the transfer destination, and the address reload function is on.

Table 11.9 indicates the transfer conditions and the setting values of each of the registers.

Table 11.9 Transfer Conditions and Register Set Values for Transfer between A/D Converter and On-Chip Memory

| Transfer Conditions | Register | Value |
| :--- | :--- | :--- |
| Transfer source: on-chip A/D converter ch0 | SAR2 | H'FFFF83F0 |
| Transfer destination: on-chip memory | DAR2 | H'FFFFF000 $^{\text {Transfer count: } 128 \text { times (reload count 32 times) }}$ |
| DMATCR2 | H'00000080 $^{\text {Transfer source address: incremented }}$ | CHCR2 |
| Transfer destination address: incremented |  | H'00085B25 $^{\text {Transfer request source: A/D converter }}$ |
| Bus mode: burst |  |  |
| Transfer unit: byte |  |  |
| Interrupt request generation at end of transfer |  | H'0101 |
| Channel priority ranking: $0>2>3>1$ | DMAOR |  |

When address reload is on, the SAR value returns to its initially established value every four transfers. In the above example, when a transfer request is input from the $\mathrm{A} / \mathrm{D}$ converter, the byte size data is first read in from the H'FFFF83F0 register of AD0 and that data is written to the onchip memory address H'FFFFF001. Because a byte size transfer was performed, the SAR and DAR values at this point are H'FFFF83F1 and H'FFFFF001, respectively. Also, because this is a burst transfer, the bus rights remain secured, so continuous data transfer is possible.

When four transfers are completed, if the address reload is off, execution continues with the fifth and sixth transfers and the SAR value continues to increment from H'FFFF83F3 to H'FFFF83F4 to H'FFFF83F5 and so on. However, when the address reload is on, the DMAC transfer is halted upon completion of the fourth one and the bus right request signal to the CPU is cleared. At this time, the value stored in SAR is not H'FFFF83F3-H'FFFF83F4, but H'FFFF83F3-H'FFFF83F0, a return to the initially established address. The DAR value always continues to be decremented regardless of whether the address reload is on or off.

The DMAC internal status, due to the above operation after completion of the fourth transfer, is indicated in table 11.10 for both address reload on and off.

Table 11.10 DMAC Internal Status

| Item | Address Reload On | Address Reload Off |
| :--- | :--- | :--- |
| SAR | H'FFFF83F0 $^{\prime}$ | H'FFFF83F4 |
| DAR | H'FFFFF004 $^{\prime}$ H'0000007C | H'FFFFF004 |
| DMATCR | Released | H'$^{\prime} 0000007 \mathrm{C}$ |
| Bus rights | Halted | Maintained |
| DMAC operation | Not issued | Processing continues |
| Interrupts | Not issued |  |
| Transfer request source flag clear | Executed | Not executed |

Notes: 1. Interrupts are executed until the DMATCR value becomes 0 , and if the IE bit of the CHCR is set to 1 , are issued regardless of whether the address reload is on or off.
2. If transfer request source flag clears are executed until the DMATCR value becomes 0 , they are executed regardless of whether the address reload is on or off.
3. Designate burst mode when using the address reload function. There are cases where abnormal operation will result if it is executed in cycle steal mode.
4. Designate a multiple of four for the TCR value when using the address reload function. There are cases where abnormal operation will result if anything else is designated.

To execute transfers after the fifth one when the address reload is on, make the transfer request source issue another transfer request signal.

### 11.4.4 Example of DMA Transfer between A/D Converter and Internal Memory (Address Reload On) (A Mask)

In this example the on-chip $A / D$ converter (A/D1) is the transfer source and the internal memory is the transfer destination, and the address reload on.

Table 11.11 indicates the transfer conditions and the setting values of each of the registers.

Table 11.11 Transfer Conditions and Register Set Values for Transfer between A/D Converter (A/D1) and Internal Memory

| Transfer Conditions | Register | Value |
| :---: | :---: | :---: |
| Transfer source: on-chip A/D converter (A/D1) | SAR2 | H'FFFF8408 |
| Transfer destination: internal memory | DAR2 | H'FFFFF000 |
| Transfer count: 128 times (reload count 32 times) | DMATCR2 | H'00000080 |
| Transfer source address: incremented | CHCR2 | H'00085B25 |
| Transfer target address: incremented |  |  |
| Transfer request source: A/D converter (A/D1) |  |  |
| Bus mode: burst |  |  |
| Transfer unit: byte |  |  |
| Interrupt request generated at end of transfer |  |  |
| Channel priority sequence: $0>2>3>1$ | DMAOR | H'0101 |

When address reload is on, the SAR value returns to its initially established value every four transfers. In the above example, when a transfer request is input from the A/D converter (A/D1), the byte size data is first read in from the H'FFFF8408 register and that data is written to the onchip memory address H'FFFFF001. Because a byte size transfer was performed, the SAR and DAR values at this point are H'FFFF8409 and H'FFFFF001, respectively. Also, because this is a burst transfer, the bus rights remain secured, so continuous data transfer is possible.

When four transfers are completed, if the address reload is off, execution continues with the fifth and sixth transfers and the SAR value continues to increment from H'FFFF840B to H'FFFF840C to H'FFFF840D and so on. However, when the address reload is on, the DMAC transfer is halted upon completion of the fourth transfer and the bus right request signal to the CPU is cleared. At this time, the values stored in SAR are not H'FFFF840B-H'FFFF840C, but H'FFFF840BH'FFFF8408, a return to the initially established address. The DAR value always continues to be decremented regardless of whether the address reload is on or off.

The DMAC internal status, due to the above operation after completion of the fourth transfer, is indicated in table 11.12 for both address reload on and off.

Table 11.12 DMAC Internal Status

| Item | Address Reload On | Address Reload Off |
| :---: | :---: | :---: |
| SAR | H'FFFF8408 | H'FFFF840C |
| DAR | H'FFFFF004 | H'FFFFF004 |
| DMATCR | H'0000007C | H'0000007C |
| Bus rights | Released | Maintained |
| DMAC operation | Halted | Processing continues |
| Interrupts | Not issued | Not issued |
| Transfer request source flag clear | Executed | Not executed |

Notes: 1. Interrupts are executed until the DMATCR value becomes 0 , and if the IE bit of the CHCR is set to 1 , are issued regardless of whether the address reload is on or off.
2. If transfer request source flag clears are executed until the DMATCR value becomes 0 , they are executed regardless of whether the address reload is on or off.
3. Designate burst mode when using the address reload function. There are cases where abnormal operation will result if it is executed in cycle steal mode.
4. Designate a multiple of four for the TCR value when using the address reload function. There are cases where abnormal operation will result if anything else is designated.

To execute more than four transfers with the address reload on, make the transfer request source issue another transfer request signal.

### 11.4.5 Example of DMA Transfer between External Memory and SCI1 Send Side (Indirect Address On)

In this example, DMAC channel 3 is used, an indirect address designated external memory is the transfer source and the SCI1 sending side is the transfer destination.

Table 11.13 indicates the transfer conditions and the setting values of each of the registers.

Table 11.13 Transfer Conditions and Register Set Values for Transfer between External Memory and SCI1 Sending Side

| Transfer Conditions | Register | Value |
| :---: | :---: | :---: |
| Transfer source: external memory | SAR3 | H'00400000 |
| Value stored in address H'00400000 | - | H'00450000 |
| Value stored in address H'00450000 | - | H'55 |
| Transfer destination: on-chip SCI TDR1 | DAR3 | H'FFFF81B3 |
| Transfer count: 10 times | DMATCR3 | H'0000000A |
| Transfer source address: incremented | CHCR3 | H'00011E01 |
| Transfer destination address: fixed |  |  |
| Transfer request source: SCI1 (TDR1) |  |  |
| Bus mode: cycle steal |  |  |
| Transfer unit: byte |  |  |
| Interrupt request not generated at end of transfer |  |  |
| Channel priority ranking: $0>1>2>3$ | DMAOR | H'0001 |

When indirect address mode is on, the data stored in the address established in SAR is not used as the transfer source data. In the case of indirect addressing, the value stored in the SAR address is read, then that value is used as the address and the data read from that address is used as the transfer source data, then that data is stored in the address designated by the DAR.

In the table 11.13 example, when a transfer request from the TDR1 of SCI1 is generated, a read of the address located at $\mathrm{H}^{\prime} 00400000$, which is the value set in SAR3, is performed first. The data $H^{\prime} 00450000$ is stored at this H'00400000 address, and the DMAC first reads this H'00450000 value. It then uses this read value of $\mathrm{H}^{\prime} 00450000$ as an address and reads the value of $\mathrm{H}^{\prime} 55$ that is stored in the H'00450000 address. It then writes the value H'55 to the address H'FFFF81B3 designated by DAR3 to complete one indirect address transfer.

With indirect addressing, the first executed data read from the address established in SAR3 always results in a longword size transfer regardless of the TS0, TS1 bit designations for transfer data size. However, the transfer source address fixed and increment or decrement designations are as according to the SM0, SM1 bits. Consequently, despite the fact that the transfer data size designation is byte in this example, the SAR3 value at the end of one transfer is $\mathrm{H}^{\prime} 00400004$. The write operation is exactly the same as an ordinary dual address transfer write operation.

## Renesns

### 11.5 Cautions on Use

1. Other than the DMA operation register (DMAOR) accessing in word (16-bit) units, access all registers in word (16-bit) or longword (32-bit) units.
2. When rewriting the RS0-RS3 bits of CHCR0-CHCR3, first clear the DE bit to 0 (set the DE bit to 0 before doing rewrites with a CHCR byte address).
3. When an NMI interrupt is input, the NMIF bit of the DMAOR is set even when the DMAC is not operating.
4. Set the DME bit of the DMAOR to 0 and make certain that any DMAC received transfer request processing has been completed before entering standby mode.
5. Do not access the DMAC, DTC, BSC, or UBC on-chip peripheral modules from the DMAC.
6. When activating the DMAC, do the CHCR or DMAOR setting as the final step. There are instances where abnormal operation will result if any other registers are established last.
7. After the DMATCR count becomes 0 and the DMA transfer ends normally, always write a 0 to the DMATCR, even when executing the maximum number of transfers on the same channel. There are instances where abnormal operation will result if this is not done.
8. Designate burst mode as the transfer mode when using the address reload function. There are instances where abnormal operation will result in cycle steal mode.
9. Designate a multiple of four for the DMATCR value when using the address reload function. There are instances where abnormal operation will result if anything else is designated.
10. When detecting external requests by falling edge, maintain the external request pin at high level when performing the DMAC establishment.
11. When operating in single address mode, establish an external address as the address. There are instances where abnormal operation will result if an internal address is established.
12. Do not access DMAC register empty addresses (H'FFFF86B2-H'FFFF86BF). Operation cannot be guaranteed when empty addresses are accessed.

## Section 12 Multifunction Timer Pulse Unit (MTU)

### 12.1 Overview

The SuperH microprocessor has an on-chip 16-bit multifunction timer pulse unit (MTU) with five channels of 16-bit timers.

### 12.1.1 Features

- Can process a maximum of sixteen different pulse outputs and inputs.
- Has sixteen timer general registers (TGR): four each for channels 0,3 , and 4, and two each for channels 1 and 2 that can be set to function independently as output compare or input capture. The channel 0,3 , and 4 TGRC and TGRD registers can be used as buffer registers.
- Can select eight counter input clock sources for all channels
- All channels can be set for the following operating modes:
- Compare match waveform output: 0 output/1 output/toggle output selectable.
- Input capture function: Selectable rising edge, falling edge, or both rising and falling edge detection.
- Counter clearing function: Counters can be cleared by a compare-match or input capture.
- Synchronizing mode: Two or more timer counters (TCNT) can be written to simultaneously. Two or more timer counters can be simultaneously cleared by a comparematch or input capture. Counter synchronization functions enable synchronized register input/output.
- PWM mode: PWM output can be provided with any duty cycle. When combined with the counter synchronizing function, enables up to twelve-phase PWM output. (With channels $0-2$ set to PWM mode 2, channels $3-4$, and channels $0-4$ synchronized with TGR3A of channel 3 as the sync register (channels $0-4$ phase output: $4,2,2,2,2$ ).)
- Channels 0,3 , and 4 can be set for buffer operation
- Input capture register double buffer configuration possible
- Output compare register automatic re-write possible
- Channels 1 and 2 can be independently set to the phase counting mode
- Two-phase encoder pulse up/down count possible
- Cascade connection operation
- Can be operated as a 32-bit counter by using the channel 2 input clock for channel 1 overflow/underflow
- Channels 3 and 4 can be set in the following modes:
- Reset-synchronized PWM mode: By combining channels 3 and 4, a sawtooth wave comparator type six-phase PWM waveform can be output.


## Renesns

- Complementary PWM mode: By combining channels 3 and 4, a triangle wave comparator type six-phase PWM output is possible with non-overlapping times.
- High speed access via internal 16-bit bus
- Twenty-three interrupt sources
- Channels 0,3 , and 4 have four compare-match/input capture interrupts and one overflow interrupt which can be requested independently.
- Channels 1 and 2 have two compare-match/input capture interrupts, one overflow interrupt, and one underflow interrupt which can be requested independently.
- Automatic transfer of register data

Block transfer, 1-word data transfers and 1-byte data transfers are possible through DTC or DMAC activation.

- A/D converter conversion start trigger can be generated
- Channels 0-4 compare-match/input capture signals can be used as A/D converter conversion start triggers.

Table 12.1 summarizes the MTU functions.
Table 12.1 MTU Functions

| Item | Channel 0 | Channel 1 | Channel 2 | Channel 3 | Channel 4 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Counter clocks | Internal: $\phi / 1, \phi / 4, \phi / 16, \phi / 64, \phi / 256, \phi / 1024$ |  |  |  |  |
| General registers | TGROA | TGR1A | TGR2A | TGR3A | TGR4A |
|  | TGR0B | TGR1B | TGR2B | TGR3B | TGR4B |
| General registers/buffer registers | TGROC TGROD | No | No | TGR3C TGR3D | $\begin{aligned} & \text { TGR4C } \\ & \text { TGR4D } \end{aligned}$ |
| Input/output pins | TIOCOA | TIOC1A | TIOC2A | TIOC3A | TIOC4A |
|  | TIOCOB | TIOC1B | TIOC2B | TIOC3B | TIOC4B |
|  | TIOCOC |  |  | TIOC3C | TIOC4C |
|  | TIOCOD |  |  | TIOC3D | TIOC4D |
| Counter clear function | TGR compare- TGR compare- TGR compare- TGR compare- TGR comparematch or input match or input match or input match or input match or input capture capture capture capture capture |  |  |  |  |
| Compare 0 match output 1 | Yes | Yes | Yes | Yes | Yes |
|  | Yes | Yes | Yes | Yes | Yes |
|  | Toggle Yes | Yes | Yes | Yes | Yes |
| Input capture function | Yes | Yes | Yes | Yes | Yes |
| Synchronization | Yes | Yes | Yes | Yes | Yes |
| Buffer operation | Yes | No | No | Yes | No |
| PWM mode 1 | Yes | Yes | Yes | Yes | Yes |
| PWM mode 2 | Yes | Yes | Yes | No | No |
| Phase counting mode | No | Yes | Yes | No | No |
| Reset-synchronized PWM mode | No | No | No | Yes | Yes |
| Complementary PWM mode | No | No | No | Yes | Yes |
| DMAC activation | TGROA compare match or input capture | TGR1A compare match or input capture | TGR2A compare match or input capture | TGR3A compare match or input capture | TGR4A compare match or input capture |

Table 12.1 MTU Functions (cont)

| Item | Channel 0 | Channel 1 | Channel 2 | Channel 3 | Channel 4 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Hard DTC activation | TGR comparematch or input capture | TGR compare match or input capture | TGR compare match or input capture | - TGR compare match or input capture | - TGR comparematch or input capture and TCNT4 overflow/ underflow |
| A/D conversion start trigger | TGROA compare match or input capture | TGR1A compare match or input capture | TGR2A compare match or input capture | TGR3A compare match or input capture | TGR4A compare match or input capture |
| Interrupt sources | Compare match/input capture 0A | Compare match/input capture 1A | Compare match/input capture 2A | Compare match/input capture 3A | Compare match/input capture 4A |
|  | Compare match/input capture 0B | Compare match/input capture 1B | Compare match/input capture 2B | Compare match/input capture 3B | Compare match/input capture 4B |
|  | Compare match/input capture 0C | Overflow | Overflow | Compare match/input capture 3C | Compare match/input capture 4C |
|  | Compare match/input capture 0D | Underflow | Underflow | Compare match/input capture 3D | Compare match/input capture 4D |
|  | Overflow | - | - | Overflow | Overflow/ underflow |

### 12.1.2 Block Diagram

Figure 12.1 is the block diagram of the MTU.


Figure 12.1 MTU Block Diagram

### 12.1.3 Pin Configuration

Table 12.2 summarizes the MTU pins.
Table 12.2 Pin Configuration

| Channel | Name | Pin Name | I/O | Function |
| :---: | :---: | :---: | :---: | :---: |
| Shared | Clock input A | TCLKA | 1 | Clock A input pin (A-phase input pin in channel 1 phase counting mode) |
|  | Clock input B | TCLKB | I | Clock B input pin (B-phase input pin in channel 1 phase counting mode) |
|  | Clock input C | TCLKC | I | Clock C input pin (A-phase input pin in channel 2 phase counting mode) |
|  | Clock input D | TCLKD |  | Clock D input pin (B-phase input pin in channel 2 phase counting mode) |
| 0 | Input capture/output compare-match 0A | TIOCOA | I/O | TGROA input capture input/output compare output/PWM output pin |
|  | Input capture/output compare-match 0B | TIOCOB | I/O | TGROB input capture input/output compare output/PWM output pin |
|  | Input capture/output compare-match 0C | TIOCOC | I/O | TGROC input capture input/output compare output/PWM output pin |
|  | Input capture/output compare-match 0D | TIOCOD | I/O | TGROD input capture input/output compare output/PWM output pin |
| 1 | Input capture/output compare-match 1A | TIOC1A | I/O | TGR1A input capture input/output compare output/PWM output pin |
|  | Input capture/output compare-match 1B | TIOC1B | I/O | TGR1B input capture input/output compare output/PWM output pin |
| 2 | Input capture/output compare-match 2A | TIOC2A | I/O | TGR2A input capture input/output compare output/PWM output pin |
|  | Input capture/output compare-match 2B | TIOC2B | I/O | TGR2B input capture input/output compare output/PWM output pin |

Table 12.2 Pin Configuration (cont)

| Channel | Name | Pin Name | I/O | Function |
| :---: | :---: | :---: | :---: | :---: |
| 3 | Input capture/output compare-match 3A | TIOC3A | I/O | TGR3A input capture input/output compare output/PWM output pin <br> In complementary PWM/reset synchronous PWM mode, 1/2 PWM period toggle output pin |
|  | Input capture/output compare-match 3B | TIOC3B | I/O | TGR3B input capture input/output compare output pin <br> In complementary PWM/reset synchronous PWM mode, PWM output/U phase output pin |
|  | Input capture/output compare-match 3C | TIOC3C | I/O | TGR3C input capture input/output compare output/PWM output pin <br> In complementary PWM/reset synchronous PWM mode |
|  | Input capture/output compare-match 3D | TIOC3D | I/O | TGR3D input capture input/output compare output pin <br> In complementary PWM/reset synchronous PWM mode, PWM output/U phase output pin |
| 4 | Input capture/output compare-match 4A | TIOC4A | I/O | TGR4A input capture input/output compare output/PWM output pin <br> In complementary PWM/reset synchronous PWM mode, PWM output/V phase output pin |
|  | Input capture/output compare-match 4B | TIOC4B | I/O | TGR4B input capture input/output compare output pin <br> In complementary PWM/reset synchronous PWM mode, PWM output/W phase output pin |
|  | Input capture/output compare-match 4C | TIOC4C | I/O | TGR4C input capture input/output compare output/PWM output pin <br> In complementary PWM/reset synchronous PWM mode, PWM output/V phase output pin |
|  | Input capture/output compare-match 4D | TIOC4D | I/O | TGR4D input capture input/output compare output pin <br> In complementary PWM/reset synchronous PWM mode, PWM output/W phase output pin |

Note: The TIOC pins output undefined values when they are set to input capture and timer output by the pin function controller (PFC).

### 12.1.4 Register Configuration

Table 12.3 summarizes the MTU register configuration.
Table 12.3 Register Configuration

| Channel | Name | Abbreviation | R/W | Initial Value | Address | Access Size (Bits) *1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Shared | Timer start register | TSTR | R/W | H'00 | H'FFFF8240 | 8, 16, 32 |
|  | Timer synchro register | TSYR | R/W | H'00 | H'FFFF8241 |  |
| 0 | Timer control register 0 | TCR0 | R/W | H'00 | H'FFFF8260 |  |
|  | Timer mode register 0 | TMDR0 | R/W | H'C0 | H'FFFF8261 |  |
|  | Timer I/O control register OH | TIOROH | R/W | H'00 | H'FFFF8262 |  |
|  | Timer I/O control register OL | TIOR0L | R/W | H'00 | H'FFFF8263 |  |
|  | Timer interrupt enable register 0 | TIERO | R/W | H'40 | H'FFFF8264 |  |
|  | Timer status register 0 | TSR0 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'C0 | H'FFFF8265 |  |
|  | Timer counter 0 | TCNT0 | R/W | H'0000 | H'FFFF8266 | 16, 32 |
|  | General register 0A | TGR0A | R/W | H'FFFF | H'FFFF8268 |  |
|  | General register 0B | TGR0B | R/W | H'FFFF | H'FFFF826A |  |
|  | General register 0C | TGR0C | R/W | H'FFFF | H'FFFF826C |  |
|  | General register 0D | TGR0D | R/W | H'FFFF | H'FFFF826E |  |
| 1 | Timer control register 1 | TCR1 | R/W | H'00 | H'FFFF8280 | 8, 16, 32 |
|  | Timer mode register 1 | TMDR1 | R/W | H'C0 | H'FFFF8281 |  |
|  | Timer I/O control register 1 | TIOR1 | R/W | H'00 | H'FFFF8282 |  |
|  | Timer interrupt enable register 1 | TIER1 | R/W | H'40 | H'FFFF8284 |  |
|  | Timer status register 1 | TSR1 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'C0 | H'FFFF8285 |  |
|  | Timer counter 1 | TCNT1 | R/W | H'0000 | H'FFFF8286 | 16, 32 |
|  | General register 1A | TGR1A | R/W | H'FFFF | H'FFFF8288 |  |
|  | General register 1B | TGR1B | R/W | H'FFFF | H'FFFF828A |  |

Table 12.3 Register Configuration (cont)

| Channel | Name | Abbreviation | R/W | Initial Value | Address | Access Size (Bits) ${ }^{* 1}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2 | Timer control register 2 | TCR2 | R/W | H'00 | H'FFFF82A0 | 8, 16, 32 |
|  | Timer mode register 2 | TMDR2 | R/W | H'C0 | H'FFFF82A1 |  |
|  | Timer I/O control register 2 | TIOR2 | R/W | H'00 | H'FFFF82A2 |  |
|  | Timer interrupt enable register 2 | TIER2 | R/W | H'40 | H'FFFF82A4 |  |
|  | Timer status register 2 | TSR2 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'C0 | H'FFFF82A5 |  |
|  | Timer counter 2 | TCNT2 | R/W | H'0000 | H'FFFF82A6 | 16, 32 |
|  | General register 2A | TGR2A | R/W | H'FFFF | H'FFFF82A8 |  |
|  | General register 2B | TGR2B | R/W | H'FFFF | H'FFFF82AA |  |
| 3 | Timer control register 3 | TCR3 | R/W*3 | H'00 | H'FFFF8200 | 8, 16, 32 |
|  | Timer mode register 3 | TMDR3 | R/W*3 | H'C0 | H'FFFF8202 |  |
|  | Timer I/O control register 3 H | TIOR3H | R/W*3 | H'OO | H'FFFF8204 |  |
|  | Timer I/O control register 3L | TIOR3L | $\mathrm{R} / \mathrm{W}^{* 3}$ | H'OO | H'FFFF8205 |  |
|  | Timer interrupt enable register 3 | TIER3 | R/W*3 | H'40 | H'FFFF8208 |  |
|  | Timer status register 3 | TSR3 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'C0 | H'FFFF822C | 8, 16, 32 |
|  | Timer counter 3 | TCNT3 | R/W*3 | H'0000 | H'FFFF8210 | 16, 32 |
|  | General register 3A | TGR3A | R/W*3 | H'FFFF | H'FFFF8218 |  |
|  | General register 3B | TGR3B | R/W*3 | H'FFFF | H'FFFF821A |  |
|  | General register 3C | TGR3C | R/W | H'FFFF | H'FFFF8224 | 16, 32 |
|  | General register 3D | TGR3D | R/W | H'FFFF | H'FFFF8226 |  |
| 4 | Timer control register 4 | TCR4 | R/W*3 | H'OO | H'FFFF8201 | 8, 16, 32 |
|  | Timer mode register 4 | TMDR4 | R/W*3 | H'C0 | H'FFFF8203 |  |
|  | Timer I/O control register 4H | TIOR4H | R/W*3 | H'OO | H'FFFF8206 |  |
|  | Timer I/O control register 4L | TIOR4L | R/W*3 | H'OO | H'FFFF8207 |  |
|  | Timer interrupt enable register 4 | TIER4 | R/W*3 | H'40 | H'FFFF8209 |  |
|  | Timer status register 4 | TSR4 | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'C0 | H'FFFF822D | 8, 16, 32 |

Table 12.3 Register Configuration (cont)

| Channel | Name | Abbreviation | R/W | Initial Value | Address | Access Size (Bits) ${ }^{* 1}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 4 (cont) | Timer counter 4 | TCNT4 | R/W*3 | H'0000 | H'FFFF8212 | 16, 32 |
|  | General register 4A | TGR4A | R/W*3 | H'FFFF | H'FFFF821C |  |
|  | General register 4B | TGR4B | R/W*3 | H'FFFF | H'FFFF821E |  |
|  | General register 4C | TGR4C | R/W | H'FFFF | H'FFFF8228 | 16, 32 |
|  | General register 4D | TGR4D | R/W | H'FFFF | H'FFFF822A |  |
| 3 and 4 | Timer output master enable register | TOER | R/W*3 | $\mathrm{H}^{\prime} \mathrm{CO}$ | H'FFFF820A | 8, 16, 32 |
|  | Timer output control register | TOCR | R/W*3 | H'00 | H'FFFF820B |  |
|  | Timer gate control register | TGCR | R/W*3 | H'80 | H'FFFF820D |  |
|  | Timer cycle data register | TCDR | R/W*3 | H'FFFF | H'FFFF8214 | 16, 32 |
|  | Timer dead time data register | TDDR | R/W*3 | H'FFFF | H'FFFF8216 |  |
|  | Timer subcounter | TCNTS | R | H'0000 | H'FFFF8220 | 16, 32 |
|  | Timer cycle buffer register | TCBR | R/W | H'FFFF | H'FFFF8222 |  |

Notes: *1 16-bit registers (TCNT, TGR) cannot be read or written in 8-bit units.
*2 Write 0 to clear flags.
*3 If the MTURWE bit of bus control register 1 (BCR) in the bus state controller (BSC) is 0 cleared, access becomes impossible (undefined read/write disabled).

### 12.2 MTU Register Descriptions

### 12.2.1 Timer Control Register (TCR)

The TCR is an 8-bit read/write register for controlling the TCNT counter for each channel. The MTU has five TCR registers, one for each of the channels 0 to 4 . TCR is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset or the standby mode. Manual reset does not initialize TCR.

## Channels 0, 3, 4: TCR0, TCR3, TCR4:



Channels 1, 2: TCR1, TCR2:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | CCLR1 | CCLR0 | CKEG1 | CKEG0 | TPSC2 | TPSC1 | TPSC0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bits 7-5-Counter Clear 2, 1, 0 (CCLR2, CCLR1, CCLR0): Select the counter clear source for the TCNT counter.


## Channels 0, 3, 4:

| Bit 7: CCLR2 | Bit 6: CCLR1 | Bit 5: CCLRO | Description |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | TCNT clear disabled (initial value) |
|  |  | 1 | TCNT is cleared by TGRA compare-match or input capture |
|  | 1 | 0 | TCNT is cleared by TGRB compare-match or input capture |
|  |  | 1 | Synchronizing clear: TCNT is cleared in synchronization with clear of other channel counters operating in sync.*1 |
| 1 | 0 | 0 | TCNT clear disabled |
|  |  | 1 | TCNT is cleared by TGRC compare-match or input capture*2 |
|  | 1 | 0 | TCNT is cleared by TGRD compare-match or input capture*2 |
|  |  | 1 | Synchronizing clear: TCNT is cleared in synchronization with clear of other channel counters operating in sync*1 |

Notes: *1 Setting the SYNC bit of the TSYR to 1 sets the synchronization.

* 2 When TGRC or TGRD are functioning as buffer registers, TCNT is not cleared because the buffer registers have priority and compare-match/input captures do not occur.

Channels 1, 2:

| Bit 7: <br> Reserved | Bit 6: <br> CCLR1 | Bit 5: <br> CCLR0 | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | TCNT clear disabled (initial value) |
|  | 1 | 0 | TCNT is cleared by TGRA compare-match or input capture |
|  | 1 | TCNT is cleared by TGRB compare-match or input capture <br> clear of other channel counters operating in sync*2 |  |

Notes: *1 The bit 7 of channels 1 and 2 is reserved. It always reads 0 , and cannot be modified.
*2 Setting the SYNC bit of the TSYR to 1 sets the synchronization.

- Bits 4-3-Clock Edge 1, 0 (CKEG1 and CKEG0): CKEG1 and CKEG0 select the input clock edges. When counting is done on both edges of the internal clock the input clock frequency becomes $1 / 2$ (Example: both edges of $\phi / 4=$ rising edge of $\phi / 2$ ). When phase count mode is used with channels 1,2 , these settings are ignored, as the phase count mode settings have priority.

Bit 4: Bit 3:
CKEG1 CKEG0 Description

| 0 | 0 | Count on rising edges (initial value) |
| :--- | :--- | :--- |
|  | 1 | Count on falling edges |
| Notes: | $\mathrm{X} \quad$ Count on both rising and falling edges |  |
| 2. $\mathrm{X}: 0$ or 1 , donternal care. clock edge selection is effective when the input clock is $\phi / 4$ or slower. When $\varnothing / 1$ |  |  |
|  | or the overflow/underflow of another channel is selected for the input clock, although <br> values can be written, counter operation complies with the initial value (count on rising <br> edges). |  |

- Bits 2-0-Timer Prescaler 2-0 (TPSC2-TPSC0): TPSC2-TPSC0 select the counter clock source for the TCNT. An independent clock source can be selected for each channel. Table 12.4 shows the possible settings for each channel.

Table 12.4 MTU Clock Sources

| Channel | Internal Clock |  |  |  |  |  | Other Channel Overflow/ Underflow | External Clock |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ¢/1 | $\phi / 4$ | $\begin{aligned} & \phi / \\ & 16 \end{aligned}$ | $\begin{aligned} & \phi / \\ & 64 \end{aligned}$ | $\begin{aligned} & \phi / \\ & 256 \end{aligned}$ | $\begin{aligned} & \hline \phi / \\ & 1024 \end{aligned}$ |  | $\begin{aligned} & \hline \text { TCL } \\ & \text { KA } \end{aligned}$ | $\begin{aligned} & \hline \text { TCL } \\ & \text { KB } \end{aligned}$ | $\begin{aligned} & \mathrm{TCL} \\ & \mathrm{KC} \end{aligned}$ | $\begin{aligned} & \text { TCL } \\ & \text { KD } \end{aligned}$ |
| 0 | O | O | O | O | X | X | X | O | O | O | O |
| 1 | 0 | 0 | O | O | 0 | X | 0 | 0 | 0 | X | X |
| 2 | 0 | 0 | O | O | X | 0 | X | 0 | 0 | 0 | X |
| 3 | 0 | 0 | O | O | 0 | 0 | X | 0 | 0 | X | X |
| 4 | 0 | 0 | O | O | 0 | 0 | X | 0 | 0 | X | X |

Channel 0:

| Bit 2: <br> TPSC2 | Bit 1: <br> TPSC1 | Bit 0: <br> TPSC0 | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Internal clock: count with $\phi / 1$ (initial value) |
|  |  | 1 | Internal clock: count with $\phi / 4$ |
| 1 | 0 | 0 | Internal clock: count with $\phi / 16$ |
| 1 | 0 | Internal clock: count with $\phi / 64$ |  |
|  |  | 0 | External clock: count with the TCLKA pin input |

## Channel 1:

| Bit 2: <br> TPSC2 | Bit 1: <br> TPSC1 | Bit 0: <br> TPSC0 | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Internal clock: count with $\phi / 1$ (initial value) |
|  |  | 1 | Internal clock: count with $\phi / 4$ |
|  | 1 | 0 | Internal clock: count with $\phi / 16$ |
| 1 | 0 | 0 | Internal clock: count with $\phi / 64$ |
|  |  | 1 | External clock: count with the TCLKA pin input |
| 1 | 0 | Internal clock: count with $\phi / 256$ |  |

Note: These settings are ineffective when channel 1 is in phase counting mode.

## Channel 2:

| Bit 2: <br> TPSC2 | Bit 1: <br> TPSC1 | Bit 0: <br> TPSC0 | Description |
| :--- | :--- | :--- | :--- |

Note: These settings are ineffective when channel 2 is in phase counting mode.

Channel 3:

| Bit 2: <br> TPSC2 | Bit 1: <br> TPSC1 | Bit 0: <br> TPSCO | Description |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Internal clock: count with $\phi / 1$ (initial value) |
|  |  | 1 | Internal clock: count with $\phi / 4$ |
|  | 1 | 0 | Internal clock: count with $\phi / 16$ |
| 1 | 0 | 0 | Internal clock: count with $\phi / 64$ |
|  |  | 1 | Internal clock: count with $\phi / 256$ |
|  | 1 | 0 | Internal clock: count with $\phi / 1024$ |

Channel 4:
Bit 2: Bit 1: Bit 0:
TPSC2 TPSC1 TPSC0 Description

| 0 | 0 | 0 | Internal clock: count with $\phi / 1$ (initial value) |
| :---: | :---: | :---: | :---: |
|  |  | 1 | Internal clock: count with $\phi / 4$ |
|  | 1 | 0 | Internal clock: count with $\phi / 16$ |
|  |  | 1 | Internal clock: count with $\phi / 64$ |
| 1 | 0 | 0 | Internal clock: count with $\phi / 256$ |
|  |  | 1 | Internal clock: count with $\phi / 1024$ |
|  | 1 | 0 | External clock: count with the TCLKA pin input |
|  |  | 1 | External clock: count with the TCLKB pin input |

### 12.2.2 Timer Mode Register (TMDR)

The TMDR is an 8 -bit read/write register that sets the operating mode for each channel. The MTU has five TMDR registers, one for each channel. TMDR is initialized to H'C0 by a power-on reset or the standby mode. Manual reset does not initialize TMDR.

## Channels 0, 3, 4: TMDR0, TMDR3, TMDR4:



## Channels 1, 2: TMDR1, TMDR2:



- Bits 7, 6-Reserved: These bits are reserved. They always read as 1, and cannot be modified.
- Bit 5—Buffer Operation B (BFB): Designates whether to use the TGRB register for normal operation, or buffer operation in combination with the TGRD register. When using TGRD as a buffer register, no TGRD register input capture/output compares are generated.
This bit is reserved in channels 1 and 2 , which have no TGRD registers. It is always read as 0 , and cannot be modified.

Bit 5: BFB
Description

| 0 | TGRB operates normally (initial value) |
| :--- | :--- |
| 1 | TGRB and TGRD buffer operation |

- Bit 4—Buffer Operation A (BFA): Designates whether to use the TGRA register for normal operation, or buffer operation in combination with the TGRC register. When using TGRC as a buffer register, no TGRC register input capture/output compares are generated.
This bit is reserved in channels 1 and 2 , which have no TGRC registers. It is always read as 0 , and cannot be modified.

Bit 4: BFA Description

| 0 | TGRA operates normally (initial value) |
| :--- | :--- |
| 1 | TGRA and TGRC buffer operation |

- Bits 3-0-Modes 3-0 (MD3-MD0): These bits set the timer operation mode.

| Bit 3: MD3 | Bit 2: MD2 | Bit 1: MD1 | $\begin{aligned} & \text { Bit 0: } \\ & \text { MDO } \end{aligned}$ | Description |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | Normal operation (initial value) |
|  |  |  | 1 | Reserved (do not set) |
|  |  | 1 | 0 | PWM mode 1 |
|  |  |  | 1 | PWM mode $2^{* 1}$ |
|  | 1 | 0 | 0 | Phase counting mode $1^{* 2}$ |
|  |  |  | 1 | Phase counting mode 2*2 |
|  |  | 1 | 0 | Phase counting mode 3*2 |
|  |  |  | 1 | Phase counting mode 4*2 |
| 1 | 0 | 0 | 0 | Reset synchronous PWM mode*3 |
|  |  |  | 1 | Reserved (do not set) |
|  |  | 1 | 0 | Reserved (do not set) |
|  |  |  | 1 | Reserved (do not set) |
|  | 1 | 0 | 0 | Reserved (do not set) |
|  |  |  | 1 | Complementary PWM mode 1 (transmit at peak)*3 |
|  |  | 1 | 0 | Complementary PWM mode 2 (transmit at valley)*3 |
|  |  |  | 1 | Complementary PWM mode 3 (transmit at peak and valley)*3 |

Notes: *1 PWM mode 2 can not be set for channels 3, 4.
*2 Phase measurement mode can not be set for channels 0, 3, 4 .
*3 Reset synchronous PWM mode, complementary PWM mode can only be set for channel 3 . When channel 3 is set to reset synchronous PWM mode or complementary PWM mode, the channel 4 settings become ineffective and automatically conform to the channel 3 settings. However, do not set channel 4 to reset synchronous PWM mode or complementary PWM mode. Reset synchronous PWM mode and complementary PWM mode can not be set for channels $0,1,2$.

## Renesns

### 12.2.3 Timer I/O Control Register (TIOR)

The TIOR is a register that controls the TGR. The MTU has eight TIOR registers, two each for channels 0,3 , and 4 , and one each for channels 1 and 2 . TIOR is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset or the standby mode. Manual reset does not initialize TIOR.

## Channels 0, 3, 4: TIOR0H, TIOR3H, TIOR4H

## Channels 1, 2: TIOR1, TIOR2:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | IOB3 | IOB2 | IOB1 | IOB0 | IOA3 | IOA2 | IOA1 | IOAO |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bits 7-4-I/O Control B3-B0 (IOB3-IOB0): These bits set the TGRB register function.
- Bits 3-0-I/O Control A3-B0 (IOA3-IOA0): These bits set the TGRA register function.


## Channels $0,3,4$ : TIOR0L, TIOR3L, TIOR4L:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | IOD3 | IOD2 | IOD1 | IOD0 | IOC3 | IOC2 | IOC1 | IOC0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: When the TGRC or TGRD registers are set for buffer operation, these settings become ineffective and the operation is as a buffer register.

- Bits 7-4-I/O Control D3-D0 (IOD3-IOD0): These bits set the TGRD register function.
- Bits 3-0-I/O Control C3-C0 (IOC3-IOC0): These bits set the TGRC register function.


## Channel 0 (TIOR0H Register):

- Bits 7-4—I/O Control B3-B0 (IOB3-IOB0): These bits set the TGR0B register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOB3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOB2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 5: } \\ & \text { IOB1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 4: } \\ & \text { IOBO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGROB is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output <br> is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGROB is an input capture register | Capture input source is the TIOCOB pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  | Capture input source is channel $1 /$ count clock | Input capture |
|  |  |  | 1 |  |  | on TCNT1 |
|  |  | 1 | 0 |  |  | count up/count down |
|  |  |  | 1 |  |  |  |

- Bits 3-0-I/O Control A3-A0 (IOA3-IOA0): These bits set the TGR0A register function.
$\left.\begin{array}{lllllll}\begin{array}{llll}\text { Bit 3: } \\ \text { IOA3 }\end{array} & \begin{array}{l}\text { Bit 2: } \\ \text { IOA2 }\end{array} & \begin{array}{l}\text { Bit 1: } \\ \text { IOA1 }\end{array} & \begin{array}{l}\text { Bit 0: } \\ \text { IOAO }\end{array} & \text { Description }\end{array}\right)$


## Channel 0 (TIOR0L Register):

- Bits 7-4—I/O Control D3-D0 (IOD3-IOD0): These bits set the TGR0D register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOD3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOD2 } \end{aligned}$ | Bit 5 : IOD1 | $\begin{aligned} & \text { Bit 4: } \\ & \text { IODO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGROD is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGROD is an input capture register | Capture input source is the TIOCOD pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  | Capture input source is channel $1 /$ count clock | Input capture |
|  |  |  | 1 |  |  | on TCNT1 |
|  |  | 1 | 0 |  |  | count up/count down |
|  |  |  | 1 |  |  |  |

Note: When the BFB bit of TMDRO is set to 1 and TGROD is being used as a buffer register, these settings become ineffective and input capture/output compares do not occur.

- Bits 3-0-I/O Control C3-C0 (IOC3-IOC0): These bits set the TGR0C register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOC3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \\ & \text { IOC2 } \end{aligned}$ | Bit 1: | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOCO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGROC is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGROC is an input capture register | Capture input source is the TIOCOC pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  | Capture input source is channel $1 /$ count clock | Input capture |
|  |  |  | 1 |  |  | on TCNT1 |
|  |  | 1 | 0 |  |  | count up/count down |
|  |  |  | 1 |  |  |  |

Note: When the BFA bit of TMDRO is set to 1 and TGROC is being used as a buffer register, these settings become ineffective and input capture/output compares do not occur.

## Channel 1 (TIOR1 Register):

- Bits 7-4—I/O Control B3-B0 (IOB3-IOB0): These bits set the TGR1B register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOB3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOB2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 5: } \\ & \text { IOB1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 4: } \\ & \text { IOBO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR1B is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR1B is an input capture register | Capture input source is the TIOC1B pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  | Capture input source TGROC compare/match input capture | Input capture |
|  |  |  | 1 |  |  | on channel TGR0C |
|  |  | 1 | 0 |  |  | compare-match/input |
|  |  |  | 1 |  |  | capture generation |

- Bits 3-0-I/O Control A3-A0 (IOA3-IOA0): These bits set the TGR1A register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOA3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \end{aligned}$ | Bit 1: <br> IOA1 | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOAO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR1A is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR1A is an input capture register | Capture input source is the TIOC1A pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  | Capture input | Input capture |
|  |  |  | 1 |  | source is TGR0A | on channel 0/TGROA |
|  |  | 1 | 0 |  | comparematch/input | compare-match/input capture generation |
|  |  |  | 1 |  | capture |  |

## Channel 2 (TIOR2 Register):

- Bits 7-4—I/O Control B3-B0 (IOB3-IOB0): These bits set the TGR2B register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IRR } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOB2 } \end{aligned}$ | Bit 5: IOB1 | Bit 4: <br> IOBO | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR2B is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR2B <br> is an input capture register | Capture input source is the TIOC2B pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

- Bits 3-0-I/O Control A3-A0 (IOA3-IOA0): These bits set the TGR2A register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOA3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \\ & \text { IOA2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 1: } \\ & \text { IOA1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOAO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR2A is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR2A is an input capture register | Capture input source is the TIOC2A pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

## Channel 3 (TIOR3H Register):

- Bits 7-4—I/O Control B3-B0 (IOB3-IOB0): These bits set the TGR3B register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOB3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOB2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 5: } \\ & \text { IOB1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 4: } \\ & \text { IOBO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR3B is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR3B <br> is an input capture register | Capture input source is the TIOC3B pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

- Bits 3-0-I/O Control A3-A0 (IOA3-IOA0): These bits set the TGR3A register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOA3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \\ & \text { IOA2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 1: } \\ & \text { IOA1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOAO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR3A is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR3A is an input capture register | Capture input source is the TIOC3A pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

## Channel 3 (TIOR3L Register):

- Bits 7-4—I/O Control D3-D0 (IOD3-IOD0): These bits set the TGR4D register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOD3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOD2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 5: } \\ & \text { IOD1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 4: } \\ & \text { IODO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR3D is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR3D <br> is an input capture register | Capture input source is the TIOC3D pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

Note: When the BFB bit of TMDR3 is set to 1 and TGR3D is being used as a buffer register, these settings become ineffective and input capture/output compares do not occur.

- Bits 3-0-I/O Control C3-C0 (IOC3-IOC0): These bits set the TGR4C register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOC3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \\ & \text { IOC2 } \end{aligned}$ | Bit 1: <br> IOC1 | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOCO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR3C is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output <br> is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR3C is an input capture register | Capture input source is the TIOC3C pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

Note: When the BFA bit of TMDR3 is set to 1 and TGR3C is being used as a buffer register, these settings become ineffective and input capture/output compares do not occur.

## Channel 4 (TIOR4H Register):

- Bits 7-4—I/O Control B3-B0 (IOB3-IOB0): These bits set the TGR4B register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOB3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOB2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 5: } \\ & \text { IOB1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 4: } \\ & \text { IOBO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR4B is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output <br> is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR4B is an input capture register | Capture input source is the TIOC4B pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

- Bits 3-0-I/O Control A3-A0 (IOA3-IOA0): These bits set the TGR4A register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOA3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \\ & \text { IOA2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 1: } \\ & \text { IOA1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOAO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR4A is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR4A is an input capture register | Capture input source is the TIOC4A pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

## Channel 4 (TIOR4L Register):

- Bits 7-4—I/O Control D3-D0 (IOD3-IOD0): These bits set the TGR4D register function.

| $\begin{aligned} & \text { Bit 7: } \\ & \text { IOD3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 6: } \\ & \text { IOD2 } \end{aligned}$ | $\begin{aligned} & \text { Bit 5: } \\ & \text { IOD1 } \end{aligned}$ | $\begin{aligned} & \text { Bit 4: } \\ & \text { IODO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR4D <br> is an <br> output <br> compare <br> register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR4D <br> is an input capture register | Capture input source is the TIOC4D pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

Note: When the BFB bit of TMDR4 is set to 1 and TGR4D is being used as a buffer register, these settings become ineffective and input capture/output compares do not occur.

- Bits 3-0-I/O Control C3-C0 (IOC3-IOC0): These bits set the TGR4C register function.

| $\begin{aligned} & \text { Bit 3: } \\ & \text { IOC3 } \end{aligned}$ | $\begin{aligned} & \text { Bit 2: } \\ & \text { IOC2 } \end{aligned}$ | Bit 1: IOC1 | $\begin{aligned} & \text { Bit 0: } \\ & \text { IOCO } \end{aligned}$ | Description |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | TGR4C is an output compare register | Output disabled (initial value) |  |
|  |  |  | 1 |  | Initial <br> output is 0 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
|  | 1 | 0 | 0 |  | Output disabled |  |
|  |  |  | 1 |  | Initial output is 1 | Output 0 on compare-match |
|  |  | 1 | 0 |  |  | Output 1 on compare-match |
|  |  |  | 1 |  |  | Toggle output on compare-match |
| 1 | 0 | 0 | 0 | TGR4C is an input capture register | Capture <br> input source <br> is the <br> TIOC4C pin | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |
|  | 1 | 0 | 0 |  |  | Input capture on rising edge |
|  |  |  | 1 |  |  | Input capture on falling edge |
|  |  | 1 | 0 |  |  | Input capture on both edges |
|  |  |  | 1 |  |  |  |

Note: When the BFA bit of TMDR4 is set to 1 and TGR4C is being used as a buffer register, these settings become ineffective and input capture/output compares do not occur.

### 12.2.4 Timer Interrupt Enable Register (TIER)

The TIER is an 8-bit register that controls the enable/disable of interrupt requests for each channel. The MTU has five TIER registers, one each for channel. TIER is initialized to H'40 by a reset or by standby mode.

## Channel 0: TIER0:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TTGE | - | - | TCIEV | TGIED | TGIEC | TGIEB | TGIEA |  |
|  | Initial value: | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | $R$ | $R$ | R/W | R/W | R/W | R/W | R/W |  |

## Channels 1, 2: TIER1, TIER2:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TTGE | - | TCIEU | TCIEV | - | - | TGIEB | TGIEA |
|  | Thitial value: | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| 0 |  |  |  |  |  |  |  |  |
| R/W: | R/W | $R$ | R/W | R/W | $R$ | $R$ | R/W | R/W |

Channels 3, 4: TIER3, TIER4:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TTGE | - | - | TCIEV | TGIED | TGIEC | TGIEB | TGIEA |
| Initial value: | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | $R$ | $R$ | R/W | R/W | R/W | R/W | R/W |

- Bit 7—A/D Conversion Start Request Enable (TTGE): Enables or disables generation of an A/D conversion start request by a TGRA register input capture/compare-match.


## Bit 7: TTGE Description

| 0 | Disable A/D conversion start requests (initial value) |
| :--- | :--- |
| 1 | Enable A/D conversion start request generation |

- Bit 6—Reserved: This bit is reserved. It always reads as 0 , and cannot be modified.
- Bit 5—Underflow Interrupt Enable (TCIEU): Enables or disables interrupt requests when the underflow flag (TCFU) of the channel 1, 2 timer status register (TSR) is set to 1 .
This bit is reserved for channels 0,3 , and 4 . It always reads as 0 . The write value should always be 1 .

Bit 5: TCIEU Description

| 0 | Disable UDF interrupt requests (TCIU) (initial value) |
| :--- | :--- |
| 1 | Enable UDF interrupt requests (TCIU) |

- Bit 4—Overflow Interrupt Enable (TCIEV): Enables or disables interrupt requests when the overflow flag TCFV of the timer status register (TSR) is set to 1 .

| Bit 4: TCIEV | Description |
| :--- | :--- |
| 0 | Disable TCFV interrupt requests (TCIV) (initial value) |
| 1 | Enable TCFV interrupt requests (TCIV) |

- Bit 3-TGR Interrupt Enable D (TGIED): Enables or disables interrupt TGFD requests when the TGFD bit of the channel $0,3,4$ TSR register is set to 1 .
This bit is reserved for channels 1 and 2. It always reads as 1 . The write value should always be 1 .

| Bit 3: TGIED | Description |
| :--- | :--- |
| 0 | Disable interrupt requests (TGID) due to the TGFD bit (initial value) |
| 1 | Enable interrupt requests (TGID) due to the TGFD bit |

- Bit 2—TGR Interrupt Enable C (TGIEC): Enables or disables TGFC interrupt requests when the TGFC bit of the channel $0,3,4$ TSR register is set to 1 .
This bit is reserved for channels 1 and 2. It always reads as 1 . The write value should always be 1 .

| Bit 2: TGIEC | Description |
| :--- | :--- |
| 0 | Disable interrupt requests (TGIC) due to the TGFC bit (initial value) |
| 1 | Enable interrupt requests (TGIC) due to the TGFC bit |

- Bit 1—TGR Interrupt Enable B (TGIEB): Enables or disables TGFB interrupt requests when the TGFB bit of the TSR register is set to 1 .

| Bit 1: TGIEB | Description |
| :--- | :--- |
| 0 | Disable interrupt requests (TGIB) due to the TGFB bit (initial value) |
| 1 | Enable interrupt requests (TGIB) due to the TGFB bit |

- Bit 0—TGR Interrupt Enable A (TGIEA): Enables or disables TGFA interrupt requests when the TGFA bit of the TSR register is set to 1 .

| Bit 0: TGIEA | Description |
| :--- | :--- |
| 0 | Disable interrupt requests (TGIA) due to the TGFA bit (initial value) |
| 1 | Enable interrupt requests (TGIA) due to the TGFA bit |

### 12.2.5 Timer Status Register (TSR)

The timer status register (TSR) is an 8-bit register that indicates the status of each channel. The MTU has five TSR registers, one each for channel. TSR is initialized to H'C0 by a power-on reset or by standby mode. This register is not initialized by a manual reset.

## Channel 0: TSR0:

| Bit: | 7 | 6 | 5 | 4 | 0 | 2 | 1 |  |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 |  |  |  |  |  |  |  |  |
|  | - | - | - | TCFV | TGFD | TGFC | TGFB | TGFA |
| Initial value: | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | R | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ |

Note: * Only 0 writes to clear the flags are possible.

Channels 1, 2: TSR1, TSR2:

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TCFD | - | TCFU | TCFV | - | - | TGFB | TGFA |
| Initial value: | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | R | R | $\mathrm{R} /(\mathrm{W})^{*}$ | R/(W)* |

Note: * Only 0 writes to clear the flags are possible.

## Channels 3, 4: TSR3, TSR4:

| Bit: | 7 | 6 | 5 | 4 |  |  |  |  |  |  |  | 3 |  | 2 |  | 1 |  | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TCFD | - | - | TCFV | TGFD | TGFC | TGFB | TGFA |  |  |  |  |  |  |  |  |  |  |
| Initial value: | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |  |  |  |  |  |  |  |
| $\mathrm{R} / \mathrm{W}:$ | R | R | R | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ |  |  |  |  |  |  |  |  |  |  |

Note: * Only 0 writes to clear the flags are possible.

- Bit 7-Count Direction Flag (TCFD): This status flag indicates the count direction of the channel 1, 2, 3, 4 TCNT counters.
This bit is reserved in channel 0 . This bit always reads as 1 . The write value should always be 1.

Bit 7: TCFD Description

| 0 | TCNT counts down |
| :--- | :--- |
| 1 | TCNT counts up (initial value) |

- Bit 6—Reserved: This bit always reads as 1 . The write value should always be 1 .
- Bit 5—Underflow Flag (TCFU): This status flag indicates the occurrence of a channel 1, 2 TCNT counter underflow.
This bit is reserved in channels 0,3 , and 4 . This bit always reads as 0 . The write value should always be 0 .

Bit 5: TCFU Description

| 0 | Clear condition: With TCFU=1, a 0 write to TCFU after reading it (initial <br> value) |
| :--- | :--- |
| 1 | Set condition: When the TCNT value underflows (H'0000 $\rightarrow$ H'FFFF) |

- Bit 4—Overflow Flag (TCFV): This status flag indicates the occurrence of a TCNT counter overflow.


## Bit 4: TCFV Description

| 0 | Clear condition: With TCFV $=1$, a 0 write to TCFV after reading it ${ }^{* 1}$ <br> (initial value) |
| :--- | :--- |
| 1 | Set condition: When the TCNT value overflows (H'FFFF $\rightarrow \mathrm{H}^{\prime} 0000$ )*2 |

Notes: *1 For channel 4, this flag is cleared by DTC transfer due to TCFV.
*2 For channel 4, this flag is also set when the TCNT value underflows ( $\mathrm{H}^{\prime} 0001 \rightarrow \mathrm{H}^{\prime} 0000$ ) in complementary PWM mode.

- Bit 3—Input Capture/Output Compare Flag D (TGFD): This status flag indicates the occurrence of a channel 0,3 , or 4 TGRD register input capture or compare-match. This bit is reserved in channels 1 and 2 . It always reads as 0 . The write value should always be 0.

| Bit 3: TGFD | Description |
| :---: | :---: |
| 0 | Clear condition: With TGFD $=1$, a 0 write to TGFD following a read (Cleared by DTC transfer due to TGFD) (initial value) |
| 1 | Set conditions: <br> - When TGRD is functioning as an output compare register (TCNT = TGRD) <br> - When TGRD is functioning as input capture (the TCNT value is sent to TGRD by the input capture signal) |

- Bit 2—Input Capture/Output Compare Flag C (TGFC): This status flag indicates the occurrence of a channel 0,3 , or 4 TGRC register input capture or compare-match.
This bit is reserved for channels 1 and 2. It always reads as 0 . The write value should always be 0 .
Bit 2: TGFC Description

| 0 | Clear condition: |
| :--- | :--- |
|  | With TGFC = 1, a 0 write to TGFC following a read (Cleared by DTC |
| transfer due to TGFC) (initial value) |  |

1 Set conditions:

- When TGRC is functioning as an output compare register (TCNT = TGRC)
- When TGRC is functioning as input capture (the TCNT value is sent to TGRC by the input capture signal)
- Bit 1—Input Capture/Output Compare Flag B (TGFB): This status flag indicates the occurrence of a TGRB register input capture or compare-match.
Bit 1: TGFB Description

| 0 | Clear condition: With TGFB $=1$, a 0 write to TGFB following a read <br> (Cleared by DTC transfer due to TGFB) (initial value) |
| :--- | :--- |
| 1 | Set conditions: |
| - When TGRB is functioning as an output compare register |  |
|  | $\quad$ (TCNT = TGRB) |
|  | - When TGRB is functioning as input capture (the TCNT value is sent |
|  | to TGRB by the input capture signal) |

- Bit 0—Input Capture/Output Compare Flag A (TGFA): This status flag indicates the occurrence of a TGRA register input capture or compare-match.

| Bit 0: TGFA | Description |
| :---: | :---: |
| 0 | Clear condition: With TGFA = 1, a 0 write to TGFA following a read (Cleared by DMAC transfer due to TGFA) (initial value) |
| 1 | Set conditions: <br> - When TGRA is functioning as an output compare register (TCNT = TGRA) <br> - When TGRA is functioning as input capture (the TCNT value is sent to TGRA by the input capture signal) |

## Renesns

### 12.2.6 Timer Counters (TCNT)

The timer counters (TCNT) are 16-bit counters, with one for each channel, for a total of five.
The TCNT are initialized to H'0000 by a power-on reset and when in standby mode. Manual reset does not initialize TCNT. Accessing the TCNT counters in 8-bit units is prohibited. Always access in 16-bit units.

Channel Abbreviation Function

| 0 | TCNT0 | Increment counter |
| :--- | :--- | :--- |
| 1 | TCNT1 | Increment/decrement counter*1 |
| 2 | TCNT2 | Increment/decrement counter*1 |
| 3 | TCNT3 | Increment/decrement counter*2 |
| 4 | TCNT4 | Increment/decrement counter*2 |

Notes: *1 Can only be used as an increment/decrement counter in phase counting mode, with other channel overflow/underflow counting. It becomes an increment counter in all other cases.
*2 Can only be used as an increment counter in complementary PWM mode. It becomes an increment counter in all other cases.


### 12.2.7 Timer General Register (TGR)

Each timer general register (TGR) is a 16-bit register that can function as either an output compare register or an input capture register. There are a total of sixteen TGR, four each for channels 0,3 , and 4 , and two each for channels 1 and 2. The TGRC and TGRD of channels 0,3 , and 4 can be set to operate as buffer registers. The TGR register and buffer register combinations are TGRA with TGRC, and TGRB with TGRD.

The TGRs are initialized to H'FFFF by a power-on reset or in standby mode. Manual reset does not initialize TGR. Accessing of the TGRs in 8-bit units is disabled; they may only be accessed in 16-bit units.


### 12.2.8 Timer Start Register (TSTR)

The timer start register (TSTR) is an 8-bit read/write register that starts and stops the timer counters (TCNT) of channels $0-4$. TSTR is initialized to $\mathrm{H}^{\prime} 00$ upon power-on reset or standby mode. Manual reset does not initialize TSTR.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | CST4 | CST3 | - | - | - | CST2 | CST1 | CST0 |

- Bits 7, 6, 2-0—Counter Start 4-0 (CST4-CST0): Select the start and stop of the timer counters (TCNT). The counter start to channel and bit to channel correspondence are indicated in the tables below.

- Bits 5-3-Reserved: These bits always read as 0 . The write value should always be 0 .


### 12.2.9 Timer Synchro Register (TSYR)

The timer synchro register (TSYR) is an 8-bit read/write register that selects independent or synchronous TCNT counter operation for channels $0-4$. Channels for which 1 is set in the corresponding bit will be synchronized. TSYR is initialized to $\mathrm{H}^{\prime} 00$ upon power-on reset or standby mode. Manual reset does not initialize TSYR.

| Bit: | 7 | 6 | 5 | 4 | 3 |  | 2 |  |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 0 |  |  |  |  |  |  |  |
|  | SYNC4 | SYNC3 | - | - | - | SYNC2 | SYNC1 | SYNC0 |
|  | Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bits 7, 6, 2-0—Timer Synchronization 4-0 (SYNC4-SYNC0): Selects operation independent of, or synchronized to, other channels. Synchronous operation allows synchronous clears due to multiple TCNT synchronous presets and other channel counter clears. A minimum of two channels must have SYNC bits set to 1 for synchronous operation. For synchronization clearing, it is necessary to set the TCNT counter clear sources (the CCLR2-CCLR0 bits of the TCR register), in addition to the SYNC bit. The counter start to channel and bit-to-channel correspondence are indicated in the tables below.

| Counter Start | Channel |
| :--- | :--- |
| SYNC4 | Channel 4 (TCNT4) |
| SYNC3 | Channel 3 (TCNT3) |
| SYNC2 | Channel 2 (TCNT2) |
| SYNC1 | Channel 1 (TCNT1) |
| SYNC0 | Channel 0 (TCNT0) |

## Bit n: SYNCn Description

| 0 | Timer counter (TCNTn) independent operation (initial value) <br> (TCNTn preset/clear unrelated to other channels) |
| :--- | :--- |
| 1 | Timer counter synchronous operation*1 |
|  | TCNTn synchronous preset/ synchronous clear*2 ${ }^{*}$ possible |

Notes: $\mathrm{n}=4$ to 0 . However, SYNC4 is bit 7 , SYNC3 is bit 6 .
*1 Minimum of two channel SYNC bits must be set to 1 for synchronous operation.
*2 TCNT counter clear sources (CCLR2-CCLR0 bits of the TCR register) must be set in addition to the SYNC bit in order to have clear synchronization.

- Bits 5-3-Reserved: These bits always read as 0 . The write value should always be 0 .


### 12.2.10 Timer Output Master Enable Register (TOER)

The timer output master enable register (TOER) enables/disables output settings for output pins TIOC4D, TIOC4C, TIOC3D, TIOC4B, TIOC4A, and TIOC3B. These pins do not output correctly if the TOER bits have not been set. Set TOER of CH3 and CH2 prior to setting TIOR of CH3 and CH 4 . The TOER is an 8 -bit read/write register. The register is initialized to $\mathrm{H}^{\prime} \mathrm{C} 0$ by a power-on reset or in standby mode. Manual reset does not initialize TOER.

| Bit: | 7 | 6 | 5 | 4 | 3 |  |  |  |  |  |  |  | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | OE4D | OE4C | OE3D | OE4B | OE4A | OE3B |  |  |  |  |  |  |  |
| Initial value: | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |  |  |  |  |
| R/W: | $R$ | $R$ | $R / W$ | $R / W$ | R/W | R/W | R/W | R/W |  |  |  |  |  |  |  |

- Bits 7-6-Reserved: These bits always read as 1 . The write value should always be 1 .
- Bit 5—Master Enable TIOC4D (OE4D): Enables or disables the TIOC4D pin MTU output.

| Bit 5: OE4D | Description |
| :--- | :--- |
| 0 | Disable TIOC4D pin MTU output (initial value) |
| 1 | Enable TIOC4D pin MTU output |

- Bit 4—Master Enable TIOC4C (OE4C): Enables or disables the TIOC4C pin MTU output.

| Bit 4: OE4C | Description |
| :--- | :--- |
| 0 | Disable TIOC4C pin MTU output (initial value) |
| 1 | Enable TIOC4C pin MTU output |

- Bit 3—Master Enable TIOC3D (OE3D): Enables or disables the TIOC3D pin MTU output.

| Bit 3: OE3D | Description |
| :--- | :--- |
| 0 | Disable TIOC3D pin MTU output (initial value) |
| 1 | Enable TIOC3D pin MTU output |

- Bit 2—Master Enable TIOC4B (OE4B): Enables or disables the TIOC4B pin MTU output.

| Bit 2: OE4B | Description |
| :--- | :--- |
| 0 | Disable TIOC4B pin MTU output (initial value) |
| 1 | Enable TIOC4B pin MTU output |

- Bit 1—Master Enable TIOC4A (OE4A): Enables or disables the TIOC4A pin MTU output.

| Bit 1: OE4A | Description |
| :--- | :--- |
| 0 | Disable TIOC4A pin MTU output (initial value) |
| 1 | Enable TIOC4A pin MTU output |

- Bit 0—Master Enable TIOC3B (OE3B): Enables or disables the TIOC3B pin MTU output.
Bit 0: OE3B Description

| 0 | Disable TIOC3B pin MTU output (initial value) |
| :--- | :--- |
| 1 | Enable TIOC3B pin MTU output |

### 12.2.11 Timer Output Control Register (TOCR)

The timer output control register (TOCR) enables/disables PWM synchronized toggle output in complementary PWM mode and reset sync PWM mode, and controls output level inversion of PWM output. The TOCR is initialized to H'00 by a power-on reset or in the standby mode. Manual reset does not initialize TOCR. These register settings are ineffective for anything other than complementary PWM mode/reset-synchronized PWM mode.

| Bit: | 7 | 6 |  | 5 | 4 | 3 |  | 2 | 1 |  | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | PSYE | - | - | - | - | OLSN | OLSP |  |  |  |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |
| R/W: | $R$ | $R / W$ | $R$ | $R$ | $R$ | $R$ | $R / W$ | $R / W$ |  |  |  |

- Bits 7, 5-2—Reserved: These bits always read as 1 . The write value should always be 1 .
- Bit 6—PWM Synchronous Output Enable (PSYE): Selects the enable/disable of toggle output synchronized with the PWM period.

| Bit 6: PSYE | Description |
| :--- | :--- |
| 0 | Toggle output synchronous with PWM period disabled (initial value) |
| 1 | Toggle output synchronous with PWM period enabled |

- Bit 1—Output Level Select N (OLSN): Selects the reverse phase output level of the complementary PWM mode or reset-synchronized PWM mode.

|  |  | Compare Match Output |  |  |
| :--- | :--- | :--- | :--- | :--- |
| OLSN | Initial Output | Active Level | Increment Count | Decrement Count |
| 0 | High level* | Low level | High level | Low level (initial <br> value) |
| 1 | Low level* $^{*}$ | High level | Low level | High level |

Note: * The reverse phase waveform initial output value changes to active level after elapse of the dead time after count start.

- Bit 0—Output Level Select P (OLSP): Selects the positive phase output level of the complementary PWM mode or reset-synchronized PWM mode.

|  |  |  | Compare Match Output |  |
| :--- | :--- | :--- | :--- | :--- |
| OLSP | Initial Output | Active Level | Increment Count | Decrement Count |
| 0 | High level | Low level | Low level | High level (initial value) |
| 1 | Low level | High level | High level | Low level |

Figure 12.2 shows an example of complementary PWM mode output ( 1 phase) when $\operatorname{OLSN}=1$, $\mathrm{OLSP}=1$.


Figure 12.2 Complementary PWM Mode Output Level Example

### 12.2.12 Timer Gate Control Register (TGCR)

The timer gate control register (TGCR) is an 8-bit read/write register that controls the waveform output necessary for brushless DC motor control in complementary PWM mode/resetsynchronized PWM mode. The TGCR is initialized to $\mathrm{H}^{\prime} 80$ by a power-on reset or in the standby mode. Manual reset does not initialize TGCR. These register settings are ineffective for anything other than complementary PWM mode/reset-synchronized PWM mode.

| Bit: | 7 | 6 | 5 | 4 | 3 |  | 2 |  |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 0 |  |  |  |  |  |  |  |
|  | - | BDC | N | P | FB | WF | VF | UF |
| Initial value: | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 7—Reserved: This bit always reads as 1 . The write value should always be 1 .
- Bit 6—Brushless DC Motor (BDC): Selects gate signal output/chopping output function for brushless DC motor control.

| Bit 6: BDC | Description |
| :--- | :--- |
| 0 | Ordinary output (initial value) |
| 1 | Gate signal/chopping output for brushless DC motor |

- Bit 5—Reverse Phase Output (N): Selects whether to output gate signals directly to the reverse phase pin (TIOC3D, TIOC4C, and TIOC4D) output, or to output by chopping the gate signal and the complementary PWM/reset-synchronized PWM output.

| Bit 5: $\mathbf{N}$ | Description |
| :--- | :--- |
| 0 | Output gate signals directly to reverse phase pin output (initial value) |
| 1 | Output chopped gate signal and complementary PWM /reset- <br> synchronized PWM output to reverse phase pin output |

- Bit 4—Positive Phase Output (P): Selects whether to output gate signals directly to the positive phase pin (TIOC3B, TIOC4A, and TIOC4B) output, or to output by chopping the gate signal and the complementary PWM/reset-synchronized PWM output.

Bit 4: $\mathbf{P}$

## Description

| 0 | Output gate signals directly to positive phase pin output (initial value) |
| :--- | :--- |
| 1 | Output chopped gate signal and complementary PWM /reset- <br> synchronized PWM output to positive phase pin output |

- Bit 3—Feedback Input (FB): Selects whether to use external input or register input for the feedback input to generate gate signals.


## Bit 3: FB

## Description

0
Feedback input is external input (initial value) (Input sources are channel 0 TGRA, TGRB, TGRC input capture signals)

1 Feedback input is register input (TGCR's UF, VF, WF settings)

- Bits 2-0-Output Phase Switch 2-0 (WF, VF, UF): These bits set the positive phase/negative phase output phase on or off state. The setting of these bits is valid only when the FB bit in this register is set to 1 . In this case, the setting of bits $2-0$ is a substitute for external input.

Function

| Bit 2: WF | Bit 1: VF | Bit 0: UF | TIOC3B | TIOC4A | TIOC4B | TIOC3D | TIOC4C | TIOC4D |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | U Phase | V <br> Phase | W Phase | U <br> Phase | V Phase | W Phase |  |
| 0 | 0 | 0 | Off | Off | Off | Off | Off | Off | Initial value |
|  |  | 1 | On | Off | Off | Off | Off | On | - |
|  | 1 | 0 | Off | On | Off | On | Off | Off | - |
|  |  | 1 | Off | On | Off | Off | Off | On | - |
| 1 | 0 | 0 | Off | Off | On | Off | On | Off | - |
|  |  | 1 | On | Off | Off | Off | On | Off | - |
|  | 1 | 0 | Off | Off | On | On | Off | Off | - |
|  |  | 1 | Off | Off | Off | Off | Off | Off | - |

### 12.2.13 Timer Subcounter (TCNTS)

The timer subcounter (TCNTS) is a 16-bit read-only counter that is used only in complementary PWM mode. The TCNTS counter is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset or in standby mode. Manual reset does not initialize TCNTS. Accessing the TCNTS counter in 8 -bit units is prohibited. Always access in 16-bit units.


### 12.2.14 Timer Dead Time Data Register (TDDR)

The timer dead time data register (TDDR) is a 16-bit register, used only in complementary PWM mode, that specifies the TCNT3 and TCNT4 counter offset values. In complementary PWM mode, when the TCNT3 and TCNT4 counters are cleared and then restarted, the TDDR register value is loaded into the TCNT3 counter and the count operation starts. The TDDR register is initialized to H'FFFF by a power-on reset or in standby mode. Manual reset does not initialize TDDR. Accessing the TDDR in 8-bit units is prohibited. Always access in 16-bit units.


### 12.2.15 Timer Period Data Register (TCDR)

The timer period data register (TCDR) is a 16-bit register used only in complementary PWM mode. Set the PWM carrier sync value as the TCDR register value. This register is constantly compared with the TCNTS counter in complementary PWM mode, and when a match occurs the TCNTS counter switches direction (decrement to increment).

The TCDR register is initialized to H'FFFF by a reset or in standby mode. Manual reset does not initialize TCDR. Accessing the TCDR in 8-bit units is prohibited. Always access in 16-bit units.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value: | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| Initial value: | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

### 12.2.16 Timer Period Buffer Register (TCBR)

The timer period buffer register (TCBR) is a 16-bit register used only in complementary PWM mode. It functions as a buffer register for the TCDR register. The TCBR register values are transferred to the TCDR register with the transfer timing established in the TMDR register. The TCBR register is initialized to H'FFFF by a power-on reset or in standby mode. Manual reset does not initialize TCBR. Accessing the TCBR in 8-bit units is prohibited. Always access in 16-bit units.


### 12.3 Bus Master Interface

### 12.3.1 16-Bit Registers

The timer counters (TCNT) and general registers (TGR) are 16-bit registers. A 16-bit data bus to the bus master enables 16 -bit read/writes. 8-bit read/write is not possible. Always access in 16 -bit units. Figure 12.3 shows an example of 16 -bit register access operation.


Figure 12.3 16-Bit Register Access Operation (Bus Master $\leftrightarrow$ TCNT (16 Bit))

### 12.3.2 8-Bit Registers

All registers other than the TCNT and general registers (TGR) are 8-bit registers. These are connected to the CPU by a 16 -bit data bus, so 16 -bit read/writes and as 8 -bit read/writes are both possible (figures $12.4,12.5$, and 12.6).


Figure 12.4 8-Bit Register Access Operation (Bus Master $\leftrightarrow$ TCR (Upper 8 Bits))


Figure 12.5 8-Bit Register Access Operation (Bus Master $\leftrightarrow$ TMDR (Lower 8 Bits))


Figure 12.6 8-Bit Register Access Operation (Bus Master $\leftrightarrow$ TCR, TMDR (16 Bit))

### 12.4 Operation

### 12.4.1 Overview

The operation modes are described below.
Ordinary Operation: Each channel has a timer counter (TCNT) and general register (TGR). The TCNT is an upcounter and can also operate as a free-running counter, periodic counter or external event counter. General registers (TGR) can be used as output compare registers or input capture registers.

Synchronized Operation: The TCNT of a channel set for synchronized operation does a synchronized preset. When any TCNT of a channel operating in the synchronized mode is rewritten, the TCNTs of other channels are simultaneously rewritten as well. The timer synchronization bits of the TSYR registers of multiple channels set for synchronous operation can be set to clear the TCNTs simultaneously.

Buffer Operation: When TGR is an output compare register, the buffer register value of the corresponding channel is transferred to the TGR when a compare-match occurs. When TGR is an input capture register, the TCNT counter value is transferred to the TGR when an input capture occur simultaneously the value previously stored in the TGR is transferred to the buffer register.

Cascade Connection Operation: The channel 1 and channel 2 counters (TCNT1 and TCNT2) can be connected together to operate as a 32-bit counter.

PWM Mode: In PWM mode, a PWM waveform is output. The output level can be set by the TIOR register. Each TGR can be set for PWM waveform output with a duty cycle between $0 \%$ and $100 \%$.

Phase Counting Mode: In phase counting mode, the phase differential between two clocks input from the channel 1 and channel 2 external clock input pins is detected and the TCNT counter operates as an up/down counter. In phase counting mode, the corresponding TCLK pins become clock inputs and TCNT functions as an up/down counter. It can be used as a two-phase encoder pulse input.

Reset-Synchronized PWM Mode: Three-phase positive and negative PWM waveforms can be obtained using channels 3 and 4 (the three phases of the PWM waveform share a transition point on one side). When set for reset-synchronized PWM mode, TGR3A, TGR3B, TGR4A, and TGR4B automatically become output compare registers. The TIOC3A, TIOC3B, TIOC4A, TIOC4B, TIOC4C, and TIOC4D pins also become PWM output pins, and TCNT3 and TCNT4 become upcounters. TCNT4, TGR4A, and TGR4B are isolated from TCNT4.

Complementary PWM Mode: Three-phase complementary positive and negative PWM waveforms whose positive and negative phases do not overlap can be obtained using channels 3 and 4. When set for complementary PWM mode, TGR3A, TGR3B, TGR4A, and TGR4B become output compare registers. The TIOC3B, TIOC3D, TIOC4A, TIOC4B, TIOC4C, and TIOC4D pins also automatically become PWM output pins while TCNT3 and TCNT4 become up/down counters.

### 12.4.2 Basic Functions

Always select MTU external pin set function using the pin function controller (PFC).
Counter Operation: When a start bit (CST0-CST4) in the timer start register (TSTR) is set to 1 , the corresponding timer counter (TCNT) starts counting. There are two counting modes: a freerunning mode and a periodic mode.

To select the counting operation (figure 12.7):

1. Set bits TPSC2-TPSC0 in the TCR to select the counter clock. At the same time, set bits CKEG1 and CKEG0 in the TCR to select the desired edge of the input clock.
2. To operate as a periodic counter, set the CCLR2-CCLR0 bits in the TCR to select TGR as a clearing source for the TCNT.
3. Set the TGR selected in step 2 as an output compare register using the timer I/O control register (TIOR).
4. Write the desired cycle value in the TGR selected in step 2.
5. Set the CST bit in the TSTR to 1 to start counting.


Figure 12.7 Procedure for Selecting the Counting Operation
Free-Running Counter Operation Example: A reset of the MTU timer counters (TCNT) leaves them all in the free-running mode. When a bit in the TSTR is set to 1 , the corresponding timer counter operates as a free-running counter and begins to increment. When the count overflows from H'FFFF-H'0000, the TCFV bit in the timer status register (TSR) is set to 1 . If the TCIEV bit in the timer's corresponding timer interrupt enable register (TIER) is set to 1 , the MTU will make an interrupt request to the interrupt controller. After the TCNT overflows, counting continues from $\mathrm{H}^{\prime} 0000$. Figure 12.8 shows an example of free-running counter operation.


Figure 12.8 Free-Running Counter Operation

Periodic Counter Operation Example: Periodic counter operation is obtained for a given channel's TCNT by selecting compare-match as a TCNT clear source. Set the TGR register for period setting to output compare register and select counter clear upon compare-match using the CCLR2-CCLR0 bits of the timer control register (TCR). After these settings, the TCNT begins incrementing as a periodic counter when the corresponding bit of TSTR is set to 1 . When the count matches the TGR register value, the TGF bit in the TSR is set to 1 and the counter is cleared to $H^{\prime} 0000$. If the TGIE bit of the corresponding TIER is set to 1 at this point, the MTU will make an interrupt request to the interrupt controller. After the compare-match, TCNT continues counting from $\mathrm{H}^{\prime} 0000$. Figure 12.9 shows an example of periodic counting.


Figure 12.9 Periodic Counter Operation
Compare-Match Waveform Output Function: The MTU can output 0 level, 1 level, or toggle output from the corresponding output pins upon compare-matches.

Procedure for selecting the compare-match waveform output operation (figure 12.10):

1. Set the TIOR to select 0 output or 1 output for the initial value, and 0 output, 1 output, or toggle output for compare-match output. The TIOC pin will output the set initial value until the first compare-match occurs.
2. Set a value in the TGR to select the compare-match timing.
3. Set the CST bit in the TSTR to 1 to start counting.


Figure 12.10 Procedure for Selecting Compare Match Waveform Output Operation
Waveform Output Operation (0 Output/1 Output): Figure 12.11 shows 0 output/ 1 output. In the example, TCNT is a free-running counter, 1 is output upon compare-match A and 0 is output upon compare-match B. When the pin level matches the set level, the pin level does not change.


Figure 12.11 Example of 0 Output/1 Output
Waveform Output Operation (Toggle Output): Figure 12.12 shows the toggle output. In the example, the TCNT operates as a periodic counter cleared by compare-match B, with toggle output at both compare-match A and compare-match B.


Figure 12.12 Example of Toggle Output
Input Capture Function: In the input capture mode, the TCNT value is transferred into the TGR register when the input edge is detected at the input capture/output compare pin (TIOC).

Detection can take place on the rising edge, falling edge, or both edges. Channels 0 and 1 can use other channel counter input clocks or compare-match signals as input capture sources.

The procedure for selecting the input capture operation (figure 12.13) is:

1. Set the TIOR to select the input capture function of the TGR, then select the input capture source, and rising edge, falling edge, or both edges as the input edge.
2. Set the CST bit in the TSTR to 1 to start the TCNT counting.


Figure 12.13 Procedure for Selecting Input Capture Operation

Input Capture Operation: Figure 12.14 shows input capture. The falling edge of TIOCB and both edges of TIOCA are selected as input capture input edges. In the example, TCNT is set to clear at the input capture of the TGRB register.


Figure 12.14 Input Capture Operation

### 12.4.3 Synchronous Operation

In the synchronizing mode, two or more timer counters can be rewritten simultaneously (synchronized preset). Multiple timer counters can also be cleared simultaneously using TCR settings (synchronized clear).

The synchronizing mode can increase the number of TGR registers for a single time base. All five channels can be set for synchronous operation.

## Procedure for Selecting the Synchronizing Mode (Figure 12.15):

1. Set 1 in the SYNC bit of the timer synchro register (TSYR) to use the corresponding channel in the synchronizing mode.
2. When a value is written in the TCNT in any of the synchronized channels, the same value is simultaneously written in the TCNT in the other channels.
3. Set the counter to clear with output compare/input capture using bits CCLR2-CCLR0 in the TCR.
4. Set the counter clear source to synchronized clear using the CCLR2-CCLR0 bits of the TCR.
5. Set the CST bits for the corresponding channels in the TSTR to 1 to start counting in the TCNT.


Figure 12.15 Procedure for Selecting Synchronizing Operation

Synchronized Operation: Figure 12.16 shows an example of synchronized operation. Channels 0,1 , and 2 are set to synchronized operation and PWM mode 1 . Channel 0 is set for a counter clear upon compare-match with TGR0B. Channels 1 and 2 are set for synchronous counter clears by synchronous presets and TGR0B register compare-matches. Accordingly, a three-phase PWM waveform with the data set in the TGR0B register as its PWM period is output from the TIOC0A, TIOC1A, and TIOC2A pins.

See section 12.4.6, PWM Mode, for details on the PWM mode.


Figure 12.16 Synchronized Operation Example

### 12.4.4 Buffer Operation

Buffer operation is a function of channels 0,3 , and 4 . TGRC and TGRD can be used as buffer registers. Table 12.5 shows the register combinations for buffer operation.

Table 12.5 Register Combinations

| Channel | General Register | Buffer Register |
| :--- | :--- | :--- |
| 0 | TGR0A | TGR0C |
|  | TGR0B | TGR0D |
| 3 | TGR3A | TGR3C |
|  | TGR3B | TGR3D |
| 4 | TGR4A | TGR4C |
|  | TGR4B | TGR4D |

The buffer operation differs, depending on whether the TGR has been set as an input capture register or an output compare register.

When TGR Is an Output Compare Register: When a compare-match occurs, the corresponding channel buffer register value is transferred to the general register. Figure 12.17 shows an example.


Figure 12.17 Compare Match Buffer Operation
When TGR Is an Input Capture Register: When an input capture occurs, the timer counter (TCNT) value is transferred to the general register (TGR), and the value that had been held up to that time in the TGR is transferred to the buffer register (figure 12.18).


Figure 12.18 Input Capture Buffer Operation

## Procedure for Setting Buffer Mode (Figure 12.19):

1. Use the timer I/O control register (TIOR) to set the TGR as either an input capture or output compare register.
2. Use the timer mode register (TMDR) BFA, and BFB bits to set the TGR for buffer mode.
3. Set the CST bit in the TSTR to 1 to start the count operation.


Figure 12.19 Buffer Operation Setting Procedure
Buffer Operation Examples-when TGR Is an Output Compare Register: Figure 12.20 shows an example of channel 0 set to PWM mode 1, and the TGRA and TGRC registers set for buffer operation.

The TCNT counter is cleared by a compare-match B , and the output is a 1 upon compare-match A and 0 output upon compare-match B. Because buffer mode is selected, a compare-match A changes the output, and the buffer register TGRC value is simultaneously transferred to the general register TGRA. This operation is repeated with each occurrence of a compare-match A.

See section 12.4.6, PWM Mode, for details on the PWM mode.


Figure 12.20 Buffer Operation Example (Output Compare Register)
Buffer Operation Examples-when TGR Is an Input Capture Register: Figure 12.21 shows an example of TGRA set as an input capture register with the TGRA and TGRB registers set for buffer operation.

The TCNT counter is cleared by a TGRA register input capture, and the TIOCA pin input capture input edge is selected as both rising and falling edge. Because buffer mode is selected, an input capture A causes the TCNT counter value to be stored in the TGRA register, and the value that was stored in the TGRA up until that time is simultaneously transferred to the TGRC register.


Figure 12.21 Buffer Operation Example (Input Capture Register)

### 12.4.5 Cascade Connection Mode

Cascade connection mode is a function that connects the 16-bit counters of two channels together to act as a 32-bit counter.

This function operates by using the TPSC2-TPSC0 bits of the TCR register to set the channel 1 counter clock to count by TCNT2 counter overflow/underflow.

Note: When channel 1 is set to phase counting mode, the counter clock settings become ineffective.

Table 12.6 shows the cascade connection combinations.

Table 12.6 Cascade Connection Combinations

| Combination | Upper 16 Bits | Lower 16 Bits |
| :--- | :--- | :--- |
| Channel 1, channel 2 | TCNT1 | TCNT2 |

Procedure for Setting Cascade Connection Mode (Figure 12.22):

1. Set the TPSC2-TPSC 0 bits of the channel 1 timer control register (TCR) to B'111 to select "count by TCNT2 overflow/underflow."
2. Set the CST bits corresponding to the upper and lower 16 bits in the TSTR to 1 to start the count operation.


Figure 12.22 Procedure for Selecting Cascade Connection Mode

Cascade Connection Operation Examples—Phase Counting Mode: Figure 12.23 shows an example of operation when the TCNT1 counter is set to count on TCNT2 overflow/underflow and channel 2 is set to phase counting mode.

The TCNT1 counter increments with a TCNT2 counter overflow and decrements with a TCNT2 underflow.


Figure 12.23 Cascade Connection Operation Example (Phase Counting Mode)

### 12.4.6 PWM Mode

PWM mode outputs the various PWM waveforms from output pins. Output levels of 0 output, 1 output, or toggle output can be selected as the output level for the compare-match of each TGR.

A period can be set for a register by using the TGR compare-match as a counter clear source. All five channels can be independently set to PWM mode. Synchronous operation is also possible.

There are two PWM modes:

- PWM mode 1

Generates PWM output using the TGRA and TGRB registers, and TGRC and TGRD registers as pairs. The initial output values are those established in the TGRA and TGRC registers. When the values set in TGR registers being used as a pair are equal, output values will not change even if a compare-match occurs.
A maximum of 8-phase PWM output is possible for PWM mode 1.

- PWM mode 2

Generates PWM output using one TGR register as a period register and another as a duty cycle register. The output value of each pin upon a counter clear is the initial value established by the TIOR register. When the values set in the period register and duty register are equal, output values will not change even if a compare-match occurs. PWM mode 2 can be set only for channels 0,1 , and 2 .

Table 12.7 lists the combinations of PWM output pins and registers.
Table 12.7 Combinations of PWM Output Pins and Registers

|  |  |  | Output Pin |  |
| :--- | :--- | :--- | :--- | :---: |
| Channel | Register | PWM Mode 1 | PWM Mode 2 |  |
| 0 (AB pair) | TGR0A | TIOC0A | TIOC 0A |  |
|  | TGR0B |  | TIOC 0B |  |
| 0 (CD pair) | TGR0C | TIOC0C | TIOC 0C |  |
|  | TGR0D |  | TIOC 0D |  |
| 1 | TGR1A | TIOC1A | TIOC 1A |  |
|  | TGR1B |  | TIOC 1B |  |
| 2 | TGR2A | TIOC2A | TIOC 2A |  |
| 3 (AB pair) | TGR2B |  | TIOC 2B |  |
| 3 (CD pair) | TGR3A | TIOC3A | Setting not possible |  |
|  | TGR3B |  |  |  |
| 4 (AB pair) | TGR3D | TIOC3C |  |  |
| 4 (CD pair) | TGR4A |  |  |  |

Note: PWM output of the period setting TGR is not possible in PWM mode 2.

## Procedure for Selecting the PWM Mode (Figure 12.24):

1. Set bits TPSC2-TPSC0 in the TCR to select the counter clock source. At the same time, set bits CKEG1 and CKEG0 in the TCR to select the desired edge of the input clock.
2. Set bits CCLR2-CCLR0 in the TCR to select the TGR to be used as a counter clear source.
3. Set the period in the TGR selected in step 2, and the duty cycle in another TGR.
4. Using the timer I/O control register (TIOR), set the TGR selected in step 3 to act as an output compare register, and select the initial value and output value.
5. Set the MD3-MD 0 bits in TMDR to select the PWM mode.
6. Set the CST bit in the TSTR to 1 to let the TCNT start counting.


Figure 12.24 Procedure for Selecting the PWM Mode
PWM Mode Operation Examples—PWM Mode 1 (Figure 12.25): A TGRA register comparematch is used as a TCNT counter clear source, the TGRA register initial output value and output compare output value are both 0 , and the TGRB register output compare output value is a 1 . In this example, the value established in the TGRA register becomes the period and the value established in the TGRB register becomes the duty cycle.


Figure 12.25 PWM Mode Operation Example (Mode 1)

PWM Mode Operation Examples-PWM Mode 2 (Figure 12.26): Channels 0 and 1 are set for synchronous operation, TGR1B register compare-match is used as a TCNT counter clear source, the other TGR register initial output value is 0 and output compare output value is 1 , and a 5 -phase PWM waveform is output. In this example, the value established in the TGR1B register becomes the period and the value established in the other TGR register becomes the duty cycle.


Figure 12.26 PWM Mode Operation Example (Mode 2)
0\% Duty Cycle: Figure 12.27 shows an example of a $0 \%$ duty cycle PWM waveform output in PWM mode.


Figure 12.27 PWM Mode Operation Example (0\% Duty Cycle)
$\mathbf{1 0 0 \%}$ Duty Cycle: Figure 12.28 shows an example of a $100 \%$ duty cycle PWM waveform output in PWM mode.

In PWM mode, when setting cycle = duty cycle the output waveform does not change, nor is there a change of waveform for the first pulse immediately after clearing the counter.


Figure 12.28 PWM Mode Operation Example ( $\mathbf{1 0 0 \%}$ Duty Cycle)

### 12.4.7 Phase Counting Mode

The phase counting mode detects the phase differential of two external clock inputs and counts the TCNT counter up or down. This mode can be set for channels 1 and 2.

When set in the phase counting mode, an external clock is selected for the counter input clock, regardless of the settings of the TPSC2-TPSC0 bits of TCR or the CKEG1 and CKEG0 bits. TCNT also becomes an up/down counter. Since the TCR CCLR1/CCLR0 bits, TIOR, TIER, and TGR functions are all enabled, input capture and compare-match functions and interrupt sources can be used.

When the TCNT counter is incrementing, an overflow sets the TSR register TCFV (overflow flag). When it is decrementing, an underflow sets the TCFU (underflow flag).

