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PRELIMINARY

california micro devices

## CM3202

## DDR V<sub>DDQ</sub> and Termination Voltage Regulator

### **Features**

- Two linear regulators -Maximum 2A current from VDDQ -Source and sink up to 2A V<sub>TT</sub> current
- 1.7V to 2.8V adjustable V<sub>DDQ</sub> output voltage
- 500mV typical  $V_{DDQ}$  dropout voltage at 2A
- V<sub>TT</sub> tracking at 50% of V<sub>DDO</sub>
- Excellent load and line regulation, low noise
- Fast transient response
- Meet JEDEC DDR-I and DDR-II memory power spec.
- Linear regulator design requires no inductors and has low external component count
- Integrated power MOSFETs
- Dual purpose ADJ/Shutdown pin
- Built-in over-current limit with short-circuit foldback and thermal shutdown for  $V_{DDQ}$  and  $V_{TT}$
- Fast transient response
- 5mA quiescent current
- TDFN-8 and SOIC-8 packages for high performance thermal dissipation and easy PC board layout
- Optional RoHS Compliant Lead-free packaging

## Applications

- DDR memory and active termination buses
- **Desktop Computers, Servers**
- Residential and Enterprise Gateways
- **DSL Modems**
- **Routers and Switchers**
- **DVD** recorders
- 3D AGP cards
- LCD TV and STB

### **Typical Application**

## **Product Description**

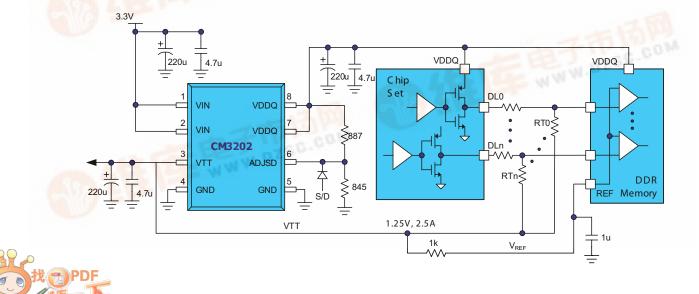
The CM3202 is a dual-output low noise linear regulator designed to meet SSTL-2 and SSTL-3 specifications for DDR-SDRAM VDDQ supply and termination voltage V<sub>TT</sub> supply. With integrated power MOSFET's, the CM3202 can source up to 2A of V<sub>DDQ</sub> current, and source or sink up to 2A V<sub>TT</sub> current. The typical dropout voltage for V<sub>DDQ</sub> is 500 mV at 2A load current.

The CM3202 provides fast response to transient load changes. Load regulation is excellent, less than 1%, from no load to full load. It also has built-in over-current limits and thermal shutdown at 170°C.

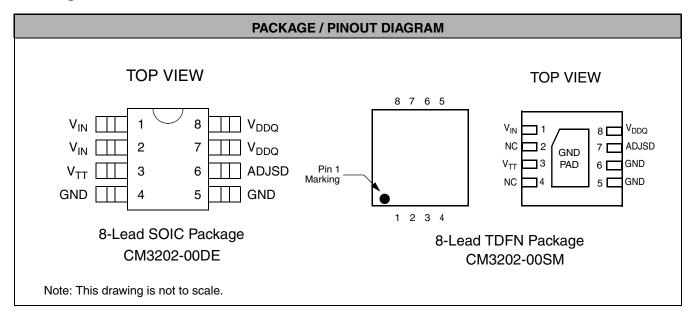
The CM3202 is packaged in an easy-to-use TDFN-8 and SOIC-8. Low thermal resistance (55°C/W) allows it to withstand 1.55W (1) dissipation at 85°C ambient. It can operate over the industrial ambient temperature range of -40°C to 85°C.

Note<sup>(1)</sup>:

If TDFN-8 is mounted on a double-sided printed circuit board with two square inches of copper area, it can to withstand 2W dissipation WWW.DZSC.COM at 85°C ambient then.



