

## **RF9904**

#### INTEGRATED TRANSCEIVER

#### **Typical Applications**

- Digital Communication Systems
- Spread Spectrum Communication Systems
   POS Terminals
- 915 MHz Cordless Phones

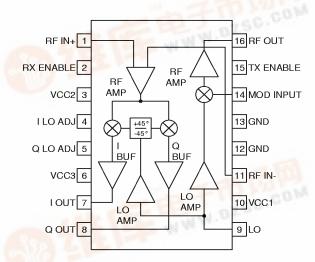
- 915 MHz ISM Band Systems
- Commercial Handheld Systems

### **Product Description**

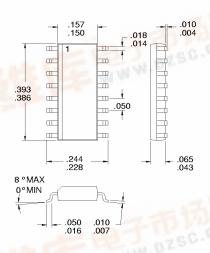
The RF9904 transceiver IC includes a transmit BPSK modulator, and a receive quadrature demodulator, as well as a splitter for the local oscillator signal, and separate transmit and receive LO buffer amplifiers. The IC operates over the 700 to 1100MHz frequency range, which accommodates 902 to 928MHz ISM band spread-spectrum systems as well as first IF stages for higher-frequency systems. The IC contains all circuits required for quadrature demodulation, using an integrated LO phase shift network, as well as all circuits for bi-phase modulation. Both modulation and demodulation are direct (at the carrier frequency), so only a single external LO source is needed. This chip is designed to be operated together with the RF2403 Front End IC.

#### **Optimum Technology Matching® Applied**

Si BJT ☐ GaAs HBT ☐ GaAs MESFET Si Bi-CMOS



**Functional Block Diagram** 



Package Style: SOP-16

#### **Features**

- Single 4.0V to 6.5V Power Supply
- Direct Modulation and Demodulation
- Separate RX and TX Power Down Mode
- DC to 15MHz Baseband Frequency
- 700 MHz to 1100 MHz RF Frequency
- Compatible With RF2403 Front End IC

#### **Ordering Information**

RF9904 Integrated Transceiver

RF9904 PCBA Fully Assembled Evaluation Board

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# RF9904

**Absolute Maximum Ratings** 

Parameter	Rating	Unit			
Supply Voltage	-0.5 to +7.0	$V_{DC}$			
RX/TX Enable Voltage	-0.5 to $V_{ m CC}$	V			
Input RF Power	+10	dBm			
Operating Ambient Temperature	-40 to +85	°C			
Storage Temperature	-40 to +150	°C			



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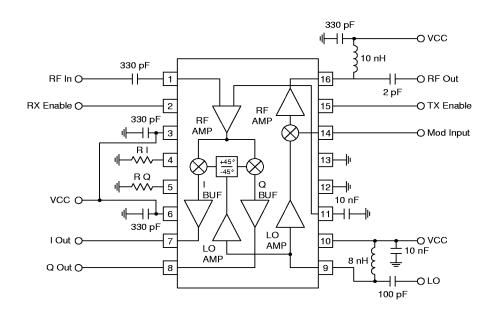
Daniel de la constant		Specification		Unit	Condition	
Parameter	Min.	Тур.	Typ. Max.			
Overall					T=25 °C, V <sub>CC</sub> =5.0 V, Freq=915MHz	
Frequency Range		700 to 1100		MHz		
LO Input Level	-5	О	+5	dBm		
LO Input VSWR			1.4:1			
Receiver Section					P <sub>IN</sub> =-30dBm, F <sub>IN</sub> =916MHz, P <sub>I O</sub> =-30dBm, F <sub>I O</sub> =915MHz	
Input VSWR			1.4:1			
Noise Figure		10		dB		
Voltage Gain	1	4	7	dB		
Input P <sub>1dB</sub>		-2		dBm		
Input IP <sub>3</sub>		+8		dBm		
Baseband Frequency		15		MHz	3 dB Bandwidth, Load=1MΩ  1pF	
I/Q Amplitude Error		±0.5	±1.0	dB		
I/Q Phase Error		±1	±5	0		
I/Q Output Impedance		50		Ω		
I/Q Output DC Level		3.5		v		
DC Offset Between I and Q		5		mV		
Max I/Q Output Level		1.0		V <sub>PP</sub>		
Transmitter Section					F <sub>MOD</sub> =100 kHz, V <sub>MOD</sub> =0.5 V <sub>PP</sub>	
Modulation Frequency		15		MHz		
Mod Input Impedance		6		kΩ		
Modulation Voltage		0.5		V <sub>PP</sub>		
Allowed DC at Mod Input	-1.0		1.0	l v		
RF Output Power	-11	-10		dBm		
Carrier Suppression	-26			dBc		
Image Suppression		-22		dBc		
Second Harmonic of RF				dBc		
Output Impedance				Ω		
Power Down Control						
"OFF" Voltage	1		0.5	V	Voltage supplied to the input; device is "OFF	
"ON" Voltage	V <sub>CC</sub> -1			\	Voltage supplied to the input; device is "ON"	
"OFF" to "ON" time			100	ns		
Power Supply						
Operating Voltage	4.0		6.5	V		
RX Current Consumption	20	25	35	mA		
TX Current Consumption	7	10	20	mA		
Power Down Current	1		200	μΑ	V <sub>CC</sub> =5.0 V, RX ENABLE=TX ENABLE=0 V	

