19-1781; Rev 3; 5/04

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Single/Dual/Quad, High-Side Current-Sense **Amplifiers with Internal Gain**

General Description

The MAX4376/MAX4377/MAX4378 single, dual, and quad precision high-side current-sense amplifiers are available in space-saving packages. They feature buffered voltage outputs that eliminate the need for gain-setting resistors and are ideal for today's notebook computers, cell phones, and other systems where current monitoring is critical. These precision devices are offered in three fixed-gain versions of 20, 50, and 100:

GAIN	SUFFIX
20	Т
50	F
100	Н

For example, MAX4376TAUK is a single high-side amplifier with a gain of 20.

High-side current monitoring is especially useful in battery-powered systems since it does not interfere with the ground path of the battery charger. The input common-mode range of 0 to +28V is independent of the supply voltage and ensures that the current-sense feedback remains viable even when connected to a battery pack in deep discharge.

The full-scale current reading can be set by choosing the appropriate voltage gain and external-sense resistor. This capability offers a high level of integration and flexibility, resulting in a simple and compact currentsense solution.

The MAX4376/MAX4377/MAX4378 operate over a supply voltage range of +3V to +28V, draw 1mA of supply current per amplifier, and operate over the full automotive temperature range of -40°C to +125°C. These devices have a wide bandwidth of 2MHz, making them suitable for use inside battery-charger control loops. The buffered outputs drive up to 2mA of output current into a ground-referenced load.

The MAX4376 is available in a tiny 5-pin SOT23 package. The MAX4377/MAX4378 are available in spacesaving 8-pin µMAX and 14-pin TSSOP packages, respectively.

Notebook Computers	Portak
Current-Limited Power	Syster
Supplies	Cell P
Fuel Gauges in PC	Smart

General-System/Board-Level Current Monitoring **Battery Chargers**

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ble/Battery-Powered ms hones rt Battery Packages Automotive Current Detect Power Management Systems PA Bias Control

Applications

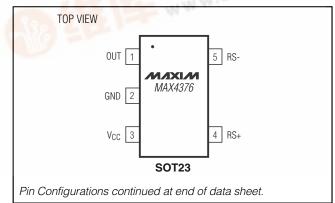
- Features
- Low-Cost, Single/Dual/Quad, High-Side Current-Sense Amplifiers
- ±0.5% Typical Full-Scale Accuracy
- +3V to +28V Supply Operation
- Adjustable Current-Sense Capability with External Sense Resistor
- Buffered Output Voltage with 2mA Drive
- 1mA (typ) Supply Current
- 2.0MHz Bandwidth (Gain = +20V/V)
- Automotive Temperature Range (-40°C to +125°C)
- Full 0 to 28V Common-Mode Range, Independent of Supply Voltage
- Three Gain Versions Available +20V/V (MAX437_T) +50V/V (MAX437 F) +100V/V (MAX437_H)
- Available in Space-Saving 5-Pin SOT23 (Single), 8-Pin µMAX (Dual), and 14-Pin TSSOP (Quad)

Ordering Information

		and the second se		
PART	GAIN (+V/V)	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4376TAUK-T	20	-40°C to +125°C	5 SOT23-5	ADOG
MAX4376FAUK-T	50	-40°C to +125°C	5 SOT23-5	ADOH
MAX4376HAUK-T	100	-40°C to +125°C	5 SOT23-5	ADOI
MAX4376TASA	20	-40°C to +125°C	8 SO	_
MAX4376FASA	50	-40°C to +125°C	8 SO	_
MAX4376HASA	100	-40°C to +125°C	8 SO	—

Ordering Information continued at end of data sheet.

