



# AD7523, AD7533

August 1997

## 8-Bit, Multiplying D/A Converters

### Features

- 8-Bit, 9-Bit and 10-Bit Linearity
- Low Gain and Linearity Temperature Coefficients
- Full Temperature Range Operation
- Static Discharge Input Protection
- TTL/CMOS Compatible
- Supply Range.....+5V to +15V
- Fast Settling Time at 25°C ..... 150ns (Max)
- Four Quadrant Multiplication
- AD7533 Direct AD7520 Equivalent

### Description

The AD7523 and AD7533 are monolithic, low cost, high performance, 8-bit and 10-bit accurate, multiplying digital-to-analog converter (DAC), in a 16 pin DIP.

Intersil' thin film resistors on CMOS circuitry provide 10-bit resolution (8-bit, 9-bit and 10-bit accuracy), with TTL/CMOS compatible operation.

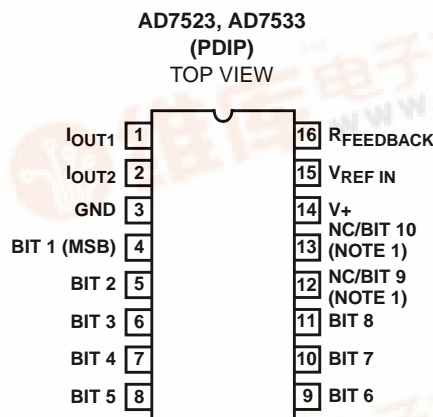
The AD7523 and AD7533s accurate four quadrant multiplication, full military temperature range operation, full input protection from damage due to static discharge by clamps to V+ and GND, and very low power dissipation make it a very versatile converter.

Low noise audio gain controls, motor speed controls, digitally controlled gain and digital attenuators are a few of the wide range of applications of the AD7523 and AD7533.

### Ordering Information

PART NUMBER	LINEARITY (INL, DNL)	TEMP. RANGE (°C)	PACKAGE	PKG. NO.
AD7523JN, AD7533JN	0.2% (8-Bit)	0 to 70	16 Ld PDIP	E16.3
AD7523KN, AD7533KN	0.1% (9-Bit)	0 to 70	16 Ld PDIP	E16.3
AD7523LN, AD7533LN	0.05% (10-Bit)	0 to 70	16 Ld PDIP	E16.3

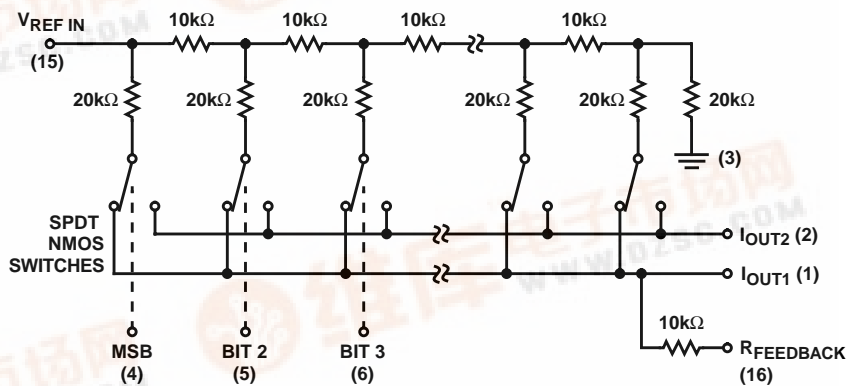
### Pinout



NOTE:

1. NC for AD7523 only.

### Functional Block Diagram



NOTE: Switches shown for digital inputs "High"



## AD7523, AD7533

### Absolute Maximum Ratings

Supply Voltage (V+ to GND)	+17V
V <sub>REF</sub>	±25V
Digital Input Voltage Range	V+ to GND
Output Voltage Compliance	-100mV to V+

### Thermal Information

Thermal Resistance (Typical, Note 1)	$\theta_{JA}$ (°C/W)
PDIP Package	100
Maximum Junction Temperature (Plastic Package)	150°C
Maximum Storage Temperature	-65°C to 150°C
Maximum Lead Temperature (Soldering 10s)	300°C

### Operating Conditions

Temperature Range	.0°C to 70°C
JN, KN, LN Versions	.0°C to 70°C

**CAUTION:** Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

#### NOTE:

- $\theta_{JA}$  is measured with the component mounted on an evaluation PC board in free air.