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Pin	Function	Description	Interface Schematic
1	RF IN+	$50\Omega$ RF input for the receiver. DC voltage at this point is about 3 V, and an external DC blocking capacitor is required.	RF IN+ O
2	RX ENABLE	Enables the RX circuits (RX LO amplifiers, demodulators, buffer amplifiers, and RF amplifier) when high.	RX ENABLE O
3	VCC2	Power supply pin for the TX output stage, I LO amplifier, and I demodulator. An external RF decoupling capacitor should be placed as close to the pin as possible.	
4	I LO ADJ	This pin is used to adjust the balance of the bias for the I LO amplifier by adjusting the value of the external resistor (Ri) to ground. Typical values are between $0\Omega$ and $200\Omega$ . The optimum value for a given application is dependent on the LO frequency, LO level, and power supply voltage. For a given operating point an optimum value can be found which results in minimum phase and amplitude error, and a minimum sensitivity of the LO level. In other words, optimizing the value for minimum phase and amplitude error alone is not enough, since a value may be found which makes the phase and/or amplitude error depend heavily on LO level; thus a small change would result in a large change of phase/amplitude error. There are combinations of $\rm R_I$ and $\rm R_Q$ that result in low amplitude and phase error, as well as maximum tolerance for LO level and frequency changes. Some values are given in the application schematic as a starting point.	LO ADJ O
5	Q LO ADJ	Same as pin 4, for the Q mixer.	See pin 4.
6	VCC3	Power supply pin for the second TX LO amplifier stage, modulator, Q LO amplifier and Q demodulator. An external RF decoupling capacitor should be placed as close to the pin as possible.	
7	IOUT	Baseband output of the I demodulator. This is a push-pull output, with a $50\Omega$ impedance. The DC level at this point is about 3.5 V when running of a 5 V power supply.	$V_{CC}$ $25 \Omega$ $25 \Omega$ $25 \Omega$ $25 \Omega$ $25 \Omega$
8	Q OUT	Baseband output of the Q demodulator.	See pin 7.
			L '

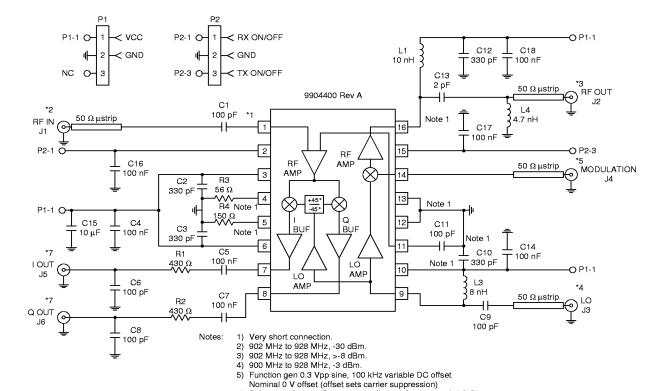
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**Application Schematic** 

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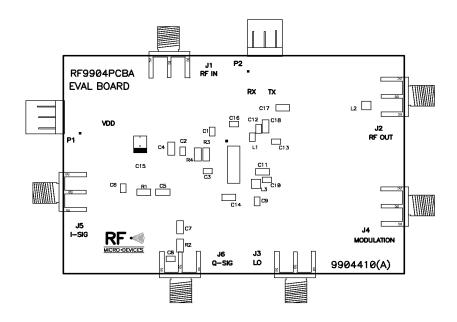


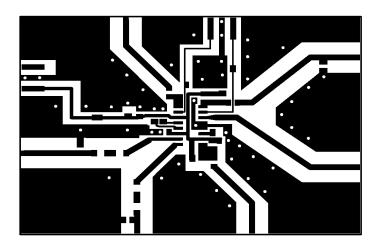
6) RX meas: Gain 1 dB comp noise figure I/Q amp match I/Q Phase error
7) TX meas: Prep RF output carrier suppression mod input for 20 dB mod harmonic suppression.
8) 90 deg phase between I and Q (offset RF from LO by 100 kHz)

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# **Evaluation Board Layout**





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