The TSR register TCFD bit is a count direction flag. Read the TCFD flag to confirm whether the TCNT is incrementing or decrementing.

Table 12.8 shows the correspondence between channels and external clock pins.
Table 12.8 Phase Counting Mode Clock Input Pins

| Channel | A Phase Input Pin | B Phase Input Pin |
| :--- | :--- | :--- |
| 1 | TCLKA | TCLKB |
| 2 | TCLKC | TCLKD |

Procedure for Selecting the Phase Counting Mode (Figure 12.29):

1. Set the MD3-MD0 bits of the timer mode register (TMDR) to select the phase counting mode.
2. Set the CST bit of the timer start register (TSTR) to 1 to start the count.


Figure 12.29 Procedure for Selecting the Phase Counting Mode
Phase Counting Operation Examples: The phase counting mode uses the phase difference between two external clocks to increment/decrement the TCNT counter. There are 4 modes, depending on the count conditions.

Phase Counting Mode 1: Figure 12.30 shows an example of phase counting mode 1 operation. Table 12.9 lists the up counting and down counting conditions for the TCNT.


Figure 12.30 Phase Counting Mode 1 Operation
Table 12.9 Phase Count Mode 1 Up/Down Counting Conditions

| TCLKA (Channel 1) TCLKC (Channel 2) | TCLKB (Channel 1) TCLKD (Channel 2) | Operation |
| :---: | :---: | :---: |
| 1 (high level) | Rising edge | Increment |
| 0 (low level) | Falling edge |  |
| Rising edge | 0 (low level) |  |
| Falling edge | 1 (high level) |  |
| 1 (high level) | Falling edge | Decrement |
| 0 (low level) | Rising edge |  |
| Rising edge | 1 (high level) |  |
| Falling edge | 0 (low level) |  |

Phase Count Mode 2: Figure 12.31 shows an example of phase counting mode 2 operation. Table 12.10 lists the up counting and down counting conditions for the TCNT.


Figure 12.31 Phase Counting Mode 2 Operation
Table 12.10 Phase Count Mode 2 Up/Down Counting Conditions

| TCLKA (Channel 1) | TCLKB (Channel 1) |  |
| :--- | :--- | :--- |
| TCLKC (Channel 2) | TCLKD (Channel 2) | Operation |
| 1 (high level) | Rising edge | Does not count (don't care) |
| 0 (low level) | Falling edge | Does not count (don't care) |
| Rising edge | 0 (low level) | Does not count (don't care) |
| Falling edge | 1 (high level) | Increment |
| 1 (high level) | Falling edge | Does not count (don't care) |
| 0 (low level) | Rising edge | Does not count (don't care) |
| Rising edge | 1 (high level) | Does not count (don't care) |
| Falling edge | 0 (low level) | Decrement |

Phase Count Mode 3: Figure 12.32 shows an example of phase counting mode 3 operation. Table 12.11 lists the up counting and down counting conditions for the TCNT.


Figure 12.32 Phase Counting Mode 3 Operation
Table 12.11 Phase Count Mode 3 Up/Down Counting Conditions

| TCLKA (Channel 1) | TCLKB (Channel 1) |  |
| :--- | :--- | :--- |
| TCLKC (Channel 2) | TCLKD (Channel 2) | Operation |
| 1 (high level) | Rising edge | Does not count (don't care) |
| 0 (low level) | Falling edge | Does not count (don't care) |
| Rising edge | 0 (low level) | Does not count (don't care) |
| Falling edge | 1 (high level) | Increment |
| 1 (high level) | Falling edge | Decrement |
| 0 (low level) | Rising edge | Does not count (don't care) |
| Rising edge | 1 (high level) | Does not count (don't care) |
| Falling edge | 0 (low level) | Does not count (don't care) |

Phase Count Mode 4: Figure 12.33 shows an example of phase counting mode 4 operation. Table 12.12 lists the up counting and down counting conditions for the TCNT.


Figure 12.33 Phase Counting Mode 4 Operation
Table 12.12 Phase Count Mode 4 Up/Down Counting Conditions

| TCLKA (Channel 1) TCLKC (Channel 2) | TCLKB (Channel 1) <br> TCLKD (Channel 2) | Operation |
| :---: | :---: | :---: |
| 1 (high level) | Rising edge | Increment |
| 0 (low level) | Falling edge |  |
| Rising edge | 0 (low level) | Does not count (don't care) |
| Falling edge | 1 (high level) |  |
| 1 (high level) | Falling edge | Decrement |
| 0 (low level) | Rising edge |  |
| Rising edge | 1 (high level) | Does not count (don't care) |
| Falling edge | 0 (low level) |  |

Phase Counting Mode Application Example: Figure 12.34 shows an example where channel 1 is set to phase counting mode and is teamed with channel 0 to input a two-phase encoder pulse for a servo motor to accurately detect position and speed.

Channel 1 is set to phase counting mode 1 , and the encoder pulse A phase and B phase are input to the TCLKA and TCLKB pins.

Channel 0 is set so that the TCNT counter is cleared on a TGR0C register compare-match, and the TGR0A and TGR0C registers are used with the compare-match function to establish the speed control and position control periods. The TGR0B register is used with the input capture function, and the TGR0B and TGR0D registers are employed for buffer operation. The channel 1 counter
input clock is used as the TGR0B register input capture source, and a pulse width of four times the 2-phase encoder pulse is detected.

The channel 1 TGR1A and TGR1B registers are set for the input capture function, the channel 0 TGR0A and TGR0C register compare-match is used as an input capture source, and all of the control period increment and decrement values are stored.


Figure 12.34 Phase Count Mode Application Example

### 12.4.8 Reset-Synchronized PWM Mode

In the reset-synchronized PWM mode, three-phase output of positive and negative PWM waveforms that share a common wave transition point can be obtained using channels 3 and 4 .

When set for reset-synchronized PWM mode, the TIOC3B, TIOC3D, TIOC4A, TIOC4C, TIOC4B, and TIOC4D pins become PWM output pins and TCNT3 becomes an upcounter.

Table 12.13 shows the PWM output pins used. Table 12.14 shows the settings of the registers.
Table 12.13 Output Pins for Reset-Synchronized PWM Mode

| Channel | Output Pin | Description |
| :--- | :--- | :--- |
| 3 | TIOC3B | PWM output 1 |
| TIOC3D | PWM output 1' (negative-phase waveform of PWM output 1) |  |
| 4 | TIOC4A | PWM output 2 |
| TIOC4C | PWM output 2' (negative-phase waveform of PWM output 2) |  |
|  | TIOC4B | PWM output 3 |
|  | TIOC4D | PWM output 3' (negative-phase waveform of PWM output 3) |

Table 12.14 Register Settings for Reset-Synchronized PWM Mode
Register Description of Contents

| TCNT3 | Initial setting of H'0000 |
| :--- | :--- |
| TCNT4 | Initial setting of H'0000 |
| TGR3A | Set count cycle for TCNT3 |
| TGR3B | Sets the turning point for PWM waveform output by the TIOC3B and TIOC3D pins |
| TGR4A | Sets the turning point for PWM waveform output by the TIOC4A and TIOC4C pins |
| TGR4B | Sets the turning point for PWM waveform output by the TIOC4B and TIOC4D pins |

Procedure for Selecting the Reset-Synchronized PWM Mode (Figure 12.35):

1. Clear the CST3 and CST4 bits in the TSTR to 0 to halt TCNT3 and TCNT4. The resetsynchronized PWM mode must be set up while TCNT3 and TCNT4 are halted.
2. Set bits TPSC2-TPSC0 and CKEG1 and CKEG0 in the TCR to select the counter clock and clock edge for channel 3.
3. Set bits CCLR2-CCLR0 in the TCR3 to select TGRA compare-match as a counter clear source.
4. When performing brushless DC motor control, set bit BDC in the timer gate control register (TGCR) and set the feedback signal input source and output chopping or gate signal direct output.
5. Reset TCNT3 and TCNT4 to H'0000.
6. TGR3A is the period register. Set the waveform period value in TGR3A. Set the transition times of the PWM output waveforms in TGR3B, TGR4A, and TGR4B. Set times within the compare-match range of TCNT3. $\mathrm{X} \leq$ TGR3A (X: set value). With $\mathrm{X}=$ TGRA (cycle $=$ duty cycle), the output waveform goes into toggle operation at the point where TCNT3 $=$ TGR3A $=$ X.
7. Select enabling/disabling of toggle output synchronized with the PMW cycle using bit PSYE in the timer output control register (TOCR), and set the PWM output level with bits OLSP and OLSN.
8. Set bits MD3-MD0 in TMDR3 to B' 1000 to select the reset-synchronized PWM mode. TIOC3B, TIOC3D, TIOC4A, TIOC4B, TIOC4C, and TIOC4D become PWM output pins.
9. Set the CST3 bit in the TSTR to 1 to start the count operation.
10. Set the STR3 bit in the TSTR to 1 to let the TCNT3 start counting.


Figure 12.35 Procedure for Selecting the Reset-Synchronized PWM Mode

Reset-Synchronized PWM Mode Operation: Figure 12.36 shows an example of operation in the reset-synchronized PWM mode. TCNT3 and TCNT4 operate as upcounters. The counter is cleared when a TCNT3 and TGR3A compare-match occurs, and then begins incrementing from H'0000. The PWM output pin output toggles with each occurrence of a TGR3B, TGR4A, TGR4B compare-match, and upon counter clears.


Figure 12.36 Reset-Synchronized PWM Mode Operation Example (When the TOCR's OLSN = 1 and $\mathrm{OLSP}=1$ )

### 12.4.9 Complementary PWM Mode

In the complementary PWM mode, three-phase output of non-overlapping positive and negative PWM waveforms can be obtained using channels 3 and 4.

In complementary PWM mode, TIOC3B, TIOC3D, TIOC4A, TIOC4B, TIOC4C, and TIOC4D pins become PWM output pins, the TIOC3A pin can be set for toggle output synchronized with the PWM period. TCNT3 and TCNT4 function as increment/decrement counters.

Table 12.15 shows the PWM output pins used. Table 12.16 shows the settings of the registers.
A function to directly cut off the PWM output by using an external signal is supported as a port function.

Table 12.15 Output Pins for Complementary PWM Mode

| Channel | Output Pin | Description |
| :--- | :--- | :--- |
| 3 | TIOC3A | Toggle output synchronized with PWM period (or I/O port) |
| TIOC3B | PWM output 1 |  |
| TIOC3C | I/O port (Avoid setting this pin as a timer I/O pin in the <br> complementary PWM mode.) |  |
|  | TIOC3D | PWM output 1 (non-overlapping negative-phase waveform of <br> PWM output 1) |
| TIOC4A | PWM output 2 |  |
| TIOC4B | PWM output 3 |  |
| TIOC4C | PWM output 2 (non-overlapping negative-phase waveform of <br> PWM output 2) |  |
|  | TIOC4D | PWM output 3 (non-overlapping negative-phase waveform of <br> PWM output 3) |

Table 12.16 Register Settings for Complementary PWM Mode

| Channel | Counter/Register | Description | Read/Write from CPU |
| :---: | :---: | :---: | :---: |
| 3 | TCNT3 | Start of up-count from value set in dead time register | Maskable by BSC/BCR1 setting* |
|  | TGR3A | Set TCNT3 upper limit value ( $1 / 2$ carrier cycle + dead time) | Maskable by BSC/BCR1 setting* |
|  | TGR3B | PWM output 1 compare register | Maskable by BSC/BCR1 setting* |
|  | TGR3C | TGR3A buffer register | Always readable/writable |
|  | TGR3D | PWM output 1/TGR3B buffer register | Always readable/writable |
| 4 | TCNT4 | Up-count start, initialized to H'0000 | Maskable by BSC/BCR1 setting* |
|  | TGR4A | PWM output 2 compare register | Maskable by BSC/BCR1 setting* |
|  | TGR4B | PWM output 3 compare register | Maskable by BSC/BCR1 setting* |
|  | TGR4C | PWM output 2/TGR4A buffer register | Always readable/writable |
|  | TGR4D | PWM output 3/TGR4B buffer register | Always readable/writable |
| Timer dead time data register (TDDR) |  | Set TCNT4 and TCNT3 offset value (dead time value) | Maskable by BSC/BCR1 setting* |
| Timer cycle data register (TCDR) |  | Set TCNT4 upper limit value (1/2 carrier cycle) | Maskable by BSC/BCR1 setting* |
| Timer cycle buffer register (TCBR) |  | TCDR buffer register | Always readable/writable |
| Subcounter (TCNTS) |  | Subcounter for dead time generation | Read-only |
| Temporary register 1 (TEMP1) |  | PWM output 1/TGR3B temporary register | Not readable/writable |
| Temporary register 2 (TEMP2) |  | PWM output 2/TGR4A temporary register | Not readable/writable |
| Temporary register 3 (TEMP3) |  | PWM output 3/TGR4B temporary register | Not readable/writable |

Note: * Access can be enabled or disabled according to the setting of bit 13 (MTURWE) in BSC/BCR1 (bus controller/bus control register 1).

## Renesns



Figure 12.37 Block Diagram of Channels 3 and 4 in Complementary PWM Mode

Example of Complementary PWM Mode Setting Procedure: An example of the complementary PWM mode setting procedure is shown in figure 12.38.

1. Clear bits CST3 and CST4 in the timer start register (TSTR) to 0 , and halt timer counter (TCNT) operation. Perform complementary PWM mode setting when TCNT3 and TCNT4 are stopped.
2. Set the same counter clock and clock edge for channels 3 and 4 with bits TPSC2-TPSC0 and bits CKEG1 and CKEG0 in the timer control register (TCR). Use bits CCLR2-CCLR0 to set synchronous clearing only when restarting by a synchronous clear from another channel during complementary PWM mode operation.
3. When performing brushless DC motor control, set bit BDC in the timer gate control register (TGCR) and set the feedback signal input source and output chopping or gate signal direct output.
4. Set the dead time in TCNT3. Set TCNT4 to H'0000.
5. Set only when restarting by a synchronous clear from another channel during complementary PWM mode operation. In this case, synchronize the channel generating the synchronous clear with channels 3 and 4 using the timer synchro register (TSYR).
6. Set the output PWM duty in the duty registers (TGR3B, TGR4A, TGR4B) and buffer registers (TGR3D, TGR4C, TGR4D). Set the same initial value in each corresponding TGR.
7. Set the dead time in the dead time register (TDDR), $1 / 2$ the carrier cycle in the carrier cycle data register (TCDR) and carrier cycle buffer register (TCBR), and $1 / 2$ the carrier cycle plus the dead time in TGR3A and TGR3C.
8. Select enabling/disabling of toggle output synchronized with the PWM cycle using bit PSYE in the timer output control register (TOCR), and set the PWM output level with bits OLSP and OLSN.
9. Select complementary PWM mode in timer mode register 3 (TMDR3). Pins TIOC3A, TIOC3B, TIOC3D, TIOC4A, TIOC4B, TIOC4C, and TIOC4D function as output pins. Do not set in TMDR4.
10. Set enabling/disabling of PWM waveform output pin output in the timer output master enable register (TOER).
11. Set bits CST3 and CST4 in TSTR to 1 simultaneously to start the count operation.


Figure 12.38 Example of Complementary PWM Mode Setting Procedure

Outline of Complementary PWM Mode Operation: In complementary PWM mode, 6-phase PWM output is possible. Figure 12.39 illustrates counter operation in complementary PWM mode, and figure 12.40 shows an example of complementary PWM mode operation.

- Counter operation

In complementary PWM mode, three counters-TCNT3, TCNT4, and TCNTS—perform up/down-count operations.
TCNT3 is automatically initialized to the value set in TDDR when complementary PWM mode is selected and the CST bit in TSTR is 0 .

When the CST bit is set to 1, TCNT3 counts up to the value set in TGR3A, then switches to down-counting when it matches TGR3A,. When the TCNT3 value matches TDDR, the counter switches to up-counting, and the operation is repeated in this way.
TCNT4 is initialized to $\mathrm{H}^{\prime} 0000$.
When the CST bit is set to 1 , TCNT4 counts up in synchronization with TCNT3, and switches to down-counting when it matches TCDR. On reaching H'0000, TCNT4 switches to upcounting, and the operation is repeated in this way.
TCNTS is a read-only counter. It need not be initialized.
When TCNT3 matches TCDR during TCNT3 and TCNT4 up/down-counting, down-counting is started, and when TCNTS matches TCDR, the operation switches to up-counting. When TCNTS matches TGR3A, it is cleared to H'0000.

When TCNT4 matches TDDR during TCNT3 and TCNT4 down-counting, up-counting is started, and when TCNTS matches TDDR, the operation switches to down-counting. When TCNTS reaches H'0000, it is set with the value in TGR3A.
TCNTS is compared with the compare register and temporary register in which the PWM duty is set during the count operation only.


Figure 12.39 Complementary PWM Mode Counter Operation

- Register operation

In complementary PWM mode, nine registers are used, comprising compare registers, buffer registers, and temporary registers. Figure 12.40 shows an example of complementary PWM mode operation.
The registers which are constantly compared with the counters to perform PWM output are TGR3B, TGR4A, and TGR4B. When these registers match the counter, the value set in bits OLSN and OLSP in the timer output control register (TOCR) is output.
The buffer registers for these compare registers are TGR3D, TGR4C, and TGR4D.
Between a buffer register and compare register there is a temporary register. The temporary registers cannot be accessed by the CPU .
Data in a compare register is changed by writing the new data to the corresponding buffer register. The buffer registers can be read or written at any time.
The data written to a buffer register is constantly transferred to the temporary register in the Ta interval. Data is not transferred to the temporary register in the Tb interval. Data written to a buffer register in this interval is transferred to the temporary register at the end of the Tb interval.
The value transferred to a temporary register is transferred to the compare register when TCNTS for which the Tb interval ends matches TGR3A when counting up, or H'0000 when counting down. The timing for transfer from the temporary register to the compare register can be selected with bits MD3-MD0 in the timer mode register (TMDR). Figure 12.40 shows an example in which the mode is selected in which the change is made in the trough. In the tb interval (tb2 in figure 12.40) in which data transfer to the temporary register is not performed, the temporary register has the same function as the compare register, and is compared with the counter. In this interval, therefore, there are two compare registers for onephase output, with the compare register containing the pre-change data, and the temporary register containing the new data. In this interval, the three counters-TCNT3, TCNT4, and TCNTS—and two registers-compare register and temporary register-are compared, and PWM output controlled accordingly.


Figure 12.40 Example of Complementary PWM Mode Operation

- Initialization

In complementary PWM mode, there are six registers that must be initialized.
Before setting complementary PWM mode with bits MD3-MD0 in the timer mode register (TMDR), the following initial register values must be set.
TGR3C operates as the buffer register for TGR3A, and should be set with $1 / 2$ the PWM carrier cycle + dead time Td. The timer cycle buffer register (TCBR) operates as the buffer register for the timer cycle data register (TCDR), and should be set with $1 / 2$ the PWM carrier cycle. Set dead time Td in the timer dead time data register (TDDR).
Set the respective initial PWM duty values in buffer registers TGR3D, TGR4C, and TGR4D. The values set in the five buffer registers excluding TDDR are transferred simultaneously to the corresponding compare registers when complementary PWM mode is set.
Set TCNT4 to H'0000 before setting complementary PWM mode.
Table 12.17 Registers and Counters Requiring Initialization

| Register/Counter | Set Value |
| :--- | :--- |
| TGR3C | $1 / 2$ PWM carrier cycle + dead time Td |
| TDDR | Dead time Td |
| TCBR | $1 / 2$ PWM carrier cycle |
| TGR3D, TGR4C, TGR4D | Initial PWM duty value for each phase |
| TCNT4 | H'0000 |

Note: The TGR3C set value must be the sum of $1 / 2$ the PWM carrier cycle set in TCBR and dead time Td set in TDDR.

- PWM output level setting

In complementary PWM mode, the PWM pulse output level is set with bits OLSN and OLSP in the timer output control register (TOCR).
The output level can be set for each of the three positive phases and three negative phases of 6phase output.
Complementary PWM mode should be cleared before setting or changing output levels.

- Dead time setting

In complementary PWM mode, PWM pulses are output with a non-overlapping relationship between the positive and negative phases. This non-overlap time is called the dead time. The non-overlap time is set in the timer dead time data register (TDDR). The value set in TDDR is used as the TCNT3 counter start value, and creates non-overlap between TCNT3 and TCNT4. Complementary PWM mode should be cleared before changing the contents of TDDR.

- PWM cycle setting

In complementary PWM mode, the PWM pulse cycle is set in two registers-TGR3A, in which the TCNT3 upper limit value is set, and TCDR, in which the TCNT4 upper limit value is set. The settings should be made so as to achieve the following relationship between these two registers:
TGR3A set value = TCDR set value + TDDR set value

The TGR3A and TCDR settings are made by setting the values in buffer registers TGR3C and TCBR. The values set in TGR3C and TCBR are transferred simultaneously to TGR3A and TCDR in accordance with the transfer timing selected with bits MD3-MD0 in the timer mode register (TMDR).
The updated PWM cycle is reflected from the next cycle when the data update is performed at the crest, and from the current cycle when performed in the trough. Figure 12.41 illustrates the operation when the PWM cycle is updated at the crest.
See the following section, Register data updating, for the method of updating the data in each buffer register.


Figure 12.41 Example of PWM Cycle Updating

- Register data updating

In complementary PWM mode, the buffer register is used to update the data in a compare register. The update data can be written to the buffer register at any time. There are five PWM duty and carrier cycle registers that have buffer registers and can be updated during operation. There is a temporary register between each of these registers and its buffer register. When subcounter TCNTS is not counting, if buffer register data is updated, the temporary register value is also rewritten. Transfer is not performed from buffer registers to temporary registers when TCNTS is counting; in this case, the value written to a buffer register is transferred after TCNTS halts.
The temporary register value is transferred to the compare register at the data update timing set with bits MD3-MD0 in the timer mode register (TMDR). Figure 12.42 shows an example of data updating in complementary PWM mode. This example shows the mode in which data updating is performed at both the counter crest and trough.
When rewriting buffer register data, a write to TGR4D must be performed at the end of the update. Data transfer from the buffer registers to the temporary registers is performed simultaneously for all five registers after the write to TGR4D.
A write to TGR4D must be performed after writing data to the registers to be updated, even when not updating all five registers, or when updating the TGR4D data. In this case, the data written to TGR4D should be the same as the data prior to the write operation.


Figure 12.42 Example of Data Update in Complementary PWM Mode

- Initial output in complementary PWM mode

In complementary PWM mode, the initial output is determined by the setting of bits OLSN and OLSP in the timer output control register (TOCR).
This initial output is the PWM pulse non-active level, and is output from when complementary PWM mode is set with the timer mode register (TMDR) until TCNT4 exceeds the value set in the dead time register (TDDR). Figure 12.43 shows an example of the initial output in complementary PWM mode.
An example of the waveform when the initial PWM duty value is smaller than the TDDR value is shown in figure 12.44.


Figure 12.43 Example of Initial Output in Complementary PWM Mode (1)


Figure 12.44 Example of Initial Output in Complementary PWM Mode (2)

- Complementary PWM mode PWM output generation method

In complementary PWM mode, 3-phase output is performed of PWM waveforms with a nonoverlap time between the positive and negative phases. This non-overlap time is called the dead time.
A PWM waveform is generated by output of the output level selected in the timer output control register in the event of a compare-match between a counter and data register. While TCNTS is counting, data register and temporary register values are simultaneously compared to create consecutive PWM pulses from 0 to $100 \%$. The relative timing of on and off comparematch occurrence may vary, but the compare-match that turns off each phase takes precedence to secure the dead time and ensure that the positive phase and negative phase on times do not overlap. Figures 12.45 to 12.47 show examples of waveform generation in complementary PWM mode.
The positive phase/negative phase off timing is generated by a compare-match with the solidline counter, and the on timing by a compare-match with the dotted-line counter operating with a delay of the dead time behind the solid-line counter. In the T1 period, compare-match a that turns off the negative phase has the highest priority, and compare-matches occurring prior to a are ignored. In the T2 period, compare-match $\mathbf{c}$ that turns off the positive phase has the highest priority, and compare-matches occurring prior to $\mathbf{c}$ are ignored.
In normal cases, compare-matches occur in the order $\mathbf{a} \rightarrow \mathbf{b} \rightarrow \mathbf{c} \rightarrow \mathbf{d}$ (or $\mathbf{c} \rightarrow \mathbf{d} \rightarrow \mathbf{a}^{\prime} \rightarrow \mathbf{b}^{\prime}$ ), as shown in figure 12.45 .
If compare-matches deviate from the $\mathbf{a} \rightarrow \mathbf{b} \rightarrow \mathbf{c} \rightarrow \mathbf{d}$ order, since the time for which the negative phase is off is less than twice the dead time, the figure shows the positive phase as not being turned on. If compare-matches deviate from the $\mathbf{c} \rightarrow \mathbf{d} \rightarrow \mathbf{a}^{\prime} \rightarrow \mathbf{b}^{\prime}$ order, since the time for which the positive phase is off is less than twice the dead time, the figure shows the negative phase as not being turned on.
If compare-match $\mathbf{c}$ occurs first following compare-match $\mathbf{a}$, as shown in figure 12.46, compare-match $\mathbf{b}$ is ignored, and the negative phase is turned off by compare-match $\mathbf{d}$. This is because turning off of the positive phase has priority due to the occurrence of compare-match $\mathbf{c}$ (positive phase off timing) before compare-match $\mathbf{b}$ (positive phase on timing) (consequently, the waveform does not change since the positive phase goes from off to off).
Similarly, in the example in figure 12.47, compare-match a' with the new data in the temporary register occurs before compare-match $\mathbf{c}$, but other compare-matches occurring up to $\mathbf{c}$, which turns of the positive phase, are ignored. As a result, the positive phase is not turned on.
Thus, in complementary PWM mode, compare-matches at turn-off timings take precedence, and turn-on timing compare-matches that occur before a turn-off timing compare-match are ignored.


Figure 12.45 Example of Complementary PWM Mode Waveform Output (1)


Figure 12.46 Example of Complementary PWM Mode Waveform Output (2)


Figure 12.47 Example of Complementary PWM Mode Waveform Output (3)


Figure $\mathbf{1 2 . 4 8}$ Example of Complementary PWM Mode $\mathbf{0 \%}$ and $\mathbf{1 0 0 \%}$ Waveform Output (1)


Figure 12.49 Example of Complementary PWM Mode $\mathbf{0 \%}$ and $\mathbf{1 0 0 \%}$ Waveform Output (2)


Figure 12.50 Example of Complementary PWM Mode $\mathbf{0 \%}$ and $\mathbf{1 0 0 \%}$ Waveform Output (3)


Figure $\mathbf{1 2 . 5 1}$ Example of Complementary PWM Mode $\mathbf{0 \%}$ and $\mathbf{1 0 0 \%}$ Waveform Output (4)


Figure 12.52 Example of Complementary PWM Mode $\mathbf{0 \%}$ and $\mathbf{1 0 0 \%}$ Waveform Output (5) 370

- Complementary PWM mode $0 \%$ and $100 \%$ duty output

In complementary PWM mode, $0 \%$ and $100 \%$ duty cycles can be output as required. Figures 12.48 to 12.52 show output examples.
$100 \%$ duty output is performed when the data register value is set to $\mathrm{H}^{\prime} 0000$. The waveform in this case has a positive phase with a $100 \%$ on-state. $0 \%$ duty output is performed when the data register value is set to the same value as TGR3A. The waveform in this case has a positive phase with a $100 \%$ off-state.
On and off compare-matches occur simultaneously, but if a turn-on compare-match and turnoff compare-match for the same phase occur simultaneously, both compare-matches are ignored and the waveform does not change.

- Toggle output synchronized with PWM cycle

In complementary PWM mode, toggle output can be performed in synchronization with the PWM carrier cycle by setting the PSYE bit to 1 in the timer output control register (TOCR). An example of a toggle output waveform is shown in figure 12.53.
This output is toggled by a compare-match between TCNT3 and TGR3A and a compare-match between TCNT4 and H'0000.
The output pin for this toggle output is the TIOC3A pin. The initial output is 1.


Figure 12.53 Example of Toggle Output Waveform Synchronized with PWM Output

- Counter clearing by another channel

In complementary PWM mode, by setting a mode for synchronization with another channel by means of the timer synchro register (TSYR), and selecting synchronous clearing with bits CCLR2-CCLR0 in the timer control register (TCR), it is possible to have TCNT3, TCNT4, and TCNTS cleared by another channel.
Figure 12.54 illustrates the operation.
Use of this function enables counter clearing and restarting to be performed by means of an external signal.


Figure 12.54 Counter Clearing Synchronized with Another Channel

- Example of AC synchronous motor (brushless DC motor) drive waveform output In complementary PWM mode, a brushless DC motor can easily be controlled using the timer gate control register (TGCR). Figures 12.55 to 12.58 show examples of brushless DC motor drive waveforms created using TGCR.
When output phase switching for a 3-phase brushless DC motor is performed by means of external signals detected with a Hall element, etc., clear the FB bit in TGCR to 0 . In this case, the external signals indicating the polarity position are input to channel 0 timer input pins TIOC0A, TIOC0B, and TIOC0C (set with PFC). When an edge is detected at pin TIOC0A, TIOC0B, or TIOC0C, the output on/off state is switched automatically.
When the FB bit is 1, the output on/off state is switched when the UF, VF, or WF bit in TGCR is cleared to 0 or set to 1 .
The drive waveforms are output from the complementary PWM mode 6-phase output pins. With this 6-phase output, in the case of on output, it is possible to use complementary PWM mode output and perform chopping output by setting the N bit or P bit to 1 . When the N bit or $P$ bit is 0 , level output is selected.
The 6-phase output active level (on output level) can be set with the OLSN and OLSP bits in the timer output control register (TOCR) regardless of the setting of the N and P bits. When using this mode, set the 6-phase output waveform to High active (Low active is also permitted for A masks).


Figure 12.55 Example of Output Phase Switching by External Input (1)


Figure 12.56 Example of Output Phase Switching by External Input (2)


Figure 12.57 Example of Output Phase Switching by Means of UF, VF, WF Bit Settings (1)


Figure 12.58 Example of Output Phase Switching by Means of UF, VF, WF Bit Settings (2)

- $\mathrm{A} / \mathrm{D}$ conversion start request setting

In complementary PWM mode, an A/D conversion start request can be issued using a TGR3A compare-match or a compare-match on a channel other than channels 3 and 4.
When start requests using a TGR3A compare-match are set, A/D conversion can be started at the center of the PWM pulse.
A/D conversion start requests can be set by setting the TTGE bit to 1 in the timer interrupt enable register (TIER).

Complementary PWM Mode Output Protection Function: Complementary PWM mode output has the following protection functions.

- Register and counter miswrite prevention function

With the exception of the buffer registers, which can be rewritten at any time, access by the CPU can be enabled or disabled for the mode registers, control registers, compare registers, and counters used in complementary PWM mode by means of bit 13 in the bus controller's bus control register 1 (BCR1). The registers and counters concerned are listed in table 12.3.
This function enables miswriting due to CPU runaway to be prevented by disabling CPU access to the mode registers, control registers, and counters.

- Halting of PWM output by external signal

The 6-phase PWM output pins can be set automatically to the high-impedance state by inputting specified external signals. There are four external signal input pins.
See section 12.9, Port Output Enable (POE), for details.

- Halting of PWM output when oscillator is stopped

If it is detected that the clock input to the SH7040 chip has stopped, the 6-phase PWM output pins automatically go to the high-impedance state. The pin states are not guaranteed when the clock is restarted.
See section 4.4, Oscillator Halt Function, for details.

### 12.5 Interrupts

### 12.5.1 Interrupt Sources and Priority Ranking

The MTU has three interrupt sources: TGR register compare-match/input captures, TCNT counter overflows and TCNT counter underflows. Because each of these three types of interrupts are allocated its own dedicated status flag and enable/disable bit, the issuing of interrupt request signals to the interrupt controller can be independently enabled or disabled.

When an interrupt source is generated, the corresponding status flag in the timer status register (TSR) is set to 1 . If the corresponding enable/disable bit in the timer input enable register (TIER) is set to 1 at this time, the MTU makes an interrupt request of the interrupt controller. The interrupt request is canceled by clearing the status flag to 0 .

The channel priority order can be changed with the interrupt controller. The priority ranking within a channel is fixed. For more information, see section 6, Interrupt Controller (INTC).

Table 12.17 lists the MTU interrupt sources.
Input Capture/Compare Match Interrupts: If the TGIE bit of the timer input enable register (TIER) is already set to 1 when the TGF flag in the timer status register (TSR) is set to 1 by a TGR register input capture/compare-match of any channel, an interrupt request is sent to the interrupt controller. The interrupt request is canceled by clearing the TGF flag to 0 . The MTU has 16 input capture/compare-match interrupts; four each for channels 0,3 , and 4 , and two each for channels 1 and 2.

Overflow Interrupts: If the TCIEV bit of the TIER is already set to 1 when the TCFV flag in the TSR is set to 1 by a TCNT counter overflow of any channel, an interrupt request is sent to the interrupt controller. The interrupt request is canceled by clearing the TCFV flag to 0 . The MTU has five overflow interrupts, one for each channel.

Underflow Interrupts: If the TCIEU bit of the TIER is already set to 1 when the TCFU flag in the TSR is set to 1 by a TCNT counter underflow of any channel, an interrupt request is sent to the interrupt controller. The interrupt request is canceled by clearing the TCFU flag to 0 . The MTU has two underflow interrupts, one each for channels 1 and 2.

Table 12.17 MTU Interrupt Sources


Note: * Indicates the initial status following reset. The ranking of channels can be altered using the interrupt controller.

### 12.5.2 DTC/DMAC Activation

DTC Activation: The TGR register input capture/compare-match interrupt of any channel can be used as a source to activate the on-chip data transfer controller (DTC). For details, refer to section 8, Data Transfer Controller (DTC).

The MTU has 17 input capture/compare-match interrupts that can be used as DTC activation sources, four each for channels 0 and 3, two each for channels 1 and 2 , and five for channel 4 .

DMAC Activation: The TGRA register input capture/compare-match interrupt of any channel can be used as a source to activate the on-chip DMAC. For details, refer to section 11, Direct Memory Access Controller (DMAC).

The MTU has 5 TGRA register input capture/compare-match interrupts, one for any channel, that can be used as DMAC activation sources.

### 12.5.3 A/D Converter Activation

The TGRA register input capture/compare-match of any channel can be used to activate the onchip A/D converter.

If the TTGE bit of the TIER is already set to 1 when the TGFA flag in the TSR is set to 1 by a TGRA register input capture/compare-match of any of the channels, an A/D conversion start request is sent to the $A / D$ converter. If the MTU conversion start trigger is selected at such a time on the $A / D$ converter side when this happens, the $A / D$ conversion starts.

The MTU has 5 TGRA register input capture/compare-match interrupts, one for each channel, that can be used as A/D converter activation sources.

### 12.6 Operation Timing

### 12.6.1 Input/Output Timing

TCNT Count Timing: Count timing for the TCNT counter with internal clock operation is shown in figure 12.59. Count timing with external clock operation (normal mode) is shown in figure 12.60 , and figure 12.61 shows count timing with external clock operation (phase counting mode).


Figure 12.59 TCNT Count Timing during Internal Clock Operation


Figure 12.60 TCNT Count Timing during External Clock Operation (Normal Mode)


Figure 12.61 TCNT Count Timing during External Clock Operation (Phase Counting Mode)

Output Compare Output Timing: The compare-match signal is generated at the final state of TCNT and TGR matching. When a compare-match signal is issued, the output value set in TIOR or TOCR is output to the output compare output pin (TIOC pin). After TCNT and TGR matching, a compare-match signal is not issued until immediately before the TCNT input clock.

Output compare output timing (normal mode and PWM mode) is shown in figure 12.62. See figure 12.63 for output compare output timing in complementary PWM mode and reset sync PWM mode.


Figure 12.62 Output Compare Output Timing (Normal Mode/PWM Mode)


Figure 12.63 Output Compare Output Timing (Complementary PWM Mode/Reset Sync PWM Mode)

Input Capture Signal Timing: Figure 12.64 illustrates input capture timing.


Figure 12.64 Input Capture Input Signal Timing

Counter Clearing Timing Due to Compare-Match/Input Capture: Timing for counter clearing due to compare-match is shown in figure 12.65. Figure 12.66 shows the timing for counter clearing due to input capture.


Figure 12.65 Counter Clearing Timing (Compare-Match)


Figure 12.66 Counter Clearing Timing (Input Capture)

Buffer Operation Timing: Compare-match buffer operation timing is shown in figure 12.67.
Figure 12.68 shows input capture buffer operation timing.


Figure 12.67 Buffer Operation Timing (Compare-Match)


Figure 12.68 Buffer Operation Timing (Input Capture)

### 12.6.2 Interrupt Signal Timing

Setting TGF Flag Timing during Compare-Match: Figure 12.69 shows timing for the TGF flag of the timer status register (TSR) due to compare-match, as well as TGI interrupt request signal timing.


Figure 12.69 TGI Interrupt Timing (Compare Match)
Setting TGF Flag Timing during Input Capture: Figure 12.70 shows timing for the TGF flag of the timer status register (TSR) due to input capture, as well as TGI interrupt request signal timing.


Figure 12.70 TGI Interrupt Timing (Input Capture)
Setting Timing for Overflow Flag (TCFV)/Underflow Flag (TCFU): Figure 12.71 shows timing for the TCFV flag of the timer status register (TSR) due to overflow, as well as TCIV interrupt request signal timing. Figure 12.72 shows timing for the TCFU flag of the timer status register (TSR) due to underflow, as well as TCIU interrupt request signal timing. Figure 12.73 shows timing for the TCFV flag of TSR4 due to underflow in complementary PWM mode, as well as TCIV interrupt request signal timing.


Figure 12.71 TCIV Interrupt Setting Timing


Figure 12.72 TCIU Interrupt Setting Timing


Figure 12.73 TCIV Interrupt Setting Timing (TSR4, Complementary PWM Mode)

Status Flag Clearing Timing: The status flag is cleared when the CPU reads a 1 status followed by a 0 write. For DTC/DMA controller activation, clearing can also be done automatically. Figure 12.74 shows the timing for status flag clearing by the CPU . Figure 12.75 shows timing for clearing due to the DTC/DMA controller.


Figure 12.74 Timing of Status Flag Clearing by the CPU


Figure 12.75 Timing of Status Flag Clearing by DTC/DMAC Activation

### 12.7 Notes and Precautions

This section describes contention and other matters requiring special attention during MTU operations.

### 12.7.1 Input Clock Limitations

The input clock pulse width, in the case of single edge, must be 1.5 states or greater, and 2.5 states or greater for both edges. Normal operation cannot be guaranteed with lesser pulse widths.

In phase counting mode, the phase difference between the two input clocks and the overlap must be 1.5 states or greater for each, and the pulse width must be 2.5 states or greater. Input clock conditions for phase counting mode are shown in figure 12.76.


Figure 12.76 Phase Difference, Overlap, and Pulse Width in Phase Count Mode

### 12.7.2 Note on Cycle Setting

When setting a counter clearing by compare-match, clearing is done in the final state when TCNT matches the TGR value (update timing for count value on TCNT match). The actual number of states set in the counter is given by the following equation:

$$
\mathrm{f}=\frac{\phi}{(\mathrm{N}+1)}
$$

(f: counter frequency, $\phi$ : operating frequency, N : value set in the TGR)

### 12.7.3 Contention between TCNT Write and Clear

If a counter clear signal is issued in the $\mathrm{T}_{2}$ state during the TCNT write cycle, TCNT clearing has priority, and TCNT write is not conducted (figure 12.77).


Figure 12.77 TCNT Write and Clear Contention

### 12.7.4 Contention between TCNT Write and Increment

If a count-up signal is issued in the $\mathrm{T}_{2}$ state during the TCNT write cycle, TCNT write has priority, and the counter is not incremented (figure 12.78).


Figure 12.78 TCNT Write and Increment Contention

### 12.7.5 Contention between Buffer Register Write and Compare Match

If a compare-match occurs in the $\mathrm{T}_{2}$ state of the TGR write cycle, data is transferred by the buffer operation from the buffer register to the TGR. Data to be transferred differs depending on channels 0 and 3 and 4: data on channel 0 is that after write, and on channels 3 and 4, before write (figures 12.79 and 12.80).


Figure 12.79 TGR Write and Compare-Match Contention (Channel 0)


Figure 12.80 TGR Write and Compare-Match Contention (Channels 3 and 4)

### 12.7.6 Contention between TGR Read and Input Capture

If an input capture signal is issued in the $T_{1}$ state of the TGR read cycle, the read data is that after input capture transfer (figure 12.81).


Figure 12.81 TGR Read and Input Capture Contention

### 12.7.7 Contention between TGR Write and Input Capture

If an input capture signal is issued in the $\mathrm{T}_{2}$ state of the TGR read cycle, input capture has priority, and TGR write does not occur (figure 12.82).


Figure 12.82 TGR Write and Input Capture Contention

### 12.7.8 Contention between Buffer Register Write and Input Capture

If an input capture signal is issued in the $\mathrm{T}_{2}$ state of the buffer write cycle, write to the buffer register does not occur, and buffer operation takes priority (figure 12.83).


Figure 12.83 Buffer Register Write and Input Capture Contention

### 12.7.9 Contention between TGR Write and Compare Match

If a compare-match occurs in the $\mathrm{T}_{2}$ state of the TGR write cycle, data is written to the TGR and a compare-match signal is issued (figure 12.84).


Figure 12.84 TGR Write and Compare Match Contention

### 12.7.10 TCNT2 Write and Overflow/Underflow Contention in Cascade Connection

With timer counters TCNT1 and TCNT2 in a cascade connection, when a contention occurs during TCNT1 count (during a TCNT2 overflow/underflow) in the $\mathrm{T}_{2}$ state of the TCNT2 write cycle, the write to TCNT2 is conducted, and the TCNT1 count signal is prohibited. At this point, if there is match with TGR1A or TGR1B and the TCNT1 value, a compare signal is issued. The timing is shown in figure 12.85 .

For cascade connections, be sure to synchronize settings for channels 1 and 2 when setting TCNT clearing.


Figure 12.85 TCNT2 Write and Overflow/Underflow Contention with Cascade Connection

### 12.7.11 Counter Value during Complementary PWM Mode Stop

When counting operation is suspended with TCNT3 and TCNT4 in complementary PWM mode, TCNT3 has the timer dead time register (TDDR) value, and TCNT4 is held at $\mathrm{H}^{\prime} 0000$.

When restarting complementary PWM mode, counting begins automatically from the initialized state (figure 12.86).

When counting begins in another operating mode, be sure that TCNT3 and TCNT4 are set to the initial values.


Figure 12.86 Counter Value during Complementary PWM Mode Stop

### 12.7.12 Buffer Operation Setting in Complementary PWM Mode

In complementary PWM mode, conduct rewrites by buffer operation for the PWM cycle setting register (TGR3A), PWM carrier cycle setting register (TCDR) and duty setting registers (TGR3B, TRG4A, and TGR4B).

In complementary PWM mode, channel 3 and channel 4 buffers operate in accordance with bit settings BFA and BFB of TMDR3. When TMDR3's BFA bit is set to 1 , TGR3C functions as a buffer register for TGR3A. At the same time, TGR4C functions as the buffer register for TRG4A, while the TCBR functions as the TCDR's buffer register.

### 12.7.13 Reset Sync PWM Mode Buffer Operation and Compare Match Flag

When setting buffer operation for reset sync PWM mode, set the BFA and BFB bits of TMDR4 to 0 . The TIOC4C pin will be unable to produce its waveform output if the BFA bit of TMDR4 is set to 1 .

In reset sync PWM mode, the channel 3 and channel 4 buffers operate in accordance with the BFA and BFB bit settings of TMDR3. For example, if the BFA bit of TMDR3 is set to 1, TGR3C functions as the buffer register for TGR3A. At the same time, TGR4C functions as the buffer register for TRG4A.

When setting buffer operation for reset sync PWM mode, take particular care, since comparematch flag TGFC bit and TGFD bit operations differ with TSR3 and TSR4.

The TGFC bit and TGFD bit of TSR3 are not set when TGR3C and TGR3D are operating as buffer registers. On the other hand, TSR4's TGFC and TGFD bits are set even when TGR4C and TGR4D are operating as buffer registers.

When buffer operation has been set for reset sync PWM mode, set the timer interrupt enable register's (TIER4) TGIEC and TGIED bits to 0 , to prohibit interrupt output.

Figure 12.87 shows an example of operations for TGR3, TGR4, TIOC3, and TIOC4, with TMDR3's BFA and BFB bits set to 1 , and TMDR4's BFA and BFB bits set to 0 .


Figure 12.87 Buffer Operation and Compare-Match Flags in Reset Sync PWM Mode

- A mask operation

For A mask, the above operation is modified as follows:
When setting buffer operation for reset sync PWM mode, set the BFA and BFB bits of TMDR4 to 0 . The TIOC4C pin will be unable to produce its waveform output if the BFA bit of TMDR4 is set to 1 .
In reset sync PWM mode, the channel 3 and channel 4 buffers operate in accordance with the BFA and BFB bit settings of TMDR3. For example, if the BFA bit of TMDR3 is set to 1 ,
TGR3C functions as the buffer register for TGR3A. At the same time, TGR4C functions as the buffer register for TRG4A.
When setting buffer operation for reset sync PWM mode, the compare-match flag TGFC bit and TGFD bit operations will be the same for TSR3 and TSR4.
The TGFC bit and TGFD bit of TSR3 are not set when TGR3C and TGR3D are operating as buffer registers. The TGFC bit and TGFD bit of TSR4 are not set when TGR4C and TGR4D are operating as buffer registers.

When setting the buffer operation in the reset synchronous PWM mode it is not necessary to set the timer interrupt enable register's (TIER4) TGIEC and TGIED bits to 0 , to prohibit interrupt output.
Figure 12.88 shows an example of operations for TGR3, TGR4, TIOC3, and TIOC4, with TMDR3's BFA and BFB bits set to 1 , and TMDR4's BFA and BFB bits set to 0 .


Figure 12.88 Buffer Operation and Compare-Match Flags in Reset Sync PWM Mode (for A Mask)

### 12.7.14 Overflow Flags in Reset Sync PWM Mode

When set to reset sync PWM mode, TCNT3 and TCNT4 start counting when the CST3 bit of TSTR is set to 1 . At this point, TCNT4's count clock source and count edge obey the TCR3 setting.

In reset sync PWM mode, with cycle register TGR3A's set value at H'FFFF, take care when specifying TGR3A compare-match for the counter clear source, since the operation of the overflow flag (TCFV bit) differs with TSR3 and TSR4.

When TCNT3 and TCNT4 count up to H'FFFF, a compare-match occurs with TGR3A, and TCNT3 and TCNT4 are both cleared. At this point, TSR3's TCFV bit is not set, but the TCFV bit of TSR4 is set.

This can be avoided by sync setting for channel 3 and channel 4. Set the SYNC3 and SYNC4 bits of the timer sync register (TSYR) to 1, compare-match with TGR3A by TCR3 for the counter clear source, and sync clear with TCR4. This gives the sync setting for channels 3 and 4 .

Figure 12.89 shows a TCFV bit operation example in reset sync PWM mode with a set value for cycle register TGR3A of H'FFFF, when a TGR3A compare-match has been specified without synchronous setting for the counter clear source.


Figure 12.89 Reset Sync PWM Mode Overflow Flag

- A mask operation

For A mask, the above operation is modified as follows:
When set to reset sync PWM mode, TCNT3 and TCNT4 start counting when the CST3 bit of TSTR is set to 1. At this point, TCNT4's count clock source and count edge obey the TCR3 setting.
In reset sync PWM mode, with cycle register TGR3A's set value at H'FFFF and specifying TGR3A compare-match for the counter clear source, the operation of the overflow flag TCFV bit for both TSR3 and TSR4 will be the same.
When TCNT3 and TCNT4 count up to H'FFFF, a compare-match occurs with TGR3A, and TCNT3 and TCNT4 are both cleared. At this point, TCFV bits for TSR3 and TSR4 are not set.
Figure 12.90 shows a TCFV bit operation example in reset sync PWM mode with a set value for cycle register TGR3A of H'FFFF, when a TGR3A compare-match has been specified without synchronous setting for the counter clear source.


Figure. 12.90 Reset Sync PWM Mode Overflow Flag (for A Mask)

### 12.7.15 Notes on Compare Match Flags in Complementary PWM Mode

In complementary PWM mode, buffer register compare-match flags can be set only for compare with three counters (TCNT3, TCNT4, and TCNTS).

Note that when the buffer register set value is dead time (Td), 2Td, TGR3A - Td, or TGR3A - 2Td, the buffer register compare-match flag may not be set.

Figure 12.91 gives a description when TGR3B is the specified duty setting register, TGR3D the buffer register with TGR3A - Td as the buffer register set value.


Figure 12.91 Special Properties of Compare Match Flag in Complementary PWM Mode

## renesns

- A mask operation

For A mask, the above operation is modified as follows:
In complementary PWM mode, buffer register compare-match flags can be set only for compare with three counters (TCNT3, TCNT4, and TCNTS).
Special properties of compare match flag disappear and compare match flags of buffer registers are set to all set values of buffer registers.
Figure 12.92 shows an example when setting the duty setting register to TGR3B, buffer register to TGR3D and Buffer register to TGR3A-Td.


Figure. 12.92 Special Properties of Compare Match Flag in Complementary PWM Mode (for A Mask)

### 12.7.16 Contention between Overflow/Underflow and Counter Clearing

If overflow/underflow and counter clearing occur simultaneously, the TCFV/TCFU flag in TSR is not set and TCNT clearing takes precedence.

Figure 12.93 shows the operation timing when a TGR compare-match is specified as the clearing source, and H'FFFF is set in TGR.


Figure 12.93 Contention between Overflow and Counter Clearing

### 12.7.17 Contention between TCNT Write and Overflow/Underflow

If there is an up-count or down-count in the T2 state of a TCNT write cycle, and overflow/underflow occurs, the TCNT write takes precedence and the TCFV/TCFU flag in TSR is not set .

Figure 12.94 shows the operation timing in this case.


Figure 12.94 Contention between TCNT Write and Overflow

### 12.7.18 Cautions on Transition from Normal Operation or PWM Mode 1 to ResetSynchronous PWM Mode

When making a transition from channel 3 or 4 normal operation or PWM mode 1 to resetsynchronous PWM mode, if the counter is halted with the output pins (TIOC3B, TIOC3D, TIOC4A, TIOC4C, TIOC4B, TIOC4D) in the high-impedance state, followed by the transition to reset-synchronous PWM mode and operation in that mode, the initial pin output will not be correct.

When making a transition from normal operation to reset-synchronous PWM mode, write H'11 to registers TIOR3H, TIOR3L, TIOR4H, and TIOR4L to initialize the output pins to low level output, then set an initial register value of $\mathrm{H}^{\prime} 00$ before making the mode transition.

When making a transition from PWM mode 1 to reset-synchronous PWM mode, first switch to normal operation, then initialize the output pins to low level output and set an initial register value of $\mathrm{H}^{\prime} 00$ before making the transition to reset-synchronous PWM mode.

### 12.7.19 Output Level in Complementary PWM Mode and Reset-Synchronous PWM Mode

When channels 3 and 4 are in complementary PWM mode or reset-synchronous PWM mode, the PWM waveform output level is set with the OLSP and OLSN bits in the timer output control register (TOCR). In the case of complementary PWM mode or reset-synchronous PWM mode, TIOR should be set to $\mathrm{H}^{\prime} 00$.

### 12.7.20 Cautions on Using the Chopping Function in Complementary PWM Mode or Reset Synchronous PWM Mode (A Mask Excluded)

When channels 3 and 4 are in complementary PWM mode or reset-synchronous PWM mode and using the chopping output function, setting the PWM waveform output level to low active with the OLSP and OLSN bits in the timer output control register (TOCR) will output an incorrect gate signal or chopping output.

When channels 3 and 4 are in complementary PWM mode or reset-synchronous PWM mode and using the chopping output function, the PWM output level should be set to high active.

### 12.7.21 Cautions on Carrying Out Buffer Operation of Channel 0 in PWM Mode (A Mask Excluded)

In PWM mode 1, the TGRA and TGRB registers are used in pairs and PWM waveform is output to the TIOCA pin. In the same manner, the TGRC and TGRD registers are used in pairs and PWM waveform is output to the TIOCC pin. If either the TGRC or TGRD register is operating as a buffer register, the TIOCC pin cannot execute default output setting or PWM waveform output with the I/O control register (TIOR).

Note that for channel 0, the TIOCC pin allows both default output setting by TIOR and PWM output when setting buffer operation only for the TGRD register in PWM mode.

When using channel 0 in PWM mode 1 and setting buffer operation, use both the TGRC and TGRD registers as buffer registers.

### 12.7.22 Cautions on Restarting with Sync Clear of Another Channel in Complementary PWM Mode (A Mask Excluded)

The complementation PWM mode operates while waiting for the current, next and the following set values as values of the PWM duty. If clearing sync from another channel during operation, the PWM output will return to the default output and restart.

When restarting with sync clear, the following operations may occur:

1. When restarting with sync clear, the next set value is used for the PWM duty, however, the following set value may be used by mistake.
2. If sync clear and the setting of the value following the next value of PWM duty (write to TGR4D) occurs at the same time, the next set value may be overwritten.

## How to avoid 1

When selecting the mode to transfer using the crest/trough in the complementary PWM transfer mode, set the value following the next value of the PWM duty (write to TGR4D) while the temporary register is not executing comparisons. Furthermore, set the occurrence timing of sync clear while the temporary register is not executing comparisons.

When selecting the mode to transfer using the crest in the transfer mode, set the value following the next value of the PWM duty (write to TGR4D) while the temporary register is not executing comparisons and while TCNT3 and TCNT4 are counting up. Furthermore, set the occurrence timing of sync clear while the temporary register is not executing comparisons and while TCNT3 and TCNT4 are counting up.

When selecting the mode to transfer using the trough in the transfer mode, set the value following the next value of the PWM duty (write to TGR4D) while the temporary register is not executing comparisons and while TCNT3 and TCNT4 are counting down. Furthermore, set the occurrence timing of sync clear while the temporary register is not executing comparisons and while TCNT3 and TCNT4 are counting down.

## How to avoid 2

Regardless of the transfer mode, set so that the sync clear and the setting of the value following the next value (write to TGR4D) does not occur at the same time.

Figure 12.95 shows an example of the duration while the temporary register is executing comparisons. initial TB, TA, and TB indicate the duration of the temporary register comparison.


Figure. 12.95 Temporary Register Comparison Execution Time

### 12.8 MTU Output Pin Initialization

### 12.8.1 Operating Modes

The MTU has the following six operating modes. Waveform output is possible in all of these modes.

- Normal mode (channels 0 and 4 )
- PWM mode 1 (channels 0 and 4 )
- PWM mode 2 (channels 0 and 2)
- Phase counting modes 1-4 (channels 1 and 2)
- Complementary PWM mode (channels 3 and 4 )
- Reset-synchronous PWM mode (channels 3 and 4 )

The MTU output pin initialization method for each of these modes is described in this section.

### 12.8.2 Reset Start Operation

The MTU output pins (TIOC*) are initialized low by a reset and in standby mode. Since MTU pin function selection is performed by the pin function controller ( PFC ), when the PFC is set, the MTU pin states at that point are output to the ports. When MTU output is selected by the PFC immediately after a reset, the MTU output initial level, low, is output directly at the port. When the active level is low, the system will operate at this point, and therefore the PFC setting should be made after initialization of the MTU output pins is completed.

Note: Channel number and port notation are substituted for *.

### 12.8.3 Operation in Case of Re-Setting Due to Error During Operation, Etc.

If an error occurs during MTU operation, MTU output should be cut by the system. Cutoff is performed by switching the pin output to port output with the PFC and outputting the inverse of the active level. For large-current pins, output can also be cut by hardware, using port output enable (POE). The pin initialization procedures for re-setting due to an error during operation, etc., and the procedures for restarting in a different mode after re-setting, are shown below.

The MTU has six operating modes, as stated above. There are thus 36 mode transition combinations, but some transitions are not available with certain channel and mode combinations. Possible mode transition combinations are shown in table 12.18.

Table 12.18 Mode Transition Combinations

|  | After |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: |
| Before | Normal | PWM1 | PWM2 | PCM | CPWM | RPWM |  |
| Normal | $(1)$ | $(2)$ | $(3)$ | $(4)$ | $(5)$ | $(6)$ |  |
| PWM1 | $(7)$ | $(8)$ | $(9)$ | $(10)$ | $(11)$ | $(12)$ |  |
| PWM2 | $(13)$ | $(14)$ | $(15)$ | $(16)$ | None | None |  |
| PCM | $(17)$ | $(18)$ | $(19)$ | $(20)$ | None | None |  |
| CPWM | $(21)$ | $(22)$ | None | None | $(23)$ | $(24)$ |  |
| RPWM | $(25)$ | $(26)$ | None | None | $(27)$ | $(28)$ |  |

Legend:
Normal: Normal mode
PWM1: PWM mode 1
PWM2: PWM mode 2
PCM: Phase counting modes 1-4
CPWM: Complementary PWM mode
RPWM: Reset-synchronous PWM mode

The above abbreviations are used in some places in following descriptions.

### 12.8.4 Overview of Initialization Procedures and Mode Transitions in Case of Error during Operation, Etc.

- When making a transition to a mode (Normal, PWM1, PWM2, PCM) in which the pin output level is selected by the timer I/O control register (TIOR) setting, initialize the pins by means of a TIOR setting.
- In PWM mode 1, since a waveform is not output to the TIOC*B (TIOC *D) pin, setting TIOR will not initialize the pins. If initialization is required, carry it out in normal mode, then switch to PWM mode 1.
- In PWM mode 2, since a waveform is not output to the cycle register pin, setting TIOR will not initialize the pins. If initialization is required, carry it out in normal mode, then switch to PWM mode 2.
- In normal mode or PWM mode 2, if TGRC and TGRD operate as buffer registers, setting TIOR will not initialize the buffer register pins. If initialization is required, clear buffer mode, carry out initialization, then set buffer mode again.
- In PWM mode 1, if either TGRC or TGRD operates as a buffer register, setting TIOR will not initialize the TGRC pin. To initialize the TGRC pin, clear buffer mode, carry out initialization, then set buffer mode again.
- When making a transition to a mode (CPWM, RPWM) in which the pin output level is selected by the timer output control register (TOCR) setting, switch to normal mode and perform initialization with TIOR, then restore TIOR to its initial value, and temporarily disable channel 3 and 4 output with the timer output master enable register (TOER). Then operate the unit in accordance with the mode setting procedure (TOCR setting, TMDR setting, TOER setting).

Pin initialization procedures are described below for the numbered combinations in table 12.19. The active level is assumed to be low.

Note: Channel number is substituted for * indicated in this article.
(1) Operation when Error Occurs during Normal Mode Operation, and Operation is

Restarted in Normal Mode: Figure 12.96 shows an explanatory diagram of the case where an error occurs in normal mode and operation is restarted in normal mode after re-setting.


Figure 12.96 Error Occurrence in Normal Mode, Recovery in Normal Mode

1. After a reset, MTU output is low and ports are in the high-impedance state.
2. After a reset, the TMDR setting is for normal mode.
3. For channels 3 and 4, enable output with TOER before initializing the pins with TIOR.
4. Initialize the pins with TIOR. (The example shows initial high output, with low output on compare-match occurrence.)
5. Set MTU output with the PFC.
6. The count operation is started by TSTR.
7. Output goes low on compare-match occurrence.
8. An error occurs.
9. Set port output with the PFC and output the inverse of the active level.
10. The count operation is stopped by TSTR.
11. Not necessary when restarting in normal mode.
12. Initialize the pins with TIOR.
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(2) Operation when Error Occurs during Normal Mode Operation, and Operation is

Restarted in PWM Mode 1: Figure 12.97 shows an explanatory diagram of the case where an error occurs in normal mode and operation is restarted in PWM mode 1 after re-setting.


Figure 12.97 Error Occurrence in Normal Mode, Recovery in PWM Mode 1
1 to 10 are the same as in figure 12.96.
11. Set PWM mode 1.
12. Initialize the pins with TIOR. (In PWM mode 1 , the TIOC*B side is not initialized. If initialization is required, initialize in normal mode, then switch to PWM mode 1.)
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(3) Operation when Error Occurs during Normal Mode Operation, and Operation is

Restarted in PWM Mode 2: Figure 12.98 shows an explanatory diagram of the case where an error occurs in normal mode and operation is restarted in PWM mode 2 after re-setting.


Figure 12.98 Error Occurrence in Normal Mode, Recovery in PWM Mode 2
1 to 10 are the same as in figure 12.96 .
11. Set PWM mode 2.
12. Initialize the pins with TIOR. (In PWM mode 2, the cycle register pins are not initialized. If initialization is required, initialize in normal mode, then switch to PWM mode 2.)
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.

Note: PWM mode 2 can only be set for channels $0-2$, and therefore TOER setting is not necessary.
(4) Operation when Error Occurs during Normal Mode Operation, and Operation is Restarted in Phase Counting Mode: Figure 12.99 shows an explanatory diagram of the case where an error occurs in normal mode and operation is restarted in phase counting mode after resetting.


Figure 12.99 Error Occurrence in Normal Mode, Recovery in Phase Counting Mode
1 to 10 are the same as in figure 12.96 .
11. Set phase counting mode.
12. Initialize the pins with TIOR.
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.

Note: Phase counting mode can only be set for channels 1 and 2, and therefore TOER setting is not necessary.
(5) Operation when Error Occurs during Normal Mode Operation, and Operation is Restarted in Complementary PWM Mode: Figure 12.100 shows an explanatory diagram of the case where an error occurs in normal mode and operation is restarted in complementary PWM mode after re-setting.


Figure 12.100 Error Occurrence in Normal Mode, Recovery in Complementary PWM Mode

1 to 10 are the same as in figure 12.96 .
11. Initialize the normal mode waveform generation section with TIOR.
12. Disable operation of the normal mode waveform generation section with TIOR.
13. Disable channel 3 and 4 output with TOER.
14. Select the complementary PWM output level and cyclic output enabling/disabling with TOCR.
15. Set complementary PWM.
16. Enable channel 3 and 4 output with TOER.
17. Set MTU output with the PFC.
18. Operation is restarted by TSTR.
(6) Operation when Error Occurs during Normal Mode Operation, and Operation is Restarted in Reset-Synchronous PWM Mode: Figure 12.101 shows an explanatory diagram of the case where an error occurs in normal mode and operation is restarted in reset-synchronous PWM mode after re-setting.


Figure 12.101 Error Occurrence in Normal Mode, Recovery in Reset-Synchronous PWM Mode

1 to 13 are the same as in figure 12.100 .
14. Select the reset-synchronous PWM output level and cyclic output enabling/disabling with TOCR.
15. Set reset-synchronous PWM.
16. Enable channel 3 and 4 output with TOER.
17. Set MTU output with the PFC.
18. Operation is restarted by TSTR.
(7) Operation when Error Occurs during PWM Mode 1 Operation, and Operation is

Restarted in Normal Mode: Figure 12.102 shows an explanatory diagram of the case where an error occurs in PWM mode 1 and operation is restarted in normal mode after re-setting.


Figure 12.102 Error Occurrence in PWM Mode 1, Recovery in Normal Mode

1. After a reset, MTU output is low and ports are in the high-impedance state.
2. Set PWM mode 1.
3. For channels 3 and 4 , enable output with TOER before initializing the pins with TIOR.
4. Initialize the pins with TIOR. (The example shows initial high output, with low output on compare-match occurrence. In PWM mode 1, the TIOC*B side is not initialized.)
5. Set MTU output with the PFC.
6. The count operation is started by TSTR.
7. Output goes low on compare-match occurrence.
8. An error occurs.
9. Set port output with the PFC and output the inverse of the active level.
10. The count operation is stopped by TSTR.
11. Set normal mode.
12. Initialize the pins with TIOR.
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(8) Operation when Error Occurs during PWM Mode 1 Operation, and Operation is

Restarted in PWM Mode 1: Figure 12.103 shows an explanatory diagram of the case where an error occurs in PWM mode 1 and operation is restarted in PWM mode 1 after re-setting.


Figure 12.103 Error Occurrence in PWM Mode 1, Recovery in PWM Mode 1
1 to 10 are the same as in figure 12.102.
11. Not necessary when restarting in PWM mode 1.
12. Initialize the pins with TIOR. (In PWM mode 1, the TIOC*B side is not initialized.)
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(9) Operation when Rrror Occurs during PWM Mode 1 Operation, and Operation is

Restarted in PWM Mode 2: Figure 12.104 shows an explanatory diagram of the case where an error occurs in PWM mode 1 and operation is restarted in PWM mode 2 after re-setting.


Figure 12.104 Error Occurrence in PWM Mode 1, Recovery in PWM Mode 2
1 to 10 are the same as in figure 12.102 .
11. Set PWM mode 2.
12. Initialize the pins with TIOR. (In PWM mode 2, the cycle register pins are not initialized.)
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.

Note: PWM mode 2 can only be set for channels $0-2$, and therefore TOER setting is not necessary.
(10) Operation when Error Occurs during PWM Mode 1 Operation, and Operation is Restarted in Phase Counting Mode: Figure 12.105 shows an explanatory diagram of the case where an error occurs in PWM mode 1 and operation is restarted in phase counting mode after resetting.


Figure 12.105 Error Occurrence in PWM Mode 1, Recovery in Phase Counting Mode
1 to 10 are the same as in figure 12.102 .
11. Set phase counting mode.
12. Initialize the pins with TIOR.
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.

Note: Phase counting mode can only be set for channels 1 and 2, and therefore TOER setting is not necessary.

## (11) Operation when Error Occurs during PWM Mode 1 Operation, and Operation is

 Restarted in Complementary PWM Mode: Figure 12.106 shows an explanatory diagram of the case where an error occurs in PWM mode 1 and operation is restarted in complementary PWM mode after re-setting.

Figure 12.106 Error Occurrence in PWM Mode 1, Recovery in Complementary PWM Mode

1 to 10 are the same as in figure 12.102 .
11. Set normal mode for initialization of the normal mode waveform generation section.
12. Initialize the PWM mode 1 waveform generation section with TIOR.
13. Disable operation of the PWM mode 1 waveform generation section with TIOR.
14. Disable channel 3 and 4 output with TOER.
15. Select the complementary PWM output level and cyclic output enabling/disabling with TOCR.
16. Set complementary PWM.
17. Enable channel 3 and 4 output with TOER.
18. Set MTU output with the PFC.
19. Operation is restarted by TSTR.
(12) Operation when Error Occurs during PWM Mode 1 Operation, and Operation is Restarted in Reset-Synchronous PWM Mode: Figure 12.107 shows an explanatory diagram of the case where an error occurs in PWM mode 1 and operation is restarted in reset-synchronous PWM mode after re-setting.


Figure 12.107 Error Occurrence in PWM Mode 1, Recovery in Reset-Synchronous PWM Mode

1 to 14 are the same as in figure 12.106 .
15. Select the reset-synchronous PWM output level and cyclic output enabling/disabling with TOCR.
16. Set reset-synchronous PWM.
17. Enable channel 3 and 4 output with TOER.
18. Set MTU output with the PFC.
19. Operation is restarted by TSTR.
(13) Operation when Error Occurs during PWM Mode 2 Operation, and Operation is

Restarted in Normal Mode: Figure 12.108 shows an explanatory diagram of the case where an error occurs in PWM mode 2 and operation is restarted in normal mode after re-setting.


Figure 12.108 Error Occurrence in PWM Mode 2, Recovery in Normal Mode

1. After a reset, MTU output is low and ports are in the high-impedance state.
2. Set PWM mode 2.
3. Initialize the pins with TIOR. (The example shows initial high output, with low output on compare-match occurrence. In PWM mode 2, the cycle register pins are not initialized. In the example, TIOC $*$ A is the cycle register.)
4. Set MTU output with the PFC.
5. The count operation is started by TSTR.
6. Output goes low on compare-match occurrence.
7. An error occurs.
8. Set port output with the PFC and output the inverse of the active level.
9. The count operation is stopped by TSTR.
10. Set normal mode.
11. Initialize the pins with TIOR.
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(14) Operation when Error Occurs during PWM Mode 2 Operation, and Operation is

Restarted in PWM Mode 1: Figure 12.109 shows an explanatory diagram of the case where an error occurs in PWM mode 2 and operation is restarted in PWM mode 1 after re-setting.


Figure 12.109 Error Occurrence in PWM Mode 2, Recovery in PWM Mode 1
1 to 9 are the same as in figure 12.108.
10. Set PWM mode 1.
11. Initialize the pins with TIOR. (In PWM mode 1, the TIOC*B side is not initialized.)
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(15) Operation when Error Occurs during PWM Mode 2 Operation, and Operation is

Restarted in PWM Mode 2: Figure 12.110 shows an explanatory diagram of the case where an error occurs in PWM mode 2 and operation is restarted in PWM mode 2 after re-setting.


Figure 12.110 Error Occurrence in PWM Mode 2, Recovery in PWM Mode 2
1 to 9 are the same as in figure 12.108.
10. Not necessary when restarting in PWM mode 2.
11. Initialize the pins with TIOR. (In PWM mode 2, the cycle register pins are not initialized.)
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(16) Operation when Error Occurs during PWM Mode 2 Operation, and Operation is Restarted in Phase Counting Mode: Figure 12.111 shows an explanatory diagram of the case where an error occurs in PWM mode 2 and operation is restarted in phase counting mode after resetting.


Figure 12.111 Error Occurrence in PWM Mode 2, Recovery in Phase Counting Mode
1 to 9 are the same as in figure 12.108 .
10. Set phase counting mode.
11. Initialize the pins with TIOR.
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(17) Operation when Error Occurs during Phase Counting Mode Operation, and Operation is Restarted in Normal Mode: Figure 12.112 shows an explanatory diagram of the case where an error occurs in phase counting mode and operation is restarted in normal mode after re-setting.


Figure 12.112 Error Occurrence in Phase Counting Mode, Recovery in Normal Mode

1. After a reset, MTU output is low and ports are in the high-impedance state.
2. Set phase counting mode.
3. Initialize the pins with TIOR. (The example shows initial high output, with low output on compare-match occurrence.)
4. Set MTU output with the PFC.
5. The count operation is started by TSTR.
6. Output goes low on compare-match occurrence.
7. An error occurs.
8. Set port output with the PFC and output the inverse of the active level.
9. The count operation is stopped by TSTR.
10. Set in normal mode.
11. Initialize the pins with TIOR.
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(18) Operation when Error Occurs during Phase Counting Mode Operation, and Operation is Restarted in PWM Mode 1: Figure 12.113 shows an explanatory diagram of the case where an error occurs in phase counting mode and operation is restarted in PWM mode 1 after re-setting.


Figure 12.113 Error Occurrence in Phase Counting Mode, Recovery in PWM Mode 1
1 to 9 are the same as in figure 12.112.
10. Set PWM mode 1.
11. Initialize the pins with TIOR. (In PWM mode 1, the TIOC $* \mathrm{~B}$ side is not initialized.)
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(19) Operation when Error Occurs during Phase Counting Mode Operation, and Operation is Restarted in PWM Mode 2: Figure 12.114 shows an explanatory diagram of the case where an error occurs in phase counting mode and operation is restarted in PWM mode 2 after re-setting.


Figure 12.114 Error Occurrence in Phase Counting Mode, Recovery in PWM Mode 2
1 to 9 are the same as in figure 12.112.
10. Set PWM mode 2.
11. Initialize the pins with TIOR. (In PWM mode 2, the cycle register pins are not initialized.)
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(20) Operation when Error Occurs during Phase Counting Mode Operation, and Operation is Restarted in Phase Counting Mode: Figure 12.115 shows an explanatory diagram of the case where an error occurs in phase counting mode and operation is restarted in phase counting mode after re-setting.


Figure 12.115 Error Occurrence in Phase Counting Mode, Recovery in Phase Counting Mode

1 to 9 are the same as in figure 12.112.
10. Not necessary when restarting in phase counting mode.
11. Initialize the pins with TIOR.
12. Set MTU output with the PFC.
13. Operation is restarted by TSTR.
(21) Operation when Error Occurs during Complementary PWM Mode Operation, and Operation is Restarted in Normal Mode: Figure 12.116 shows an explanatory diagram of the case where an error occurs in complementary PWM mode and operation is restarted in normal mode after re-setting.


Figure 12.116 Error Occurrence in Complementary PWM Mode, Recovery in Normal Mode

1. After a reset, MTU output is low and ports are in the high-impedance state.
2. Select the complementary PWM output level and cyclic output enabling/disabling with TOCR.
3. Set complementary PWM.
4. Enable channel 3 and 4 output with TOER.
5. Set MTU output with the PFC.
6. The count operation is started by TSTR.
7. The complementary PWM waveform is output on compare-match occurrence.
8. An error occurs.
9. Set port output with the PFC and output the inverse of the active level.
10. The count operation is stopped by TSTR. (MTU output becomes the complementary PWM output initial value.)
11. Set normal mode. (MTU output goes low.)
12. Initialize the pins with TIOR.
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(22) Operation when Error Occurs during Complementary PWM Mode Operation, and Operation is Restarted in PWM Mode 1: Figure 12.117 shows an explanatory diagram of the case where an error occurs in complementary PWM mode and operation is restarted in PWM mode 1 after re-setting.


Figure 12.117 Error Occurrence in Complementary PWM Mode, Recovery in PWM Mode 1

1 to 10 are the same as in figure 12.116 .
11. Set PWM mode 1. (MTU output goes low.)
12. Initialize the pins with TIOR. (In PWM mode 1, the TIOC $* \mathrm{~B}$ side is not initialized.)
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(23a) Operation when Error Occurs during Complementary PWM Mode Operation, and Operation is Restarted in Complementary PWM Mode: Figure 12.118 shows an explanatory diagram of the case where an error occurs in complementary PWM mode and operation is restarted in complementary PWM mode after re-setting (when operation is restarted using the cycle and duty settings at the time the counter was stopped).


Figure 12.118 Error Occurrence in Complementary PWM Mode, Recovery in Complementary PWM Mode

1 to 10 are the same as in figure 12.116 .
11. Set MTU output with the PFC.
12. Operation is restarted by TSTR.
13. The complementary PWM waveform is output on compare-match occurrence.
(23b) Operation when Error Occurs during Complementary PWM Mode Operation, and Operation is Restarted in Complementary PWM Mode: Figure 12.119 shows an explanatory diagram of the case where an error occurs in complementary PWM mode and operation is restarted in complementary PWM mode after re-setting (when operation is restarted using completely new cycle and duty settings).


Figure 12.119 Error Occurrence in Complementary PWM Mode, Recovery in Complementary PWM Mode

1 to 10 are the same as in figure 12.116 .
11. Set normal mode and make new settings. (MTU output goes low.)
12. Disable channel 3 and 4 output with TOER.
13. Select the complementary PWM mode output level and cyclic output enabling/disabling with TOCR.
14. Set complementary PWM.
15. Enable channel 3 and 4 output with TOER.
16. Set MTU output with the PFC.
17. Operation is restarted by TSTR.
(24) Operation when Error Occurs during Complementary PWM Mode Operation, and Operation is Restarted in Reset-Synchronous PWM Mode: Figure 12.120 shows an explanatory diagram of the case where an error occurs in complementary PWM mode and operation is restarted in reset-synchronous PWM mode.


Figure 12.120 Error Occurrence in Complementary PWM Mode, Recovery in Reset-Synchronous PWM Mode

1 to 10 are the same as in figure 12.116 .
11. Set normal mode. (MTU output goes low.)
12. Disable channel 3 and 4 output with TOER.
13. Select the reset-synchronous PWM mode output level and cyclic output enabling/disabling with TOCR.
14. Set reset-synchronous PWM.
15. Enable channel 3 and 4 output with TOER.
16. Set MTU output with the PFC.
17. Operation is restarted by TSTR.
(25) Operation when Error Occurs during Reset-Synchronous PWM Mode Operation, and Operation is Restarted in Normal Mode: Figure 12.121 shows an explanatory diagram of the case where an error occurs in reset-synchronous PWM mode and operation is restarted in normal mode after re-setting.


Figure 12.121 Error Occurrence in Reset-Synchronous PWM Mode, Recovery in Normal Mode

1. After a reset, MTU output is low and ports are in the high-impedance state.
2. Select the reset-synchronous PWM output level and cyclic output enabling/disabling with TOCR.
3. Set reset-synchronous PWM.
4. Enable channel 3 and 4 output with TOER.
5. Set MTU output with the PFC.
6. The count operation is started by TSTR.
7. The reset-synchronous PWM waveform is output on compare-match occurrence.
8. An error occurs.
9. Set port output with the PFC and output the inverse of the active level.
10. The count operation is stopped by TSTR. (MTU output becomes the reset-synchronous PWM output initial value.)
11. Set normal operating mode. (MTU positive phase output is low, and negative phase output is high.)
12. Initialize the pins with TIOR.
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(26) Operation when Error Occurs during Reset-Synchronous PWM Mode Operation, and Operation is Restarted in PWM Mode 1: Figure 12.122 shows an explanatory diagram of the case where an error occurs in reset-synchronous PWM mode and operation is restarted in PWM mode 1 after re-setting.


Figure 12.122 Error Occurrence in Reset-Synchronous PWM Mode, Recovery in PWM Mode 1

1 to 10 are the same as in figure 12.121 .
11. Set PWM mode 1. (MTU positive phase output is low, and negative phase output is high.)
12. Initialize the pins with TIOR. (In PWM mode 1, the TIOC $* \mathrm{~B}$ side is not initialized.)
13. Set MTU output with the PFC.
14. Operation is restarted by TSTR.
(27) Operation when Error Occurs during Reset-Synchronous PWM Mode Operation, and Operation is Restarted in Complementary PWM Mode: Figure 12.123 shows an explanatory diagram of the case where an error occurs in reset-synchronous PWM mode and operation is restarted in complementary PWM mode after re-setting.


Figure 12.123 Error Occurrence in Reset-Synchronous PWM Mode,
Recovery in Complementary PWM Mode
1 to 10 are the same as in figure 12.121 .
11. Disable channel 3 and 4 output with TOER.
12. Select the complementary PWM output level and cyclic output enabling/disabling with TOCR.
13. Set complementary PWM. (The MTU cyclic output pin goes low.)
14. Enable channel 3 and 4 output with TOER.
15. Set MTU output with the PFC.
16. Operation is restarted by TSTR.
(28) Operation when Error Occurs during Reset-Synchronous PWM Mode Operation, and Operation is Restarted in Reset-Synchronous PWM Mode: Figure 12.124 shows an explanatory diagram of the case where an error occurs in reset-synchronous PWM mode and operation is restarted in reset-synchronous PWM mode after re-setting.


Figure 12.124 Error Occurrence in Reset-Synchronous PWM Mode, Recovery in Reset-Synchronous PWM Mode

1 to 10 are the same as in figure 12.121 .
11. Set MTU output with the PFC.
12. Operation is restarted by TSTR.
13. The reset-synchronous PWM waveform is output on compare-match occurrence.

### 12.9 Port Output Enable (POE)

The port output enable (POE) can be used to establish a high-impedance state for high-current pins, by changing the $\overline{\mathrm{POE}}-\overline{\mathrm{POE} 3}$ pin input, depending on the output status of the high-current pins (PE09/TIOC3B, PE11/TIOC3D, PE12/TIOC4A, PE13/TIOC4B/MRES, PE14/TIOC4C/DACK0/AH, PE15/TIOC4D/DACK1/(IRQOUT). It can also simultaneously generate interrupt requests.

The high-current pins also become high-impedance regardless of whether these pin functions are selected in cases such as when the oscillator stops or in standby mode. Refer to section 4, Clock Pulse Generator (CPG), for details.

### 12.9.1 Features

- Each of the $\overline{\text { POE0 }}-\overline{\text { POE3 }}$ input pins can be set for falling edge, $\phi / 8 \times 16, \phi / 16 \times 16$, or $\phi / 128 \times$ 16 low-level sampling.
- High-current pins can be set to high-impedance state by $\overline{\mathrm{POE} 0}-\overline{\mathrm{POE} 3}$ pin falling-edge or lowlevel sampling.
- High-current pins can be set to high-impedance state when the high-current pin output levels are compared and simultaneous low-level output continues for one cycle or more (except in the 33.3 MHz version).
- Interrupts can be generated by input-level sampling or output-level comparison results.


### 12.9.2 Block Diagram

The POE has input-level detection circuitry and output-level detection circuitry, as shown in the block diagram of figure 12.125 .


Figure 12.125 POE Block Diagram

### 12.9.3 Pin Configuration

Table 12.18 shows the POE pins.
Table 12.18 Pin Configuration

| Name | Abbreviation | I/O | Description |
| :--- | :--- | :--- | :--- |
| Port output enable input pins | $\overline{\text { POE0 }}-\overline{\text { POE3 }}$ | Input | Input request signals to make high- <br> current pins high-impedance state |

Table 12.19 shows output-level comparisons with pin combinations.
Table 12.19 Output Level Comparisons

| Pin Combination | I/O | Description |
| :--- | :--- | :--- |
| PE09/TIOC3B and <br> PE11/TIOC3D | Output | All high-current pins are made high-impedance <br> state when the pins simultaneously output low-level <br> for longer than 1 cycle. |
| PE12/TIOC4A and | Output | All high-current pins are made high-impedance <br> state when the pins simultaneously output low-level <br> for longer than 1 cycle. |
| PE14/TIOC4C/DACK0/AH | Output | All high-current pins are made high-impedance <br> state when the pins simultaneously output low-level <br> for longer than 1 cycle. |

### 12.9.4 Register Configuration

The POE has the two registers shown in table 12.20. The input level control/status register (ICSR) controls both $\overline{\text { POE } 0}-\overline{\mathrm{POE} 3}$ pin input signal detection and interrupts. The output level control/status register (OCSR) controls both the enable/disable of output comparison and interrupts.

Table 12.20 Input Level Control/Status Register Configuration

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Input level control/status register | ICSR | R/(W) ${ }^{* 1}$ | H'0000 | H'FFFF83C0 H'FFFF83C1 | 8, 16, 32 |
| Output level control/status register | OCSR | $\mathrm{R} /(\mathrm{W})^{* 2}$ | H'0000 | H'FFFF83C2 H'FFFF83C3 | 8,16, 32 |

Notes: *1 Only 0 writes to bits $15-12$ are possible to clear the flags.
*2 Only 0 writes to bits 15 are possible to clear the flags.

### 12.10 POE Register Descriptions

### 12.10.1 Input Level Control/Status Register (ICSR)

The input level control/status register (ICSR) is a 16-bit read/write register that selects the $\overline{\text { POE0 }}-$ $\overline{\text { POE3 }}$ pin input modes, controls the enable/prohibit of interrupts, and indicates status. If any of the POE3F-POE0F bits are set to 1 , the high current pins become high impedance state.

ICSR is initialized to $\mathrm{H}^{\prime} 0000$ by power-on resets; however, it is not initialized for manual resets, standby mode, or sleep mode, so the previous data is maintained.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | POE3F | POE2F | POE1F | POE0F | - | - | - | PIE |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | $\mathrm{R} /(\mathrm{W})^{*}$ | R/(W)* | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | R | R | R | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | POE3M1 | POE3M0 | POE2M1 | POE2M0 | POE1M1 | POE1M0 | POE0M1 | POEOM0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Note: * Only 0 writes are possible to clear the flags.

- Bit 15-POE3 Flag (POE3F): This flag indicates that a high impedance request has been input to the $\overline{\mathrm{POE} 3} \mathrm{pin}$.

| Bit 15: POE3F | Description |
| :--- | :--- |
| 0 | Clear condition: By writing 0 to POE3F after reading a POE3F $=1$ <br> (initial value) |
| 1 | Set condition: When the input set by ICSR bits 7 and 6 occurs at <br> the $\overline{\text { POE3 }}$ pin |

- Bit 14—POE2 Flag (POE2F): This flag indicates that a high impedance request has been input to the $\overline{\mathrm{POE} 2} \mathrm{pin}$.

Bit 14: POE2F Description

| 0 | Clear condition: By writing 0 to POE2F after reading a POE2F $=1$ <br> (initial value) |
| :--- | :--- |
| 1 | Set condition: When the input set by ICSR bits 5 and 4 occurs at <br> the $\overline{\text { POE2 }}$ pin |

- Bit 13—POE1 Flag (POE1F): This flag indicates that a high impedance request has been input to the $\overline{\mathrm{POE} 1}$ pin.

| Bit 13: POE1F | Description |
| :--- | :--- |
| 0 | Clear condition: By writing 0 to POE1F after reading a POE1F $=1$ <br> (initial value) |
| 1 | Set condition: When the input set by ICSR bits 3 and 2 occurs at <br> the $\overline{\text { POE1 }}$ pin |

- Bit 12—POE0 Flag (POE0F): This flag indicates that a high impedance request has been input to the $\overline{\mathrm{POE}} \mathrm{pin}$.

Bit 12: POEOF Description

| 0 | Clear condition: By writing 0 to POEOF after reading a POEOF $=1$ <br> (initial value) |
| :--- | :--- |
| 1 | Set condition: When the input set by ICSR bits 1 and 0 occurs at <br> the $\overline{\text { POEO }}$ pin |

- Bits 11-9—Reserved: These bits always read as 0 . The write value should always be 0 .
- Bit 8—Port Interrupt Enable (PIE): Enables or disables interrupt requests when any of the POE 0 F-POE3F bits of the ICSR are set to 1 .

Bit 8: PIE

## Description

| 0 | Interrupt requests disabled (initial value) |
| :--- | :--- |
| 1 | Interrupt requests enabled |

- Bits 7 and 6-POE3 Mode 1, 0 (POE3M1 and POE3M0): These bits select the input mode of the $\overline{\mathrm{POE} 3}$ pin.
\(\left.$$
\begin{array}{lll}\begin{array}{l}\text { Bit 7: } \\
\text { POE3M1 }\end{array}
$$ \& \begin{array}{l}Bit 6: <br>

POE3M0\end{array} \& Description\end{array}\right]\)| 0 | 0 | Accept request on falling edge of $\overline{\text { POE3 input. (initial value) }}$ |
| :--- | :--- | :--- |
| 1 | Accept request when $\overline{\text { POE3 }}$ input has been sampled for 16 <br> $\phi / 8$ clock pulses, and all are low level. |  |
| 1 | 0 | Accept request when $\overline{\text { POE3 }}$ input has been sampled for 16 <br> $\phi / 16$ clock pulses, and all are low level. |
|  | 1 | Accept request when $\overline{\text { POE3 }}$ input has been sampled for 16 <br> $\phi / 128 ~ c l o c k ~ p u l s e s, ~ a n d ~ a l l ~ a r e ~ l o w ~ l e v e l . ~$ |

- Bits 5 and 4—POE2 Mode 1, 0 (POE2M1 and POE2M0): These bits select the input mode of the $\overline{\mathrm{POE} 2} \mathrm{pin}$.

| Bit 5: POE2M1 | Bit 4: <br> POE2M0 | Description |
| :---: | :---: | :---: |
| 0 | 0 | Accept request on falling edge of $\overline{\text { POE2 }}$ input. (initial value) |
|  | 1 | Accept request when $\overline{\mathrm{POE} 2}$ input has been sampled for 16 $\phi / 8$ clock pulses, and all are low level. |
| 1 | 0 | Accept request when $\overline{\mathrm{POE} 2}$ input has been sampled for 16 $\phi / 16$ clock pulses, and all are low level. |
|  | 1 | Accept request when $\overline{\text { POE2 }}$ input has been sampled for 16 $\phi / 128$ clock pulses, and all are low level. |

- Bits 3 and 2—POE1 Mode 1, 0 (POE1M1 and POE1M0): These bits select the input mode of the $\overline{\text { POE1 }}$ pin.

| Bit 3: <br> POE1M1 | Bit 2: <br> POE1M0 | Description |
| :--- | :--- | :--- | | 0 | 0 | Accept request on falling edge of $\overline{\text { POE1 input. (initial value) }}$ |
| :--- | :--- | :--- |
|  | 1 | Accept request when $\overline{\text { POE }}$ input has been sampled for 16 <br> $\phi / 8$ clock pulses, and all are low level. |
| 1 | 0 | Accept request when $\overline{\text { POE } 1}$ input has been sampled for 16 <br> $\phi / 16$ clock pulses, and all are low level. |
|  | 1 | Accept request when $\overline{\text { POE } 1}$ input has been sampled for 16 <br> $\phi / 128$ clock pulses, and all are low level. |

- Bits 1 and 0—POE0 Mode 1, 0 (POE0M1 and POE0M0): These bits select the input mode of the $\overline{\text { POE } 0}$ pin.

| Bit 1: POE0M1 | Bit 0: POEOMO | Description |
| :---: | :---: | :---: |
| 0 | 0 | Accept request on falling edge of $\overline{\text { POEO }}$ input. (initial value) |
|  | 1 | Accept request when $\overline{\text { POEO }}$ input has been sampled for 16 $\phi / 8$ clock pulses, and all are low level. |
| 1 | 0 | Accept request when $\overline{\text { POEO }}$ input has been sampled for 16 $\phi / 16$ clock pulses, and all are low level. |
|  | 1 | Accept request when $\overline{\mathrm{POEO}}$ input has been sampled for 16 $\phi / 128$ clock pulses, and all are low level. |

### 12.10.2 Output Level Control/Status Register (OCSR)

The output level control/status register (OCSR) is a 16-bit read/write register that controls the enable/disable of both output level comparison and interrupts, and indicates status. If the OSF bit is set to 1 , the high current pins become high impedance.

OCSR is initialized to $\mathrm{H}^{\prime} 0000$ by an external power-on reset; however, it is not initialized for manual resets, reset by WDT standby mode, or sleep mode, so the previous data is maintained.


Note: * Only 0 writes are possible to clear the flag.

- Bit 15-Output Short Flag (OSF): This flag indicates that among the three pairs of 2 phase outputs compared, the outputs of at least one pair have simultaneously become Low level output.