### Electrical Specifications $V_+ = +15V$ , $V_{REF} = +10V$ , $V_{OUT1} = V_{OUT2} = 0V$ , Unless Otherwise Specified

PARAMETER		TEST CONDITIONS	AD7523				AD7533				UNITS
			$T_A$ 25°C		$T_A$ MIN-MAX		$T_A$ 25°C		$T_A$ MIN-MAX		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
<b>SYSTEM PERFORMANCE</b>											
Resolution			8	-	8	-	10	-	10	-	Bits
Nonlinearity	J	$-10V \leq V_{REF} \leq +10V$ $V_{OUT1} = V_{OUT2} = 0V$ (Notes 2, 3, 6)	-	±0.2	-	±0.2	-	±0.2	-	±0.2	% of FSR
	K, T		-	±0.1	-	±0.1	-	±0.1	-	±0.1	% of FSR
	L		-	±0.05	-	±0.05	-	±0.05	-	±0.05	% of FSR
Monotonicity			Guaranteed				Guaranteed				
Gain Error		All Digital Inputs High (Note 3)	-	±1.5	-	±1.8	-	±1.4	-	±1.8	% of FSR
Nonlinearity Tempco		$-10V \leq V_{REF} \leq +10V$ (Notes 3, 4)	-	±2	-	±2	-	±2	-	±2	ppm of FSR/°C
Gain Error Tempco			-	±10	-	±10	-	±10	-	±10	ppm of FSR/°C
Output Leakage Current (Either Output)		$V_{OUT1} = V_{OUT2} = 0$	-	±50	-	±200	-	±50	-	±200	nA
<b>DYNAMIC CHARACTERISTICS</b>											
Power Supply Rejection		$V_+ = 14.0V$ to $15.0V$ (Note 3)	-	±0.02	-	±0.03	-	±0.005	-	±0.008	% of FSR/% of $\Delta V_+$
Output Current Settling Time		To 0.2% of FSR, $R_L = 100\Omega$ (Note 4)	-	150	-	200	-	600	-	800	ns
Feedthrough Error		$V_{REF} = 20V_{P-P}$ , 200kHz Sine Wave, All Digital Inputs Low (Note 4)	-	±1/2	-	±1	-	±0.05	-	±0.1	LSB
<b>REFERENCE INPUTS</b>											
Input Resistance (Pin 15)		All Digital Inputs High $I_{OUT1}$ at Ground (Note 4)	5	-	5	-	5	-	5	-	k $\Omega$
			-	20	-	20	-	20	-	20	k $\Omega$
Temperature Coefficient			-	-500	-	-500	-	-300	-	-300	ppm/°C

## AD7523, AD7533

### Electrical Specifications $V_+ = +15V$ , $V_{REF} = +10V$ , $V_{OUT1} = V_{OUT2} = 0V$ , Unless Otherwise Specified (Continued)

PARAMETER		TEST CONDITIONS	AD7523				AD7533				UNITS
			$T_A$ 25°C		$T_A$ MIN-MAX		$T_A$ 25°C		$T_A$ MIN-MAX		
			MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
<b>ANALOG OUTPUT</b>											
Output Capacitance	$C_{OUT1}$	All Digital Inputs High (Note 4)	-	100	-	100	-	100	-	100	pF
	$C_{OUT2}$		-	30	-	30	-	35	-	35	pF
	$C_{OUT1}$	All Digital Inputs Low (Note 4)	-	30	-	30	-	35	-	35	pF
	$C_{OUT2}$		-	100	-	100	-	100	-	100	pF
<b>DIGITAL INPUTS</b>											
Low State Threshold, $V_{IL}$			-	0.8	-	0.8	-	0.8	-	0.8	V
High State Threshold, $V_{IH}$			2,4	-	2,4	-	2.4	-	2.4	-	V
Input Current (Low or High), $I_{IL}$ , $I_{IH}$		$V_{IN} = 0V$ or $+15V$	-	$\pm 1$	-	$\pm 1$	-	$\pm 1$	-	$\pm 1$	$\mu A$
Input Coding		See Tables 1 and 3	Binary/Offset Binary				Binary/Offset Binary				
Input Capacitance		(Note 4)	-	4	-	4	-	4	-	4	pF
<b>POWER SUPPLY CHARACTERISTICS</b>											
Power Supply Voltage Range		(Note 6)	+5 to +16				+5 to +16				V
$I_+$		All Digital Inputs High or Low (Excluding Ladder Network)	-	2	-	2.5	-	2	-	2.5	mA

**NOTES:**

2. Full Scale Range (FSR) is 10V for unipolar and  $\pm 10V$  for bipolar modes.
3. Using internal feedback resistor,  $R_{FEEDBACK}$ .
4. Guaranteed by design or characterization and not production tested.
5. Accuracy not guaranteed unless outputs at ground potential.
6. Accuracy is tested and guaranteed at  $V_+ = +15V$ , only.