Pin Configurations



Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at

ABSOLUTE MAXIMUM RATINGS

V _{CC} , RS+, RS- to GND OUT to GND Differential Input Voltage (V _{RS+} - V _{RS-})	0.3V to (V _{CC} + 0.3V)
Output Short Circuit to V _{CC}	Continuous
Output Short Circuit to GND	
Current into Any Pin	
Continuous Power Dissipation ($T_A = +70^\circ$	C)
5-Pin SOT23 (derate 7.1mW/°C above	+70°C)571mW
8-Pin µMAX (derate 4.5mW/°C above +	70°C)362mW
8-Pin SO (derate 5.88mW/°C above +7	0°C)471mW

14-Pin SO (derate 8.33mW/°C above +70°C)	667mW
14-Pin TSSOP (derate 9.1mW/°C above +70°C)	727mW
Operating Temperature Range40°C to	o +125°C
Junction Temperature	+150°C
Storage Temperature Range65°C to	o +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{RS+} = 0 \text{ to } 28V, V_{SENSE} = (V_{RS+} - V_{RS-}) = 0V, V_{CC} = +3.0V \text{ to } +28V, R_L = \infty, T_A = T_{MIN} \text{ to } T_{MAX}$, unless otherwise noted. Typical values are at $T_A = 25^{\circ}$ C.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS			ТҮР	MAX	UNITS
Operating Voltage Range	VCC	Guaranteed by PSR test		3		28	V
Common-Mode Input Range	VCM	Guaranteed by tot	tal OUT voltage error test	0		28	V
Common-Mode Rejection	CMR	$2V \le V_{RS+} \le 28V, V$	VSENSE = 100mV		90		dB
Supply Current per Amplifier	Icc	VSENSE = 5mV, VF	_{RS+} > 2.0V, V _{CC} = 12V		1	2.2	mA
Leakage Current	I _{RS+} , I _{RS-}	$V_{CC} = 0V, V_{RS+} =$	28V			8	μΑ
	1= -	$V_{RS+} > 2.0V$		0		60	
	I _{RS+}	$V_{RS+} \le 2.0V$		-400		60	
Input Bias Current	I _{RS-}	$V_{RS+} > 2.0V$		0		120	μA
	IRS-	$V_{RS+} \le 2.0V$		-800		120	
Full-Scale Sense Voltage	VSENSE				150		mV
	I _{OUT} ≤ 2mA	I _{OUT} ≤ 2mA	$V_{\text{SENSE}} = 100 \text{mV},$ $V_{\text{CC}} = 12 \text{V}, V_{\text{RS+}} = 12 \text{V}$			±6.75	%
			$V_{SENSE} = 100mV,$ $V_{CC} = 12V, T_A = +25^{\circ}C$		±0.5	±3.25	
			V _{SENSE} = 100mV, V _{CC} = 28V, V _{RS+} = 28V			±11	
Total OUT Voltage Error (Note 2)			$V_{\text{SENSE}} = 100\text{mV},$ $V_{\text{CC}} = 28\text{V}, V_{\text{RS+}} = 28\text{V},$ $T_{\text{A}} = +25^{\circ}\text{C}$		±0.5	±5	
		$V_{SENSE} = 100 \text{mV},$ $V_{CC} = 12 \text{V}, V_{RS+} = 0.1 \text{V}$		±9	±32		
			$V_{\text{SENSE}} = 6.25 \text{mV},$ $V_{\text{CC}} = 12 \text{V}, V_{\text{RS+}} = 12 \text{V}$ (Note 3)		±7		
OUT High Voltage (Note 4)	(Vcc - Vout)	V _{CC} = 3V, I _{OUT} = 2mA			0.9	1.2	V
OUT Low Voltage	V _{OL}	I _{OUT} = 200μA, V _C V _{SENSE} = 0V, T _A =			25	40	mV

ELECTRICAL CHARACTERISTICS (continued)

(V_{RS+} = 0 to 28V, V_{SENSE} = (V_{RS+} - V_{RS-}) = 0V, V_{CC} = +3.0V to +28V, R_L = ∞ , T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = 25°C.) (Note 1)