Bit 15: OSF Description

| 0 | Clear condition: By writing 0 to OSF after reading an OSF = 1 (initial value) |
| :--- | :--- |
| 1 | Set condition: When any one pair of the 2-phase outputs simultaneously <br> become Low level |

- Bits 14-10—Reserved: These bits always read as 0 . The write value should always be 0 .
- Bit 9—Output Level Compare Enable (OCE): This bit enables the start of output level comparisons. When setting this bit, pay special attention to the output pin combinations shown in table 12.19. When 0 is output, the OSF bit is set to 1 at the same time this bit is set, and output goes to high impedance. Accordingly, bits 15-11 and bit 9 of the port $E$ data register (PEDR) are set to 1 . For the MTU output comparison, set the bit to 1 after setting the MTU's output pins with the PFC. Set this bit only when using pins as outputs.
When the OCE bit is set to 1 , if OIE $=0$ a high-impedance request will not be issued even if OSF is set to 1 . Therefore, in order to have a high-impedance request issued according to the result of the output comparison, the OIE bit must be set to 1 . When $\mathrm{OCE}=1$ and $\mathrm{OIE}=1$, an interrupt request will be generated at the same time as the high-impedance request; however, this interrupt can be masked by means of an interrupt controller (INTC) setting.

Bit 9: OCE Description

| 0 | Output level compare disabled (initial value) |
| :--- | :--- |
| 1 | Output level compare enabled; makes an output high impedance request <br> when $O S F=1$. |

- Bit 8—Output Short Interrupt Enable (OIE): Makes interrupt requests when the OSF bit of the OCSR is set.
Bit 8: OIE Description

| 0 | Interrupt requests disabled (initial value) |
| :--- | :--- |
| 1 | Interrupt requests enabled |

- Bits 7-0—Reserved: Always read as 0 , and cannot be modified.


### 12.11 Operation

### 12.11.1 Input Level Detection Operation

If the input conditions set by the ICSR occur on any of the $\overline{\text { POE }}$ pins, all high-current pins become high-impedance state.

Falling Edge Detection: When a change from high to low level is input to the $\overline{\mathrm{POE}}$ pins.
Low-Level Detection: Figure 12.126 shows the low-level detection operation. Sixteen continuous low levels are sampled with the sampling clock established by the ICSR. If even one high level is detected during this interval, the low level is not accepted.


Figure 12.126 Low-Level Detection Operation

### 12.11.2 Output-Level Compare Operation

Figure 12.127 shows an example of the output-level compare operation for the combination of PE09/TIOC3B and PE11/TIOC3D. The operation is the same for the other pin combinations.


Figure 12.127 Output-Level Detection Operation

### 12.11.3 Release from High-Impedance State

High-current pins that have entered high-impedance state due to input-level detection can be released either by returning them to their initial state with a power-on reset, or by clearing all of the bit 12-15 (POE0F-POE3F) flags of the ICSR. High-current pins that have become highimpedance due to output-level detection can be released either by returning them to their initial state with a power-on reset, or by first clearing bit 9 (OCE) of the OCSR to disable output-level compares, then clearing the bit 15 (OSF) flag. However, when returning from high-impedance state by clearing the OSF flag, always do so only after outputting a high level from the highcurrent pins (TIOC3B, TIOC3D, TIOC4A, TIOC4B, TIOC4C, and TIOC4D). High-level outputs can be achieved by setting the MTU internal registers. See section 12.2, MTU Register Descriptions, for details.

### 12.11.4 POE timing

Figure 12.128 shows an example of timing from $\overline{\text { POE }}$ input to high impedance of pin.


Figure 12.128 Last Transition Edge Detection Operation

### 12.11.5 Usage Notes

To perform POE level detection, first set POE input to high level.

## Section 13 Watchdog Timer (WDT)

### 13.1 Overview

The watchdog timer (WDT) is a 1-channel timer for monitoring system operations. If a system encounters a problem (crashes, for example) and the timer counter overflows without being rewritten correctly by the CPU, an overflow signal ( $\overline{\mathrm{WDTOVF}}$ ) is output externally. The WDT can simultaneously generate an internal reset signal for the entire chip.

When the watchdog function is not needed, the WDT can be used as an interval timer. In the interval timer operation, an interval timer interrupt is generated at each counter overflow. The WDT is also used in recovering from the standby mode.

### 13.1.1 Features

- Works in watchdog timer mode or interval timer mode.
- Outputs WDTOVF in the watchdog timer mode. When the counter overflows in the watchdog timer mode, overflow signal $\overline{\text { WDTOVF }}$ is output externally. You can select whether to reset the chip internally when this happens. Either the power-on reset or manual reset signal can be selected as the internal reset signal.
- Generates interrupts in the interval timer mode. When the counter overflows, it generates an interval timer interrupt.
- Clears standby mode.
- Works with eight counter input clocks.


### 13.1.2 Block Diagram

Figure 13.1 is the block diagram of the WDT.


Figure 13.1 WDT Block Diagram

### 13.1.3 Pin Configuration

Table 13.1 shows the pin configuration.
Table 13.1 Pin Configuration

| Pin | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Watchdog timer overflow | $\overline{\text { WDTOVF }}$ | O | Outputs the counter overflow signal in the <br> watchdog timer mode |

### 13.1.4 Register Configuration

Table 13.2 summarizes the three WDT registers. They are used to select the clock, switch the WDT mode, and control the reset signal.

Table 13.2 WDT Registers

| Name | Abbreviation | R/W | Initial Value | Address |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Write*1 | Read*2 |
| Timer control/status register | TCSR | $\mathrm{R} /(\mathrm{W})^{* 3}$ | H'18 | H'FFFF8610 | H'FFFF8610 |
| Timer counter | TCNT | R/W | H'00 |  | H'FFFF8611 |
| Reset control/status register | RSTCSR | $\mathrm{R} /(\mathrm{W})^{* 3}$ | H'1F | H'FFFFF8612 | H'FFFF8613 |

Notes: *1 Write by word transfer. It cannot be written in byte or longword.
*2 Read by byte transfer. It cannot be read in word or longword.
*3 Only 0 can be written in bit 7 to clear the flag.

### 13.2 Register Descriptions

### 13.2.1 Timer Counter (TCNT)

The TCNT is an 8-bit read/write upcounter. (The TCNT differs from other registers in that it is more difficult to write to. See section 13.2.4, Register Access, for details.) When the timer enable bit (TME) in the timer control/status register (TCSR) is set to 1 , the watchdog timer counter starts counting pulses of an internal clock selected by clock select bits 2-0 (CKS2-CKS0) in the TCSR. When the value of the TCNT overflows (changes from H'FF to H'00), a watchdog timer overflow signal ( $\overline{\mathrm{WDTOVF}}$ ) or interval timer interrupt (ITI) is generated, depending on the mode selected in the WT/ITT bit of the TCSR.

The TCNT is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset and when the TME bit is cleared to 0 . It is not initialized in the standby mode. The TCNT is not initialized by a manual reset from an external source ( $\overline{\mathrm{MRES}})$, but is initialized by a manual reset from the WDT.


### 13.2.2 Timer Control/Status Register (TCSR)

The timer control/status register (TCSR) is an 8-bit read/write register. (The TCSR differs from other registers in that it is more difficult to write to. See section 13.2.4, Register Access, for details.) Its functions include selecting the timer mode and clock source.

Bits $7-5$ are initialized to 000 by a power-on reset or in standby mode. Bits $2-0$ are initialized to 000 by a power-on reset, but retain their values in the standby mode. These bits are not initialized by a manual reset from an external source ( $\overline{\mathrm{MRES}}$ ), but are initialized by a manual reset from the WDT.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | OVF | WT/IT | TME | - | - | CKS2 | CKS1 | CKS0 |
| Initial value: | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 0 |
| R/W: | $\mathrm{R} /(\mathrm{W})$ | R/W | R/W | R | R | R/W | R/W | R/W |

- Bit 7—Overflow Flag (OVF): Indicates that the TCNT has overflowed from H'FF to H'00 in the interval timer mode. It is not set in the watchdog timer mode.


## Bit 7: OVF Description

| 0 | No overflow of TCNT in interval timer mode (initial value) <br> Cleared by reading OVF, then writing 0 in OVF |
| :--- | :--- |
| 1 | TCNT overflow in the interval timer mode |

- Bit 6—Timer Mode Select (WT/ITT): Selects whether to use the WDT as a watchdog timer or interval timer. When the TCNT overflows, the WDT either generates an interval timer interrupt (ITI) or generates a $\overline{\text { WDTOVF }}$ signal, depending on the mode selected.

| Bit 6: WT//TT | Description |
| :--- | :--- |
| 0 | Interval timer mode: interval timer interrupt request to the CPU when <br> TCNT overflows (initial value) |
| 1 | Watchdog timer mode: $\overline{\text { WDTOVF signal output externally when TCNT }}$ <br> overflows. (Section 13.2.3, Reset Control/Status Register (RSTCSR), <br> describes in detail what happens when TCNT overflows in the watchdog <br> timer mode.) |

- Bit 5-Timer Enable (TME): Enables or disables the timer.

| Bit 5: TME | Description |
| :--- | :--- |
| 0 | Timer disabled: TCNT is initialized to H'00 and count-up stops (initial <br> value) |
| 1 | Timer enabled: TCNT starts counting. A WDTOVF signal or interrupt is <br> generated when TCNT overflows. |

- Bits 4 and 3—Reserved: These bits always read as 1 . The write value should always be 1 .
- Bits 2-0: Clock Select 2-0 (CKS2-CKS0): These bits select one of eight internal clock sources for input to the TCNT. The clock signals are obtained by dividing the frequency of the system clock ( $\phi$ ).

|  |  | Description |  |  |
| :--- | :--- | :--- | :--- | :--- |
| Bit 2: CKS2 | Bit 1: CKS1 | Bit 0: CKSO | Clock Source | Overflow Interval* <br> ( $\phi=\mathbf{2 8 . 7 ~ M H z )}$ |
| 0 | 0 | 0 | $\phi / 2$ (initial value) | $17.9 \mu \mathrm{~s}$ |
| 0 | 0 | 1 | $\phi / 64$ | $573.4 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | $\phi / 128$ | 1.1 ms |
| 0 | 1 | 1 | $\phi / 256$ | 2.3 ms |
| 1 | 0 | 0 | $\phi / 512$ | 4.6 ms |
| 1 | 0 | 1 | $\phi / 1024$ | 9.2 ms |
| 1 | 1 | 0 | $\phi / 4096$ | 36.7 ms |
| 1 | 1 | 1 | $\phi / 8192$ | 73.4 ms |

Note: * The overflow interval listed is the time from when the TCNT begins counting at H'00 until an overflow occurs.

### 13.2.3 Reset Control/Status Register (RSTCSR)

The RSTCSR is an 8-bit readable and writable register. (The RSTCSR differs from other registers in that it is more difficult to write. See section 13.2.4, Register Access, for details.) It controls output of the internal reset signal generated by timer counter (TCNT) overflow and selects the internal reset signal type. RSTCR is initialized to H'1F by input of a reset signal from the $\overline{\mathrm{RES}}$ pin, but is not initialized by the internal reset signal generated by the overflow of the WDT. It is initialized to $\mathrm{H}^{\prime} 1 \mathrm{~F}$ in standby mode.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | WOVF | RSTE | RSTS | - | - | - | - | - |
| Initial value: | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 |
| R/W: | $\mathrm{R} /(\mathrm{W})^{*}$ | R/W | R/W | R | R | R | R | R |

Note: * Only 0 can be written in bit 7 to clear the flag.

- Bit 7—Watchdog Timer Overflow Flag (WOVF): Indicates that the TCNT has overflowed ( $\mathrm{H}^{\prime} \mathrm{FF}-\mathrm{H}^{\prime} 00$ ) in the watchdog timer mode. It is not set in the interval timer mode.

| Bit 7: WOVF | Description |
| :--- | :--- |
| 0 | No TCNT overflow in watchdog timer mode (initial value) |
|  | Cleared when software reads WOVF, then writes 0 in WOVF |
| 1 | Set by TCNT overflow in watchdog timer mode |

- Bit 6—Reset Enable (RSTE): Selects whether to reset the chip internally if the TCNT overflows in the watchdog timer mode.

| Bit 6: RSTE | Description |
| :--- | :--- |
| 0 | Not reset when TCNT overflows (initial value). LSI not reset internally, <br> but TCNT and TCSR reset within WDT. |
| 1 | Reset when TCNT overflows |

- Bit 5—Reset Select (RSTS): Selects the type of internal reset generated if the TCNT overflows in the watchdog timer mode.

Bit 5: RSTS Description

| 0 | Power-on reset (initial value) |
| :--- | :--- |
| 1 | Manual reset |

- Bits 4-0—Reserved: These bits always read as 1 . The write value should always be 1 .


### 13.2.4 Register Access

The watchdog timer's TCNT, TCSR, and RSTCSR registers differ from other registers in that they are more difficult to write to. The procedures for writing and reading these registers are given below.

Writing to the TCNT and TCSR: These registers must be written by a word transfer instruction. They cannot be written by byte transfer instructions.

The TCNT and TCSR both have the same write address. The write data must be contained in the lower byte of the written word. The upper byte must be H'5A (for the TCNT) or H'A5 (for the TCSR) (figure 13.2). This transfers the write data from the lower byte to the TCNT or TCSR.


Figure 13.2 Writing to the TCNT and TCSR
Writing to the RSTCSR: The RSTCSR must be written by a word access to address H'FFFF8612. It cannot be written by byte transfer instructions.

Procedures for writing 0 in WOVF (bit 7) and for writing to RSTE (bit 6) and RSTS (bit 5) are different, as shown in figure 13.3.

To write 0 in the WOVF bit, the write data must be H'A5 in the upper byte and H'00 in the lower byte. This clears the WOVF bit to 0 . The RSTE and RSTS bits are not affected. To write to the RSTE and RSTS bits, the upper byte must be H'5A and the lower byte must be the write data. The values of bits 6 and 5 of the lower byte are transferred to the RSTE and RSTS bits, respectively. The WOVF bit is not affected.


Figure 13.3 Writing to the RSTCSR
Reading from the TCNT, TCSR, and RSTCSR: TCNT, TCSR, and RSTCSR are read like other registers. Use byte transfer instructions. The read addresses are H'FFFF8610 for the TCSR, H'FFFF8611 for the TCNT, and H'FFFF8613 for the RSTCSR.

### 13.3 Operation

### 13.3.1 Watchdog Timer Mode

To use the WDT as a watchdog timer, set the WT/ $\overline{\mathrm{IT}}$ and TME bits of the TCSR to 1 . Software must prevent TCNT overflow by rewriting the TCNT value (normally by writing H'00) before overflow occurs. No TCNT overflows will occur while the system is operating normally, but if the TCNT fails to be rewritten and overflows occur due to a system crash or the like, a $\overline{\text { WDTOVF }}$ signal is output externally (figure 13.4). The $\overline{\mathrm{WDTOVF}}$ signal can be used to reset the system. The $\overline{\text { WDTOVF }}$ signal is output for $128 \phi$ clock cycles.

If the RSTE bit in the RSTCSR is set to 1 , a signal to reset the chip will be generated internally simultaneous to the $\overline{\text { WDTOVF }}$ signal when TCNT overflows. Either a power-on reset or a manual reset can be selected by the RSTS bit. The internal reset signal is output for $512 \phi$ clock cycles.

When a watchdog overflow reset is generated simultaneously with a reset input at the $\overline{\mathrm{RES}}$ pin, the $\overline{\mathrm{RES}}$ reset takes priority, and the WOVF bit is cleared to 0 .

The following are not initialized a WDT reset signal:

- The MTU's POE (Port Output Enable) function register
- PFC (Pin Function Controller) function register
- I/O port register

Initializing is only possible by external power-on reset.


Figure 13.4 Operation in the Watchdog Timer Mode

### 13.3.2 Interval Timer Mode

To use the WDT as an interval timer, clear WT/IT to 0 and set TME to 1 . An interval timer interrupt (ITI) is generated each time the timer counter overflows. This function can be used to generate interval timer interrupts at regular intervals (figure 13.5).


Figure 13.5 Operation in the Interval Timer Mode

### 13.3.3 Clearing the Standby Mode

The watchdog timer has a special function to clear the standby mode with an NMI interrupt. When using the standby mode, set the WDT as described below.

Before Transition to the Standby Mode: The TME bit in the TCSR must be cleared to 0 to stop the watchdog timer counter before it enters the standby mode. The chip cannot enter the standby mode while the TME bit is set to 1 . Set bits CKS2-CKS0 so that the counter overflow interval is equal to or longer than the oscillation settling time. See sections 25.3, and 26.3, AC Characteristics, for the oscillation settling time.

Recovery from the Standby Mode: When an NMI request signal is received in standby mode, the clock oscillator starts running and the watchdog timer starts incrementing at the rate selected by bits CKS2-CKS0 before the standby mode was entered. When the TCNT overflows (changes from H'FF to $\mathrm{H}^{\prime} 00$ ), the clock is presumed to be stable and usable; clock signals are supplied to the entire chip and the standby mode ends.

For details on the standby mode, see section 24, Power-Down State.

### 13.3.4 Timing of Setting the Overflow Flag (OVF)

In the interval timer mode, when the TCNT overflows, the OVF flag of the TCSR is set to 1 and an interval timer interrupt is simultaneously requested (figure 13.6).


Figure 13.6 Timing of Setting the OVF

### 13.3.5 Timing of Setting the Watchdog Timer Overflow Flag (WOVF)

When the TCNT overflows in the watchdog timer mode, the WOVF bit of the RSTCSR is set to 1 and a $\overline{\text { WDTOVF }}$ signal is output. When the RSTE bit is set to 1 , TCNT overflow enables an internal reset signal to be generated for the entire chip (figure 13.7).


Figure 13.7 Timing of Setting the WOVF Bit

### 13.4 Notes on Use

### 13.4.1 TCNT Write and Increment Contention

If a timer counter (TCNT) increment clock pulse is generated during the $\mathrm{T}_{3}$ state of a write cycle to the TCNT, the write takes priority and the timer counter is not incremented (figure 13.8).


Figure 13.8 Contention between TCNT Write and Increment

### 13.4.2 Changing CKS2-CKS0 Bit Values

If the values of bits CKS2-CKS0 are altered while the WDT is running, the count may increment incorrectly. Always stop the watchdog timer (by clearing the TME bit to 0 ) before changing the values of bits CKS2-CKS0.

### 13.4.3 Changing between Watchdog Timer/Interval Timer Modes

To prevent incorrect operation, always stop the watchdog timer (by clearing the TME bit to 0 ) before switching between interval timer mode and watchdog timer mode.

### 13.4.4 System Reset With $\overline{\text { WDTOVF }}$

If a $\overline{\text { WDTOVF }}$ signal is input to the $\overline{\text { RES }}$ pin, the LSI cannot initialize correctly.
Avoid logical input of the $\overline{\text { WDTOVF }}$ output signal to the $\overline{\mathrm{RES}}$ input pin. To reset the entire system with the $\overline{\mathrm{WDTOVF}}$ signal, use the circuit shown in figure 13.9.


Figure 13.9 Example of a System Reset Circuit with a $\overline{\text { WDTOVF }}$ Signal

### 13.4.5 Internal Reset with the Watchdog Timer

If the RSTE bit is cleared to 0 in the watchdog timer mode, the LSI will not reset internally when a TCNT overflow occurs, but the TCNT and TCSR in the WDT will reset.

## Section 14 Serial Communication Interface (SCI)

### 14.1 Overview

The SH7040 Series has a serial communication interface (SCI) with two independent channels, both of which possess the same functions.

The SCI supports both asynchronous and clock synchronous serial communication. It also has a multiprocessor communication function for serial communication among two or more processors.

### 14.1.1 Features

- Select asynchronous or clock synchronous as the serial communications mode.
- Asynchronous mode: Serial data communications are synched by start-stop in character units. The SCI can communicate with a universal asynchronous receiver/transmitter (UART), an asynchronous communication interface adapter (ACIA), or any other chip that employs a standard asynchronous serial communication. It can also communicate with two or more other processors using the multiprocessor communication function. There are twelve selectable serial data communication formats.
Data length: seven or eight bits
Stop bit length: one or two bits
Parity: even, odd, or none
Multiprocessor bit: one or none
Receive error detection: parity, overrun, and framing errors
Break detection: by reading the RxD level directly when a framing error occurs
- Clocked synchronous mode: Serial data communication is synchronized with a clock signal. The SCI can communicate with other chips having a clock synchronous communication function. There is one serial data communication format.
Data length: eight bits
Receive error detection: overrun errors
- Full duplex communication: The transmitting and receiving sections are independent, so the SCI can transmit and receive simultaneously. Both sections use double buffering, so continuous data transfer is possible in both the transmit and receive directions.
- On-chip baud rate generator with selectable bit rates.
- Internal or external transmit/receive clock source: baud rate generator (internal) or SCK pin (external).
- Four types of interrupts: Transmit-data-empty, transmit-end, receive-data-full, and receiveerror interrupts are requested independently. The transmit-data-empty and receive-data-full interrupts can start the direct memory access controller (DMAC)/data transfer controller (DTC) to transfer data.


### 14.1.2 Block Diagram

Figure 14.1 shows a block diagram of the SCI.


Figure 14.1 SCI Block Diagram

### 14.1.3 Pin Configuration

Table 14.1 summarizes the SCI pins by channel.
Table 14.1 SCI Pins

| Channel | Pin Name | Abbreviation | Input/Output | Function |
| :--- | :--- | :--- | :--- | :--- |
| 0 | Serial clock pin | SCK0 | Input/output | SCI0 clock input/output |
|  | Receive data pin | RxD0 | Input | SCI0 receive data input |
|  | Transmit data pin | TxD0 | Output | SCI0 transmit data output |
| 1 | Serial clock pin | SCK1 | Input/output | SCI1 clock input/output |
|  | Receive data pin | RxD1 | Input | SCI1 receive data input |
|  | Transmit data pin | TxD1 | Output | SCI1 transmit data output |

### 14.1.4 Register Configuration

Table 14.2 summarizes the SCI internal registers. These registers select the communication mode (asynchronous or clock synchronous), specify the data format and bit rate, and control the transmitter and receiver sections.

Table 14.2 Registers

| Channel | Name | Abbreviation | R/W | Initial Value | Address*2 | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | Serial mode register | SMR0 | R/W | H'00 | H'FFFF81A0 | 8, 16 |
|  | Bit rate register | BRRO | R/W | H'FF | H'FFFF81A1 | 8, 16 |
|  | Serial control register | SCR0 | R/W | H'00 | H'FFFF81A2 | 8, 16 |
|  | Transmit data register | TDR0 | R/W | H'FF | H'FFFF81A3 | 8, 16 |
|  | Serial status register | SSR0 | $\mathrm{R} /(\mathrm{W})^{* 1}$ | H'84 | H'FFFF81A4 | 8,16 |
|  | Receive data register | RDR0 | R | H'00 | H'FFFF81A5 | 8, 16 |
| 1 | Serial mode register | SMR1 | R/W | H'00 | H'FFFF81B0 | 8,16 |
|  | Bit rate register | BRR1 | R/W | H'FF | H'FFFF81B1 | 8, 16 |
|  | Serial control register | SCR1 | R/W | H'00 | H'FFFF81B2 | 8,16 |
|  | Transmit data register | TDR1 | R/W | H'FF | H'FFFF81B3 | 8, 16 |
|  | Serial status register | SSR1 | $\mathrm{R} /(\mathrm{W})^{* 1}$ | H'84 | H'FFFF81B4 | 8, 16 |
|  | Receive data register | RDR1 | R | H'00 | H'FFFF81B5 | 8,16 |

Notes: *1 The only value that can be written is a 0 to clear the flags.
*2 Do not access empty addresses.

## Renesns

### 14.2 Register Descriptions

### 14.2.1 Receive Shift Register (RSR)

The receive shift register (RSR) receives serial data. Data input at the RxD pin is loaded into the RSR in the order received, LSB (bit 0) first, converting the data to parallel form. When one byte has been received, it is automatically transferred to the RDR.

The CPU cannot read or write the RSR directly.


### 14.2.2 Receive Data Register (RDR)

The receive data register (RDR) stores serial receive data. The SCI completes the reception of one byte of serial data by moving the received data from the receive shift register (RSR) into the RDR for storage. The RSR is then ready to receive the next data. This double buffering allows the SCI to receive data continuously.

The CPU can read but not write the RDR. The RDR is initialized to H'00 by a power-on reset or in standby mode. Manual reset does not initialize RDR.


### 14.2.3 Transmit Shift Register (TSR)

The transmit shift register (TSR) transmits serial data. The SCI loads transmit data from the transmit data register (TDR) into the TSR, then transmits the data serially from the TxD pin, LSB (bit 0) first. After transmitting one data byte, the SCI automatically loads the next transmit data from the TDR into the TSR and starts transmitting again. If the TDRE bit of the SSR is 1 , however, the SCI does not load the TDR contents into the TSR.

The CPU cannot read or write the TSR directly.


### 14.2.4 Transmit Data Register (TDR)

The transmit data register (TDR) is an 8-bit register that stores data for serial transmission. When the SCI detects that the transmit shift register (TSR) is empty, it moves transmit data written in the TDR into the TSR and starts serial transmission. Continuous serial transmission is possible by writing the next transmit data in the TDR during serial transmission from the TSR.

The CPU can always read and write the TDR. The TDR is initialized to H'FF by a power-on reset or in standby mode. Manual reset does not initialize TDR.


### 14.2.5 Serial Mode Register (SMR)

The serial mode register (SMR) is an 8-bit register that specifies the SCI serial communication format and selects the clock source for the baud rate generator.

The CPU can always read and write the SMR. The SMR is initialized to H'00 by a power-on reset or in standby mode. Manual reset does not initialize SMR.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{C} / \overline{\mathrm{A}}$ | CHR | PE | $\mathrm{O} / \overline{\mathrm{E}}$ | STOP | MP | CKS1 | CKS0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 7—Communication Mode (C/A): Selects whether the SCI operates in the asynchronous or clock synchronous mode.

Bit 7: C/A Description

| 0 | Asynchronous mode (initial value) |
| :--- | :--- |
| 1 | Clocked synchronous mode |

- Bit 6-Character Length (CHR): Selects 7-bit or 8-bit data in the asynchronous mode. In the clock synchronous mode, the data length is always eight bits, regardless of the CHR setting.

| Bit 6: CHR | Description |
| :--- | :--- |
| 0 | Eight-bit data (initial value) |
| 1 | Seven-bit data. (When 7-bit data is selected, the MSB (bit 7) of the <br> transmit data register is not transmitted.) |

- Bit 5—Parity Enable (PE): Selects whether to add a parity bit to transmit data and to check the parity of receive data, in the asynchronous mode. In the clock synchronous mode, a parity bit is neither added nor checked, regardless of the PE setting.

| Bit 5: PE | Description |
| :--- | :--- |
| 0 | Parity bit not added or checked (initial value) |
| 1 | Parity bit added and checked. When PE is set to 1, an even or odd <br> parity bit is added to transmit data, depending on the parity mode $(\mathrm{O} / \overline{\mathrm{E}})$ <br> setting. Receive data parity is checked according to the even/odd $(\mathrm{O} / \overline{\mathrm{E}})$ <br> mode setting. |

- Bit 4—Parity Mode (O/E): Selects even or odd parity when parity bits are added and checked. The $\mathrm{O} / \overline{\mathrm{E}}$ setting is used only in asynchronous mode and only when the parity enable bit (PE) is set to 1 to enable parity addition and check. The $O / \bar{E}$ setting is ignored in the clock synchronous mode, or in the asynchronous mode when parity addition and check is disabled.

| Bit 4: O/E | Description |
| :--- | :--- |
| 0 | Even parity (initial value). If even parity is selected, the parity bit is <br> added to transmit data to make an even number of 1 s in the transmitted <br> character and parity bit combined. Receive data is checked to see if it <br> has an even number of 1 s in the received character and parity bit <br> combined. |
| 1 | Odd parity. If odd parity is selected, the parity bit is added to transmit <br> data to make an odd number of 1s in the transmitted character and <br> parity bit combined. Receive data is checked to see if it has an odd <br> number of 1s in the received character and parity bit combined. |

- Bit 3—Stop Bit Length (STOP): Selects one or two bits as the stop bit length in the asynchronous mode. This setting is used only in the asynchronous mode. It is ignored in the clock synchronous mode because no stop bits are added.
In receiving, only the first stop bit is checked, regardless of the STOP bit setting. If the second stop bit is 1 , it is treated as a stop bit, but if the second stop bit is 0 , it is treated as the start bit of the next incoming character.

| Bit 3: STOP | Description |
| :--- | :--- |
| 0 | One stop bit (initial value). In transmitting, a single bit of 1 is added at <br> the end of each transmitted character. |
| 1 | Two stop bits. In transmitting, two bits of 1 are added at the end of each <br> transmitted character. |

- Bit 2—Multiprocessor Mode (MP): Selects multiprocessor format. When multiprocessor format is selected, settings of the parity enable (PE) and parity mode ( $\mathrm{O} / \overline{\mathrm{E}}$ ) bits are ignored. The MP bit setting is used only in the asynchronous mode; it is ignored in the clock synchronous mode. For the multiprocessor communication function, see section 14.3.3, Multiprocessor Communication.

Bit 2: MP Description

| 0 | Multiprocessor function disabled (initial value) |
| :--- | :--- |
| 1 | Multiprocessor format selected |

- Bits 1 and 0 -Clock Select 1 and 0 (CKS1 and CKS0): These bits select the internal clock source of the on-chip baud rate generator. Four clock sources are available; $\phi, \phi / 4, \phi / 16$, or $\phi / 64$. For further information on the clock source, bit rate register settings, and baud rate, see section 14.2.8, Bit Rate Register (BRR).

Bit 1: CKS1 Bit 0: CKSO Description

| 0 | 0 | $\phi$ (initial value) |
| :--- | :--- | :--- |
|  | 1 | $\phi / 4$ |
| 1 | 0 | $\phi / 16$ |
|  | $\phi / 64$ |  |

### 14.2.6 Serial Control Register (SCR)

The serial control register (SCR) operates the SCI transmitter/receiver, selects the serial clock output in the asynchronous mode, enables/disables interrupt requests, and selects the transmit/receive clock source. The CPU can always read and write the SCR. The SCR is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset or in standby mode. Manual reset does not initialize SCR.

| Bit: | 7 | 6 | 5 | 4 | 3 |  |  |  |  |  |  |  | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TIE | RIE | TE | RE | MPIE | TEIE | CKE1 | CKE0 |  |  |  |  |  |  |  |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |  |  |  |  |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |  |  |  |  |  |  |  |

- Bit 7—Transmit Interrupt Enable (TIE): Enables or disables the transmit-data-empty interrupt (TxI) requested when the transmit data register empty bit (TDRE) in the serial status register (SSR) is set to 1 by transfer of serial transmit data from the TDR to the TSR.

| Bit 7: TIE | Description |
| :--- | :--- |
| 0 | Transmit-data-empty interrupt request (TxI) is disabled (initial value). <br> The TxI interrupt request can be cleared by reading TDRE after it has <br> been set to 1, then clearing TDRE to 0, or by clearing TIE to 0. |
| 1 | Transmit-data-empty interrupt request (TxI) is enabled |

- Bit 6—Receive Interrupt Enable (RIE): Enables or disables the receive-data-full interrupt (RxI) requested when the receive data register full bit (RDRF) in the serial status register (SSR) is set to 1 by transfer of serial receive data from the RSR to the RDR. It also enables or disables receive-error interrupt (ERI) requests.
Bit 6: RIE Description

0
Receive-data-full interrupt (RxI) and receive-error interrupt (ERI) requests are disabled (initial value). RxI and ERI interrupt requests can be cleared by reading the RDRF flag or error flag (FER, PER, or ORER) after it has been set to 1 , then clearing the flag to 0 , or by clearing RIE to 0.

| 1 | Receive-data-full interrupt (RxI) and receive-error interrupt (ERI) <br> requests are enabled. |
| :--- | :--- |

- Bit 5-Transmit Enable (TE): Enables or disables the SCI serial transmitter.

| Bit 5: TE | Description |
| :--- | :--- |
| 0 | Transmitter disabled (initial value). The transmit data register empty bit <br> (TDRE) in the serial status register (SSR) is locked at 1. |
| 1 | Transmitter enabled. Serial transmission starts when the transmit data <br> register empty (TDRE) bit in the serial status register (SSR) is cleared to <br> 0 after writing of transmit data into the TDR. Select the transmit format <br> in the SMR before setting TE to 1. |

- Bit 4—Receive Enable (RE): Enables or disables the SCI serial receiver.


## Bit 4: RE

## Description

0
Receiver disabled (initial value). Clearing RE to 0 does not affect the receive flags (RDRF, FER, PER, ORER). These flags retain their previous values.

1
Receiver enabled. Serial reception starts when a start bit is detected in the asynchronous mode, or synchronous clock input is detected in the clock synchronous mode. Select the receive format in the SMR before setting RE to 1 .

- Bit 3-Multiprocessor Interrupt Enable (MPIE): Enables or disables multiprocessor interrupts. The MPIE setting is used only in the asynchronous mode, and only if the multiprocessor mode bit (MP) in the serial mode register (SMR) is set to 1 during reception. The MPIE setting is ignored in the clock synchronous mode or when the MP bit is cleared to 0 .

Bit 3: MPIE Description
Multiprocessor interrupts are disabled (normal receive operation) (initial value). MPIE is cleared when the MPIE bit is cleared to 0 , or the multiprocessor bit (MPB) is set to 1 in receive data.
1 Multiprocessor interrupts are enabled. Receive-data-full interrupt requests (RxI), receive-error interrupt requests (ERI), and setting of the RDRF, FER, and ORER status flags in the serial status register (SSR) are disabled until data with the multiprocessor bit set to 1 is received.
The SCI does not transfer receive data from the RSR to the RDR, does not detect receive errors, and does not set the RDRF, FER, and ORER flags in the serial status register (SSR). When it receives data that includes MPB $=1$, MPB is set to 1 , and the SCI automatically clears MPIE to 0 , generates RxI and ERI interrupts (if the TIE and RIE bits in the SCR are set to 1), and allows the FER and ORER bits to be set.

- Bit 2—Transmit-End Interrupt Enable (TEIE): Enables or disables the transmit-end interrupt (TEI) requested if TDR does not contain valid transmit data when the MSB is transmitted.

Bit 2: TEIE Description

| 0 | Transmit-end interrupt (TEI) requests are disabled.* (initial value) |
| :--- | :--- |
| 1 | Transmit-end interrupt (TEI) requests are enabled. ${ }^{*}$ |
| Note: * The TEI request can be cleared by reading the TDRE bit in the serial status register (SSR) |  | after it has been set to 1 , then clearing TDRE to 0 and clearing the transmit end (TEND) bit to 0 ; or by clearing the TEIE bit to 0 .

- Bits 1 and 0—Clock Enable 1 and 0 (CKE1 and CKE0): These bits select the SCI clock source and enable or disable clock output from the SCK pin. Depending on the combination of CKE1 and CKE0, the SCK pin can be used for serial clock output, or serial clock input. Select the SCK pin function by using the pin function controller (PFC).
The CKE0 setting is valid only in the asynchronous mode, and only when the SCI is internally clocked $(\mathrm{CKE} 1=0)$. The CKE0 setting is ignored in the clock synchronous mode, or when an external clock source is selected $($ CKE1 $=1)$. Select the SCI operating mode in the serial mode register (SMR) before setting CKE1 and CKE0. For further details on selection of the SCI clock source, see table 14.9 in section 14.3, Operation.


## Bit 1: Bit 0:

CKE1 CKE0 Description*1

| 0 | 0 | Asynchronous mode | Internal clock, SCK pin used for input pin (input signal is ignored) or output pin (output level is undefined) ${ }^{* 2}$ |
| :---: | :---: | :---: | :---: |
|  |  | Clock synchronous mode | Internal clock, SCK pin used for synchronous clock output*2 |
| 0 | 1 | Asynchronous mode | Internal clock, SCK pin used for clock output*3 |
|  |  | Clock synchronous mode | Internal clock, SCK pin used for synchronous clock output |
| 1 | 0 | Asynchronous mode | External clock, SCK pin used for clock input*4 |
|  |  | Clock synchronous mode | External clock, SCK pin used for synchronous clock input |
| 1 | 1 | Asynchronous mode | External clock, SCK pin used for clock input*4 |
|  |  | Clock synchronous mode | External clock, SCK pin used for synchronous clock input |

Notes: *1 The SCK pin is multiplexed with other functions. Use the pin function controller (PFC) to select the SCK function for this pin, as well as the I/O direction.
*2 Initial value.
*3 The output clock frequency is the same as the bit rate.
*4 The input clock frequency is 16 times the bit rate.

### 14.2.7 Serial Status Register (SSR)

The serial status register (SSR) is an 8-bit register containing multiprocessor bit values, and status flags that indicate SCI operating status.

The CPU can always read and write the SSR, but cannot write 1 in the status flags (TDRE, RDRF, ORER, PER, and FER). These flags can be cleared to 0 only if they have first been read (after being set to 1 ). Bits 2 (TEND) and 1 (MPB) are read-only bits that cannot be written. The SSR is initialized to $\mathrm{H}^{\prime} 84$ by a power-on reset or in standby mode. Manual reset does not initialize SSR.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TDRE | RDRF | ORER | FER | PER | TEND | MPB | MPBT |
| Initial value: | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| R/W: | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | $\mathrm{R} /(\mathrm{W})^{*}$ | R | R | R/W |

Note: * The only value that can be written is a 0 to clear the flag.

- Bit 7—Transmit Data Register Empty (TDRE): Indicates that the SCI has loaded transmit data from the TDR into the TSR and new serial transmit data can be written in the TDR.

Bit 7: TDRE Description

| 0 | TDR contains valid transmit data |
| :--- | :--- |
|  | TDRE is cleared to 0 when software reads TDRE after it has been set to 1, then <br> writes 0 in TDRE or the DMAC or DTC writes data in TDR |
| 1 | TDR does not contain valid transmit data (initial value) <br> TDRE is set to 1 when the chip is power-on reset or enters standby mode, the TE <br> bit in the serial control register (SCR) is cleared to 0 , or TDR contents are loaded <br> into TSR, so new data can be written in TDR |

- Bit 6—Receive Data Register Full (RDRF): Indicates that RDR contains received data.
\(\left.$$
\begin{array}{ll}\text { Bit 6: RDRF } & \text { Description } \\
\hline 0 & \text { RDR does not contain valid received data (initial value) } \\
& \begin{array}{l}\text { RDRF is cleared to } 0 \text { when the chip is power-on reset or enters standby mode, } \\
\text { software reads RDRF after it has been set to 1, then writes } 0 \text { in RDRF, or the }\end{array}
$$ <br>

\& DMAC or DTC reads data from RDR\end{array}\right]\)| RDR contains valid received data |
| :--- |
| 1 |
| RDRF is set to 1 when serial data is received normally and transferred from RSR |
| to RDR |

Note: The RDR and RDRF are not affected by detection of receive errors or by clearing of the RE bit to 0 in the serial control register. They retain their previous contents. If RDRF is still set to 1 when reception of the next data ends, an overrun error (ORER) occurs and the received data is lost.

- Bit 5-Overrun Error (ORER): Indicates that data reception ended abnormally due to an overrun error.


## Bit 5: ORER Description

$0 \quad$ Receiving is in progress or has ended normally (initial value). Clearing the RE bit to 0 in the serial control register does not affect the ORER bit, which retains its previous value.
ORER is cleared to 0 when the chip is power-on reset or enters standby mode or software reads ORER after it has been set to 1 , then writes 0 in ORER
1 A receive overrun error occurred. RDR continues to hold the data received before the overrun error, so subsequent receive data is lost. Serial receiving cannot continue while ORER is set to 1 . In the clock synchronous mode, serial transmitting is disabled.
ORER is set to 1 if reception of the next serial data ends when RDRF is set to 1

- Bit 4—Framing Error (FER): Indicates that data reception ended abnormally due to a framing error in the asynchronous mode.


## Bit 4: FER Description

| 0 | Receiving is in progress or has ended normally (initial value). Clearing the RE bit <br> to 0 in the serial control register does not affect the FER bit, which retains its <br> previous value. <br> FER is cleared to 0 when the chip is power-on reset or enters standby mode or <br> software reads FER after it has been set to 1 , then writes 0 in FER |
| :--- | :--- |
| 1 | A receive framing error occurred. When the stop bit length is two bits, only the <br> first bit is checked to see if it is a 1. The second stop bit is not checked. When a <br> framing error occurs, the SCI transfers the receive data into the RDR but does <br> not set RDRF. Serial receiving cannot continue while FER is set to 1. In the clock <br> synchronous mode, serial transmitting is also disabled. <br> FER is set to 1 if the stop bit at the end of receive data is checked and found to <br> be 0 |

- Bit 3—Parity Error (PER): Indicates that data reception (with parity) ended abnormally due to a parity error in the asynchronous mode.


## Bit 3: PER Description

| 0 | Receiving is in progress or has ended normally (initial value). Clearing the RE bit <br> to 0 in the serial control register does not affect the PER bit, which retains its <br> previous value. |
| :--- | :--- |
| PER is cleared to 0 when the chip is power-on reset or enters standby mode or |  |
| software reads PER after it has been set to 1 , then writes 0 in PER |  |

- Bit 2—Transmit End (TEND): Indicates that when the last bit of a serial character was transmitted, the TDR did not contain valid data, so transmission has ended. TEND is a readonly bit and cannot be written.

Bit 2: TEND Description

| 0 | Transmission is in progress |
| :--- | :--- |
|  | TEND is cleared to 0 when software reads TDRE after it has been set to 1, then |
| writes 0 in TDRE, or the DMAC or DTC writes data in TDR |  |

- Bit 1—Multiprocessor Bit (MPB): Stores the value of the multiprocessor bit in receive data when a multiprocessor format is selected for receiving in the asynchronous mode. The MPB is a read-only bit and cannot be written.


## Bit 1: MPB Description

$0 \quad$ Multiprocessor bit value in receive data is 0 (initial value). If RE is cleared to 0 when a multiprocessor format is selected, the MPB retains its previous value.
$1 \quad$ Multiprocessor bit value in receive data is 1

- Bit 0—Multiprocessor Bit Transfer (MPBT): Stores the value of the multiprocessor bit added to transmit data when a multiprocessor format is selected for transmitting in the asynchronous mode. The MPBT setting is ignored in the clock synchronous mode, when a multiprocessor format is not selected, or when the SCI is not transmitting.

Bit 0: MPBT Description

| 0 | Multiprocessor bit value in transmit data is 0 (initial value) |
| :--- | :--- |
| 1 | Multiprocessor bit value in transmit data is 1 |

### 14.2.8 Bit Rate Register (BRR)

The bit rate register (BRR) is an 8-bit register that, together with the baud rate generator clock source selected by the CKS1 and CKS0 bits in the serial mode register (SMR), determines the serial transmit/receive bit rate.

The CPU can always read and write the BRR. The BRR is initialized to H'FF by a power-on reset or in standby mode. Each channel has independent baud rate generator control, so different values can be set in the two channels. Manual reset does not initialize BRR.


Table 14.3 lists examples of BRR settings in the asynchronous mode; table 14.4 lists examples of BBR settings in the clock synchronous mode.

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 4 |  |  | 4.9152 |  |  | 6 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 2 | 70 | 0.03 | 2 | 86 | 0.31 | 2 | 106 | -0.44 |
| 150 | 1 | 207 | 0.16 | 1 | 255 | 0.00 | 2 | 77 | 0.16 |
| 300 | 1 | 103 | 0.16 | 1 | 127 | 0.00 | 1 | 155 | 0.16 |
| 600 | 0 | 207 | 0.16 | 0 | 255 | 0.00 | 1 | 77 | 0.16 |
| 1200 | 0 | 103 | 0.16 | 0 | 127 | 0.00 | 0 | 155 | 0.16 |
| 2400 | 0 | 51 | 0.16 | 0 | 63 | 0.00 | 0 | 77 | 0.16 |
| 4800 | 0 | 25 | 0.16 | 0 | 31 | 0.00 | 0 | 38 | 0.16 |
| 9600 | 0 | 12 | 0.16 | 0 | 15 | 0.00 | 0 | 19 | -2.34 |
| 14400 | 0 | 8 | -3.55 | 0 | 10 | -3.03 | 0 | 12 | 0.16 |
| 19200 | 0 | 6 | -6.99 | 0 | 7 | 0.00 | 0 | 9 | -2.34 |
| 28800 | 0 | 3 | 8.51 | 0 | 4 | 6.67 | 0 | 6 | -6.99 |
| 31250 | 0 | 3 | 0.00 | 0 | 4 | -1.70 | 0 | 5 | 0.00 |
| 38400 | 0 | 2 | 8.51 | 0 | 3 | 0.00 | 0 | 4 | -2.34 |

## Renesns

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 7.3728 |  |  | 8 |  |  | 9.8304 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 2 | 130 | -0.07 | 2 | 141 | 0.03 | 2 | 174 | -0.26 |
| 150 | 2 | 95 | 0.00 | 2 | 103 | 0.16 | 2 | 127 | 0.00 |
| 300 | 1 | 191 | 0.00 | 1 | 207 | 0.16 | 1 | 255 | 0.00 |
| 600 | 1 | 95 | 0.00 | 1 | 103 | 0.16 | 1 | 127 | 0.00 |
| 1200 | 0 | 191 | 0.00 | 0 | 207 | 0.16 | 0 | 255 | 0.00 |
| 2400 | 0 | 95 | 0.00 | 0 | 103 | 0.16 | 0 | 127 | 0.00 |
| 4800 | 0 | 47 | 0.00 | 0 | 51 | 0.16 | 0 | 63 | 0.00 |
| 9600 | 0 | 23 | 0.00 | 0 | 25 | 0.16 | 0 | 31 | 0.00 |
| 14400 | 0 | 15 | 0.00 | 0 | 16 | 2.12 | 0 | 20 | 1.59 |
| 19200 | 0 | 11 | 0.00 | 0 | 12 | 0.16 | 0 | 15 | 0.00 |
| 28800 | 0 | 7 | 0.00 | 0 | 8 | -3.55 | 0 | 10 | -3.03 |
| 31250 | 0 | 6 | 5.33 | 0 | 7 | 0.00 | 0 | 9 | -1.70 |
| 38400 | 0 | 5 | 0.00 | 0 | 6 | -6.99 | 0 | 7 | 0.00 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 10 |  |  | 11.0592 |  |  | 12 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 2 | 177 | -0.25 | 2 | 195 | 0.19 | 2 | 212 | 0.03 |
| 150 | 2 | 129 | 0.16 | 2 | 143 | 0.00 | 2 | 155 | 0.16 |
| 300 | 2 | 64 | 0.16 | 2 | 71 | 0.00 | 2 | 77 | 0.16 |
| 600 | 1 | 129 | 0.16 | 1 | 143 | 0.00 | 1 | 155 | 0.16 |
| 1200 | 1 | 64 | 0.16 | 1 | 71 | 0.00 | 1 | 77 | 0.16 |
| 2400 | 0 | 129 | 0.16 | 0 | 143 | 0.00 | 0 | 155 | 0.16 |
| 4800 | 0 | 64 | 0.16 | 0 | 71 | 0.00 | 0 | 77 | 0.16 |
| 9600 | 0 | 32 | -1.36 | 0 | 35 | 0.00 | 0 | 38 | 0.16 |
| 14400 | 0 | 21 | -1.36 | 0 | 23 | 0.00 | 0 | 25 | 0.16 |
| 19200 | 0 | 15 | 1.73 | 0 | 17 | 0.00 | 0 | 19 | -2.34 |
| 28800 | 0 | 10 | -1.36 | 0 | 11 | 0.00 | 0 | 12 | 0.16 |
| 31250 | 0 | 9 | 0.00 | 0 | 10 | 0.54 | 0 | 11 | 0.00 |
| 38400 | 0 | 7 | 1.73 | 0 | 8 | 0.00 | 0 | 9 | -2.34 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 12.288 |  |  | 14 |  |  | 14.7456 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 2 | 217 | 0.08 | 2 | 248 | -0.17 | 3 | 64 | 0.70 |
| 150 | 2 | 159 | 0.00 | 2 | 181 | 0.16 | 2 | 191 | 0.00 |
| 300 | 2 | 79 | 0.00 | 2 | 90 | 0.16 | 2 | 95 | 0.00 |
| 600 | 1 | 159 | 0.00 | 1 | 181 | 0.16 | 1 | 191 | 0.00 |
| 1200 | 1 | 79 | 0.00 | 1 | 90 | 0.16 | 1 | 95 | 0.00 |
| 2400 | 0 | 159 | 0.00 | 0 | 181 | 0.16 | 0 | 191 | 0.00 |
| 4800 | 0 | 79 | 0.00 | 0 | 90 | 0.16 | 0 | 95 | 0.00 |
| 9600 | 0 | 39 | 0.00 | 0 | 45 | -0.93 | 0 | 47 | 0.00 |
| 14400 | 0 | 26 | -1.23 | 0 | 29 | 1.27 | 0 | 31 | 0.00 |
| 19200 | 0 | 19 | 0.00 | 0 | 22 | -0.93 | 0 | 23 | 0.00 |
| 28800 | 0 | 12 | 2.56 | 0 | 14 | 1.27 | 0 | 15 | 0.00 |
| 31250 | 0 | 11 | 2.40 | 0 | 13 | 0.00 | 0 | 14 | -1.70 |
| 38400 | 0 | 9 | 0.00 | 0 | 10 | 3.57 | 0 | 11 | 0.00 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 16 |  |  | 17.2032 |  |  | 18 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 70 | 0.03 | 3 | 75 | 0.48 | 3 | 79 | -0.12 |
| 150 | 2 | 207 | 0.16 | 2 | 223 | 0.00 | 2 | 233 | 0.16 |
| 300 | 2 | 103 | 0.16 | 2 | 111 | 0.00 | 2 | 116 | 0.16 |
| 600 | 1 | 207 | 0.16 | 1 | 223 | 0.00 | 1 | 233 | 0.16 |
| 1200 | 1 | 103 | 0.16 | 1 | 111 | 0.00 | 1 | 116 | 0.16 |
| 2400 | 0 | 207 | 0.16 | 0 | 223 | 0.00 | 0 | 233 | 0.16 |
| 4800 | 0 | 103 | 0.16 | 0 | 111 | 0.00 | 0 | 116 | 0.16 |
| 9600 | 0 | 51 | 0.16 | 0 | 55 | 0.00 | 0 | 58 | -0.69 |
| 14400 | 0 | 34 | -0.79 | 0 | 36 | 0.90 | 0 | 38 | 0.16 |
| 19200 | 0 | 25 | 0.16 | 0 | 27 | 0.00 | 0 | 28 | 1.02 |
| 28800 | 0 | 16 | 2.12 | 0 | 18 | -1.75 | 0 | 19 | -2.34 |
| 31250 | 0 | 15 | 0.00 | 0 | 16 | 1.20 | 0 | 17 | 0.00 |
| 38400 | 0 | 12 | 0.16 | 0 | 13 | 0.00 | 0 | 14 | -2.34 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 18.432 |  |  | 19.6608 |  |  | 20 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 81 | -0.22 | 3 | 86 | 0.31 | 3 | 88 | -0.25 |
| 150 | 2 | 239 | 0.00 | 2 | 255 | 0.00 | 3 | 64 | 0.16 |
| 300 | 2 | 119 | 0.00 | 2 | 127 | 0.00 | 2 | 129 | 0.16 |
| 600 | 1 | 239 | 0.00 | 1 | 255 | 0.00 | 2 | 64 | 0.16 |
| 1200 | 1 | 119 | 0.00 | 1 | 127 | 0.00 | 1 | 129 | 0.16 |
| 2400 | 0 | 239 | 0.00 | 0 | 255 | 0.00 | 1 | 64 | 0.16 |
| 4800 | 0 | 119 | 0.00 | 0 | 127 | 0.00 | 0 | 129 | 0.16 |
| 9600 | 0 | 59 | 0.00 | 0 | 63 | 0.00 | 0 | 64 | 0.16 |
| 14400 | 0 | 39 | 0.00 | 0 | 42 | -0.78 | 0 | 42 | 0.94 |
| 19200 | 0 | 29 | 0.00 | 0 | 31 | 0.00 | 0 | 32 | -1.36 |
| 28800 | 0 | 19 | 0.00 | 0 | 20 | 1.59 | 0 | 21 | -1.36 |
| 31250 | 0 | 17 | 2.40 | 0 | 19 | -1.70 | 0 | 19 | 0.00 |
| 38400 | 0 | 14 | 0.00 | 0 | 15 | 0.00 | 0 | 15 | 1.73 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 22 |  |  | 22.1184 |  |  | 24 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 97 | -0.35 | 3 | 97 | 0.19 | 3 | 106 | -0.44 |
| 150 | 3 | 71 | -0.54 | 3 | 71 | 0.00 | 3 | 77 | 0.16 |
| 300 | 2 | 142 | 0.16 | 2 | 143 | 0.00 | 2 | 155 | 0.16 |
| 600 | 2 | 71 | -0.54 | 2 | 71 | 0.00 | 2 | 77 | 0.16 |
| 1200 | 1 | 142 | 0.16 | 1 | 143 | 0.00 | 1 | 155 | 0.16 |
| 2400 | 1 | 71 | -0.54 | 1 | 71 | 0.00 | 1 | 77 | 0.16 |
| 4800 | 0 | 142 | 0.16 | 0 | 143 | 0.00 | 0 | 155 | 0.16 |
| 9600 | 0 | 71 | -0.54 | 0 | 71 | 0.00 | 0 | 77 | 0.16 |
| 14400 | 0 | 47 | -0.54 | 0 | 47 | 0.00 | 0 | 51 | 0.16 |
| 19200 | 0 | 35 | -0.54 | 0 | 35 | 0.00 | 0 | 38 | 0.16 |
| 28800 | 0 | 23 | -0.54 | 0 | 23 | 0.00 | 0 | 25 | 0.16 |
| 31250 | 0 | 21 | 0.00 | 0 | 21 | 0.54 | 0 | 23 | 0.00 |
| 38400 | 0 | 17 | -0.54 | 0 | 17 | 0.00 | 0 | 19 | -2.34 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 24.576 |  |  | 25.8048 |  |  | 26 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 108 | 0.08 | 3 | 114 | -0.40 | 3 | 114 | 0.36 |
| 150 | 3 | 79 | 0.00 | 3 | 83 | 0.00 | 3 | 84 | -0.43 |
| 300 | 2 | 159 | 0.00 | 2 | 167 | 0.00 | 2 | 168 | 0.16 |
| 600 | 2 | 79 | 0.00 | 2 | 83 | 0.00 | 2 | 84 | -0.43 |
| 1200 | 1 | 159 | 0.00 | 1 | 167 | 0.00 | 1 | 168 | 0.16 |
| 2400 | 1 | 79 | 0.00 | 1 | 83 | 0.00 | 1 | 84 | -0.43 |
| 4800 | 0 | 159 | 0.00 | 0 | 167 | 0.00 | 0 | 168 | 0.16 |
| 9600 | 0 | 79 | 0.00 | 0 | 83 | 0.00 | 0 | 84 | -0.43 |
| 14400 | 0 | 52 | 0.63 | 0 | 55 | 0.00 | 0 | 55 | 0.76 |
| 19200 | 0 | 39 | 0.00 | 0 | 41 | 0.00 | 0 | 41 | 0.76 |
| 28800 | 0 | 26 | -1.23 | 0 | 27 | 0.00 | 0 | 27 | 0.76 |
| 31250 | 0 | 24 | -1.70 | 0 | 25 | -0.75 | 0 | 25 | 0.00 |
| 38400 | 0 | 19 | 0.00 | 0 | 20 | 0.00 | 0 | 20 | 0.76 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 27.0336 |  |  | 28 |  |  | 29.4912 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 119 | 0.00 | 3 | 123 | 0.23 | 3 | 130 | -0.07 |
| 150 | 3 | 87 | 0.00 | 3 | 90 | 0.16 | 3 | 95 | 0.00 |
| 300 | 2 | 175 | 0.00 | 2 | 181 | 0.16 | 2 | 191 | 0.00 |
| 600 | 1 | 87 | 0.00 | 2 | 90 | 0.16 | 2 | 95 | 0.00 |
| 1200 | 1 | 175 | 0.00 | 1 | 181 | 0.16 | 1 | 191 | 0.00 |
| 2400 | 1 | 87 | 0.00 | 1 | 90 | 0.16 | 1 | 95 | 0.00 |
| 4800 | 0 | 175 | 0.00 | 0 | 181 | 0.16 | 0 | 191 | 0.00 |
| 9600 | 0 | 87 | 0.00 | 0 | 90 | 0.16 | 0 | 95 | 0.00 |
| 14400 | 0 | 58 | -0.56 | 0 | 60 | -0.39 | 0 | 63 | 0.00 |
| 19200 | 0 | 43 | 0.00 | 0 | 45 | 0.93 | 0 | 47 | 0.00 |
| 28800 | 0 | 28 | 1.15 | 0 | 29 | 1.27 | 0 | 31 | 0.00 |
| 31250 | 0 | 26 | 0.12 | 0 | 27 | 0.00 | 0 | 28 | 1.69 |
| 38400 | 0 | 21 | 0.00 | 0 | 22 | -0.93 | 0 | 23 | 0.00 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 30 |  |  | 31.9488 |  |  | 32 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 132 | 0.13 | 3 | 141 | -0.13 | 3 | 141 | 0.03 |
| 150 | 3 | 97 | -0.35 | 3 | 103 | 0.00 | 3 | 103 | 0.16 |
| 300 | 2 | 194 | 0.16 | 2 | 207 | 0.00 | 2 | 207 | 0.16 |
| 600 | 2 | 97 | -0.35 | 2 | 103 | 0.00 | 2 | 103 | 0.16 |
| 1200 | 1 | 194 | 0.16 | 1 | 207 | 0.00 | 1 | 207 | 0.16 |
| 2400 | 1 | 97 | -0.35 | 1 | 103 | 0.00 | 1 | 103 | 0.16 |
| 4800 | 0 | 194 | 0.16 | 0 | 207 | 0.00 | 0 | 207 | 0.16 |
| 9600 | 0 | 97 | -0.35 | 0 | 103 | 0.00 | 0 | 103 | 0.16 |
| 14400 | 0 | 64 | 0.16 | 0 | 68 | 0.48 | 0 | 68 | 0.64 |
| 19200 | 0 | 48 | -0.35 | 0 | 51 | 0.00 | 0 | 51 | 0.16 |
| 28800 | 0 | 32 | -1.36 | 0 | 34 | -0.95 | 0 | 34 | -0.79 |
| 31250 | 0 | 29 | 0.00 | 0 | 31 | -0.16 | 0 | 31 | 0.00 |
| 38400 | 0 | 23 | 1.73 | 0 | 25 | 0.00 | 0 | 25 | 0.16 |

Table 14.3 Bit Rates and BRR Settings in Asynchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 33 |  |  | 33.1776 |  |  | 33.3333 |  |  |
|  | n | N | Error (\%) | n | N | Error (\%) | n | N | Error (\%) |
| 110 | 3 | 145 | 0.33 | 3 | 146 | 0.19 | 3 | 147 | -0.02 |
| 150 | 3 | 106 | 0.39 | 3 | 107 | 0.00 | 3 | 108 | -0.45 |
| 300 | 2 | 214 | -0.07 | 2 | 215 | 0.00 | 2 | 216 | 0.01 |
| 600 | 2 | 106 | 0.39 | 2 | 107 | 0.00 | 2 | 108 | -0.45 |
| 1200 | 1 | 214 | -0.07 | 1 | 215 | 0.00 | 1 | 216 | 0.01 |
| 2400 | 1 | 106 | 0.39 | 1 | 107 | 0.00 | 1 | 108 | -0.45 |
| 4800 | 0 | 214 | -0.07 | 0 | 215 | 0.00 | 0 | 216 | 0.01 |
| 9600 | 0 | 106 | 0.39 | 0 | 107 | 0.00 | 0 | 108 | -0.45 |
| 14400 | 0 | 71 | -0.54 | 0 | 91 | 0.00 | 0 | 91 | 0.47 |
| 19200 | 0 | 53 | -0.54 | 0 | 53 | 0.00 | 0 | 53 | 0.47 |
| 28800 | 0 | 35 | -0.54 | 0 | 35 | 0.00 | 0 | 35 | 0.47 |
| 31250 | 0 | 32 | 0.00 | 0 | 32 | 0.54 | 0 | 32 | 1.01 |
| 38400 | 0 | 26 | -0.54 | 0 | 26 | 0.00 | 0 | 26 | 0.47 |

Table 14.4 Bit Rates and BRR Settings in Clocked Synchronous Mode

| Bit Rate (Bits/s) | $\phi(\mathrm{MHz})$ |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 4 |  | 8 |  | 10 |  | 12 |  |
|  | n | N | n | N | n | N | n | N |
| 110 | 3 | 141 |  |  |  |  |  |  |
| 250 | 2 | 249 | 3 | 124 | 3 | 155 | 3 | 187 |
| 500 | 2 | 124 | 2 | 249 | 3 | 77 | 3 | 93 |
| 1k | 1 | 249 | 2 | 124 | 2 | 155 | 2 | 187 |
| 2.5k | 1 | 99 | 1 | 199 | 1 | 249 | 2 | 74 |
| 5k | 0 | 199 | 1 | 99 | 1 | 124 | 1 | 149 |
| 10k | 0 | 99 | 0 | 199 | 0 | 249 | 1 | 74 |
| 25k | 0 | 39 | 0 | 79 | 0 | 99 | 0 | 119 |
| 50k | 0 | 19 | 0 | 39 | 0 | 49 | 0 | 59 |
| 100k | 0 | 9 | 0 | 19 | 0 | 24 | 0 | 29 |
| 250k | 0 | 3 | 0 | 7 | 0 | 9 | 0 | 11 |
| 500k | 0 | 1 | 0 | 3 | 0 | 4 | 0 | 5 |
| 1M | 0 | 0* | 0 | 1 | - | - | 0 | 2 |
| 2.5 M |  |  |  |  | 0 | 0* | 0 | 0* |
| 5M |  |  |  |  |  |  |  |  |

Table 14.4 Bit Rates and BRR Settings in Clocked Synchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 16 |  | 20 |  | 24 |  | 28 |  |
|  | n | N | n | N | n | N | n | N |
| 110 |  |  |  |  |  |  |  |  |
| 250 | 3 | 249 |  |  |  |  |  |  |
| 500 | 3 | 124 | 3 | 155 | 3 | 187 | 3 | 218 |
| 1k | 2 | 249 | 3 | 77 | 3 | 93 | 3 | 108 |
| 2.5k | 2 | 99 | 2 | 124 | 2 | 149 | 2 | 174 |
| 5k | 1 | 199 | 1 | 249 | 2 | 74 | 2 | 87 |
| 10k | 1 | 99 | 1 | 124 | 1 | 149 | 1 | 174 |
| 25k | 0 | 159 | 0 | 199 | 0 | 239 | 1 | 69 |
| 50k | 0 | 79 | 0 | 99 | 0 | 119 | 0 | 139 |
| 100k | 0 | 39 | 0 | 49 | 0 | 59 | 0 | 69 |
| 250k | 0 | 15 | 0 | 19 | 0 | 23 | 0 | 27 |
| 500k | 0 | 7 | 0 | 9 | 0 | 11 | 0 | 13 |
| 1M | 0 | 3 | 0 | 4 | 0 | 5 | 0 | 6 |
| 2.5M | - | - | 0 | 1 | - | - | 0 | 2 |
| 3.5M |  |  | - | - | - | - | 0 | 1 |
| 5M |  |  | 0 | 0* | - | - | - | - |
| 7M |  |  |  |  | - | - | 0 | 0* |

Table 14.4 Bit Rates and BRR Settings in Clocked Synchronous Mode (cont)

| Bit Rate (Bits/s) | $\phi$ (MHz) |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 30 |  | 32 |  | 33 |  | 33.3333 |  |
|  | n | N | n | N | n | N | n | N |
| 110 |  |  |  |  |  |  |  |  |
| 250 |  |  |  |  |  |  |  |  |
| 500 | 3 | 233 | 3 | 249 |  |  |  |  |
| 1k | 3 | 116 | 3 | 124 | 3 | 128 | 3 | 129 |
| 2.5k | 2 | 187 | 2 | 199 | 2 | 205 | 2 | 207 |
| 5k | 2 | 93 | 2 | 99 | 2 | 102 | 2 | 103 |
| 10k | 1 | 187 | 1 | 199 | 1 | 205 | 1 | 207 |
| 25k | 1 | 74 | 1 | 79 | 1 | 82 | 1 | 82 |
| 50k | 0 | 149 | 0 | 159 | 0 | 164 | 0 | 166 |
| 100k | 0 | 74 | 0 | 79 | 0 | 82 | 0 | 82 |
| 250k | 0 | 29 | 0 | 31 | 0 | 32 | 0 | 32 |
| 500k | 0 | 14 | 0 | 15 | 0 | 16 | 0 | 16 |
| 1M | 0 | 7 | 0 | 7 | 0 | 7 | 0 | 7 |
| 2.5 M | 0 | 2 | 0 | 2 | 0 | 2 | - | - |
| 5M | - | - | - | - | - | - | - | - |
| 7M | - | - | - | - | - | - | - | - |

Note: Settings with an error of $1 \%$ or less are recommended.
Legend
Blank: No setting available
-: Setting possible, but error occurs
*: Continuous transmission/reception is not possible.
The BRR setting is calculated as follows:
Asynchronous mode:
$N=\frac{\phi}{64 \times 2^{2 n-1} \times B} \times 10^{6}-1$
Synchronous mode:

$$
N=\frac{\phi}{64 \times 2^{2 n-1} \times B} \times 10^{6}-1
$$

B: Bit rate (bit/s)
N : Baud rate generator BRR setting $(0 \leq \mathrm{N} \leq 255)$
$\phi$ : Operating frequency ( MHz )
n : Baud rate generator input clock ( $\mathrm{n}=0$ to 3 )
(See the following table for the clock sources and value of $n$.)

|  |  | SMR Settings |  |
| :--- | :--- | :--- | :--- |
| $\mathbf{n}$ | Clock Source | CKS1 | CKS2 |
| 0 | $\phi$ | 0 | 0 |
| 1 | $\phi / 4$ | 0 | 1 |
| 2 | $\phi / 16$ | 1 | 0 |
| 3 | $\phi / 64$ | 1 | 1 |

The bit rate error in asynchronous mode is calculated as follows:

$$
\text { Error }(\%)=\left\{\frac{\phi \times 10^{6}}{(N+1) \times B \times 64 \times 2^{2 n-1}}-1\right\} \times 100
$$

Table 14.5 indicates the maximum bit rates in the asynchronous mode when the baud rate generator is being used for various frequencies. Tables 14.6 and 14.7 show the maximum rates for external clock input.

Table 14.5 Maximum Bit Rates for Various Frequencies with Baud Rate Generator (Asynchronous Mode)

| $\phi$ (MHz) | Maximum Bit Rate (Bits/s) | Settings |  |
| :---: | :---: | :---: | :---: |
|  |  | n | N |
| 4 | 125000 | 0 | 0 |
| 4.9152 | 153600 | 0 | 0 |
| 6 | 187500 | 0 | 0 |
| 7.3728 | 230400 | 0 | 0 |
| 8 | 250000 | 0 | 0 |
| 9.8304 | 307200 | 0 | 0 |
| 10 | 312500 | 0 | 0 |
| 11.0592 | 345600 | 0 | 0 |
| 12 | 375000 | 0 | 0 |
| 12.288 | 384000 | 0 | 0 |
| 14 | 437500 | 0 | 0 |
| 14.7456 | 460800 | 0 | 0 |
| 16 | 500000 | 0 | 0 |
| 17.2032 | 537600 | 0 | 0 |
| 18 | 562500 | 0 | 0 |
| 18.432 | 576000 | 0 | 0 |
| 19.6608 | 614400 | 0 | 0 |
| 20 | 625000 | 0 | 0 |
| 22 | 687500 | 0 | 0 |
| 22.1184 | 691200 | 0 | 0 |
| 24 | 750000 | 0 | 0 |
| 24.576 | 768000 | 0 | 0 |
| 25.8048 | 806400 | 0 | 0 |
| 26 | 812500 | 0 | 0 |
| 27.0336 | 844800 | 0 | 0 |
| 28 | 875000 | 0 | 0 |
| 29.4912 | 921600 | 0 | 0 |
| 30 | 937500 | 0 | 0 |
| 31.9488 | 998400 | 0 | 0 |
| 32 | 1000000 | 0 | 0 |
| 33 | 1031250 | 0 | 0 |
| 33.1776 | 1036800 | 0 | 0 |
| 33.3333 | 1041666 | 0 | 0 |
| 498 -RENESN |  |  |  |
|  |  |  |  |  |

Table 14.6 Maximum Bit Rates during External Clock Input (Asynchronous Mode)

| $\phi$ (MHz) | External Input Clock (MHz) | Maximum Bit Rate (Bits/s) |
| :---: | :---: | :---: |
| 4 | 1.0000 | 62500 |
| 4.9152 | 1.2288 | 76800 |
| 6 | 1.5000 | 93750 |
| 7.3728 | 1.8432 | 115200 |
| 8 | 2.0000 | 125000 |
| 9.8304 | 2.4576 | 153600 |
| 10 | 2.5000 | 156250 |
| 11.0592 | 2.7648 | 172800 |
| 12 | 3.0000 | 187500 |
| 12.288 | 3.0720 | 192000 |
| 14 | 3.5000 | 218750 |
| 14.7456 | 3.6864 | 230400 |
| 16 | 4.0000 | 250000 |
| 17.2032 | 4.3008 | 268800 |
| 18 | 4.5000 | 281250 |
| 18.432 | 4.6080 | 288000 |
| 19.6608 | 4.9152 | 307200 |
| 20 | 5.0000 | 312500 |
| 22 | 5.5000 | 343750 |
| 22.1184 | 5.5296 | 345600 |
| 24 | 6.0000 | 375000 |
| 24.576 | 6.1440 | 384000 |
| 25.8048 | 6.4512 | 403200 |
| 26 | 6.5000 | 406250 |
| 27.0336 | 6.7584 | 422400 |
| 28 | 7.0000 | 437500 |
| 29.4912 | 7.3728 | 460800 |
| 30 | 7.5000 | 468750 |
| 31.9488 | 7.9872 | 499200 |
| 32 | 8.0000 | 500000 |
| 33 | 8.2500 | 515625 |
| 33.1776 | 8.2944 | 518400 |
| 33.3333 | 8.3333 | 520832.8125 |

Table 14.7 Maximum Bit Rates during External Clock Input (Clock Synchronous Mode)

| $\phi(\mathbf{M H z})$ | External Input Clock (MHz) | Maximum Bit Rate (Bits/s) |
| :--- | :--- | :--- |
| 4 | 0.6667 | 666666.7 |
| 6 | 1.0000 | 1000000.0 |
| 8 | 1.3333 | 1333333.3 |
| 10 | 1.6667 | 1666666.7 |
| 12 | 2.0000 | 2000000.0 |
| 14 | 2.3333 | 2333333.3 |
| 16 | 2.6667 | 2666666.7 |
| 18 | 3.0000 | 3000000.0 |
| 20 | 3.3333 | 3333333.3 |
| 22 | 3.6667 | 3666666.7 |
| 24 | 4.0000 | 4000000.0 |
| 26 | 4.3333 | 4333333.3 |
| 28 | 4.6667 | 4666666.7 |
| 30 | 5.0000 | 5000000.0 |
| 32 | 5.3333 | 5333333.3 |
| 33.3333 | 5.5556 | 5555550.0 |

### 14.3 Operation

### 14.3.1 Overview

For serial communication, the SCI has an asynchronous mode in which characters are synchronized individually, and a clock synchronous mode in which communication is synchronized with clock pulses. Asynchronous/clock synchronous mode and the transmission format are selected in the serial mode register (SMR), as shown in table 14.8. The SCI clock source is selected by the C/A bit in the serial mode register (SMR) and the CKE1 and CKE0 bits in the serial control register (SCR), as shown in table 14.9.

## Asynchronous Mode:

- Data length is selectable: seven or eight bits.
- Parity and multiprocessor bits are selectable, as well as the stop bit length (one or two bits). These selections determine the transmit/receive format and character length.
- In receiving, it is possible to detect framing errors (FER), parity errors (PER), overrun errors (ORER), and the break state.
- An internal or external clock can be selected as the SCI clock source.
- When an internal clock is selected, the SCI operates using the on-chip baud rate generator clock, and can output a clock with a frequency matching the bit rate.
- When an external clock is selected, the external clock input must have a frequency 16 times the bit rate. (The on-chip baud rate generator is not used.)


## Clock Synchronous Mode:

- The communication format has a fixed 8-bit data length.
- In receiving, it is possible to detect overrun errors (ORER).
- An internal or external clock can be selected as the SCI clock source.
- When an internal clock is selected, the SCI operates using the on-chip baud rate generator clock, and outputs a synchronous clock signal to external devices.
- When an external clock is selected, the SCI operates on the input synchronous clock. The on-chip baud rate generator is not used.

Table 14.8 Serial Mode Register Settings and SCI Communication Formats

| Mode | SMR Settings |  |  |  |  | SCI Communication Format |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \overline{\text { Bit } 7} \\ & \mathrm{C} / \overline{\mathbf{A}} \end{aligned}$ | $\begin{aligned} & \hline \text { Bit } 6 \\ & \text { CHR } \end{aligned}$ | $\begin{aligned} & \hline \text { Bit } 5 \\ & \text { PE } \end{aligned}$ | $\begin{aligned} & \text { Bit } 2 \\ & \text { MP } \end{aligned}$ | $\begin{aligned} & \text { Bit } 3 \\ & \text { STOP } \end{aligned}$ | Data Length | Parity Bit | Multiprocessor Bit | Stop Bit Length |
| Asynchronous | 0 | 0 | 0 | 0 | 0 | 8-bit | Not set | Not set | 1 bit |
|  |  |  |  |  | 1 |  |  |  | 2 bits |
|  |  |  | 1 |  | 0 |  | Set |  | 1 bit |
|  |  |  |  |  | 1 |  |  |  | 2 bits |
|  |  | 1 | 0 |  | 0 | 7-bit | Not set |  | 1 bit |
|  |  |  |  |  | 1 |  |  |  | 2 bits |
|  |  |  | 1 |  | 0 |  | Set |  | 1 bit |
|  |  |  |  |  | 1 |  |  |  | 2 bits |
| Asynchronous (multiprocessor format) |  | 0 | * | 1 | 0 | 8-bit | Not set | Set | 1 bit |
|  |  |  | * |  | 1 |  |  |  | 2 bits |
|  |  | 1 | * |  | 0 | 7-bit |  |  | 1 bit |
|  |  |  | * |  | 1 |  |  |  | 2 bits |
| Clock synchronous | 1 | * | * | * | * | 8-bit |  | Not set | None |

Note: Asterisks $\left({ }^{*}\right)$ in the table indicate don't-care bits.

Table 14.9 SMR and SCR Settings and SCI Clock Source Selection

| Mode | $\begin{aligned} & \frac{\text { SMR }}{} \frac{\text { Bit } 7}{\text { C } / \bar{A}} \end{aligned}$ | SCR Settings |  | SCI Transmit/Receive Clock |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Bit 1 <br> CKE1 | $\begin{aligned} & \hline \text { Bit } 0 \\ & \text { CKEO } \end{aligned}$ | Clock Source | SCK Pin Function* |
| Asynchronous | 0 | 0 | 0 | Internal | SCI does not use the SCK pin |
|  |  |  | 1 |  | Outputs a clock with frequency matching the bit rate |
|  |  | 1 | 0 | External | Inputs a clock with frequency 16 times the bit rate |
|  |  |  | 1 |  |  |
| Clock synchronous | 1 | 0 | 0 | Internal | Outputs the synchronous clock |
|  |  |  | 1 |  |  |
|  |  | 1 | 0 | External | Inputs the synchronous clock |
|  |  |  | 1 |  |  |

Note: * Select the function in combination with the pin function controller (PFC).

### 14.3.2 Operation in Asynchronous Mode

In the asynchronous mode, each transmitted or received character begins with a start bit and ends with a stop bit. Serial communication is synchronized one character at a time.

The transmitting and receiving sections of the SCI are independent, so full duplex communication is possible. The transmitter and receiver are both double buffered, so data can be written and read while transmitting and receiving are in progress, enabling continuous transmitting and receiving.

Figure 14.2 shows the general format of asynchronous serial communication. In asynchronous serial communication, the communication line is normally held in the marking (high) state. The SCI monitors the line and starts serial communication when the line goes to the space (low) state, indicating a start bit. One serial character consists of a start bit (low), data (LSB first), parity bit (high or low), and stop bit (high), in that order.

When receiving in the asynchronous mode, the SCI synchronizes on the falling edge of the start bit. The SCI samples each data bit on the eighth pulse of a clock with a frequency 16 times the bit rate. Receive data is latched at the center of each bit.


Figure 14.2 Data Format in Asynchronous Communication (Example: 8-bit Data with Parity and Two Stop Bits)

Transmit/Receive Formats: Table 14.10 shows the 11 communication formats that can be selected in the asynchronous mode. The format is selected by settings in the serial mode register (SMR).

Table 14.10 Serial Communication Formats (Asynchronous Mode)

| SMR Bits |  |  |  | Serial Transmit/Receive Format and Frame Length |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CHR | PE | MP | STOP | 1 | 23 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 |
| 0 | 0 | 0 | 0 | START |  |  | 8-Bi | it dat |  |  |  | STOP |  |  |
| 0 | 0 | 0 | 1 | START |  |  | 8-Bi | it dat |  |  |  | STOP | STOP |  |
| 0 | 1 | 0 | 0 | START |  |  | 8-Bi | it dat |  |  |  | P | STOP |  |
| 0 | 1 | 0 | 1 | START |  |  | 8-Bi | it dat |  |  |  | P | STOP | STOP |
| 1 | 0 | 0 | 0 | START |  |  | it da |  |  |  | STOP |  |  |  |
| 1 | 0 | 0 | 1 | START |  |  | Bit da |  |  |  | STOP | STOP |  |  |
| 1 | 1 | 0 | 0 | START |  |  | Bit dat |  |  |  | P | STOP |  |  |
| 1 | 1 | 0 | 1 | START |  |  | it da |  |  |  | P | STOP | STOP |  |
| 0 | - | 1 | 0 | START |  |  | 8-Bi | it dat |  |  |  | MPB | STOP |  |
| 0 | - | 1 | 1 | START |  |  | 8-Bi | it dat |  |  |  | MPB | STOP | STOP |
| 1 | - | 1 | 0 | START |  |  | Bit da |  |  |  | MPB | STOP |  |  |
| 1 | - | 1 | 1 | START |  |  | it da |  |  |  | MPB | STOP | STOP |  |

-: Don't care bits.
Note: START: Start bit
STOP: Stop bit
$P$ : Parity bit
MPB: Multiprocessor bit

Clock: An internal clock generated by the on-chip baud rate generator or an external clock input from the SCK pin can be selected as the SCI transmit/receive clock. The clock source is selected by the C/A bit in the serial mode register (SMR) and bits CKE1 and CKE0 in the serial control register (SCR) (table 14.9).

When an external clock is input at the SCK pin, it must have a frequency equal to 16 times the desired bit rate.

When the SCI operates on an internal clock, it can output a clock signal at the SCK pin. The frequency of this output clock is equal to the bit rate. The phase is aligned as in figure 14.3 so that the rising edge of the clock occurs at the center of each transmit data bit.


Figure 14.3 Output Clock and Communication Data Phase Relationship (Asynchronous Mode)

SCI Initialization (Asynchronous Mode): Before transmitting or receiving, clear the TE and RE bits to 0 in the serial control register (SCR), then initialize the SCI as follows.

When changing the operation mode or communication format, always clear the TE and RE bits to 0 before following the procedure given below. Clearing TE to 0 sets TDRE to 1 and initializes the transmit shift register (TSR). Clearing RE to 0 , however, does not initialize the RDRF, PER, FER, and ORER flags and receive data register (RDR), which retain their previous contents.

When an external clock is used, the clock should not be stopped during initialization or subsequent operation. SCI operation becomes unreliable if the clock is stopped.

Figure 14.4 is a sample flowchart for initializing the SCI. The procedure is as follows (the steps correspond to the numbers in the flowchart):

1. Select the clock source in the serial control register (SCR). Leave RIE, TIE, TEIE, MPIE, TE, and RE cleared to 0 . If clock output is selected in asynchronous mode, clock output starts immediately after the setting is made to SCR.
2. Select the communication format in the serial mode register (SMR).
3. Write the value corresponding to the bit rate in the bit rate register (BRR) unless an external clock is used.
4. Wait for at least the interval required to transmit or receive one bit, then set TE or RE in the serial control register (SCR) to 1. Also set RIE, TIE, TEIE, and MPIE as necessary. Setting TE or RE enables the SCI to use the TxD or RxD pin. The initial states are the marking transmit state, and the idle receive state (waiting for a start bit).


Figure 14.4 Sample Flowchart for SCI Initialization
Transmitting Serial Data (Asynchronous Mode): Figure 14.5 shows a sample flowchart for transmitting serial data. The procedure is as follows (the steps correspond to the numbers in the flowchart):

1. SCI initialization: Set the TxD pin using the PFC.
2. SCI status check and transmit data write: Read the serial status register (SSR), check that the TDRE bit is 1 , then write transmit data in the transmit data register (TDR) and clear TDRE to 0 .
3. Continue transmitting serial data: Read the TDRE bit to check whether it is safe to write (if it reads 1); if so, write data in TDR, then clear TDRE to 0 . When the DMAC or the DTC is started by a transmit-data-empty interrupt request (TxI) in order to write data in TDR, the TDRE bit is checked and cleared automatically.
4. To output a break at the end of serial transmission, first clear the port data register (DR) to 0 , then clear the TE to 0 in SCR and use the PFC to establish the TxD pin as an output port.


Figure 14.5 Sample Flowchart for Transmitting Serial Data

In transmitting serial data, the SCI operates as follows:

1. The SCI monitors the TDRE bit in the SSR. When TDRE is cleared to 0 , the SCI recognizes that the transmit data register (TDR) contains new data, and loads this data from the TDR into the transmit shift register (TSR).
2. After loading the data from the TDR into the TSR, the SCI sets the TDRE bit to 1 and starts transmitting. If the transmit-data-empty interrupt enable bit (TIE) is set to 1 in the SCR, the SCI requests a transmit-data-empty interrupt (TxI) at this time.
Serial transmit data is transmitted in the following order from the TxD pin:
a. Start bit: one 0 bit is output.
b. Transmit data: seven or eight bits of data are output, LSB first.
c. Parity bit or multiprocessor bit: one parity bit (even or odd parity) or one multiprocessor bit is output. Formats in which neither a parity bit nor a multiprocessor bit is output can also be selected.
d. Stop bit: one or two 1 bits (stop bits) are output.
e. Marking: output of 1 bits continues until the start bit of the next transmit data.
3. The SCI checks the TDRE bit when it outputs the stop bit. If TDRE is 0 , the SCI loads new data from the TDR into the TSR, outputs the stop bit, then begins serial transmission of the next frame. If TDRE is 1 , the SCI sets the TEND bit to 1 in the SSR, outputs the stop bit, then continues output of 1 bits (marking). If the transmit-end interrupt enable bit (TEIE) in the SCR is set to 1 , a transmit-end interrupt (TEI) is requested.

Figure 14.6 shows an example of SCI transmit operation in the asynchronous mode.


Figure 14.6 SCI Transmit Operation in Asynchronous Mode
Receiving Serial Data (Asynchronous Mode): Figures 14.7 and 14.8 show a sample flowchart for receiving serial data. The procedure is as follows (the steps correspond to the numbers in the flowchart).

1. SCI initialization: Set the RxD pin using the PFC
2. Receive error handling and break detection: If a receive error occurs, read the ORER, PER, and FER bits of the SSR to identify the error. After executing the necessary error handling, clear ORER, PER, and FER all to 0 . Receiving cannot resume if ORER, PER, or FER remain set to 1 . When a framing error occurs, the RxD pin can be read to detect the break state.
3. SCI status check and receive-data read: Read the serial status register (SSR), check that RDRF is set to 1 , then read receive data from the receive data register (RDR) and clear RDRF to 0 . The RxI interrupt can also be used to determine if the RDRF bit has changed from 0 to 1 .
4. Continue receiving serial data: Read the RDR and RDRF bit and clear RDRF to 0 before the stop bit of the current frame is received. If the DMAC or the DTC is started by a receive-datafull interrupt ( RxI ) to read RDR, the RDRF bit is cleared automatically so this step is unnecessary.


Figure 14.7 Sample Flowchart for Receiving Serial Data (1)


Figure 14.8 Sample Flowchart for Receiving Serial Data (2)

In receiving, the SCI operates as follows:

1. The SCI monitors the communication line. When it detects a start bit (0), the SCI synchronizes internally and starts receiving.
2. Receive data is shifted into the RSR in order from the LSB to the MSB.
3. The parity bit and stop bit are received. After receiving these bits, the SCI makes the following checks:
a. Parity check. The number of 1 s in the receive data must match the even or odd parity setting of the O/E bit in the SMR.
b. Stop bit check. The stop bit value must be 1 . If there are two stop bits, only the first stop bit is checked.
c. Status check. RDRF must be 0 so that receive data can be loaded from the RSR into the RDR.

If the data passes these checks, the SCI sets RDRF to 1 and stores the received data in the RDR. If one of the checks fails (receive error), the SCI operates as indicated in table 14.11.
Note: When a receive error occurs, further receiving is disabled. While receiving, the RDRF bit is not set to 1 , so be sure to clear the error flags.
4. After setting RDRF to 1 , if the receive-data-full interrupt enable bit (RIE) is set to 1 in the SCR, the SCI requests a receive-data-full interrupt (RxI). If one of the error flags (ORER, PER, or FER) is set to 1 and the receive-data-full interrupt enable bit (RIE) in the SCR is also set to 1 , the SCI requests a receive-error interrupt (ERI).

Figure 14.9 shows an example of SCI receive operation in the asynchronous mode.
Table 14.11 Receive Error Conditions and SCI Operation

| Receive Error | Abbreviation | Condition | Data Transfer |
| :--- | :--- | :--- | :--- |
| Overrun error | ORER | Receiving of next data ends while <br> RDRF is still set to 1 in SSR | Receive data not loaded <br> from RSR into RDR |
| Framing error | FER | Stop bit is 0 | Receive data loaded from <br> RSR into RDR |
| Parity error | PER | Parity of receive data differs from <br> even/odd parity setting in SMR | Receive data loaded from <br> RSR into RDR |



Figure 14.9 SCI Receive Operation

### 14.3.3 Multiprocessor Communication

The multiprocessor communication function enables several processors to share a single serial communication line for sending and receiving data. The processors communicate in the asynchronous mode using a format with an additional multiprocessor bit (multiprocessor format).

In multiprocessor communication, each receiving processor is addressed by a unique ID. A serial communication cycle consists of an ID-sending cycle that identifies the receiving processor, and a data-sending cycle. The multiprocessor bit distinguishes ID-sending cycles from data-sending cycles. The transmitting processor starts by sending the ID of the receiving processor with which it wants to communicate as data with the multiprocessor bit set to 1 . Next the transmitting processor sends transmit data with the multiprocessor bit cleared to 0 .

Receiving processors skip incoming data until they receive data with the multiprocessor bit set to 1. When they receive data with the multiprocessor bit set to 1 , receiving processors compare the data with their IDs. The receiving processor with a matching ID continues to receive further incoming data. Processors with IDs not matching the received data skip further incoming data until they again receive data with the multiprocessor bit set to 1 . Multiple processors can send and receive data in this way.

Figure 14.10 shows the example of communication among processors using the multiprocessor format.

Communication Formats: Four formats are available. Parity-bit settings are ignored when the multiprocessor format is selected. For details see table 14.8.

Clock: See the description in the asynchronous mode section.


Figure 14.10 Communication among Processors Using Multiprocessor Format
Transmitting Multiprocessor Serial Data: Figure 14.11 shows a sample flowchart for transmitting multiprocessor serial data. The procedure is as follows (the steps correspond to the numbers in the flowchart):

1. SCI initialization: Set the TxD pin using the PFC.
2. SCI status check and transmit data write: Read the serial status register (SSR), check that the TDRE bit is 1 , then write transmit data in the transmit data register (TDR). Also set MPBT (multiprocessor bit transfer) to 0 or 1 in SSR. Finally, clear TDRE to 0 .
3. Continue transmitting serial data: Read the TDRE bit to check whether it is safe to write (if it reads 1); if so, write data in TDR, then clear TDRE to 0 . When the DMAC or the DTC is started by a transmit-data-empty interrupt request (TxI) to write data in TDR, the TDRE bit is checked and cleared automatically.
4. Output a break at the end of serial transmission: Set the data register $(\mathrm{DR})$ of the port to 0 , then clear TE to 0 in SCR and set the TxD pin function as output port with the PFC.


Figure 14.11 Sample Flowchart for Transmitting Multiprocessor Serial Data

In transmitting serial data, the SCI operates as follows:

1. The SCI monitors the TDRE bit in the SSR. When TDRE is cleared to 0 the SCI recognizes that the transmit data register (TDR) contains new data, and loads this data from the TDR into the transmit shift register (TSR).
2. After loading the data from the TDR into the TSR, the SCI sets the TDRE bit to 1 and starts transmitting. If the transmit-data-empty interrupt enable bit (TIE) in the SCR is set to 1 , the SCI requests a transmit-data-empty interrupt (TxI) at this time.
Serial transmit data is transmitted in the following order from the TxD pin:
a. Start bit: one 0 bit is output.
b. Transmit data: seven or eight bits are output, LSB first.
c. Multiprocessor bit: one multiprocessor bit (MPBT value) is output.
d. Stop bit: one or two 1 bits (stop bits) are output.
e. Marking: output of 1 bits continues until the start bit of the next transmit data.
3. The SCI checks the TDRE bit when it outputs the stop bit. If TDRE is 0 , the SCI loads data from the TDR into the TSR, outputs the stop bit, then begins serial transmission of the next frame. If TDRE is 1 , the SCI sets the TEND bit in the SSR to 1 , outputs the stop bit, then continues output of 1 bits in the marking state. If the transmit-end interrupt enable bit (TEIE) in the SCR is set to 1 , a transmit-end interrupt (TEI) is requested at this time.

Figure 14.12 shows an example of SCI receive operation in the multiprocessor format.


Figure 14.12 SCI Multiprocessor Transmit Operation
Receiving Multiprocessor Serial Data: Figure 14.13 shows a sample flowchart for receiving multiprocessor serial data. The procedure for receiving multiprocessor serial data is listed below.