### **Package Pinout**



## **Ordering Information**

PART NUMBERING INFORMATION					
		Lead-free Finish			
Pins	Package	Ordering Part Number <sup>1</sup>	Part Marking		
8	TDFN	CM3202-00DE	CM3202		
8	SOIC	CM3202-00SM	CM3202		

Note 1: Parts are shipped in Tape & Reel form unless otherwise specified.

## **Specifications**

ABSOLUTE MAXIMUM RATINGS						
PARAMETER	RATING	UNITS				
VIN to GND	[GND - 0.3] to +6.0	V				
Pin Voltages V <sub>DDQ</sub> ,V <sub>TT</sub> to GND ADJSD to GND	[GND - 0.3] to +6.0 [GND - 0.3] to +6.0	V V				
Storage Temperature Range	-65 to +150	°C				
Operating Temperature Range	-40 to +85	°C				
Lead Temperature (Soldering, 10s)	300	°C				

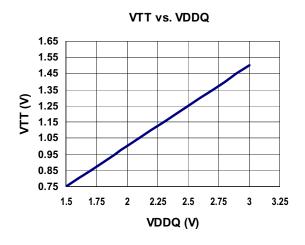
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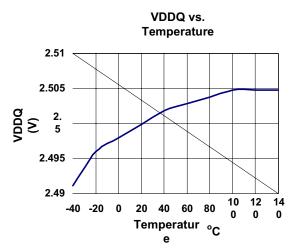
## Specifications (cont'd)

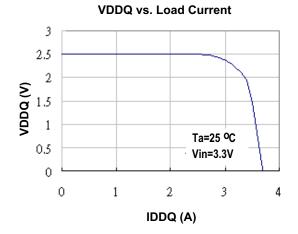
$v_{\rm INI} = 3.3 V$ typics	<b>ELECTRICAL OPERA</b> al values are at $T_{A} = 25^{\circ}C$ (upless of					
SYMBOL	al values are at T <sub>A</sub> = 25°C (unless otherwise specified)   PARAMETER   CONDITIONS			ТҮР	МАХ	UNI
VIN						
V <sub>IN</sub>	Supply Voltage Range		3	3.3	3.6	V
V <sub>UVLO</sub>	Under-voltage Lockout	All outputs are no load		2.5		V
	UVLO Hysterisis			200		m∖
Ι <sub>Q</sub>	Quiescent Current	$V_{DDQ} = 0V, V_{TT} = 0V,$ ADJSD = 3.3V (shutdown)		200		mA
		$V_{DDQ} = 2.5V, V_{TT} = 1.25V$ , (no load)		5		mA
V <sub>DDQ</sub> Regulator					1	
	Output Current Limit	V <sub>OUT</sub> = 2.5V		2.8		A
V <sub>REF</sub>	Reference Voltage		1.235	1.250	1.265	V
I <sub>BIAS</sub>	Input Bias Current (I <sub>ADJ</sub> )	VADJSD = V <sub>REF</sub>		30	200	nA
V <sub>R LOAD</sub>	Load Regulation	$I_0 = 10$ mA to 2A		1		%
V <sub>R LINE</sub>	Line Regulation	$V_{IN} = 3.15V$ to 3.5V, $I_O = 10mA$		1		%
V <sub>DROPOUT</sub>	Dropout Voltage	V <sub>IN</sub> = 3.15V, I <sub>O</sub> = 2A		500		m∖
V <sub>TT</sub> Regulator	1					<u>I</u>
	Output Current Limit (Source)	V <sub>OUT</sub> = 1.25V		2.8		A
	Output Current Limit (Sink)	V <sub>OUT</sub> = 1.25V		2.8		A
V <sub>R VTTLOAD</sub>	Load Regulation	$I_{O} = 0A \text{ to } 2A$		1		%
		$I_0 = 0A \text{ to } -2A$		1		%
Over Temperatu	Ire Protection					L
-	Thermal Shutdown Temperature			170		°C
	Thermal Shutdown Hysteresis			50		°C