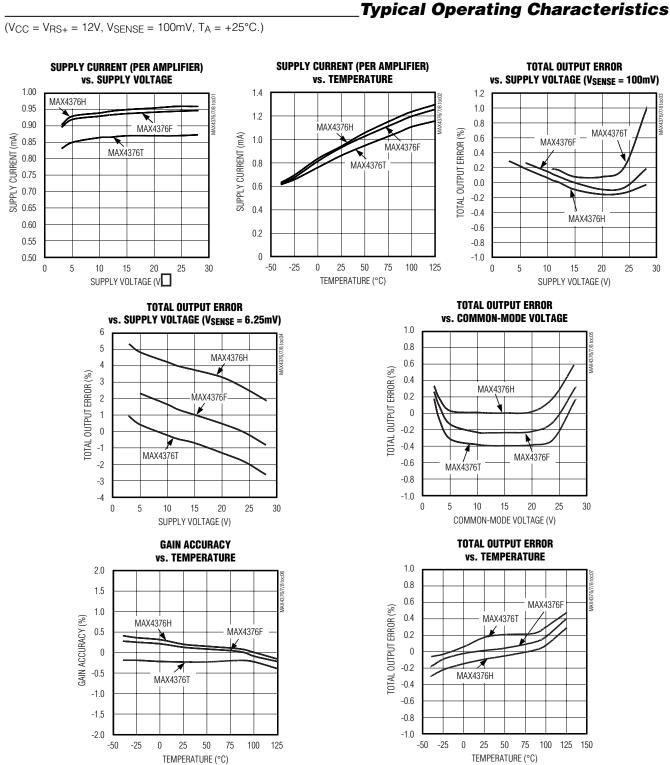
PARAMETER	SYMBOL	co		NS	MIN	ТҮР	MAX	UNITS
				_{SE} = 100mV = +20V/V)		2		
Bandwidth	BW	$V_{CC} = 12V$ $V_{RS+} = 12V$		SE = 100mV = +50V/V)		1.7	MHz	
		C _{LOAD} = 15pF		SE = 100mV = +100V/V)		1.2		
			VSENS	_E = 6.25mV (Note 3)		0.5		
Slew Rate	SR	V _{SENSE} = 20mV to	o 100mV	, C _{LOAD} = 15pF		10		V/µs
		MAX437_T				+20		
Gain	Av	MAX437_F				+50		V/V
		MAX437_H				+100		
Gain Accuracy		$V_{\text{SENSE}} = 10 \text{mV} \text{ to } 150 \text{mV},$ $V_{\text{CC}} = 12 \text{V}, I_{\text{OUT}} = 2 \text{mA},$		$T_A = T_{MIN}$ to T_{MAX}			±5.5	
	ΔA _V	$v_{CC} = 12v, 1001 = 2$ gain = 20 and 50	2111A,	$T_A = +25^{\circ}C$		±0.5	±2.5	C(
		VSENSE = 10mV to 150r		$T_A = T_{MIN}$ to T_{MAX}			5.5	%
		V _{CC} = 20V, I _{OUT} = gain = 100	= 2mA,	T _A = +25°C		±0.5	±2.5	
OUT Setting Time to 1% of Final		V _{CC} = 12V, V _{RS+} :	= 12V,	V _{SENSE} = 6.25mV to 100mV		400		
Value		C _{LOAD} = 15pF		$V_{SENSE} = 100mV$ to 6.25mV		800		ns
Maximum Capacitive Load	CLOAD	No sustained osci	llation			1000		рF
Output Resistance	ROUT	$V_{SENSE} = 100 mV$				5		Ω
Power-Supply Rejection	PSR	$V_{RS+} > 2V, V_{OUT} =$	= 1.6V, \	$V_{\rm CC} = 3V \text{ to } 28V$	66	90		dB
Power-Up Time to 1% of Final Value		V _{SENSE} = 100mV, C _{LOAD} = 15pF			2		μs	
Saturation Recovery Time to 1% of Final Value		$V_{CC} = 12V, V_{RS+} = 12V, C_{LOAD} = 15pF,$ $V_{SENSE} = 100mV$			1		μs	
Reverse Recovery Time to 1% of Final Value		V _{CC} = 12V, V _{RS} = V _{SENSE} = -100mV				1		μs

Note 1: All devices are 100% production tested at $T_A = +25$ °C. All temperature limits are guaranteed by design.