1. SCI initialization: Set the RxD pin using the PFC.
2. ID receive cycle: Set the MPIE bit in the serial control register (SCR) to 1 .
3. SCI status check and compare to ID reception: Read the serial status register (SSR), check that RDRF is set to 1 , then read data from the receive data register (RDR) and compare with the processor's own ID. If the ID does not match the receive data, set MPIE to 1 again and clear RDRF to 0 . If the ID matches the receive data, clear RDRF to 0 .
4. Receive error handling and break detection: If a receive error occurs, read the ORER and FER bits in SSR to identify the error. After executing the necessary error processing, clear both ORER and FER to 0 . Receiving cannot resume if ORER or FER remain set to 1 . When a framing error occurs, the RxD pin can be read to detect the break state.
5. SCI status check and data receiving: Read SSR, check that RDRF is set to 1 , then read data from the receive data register (RDR).


Figure 14.13 Sample Flowchart for Receiving Multiprocessor Serial Data


Figure 14.13 Sample Flowchart for Receiving Multiprocessor Serial Data (cont)

Figures 14.14 and 14.15 show examples of SCI receive operation using a multiprocessor format.


Figure 14.14 SCI Receive Operation (ID Does Not Match)


Figure 14.15 Example of SCI Receive Operation (ID Matches)

### 14.3.4 Clock Synchronous Operation

In the clock synchronous mode, the SCI transmits and receives data in synchronization with clock pulses. This mode is suitable for high-speed serial communication.

The SCI transmitter and receiver are independent, so full duplex communication is possible while sharing the same clock. The transmitter and receiver are also double buffered, so continuous transmitting or receiving is possible by reading or writing data while transmitting or receiving is in progress.

Figure 14.16 shows the general format in clock synchronous serial communication.


Note: * High except in continuous transmitting or receiving.
Figure 14.16 Data Format in Clock Synchronous Communication
In clock synchronous serial communication, each data bit is output on the communication line from one falling edge of the serial clock to the next. Data are guaranteed valid at the rising edge of the serial clock. In each character, the serial data bits are transmitted in order from the LSB (first) to the MSB (last). After output of the MSB, the communication line remains in the state of the MSB. In the clock synchronous mode, the SCI transmits or receives data by synchronizing with the falling edge of the synchronization clock.

Communication Format: The data length is fixed at eight bits. No parity bit or multiprocessor bit can be added.

Clock: An internal clock generated by the on-chip baud rate generator or an external clock input from the SCK pin can be selected as the SCI transmit/receive clock. The clock source is selected by the C// bit in the serial mode register (SMR) and bits CKE1 and CKE0 in the serial control register (SCR). See table 14.9.

When the SCI operates on an internal clock, it outputs the clock signal at the SCK pin. Eight clock pulses are output per transmitted or received character. When the SCI is not transmitting or receiving, the clock signal remains in the high state.

Note: An overrun error occurs only during the receive operation, and the sync clock is output until the RE bit is cleared to 0 . When you want to perform a receive operation in onecharacter units, select external clock for the clock source.

SCI Initialization (Clock Synchronous Mode): Before transmitting or receiving, software must clear the TE and RE bits to 0 in the serial control register (SCR), then initialize the SCI as follows.

When changing the mode or communication format, always clear the TE and RE bits to 0 before following the procedure given below. Clearing TE to 0 sets TDRE to 1 and initializes the transmit shift register (TSR). Clearing RE to 0 , however, does not initialize the RDRF, PER, FER, and ORER flags and receive data register (RDR), which retain their previous contents.

Figure 14.17 is a sample flowchart for initializing the SCI.

1. Select the clock source in the serial control register (SCR). Leave RIE, TIE, TEIE, MPIE, TE, and RE cleared to 0 .
2. Select the communication format in the serial mode register (SMR).
3. Write the value corresponding to the bit rate in the bit rate register (BRR) unless an external clock is used.
4. Wait for at least the interval required to transmit or receive one bit, then set TE or RE in the serial control register (SCR) to 1 . Also set RIE, TIE, TEIE, and MPIE. The TxD, RxD pins becomes usable in response to the PFC corresponding bits and the TE, RE bit settings.


Figure 14.17 Sample Flowchart for SCI Initialization

Transmitting Serial Data (Synchronous Mode): Figure 14.18 shows a sample flowchart for transmitting serial data and indicates the procedure to follow.

1. SCI initialization: Set the TxD pin function with the PFC.
2. SCI status check and transmit data write: Read SSR, check that the TDRE flag is 1 , then write transmit data in TDR and clear the TDRE flag to 0 .
3. To continue transmitting serial data: After checking that the TDRE flag is 1, indicating that data can be written, write data in TDR, then clear the TDRE flag to 0 . When the DMAC or DTC is activated by a transmit-data-empty interrupt request (TxI) to write data in TDR, the TDRE flag is checked and cleared automatically.


Figure 14.18 Sample Flowchart for Serial Transmitting

Figure 14.19 shows an example of SCI transmit operation.


Figure 14.19 Example of SCI Transmit Operation
SCI serial transmission operates as follows.

1. The SCI monitors the TDRE bit in the SSR. When TDRE is cleared to 0 the SCI recognizes that the transmit data register (TDR) contains new data and loads this data from the TDR into the transmit shift register (TSR).
2. After loading the data from the TDR into the TSR, the SCI sets the TDRE bit to 1 and starts transmitting. If the transmit-data-empty interrupt enable bit (TIE) in the SCR is set to 1 , the SCI requests a transmit-data-empty interrupt (TxI) at this time.
If clock output mode is selected, the SCI outputs eight synchronous clock pulses. If an external clock source is selected, the SCI outputs data in synchronization with the input clock. Data are output from the TxD pin in order from the LSB (bit 0) to the MSB (bit 7).
3. The SCI checks the TDRE bit when it outputs the MSB (bit 7). If TDRE is 0 , the SCI loads data from the TDR into the TSR, then begins serial transmission of the next frame. If TDRE is 1, the SCI sets the TEND bit in the SSR to 1, transmits the MSB, then holds the transmit data pin (TxD) in the MSB state. If the transmit-end interrupt enable bit (TEIE) in the SCR is set to 1 , a transmit-end interrupt (TEI) is requested at this time.
4. After the end of serial transmission, the SCK pin is held in the high state.

Receiving Serial Data (Clock Synchronous Mode): Figures 14.20 and 14.21 shows a sample flowchart for receiving serial data. When switching from the asynchronous mode to the clock synchronous mode, make sure that ORER, PER, and FER are cleared to 0. If PER or FER is set to 1 , the RDRF bit will not be set and both transmitting and receiving will be disabled.

The procedure for receiving serial data is listed below:

1. SCI initialization: Set the RxD pin using the PFC.
2. Receive error handling: If a receive error occurs, read the ORER bit in SSR to identify the error. After executing the necessary error handling, clear ORER to 0 . Transmitting/receiving cannot resume if ORER remains set to 1 .
3. SCI status check and receive data read: Read the serial status register (SSR), check that RDRF is set to 1 , then read receive data from the receive data register (RDR) and clear RDRF to 0 . The RxI interrupt can also be used to determine if the RDRF bit has changed from 0 to 1 .
4. Continue receiving serial data: Read RDR, and clear RDRF to 0 before the frame MSB (bit 7) of the current frame is received. If the DMAC or the DTC is started by a receive-data-full interrupt (RxI) to read RDR, the RDRF bit is cleared automatically so this step is unnecessary.


Figure 14.20 Sample Flowchart for Serial Receiving (1)


Figure 14.21 Sample Flowchart for Serial Receiving (2)
Figure 14.22 shows an example of the SCI receive operation.


Figure 14.22 Example of SCI Receive Operation
In receiving, the SCI operates as follows:

1. The SCI synchronizes with serial clock input or output and initializes internally.
2. Receive data is shifted into the RSR in order from the LSB to the MSB. After receiving the data, the SCI checks that RDRF is 0 so that receive data can be loaded from the RSR into the RDR. If this check passes, the SCI sets RDRF to 1 and stores the received data in the RDR. If the check does not pass (receive error), the SCI operates as indicated in table 14.11 and no further transmission or reception is possible. If the error flag is set to 1 , the RDRF bit is not set
to 1 during reception, even if the RDRF bit is 0 cleared. When restarting reception, be sure to clear the error flag.
3. After setting RDRF to 1 , if the receive-data-full interrupt enable bit (RIE) is set to 1 in the SCR, the SCI requests a receive-data-full interrupt (RxI). If the ORER bit is set to 1 and the receive-data-full interrupt enable bit (RIE) in the SCR is also set to 1 , the SCI requests a receive-error interrupt (ERI).

Transmitting and Receiving Serial Data Simultaneously (Clock Synchronous Mode): Figure 14.23 shows a sample flowchart for transmitting and receiving serial data simultaneously. The procedure is as follows (the steps correspond to the numbers in the flowchart):

1. SCI initialization: Set the TxD and RxD pins using the PFC.
2. SCI status check and transmit data write: Read the serial status register (SSR), check that the TDRE bit is 1 , then write transmit data in the transmit data register (TDR) and clear TDRE to 0 . The TxI interrupt can also be used to determine if the TDRE bit has changed from 0 to 1 .
3. Receive error handling: If a receive error occurs, read the ORER bit in SSR to identify the error. After executing the necessary error processing, clear ORER to 0 . Transmitting/receiving cannot resume if ORER remains set to 1 .
4. SCI status check and receive data read: Read the serial status register (SSR), check that RDRF is set to 1 , then read receive data from the receive data register (RDR) and clear RDRF to 0 . The RxI interrupt can also be used to determine if the RDRF bit has changed from 0 to 1 .
5. Continue transmitting and receiving serial data: Read the RDRF bit and RDR, and clear RDRF to 0 before the frame MSB (bit 7) of the current frame is received. Also read the TDRE bit to check whether it is safe to write (if it reads 1 ); if so, write data in TDR, then clear TDRE to 0 before the MSB (bit 7) of the current frame is transmitted. When the DMAC or the DTC is started by a transmit-data-empty interrupt request (TxI) to write data in TDR, the TDRE bit is checked and cleared automatically. When the DMAC or the DTC is started by a receive-datafull interrupt (RxI) to read RDR, the RDRF bit is cleared automatically.

Note: In switching from transmitting or receiving to simultaneous transmitting and receiving, simultaneously clear both TE and RE to 0 , then simultaneously set both TE and RE to 1 .


Figure 14.23 Sample Flowchart for Serial Transmission

### 14.4 SCI Interrupt Sources and the DMAC/DTC

The SCI has four interrupt sources: transmit-end (TEI), receive-error (ERI), receive-data-full (RxI), and transmit-data-empty (TxI). Table 14.12 lists the interrupt sources and indicates their priority. These interrupts can be enabled and disabled by the TIE, RIE, and TEIE bits in the serial control register (SCR). Each interrupt request is sent separately to the interrupt controller.

TxI is requested when the TDRE bit in the SSR is set to 1 . TxI can start the direct memory access controller (DMAC) or the data transfer controller (DTC) to transfer data. TDRE is automatically cleared to 0 when the DMAC or the DTC writes data in the transmit data register (TDR).

RxI is requested when the RDRF bit in the SSR is set to 1 . RxI can start the DMAC or the DTC to transfer data. RDRF is automatically cleared to 0 when the DMAC or the DTC reads the receive data register (RDR).

ERI is requested when the ORER, PER, or FER bit in the SSR is set to 1 . ERI cannot start the DMAC or the DTC.

TEI is requested when the TEND bit in the SSR is set to 1 . TEI cannot start the DMAC or the DTC. Where the TxI interrupt indicates that transmit data writing is enabled, the TEI interrupt indicates that the transmit operation is complete.

Table 14.12 SCI Interrupt Sources

| Interrupt Source | Description | DMAC/DTC Activation | Priority |
| :--- | :--- | :--- | :--- |
| ERI | Receive error (ORER, PER, or FER) | No | High |
| RxI | Receive data full (RDRF) | Yes | (TDE) |
| TxI | Transmit data empty (TDRE) | Yes | Low |
| TEI | Transmit end (TEND) | No | Lown |

### 14.5 Notes on Use

Sections 14.5.1 through 14.5.9 provide information for using the SCI.

### 14.5.1 TDR Write and TDRE Flags

The TDRE bit in the serial status register (SSR) is a status flag indicating loading of transmit data from TDR into TSR. The SCI sets TDRE to 1 when it transfers data from TDR to TSR. Data can be written to TDR regardless of the TDRE bit status. If new data is written in TDR when TDRE is 0 , however, the old data stored in TDR will be lost because the data has not yet been transferred to the TSR. Before writing transmit data to the TDR, be sure to check that TDRE is set to 1 .

### 14.5.2 Simultaneous Multiple Receive Errors

Table 14.13 indicates the state of the SSR status flags when multiple receive errors occur simultaneously. When an overrun error occurs, the RSR contents cannot be transferred to the RDR, so receive data is lost.

Table 14.13 SSR Status Flags and Transfer of Receive Data

|  |  | SSR Status Flags |  | Receive Data <br> Transfer |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Receive Error Status | 1 | 1 | 0 | 0 | X |
| RDRF | ORER | FER | PER | RSR $\rightarrow$ RDR |  |
| Framing error | 0 | 0 | 1 | 0 | O |
| Parity error | 0 | 0 | 0 | 1 | O |
| Overrun error + framing error | 1 | 1 | 1 | 0 | X |
| Overrun error + parity error | 1 | 1 | 0 | 1 | X |
| Framing error + parity error | 0 | 0 | 1 | 1 | O |
| Overrun error + framing error + parity <br> error | 1 | 1 | 1 | 1 | X |
| N |  |  |  |  |  |

Notes: O = Receive data is transferred from RSR to RDR.
$X=$ Receive data is not transferred from RSR to RDR.

### 14.5.3 Break Detection and Processing

Break signals can be detected by reading the RxD pin directly when a framing error (FER) is detected. In the break state, the input from the RxD pin consists of all 0s, so FER is set and the parity error flag (PER) may also be set. In the break state, the SCI receiver continues to operate, so if the FER bit is cleared to 0 , it will be set to 1 again.

### 14.5.4 Sending a Break Signal

The TxD pin becomes a general I/O pin with the I/O direction and level determined by the I/O port data register (DR) and pin function controller (PFC) control register (CR). These conditions allow break signals to be sent. The DR value is substituted for the marking status until the PFC is set. Consequently, the output port is set to initially output a 1 . To send a break in serial transmission, first clear the DR to 0 , then establish the TxD pin as an output port using the PFC. When TE is cleared to 0 , the transmission section is initialized regardless of the present transmission status.

### 14.5.5 Receive Error Flags and Transmitter Operation (Clock Synchronous Mode Only)

When a receive error flag (ORER, PER, or FER) is set to 1 , the SCI will not start transmitting even if TDRE is set to 1 . Be sure to clear the receive error flags to 0 before starting to transmit. Note that clearing RE to 0 does not clear the receive error flags.

### 14.5.6 Receive Data Sampling Timing and Receive Margin in the Asynchronous Mode

In the asynchronous mode, the SCI operates on a base clock of 16 times the bit rate frequency. In receiving, the SCI synchronizes internally with the falling edge of the start bit, which it samples on the base clock. Receive data is latched on the rising edge of the eighth base clock pulse (figure 14.24).


Figure 14.24 Receive Data Sampling Timing in Asynchronous Mode
The receive margin in the asynchronous mode can therefore be expressed as:

$$
M=\left|\left(0.5-\frac{1}{2 N}\right)-(L-0.5) F-\frac{|D-0.5|}{N}(1+F)\right| \times 100 \%
$$

M : Receive margin (\%)
N : Ratio of clock frequency to bit rate $(\mathrm{N}=16)$
D: Clock duty cycle ( $\mathrm{D}=0-1.0$ )
L : Frame length (L = 9-12)
F: Absolute deviation of clock frequency
From the equation above, if $\mathrm{F}=0$ and $\mathrm{D}=0.5$ the receive margin is $46.875 \%$ :

$$
\begin{aligned}
D & =0.5, F=0 \\
M & =(0.5-1 /(2 \times 16)) \times 100 \% \\
& =46.875 \%
\end{aligned}
$$

This is a theoretical value. A reasonable margin to allow in system designs is 20-30\%.

### 14.5.7 Constraints on DMAC/DTC Use

- When using an external clock source for the synchronization clock, update the TDR with the DMAC or the DTC, and then after five system clocks or more elapse, input a transmit clock. If a transmit clock is input in the first four system clocks after the TDR is written, an error may occur (figure 14.25).
- Before reading the receive data register (RDR) with the DMAC/DTC, select the receive-datafull interrupt of the SCI as a start-up source.


Note: During external clock operation, an error may occur if $t$ is $4 \phi$ or less.
Figure 14.25 Example of Clock Synchronous Transmission with DMAC

### 14.5.8 Cautions for Clock Synchronous External Clock Mode

- Set $\mathrm{TE}=\mathrm{RE}=1$ only when the external clock SCK is 1 .
- Do not set $\mathrm{TE}=\mathrm{RE}=1$ until at least four clocks after the external clock SCK has changed from 0 to 1 .
- When receiving, RDRF is 1 when RE is set to zero 2.5-3.5 clocks after the rising edge of the RxD D7 bit SCK input, but it cannot be copied to RDR.


### 14.5.9 Caution for Clock Synchronous Internal Clock Mode

When receiving, RDRF is 1 when RE is set to zero 1.5 clocks after the rising edge of the RxD D7 bit SCK output, but it cannot be copied to RDR.

## Section 15 High Speed A/D Converter (Excluding A Mask)

### 15.1 Overview

The high speed A/D converter has 10-bit resolution, and can select from a maximum of eight channels of analog inputs.

### 15.1.1 Features

The high speed A/D converter has the following features:

- 10-bit resolution
- Eight input channels
- Analog conversion voltage range setting is selectable
- Using the reference voltage pin (AVref) as an analog standard voltage (Vref), conversion of analog input from 0 to Vref (only with SH7043).
- High-speed conversion
— Minimum conversion time: 2.9 s per channel (for $28-\mathrm{MHz}$ operation)
- $1.4 \mu \mathrm{~s}$ per channel during continuous conversion
- Multiple conversion modes
- Select mode/group mode
- Single mode/scan mode
- Buffered operation possible
- 2 channel simultaneous sampling possible
- Three types of conversion start
- Software, timer conversion start trigger (MTU), or ADTRG pin can be selected.
- Eight data registers
- Conversion results stored in 16-bit data registers corresponding to each channel.
- Sample and hold function
- $\mathrm{A} / \mathrm{D}$ conversion end interrupt generation
- An A/D conversion end interrupt (ADI) request can be generated on completion of A/D conversions


### 15.1.2 Block Diagram

Figure 15.1 is the block diagram of the high speed $\mathrm{A} / \mathrm{D}$ converter.


Figure 15.1 High Speed A/D Converter Block Diagram

### 15.1.3 Pin Configuration

Table 15.1 shows the input pins used by the high speed A/D converter.
The $A V_{c c}$ and $A V_{s s}$ pins are for the $A / D$ converter internal analog section power supply. The $A V_{\text {ref }} \operatorname{pin}$ is for the $A / D$ conversion standard voltage.

Table 15.1 Pin Configuration

| Pin | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Analog supply | $\mathrm{AV}_{\mathrm{CC}}$ | I | Analog section power supply |
| Analog ground | $\mathrm{AV}_{\mathrm{SS}}$ | I | Analog section ground and A/D conversion <br> reference voltage |
| Reference voltage | $\mathrm{AV}_{\text {ref }}$ | I | A/D conversion standard voltage <br> (SH7043 only) |
| Analog input 0 | AN0 | I | Analog input channel 0 |
| Analog input 1 | AN1 | I | Analog input channel 1 |
| Analog input 2 | AN2 | I | Analog input channel 2 |
| Analog input 3 | AN3 | I | Analog input channel 3 |
| Analog input 4 | AN4 | I | Analog input channel 4 |
| Analog input 5 | AN5 | I | Analog input channel 5 |
| Analog input 6 | AN6 | I | Analog input channel 6 |
| Analog input 7 | AN7 | I | Analog input channel 7 |
| A/D external trigger input | $\overline{\text { ADTRG }}$ | I | External trigger for A/D conversion start |

### 15.1.4 Register Configuration

Table 15.2 shows the configuration of the high speed A/D converter registers.
Table 15.2 Register Configuration

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A/D data register A | ADDRA | R | H'0000 | H'FFFF83F0 | 8,16 |
| $A / D$ data register $B$ | ADDRB | R | H'0000 | H'FFFF83F2 |  |
| A/D data register $C$ | ADDRC | R | H'0000 | H'FFFF83F4 |  |
| $A / D$ data register $D$ | ADDRD | R | H'0000 | H'FFFF83F6 |  |
| A/D data register $E$ | ADDRE | R | H'0000 | H'FFFF83F8 |  |
| A/D data register $F$ | ADDRF | R | H'0000 | H'FFFF83FA |  |
| A/D data register $G$ | ADDRG | R | H'0000 | H'FFFF83FC |  |
| A/D data register H | ADDRH | R | H'0000 | H'FFFF83FE |  |
| A/D control/status register | ADCSR | R/(W) |  | H'FFFF83E0 |  |
| A/D control register | ADCR | R/W | H'00 | H'FFFF83E1 |  |

Note: * Only 0 can be written to bit 7 to clear the flag.

### 15.2 Register Descriptions

### 15.2.1 A/D Data Registers A-H (ADDRA-ADDRH)

The ADDR are 16-bit read only registers for storing A/D conversion results. There are eight of these registers, ADDRA through ADDRH.

The A/D converted data is 10 -bit data which is sent to the ADDR for the corresponding converted channel for storage. The lower 8 bits of the A/D converted data are transferred to and stored in the lower byte (bits 7-0) of the ADDR, and the upper 2 bits are stored into the upper byte (bits 9,8 ). Bits 15-10 always read as 0 . Data reads can be either byte or word. The upper 8 bits of the converted data are transferred upon byte data reads. Additionally, buffered operation is possible by combining ADDRA-ADDRD.

Table 15.3 shows the correspondence between the analog input channels and the ADDR.
The ADDR are initialized to $\mathrm{H}^{\prime} 0000$ by power-on reset or in standby mode. Manual reset does not initialize ADDR.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | AD9 | AD8 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | R | R | R | R | R | R |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | R | R | R | R | R | R |

Table 15.3 Analog Input Channel and ADDR Correspondence

| Analog Input Channel | A/D Data Register |
| :--- | :--- |
| AN0 | ADDRA* |
| AN1 | ADDRB* $^{*}$ |
| AN2 | ADDRC* |
| AN3 | ADDRD* |
| AN4 | ADDRE |
| AN5 | ADDRF |
| AN6 | ADDRG |
| AN7 | ADDRH |

Note: * Except during buffer operation

### 15.2.2 A/D Control/Status Register (ADCSR)

The ADCSR is an 8-bit read/write register used for A/D conversion operation control and to indicate status.

The ADCSR is initialized to $\mathrm{H}^{\prime} 00$ by power-on reset or in standby mode. Manual reset does not initialize ADCSR.


Note: * The only value that can be written is a 0 to clear the flag.

- Bit 7—A/D End Flag (ADF): This status flag indicates that A/D conversion has ended.

| Bit 7: ADF | Description |
| :---: | :---: |
| 0 | Clear conditions (initial value) <br> - With ADF $=1$, by reading the ADF flag then writing 0 in ADF <br> - When the DTC or DMAC are activated by an ADI interrupt |
| 1 | Set conditions <br> - Single mode: When A/D conversion ends after conversion for all designated channels (during buffer operation, this is not set until operation of the specified buffer has ended) <br> - Scan mode: After one round of A/D conversion for all specified channels |

- Bit 6—A/D Interrupt Enable (ADIE): Enables or disables interrupt requests (ADI) after A/D conversion ends. Set the ADIE bit while conversion is suspended.

| Bit 6: ADIE | Description |
| :--- | :--- |
| 0 | Disables interrupt requests (ADI) after A/D conversion ends (initial <br> value) |
| 1 | Enables interrupt requests (ADI) after A/D conversion ends |

- Bit 5—A/D Start (ADST): Selects start or stop for A/D conversion. A 1 is maintained during A/D conversions.
The ADST bit can be set to 1 by software, timer conversion start triggers, or an A/D external trigger input pin ( $\overline{\text { ADTRG }}$ ).

| Bit 5: ADST | Description |
| :--- | :--- |
| 0 | A/D conversion halted (initial value) |
| 1 | Single mode: Start A/D conversion. Automatically cleared to 0 after <br> conversion for the designated channel ends. <br> Scan mode: Start A/D conversion. Continuous conversion until 0 <br> cleared by software. |

- Bit 4—Clock Select (CKS): Sets the A/D conversion time. Set, according to the operating frequency, to give a conversion time of at least $2 \mu \mathrm{~s}$ ( 5 V version) or $4 \mu \mathrm{~s}$ ( 3.3 V version). Make conversion time changes only while conversion is halted.

Bit 4: CKS
0

## Description

Conversion time $=40$ states (A/D converter standard clock $=\phi / 2$ ) (initial value)

1 Conversion time $=80$ states (when $\phi / 4$ is selected)

- Bit 3-Group Mode (GRP): Designates either select mode or group mode for the A/D conversion channel selection.
Set the GRP bit only while conversion is halted.
Bit 3: GRP Description

| 0 | Select mode (initial value) |
| :--- | :--- |
| 1 | Group mode |

- Bits 2-0—Channel Select 2-0 (CH2-CH0): These bits, along with the GRP bit, select the analog input channel.
Set the input channel only while conversion is halted.


## Description

| Bit 2: $\mathbf{C H} 2$ | Bit 1: $\mathbf{C H 1}$ | Bit 0: $\mathbf{C H}$ | Select Mode (GRP = $\mathbf{0}$ ) | Group Mode (GRP = 1) |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | AN0 (initial value) | AN0 |
| 0 | 0 | 1 | AN1 | AN0-AN1 |
| 0 | 1 | 0 | AN2 | AN0-AN2 |
| 0 | 1 | 1 | AN3 | AN0-AN3 |
| 1 | 0 | 0 | AN4 | AN0-AN4 |
| 1 | 0 | 1 | AN5 | AN0-AN5 |
| 1 | 1 | 0 | AN6 | AN0-AN6 |
| 1 | 1 | 1 | AN7 | AN0-AN7 |

## 15．2．3 A／D Control Register（ADCR）

The ADCR is an 8－bit read／write register used for A／D conversion operation control．The ADCR is initialized to $\mathrm{H}^{\prime} 00$ by power－on reset or in standby mode．Manual reset does not initialize．

| Bit： | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | PWR | TRGS1 | TRGS0 | SCAN | DSMP | BUFE1 | BUFE0 |  |
| Initial value： | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |
| R／W： | $R$ | R／W | R／W | R／W | R／W | R／W | R／W | R／W |  |

－Bit 7—Reserved：This bit always reads as 0 ．The write value should always be 0 ．
－Bit 6—Power（PWR）：Designates the conversion start mode for the high speed A／D converter． Setting the PWR bit to 1 sets high speed start mode，and a 0 sets to low power conversion mode．See section 15．4．7，Conversion Start Modes，for details on the conversion start operation．
Set the PWR bit only while conversion is halted．
Bit 6：PWR Description

| 0 | Low power conversion mode（initial value） |
| :--- | :--- |
| 1 | High speed start mode |

－Bits 5 and 4—Timer Trigger Select 1， 0 （TRGS1，TRGS0）：These bits enable or prohibit A／D conversion starts by trigger signals．
Set the TRGS1，TRGS0 bits only while conversion is halted．
Bit 5：TRGS1 Bit 4：TRGSO Description

| 0 | 0 | Enable A／D conversion start by software（initial value） |
| :--- | :--- | :--- |
| 0 | 1 | Enables A／D conversion start by MTU conversion start trigger |
| 1 | 0 | Reserved |
| 1 | 1 | Enables A／D conversion start by external trigger pin（⿳亠二口ADTRG） |

－Bit 3—Scan Mode（SCAN）：Selects either single mode or scan mode for the A／D conversion operation mode．See section 15．4，Operation，for details on single mode and scan mode operation．
Set the SCAN bit only while conversion is halted．

| Bit 3：SCAN | Description |
| :--- | :--- |
| 0 | Single mode（initial value） |
| 1 | Scan mode |

- Bit 2—Simultaneous Sampling (DSMP): Enables or disables the simultaneous sampling of two channels. See section 15.4.6, Simultaneous Sampling Operation, for details on simultaneous sampling.
Set the DSMP bit only while conversion is halted.


## Bit 2: DSMP Description

| 0 | Normal sampling operation (initial value) |
| :--- | :--- |
| 1 | Simultaneous sampling operation |

- Bits 1-0—Buffer Enable 1, 0 (BUFE1, BUFE0): These bits select whether to use the ADDRB-ADDRD as buffer registers.

Set the BUFE1 and BUFE0 bits only while conversion is halted.

| Bit 1: BUFE1 | Bit 0: BUFE0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | Normal operation (initial value) |
| 0 | 1 | ADDRA and ADDRB buffer operation: conversion result $\rightarrow$ <br> ADDRA $\rightarrow$ ADDRB (ADDRB is the buffer register) |
| 1 | 0 | ADDRA and ADDRC, also ADDRB and ADDRD buffer <br> operation: conversion result $1 \rightarrow$ ADDRA $\rightarrow$ ADDRC, <br> conversion result $2 \rightarrow$ ADDRB $\rightarrow$ ADDRD (ADDRC and <br> ADDRD are buffer registers) |
| 1 | 1 | ADDRA-ADDRD buffer operation: conversion result $\rightarrow$ <br> ADDRA $\rightarrow$ ADDRB $\rightarrow$ ADDRC $\rightarrow$ ADDRD (ADDRB-ADDRD <br> are buffer registers) |

### 15.3 Bus Master Interface

The ADDRA-ADDRH are 16-bit registers with a 16 -bit width data bus to the bus master. The bus master can read from ADDRA-ADDRH in either word or byte units.

When an ADDR is read in word units, the ADDR contents are transferred to the bus master 16 bits at a time. In byte unit reads, the contents of the most significant eight bits (AD9-AD2) of the converted data (AD9-AD0) are transferred to the bus master.

Figures 15.2 and 15.3 shows an example of the ADDR read operation.


Figure 15.2 ADDR Read Operation (1)


Figure 15.3 ADDR Read Operation (2)

### 15.4 Operation

- The high speed A/D converter has 10 -bit resolution.
- In addition to the four operating modes of select or group, and single or scan can be set in combination with buffer operation and simultaneous sampling operation.
- Select mode uses one channel and group mode selects multiple channels.
- One start in the single mode performs conversions on all selected channels, and one start in the scan mode performs repeated conversions until stopped by software.
- In buffer operation, the previous conversion result is saved in a buffer register at the end of a conversion for the relevant channel.
- In simultaneous sampling operation, the analog input voltages of two channels are sampled simultaneously then converted in order.
- Software, a timer conversion start trigger (MTU), or an ADTRG input can be selected as the conversion start condition.
- High speed start mode or low power conversion mode can be selected for A/D conversion using the PWR bit setting.
- When changing the operation mode or input channel, rewrite the ADCSR, ADCR while the ADST bit is cleared to 0 . After rewriting the ADCSR, ADCR, A/D conversion will be restarted when the ADST bit is set to 1 . Operation mode or input channel changes can be made simultaneously with ADST bit setting. When stopping an A/D conversion before completion, 0 clear the ADST bit.


### 15.4.1 Select-Single Mode

Choose select-single mode when doing A/D conversions for one channel only.
When the ADST bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion is started according to the designated conversion start conditions. The ADST bit is held to 1 during the A/D conversion and is automatically cleared to 0 upon completion.

The ADF flag is also set to 1 at the end of conversion. If the ADIE bit is set to 1 at this time, an ADI interrupt request is generated. The ADF flag is cleared by reading the ADCSR, then writing a 0.

Figure 15.4 shows an example of operation in the select-single mode when AN1 is selected.


Figure 15.4 A/D Converter Operation Example (Select-Single Mode)

### 15.4.2 Select-Scan Mode

Choose select-scan mode when doing repeated A/D conversions for one channel. This is useful when doing continuous monitoring of the analog input of one channel.

When the ADST bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion is started according to the designated conversion start conditions. The ADST bit is held to 1 until 0 cleared by software. A/D conversion for the selected input channel is repeated during that interval.

The ADF flag is set to 1 at the end of the first conversion. At this point, if the ADIE bit is set, an ADI interrupt request is issued, and the $\mathrm{A} / \mathrm{D}$ converter is halted. With the $\mathrm{A} / \mathrm{D}$ converter in stop mode due to an ADI interrupt request, conversion is restarted when the ADF flag is cleared to 0 . The ADF flag is cleared by reading the ADCSR then writing a 0 .

Figure 15.5 shows an example of operation in the select-scan mode when AN1 is selected.


Figure 15.5 A/D Converter Operation Example (Select-Scan Mode)

### 15.4.3 Group-Single Mode

Choose group-single mode when doing A/D conversions for multiple channels.
When the ADST bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion is started according to the designated conversion start conditions. The ADST bit is held to 1 during A/D conversion and is automatically cleared to 0 when all conversions for the designated input channels are completed.

The ADF flag is set to 1 when all conversions for the designated input channels are completed. If the ADIE bit is set to 1 at this time, an ADI interrupt request is generated. The ADF flag is cleared by reading the ADCSR then writing a 0 .

Figure 15.6 shows an example of operation in the group-single mode when AN0-AN2 are selected.


Figure 15.6 A/D Converter Operation Example (Group-Single Mode)

### 15.4.4 Group-Scan Mode

Choose group-scan mode when doing repeated A/D conversions for multiple channels. This is useful when doing continuous monitoring of the analog inputs of multiple channels.

When the ADST bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion is started according to the designated conversion start conditions. The ADST bit is held to 1 until 0 cleared by software. A/D conversion for the selected input channels is repeated during that interval.

The ADF flag is set to 1 at the completion of the first conversions of all the designated input channels. At this point, if the ADIE bit is set to 1 , an ADI interrupt request is issued, and the A/D converter is temporarily halted. With the A/D converter in stop mode due to an ADI interrupt request, conversion is restarted when the ADF flag is cleared to 0 . The ADF flag is cleared by reading the ADCSR, then writing a 0 .

## Renesns

Figure 15.7 shows an example of operation in the group-scan mode when AN0-AN2 are selected.


Figure 15.7 A/D Converter Operation Example (Group-Scan Mode)

### 15.4.5 Buffer Operation

When conversion ends on the relevant channel, the conversion result is stored in the ADDR, and simultaneously, the previously stored result is transferred to another ADDR. Buffer operation can be selected from the following:

- AN0 $\rightarrow$ ADDRA $\rightarrow$ ADDRB (Two-stage, one-group operation)
- AN0 $\rightarrow$ ADDRA $\rightarrow$ ADDRC, AN1 $\rightarrow$ ADDRB $\rightarrow$ ADDRD (Two-stage, two-group operation)
- AN0 $\rightarrow$ ADDRA $\rightarrow$ ADDRB $\rightarrow$ ADDRC $\rightarrow$ ADDRD (Four-stage, one-group operation)

To use in combination with simultaneous sampling, set GRP $=1$, $\mathrm{BUFE} 1, \mathrm{BUFE} 0=\mathrm{B}^{\prime} 10$, and $\mathrm{CH} 2=0$. Buffer operation timing is shown in figure 15.8.


Figure 15.8 Buffer Operation Example (Select Scan Mode: Two-Stage One-Group Operation, When CH2-CH0 = B'001)

Buffer-Only Operation: When performing conversion only on the analog input channels specified by the BUFE1 and BUFE0 bits, select group mode, and you can select the ADF flag setting conditions with the $\mathrm{CH} 2-\mathrm{CH} 0$ bits.

Table 15.4 shows conversion during buffer operation and ADF flag setting conditions. The ADF flag is set at the point in the table when the final conversion has ended. In single mode, conversion is halted after the ADF flag is set to 1 . In scan mode, conversion continues, and the converted data is stored in sequence in the buffer registers specified by the BUFE1 and BUFE0 bits.

## renesns

When the ADF flag is set to 1 , if the ADIE bit is also set to 1 , an ADI interrupt is issued. After the ADCSR is read, the ADF flag is cleared by a 0 write.

With select single mode, the A/D converter goes into standby mode at the end of every conversion cycle. The A/D converter is restarted by software, a timer trigger, or external trigger. When the number of conversion cycles shown in table 15.4 have ended, the ADF flag is set to 1 .

Table 15.4 Conversion Channel and ADF Flag Setting/Clearing Conditions during Buffer Operation 1

| Channel Setting |  |  | Sampling Channel |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CH2 | CH1 | CHO | BUFE1, BUFE0 = B'01 | BUFE1, BUFE0 = B'10 | BUFE1, BUFE0 = B'11 |
| 0 | 0 | 0 | AN0 1 time (ADDRA) | AN0, AN1 1 time (ADDRB) | AN0 1 time (ADDRA) |
|  |  | 1 | AN0 2 times (ADDRB) |  | AN0 2 times (ADDRB) |
|  | 1 | 0 | * | ANO, AN1 2 times (ADDRD) | AN0 3 times (ADDRC) |
|  |  | 1 | * |  | AN0 4 times (ADDRD) |
| 1 | - | - | * | * | * |

Combined Group Mode and Buffer Operation: Continuous conversion is possible on analog input channels (AN0 and AN1) specified by bits BUFE1 and BUFE0 as well as AN4-AN7 due to setting of bits CH2-CH0.

Table 15.5 shows conversion during buffer operation and ADF flag setting conditions. The ADF flag is set at the point in the table when the final conversion has ended. In this case, conversion is performed on the analog input corresponding with the ADDR specified in the buffer register. For example, when BUFE1 and BUFE0 $=\mathrm{B}^{\prime} 11$ and $\mathrm{CH} 2-\mathrm{CH} 0=\mathrm{B}^{\prime} 110$, conversion results are stored in ADDRA and ADDRE-ADDRG. Also, contents of ADDRA-ADDRC before the start of conversion are transferred to ADDRB-ADDRD.

In single mode, conversion is halted after the ADF flag has been set to 1 . Conversion continues in scan mode.

Table 15.5 Conversion Channel and ADF Flag Setting/Clearing Conditions During Buffer Operation 2

| Channel Setting |  |  | Sampling Channel |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CH2 | CH1 | CHO | BUFE1, BUFE0 = B'01 | BUFE1, BUFE0 = B'10 | BUFE1, BUFE0 = B'11 |
| 0 | 0 | - | * | * | * |
|  | 1 | 0 | ANO, AN2 (ADDRC) |  |  |
|  |  | 1 | AN0, AN2, AN3 (ADDRD) |  |  |
| 1 | 0 | 0 | AN0, AN2-AN4 (ADDRE) | ANO, AN1, AN4 (ADDRE) | AN0, AN4 (ADDRE) |
|  |  | 1 | AN0, AN2-AN5 (ADDRF) | AN0, AN1, AN4, AN5 (ADDRF) | AN0, AN4, AN5 (ADDRF) |
|  | 1 | 0 | $\begin{aligned} & \text { ANO, AN2-AN6 } \\ & \text { (ADDRG) } \end{aligned}$ | AN0, AN1, AN4-AN6 (ADDRG) | AN0, AN4-AN6 (ADDRG) |
|  |  | 1 | AN0, AN2-AN7 (ADDRH) | AN0, AN1, AN4-AN7 (ADDRH) | $\begin{aligned} & \text { ANO, AN4-AN7 } \\ & \text { (ADDRH) } \end{aligned}$ |

Note: * See table 15.4.

ADF Flag Clearing: When the DTC and DMAC are started up due to an A/D conversion end interrupt, the ADF flag is cleared when the ADDR specified in table 15.4 or 15.5 has been read.

Resetting the Number of Buffer Operations: Clear the BUFE1 and BUFE0 bits to B'00 in conversion standby mode or when the converter has been halted. The number of buffer operations is cleared to 0 .

Updating Buffer Operations: Clear the BUFE1 and BUFE0 bits to B'00 in conversion standby mode or when the converter has been halted. Thereafter, set BUFE1 and BUFE0, and the buffer operations shown in tables 15.4 and 15.5 are performed when conversion is resumed.

### 15.4.6 Simultaneous Sampling Operation

With simultaneous sampling, continuous conversion is conducted with sampling of the input voltages on two channels at the same time. Simultaneous sampling is valid in group mode. Channels for sampling are determined by the CH 2 and CH 1 bits of the RDSCR. The combinations are shown in table 15.6. For example, if GRP $=1$ when CH 2 and $\mathrm{CH} 1=\mathrm{B}^{\prime} 11$, sampling occurs in order in the following pairs: AN0, AN1 $\rightarrow$ AN2, AN3 $\rightarrow$ AN4, AN5 $\rightarrow$ AN6, AN7. Sampling timing is shown in figure 15.9.

## Renesns

Table 15.6 Simultaneous Sampling Channels

| Channel Setting |  |  |
| :--- | :--- | :--- |
| CH2 | CH1 | Sampling Channels, GRP $\mathbf{1}$ |
| 0 | 0 | AN0, AN1 |
| 1 | 0 | AN0, AN1 $\rightarrow$ AN2, AN3 |
| 1 | 1 | AN0, AN1 $\rightarrow$ AN2, AN3 $\rightarrow$ AN4, AN5 |



Figure 15.9 Simultaneous Sampling Operation (Group Single Mode)

### 15.4.7 Conversion Start Modes

The conversion start mode of the high speed A/D converter is set by the PWR bit of the ADCSR. When the PWR bit is cleared to 0 , low-power conversion mode is set and the internal analog circuit becomes inactive. High-speed start mode is set by setting the PWR bit to 1, and the analog circuit becomes active.

In the low-power conversion mode, power is applied to the analog circuitry simultaneous to the conversion start (ADST set). When 200 cycles of the reference clock have elapsed, conversion becomes possible for the analog circuit and the first A/D conversion begins. When performing consecutive conversions, the second and later conversions are executed in 10 cycles. Select the basic clock with the CKS bit of the ADCSR. When the A/D conversion ends, ADST is cleared to 0 and the analog circuit power supply is automatically cut off. Because the analog circuit is only active during the $\mathrm{A} / \mathrm{D}$ conversion operation period in this mode, current consumption can be reduced.

In high-speed start mode, ADST is cleared to 0 when $\mathrm{A} / \mathrm{D}$ conversion ends. Power continues to be supplied to the analog circuitry, and conversion-ready status is maintained. Conversion is restarted immediately by resetting ADST to 1 . However, the first conversion after power-on begins 200 cycles after setting ADST. Clear the PWR bit to 0 to switch off the analog power supply. When performing consecutive conversions, the second and later conversions are executed in 20 cycles. Because the analog circuit is always active in this mode, A/D conversion can be executed at high speed.

Figures 15.10 and 15.11 show examples of conversion start operation timing.


Figure 15.10 Conversion Start Operation (Low-Power Conversion Mode)


Figure 15.11 Conversion Start Operation (High-Speed Start Mode)

### 15.4.8 Conversion Start by External Input

A/D conversions can be started by trigger signals generated by timer conversion start triggers or $\overline{\text { ADTRG }}$ inputs. When a trigger signal designated by the TRGS1 and TRGS0 bits of the ADCR occurs, the ADST bit of the ADCSR is set to 1 and A/D conversion is started.

The other operations are the same as when the ADST bit is set to 1 by software. Figure 15.12 shows an example of the timing when the ADST bit is set by an external input.


Figure 15.12 Conversion Start by $\overline{\text { ADTRG }}$ Conversion Start Trigger

### 15.4.9 A/D Conversion Time

The high speed A/D converter has an on-chip sample and hold circuit. The high speed A/D converter samples the input at time $\mathrm{t}_{\mathrm{D}}$ after the ADST bit is set to 1 , and then starts the conversion.

The A/D conversion time $t_{\text {Conv }}$ is the sum of the conversion start delay time $t_{\mathrm{D}}$, the input sampling time $\mathrm{t}_{\mathrm{SPL}}$, and the operating time $\mathrm{t}_{\mathrm{cP}}$. This conversion time is not a set value, but is decided by the $t_{D}$ ADCSR write timing, or the timer conversion start trigger generation timing.

Figure 15.13 shows an example of A/D conversion timing. Table 15.7 lists A/D conversion times.


Figure 15.13 A/D Conversion Timing

## Table 15.7 A/D Conversion Times

| Time | Symbol | CKS $=0$ |  |  | CKS $=1$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max | Min | Typ | Max |
| A/D conversion start delay time | $\mathrm{t}_{\mathrm{D}}$ | 1.5 | 1.5 | 1.5 | 1.5 | 1.5 | 1.5 |
| Input sampling time | $\mathrm{t}_{\text {SPL }}$ | 20 | 20 | 20 | 40 | 40 | 40 |
| A/D conversion time | $\mathrm{t}_{\text {conv }}$ | 42.5 | 42.5 | 42.5 | 82.5 | 82.5 | 82.5 |

Notes: 1. Unit: states
2. Table entries are for when $\operatorname{ADST}=1$. If 200 states have not elapsed since the PWR bit has been set, no conversions are done until after those 200 states have occurred. When PWR $=0$, add 200 states to the first A/D conversion start delay time. When continuously executing conversion, tcp for the second time and following is 20 cycle when CKS=0 and 40 cycle when CKS $=1$.

The CKS bit of the ADCSR is the operation time $t_{\text {CoNv }}$, but set so that this is $2 \mu \mathrm{~s}$ or greater. Table 15.8 shows the operating frequency and CKS bit settings.

Table 15.8 Operating Frequency and CKS Bit Settings

|  | Conversion Time <br> (States) | Minimum Conversion Time ( $\boldsymbol{\mu s}$ ) |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| CKS | $\mathbf{2 8} \mathbf{~ M H z}$ | $\mathbf{2 0} \mathbf{~ M H z}$ | $\mathbf{1 6} \mathbf{~ M H z}$ | $\mathbf{1 0} \mathbf{~ M H z}$ | $\mathbf{8} \mathbf{~ M H z}$ |  |
| 0 | 42.5 | - | 2.1 | 2.6 | 4.3 | 5.3 |
| 1 | 82.5 | 2.9 | 4.2 | 5.0 | 8.3 | 10.3 |

Note: The indication "-" means the setting is not available.

### 15.5 Interrupts

The high speed $\mathrm{A} / \mathrm{D}$ converter generates an $\mathrm{A} / \mathrm{D}$ conversion end interrupt (ADI) upon completion of A/D conversions. The ADI interrupt request can be enabled or disabled by the ADIE bit of the ADCSR.

The DTC or DMAC can be activated by ADI interrupts. When converted data is read by the DTC or DMAC upon an ADI interrupt, consecutive conversions can be done without software responsibility.

Table 15.9 lists the high speed A/D converter interrupt sources.
During scan mode, if the ADIE bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion is temporarily suspended immediately when the ADF flag is set to 1 . A/D conversion is restarted when the ADF flag is cleared to 0 .

When the DTC or DMAC are activated by an ADI interrupt, the ADF flag is cleared to 0 when the final specified data register is read.

Table 15.9 High Speed A/D Converter Interrupt Sources
Interrupt Source

## Description

DTC, DMAC Activation
ADI Interrupt caused by conversion end Possible

### 15.6 Notes on Use

Take note of the following for the A/D converter.

1. Analog input voltage range

During A/D conversions, see that the voltage applied to the analog input pins AN0-AN7 is within the range Avss $\leq$ AN0-AN7 $\leq$ AVcc.
2. $A V c c$ and $A V s s$ input voltages

The AVcc and AVss input voltage must be $\mathrm{AVcc}=\mathrm{Vcc} \pm 10 \%$, $\mathrm{AVss}=\mathrm{Vss}$. When not using the A/D converter, use AVcc $=$ Vcc, AVss $=$ Vss. During the standby mode, use $\mathrm{V}_{\text {RAM }} \leq \mathrm{Avcc}$ $\leq 5.5 \mathrm{~V}$, AVss $=$ Vss. $\mathrm{V}_{\text {RAM }}$ is the RAM standby voltage.
3. AVref input voltage

The analog standard voltage $A V r e f\left(\mathrm{AV}_{\text {ref }}\right)$ must be $A v r e f \leq A V c c$. When not using the $A / D$ converter, use $A V_{\text {ref }}=V c c$. During the standby mode, use $V_{\text {RAM }} \leq A V r e f \leq A V c c . V_{\text {RAM }}$ is the RAM standby voltage.
4. Input ports

The time constant for the circuit connecting to the input port must be shorter than the sampling time of the A/D converter. Input voltage may not be sampled sufficiently when the time constant of the circuit is long.
5. Conversion start modes

Depending on the PWR bit setting, the demand for A/D conversion will differ for the highspeed start mode and low-demand conversion mode.
6. Analog input pins handling

Connect a protection circuit as shown in figure 15.14 to prevent analog input pins (AN0-AN7) from being destroyed due to abnormal voltage from surge, etc. This circuit is also equipped with a CR filter to control errors due to noise. The circuit shown in the diagram is only an example and the number of circuits is to be determined by considering the actual condition of use.

Figure 15.15 shows an equivalent circuit of analog input pins and table 15.10 shows the specification of the analog input pins.


Notes: Numbers are only to be noted as reference value
*1

*2 Rin: Input impedance
Figure 15.14 Example of a Protection Circuit for the Analog Input Pins


Figure 15.15 Equivalent Circuit of Analog Input Pins

Table 15.10 Analog Input Pin Specification

| Item | Min | Max | Unit |
| :--- | :--- | :--- | :--- |
| Analog input capacity | - | 20 | pF |
| Permitted source impedance | - | 1 | $\mathrm{k} \Omega$ |

## Section 16 Mid-Speed A/D Converter (A Mask)

### 16.1 Overview

The mid-speed A/D converter has 10 bit resolution, and can select from a maximum of eight channels of analog input.

The mid-speed A/D converter is structured by two independent modules (A/D0 and A/D1)

### 16.1.1 Features

The mid-speed A/D converter has the following features:

- 10-bit resolution
- Eight input channels (four channels times two)
- Analog conversion voltage range setting is selectable
- Using the standard voltage pin (AVref) as an analog standard voltage (Vref), conversion of analog input from 0V to Vref (only with SH7041A, SH7043A, and SH7045).
(Connected to $\mathrm{AV}_{\mathrm{CC}}$ internally in the SH7040A, SH7042A, and SH7044.)
- High speed conversion
- Minimum conversion time: per channel
- Operation frequency: $\mathrm{f} \leq 20 \mathrm{MHz}, \mathrm{CKS}=0,1$ $6.7 \mu \mathrm{~s}(20 \mathrm{MHz}, \mathrm{CKS}=1)$
- Operation frequency: f>20MHz, CKS=0 $9.3 \mu \mathrm{~s}$ ( $28.7 \mathrm{MHz}, \mathrm{CKS}=0$ )
- Multiple conversion modes
- Single mode/scan mode
- 2 channel simultaneous conversion
- Three types of conversion start
- Software, timer conversion start trigger (MTU), or ADTRG pin can be selected.
- Eight data registers
- Conversion results stored in 16-bit data registers corresponding to each channel.
- Sample and hold function
- A/D conversion end interrupt generation
- An A/D conversion end interrupt (ADI) can be generated on completion of A/D conversion.
- Furthermore, ADI0 (A/D0 interrupt request) can activate DTC and ADI1 (A/D1 interrupt request) can activate DMAC.


### 16.1.2 Block Diagram

Figure 16.1 is the block diagram of the mid-speed $\mathrm{A} / \mathrm{D}$ converter.
$\mathrm{AV}_{\mathrm{CC}}, \mathrm{AVref}$ and $\mathrm{AV}_{\mathrm{SS}}$ pins of both $\mathrm{A} / \mathrm{D}$ are common in LSI.


Figure 16.1 Mid-Speed A/D Converter Block Diagram

### 16.1.3 Pin Configuration

Table 16.1 shows the input pins used with the mid-speed A/D converter.
The $A V_{C C}$ and $A V_{S S}$ pins are for the mid-speed $A / D$ converter internal analog section power supply. AVref pin is the A/D conversion standard voltage.

Table 16.1 Pin Configuration

| Pin | Abbreviation | 1/0 | Function |
| :---: | :---: | :---: | :---: |
| Analog supply | $\mathrm{Av}_{\mathrm{cc}}$ | 1 | Analog section power supply |
| Analog ground | $\mathrm{AV}_{\text {ss }}$ | 1 | Analog section ground and $A / D$ conversion standard voltage |
| Standard voltage | AVref* | I | A/D conversion standard voltage (SH7041A, SH7043A, and SH7045 only) |
| A/D0 Analog input 0 | AND | I | Analog input channel 0 |
| Analog input 1 | AN1 | 1 | Analog input channel 1 |
| Analog input 2 | AN2 | I | Analog input channel 2 |
| Analog input 3 | AN3 | 1 | Analog input channel 3 |
| A/D1 Analog input 4 | AN4 | I | Analog input channel 4 |
| Analog input 5 | AN5 | I | Analog input channel 5 |
| Analog input 6 | AN6 | 1 | Analog input channel 6 |
| Analog input 7 | AN7 | I | Analog input channel 7 |
| A/D external trigger input | $\overline{\text { ADTRG }}$ | 1 | External trigger for A/D conversion start |

### 16.1.4 Register Configuration

Table 16.2 shows the register configuration of the mid-speed $A / D$ converter.
Table 16.2 Register Configuration

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A/D0 data register AH | ADDRAOH | R | H'00 | H'FFFF8400 | 8, 16 |
| A/D0 data register AL | ADDRAOL | R | H'00 | H'FFFF8401 | 8 |
| A/D0 data register BH | ADDRB0H | R | H'00 | H'FFFF8402 | 8,16 |
| A/D0 data register BL | ADDRB0L | R | H'00 | H'FFFF8403 | 8 |
| A/DO data register CH | ADDRCOH | R | H'00 | H'FFFF8404 | 8, 16 |
| A/D0 data register CL | ADDRCOL | R | H'00 | H'FFFF8405 | 8 |
| A/DO data register DH | ADDRDOH | R | H'00 | H'FFFF8406 | 8, 16 |
| A/D0 data register DL | ADDRD0L | R | H'00 | H'FFFF8407 | 8 |
| A/D0 control/status register | ADCSR0 | $\mathrm{R} / \mathrm{W})^{*}$ | H'00 | H'FFFF8410 | 8, 16 |
| A/D0 control register | ADCR0 | R/W | H'7F | H'FFFF8412 | 8,16 |
| A/D1 data register AH | ADDRA1H | R | H'00 | H'FFFF8408 | 8,16 |
| A/D1 data register AL | ADDRA1L | R | H'00 | H'FFFF8409 | 8 |
| A/D1 data register BH | ADDRB1H | R | H'00 | H'FFFF840A | 8,16 |
| A/D1 data register BL | ADDRB1L | R | H'00 | H'FFFF840B | 8 |
| A/D1 data register CH | ADDRC1H | R | H'00 | H'FFFF840C | 8,16 |
| A/D1 data register CL | ADDRC1L | R | H'00 | H'FFFF840D | 8 |
| A/D1 data register DH | ADDRD1H | R | H'00 | H'FFFF840E | 8,16 |
| A/D1 data register DL | ADDRD1L | R | H'00 | H'FFFF840F | 8 |
| A/D1 control/status register | ADCSR1 | R/(W)* | H'00 | H'FFFF8411 | 8 |
| A/D1 control register | ADCR1 | R/W | H'7F | H'FFFF8413 | 8 |

Note: * Only 0 can be written to bit 7 to clear the flag.

### 16.2 Register Descriptions

### 16.2.1 A/D Data Register A-D (ADDRA0-ADDRD0, ADDRA1-ADDRD1)

$\mathrm{A} / \mathrm{D}$ registers are special registers that read stored results of $\mathrm{A} / \mathrm{D}$ conversion in 16 bits. There are eight registers: ADDRA0-ADDRD0 (A/D0) and ADDRA1-ADDRD1 (A/D1).

The A/D converted data is 10 bit data which is to the ADDR of the corresponding converted channel for storage. The upper 8 bits of the A/D converted data correspond to the upper byte of the ADDR and the lower 2 bits correspond to the lower byte. Bits 5-0 of the lower byte of ADDR are reserved and always read 0 . Analog input channels and correspondence to ADDR are shown in table 16.3.

ADDR can always be read from the CPU. The upper byte may be read directly. The lower byte is transferred through the temporary register (TEMP). For details, see section 16.3, Interface with CPU.

ADDR is initialized to $\mathrm{H}^{\prime} 0000$ during power-on reset or standby mode. ADDR will not be initialized by manual reset.

| Bit | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADDRn | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 | - | - | - | - | - | - |
| Initial value | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| $\begin{array}{r} \mathrm{R} / \mathrm{W}: \\ (\mathrm{n}=\mathrm{A} \text { to } \end{array}$ | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R | R |

Table 16.3 Analog Input Channel and ADDRA-ADDRD Correspondence

| Analog Input Channel | A/D Data Register | Module |
| :--- | :--- | :--- |
| AN0 | ADDRA0 | A/D0 |
| AN1 | ADDRB0 |  |
| AN2 | ADDRC0 |  |
| AN3 | ADDRD0 |  |
| AN4 | ADDRA1 |  |
| AN5 | ADDRB1 |  |
| AN6 | ADDRC1 |  |
| AN7 | ADDRD1 |  |

### 16.2.2 A/D Control/Status Register (ADCSR0, ADCSR1)

The A/D control/status registers (ADCSR0, 1) are registers that can read/write in 8 bits and control A/D converter operations such as mode selection. There are the ADCSR0 (A/D0) and ADCSR1 (A/D1).

The ADCSR is initialized to $\mathrm{H}^{\prime} 00$ during power-on reset or standby mode. Manual reset does not initialize ADCSR.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ADF | ADIE | ADST | SCAN | CKS | - | CH 1 | CHO |
| Initial value : | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| R/W : | $\mathrm{R} /(\mathrm{W})^{*}$ | R/W | R/W | R/W | R/W | R | R/W | R/W |

Note: * Only 0 can be written to clear the flag.

- Bit 7—A/D End Flag (ADF): Status flag that indicates end of A/D conversion.

Bit 7:
ADF Description
0 [Clear conditions]
(Initial value)

1. Writing 0 to ADF after reading ADF with $A D F=1$
2. When registers of the mid-speed converter are accessed after the DMAC and DTC are activated by ADI interrupt.
1 [Set conditions]
3. Single mode: When $A / D$ conversion is complete
4. Scan mode: When A/D conversion of all designated channels are complete

- Bit 6-A/D Interrupt Enable (ADIE): Enables or disables interrupt request (ADI) due to completion of A/D conversion.


## Bit 6:

ADIE Description

| 0 | Disables interrupt request (ADI) due to completion of A/D conversion | (Initial value) |
| :--- | :--- | :--- |
| 1 | Enables interrupt request (ADI) due to completion of A/D conversion |  |

- Bit 5—A/D Start (ADST): Selects start/end of A/D conversion. A1 is maintained during A/D conversion start. It is also possible to set a 1 by the A/D conversion trigger input pin (ADTRG).


## Bit 5:

## ADST Description

0 A/D conversion halted $\quad$ (Initial value)
1 1. Single mode: Starts A/D conversion. Automatically clears to 0 when conversion of the designated channel is complete
2. Scan mode: Starts A/D conversion. Continuous conversion until cleared to 0 by the software

- Bit 4—Scan Mode (SCAN): Selects the A/D conversion mode from single mode and scan mode. For operations during single/scan mode, see section 16.4, Operation. When switching modes, proceed while ADST $=0$.

Bit 4:
SCAN Description

| 0 | Single mode | (Initial value) |
| :--- | :--- | :--- |
| 1 | Scan mode |  |

- Bit 3-Clock Select (CKS): Sets the A/D conversion time. Proceed conversion time switch while adst=0. Always set $\mathrm{CKS}=0$ when operating frequency exceeds 20 MHz .


## Bit 3:

CKS Description

| 0 | Conversion time $=266$ states $(\max )$ | (Initial value) |
| :--- | :--- | :---: |
| 1 | Conversion time $=134$ states $(\max )$ |  |

- Bit 2—Reserved bit: Bit 2 always reads 0 . Furthermore, always write 0 .
- Bits 1, 0-Channel select 1, $0(\mathrm{CH} 1, \mathrm{CH} 0)$ : Selects the analog input channel along with the SCAN bit. Switch channels while $\operatorname{ADST}=0$.

Channel
Selection
Description

|  |  | Single mode |  |  | Scan mode |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: |
| CH1 | CH0 | A/D0 | A/D1 | A/D0 | A/D1 |  |  |
| 0 | 0 | AN0 <br> (Initial value) | AN4 <br> (Initial value) | AN0 | AN4 |  |  |
|  | 1 | AN1 | AN5 | AN0, AN1 | AN4, AN5 |  |  |
| 1 | 0 | AN2 | AN6 | AN0, AN1 | AN4-AN6 |  |  |
|  | 1 | AN3 | AN7 | AN0-AN3 | AN4-AN7 |  |  |

### 16.2.3 A/D Control Register (ADCR0, ADCR1)

A/D control registers (ADCR0,1) are registers that can read/write in 8 bits and enables or disables A/D conversion start of the external trigger input. There are the ADCR0 (A/D0) and ADCR1 (A/D1).

ADCR is initialized to $\mathrm{H}^{\prime} 7 \mathrm{~F}$ during power-on reset and standby mode. Manual reset does not initialize ADCR.

| Bit : | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Initial value : | TRGE | - | - | - | - | - | - | - |
|  | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| R/W | R/W | R | R | R | R | R | R | R |

- Bit 7—Trigger Enable (TRGE): Enables or disables A/D conversion start of input from external or MTU trigger.

Bit 7:
TRGE Description

| 0 | Disables A/D conversion start of external or MTU trigger | (Initial value) |
| :--- | :--- | ---: |
| 1 | Starts A/D conversion on last transition edge of A/D conversion trigger input pin <br> $(\overline{\text { ADTRG }})$ or MTU trigger. |  |

$\mathrm{A} / \mathrm{D} 0$ and $\mathrm{A} / \mathrm{D} 1$ are common for external trigger pin and MTU trigger.
$\mathrm{A} / \mathrm{D} 0$ and $\mathrm{A} / \mathrm{D} 1$ settings are of logical sum.

- Bits 6-0—Reserved bits: These bits always read as 1 . The write value should always be 1 .


### 16.3 Interface with CPU

Although A/D data register ADDR (ADDRA0-ADDRD0, ADDRA1-ADDRD1) are 16-bit registers, the bus width within the chip that integrates with the CPU is 8 -bits. So, upper and lower data of the ADDR must be read separately.

To avoid change in data while reading the upper/lower 2 bytes of ADDR, the lower byte data is read through the temporary register (TEMP). The upper byte data can be read directly.

The procedure for reading data from ADDR is as follows: First, read the upper byte data from ADDR. At this time, the upper byte data is read directly into the CPU and the lower byte data is transferred to TEMP of the mid-speed A/D converter. Next, read the lower byte to read the TEMP contents into the CPU.

When reading the ADDR in byte size, read the upper byte before the lower byte. Furthermore, it is possible to read only the upper byte, however, please note that contents are not guaranteed when reading only the lower byte. In addition, when reading ADDR in word size, upper byte is automatically read before the lower byte.

Figure 16.2 shows the data flow when reading from ADDR.


Figure 16.2 ADDR Access Operation (During Reading of (H'AA40))

### 16.4 Operation

The mid-speed converter operates using the continuous comparison method and is equipped with 10 -bit resolution. Operations for the single and scan modes are explained below.

### 16.4.1 Single Mode (SCAN=0)

The single mode is selected when executing A/D conversion for one channel only. A/D conversion is initiated when the ADST bit of the $\mathrm{A} / \mathrm{D}$ control/status register is set to 1 by the software or external trigger input. The ADST bit is held to 1 during the $\mathrm{A} / \mathrm{D}$ conversion and is automatically cleared to 0 upon completion.

When conversion is complete, the ADF bit of ADCSR is set to 1 . At this time, if the ADIE bit of ADCSR is $1, \mathrm{ADI}$ interrupt request occurs.

The ADF bit can be cleared by writing 0 after reading ADF=1.
To switch modes or analog input channels during A/D conversion, clear the ADST bit to 0 and stop A/D conversion to avoid malfunction. After switching (mode/channel change and ADST bit setting can be made at the same time), set the ADST bit to 1 to restart A/D conversion.

An example of operation when channel 1 (AN1) is selected in the single mode is shown in figure 16.3 (the bit specification in the example is the ADCSR0 register).

1. Set operation mode to single mode (SCAN=0), input channel to $\mathrm{AN} 1(\mathrm{CH} 1=0, \mathrm{CH} 0=1)$ and $\mathrm{A} / \mathrm{D}$ interrupt request to enable (ADIE) then start A/D conversion (ADST=1).
2. When $\mathrm{A} / \mathrm{D}$ conversion is complete, $\mathrm{A} / \mathrm{D}$ conversion result is transferred to ADDRB0. At the same time, $\mathrm{ADF}=1$ will become $\mathrm{ADF}=0$ and the mid-speed converter will standby for conversion.
3. Since $\mathrm{ADF}=1$ and $\mathrm{ADIE}=1, \mathrm{ADI}$ interrupt request will occur.
4. The $\mathrm{A} / \mathrm{D}$ interrupt process routine will start.
5. After reading $\mathrm{ADF}=1$, write 0 to ADF .
6. Read the $\mathrm{A} / \mathrm{D}$ conversion result (ADDRB0) and process.
7. End A/D interrupt process routine execution. When ADST bit is set to $1, \mathrm{~A} / \mathrm{D}$ conversion starts, following steps (2) to (7) above.


Figure 16.3 Operation Example of Mid-speed A/D Converter (Single Mode, Channel 1 Selected)

### 16.4.2 Scan Mode (SCAN=1)

The scan mode is optimal for monitoring analog input of multiple channels (including channel 1). A/D conversion is started from channel 1 (AN0 for A/D0 and AN4 for A/D1) of the group when the ADST bit of the A/D control/status register (ADCSR) is set to 1 by the software or external trigger input.

When multiple channels are selected, A/D conversion of channel 2 (AN1 or AN5) is initiated immediately after completion of the channel 1 conversion.

To switch modes or analog input channels during A/D conversion, clear the ADST bit to 0 and stop A/D conversion to avoid malfunction. After switching (mode/channel change and ADST bit setting can be made at the same time), set ADST bit to 1 to restart A/D conversion from channel 1.

An example of operation when three channels of A/D0 (AN0-2) are selected for A/D conversion is shown in figure 16.4 (the bit specification in the example is the ADCSR0 register).

1. Set operation mode to scan mode (SCAN=1), set analog channels to AN0-2 $(\mathrm{CH} 1=1, \mathrm{CH} 0=0)$ then start A/D conversion (ADST=1).
2. When $\mathrm{A} / \mathrm{D}$ conversion for channel 1 is complete, $\mathrm{A} / \mathrm{D}$ conversion result is transferred to ADDRA0.
Next, channel 2 (AN1) will automatically be selected and conversion will begin.
3. In the same manner, channel 3 will be converted (AN2).
4. When conversion of all of the selected channels (AN0-AN2) are complete, ADF will become 1 and channel 1 (ANO) will again be selected and conversion will begin.
At this time, if the ADIE bit is set to 1, ADI interrupt request will occur after completing A/D conversion.
5. Steps (2) to (4) will be repeated while ADST bit is set to 1 .

A/D conversion will stop when setting the ADST bit to 0 . When setting the ADST bit to 1 , A/D conversion will start again from channel 1 (AN0).


Figure 16.4 Operation Example of Mid-speed A/D Converter (Scan Mode, Three Channels Selected) (AN0-AN2)

### 16.4.3 Input Sampling and A/D Conversion Time

The mid-speed $A / D$ converter is equipped with a sample and hold circuit. The mid-speed $A / D$ converter samples input after $t_{D}$ hours has elapsed since setting the ADST bit of the A/D control/status register (ADCSR) to 1 , then begins conversion. The A/D conversion timing is shown in table 16.4.

The A/D conversion time, as shown in figure 16.5, includes both $t_{D}$ and input sampling time. Here, $\mathrm{t}_{\mathrm{D}}$ is determined by the write timing to ADCSR and is not constant. Thus the conversion time changes in the range shown in table 16.4.

The conversion time shown in table 16.4 is the time for the first conversion. For the second conversion and after, the time will be 256 state (fixed) for $\mathrm{CKS}=0$ and 128 state (fixed) for CKS=1.


Figure 16.5 A/D Conversion Timing

Table 16.4 A/D Conversion Time (Single Mode)

| Notation |  | CKS=0 |  |  | CKS=1 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max | Min | Typ | Max |
| A/D conversion start | $t_{\text {D }}$ | 10 | - | 17 | 6 | - | 9 |
| Input sampling time | $\mathrm{t}_{\text {SPL }}$ | - | 64 | - | - | 32 | - |
| A/D conversion time | $\mathrm{t}_{\text {canv }}$ | 259 | - | 266 | 131 | - | 134 |

Note: Numbers in the table are in states ( $\mathrm{t}_{\text {cyc }}$ ).

### 16.4.4 External Trigger Input Timing

It is possible to start $\mathrm{A} / \mathrm{D}$ conversion from an external trigger input. External trigger input is input from the $\overline{\text { ADTRG }}$ pin or MTU when the TRGE bit of the A/D control register (ADCR) is set to 1 .

A/D conversion is started when the ADST bit of the A/D control/status register (ADCSR) is set to 1 by the $\overline{\text { ADTRG }}$ input pin last transition edge or MTU trigger. Other operations, regardless of whether in the single or scan mode, are the same as when setting the ADST bit to 1 with the software.

Figure 16.6 shows an example of external trigger input timing.


Figure 16.6 External Trigger Input Timing

### 16.5 Interrupt and DMA, DTC Transfer Requests

The mid-speed $A / D$ converter generates $A / D$ conversion complete interrupt when completing $A / D$ conversion.

The ADI interrupt request can be enabled or disabled by the ADIE bit of ADCSR. It is also possible to activate DMA or DTC transfer by the ADI interrupt request. It is possible to activate DTC with ADI0 interrupt of A/D0 and activate DMAC with ADI1 interrupt of A/D1. Table 16.5 shows the interrupt factors of the mid-speed A/D converter.

Table 16.5 Mid-speed A/D Converter Interrupt Factors
Mid-speed
A/D converter Interrupt Factor Content DTC DMAC

| A/D0 | ADI0 | Interrupt by conversion complete $O$ | $\times$ |
| :--- | ---: | :--- | :--- |
| A/D1 | ADI1 | $\times$ | O |
| O: activation enabled |  |  |  |
| $\times:$ activation disabled |  |  |  |

When accessing the A/D0 register with DTC activated by ADI0 interrupt, the ADF bit of the A/D0 control/status register (ADCSR0) will automatically be cleared to 0 . Furthermore, it is possible to automatically clear the ADF bit of ADCSR1 by register access of A/D1 with activated DMAC of ADI1 interrupt. For details on the automatic clearing operation of this interrupt factor, see section 8, Data Transfer Controller (DTC).

### 16.6 A/D Conversion Precision Definitions

The medium-speed A/D converter converts analog values input from analog input channels to 10 bit digital values by comparing them with an analog reference voltage. In this operation, the absolute precision of the $\mathrm{A} / \mathrm{D}$ conversion (i.e. the deviation between the input analog value and the output digital value) includes the following kinds of error.
(1) Offset error
(2) Full-scale error
(3) Quantization error
(4) Nonlinearity error

The above four kinds of error are described below with reference to figure 16.7. For the sake of clarity, this figure shows 3-bit medium-speed A/D conversion rather than 10-bit medium-speed A/D conversion. Offset error (see figure 16.7 (1)) is the deviation between the actual A/D conversion characteristic and the ideal A/D conversion characteristic when the digital output value changes from the minimum value (zero voltage) of 0000000000 ( 000 in the figure) to 0000000001 (001 in the figure ). Full-scale error (see figure 16.7 (2)) is the deviation between the actual A/D conversion characteristic and the ideal A/D conversion characteristic when the digital output value changes from 1111111110 (110 in the figure) to the maximum value (full-scale voltage) of 111111111 (111 in the figure). Quantization error is the deviation inherent in the medium-speed A/D converter, given by $1 / 2 \mathrm{LSB}$ (see figure 16.7 (3)). Nonlinearity error is the deviation between the actual $\mathrm{A} / \mathrm{D}$ conversion characteristic and the ideal $\mathrm{A} / \mathrm{D}$ conversion characteristic from zero voltage to full-scale voltage (see figure 16.7 (4)). This does not include offset error, full-scale error, and quantization error.


Figure 16.7 A/D Conversion Precision Definitions

### 16.7 Usage Notes

The following points should be noted when using the mid-speed A/D converter.

### 16.7.1 Analog Voltage Settings

(1) Analog input voltage range

The voltage applied to analog input pins during A/D conversion should be in the range AVSS $\leq \mathrm{ANn} \leq \mathrm{AV}_{\text {ref }}$ ( $\mathrm{n}=0$ to 7 ).
(2) $A V_{C C}$ and $A V_{S S}$ input voltages

For the $A V_{C C}$ and $A V_{S S}$ input voltages, set $A V_{C C}=V_{C C} \pm 10 \%$ and $A V_{S S}=V_{S S}$. When the medium-speed $A / D$ converter is not used, set $A V_{C C}=V_{C C}$ and $A V_{S S}=V_{S S}$.
(3) AVref input voltage

For the $A V_{\text {ref }}$ pin input voltage analog reference, set $A V_{\text {ref }} \leq A V_{C C}$. When the medium-speed $A / D$ converter is not used, set $A V_{\text {ref }}=A V_{C C}$.
(4) $\mathrm{AV}_{\mathrm{CC}}$ and $\mathrm{AV}_{\text {ref }}$ must be connected to the power supply $\left(\mathrm{V}_{\mathrm{CC}}\right)$ even if the medium-speed $\mathrm{A} / \mathrm{D}$ converter is not used or is in standby mode.

### 16.7.2 Handling of Analog Input Pins

To prevent damage from surges and other abnormal voltages at the analog input pins (AN0-AN7), connect a protection circuit such as that shown in figure 16.8. This circuit also includes a CR filter function that suppresses error due to noise. The circuit shown here is only a design example; circuit constants must be decided on the basis of the actual operating conditions.

Figure 16.9 shows an equivalent circuit for the analog input pins, and table 16.6 summarizes the analog input pin specifications.


Figure 16.8 Example of Analog Input Pin Protection Circuit


Figure 16.9 Equivalent Circuit for the Analog Input Pins

Table 16.6 Analog Pin Specifications

| Item | Min | Max | Unit |
| :--- | :--- | :--- | :--- |
| Analog input capacitance | - | 20 | pF |
| Permissible signal source impedance | - | 1 | $\mathrm{k} \Omega$ |

## Section 17 Compare Match Timer (CMT)

### 17.1 Overview

The SH7040 series has an on-chip compare match timer (CMT) configured of 16-bit timers for two channels. The CMT has 16 -bit counters and can generate interrupts at set intervals.

### 17.1.1 Features

The CMT has the following features:

- Four types of counter input clock can be selected
— One of four internal clocks $(\phi / 8, \phi / 32, \phi / 128, \phi / 512)$ can be selected independently for each channel.
- Interrupt sources
- A compare match interrupt can be requested independently for each channel.


### 17.1.2 Block Diagram

Figure 17.1 shows a block diagram of the CMT.


Figure 17.1 CMT Block Diagram

### 17.1.3 Register Configuration

Table 17.1 summarizes the CMT register configuration.
Table 17.1 Register Configuration

| Channel | Name | Abbreviation | R/W | Initial Value | Address | Access Size (Bits) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Shared | Compare match timer start register | CMSTR | R/W | H'0000 | H'FFFF83D0 | 8,16, 32 |
| 0 | Compare match timer control/status register 0 | CMCSR0 | R/(W) ${ }^{\text {* }}$ | H'0000 | H'FFFF83D2 | 8,16, 32 |
|  | Compare match timer counter 0 | CMCNTO | R/W | H'0000 | H'FFFF83D4 | 8, 16, 32 |
|  | Compare match timer constant register 0 | CMCORO | R/W | H'FFFF | H'FFFF83D6 | 8,16, 32 |
| 1 | Compare match timer control/status register 1 | CMCSR1 | R/(W) ${ }^{\text {* }}$ | H'0000 | H'FFFF83D8 | 8,16, 32 |
|  | Compare match timer counter 1 | CMCNT1 | R/W | H'0000 | H'FFFF83DA | 8, 16, 32 |
|  | Compare match timer constant register 1 | CMCOR1 | R/W | H'FFFF | H'FFFF83DC | 8, 16, 32 |
| Note: * | The only value that can the flags. | be written to th | $\overline{\mathrm{MCSRO}}$ | nd CM | $\overline{\text { SR1 CMF bi }}$ | s a 0 to clear |

### 17.2 Register Descriptions

### 17.2.1 Compare Match Timer Start Register (CMSTR)

The compare match timer start register (CMSTR) is a 16-bit register that selects whether to operate or halt the channel 0 and channel 1 counters (CMCNT). It is initialized to H'0000 by power-on resets and by standby mode. Manual reset does not initialize CMSTR.


- Bits $15-2$-Reserved: These bits always read as 0 . The write value should always be 0 .
- Bit 1 -Count Start 1 (STR1): Selects whether to operate or halt compare match timer counter 1.


## Bit 1: STR1 Description

| 0 | CMCNT1 count operation halted (initial value) |
| :--- | :--- |
| 1 | CMCNT1 count operation |

- Bit 0—Count Start 0 (STR0): Selects whether to operate or halt compare match timer counter 0.
Bit 0: STRO Description

| 0 | CMCNT0 count operation halted (initial value) |
| :--- | :--- |
| 1 | CMCNT0 count operation |

### 17.2.2 Compare Match Timer Control/Status Register (CMCSR)

The compare match timer control/status register (CMCSR) is a 16-bit register that indicates the occurrence of compare matches, sets the enable/disable of interrupts, and establishes the clock used for incrementation. It is initialized to $\mathrm{H}^{\prime} 0000$ by power-on resets and by standby mode. Manual reset does not initialize CMCSR.


| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | CMF | CMIE | - | - | - | - | CKS1 | CKS0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | $R /(W)^{*}$ | R/W | $R$ | $R$ | $R$ | $R$ | R/W | R/W |

Note: * The only value that can be written is a 0 to clear the flag.

- Bits $15-8$ and 5-2-Reserved: These bits always read as 0 . The write value should always be 0.
- Bit 7—Compare Match Flag (CMF): This flag indicates whether or not the CMCNT and CMCOR values have matched.

| Bit 7: CMF | Description |
| :--- | :--- |
| 0 | CMCNT and CMCOR values have not matched (initial status) <br> Clear condition: Write a 0 to CMF after reading a 1 from it |
| 1 | CMCNT and CMCOR values have matched |

- Bit 6-Compare Match Interrupt Enable (CMIE): Selects whether to enable or disable a compare match interrupt (CMI) when the CMCNT and CMCOR values have matched $(\mathrm{CMF}=$ $1)$.


## Bit 6: CMIE <br> Description

| 0 | Compare match interrupts (CMI) disabled (initial status) |
| :--- | :--- |
| 1 | Compare match interrupts (CMI) enabled |

- Bits 1, 0—Clock Select 1, 0 (CKS1, CKS0): These bits select the clock input to the CMCNT from among the four internal clocks obtained by dividing the system clock $(\phi)$. When the STR bit of the CMSTR is set to 1 , the CMCNT begins incrementing with the clock selected by CKS1 and CKS0.

| Bit 1: CKS1 | Bit 0: CKSO | Description |
| :--- | :--- | :--- |
| 0 | 0 | $\phi / 8$ (initial status) |
|  | 1 | $\phi / 32$ |
| 1 | 0 | $\phi / 128$ |
|  | 1 | $\phi / 512$ |

### 17.2.3 Compare Match Timer Counter (CMCNT)

The compare match timer counter (CMCNT) is a 16-bit register used as an upcounter for generating interrupt requests.

When an internal clock is selected with the CKS1, CKS0 bits of the CMCSR register and the STR bit of the CMSTR is set to 1 , the CMCNT begins incrementing with that clock. When the CMCNT value matches that of the compare match timer constant register (CMCOR), the CMCNT is cleared to H'0000 and the CMF flag of the CMCSR is set to 1. If the CMIE bit of the CMCSR is set to 1 at this time, a compare match interrupt (CMI) is requested.

The CMCNT is initialized to $\mathrm{H}^{\prime} 0000$ by power-on resets and by standby mode. Manual reset does not initialize CMCNT.


### 17.2.4 Compare Match Timer Constant Register (CMCOR)

The compare match timer constant register (CMCOR) is a 16-bit register that sets the compare match period with the CMCNT.

The CMCOR is initialized to H'FFFF by power-on resets and by standby mode. There is no initializing with manual reset.


### 17.3 Operation

### 17.3.1 Period Count Operation

When an internal clock is selected with the CKS1, CKS0 bits of the CMCSR register and the STR bit of the CMSTR is set to 1 , the CMCNT begins incrementing with the selected clock. When the CMCNT counter value matches that of the compare match constant register (CMCOR), the CMCNT counter is cleared to $\mathrm{H}^{\prime} 0000$ and the CMF flag of the CMCSR register is set to 1 . If the CMIE bit of the CMCSR register is set to 1 at this time, a compare match interrupt (CMI) is requested. The CMCNT counter begins counting up again from H'0000.

Figure 17.2 shows the compare match counter operation.


Figure 17.2 Counter Operation

### 17.3.2 CMCNT Count Timing

One of four clocks ( $\phi / 8, \phi / 32, \phi / 128, \phi / 512$ ) obtained by dividing the system clock (CK) can be selected by the CKS1, CKS0 bits of the CMCSR. Figure 17.3 shows the timing.


Figure 17.3 Count Timing

### 17.4 Interrupts

### 17.4.1 Interrupt Sources and DTC Activation

The CMT has a compare match interrupt for each channel, with independent vector addresses allocated to each of them. The corresponding interrupt request is output when the interrupt request flag CMF is set to 1 and the interrupt enable bit CMIE has also been set to 1 .

When activating CPU interrupts by interrupt request, the priority between the channels can be changed by using the interrupt controller settings. See section 6, Interrupt Controller (INTC), for details.

Interrupt requests can also be used as data transfer controller (DTC) activating sources. In this case, channel priorities are fixed. See section 8, Data Transfer Controller (DTC), for details.

### 17.4.2 Compare Match Flag Set Timing

The CMF bit of the CMCSR register is set to 1 by the compare match signal generated when the CMCOR register and the CMCNT counter match. The compare match signal is generated upon the final state of the match (timing at which the CMCNT counter matching count value is updated). Consequently, after the CMCOR register and the CMCNT counter match, a compare match signal will not be generated until a CMCNT counter input clock occurs. Figure 17.4 shows the CMF bit set timing.


Figure 17.4 CMF Set Timing

### 17.4.3 Compare Match Flag Clear Timing

The CMF bit of the CMCSR register is cleared either by writing a 0 to it after reading a 1 , or by a clear signal after a DTC transfer. Figure 17.5 shows the timing when the CMF bit is cleared by the CPU


Figure 17.5 Timing of CMF Clear by the CPU

### 17.5 Notes on Use

Take care that the contentions described in sections 17.5.1-17.5.3 do not arise during CMT operation.

### 17.5.1 Contention between CMCNT Write and Compare Match

If a compare match signal is generated during the $\mathrm{T}_{2}$ state of the CMCNT counter write cycle, the CMCNT counter clear has priority, so the write to the CMCNT counter is not performed. Figure 17.6 shows the timing.


Figure 17.6 CMCNT Write and Compare Match Contention

### 17.5.2 Contention between CMCNT Word Write and Incrementation

If an increment occurs during the $\mathrm{T}_{2}$ state of the CMCNT counter word write cycle, the counter write has priority, so no increment occurs. Figure 17.7 shows the timing.


Figure 17.7 CMCNT Word Write and Increment Contention

### 17.5.3 Contention between CMCNT Byte Write and Incrementation

If an increment occurs during the $\mathrm{T}_{2}$ state of the CMCNT byte write cycle, the counter write has priority, so no increment of the write data results on the writing side. The byte data on the side not performing the writing is also not incremented, so the contents are those before the write.

Figure 17.8 shows the timing when an increment occurs during the $\mathrm{T}_{2}$ state of the CMCNTH write cycle.


Figure 17.8 CMCNT Byte Write and Increment Contention

## Section 18 Pin Function Controller

### 18.1 Overview

The pin function controller (PFC) is composed of registers for selecting the function of multiplexed pins and the direction of input/output. Table 18.1 lists the SH7040 Series's multiplexed pins. The multiplex pin functions have restrictions dependent on the operating mode. Table 18.2 lists the pin functions and initial values for each operating mode.