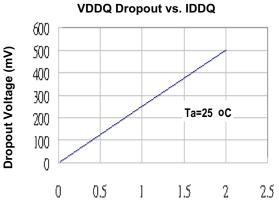
## CM3202

## **Typical Operating Characteristics**

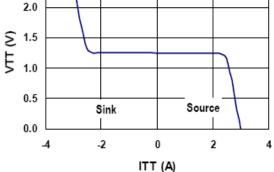






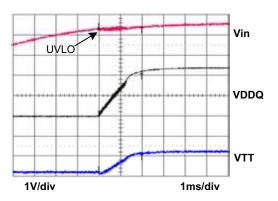


VTT vs. Load Current





IDDQ (A)



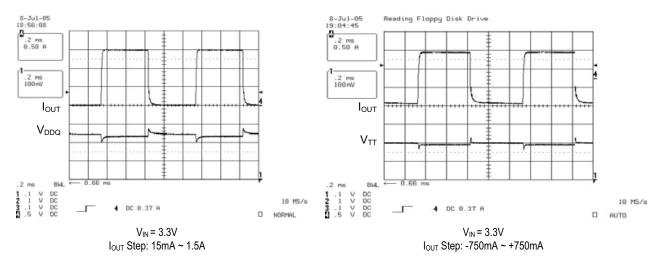
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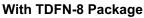
## Typical Operating Characteristics

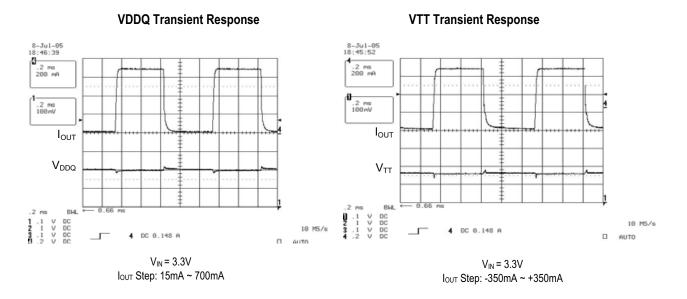
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VTT Transient Response

#### VDDQ Transient Response







#### With SOIC-8 Package

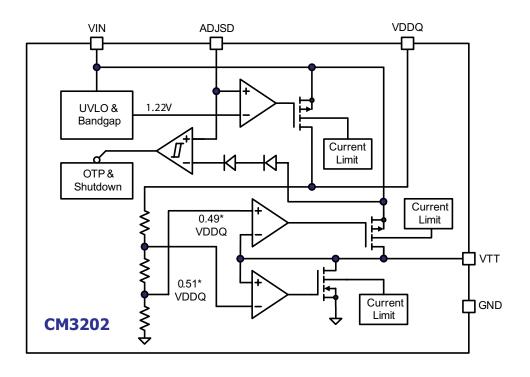
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## Pin Description Pin Descriptions

PIN DESCRIPTIONS					
PIN(S) TDFN 8	PIN(S) SOIC 8	NAME	DESCRIPTION		
1	1, 2	V <sub>IN</sub>	Input voltage pin, typically 3.3V from the silver box.		
2, 4		NC			
3	3	$V_{TT}$	$V_{\text{TT}}$ regulator output voltage pin, which is preset to 50% of $V_{\text{DDQ}}$ .		
5, 6	4, 5	GND	Ground pin. The back tab is also ground and serves as the package heatsink. It should be soldered to the circuit board copper to remove excess heat from the IC.		
7	6	ADJSD	This pin is for $V_{DDQ}$ output voltage Adjustment. The $V_{DDQ}$ output voltage is set using an external resistor divider connected to ADJSD. The output voltage is determined by the following formula $V_{DDQ} = 1.25V \times \frac{R1 + R2}{R1}$ where R1 is the ground-side resistor and R2 is the upper resistor of the divider. Connect these resistors to the $V_{DDQ}$ output at the point of regulation. In addition, functions as a Shutdown pin. Apply a voltage higher than $V_{IN}$ - 0.6V to this pin to simultaneously shutdown both $V_{DDQ}$ and $V_{TT}$ outputs. The outputs are restored when the voltage is lowered below $V_{IN}$ -0.6V. A low-leakage diode in series with the Shutdown signal is recommended to avoid interference with the voltage adjustment setting.		
8	7, 8	V <sub>DDQ</sub>	V <sub>DDQ</sub> regulator output voltage pin.		