Note 2: Total OUT Voltage Error is the sum of gain and offset errors.

Note 3: 6.25mV = 1/16 of 100mV full-scale sense voltage.

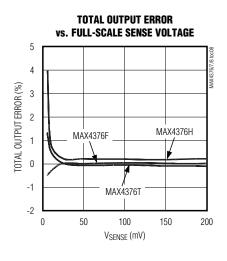
Note 4: V_{SENSE} such that V_{OUT} is in saturation.



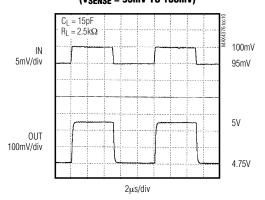
MAX4376/MAX4377/MAX4378



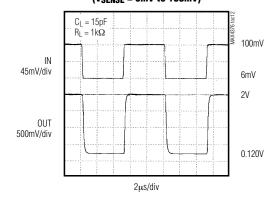
 $(V_{CC} = V_{RS+} = 12V, V_{SENSE} = 100mV, T_A = +25^{\circ}C.)$

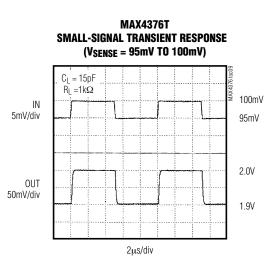




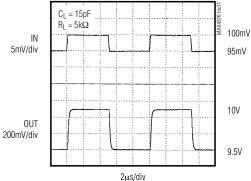




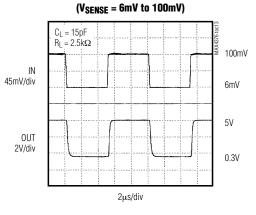


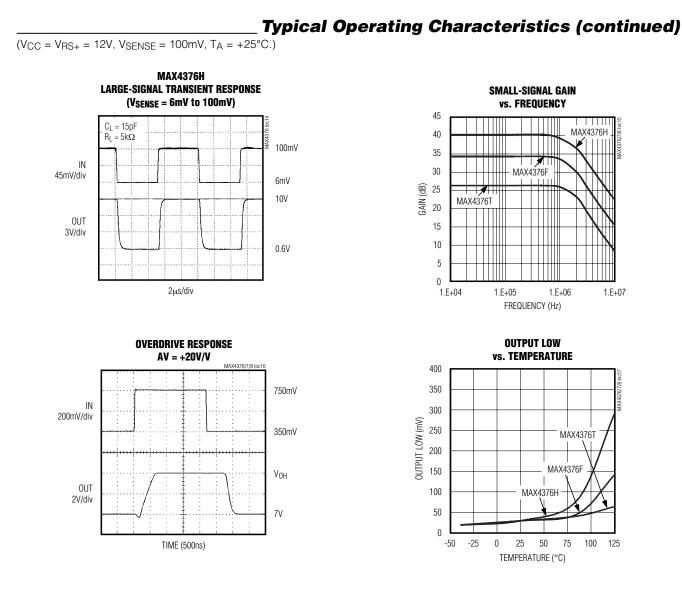


MAX4376H SMALL-SIGNAL TRANSIENT RESPONSE (Vsense = 95mV to 100mV)