Table 18.1 Multiplexed Pins

| Port | Function 1 (Related Module) | Function 2 <br> (Related Module) | Function 3 (Related Module) | Function 4 (Related Module) | $\begin{aligned} & \text { FP. } \\ & 112 \end{aligned}$ |  | $\begin{aligned} & \text { TFP- } \\ & 120 \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | PA23 I/O (port) | $\overline{\text { WRHH output (BSC) }}$ | - | - | - | 1 |  |
|  | PA22 I/O (port) | $\overline{\text { WRHL output (BSC) }}$ | - | - | - | 3 | - |
|  | PA21 I/O (port) | CASHH output (BSC) |  | - | - | 4 |  |
|  | PA20 I/O (port) | CASHL output (BSC) | - | - | - | 29 | - |
|  | PA19 I/O (port) | $\overline{\text { BACK output (BSC) }}$ | DRAK1 output (DMAC) | - | - | 30 | - |
|  | PA18 I/O (port) | $\overline{\mathrm{BREQ}}$ input (BSC) | DRAKO output (DMAC) | - | - | 33 | - |
|  | PA17 /O (port) | $\overline{\text { WAIT input (BSC) }}$ | - | - | - |  | - |
|  | PA16 I/O (port) | $\overline{\mathrm{AH}}$ output (BSC) | - | - | - |  | - |
|  | PA15 I/O (port) | CK output (CPG) | - | - | 83 |  |  |
|  | PA14 I/O (port) | $\overline{\mathrm{RD}}$ output (BSC) | - | - | 34 | 43 | 37 |
|  | PA13 I/O (port) | $\overline{\text { WRH output (BSC) }}$ | - | - | 36 | 47 | 39 |
|  | PA12 I/O (port) | $\overline{\text { WRL }}$ output (BSC) | - | - | 38 | 48 | 41 |
|  | PA11 I/O (port) | $\overline{\text { CS1 }}$ output (BSC) | - | - | 40 | 49 | 43 |
|  | PA10 I/O (port) | CS0 output (BSC) | - | - | 41 | 50 | 44 |
|  | PA9 I/O (port) | TCLKD input (MTU) | $\overline{\mathrm{IRQ3}}$ (INTC) | - | 42 | 51 | 45 |
|  | PA8 I/O (port) | TCLKC input (MTU) | $\overline{\mathrm{IRQ2}}$ (INTC) | - | 43 | 52 | 46 |
|  | PA7 I/O (port) | TCLKB input (MTU) | $\overline{\mathrm{CS3}}$ output (BSC) |  | 44 | 53 | 47 |
|  | PA6 I/O (port) | TCLKA input (MTU) | $\overline{\text { CS2 output (BSC) }}$ |  | 45 | 54 | 48 |
|  | PA5 I/O (port) | SCK1 I/O (SCI) | DREQ1 input (DMAC) | $\overline{\text { IRQ1 }}$ input (INTC) |  |  | 49 |

Table 18.1 Multiplexed Pins (cont)

| Port | Function 1 (Re-lated Module) | Function 2 (Related Module) | Function 3 (Related Module) | Function 4 (Related Module) | $\begin{array}{lll} \text { FP- FP- TFP- } \\ 112 & 144 & 120 \end{array}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | PA4 I/O (port) | TxD1 output (SCI) | - | - | 47 |  |  |
|  | PA3 I/O (port) | RxD1 input (SCI) | - | - | 48 | 133 | 51 |
|  | PA2 I/O (port) | SCKO I/O (SCI) | $\overline{\text { DREQ0 }}$ input (DMAC) | $\overline{\text { IRQO }}$ input (INTC) | 49 | 132 | 52 |
|  | PA1 I/O (port) | TxD0 output (SCI) | - | - | 50 | 131 | 53 |
|  | PAO I/O (port) | RxD0 input (SCI) | - | - | 51 | 130 | 54 |
| B | PB9 I/O (port) | $\overline{\text { IRQ7 input (INTC) }}$ | A21 output (BSC) | $\overline{\text { ADTRG input (A/D) }}$ | 32 | 41 | 35 |
|  | PB8 I/O (port) | $\overline{\mathrm{IRQ6}}$ input (INTC) | A20 output (BSC) | $\overline{\text { WAIT input (BSC) }}$ | 31 | 39 | 34 |
|  | PB7 I/O (port) | $\overline{\mathrm{IRQ5}}$ input (INTC) | A19 output (BSC) | $\overline{\mathrm{BREQ}}$ input (BSC) | 30 | 38 | 33 |
|  | PB6 I/O (port) | $\overline{\mathrm{IRQ4}}$ input (INTC) | A18 output (BSC) | $\overline{\text { BACK }}$ output (BSC) | 29 | 37 | 32 |
|  | PB5 I/O (port) | IRQ3 input (INTC) | $\overline{\text { POE3 input (port) }}$ | RDWR output (BSC) |  | 36 | 29 |
|  | PB4 I/O (port) | $\overline{\text { IRQ2 }}$ input (INTC) | $\overline{\text { POE2 input (port) }}$ | $\overline{\text { CASH output (BSC) }}$ | 26 | 34 | 27 |
|  | PB3 I/O (port) | $\overline{\mathrm{IRQ1}}$ input (INTC) | $\overline{\text { POE1 input (port) }}$ | $\overline{\text { CASL output (BSC) }}$ | 25 | 32 | 26 |
|  | PB2 I/O (port) | $\overline{\text { IRQ0 input (INTC) }}$ | $\overline{\text { POEO input (port) }}$ | $\overline{\text { RAS output (BSC) }}$ | 24 | 31 | 25 |
|  | PB1 I/O (port) | A17 input (BSC) | - | - | 22 | 27 | 23 |
|  | PB0 I/O (port) | A16 output (BSC) | - | - | 20 | 25 | 21 |
| C | PC15 I/O (port) | A15 output (BSC) | - | - | 19 | 24 | 20 |
|  | PC14 I/O (port) | A14 output (BSC) | - | - | 18 | 23 | 19 |
|  | PC13 I/O (port) | A13 output (BSC) | - | - | 17 | 22 | 18 |
|  | PC12 I/O (port) | A12 output (BSC) | - | - | 16 | 21 | 17 |
|  | PC11 I/O (port) | A11 output (BSC) | - | - | 15 | 20 | 16 |
|  | PC10 I/O (port) | A10 output (BSC) | - | - | 14 | 19 | 15 |
|  | PC9 I/O (port) | A9 output (BSC) | - | - | 13 | 18 | 14 |
|  | PC8 I/O (port) | A8 output (BSC) | - | - | 12 | 17 | 13 |
|  | PC7 I/O (port) | A7 output (BSC) | - | - | 11 | 16 | 12 |

Table 18.1 Multiplexed Pins (cont)

| Port | Function 1 (Related Module) | Function 2 (Related Module) | Function 3 (Related Module) | Function 4 (Related Module) |  |  | $\begin{aligned} & \text { TFP- } \\ & 120 \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| C | PC6 I/O (port) | A6 output (BSC) | - | - | 10 | 15 | 11 |
|  | PC5 I/O (port) | A5 output (BSC) | - | - | 9 | 13 | 10 |
|  | PC4 I/O (port) | A4 output (BSC) | - | - | 8 | 11 | 9 |
|  | PC3 I/O (port) | A3 output (BSC) | - | - | 7 | 10 | 8 |
|  | PC2 I/O (port) | A2 output (BSC) | - | - | 6 | 9 | 7 |
|  | PC1 I/O (port) | A1 output (BSC) | - | - | 5 | 8 | 6 |
|  | PCO I/O (port) | A0 output (BSC) | - | - | 4 | 7 | 5 |
| D | PD31 I/O (port) | D31 I/O (BSC) | $\overline{\text { ADTRG input (A/D) }}$ | - | - | 45 | - |
|  | PD30 I/O (port) | D30 I/O (BSC) | IRQOUT output (INTC) | - | - | 46 | - |
|  | PD29 I/O (port) | D29 I/O (BSC) | $\overline{\mathrm{CS3}}$ output (BSC) | - | - | 56 | - |
|  | PD28 I/O (port) | D28 I/O (BSC) | $\overline{\mathrm{CS} 2}$ output (BSC) | - | - | 57 | - |
|  | PD27 I/O (port) | D27 I/O (BSC) | DACK1 output (DMAC) | - | - | 58 | - |
|  | PD26 I/O (port) | D26 I/O (BSC) | DACK0 output (DMAC) | - | - | 59 | - |
|  | PD25 I/O (port) | D25 I/O (BSC) | $\overline{\text { DREQ1 }}$ input (DMAC) | - | - | 60 | - |
|  | PD24 I/O (port) | D24 I/O (BSC) | $\overline{\text { DREQO input }}$ (DMAC) | - | - | 62 | - |
|  | PD23 I/O (port) | D23 I/O (BSC) | $\overline{\text { IRQ7 input (INTC) }}$ | - | - | 64 | - |
|  | PD22 I/O (port) | D22 I/O (BSC) | $\overline{\text { IRQ6 }}$ input (INTC) | - | - | 65 | - |
|  | PD21 I/O (port) | D21 I/O (BSC) | $\overline{\text { IRQ5 input (INTC) }}$ | - | - | 66 | - |
|  | PD20 I/O (port) | D20 I/O (BSC) | $\overline{\text { IRQ4 }}$ input (INTC) | - | - | 67 | - |
|  | PD19 I/O (port) | D19 I/O (BSC) | IRQ3 input (INTC) | - | - | 68 | - |
|  | PD18 I/O (port) | D18 I/O (BSC) | $\overline{\text { IRQ2 }}$ input (INTC) | - | - | 69 | - |
|  | PD17 I/O (port) | D17 I/O (BSC) | $\overline{\text { IRQ1 }}$ input (INTC) | - | - | 70 | - |
|  | PD16 I/O (port) | D16 I/O (BSC) | $\overline{\text { IRQ0 }}$ input (INTC) | - | - | 72 | - |
|  | PD15 I/O (port) | D15 I/O (BSC) | - | - | 52 | 73 | 55 |
|  | PD14 I/O (port) | D14 I/O (BSC) | - | - | 53 | 74 | 56 |

Table 18.1 Multiplexed Pins (cont)
Function 1 Function 2 Function 3 Function $4 \quad$ FP- FP- TFPPort (Related Module) (Related Module) (Related Module) (Related Module) 112144120

| D | PD13 I/O (port) | D13 I/O (BSC) | - | - | 54 | 75 | 57 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PD12 I/O (port) | D12 I/O (BSC) | - | - | 56 | 76 | 59 |
|  | PD11 I/O (port) | D11 I/O (BSC) | - | - | 57 | 78 | 62 |
|  | PD10 I/O (port) | D10 I/O (BSC) | - | - | 58 | 80 | 63 |
|  | PD9 I/O (port) | D9 I/O (BSC) | - | - | 59 | 81 | 64 |
|  | PD8 I/O (port) | D8 I/O (BSC) | - | - | 60 | 82 | 65 |
|  | PD7 I/O (port) | D7 I/O (BSC) | - | - | 62 | 83 | 67 |
|  | PD6 I/O (port) | D6 I/O (BSC) | - | - | 63 | 84 | 68 |
|  | PD5 I/O (port) | D5 I/O (BSC) | - | - | 64 | 86 | 69 |
|  | PD4 I/O (port) | D4 I/O (BSC) | - | - | 66 | 88 | 71 |
|  | PD3 I/O (port) | D3 I/O (BSC) | - | - | 67 | 89 | 72 |
|  | PD2 I/O (port) | D2 I/O (BSC) | - | - | 68 | 90 | 73 |
|  | PD1 I/O (port) | D1 I/O (BSC) | - | - | 69 | 91 | 74 |
|  | PDO I/O (port) | D0 I/O (BSC) | - | - | 70 | 92 | 75 |
| E | PE15 I/O (port) | TIOC4D I/O (MT |  |  | 2 | 5 | 3 |
|  | PE14 I/O (port) | TIOC4C I/O (MT |  |  | 1 | 2 | 2 |
|  | PE13 I/O (port) | TIOC4B I/O (MT | $\bar{M}$ |  | 112 | 144 | 120 |
|  | PE12 I/O (port) | TIOC4A I/O (MT |  | - | 111 | 143 | 119 |
|  | PE11 I/O (port) | TIOC3D I/O (MT |  | - | 110 | 142 | 118 |
|  | PE10 I/O (port) | TIOC3C I/O (MT) |  | - | 108 | 140 | 116 |
|  | PE9 I/O (port) | TIOC3B I/O (MT |  | - | 107 | 139 | 115 |
|  | PE8 I/O (port) | TIOC3A I/O (MT | - | - | 106 | 138 | 114 |
|  | PE7 I/O (port) | TIOC2B I/O (MT) | - | - | 105 | 137 | 113 |
|  | PE6 I/O (port) | TIOC2A I/O (MT) | - | - | 104 | 116 | 112 |
|  | PE5 I/O (port) | TIOC1B I/O (MT) | - | - | 102 | 115 | 109 |
|  | PE4 I/O (port) | TIOC1A I/O (MT | - | - | 89 |  |  |

Table 18.1 Multiplexed Pins (cont)

| Port | Function 1 (Related Module) | Function 2 (Related Module) | Function 3 (Related Module) | Function 4 (Related Module) | $\begin{aligned} & \text { FP- } \\ & 112 \end{aligned}$ | $\begin{aligned} & \text { FP- TFP- } \\ & 144120 \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| E | PE3 I/O (port) | TIOCOD I/O (MTU) | DRAK1 output (DMAC) | - | 88 | 11395 |
|  | PE2 I/O (port) | TIOCOC I/O (MTU) | DREQ1 input (DMAC) | - | 87 | 11194 |
|  | PE1 I/O (port) | TIOCOB I/O (MTU) | DRAKO output (DMAC) | - | 86 | 11093 |
|  | PEO I/O (port) | TIOCOA I/O (MTU) | $\overline{\text { DREQ0 }}$ input (DMAC) | - | 85 | 10992 |
| F | PF7 input (port) | AN7 input (A/D) | - | - | 99 | 126106 |
|  | PF6 input (port) | AN6 input (A/D) | - | - | 98 | 125105 |
|  | PF5 input (port) | AN5 input (A/D) | - | - | 96 | 123103 |
|  | PF4 input (port) | AN4 input (A/D) | - | - | 95 | 122102 |
|  | PF3 input (port) | AN3 input (A/D) | - | - | 94 | 121101 |
|  | PF2 input (port) | AN2 input (A/D) | - | - | 93 | 120100 |
|  | PF1 input (port) | AN1 input (A/D) | - | - | 92 | 11999 |
|  | PF0 input (port) | ANO input (A/D) | - | - | 91 | 11898 |

Table 18.2 Pin Arrangement by Mode

Table 18.2 Pin Arrangement by Mode (cont)

Table 18.2 Pin Arrangement by Mode (cont)

| Pin No. |  |  | Pin Name |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TFP120 FP144 |  | FP112 | $\frac{\text { On-Chip ROM Disabled }}{\text { MPU Mode0 }}$ |  | MPU Mode 1 |  | On-Chip ROM Enabled MPU Mode2 |  | Single Chip Mode |  | PROM Mode |
|  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | Intital Function | PFC Selected Function Possibilities | Intitia Function | PFC Selected Function Possibilitities | Initial Function | PFC Selected Function Possibilitities | Initial Function | PFC Selected Function Possibilitities |  |
|  | 127 |  | - | AVREF | AVREF | AVREF | AVREF | AVREF | AVREF | AVREF | AVREF |  |
| 98 | 118 | 91 | PFO/ANO | PFO/ANO | PFO/ANO | PFO/ANO | PFO/ANO | PFO/ANO | PFO/AN0 | PFO/ANO | VSS |
| 99 | 119 | 92 | PF 1/AN1 | PF1/AN1 | PF1/AN1 | PF1/AN1 | PF1/AN1 | PF1/AN1 | PF1/AN1 | PF1/AN1 | vss |
| 100 | 120 | 93 | PF2/AN2 | PF2/AN2 | PF2/AN2 | PF2/AN2 | PF2/AN2 | PF2/AN2 | PF2/AN2 | PF2/AN2 | vss |
| 101 | 121 | 94 | PF3/AN3 | PF3/AN3 | PF3/AN3 | PF3/AN3 | PF3/AN3 | PF3/AN3 | PF3/AN3 | PF3/AN3 | vSs |
| 102 | 122 | 95 | PF4/AN4 | PF4/AN4 | PF4/AN4 | PF4/AN4 | PF4/AN4 | PF4/AN4 | PF4/AN4 | PF4/AN4 | vss |
| 103 | 123 | 96 | PF5/AN5 | PF5/AN5 | PF5/AN5 | PF5/AN5 | PF5/AN5 | PF5/AN5 | PF5/AN5 | PF5/AN5 | vss |
| 105 | 125 | 98 | PF6/AN6 | PF6/AN6 | PF6/AN6 | PF6/AN6 | PF6/AN6 | PF6/AN6 | PF6/AN6 | PF6/AN6 | vss |
| 106 | 126 | 99 | PF7/AN7 | PF7/AN7 | PF7/AN7 | PF7/AN7 | PF7/AN7 | PF7/AN7 | PF7/AN7 | PF7/AN7 | vss |
| 92 | 109 | 85 | PE0 | PEO/TIOCOADREQ0 | PE0 | PEO/TIOCOADREQ0 | PE0 | PEO/TIOCOADREQ0 | PE0 | PEO/TIOCOA | NC |
| 93 | 110 | 86 | PE1 | PE1/TIOCOB/DRAK0 | PE1 | PE1/TIOCOB/DRAK0 | PE1 | PE1/TIOCOB/DRAK0 | PE1 | PE1/TIOCOB | NC |
| 94 | 111 | 87 | PE2 | PE2TIOCOC/DREQ1 | PE2 | PE2/TIOC0C/DREQ1 | PE2 | PE2TIOCOC/DREQ1 | PE2 | PE2/TIOCOC | NC |
| 95 | 113 | 88 | PE3 | PE3/TIOCOD/DRAK1 | PE3 | PE3/TIOCOD/DRAK1 | PE3 | PE3/TIOCOD/DRAK1 | PE3 | PE3/TIOCOD | NC |
| 96 | 114 | 89 | PE4 | PE4/TIOC1A | PE4 | PE4TIOC1A | PE4 | PE4TIOC1A | PE4 | PE4/TIOC1A | NC |
| 109 | 115 | 102 | PE5 | PE5/TIOC1B | PE5 | PE5/TIOC1B | PE5 | PE5/TIOC1B | PE5 | PE5/TIOC1B | NC |
| 112 | 116 | 104 | PE6 | PE6/TIOC2A | PE6 | PE6/TIOC2A | PE6 | PE6/TIOC2A | PE6 | PE6/TIOC2A | NC |
| 113 | 137 | 105 | PE7 | PE7/TIOC2B | PE7 | PE7/TIOC2B | PE7 | PE7/TIOC2B | PE7 | PE7/TIOC2B | NC |
| 114 | 138 | 106 | PE8 | PE8/TIOC3A | PE8 | PE8/TIOC3A | PE8 | PE8/TIOC3A | PE8 | PE8/TIOC3A | NC |
| 115 | 139 | 107 | PE9 | РЕ9/ТІОСЗВ | PE9 | PE9/TIOC3B | PE9 | РЕ9/ТОССЗВ | PE9 | PE9/TIOC3B | NC |
| 116 | 140 | 108 | PE10 | PE10/TIOC3C | PE10 | PE10/TIOC3C | PE10 | PE10/TIOC3C | PE10 | PE10/TIOC3C | NC |
| 118 | 142 | 110 | PE11 | PE11/TIOC3D | PE11 | PE11/TIOC3D | PE11 | PE11/TIOC3D | PE11 | PE11/TIOC3D | NC |
| 119 | 143 | 111 | PE12 | PE12/TIOC4A | PE12 | PE12TIOC4A | PE12 | PE12/TIOC4A | PE12 | PE12TIOC4A | NC |
| 120 | 144 | 112 | PE13 | PE13/TIOC4BMRES | PE13 | PE13TIOC4BMRES | PE13 | PE13/TIOC4BMRES | PE13 | PE13TIOC4BMRES | NC |
| 2 | 2 | 1 | PE14 | PE14/TIOC4C/DACK0/AH | PE14 | PE14/TIOC4C/DACKO/AH | PE14 | PE14/TIOC4C/DACK0/AH | PE14 | PE14/TIOC4C | NC |
|  | 5 | 2 | PE15 | PE15/TIOC4D/DACK1/RQOU | TPE15 | PE15/TIOC4D/DACK1/RQOU |  | PE15/TIOC4D/DACK1/RQOU |  | PE15/TIOC4D | CE |

### 18.2 Register Configuration

Table 18.3 summarizes the registers of the pin function controller.
Table 18.3 Pin Function Controller Registers

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Port A I/O register H | PAIORH | R/W | H'0000 | H'FFFF8384 H'FFFF8385 | 8, 16, 32 |
| Port A I/O register L | PAIORL | R/W | H'0000 | H'FFFF8386 H'FFFF8387 | 8, 16, 32 |
| Port A control register H | PACRH | R/W | H'0000 | H'FFFF8388 H'FFFF8389 | 8, 16, 32 |
| Port A control register L1 | PACRL1 | R/W | $\begin{aligned} & \text { H'0000* } \\ & \text { H'4000 } \end{aligned}$ | H'FFFF838C H'FFFF838D | 8, 16, 32 |
| Port A control register L2 | PACRL2 | R/W | H'0000 | H'FFFF838E H'FFFF838F | 8, 16, 32 |
| Port B I/O register | PBIOR | R/W | H'0000 | H'FFFF8394 H'FFFF8395 | 8, 16, 32 |
| Port B control register 1 | PBCR1 | R/W | H'0000 | H'FFFF8398 <br> H'FFFF8399 | 8, 16, 32 |
| Port B control register 2 | PBCR2 | R/W | H'0000 | H'FFFF839A H'FFFF839B | 8, 16, 32 |
| Port C I/O register | PCIOR | R/W | H'0000 | H'FFFF8396 H'FFFF8397 | 8, 16, 32 |
| Port C control register | PCCR | R/W | H'0000 | H'FFFF839C H'FFFF839D | 8, 16, 32 |
| Port D I/O register H | PDIORH | R/W | H'0000 | H'FFFF83A4 H'FFFF83A5 | 8, 16, 32 |
| Port D I/O register L | PDIORL | R/W | H'0000 | H'FFFF83A6 H'FFFF83A7 | 8, 16, 32 |
| Port D control register H1 | PDCRH1 | R/W | H'0000 | H'FFFF83A8 H'FFFF83A9 | 8, 16, 32 |
| Port D control register H2 | PDCRH2 | R/W | H'0000 | H'FFFF83AA H'FFFF83AB | 8, 16, 32 |
| Port D control register L | PDCRL | R/W | H'0000 | H'FFFF83AC H'FFFF83AD | 8, 16, 32 |
| Port E I/O register | PEIOR | R/W | H'0000 | H'FFFF83B4 <br> H'FFFF83B5 | 8, 16, 32 |
| Port E control register 1 | PECR1 | R/W | H'0000 | H'FFFF83B8 H'FFFF83B9 | 8,16, 32 |
| Port E control register 2 | PECR2 | R/W | H'0000 | H'FFFF83BA H'FFFF83BB | 8, 16, 32 |
| IRQOUT function control register | IFCR | R/W | H'0000 | H'FFFF83C8 H'FFFF83C9 | 8, 16, 32 |

Note: * The port A control register L1 initial value varies depending on the operating mode.

## Renesns

### 18.3 Register Descriptions

### 18.3.1 Port A I/O Register H (PAIORH)

The port A I/O register H (PAIORH) is a 16-bit read/write register that selects input or output for the most significant 8 pins of port A. Bits PA23IOR-PA16IOR correspond to pins PA23/WRHH PA16/AH. PAIORH is enabled when the port A pins function as general input/outputs (PA23PA16). For other functions, it is disabled.

For port A pin functions PA23-PA16, a given pin in port A is an output pin if its corresponding PAIORH bit is set to 1 , and an input pin if the bit is cleared to 0 .

PAIORH is initialized to H'0000 by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

The settings for this register are effective only for the 144-pin version. There are no corresponding pins for this register in the 112-pin and 120-pin versions. However, read/writes are possible.


## Renesns

### 18.3.2 Port A I/O Register L (PAIORL)

The port A I/O register L (PAIORL) is a 16-bit read/write register that selects input or output for the least significant 16 pins of port A. Bits PA15IOR-PA0IOR correspond to pins PA15/CKPA0/RXD0. PAIORL is enabled when the port A pins function as general input/outputs (PA15PA0), or with the serial clock (SCK1, SCK0). For other functions, it is disabled.

When the port A pin functions PA15-PA0 are SCK1, SCK0, a given pin in port A is an output pin if its corresponding PAIORL bit is set to 1 , and an input pin if the bit is cleared to 0 .

PAIORL is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.


### 18.3.3 Port A Control Register H (PACRH)

PACRH is a 16-bit read/write register that selects the multiplex pin function for the eight most significant pins of port A. PACRH selects the PA23/WRHH$-\mathrm{PA} 16 / \overline{\mathrm{AH}}$ pin functions.

The eight most significant pins of port A have bus control signals ( $\overline{\mathrm{WRHH}}, \overline{\mathrm{WRHL}}, \overline{\mathrm{CASHH}}$, $\overline{\mathrm{CASHL}}, \overline{\mathrm{BACK}}, \overline{\mathrm{BREQ}}, \overline{\mathrm{WAIT}}, \overline{\mathrm{AH}}$ ) and DMAC control signals (DRAK1, DRAK0), but there are instances when the register settings that select these pin functions will be ignored. Refer to table 18.2, Pin Arrangement by Mode.

PACRH is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset but is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

The settings for this register are effective only for the 144-pin version. There are no corresponding pins for this register in the 112-pin and 120-pin versions. However, read/writes are possible.


- Bit 15 -Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit $14 —$ PA23 Mode (PA23MD): Selects the function of the PA23/WRHH pin.


## Bit 14: PA23MD Description

| 0 | General input/output (PA23) (initial value) ( $\overline{\mathrm{WRHH}}$ in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Most significant byte write output $\overline{\mathrm{WRHH}})($ PA23 in single chip mode) |

- Bit 13 —Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 12—PA22 Mode (PA22MD): Selects the function of the PA22/WRHL pin.


## Bit 12: PA22MD Description

| 0 | General input/output (PA22) (initial value) ( $\overline{\text { WRHL }}$ in on-chip ROM invalid mode) |
| :--- | :--- |
|  | Write output ( $\overline{\mathrm{WRHL}})(\mathrm{PA} 22$ in single chip mode) |

1
Write output ( $\overline{\mathrm{WRHL}})$ (PA22 in single chip mode)

- Bit 11 —Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 10—PA21 Mode (PA21MD): Selects the function of the PA21/ $\overline{\mathrm{CASHH}}$ pin.


## Bit 10: PA21MD Description

| 0 | General input/output (PA21) (initial value) |
| :--- | :--- |
| 1 | Column address output ( $\overline{\text { CASHH }})($ PA21 in single chip mode) |

- Bit 9—Reserved: Always reads as 0 . The write values should always be 0 .
- Bit 8—PA20 Mode (PA20MD): Selects the function of the PA20/CASHL $p$ in.

Bit 8: PA20MD Description

| 0 | General input/output (PA20) (initial value) |
| :--- | :--- |
| 1 | Column address output ( $\overline{\text { CASHL }})($ PA20 in single chip mode) |

- Bits 7 and 6-PA19 Mode 1, 0 (PA19MD1 and PA19MD0): These bits select the function of the PA19/BACK $/ D R A K 1$ pin.

| Bit 7: <br> PA19MD1 | Bit 6: <br> PA19MD0 | Description |
| :--- | :--- | :--- |

- Bits 5 and 4—PA18 Mode 1, 0 (PA18MD1 and PA18MD0): These bits select the function of the PA18/BREQ $/ \mathrm{DRAK} 0$ pin.

| Bit 5: <br> PA18MD1 | Bit 4: <br> PA18MD0 | Description |
| :--- | :--- | :--- |

- Bit 3—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 2—PA17 Mode (PA17MD): Selects the function of the PA17/ $\overline{\mathrm{WAIT}}$ pin.


## Bit 2: PA17MD Description

| 0 | General input/output (PA17) (initial value) |
| :--- | :--- |
| 1 | Wait state request input ("WAIT) (PA17 in single chip mode) |

- Bit $1 —$ Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 0—PA16 Mode (PA16MD): Selects the function of the PA16/ $\overline{\mathrm{AH}}$ pin.

Bit 0: PA16MD Description

| 0 | General input/output (PA16) (initial value) |
| :--- | :--- |
| 1 | Address hold output $(\overline{\mathrm{AH}})($ PA16 in single chip mode) |

### 18.3.4 Port A Control Registers L1, L2 (PACRL1 and PACRL2)

PACRL1 and PACRL2 are 16-bit read/write registers that select the functions of the least significant sixteen multiplexed pins of port A. PACRL1 selects the function of the PA15/CKPA8/TCLKC/IRQ2 pins of port A; PACRL2 selects the function of the PA7/TCLKB/ $\overline{\mathrm{CS} 3}-$ PA0/RXD0 pins of port A.

Port A has bus control signals ( $\overline{\mathrm{RD}}, \overline{\mathrm{WRH}}, \overline{\mathrm{WRL}}, \overline{\mathrm{CSO}}-\overline{\mathrm{CS} 3}, \overline{\mathrm{AH}})$ and DMAC control signals (DREQ0-DREQ1), but there are instances when the register settings that select these pin functions will be ignored, depending on the operation mode. Refer to table 18.2, Pin Arrangement by Mode, for details.

PACRL1 is initialized by external power-on reset to $\mathrm{H}^{\prime} 4000$ in extended mode, and to $\mathrm{H}^{\prime} 0000$ in single chip mode. PACRL2 is initialized by external power-on reset to $\mathrm{H}^{\prime} 0000$. Neither register is initialized by manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

## Port A Control Register L1 (PACRL1):



Note: * Bit 14 is initialized to 1 in extended mode.

- Bit 15 -Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit $14 —$ PA15 Mode (PA15MD): Selects the function of the PA15/CK pin.

Bit 14: PA15MD Description

| 0 | General input/output (PA15) (single chip mode initial value) |
| :--- | :--- |
| 1 | Clock output (CK) (extended mode initial value) |

- Bit 13 —Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 12—PA14 Mode (PA14MD): Selects the function of the PA14/ $\overline{\text { RD }}$ pin.

Bit 12: PA14MD Description

| 0 | General input/output (PA14) (initial value) ( $\overline{\mathrm{RD}}$ in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Read output ( $\overline{\mathrm{RD}}$ ) (PA14 in single chip mode) |

- Bit 11 —Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 10—PA13 Mode (PA13MD): Selects the function of the PA13/ $\overline{\mathrm{WRH}}$ pin.

Bit 10: PA13MD Description
$1 \quad$ Most significant side write output ( $\overline{\mathrm{WRH}}$ ) (PA13 in single chip mode)

- Bit 9—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 8—PA12 Mode (PA12MD): Selects the function of the PA12/WRL pin.

Bit 8: PA12MD Description

| 0 | General input/output (PA12) (initial value) ( $\overline{\mathrm{WRL}}$ in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Least significant side write output ( $\overline{\mathrm{WRL}})(\mathrm{PA12}$ in single chip mode) |

- Bit 7—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 6—PA11 Mode (PA11MD): Selects the function of the PA11/ $\overline{\mathrm{CS} 1}$ pin.


## Bit 6: PA11MD Description

| 0 | General input/output (PA11) (initial value) ( $\overline{\mathrm{CS} 1}$ in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Chip select output ( $\overline{\mathrm{CS} 1}$ ) (PA11 in single chip mode) |

- Bit 5—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 4—PA10 Mode (PA10MD): Selects the function of the PA10/ $\overline{\mathrm{CS} 0}$ pin.


## Bit 4: PA10MD Description

| 0 | General input/output (PA10) (initial value) ( $\overline{\mathrm{CSO}}$ in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Chip select output ( $\overline{\mathrm{CSO}})(\mathrm{PA10}$ in single chip mode) |

- Bits 3 and 2—PA9 Mode 1, 0 (PA9MD1 and PA9MD0): These bits select the function of the PA9/TCLKD/ $\overline{\mathrm{IRQ} 3}$ pin.

| Bit 3: <br> PA9MD1 | Bit 2: <br> PA9MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PA9) (initial value) |
|  | 1 | MTU timer clock input (TCLKD) |
| 1 | 0 | Interrupt request input ( $\overline{\mathrm{IRQ3}})$ |
| 1 | Reserved |  |

- Bits 1 and 0—PA8 Mode 1, 0 (PA8MD1 and PA8MD0): These bits select the function of the PA8/TCLKC/IRQ2 pin.

| Bit 1: <br> PA8MD1 | Bit 0: <br> PA8MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PA8) (initial value) |
|  | 1 | MTU timer clock input (TCLKC) |
| 1 | 0 | Interrupt request input (IRQ2) |
| 1 | Reserved |  |

Port A Control Register L2 (PACRL2):

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PA7 <br> MD1 | PA7 <br> MD0 | PA6 <br> MD1 | PA6 <br> MD0 | PA5 <br> MD1 | PA5 <br> MD0 | - | PA4MD |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R | R/W |


| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | PA3MD | PA2 <br> MD1 | PA2 <br> MD0 | - | PA1MD | - | PAOMD |
|  |  |  |  |  |  |  |  |  |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | $R$ | R/W | R/W | R/W | $R$ | R/W | $R$ | R/W |

- Bits 15 and $14 —$ PA7 Mode 1, 0 (PA7MD1 and PA7MD0): These bits select the function of the PA7/TCLKB/ $\overline{\mathrm{CS} 3}$ pin.

| Bit 15: <br> PA7MD1 | Bit 14: <br> PA7MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PA7) (initial value) |
|  | 1 | MTU timer clock input (TCLKB) |
| 1 | 0 | Chip select output ( $\overline{\mathrm{CS3}})$ (PA7 in single chip mode) |
| 1 |  | Reserved |

- Bits 13 and 12—PA6 Mode 1, 0 (PA6MD1 and PA6MD0): These bits select the function of the PA6/TCLKA/CS2 pin.

| Bit 13: <br> PA6MD1 | Bit 12: <br> PA6MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PA6) (initial value) |
|  | 1 | MTU timer clock input (TCLKA) |
| 1 | 0 | Chip select output ( $\overline{\mathrm{CS} 2})$ (PA6 in single chip mode) |
| 1 | Reserved |  |

- Bits 11 and 10—PA5 Mode 1, 0 (PA5MD1 and PA5MD0): These bits select the function of the PA5/SCK1/ $\overline{\mathrm{DREQ}} 1 / \overline{\mathrm{IRQ} 1} \mathrm{pin}$.

| Bit 11: <br> PA5MD1 | Bit 10: <br> PA5MD0 | Description |
| :--- | :--- | :--- |, | 0 | 0 | General input/output (PA5) (initial value) |
| :--- | :--- | :--- |
|  | 1 | Serial clock input/output (SCK1) |
| 1 | 0 | DMA transfer request received input (DREQ1) (PA5 in single <br> chip mode) |
|  | 1 | Interrupt request input (IRQ1) |

- Bit 9—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 8—PA4 Mode (PA4MD): Selects the function of the PA4/TxD1 pin.


## Bit 8: PA4MD Description

| 0 | General input/output (PA4) (initial value) |
| :--- | :--- |
| 1 | Transmit data output (TxD1) |

- Bit 7—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 6—PA3 Mode (PA3MD): Selects the function of the PA3/RxD1 pin.


## Bit 6: PA3MD Description

$0 \quad$ General input/output (PA3) (initial value)
$1 \quad$ Receive data input (RxD1)

- Bits 5 and 4—PA2 Mode 1,0 (PA2MD1 and PA2MD0): These bits select the function of the PA2/SCK0/DREQ0 $/ \overline{\text { IRQ } 0}$ pin.

| Bit 5: <br> PA2MD1 | Bit 4: <br> PA2MD0 | Description |
| :--- | :--- | :--- |, | 0 | 0 | General input/output (PA2) (initial value) |
| :--- | :--- | :--- |
|  | 1 | Serial clock input/output (SCK0) <br> 1 |
|  | 0 | DMA transfer request received input ( $\overline{\text { DREQ0 }})($ PA2 in single |

- Bit 3—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 2—PA1 Mode (PA1MD): Selects the function of the PA1/TxD0 pin.

Bit 2: PA1MD Description

| 0 | General input/output (PA1) (initial value) |
| :--- | :--- |
| 1 | Transmit data output (TxD0) |

- Bit 1 -Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 0—PA0 Mode (PA0MD): Selects the function of the PA0/RxD0 pin.

Bit 0: PAOMD Description

| 0 | General input/output (PA0) (initial value) |
| :--- | :--- |
| 1 | Receive data input (RxD0) |

### 18.3.5 Port B I/O Register (PBIOR)

The port B I/O register (PBIOR) is a 16-bit read/write register that selects input or output for the ten port B pins. Bits PB9IOR-PB0IOR correspond to the PB9/ $\overline{\mathrm{IRQ} 7 / \mathrm{A} 21 / \overline{\mathrm{ADTRG}} \text { pin to } \mathrm{PB} 0 / \mathrm{A} 16}$ pin. PBIOR is enabled when the port B pins function as input/outputs (PB9-PB0). For other functions, it is disabled.

For port B pin functions PB9-PB0, a given pin in port B is an output pin if its corresponding PBIOR bit is set to 1 , and an input pin if the bit is cleared to 0 .

PBIOR is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.


### 18.3.6 Port B Control Registers (PBCR1 and PBCR2)

PBCR1 and PBCR2 are 16-bit read/write registers that select the functions of the ten multiplexed pins of port B. PBCR1 selects the functions of the top two bits of port B ; PBCR2 selects the functions of the bottom eight bits of port $B$.

Port B has bus control signals (RDWR, $\overline{\mathrm{RAS}}, \overline{\mathrm{CASH}}, \overline{\mathrm{CASL}}, \overline{\mathrm{WAIT}}, \overline{\mathrm{BREQ}}, \overline{\mathrm{BACK}}$ ) and address outputs (A21, A20, A19, A18, A17, A16), but there are instances when the register settings that select these pin functions will be ignored, depending on the operation mode. Refer to table 18.2, Pin Arrangement by Mode, for details.

PBCR1 and PBCR2 are both initialized to H'0000 by external power-on reset but are not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

Port B Control Register 1 (PBCR1):


- Bits $15-4 —$ Reserved: These bits always read as 0 . The write value should always be 0 .
- Bits 3 and 2—PB9 Mode (PB9MD1 and PB9MD0): PB9MD1 and PB9MD0 select the function of the PB9/IRQ7/A21/ADTRG pin.

Bit 3: PB9MD1 Bit 2: PB9MD0 Description

| 0 | 0 | General input/output (PB9) (initial value) |
| :--- | :--- | :--- |
| 1 | 0 | Interrupt request input (IRQ7) |
| 1 | $\overline{1}$ | Address output (A21) (PB9 in single chip mode) |

- Bits 1 and 0—PB8 Mode (PB8MD1 and PB8MD0): PB8MD1 and PB8MD0 select the function of the $\mathrm{PB} 8 / \overline{\mathrm{IRQ} 6} / \mathrm{A} 20 / \overline{\mathrm{WAIT}}$ pin.

| Bit 1: PB8MD1 | Bit 0: PB8MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PB8) (initial value) |
|  | 1 | Interrupt request input ( $\overline{\mathrm{IRQ6}})$ |
| 1 | 0 | Address output (A20) (PB8 in single chip mode) |
| 1 | Wait state request input ("WAIT) (PB8 in single chip <br> mode) |  |

## Port B Control Register 2 (PBCR2):

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PB7MD1 | PB7MD0 | PB6MD1 | PB6MD0 | PB5MD1 | PB5MD0 | PB4MD1 | PB4MD0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | PB3MD1 | PB3MD0 | PB2MD1 | PB2MD0 | - | PB1MD | - | PBOMD |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R | R/W | R | R/W |

- Bits 15 and 14—PB7 Mode (PB7MD1 and PB7MD0): PB7MD1 and PB7MD0 select the function of the PB7/IRQ5/A19/ $\overline{\mathrm{BREQ}}$ pin.

| Bit 15: <br> PB7MD1 | Bit 14: <br> PB7MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PB7) (initial value) |
|  | 1 | Interrupt request input (쥬Q5) |
| 1 | 0 | Address output (A19) (PB7 in single chip mode) |
|  | 1 | Bus right request input ( $\overline{\mathrm{BREQ}})$ (PB7 in single chip mode) |

- Bits 13 and 12—PB6 Mode (PB6MD1 and PB6MD0): PB6MD1 and PB6MD0 select the function of the PB6//[RQ4/A18/BACK pin.

| Bit 13: <br> PB6MD1 | Bit 12: <br> PB6MD0 | Description |
| :--- | :--- | :--- |, | 0 | 0 | General input/output (PB6) (initial value) |
| :--- | :--- | :--- |
|  | 1 | Interrupt request input ( $\overline{\mathrm{RQQ} 4})$ |
| 1 | 0 | Address output (A18) (PB6 in single chip mode) |
|  | 1 | Bus right request output ( $\overline{\mathrm{BACK}})$ (PB6 in single chip mode) |

- Bits 11 and 10—PB5 Mode (PB5MD1 and PB5MD0): PB5MD1 and PB5MD0 select the function of the PB5/IRQ3/POE3/RDWR pin.

| Bit 11: <br> PB5MD1 | Bit 10: <br> PB5MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PB5) (initial value) |
|  | 1 | Interrupt request input ( $\overline{\text { RQS3 }})$ |
| 1 | 0 | Port output enable ( $\overline{\text { POE3 }})$ |
| 1 | Read/write output (RDWR) |  |

- Bits 9 and 8—PB4 Mode (PB4MD1 and PB4MD0): PB4MD1 and PB4MD0 select the function of the PB4/ $\overline{\mathrm{IRQ} 2} / \overline{\mathrm{POE} 2} / \overline{\mathrm{CASH}}$ pin.

Bit 9: PB4MD1 Bit 8: PB4MDO Description

| 0 | 0 | General input/output (PB4) (initial value) |
| :--- | :--- | :--- |
|  | 1 | Interrupt request input ( $\overline{\mathrm{RQ} 2})$ |
| 1 | 0 | Port output enable ( $\overline{\text { POE2 }})$ |
| 1 | Column address strobe ( $\overline{\text { CASH }})$ (PB4 in single chip mode) |  |

- Bits 7 and 6—PB3 Mode (PB3MD1 and PB3MD0): PB3MD1 and PB3MD0 select the function of the $\mathrm{PB} 3 / \overline{\mathrm{IRQ}} 1 / \overline{\mathrm{POE}} / / \overline{\mathrm{CASL}}$ pin.

Bit 7: PB3MD1 Bit 6: PB3MD0 Description

| 0 | 0 | General input/output (PB3) (initial value) |
| :---: | :---: | :---: |
|  | 1 | Interrupt request input (IRQ1) |
| 1 | 0 | Port output enable ( $\overline{\text { POE1 }}$ ) |
|  | 1 | Column address strobe ('CASL) (PB3 in single chip mode) |

- Bits 5 and 4—PB2 Mode (PB2MD1 and PB2MD0): PB2MD1 and PB2MD0 select the function of the $\mathrm{PB} 2 / \overline{\mathrm{IRQ} 0} / \overline{\mathrm{POE} 0} / \overline{\mathrm{RAS}}$ pin.

Bit 5: PB2MD1 Bit 4: PB2MD0 Description

| 0 | 0 | General input/output (PB2) (initial value) |
| :--- | :--- | :--- |
|  | 1 | Interrupt request input ( $\overline{\mathrm{RQQ}})$ |
| 1 | 0 | Port output enable ( $\overline{\text { POE0 }})$ |
|  | 1 | Row address strobe ( $\overline{\mathrm{RAS}})$ (PB2 in single chip mode) |

- Bit 3—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 2—PB1 Mode (PB1MD): Selects the function of the PB1/A17 pin.

Bit 2: PB1MD Description

| 0 | General input/output (PB1) (initial value) (A17 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A17) (PB1 in single chip mode) |

- Bit 1 —Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 0—PB0 Mode (PB0MD): Selects the function of the PB0/A16 pin.

Bit 0: PAOMD Description

| 0 | General input/output (PB0) (initial value) (A16 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A16) (PBO in single chip mode) |

### 18.3.7 Port C I/O Register (PCIOR)

The port C I/O register (PCIOR) is a 16 -bit read/write register that selects input or output for the 16 port C pins. Bits PC15IOR-PC0IOR correspond to pins PC15/A15 to PC0/A0. PCIOR is enabled when the port C pins function as general input/outputs ( $\mathrm{PC} 15-\mathrm{PC} 0$ ). For other functions, it is disabled.

When the port C pin functions are as $\mathrm{PC} 15-\mathrm{PC} 0$, a given pin in port C is an output pin if its corresponding PCIOR bit is set to 1 , and an input pin if the bit is cleared to 0 .

PCIOR is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous values are maintained.


| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PC7 <br> IOR | PC6 <br> IOR | PC5 <br> IOR | PC4 <br> IOR | PC3 <br> IOR | PC2 <br> IOR | PC1 <br> IOR | PC0 <br> IOR |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

### 18.3.8 Port C Control Register (PCCR)

PCCR is a 16-bit read/write register that selects the functions for the sixteen port C multiplexed pins. There are instances when these register settings will be ignored, depending on the operation mode. Refer to table 18.2, Pin Arrangement by Mode, for details.

PCCR is initialized to $\mathrm{H}^{\prime} 0000$ by power-on resets but is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { PC15 } \\ & \text { MD } \end{aligned}$ | $\begin{gathered} \text { PC14 } \\ \text { MD } \end{gathered}$ | $\begin{aligned} & \text { PC13 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC12 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC11 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC10 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC9 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC8 } \\ & \text { MD } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |


| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { PC7 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC6 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC5 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC4 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC3 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC2 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PC1 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PCO } \\ & M D \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 15—PC15 Mode (PC15MD): Selects the function of the PC15/A15 pin.

Bit 15: PC15MD Description

| 0 | General input/output (PC15) (initial value) (A15 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A15) (PC15 in single chip mode) |

- Bit 14—PC14 Mode (PC14MD): Selects the function of the PC14/A14 pin.

Bit 14: PC14MD Description

| 0 | General input/output (PC14) (initial value) (A14 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A14) (PC14 in single chip mode) |

- Bit 13—PC13 Mode (PC13MD): Selects the function of the PC13/A13 pin.


## Bit 13: PC13MD Description

| 0 | General input/output (PC13) (initial value) (A13 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A13) (PC13 in single chip mode) |

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- Bit 12—PC12 Mode (PC12MD): Selects the function of the PC12/A12 pin.


## Bit 12: PC12MD Description

| 0 | General input/output (PC12) (initial value) (A12 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A12) (PC12 in single chip mode) |

- Bit 11—PC11 Mode (PC11MD): Selects the function of the PC11/A11 pin.

Bit 11: PC11MD Description

| 0 | General input/output (PC11) (initial value) (A11 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A11) (PC11 in single chip mode) |

- Bit 10—PC10 Mode (PC10MD): Selects the function of the PC10/A10 pin.

Bit 10: PC10MD Description

| 0 | General input/output (PC10) (initial value) (A10 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A10) (PC10 in single chip mode) |

- Bit 9—PC9 Mode (PC9MD): Selects the function of the PC9/A9 pin.


## Bit 9: PC9MD Description

| 0 | General input/output (PC9) (initial value) (A9 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A9) (PC9 in single chip mode) |

- Bit 8—PC8 Mode (PC8MD): Selects the function of the PC8/A8 pin.

Bit 8: PC8MD Description

| 0 | General input/output (PC8) (initial value) (A8 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A8) (PC8 in single chip mode) |

- Bit 7—PC7 Mode (PC7MD): Selects the function of the PC7/A7 pin.

Bit 7: PC7MD Description

| 0 | General input/output (PC7) (initial value) (A7 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A7) (PC7 in single chip mode) |

- Bit 6-PC6 Mode (PC6MD): Selects the function of the PC6/A6 pin.

Bit 6: PC6MD Description

| 0 | General input/output (PC6) (initial value) (A6 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A6) (PC6 in single chip mode) |

- Bit 5-PC5 Mode (PC5MD): Selects the function of the PC5/A5 pin.

Bit 5: PC5MD Description

| 0 | General input/output (PC5) (initial value) (A5 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A5) (PC5 in single chip mode) |

- Bit 4—PC4 Mode (PC4MD): Selects the function of the PC4/A4 pin.

Bit 4: PC4MD Description

| 0 | General input/output (PC4) (initial value) (A4 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A4) (PC4 in single chip mode) |

- Bit 3—PC3 Mode (PC3MD): Selects the function of the PC3/A3 pin.


## Bit 3: PC3MD Description

| 0 | General input/output (PC3) (initial value) (A3 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A3) (PC3 in single chip mode) |

- Bit 2—PC2 Mode (PC2MD): Selects the function of the PC2/A2 pin.

Bit 2: PC2MD Description

| 0 | General input/output (PC2) (initial value) (A2 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A2) (PC2 in single chip mode) |

- Bit 1—PC1 Mode (PC1MD): Selects the function of the PC1/A1 pin.

Bit 1: PC1MD Description

| 0 | General input/output (PC1) (initial value) (A1 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (A1) (PC1 in single chip mode) |

- Bit 0—PC0 Mode (PC0MD): Selects the function of the PC0/A0 pin.

Bit 0: PCOMD Description

| 0 | General input/output (PCO) (initial value) (A0 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Address output (AO) (PCO in single chip mode) |

### 18.3.9 Port D I/O Register H (PDIORH)

The port D I/O register H (PDIORH) is a 16-bit read/write register that selects input or output for the most significant sixteen port D pins. Bits PD31IOR-PD16IOR correspond to the PD31/D31/ $\overline{\text { ADTRG }}$ pin to PD16/D16/IRQ0 pin. PDIORH is enabled when the port D pins function as general input/outputs (PD31-PD16). For other functions, it is disabled.

For port D pin functions PD31-PD16, a given pin in port D is an output pin if its corresponding PDIORH bit is set to 1 , and an input pin if the bit is cleared to 0 .

PDIORH is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

The settings for this register are effective only for the 144-pin version. There are no corresponding pins for this register in the 112 -pin and 120 -pin versions. However, read/writes are possible.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |$c$

### 18.3.10 Port D I/O Register L (PDIORL)

The port D I/O register L (PDIORL) is a 16-bit read/write register that selects input or output for the least significant sixteen port D pins. Bits PD15IOR-PD0IOR correspond to the PD15/D15 pin to PD0/D0 pin. PDIORL is enabled when the port D pins function as general input/outputs (PD15-PD0). For other functions, it is disabled.

For port D pin functions PD15-PD0, a given pin in port D is an output pin if its corresponding PDIORL bit is set to 1 , and an input pin if the bit is cleared to 0 .

PDIORL is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.


### 18.3.11 Port D Control Registers H1, H2 (PDCRH1 and PDCRH2)

PDCRH1 and PDCRH2 are 16-bit read/write registers that select the functions of the most significant sixteen multiplexed pins of port D. PDCRH1 selects the functions of the PD31/D31/ $\overline{\text { ADTRG }}-P D 24 / \mathrm{D} 24 / \overline{\mathrm{DREQ} 0}$ pins of port D; PDCRH2 selects the functions of the PD23/D23/쥬Q7-PD16/D16/IRQ0 pins of port D. There are instances when these register settings will be ignored, depending on the operation mode. Refer to table 18.2, Pin Arrangement by Mode, for details.

The settings for this register are effective only for the 144 -pin version. There are no corresponding pins for this register in the 112-pin and 120-pin versions. However, read/writes are possible.

PDCRH1 and PDCRH2 are both initialized to H'0000 by external power-on reset but are not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

## Port D Control Register H1 (PDCRH1):

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { PD31 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD31 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD30 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD30 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD29 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { PD29 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD28 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD28 } \\ & \text { MD0 } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 10 |  |
|  | $\begin{aligned} & \text { PD27 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD27 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD26 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD26 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD25 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD25 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD24 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD24 } \\ & \text { MD0 } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bits 15 and 14—PD31 Mode 1, 0 (PD31MD1 and PD31MD0): These bits select the function of the PD31/D31/ADTRG pin.

| Bit 15: <br> PD31MD1 | Bit 14: <br> PD31MD0 | Description |
| :--- | :--- | :--- |

- Bits 13 and 12—PD30 Mode 1, 0 (PD30MD1 and PD30MD0): These bits select the function of the PD30/D30/IRQOUT $p i n$.

| Bit 13: <br> PD30MD1 | Bit 12: <br> PD30MD0 | Description |
| :--- | :--- | :--- |

- Bits 11 and 10—PD29 Mode 1, 0 (PD29MD1 and PD29MD0): These bits select the function of the PD29/D29/CS3 pin.

| Bit 11: <br> PD29MD1 | Bit 10: <br> PD29MD0 | Description |
| :--- | :--- | :--- | | 0 | 0 | General input/output (PD29) (initial value) (D29 with no ROM <br> and CS0 $=32$ bit width) |
| :--- | :--- | :--- |
|  | 1 | Data input/output (D29) (PD29 in single chip mode) |
| 1 | 0 | Chip select output (CS3) (PD29 in single chip mode, D29 with <br> no ROM and CS0 $=32$ bit width) |
|  | 1 | Reserved |

- Bits 9 and 8—PD28 Mode 1, 0 (PD28MD1 and PD28MD0): These bits select the function of the PD28/D28/ $\overline{\mathrm{CS} 2}$ pin.

| Bit 9: <br> PD28MD1 | Bit 8: <br> PD28MD0 | Description |
| :--- | :--- | :--- |

- Bits 7 and 6-PD27 Mode 1, 0 (PD27MD1 and PD27MD0): These bits select the function of the PD27/D27/DACK1 pin.

| Bit 7: <br> PD27MD1 | Bit 6: <br> PD27MD0 | Description |
| :--- | :--- | :--- |$\quad$| 0 | 0 | General input/output (PD27) (initial value) (D27 with no ROM <br> and CSO $=32$ bit width) |
| :--- | :--- | :--- |
|  | 1 | Data input/output (D27) (PD27 in single chip mode) |
| 1 | 0 | DMA transfer request received output (DACK1) (PD27 in single <br> chip mode, and D27 with no ROM and CS0 $=32$ bit width) |
|  | 1 | Reserved |

- Bits 5 and 4—PD26 Mode 1, 0 (PD26MD1 and PD26MD0): These bits select the function of the PD26/D26/DACK0 pin.

| Bit 5: <br> PD26MD1 | Bit 4: <br> PD26MD0 | Description |
| :--- | :--- | :--- |

- Bits 3 and 2—PD25 Mode 1, 0 (PD25MD1 and PD25MD0): These bits select the function of the PD25/D25/DREQ1 $p$ in.

| Bit 3: <br> PD25MD1 | Bit 2: <br> PD25MD0 | Description |
| :--- | :--- | :--- |

- Bits 1 and 0—PD24 Mode 1, 0 (PD24MD1 and PD24MD0): These bits select the function of


| Bit 1: <br> PD24MD1 | Bit 0: <br> PD24MD0 | Description |
| :--- | :--- | :--- |

## Port D Control Register H2 (PDCRH2):

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { PD23 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD23 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD22 } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { PD22 } \\ & \text { MDO } \end{aligned}$ | $\begin{aligned} & \text { PD21 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD21 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD20 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD20 } \\ & \text { MD0 } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | $\begin{aligned} & \text { PD19 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD19 } \\ & \text { MDO } \end{aligned}$ | $\begin{aligned} & \text { PD18 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD18 } \\ & \text { MD0 } \end{aligned}$ | $\begin{aligned} & \text { PD17 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD17 } \\ & \text { MDO } \end{aligned}$ | $\begin{aligned} & \text { PD16 } \\ & \text { MD1 } \end{aligned}$ | $\begin{aligned} & \text { PD16 } \\ & \text { MD0 } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bits 15 and 14—PD23 Mode 1, 0 (PD23MD1 and PD23MD0): These bits select the function of the PD23/D23/IRQ7 pin.

| Bit 15: <br> PD23MD1 | Bit 14: <br> PD23MD0 | Description |
| :--- | :--- | :--- |

- Bits 13 and 12-PD22 Mode 1, 0 (PD22MD1 and PD22MD0): These bits select the function of the PD22/D22/IRQ6 pin.

| Bit 13: <br> PD22MD1 | Bit 12: <br> PD22MD0 | Description |
| :--- | :--- | :--- |

- Bits 11 and $10-\mathrm{PD} 21$ Mode 1, 0 (PD21MD1 and PD21MD0): These bits select the function of the PD21/D21/IRQ5 pin.

| Bit 11: <br> PD21MD1 | Bit 10: <br> PD21MD0 | Description |
| :--- | :--- | :--- |

- Bits 9 and 8—PD20 Mode 1, 0 (PD20MD1 and PD20MD0): These bits select the function of the PD20/D20/IRQ4 pin.

| Bit 9: <br> PD20MD1 | Bit 8: <br> PD20MD0 | Description |
| :--- | :--- | :--- |

- Bits 7 and 6-PD19 Mode 1, 0 (PD19MD1 and PD19MD0): These bits select the function of the PD19/D19/ $\overline{\text { IRQ3 }}$ pin.

| Bit 7: <br> PD19MD1 | Bit 6: <br> PD19MD0 | Description |
| :--- | :--- | :--- |

- Bits 5 and 4—PD18 Mode 1, 0 (PD18MD1 and PD18MD0): These bits select the function of the PD18/D18/IRQ2 pin.

| Bit 5: <br> PD18MD1 | Bit 4: <br> PD18MD0 | Description |
| :--- | :--- | :--- |

- Bits 3 and 2—PD17 Mode 1, 0 (PD17MD1 and PD17MD0): These bits select the function of the PD17/D17/IRQ1 pin.

| Bit 3: <br> PD17MD1 | Bit 2: <br> PD17MD0 | Description |
| :--- | :--- | :--- |

- Bits 1 and 0—PD16 Mode 1, 0 (PD16MD1 and PD16MD0): These bits select the function of the PD16/D16/IRQ0 pin.

| Bit 1: <br> PD16MD1 | Bit 0: <br> PD16MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PD16) (initial value) (D16 with no ROM <br> and CS0 $=32$ bit width) |
|  | 1 | Data input/output (D16) (PD16 in single chip mode) |
| 1 | 0 | Interrupt request input (IRQ0) |
| 1 | Reserved |  |

### 18.3.12 Port D Control Register L (PDCRL)

PDCRL is a 16-bit read/write register that selects the multiplexed pin functions for the least significant sixteen port D pins. There are instances when these register settings will be ignored, depending on the operation mode.

## On-Chip ROM-Disabled Extended Mode:

- 144-pin version:
— Mode 0 (16-bit bus): Port D pins are data I/O pins; PDCRL settings are disabled.
— Mode 1 (32-bit bus): Port D pins are data I/O pins; PDCRL settings are disabled.
- 112-pin and 120-pin versions:
— Mode 0 ( 8 -bit bus): Port D pins are data I/O pins; PDCRL settings are disabled.
- Mode 1 (16-bit bus): Port D pins are data I/O pins; PDCRL settings are disabled.

On-Chip ROM-Enabled Extended Mode: The port D pins are shared as data I/O pins and general I/O pins; PDCRL settings are enabled.

Single Chip Mode: The port D pins are general I/O pins; PDCRL settings are disabled.
PDCRL is initialized to H'0000 by external power-on reset but is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

## Port D Control Register L (PDCRL)

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { PD15 } \\ & \text { MD } \end{aligned}$ | $\begin{gathered} \text { PD14 } \\ \text { MD } \end{gathered}$ | $\begin{aligned} & \text { PD13 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD12 } \\ & \text { MD } \end{aligned}$ | $\begin{gathered} \text { PD11 } \\ \text { MD } \end{gathered}$ | $\begin{gathered} \text { PD10 } \\ \text { MD } \end{gathered}$ | $\begin{aligned} & \text { PD9 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD8 } \\ & \text { MD } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | $\begin{aligned} & \text { PD7 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD6 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD5 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD4 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD3 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD2 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PD1 } \\ & \text { MD } \end{aligned}$ | $\begin{aligned} & \text { PDO } \\ & \text { MD } \end{aligned}$ |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 15—PD15 Mode (PD15MD): Selects the function of the PD15/D15 pin.

Bit 15: PD15MD Description

| 0 | General input/output (PD15) (initial value) (D15 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D15) (PD15 in single chip mode) |

- Bit 14—PD14 Mode (PD14MD): Selects the function of the PD14/D14 pin.


## Bit 14: PD14MD Description

| 0 | General input/output (PD14) (initial value) (D14 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D14) (PD14 in single chip mode) |

- Bit 13—PD13 Mode (PD13MD): Selects the function of the PD13/D13 pin.

Bit 13: PD13MD Description

| 0 | General input/output (PD13) (initial value) (D13 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D13) (PD13 in single chip mode) |

- Bit 12—PD12 Mode (PD12MD): Selects the function of the PD12/D12 pin.

Bit 12: PD12MD Description

| 0 | General input/output (PD12) (initial value) (D12 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D12) (PD12 in single chip mode) |

- Bit 11—PD11 Mode (PD11MD): Selects the function of the PD11/D11 pin.

Bit 11: PD11MD Description

| 0 | General input/output (PD11) (initial value) (D11 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D11) (PD11 in single chip mode) |

- Bit 10—PD10 Mode (PD10MD): Selects the function of the PD10/D10 pin.

Bit 10: PD10MD Description

| 0 | General input/output (PD10) (initial value) (D10 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D10) (PD10 in single chip mode) |

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- Bit 9—PD9 Mode (PD9MD): Selects the function of the PD9/D9 pin.


## Bit 9: PD9MD Description

| 0 | General input/output (PD9) (initial value) (D9 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D9) (PD9 in single chip mode) |

- Bit 8—PD8 Mode (PD8MD): Selects the function of the PD8/D8 pin.


## Bit 8: PD8MD Description

| 0 | General input/output (PD8) (initial value) (D8 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D8) (PD8 in single chip mode) |

- Bit 7—PD7 Mode (PD7MD): Selects the function of the PD7/D7 pin.

| Bit 7: PD7MD | Description |
| :--- | :--- |
| 0 | General input/output (PD7) (initial value) (D7 in on-chip ROM invalid mode) |
| 1 | Data input/output (D7) (PD7 in single chip mode) |

- Bit 6—PD6 Mode (PD6MD): Selects the function of the PD6/D6 pin.


## Bit 6: PD6MD Description

| 0 | General input/output (PD6) (initial value) (D6 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D6) (PD6 in single chip mode) |

- Bit 5—PD5 Mode (PD5MD): Selects the function of the PD5/D5 pin.


## Bit 5: PD5MD Description

| 0 | General input/output (PD5) (initial value) (D5 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D5) (PD5 in single chip mode) |

- Bit 4—PD4 Mode (PD4MD): Selects the function of the PD4/D4 pin.


## Bit 4: PD4MD Description

$0 \quad$ General input/output (PD4) (initial value) (D4 in on-chip ROM invalid mode)
$1 \quad$ Data input/output (D4) (PD4 in single chip mode)

- Bit 3—PD3 Mode (PD3MD): Selects the function of the PD3/D3 pin.


## Bit 3: PD3MD Description

| 0 | General input/output (PD3) (initial value) (D3 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D3) (PD3 in single chip mode) |

- Bit 2—PD2 Mode (PD2MD): Selects the function of the PD2/D2 pin.


## Bit 2: PD2MD Description

| 0 | General input/output (PD2) (initial value) (D2 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D2) (PD2 in single chip mode) |

- Bit 1—PD1 Mode (PD1MD): Selects the function of the PD1/D1 pin.


## Bit 1: PD1MD Description

| 0 | General input/output (PD1) (initial value) (D1 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D1) (PD1 in single chip mode) |

- Bit 0—PD0 Mode (PD0MD): Selects the function of the PD0/D0 pin.

Bit 0: PDOMD Description

| 0 | General input/output (PDO) (initial value) (D0 in on-chip ROM invalid mode) |
| :--- | :--- |
| 1 | Data input/output (D0) (PD0 in single chip mode) |

### 18.3.13 Port E I/O Register (PEIOR)

The port E I/O register (PEIOR) is a 16-bit read/write register that selects input or output for the 16 port E pins. Bits PE15IOR-PE0IOR correspond to pins PE15/TIOC4D/DACK1/IRQOUT PE0/TIOC0A/DREQ0. PEIOR is enabled when the port E pins function as general input/outputs (PE15-PE0) or TIOC pin of the MTU. For other functions, it is disabled.

When the port E pin functions are as PE15-PE0, or TIOC pin of the MTU, a given pin in port E is an output pin if its corresponding PEIOR bit is set to 1 , and an input pin if the bit is cleared to 0 .

PEIOR is initialized to H'0000 by external power-on reset; however, it is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.


### 18.3.14 Port E Control Registers 1, 2 (PECR1 and PECR2)

PECR1 and PECR2 are 16-bit read/write registers that select the functions of the sixteen multiplexed pins of port E. PECR1 selects the functions of the upper eight bit pins of port E; PECR2 selects the function of the lower eight bit pins of port E .

Port E has a bus control signal $(\overline{\mathrm{AH}})$ and DMAC control signals (DACK1, DACK0, DRAK1, DRAK0), but there are instances when the register settings that select these pin functions will be ignored, depending on the operation mode. Refer to table 18.2, Pin Arrangement by Mode, for details.

PECR1 and PECR2 are both initialized to H'0000 by external power-on reset but are not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.

## Port E Control Register 1 (PECR1):

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PE15 <br> MD1 | PE15 MDO | PE14 <br> MD1 | PE14 MDO | PE13 <br> MD1 | PE13 MDO | - | PE12MD |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | - | PE11MD | - | PE10MD | - | PE9MD | - | PE8MD |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R/W | R | R/W | R | R/W | R | R/W |

- Bits 15 and 14—PE15 Mode 1, 0 (PE15MD1 and PE15MD0): These bits select the function of the PE15/TIOC4D/DACK1/IRQOUT pin.
\(\left.$$
\begin{array}{lll}\begin{array}{l}\text { Bit 15: } \\
\text { PE15MD1 }\end{array}
$$ \& \begin{array}{l}Bit 14: <br>

PE15MD0\end{array} \& Description\end{array}\right]\)| 0 | 0 | Input/output (PE15) (initial value) |
| :--- | :--- | :--- |
|  | 1 | MTU input capture input/output compare output (TIOC4D) <br> 1 |
|  | 0 | DMAC request received output (DACK1) (PE15 in single chip <br> mode) |
| 1 | Interrupt request output (IRQOUT) <br> (Reserved in single chip mode) |  |

- Bits 13 and 12-PE14 Mode 1, 0 (PE14MD1 and PE14MD0): These bits select the function of the PE14/TIOC4C/DACK0/AH pin.
\(\left.$$
\begin{array}{lll}\begin{array}{l}\text { Bit 13: } \\
\text { PE14MD1 }\end{array}
$$ \& \begin{array}{l}Bit 12: <br>

PE14MD0\end{array} \& Description\end{array}\right]\)| 0 | 0 | Input/output (PE14) (initial value) |
| :--- | :--- | :--- |
|  | 1 | MTU input capture input/output compare output (TIOC4C) |
| 1 | 0 | DMAC request received output (DACKO) (PE14 in single chip <br> mode) |
|  | 1 | Address hold output $(\overline{\mathrm{AH}})($ PE14 in single chip mode) |

- Bits 11 and 10—PE13 Mode 1, 0 (PE13MD1 and PE13MD0): These bits select the function of the PE13/TIOC4B/MRES pin.

| Bit 11: <br> PE13MD1 | Bit 10: <br> PE13MD0 | Description |
| :--- | :--- | :--- |
| 0 | 0 | General input/output (PE13) (initial value) |
|  | 1 | MTU input capture input/output compare output (TIOC4B) |
| 1 | 0 | Manual reset input ( $\overline{\text { MRES }})$ |
| 1 | Reserved |  |

- Bit 9—Reserved: This bit always reads as 0 . The write values should always be 0 .
- Bit 8—PE12 Mode (PE12MD): Selects the function of the PE12/TIOC4A pin.


## Bit 8: PE12MD Description

| 0 | General input/output (PE12) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC4A) |

- Bit 7—Reserved: This bit always reads as 0 . The write values should always be 0 .
- Bit 6—PE11 Mode (PE11MD): Selects the function of the PE11/TIOC3D pin.


## Bit 6: PE11MD Description

| 0 | General input/output (PE11) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC3D) |

- Bit 5—Reserved: This bit always reads as 0 . The write values should always be 0 .
- Bit 4—PE10 Mode (PE10MD): Selects the function of the PE10/TIOC3C pin.

Bit 4: PE10MD Description

| 0 | General input/output (PE10) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC3C) |

- Bit 3—Reserved: This bit always reads as 0 . The write values should always be 0 .
- Bit 2—PE9 Mode (PE9MD): Selects the function of the PE9/TIOC3B pin.

Bit 2: PE9MD Description

| 0 | General input/output (PE9) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC3B) |

- Bit $1 —$ Reserved: This bit always reads as 0 . The write values should always be 0 .
- Bit 0—PE8 Mode (PE8MD): Selects the function of the PE8/TIOC3A pin.

Bit 0: PE8MD Description

| 0 | General input/output (PE8) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC3A) |

Port E Control Register 2 (PECR2):

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 |  | 9 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 |  |  |  |  |  |  |  |  |
|  | - | PE7MD | - | PE6MD | - | PE5MD | - | PE4MD |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | $R$ | R/W | $R$ | $R / W$ | $R$ | $R / W$ | $R$ | R/W |


| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PE3 | PE3 | PE2 | PE2 | PE1 | PE1 | PE0 | PE0 |
|  | MD1 | MD0 | MD1 | MD0 | MD1 | MD0 | MD1 | MD0 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 15 -Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 14—PE7 Mode (PE7MD): Selects the function of the PE7/TIOC2B pin.


## Bit 14: PE7MD Description

| 0 | General input/output (PE7) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC2B) |

- Bit 13 -Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 12—PE6 Mode (PE6MD): Selects the function of the PE6/TIOC2A pin.


## Bit 12: PE6MD Description

| 0 | General input/output (PE6) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC2A) |

- Bit 11 -Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 10—PE5 Mode (PE5MD): Selects the function of the PE5/TIOC1B pin.


## Bit 10: PE5MD Description

| 0 | General input/output (PE5) (initial value) |
| :--- | :--- |
| 1 | MTU input capture input/output compare output (TIOC1B) |

- Bit 9—Reserved: This bit always reads as 0 . The write value should always be 0 .
- Bit 8—PE4 Mode (PE4MD): Selects the function of the PE4/TIOC1A pin.


## Bit 8: PE4MD Description

$0 \quad$ General input/output (PE4) (initial value)
1 MTU input capture input/output compare output (TIOC1A)

- Bits 7 and 6-PE3 Mode 1, 0 (PE3MD1 and PE3MD0): These bits select the function of the PE3/TIOC0D/DRAK1 pin.

| Bit 7: <br> PE3MD1 | Bit 6: <br> PE3MD0 | Description |
| :--- | :--- | :--- | | 0 | 0 | General input/output (PE3) (initial value) |
| :--- | :--- | :--- |
|  | 1 | MTU input capture input/output compare output (TIOC0D) |
| 1 | 0 | DREQ1 request received output (DRAK1) (PE3 in single chip <br> mode) |
|  | 1 | Reserved |

- Bits 5 and 4—PE2 Mode 1, 0 (PE2MD1 and PE2MD0): These bits select the function of the PE2/TIOC0C/DREQ1 pin.

| Bit 5: <br> PE2MD1 | Bit 4: <br> PE2MD0 | Description |
| :--- | :--- | :--- |

- Bits 3 and 2—PE1 Mode 1, 0 (PE1MD1 and PE1MD0): These bits select the function of the PE1/TIOC0B/DRAK0 pin.

| Bit 3: <br> PE1MD1 | Bit 2: <br> PE1MD0 | Description |
| :--- | :--- | :--- |

- Bits 1 and 0—PE0 Mode 1, 0 (PE0MD1 and PE0MD0): These bits select the function of the PE0/TIOC0A/ $\overline{\mathrm{DREQ} 0}$ pin.

| Bit 1: <br> PEOMD1 | Bit 0: <br> PEOMD0 | Description |
| :--- | :--- | :--- | | 0 | 0 | General input/output (PEO) (initial value) |
| :--- | :--- | :--- |
|  | 1 | MTU input capture input/output compare output (TIOCOA) |
| 1 | 0 | $\overline{\text { DREQ0 request receive input (PEO in single chip mode) }}$ |
|  | 1 | Reserved |

### 18.3.15 IRQOUT Function Control Register (IFCR)

The IFCR is a 16 -bit read/write register used to control output when the multiplexed pins are established as $\overline{\text { IRQOUT }}$ outputs by the port D control register (PDCRH1) or port E control register (PECR1). When PDCRH1 or PECR1 are set for any other function, the settings of this register have no effect on the pin functions.

The IFCR is initialized to $\mathrm{H}^{\prime} 0000$ by external power-on reset but is not initialized for manual resets, reset by WDT, standby mode, or sleep mode, so the previous data is maintained.


- Bits 3 and 2—IRQOUT Mode 3, 2 (IRQMD3 and IRQMD2): These bits select the $\overline{\text { IRQOUT }}$ pin function when the PDCRH1 bits 13 and 12 (PD30MD1, PD30MD0) are set to ( 1,0 ). These bit settings are effective only for the 144 pin version. Reads and writes are also possible in the 112-pin and 120-pin versions, but they have no effect on the pin functions.


## Bit 3: IRQMD3 Bit 2: IRQMD2 Description

| 0 | 0 | Interrupt request received output (initial value) |
| :--- | :--- | :--- |
|  | 1 | Refresh signal output |
| 1 | 0 | Interrupt request received, or refresh signal output <br> (which of the two is output depends on the operation <br> status at the time) |
|  | Always high level output |  |

- Bits 1 and 0—IRQOUT Mode 1, 0 (IRQMD1 and IRQMD0): These bits select the $\overline{\text { IRQOUT }}$ pin function when the PECR1 bits 1 and 0 (PE15MD1, PE15MD0) are set to (1, 1).

| Bit 1: IRQMD1 | Bit 0: IRQMDO | Description |
| :--- | :--- | :--- |
| 0 | 0 | Interrupt request received output (initial value) |
|  | 1 | Refresh signal output |
| 1 | 0 | Interrupt request received, or refresh signal output <br> (which of the two is output depends on the operation <br> status at the time) |
|  | 1 | Always high level output |

### 18.4 Cautions on Use

For the I/O ports and pins with multiplexing of DREQ or IRQ, switching from the port input Low level condition to IRQ or DREQ edge detection will detect the concerned edge.

## Section 19 I/O Ports (I/O)

### 19.1 Overview

There are six ports, A, B , C, D, E, and F. The pins of the ports are multiplexed for use as generalpurpose I/Os (the port F pins are general input) or for other functions. Use the pin function controller ( PFC ) to select the function of multiplexed pins. The ports each have one data register for storing pin data. The initialize function after power-on reset differs depending on the operating mode of each pin. See table 18.2, Pin Arrangement by Mode, for details.

### 19.2 Port A

There are two versions of port A:

- FP-112/TFP-120
- FP-144

In the FP-112 and TFP-120 versions, port A is a 16-pin input/output port, as listed in table 19.1.

Table 19.1 Port A, FP-112/TFP-120 Version

| ROM Disabled Extended Mode (Modes 0, 1) | ROM Enabled Extended Mode (Mode 2) | Single Chip Mode |
| :---: | :---: | :---: |
| PA15 (I/O)/CK (output) | PA15 (I/O)/CK (output) | PA15 (I/O)/CK (output) |
| $\overline{\mathrm{RD}}$ (output) | PA14 (I/O)/ $\overline{\mathrm{RD}}$ (output) | PA14 (I/O) |
| $\overline{\text { WRH (output) }}$ | PA13 (I/O)/ $\overline{\text { WRH }}$ (output) | PA13 (I/O) |
| $\overline{\text { WRL }}$ (output) | PA12 (I/O)/WRL (output) | PA12 (I/O) |
| $\overline{\text { CS1 }}$ (output) | PA11 (I/O)/CS1 (output) | PA11 (I/O) |
| $\overline{\mathrm{CSO}}$ (output) | PA10 (I/O)/CS0 (output) | PA10 (I/O) |
| PA9 (I/O)/TCLKD (input)//RQ3 (input) | PA9 (I/O)/TCLKD (input)//RQ3 (input) | PA9 (I/O)/TCLKD (input)//RQ3 (input) |
| PA8 (I/O)/TCLKC (input)/IRQ2 (input) | PA8 (I/O)/TCLKC (input)/IRQ2 (input) | PA8 (I/O)/TCLKC (input)/IRQ2 (input) |
| PA7 (I/O)/TCLKB (input)/ $\overline{\mathrm{CS3}}$ (output) | PA7 (I/O)/TCLKB (input)/ $\overline{\text { CS3 }}$ (output) | PA7 (I/O)/TCLKB (input) |
| PA6 (I/O)/TCLKA (input)/ $\overline{\mathrm{CS} 2}$ (output) | PA6 (I/O)/TCLKA (input)/ $\overline{\mathrm{CS} 2}$ (output) | PA6 (I/O)/TCLKA (input) |
| PA5 (I/O)/SCK1 (I/O)/DREQ1 (input)/IRQ1 (input) | PA5 (I/O)/SCK1 (I/O)/ $\overline{\text { DREQ1 }}$ (input)/IRQ1 (input) | PA5 (I/O)/SCK1 (I/O)//IRQ1 (input) |
| PA4 (I/O)/TXD1 (output) | PA4 (I/O)/TXD1 (output) | PA4 (I/O)/TXD1 (output) |
| PA3 (I/O)/RXD1 (input) | PA3 (I/O)/RXD1 (input) | PA3 (I/O)/RXD1 (input) |
| PA2 (I/O)/SCK0 (I/O)// $\overline{\text { DREQ0 }}$ (input)/IRQ0 (input) | PA2 (I/O)/SCK0 (I/O)/ $\overline{\text { DREQ0 }}$ (input)/IRQ0 (input) | $\begin{aligned} & \mathrm{PA2}(\mathrm{I} / \mathrm{O}) / \mathrm{SCKO}(\mathrm{I} / \mathrm{O}) / \overline{\mathrm{IRQO}} \\ & \text { (input) } \end{aligned}$ |
| PA1 (I/O)/TXD0 (output) | PA1 (I/O)/TXD0 (output) | PA1 (I/O)/TXD0 (output) |
| PA0 (I/O)/RXD0 (input) | PA0 (I/O)/RXD0 (input) | PA0 (I/O)/RXD0 (input) |

In the FP-144 version, port A is a 24 -pin input/output port, as listed in table 19.2.

Table 19.2 Port A, FP-144 Version

| ROM Disabled Extended Mode (Modes 0, 1) | ROM Enabled Extended Mode (Mode 2) | Single Chip Mode |
| :---: | :---: | :---: |
| $\overline{\text { WRHH }}$ (output) | PA23 (I/O)/ $\overline{\mathrm{WRHH}}$ (output) | PA23 (I/O) |
| $\overline{\text { WRHL }}$ (output) | PA22 (I/O) $\overline{\text { WRHL }}$ (output) | PA22 (I/O) |
| PA21 (I/O)/CASHH (output) | PA21 (I/O)/CASHH (output) | PA21 (I/O) |
| PA20 (I/O)//CASHL (output) | PA20 (I/O)/CASHL (output) | PA20 (I/O) |
| PA19 (I/O)/BACK (output)/ DRAK1 (output) | PA19 (I/O)/BACK (output)/ DRAK1 (output) | PA19 (I/O) |
| PA18 (I/O)/ $\overline{\text { BREQ }}$ (input)/ DRAKO (output) | PA18 (I/O)/ $\overline{\text { BREQ }}$ (input)/ DRAKO (output) | PA18 (I/O) |
| PA17 (I/O)/WAIT (input) | PA17 (I/O)/ $\overline{\text { WAIT }}$ (input) | PA17 (I/O) |
| PA16 (I/O)/ $\overline{\text { AH }}$ (output) | PA16 (I/O)/ $\overline{\text { AH }}$ (output) | PA16 (I/O) |
| PA15 (I/O)/CK (output) | PA15 (I/O)/CK (output) | PA15 (I/O)/CK (output) |
| $\overline{\mathrm{RD}}$ (output) | PA14 (I/O)/ $\overline{\mathrm{RD}}$ (output) | PA14 (I/O) |
| $\overline{\text { WRH (output) }}$ | PA13 (I/O) $\overline{\text { WRH (output) }}$ | PA13 (I/O) |
| $\overline{\text { WRL }}$ (output) | PA12 (I/O) $\overline{\text { WRL }}$ (output) | PA12 (I/O) |
| $\overline{\text { CS1 }}$ (output) | PA11 (I/O)//CS1 (output) | PA11 (I/O) |
| $\overline{\mathrm{CSO}}$ (output) | PA10 (I/O)/CS0 (output) | PA10 (I/O) |
| PA9 (I/O)/TCLKD (input)/IRQ3 (input) | PA9 (I/O)/TCLKD (input)//[RQ3 (input) | PA9 (I/O)/TCLKD (input)/IRQ3 (input) |
| PA8 (I/O)/TCLKC (input)//RQ2 (input) | PA8 (I/O)/TCLKC (input)//[RQ2 (input) | PA8 (I/O)/TCLKC (input)/IRQ2 (input) |
| PA7 (I/O)/TCLKB (input)/ $\overline{\text { CS3 }}$ (output) | PA7 (I/O)/TCLKB (input)/ $\overline{\text { CS3 }}$ (output) | PA7 (I/O)/TCLKB (input) |
| PA6 (I/O)/TCLKA (input)/ $\overline{\text { CS2 }}$ (output) | PA6 (I/O)/TCLKA (input)/ $\overline{\text { CS2 }}$ (output) | PA6 (I/O)/TCLKA (input) |
| PA5 (I/O)/SCK1 (I/O)/ㄷREQ1 (input)/IRQ1 (input) | PA5 (I/O)/SCK1 (I/O)/ㅁREQ1 (input)//RQ1 (input) | PA5 (I/O)/SCK1 (I/O)///RQ1 (input) |
| PA4 (I/O)/TXD1 (output) | PA4 (I/O)/TXD1 (output) | PA4 (I/O)/TXD1 (output) |
| PA3 (I/O)/RXD1 (input) | PA3 (I/O)/RXD1 (input) | PA3 (I/O)/RXD1 (input) |
| PA2 (I/O)/SCK0 (I/O)// $\overline{\text { DREQ0 }}$ (input)/IRQ0 (input) | PA2 (I/O)/SCK0 (I/O)//̄REQ0 (input)//RQ0 (input) | $\begin{aligned} & \mathrm{PA2}(\mathrm{I} / \mathrm{O}) / \mathrm{SCKO}(\mathrm{I} / \mathrm{O}) / \overline{\mathrm{RQQ}} \\ & \text { (input) } \end{aligned}$ |
| PA1 (I/O)/TXD0 (output) | PA1 (I/O)/TXD0 (output) | PA1 (I/O)/TXD0 (output) |
| PA0 (I/O)/RXD0 (input) | PA0 (I/O)/RXD0 (input) | PA0 (I/O)/RXD0 (input) |

### 19.2.1 Register Configuration

Table 19.3 summarizes the port A register.
Table 19.3 Port A Register

| Name Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: |
| Port A data register H PADRH | R/W | H'0000 | H'FFFF8380 H'FFFF8381 | 8, 16, 32 |
| Port A data register L PADRL | R/W | H'0000 | H'FFFF8382 H'FFFF8383 | 8,16, 32 |

### 19.2.2 Port A Data Register H (PADRH)

PADRH is a 16-bit read/write register that stores data for port A. The bits PA23DR-PA16DR correspond to the PA23/WRHH $-\mathrm{PA} 16 / \overline{\mathrm{AH}}$ pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PADRH; when PADRH is read, the register value will be output regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PADRH is read. When a value is written to PADRH, that value can be written into PADRH, but it will not affect the pin status. Table 19.4 shows the read/write operations of the port A data register.

PADRH is initialized by an external power-on reset. However, PADRH is not initialized for manual reset, reset by WDT, standby mode, or sleep mode.

These register settings function only for the 144-pin version. There are no pins corresponding to this register in the 112-pin version. However, read/writes are possible.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | - | - |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | R | R | R | R | R | R |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | PA23DR | PA22DR | PA21DR | PA20DR | PA19DR | PA18DR | PA17DR | PA16DR |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

### 19.2.3 Port A Data Register L (PADRL)

PADRL is a 16-bit read/write register that stores data for port A. The bits PA15DR-PA0DR correspond to the PA15/CK-PA0/RXD0 pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PADRL; when PADRL is read, the register value will be output regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PADRL is read. When a value is written to PADRL, that value can be written into PADRL, but it will not affect the pin status. Table 19.4 shows the read/write operations of the port A data register.

PADRL is initialized by an external power-on reset. However, PADRL is not initialized for manual reset, reset by WDT, standby mode, or sleep mode.


Table 19.4 Read/Write Operation of the Port A Data Register (PADR)

| PAIOR | Pin Status | Read | Write |
| :--- | :--- | :--- | :--- |
| 0 | Ordinary input | Pin status | Can write to PADR, but it has no effect on pin status |
|  | Other function | Pin status | Can write to PADR, but it has no effect on pin status |
| 1 | Ordinary output | PADR value | Value written is output by pin |
|  | Other function | PADR value | Can write to PADR, but it has no effect on pin status |

### 19.3 Port B

Port B is a 10-pin input/output port as listed in table 19.5.
Table 19.5 Port B

| ROM Disabled Extended Mode (Modes 0, 1) | ROM Enabled Extended Mode (Mode 2) | Single Chip Mode |
| :---: | :---: | :---: |
| PB9 (I/O)//쥬Q7 (input)/A21 (output)/ADTRG (input) | PB9 (I/O)//ㅈRQ7 (input)/A21 (output)/ADTRG (input) | PB9 (I/O)///RQ7 (input)/ $\overline{\text { ADTRG }}$ (input) |
| PB8 (I/O)//IRQ6 (input)/A20 (output)/WAIT (input) | PB8 (I/O)/ $\overline{\mathrm{RQ} 6}$ (input)/A20 (output)/WAIT (input) | PB8 (I/O)//]RQ6 (input) |
| PB7 (I/O)//IRQ5 (input)/A19 (output)/BREQ (input) | PB7 (I/O)/IRQ5 (input)/A19 (output)/BREQ (input) | PB7 (I/O)//]RQ5 (input) |
| PB6 (I/O)/IRQ4 (input)/A18 (output)/BACK (output) | PB6 (I/O)/IRQ4 (input)/A18 (output)/BACK (input) | PB6 (I/O)//IRQ4 (input) |
| PB5 (I/O)/ $\overline{\mathrm{RQ} 3}$ (input)/ $\overline{\text { POE3 }}$ (input)/RDWR (output) |  | $\begin{aligned} & \begin{array}{l} \text { PB5 (I/O)//ㅈRQ3 }(\text { input } / \overline{\mathrm{POE3}} \\ \text { (input) } \end{array} \end{aligned}$ |
| PB4 (I/O)/ $\overline{\mathrm{RQ} 2}$ (input)// $\overline{\text { POE2 }}$ (input)/CASH (output) | $\begin{aligned} & \text { PB4 (I/O)/ } \overline{\mathrm{RQQ} 2} \text { (input)// } \overline{\mathrm{POE} 2} \\ & \text { (input)/ } / \overline{\mathrm{CASH}} \text { (output) } \end{aligned}$ | (input) |
| PB3 (I/O)///RQ1 (input)//POE1 (input)/CASL (output) |  |  |
| PB2 (I/O)/ $\overline{\mathrm{RQQ0}}$ (input)// $\overline{\text { POE0 }}$ (input)/RAS (output) | PB2 (I/O)// $\overline{\mathrm{RQO}}$ (input)/ $\overline{\text { POE0 }}$ (input)/RAS (output) | PB2 (I/O)///RQ0 (input)// $\overline{\text { POE0 }}$ (input) |
| A17 (output) | PB1 (I/O)/A17 (output) | PB1 (I/O) |
| A16 (output) | PB0 (I/O)/A16 (output) | PB0 (I/O) |

### 19.3.1 Register Configuration

Table 19.6 summarizes the port B register.
Table 19.6 Port B Register

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Port B data register | PBDR | R/W | H'0000 | H'FFFF8390 | 8, 16, 32 |
|  |  |  |  | H'FFFF8391 |  |

### 19.3.2 Port B Data Register (PBDR)

PBDR is a 16-bit read/write register that stores data for port B. The bits PB9DR-PB0DR correspond to the PB9/ $\overline{\mathrm{IRQ} 7} / \mathrm{A} 21 / \overline{\mathrm{ADTRG}}-\mathrm{PB} 0 / \mathrm{A} 16$ pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PBDR; when PBDR is read, the register value will be read regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PBDR is read. When a value is written to PBDR, that value can be written into PBDR, but it will not affect the pin status. Table 19.7 shows the read/write operations of the port B data register.

PBDR is initialized by an external power-on reset. However, PBDR is not initialized for a manual reset, reset by WDT, standby mode, or sleep mode.


Table 19.7 Read/Write Operation of the Port B Data Register (PBDR)

| PBIOR | Pin Status | Read | Write |
| :--- | :--- | :--- | :--- |
| 0 | Ordinary input | Pin status | Can write to PBDR, but it has no effect on pin status |
|  | Other function | Pin status | Can write to PBDR, but it has no effect on pin status |
| 1 | Ordinary output | PBDR value | Value written is output by pin |
|  | Other function | PBDR value | Can write to PBDR, but it has no effect on pin status |

### 19.4 Port C

Port C is a 16 pin input/output port as listed in table 19.8.
Table 19.8 Port C

| ROM Disabled Extended <br> Mode (Modes 0, 1) | ROM Enabled Extended <br> Mode (Mode 2) | Single Chip Mode |
| :--- | :--- | :--- |
| A15 (output) | PC15 (I/O)/A15 (output) | PC15 (I/O) |
| A14 (output) | PC14 (I/O)/A14 (output) | PC14 (I/O) |
| A13 (output) | PC13 (I/O)/A13 (output) | PC13 (I/O) |
| A12 (output) | PC12 (I/O)/A12 (output) | PC12 (I/O) |
| A11 (output) | PC11 (I/O)/A11 (output) | PC11 (I/O) |
| A10 (output) | PC10 (I/O)/A10 (output) | PC10 (I/O) |
| A9 (output) | PC9 (I/O)/A9 (output) | PC9 (I/O) |
| A8 (output) | PC8 (I/O)/A8 (output) | PC8 (I/O) |
| A7 (output) | PC7 (I/O)/A7 (output) | PC7 (I/O) |
| A6 (output) | PC6 (I/O)/A6 (output) | PC6 (I/O) |
| A5 (output) | PC5 (I/O)/A5 (output) | PC5 (I/O) |
| A4 (output) | PC4 (I/O)/A4 (output) | PC4 (I/O) |
| A3 (output) | PC3 (I/O)/A3 (output) | PC3 (I/O) |
| A2 (output) | PC2 (I/O)/A2 (output) | PC2 (I/O) |
| A1 (output) | PC1 (I/O)/A1 (output) | PC1 (I/O) |
| A0 (output) | PC0 (I/O)/A0 (output) | PC0 (I/O) |
|  |  |  |

### 19.4.1 Register Configuration

Table 19.9 summarizes the port C register.
Table 19.9 Port C Register

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Port C data register | PCDR | R/W | H'0000 | H'FFFF8392 | $8,16,32$ |
|  |  |  |  | H'FFFF8393 $^{\prime}$ |  |

### 19.4.2 Port C Data Register (PCDR)

PCDR is a 16-bit read/write register that stores data for port C. The bits PC15DR-PC0DR correspond to the PC15/A15-PC0/A0 pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PCDR; when PCDR is read, the register value will be read regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PCDR is read. When a value is written to PCDR, that value can be written into PCDR, but it will not affect the pin status. Table 19.10 shows the read/write operations of the port C data register.

PCDR is initialized by an external power-on reset. However, PCDR is not initialized for a manual reset, reset by WDT, standby mode, or sleep mode.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PC15DR | PC14DR | PC13DR | PC12DR | PC11DR | PC10DR | PC9DR | PC8DR |
|  | Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  |  |  |  |  |  |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | PC7DR | PC6DR | PC5DR | PC4DR | PC3DR | PC2DR | PC1DR | PCODR |

Table 19.10 Read/Write Operation of the Port C Data Register (PCDR)

| PCIOR | Pin Status | Read | Write |
| :--- | :--- | :--- | :--- |
| 0 | Ordinary input | Pin status | Can write to PCDR, but it has no effect on pin status |
|  | Other function | Pin status | Can write to PCDR, but it has no effect on pin status |
| 1 | Ordinary output | PCDR value | Value written is output by pin |
|  | Other function | PCDR value | Can write to PCDR, but it has no effect on pin status |

### 19.5 Port D

There are two versions of port D:

- FP-112
- FP-144

In the FP-112 version, port D is a 16-pin input/output port, as shown in table 19.11.
Table 19.11 Port D, FP-112 Version

| Extended Mode <br> Without ROM <br> (Mode 0) | Extended Mode <br> Without ROM <br> (Mode 1) | Extended Mode With <br> ROM (Mode 2) | Single Chip Mode |
| :--- | :--- | :--- | :--- |
| D15 (I/O) | D15 (I/O) | PD15 (I/O)/D15 (I/O) | PD15 (I/O) |
| D14 (I/O) | D14 (I/O) | PD14 (I/O)/D14 (I/O) | PD14 (I/O) |
| D13 (I/O) | D13 (I/O) | PD13 (I/O)/D13 (I/O) | PD13 (I/O) |
| D12 (I/O) | D12 (I/O) | PD12 (I/O)/D12 (I/O) | PD12 (I/O) |
| D11 (I/O) | D11 (I/O) | PD11 (I/O)/D11 (I/O) | PD11 (I/O) |
| D10 (I/O) | D10 (I/O) | PD10 (I/O)/D10 (I/O) | PD10 (I/O) |
| D9 (I/O) | D9 (I/O) | PD9 (I/O)/D9 (I/O) | PD9 (I/O) |
| D8 (I/O) | D8 (I/O) | PD8 (I/O)/D8 (I/O) | PD8 (I/O) |
| D7 (I/O) | D7 (I/O) | PD7 (I/O)/D7 (I/O) | PD7 (I/O) |
| D6 (I/O) | D6 (I/O) | PD6 (I/O)/D6 (I/O) | PD6 (I/O) |
| D5 (I/O) | D5 (I/O) | PD5 (I/O)/D5 (I/O) | PD5 (I/O) |
| D4 (I/O) | D4 (I/O) | PD4 (I/O)/D4 (I/O) | PD4 (I/O) |
| D3 (I/O) | D3 (I/O) | PD3 (I/O)/D3 (I/O) | PD3 (I/O) |
| D2 (I/O) | D2 (I/O) | PD2 (I/O)/D2 (I/O) | PD2 (I/O) |
| D1 (I/O) | D1 (I/O) | PD1 (I/O)/D1 (I/O) | PD1 (I/O) |
| D0 (I/O) | D0 (I/O) | PD0 (I/O)/D0 (I/O) | PD0 (I/O) |

In the FP-144 version, port D is a 32-pin input/output port, as listed in table 19.12.