### CM3202

## **Application Information (cont'd)**



## **Application Information**

#### Powering DDR Memory

Double-Data-Rate (DDR) memory has provided a huge step in performance for personal computers, servers and graphic systems. As is apparent in its name, DDR operates at double the data rate of earlier RAM, with two memory accesses per cycle versus one. DDR SDRAM's transmit data at both the rising falling edges of the memory bus clock.

DDR's use of Stub Series Terminated Logic (SSTL) topology improves noise immunity and power-supply rejection, while reducing power dissipation. To achieve this performance improvement, DDR requires more complex power management architecture than previous RAM technology.

Unlike the conventional DRAM technology, DDR SDRAM uses differential inputs and a reference voltage for all interface signals. This increases the data bus bandwidth, and lowers the system power consumption. Power consumption is reduced by lower operating voltage, a lower signal voltage swing associated with Stub Series Terminated Logic (SSTL\_2) and by the use of a termination voltage, V<sub>TT</sub>. SSTL\_2 is an

industry standard, defined in JEDEC document JESD8-9. SSTL\_2 maintains high-speed data bus signal integrity by reducing transmission reflections. JEDEC further defines the DDR SDRAM specification in JESD79C.

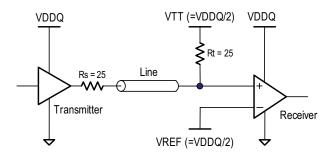
DDR memory requires three tightly regulated voltages:  $V_{DDQ}$ ,  $V_{TT}$ , and  $V_{REF}$  (see Figure 1). In a typical SSTL\_2 receiver, the higher current  $V_{DDQ}$  supply voltage is normally 2.5V with a tolerance of ±200 mV. The active bus termination voltage,  $V_{TT}$ , is half of  $V_{DDQ}$ .  $V_{REF}$  is a reference voltage that tracks half of  $V_{DDQ}$ , ± 1%, and is compared with the  $V_{TT}$  terminated signal at the receiver.  $V_{TT}$  must be within ±40 mV of  $V_{REF}$ 

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## Application Info (cont'd)



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Figure 1. Typical DDR terminations, Class II

The  $V_{TT}$  power requirement is proportional to the number of data lines and the resistance of the termination resistor, but does not vary with memory size. In a typical DDR data bus system each data line termination may momentarily consume 16.2 mA to achieve the 405 mV minimum over  $V_{TT}$  needed at the receiver:

$$I_{\text{terminaton}} = \frac{405 \text{mV}}{\text{Rt}(25\Omega)} = 16.2 \text{mA}$$

A typical 64 Mbyte SSTL-2 memory system, with 128 terminated lines, has a worst-case maximum V<sub>TT</sub> supply current up to  $\pm$  2.07A. However, a DDR memory system is dynamic, and the theoretical peak currents only occur for short durations, if they ever occur at all. These high current peaks can be handled by the V<sub>TT</sub> external capacitor. In a real memory system, the continuous average V<sub>TT</sub> current level in normal operation is less than  $\pm$  200 mA.

The  $V_{DDQ}$  power supply, in addition to supplying current to the memory banks, could also supply current to controllers and other circuitry. The current level typically stays within a range of 0.5A to 1A, with peaks up to 2A or more, depending on memory size and the computing operations being performed.

The tight tracking requirements and the need for  $V_{TT}$  to sink, as well as source, current provide unique challenges for powering DDR SDRAM.