MAX4376F LARGE-SIGNAL TRANSIENT RESPONSE





_Pin Description

	Р	IN			FUNCTION	
MAX4376	MAX4376	MAX4377	MAX4378	NAME		
SOT23-5	SO-8	µMAX-8/ SO-8	SO-14/ TSSOP-14	NAME		
1	4	1, 7	1, 7, 8, 14	OUT, OUT_	Output Voltage. V_{OUT} is proportional to the magnitude of the sense voltage ($V_{RS+} - V_{RS-}$). V_{OUT} is approximately zero when $V_{RS-} > V_{RS-} +$ (no phase reversal).	
2	3	4	11	GND	Ground	
3	1	8	4	VCC	Supply Voltage	
4	8	3, 5	3, 5, 10, 12	RS+, RS_+	Power connection to the external sense resistor	
5	6	2, 6	2, 6, 9, 13	RS-, RS	Load-side connection to the external sense resistor	
—	2, 5, 7	_	_	N.C.	No Connection. Not internally connected.	

Detailed Description

The MAX4376/MAX4377/MAX4378 high-side currentsense amplifiers feature a 0 to +28V input common-mode range that is independent of supply voltage. This feature allows the monitoring of current out of a battery in deep discharge and also enables high-side current sensing at voltages greater than the supply voltage (V_{CC}).

The MAX4376/MAX4377/MAX4378 operate as follows: current from the source flows through R_{SENSE} to the load (Figure 1). Since the internal sense amplifier's inverting input has high impedance, negligible current flows through RG2 (neglecting the input bias current). Therefore, the sense amplifier's inverting-input voltage equals V_{SOURCE} - (I_{LOAD})(R_{SENSE}).

The amplifier's open-loop gain forces its noninverting input to the same voltage as the inverting input. Therefore, the drop across RG1 equals (I_{LOAD}) (RSENSE). Since I_{RG1} flows through RG1, I_{RG1} = (I_{LOAD})(RSENSE)/RG1. The internal current mirror multiplies I_{RG1} by a current gain factor, β , to give I_{RGD} = $\beta \times$ I_{RG1}. Solving I_{RGD} = $\beta \times$ (I_{LOAD})(RSENSE)/RG1. Therefore:

 $V_{OUT} = \beta \times (RGD/RG1)(R_{SENSE} \times I_{LOAD}) \times amp gain$

where amp gain is 2, 5, or 10.

The part's gain equals ($\beta \times RGD / RG1$) x amp gain.

Therefore:

 $V_{OUT} = (GAIN)(R_{SENSE})(I_{LOAD})$

where GAIN = 20 for $MAX437_T$.

 $GAIN = 50 \text{ for } MAX437_F.$

 $GAIN = 100 \text{ for } MAX437_H.$

Set the full-scale output range by selecting R_{SENSE} and the appropriate gain version of the MAX4376/MAX4377./MAX4378.

_Applications Information

Recommended Component Values

The MAX4376/MAX4377/MAX4378 sense a wide variety of currents with different sense resistor values. Table 1 lists common resistor values for typical operation of the MAX4376/MAX4377/MAX4378.

Choosing RSENSE

To measure lower currents more accurately, use a high value for R_{SENSE}. The high value develops a higher sense voltage that reduces offset voltage errors of the internal op amp.

In applications monitoring very high currents, RSENSE must be able to dissipate the I²R losses. If the resistor's rated power dissipation is exceeded, its value may drift or it may fail altogether, causing a differential voltage across the terminals in excess of the absolute maximum ratings.

If ISENSE has a large high-frequency component, minimize the inductance of RSENSE. Wire-wound resistors have the highest inductance, metal-film resistors are somewhat better, and low-inductance metal-film resistors are best suited for these applications.