### Definition of Terms

**Nonlinearity:** Error contributed by deviation of the DAC transfer function from a “best straight line” through the actual plot of transfer function. Normally expressed as a percentage of full scale range or in (sub)multiples of 1 LSB.

**Resolution:** It is addressing the smallest distinct analog output change that a D/A converter can produce. It is commonly expressed as the number of converter bits. A converter with resolution of  $n$  bits can resolve output changes of  $2^{-N}$  of the full-scale range, e.g.,  $2^{-N} V_{REF}$  for a unipolar conversion. Resolution by no means implies linearity.

**Settling Time:** Time required for the output of a DAC to settle to within specified error band around its final value (e.g.,  $1/2$  LSB) for a given digital input change, i.e., all digital inputs LOW to HIGH and HIGH to LOW.

**Gain Error:** The difference between actual and ideal analog output values at full-scale range, i.e., all digital inputs at HIGH state. It is expressed as a percentage of full scale range or in (sub)multiples of 1 LSB.

**Feedthrough Error:** Error caused by capacitive coupling from  $V_{REF}$  to  $I_{OUT1}$  with all digital inputs LOW.

**Output Capacitance:** Capacitance from  $I_{OUT1}$ , and  $I_{OUT2}$  terminals to ground.

**Output Leakage Current:** Current which appears on  $I_{OUT1}$ , terminal when all digital inputs are LOW or on  $I_{OUT2}$  terminal when all digital inputs are HIGH.

For further information on the use of this device, see the following Application Notes:

NOTE #	DESCRIPTION	AnswerFAX DOC. #
AN002	“Principles of Data Acquisition and Conversion”	9002
AN018	“Do’s and Don’ts of Applying A/D Converters”	9018
AN042	“Interpretation of Data Conversion Accuracy Specifications”	9042



## AD7523, AD7533

### Unipolar Binary Operation - AD7533 (10-Bit DAC)

The circuit configuration for operating the AD7533 in unipolar mode is shown in Figure 2. With positive and negative  $V_{REF}$  values the circuit is capable of 2-Quadrant multiplication. The "Digital Input Code/Analog Output Value" table for unipolar mode is given in Table 2.

**TABLE 2. UNIPOLAR BINARY CODE - AD7533**

DIGITAL INPUT MSB LSB	(NOTE 1) NOMINAL ANALOG OUTPUT
1111111111	$-V_{REF} \left( \frac{1023}{1024} \right)$
1000000001	$-V_{REF} \left( \frac{513}{1024} \right)$
1000000000	$-V_{REF} \left( \frac{512}{1024} \right) = -\frac{V_{REF}}{2}$
0111111111	$-V_{REF} \left( \frac{511}{1024} \right)$
0000000001	$-V_{REF} \left( \frac{1}{1024} \right)$
0000000000	$-V_{REF} \left( \frac{0}{1024} \right) = 0$

**NOTES:**

- $V_{OUT}$  as shown in the Functional Diagram.
- Nominal Full Scale for the circuit of Figure 2 is given by:  

$$FS = -V_{REF} \left( \frac{1023}{1024} \right)$$
- Nominal LSB magnitude for the circuit of Figure 2 is given by:  

$$LSB = V_{REF} \left( \frac{1}{1024} \right)$$

### Zero Offset Adjustment

- Connect all digital inputs to GND.
- Adjust the offset zero adjust trimpot of the output operational amplifier for  $0V \pm 1mV$  (Max) at  $V_{OUT}$ .

### Gain Adjustment

- Connect all digital inputs to  $V+$ .
- Monitor  $V_{OUT}$  for a  $-V_{REF} (1 - 1/2^{10})$  reading.

- To increase  $V_{OUT}$ , connect a series resistor,  $R_2$ , ( $0\Omega$  to  $250\Omega$ ) in the  $I_{OUT1}$  amplifier feedback loop.
- To decrease  $V_{OUT}$ , connect a series resistor,  $R_1$ , ( $0\Omega$  to  $250\Omega$ ) between the reference voltage and the  $V_{REF}$  terminal.