#### CM3202 Regulator

The CM3202 dual output linear regulator provides all of the power requirements of DDR memory by combining two linear regulators into a single TDFN-8 or SOIC-8 package.  $V_{DDQ}$  regulator can supply up to 2A current,

and the two-quadrant V<sub>TT</sub> termination regulator has current sink and source capability to ±2A. The V<sub>DDQ</sub> linear regulator uses a PMOS pass element for a very low dropout voltage, typically 500mV at a 2A output. The output voltage of V<sub>DDQ</sub> can be set by an external voltage divider. The second output, V<sub>TT</sub>, is regulated at V<sub>DDQ</sub>/2 by an internal resistor divider. The V<sub>TT</sub> regulator can source, as well as sink, up to 2A current. The CM3202 is designed for optimal operation from a nominal 3.3VDC bus, but can work with VIN as high as 5V. When operating at higher VIN voltages, attention must be given to the increased package power dissipation and proportionally increased heat generation.

 $V_{REF}$  is typically routed to inputs with high impedance, such as a comparator, with little current draw. An adequate  $V_{REF}$  can be created with a simple voltage divider of precision, matched resistors from  $V_{DDQ}$  to ground. A small ceramic bypass capacitor can also be added for improved noise performance.

#### Input and Output Capacitors

The CM3202 requires that at least a 220 $\mu$ F electrolytic capacitor be located near the V<sub>IN</sub> pin for stability and to maintain the input bus voltage during load transients. An additional 4.7 $\mu$ F ceramic capacitor between the V<sub>IN</sub> and the GND, located as close as possible to those pins, is recommended to ensure stability.

A minimum of a 220 $\mu$ F electrolytic capacitor is recommended for the V<sub>DDQ</sub> output. An additional 4.7 $\mu$ F ceramic capacitor between the V<sub>DDQ</sub> and GND, located very close to those pins, is recommended.

A minimum of a 220 $\mu$ F, electrolytic capacitor is recommended for the V<sub>TT</sub> output. This capacitor should have low ESR to achieve best output transient response. SP or OSCON capacitors provide low ESR at high frequency, and thus are a good choice. In addition, place a 4.7 $\mu$ F ceramic capacitor between the V<sub>TT</sub> pin and GND, located very close to those pins. The total ESR must be low enough to keep the transient within the V<sub>TT</sub> window of 40mV during the transition for source to sink. An average current step of ± 0.5A requires:

$$\mathrm{ESR} < \frac{40\mathrm{mV}}{1\mathrm{A}} = 40\mathrm{m}\Omega$$

Both outputs will remain stable and in regulation even during light or no load conditions.

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## Application Info (cont'd)

#### Adjusting V<sub>DDQ</sub> Output Voltage

The CM3202 internal bandgap reference is set at 1.25V. The  $V_{DDQ}$  voltage is adjustable by using a resistor divider, R1 and R2:

$$V_{DDQ} = V_{ADJ} \times \frac{R1 + R2}{R1}$$

where  $V_{ADJ} = 1.25V (\pm 1\%)$ . For best regulator stability, we recommend that R1 and R2 not exceed  $10k\Omega$  each.

#### Shutdown

ADJSD also serves as a shutdown pin. When this is pulled high, > ( $V_{IN}$  - 0.6V), the  $V_{DDQ}$  output is turned off and both source and sink MOSFET's of the  $V_{TT}$  regulator are set to a high impedance state. During shutdown, the quiescent current is reduced to less than 3mA, independent of output load.

It is recommended that a 1N914 or equivalent low leakage diode be placed between ADJSD Pin and an external shutdown signal to prevent interference with the ADJ pin's normal operation. When the diode anode is pulled low, or left open, the CM3202 is again enabled.