Bidirectional Current-Sense Amplifier

Systems such as laptop computers and other devices that have internal charge circuitry require a precise bidirectional current-sense amplifier to monitor accurately the battery's current regardless of polarity. Figure 2 shows the MAX4377 used as a bidirectional current



Table 1. Recommended Component Values

FULL-SCALE LOAD CURRENT, I _{LOAD} (A)	CURRENT-SENSE RESISTOR, R _{SENSE} (m Ω)	GAIN (+V/V)	FULL-SCALE OUTPUT VOLTAGE (FULL-SCALE V _{SENSE} = 100mV), Vout (V)
0.1	1000	20	2.0
		50	5.0
		100	10.0
1	100	20	2.0
		50	5.0
		100	10.0
5	20	20	2.0
		50	5.0
		100	10.0
10	10	20	2.0
		50	5.0
		100	10.0

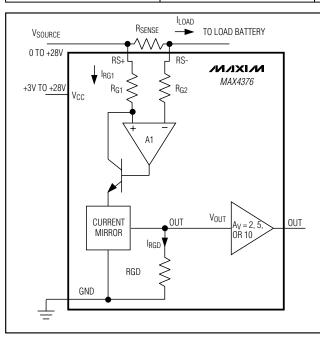


Figure 1. Functional Diagram

monitor. This is useful for implementing either smart battery packs or fuel gauges.

Current Source Circuit

Figure 3 shows a block diagram using the MAX4376 with a switching regulator to make a current source.

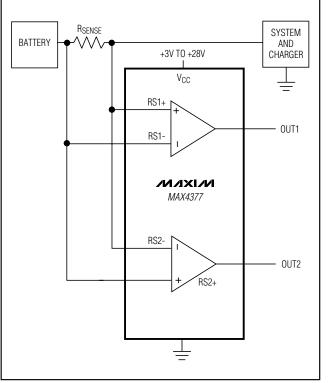
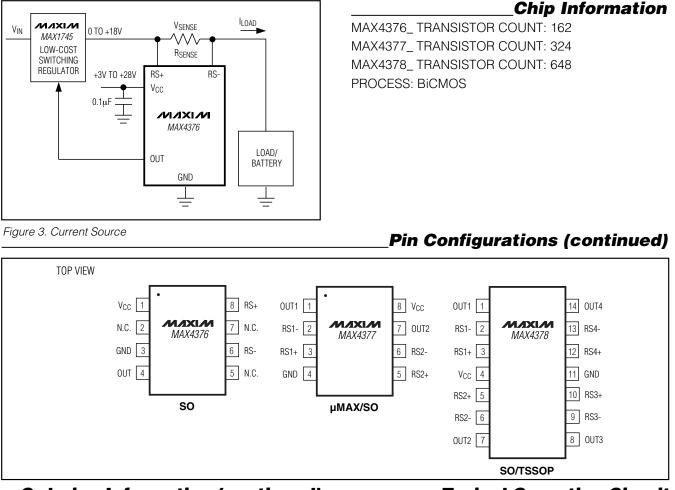


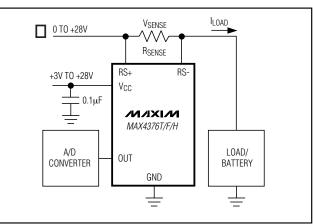
Figure 2. Bidirectional Current Monitor



Ordering Information (continued)

PART	GAIN (+V/V)	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4377 TAUA	20	-40°C to +125°C	8 µMAX	_
MAX4377FAUA	50	-40°C to +125°C	8 µMAX	_
MAX4377HAUA	100	-40°C to +125°C	8 µMAX	_
MAX4377TASA	20	-40°C to +125°C	8 SO	_
MAX4377FASA	50	-40°C to +125°C	8 SO	_
MAX4377HASA	100	-40°C to +125°C	8 SO	_
MAX4378TAUD	20	-40°C to +125°C	14 TSSOP	_
MAX4378FAUD	50	-40°C to +125°C	14 TSSOP	_
MAX4378HAUD	100	-40°C to +125°C	14 TSSOP	_
MAX4378TASD	20	-40°C to +125°C	14 SO	_
MAX4378FASD	50	-40°C to +125°C	14 SO	_
MAX4378HASD	100	-40°C to +125°C	14 SO	_

Typical Operating Circuit



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