Table 19.12 Port D, FP-144 Version

| Extended Mode Without ROM (Mode 0) | Extended Mode Without ROM (Mode 1) | Extended Mode With ROM (Mode 2) | Single Chip Mode |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { PD31 (I/O)/D31 (I/O)/ } \\ & \hline \text { ADTRG } \text { (input) } \end{aligned}$ | D31 (I/O) | PD31 (I/O)/D31 (I/O)/ <br> ADTRG (input) | PD31 (I/O)// $\overline{\text { ADTRG }}$ (input) |
| $\begin{aligned} & \text { PD30 (I/O)/D30 (I/O)/ } \\ & \hline \text { IRQOUT (output) } \end{aligned}$ | D30 (1/O) | $\begin{aligned} & \text { PD30 (I/O)/D30 (I/O)/ } \\ & \hline \text { RQOUT (output) } \end{aligned}$ | PD30 (I/O)///RQOUT (output) |
| PD29 (I/O)/D29 (I/O)/ $\overline{\mathrm{CS} 3}$ (output) | D29 (I/O) | PD29 (I/O)/D29 (I/O)/ $\overline{\mathrm{CS} 3}$ (output) | PD29 (I/O) |
| PD28 (I/O)/D28 (I/O)/ $\overline{\mathrm{CS} 2}$ (output) | D28 (I/O) | PD28 (I/O)/D28 (I/O)/ $\overline{\mathrm{CS} 2}$ (output) | PD28 (I/O) |
| PD27 (I/O)/D27 (I/O)/ DACK1 (output) | D27 (I/O) | PD27 (I/O)/D27 (I/O)/ DACK1 (output) | PD27 (I/O) |
| PD26 (I/O)/D26 (I/O)/ DACK0 (output) | D26 (I/O) | PD26 (I/O)/D26 (I/O)/ DACK0 (output) | PD26 (I/O) |
| PD25 (I/O)/D25 (I/O)/ DREQ1 (input) | D25 (I/O) | $\begin{aligned} & \text { PD25 (I/O)/D25 (I/O)/ } \\ & \hline \text { DREQ1 (input) } \end{aligned}$ | PD25 (I/O) |
| $\begin{aligned} & \text { PD24 (I/O)/D24 (I/O)/ } \\ & \hline \text { DREQ0 (input) } \end{aligned}$ | D24 (I/O) | $\frac{\text { PD24 (I/O)/D24 (I/O)/ }}{\text { DREQ0 (input) }}$ | PD24 (I/O) |
| $\frac{\text { PD23 (I/O)/D23 (I/O)/ }}{\text { IRQ7 (input) }}$ | D23 (1/O) | $\begin{aligned} & \text { PD23 }(\mathrm{I} / \mathrm{O}) / \mathrm{D} 23(\mathrm{I} / \mathrm{O}) / \\ & \hline \text { RQ7 } \\ & \text { (input) } \end{aligned}$ | $\begin{aligned} & \mathrm{PD23}(\mathrm{I} / \mathrm{O}) / \overline{\mathrm{RQ} 7} \\ & \text { (input) } \end{aligned}$ |
| $\begin{aligned} & \text { PD22 (I/O)/D22 (I/O)/ } \\ & \hline \text { IRQ6 (input) } \end{aligned}$ | D22 (I/O) | $\begin{aligned} & \text { PD22 }(\mathrm{I} / \mathrm{O}) / \mathrm{D} 22(\mathrm{I} / \mathrm{O}) / \\ & \hline \text { RQ6 } \text { (input) } \end{aligned}$ | $\begin{aligned} & \mathrm{PD22}(\mathrm{I} / \mathrm{O}) / \overline{\mathrm{IRQ6}} \\ & \text { (input) } \end{aligned}$ |
| $\begin{aligned} & \text { PD21 (I/O)/D21 (I/O)/ } \\ & \hline \text { RQ5 (input) } \end{aligned}$ | D21 (I/O) | PD21 (I/O)/D21 (I/O)/ $\overline{\mathrm{RQ} 5}$ (input) | PD21 (I/O)//ㅈRQ5 (input) |
| $\begin{aligned} & \text { PD20 (I/O)/D20 (I/O)/ } \\ & \hline \text { IRQ4 (input) } \end{aligned}$ | D20 (1/O) | $\frac{\mathrm{PD} 20}{} \frac{(\mathrm{I} / \mathrm{O}) / \mathrm{D} 20(\mathrm{I} / \mathrm{O}) /}{\text { IRQ4 (input) }}$ | $\begin{aligned} & \mathrm{PD20}(\mathrm{I} / \mathrm{O}) / \overline{\mathrm{IRQ} 4} \\ & \text { (input) } \end{aligned}$ |
| PD19 (I/O)/D19 (I/O)/ IRQ3 (input) | D19 (I/O) | PD19 (I/O)/D19 (I/O)/ $\overline{\mathrm{RQ} 3}$ (input) | PD19 (I/O)//IRQ3 (input) |
| $\begin{aligned} & \text { PD18 (I/O)/D18 (I/O)/ } \\ & \hline \text { IRQ2 (input) } \end{aligned}$ | D18 (1/O) | $\begin{aligned} & \text { PD18 (I/O)/D18 (I/O)/ } \\ & \hline \text { IRQ2 (input) } \end{aligned}$ | PD18 (I/O)//ㅈRQ2 (input) |
| PD17 (I/O)/D17 (I/O)/ $\overline{\text { IRQ1 }}$ (input) | D17 (I/O) | PD17 (I/O)/D17 (I/O)/ $\overline{\mathrm{RQ} 1}$ (input) | PD17 (I/O)//RQ1 (input) |
| $\begin{aligned} & \text { PD16 }(\mathrm{I} / \mathrm{O}) / \mathrm{D} 16(\mathrm{I} / \mathrm{O}) / \\ & \hline \text { IRQ0 } \text { (input) } \end{aligned}$ | D16 (1/O) | $\begin{aligned} & \text { PD16 }(\mathrm{I} / \mathrm{O}) / \mathrm{D} 16(\mathrm{I} / \mathrm{O}) / \\ & \hline \text { RQ0 } \\ & \text { (input) } \end{aligned}$ | PD16 (I/O)//IRQ0 (input) |

Table 19.12 Port D, FP-144 Version (cont)

| Extended Mode <br> Without ROM <br> (Mode 0) | Extended Mode <br> Without ROM <br> (Mode 1) | Extended Mode With <br> ROM (Mode 2) | Single Chip Mode |
| :--- | :--- | :--- | :--- |
| D15 (I/O) | D15 (I/O) | PD15 (I/O)/D15 (I/O) | PD15 (I/O) |
| D14 (I/O) | D14 (I/O) | PD14 (I/O)/D14 (I/O) | PD14 (I/O) |
| D13 (I/O) | D13 (I/O) | PD13 (I/O)/D13 (I/O) | PD13 (I/O) |
| D12 (I/O) | D12 (I/O) | PD12 (I/O)/D12 (I/O) | PD12 (I/O) |
| D11 (I/O) | D11 (I/O) | PD11 (I/O)/D11 (I/O) | PD11 (I/O) |
| D10 (I/O) | D10 (I/O) | PD10 (I/O)/D10 (I/O) | PD10 (I/O) |
| D9 (I/O) | D9 (I/O) | PD9 (I/O)/D9 (I/O) | PD9 (I/O) |
| D8 (I/O) | D8 (I/O) | PD8 (I/O)/D8 (I/O) | PD8 (I/O) |
| D7 (I/O) | D7 (I/O) | PD7 (I/O)/D7 (I/O) | PD7 (I/O) |
| D6 (I/O) | D6 (I/O) | PD6 (I/O)/D6 (I/O) | PD6 (I/O) |
| D5 (I/O) | D5 (I/O) | PD5 (I/O)/D5 (I/O) | PD5 (I/O) |
| D4 (I/O) | D4 (I/O) | PD4 (I/O)/D4 (I/O) | PD4 (I/O) |
| D3 (I/O) | D3 (I/O) | PD3 (I/O)/D3 (I/O) | PD3 (I/O) |
| D2 (I/O) | D2 (I/O) | PD2 (I/O)/D2 (I/O) | PD2 (I/O) |
| D1 (I/O) | D1 (I/O) | PD1 (I/O)/D1 (I/O) | PD1 (I/O) |
| D0 (I/O) | D0 (I/O) | PD0 (I/O)/D0 (I/O) | PD0 (I/O) |

### 19.5.1 Register Configuration

Table 19.13 summarizes the port D register.
Table 19.13 Port D Register

| Name Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: |
| Port D data register H PDDRH | R/W | H'0000 | H'FFFF83A0 H'FFFF83A1 | 8, 16, 32 |
| Port D data register L PDDRL | R/W | H'0000 | H'FFFF83A2 H'FFFF83A3 | 8,16, 32 |

### 19.5.2 Port D Data Register H (PDDRH)

PDDRH is a 16 -bit read/write register that stores data for port D . The bits PD31DR-PD16DR correspond to the PD31/D31/㧱TRG-PD16/D16/ $\overline{\mathrm{IRQ} 0}$ pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PDDRH; when PDDRH is read, the register value will be read regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PDDRH is read. When a value is written to PDDRH, that value can be written into PDDRH, but it will not affect the pin status. Table 19.14 shows the read/write operations of the port D data register.

PDDRH is initialized by an external power-on reset. However, PDDRH is not initialized for a manual reset, reset by WDT, standby mode, or sleep mode.

These register settings function only for the 144-pin version. There are no pins corresponding to this register in the 112-pin version. However, read/writes are possible.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PD31DR | PD30DR | PD29DR | PD28DR | PD27DR | PD26DR | PD25DR | PD24DR |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | PD23DR | PD22DR | PD21DR | PD20DR | PD19DR | PD18DR | PD17DR | PD16DR |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

### 19.5.3 Port D Data Register L (PDDRL)

PDDRL is a 16-bit read/write register that stores data for port D. The bits PD15DR-PD0DR correspond to the PD15/D15-PD0/D0 pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PDDRL; when PDDRL is read, the register value will be read regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PDDRL is read. When a value is written to PDDRL, that value can be written into PDDRL, but it will not affect the pin status. Table 19.14 shows the read/write operations of the port D data register.

PDDRL is initialized by an external power-on reset. However, PDDRL is not initialized for a manual reset, reset by WDT, standby mode, or sleep mode.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PD15DR | PD14DR | PD13DR | PD12DR | PD11DR | PD10DR | PD9DR | PD8DR |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  |  |  |  |  |  |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | PD7DR | PD6DR | PD5DR | PD4DR | PD3DR | PD2DR | PD1DR | PD0DR |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Table 19.14 Read/Write Operation of the Port D Data Register (PDDR)

| PDIOR | Pin Status | Read | Write |
| :--- | :--- | :--- | :--- |
| 0 | Ordinary input | Pin status | Can write to PDDR, but it has no effect on pin status |
|  | Other function | Pin status | Can write to PDDR, but it has no effect on pin status |
| 1 | Ordinary output | PDDR value | Value written is output by pin |
|  | Other function | PDDR value | Can write to PDDR, but it has no effect on pin status |

### 19.6 Port E

Port E is a 16-pin input/output port, as listed in table 19.15.
Table 19.15 Port E

| Extended Modes (Modes 0, 1, 2) | Single Chip Mode |
| :--- | :--- |
| PE15 (I/O)/TIOC4D (I/O)/DACK1 <br> (output)/IRQOUT (output) | PE15 (I/O)/TIOC4D (I/O)//RQOUT (output) |
| PE14 (I/O)/TIOC4C (I/O)/DACK0 (output)/AH <br> (output) | PE14 (I/O)/TIOC4C (I/O) |
| PE13 (I/O)/TIOC4B (I/O)/MRES (input) | PE13 (I/O)/TIOC4B (I/O)/MRES (input) |
| PE12 (I/O)/TIOC4A (I/O) | PE12 (I/O)/TIOC4A (I/O) |
| PE11 (I/O)/TIOC3D (I/O) | PE11 (I/O)/TIOC3D (I/O) |
| PE10 (I/O)/TIOC3C (I/O) | PE10 (I/O)/TIOC3C (I/O) |
| PE9 (I/O)/TIOC3B (I/O) | PE9 (I/O)/TIOC3B (I/O) |
| PE8 (I/O)/TIOC3A (I/O) | PE8 (I/O)/TIOC3A (I/O) |
| PE7 (I/O)/TIOC2B (I/O) | PE7 (I/O)/TIOC2B (I/O) |
| PE6 (I/O)/TIOC2A (I/O) | PE6 (I/O)/TIOC2A (I/O) |
| PE5 (I/O)/TIOC1B (I/O) | PE5 (I/O)/TIOC1B (I/O) |
| PE4 (I/O)/TIOC1A (I/O) | PE4 (I/O)/TIOC1A (I/O) |
| PE3 (I/O)/TIOC0D (I/O)/DRAK1 (output) | PE3 (I/O)/TIOC0D (I/O) |
| PE2 (I/O)/TIOC0C (I/O)/ $\overline{\text { DREQ1 }(\text { input })}$ | PE2 (I/O)/TIOC0C (I/O) |
| PE1 (I/O)/TIOC0B (I/O)/DRAKO (output) | PE1 (I/O)/TIOC0B (I/O) |
| PE0 (I/O)/TIOC0A (I/O)/ $\overline{\text { DREQ0 (input) }}$ | PE0 (I/O)/TIOC0A (I/O) |

### 19.6.1 Register Configuration

Table 19.16 summarizes the port E register.
Table 19.16 Port E Register

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Port E data register | PEDR | R/W | H'0000 | H'FFFF83B0 | 8, 16, 32 |
|  |  |  |  | H'FFFF83B1 $^{2}$ |  |

### 19.6.2 Port E Data Register (PEDR)

PEDR is a 16-bit read/write register that stores data for port E. The bits PE15DR-PE0DR correspond to the PE15/TIOC4D/DACK1/IRQOUT $-\mathrm{PE} 0 / \mathrm{TIOC} 0 \mathrm{~A} / \overline{\mathrm{DREQ} 0}$ pins. When the pins are used as ordinary outputs, they will output whatever value is written in the PEDR; when PEDR is read, the register value will be read regardless of the pin status. When the pins are used as ordinary inputs, the pin status rather than the register value is read directly when PEDR is read. When a value is written to PEDR, that value can be written into PEDR, but it will not affect the pin status. Table 19.17 shows the read/write operations of the port E data register.

PEDR is initialized by a external power-on reset. However, PEDR is not initialized for a manual reset, reset by WDT, standby mode, or sleep mode, so the previous data is retained.

| Bit: | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PE15DR | PE14DR | PE13DR | PE12DR | PE11DR | PE10DR | PE9DR | PE8DR |
|  | Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
|  |  |  |  |  |  |  |  |  |
| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|  | PE7DR | PE6DR | PE5DR | PE4DR | PE3DR | PE2DR | PE1DR | PE0DR |

Table 19.17 Read/Write Operation of the Port E Data Register (PEDR)

| PEIOR | Pin Status | Read | Write |
| :--- | :--- | :--- | :--- |
| 0 | Ordinary input | Pin status | Can write to PEDR, but it has no effect on pin status |
|  | Other function | Pin status | Can write to PEDR, but it has no effect on pin status |
| 1 | Ordinary output | PEDR value | Value written is output by pin |
|  | Other function | PEDR value | Can write to PEDR, but it has no effect on pin status |

### 19.7 Port F

Port F is an 8 -pin input port. All modes are configured in the following way:

- PF7 (input)/AN7 (input)
- PF6 (input)/AN6 (input)
- PF5 (input)/AN5 (input)
- PF4 (input)/AN4 (input)
- PF3 (input)/AN3 (input)
- PF2 (input)/AN2 (input)
- PF1 (input)/AN1 (input)
- PF0 (input)/AN0 (input)


### 19.7.1 Register Configuration

Table 19.18 summarizes the port F register.
Table 19.18 Port F Register

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Port F data register | PFDR | R | External pin <br> dependent | H'FFFF83B3 | 8 |

### 19.7.2 Port F Data Register (PFDR)

PFDR is an 8-bit read-only register that stores data for port F . The bits PF7DR-PF0DR correspond to the PF7/AN7-PF0/AN0 pins. There are no bits $15-8$, so always access as eight bits. Any value written into these bits is ignored, and there is no effect on the status of the pins. When any of the bits are read, the pin status rather than the bit value is read directly. However, when an A/D converter analog input is being sampled, values of 1 are read out. Table 19.19 shows the read/write operations of the port F data register.

PFDR is not initialized by power-on resets, manual resets, standby mode, or sleep mode (the bits always reflect the pin status).

| Bit: | 7 | 6 | 5 | 43 |  | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PF7DR | PF6DR | PF5DR | PF4DR | PF3DR | PF2DR | PF1DR | PFODR |
| Initial value: | * | * | * | * | * | * | * | * |
| R/W: | R | R | R | R | R | R | R | R |

Note: * Initial values are dependent on the status of the pins at the time of the reads.

Table 19.19 Read/Write Operation of the Port F Data Register (PFDR)

| Pin I/O | Pin Function | Read | Write |
| :--- | :--- | :--- | :--- |
| Input | Ordinary | Pin status is read | Ignored (no effect on pin status) |
|  | ANn: analog input | 1 is read | Ignored (no effect on pin status) |
| $\mathrm{n}=7-0$ |  |  |  |

## Section 20 64/128/256kB Mask ROM

### 20.1 Overview

This LSI is available with 64 kbytes, 128 kbytes, or 256 kbytes of on-chip ROM. The on-chip ROM is connected to the CPU, direct memory access controller (DMAC) and data transfer controller (DTC) through a 32 -bit data bus (figures 20.1, 20.2, and 20.3). The CPU, DMAC, and DTC can access the on-chip ROM in 8,16 , and 32-bit widths. Data in the on-chip ROM can always be accessed in one cycle.


Figure 20.1 Mask ROM Block Diagram (64-kbyte Version)


Figure 20.2 Mask ROM Block Diagram (128-kbyte Version)


Figure 20.3 Mask ROM Block Diagram (256-kbyte Version)

The operating mode determines whether the on-chip ROM is valid or not. The operating mode is selected using mode-setting pins MD3-MD0 as shown in table 20.1. If you are using the on-chip ROM, select mode 2 or mode 3 ; if you are not, select mode 0 or 1 . The on-chip ROM is allocated to addresses $\mathrm{H}^{\prime} 00000000-\mathrm{H}^{\prime} 0000 \mathrm{FFFF}$ of memory area 0 for the 64-kbyte version, $\mathrm{H}^{\prime} 00000000$ $H^{\prime} 0001$ FFFF of memory area 0 for the 128-kbyte version and H'00000000-H'0003FFFF of memory area 0 for the 256-kbyte version.

Table 20.1 Operation Modes and ROM

|  | Mode Setting Pin |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Operation Mode | MD3 MD2 MD1 MD0 | Area 0 |  |  |

## Section 21 128kB PROM

### 21.1 Overview

This LSI has 128 kbytes of on-chip PROM.The on-chip ROM is connected to the CPU, the direct memory access controller (DMAC) and the data transfer controller (DTC) through a 32-bit data bus (figures 21.1). The CPU, DMAC, and DTC can access the on-chip ROM in 8, 16, and 32-bit widths. Data in the on-chip ROM can always be accessed in one cycle.


Figure 21.1 PROM Block Diagram
The operating mode determines whether the on-chip ROM is valid or not. The operating mode is selected using mode-setting pins MD3-MD0 as shown in table 21.1. If you are using the on-chip ROM, select mode 2 or mode 3 ; if you are not, select mode 0 or 1 . The on-chip ROM is allocated to addresses $\mathrm{H}^{\prime} 00000000-\mathrm{H}^{\prime} 0001 \mathrm{FFFF}$ of memory area 0 .

Table 21.1 Operating Modes and ROM

|  | Mode Setting Pin |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Operating Mode | MD3 MD2 MD1 MD0 Area 0 |  |  |

With the PROM version, programs can be written in the same manner as with an ordinary EPROM by setting the LSI to PROM mode and using a standard EPROM writer.

### 21.2 PROM Mode

### 21.2.1 PROM Mode Settings

When programming the on-chip PROM, set the pins as shown in figure 21.2, 21.3, or 21.4, and perform the programming in PROM mode.

### 21.2.2 Socket Adapter Pin Correspondence and Memory Map

Connect the socket adapter to the SH7040 series chip as shown in figure 21.2 or 21.3. This will allow the on-chip PROM to be programmed in the same manner as an ordinary 32-pin EPROM (HN27C101). Figures 21.2, 21.3, and 21.4 show the correspondence between the SH7040 Series pins and HN27C101 pins. Figure 21.5 is a memory map of the on-chip ROM.

| SH7042(112-pin version) |  | EPROM sock adapter$\qquad$ | et HN27C101 |  |
| :---: | :---: | :---: | :---: | :---: |
| Pin number | Pin name |  | Pin name | Pin number |
| 84 | $\overline{\mathrm{RES}} / \mathrm{V}_{\mathrm{PP}}$ |  | $\mathrm{V}_{\mathrm{PP}}$ | 1 |
| 76 | NMI |  | A9 | 26 |
| 70 | PD0/D0 |  | 1/O0 | 13 |
| 69 | PD1/D1 |  | I/O1 | 14 |
| 68 | PD2/D2 |  | I/O2 | 15 |
| 67 | PD3/D3 |  | 1/O3 | 17 |
| 66 | PD4/D4 |  | I/O4 | 18 |
| 64 | PD5/D5 |  | I/O5 | 19 |
| 63 | PD6/D6 |  | I/O6 | 20 |
| 62 | PD7/D7 |  | I/O7 | 21 |
| 4 | PC0/A0 |  | A0 | 12 |
| 5 | PC1/A1 |  | A1 | 11 |
| 6 | PC2/A2 |  | A2 | 10 |
| 7 | PC3/A3 |  | A3 | 9 |
| 8 | PC4/A4 |  | A4 | 8 |
| 9 | PC5/A5 |  | A5 | 7 |
| 10 | PC6/A6 |  | A6 | 6 |
| 11 | PC7/A7 |  | A7 | 5 |
| 12 | PC8/A8 |  | A8 | 27 |
| 25 | PB3/RD1/POE1/CASL |  | $\overline{\mathrm{OE}}$ | 24 |
| 14 | PC10/A10 |  | A10 | 23 |
| 15 | PC11/A11 |  | A11 | 25 |
| 16 | PC12/A12 |  | A12 | 4 |
| 17 | PC13/A13 |  | A13 | 28 |
| 18 | PC14/A14 |  | A14 | 29 |
| 19 | PC15/A15 |  | A15 | 3 |
| 20 | PB0/A16 |  | A16 | 2 |
| 26 | PB4/I/RQ2/POE2/CASH | С172nF | PGM | 31 |
| 2 | PE15/TIOC4D/DACK1/IRQOUT | - $\square^{2 n F}$ | $\overline{\mathrm{CE}}$ | 22 |
| 1 | PE14/TIOC4C/DACK0/AH |  | $\mathrm{V}_{\mathrm{CC}}$ | 32 |
| 28 | PB5/IRQ3/POE3/RDWR |  | $\mathrm{V}_{\mathrm{SS}}$ | 16 |
| 21, 37,65,77,103 | $\mathrm{V}_{\text {CC }}$ | - |  | PROM program |
| 79 | MD0 |  |  | power supply |
| 78 | MD1 | - |  | (12.5 V) |
| 75 | MD2 | , | A16-A0: | Address input |
| 80, 100 | $\mathrm{PLLV}_{\text {cC }} / \mathrm{AV}_{\text {cc }}$ | $\bigcirc$ | I/O7-I/O0: | Data input/ |
| 81, 82, 74 | PLLCAP, PLLV ${ }_{\text {SS }}$, EXTAL | 1 - |  | output |
| $\begin{aligned} & 3,23,27,33,39,55, \\ & 61,71,90,101,109 \end{aligned}$ | $V_{\text {SS }}$ | $\geqslant 100$ | $\overline{\text { OE: }}$ PGM: | Output enable Program enable |
| $\begin{aligned} & 91-96, \\ & 98,99 \\ & \hline \end{aligned}$ | PF0/AN0-PF7/AN7 |  | $\overline{\mathrm{CE}}$ : | Chip enable |
| 97 | $\mathrm{AV}_{\text {SS }}$ | $!$ ! |  |  |
| 73 | MD3 |  |  |  |

Figure 21.2 SH7042 Pin and HN27C101 Pin Correspondence (112-Pin Version)


Figure 21.3 SH7042 Pin and HN27C101 Pin Correspondence (120-Pin Version)

| SH7043(144-pin version) |  | EPROM socket HN27C101 |  |
| :---: | :---: | :---: | :---: |
| Pin number | Pin name | adapter - Pin name | Pin number |
| 108 | $\overline{\mathrm{RES}} / \mathrm{V}_{\mathrm{PP}}$ | $1 \square^{0.1 \mu F 1} \mathrm{~V}^{1}$ | 1 |
| 98 | NMI | 7r: A9 | 26 |
| 92 | PD0/D0 | 1/O0 | 13 |
| 91 | PD1/D1 | 1/O1 | 14 |
| 90 | PD2/D2 | 1/O2 | 15 |
| 89 | PD3/D3 | 1/O3 | 17 |
| 88 | PD4/D4 | 1/04 | 18 |
| 86 | PD5/D5 | 1/O5 | 19 |
| 84 | PD6/D6 | 1/06 | 20 |
| 83 | PD7/D7 | 1/O7 | 21 |
| 7 | PC0/A0 | A0 | 12 |
| 8 | PC1/A1 | A1 | 11 |
| 9 | PC2/A2 | A2 | 10 |
| 10 | PC3/A3 | A3 | 9 |
| 11 | PC4/A4 | A4 | 8 |
| 13 | PC5/A5 | A5 | 7 |
| 15 | PC6/A6 | A6 | 6 |
| 16 | PC7/A7 | A7 | 5 |
| 17 | PC8/A8 | A8 | 27 |
| 32 | PB3//RQ1/POE1/CASL | $\overline{\mathrm{OE}}$ | 24 |
| 19 | PC10/A10 | A10 | 23 |
| 20 | PC11/A11 | A11 | 25 |
| 21 | PC12/A12 | A12 | 4 |
| 22 | PC13/A13 | A13 | 28 |
| 23 | PC14/A14 | A14 | 29 |
| 24 | PC15/A15 | A15 | 3 |
| 25 | PB0/A16 | A16 | 2 |
| 34 | PB4//RQ2/POE2/EASH | $\bigcirc \overline{\text { PGM }}$ | 31 |
| 5 | PE15/TIOC4D/DACK1//RQOUT | - ${ }^{\text {CE }}$ | 22 |
| 2 | PE14/TIOC4D/DACK0/AH | Vr $\mathrm{V}_{\text {CC }}$ | 32 |
| 36 | PB5/IRQ3/POE3/RDWR | $\mathrm{V}_{\text {SS }}$ | 16 |
| $\begin{gathered} 12,26,40,63,77, \\ 85,99,112,135 \\ \hline \end{gathered}$ | $\mathrm{V}_{\mathrm{CC}}$ | $\quad V_{P P}:$ | PROM program power supply |
| 103 | MD0 | , | power supply $(12.5 \mathrm{~V})$ |
| 102 | MD1 | A16-A0: | Address input |
| 97 | MD2 | A/O7-I/OO: | Data input/ |
| 104, 128, 127 | $\mathrm{PLLV}_{\text {CC }}, \mathrm{AV}_{\mathrm{CC}}, \mathrm{AV}_{\text {ref }}$ | , | output |
| 105, 106, 96 | PLLCAP, PLLV ${ }_{\text {SS }}$, EXTAL | OE: | Output enable |
| $\begin{gathered} \hline 6,14,28,35,42,55,61, \\ 71,79,87,93,117,129 \\ 141 \\ \hline \end{gathered}$ | $\mathrm{V}_{\text {Ss }}$ | 有 $100 \quad \overline{\mathrm{PGM}}:$ | Program enable Chip enable |
| 118-123, 125, 126 | PF0/AN0-PF7/AN7 |  |  |
| 124 | $\mathrm{AV}_{\text {SS }}$ |  |  |
| 95 |  | $\text { 吴 } 0.1 \mu \mathrm{~F}$ |  |

Figure 21.4 SH7043 Pin and HN27C101 Pin Correspondence (144-Pin Version)

| Address for MCU modes 0, 1, 2, 3 |  | Address for PROM mode |
| :---: | :---: | :---: |
| H'00000000 |  | H'0000 |
|  | On-chip ROM space (area 0) |  |
| H'0001FFFF (128-kbyte version) |  | H'1FFFF (128-kbyte version) |

Figure 21.5 On-Chip ROM Memory Map

### 21.3 PROM Programming

The PROM mode write/verify specifications are the same as those of the standard EPROM HN27C101. However, because the page program format is not supported, do not set the PROM writer to the page programming mode. PROM writers that only support page programming mode cannot be used. When selecting a PROM writer, confirm that it supports the byte-by-byte highspeed, high-reliability programming format.

### 21.3.1 Programming Mode Selection

There are two on-chip PROM programming modes: write and verify (reads and confirms written data). The mode is selected by using the pins (table 21.2).

Table 21.2 PROM Programming Mode Selection

| Mode | Pin |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\overline{\text { CE }}$ | $\overline{\mathrm{OE}}$ | $\overline{\text { PGM }}$ | $\mathrm{V}_{\text {PP }}$ | $\mathrm{V}_{\text {cc }}$ | I/07-I/00 | A16-A0 |
| Write | 0 | 1 | 0 | $V_{\text {PP }}$ | $\mathrm{V}_{\text {c }}$ | Data input | Address input |
| Verify | 0 | 0 | 1 |  |  | Data output |  |
| Programming Prohibited | 0 | 0 | 0 |  |  | High impedance |  |
|  | 0 | 1 | 1 |  |  |  |  |
|  | 1 | 0 | 0 |  |  |  |  |
|  | 1 | 1 | 1 |  |  |  |  |

Note: 0 : low level, 1: high level, $\mathrm{V}_{\mathrm{Pp}}: \mathrm{V}_{\mathrm{PP}}$ level, $\mathrm{V}_{\mathrm{cc}}: \mathrm{V}_{\mathrm{cc}}$ level.

### 21.3.2 Write/Verify and Electrical Characteristics

Write/Verify: Writing and verification can be done using an efficient high speed, high reliability programming format. This format allows data writing that is both fast and reliable without applying voltage stress to the device. Figure 21.6 shows the basic flow of the high speed, high reliability programming format.


Figure 21.6 High-Speed, High-Reliability Programming Basic Flow

Electrical Characteristics: Tables 21.3 and 21.4 show the electrical characteristics for programming. Figure 21.7 shows the timing.

Table 21.3 DC Characteristics $\left(V_{\mathrm{CC}}=6.0 \mathrm{~V} \pm 0.25 \mathrm{~V}, \mathrm{~V}_{\mathrm{PP}}=12.5 \mathrm{~V} \pm 0.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{Ta}=\right.$ $25^{\circ} \mathrm{C} \pm \mathbf{5}^{\circ} \mathrm{C}$ )

| Item | Pin | Symbo | Min | Typ | Max | Unit | Measurement Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input high-level voltage | $\begin{aligned} & \mathrm{I} / \mathrm{O} 7-\mathrm{I} / \mathrm{O} 0, \mathrm{~A} 16-\mathrm{A} 0, \\ & \overline{\mathrm{OE}}, \overline{\mathrm{CE}}, \overline{\mathrm{PGM}} \end{aligned}$ | $\mathrm{V}_{\mathrm{IH}}$ | 2.4 | - | $\mathrm{V}_{\text {cc }}+0.3$ | V |  |
| Input low-level voltage | $\begin{aligned} & \mathrm{I} / \mathrm{O} 7-\mathrm{I} / \mathrm{O} 0, \mathrm{~A} 16-\mathrm{A} 0, \\ & \mathrm{OE}, \overline{\mathrm{CE}}, \overline{\mathrm{PGM}} \end{aligned}$ | $\mathrm{V}_{\text {IL }}$ | -0.3 | - | 0.8 | V |  |
| Output high-level voltage | I/O7-I/O0 | $\mathrm{V}_{\mathrm{OH}}$ | 2.4 | - | - | V | $\mathrm{I}_{\mathrm{OH}}=-200 \mu \mathrm{~A}$ |
| Output low-level voltage | I/O7-I/O0 | $\mathrm{V}_{\mathrm{OL}}$ | - | - | 0.45 | V | $\mathrm{I}_{\mathrm{OL}}=1.6 \mathrm{~mA}$ |
| Input leak current | $\begin{aligned} & \mathrm{I} / \mathrm{O} 7-\mathrm{I} / \mathrm{O} 0, \mathrm{~A} 16-\mathrm{A} 0, \\ & \mathrm{OE}, \overline{\mathrm{CE}}, \overline{\mathrm{PGM}} \end{aligned}$ | $\left\|I_{\text {LI }}\right\|$ | - | - | 2 | $\mu \mathrm{A}$ | $\mathrm{VIN}=5.25 \mathrm{~V} / 0.5 \mathrm{~V}$ |
| $\mathrm{V}_{\text {cC }}$ current |  | $\mathrm{I}_{\mathrm{CC}}$ | - | - | 80 | mA |  |
| $\mathrm{V}_{\mathrm{PP}}$ current |  | $\mathrm{I}_{\text {PP }}$ | - | - | 80 | mA |  |

Table 21.4 AC Characteristics $\left(V_{C C}=6.0 \mathrm{~V} \pm 0.25 \mathrm{~V}, \mathrm{~V}_{\mathrm{PP}}=\mathbf{1 2 . 5} \mathrm{V} \pm \mathbf{0 . 3} \mathrm{V}, \mathrm{V}_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{Ta}=\right.$ $\mathbf{2 5}^{\circ} \mathrm{C} \pm \mathbf{5}^{\circ} \mathrm{C}$ )

| Item | Symbol | Min | Typ | Max | Unit | Measurement <br> Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Address setup time | $\mathrm{t}_{\mathrm{AS}}$ | 2 | - | - | $\mu \mathrm{s}$ | Figure $21.6^{* 1}$ |
| $\overline{\text { OE setup time }}$ | $\mathrm{t}_{\mathrm{OES}}$ | 2 | - | - | $\mu \mathrm{s}$ |  |
| Data setup time | $\mathrm{t}_{\mathrm{DS}}$ | 2 | - | - | $\mu \mathrm{s}$ |  |
| Address hold time | $\mathrm{t}_{\mathrm{AH}}$ | 0 | - | - | $\mu \mathrm{s}$ |  |
| Data hold time | $\mathrm{t}_{\mathrm{DH}}$ | 2 | - | - | $\mu \mathrm{s}$ |  |
| Data output disable time | $\mathrm{t}_{\mathrm{DF}}{ }^{* 2}$ | - | - | 130 | ns |  |
| Vpp setup time | $\mathrm{t}_{\mathrm{VPS}}$ | 2 | - | - | $\mu \mathrm{s}$ |  |
| $\overline{\text { PGM }}$ pulse width during initial programming | $\mathrm{t}_{\mathrm{PW}}$ | 0.19 | 0.20 | 0.21 | ms |  |
| $\overline{\text { PGM }}$ pulse width during over-programming | $\mathrm{t}_{\mathrm{OPW}}{ }^{* 3}$ | 0.19 | - | 5.25 | ms |  |
| Vcc setup time | $\mathrm{t}_{\mathrm{VCS}}$ | 2 | - | - | $\mu \mathrm{s}$ |  |
| $\overline{\mathrm{CE}}$ setup time | $\mathrm{t}_{\mathrm{CES}}$ | 2 | - | - | $\mu \mathrm{s}$ |  |
| Data output delay time | $\mathrm{t}_{\mathrm{OE}}$ | 0 | - | 150 | ns |  |

Notes: *1 Input pulse level: 0.45 V to 2.4 V ; input rise, fall times $\leq 20 \mathrm{~ns}$; input timing reference levels: $0.8 \mathrm{~V}, 2.0 \mathrm{~V}$; output timing reference levels: $0.8 \mathrm{~V}, 2.0 \mathrm{~V}$.

* $2 t_{D F}$ is defined as when the output becomes open state and referencing the output level is no longer possible.
*3 $t_{\text {opw }}$ is defined by the values noted in the flowchart (figure 21.6).


Note: * $\mathrm{t}_{\mathrm{OPW}}$ is defined by the values noted in the flowchart (figure 19.6).
Figure 21.7 Write/Verify Timing

### 21.3.3 Cautions on Writing

1. Writes must always be done with the established voltage and timing. The write voltage (programming voltage) $\mathrm{V}_{\mathrm{PP}}$ is 12.5 V (when the EPROM writer is set for the HN27C101 Hitachi specifications, $\mathrm{V}_{\mathrm{PP}}$ becomes 12.5 V ). Devices will sometimes be destroyed if a voltage higher than the rated one is applied. Pay particular attention to such phenomena as EPROM writer overshoot.
2. Always confirm that the indices of the EPROM writer socket, socket adapter, and device are in agreement before programming. Devices will sometimes be destroyed due to excessive current flow if these are not connected in the proper locations.
3. Do not touch the socket adapter or device during writing. Contact faults can sometimes cause devices to be improperly written.
4. Page programming mode writes are not possible. Always set to byte programming mode.
5. Terminate the writing if a write malfunction occurs in consecutive addresses. In such cases, check for problems in the EPROM writer and/or socket adapter. There are some cases where write/verify will not be possible if using an EPROM writer with a high impedance power supply system.
6. Use a EPROM writer that conforms to the socket adapter supported by this LSI.

### 21.3.4 Post-Write Reliability

High temperature biasing (or burn-in) of devices is recommended after writing in order to improve the data retention characteristics. High temperature biasing is a method of screening that eliminates parts with faulty initial data retention by on-chip PROM memory cells within a short period of time. Figure 21.8 shows the flow from the on-chip PROM programming including screening to the installation of the device on a board.


Figure 21.8 Screening Flow
If there are any abnormalities in program write/verify or program read-out verification after high temperature biasing, please contact a Renesas Technology technical representative.

## Section 22 256kB Flash Memory (F-ZTAT)

### 22.1 Features

This LSI has 256 kbytes of on-chip flash memory. The features of the flash memory are summarized below.

- Four flash memory operating modes
- Program mode
- Erase mode
- Program-verify mode
- Erase-verify mode
- Programming/erase methods

The flash memory is programmed 32 bytes at a time. Block erase (in single-block units) can be performed. Block erasing can be performed as required on 1 kbyte, 28 kbyte, and 32 kbyte blocks.

- Programming/erase times

The flash memory programming time is 10 ms (typ.) for simultaneous 32-byte programming, equivalent to $300 \mu \mathrm{~s}$ (typ.) per byte, and the erase time is 100 ms (typ.) per block.

- Reprogramming capability

The flash memory can be reprogrammed up to 100 times.

- On-board programming modes

There are two modes in which flash memory can be programmed/erased/verified on-board:

- Boot mode
- User program mode
- Automatic bit rate adjustment

With data transfer in boot mode, this LSI's bit rate can be automatically adjusted to match the transfer bit rate of the host.

- Flash memory emulation in RAM

Flash memory programming can be emulated in real time by overlapping a part of RAM onto flash memory.

- Protect modes

There are two protect modes, hardware and software, which allow protected status to be designated for flash memory program/erase/verify operations

- Programmer mode

Flash memory can be programmed/erased in programmer mode, using a PROM programmer, as well as in on-board programming mode.

### 22.2 Overview

### 22.2.1 Block Diagram



Figure 22.1 Flash Memory Block Diagram

### 22.2.2 Mode Transition Diagram

When the mode pins and the FWP pin are set in the reset state and a reset start is executed, the microcomputer enters one of the operating modes shown in figure 22.2. In user mode, flash memory can be read but not programmed or erased.

Flash memory can be programmed and erased in boot mode, user program mode, and programmer mode.


Figure 22.2 Flash Memory Mode Transitions

### 22.2.3 Onboard Program Mode

## Boot mode

1. Initial state

The old program version or data remains written in the flash memory. The user should prepare the programming control program and new application program beforehand in the host.

3. Initializing the flash memory

To initialize (to H'FF) the flash memory, execute the erase program located in the boot program area (within RAM). During the boot mode, the entire flash memory is erased, regardless of blocks.

2. Programming control program transfer When boot mode is entered, the boot program in the LSI (originally incorporated in the chip) is started and the programming control program in the host is transferred to RAM via SCI communication. The boot program required for flash memory erasing is automatically transferred to the RAM boot program area.

4. Writing the new application program Execute the program transferred to RAM from the host and write the new application program located at the transfer destination to the flash memory.


7 Program execution state

Figure 22.3 Boot Mode

## User program mode

1. Initial state

The FWP assessment program that confirms that user program mode has been entered, and the program that will transfer the programming/erase control program from flash memory to on-chip RAM should be written into the flash memory by the user beforehand. The programming/erase control program should be prepared in the host or in the flash memory.

3. Initializing the flash memory

Execute the Programming/erase program in RAM to initialize (to H'FF) the flash memory. Erase is executed in block units, but cannot be executed in byte units.

2. Programming/erase control program transfer When user program mode is entered, user software confirms this fact, executes transfer program in the flash memory, and transfers the programming/erase control program to RAM.

4. Writing new application program Next, the new application program in the host is written into the erased flash memory blocks. Do not write to unerased blocks.

$\Delta$ Program execution state

Figure 22.4 User Program Mode

## renesas

### 22.2.4 Flash Memory Emulation in RAM

Emulation should be performed in user mode or user program mode. When the emulation block set in RAMER is accessed while the emulation function is being executed, data written in the overlap RAM is read.

- User Mode
- User Program Mode


Figure 22.5 Emulation

When overlap RAM data is confirmed, the RAMS bit is cleared, RAM overlap is released, and writes should actually be performed to the flash memory.
When the programming control program is transferred to RAM, ensure that the transfer destination and the overlap RAM do not overlap, as this will cause data in the overlap RAM to be rewritten.

- User Program Mode


Figure 22.6 Programming to the Flash Memory

### 22.2.5 Differences between Boot Mode and User Program Mode

Table 22.1 Differences between Boot Mode and User Program Mode

|  | Boot Mode | User Program Mode |
| :--- | :--- | :--- |
| Total erase | Yes | Yes |
| Block erase | No | Yes |
| Programming control program* | $(2)$ | $(1)(2)(3)$ |

(1) Erase/erase-verify
(2) Program/program-verify
(3) Emulation

Note: * To be prepared by the user according to the recommended algorithm.

### 22.2.6 Block Configuration

The flash memory is divided into seven 32 kbyte blocks, one 28 kbyte blocks, and four 1 kbyte blocks.


Figure 22.7 Block Configuration

### 22.3 Pin Configuration

The flash memory is controlled by the pins shown in table 22.2.
Table 22.2 Pin Configuration

| Pin Name | Abbreviation | I/O | Function |
| :--- | :--- | :--- | :--- |
| Power-on reset | $\overline{\text { RES }}$ | Input | Power-on reset |
| Flash write protect | FWP | Input | Flash program/erase protection by hardware |
| Mode 3 | MD3 | Input | Set operation mode of LSI |
| Mode 2 | MD2 | Input | Set operation mode of LSI |
| Mode 1 | MD1 | Input | Set operation mode of LSI |
| Mode 0 | MD0 | Input | Set operation mode of LSI |
| Transmit data | TxD1 | Output | Serial send data output |
| Receive data | RxD1 | Input | Serial receive data input |

### 22.4 Register Configuration

Registers that control the flash memory when the on-chip flash memory is valid are shown in table 22.3.

Table 22.3 Register Configuration

| Name | Abbreviation | R/W | Initial Value | Address | Access Size |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Flash memory control register 1 | FLMCR1 | R/W*1 | $\mathrm{H}^{\prime} 00{ }^{* 2}$ | H'FFFF8580 | 8 |
| Flash memory control register 2 | FLMCR2 | R/W*1 | $\mathrm{H}^{\prime} 00^{* 3}$ | H'FFFF8581 | 8 |
| Erase block register 1 | EBR1 | R/W*1 | $\mathrm{H}^{\prime} 00^{* 3}$ | H'FFFF8582 | 8 |
| Erase block register 2 | EBR2 | R/W*1 | $\mathrm{H}^{\prime} 00{ }^{* 3}$ | H'FFFF8583 | 8 |
| RAM emulation register | RAMER | R/W | H'0000 | H'FFFF8628 | 8, 16, 32 |

Notes: 1. FLMCR1, FLMCR2, EBR1, and EBR2 are 8-bit registers, and RAMER is a 16-bit register.
2. Only byte accesses are valid for FLMCR1, FLMCR2, EBR1, and EBR2, the access requiring 3 cycles. Three cycles are required for a byte or word access to RAMER, and 6 cycles for a longword access.
3. When a longword write is performed on RAMER, 0 must always be written to the lower word (address H'FFFF8630). Operation is not guaranteed if any other value is written.
*1 In modes in which the on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid. Writes are also disabled when the FWE bit is set to 1 in FLMCR1.
*2 When a low level is input to the FWP pin, the initial value is H'80.
*3 When a high level is input to the FWP pin, or if a low level is input and the SWE bit in FLMCR1 is not set, these registers are initialized to $\mathrm{H}^{\prime} 00$.

### 22.5 Description of Registers

### 22.5.1 Flash Memory Control Register 1 (FLMCR1)

FLMCR1 is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode for addresses H'00000-H'1FFFF is entered by setting SWE to 1 when FWE $=1$, then setting the EV1 or PV1 bit. Program mode for addresses H'00000-H'1FFFF is entered by setting SWE to 1 when $F W E=1$, then setting the PSU1 bit, and finally setting the P1 bit. Erase mode for addresses H'00000-H'1FFFF is entered by setting SWE to 1 when FWE $=1$, then setting the ESU1 bit, and finally setting the E1 bit. FLMCR1 is initialized in the standby mode or with power-on reset. Its initial value is $\mathrm{H}^{\prime} 80$ when a low level is input to the FWP pin, and $\mathrm{H}^{\prime} 00$ when a high level is input. When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 80$, and writes are invalid.

Writes to bits SWE, ESU1, PSU1, EV1, and PV1 are enabled only when FWE $=1$ and $\mathrm{SWE}=1$; writes to the E1 bit only when $\mathrm{FWE}=1, \mathrm{SWE}=1$, and $\mathrm{ESU} 1=1$; and writes to the P1 bit only when $\mathrm{FWE}=1, \mathrm{SWE}=1$, and PSU1 $=1$.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FWE | SWE | ESU1 | PSU1 | EV1 | PV1 | E1 | P1 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 7—Flash Write Enable Bit (FWE): Displays the state of the FWP pin which sets hardware protection against flash memory programming/erasing.

Bit 7: FWE Description

| 0 | When high level is input to the FWP pin (hardware-protect state) |
| :--- | :--- |
| 1 | When low level is input to the FWP pin |

- Bit 6-Software Write Enable Bit (SWE): Enables or disables the flash memory. This bit should be set when setting bits 5-0, FLMCR2 bits 5-0, EBR1 bits 3-0, and EBR2 bits 7-0.

Bit 6: SWE Description

| 0 | Writes disabled | (Initial value) |
| :--- | :--- | :--- |
| 1 | Writes enabled |  |
|  | [Setting condition] When FWE $=1$ |  |

- Bit 5—Erase Setup Bit 1 (ESU1): Prepares for a transition to erase mode (applicable addresses: H'00000-H'1FFFF). Do not set the SWE, PSU1, EV1, PV1, E1, or P1 bit at the same time.


## Renesns

| Bit 5: ESU1 | Description |
| :--- | :--- |
| 0 | Erase setup release (Initial value) |
| 1 | Erase setup |
|  | [Setting condition] When FWE=1 and SWE=1 |

- Bit 4—Program Setup Bit 1 (PSU1): Prepares for a transition to program mode (applicable addresses: H'00000-H'1FFFFF). Do not set the SWE, ESU1, EV1, PV1, E1, or P1 bit at the same time.

Bit 4: PSU1 Description

| 0 | Program setup release (Initial value) |
| :--- | :--- |
| 1 | Program setup |
|  | [Setting condition] When FWE $=1$ and $\mathrm{SWE}=1$ |

- Bit 3—Erase-Verify 1 (EV1): Selects erase-verify mode transition or release (applicable addresses: H'00000-H'1FFFF). Do not set the SWE, ESU1, PSU1, PV1, E1, or P1 bit at the same time.

Bit 3: EV1 Description

| 0 | Erase verify mode release (Initial value) |
| :--- | :--- |
| 1 | Transition to erase verify mode |
|  | [Setting condition] When $\mathrm{FWE}=1$ and $\mathrm{SWE}=1$ |

- Bit 2—Program-Verify 1 (PV1): Selects program-verify mode transition or release (applicable addresses: H'00000-H'1FFFF). Do not set the SWE, ESU1, PSU1, EV1, E1, or P1 bit at the same time.


## Bit 2: PV1 Description

| 0 | Program verify mode release (Initial value) |
| :--- | :--- |
| 1 | Transition to program verify mode |
| [Setting condition] When FWE=1 and SWE=1 |  |

- Bit 1—Erase 1 (E1): Selects erase mode transition or release (applicable addresses: H'00000$\left.H^{\prime} 1 \mathrm{FFFF}\right)$. Do not set the SWE, ESU1, PSU1, EV1, PV1, or P1 bit at the same time.

| Bit 1: E1 | Description |
| :--- | :--- |
| 0 | Erase mode release (Initial value) |
| 1 | Transition to erase mode |
|  | [Setting condition] When FWE=1, SWE=1, and ESU1=1 |

- Bit 0—Program 1 (P1): Selects program mode transition or release (applicable addresses: $\left.H^{\prime} 00000-H^{\prime} 1 F F F F\right)$. Do not set the SWE, PSU1, ESU1, EV1, PV1, or E1 bit at the same time.

Bit 0: P1 Description

| 0 | Program setup mode release (Initial value) |
| :--- | :--- |
| 1 | Program setup |
|  | [Setting condition] When FWE=1, SWE $=1$, and PSU1 $=1$ |

### 22.5.2 Flash Memory Control Register 2 (FLMCR2)

FLMCR2 is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode for addresses $\mathrm{H}^{\prime} 20000-\mathrm{H}^{\prime} 3 F F F F$ is entered by setting SWE (FLMCR1) to 1 when FWE $($ FLMCR1 $)=1$, then setting the EV2 or PV2 bit. Program mode for addresses $\mathrm{H}^{\prime} 20000-\mathrm{H}^{\prime} 3 \mathrm{FFFF}$ is entered by setting SWE (FLMCR1) to 1 when FWE (FLMCR1) $=1$, then setting the PSU2 bit, and finally setting the P2 bit. Erase mode for addresses $\mathrm{H}^{\prime} 20000-\mathrm{H}^{\prime} 3 \mathrm{FFFF}$ is entered by setting SWE (FLMCR1) to 1 when FWE (FLMCR1) = 1, then setting the ESU2 bit, and finally setting the E2 bit. FLMCR2 is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset, in standby mode, when a high level is input to the FWP pin, and when a low level is input to the FWP pin and the SWE bit in FLMCR1 is not set (the exception is the FLER bit, which is initialized only by a power-on reset). When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

Writes to bits ESU2, PSU2, EV2, and PV2 in FLMCR2 are enabled only when FWE $($ FLMCR1 $)=$ 1 and SWE $($ FLMCR1 $)=1$; writes to the E2 bit only when FWE $($ FLMCR1 $)=1$, SWE
$($ FLMCR1 $)=1$, and ESU2 $=1$; and writes to the P2 bit only when FWE $($ FLMCR1 $)=1$, SWE $($ FLMCR1 $)=1$, and PSU2 $=1$.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FLER | - | ESU2 | PSU2 | EV2 | PV2 | E2 | P2 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R | R | R/W | R/W | R/W | R/W | R/W | R/W |

- Bit 7—Flash Memory Error (FLER): Indicates that an error has occurred during an operation on flash memory (programming or erasing). When FLER is set to 1 , flash memory goes to the error-protection state.

Bit 7: FLER Description

| 0 | Flash memory is operating normally. (Initial value) |
| :--- | :--- |
|  | Flash memory program/erase protect (error protect) disabled |
| 1 | Indicates error during flash memory program/erase. |
|  | Flash memory program/erase protect (error protect) enabled |
|  | [Setting condition] See section 22.8.3, Error protection |

- Bit 6—Reserved bit: This bit is always read as 0 .
- Bit 5-Erase Setup Bit 2 (ESU2): Prepares for a transition to erase mode (applicable addresses: $\left.H^{\prime} 20000-H^{\prime} 3 F F F F\right)$. Do not set the PSU2, EV2, PV2, E2, or P2 bit at the same time.

Bit 5: ESU2 Description

| 0 | Erase setup release (Initial value) |
| :--- | :--- |
| 1 | Erase setup |
|  | [Setting condition] When FWE $=1$ and SWE $=1$ |

- Bit 4—Program Setup Bit 2 (PSU2): Prepares for a transition to program mode (applicable addresses: H'20000-H'3FFFF). Do not set the ESU2, EV2, PV2, E2, or P2 bit at the same time.


## Bit 4: PSU2 Description

| 0 | Program setup release (Initial value) |
| :--- | :--- |
| 1 | Program setup |
|  | [Setting condition] When $\mathrm{FWE}=1$ and $\mathrm{SWE}=1$ |

- Bit 3—Erase-Verify 2 (EV2): Selects erase-verify mode transition or release (applicable addresses: H'20000-H'3FFFF). Do not set the ESU2, PSU2, PV2, E2, or P2 bit at the same time.

Bit 3: EV2 Description

| 0 | Erase verify mode release (Initial value) |
| :--- | :--- |
| 1 | Transition to the erase verify mode |
| [Setting condition] When FWE $=1$ and SWE $=1$ |  |

- Bit 2—Program-Verify 2 (PV2): Selects program-verify mode transition or release (applicable addresses: H'20000-H'3FFFF). Do not set the ESU2, PSU2, EV2, E2, or P2 bit at the same time.

Bit 2: PV2 Description

| 0 | Program verify mode release (Initial value) |
| :--- | :--- |
| 1 | Transition to the program verify mode |
|  | [Setting condition] When FWE $=1$, and $\mathrm{SWE}=1$ |

- Bit 1—Erase 2 (E2): Selects erase mode transition or release (applicable addresses: H'20000H'3FFFF). Do not set the ESU2, PSU2, EV2, PV2, or P2 bit at the same time.


## Bit 1: E2 Description

| 0 | Erase mode release (Initial value) |
| :--- | :--- |
| 1 | Transition to the erase mode |
|  | [Setting condition] When FWE $=1, \mathrm{SWE}=1$, and $\mathrm{ESU} 2=1$ |

- Bit 0—Program 2 (P2): Selects program mode transition or release (applicable addresses: H'20000-H'3FFFF). Do not set the ESU2, PSU2, EV2, PV2, or E2 bit at the same time.

Bit 0: P2 Description

| 0 | Program mode release(Initial value) |
| :--- | :--- |
| 1 | Transition to the program mode |
| [Setting condition] When FWE $=1$, SWE $=1$, and PSU2 $=1$ |  |

### 22.5.3 Erase Block Register 1 (EBR1)

EBR1 is an 8-bit register that specifies the flash memory erase area block by block. EBR1 is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset and standby mode, when a high level is input to the FWP pin, and when a low level is input to the FWP pin and the SWE bit in FLMCR1 is not set. When a bit in EBR1 is set to 1, the corresponding block can be erased. Other blocks are erase-protected. Only one of the bits of EBR1 and EBR2 combined can be set. Do not set more than one bit. If more than one bit is set, writes to bits ESU1, ESU2, E1, and E2 will be invalid. When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

The flash memory block configuration is shown in table 22.4.


### 22.5.4 Erase Block Register 2 (EBR2)

EBR2 is an 8-bit register that specifies the flash memory erase area block by block. EBR2 is initialized to $\mathrm{H}^{\prime} 00$ by a power-on reset and standby mode, when a high level is input to the FWP pin, and when a low level is input to the FWP pin and the SWE bit in FLMCR1 is not set. When a bit in EBR2 is set to 1, the corresponding block can be erased. Other blocks are erase-protected. When on-chip flash memory is disabled, a read will return $\mathrm{H}^{\prime} 00$, and writes are invalid.

The flash memory block configuration is shown in table 22.4.

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| ---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | EB11 | EB10 | EB9 | EB8 | EB7 | EB6 | EB5 | EB4 |
| Initial value: | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| R/W: | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |

Table 22.4 Flash Memory Erase Blocks

| Block (size) | Addresses |
| :---: | :---: |
| EB0 (32kB) | H'000000-H'007FFF |
| EB1 (32kB) | H'008000-H'00FFFF |
| EB2 (32kB) | H'010000-H'017FFF |
| EB3 (32kB) | H'018000-H'01FFFF |
| EB4 (32kB) | H'020000-H'027FFF |
| EB5 (32kB) | H'028000-H'02FFFF |
| EB6 (32kB) | H'030000-H'037FFF |
| EB7 (28kB) | H'038000-H'03EFFF |
| EB8 (1kB) | H'03F000-H'03F3FF |
| EB9 (1kB) | H'03F400-H'03F7FF |
| EB10 (1kB) | H'03F800-H'03FBFF |
| EB11 (1kB) | H'03FC00-H'03FFFF |

### 22.5.5 RAM Emulation Register (RAMER)

RAMER specifies the area of flash memory to be overlapped with part of RAM when emulating real-time flash memory programming. RAMER is initialized to H'0000 by a power-on reset. It is not initialized in software standby mode. RAMER settings should be made in user mode or user program mode. (For details, see the description of the BSC.)

Flash memory area divisions are shown in table 22.5. To ensure correct operation of the emulation function, the ROM for which RAM emulation is performed should not be accessed immediately after this register has been modified. Normal execution of an access immediately after register modification is not guaranteed.


- Bits 15-3-Reserved bits: These bits are always read as 0 .
- Bit 2—RAM Select (RAMS): Specifies selection or non-selection of flash memory emulation in RAM. When RAMS $=1$, all flash memory block are program/erase-protected. This bit is ignored when the on-chip ROM is disabled.

Bit 2: RAMS Description

| 0 | Emulation not selected |
| :--- | :--- |
| 1 | Program/erase protect of all flash memory blocks is disabled (Initial value) |

- Bits 1 and 0—Flash Memory Area Selection (RAM1, RAM0): These bits are used together with bit 2 to select the flash memory area to be overlapped with RAM. (See table 22.5.)

Table 22.5 Separation of the Flash Memory Area

| Addresses | Block Name | RAMS | RAM1 | RAM0 |
| :--- | :--- | :--- | :--- | :--- |
| H'FFF800-H'FFFBFF | RAM area 1kB | 0 | $*$ | $*$ |
| H'03F000-H'03F3FF | EB8 $(1 \mathrm{kB})$ | 1 | 0 | 0 |
| H'03F400-H'03F7FF | EB9 $(1 \mathrm{kB})$ | 1 | 0 | 1 |
| H'03F800-H'03FBFF | EB10(1kB) | 1 | 1 | 0 |
| H'03FC00-H'03FFFF | EB11(1kB) | 1 | 1 | 1 |

### 22.6 On-Board Programming Mode

When pins are set to on-board programming mode and a power-on reset is executed, a transition is made to the on-board programming state in which program/erase/verify operations can be performed on the on-chip flash memory. There are two on-board programming modes: boot mode and user program mode. The pin settings for transition to each of these modes are shown in table 22.6. For a diagram of the transitions to the various flash memory modes, see figure 22.2.

Table 22.6 Setting On-Board Programming Modes

| Mode |  | PLL Multiple | FWP | MD3 | MD2 | MD1 | MDO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Boot mode | Expanded Mode | $\times 1$ | 0 | 0 | 0 | 0 | 0 |
|  | Single-chip Mode |  |  | 0 | 0 | 0 | 1 |
|  | Expanded Mode | $\times 2$ |  | 0 | 1 | 0 | 0 |
|  | Single-chip Mode |  |  | 0 | 1 | 0 | 1 |
|  | Expanded Mode | $\times 4$ |  | 1 | 0 | 0 | 0 |
|  | Single-chip Mode |  |  | 1 | 0 | 0 | 1 |
| User program mode | Expanded Mode | $\times 1$ | 0 | 0 | 0 | 1 | 0 |
|  | Single-chip Mode |  |  | 0 | 0 | 1 | 1 |
|  | Expanded Mode | $\times 2$ |  | 0 | 1 | 1 | 0 |
|  | Single-chip Mode |  |  | 0 | 1 | 1 | 1 |
|  | Expanded Mode | $\times 4$ |  | 1 | 0 | 1 | 0 |
|  | Single-chip Mode |  |  | 1 | 0 | 1 | 1 |

### 22.6.1 Boot Mode

When boot mode is used, the flash memory programming control program must be prepared in the host beforehand. The SCI to be used is set to channel asynchronous mode.

When a reset start is executed after the LSI pins have been set to boot mode in the power-on reset state, the boot program built into the LSI is started and the programming control program prepared in the host is serially transmitted to the LSI via SCI channel 1. In the LSI, the programming control program received via SCI channel 1 is written into the programming control program area in on-chip RAM. After the transfer is completed, control branches to the start address of the programming control program area and the programming control program execution state is entered (flash memory programming is performed).

The transferred programming control program must therefore include coding that follows the programming algorithm given later.

The system configuration in boot mode is shown in figure 22.8, and the boot mode execution procedure in figure 22.9.


Figure 22.8 System Configuration in Boot Mode


Figure 22.9 Boot Mode Execution Procedure


Figure 22.10 Automatic SCI Bit Rate Adjustment

When boot mode is initiated, the LSI measures the low period of the asynchronous SCI communication data ( $\mathrm{H}^{\prime} 00$ ) transmitted continuously from the host. The SCI transmit/receive format should be set as follows: 8-bit data, 1 stop bit, no parity. The LSI calculates the bit rate of the transmission from the host from the measured low period, and transmits one $\mathrm{H}^{\prime} 00$ byte to the host to indicate the end of bit rate adjustment. The host should confirm that this adjustment end indication ( $\mathrm{H}^{\prime} 00$ ) has been received normally, and transmit one $\mathrm{H}^{\prime} 55$ byte to the LSI. If reception cannot be performed normally, initiate boot mode again (reset), and repeat the above operations. Depending on the host's transmission bit rate and the LSI's system clock frequency, there will be a discrepancy between the bit rates of the host and the LSI. To ensure correct SCI operation, the host's transfer bit rate should be set to 9,600 or $4,800 \mathrm{bps}$.

Table 22.7 shows host transfer bit rates and system clock frequencies for which automatic adjustment of the LSI bit rate is possible. The boot program should be executed within this system clock range.

Table 22.7 System Clock Frequencies for which Automatic Adjustment of LSI Bit Rate is Possible

| Host Bit Rate | System Clock Frequency for which Automatic Adjustment <br> of LSI Bit Rate is Possible |
| :--- | :--- |
| $9,600 \mathrm{bps}$ | 8 to 28.7 MHz |
| $4,800 \mathrm{bps}$ | 4 to 20 MHz |

On-Chip RAM Area Divisions in Boot Mode: In boot mode, the RAM area is divided into an area used by the boot program and an area to which the programming control program is transferred via the SCI, as shown in figure 22.11. The boot program area cannot be used until the execution state in boot mode switches to the programming control program transferred from the host.


Figure 22.11 RAM Areas in Boot Mode
Note: The boot program area cannot be used until a transition is made to the execution state for the programming control program transferred to RAM. Note also that the boot program remains in this area of the on-chip RAM even after control branches to the programming control program.

### 22.6.2 User Program Mode

After setting FWP, the user should branch to, and execute, the previously prepared programming/erase control program.

As the flash memory itself cannot be read while flash memory programming/erasing is being executed, the control program that performs programming and erasing should be run in on-chip RAM or external memory.

Use the following procedure (figure 22.12) to execute the programming control program that writes to flash memory (when transferred to RAM).


Figure 22.12 User Program Mode Execution Procedure
Notes: 1. When programming and erasing, start the watchdog timer so that measures can be taken to prevent program runaway, etc. Memory cells may not operate normally if overprogrammed or overerased due to program runaway.
2. If an address at which a flash memory register resides is read in the mask ROM or ZTAT version, the value will be undefined. When a flash memory version program is used in the mask ROM or ZTAT version, the state of the FWP pin cannot be determined. A modification must therefore be made to prevent operation of the flash memory rewrite program.

### 22.7 Programming/Erasing Flash Memory

A software method, using the CPU, is employed to program and erase flash memory in the onboard programming modes. There are four flash memory operating modes: program mode, erase mode, program-verify mode, and erase-verify mode. Transitions to these modes are made by setting the PSU1, ESU1, P1, E1, PV1, and EV1 bits in FLMCR1 for addresses H'00000-H'1FFFF, or the PSU2, ESU2, P2, E2, PV2, and EV2 bits in FLMCR2 for addresses H'20000-H'3FFFF.

The flash memory cannot be read while being programmed or erased. Therefore, the program (programming control program) that controls flash memory programming/erasing should be located and executed in on-chip RAM or external memory.

Notes: 1. Operation is not guaranteed if setting/resetting of the SWE, ESU1, PSU1, EV1, PV1, E1, and P1 bits in FLMCR1, or the ESU2, PSU2, EV2, PV2, E2, and P2 bits in FLMCR2, is executed by a program in flash memory.
2. When programming or erasing, set FWP to low level (programming/erasing will not be executed if FWP is set to high level).
3. Programming should be performed in the erased state. Do not perform additional programming on previously programmed addresses.
4. Do not program addresses $\mathrm{H}^{\prime} 00000-\mathrm{H}^{\prime} 1 \mathrm{FFFF}$ and $\mathrm{H}^{\prime} 20000-\mathrm{H}^{\prime} 3 F F F F$ simultaneously. Operation is not guaranteed if this is done.

### 22.7.1 Program Mode ( $\mathbf{n}=1$ for Addresses $\mathrm{H}^{\prime} \mathbf{0 0 0 0}-\mathrm{H}^{\prime} \mathbf{1 F F F F}, \mathbf{n}=2$ for Addresses H'20000-H'3FFFF)

When writing data or programs to flash memory, the program/program-verify flowchart shown in figure 22.13 should be followed. Performing program operations according to this flowchart will enable data or programs to be written to flash memory without subjecting the device to voltage stress or sacrificing program data reliability. Programming should be carried out 32 bytes at a time.

Following the elapse of $10 \mu \mathrm{~s}$ or more after the SWE bit is set to 1 in flash memory control register 1 (FLMCR1), 32-byte program data is stored in the program data area and reprogram data area, and the 32-byte data in the program data area in RAM is written consecutively to the program address (the lower 8 bits of the first address written to must be $\mathrm{H}^{\prime} 00, \mathrm{H}^{\prime} 20, \mathrm{H}^{\prime} 40, \mathrm{H}^{\prime} 60$, $\mathrm{H}^{\prime} 80, \mathrm{H}^{\prime} \mathrm{A} 0, \mathrm{H}^{\prime} \mathrm{C} 0$, or $\left.\mathrm{H}^{\prime} \mathrm{E} 0\right)$. Thirty-two consecutive byte data transfers are performed. The program address and program data are latched in the flash memory. A 32-byte data transfer must be performed even if writing fewer than 32 bytes; in this case, H'FF data must be written to the extra addresses.

Next, the watchdog timer is set to prevent overprogramming in the event of program runaway, etc. Set a minimum value of $300 \mu \mathrm{~s}$ or more as the WDT overflow period. After this, preparation for program mode (program setup) is carried out by setting the PSUn bit in FLMCRn, and after the elapse of $50 \mu$ s or more, the operating mode is switched to program mode by setting the Pn bit in 704
renesns

FLMCRn. The time during which the Pn bit is set is the flash memory programming time. Set $200 \mu$ s as the time for one programming operation.

### 22.7.2 Program-Verify Mode ( $\mathbf{n}=1$ for Addresses $\mathbf{H}^{\prime} \mathbf{0 0 0 0 - H ' 1 F F F F , ~} \mathbf{n}=\mathbf{2}$ for Addresses H'20000-H'3FFFF)

In program-verify mode, the data written in program mode is read to check whether it has been correctly written in the flash memory.

After the elapse of a given programming time, the programming mode is exited (the Pn bit in FLMCRn is released, then the PSUn bit is released at least $10 \mu \mathrm{~s}$ later). The watchdog timer is released after the elapse of $10 \mu$ or more, and the operating mode is switched to program-verify mode by setting the PVn bit in FLMCRn. Before reading in program-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of $4 \mu \mathrm{~s}$ or more. When the flash memory is read in this state (verify data is read in 32 -bit units), the data at the latched address is read. Wait at least $2 \mu \mathrm{~s}$ after the dummy write before performing this read operation. Next, the written data is compared with the verify data, and reprogram data is computed (see figure 22.13) and transferred to the reprogram data area. After 32 bytes of data have been verified, exit program-verify mode, wait for at least $4 \mu \mathrm{~s}$, then release the SWE bit in FLMCR1. If reprogramming is necessary, set program mode again, and repeat the program/program-verify sequence as before. However, ensure that the program/program-verify sequence is not repeated more than 1,000 times on the same bits.


Figure 22.13 Program/Program Verify Flow

- Sample 32-byte programming program

The wait time set values (number of loops) are for the case where $\mathrm{f}=28.7 \mathrm{MHz}$. For other frequencies, the set value is given by the following expression:
Wait time $(\mu \mathrm{s}) \times \mathrm{f}(\mathrm{MHz}) \div 4$

## Registers Used

| R4 (input): P | Program data storage address |  |
| :---: | :---: | :---: |
| R5 (input): $\operatorname{Pr}$ | Programming destination address |  |
| R7 (output): O | OK (normal) or NG (error) |  |
| R0-3, 8-13: W | Work registers |  |
| FLMCR1 | .EQU H 80 |  |
| FLMCR2 | .EQU $\mathrm{H}^{\prime} 81$ |  |
| OK | .EQU H'0 |  |
| NG | .EQU $\mathrm{H}^{\prime} 1$ |  |
| Wait10u | .EQU 72 |  |
| Wait50u | .EQU 359 |  |
| Wait4u | .EQU 29 |  |
| Wait2u | .EQU 14 |  |
| Wait200u | .EQU 1435 |  |
| WDT_TCSR | .EQU H'FFFF8610 |  |
| WDT_573u | .EQU H'A579 |  |
| SWESET | .EQU B'01000000 |  |
| PSU1SET | .EQU B'00010000 |  |
| P1SET | .EQU B'00000001 |  |
| P1CLEAR | .EQU B'11111110 |  |
| PSU1CLEAR | .EQU B'11101111 |  |
| PVSET | .EQU B'00000100 |  |
| PVCLEAR | .EQU B'11111011 |  |
| SWECLEAR | .EQU B'10111111 |  |
| MAXVerify | .EQU 1000 |  |
| ; |  |  |
| FlashProgram | $m$.EQU \$ |  |
| MOV | \# ${ }^{\prime}$ 01, R2 | ; R2 work register (1) |
| MOV.L | .L \#PdataBuff,R0 | ; Save program data to work area |
| MOV | R4, R12 |  |
| MOV | \#8, R13 |  |
| COPY_LOOP | .EQU \$ |  |

```
    MOV.L @R12+,R1
    MOV.L R1,@R0
    ADD.L #4,R0
    ADD.L #-1,R13
    CMP/PL R13
    BT COPY_LOOP
    MOV.L #H' FFFF8500,R0 ; Initialize GBR
    LDC R0,GBR
;
MOV.L #Wait10u,R3
    MOV.L #FLMCR1,R0 ; Initialize R0 to FLCMR1 address
    OR.B #SWESET,@(R0,GBR) ; Set SWE
Wait_1 SUBC R2,R3 ; Wait 10 \mus
BF Wait_1
;
    MOV.L #H'20000,R9
    CMP/GT R5,R9
    BT Program_Start
    MOV.L #FLMCR2,R0
Program_Start .EQU $
    MOV.L #0,R9 ; Initialize n (R9) to 0
;
Program_loop .EQU $
            MOV.L #0,R10 ; Initialize m (R10) to 0
            MOV.L #32,R3 ; Write 32-byte data consecutively
            MOV.L #PdataBuff,R12
            MOV.L R5,R13
Write_Loop .EQU $
            MOV.B @R12+,R1
            MOV.B R1,@R13
            ADD.L #1,R13
            ADD.L #-1,R3
            CMP /PL R3
            BT Write_Loop
;
\begin{tabular}{lll} 
MOV.L & \#WDT_TCSR,R1 & ; Enable WDT \\
MOV.W & \#WDT_573u,R3 & ; \(573.4 \mu\) s cycle
\end{tabular}
```

|  | MOV.W | R3, @R1 |  |
| :---: | :---: | :---: | :---: |
| ; |  |  |  |
|  | MOV.L | \#Wait50u, R3 |  |
|  | OR.B | \#PSU1SET, @ (R0, GBR) | ; Set PSU |
| Wait_2 | SUBC | R2, R3 | ; Wait $50 \mu \mathrm{~s}$ |
|  | BF |  |  |
| ; |  |  |  |
|  | MOV.L | \#Wait200u, R3 |  |
|  | OR.B | \#P1SET, @ (R0, GBR) | ; Set P |
| Wait_3 | SUBC | R2, R3 | ; Wait 200 s |
|  | BF | Wait_3 |  |
| ; |  |  |  |
|  | MOV.L | \#Wait10u, R3 |  |
|  | AND.B | \#P1CLEAR, @ (R0, GBR) | ; Clear P |
| Wait_4 | SUBC | R2, R3 | ; Wait $10 \mu \mathrm{~s}$ |
|  | BF | Wait_4 |  |
| ; |  |  |  |
|  | MOV.L | \#Wait10u, R3 |  |
|  | AND.B | \#PSU1CLEAR, @ (R0, GBR) | ; Clear PSU |
| Wait_5 | SUBC | R2, R3 | ; Wait $10 \mu \mathrm{~s}$ |
|  | BF | Wait_5 |  |
| ; |  |  |  |
|  | MOV.L | \#WDT_TCSR, R1 | ; Disable WDT |
|  | MOV.W | \# ${ }^{\prime}$ A $555, R 3$ |  |
|  | MOV.W | R3, @R1 |  |
| ; |  |  |  |
|  | MOV.L | \#Wait4u, R3 |  |
|  | OR.B | \#PVSET, @ (R0, GBR) | ; Set PV |
| Wait_6 | SUBC | R2, R3 | ; Wait 4 ¢s |
|  | BF | Wait_6 |  |
| ; |  |  |  |
|  | MOV.L | PdataBuff,R3 |  |
|  | MOV.L | R4, R1 |  |
|  | MOV.L | R5, R12 |  |
|  | MOV.L | \#8,R13 |  |
|  | MOV.L | \# ${ }^{\prime}$ ' FFFFFFFF, R11 |  |
| ; |  |  |  |

```
VerifyLoop .EQU $
    MOV.L R11,@R12 ; Write H'FF to verify address
    MOV.L R11,@R3 ; Reprogram data RAM (PdataBuff) initialization
    MOV.L #Wait2u,R7
    Wait_7 SUBC R2,R7 ; Wait 2 \mus
    BF Wait_7
;
    MOV.L @R12+,R7
    MOV.L @R1+,R8
    CMP /EQ R7,R8 ; Verify
    BT Verify_OK
    MOV.L #1,R10 ; Verify NG, m<-1
    XOR R8,R7 ; Program data computation
    NOT R7,R7
    OR R7,R8
    MOV.L R8,@R3 ; Store in reprogram data RAM (PdataBuff)
Verify_OK .EQU $
    ADD.L #4,R3
    ADD.L #-1,R13
    CMP/PL R13
    BT VerifyLoop
;
    MOV.L #Wait4u,R7
    AND.B #PVCLEAR,@ (R0,GBR) ; Clear PV
Wait_8 SUBC R2,R7 ; Wait 4 \mus
    BF Wait_8
;
    CMP /PL R10 ; if m=0 then GOTO Program_OK
    BF Program_OK
    ADD #1,R9
    MOV.L #NG,R7 ; R7 <- NG (return value)
    MOV.L #MAXVerify,R12 ; if n>=MAXVerify then Program NG
    CMP /EQ R9,R12
    BT Program_end
    BRA Program_loop
    NOP
Program_OK .EQU $
```

```
MOV.L #OK,R7 % ; R7 <- OK (return value)
```


### 22.7.3 Erase Mode ( $\mathrm{n}=1$ for Addresses $\mathbf{H}^{\prime} \mathbf{0 0 0 0}-\mathbf{H}^{\prime} \mathbf{1 F F F F}$, $\mathbf{n}=\mathbf{2}$ for Addresses $\mathbf{H}^{\prime} \mathbf{2 0 0 0 0}$ H'3FFFF)

When erasing flash memory, the erase/erase-verify flowchart shown in figure 22.14 should be followed.

To perform data or program erasure, set the flash memory area to be erased in erase block register n (EBRn) at least $10 \mu \mathrm{~s}$ after setting the SWE bit to 1 in flash memory control register 1 (FLMCR1). Next, the watchdog timer is set to prevent overerasing in the event of program runaway, etc. Set $5.3 \mu$ s as the WDT overflow period. After this, preparation for erase mode (erase setup) is carried out by setting the ESUn bit in FLMCRn, and after the elapse of $200 \mu$ s or more, the operating mode is switched to erase mode by setting the En bit in FLMCRn. The time during which the En bit is set is the flash memory erase time. Set an erase time of 5 ms .

Note: With flash memory erasing, preprogramming (setting all memory data in the memory to be erased to all " 0 ") is not necessary before starting the erase procedure.

### 22.7.4 Erase-Verify Mode ( $\mathbf{n}=1$ for Addresses $\mathbf{H}^{\prime} \mathbf{0 0 0 0 0}-\mathbf{H}^{\prime} \mathbf{1 F F F F}, \mathbf{n}=2$ for Addresses $\mathbf{H}^{\prime} 20000-\mathrm{H}^{\prime}$ 3FFFF)

In erase-verify mode, data is read after memory has been erased to check whether it has been correctly erased.

After the elapse of the erase time, erase mode is exited (the En bit in FLMCRn is released, then the ESUn bit is released at least $10 \mu$ s later), the watchdog timer is released after the elapse of 10 $\mu \mathrm{s}$ or more, and the operating mode is switched to erase-verify mode by setting the EVn bit in FLMCRn. Before reading in erase-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of $20 \mu \mathrm{~s}$ or more. When the flash memory is read in this state (verify data is read in 32-bit units), the data at the latched address is read. Wait at least $2 \mu \mathrm{~s}$ after the dummy write before performing this read operation. If the read data has been erased (all " 1 "), a dummy write is performed to the next address, and erase-verify is performed. If the read data is unerased, set erase mode again and repeat the erase/erase-verify sequence in the same way. However, ensure that the erase/eraseverify sequence is not repeated more than 60 times. When verification is completed, exit eraseverify mode, and wait for at least $5 \mu \mathrm{~s}$. If erasure has been completed on all the erase blocks after completing erase-verify operations on all these blocks, release the SWE bit in FLMCR1. If there are any unerased blocks, set erase mode again, and repeat the erase/erase-verify sequence as before. However, ensure that the erase/erase-verify sequence is not repeated more than 60 times.


Figure 22.14 Erase/Erase-Verify Flowchart (Single Block Erase)

- Sample one-block erase program

The wait time set values (number of loops) are for the case where $\mathrm{f}=28.7 \mathrm{MHz}$. For other frequencies, the set value is given by the following expression:
Wait time $(\mu \mathrm{S}) \times \mathrm{f}(\mathrm{MHz}) \div 4$
The WDT overflow cycle set value is for the case where $\mathrm{f}=28.7 \mathrm{MHz}$. For other frequencies, ensure that the overflow cycle is a minimum of 5.3 ms .