#### Current Limit, Foldback and Over-temperature Protection

The CM3202 features internal current limiting with thermal protection. During normal operation,  $V_{DDQ}$  limits the output current to approximately 2A and  $V_{TT}$  limits the output current to approximately ±2A. When  $V_{TT}$  is current limiting into a hard short circuit, the output current folds back to a lower level, about 1A, until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the junction temperature of the device exceeds 170 °C (typical), the thermal protection circuitry triggers and shuts down both outputs. Once the junction temperature has cooled to below about 120 °C, the CM3202 returns to normal operation.

### **Thermal Considerations**

#### **Typical Thermal Characteristics**

The overall junction to ambient thermal resistance  $(\theta_{JA})$  for device power dissipation  $(P_D)$  primarily consists of two paths in the series. The first path is the junction to the case  $(\theta_{JC})$  which is defined by the package style and the second path is case to ambient  $(\theta_{CA})$  thermal resistance which is dependent on board layout. The final operating junction temperature for any condition can be estimated by the following thermal equation:

$$T_{JUNC} = T_{AMB} + P_D \times (\theta_{JC}) + P_D \times (\theta_{CA}) = T_{AMB} + P_D \times (\theta_{CA})$$

When a CM3202 is mounted on a double-sided printed circuit board with two square inches of copper allocated for "heat spreading," the  $\theta_{JA}$  is approximately 42.5 °C/Watt for the CM3202-00DE (TDFN-8) and 85 °C/Watt for CM3202-00SM (SOIC-8). Based on the over temperature limit of 170°C with an ambient of 85°C, the available power of the package will be:

$$P_{D(TDFN8)} = \frac{170 \text{ °C} \text{ 85°C}}{42.5 \text{ °C} / \text{W}} = 2\text{W}$$
$$P_{D(SOIC8)} = \frac{170 \text{ °C} \text{ 85°C}}{85 \text{ °C} / \text{W}} = 1\text{W}$$

#### **PCB Layout Considerations**

TheCM3202 has a heat spreader attached to the bottom of the TDFN-8 package in order for the heat to be transferred more easily from the package to the PCB. The heat spreader is a copper pad of dimensions just smaller than the package itself. By positioning the matching pad on the PCB top layer to connect to the spreader during the manufacturing, the heat will be transferred between the two pads. See the Figure 2, the CM3202-00DE (TDFN-8) and CM3202-00SM (SOIC-8) show the recommended PCB layout. Please be noted that there are six vias in the SOIC-8 package (four vias in the TDFN-8 package) on either side to allow the heat to dissipate into the ground and power planes on the inner layers of the PCB. Vias can be placed underneath the chip, but this can be resulted in

### CM3202

### **Application Info (cont'd)**

blocking of the solder. The ground and power planes need to be at least 2 square inches of copper by the vias. It also helps dissipation if the chip is positioned away from the edge of the PCB, and not near other heat-dissipating devices. A good thermal link from the PCB pad to the rest of the PCB will assure the best heat transfer from the CM3202-00DE (TDFN-8) to ambient,  $\theta_{JA}$ , of approximately 42.5 °C/W, or  $\theta_{JA}$  of approximately 85 °C/W for the CM3202-00SM (SOIC-8).