## Registers Used

R5 (input): Memory block table pointer
R7 (output): OK (normal) or NG (error)
R0-3, 6, 8-9: Work registers

| FLMCR1 | .EQU | $\mathrm{H}^{\prime} 80$ |  |
| :---: | :---: | :---: | :---: |
| FLMCR2 | .EQU | $\mathrm{H}^{\prime} 81$ |  |
| EBR1 | .EQU | $\mathrm{H}^{\prime} 82$ |  |
| EBR2 | .EQU | $\mathrm{H}^{\prime} 83$ |  |
| Wait10u | .EQU | 72 |  |
| Wait2u | .EQU | 14 |  |
| Wait200u | .EQU | 1435 |  |
| Wait5m | .EQU | 35875 |  |
| Wait20u | .EQU | 144 |  |
| Wait5u | .EQU | 36 |  |
| WDT_TCSR | .EQU | $\mathrm{H}^{\prime} \mathrm{FFFF} 8610$ |  |
| WDT_9m | .EQU | $H^{\prime}$ A57D |  |
| SWESET | .EQU | $B^{\prime} 01000000$ |  |
| ESUSET | .EQU | $B^{\prime} 00100000$ |  |
| ESET | .EQU | $B^{\prime} 00000010$ |  |
| ECLEAR | .EQU | $B^{\prime} 11111101$ |  |
| ESUCLEAR | .EQU | B' 11011111 |  |
| EVSET | .EQU | $B^{\prime} 00001000$ |  |
| EVCLEAR | .EQU | $B^{\prime} 11110111$ |  |
| SWECLEAR | .EQU | $B^{\prime} 10111111$ |  |
| MAXErase | .EQU | 60 |  |
| ; |  |  |  |
| FlashErase | .EQU | \$ |  |
| MOV.L | \# $\mathrm{H}^{\prime} \mathrm{FF}$ | 500,R0 |  |
| LDC | R0, GB |  | ; Initialize GBR |
| MOV.L | \#1, R2 |  |  |


|  | MOV.L | \#Wait10u, R3 |  |
| :---: | :---: | :---: | :---: |
|  | MOV.L | \#FLMCR1,R0 |  |
|  | OR.B | \#SWESET, © (R0, GBR) | ; Set SWE |
| EWait_1 | SUBC | R2, R3 | ; Wait 10 ¢s |
|  | BF | EWait_1 |  |
| ; |  |  |  |
|  | MOV.L | \#0,R9 | ; Initialize n (R9) to 0 |
| ; |  |  |  |
|  | MOV.B | @ (6,R5) , R0 |  |
|  | MOV.B | R0, © (EBR1, GBR) | ; Erase memory block (EBR1) setting |
|  | MOV.B | @ (7,R5) , R0 |  |
|  | MOV.B | R0, © (EBR2, GBR) | ; Erase memory block (EBR2) setting |
| ; |  |  |  |
|  | MOV.L | \#FLMCR1,R0 |  |
|  | MOV.L | @R5, R6 | ; Erase memory block start address -> R6 |
|  | MOV.L | \# ${ }^{\prime}$ 020000,R7 |  |
|  | CMP /GT | R6, R7 |  |
|  | BT | EraseLoop |  |
|  | MOV.L | \#FLMCR2,R0 |  |
| ; |  |  |  |
| EraseLoop |  | .EQU \$ |  |
|  | MOV.L | \#WDT_TCSR,R1 | ; Enable WDT |
|  | MOV.w | \#WDT_9m, R3 | ; 9.2 ms cycle |
|  | MOV.W | R3, @R1 |  |
| ; |  |  |  |
|  | MOV.L | \#Wait200u,R3 |  |
|  | OR.B | \#ESUSET, @ (R0, GBR) | ; Set ESU |
| EWait_2 | SUBC | R2, R3 | ; Wait $200 \mu \mathrm{~s}$ |
|  | BF | EWait_2 |  |
| ; |  |  |  |
|  | MOV.L | \#Wait5m, R3 |  |
|  | OR.B | \#ESET, @ (R0, GBR) | ; Set E |
| EWait_3 | SUBC | R2, R3 | ; Wait 5 ms |
|  | BF | EWait_3 |  |
| ; |  |  |  |
|  | MOV.L | \#Wait10u, R3 |  |

```
\begin{tabular}{rll}
\(\quad\) AND.B & \#ECLEAR, @(R0,GBR) & ; Clear E \\
EWait_4 \begin{tabular}{rl} 
SUBC & R2,R3 \\
BF & EWait_4
\end{tabular} & ; Wait \(10 \mu \mathrm{~s}\)
\end{tabular}
;
    MOV.L #Wait10u,R3
    AND.B #ESUCLEAR,@ (R0,GBR) ; Clear ESU
EWait_5 SUBC R2,R3 ; Wait 10 \mus
;
    MOV.L #WDT_TCSR,R1 ; Disable WDT
    MOV.W #H'A55F,R3
    MOV.W R3,@R1
;
MOV.L #Wait20u,R3
    OR.B #EVSET,@(R0,GBR) ; Set EV
EWait_6 SUBC R2,R3 ; Wait 20 \mus
    BF EWait_6
;
MOV.L @R5,R6 ; Erase memory block start address -> R6
BlockVerify_1 .EQU $ ; Erase-verify
    MOV.L #H'FFFFFFFF,R8
    MOV.L R8,@R6 ; H'FF dummy write
    MOV.L #Wait2u,R3
EWait_7 SUBC R2,R3
    BF EWait_7
;
    MOV.L @R6+,R1 ; Read verify data
    CMP/EQ R8,R1
    BF BlockVerify_NG
    MOV.L @ (8,R5),R7
    CMP /EQ R6,R7 ; Check for last address of memory block
    BF BlockVerify_1
    MOV.L #Wait5u,R3
    AND.B #EVCLEAR, @ (R0,GBR) ; Clear EV
EWait_8 SUBC R2,R3 ; Wait 5 \mus
    BF EWait_8
;
7 1 6
```

    MOV.L #OK,R7 ; R7 <- OK (return value)
    BRA FlashErase_end ; Verify OK
    NOP
    ;
BlockVerify_NG .EQU \$
ADD.L \#1,R9 ; Verify NG, n <- n + 1
MOV.L \#Wait5u,R3
AND.B \#EVCLEAR,@ (R0,GBR) ; Clear EV
EWait_9 SUBC R2,R3 ; Wait 5 \mus
BF EWait_9
MOV.L \#MAXErase,R7 ; If n > MAXErase then erase NG
CMP/EQ R7,R9
BF EraseLoop
MOV.L \#NG,R7 ; R7 <- NG (return value)
FlashErase_end .EQU \$
MOV.L \#FLMCR1,R0
AND.B \#SWECLEAR,@ (R0,GBR) ; Clear SWE
;
RTS
NOP
;
; Memory block table Memory block start address: EBR value
.ALIGN 4
Flash_BlockData .EQU \$
EB0 .DATA.L H'00000000, H'00000100
EB1 .DATA.L H'00008000, H'00000200
EB2 .DATA.L H'00010000, H'00000400
EB3 .DATA.L H'00018000, H'00000800
EB4 .DATA.L H'00020000, H'00000001
EB5 .DATA.L H'00028000, H'00000002
EB6 .DATA.L H'00030000, H'00000004
EB7 .DATA.L H'00038000, H'00000008
EB8 .DATA.L H'0003F000, H'00000010
EB9 .DATA.L H'0003F400, H'00000020
EB10 .DATA.L H'0003F800, H'00000040
EB11 .DATA.L H'0003FC00, H'00000080
Dummy .DATA.L H'00040000

```

\subsection*{22.8 Protection}

There are two kinds of flash memory program/erase protection, hardware protection and software protection.

\subsection*{22.8.1 Hardware Protection}

Hardware protection refers to a state in which programming/erasing of flash memory is forcibly disabled or aborted. Hardware protection is reset by settings in flash memory control register 1 (FLMCR1), flash memory control register 2 (FLMCR2), erase block register 1 (EBR1), and erase block register 2 (EBR2). The FLMCR1, FLMCR2, EBR1, and EBR2 settings are retained in the error-protected state. (See table 22.8.)

Table 22.8 Hardware Protection
\begin{tabular}{|c|c|c|c|}
\hline & & \multicolumn{2}{|l|}{Function} \\
\hline Item & Description & Program & Erase \\
\hline FWP pin protection & - When a high level is input to the FWP pin, FLMCR1, FLMCR2, EBR1, and EBR2 are initialized, and the program/erase-protected state is entered. & Yes & Yes \\
\hline Reset/standby protection & \begin{tabular}{l}
- In a reset (including a WDT overflow reset) and in standby mode, FLMCR1, FLMCR2, EBR1, and EBR2 are initialized, and the program/erase-protected state is entered. \\
- In a reset via the \(\overline{R E S}\) pin, the reset state is not entered unless the \(\overline{\operatorname{RES}}\) pin is held low until oscillation stabilizes after powering on. In the case of a reset during operation, hold the \(\overline{\text { RES }}\) pin low for the \(\overline{\text { RES }}\) pulse width specified in the \(A C\) Characteristics section.
\end{tabular} & Yes & Yes \\
\hline
\end{tabular}

\subsection*{22.8.2 Software Protection}

Software protection can be implemented by setting the SWE bit in FLMCR1, erase block register 1 (EBR1), erase block register 2 (EBR2), and the RAMS bit in the RAM emulation register (RAMER). When software protection is in effect, setting the P1 or E1 bit in flash memory control register 1 (FLMCR1), or the P2 or E2 bit in flash memory control register 2 (FLMCR2), does not cause a transition to program mode or erase mode. (See table 22.9.)

Software protect can be enabled by setting the SWE bit of FLMCR1, block specification register 1 (EBR1), block specification register 2 (EBR2) and the RAMS bit of the RAM emulation register. During software protect, transition cannot be made to the program mode or the erase mode even when setting P1 or E1 bits of the flash memory control register 1 (FLMCR1), or P2 or E2 bits of flash memory control register 2 (FLMCR2). (See table 22.9.)

Table 22.9 Software Protection
\begin{tabular}{|c|c|c|c|}
\hline \multirow[b]{2}{*}{Item D} & \multirow[b]{2}{*}{Description} & \multicolumn{2}{|r|}{Function} \\
\hline & & & Erase \\
\hline SWE bit protection \({ }^{\text {- }}\) & \begin{tabular}{l}
- Clearing the SWE bit to 0 in FLMCR1 sets the program/erase-protected state for all blocks. \\
(Execute in on-chip RAM or external memory.)
\end{tabular} & Yes & Yes \\
\hline Block specification protection & \begin{tabular}{l}
- Erase protection can be set for individual blocks by settings in erase block register 1 (EBR1) and erase block register 2 (EBR2). \\
- Setting EBR1 and EBR2 to H'00 places all blocks in the erase-protected state.
\end{tabular} & - & Yes \\
\hline Emulation protection & - Setting the RAMS bit to 1 in the RAM emulation register (RAMER) places all blocks in the program/erase-protected state. & Yes & Yes \\
\hline
\end{tabular}

\subsection*{22.8.3 Error Protection}

In error protection, an error is detected when microcomputer runaway occurs during flash memory programming/erasing, or operation is not performed in accordance with the program/erase algorithm, and the program/erase operation is aborted. Aborting the program/erase operation prevents damage to the flash memory due to overprogramming or overerasing.

If the SH7051 malfunctions during flash memory programming/erasing, the FLER bit is set to 1 in FLMCR2 and the error protection state is entered. The FLMCR1, FLMCR2, EBR1, and EBR2 settings are retained, but program mode or erase mode is aborted at the point at which the error occurred. Program mode or erase mode cannot be re-entered by re-setting the P1, P2, E1, or E2 bit. However, PV1, PV2, EV1, and EV2 bit setting is enabled, and a transition can be made to verify mode.

FLER bit setting conditions are as follows:
1. When flash memory is read during programming/erasing (including a vector read or instruction fetch)
2. Immediately after exception handling (excluding a reset) during programming/erasing
3. When a SLEEP instruction (including software standby) is executed during programming/erasing
4. When the bus is released during programming/erasing

Error protection is released only by a reset and in hardware standby mode.
Figure 22.15 shows the flash memory state transition diagram.


Figure 22.15 Flash Memory State Transitions

\subsection*{22.9 Flash Memory Emulation in RAM}

Making a setting in the RAM emulation register (RAMER) enables part of RAM to be overlapped onto the flash memory area so that data to be written to flash memory can be emulated in RAM in real time. After the RAMER setting has been made, accesses can be made from the flash memory area or the RAM area overlapping flash memory. Emulation can be performed in user mode and user program mode. Figure 22.16 shows an example of emulation of real-time flash memory programming.


Figure 22.16 Flowchart for Flash Memory Emulation in RAM


Figure 22.17 Example of RAM Overlap Operation

\section*{Example in which Flash Memory Block Area (EB8) is Overlapped}
1. Set bits RAMS, RAM1, and RAM0 in RAMER to \(1,0,1\), to overlap part of RAM onto the area (EB8) for which real-time programming is required.
2. Real-time programming is performed using the overlapping RAM.
3. After the program data has been confirmed, the RAMS bit is cleared, releasing RAM overlap.
4. The data written in the overlapping RAM is written into the flash memory space (EB8).

Notes: 1. When the RAMS bit is set to 1, program/erase protection is enabled for all blocks regardless of the value of RAM1 and RAM0 (emulation protection). In this state, setting the P1 or E1 bit in flash memory control register 1 (FLMCR1), or the P2 or E2 bit in flash memory control register 2 (FLMCR2), will not cause a transition to program mode or erase mode. When actually programming or erasing a flash memory area, the RAMS bit should be cleared to 0 .
2. A RAM area cannot be erased by execution of software in accordance with the erase algorithm while flash memory emulation in RAM is being used.

\subsection*{22.10 Note on Flash Memory Programming/Erasing}

In the on-board programming modes (user mode and user program mode), NMI input should be disabled to give top priority to the program/erase operations (including RAM emulation).

\subsection*{22.11 Flash Memory Programmer Mode}

Programs and data can be written and erased in programmer mode as well as in the on-board programming modes. In programmer mode, flash memory read mode, auto-program mode, autoerase mode, and status read mode are supported. In auto-program mode, auto-erase mode, and status read mode, a status polling procedure is used, and in status read mode, detailed internal signals are output after execution of an auto-program or auto-erase operation.

In programmer mode, set the mode pins to PLL x2 mode (see table 22.10) and use a 6 MHz input clock. The LSI will then operate at 12 MHz .

Table 22.10 shows the pin settings for programmer mode. For the pin names in programmer mode, see section 1.3.2, Pin Arrangement by Mode).

Table 22.10 Programming Mode Pin Settings
\begin{tabular}{ll} 
Pin Names & Settings \\
\hline Mode pin: MD3, MD2, MD1, MD0 & 1101 (PLL \(\times 2\) ) \\
\hline FWE pin & \begin{tabular}{l} 
High level input (in auto-program and auto- \\
erase modes)
\end{tabular} \\
\hline\(\overline{\text { RES pin }}\) & Power-on reset circuit \\
\hline XTAL, EXTAL, PLLVcc, PLLCAP, and PLLVss pins Oscillator circuit \\
\hline Note: \begin{tabular}{l} 
During the programming mode, polarity of the FWP pin is inverted and becomes the FWE \\
(flash write enable) pin.
\end{tabular}
\end{tabular}

\subsection*{22.11.1 Socket Adapter Pin Correspondence Diagrams}

Connect the socket adapter to the chip as shown in figures 22.19 and 22.20. This will enable conversion to a 32-pin arrangement. The on-chip ROM memory map is shown in figure 22.18, and socket adapter pin correspondence diagrams in figures 22.19 and 22.20.


Figure 22.18 On-Chip ROM Memory Map


Figure 22.19 Socket Adapter Pin Correspondence Diagram (SH7044)


Figure 22.20 Socket Adapter Pin Correspondence Diagram (SH7045)

\subsection*{22.11.2 Programmer Mode Operation}

Table 22.11 shows how the different operating modes are set when using programmer mode, and table 22.12 lists the commands used in programmer mode. Details of each mode are given below.
- Memory Read Mode

Memory read mode supports byte reads.
- Auto-Program Mode

Auto-program mode supports programming of 128 bytes at a time. Status polling is used to confirm the end of auto-programming.
- Auto-Erase Mode

Auto-erase mode supports automatic erasing of the entire flash memory. Status polling is used to confirm the end of auto-programming.
- Status Read Mode

Status polling is used for auto-programming and auto-erasing, and normal termination can be confirmed by reading the I/O6 signal. In status read mode, error information is output if an error occurs.

Table 22.11 Settings for Various Operating Modes In Programmer Mode
\begin{tabular}{lllllll} 
& \multicolumn{7}{c}{ Pin names } \\
\cline { 2 - 7 } Mode & FWE & \(\overline{\mathbf{C E}}\) & \(\overline{\mathbf{O E}}\) & \(\overline{\mathbf{W E}}\) & I/O7-0 & A17-0 \\
\hline Read & H or L & L & L & H & Data output & Ain \\
\hline Output disable & H or L & L & H & H & Hi-z & Ain \\
\hline Command write & H or L & L & H & L & Data input & *Ain \\
\hline Chip disable & H or L & H & X & X & Hi-z & Ain \\
\hline
\end{tabular}

Notes: *Ain indicates that there is also address input in auto-program mode.
1. Chip disable is not a standby state; internally, it is an operation state.
2. For command writes in auto-program and auto-erase modes, input a high level to the FWE pin.

Table 22.12 Commands of the Programmer Mode
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline & \multirow[t]{2}{*}{Number of Cycles} & \multicolumn{3}{|c|}{First Cycle} & \multicolumn{3}{|r|}{Second Cycle} \\
\hline Command Name & & Mode & Address & Data & Mode & Address & Data \\
\hline Memory read mode & 1+n & write & X & H'00 & read & RA & Dout \\
\hline Auto-program mode & 129 & write & X & H'40 & write & WA & Din \\
\hline Auto-erase mode & 2 & write & X & H'20 & write & X & H'20 \\
\hline Status read mode & 2 & write & X & H'71 & write & X & H'71 \\
\hline
\end{tabular}

Notes: 1. In auto-program mode, 129 cycles are required for command writing by a simultaneous 128-byte write.
2. In memory read mode, the number of cycles depends on the number of address write cycles ( n ).

\subsection*{22.11.3 Memory Read Mode}

Table 22.13 AC Characteristics in Transition to Memory Read Mode (Conditions: \(\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{~V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Notes \\
\hline Command write cycle & \(\mathrm{t}_{\text {nxtc }}\) & 20 & & \(\mu \mathrm{~s}\) & \\
\hline\(\overline{\overline{C E}}\) hold time & \(\mathrm{t}_{\text {ceh }}\) & 0 & & ns & \\
\hline\(\overline{\overline{C E}}\) setup time & \(\mathrm{t}_{\text {ces }}\) & 0 & ns & \\
\hline Data hold time & \(\mathrm{t}_{\text {dh }}\) & 50 & & ns \\
\hline Data setup time & \(\mathrm{t}_{\mathrm{ds}}\) & 50 & & ns \\
\hline Write pulse width & \(\mathrm{t}_{\text {wep }}\) & 70 & & ns \\
\hline\(\overline{\text { WE rise time }}\) & \(\mathrm{t}_{\mathrm{r}}\) & & 30 & ns \\
\hline\(\overline{\text { WE fall time }}\) & \(\mathrm{t}_{\mathrm{f}}\) & & 30 & ns \\
\hline
\end{tabular}


Figure 22.21 Timing Waveforms for Memory Read after Memory Write
Table 22.14 AC Characteristics in Transition from Memory Read Mode to Another Mode (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{V}_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{T}_{\mathrm{a}}=\mathbf{2 5}{ }^{\circ} \mathrm{C} \pm \mathbf{5}^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Notes \\
\hline Command write cycle & \(\mathrm{t}_{\text {nxtc }}\) & 20 & & \(\mu \mathrm{~s}\) & \\
\hline\(\overline{\mathrm{CE}}\) hold time & \(\mathrm{t}_{\text {ceh }}\) & 0 & ns & \\
\hline\(\overline{\mathrm{CE}}\) setup time & \(\mathrm{t}_{\text {ces }}\) & 0 & ns \\
\hline Data hold time & \(\mathrm{t}_{\mathrm{dh}}\) & 50 & ns \\
\hline Data setup time & \(\mathrm{t}_{\mathrm{ds}}\) & 50 & & ns \\
\hline Write pulse width & \(\mathrm{t}_{\text {wep }}\) & 70 & & ns \\
\hline\(\overline{\mathrm{WE}}\) rise time & \(\mathrm{t}_{\mathrm{r}}\) & & 30 & ns \\
\hline\(\overline{\mathrm{WE}}\) fall time & \(\mathrm{t}_{\mathrm{f}}\) & & 30 & ns \\
\hline
\end{tabular}


Figure 22.22 Timing Waveforms in Transition from Memory Read Mode to Another Mode
Table 22.15 AC Characteristics in Memory Read Mode (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Notes \\
\hline Access time & \(\mathrm{t}_{\mathrm{acc}}\) & & 20 & \(\mu \mathrm{~s}\) & \\
\hline\(\overline{\mathrm{CE}}\) output delay time & \(\mathrm{t}_{\mathrm{ce}}\) & & 150 & ns & \\
\hline\(\overline{\mathrm{OE}}\) output delay time & \(\mathrm{t}_{\mathrm{ee}}\) & & 150 & ns & \\
\hline Output disable delay time & \(\mathrm{t}_{\mathrm{df}}\) & & 100 & ns & \\
\hline Data output hold time & \(\mathrm{t}_{\mathrm{oh}}\) & 5 & & ns & \\
\hline
\end{tabular}


Figure \(22.23 \overline{\mathbf{C E}}\) and \(\overline{\mathrm{OE}}\) Enable State Read Timing Waveforms


Figure \(22.24 \overline{\mathbf{C E}}\) and \(\overline{\text { OE }}\) Clock System Read Timing Waveforms

\subsection*{22.11.4 Auto-Program Mode}
1. In auto-program mode, 128 bytes are programmed simultaneously. This should be carried out by executing 128 consecutive byte transfers.
2. A 128 -byte data transfer is necessary even when programming fewer than 128 bytes. In this case, H'FF data must be written to the extra addresses.
3. The lower 8 bits of the transfer address must be \(\mathrm{H}^{\prime} 00\) or \(\mathrm{H}^{\prime} 80\). If a value other than an effective address is input, processing will switch to a memory write operation but a write error will be flagged.
4. Memory address transfer is performed in the second cycle (figure 22.24). Do not perform transfer after the second cycle.
5. Do not perform a command write during a programming operation.
6. Perform one auto-program operation for a 128-byte block for each address. Characteristics are not guaranteed for two or more additional programming operations.
7. Confirm normal end of auto-programming by checking I/O6. Alternatively, status read mode can also be used for this purpose (I/O7 status output uses the auto-program operation end identification pin).
8. Status polling I/O6 and I/O7 pin information is retained until the next command write. As long as the next command write has not been performed, reading is possible by enabling \(\overline{\mathrm{CE}}\) and \(\overline{\mathrm{OE}}\).

Table 22.16 AC Characteristics in Auto-Program Mode (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}\) )
\begin{tabular}{lllll} 
& Symbol & Min & Max & Unit \\
Item & Notes \\
\hline Command write cycle & \(\mathrm{t}_{\mathrm{nxtc}}\) & 20 & \(\mu \mathrm{~s}\) \\
\hline\(\overline{\mathrm{CE}}\) hold time & \(\mathrm{t}_{\text {ceh }}\) & 0 & ns \\
\hline\(\overline{\mathrm{CE}}\) setup time & \(\mathrm{t}_{\text {ces }}\) & 0 & ns \\
\hline Data hold time & \(\mathrm{t}_{\mathrm{dh}}\) & 50 & ns \\
\hline Data setup time & \(\mathrm{t}_{\mathrm{ds}}\) & 50 & ns \\
\hline Write pulse width & \(\mathrm{t}_{\text {wep }}\) & 70 & ns \\
\hline Status polling start time & \(\mathrm{t}_{\text {wsts }}\) & 1 & ms \\
\hline Status polling access time & \(\mathrm{t}_{\mathrm{spa}}\) & & 150 & ns \\
\hline Address setup time & \(\mathrm{t}_{\mathrm{as}}\) & 0 & & ns \\
\hline Address hold time & \(\mathrm{t}_{\mathrm{ah}}\) & 60 & & ns \\
\hline Memory write time & \(\mathrm{t}_{\text {write }}\) & 1 & 3000 & ms \\
\hline Write setup time & \(\mathrm{t}_{\mathrm{pns}}\) & 100 & ns \\
\hline Write end setup time & \(\mathrm{t}_{\text {pnh }}\) & 100 & ns \\
\hline\(\overline{\text { WE rise time }}\) & \(\mathrm{t}_{\mathrm{r}}\) & & 30 & ns \\
\hline\(\overline{\text { WE fall time }}\) & \(\mathrm{t}_{\mathrm{f}}\) & & 30 & ns \\
\hline
\end{tabular}


Figure 22.25 Auto-Program Mode Timing Waveforms

\subsection*{22.11.5 Auto-Erase Mode}
1. Auto-erase mode supports only entire memory erasing.
2. Do not perform a command write during auto-erasing.
3. Confirm normal end of auto-erasing by checking I/O6. Alternatively, status read mode can also be used for this purpose (I/O7 status output uses the auto-erase operation end identification pin).
4. Status polling I/O6 and I/O7 pin information is retained until the next command write. As long as the next command write has not been performed, reading is possible by enabling \(\overline{\mathrm{CE}}\) and \(\overline{\mathrm{OE}}\).

Table 22.17 AC Characteristics in Auto-Erase Mode (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}\) )
\begin{tabular}{lllll} 
Item & Symbol & Min & Max & Unit \\
\hline Notes \\
\hline Command write cycle & \(\mathrm{t}_{\text {nxt }}\) & 20 & \(\mu \mathrm{~s}\) \\
\hline\(\overline{\text { CE }}\) hold time & \(\mathrm{t}_{\text {ceh }}\) & 0 & ns \\
\hline\(\overline{\text { CE }}\) setup time & \(\mathrm{t}_{\text {ces }}\) & 0 & ns \\
\hline Data hold time & \(\mathrm{t}_{\text {dh }}\) & 50 & ns \\
\hline Data setup time & \(\mathrm{t}_{\text {ds }}\) & 50 & ns \\
\hline Write pulse width & \(\mathrm{t}_{\text {wep }}\) & 70 & ns \\
\hline Status polling start time & \(\mathrm{t}_{\text {ests }}\) & 1 & ms \\
\hline Status polling access time & \(\mathrm{t}_{\text {spa }}\) & & 150 & ns \\
\hline Memory erase time & \(\mathrm{t}_{\text {erase }}\) & 100 & 40000 & ms \\
\hline Erase setup time & \(\mathrm{t}_{\text {ens }}\) & 100 & & ns \\
\hline Erase end setup time & \(\mathrm{t}_{\text {enh }}\) & 100 & ns \\
\hline\(\overline{\text { WE } \text { rise time }}\) & \(\mathrm{t}_{\mathrm{r}}\) & & 30 & ns \\
\hline\(\overline{\text { WE fall time }}\) & \(\mathrm{t}_{\mathrm{f}}\) & & 30 & ns \\
\hline
\end{tabular}


Figure 22.26 Auto-Erase Mode Timing Waveforms

\subsection*{22.11.6 Status Read Mode}

Table 22.18 AC Characteristics in Status Read Mode (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm 10 \%\), \(\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=25^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Notes \\
\hline Read time after command write & \(\mathrm{t}_{\text {std }}\) & 20 & & \(\mu \mathrm{~s}\) \\
\hline\(\overline{\overline{C E}}\) hold time & \(\mathrm{t}_{\text {ceh }}\) & 0 & ns \\
\hline\(\overline{\overline{C E}}\) setup time & \(\mathrm{t}_{\text {ces }}\) & 0 & ns \\
\hline Data hold time & \(\mathrm{t}_{\mathrm{dh}}\) & 50 & ns \\
\hline Data setup time & \(\mathrm{t}_{\mathrm{ds}}\) & 50 & & ns \\
\hline Write pulse width & \(\mathrm{t}_{\text {wep }}\) & 70 & & ns \\
\hline\(\overline{\overline{O E}}\) output delay time & \(\mathrm{t}_{\text {oe }}\) & & 150 & ns \\
\hline Disable delay time & \(\mathrm{t}_{\mathrm{df}}\) & & 100 & ns \\
\hline\(\overline{\overline{\mathrm{CE}} \text { output delay time }}\) & \(\mathrm{t}_{\mathrm{ce}}\) & & 150 & ns \\
\hline\(\overline{\overline{W E}}\) rise time & \(\mathrm{t}_{\mathrm{r}}\) & & 30 & ns \\
\hline\(\overline{\overline{W E}}\) fall time & \(\mathrm{t}_{\mathrm{f}}\) & & 30 & ns \\
\hline
\end{tabular}


Figure 22.27 Status Read Mode Timing Waveforms
Table 22.19 Return Commands for the Status Read Mode
\begin{tabular}{|c|c|c|c|c|c|c|c|c|}
\hline Pin Name & I/07 & I/O6 & I/O5 & I/O4 & I/O3 & I/O2 & I/01 & I/OO \\
\hline Attribute & Normal end identification & Command error & Programming error & Erase error & - & - & Programming or erase count exceeded & Effective address error \\
\hline Initial value & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\
\hline \multirow[t]{2}{*}{Indications} & \multirow[t]{2}{*}{\begin{tabular}{l}
Normal \\
end: 0 \\
Abnormal end: 1
\end{tabular}} & \begin{tabular}{l}
Command \\
Error: 1
\end{tabular} & \multirow[t]{2}{*}{\begin{tabular}{l}
Programming \\
Error: 1 \\
Otherwise: 0
\end{tabular}} & \begin{tabular}{l}
Erasing \\
Error: 1
\end{tabular} & \multirow[t]{2}{*}{-} & \multirow[t]{2}{*}{-} & Count exceeded: 1 & Effective address \\
\hline & & Otherwise: 0 & & Otherwise: 0 & & & Otherwise: 0 & \begin{tabular}{l}
Error: 1 \\
Otherwise: 0
\end{tabular} \\
\hline
\end{tabular}

Note: D2 and D3 are undefined at present.

\subsection*{22.11.7 Status Polling}
1. I/O7 status polling is a flag that indicates the operating status in auto-program/auto-erase mode.
2. I/O6 status polling is a flag that indicates a normal or abnormal end in auto-program/auto-erase mode.

Table 22.20 Status Polling Output Truth Table
\begin{tabular}{lllll} 
Pin Name & \begin{tabular}{l} 
During Internal \\
Operation
\end{tabular} & Abnormal End & Normal End \\
\hline I/O7 & 0 & 1 & 0 & 1 \\
\hline I/O6 & 0 & 0 & 1 & 1 \\
\hline I/O5-0 & 0 & 0 & 0 & 0 \\
\hline
\end{tabular}

\subsection*{22.11.8 Programmer Mode Transition Time}

Commands cannot be accepted during the oscillation stabilization period or the programmer mode setup period. After the programmer mode setup time, a transition is made to memory read mode.

Table 22.21 Stipulated Transition Times to Command Wait State
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Notes \\
\hline \begin{tabular}{l} 
Standby release \\
(oscillation stabilization time)
\end{tabular} & \(\mathrm{t}_{\mathrm{osc} 1}\) & 10 & ms & \\
\hline Programmer mode setup time & \(\mathrm{t}_{\mathrm{bmv}}\) & 10 & ms & \\
\hline \(\mathrm{~V}_{\mathrm{Cc}}\) hold time & \(\mathrm{t}_{\mathrm{dwn}}\) & 0 & ms & \\
\hline
\end{tabular}


Note : For the level of FWE input pin, set \(\mathrm{V}_{\mathrm{IL}}\) when using other than the automatic write mode and automatic erase mode.

Figure 22.28 Oscillation Stabilization Time and Boot Program Transfer Time

\subsection*{22.11.9 Cautions Concerning Memory Programming}
1. When programming addresses which have previously been programmed, carry out autoerasing before auto-programming.
2. When performing programming using PROM mode on a chip that has been programmed/erased in an on-board programming mode, auto-erasing is recommended before carrying out auto-programming.

Notes: 1. The flash memory is initially in the erased state when the device is shipped by Renesas Technology. For other chips for which the erasure history is unknown, it is recommended that auto-erasing be executed to check and supplement the initialization (erase) level.
2. Auto-programming should be performed once only on the same address block. Additional programming cannot be performed on previously programmed address blocks.

\section*{Section 23 RAM}

\subsection*{23.1 Overview}

The SH7040 series has 4 kbytes of on-chip RAM. The on-chip RAM is linked to the CPU and direct memory access controller (DMAC)/data transfer controller (DTC) with a 32-bit data bus (figure 23.1). The CPU can access data in the on-chip RAM in 8,16 , or 32 bit widths. The DMAC can access 8 or 16 bit widths. On-chip RAM data can always be accessed in one state, making the RAM ideal for use as a program area, stack area, or data area, which require high-speed access. The contents of the on-chip RAM are held in both the sleep and standby modes. Memory area 0 addresses H'FFFFF000-H'FFFFFFFF are allocated to the on-chip RAM.


Figure 23.1 Block Diagram of RAM

\subsection*{23.2 Operation}

The on-chip RAM is accessed by accessing addresses H'FFFFF000-H'FFFFFFFF. On-chip RAM is also used as cache memory. There are 2 kbytes of on-chip RAM space during cache use. See section 9, Cache Memory (CAC), for details.

\section*{Section 24 Power-Down State}

\subsection*{24.1 Overview}

In the power-down state, the CPU functions are halted. This enables a great reduction in power consumption.

\subsection*{24.1.1 Power-Down States}

The power-down state is effected by the following two modes:
- Sleep mode
- Standby mode

Table 24.1 describes the transition conditions for entering the modes from the program execution state as well as the CPU and peripheral function status in each mode and the procedures for canceling each mode.

Table 24.1 Power-Down State Conditions
\begin{tabular}{|c|c|c|c|c|c|c|c|c|}
\hline & & \multicolumn{6}{|c|}{State} & \\
\hline Mode & Entering Procedure & Clock & CPU & On-Chip Peripheral Modules & \begin{tabular}{l}
CPU \\
Registers
\end{tabular} & & I/O Ports & Canceling Procedure \\
\hline Sleep & Execute SLEEP instruction with SBY bit set to 0 in SBYCR & Run & Halt & Run & Held & Held & Held & \begin{tabular}{l}
- Interrupt \\
- DMAC/DTC address error \\
- Power-on reset \\
- Manual reset
\end{tabular} \\
\hline \begin{tabular}{l}
Stand- \\
by
\end{tabular} & Execute SLEEP instruction with SBY bit set to 1 in SBYCR & Halt & Halt & Halt** & Held & Held & Held or high impedance*2 & \begin{tabular}{l}
- NMI interrupt \\
- Power-on reset \\
- Manual reset
\end{tabular} \\
\hline
\end{tabular}

Notes: SBYCR: standby control register. SBY: standby bit
*1 Some bits within on-chip peripheral module registers are initialized by the standby mode; some are not. Refer to table 24.3, Register States in the Standby Mode, in section 24.4.1, Transition to Standby Mode. Also refer to the register descriptions for each peripheral module.
*2 The status of the I/O port in standby mode is set by the port high impedance bit (HIZ) of the SBYCR. Refer to section 24.2, Standby Control Register (SBYCR). For pin status other than for the I/O port, refer to Appendix C, Pin Status.

\subsection*{24.1.2 Related Register}

Table 24.2 shows the register used for power-down state control.
Table 24.2 Related Register
\begin{tabular}{llllll} 
Name & Abbreviation & R/W & Initial Value & Address & Access Size \\
\hline Standby control register & SBYCR & R/W & H'1F & H'FFFF8614 & \(8,16,32\) \\
\hline
\end{tabular}

\subsection*{24.2 Standby Control Register (SBYCR)}

The standby control register (SBYCR) is a read/write 8-bit register that sets the transition to standby mode, and the port status in standby mode. The SBYCR is initialized to H'1F when reset.
\begin{tabular}{rc|c|c|c|c|c|c|c|}
\multicolumn{1}{c}{ Bit: } & 7 & 6 & 5 & 4 & 3 & 2 & 1 & 0 \\
\cline { 2 - 9 } & SBY & HIZ & - & - & - & - & - & - \\
\hline Initial value: & 0 & 0 & 0 & 1 & 1 & 1 & 1 & 1 \\
R/W: & R/W & R/W & \(R\) & \(R\) & \(R\) & \(R\) & \(R\) & \(R\)
\end{tabular}
- Bit 7—Standby (SBY): Specifies transition to the standby mode. The SBY bit cannot be set to 1 while the watchdog timer is running (when the timer enable bit (TME) of the WDT timer control/status register (TCSR) is set to 1 ). To enter the standby mode, always halt the WDT by 0 clearing the TME bit, then set the SBY bit.
\begin{tabular}{ll} 
Bit 7: SBY & Description \\
\hline 0 & Executing SLEEP instruction puts the LSI into sleep mode (initial value) \\
\hline 1 & Executing SLEEP instruction puts the LSI into standby mode \\
\hline
\end{tabular}
- Bit 6-Port High Impedance (HIZ): In the standby mode, this bit selects whether to set the I/O port pin to high impedance or hold the pin status. The HIZ bit cannot be set to 1 when the TME bit of the WDT timer control/status register (TCSR) is set to 1 . When making the I/O port pin status high impedance, always clear the TME bit to 0 before setting the HIZ bit.
\begin{tabular}{ll} 
Bit 6: HIZ & Description \\
\hline 0 & Holds pin status while in standby mode (initial value) \\
\hline 1 & Keeps pin at high impedance while in standby mode \\
\hline
\end{tabular}
- Bits 5-0—Reserved: Bit 5 always reads as 0 . Always write 0 to bit 5. Bits \(4-0\) always read as 1. Always write 1 to these bits.

\subsection*{24.3 Sleep Mode}

\subsection*{24.3.1 Transition to Sleep Mode}

Executing the SLEEP instruction when the SBY bit of SBYCR is 0 causes a transition from the program execution state to the sleep mode. Although the CPU halts immediately after executing the SLEEP instruction, the contents of its internal registers remain unchanged. The on-chip peripheral modules continue to run during the sleep mode.

\subsection*{24.3.2 Canceling Sleep Mode}

Sleep mode is canceled by an interrupt, DMAC/DTC address error, power-on reset, or manual reset.

Cancellation by an Interrupt: When an interrupt occurs, the sleep mode is canceled and interrupt exception processing is executed. The sleep mode is not canceled if the interrupt cannot be accepted because its priority level is equal to or less than the mask level set in the CPU's status register (SR) or if an interrupt by an on-chip peripheral module is disabled at the peripheral module.

Cancellation by a DMAC/DTC Address Error: If a DMAC/DTC address error occurs, the sleep mode is canceled and DMAC/DTC address error exception processing is executed.

Cancellation by a Power-On Reset: A power-on reset resulting from setting the \(\overline{\mathrm{RES}}\) pin to low level cancels the sleep mode.

Cancellation by a Manual Reset: When the \(\overline{\text { MRES }}\) pin is set to low level while the \(\overline{\text { RES }}\) pin is at high level, a manual reset occurs and the sleep mode is canceled.

\subsection*{24.4 Standby Mode}

\subsection*{24.4.1 Transition to Standby Mode}

To enter the standby mode, set the SBY bit to 1 in SBYCR, then execute the SLEEP instruction. The LSI moves from the program execution state to the standby mode. In the standby mode, power consumption is greatly reduced by halting not only the CPU, but the clock and on-chip peripheral modules as well. CPU register contents and on-chip RAM data are held as long as the prescribed voltages are applied. The register contents of some on-chip peripheral modules are initialized, but some are not (table 24.3). The I/O port status can be selected as held or high impedance by the port high impedance bit (HIZ) of the SBYCR. For pin status other than for the I/O port, refer to Appendix C, Pin States.

Table 24.3 Register States in the Standby Mode
\begin{tabular}{lllll} 
Module & Registers Initialized & \begin{tabular}{l} 
Registers that \\
Retain Data
\end{tabular} & \begin{tabular}{l} 
Registers with \\
Undefined Contents
\end{tabular} \\
\hline \begin{tabular}{llll} 
Interrupt controller \\
(INTC)
\end{tabular} & - & All registers & - & \\
\hline \begin{tabular}{llll} 
User break controller \\
(UBC)
\end{tabular} & - & All registers & - \\
\hline \begin{tabular}{llll} 
Data transfer controller & All registers (excluding transfer \\
(DTC)
\end{tabular} & - & - \\
\hline data in memory and DTDR)
\end{tabular}
Serial communication • Receive data register (RDR) - -
- Transmit data register (TDR)
- Serial mode register (SMR)
- Serial control register (SCR)
- Serial status register (SSR)
- Bit rate register (BBR)
\begin{tabular}{llll}
\hline A/D converter (A/D) & All registers & - & - \\
\hline \begin{tabular}{l} 
Compare match timer
\end{tabular} & All registers & - & - \\
\((\) CMT \()\) & & \\
\hline
\end{tabular}

Table 24.3 Register States in the Standby Mode (cont)
\begin{tabular}{llll} 
Module & Registers Initialized & \begin{tabular}{l} 
Registers that \\
Retain Data
\end{tabular} & \begin{tabular}{l} 
Registers with \\
Undefined Contents
\end{tabular} \\
\hline \begin{tabular}{l} 
Pin function controller \\
(PFC)
\end{tabular} & - & All registers & - \\
\hline I/O port (I/O) & - & All registers & - \\
\hline \begin{tabular}{l} 
Power-down state \\
related
\end{tabular} & - & \begin{tabular}{l} 
Standby control \\
register (SBYCR)
\end{tabular} & - \\
\hline
\end{tabular}

\subsection*{24.4.2 Canceling the Standby Mode}

The standby mode is canceled by an NMI interrupt, a power-on reset, or a manual reset.
Cancellation by an NMI: Clock oscillation starts when a rising edge or falling edge (selected by the NMI edge select bit (NMIE) of the interrupt control register (ICR) of the INTC) is detected in the NMI signal. This clock is supplied only to the watchdog timer (WDT). A WDT overflow occurs if the time established by the clock select bits (CKS2-CKS0) in the TCSR of the WDT elapses before transition to the standby mode. The occurrence of this overflow is used to indicate that the clock has stabilized, so the clock is supplied to the entire chip, the standby mode is canceled, and NMI exception processing begins.

When canceling standby mode with NMI interrupts, set the CKS2-CKS0 bits so that the WDT overflow period is longer than the oscillation stabilization time.

When canceling standby mode with an NMI pin set for falling edge, be sure that the NMI pin level upon entering standby (when the clock is halted) is high level, and that the NMI pin level upon returning from standby (when the clock starts after oscillation stabilization) is low level. When canceling standby mode with an NMI pin set for rising edge, be sure that the NMI pin level upon entering standby (when the clock is halted) is low level, and that the NMI pin level upon returning from standby (when the clock starts after oscillation stabilization) is high level.

Cancellation by a Power-On Reset: A power-on reset caused by setting the \(\overline{\operatorname{RES}}\) pin to low level cancels the standby mode.

\subsection*{24.4.3 Standby Mode Application Example}

This example describes a transition to standby mode on the falling edge of an NMI signal, and a cancellation on the rising edge of the NMI signal. The timing is shown in figure 24.1.

When the NMI pin is changed from high to low level while the NMI edge select bit (NMIE) of the ICR is set to 0 (falling edge detection), the NMI interrupt is accepted. When the NMIE bit is set to 1 (rising edge detection) by an NMI exception service routine, the standby bit (SBY) of the SBYCR is set to 1 , and a SLEEP instruction is executed, standby mode is entered. Thereafter, standby mode is canceled when the NMI pin is changed from low to high level.


Figure 24.1 Standby Mode NMI Timing (Application Example)

\section*{Section 25 Electrical Characteristics (5V, 28.7 MHz Version)}

\subsection*{25.1 Absolute Maximum Ratings}

Table 25.1 shows the absolute maximum ratings.
Table 25.1 Absolute Maximum Ratings
\begin{tabular}{llll} 
Item & Symbol & Rating & Unit \\
\hline Power supply voltage & \(\mathrm{V}_{\mathrm{cC}}\) & -0.3 to +7.0 & V \\
\hline Programmable voltage (ZTAT version only) & \(\mathrm{V}_{\mathrm{PP}}\) & -0.3 to +13.5 & V \\
\hline Input voltage (other than A/D ports) & \(\mathrm{V}_{\text {in }}\) & -0.3 to \(\mathrm{V}_{\mathrm{cC}}+0.3\) & V \\
\hline Input voltage (A/D ports) & \(\mathrm{V}_{\text {in }}\) & -0.3 to \(\mathrm{AV} \mathrm{CC}+0.3\) & V \\
\hline Analog supply voltage & \(\mathrm{AV}_{\mathrm{cC}}\) & -0.3 to +7.0 & V \\
\hline Analog reference voltage (QFP-144 only) & \(\mathrm{AV}_{\text {ref }}\) & -0.3 to \(\mathrm{AV} \mathrm{CC}+0.3\) & V \\
\hline Analog input voltage & \(\mathrm{V}_{\mathrm{AN}}\) & -0.3 to \(\mathrm{AV} \mathrm{CC}+0.3\) & V \\
\hline Operating temperature & \(\mathrm{T}_{\text {opr }}\) & -20 to \(+75^{* 1}\) & \({ }^{\circ} \mathrm{C}\) \\
\hline Programming temperature (ZTAT version only) & \(\mathrm{T}_{\text {we }}\) & -20 to \(+75^{* 2}\) & \({ }^{\circ} \mathrm{C}\) \\
\hline Storage temperature & \(\mathrm{T}_{\text {stg }}\) & -55 to +125 & \({ }^{\circ} \mathrm{C}\) \\
\hline
\end{tabular}

Notes: Operating the LSI in excess of the absolute maximum ratings may result in permanent damage.
*1 Normal Products : \(\mathrm{T}_{\text {OPR }}=-40\) to \(+85^{\circ} \mathrm{C}\) for wide-temperature range products.
*2 Normal Products: \(\mathrm{T}_{\mathrm{we}}=-20\) to \(+85^{\circ} \mathrm{C}\) for wide-temperature range products.

\subsection*{25.2 DC Characteristics}

Table 25.2 DC Characteristics (Conditions: \(V_{C C}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(A V_{\text {CC }}=V_{\text {CC }} \pm \mathbf{1 0 \%}, A V_{\text {ref }}=4.5 \mathrm{~V}\) to \(A V_{\mathrm{CC}}, V_{\mathrm{SS}}=A V_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{Ta}=\mathbf{- 2 0}\) to \(+\mathbf{7 5}{ }^{\circ} \mathrm{C}\) )

Measurement
\begin{tabular}{|c|c|c|c|c|c|c|}
\hline Item & Pin Symbol & Min & Typ & Max & Unit & Conditions \\
\hline \multirow[t]{4}{*}{Input highlevel voltage} & \begin{tabular}{l}
\(\overline{R E S}\), NMI, MD3- \(V_{I H}\) \\
MD0, PA2, PA5, \\
PA6-PA9, PE0PE15, FWP
\end{tabular} & \(\mathrm{V}_{\mathrm{CC}}\) - & - & \(V_{C C}+0.3\) & V & - \\
\hline & EXTAL & \(\mathrm{V}_{\mathrm{cc}} \times\) & - & \(V_{C C}+0.3\) & V & - \\
\hline & A/D port & 2.2 & - & \(\mathrm{AV}_{\mathrm{cc}}+0.3\) & V & - \\
\hline & Other input pins & 2.2 & - & \(V_{C C}+0.3\) & V & - \\
\hline \multirow[t]{2}{*}{\begin{tabular}{l}
Input low- \\
level \\
voltage
\end{tabular}} & \[
\begin{aligned}
& \hline \overline{\mathrm{RES}}, \mathrm{NMI}, \mathrm{MD} 3-\mathrm{V}_{\mathrm{IL}} \\
& \text { MDO, PA2, PA5, } \\
& \text { PA6-PA9, PE0-- } \\
& \text { PE15, FWP }
\end{aligned}
\] & -0.3 & - & 0.5 & V & - \\
\hline & Other input pins & -0.3 & - & 0.8 & V & - \\
\hline \multicolumn{2}{|l|}{Schmitt PA2, PA5, PA6trigger input PA9, PE0-PE15 voltage} & 0.4 & - & - & V & \[
\begin{aligned}
& \mathrm{VT}^{+} \geq \mathrm{V}_{\mathrm{CC}}-0.7 \mathrm{~V}(\mathrm{~min}) \\
& \mathrm{VT}^{-} \leq 0.5 \mathrm{~V}(\max )
\end{aligned}
\] \\
\hline \multirow[t]{3}{*}{Input leak current} & \[
\begin{aligned}
& \overline{\text { RES, NMI, MD3- } ~|~ l i n ~| ~} \\
& \text { MD0, PA2, PA5, } \\
& \text { PA6-PA9, PE0- } \\
& \text { PE15,FWP }
\end{aligned}
\] & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{Vin}=0.5\) to \(\mathrm{V}_{\mathrm{cc}}-0.5 \mathrm{~V}\) \\
\hline & A/D port & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{Vin}=0.5\) to \(\mathrm{AV}_{\mathrm{cc}}-0.5 \mathrm{~V}\) \\
\hline & Other input pins & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{Vin}=0.5\) to \(\mathrm{V}_{\mathrm{cc}}-0.5 \mathrm{~V}\) \\
\hline
\end{tabular}
(except EXTAL
pin)
Three-state A21-A0, D31- \(\left|I_{\text {TSI }}\right| \quad-\quad-1.0 \quad \mu \mathrm{~A} \quad \mathrm{Vin}=0.5\) to \(\mathrm{V}_{\mathrm{CC}}-0.5 \mathrm{~V}\) leak current D0, \(\overline{\mathrm{CS} 3}-\overline{\mathrm{CSO}}\), (while off) RDWR, \(\overline{R A S}\),
\(\overline{\text { CASxx }}, \overline{\text { WRxx }}\),
\(\overline{\mathrm{RD}}\), ports \(\mathrm{A}, \mathrm{B}, \mathrm{C}\),
D, E

Table 25.2 DC Characteristics (Conditions: \(V_{C C}=\mathbf{5 . 0} V \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(A V_{C C}=V_{C C} \pm \mathbf{1 0 \%}, A V_{\text {ref }}=4.5 \mathrm{~V}\) to \(A V_{C C}, V_{S S}=A V_{S S}=\mathbf{0} V, T a=-20\) to \(+\mathbf{7 5}{ }^{\circ} \mathrm{C}\) ) (cont)


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Table 25.3 Permitted Output Current Values (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(\mathrm{AV} \mathrm{V}_{\mathrm{CC}}=\) \(\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV} \mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, A \mathrm{~V}_{\text {ref }}=4.5 \mathrm{~V}\) to \(\mathrm{AV} \mathrm{CC}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=\mathbf{0} \mathrm{V}\), \(\mathrm{Ta}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Typ & Max & Unit \\
\hline Output low-level permissible current (per pin) & \(\mathrm{I}_{\mathrm{OL}}\) & - & - & \(2.0^{*}\) & mA \\
\hline Output low-level permissible current (total) & \(\sum \mathrm{I}_{\mathrm{OL}}\) & - & - & 80 & mA \\
\hline Output high-level permissible current (per pin) & \(-\mathrm{I}_{\mathrm{OH}}\) & - & - & 2.0 & mA \\
\hline Output high-level permissible current (total) & \(\sum\left(-\mathrm{I}_{\mathrm{OH}}\right)\) & - & - & 25 & mA \\
\hline
\end{tabular}

Notes: To assure LSI reliability, do not exceed the output values listed in this table.
* PE9, PE11-PE15 are \(\mathrm{I}_{\mathrm{OL}}=15 \mathrm{~mA}\) (max). Make sure that no more than three of these pins exceed an \(\mathrm{I}_{\mathrm{oL}}\) of 2.0 mA simultaneously.

\subsection*{25.3 AC Characteristics}

\subsection*{25.3.1 Clock Timing}

Table 25.4 Clock Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\) \(\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=\mathbf{4 . 5} \mathrm{V}\) to \(\mathrm{AV}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathbf{T a}=\mathbf{- 2 0}\) to \(\left.+\mathbf{7 5}{ }^{\circ} \mathrm{C}\right)\)
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figures \\
\hline Operating frequency & \(\mathrm{f}_{\mathrm{OP}}\) & 4 & 28.7 & MHz & 25.1 \\
\hline Clock cycle time & \(\mathrm{t}_{\mathrm{cyc}}\) & 34.8 & 250 & ns \\
\hline Clock low-level pulse width & \(\mathrm{t}_{\mathrm{CL}}\) & 10 & - & ns \\
\hline Clock high-level pulse width & \(\mathrm{t}_{\mathrm{CH}}\) & 10 & - & ns \\
\hline Clock rise time & \(\mathrm{t}_{\mathrm{CR}}\) & - & 5 & ns \\
\hline Clock fall time & \(\mathrm{t}_{\mathrm{CF}}\) & - & 5 & ns \\
\hline EXTAL clock input frequency & \(\mathrm{f}_{\mathrm{EX}}\) & 4 & 10 & MHz \\
\hline EXTAL clock input cycle time & \(\mathrm{t}_{\mathrm{EXCyc}}\) & 100 & 250 & ns \\
\hline EXTAL clock low-level input pulse width & \(\mathrm{t}_{\mathrm{EXL}}\) & 40 & - & ns \\
\hline EXTAL clock high-level input pulse width & \(\mathrm{t}_{\mathrm{EXH}}\) & 40 & - & ns \\
\hline EXTAL clock input rise time & \(\mathrm{t}_{\mathrm{EXR}}\) & - & 5 & ns \\
\hline EXTAL clock input fall time & \(\mathrm{t}_{\mathrm{EXF}}\) & - & 5 & ns \\
\hline Reset oscillation settling time & \(\mathrm{t}_{\mathrm{OSC1}}\) & 10 & - & ms \\
\hline Standby return clock settling time & \(\mathrm{t}_{\mathrm{OSc2}}\) & 10 & - & ms \\
\hline
\end{tabular}


Figure 25.1 System Clock Timing


Figure 25.2 EXTAL Clock Input Timing


Figure 25.3 Oscillation Settling Time

\subsection*{25.3.2 Control Signal Timing}

Table 25.5 Control Signal Timing (Conditions: \(V_{C C}=5.0 \mathrm{~V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(A V_{C C}=V_{C C} \pm \mathbf{1 0 \%}, A V_{\text {ref }}=4.5 \mathrm{~V}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{A} V_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{Ta}=\mathbf{- 2 0}\) to \(+\mathbf{7 5}{ }^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline \(\overline{\text { RES }}\) rise/fall & \(\mathrm{t}_{\text {RES }}, \mathrm{t}_{\text {RES }}\) & - & 200 & ns & 25.4 \\
\hline \(\overline{\mathrm{RES}}\) pulse width & \(\mathrm{t}_{\text {RESW }}\) & 20 & - & \(\mathrm{t}_{\text {cy }}\) & \\
\hline MRES pulse width & \(\mathrm{t}_{\text {MRESW }}\) & 20 & - & \(\mathrm{t}_{\text {cyc }}\) & \\
\hline NMI rise/fall & \(\mathrm{t}_{\text {NMII, }}, \mathrm{t}_{\text {MMIF }}\) & - & 200 & ns & 25.5 \\
\hline \(\overline{\mathrm{RES}}\) setup time* & \(\mathrm{t}_{\text {RESS }}\) & 35 & - & ns & 25.4, \\
\hline MRES setup time* & \(t_{\text {MRESS }}\) & 35 & - & ns & 25.5 \\
\hline NMI setup time* & \(\mathrm{t}_{\text {NMIS }}\) & 35 & - & ns & \\
\hline \(\overline{\text { IRQ7-IRQ0 }}\) setup time (edge detection) & \(\mathrm{t}_{\text {IROES }}\) & 35 & - & ns & \\
\hline IRQ7-\/TQ0 setup time (level detection) & \(\mathrm{t}_{\text {IRQLS }}\) & 35 & - & ns & \\
\hline NMI hold time & \(\mathrm{t}_{\text {мMH }}\) & 35 & - & ns & 25.5 \\
\hline \(\overline{\text { IRQ7-IRQ0 }}\) hold time & \(\mathrm{t}_{\text {IROEH }}\) & 35 & - & ns & \\
\hline IRQOUT output delay time & \(\mathrm{t}_{\text {IROOD }}\) & - & 35 & ns & 25.6 \\
\hline Bus request setup time & \(\mathrm{t}_{\text {Bras }}\) & 35 & - & ns & 25.7 \\
\hline Bus acknowledge delay time 1 & \(\mathrm{t}_{\text {BACKD1 }}\) & - & 35 & ns & \\
\hline Bus acknowledge delay time 2 & \(\mathrm{t}_{\text {BACKD2 }}\) & - & 35 & ns & \\
\hline Bus three-state delay time & \(\mathrm{t}_{\text {BzD }}\) & - & 35 & ns & \\
\hline
\end{tabular}

Note: * The \(\overline{\text { RES }}, \overline{\text { MRES }}\), NMI, \(\overline{\mathrm{BREQ}}\), and \(\overline{\mathrm{IRQ7}}-\overline{\mathrm{IRQ0}}\) signals are asynchronous inputs, but when thesetup times shown here are provided, the signals are considered to have produced changes at clock rise (for \(\overline{\mathrm{RES}}, \overline{\mathrm{MRES}}, \overline{\mathrm{BREQ}}\) ) or clock fall (for NMI and \(\overline{\mathrm{RQ} 7}\) IRQO). If the setup times are not provided, recognition is delayed until the next clock rise or fall.


Figure 25.4 Reset Input Timing


Figure 25.5 Interrupt Signal Input Timing


Figure 25.6 Interrupt Signal Output Timing


Figure 25.7 Bus Right Release Timing

\subsection*{25.3.3 Bus Timing}

Table 25.6 Bus Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm\) \(10 \%, \mathrm{AV}_{\text {ref }}=4.5 \mathrm{~V}-\mathrm{AV} \mathrm{CC}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV}\) SS \(=0 \mathrm{~V}, \mathrm{Ta}=-20\) to \(+\mathbf{7 5}{ }^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline Address delay time & \(\mathrm{t}_{\text {AD }}\) & \(2^{* 3}\) & 18 & ns & \[
\begin{aligned}
& \text { 25.8, 25.9, } \\
& 25.11-25.16, \\
& 25.19
\end{aligned}
\] \\
\hline \(\overline{\mathrm{CS}}\) delay time 1 & \(\mathrm{t}_{\text {cSD } 1}\) & \(2^{* 3}\) & 21 & ns & \multirow[t]{2}{*}{25.8, 25.9, 25.19} \\
\hline \(\overline{\mathrm{CS}}\) delay time 2 & \(\mathrm{t}_{\mathrm{CSD} 2}\) & \(2^{* 3}\) & 21 & ns & \\
\hline Read strobe delay time 1 & \(\mathrm{t}_{\text {RSD1 }}\) & \(2^{* 3}\) & 18 & ns & \multirow[t]{8}{*}{\[
\begin{aligned}
& \text { 25.8, 25.9, } \\
& 25.11-25.16, \\
& -25.19
\end{aligned}
\]} \\
\hline Read strobe delay time 2 & \(\mathrm{t}_{\text {RSD2 }}\) & \(2^{* 3}\) & 18 & ns & \\
\hline Read data setup time & \(\mathrm{t}_{\text {RDS }}{ }^{* 4}\) & 15 & - & ns & \\
\hline Read data hold time & \(\mathrm{t}_{\text {RDH }}\) & 0 & - & ns & \\
\hline Write strobe delay time 1 & \(\mathrm{t}_{\text {wSD } 1}\) & \(2^{* 3}\) & 18 & ns & \\
\hline Write strobe delay time 2 & \(\mathrm{t}_{\text {WSD2 }}\) & \(2^{* 3}\) & 18 & ns & \\
\hline Write data delay time & \(t_{\text {wDD }}\) & - & 35 & ns & \\
\hline Write data hold time & \(\mathrm{t}_{\text {WOH }}\) & 0 & 10*2 & ns & \\
\hline WAIT setup time & \(\mathrm{t}_{\text {wTS }}\) & 15 & - & ns & \multirow[t]{2}{*}{\[
\begin{aligned}
& 25.10,25.15, \\
& 25.19
\end{aligned}
\]} \\
\hline \(\overline{\text { WAIT }}\) hold time & \(t_{\text {WTH }}\) & 0 & - & ns & \\
\hline \(\overline{\text { RAS }}\) delay time 1 & \(\mathrm{t}_{\text {RASD } 1}\) & \(2^{* 3}\) & 18 & ns & \multirow[t]{4}{*}{25.11-25.18} \\
\hline \(\overline{\mathrm{RAS}}\) delay time 2 & \(\mathrm{t}_{\text {RASD2 }}\) & \(2^{* 3}\) & 18 & ns & \\
\hline \(\overline{\text { CAS }}\) delay time 1 & \(\mathrm{t}_{\text {CASD } 1}\) & \(2^{* 3}\) & 18 & ns & \\
\hline \(\overline{\mathrm{CAS}}\) delay time 2 & \(\mathrm{t}_{\text {CASD } 2}\) & \(2^{* 3}\) & 18 & ns & \\
\hline Read data access time & \(\mathrm{t}_{\mathrm{ACC}}{ }^{* 1}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+2)-40\) & - & ns & \multirow[t]{2}{*}{25.8, 25.9} \\
\hline Access time from read strobe & \(\mathrm{t}_{\mathrm{OE}}{ }^{*}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+1.5)-40\) & - & ns & \\
\hline Access time from column address & \(\mathrm{t}_{\mathrm{AA}}{ }^{* 1}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+2)-40\) & - & ns & \multirow[t]{7}{*}{25.11-25.16} \\
\hline Access time from \(\overline{\mathrm{RAS}}\) & \(\mathrm{t}_{\text {RAC }}{ }^{* 1}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+\mathrm{RCD}+2.5)-40\) & & ns & \\
\hline Access time from CAS & \(\mathrm{t}_{\mathrm{CAC}}{ }^{* 1}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+1)-40\) & - & ns & \\
\hline Row address hold time & \(\mathrm{t}_{\text {RAH }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{RCD}+0.5)-15\) & - & ns & \\
\hline Row address setup time & \(\mathrm{t}_{\text {ASR }}{ }^{* 5}\) & \(\mathrm{t}_{\text {cyc }} \times 0.5-17.5\) & - & ns & \\
\hline Data input setup time & \(\mathrm{t}_{\mathrm{DS}}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{m}+0.5)-25\) & - & ns & \\
\hline Data input hold time & \(\mathrm{t}_{\mathrm{DH}}\) & 20 & - & ns & \\
\hline
\end{tabular}

Table 25.7 Bus Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm 10 \%, A \mathrm{~V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm\) \(10 \%, \mathrm{AV}_{\text {ref }}=4.5 \mathrm{~V}\) to \(\mathrm{AV} \mathrm{VC}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{Ta}=-20\) to \(\left.+75^{\circ} \mathrm{C}\right)\)
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline Write address setup time & \(\mathrm{t}_{\text {As }}\) & 0 & - & ns & 25.8-25.9 \\
\hline Write address hold time & \(t_{\text {wr }}\) & 5 & - & ns & 25.8, 25.9, 25.19 \\
\hline Write data hold time & \(\mathrm{t}_{\text {wRH }}\) & 0 & - & ns & \\
\hline Read/write strobe delay time 1 & \(\mathrm{t}_{\text {RWD1 }}\) & \(2^{* 3}\) & 18 & ns & 25.11-25.16 \\
\hline Read/write strobe delay time 2 & \(\mathrm{t}_{\text {RWD2 }}\) & \(2^{* 3}\) & 18 & ns & \\
\hline High-speed page mode CAS precharge time & \(\mathrm{t}_{\text {cP }}\) & \(\mathrm{t}_{\text {cyc }}-25\) & - & ns & 25.16 \\
\hline RAS precharge time & \(\mathrm{t}_{\text {RP }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{TPC}+1.5)-15\) & - & ns & 25.11-25.16 \\
\hline CAS setup time & \(\mathrm{t}_{\text {CSR }}\) & 10 & - & ns & 25.17, 25.18 \\
\hline AH delay time 1 & \(\mathrm{t}_{\text {AHD } 1}\) & \(2^{* 3}\) & 18 & ns & 25.19 \\
\hline AH delay time 2 & \(\mathrm{t}_{\text {AHD2 }}\) & \(2^{* 3}\) & 18 & ns & \\
\hline Multiplex address delay time & \(t_{\text {MAD }}\) & \(2^{* 3}\) & 18 & ns & \\
\hline Multiplex address hold time & \(\mathrm{t}_{\text {MAH }}\) & 0 & - & ns & \\
\hline DACK delay time & \(\mathrm{t}_{\text {DACKD1 }}\) & \(2^{* 3}\) & 21 & ns & \[
\begin{aligned}
& 25.8,25.9,25.11- \\
& 25.16,25.19
\end{aligned}
\] \\
\hline
\end{tabular}

Notes: n is the number of waits. m is 0 when the number of DRAM write cycle waits is 0 , and 1 otherwise. RCD is the set value of the RCD bit in DCR. TPC is the set value of the TPC bit in DCR.
*1 If the access time is satisfied, \(t_{\text {RDS }}\) need not be satisfied.
*2 \(t_{\text {wDH }}(\max )\) is a reference value.
*3 The delay time Min values are reference values (typ).
* \(4 \mathrm{t}_{\mathrm{RDS}}\) is a reference value.
*5 When 28.7 MHz , tASR=0ns (min)


Figure 25.8 Basic Cycle (No Waits)


Figure 25.9 Basic Cycle (Software Waits)


Figure 25.10 Basic Cycle (2 Software Waits + Wait due to \(\overline{\text { WAIT }}\) Signal)


Figure 25.11 DRAM Cycle (Normal Mode, No Waits, TPC = 0, RCD = 0)


Figure 25.12 DRAM Cycle (Normal Mode, 1 Wait, TPC=0, RCD=0)


Figure 25.13 DRAM Cycle (Normal Mode, 2 Waits, TPC=1, RCD=1)


Figure 25.14 DRAM Cycle (Normal Mode, 3 Waits, TPC = 1, RCD = \(\mathbf{1}\) )

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Figure 25.15 DRAM Cycle (Normal Mode, 2 Waits + Wait due to \(\overline{\text { WAIT }}\) Signal)


Figure 25.16 DRAM Cycle (High-Speed Page Mode)


Figure 25.17 CAS Before RAS Refresh (TRAS1 \(=0\), TRAS0 \(=0\) )


Figure 25.18 Self Refresh


Figure 25.19 Address Data Multiplex I/O Space Cycle (1 Software Wait + External Wait)

\subsection*{25.3.4 Direct Memory Access Controller Timing}

Table 25.8 Direct Memory Access Controller Timing (Conditions: \(V_{C C}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\),
\(A V_{C C}=5.0 \mathrm{~V} \pm 10 \%, A V_{C C}=V_{C C} \pm 10 \%, A V_{\text {ref }}=4.5 \mathrm{~V}\) to \(A V_{C C}, V_{S S}=A V_{S S}=0\) \(\mathrm{V}, \mathrm{Ta}=-20\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline\(\overline{\text { DREQ0 }}\) and \(\overline{\text { DREQ1 }}\) setup time & \(\mathrm{t}_{\text {DRQs }}\) & 18 & - & ns & 25.20 \\
\hline\(\overline{\text { DREQ0 } \text { and } \overline{\text { DREQ1 }} \text { hold time }}\) & \(\mathrm{t}_{\text {DRQH }}\) & 18 & - & ns & \\
\hline\(\overline{\text { DREQ0 } \text { and } \overline{\text { DREQ1 }} \text { pulse width }}\) & \(\mathrm{t}_{\text {DRQw }}\) & 1.5 & - & \(\mathrm{t}_{\text {cyc }}\) & 25.21 \\
\hline\(\overline{\text { DRAK output delay time }}\) & \(\mathrm{t}_{\text {DRAKD }}\) & - & 18 & ns & 25.22 \\
\hline
\end{tabular}


Figure 25.20 \(\overline{\text { DREQ0 }}\) and \(\overline{\text { DREQ1 }}\) Input Timing (1)


Figure 25.21 \(\overline{\text { DREQ0 }}\) and \(\overline{\text { DREQ1 }}\) Input Timing (2)


Figure 25.22 DRAK Output Delay Time

\subsection*{25.3.5 Multifunction Timer Pulse Unit Timing}

Table 25.9 Multifunction Timer Pulse Unit Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(\mathrm{AV}_{\mathrm{CC}}\) \(=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm 10 \%, \mathrm{AV}_{\text {ref }}=4.5 \mathrm{~V}\) to \(\mathrm{AV}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}\), \(\mathbf{T a}\) \(=-20\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline Output compare output delay time & \(\mathrm{t}_{\text {TOCD }}\) & - & 100 & ns & 25.23 \\
\hline Input capture input setup time & \(\mathrm{t}_{\text {TICS }}\) & 30 & - & ns & \\
\hline Timer input setup time & \(\mathrm{t}_{\text {TCKS }}\) & 35 & - & ns & \\
\hline \begin{tabular}{l} 
Timer clock pulse width (single \\
edge specification)
\end{tabular} & \(\mathrm{t}_{\text {TCKWH/L }}\) & 1.5 & - & \(\mathrm{t}_{\text {cyc }}\) & 25.24 \\
\begin{tabular}{lllll} 
Timer clock pulse width (both \\
\begin{tabular}{l} 
edges specified)
\end{tabular} & \(\mathrm{t}_{\text {TCKWH/L }}\) & 2.5 & - & \(\mathrm{t}_{\text {cyc }}\) \\
\begin{tabular}{l} 
Timer clock pulse width (phase \\
measurement mode)
\end{tabular} & \(\mathrm{t}_{\text {TCKWH/L }}\) & 2.5 & - & \(\mathrm{t}_{\text {cyc }}\) \\
\hline
\end{tabular} & \\
\hline
\end{tabular}


Figure 25.23 MTU I/O Timing


Figure 25.24 MTU Clock Input Timing

\subsection*{25.3.6 I/O Port Timing}

Table 25.10 I/O Port Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\) \(\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=4.5 \mathrm{~V}\) to \(\mathrm{AV}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{Ta}=-\mathbf{2 0}\) to \(\left.+75^{\circ} \mathrm{C}\right)\)
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline Port output data delay time & \(\mathrm{t}_{\text {PWD }}\) & - & 100 & ns & 25.25 \\
\hline Port input hold time & \(\mathrm{t}_{\text {PRH }}\) & 35 & - & ns & \\
\hline Port input setup time & \(\mathrm{t}_{\text {PRS }}\) & 35 & - & ns \\
\hline
\end{tabular}


Figure 25.25 I/O Port I/O Timing

\subsection*{25.3.7 Watchdog Timer Timing}

Table 25.11 Watchdog Timer Timing (Conditions: \(V_{C C}=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(A V_{C C}=V_{C C} \pm 10 \%, A V_{\text {ref }}=4.5 \mathrm{~V}\) to \(A V_{C C}, V_{S S}=A V_{S S}=0 \mathrm{~V}, \mathrm{Ta}=-20\) to \(\left.+75^{\circ} \mathrm{C}\right)\)
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline\(\overline{\text { WDTOVF }}\) delay time & \(\mathrm{t}_{\text {wovd }}\) & - & 100 & ns & 25.26 \\
\hline
\end{tabular}


Figure 25.26 Watchdog Timer Timing

\subsection*{25.3.8 Serial Communication Interface Timing}

Table 25.12 Serial Communication Interface Timing (Conditions: \(V_{C C}=\mathbf{5 . 0} \mathbf{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}\) \(=5.0 \mathrm{~V} \pm 10 \%, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm 10 \%, \mathrm{AV}_{\text {ref }}=4.5 \mathrm{~V}\) to \(\mathrm{AV} \mathrm{V}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}\), \(\mathbf{T a}\) \(=-20\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline Input clock cycle & \(\mathrm{t}_{\mathrm{scyc}}\) & 4 & - & \(\mathrm{t}_{\mathrm{cyc}}\) & 25.27 \\
\hline Input clock cycle (clock sync) & \(\mathrm{t}_{\mathrm{scyc}}\) & 6 & - & \(\mathrm{t}_{\mathrm{cyc}}\) & \\
\hline Input clock pulse width & \(\mathrm{t}_{\mathrm{sckw}}\) & 0.4 & 0.6 & \(\mathrm{t}_{\mathrm{scyc}}\) & \\
\hline Input clock rise time & \(\mathrm{t}_{\mathrm{sckr}}\) & - & 1.5 & \(\mathrm{t}_{\mathrm{cyc}}\) & \\
\hline Input clock fall time & \(\mathrm{t}_{\mathrm{sck}}\) & - & 1.5 & \(\mathrm{t}_{\mathrm{cyc}}\) & \\
\hline Transmit data delay time (clock sync) & \(\mathrm{t}_{\mathrm{TXD}}\) & - & 100 & ns & 2025.28 \\
\hline Receive data setup time (clock sync) & \(\mathrm{t}_{\mathrm{Rxs}}\) & 100 & - & ns & \\
\hline Receive data hold time (clock sync) & \(\mathrm{t}_{\mathrm{RXH}}\) & 100 & - & ns & \\
\hline
\end{tabular}


Figure 25.27 Input Clock Timing


Figure 25.28 SCI I/O Timing (Clock Sync Mode)

\subsection*{25.3.9 High-speed A/D Converter Timing (excluding A mask)}

Table 25.13 High-speed A/D Converter Timing (Conditions: \(V_{C C}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathbf{5 . 0}\) \(\mathrm{V} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{A} \mathrm{V}_{\text {ref }}=\mathbf{4 . 5} \mathrm{V}\) to \(A \mathrm{~V}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{Ta}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{Item} & Symbol & Min & Typ & Max & Unit & Figure \\
\hline \multicolumn{2}{|l|}{External trigger input pulse width} & \(\mathrm{t}_{\text {TRGW }}\) & 2 & - & - & \(\mathrm{t}_{\text {cyc }}\) & 25.29 \\
\hline \multicolumn{2}{|l|}{External trigger input start delay time} & \(\mathrm{t}_{\text {TRGS }}\) & 50 & - & - & ns & \\
\hline \multirow[t]{2}{*}{A/D conversion start delay time} & CKS = 0 & \(\mathrm{t}_{\mathrm{D}}\) & 1.5 & 1.5 & 1.5 & \multirow[t]{6}{*}{\(\mathrm{t}_{\text {cyc }}\)} & \multirow[t]{6}{*}{25.30} \\
\hline & CKS \(=1\) & & 1.5 & 1.5 & 1.5 & & \\
\hline \multirow[t]{2}{*}{Input sampling time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {SPL }}\)} & 20 & 20 & 20 & & \\
\hline & CKS \(=1\) & & 40 & 40 & 40 & & \\
\hline \multirow[t]{2}{*}{A/D conversion time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {conv }}\)} & 42.5 & 42.5 & 42.5 & & \\
\hline & CKS = 1 & & 82.5 & 82.5 & 82.5 & & \\
\hline
\end{tabular}


Figure 25.29 External Trigger Input Timing


Figure 25.30 Analog Conversion Timing

\subsection*{25.3.10 Mid-speed Converter Timing (A mask)}

Table 25.14 shows Mid-speed converter timing
Table 25.14 Mid-speed Converter Timing (Conditions:Vcc=5.0V \(\pm 10 \%\), AVcc=5.0V \(\pm 10 \%\), AVcc \(=\) Vcc \(\pm \mathbf{1 0 \%}\), AVref=4.5V to Avcc, Vss=AVss=0V, Ta=-20 to \(\pm \mathbf{7 5}{ }^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{Item} & Symbo & Min & Typ & Max & Unit & Figure \\
\hline \multicolumn{2}{|l|}{External trigger input pulse width} & \(\mathrm{t}_{\text {TRGW }}\) & 2 & - & - & \(\mathrm{t}_{\text {cy }}\) & 25.31 \\
\hline \multicolumn{2}{|l|}{External trigger input start delay time} & \(\mathrm{t}_{\text {TRGS }}\) & 50 & - & - & ns & \\
\hline \multirow[t]{2}{*}{A/D conversion start delay time} & CKS \(=0\) & \(\mathrm{t}_{\text {D }}\) & 10 & - & 17 & \multirow[t]{6}{*}{\(\mathrm{t}_{\text {cyc }}\)} & \multirow[t]{6}{*}{25.32} \\
\hline & CKS \(=1\) & & 6 & - & 9 & & \\
\hline \multirow[t]{2}{*}{Input sampling time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {SPL }}\)} & - & 64 & - & & \\
\hline & CKS \(=1\) & & - & 32 & - & & \\
\hline \multirow[t]{2}{*}{A/D conversion time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {conv }}\)} & 259 & - & 266 & & \\
\hline & CKS \(=1\) & & 131 & - & 134 & & \\
\hline
\end{tabular}


Figure 25.31 External Trigger Input Timing


Figure 25.32 Analog Conversion Timing

\subsection*{25.3.11 Measuring Conditions for AC Characteristics}
- Input reference levels:
— High level: 2.2 V
— Low level: 0.8 V
- Output reference levels:
— High level: 2.0 V
— Low level: 0.8 V


Note: \(C_{L}\) is set with the following pins, including the total capacitance of the measurement equipment etc:
\(30 \mathrm{pF}: \quad \mathrm{CK}, \overline{\mathrm{RAS}}, \overline{\mathrm{CASxx}}, \mathrm{RDWR}, \overline{\mathrm{CSO}}-\overline{\mathrm{CS} 3}, \overline{\mathrm{AH}}, \overline{\mathrm{BREQ}}, \overline{\mathrm{BACK}}, \mathrm{DACK0}\), DACK1, and IRQOUT
\(50 \mathrm{pF}: ~ A 21-A 0, \mathrm{D} 31-\mathrm{D} 0, \overline{\mathrm{RD}}, \overline{\mathrm{WRxx}}\)
70 pF : Port output and peripheral module output pins other than the above.
\(\mathrm{I}_{\mathrm{OL}}\), \(\mathrm{I}_{\mathrm{OH}}\) : See table 25.3, Permitted Output Current Values.

Figure 25.33 Output Load Circuit

\subsection*{25.4 A/D Converter Characteristics}

Table 25.15 A/D Converter Timing (excluding A mask) (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{5 . 0} \mathrm{V} \pm \mathbf{1 0 \%}\), \(A V_{C C}=5.0 \mathrm{~V} \pm 10 \%, A V_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm 10 \%, A V_{\text {ref }}=4.5 \mathrm{~V}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=A \mathrm{~V}_{\mathrm{SS}}=0\) \(\mathrm{V}, \mathrm{Ta}=-20\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{lllll} 
& \multicolumn{3}{c}{\(\mathbf{2 8 . 7 ~ M H z}\)} & \\
\cline { 2 - 4 } Item & Min & Typ & \(\mathbf{M a x}\) & Unit \\
\hline Resolution & 10 & 10 & 10 & Bits \\
\hline Conversion time (when CKS =1) & - & - & 2.9 & \(\mu \mathrm{~s}\) \\
\hline Analog input capacitance & - & - & 20 & pF \\
\hline \begin{tabular}{l} 
Permitted signal source \\
impedance
\end{tabular} & - & - & 1 & \(\mathrm{k} \Omega\) \\
\hline Non-linear error* & - & - & \(\pm 8\) & LSB \\
\hline Offset error* & - & - & \(\pm 8\) & LSB \\
\hline Full-scale error* & - & - & \(\pm 8\) & LSB \\
\hline Quantization error* & - & - & \(\pm 0.5\) & LSB \\
\hline Absolute error (when CKS \(=1\) ) & - & - & \(\pm 15\) & LSB \\
\hline
\end{tabular}

Note: * Reference values

Table 25.16 A/D Converter Timing (A mask) (Condition: Vcc=5.0 \(\pm \mathbf{1 0 \%}\), AVcc=5.0 \(\pm \mathbf{1 0 \%}\), AVcc \(=\) Vcc \(\pm \mathbf{1 0 \%}\), AVref \(=4.5 \mathrm{~V}\) to \(\mathrm{AVcc}, \mathrm{Vss}=\mathrm{AVss}=0 \mathrm{~V}, \mathrm{Ta}=-20\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Item} & \multicolumn{3}{|c|}{28.7 MHz} & \multicolumn{3}{|c|}{20 MHz} & \multirow[b]{2}{*}{Unit} \\
\hline & Min & Typ & Max & Min & Typ & Max & \\
\hline Resolution & 10 & 10 & 10 & 10 & 10 & 10 & Bits \\
\hline Conversion time (when CKS=0) & - & - & 9.3 & - & - & 13.4 & \(\mu \mathrm{s}\) \\
\hline Analog input capacity & - & - & 20 & - & - & 20 & pF \\
\hline Permission signal source impedance & - & - & 1 & - & - & 1 & k \(\Omega\) \\
\hline Non-linearity error* & - & - & \(\pm 3\) & - & - & \(\pm 3\) & LSB \\
\hline Offset error* & - & - & \(\pm 3\) & - & - & \(\pm 3\) & LSB \\
\hline Full scale error* & - & - & \(\pm 3\) & - & - & \(\pm 3\) & LSB \\
\hline Quantize error* & - & - & \(\pm 0.5\) & - & - & \(\pm 0.5\) & LSB \\
\hline Absolute error & - & - & \(\pm 4\) & - & - & \(\pm 4\) & LSB \\
\hline
\end{tabular}

Note: * Reference value

\section*{Section 26 Electrical Characteristics (3.3V, 16.7 MHz Version)}

\subsection*{26.1 Absolute Maximum Ratings}

Table 26.1 Absolute Maximum Ratings
\begin{tabular}{llll} 
Item & Symbol & Rating & Unit \\
\hline Power supply voltage & \(\mathrm{V}_{\mathrm{CC}}\) & -0.3 to +7.0 & V \\
\hline Programmable voltage (ZTAT version only) & \(\mathrm{V}_{\mathrm{PP}}\) & -0.3 to +13.5 & V \\
\hline Input voltage (other than A/D ports) & \(\mathrm{V}_{\text {in }}\) & -0.3 to \(\mathrm{V}_{\mathrm{cc}}+0.3\) & V \\
\hline Input voltage (A/D ports) & \(\mathrm{V}_{\text {in }}\) & -0.3 to \(\mathrm{AV}_{\mathrm{cc}}+0.3\) & V \\
\hline Analog supply voltage & \(\mathrm{AV}_{\mathrm{cc}}\) & -0.3 to +7.0 & V \\
\hline Analog reference voltage (QFP-144 only) & \(\mathrm{AV}_{\text {ref }}\) & -0.3 to \(\mathrm{AV}_{\mathrm{cc}}+0.3\) & V \\
\hline Analog input voltage & \(\mathrm{V}_{\mathrm{AN}}\) & -0.3 to \(\mathrm{AV}_{\mathrm{cc}}+0.3\) & V \\
\hline Operating temperature & \(\mathrm{T}_{\text {opr }}\) & -20 to +75 & \({ }^{\circ} \mathrm{C}\) \\
\hline Programming temperature (ZTAT version only) & \(\mathrm{T}_{\text {we }}\) & -20 to +75 & \({ }^{\circ} \mathrm{C}\) \\
\hline Storage temperature & \(\mathrm{T}_{\text {stg }}\) & -55 to +125 & \({ }^{\circ} \mathrm{C}\)
\end{tabular}

Note: Operating the LSI in excess of the absolute maximum ratings may result in permanent damage.

\subsection*{26.2 DC Characteristics}

Table 26.2 DC Characteristics (Conditions: \(V_{C C}=3.0^{* 1}\) to 3.6V, \(A V_{C C}=3.0^{* 1}\) to 3.6 V , \(A V_{C C}=V_{C C} \pm \mathbf{1 0 \%}, A V_{\text {ref }}=3.0^{* 1}\) to \(A V_{C C}, V_{S S}=A V_{S S}=0 V, T_{a}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline Item & Pin & Symbol & Min & Typ & Max & & Measurement Conditions \\
\hline \multirow[t]{4}{*}{Input highlevel voltage} & \(\overline{R E S}, \mathrm{NMI}\), MD3-0, PA2, PA5, PA6-PA9, PA0-PE15, FWP & \multirow[t]{4}{*}{\(\mathrm{V}_{\text {IH }}\)} & \multicolumn{2}{|l|}{\(\mathrm{V}_{\text {cc }} \times 0.9\) -} & \(\mathrm{V}_{\text {cc }}+0.3\) & V & \\
\hline & EXTAL & & \multicolumn{2}{|l|}{\(\mathrm{V}_{\text {cc }} \times 0.9-\)} & \(\mathrm{V}_{\mathrm{CC}}+0.3\) & V & \\
\hline & A/D port & & \multicolumn{2}{|l|}{\(\mathrm{V}_{\text {cc }} \times 0.7-\)} & \(\mathrm{AV}_{\mathrm{cc}}+0.3\) & V & \\
\hline & Other input pins & & \multicolumn{2}{|l|}{\(\mathrm{V}_{\mathrm{CC}} \times 0.7\) -} & \(\mathrm{V}_{\text {cc }}+0.3\) & V & \\
\hline \multirow[t]{2}{*}{Input lowlevel voltage} & \(\overline{R E S}, \mathrm{NMI}\), MD3-0, PA2, PA5, PA6-PA9, PA0-PE15, FWP & \multirow[t]{2}{*}{VIL} & -0.3 & - & \(\mathrm{V}_{C C} \times 0.1\) & V & \\
\hline & Other input pins & & -0.3 & - & \(\mathrm{V}_{\text {cc }} \times 0.2\) & V & \\
\hline Schmitt trigger inpu voltage & \[
\begin{aligned}
& \text { PA2, PA5, PA6- } \\
& \text { tPA9, } \\
& \text { PE0-PE15 }
\end{aligned}
\] & \[
V_{T}^{+}-V_{T}^{-}
\] & \[
\begin{aligned}
& V_{c c} \times \\
& 0.07
\end{aligned}
\] & - & - & & \[
\frac{\mathrm{VT}^{+} \geq \mathrm{V}_{\mathrm{CC}} \times 0.9 \mathrm{~V} \text { (min) }}{\mathrm{VT}^{-} \leq \mathrm{V}_{\mathrm{Cc}} \times 0.2 \mathrm{~V}(\max )}
\] \\
\hline \multirow[t]{3}{*}{Input leak current} & \[
\begin{aligned}
& \overline{\mathrm{RES}}, \mathrm{NMI}, \mathrm{MD} 3- \\
& \text { 0, PA2, PA5, } \\
& \text { PA6-PA9, PE0- } \\
& \text { PE15,FWP }
\end{aligned}
\] & \multirow[t]{3}{*}{\[
\left|I_{\text {in }}\right|
\]} & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{V}_{\text {in }}=0.5\) to \(\mathrm{V}_{\mathrm{cc}}-0.5 \mathrm{~V}\) \\
\hline & A/D port & & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{V}_{\text {in }}=0.5\) to \(\mathrm{AV}_{\text {cC }}-0.5 \mathrm{~V}\) \\
\hline & Other input pins (except EXTAL pin) & & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{V}_{\text {in }}=0.5\) to \(\mathrm{V}_{\text {cc }}-0.5 \mathrm{~V}\) \\
\hline Three-state leak current (while off) & A21-A0, D31D0, CS3-CSO, RDWR, \(\overline{R A S}\), \(\overline{\text { CASxx }}\), WRxx, RD, Ports A, B, C, D, E & \(\left|\mathrm{I}_{\text {TS }}\right|\) & - & - & 1.0 & \(\mu \mathrm{A}\) & \(\mathrm{V}_{\text {in }}=0.5\) to \(\mathrm{V}_{\mathrm{CC}}-0.5 \mathrm{~V}\) \\
\hline
\end{tabular}

Table 26.2 DC Characteristics (Conditions: \(V_{C C}=3.0^{* 1}\) to 3.6V, \(\mathrm{AV}_{\mathrm{CC}}=3 . \mathbf{0}^{* 1}\) to \(\mathbf{3 . 6} \mathrm{V}, \mathrm{AV}\) CC

\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline Item & Pin & Symbo & Min & Typ & Max & Unit & Measurement Conditions \\
\hline \multirow[t]{2}{*}{Output high-level voltage} & \multirow[t]{2}{*}{All output pins} & \multirow[t]{2}{*}{\(\mathrm{V}_{\text {OH }}\)} & \(\mathrm{V}_{\text {cc }}-0.5\) & - & - & V & \(\mathrm{I}_{\text {OH }}=-200 \mu \mathrm{~A}\) \\
\hline & & & \(\mathrm{V}_{\text {cc }}-1.0\) & - & - & V & \(\mathrm{I}_{\mathrm{OH}}=-1 \mathrm{~mA}\) \\
\hline \multicolumn{2}{|l|}{Output low- All output pins level voltage} & \(\mathrm{V}_{\text {oL }}\) & - & - & 0.4 & V & \(\mathrm{I}_{\mathrm{OL}}=1.6 \mathrm{~mA}\) \\
\hline \multirow[t]{3}{*}{Input capacitance} & \(\overline{\text { RES }}\) & \multirow[t]{3}{*}{\(\mathrm{C}_{\text {in }}\)} & - & - & 80*2 & pF & \(\mathrm{Vin}=0 \mathrm{~V}\) \\
\hline & NMI & & - & - & 50 & pF & \(\mathrm{f}=1 \mathrm{MHz}\) \\
\hline & All other input pins & & - & - & 20 & pF & \(\mathrm{Ta}=25^{\circ} \mathrm{C}\) \\
\hline \multirow[t]{4}{*}{Current consumption} & During normal operations & \multirow[t]{4}{*}{\(\mathrm{I}_{\mathrm{cc}}\)} & - & 80 & 130 & mA & \(\mathrm{f}=16.7 \mathrm{MHz}\) \\
\hline & During sleep mode & & - & 70 & 110 & mA & \(\mathrm{f}=16.7 \mathrm{MHz}\) \\
\hline & During standby mode & & - & 0.01 & 5 & \(\mu \mathrm{A}\) & \(\mathrm{T}_{\mathrm{a}} \leq 50^{\circ} \mathrm{C}\) \\
\hline & & & - & - & 20 & \(\mu \mathrm{A}\) & \(50^{\circ} \mathrm{C}<\mathrm{Ta}\) \\
\hline \multirow[t]{2}{*}{Analog supply current} & & \(\mathrm{Al}_{\text {cc }}\) & - & 4 & 8 & mA & \(f=16.7 \mathrm{MHz}\) \\
\hline & & \(\mathrm{Al}_{\text {ref }}\) & - & 0.5 & \(1^{* 3}\) & mA & QFP144 version only \\
\hline RAM standby voltage & & \(V_{\text {RAM }}\) & 2.0 & - & - & V & \\
\hline
\end{tabular}

Notes: 1. Do not release \(A V_{\text {cc }}, \mathrm{AV}_{\text {ss }}\) and \(\mathrm{AV}_{\text {ref }}\) (SH7041, SH7043 and SH7045 only) pins when not using the A/D converter (including standby).
Connect AV \({ }_{\text {cc }}\) (SH7041,SH7043,SH7045 only) and \(\mathrm{AV}_{\text {ref }}(\mathrm{SH} 7041, \mathrm{SH} 7043\) and SH7045 only) pins to \(\mathrm{V}_{\mathrm{cc}}\) and \(A V_{\text {ss }}\) pin to \(\mathrm{V}_{\mathrm{ss}}\).
2. The value for consumed current is with conditions of \(\mathrm{V}_{I H} \min =\mathrm{V}_{\mathrm{CC}}-0.5 \mathrm{~V}\) and \(\mathrm{V}_{\text {IL }} \max =\) 0.5 V , with no burden on any of the output pins.
3. The ZTAT and mask versions have the same functions, and the electrical characteristics of both are within specification, but characteristic-related performance values, operating margins, noise margins, noise emission, etc., are different. Caution is therefore required in carrying out system design, and when switching between ZTAT and mask versions.
*1 SH7042/43 ZTAT (excluding A mask) are 3.2 V .
*2 110pF for A mask
*3 2 mA in the A mask version of MASK products.

\section*{Renesns}

Table 26.3 Permitted Output Current Values (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to 3.6V, \(\mathrm{AV}_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to \(\mathbf{3 . 6 V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=\mathbf{3 . 0} \mathbf{0}^{*}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} V_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Typ & Max & Unit \\
\hline Output low-level permissible current (per pin) & \(\mathrm{I}_{\mathrm{OL}}\) & - & - & 2.0 & mA \\
\hline Output low-level permissible current (total) & \(\sum \mathrm{I}_{\mathrm{OL}}\) & - & - & 80 & mA \\
\hline Output high-level permissible current (per pin) & \(-\mathrm{I}_{\mathrm{OH}}\) & - & - & 2.0 & mA \\
\hline Output high-level permissible current (total) & \(\sum\left(-\mathrm{I}_{\mathrm{OH}}\right)\) & - & - & 25 & mA \\
\hline
\end{tabular}

Notes: To assure LSI reliability, do not exceed the output values listed in this table.
* SH7042/43 ZTAT (excluding A mask) are 3.2V.

\subsection*{26.3 AC Characteristics}

\subsection*{26.3.1 Clock Timing}

Table 26.4 Clock Timing (Conditions: \(V_{C C}=3.0^{*}\) to \(3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.0^{*}\) to \(3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm\) \(\mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(\mathrm{AV} \mathrm{CC}, \mathrm{V}_{\text {SS }}=A V_{\text {SS }}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=-20\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
& Symbol & Min & Max & Unit & Figures \\
\hline Operating frequency & \(\mathrm{f}_{\mathrm{OP}}\) & 4 & 16.7 & MHz & 26.1 \\
\hline Clock cycle time & \(\mathrm{t}_{\mathrm{cyc}}\) & 60 & 250 & ns \\
\hline Clock low-level pulse width & \(\mathrm{t}_{\mathrm{CL}}\) & 10 & - & ns \\
\hline Clock high-level pulse width & \(\mathrm{t}_{\mathrm{CH}}\) & 10 & - & ns \\
\hline Clock rise time & \(\mathrm{t}_{\mathrm{CR}}\) & - & 5 & ns \\
\hline Clock fall time & \(\mathrm{t}_{\mathrm{CF}}\) & - & 5 & ns \\
\hline EXTAL clock input frequency & \(\mathrm{f}_{\mathrm{EX}}\) & 4 & 10 & MHz & 26.2 \\
\hline EXTAL clock input cycle time & \(\mathrm{t}_{\mathrm{EXCyc}}\) & 100 & 250 & ns \\
\hline EXTAL clock low-level input pulse width & \(\mathrm{t}_{\mathrm{ExL}}\) & 40 & - & ns \\
\hline EXTAL clock high-level input pulse width & \(\mathrm{t}_{\mathrm{EXH}}\) & 40 & - & ns \\
\hline EXTAL clock input rise time & \(\mathrm{t}_{\mathrm{EXR}}\) & - & 5 & ns \\
\hline EXTAL clock input fall time & \(\mathrm{t}_{\mathrm{EXF}}\) & - & 5 & ns \\
\hline Reset oscillation settling time & \(\mathrm{t}_{\mathrm{OSC} 1}\) & 10 & - & ms \\
\hline Standby return clock settling time & \(\mathrm{t}_{\mathrm{OSC} 2}\) & 10 & - & ms \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.1 System Clock Timing


Figure 26.2 EXTAL Clock Input Timing


Figure 26.3 Oscillation Settling Time

\subsection*{26.3.2 Control Signal Timing}

Table 26.5 Control Signal Timing (Conditions: \(V_{C C}=3.0^{* 1}\) to 3.6 V , \(\mathrm{AV}_{\mathrm{CC}}=3 . \mathbf{0}^{* 1}\) to 3.6V, \(A V_{C C}=V_{C C} \pm \mathbf{1 0 \%}, A V_{\text {ref }}=\mathbf{3 . 0} \mathbf{0}^{* 1}\) to \(A V_{C C}, V_{S S}=A V_{S S}=0 V, T_{\mathrm{a}}=\mathbf{- 2 0}\) to \(\left.+\mathbf{7 5}{ }^{\circ} \mathrm{C}\right)\)
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline \(\overline{\mathrm{RES}}\) rise/fall & \(\mathrm{t}_{\text {RES } \mathrm{r}}, \mathrm{t}_{\text {RES }}\) & - & 200 & ns & \multirow[t]{3}{*}{26.4} \\
\hline \(\overline{\mathrm{RES}}\) pulse width & \(\mathrm{t}_{\text {RESW }}\) & 20 & - & \(\mathrm{t}_{\text {cyc }}\) & \\
\hline \(\overline{\text { MRES }}\) pulse width & \(\mathrm{t}_{\text {MRESW }}\) & 20 & - & \(\mathrm{t}_{\text {cyc }}\) & \\
\hline NMI rise/fall & \(\mathrm{t}_{\text {MMİ, }}, \mathrm{t}_{\text {MMIF }}\) & - & 200 & ns & 26.5 \\
\hline \(\overline{\mathrm{RES}}\) setup time \({ }^{* 1}\) & \(\mathrm{t}_{\text {RESS }}\) & 100 & - & ns & 26.4 \\
\hline \(\overline{\text { MRES }}\) setup time* \({ }^{* 1}\) & \(\mathrm{t}_{\text {MRESS }}\) & 100 & - & ns & \multirow[t]{4}{*}{26.5} \\
\hline NMI setup time (during edge detection) & \(\mathrm{t}_{\text {NMIS }}\) & 100 & - & ns & \\
\hline \(\overline{\mathrm{IRQ}}-\overline{\mathrm{IRQ0}}\) setup time (edge detection)*2 & \(\mathrm{t}_{\text {IRQES }}\) & 100 & - & ns & \\
\hline \(\overline{\mathrm{IRQ}}-\overline{\mathrm{IRQ0}}\) setup time (level detection)*2 & \(\mathrm{t}_{\text {IRaLS }}\) & 100 & - & ns & \\
\hline NMI hold time & \(\mathrm{t}_{\text {NMIH }}\) & 50 & - & ns & \multirow[t]{2}{*}{26.5} \\
\hline \(\overline{\text { IRQ7-IRQ0 }}\) hold time & \(\mathrm{t}_{\text {IRQEH }}\) & 50 & - & ns & \\
\hline IRQOUT output delay time & \(\mathrm{t}_{\text {IROOD }}\) & - & 50 & ns & 26.6 \\
\hline Bus request setup time & \(\mathrm{t}_{\text {Bras }}\) & 35 & - & ns & \multirow[t]{4}{*}{26.7} \\
\hline Bus acknowledge delay time 1 & \(\mathrm{t}_{\text {BACKD1 }}\) & - & 35 & ns & \\
\hline Bus acknowledge delay time 2 & \(\mathrm{t}_{\text {BACKD2 }}\) & - & 35 & ns & \\
\hline Bus three state delay time & \(\mathrm{t}_{\text {BzD }}\) & - & 35 & ns & \\
\hline
\end{tabular}

Notes: *1 SH7042/43 ZTAT (excluding A mask) are 3.2V.
*2 The \(\overline{\mathrm{RES}}, \overline{\mathrm{MRES}}, \mathrm{NMI}, \overline{\mathrm{BREQ}}\), and \(\overline{\mathrm{RQ} 7}-\overline{\mathrm{IRQ0}}\) signals are asynchronous inputs, but when the setup times shown here are provided, the signals are considered to have produced changes at clock rise (for RES, MRES, BREQ) or clock fall (for NMI and \(\overline{\mathrm{IRQ7}}-\overline{\mathrm{RQO}})\). If the setup times are not provided, recognition is delayed until the next clock rise or fall.


Figure 26.4 Reset Input Timing


Figure 26.5 Interrupt Signal Input Timing


Figure 26.6 Interrupt Signal Output Timing


Figure 26.7 Bus Right Release Timing

\subsection*{26.3.3 Bus Timing}

Table 26.6 Bus Timing (Conditions: \(V_{C C}=3.0^{* 1}\) to 3.6V, \(A V_{C C}=3.0^{* 1}\) to \(3.6 \mathrm{~V}, A V_{C C}=V_{C C}\) \(\pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=\mathbf{3 . 0}{ }^{* 1}\) to \(\mathrm{AV}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=\mathbf{0 V}, \mathrm{T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(\left.+75^{\circ} \mathrm{C}\right)\)
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline Address delay time & \(t_{\text {AD }}\) & \(3^{* 4}\) & 35 & ns & 26.8, 9, 11-16, 19 \\
\hline \(\overline{\mathrm{CS}}\) delay time 1 & \(\mathrm{t}_{\text {CSD1 }}\) & \(3^{* 4}\) & 35 & ns & 26.8, 9, 19 \\
\hline \(\overline{\mathrm{CS}}\) delay time 2 & \(\mathrm{t}_{\text {CSD2 }}\) & \(3^{* 4}\) & 35 & ns & \\
\hline Read strobe delay time 1 & \(\mathrm{t}_{\text {RSD1 }}\) & \(3^{* 4}\) & 35 & ns & 26.8, 9, 11-16, 19 \\
\hline Read strobe delay time 2 & \(\mathrm{t}_{\text {RSD2 }}\) & \(3^{* 4}\) & 35 & ns & \\
\hline Read data setup time & \(\mathrm{t}_{\mathrm{RDS}}{ }^{* 5}\) & 25 & - & ns & \\
\hline Read data hold time & \(\mathrm{t}_{\text {RDH }}\) & 0 & - & ns & \\
\hline Write strobe delay time 1 & \(\mathrm{t}_{\text {WSD1 }}\) & \(3^{* 4}\) & 35 & ns & \\
\hline Write strobe delay time 2 & \(\mathrm{t}_{\text {WSD2 }}\) & 3*5 & 35 & ns & \\
\hline Write data delay time & \(\mathrm{t}_{\text {WDD }}\) & - & 45 & ns & \\
\hline Write data hold time & \(\mathrm{t}_{\text {WDH }}\) & 0 & \(25^{* 3}\) & ns & \\
\hline \(\overline{\text { WAIT }}\) setup time & \(\mathrm{t}_{\text {wTS }}\) & 15 & - & ns & 26.10,15, 19 \\
\hline \(\overline{\text { WAIT }}\) hold time & \(\mathrm{t}_{\text {WTH }}\) & 0 & - & ns & \\
\hline \(\overline{\text { RAS }}\) delay time 1 & \(\mathrm{t}_{\text {RASD1 }}\) & \(3^{* 4}\) & 35 & ns & 26.11-18 \\
\hline \(\overline{\text { RAS }}\) delay time 2 & \(\mathrm{t}_{\text {RASD2 }}\) & \(3^{* 4}\) & 35 & ns & \\
\hline \(\overline{\text { CAS }}\) delay time 1 & \(\mathrm{t}_{\text {CASD1 }}\) & \(3^{* 4}\) & 35 & ns & \\
\hline \(\overline{\text { CAS }}\) delay time 2 & \(\mathrm{t}_{\text {CASD2 }}\) & \(3^{* 4}\) & 35 & ns & \\
\hline Read data access time & \(\mathrm{t}_{\mathrm{ACC}}{ }^{\text {*2 }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+2)-45\) & - & ns & 26.8, 9 \\
\hline Access time from read strobe & \(\mathrm{t}_{\mathrm{OE}}{ }^{*}{ }^{\text {2 }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+1.5)-40\) & - & ns & \\
\hline Access time from column address & \(\mathrm{t}_{\text {AA }}{ }^{*}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+2)-45\) & - & ns & 26.11-16 \\
\hline Access time from \(\overline{\text { RAS }}\) & \(\mathrm{t}_{\mathrm{RAC}}{ }^{* 2}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+\mathrm{RCD}+2.5)-40\) & - & ns & \\
\hline Access time from \(\overline{\mathrm{CAS}}\) & \(\mathrm{t}_{\mathrm{CAC}}{ }^{\text {*2 }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{n}+1)-40\) & - & ns & \\
\hline Row address hold time & \(\mathrm{t}_{\text {RAH }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{RCD}+0.5)-15\) & - & ns & \\
\hline Row address setup time & \(\mathrm{t}_{\text {ASR }}\) & 0 & - & ns & \\
\hline Data input setup time & \(\mathrm{t}_{\mathrm{DS}}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\mathrm{m}+0.5)-27\) & - & ns & \\
\hline Data input hold time & \(\mathrm{t}_{\mathrm{DH}}\) & 20 & - & ns & \\
\hline
\end{tabular}

Notes: \(n\) is the wait number. \(m\) is 1 unless the DRAM write cycle wait number is 0 , then \(m\) is 0 . RCD is the set value of the RCD bit of DCR.
*1 SH7042/43 ZTAT (excluding A mask) are 3.2V.
*2 If the access time is satisfied, then the \(\mathrm{t}_{\mathrm{RDS}}\) need not be satisfied.
*3 \(t_{\text {wDH }}(\max )\) is a reference value.
*4 The delay time min values are reference values (typ).
* \(5 t_{\text {RDS }}\) is a reference value.

\section*{renesas}

Table 26.7 Bus Timing (Conditions: \(V_{C C}=3.0^{* 1}\) to 3.6V, \(A V_{C C}=3.0^{* 1}\) to 3.6V, \(A V_{C C}=V_{C C}\) \(\pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=\mathbf{3 . 0}{ }^{* 1}\) to \(\mathrm{AV}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathbf{A V} V_{\mathrm{SS}}=\mathbf{0} \mathrm{V}, \mathrm{T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+\mathbf{7 5}{ }^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline Write address setup time & \(\mathrm{t}_{\text {AS }}\) & 0 & - & ns & 26.8, 9 \\
\hline Write address hold time & \(t_{\text {wR }}\) & 5 & - & ns & 26.8, 9, 19 \\
\hline Write data hold time & \(\mathrm{t}_{\text {wRH }}\) & 0 & - & ns & \\
\hline Read/write strobe delay time 1 & \(\mathrm{t}_{\text {RWD1 }}\) & \(3^{* 2}\) & 27 & ns & 26.11-16 \\
\hline Read/write strobe delay time 2 & \(\mathrm{t}_{\text {RWD2 }}\) & \(3^{* 2}\) & 27 & ns & \\
\hline High-speed page mode \(\overline{\mathrm{CAS}}\) & \(\mathrm{t}_{\mathrm{CP}}\) & \(\mathrm{t}_{\text {cyc }}-35\) & - & ns & 26.16 \\
\hline \(\overline{\mathrm{RAS}}\) precharge time & \(\mathrm{t}_{\text {RP }}\) & \(\mathrm{t}_{\mathrm{cyc}} \times(\) TPC +1.5\()-20\) & - & ns & 26.11-16 \\
\hline \(\overline{\mathrm{CAS}}\) setup time & \(\mathrm{t}_{\text {CSR }}\) & 10 & - & ns & 26.17, 18 \\
\hline \(\overline{\mathrm{AH}}\) delay time 1 & \(\mathrm{t}_{\text {AHD } 1}\) & \(3^{* 2}\) & 40 & ns & 26.19 \\
\hline \(\overline{\mathrm{AH}}\) delay time 2 & \(\mathrm{t}_{\text {AHO2 }}\) & \(3^{* 2}\) & 40 & ns & \\
\hline Multiplex address delay time & \(\mathrm{t}_{\text {MAD }}\) & \(3^{* 2}\) & 35 & ns & \\
\hline Multiplex address hold time & \(\mathrm{t}_{\text {MAH }}\) & 0 & - & ns & \\
\hline DACK delay time 1 & \(\mathrm{t}_{\text {DACKD1 }}\) & 3*2 & 45 & ns & 26.8, 9, 11-16,19 \\
\hline
\end{tabular}

Notes: TPC is the set value of the TPC bit in DCR.
* 1 SH7042/43 ZTAT (excluding A mask) are 3.2V
*2 Min values for delay time are reference values (typ)


Figure 26.8 Basic Cycle (No Waits)


Figure 26.9 Basic Cycle (Software Waits)


Figure 26.10 Basic Cycle (2 Software Waits + Wait due to \(\overline{\text { WAIT }}\) Signal)


Figure 26.11 DRAM Cycle (Normal Mode, No Wait, TPC=0, RCD=0)


Figure 26.12 DRAM Cycle (Normal Mode, 1 Wait, \(\mathbf{T P C}=\mathbf{0}, \mathbf{R C D}=0\) )


Figure 26.13 DRAM Cycle (Normal Mode, 2 Waits, \(\mathbf{T P C}=1, \mathbf{R C D}=1\) )


Figure 26.14 DRAM Cycle (Normal Mode, 3 Waits, \(\mathbf{T P C}=1\), RCD = 1)


Figure 26.15 DRAM Cycle (Normal Mode, 2 Waits + Wait due to \(\overline{\text { WAIT }}\) Signal)


Figure 26.16 DRAM Cycle (High-Speed Page Mode)


Figure 26.17 CAS Before RAS Refresh (TRAS1 \(=0\), TRAS0 \(=0\) )


Figure 26.18 Self Refresh


Figure 26.19 Address Data Multiplex I/O Space Cycle (1 Software Wait + External Wait)

\subsection*{26.3.4 Direct Memory Access Controller Timing}

Table 26.8 Direct Memory Access Controller Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to \(\mathbf{3 . 6 V}\), \(\mathrm{AV}_{\mathrm{CC}}\) \(=3.0^{*}\) to \(3.6 \mathrm{~V}, A V_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm 10 \%, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=A \mathrm{~V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=\) -20 to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline\(\overline{\text { DREQ0, }} \overline{\text { DREQ1 }}\) setup time & \(\mathrm{t}_{\text {DRQS }}\) & 35 & - & ns & 26.20 \\
\hline\(\overline{\text { DREQ0, }} \overline{\text { DREQ1 }}\) hold time & \(\mathrm{t}_{\text {DRQH }}\) & 35 & - & ns & \\
\hline\(\overline{\text { DREQ0, }} \overline{\text { DREQ1 }}\) pulse width & \(\mathrm{t}_{\text {DRQW }}\) & 1.5 & - & \(\mathrm{t}_{\text {cyc }}\) & 26.21 \\
\hline DRAK output delay time & \(\mathrm{t}_{\text {DRAKD }}\) & - & 35 & ns & 26.22 \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.20 \(\overline{\text { DREQ0 }}\) and \(\overline{\text { DREQ1 }}\) Input Timing (1)


Figure 26.21 \(\overline{\text { DREQ0 }}\) and \(\overline{\text { DREQ1 }}\) Input Timing (2)


Figure 26.22 DRAK Output Delay Time

\subsection*{26.3.5 Multifunction Timer Pulse Unit Timing}

Table 26.9 Multifunction Timer Pulse Unit Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to 3.6V, \(\mathrm{AV}_{\mathrm{CC}}=\) \(3.0^{*}\) to \(3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm 10 \%, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(\mathrm{AV} \mathrm{CC}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{AV} \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=\) -20 to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|}
\hline Item & Symbol & Min & Max & Unit & Figure \\
\hline Output compare output delay time & \(\mathrm{t}_{\text {TOCD }}\) & - & 100 & ns & \multirow[t]{2}{*}{26.23} \\
\hline Input capture input setup time & \(t_{\text {TICS }}\) & 100 & - & ns & \\
\hline Timer input setup time & \(\mathrm{t}_{\text {TCKS }}\) & 100 & - & ns & \multirow[t]{4}{*}{26.24} \\
\hline Timer clock pulse width (single edge specification) & \(\mathrm{t}_{\text {TCKWHLL }}\) & 1.5 & - & \(\mathrm{t}_{\text {cyc }}\) & \\
\hline Timer clock pulse width (both edges specified) & \(\mathrm{t}_{\text {TcKWH/L }}\) & 2.5 & - & \(\mathrm{t}_{\text {cyc }}\) & \\
\hline Timer clock pulse width (phase measurement mode) & \(\mathrm{t}_{\text {TcKwH/L }}\) & 2.5 & - & \(\mathrm{t}_{\text {cyc }}\) & \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.23 MTU I/O Timing


Figure 26.24 MTU Clock Input Timing

\subsection*{26.3.6 I/O Port Timing}

Table 26.10 I/O Port Timing (Conditions: \(V_{C C}=3.0^{*}\) to \(3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.0^{*}\) to \(3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=\) \(\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\mathrm{A} \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline Port output data delay time & \(\mathrm{t}_{\text {PWD }}\) & - & 100 & ns & 26.25 \\
\hline Port input hold time & \(\mathrm{t}_{\text {PRH }}\) & 100 & - & ns & \\
\hline Port input setup time & \(\mathrm{t}_{\text {PRS }}\) & 100 & - & ns & \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.25 I/O Port I/O Timing

\subsection*{26.3.7 Watchdog Timer Timing}

Table 26.11 Watchdog Timer Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3.0^{*}\) to \(3.6 \mathrm{~V}, \mathrm{AV}_{\mathrm{CC}}=3.0^{*}\) to 3.6V, \(A V_{C C}=V_{C C} \pm \mathbf{1 0 \%}, A V_{\text {ref }}=3.0^{*}\) to \(A V_{C C}, V_{S S}=A V_{S S}=0 V, T_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline\(\overline{\text { WDTOVF }}\) delay time & \(\mathrm{t}_{\text {wovd }}\) & - & 100 & ns & 26.26 \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.26 Watchdog Timer Timing

\subsection*{26.3.8 Serial Communication Interface Timing}

Table 26.12 Serial Communication Interface Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to 3.6V, \(\mathrm{AV}_{\mathrm{CC}}=\) \(3.0^{*}\) to \(\mathbf{3 . 6 V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(\mathrm{AV} V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=A V_{\mathrm{SS}}=\mathbf{0 V}, \mathrm{T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{llllll} 
Item & Symbol & Min & Max & Unit & Figure \\
\hline Input clock cycle & \(\mathrm{t}_{\text {scyc }}\) & 4 & - & \(\mathrm{t}_{\mathrm{cyc}}\) & 26.27 \\
\cline { 1 - 5 } Input clock cycle (clock sync) & \(\mathrm{t}_{\text {scyc }}\) & 6 & - & \(\mathrm{t}_{\mathrm{cyc}}\) & \\
\hline Input clock pulse width & \(\mathrm{t}_{\text {sckw }}\) & 0.5 & 0.6 & \(\mathrm{t}_{\text {scyc }}\) & \\
\hline Input clock rise time & \(\mathrm{t}_{\text {sckr }}\) & - & 1.5 & \(\mathrm{t}_{\mathrm{cyc}}\) & \\
\hline Input clock fall time & \(\mathrm{t}_{\text {sckf }}\) & - & 1.5 & \(\mathrm{t}_{\mathrm{cyc}}\) & \\
\hline Transmit data delay time (clock sync) & \(\mathrm{t}_{\mathrm{TXD}}\) & - & 100 & ns & 26.28 \\
\hline Receive data setup time (clock sync) & \(\mathrm{t}_{\mathrm{Rxs}}\) & 100 & - & ns & \\
\hline Receive data hold time (clock sync) & \(\mathrm{t}_{\mathrm{RXH}}\) & 100 & - & ns & \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.27 Input Clock Timing


Figure 26.28 SCI I/O Timing (Clock Sync Mode)

\subsection*{26.3.9 High-speed A/D Converter Timing (excluding A mask)}

Table 26.13 High-speed A/D Converter Timing (Conditions: \(V_{C C}=3.0^{*}\) to 3.6V, \(A V_{C C}=3.0^{*}\) to \(\mathbf{3 . 6 V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=A V_{\mathrm{SS}}=\mathbf{0 V}, \mathrm{T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{Item} & Symbol & Min & Typ & Max & Unit & Figure \\
\hline \multicolumn{2}{|l|}{External trigger input pulse width} & \(\mathrm{t}_{\text {traw }}\) & 2 & - & - & \(\mathrm{t}_{\text {cyc }}\) & 26.29 \\
\hline \multicolumn{2}{|l|}{External trigger input start delay time} & \(\mathrm{t}_{\text {TRGS }}\) & 50 & - & - & ns & \\
\hline \multirow[t]{2}{*}{A/D conversion start delay time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\mathrm{D}}\)} & 1.5 & 1.5 & 1.5 & \multirow[t]{6}{*}{\(\mathrm{t}_{\text {cyc }}\)} & \multirow[t]{6}{*}{26.30} \\
\hline & CKS \(=1\) & & 1.5 & 1.5 & 1.5 & & \\
\hline \multirow[t]{2}{*}{Input sampling time} & CKS \(=0\) & \multirow[t]{2}{*}{\(t_{\text {SPL }}\)} & 20 & 20 & 20 & & \\
\hline & CKS \(=1\) & & 40 & 40 & 40 & & \\
\hline \multirow[t]{2}{*}{A/D conversion time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {conv }}\)} & 42.5 & 42.5 & 42.5 & & \\
\hline & CKS = 1 & & 82.5 & 82.5 & 82.5 & & \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.29 External Trigger Input Timing


Figure 26.30 Analog Conversion Timing

\subsection*{26.3.10 Mid-speed Converter Timing (A mask)}

Table 26.14 Mid-speed A/D Converter Timing (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to 3.6V, \(A V_{\mathrm{CC}}=3 . \mathbf{0}^{*}\) to \(\mathbf{3 . 6 V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=3.0^{*}\) to \(\mathrm{AV}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=A V_{\mathrm{SS}}=\mathbf{0 V}, \mathrm{T}_{\mathrm{a}}=\mathbf{- 2 0}\) to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{Item} & Symbol & Min & Typ & Max & Unit & Figure \\
\hline \multicolumn{2}{|l|}{External trigger input pulse width} & \(\mathrm{t}_{\text {tRGW }}\) & 2 & - & - & \(\mathrm{t}_{\text {cyc }}\) & 26.31 \\
\hline \multicolumn{2}{|l|}{External trigger input start delay time} & \(\mathrm{t}_{\text {TRGS }}\) & 50 & - & - & ns & \\
\hline \multirow[t]{2}{*}{A/D conversion start delay time} & CKS \(=0\) & \(\mathrm{t}_{\mathrm{D}}\) & 10 & - & 17 & \multirow[t]{6}{*}{\(\mathrm{t}_{\text {cyc }}\)} & \multirow[t]{6}{*}{26.32} \\
\hline & CKS \(=1\) & & 6 & - & 9 & & \\
\hline \multirow[t]{2}{*}{Input sampling time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {SPL }}\)} & - & 64 & - & & \\
\hline & CKS \(=1\) & & - & 32 & - & & \\
\hline \multirow[t]{2}{*}{A/D conversion time} & CKS \(=0\) & \multirow[t]{2}{*}{\(\mathrm{t}_{\text {conv }}\)} & 259 & - & 266 & & \\
\hline & CKS \(=1\) & & 131 & - & 134 & & \\
\hline
\end{tabular}

Note: * SH7042/43 ZTAT (excluding A mask) are 3.2V.


Figure 26.31 External Trigger Input Timing


Figure 26.32 Analog Conversion Timing

\subsection*{26.3.11 Measurement Conditions for AC Characteristic}
- Input reference levels:
— High level: 2.2 V
— Low level: 0.8 V
- Output reference levels:
— High level: 2.0 V
_ Low level: 0.8 V


Note: \(C_{L}\) is set with the following pins, including the total capacitance of the measurement equipment etc:
\(30 \mathrm{pF}: \quad \mathrm{CK}, \overline{\mathrm{RAS}}, \overline{\mathrm{CASxx}}, \mathrm{RDWR}, \overline{\mathrm{CSO}}-\overline{\mathrm{CS} 3}, \overline{\mathrm{AH}}, \overline{\mathrm{BREQ}}, \overline{\mathrm{BACK}}, \mathrm{DACKO}\), DACK1, and IRQOUT
\(50 \mathrm{pF}:\) A21-A0, D31-D0, \(\overline{\mathrm{RD}}, \overline{\mathrm{WRxx}}\)
70 pF : Port output and peripheral module output pins other than the above.
\(\mathrm{I}_{\mathrm{OL}}\), \(\mathrm{I}_{\mathrm{OH}}:\) See table 26.3, Permitted Output Current Values.

Figure 26.33 Output Load Circuit

\subsection*{26.4 A/D Converter Characteristics}

Table 26.15 A/D Converter Characteristics (excluding A mask) (Conditions: \(\mathrm{V}_{\mathrm{CC}}=\mathbf{3 . 0}^{* 1}\) to \(\mathbf{3 . 6 V}, \mathrm{AV}_{\mathrm{CC}}=3 . \mathbf{0}^{* 1}\) to \(\mathbf{3 . 6} \mathrm{V}, \mathrm{AV}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm \mathbf{1 0 \%}, \mathrm{AV}_{\text {ref }}=3 . \mathbf{0}^{* 1}\) to \(\mathrm{AV} \mathrm{VC}_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=\) \(A V_{S S}=0 V, T_{a}=-20\) to \(\left.+75^{\circ} \mathrm{C}\right)\)
\begin{tabular}{lllll} 
& \multicolumn{3}{c}{ 16.7MHz } & \\
\cline { 2 - 5 } Item & Min & Typ & Max & Unit \\
\hline Resolution & 10 & 10 & 10 & bit \\
\hline Conversion time (when CKS =1) & - & - & 5 & \(\mu \mathrm{~s}\) \\
\hline Analog input capacity & - & - & 20 & pF \\
\hline Permission signal source impedance & - & - & 1 & \(\mathrm{k} \Omega\) \\
\hline Non-linearity error*2 & - & - & \(\pm 15\) & LSB \\
\hline Offset error*2 & - & - & \(\pm 15\) & LSB \\
\hline Full scale error*2 & - & - & \(\pm 15\) & LSB \\
\hline Quantize error*2 & - & - & \(\pm 0.5\) & LSB \\
\hline Absolute error & - & - & \(\pm 31\) & LSB \\
\hline N & & & &
\end{tabular}

Notes: *1 SH7042/43 ZTAT (excluding A mask) are 3.2V.
*2 Reference values

Table 26.16 A/D Converter Characteristics (A mask) (Conditions: \(\mathrm{V}_{\mathrm{CC}}=3 . \mathbf{0}^{* 1}\) to 3.6V, \(\mathrm{AV}_{\mathrm{CC}}\) \(=3.0^{* 1}\) to \(3.6 \mathrm{~V}, A V_{\mathrm{CC}}=\mathrm{V}_{\mathrm{CC}} \pm 10 \%, A V_{\text {ref }}=3.0^{* 1}\) to \(A V_{\mathrm{CC}}, \mathrm{V}_{\mathrm{SS}}=A V_{\mathrm{SS}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{a}}=\) -20 to \(+75^{\circ} \mathrm{C}\) )
\begin{tabular}{|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Item} & \multicolumn{3}{|c|}{16.7MHz} & \multirow[b]{2}{*}{Unit} \\
\hline & min & typ & max & \\
\hline Resolution & 10 & 10 & 10 & bit \\
\hline Conversion time (when CKS = 0) & - & - & 16.0 & \(\mu \mathrm{s}\) \\
\hline Analog input capacity & - & - & 20 & pF \\
\hline Permission signal source impedance & - & - & 1 & \(\mathrm{k} \Omega\) \\
\hline Non-linearity error*2 & - & - & \(\pm 4\) & LSB \\
\hline Offset error*2 & - & - & \(\pm 4\) & LSB \\
\hline Full scale error*2 & - & - & \(\pm 4\) & LSB \\
\hline Quantize error*2 & - & - & \(\pm 0.5\) & LSB \\
\hline Absolute error & - & - & \(\pm 6\) & LSB \\
\hline
\end{tabular}

Notes: *1 SH7042/43 ZTAT (excluding A mask) are 3.2V.
*2 Reference values

\section*{Appendix A On-Chip Supporting Module Registers}

\section*{A. 1 Addresses}

Table A. 1 On-Chip I/O Register Addresses

\begin{tabular}{lllllllllll}
\hline H'FFFF81A0 & SMR0 & C/A & CHR & PE & O/E & STOP & MP & CKS1 & CKS0 & SCI \\
\hline H'FFFF81A1 & BRR0 & & & & & & & & \\
\hline H'FFFF81A2 & SCR0 & TIE & RIE & TE & RE & MPIE & TEIE & CKE1 & CKE0 \\
\hline H'FFFF81A3 & TDR0 & & & & & & & & \\
\hline H'FFFF81A4 & SSR0 & TDRE & RDRF & ORER & FER & PER & TEND & MPB & MPBT \\
\hline H'FFFF81A5 RDR0 & & & & & & & & \\
\hline \begin{tabular}{l} 
H'FFFF81A6 \\
to \\
H'FFFF81AF
\end{tabular} & - & - & - & - & - & - & - & - \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{Register Abbr.} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF81B0 & SMR1 & C/A & CHR & PE & O/E & STOP & MP & CKS1 & CKS0 & SCI \\
\hline H'FFFF81B1 & BRR1 & & & & & & & & & \\
\hline H'FFFF81B2 & SCR1 & TIE & RIE & TE & RE & MPIE & TEIE & CKE1 & CKEO & \\
\hline H'FFFF81B3 & TDR1 & & & & & & & & & \\
\hline H'FFFF81B4 & SSR1 & TDRE & RDRF & ORER & FER & PER & TEND & MPB & MPBT & \\
\hline H'FFFF81B5 & RDR1 & & & & & & & & & \\
\hline H'FFFF81B6 to H'FFFF81FF & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8200 & TCR3 & CCLR2 & CCLR1 & CCLRO & CKEG1 & CKEGO & TPSC2 & TPSC1 & TPSC0 & MTU \\
\hline H'FFFF8201 & TCR4 & CCLR2 & CCLR1 & CCLR0 & CKEG1 & CKEGO & TPSC2 & TPSC1 & TPSC0 & \\
\hline H'FFFF8202 & TMDR3 & - & - & BFB & BFA & MD3 & MD2 & MD1 & MDO & \\
\hline H'FFFF8203 & TMDR4 & - & - & BFB & BFA & MD3 & MD2 & MD1 & MD0 & \\
\hline H'FFFF8204 & TIOR3H & IOB3 & IOB2 & IOB1 & IOBO & IOA3 & IOA2 & IOA1 & IOAO & \\
\hline H'FFFF8205 & TIOR3L & IOD3 & IOD2 & IOD1 & IODO & IOC3 & \(1 \mathrm{OC2}\) & IOC1 & IOCO & \\
\hline H'FFFF8206 & TIOR4H & IOB3 & IOB2 & IOB1 & IOBO & IOA3 & IOA2 & IOA1 & IOAO & \\
\hline H'FFFF8207 & TIOR4L & IOD3 & IOD2 & IOD1 & IODO & IOC3 & IOC2 & IOC1 & IOCO & \\
\hline H'FFFF8208 & TIER3 & TTGE & - & - & TCIEV & TGIED & TGIEC & TGIEB & TGIEA & \\
\hline H'FFFF8209 & TIER4 & TTGE & - & - & TCIEV & TGIED & TGIEC & TGIEB & TGIEA & \\
\hline H'FFFF820A & TOER & - & - & OE4D & OE4C & OE3D & OE4B & OE4A & OE3B & \\
\hline H'FFFF820B & TOCR & - & PSYE & - & - & - & - & OLSN & OLSP & \\
\hline H'FFFF820C & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF820D & TGCR & - & BDC & N & P & FB & WF & VF & UF & \\
\hline H'FFFF820E & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF820F & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8210 & \multirow[t]{2}{*}{TCNT3} & & & & & & & & & \\
\hline H'FFFF8211 & & & & & & & & & & \\
\hline H'FFFF8212 & \multirow[t]{2}{*}{TCNT4} & & & & & & & & & \\
\hline H'FFFF8213 & & & & & & & & & & \\
\hline H'FFFF8214 & \multirow[t]{2}{*}{TCDR} & & & & & & & & & \\
\hline H'FFFF8215 & & & & & & & & & & \\
\hline H'FFFF8216 & \multirow[t]{2}{*}{TDDR} & & & & & & & & & \\
\hline H'FFFF8217 & & & & & & & & & & \\
\hline H'FFFF8218 & \multirow[t]{2}{*}{TGR3A} & & & & & & & & & \\
\hline H'FFFF8219 & & & & & & & & & & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)


Table A. 1 On-Chip I/O Register Addresses (cont)


Table A. 1 On-Chip I/O Register Addresses (cont)


Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{Register Abbr.} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF838A & - & - & - & - & - & - & - & - & - & PFC \\
\hline H'FFFF838B & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF838C & PACRL1 & - & PA15MD & - & PA14MD & - & PA13MD & - & PA12MD & \\
\hline H'FFFF838D & & - & PA11MD & - & PA10MD & PA9MD1 & PA9MD0 & PA8MD1 & PA8MD0 & \\
\hline H'FFFF838E & PACRL2 & PA7MD1 & PA7MD0 & PA6MD1 & PA6MD0 & PA5MD1 & PA5MD0 & - & PA4MD & \\
\hline H'FFFF838F & & - & PA3MD & PA2MD1 & PA2MD0 & - & PA1MD & - & PAOMD & \\
\hline H'FFFF8390 & PBDR & - & - & - & - & - & - & PB9DR & PB8DR & 1/0 \\
\hline H'FFFF8391 & & PB7DR & PB6DR & PB5DR & PB4DR & PB3DR & PB2DR & PB1DR & PB0DR & \\
\hline H'FFFF8392 & PCDR & PC15DR & PC14DR & PC13DR & PC12DR & PC11DR & PC10DR & PC9DR & PC8DR & \\
\hline H'FFFF8393 & & PC7DR & PC6DR & PC5DR & PC4DR & PC3DR & PC2DR & PC1DR & PCODR & \\
\hline H'FFFF8394 & PBIOR & - & - & - & - & - & - & PB9IOR & PB8IOR & PFC \\
\hline H'FFFF8395 & & PB7IOR & PB6IOR & PB5IOR & PB4IOR & PB3IOR & PB2IOR & PB1IOR & PBOIOR & \\
\hline H'FFFF8396 & PCIOR & PC15IOR & PC14IOR & PC13IOR & PC12IOR & PC11IOR & PC10IOR & PC9IOR & PC8IOR & \\
\hline H'FFFF8397 & & PC7IOR & PC6IOR & PC5IOR & PC4IOR & PC3IOR & PC2IOR & PC1IOR & PCOIOR & \\
\hline H'FFFF8398 & PBCR1 & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8399 & & - & - & - & - & PB9MD1 & PB9MD0 & PB8MD1 & PB8MD0 & \\
\hline H'FFFF839A & PBCR2 & PB7MD1 & PB7MD0 & PB6MD1 & PB6MD0 & PB5MD1 & PB5MD0 & PB4MD1 & PB4MD0 & \\
\hline H'FFFF839B & & PB3MD1 & PB3MD0 & PB2MD1 & PB2MD0 & - & PB1MD & - & PBOMD & \\
\hline H'FFFF839C & PCCR & PC15MD & PC14MD & PC13MD & PC12MD & PC11MD & PC10MD & PC9MD & PC8MD & \\
\hline H'FFFF839D & & PC7MD & PC6MD & PC5MD & PC4MD & PC3MD & PC2MD & PC1MD & PCOMD & \\
\hline H'FFFF839E & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF839F & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83A0 & PDDRH & PD31DR & PD30DR & PD29DR & PD28DR & PD27DR & PD26DR & PD25DR & PD24DR & I/O \\
\hline H'FFFF83A1 & & PD23DR & PD22DR & PD21DR & PD20DR & PD19DR & PD18DR & PD17DR & PD16DR & \\
\hline H'FFFF83A2 & PDDRL & PD15DR & PD14DR & PD13DR & PD12DR & PD11DR & PD10DR & PD9DR & PD8DR & \\
\hline H'FFFF83A3 & & PD7DR & PD6DR & PD5DR & PD4DR & PD3DR & PD2DR & PD1DR & PDODR & \\
\hline H'FFFF83A4 & PDIORH & PD31IOR & PD30IOR & PD29IOR & PD28IOR & PD27IOR & PD26IOR & PD25IOR & PD24IOR & PFC \\
\hline H'FFFF83A5 & & PD23IOR & PD22IOR & PD21IOR & PD20IOR & PD19IOR & PD18IOR & PD17IOR & PD16IOR & \\
\hline H'FFFF83A6 & PDIORL & PD15IOR & PD14IOR & PD13IOR & PD12IOR & PD11IOR & PD10IOR & PD9IOR & PD8IOR & \\
\hline H'FFFF83347 & & PD7IOR & PD6IOR & PD5IOR & PD4IOR & PD3IOR & PD2IOR & PD1IOR & PDOIOR & \\
\hline H'FFFF83A8 & PDCRH1 & PD31MD1 & PD31MD0 & PD30MD1 & PD30MD0 & PD29MD1 & PD29MD0 & PD28MD1 & PD28MD0 & \\
\hline H'FFFF83A9 & & PD27MD1 & PD27MD0 & PD26MD1 & PD26MD0 & PD25MD1 & PD25MD0 & PD24MD1 & PD24MD0 & \\
\hline H'FFFF83AA & PDCRH2 & PD23MD1 & PD23MD0 & PD22MD1 & PD22MD0 & PD21MD1 & PD21MD0 & PD20MD1 & PD20MD0 & \\
\hline H'FFFFF83AB & & PD19MD1 & PD19MD0 & PD18MD1 & PD18MD0 & PD17MD1 & PD17MD0 & PD16MD1 & PD16MD0 & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{Register Abbr.} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF83AC & PDCRL & PD15MD & PD14MD & PD13MD & PD12MD & PD11MD & PD10MD & PD9MD & PD8MD & PFC \\
\hline H'FFFF83AD & & PD7MD & PD6MD & PD5MD & PD4MD & PD3MD & PD2MD & PD1MD & PDOMD & \\
\hline H'FFFF83AE & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83AF & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83B0 & PEDR & PE15DR & PE14DR & PE13DR & PE12DR & PE11DR & PE10DR & PE9DR & PE8DR & I/O \\
\hline H'FFFF83B1 & & PE7DR & PE6DR & PE5DR & PE4DR & PE3DR & PE2DR & PE1DR & PEODR & \\
\hline H'FFFF83B2 & PFDR & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83B3 & & PF7DR & PF6DR & PF5DR & PF4DR & PF3DR & PF2DR & PF1DR & PFODR & \\
\hline H'FFFF83B4 & PEIOR & PE15IOR & PE14IOR & PE13IOR & PE12IOR & PE11IOR & PE10IOR & PE9IOR & PE8IOR & PFC \\
\hline H'FFFF83B5 & & PE7IOR & PE6IOR & PE5IOR & PE4IOR & PE3IOR & PE2IOR & PE1IOR & PEOIOR & \\
\hline H'FFFF83B6 & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83B7 & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83B8 & PECR1 & PE15MD1 & PE15MD0 & PE14MD1 & PE14MD0 & PE13MD1 & PE13MD0 & - & PE12MD & \\
\hline H'FFFF83B9 & & - & PE11MD & - & PE10MD & - & PE9MD & - & PE8MD & \\
\hline H'FFFF83BA & PECR2 & - & PE7MD & - & PE6MD & - & PE5MD & - & PE4MD & \\
\hline H'FFFF83BB & & PE3MD1 & PE3MD0 & PE2MD1 & PE2MD0 & PE1MD1 & PE1MD0 & PE0MD1 & PEOMDO & \\
\hline H'FFFF83BC to H'FFFF83BF & - & - & - & - & - & - & - & - & - & - \\
\hline H'FFFF83C0 & ICSR & POE3F & POE2F & POE1F & POE0F & - & - & - & PIE & MTU \\
\hline H'FFFF83C1 & & POE3M1 & POE3M0 & POE2M1 & POE2M0 & POE1M1 & POE1M0 & POE0M1 & POEOMO & \\
\hline H'FFFF83C2 & OCSR & OSF & - & - & - & - & - & OCE & OIE & \\
\hline H'FFFF83C3 & & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83C4 to H'FFFF833C7 & - & - & - & - & - & - & - & - & - & - \\
\hline H'FFFF83C8 & IFCR & - & - & - & - & - & - & - & - & PFC \\
\hline H'FFFF83C9 & & - & - & - & - & IRQMD3 & IRQMD2 & IRQMD1 & IRQMD0 & \\
\hline H'FFFF83CA to H'FFFF83CF & - & - & - & - & - & - & - & - & - & - \\
\hline H'FFFF83D0 & CMSTR & - & - & - & - & - & - & - & - & CMT \\
\hline H'FFFF83D1 &  & - & - & - & - & - & - & STR1 & STR0 & \\
\hline H'FFFF83D2 & CMCSR0 & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83D3 &  & CMF & CMIE & - & - & - & - & CKS1 & CKSO & \\
\hline H'FFFF83D4 & CMCNTO & & & & & & & & & \\
\hline H'FFFF83D5 & & & & & & & & & & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{Register Abbr.} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF833D6 & \multirow[t]{2}{*}{CMCORO} & & & & & & & & & \multirow[t]{10}{*}{CMT} \\
\hline H'FFFF83D7 & & & & & & & & & & \\
\hline H'FFFF83D8 & \multirow[t]{2}{*}{CMCSR1} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83D9 & & CMF & CMIE & - & - & - & - & CKS1 & CKSO & \\
\hline H'FFFF833DA & \multirow[t]{2}{*}{CMCNT1} & & & & & & & & & \\
\hline H'FFFFF83DB & & & & & & & & & & \\
\hline H'FFFF83DC & \multirow[t]{2}{*}{CMCOR1} & & & & & & & & & \\
\hline \multicolumn{9}{|l|}{H'FFFF83DD} & & \\
\hline H'FFFF833DE & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83DF & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF83E0 & ADCSR & ADF & ADIE & ADST & CKS & GRP & CH 2 & CH 1 & CHO & A/D \\
\hline H'FFFF83E1 & ADCR & - & PWR & TRGS1 & TRGS0 & SCAN & DSMP & BUFE1 & BUFE0 & (High \\
\hline \begin{tabular}{l}
H'FFFF83E2 \\
to \\
H'FFFF83EF
\end{tabular} & - & - & - & - & - & - & - & - & - & \begin{tabular}{l}
speed) \\
(Excl. A mask)
\end{tabular} \\
\hline H'FFFF83F0 & \multirow[t]{2}{*}{ADDRA} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83F1 & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83F2 & \multirow[t]{2}{*}{ADDRB} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83F3 & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83F4 & \multirow[t]{2}{*}{ADDRC} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83F5 & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83F6 & \multirow[t]{2}{*}{ADDRD} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83F7 & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83F8 & \multirow[t]{2}{*}{ADDRE} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83F9 & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83FA & \multirow[t]{2}{*}{ADDRF} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83FB & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83FC & \multirow[t]{2}{*}{ADDRG} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83FD & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF83FE & \multirow[t]{2}{*}{ADDRH} & - & - & - & - & - & - & AD9 & AD8 & \\
\hline H'FFFF83FF & & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & AD1 & AD0 & \\
\hline H'FFFF8400 & \multirow[t]{2}{*}{ADDRAO} & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & A/D(Mid- \\
\hline H'FFFF8401 & & AD1 & AD0 & - & - & - & - & - & - & speed) \\
\hline H'FFFF8402 & \multirow[t]{2}{*}{ADDRB0} & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & (A mask \\
\hline H'FFFF8403 & & AD1 & AD0 & - & - & - & - & - & - & \\
\hline H'FFFF8404 & \multirow[t]{2}{*}{ADDRC0} & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & \\
\hline H'FFFF8405 & & AD1 & AD0 & - & - & - & - & - & - & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{\begin{tabular}{l}
Register \\
Abbr.
\end{tabular}} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF8406 & ADDRD0 & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & A/D(Mid- \\
\hline H'FFFF8407 & & AD1 & AD0 & - & - & - & - & - & - & speed) \\
\hline H'FFFF8408 & ADDRA1 & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & (A mask \\
\hline H'FFFF8409 & & AD1 & AD0 & - & - & - & - & - & - & only) \\
\hline H'FFFF840A & ADDRB1 & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & \\
\hline H'FFFF840B & & AD1 & AD0 & - & - & - & - & - & - & \\
\hline H'FFFF840C & ADDRC1 & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & \\
\hline H'FFFF840D & & AD1 & AD0 & - & - & - & - & - & - & \\
\hline H'FFFF840E & ADDRD1 & AD9 & AD8 & AD7 & AD6 & AD5 & AD4 & AD3 & AD2 & \\
\hline H'FFFF840F & & AD1 & ADO & - & - & - & - & - & - & \\
\hline H'FFFF8410 & ADCSR0 & ADF & ADIE & ADST & SCAN & CKS & CH 2 & CH1 & CHO & \\
\hline H'FFFF8411 & ADCSR1 & ADF & ADIE & ADST & SCAN & CKS & CH 2 & CH1 & CHO & \\
\hline H'FFFF8412 & AADCR0 & TRGE & - & - & - & - & - & - & - & \\
\hline H'FFFF8413 & AADCR1 & TRGE & - & - & - & - & - & - & - & \\
\hline H'FFFF8414 to H'FFFF857F & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8580 & FLMCR1 & FWE & SWE & ESU1 & PSU1 & EV1 & PV1 & E1 & P1 & FLASH \\
\hline H'FFFF8581 & FLMCR2 & FLER & - & ESU2 & PSU2 & EV2 & PV2 & E2 & P2 & (F-ZTAT \\
\hline H'FFFF8582 & EBR1 & - & - & - & - & EB3 & EB2 & EB1 & EB0 & version- \\
\hline H'FFFF8583 & EBR2 & EB11 & EB10 & EB9 & EB8 & EB7 & EB6 & EB5 & EB4 & only) \\
\hline H'FFFF8584 to H'FFFF859F & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8600 & UBARH & UBA31 & UBA30 & UBA29 & UBA28 & UBA27 & UBA26 & UBA25 & UBA24 & UBC \\
\hline H'FFFF8601 & & UBA23 & UBA22 & UBA21 & UBA20 & UBA19 & UBA18 & UBA17 & UBA16 & \\
\hline H'FFFF8602 & UBARL & UBA15 & UBA14 & UBA13 & UBA12 & UBA11 & UBA10 & UBA9 & UBA8 & \\
\hline H'FFFF8603 & & UBA7 & UBA6 & UBA5 & UBA4 & UBA3 & UBA2 & UBA1 & UBAO & \\
\hline H'FFFF8604 & UBAMRH & UBM31 & UBM30 & UBM29 & UBM28 & UBM27 & UBM26 & UBM25 & UBM24 & \\
\hline H'FFFF8605 & & UBM23 & UBM22 & UBM21 & UBM20 & UBM19 & UBM18 & UBM17 & UBM16 & \\
\hline H'FFFF8606 & UBAMRL & UBM15 & UBM14 & UBM13 & UBM12 & UBM11 & UBM10 & UBM9 & UBM8 & \\
\hline H'FFFF8607 & & UBM7 & UBM6 & UBM5 & UBM4 & UBM3 & UBM2 & UBM1 & UBM0 & \\
\hline H'FFFF8608 & UBBR & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8609 & & CP1 & CP0 & ID1 & ID0 & RW1 & RW0 & SZ1 & SZO & \\
\hline H'FFFF860A to H'FFFF860F & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8610 & TCSR & OVF & WT/IT & TME & - & - & CKS2 & CKS1 & CKS0 & WDT \\
\hline
\end{tabular}

\section*{renesns}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{Register Abbr.} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF8610 & TCNT*1 & & & & & & & & & WDT \\
\hline H'FFFF8611 & TCNT*2 & & & & & & & & & \\
\hline H'FFFF8612 & RSTCSR*1 & WOVF & RSTE & RSTS & - & - & - & - & - & \\
\hline H'FFFF8613 & RSTCSR*2 & WOVF & RSTE & RSTS & - & - & - & - & - & \\
\hline H'FFFF8614 & SBYCR & SBY & HIZ & - & - & - & - & - & - & Powerdown state \\
\hline H'FFFF86615 to H'FFFFF861F & - & - & - & - & - & - & - & - & - & BSC \\
\hline H'FFFF8620 & BCR1 & - & - & MTURWE & - & - & - & - & IOE & \\
\hline H'FFFF8621 & & A3LG & A2LG & A1LG & A0LG & A3SZ & A2SZ & A1SZ & A0SZ & \\
\hline H'FFFF8622 & BCR2 & IW31 & IW30 & IW21 & IW20 & IW11 & IW10 & IW01 & IW00 & \\
\hline H'FFFF8623 & & CW3 & CW2 & CW1 & cwo & SW3 & SW2 & SW1 & swo & \\
\hline H'FFFF8624 & WCR1 & W33 & W32 & W31 & W30 & W23 & W22 & W21 & W20 & \\
\hline H'FFFF8625 & & W13 & W12 & W11 & W10 & W03 & W02 & W01 & woo & \\
\hline H'FFFF8626 & WCR2 & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8627 & & - & - & DDW1 & DDW0 & DSW3 & DSW2 & DSW1 & DSW0 & \\
\hline H'FFFF8628 & RAMER & - & - & - & - & - & - & - & - & \[
\begin{aligned}
& \text { FLASH } \\
& \text { (F-ZTAT }
\end{aligned}
\] \\
\hline H'FFFF8629 & & - & - & - & - & - & RAMS & RAM1 & RAMO & \begin{tabular}{l}
version \\
only)
\end{tabular} \\
\hline H'FFFF862A & DCR & TPC & RCD & TRAS1 & TRAS0 & DWW1 & DWW0 & DWR1 & DWR0 & BSC \\
\hline H'FFFF862B & & DIW & - & BE & RASD & SZ1 & SZ0 & AMX1 & AMX0 & \\
\hline H'FFFF862C & RTCSR & - & - & - & - & - & - & - & - & \\
\hline H'FFFF862D & & - & CMF & CMIE & CKS2 & CKS1 & CKS0 & RFSH & RMD & \\
\hline H'FFFF862E & RTCNT & - & - & - & - & - & - & - & - & \\
\hline \multicolumn{10}{|l|}{H'FFFF862F} & \\
\hline H'FFFF8630 & RTCOR & - & - & - & - & - & - & - & - & \\
\hline \multicolumn{11}{|l|}{H'FFFF8631} \\
\hline \[
\begin{aligned}
\hline \text { Notes: } & * 1 \\
& * 2
\end{aligned}
\] & Write ad Read ad Watchdo & dress. dress. g Time & or deta (WDT) & see sect & ion 13.2 & 4, Regi & Acce & in se & n 13, & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{\begin{tabular}{l}
Register \\
Abbr.
\end{tabular}} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline \begin{tabular}{l}
H'FFFF8632 \\
to \\
H'FFFF86AF
\end{tabular} & - & - & - & - & - & - & - & - & - & BSC \\
\hline H'FFFFF86B0 & \multirow[t]{2}{*}{DMAOR} & - & - & - & - & - & - & PR1 & PR0 & \multirow[t]{3}{*}{DMAC} \\
\hline H'FFFF86B1 & & - & - & - & - & - & AE & NMIF & DME & \\
\hline H'FFFF86B2 to H'FFFF86BF & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86C0 & \multirow[t]{4}{*}{SARO} & & & & & & & & & \\
\hline H'FFFF86C1 & & & & & & & & & & \\
\hline H'FFFF86C2 & & & & & & & & & & \\
\hline H'FFFF86C3 & & & & & & & & & & \\
\hline H'FFFF86C4 & \multirow[t]{4}{*}{DARO} & & & & & & & & & \\
\hline H'FFFF86C5 & & & & & & & & & & \\
\hline H'FFFF86C6 & & & & & & & & & & \\
\hline H'FFFF86C7 & & & & & & & & & & \\
\hline H'FFFF86C8 & \multirow[t]{4}{*}{DMATCRO} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86C9 & & & & & & & & & & \\
\hline H'FFFF86CA & & & & & & & & & & \\
\hline H'FFFF86CB & & & & & & & & & & \\
\hline H'FFFF86CC & \multirow[t]{4}{*}{CHCRO} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86CD & & - & - & - & DI & RO & RL & AM & AL & \\
\hline H'FFFF86CE & & DM1 & DM0 & SM1 & SM0 & RS3 & RS2 & RS1 & RS0 & \\
\hline H'FFFF86CF & & - & DS & TM & TS1 & TS0 & IE & TE & DE & \\
\hline H'FFFF86D0 & \multirow[t]{4}{*}{SAR1} & & & & & & & & & \\
\hline H'FFFF86D1 & & & & & & & & & & \\
\hline H'FFFF86D2 & & & & & & & & & & \\
\hline H'FFFF86D3 & & & & & & & & & & \\
\hline H'FFFF86D4 & \multirow[t]{4}{*}{DAR1} & & & & & & & & & \\
\hline H'FFFF86D5 & & & & & & & & & & \\
\hline H'FFFF86D6 & & & & & & & & & & \\
\hline H'FFFF86D7 & & & & & & & & & & \\
\hline H'FFFF86D8 & \multirow[t]{4}{*}{DMATCR1} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86D9 & & & & & & & & & & \\
\hline H'FFFF86DA & & & & & & & & & & \\
\hline H'FFFF86DB & & & & & & & & & & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{Register Abbr.} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF86DC & \multirow[t]{4}{*}{CHCR1} & - & - & - & - & - & - & - & - & \multirow[t]{4}{*}{DMAC} \\
\hline H'FFFF86DD & & - & - & - & DI & RO & RL & AM & AL & \\
\hline H'FFFF86DE & & DM1 & DM0 & SM1 & SM0 & RS3 & RS2 & RS1 & RS0 & \\
\hline H'FFFF86DF & & - & DS & TM & TS1 & TS0 & IE & TE & DE & \\
\hline H'FFFF86E0 & \multirow[t]{4}{*}{SAR2} & & & & & & & & & \\
\hline H'FFFF86E1 & & & & & & & & & & \\
\hline H'FFFF86E2 & & & & & & & & & & \\
\hline H'FFFF86E3 & & & & & & & & & & \\
\hline H'FFFF86E4 & \multirow[t]{4}{*}{DAR2} & & & & & & & & & \\
\hline H'FFFF86E5 & & & & & & & & & & \\
\hline H'FFFF86E6 & & & & & & & & & & \\
\hline H'FFFF86E7 & & & & & & & & & & \\
\hline H'FFFF86E8 & \multirow[t]{4}{*}{DMATCR2} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86E9 & & & & & & & & & & \\
\hline H'FFFF86EA & & & & & & & & & & \\
\hline H'FFFF86EB & & & & & & & & & & \\
\hline H'FFFF86EC & \multirow[t]{4}{*}{CHCR2} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86ED & & - & - & - & DI & RO & RL & AM & AL & \\
\hline H'FFFF86EE & & DM1 & DM0 & SM1 & SM0 & RS3 & RS2 & RS1 & RS0 & \\
\hline H'FFFF86EF & & - & DS & TM & TS1 & TS0 & IE & TE & DE & \\
\hline H'FFFF86F0 & \multirow[t]{4}{*}{SAR3} & & & & & & & & & \\
\hline H'FFFF86F1 & & & & & & & & & & \\
\hline H'FFFF86F2 & & & & & & & & & & \\
\hline H'FFFF86F3 & & & & & & & & & & \\
\hline H'FFFF86F4 & \multirow[t]{4}{*}{DAR3} & & & & & & & & & \\
\hline H'FFFF86F5 & & & & & & & & & & \\
\hline H'FFFF86F6 & & & & & & & & & & \\
\hline H'FFFF86F7 & & & & & & & & & & \\
\hline H'FFFF86F8 & \multirow[t]{4}{*}{DMATCR3} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86F9 & & & & & & & & & & \\
\hline H'FFFF86FA & & & & & & & & & & \\
\hline H'FFFF86FB & & & & & & & & & & \\
\hline H'FFFF86FC & \multirow[t]{4}{*}{CHCR3} & - & - & - & - & - & - & - & - & \\
\hline H'FFFF86FD & & - & - & - & DI & RO & RL & AM & AL & \\
\hline H'FFFF86FE & & DM1 & DM0 & SM1 & SM0 & RS3 & RS2 & RS1 & RS0 & \\
\hline H'FFFF86FF & & - & DS & TM & TS1 & TS0 & IE & TE & DE & \\
\hline
\end{tabular}

Table A. 1 On-Chip I/O Register Addresses (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Address} & \multirow[t]{2}{*}{\begin{tabular}{l}
Register \\
Abbr.
\end{tabular}} & \multicolumn{8}{|c|}{Bit Names} & \multirow[b]{2}{*}{Module} \\
\hline & & Bit 7 & Bit 6 & Bit 5 & Bit 4 & Bit 3 & Bit 2 & Bit 1 & Bit 0 & \\
\hline H'FFFF8700 & DTEA & DTE7 & DTE6 & DTE5 & DTE4 & DTE3 & DTE2 & DTE1 & DTE0 & DTC \\
\hline H'FFFF8701 & DTEB & DTE7 & DTE6 & DTE5 & DTE4 & DTE3 & DTE2 & DTE1 & DTE0 & \\
\hline H'FFFF8702 & DTEC & DTE7 & DTE6 & DTE5 & DTE4 & DTE3 & DTE2 & DTE1 & DTE0 & \\
\hline H'FFFF8703 & DTED & DTE7 & DTE6 & DTE5 & DTE4 & DTE3 & DTE2 & DTE1 & DTE0 & \\
\hline H'FFFF8704 & DTEE & DTE7 & DTE6 & DTE5 & DTE4 & DTE3 & DTE2 & DTE1 & DTE0 & \\
\hline H'FFFF8705 & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8706 & DTCSR & - & - & - & - & - & NMIF & AE & SWDTE & \\
\hline H'FFFF8707 & & DTVEC7 & DTVEC6 & DTVEC5 & DTVEC4 & DTVEC3 & DTVEC2 & DTVEC1 & DTVEC0 & \\
\hline H'FFFF8708 & DTBR & & & & & & & & & \\
\hline H'FFFF8709 & & & & & & & & & & \\
\hline H'FFFF870A to H'FFFF873F & - & - & - & - & - & - & - & - & - & \\
\hline H'FFFF8740 & CCR & - & - & - & - & - & - & - & - & CAC \\
\hline H'FFFF8741 & & - & - & - & CEDRAM & CECS3 & CECS2 & CECS1 & CECSO & \\
\hline \begin{tabular}{l}
H'FFFF8742 \\
to \\
H'FFFF87FF
\end{tabular} & - & - & - & - & - & - & - & - & - & \\
\hline
\end{tabular}

\section*{Appendix B Block Diagrams}


Figure B. 1 PAn/RXDm Block Diagram


Figure B. 2 PAn/SCKm \(/ \overline{\text { DREQm }} / \overline{\text { IRQm }}\) Block Diagram


Figure B. 3 PA6/TCLKA/ \(\overline{\text { CS2 }}\), PA7/TCLKB/ \(/ \overline{\text { CS3 }}\) (ZTAT, Mask) Block Diagram


Figure B. 4 PA7/TCLKB/CS3 Block Diagram (F-ZTAT Version)


Figure B. 5 PAn/TCLKm \(/ \overline{\overline{\text { IRQx }}}\) Block Diagram (ZTAT, Mask)


Figure B. 6 PAn/TCLKm/ \(\overline{\text { IRQx }}\) Block Diagram (F-ZTAT Version)


Figure B. 7 PAn/TXDm Block Diagram


Figure B. 8 PA15/CK Block Diagram


Figure B. 9 PA18/DRAK0/BREQ Block Diagram


Figure B. 10 PA19/DRAQ1/BACK Block Diagram


Figure B. 11 PAn/XXX Block Diagram


Figure B. 12 PAn/XXXX Block Diagram


Figure B. 13 PA17/WAIT Block Diagram


Figure B. 14 PB0/A16 Block Diagram


Figure B. 15 PB0/A16 Block Diagram (F-ZTAT Version)


Figure B. 16 PB1/A17 Block Diagram


Figure B. 17 PB6/IRQ4 \(/\) A18/BACK Block Diagram


Figure B. 18 PBn \(/ \overline{\text { IRQm }} / \overline{\text { POEm }} / \overline{\text { CASx }}\) Block Diagram


Figure B. 19 PB4/ㄷRQ2/POE2 \(/ \overline{\mathrm{CASH}}, \mathrm{PB} 3 / \overline{\mathrm{IRQ}} / \overline{\mathrm{POE} 1} / \overline{\mathrm{CASL}}\)
Block Diagram (F-ZTAT Version)


Figure B. 20 PBn \(/ \overline{\text { IRQm }} / \mathbf{X X X} /\) YYY Block Diagram


Figure B. 21 PBn/IRQm/XXXX/YYYY Block Diagram


Figure B. 22 PCn/An Block Diagram


Figure B. 23 PCn/An Block Diagram (F-ZTAT Version)


Figure B. 24 PDn/Dn Block Diagram


Figure B. 25 PDn/Dn Block Diagram (F-ZTAT Version)
\[
\begin{aligned}
& n=16-23 \\
& m=0-7 \\
& \text { PDR: Port D read signal } \\
& \text { PDW: Port D write signal }
\end{aligned}
\]

RES: Reset signal
Dout: Data output timing signal
Din: Data bus input timing signal

Figure B. 26 PDn/Dn/IRQm Block Diagram ( \(\mathbf{n}=16\)-23)


Figure B. 27 PDn/Dn/ \(\overline{\text { DREQm }}\) Block Diagram \((n=24,25)\)


Figure B. 28 PDn/Dn/DACKm Block Diagram ( \(\mathbf{n}=\mathbf{2 6}\), 27)


Figure B. 29 PDn/Dn/ \(\overline{\text { CSm }}\) Block Diagram \((\mathbf{n}=28,29)\)


Figure B. 30 PDn/Dn/IRQOUT Block Diagram ( \(\mathbf{n}=\mathbf{3 0}\) )


Figure B. 31 PD31/D31/ADTRG Block Diagram


Figure B. 32 PDn/Dn Block Diagram


Figure B. 33 PE13/TIOC4B/MRES Block Diagram


Figure B. 34 PE14/TIOC4C/DACK0/AH Block Diagram


Figure B. 35 PEn/TIOC4D/DACK1//(IRQOUT Block Diagram


Figure B. 36 PEn/TIOCXX Block Diagram


Figure B. 37 PEn/TIOCXX/DRAKm Block Diagram


Figure B. 38 PEn/TIOCXX/DREQm Block Diagram


Figure B. 39 PFn/ANn Block Diagram

\section*{Appendix C Pin States}

Table C. 1 Pin Modes During Reset, Power-Down, and Bus Right Release Modes (144 Pin)
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|r|}{\multirow[b]{2}{*}{Pin Function}} & \multicolumn{5}{|c|}{Pin modes} & \multirow[b]{3}{*}{Standby in Bus Right Release} \\
\hline & & \multicolumn{2}{|c|}{Reset} & \multicolumn{2}{|l|}{Power-Down} & \multirow[t]{2}{*}{Bus Right Release} & \\
\hline Class & Pin Name & Pow & Manual & Stan & Sleep & & \\
\hline Clock & CK & \(\bigcirc\) & O & \(\mathrm{H}^{* 1}\) & O & \(\bigcirc\) & O \\
\hline System & \(\overline{R E S}\) & I & 1 & 1 & I & I & 1 \\
\hline control & \(\overline{\text { MRES }}\) & \(\mathrm{Z}^{* 4}\) & 1 & Z & I & I & Z \\
\hline & \(\overline{\text { WDTOVF }}\) & O*3 & O*3 & 0 & O & 0 & 0 \\
\hline & \(\overline{\text { BREQ }}\) & \(\mathrm{Z}^{* 4}\) & 1 & Z & I & I & I \\
\hline & \(\overline{\overline{B A C K}}\) & \(\mathrm{Z}^{* 4}\) & 0 & Z & 0 & L & L \\
\hline Interrupt & NMI & I & I & I & I & I & I \\
\hline & \(\overline{\overline{\text { RQ0-IRQ7 }}}\) & \(\mathrm{Z}^{* 4}\) & 1 & Z & 1 & 1 & Z \\
\hline & IRQOUT (PD30) & \(\mathrm{Z}^{* 4}\) & 0 & \(\mathrm{H}^{* 1}\) & H & 0 & \(\mathrm{H}^{* 1}\) \\
\hline & IRQOUT (PE15) & \(\mathrm{Z}^{* 4}\) & 0 & Z & H & O & Z \\
\hline Address bus & A0-A21 & O*2 & 0 & Z & O & Z & Z \\
\hline Data bus & D0-D31 & \(\mathrm{Z}^{* 4}\) & I/O & Z & I/O & Z & Z \\
\hline Bus & WAIT & \(\mathrm{Z}^{* 4}\) & 1 & Z & I & Z & Z \\
\hline control & RD/WR, \(\overline{\text { RAS }}\) & \(\mathrm{Z}^{* 4}\) & 0 & 0 & 0 & Z & Z \\
\hline & \[
\begin{aligned}
& \overline{\mathrm{CASH}}, \overline{\mathrm{CASL}}, \\
& \overline{\mathrm{CASLH}}, \overline{\mathrm{CASLL}}
\end{aligned}
\] & \(\mathrm{Z}^{* 4}\) & 0 & O & 0 & Z & Z \\
\hline & \(\overline{\overline{R D}}\) & H & 0 & Z & 0 & Z & Z \\
\hline & \(\overline{\overline{C S 0}}, \overline{\mathrm{CS} 1}\) & H & 0 & Z & 0 & Z & Z \\
\hline & \(\overline{\overline{\mathrm{CS} 2}, \overline{\mathrm{CS3}}}\) & \(\mathrm{Z}^{* 4}\) & 0 & Z & 0 & Z & Z \\
\hline & \(\overline{\text { WRHH, }} \overline{\text { WRHL }}\), WRH, WRL & H & 0 & Z & 0 & Z & Z \\
\hline & \(\overline{\overline{\text { AH }}}\) & \(\mathrm{Z}^{* 4}\) & 0 & Z & 0 & Z & Z \\
\hline DMAC & DACKO, DACK1 (PD26, PD27) & \(\mathrm{Z}^{* 4}\) & 0 & O*1 & 0 & 0 & O*1 \\
\hline & DACKO, DACK1 (PE14, PE15) & \(\mathrm{Z}^{* 4}\) & 0 & Z & 0 & 0 & Z \\
\hline & DRAK0, DRAK1 & \(\mathrm{Z}^{* 4}\) & 0 & O*1 & O & O & O*1 \\
\hline & DREQ0, \(\overline{\text { DREQ1 }}\) & \(\mathrm{Z}^{* 4}\) & I & Z & I & I & Z \\
\hline
\end{tabular}

Table C. 1 Pin Modes During Reset, Power-Down, and Bus Right Release Modes (144 Pin) (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|r|}{\multirow[b]{2}{*}{Pin Function}} & \multicolumn{5}{|c|}{Pin modes} & \multirow[b]{3}{*}{Standby in Bus Right Release} \\
\hline & & \multicolumn{2}{|c|}{Reset} & \multicolumn{2}{|l|}{Power-Down} & \multirow[t]{2}{*}{Bus Right Release} & \\
\hline Class & Pin Name & \multicolumn{2}{|l|}{Power-OnManual} & \multicolumn{2}{|l|}{Standby Sleep} & & \\
\hline \multirow[t]{3}{*}{MTU} & \[
\begin{aligned}
& \text { TIOC0A-TIOC0D, } \\
& \text { TIOC1A-TIOC1D, } \\
& \text { TIOC2A-TIOC2D, } \\
& \text { TIOC3A, TIOC3C }
\end{aligned}
\] & \(Z^{* 4}\) & I/O & \(\mathrm{K}^{* 1}\) & I/O & I/O & \(\mathrm{K}^{* 1}\) \\
\hline & \[
\begin{aligned}
& \text { TIOC3B,TIOC3D, } \\
& \text { TIOC4A-TIOC4D }
\end{aligned}
\] & \(z^{* 4}\) & I/O & Z & I/O & I/O & Z \\
\hline & TCLKA-TCLKD & \(\mathrm{Z}^{* 4}\) & I & Z & I & I & Z \\
\hline Port control & \(\overline{\text { POE0-- }} \overline{\text { POE3 }}\) & \(Z^{* 4}\) & I & Z & I & I & Z \\
\hline \multirow[t]{3}{*}{SCI} & SCK0-SCK1 & \(\mathrm{Z}^{* 4}\) & I/O & Z & I/O & I/O & Z \\
\hline & TXD0-TCD1 & \(Z^{* 4}\) & 0 & O*1 & 0 & 0 & O*1 \\
\hline & RXD0-RXD1 & Z*4 & 1 & Z & I & 1 & Z \\
\hline \multirow[t]{2}{*}{A/D converter} & \(\overline{\text { ADTRG }}\) & \(\mathrm{Z}^{* 4}\) & 1 & Z & 1 & 1 & Z \\
\hline & AN0-AN7 & Z & 1 & Z & I & 1 & Z \\
\hline \multirow[t]{7}{*}{I/O Port} & PA0-PA23 & \(Z^{* 4}\) & I/O & \(K^{* 1}\) & K & I/O & \(\mathrm{K}^{* 1}\) \\
\hline & PB0-PB9 & & & & & & \\
\hline & PC0-PC15 & & & & & & \\
\hline & PD0-PD31 & & & & & & \\
\hline & PE0-PE8,PE10 & & & & & & \\
\hline & PE9,PE11-PE15 & \(Z^{* 4}\) & I/O & Z & K & I/O & Z \\
\hline & PF0-PF17 & Z & 1 & Z & I & 1 & Z \\
\hline
\end{tabular}

Notes: 1. There are instances where bus right release and transition to software standby mode occur simultaneously due to the timing between \(\overline{\mathrm{BREQ}}\) and internal operations. In such cases, standby mode results, but the standby state may be different.
The initial pin states depend on the mode. See section 18, Pin Function Controller, for details.
2. I: Input, O: Output, H: High-level output, L: Low-level output, Z: High impedance, K: Input pin with high impedance, output pin mode maintained.
*1 If the standby control register port high-impedance bits are set to 1 , output pins become high impedance.
*2 A21-A18 will become input ports after power-on reset.
*3 Input in the SH7044/SH7045 F-ZTAT version.
*4 General use I/O ports PAn, PBn, PCn, PDn, and PEn, as well as pins multiplexed with them, are unstable during the \(\overline{\operatorname{RES}}\) setup time ( \(\mathrm{t}_{\text {RESS }}\) ) immediately after the \(\overline{\mathrm{RES}}\) pin goes to low level.

Table C. 2 Pin Modes During Reset, Power-Down, and Bus Right Release Modes (112 Pin, 120 Pin)


Table C. 2 Pin Modes During Reset, Power-Down, and Bus Right Release Modes (112 Pin, 120 Pin ) (cont)
\begin{tabular}{|c|c|c|c|c|c|}
\hline \multicolumn{2}{|r|}{\multirow[b]{2}{*}{Pin Function}} & \multicolumn{3}{|c|}{Pin modes} & \multirow[b]{3}{*}{Standby in Bus Right Release} \\
\hline & & Reset & Power-Down & \multirow[t]{2}{*}{Bus Right Release} & \\
\hline Class & Pin Name & Power-On Manual & Standby Sleep & & \\
\hline Port control & \(\overline{\text { POE0-- }}\) POE3 & \(Z^{* 4}\) I & Z I & I & Z \\
\hline SCI & SCK0-SCK1 & \(Z^{* 4} \quad \mathrm{I} / \mathrm{O}\) & Z I/O & I/O & Z \\
\hline & TXD0-TCD1 & \(\mathrm{Z}^{* 4}\) O & O*1 O & 0 & O*1 \\
\hline & RXD0-RXD1 & \(Z^{* 4}\) I & Z I & I & Z \\
\hline A/D & \(\overline{\text { ADTRG }}\) & \(Z^{* 4}\) I & Z I & I & Z \\
\hline converter & & & & & \\
\hline control & AN0-AN7 & Z I & Z I & 1 & Z \\
\hline I/O Port & PA0-PA15 & \(Z^{* 4} \quad \mathrm{I} / \mathrm{O}\) & \(\mathrm{K}^{* 1} \mathrm{~K}\) & I/O & \(\mathrm{K}^{* 1}\) \\
\hline & PB0-PB9 & & & & \\
\hline & PC0-PC15 & & & & \\
\hline & PD0-PD15 & & & & \\
\hline & PE0-PE8-PE10 & & & & \\
\hline & PE9,PE11-PE15 & \(Z^{* 4} \quad 1 / 0\) & Z K & 1/O & Z \\
\hline & PF0-PF7 & Z I & Z I & 1 & Z \\
\hline
\end{tabular}

Notes: 1. There are instances where bus right release and transition to software standby mode occur simultaneously due to the timing between \(\overline{\mathrm{BREQ}}\) and internal operations. In such cases, standby mode results, but the standby state may be different.
The initial pin states depend on the mode. See section 18, Pin Function Controller, for details.
2. I: Input, O: Output, H: High-level output, L: Low-level output, Z: High impedance, K: Input pin with high impedance, output pin mode maintained.
*1 If the standby control register port high-impedance bits are set to 1 , output pins become high impedance.
*2 A21-A18 will become input ports after power-on reset.
*3 Input in the SH7044/SH7045 F-ZTAT version.
*4 General use I/O ports PAn, PBn, PCn, PDn, and PEn, as well as pins multiplexed with them, are unstable during the \(\overline{\operatorname{RES}}\) setup time ( \(\mathrm{t}_{\text {RESS }}\) ) immediately after the \(\overline{\mathrm{RES}}\) pin goes to low level.

Table C. 3 Pin Settings for On-Chip Peripheral Modules
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{3}{*}{Pin Name} & & \multirow[b]{3}{*}{On-Chip ROM} & \multirow[b]{3}{*}{On-Chip RAM} & \multicolumn{4}{|c|}{On-Chip Peripheral Module} \\
\hline & & & & \multirow[b]{2}{*}{8-Bit Space} & \multicolumn{3}{|c|}{16-Bit Space} \\
\hline & & & & & Upper Byte & Lower Byte & Word/ Longword \\
\hline \(\overline{\mathrm{CSO}}-\overline{\mathrm{CS3}}\) & & H & H & H & H & H & H \\
\hline RAS \({ }^{*}\) & & H & H & H & H & H & H \\
\hline \(\overline{\mathrm{CASHH}}^{* 2}\) & & H & H & H & H & H & H \\
\hline \(\overline{\text { CASHL }}^{* 2}\) & & H & H & H & H & H & H \\
\hline \(\overline{\text { CASLH }}^{* 2}\) & & H & H & H & H & H & H \\
\hline \(\overline{\text { CASLL }}^{* 2}\) & & H & H & H & H & H & H \\
\hline RD/WR & & H & H & H & H & H & H \\
\hline \(\overline{\mathrm{AH}}\) & & L & L & L & L & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\mathrm{RD}}\)} & R & H & H & H & H & H & H \\
\hline & W & - & H & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHH }}\)} & R & H & H & H & H & H & H \\
\hline & W & - & H & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHL }}\)} & R & H & H & H & H & H & H \\
\hline & W & - & H & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRLH }}\)} & R & H & H & H & H & H & H \\
\hline & W & - & H & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRLL }}\)} & R & H & H & H & H & H & H \\
\hline & W & - & H & H & H & H & H \\
\hline A21-A0 & & Address & Address & Address & Address & Address & Address \\
\hline D31-D24 & & High-Z & High-Z & High-Z & High-Z & High-Z & High-Z \\
\hline D23-D16 & & High-Z & High-Z & High-Z & High-Z & High-Z & High-Z \\
\hline D15-D8 & & High-Z & High-Z & High-Z & High-Z & High-Z & High-Z \\
\hline D7-D0 & & High-Z & High-Z & High-Z & High-Z & High-Z & High-Z \\
\hline
\end{tabular}

Notes: R: Read, W: Write
*1 L asserted in RAS down state or refresh state.
*2 \(L\) asserted in refresh state.

Table C. 4 Pin Settings for Normal External Space
\begin{tabular}{|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Pin Name} & & \multirow[b]{2}{*}{8-Bit Space} & \multicolumn{3}{|c|}{16-Bit Space} \\
\hline & & & Upper Word & Lower Word & Word/Longword \\
\hline \(\overline{\text { CS0 }}\) - \(\overline{\mathrm{CS}} 3\) & & Valid & Valid & Valid & Valid \\
\hline RAS \({ }^{*}\) & & H & H & H & H \\
\hline \(\overline{\text { CASHH }}^{*}\) & & H & H & H & H \\
\hline \(\overline{\text { CASHL }}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASLH }}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASLL }}^{* 2}\) & & H & H & H & H \\
\hline RD/WR & & H & H & H & H \\
\hline \(\overline{\mathrm{AH}}\) & & L & L & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\mathrm{RD}}\)} & R & L & L & L & L \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHH }}\)} & R & H & H & H & H \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHL }}\)} & R & H & H & H & H \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRLH }}\)} & R & H & H & H & H \\
\hline & W & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRLL }}\)} & R & H & H & H & H \\
\hline & W & L & H & L & L \\
\hline A21-A0 & & Address & Address & Address & Address \\
\hline D31-D24 & & High-Z & High-Z & High-Z & High-Z \\
\hline D23-D16 & & High-Z & High-Z & High-Z & High-Z \\
\hline D15-D8 & & High-Z & Data & High-Z & Data \\
\hline D7-D0 & & Data & High-Z & Data & Data \\
\hline
\end{tabular}

Table C. 4 Pin Settings for Normal External Space (cont)
\begin{tabular}{|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Pin Name} & & \multicolumn{7}{|c|}{32-Bit Space} \\
\hline & & MSB & 2nd Byte & 3rd Byte & LSB & Upper Word & Lower Word & Longword \\
\hline \(\overline{\overline{C S 0}}-\overline{\mathrm{CS3}}\) & & Valid & Valid & Valid & Valid & Valid & Valid & Valid \\
\hline \(\overline{\text { RAS }}^{* 1}\) & & H & H & H & H & H & H & H \\
\hline \(\overline{\text { CASHH }}^{* 2}\) & & H & H & H & H & H & H & H \\
\hline \(\overline{\text { CASHL }}^{* 2}\) & & H & H & H & H & H & H & H \\
\hline \(\overline{\text { CASLH }}^{*}{ }^{2}\) & & H & H & H & H & H & H & H \\
\hline \(\overline{\text { CASLL }}^{* 2}\) & & H & H & H & H & H & H & H \\
\hline RD/WR & & H & H & H & H & H & H & H \\
\hline \(\overline{\mathrm{AH}}\) & & L & L & L & L & L & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\mathrm{RD}}\)} & R & L & L & L & L & L & L & L \\
\hline & W & H & H & H & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHH }}\)} & R & H & H & H & H & H & H & H \\
\hline & W & L & H & H & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R H L}}\)} & R & H & H & H & H & H & H & H \\
\hline & W & H & L & H & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R H}}\)} & R & H & H & H & H & H & H & H \\
\hline & W & H & H & L & H & H & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R L}}\)} & R & H & H & H & H & H & H & H \\
\hline & W & H & H & H & L & H & L & L \\
\hline A21-A0 & & Address & Address & Address & Address & Address & Address & Address \\
\hline D31-D24 & & Data & High-Z & High-Z & High-Z & Data & High-Z & Data \\
\hline D23-D16 & & High-Z & Data & High-Z & High-Z & Data & High-Z & Data \\
\hline D15-D8 & & High-Z & High-Z & Data & High-Z & High-Z & Data & Data \\
\hline D7-D0 & & High-Z & High-Z & High-Z & Data & High-Z & Data & Data \\
\hline
\end{tabular}

Notes: 1. R: Read, W: Write
2. Valid: Chip select signal corresponding with accessed area is low; chip select signal in other cases is high.
* 1 L asserted in RAS down mode or refresh mode.
*2 \(L\) asserted in refresh mode.

Table C. 5 Pin Settings for Multiplex I/O Space
16-Bit Space
\begin{tabular}{|c|c|c|c|c|c|}
\hline Pin Name & & 8-Bit Space & Upper Byte & Lower Byte & Word/Longword \\
\hline \(\overline{\overline{C S 0}} \overline{\text { CS2 }}\) & & H & H & H & H \\
\hline \(\overline{\text { CS3 }}\) & & L & L & L & L \\
\hline \(\overline{\mathrm{RAS}}^{* 1}\) & & H & H & H & H \\
\hline \(\overline{\mathrm{CASHH}}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASHL }}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASLH }}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASLL }}^{* 2}\) & & H & H & H & H \\
\hline RD/WR & & H & H & H & H \\
\hline \(\overline{\mathrm{AH}}\) & & Valid & Valid & Valid & Valid \\
\hline \multirow[t]{2}{*}{\(\overline{\mathrm{RD}}\)} & R & L & L & L & L \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHH }}\)} & R & H & H & H & H \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R H L}}\)} & R & H & H & H & H \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R H}}\)} & R & H & H & H & H \\
\hline & W & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R L}}\)} & R & H & H & H & H \\
\hline & W & L & H & L & L \\
\hline A21-A0 & & Address & Address & Address & Address \\
\hline D31-D24 & & High-Z & High-Z & High-Z & High-Z \\
\hline D23-D16 & & High-Z & High-Z & High-Z & High-Z \\
\hline D15-D8 & & High-Z & Address/Data & Address & Address/Data \\
\hline D7-D0 & & Address/Data & Address & Address/Data & Address/Data \\
\hline
\end{tabular}

Notes: 1. R: Read, W: Write
2. Valid: High output in accordance with \(\overline{\mathrm{AH}}\) timing.
*1 L asserted in RAS down mode or refresh mode.
*2 L asserted in refresh mode.

Table C. 6 Pin Settings for DRAM Space
\begin{tabular}{|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Pin Name} & & \multirow[b]{2}{*}{8-Bit Space} & \multicolumn{3}{|c|}{16-Bit Space} \\
\hline & & & Upper Word & Lower Word & Word/Longword \\
\hline \(\overline{\mathrm{CSO}}-\overline{\mathrm{CS3}}\) & & H & H & H & H \\
\hline \(\overline{\text { RAS }}^{* 1}\) & & Valid & Valid & Valid & Valid \\
\hline \(\overline{\mathrm{CASHH}}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASHL }}^{* 2}\) & & H & H & H & H \\
\hline \(\overline{\text { CASLH }}^{* 2}\) & & H & Valid & H & Valid \\
\hline \(\overline{\text { CASLL }}^{* 2}\) & & Valid & H & Valid & Valid \\
\hline \multirow[t]{2}{*}{RD/WR} & R & H & H & H & H \\
\hline & W & L & L & L & L \\
\hline \(\overline{\mathrm{AH}}\) & & L & L & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\mathrm{RD}}\)} & R & L & L & L & L \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHH }}\)} & R & H & H & H & H \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHL }}\)} & R & H & H & H & H \\
\hline & W & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRH }}\)} & R & H & H & H & H \\
\hline & W & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R L}}\)} & R & H & H & H & H \\
\hline & W & L & H & L & L \\
\hline A21-A0 & & Address & Address & Address & Address \\
\hline D31-D24 & & High-Z & High-Z & High-Z & High-Z \\
\hline D23-D16 & & High-Z & High-Z & High-Z & High-Z \\
\hline D15-D8 & & High-Z & Data & High-Z & Data \\
\hline D7-D0 & & Data & High-Z & Data & Data \\
\hline
\end{tabular}

Table C. 6 Pin Settings for DRAM Space (cont)
32-Bit Space
\begin{tabular}{|c|c|c|c|c|c|c|c|c|}
\hline \multirow[b]{2}{*}{Pin Name} & & \multicolumn{7}{|c|}{32-Bi Space} \\
\hline & & MSB & 2nd Byte & 3rd Byte & LSB & Upper Word & Lower Word & Longword \\
\hline \(\overline{\overline{C S O}} \overline{\mathrm{CS3}}\) & & H & H & H & H & H & H & H \\
\hline RAS \({ }^{* 1}\) & & Valid & Valid & Valid & Valid & Valid & Valid & Valid \\
\hline \(\overline{\mathrm{CASHH}}^{*}\) & & Valid & H & H & H & Valid & H & Valid \\
\hline \(\overline{\text { AASHL }}^{* 2}\) & & H & Valid & H & H & Valid & H & Valid \\
\hline \(\overline{\text { CASLH }}^{* 2}\) & & H & H & Valid & H & H & Valid & Valid \\
\hline \(\overline{\text { CASLL }}^{* 2}\) & & H & H & H & Valid & H & Valid & Valid \\
\hline \multirow[t]{2}{*}{RD/WR} & R & H & H & H & H & H & H & H \\
\hline & W & L & L & L & L & L & L & L \\
\hline \(\overline{\text { AH }}\) & & L & L & L & L & L & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{R D}}\)} & R & L & L & L & L & L & L & L \\
\hline & W & H & H & H & H & H & H & H \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHH }}\)} & R & H & H & H & H & H & H & H \\
\hline & W & L & H & H & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\text { WRHL }}\)} & R & H & H & H & H & H & H & H \\
\hline & W & H & L & H & H & L & H & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R H}}\)} & R & H & H & H & H & H & H & H \\
\hline & W & H & H & L & H & H & L & L \\
\hline \multirow[t]{2}{*}{\(\overline{\overline{W R L}}\)} & R & H & H & H & H & H & H & H \\
\hline & W & H & H & H & L & H & L & L \\
\hline A21-A0 & & Address & Address & Address & Address & Address & Address & Address \\
\hline D31-D24 & & Data & High-Z & High-Z & High-Z & Data & High-Z & Data \\
\hline D23-D16 & & High-Z & Data & High-Z & High-Z & Data & High-Z & Data \\
\hline D15-D8 & & High-Z & High-Z & Data & High-Z & High-Z & Data & Data \\
\hline D7-D0 & & High-Z & High-Z & High-Z & Data & High-Z & Data & Data \\
\hline
\end{tabular}

Notes: 1. R: Read, W: Write
2. Valid: Chip select signal corresponding with accessed area is low; chip select signal in other cases is high.
*1 Asserted in RAS down mode or refresh mode.
*2 Asserted in refresh mode.

\section*{Appendix D Notes when Converting the F-ZTAT Application Software to the Mask-ROM Versions}

Please note the following when converting the F-ZTAT application software to the mask-ROM versions.

The values read from the internal registers for the flash ROM of the mask-ROM version and F ZTAT version differ as follows.
\begin{tabular}{llll} 
& & \multicolumn{1}{l}{ Status } & \\
\cline { 3 - 4 } Register & Bit & F-ZTAT Version & Mask-ROM Version \\
\hline FLMCR1 & \begin{tabular}{l} 
0: Application software \\
running \\
1: Programming
\end{tabular} & \begin{tabular}{l} 
0: Is not read out \\
\end{tabular} & \begin{tabular}{l} 
1: Application software \\
running
\end{tabular} \\
\hline Note: \begin{tabular}{l} 
This difference applies to all the F-ZTAT versions and all the mask-ROM versions that have \\
different ROM size.
\end{tabular} &
\end{tabular}

\section*{Appendix E Product Code Lineup}

Table E. 1 SH7040, SH7041, SH7042, SH7043, SH7044, and SH7045 Product Lineup
\begin{tabular}{|c|c|c|c|c|c|c|}
\hline Product Type & & Mask Version & Product Code & Mark Code & Package & Order Model No.*2 \\
\hline \multirow[t]{10}{*}{SH7040A} & \multirow[t]{5}{*}{Mask ROM verion} & \multirow[t]{5}{*}{A MASK} & HD6437040AF28 & HD6437040A (***)F28 & QFP2020-112 & HD6437040A*** \\
\hline & & & HD6437040AVF16 & HD6437040A (***)VF16 & QFP2020-112 & HD6437040A*** \\
\hline & & & HD6437040AVX16 & HD6437040A(***)VX16 & TQFP1414-120 & HD6437040A***X \\
\hline & & & HD6437040ACF28 & HD6437040A (***)CF28 & QFP2020-112Cu*1 & HD6437040A***CF \\
\hline & & & HD6437040AVCF16 & HD6437040A(***)VCF16 & QFP2020-112Cu*1 & HD6437040A***CF \\
\hline & \multirow[t]{5}{*}{ROM less version} & \multirow[t]{5}{*}{A MASK} & HD6417040AF28 & HD6417040AF28 & QFP2020-112 & HD6417040AF28 \\
\hline & & & HD6417040AVF16 & HD6417040AVF16 & QFP2020-112 & HD6417040AVF16 \\
\hline & & & HD6417040AVX16 & HD6417040AVX16 & TQFP1414-120 & HD6417040AVX16 \\
\hline & & & HD6417040ACF28 & HD6417040ACF28 & QFP2020-112Cu*1 & HD6417040ACF28 \\
\hline & & & HD6417040AVCF16 & HD6417040AVCF16 & QFP2020-112Cu*1 & HD6417040AVCF16 \\
\hline \multirow[t]{8}{*}{SH7041A} & \multirow[t]{4}{*}{\begin{tabular}{l}
Mask ROM \\
version
\end{tabular}} & \multirow[t]{4}{*}{A MASK} & HD6437041AF28 & HD6437041A(***)F28 & QFP2020-144 & HD6437041A***F \\
\hline & & & HD6437041AVF16 & HD6437041A(***)VF16 & QFP2020-144 & HD6437041A***F \\
\hline & & & HD6437041ACF28 & HD6437041A (***)CF28 & QFP2020-144Cu*1 & HD6437041A***CF \\
\hline & & & HD6437041AVCF16 & HD6437041A(***)VCF16 & QFP2020-144Cu*1 & HD6437041A***CF \\
\hline & \multirow[t]{4}{*}{ROM less verion} & \multirow[t]{4}{*}{A MASK} & HD6417041AF28 & HD6417041AF28 & QFP2020-144 & HD6417041AF28 \\
\hline & & & HD6417041AVF16 & HD6417041AVF16 & QFP2020-144 & HD6417041AVF16 \\
\hline & & & HD6417041ACF28 & HD6417041ACF28 & QFP2020-144Cu*1 & HD6417041ACF28 \\
\hline & & & HD6417041AVCF16 & HD6417041AVCF16 & QFP2020-144Cu*1 & HD6417041AVCF16 \\
\hline \multirow[t]{4}{*}{SH7042} & Mask ROM & \multirow[t]{2}{*}{-} & HD6437042F28 & HD6437042 (***)F28 & QFP2020-112 & HD6437042***F \\
\hline & version & & HD6437042VF16 & HD6437042 (***)VF16 & QFP2020-112 & HD6437042***F \\
\hline & Z-TAT & \multirow[t]{2}{*}{-} & HD6477042F28 & HD6477042F28 & QFP2020-112 & HD6477042F28 \\
\hline & version & & HD6477042VF16 & HD6477042VF16 & QFP2020-112 & HD6477042VF16 \\
\hline \multirow[t]{5}{*}{SH7042A} & \multirow[t]{5}{*}{Mask ROM version} & \multirow[t]{5}{*}{A MASK} & HD6437042AF28 & HD6437042A (***)F28 & QFP2020-112 & HD6437042A**F \\
\hline & & & HD6437042AVF16 & HD6437042A(***)VF16 & QFP2020-112 & HD6437042A*** \\
\hline & & & HD6437042AVX16 & HD6437042A(***)VX16 & TQFP1414-120 & HD6437042A***X \\
\hline & & & HD6437042ACF28 & HD6437042A (***)CF28 & QFP2020-112Cu*1 & HD6437042A***CF \\
\hline & & & HD6437042AVCF16 & HD6437042A (***)VCF16 & QFP2020-112Cu*1 & HD6437042A***CF \\
\hline
\end{tabular}

Table E. 1 SH7040, SH7041, SH7042, SH7043, SH7044, and SH7045 Product Lineup (cont)
\begin{tabular}{|c|c|c|c|c|c|c|}
\hline Product Type & & \begin{tabular}{l}
Mask \\
Version
\end{tabular} & Product Code & Mark Code & Package & Order Model No.*2 \\
\hline \multirow[t]{5}{*}{SH7042A} & \multirow[t]{5}{*}{Z-TAT version} & \multirow[t]{5}{*}{A MASK} & HD6477042AF28 & HD6477042AF28 & QFP2020-112 & HD6477042AF28 \\
\hline & & & HD6477042AVF16 & HD6477042AVF16 & QFP2020-112 & HD6477042AVF16 \\
\hline & & & HD6477042AVX16 & HD6477042AVX16 & TQFP1414-120 & HD6477042AVX16 \\
\hline & & & HD6477042ACF28 & HD6477042ACF28 & QFP2020-112Cu*1 & HD6477042ACF28 \\
\hline & & & HD6477042AVCF16 & HD6477042AVCF16 & QFP2020-112Cu*1 & HD6477042AVCF16 \\
\hline \multirow[t]{4}{*}{SH7043} & Mask ROM & \multirow[t]{2}{*}{-} & HD6437043F28 & HD6437043(***)F28 & QFP2020-144 & HD6437043***F \\
\hline & version & & HD6437043VF16 & HD6437043(***)VF16 & QFP2020-144 & HD6437043***F \\
\hline & Z-TAT & \multirow[t]{2}{*}{-} & HD6477043F28 & HD6477043F28 & QFP2020-144 & HD6477043F28 \\
\hline & version & & HD6477043VF16 & HD6477043VF16 & QFP2020-144 & HD6477043VF16 \\
\hline \multirow[t]{8}{*}{SH7043A} & \multirow[t]{4}{*}{Mask ROM version} & \multirow[t]{4}{*}{A MASK} & HD6437043AF28 & HD6437043A(***)F28 & QFP2020-144 & HD6437043A***F \\
\hline & & & HD6437043AVF16 & HD6437043A(***)VF16 & QFP2020-144 & HD6437043A***F \\
\hline & & & HD6437043ACF28 & HD6437043A(***)CF28 & QFP2020-144Cu*1 & HD6437043A***CF \\
\hline & & & HD6437043AVCF16 & HD6437043A(***)VCF16 & QFP2020-144Cu*1 & HD6437043A***CF \\
\hline & \multirow[t]{4}{*}{\begin{tabular}{l}
Z-TAT \\
version
\end{tabular}} & \multirow[t]{4}{*}{A MASK} & HD6477043AF28 & HD6477043AF28 & QFP2020-144 & HD6477043AF28 \\
\hline & & & HD6477043AVF16 & HD6477043AVF16 & QFP2020-144 & HD6477043AVF16 \\
\hline & & & HD6477043ACF28 & HD6477043ACF28 & QFP2020-144Cu*1 & HD6477043ACF28 \\
\hline & & & HD6477043AVCF16 & HD6477043AVCF16 & QFP2020-144Cu*1 & HD6477043AVCF16 \\
\hline \multirow[t]{2}{*}{SH7044} & Mask ROM version & \multirow[t]{2}{*}{A MASK} & HD6437044F28 & HD6437044(***)F28 & QFP2020-112 & HD6437044***F \\
\hline & F-ZTAT version & & HD64F7044F28 & HD64F7044F28 & QFP2020-112 & HD64F7044F28 \\
\hline \multirow[t]{2}{*}{SH7045} & Mask ROM version & \multirow[t]{2}{*}{A MASK} & HD6437045F28 & HD6437045(***)F28 & QFP2020-144 & HD6437045***F \\
\hline & F-ZTAT version & & HD64F7045F28 & HD64F7045F28 & QFP2020-144 & HD64F7045F28 \\
\hline
\end{tabular}
(***) is the ROM code.
Notes: *1 Package with Copper used as the lead material.
*2 *** in the Order Model No. is the ROM code, consisting of a letter and a two-digit number (ex. E00). The letter indicates the voltage and frequency, as shown below.
- E, F, G, H: \(5.0 \mathrm{~V}, 28 \mathrm{MHz}\)
- P, Q, R: \(3.3 \mathrm{~V}, 16 \mathrm{MHz}\)

\section*{Appendix F Package Dimensions}

Package dimensions of the SH7040, SH7042, SH7044 (FP-112) are shown in figures F. 1 and F.2. Package dimensions of the SH7040, SH7042 (TFP-120) are shown in figure F.3.
Package dimensions of the SH7041, SH7043, SH7045 (FP-144) are shown in figures F. 4 and F. 5 .


Figure F. 1 Package Dimensions (FP-112)

This package (FP-112B) uses copper leads.


Figure F. 2 Package Dimensions (FP-112B)


Figure F. 3 Package Dimensions (TFP-120)


Figure F. 4 Package Dimensions (FP-144J)

This package (FP-144G) uses copper leads.


Figure F. 5 Package Dimensions (FP-144G)

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Publication Date: 1st Edition, February, 1997
Rev.6.00, May 26, 2003
Published by: Sales Strategic Planning Div.
Renesas Technology Corp.
Edited by: Technical Documentation \& Information Department Renesas Kodaira Semiconductor Co., Ltd.
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