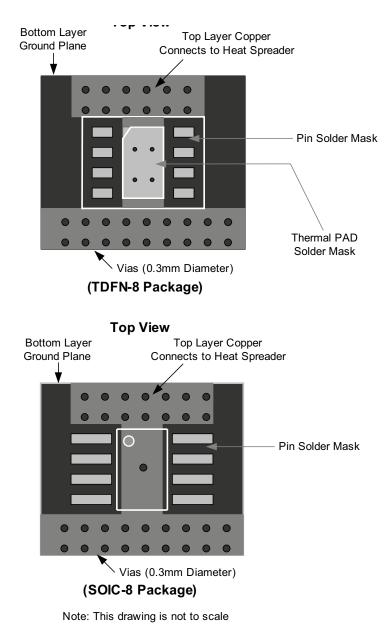


Figure 2. Thermal Layout

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## CM3202

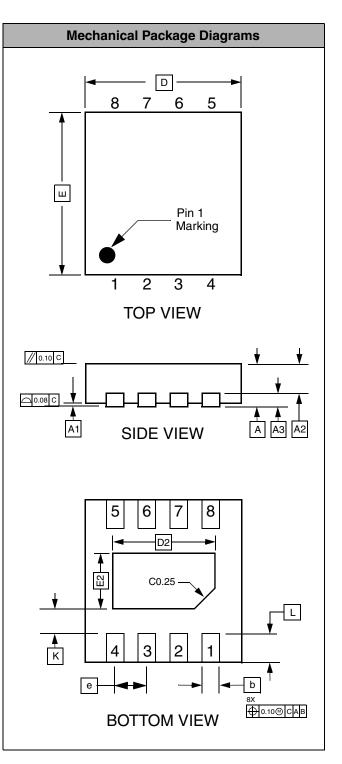
### **Mechanical Details**

#### **TDFN-08 Mechanical Specifications**

The CM3202-00DE is supplied in an 8-lead, 0.65mm pitch TDFN package. Dimensions are presented below.

PACKAGE DIMENSIONS						
Package	TDFN					
JEDEC No.	MO-229 (Var. WEEC-1) <sup>=</sup>					
Leads				6		
Dim.	N	lillimete	rs	Inches		
Dini.	Min	Nom	Max	Min	Nom	Max
Α	0.70	0.75	0.80	0.028	0.030	0.031
A1	0.00	0.02	0.05	0.000	0.001	0.002
A2	0.45	0.55	0.65	0.018	0.022	0.026
A3		0.20			0.008	
b	0.25	0.30	0.35	0.010	0.012	0.014
D		3.00			0.118	
D2	1.90	2.00	2.10	0.075	0.079	0.083
E		3.00			0.118	
E2	1.60	1.70	1.80	0.063	0.067	0.071
е		0.65			0.026	
К	0.20			0.008		
L	0.20	0.30	0.45	0.008	0.012	0.018
# per tape and	3000 pieces					
reel						
Controlling dimension: millimeters						

<sup>=</sup>This package is compliant with JEDEC standard MO-229, variation VEEC-1 with exception of the "D2", "E2" and "b" dimensions as called out in the table above.



#### Package Dimensions for 8-Lead TDFN

### Mechanical Details (cont'd)

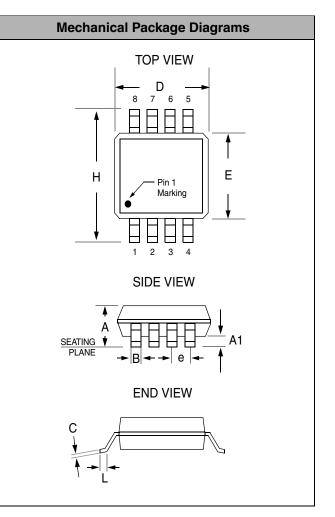
#### **SOIC-8 Mechanical Specifications**

Dimensions for CM3202-00SM devices packaged in 8lead SOIC packages with an intagrated heatslug are presented below.

PACKAGE DIMENSIONS					
Package	SOIC-8				
Leads	8				
Dimensions	Millir	neters	Inches		
Dimensions	Min	Max	Min	Max	
Α	1.30	1.62	0.051	0.064	
A <sub>1</sub>	0.03	0.10	0.001	0.004	
В	0.33	0.51	0.013	0.020	
С	0.18	0.25	0.007	0.010	
D	4.83	5.00	0.190	0.197	
E	3.81	3.99	0.150	0.157	
е	1.27 BSC 0.050 BSC				
н	5.79	6.20	0.228	0.244	
L	0.41	1.27	0.016	0.050	
# per tube	100 pieces*				
# per tape and reel	2500 pieces				
Controlling dimension: inches					

\* This is an approximate number which may vary.

\*\* Centered on package centerline.



Package Dimensions for SOIC-8