### Bipolar (Offset Binary) Operation - AD7523

The circuit configuration for operating the AD7523 in the bipolar mode is given in Figure 3. Using offset binary digital input codes and positive and negative reference voltage values, Four-Quadrant multiplication can be realized. The "Digital Input Code/Analog Output Value" table for bipolar mode is given in Table 3.)

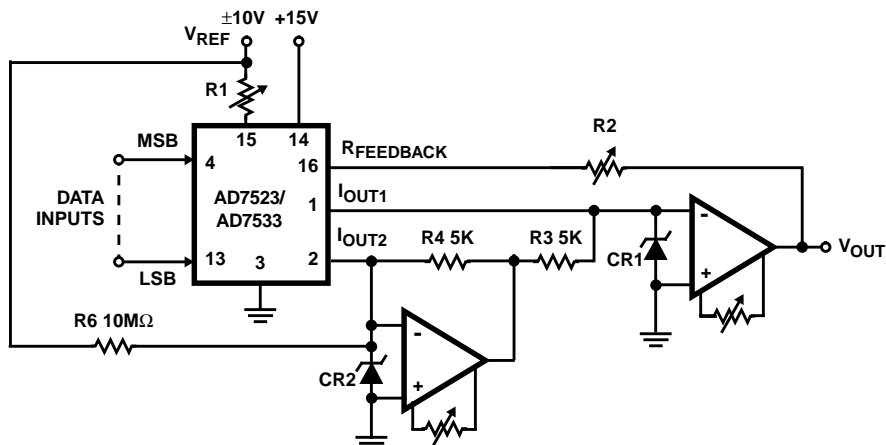
A "Logic 1" input at any digital input forces the corresponding ladder switch to steer the bit current to  $I_{OUT1}$  bus. A "Logic 0" input forces the bit current to  $I_{OUT2}$  bus. For any code the  $I_{OUT1}$  and  $I_{OUT2}$  bus currents are complements of one another. The current amplifier at  $I_{OUT2}$  changes the polarity of  $I_{OUT2}$  current and the transconductance amplifier at  $I_{OUT}$  output sums the two currents. This configuration doubles the output range. The difference current resulting at zero offset binary code, (MSB = "Logic 1", all other bits = "Logic 0"), is corrected by using an external resistor, ( $10M\Omega$ ), from  $V_{REF}$  to  $I_{OUT2}$  (Figure 3).

**TABLE 3. BIPOLAR (OFFSET BINARY) CODE - AD7523**

DIGITAL INPUT MSB LSB	ANALOG OUTPUT
1111111111	$-V_{REF} \left( \frac{127}{128} \right)$
100000001	$-V_{REF} \left( \frac{1}{128} \right)$
100000000	0
0111111111	$+V_{REF} \left( \frac{1}{128} \right)$
000000001	$+V_{REF} \left( \frac{127}{128} \right)$
000000000	$+V_{REF} \left( \frac{128}{128} \right)$

**NOTE:**

- $1 \text{ LSB} = (2^{-7})(V_{REF}) = \left( \frac{1}{128} \right)(V_{REF})$ .



**FIGURE 3. BIPOLAR OPERATION (4-QUADRANT MULTIPLICATION)**

## AD7523, AD7533

### Offset Adjustment

1. Adjust  $V_{REF}$  to approximately +10V.
2. Connect all digital inputs to "Logic 1".
3. Adjust  $I_{OUT2}$  amplifier offset adjust trimpot for  $0V \pm 1mV$  at  $I_{OUT2}$  amplifier output.
4. Connect MSB (Bit 1) to "Logic 1" and all other bits to "Logic 0".
5. Adjust  $I_{OUT1}$  amplifier offset adjust trimpot for  $0V \pm 1mV$  at  $V_{OUT}$ .

### Gain Adjustment

1. Connect all digital inputs to  $V+$ .
2. Monitor  $V_{OUT}$  for a  $-V_{REF} (1\frac{1}{2}^8)$  volts reading.
3. To increase  $V_{OUT}$ , connect a series resistor, R2, of up to  $250\Omega$  between  $V_{OUT}$  and  $R_{FEEDBACK}$ .
4. To decrease  $V_{OUT}$ , connect a series resistor, R1, of up to  $250\Omega$  between the reference voltage and the  $V_{REF}$  terminal.

### Bipolar (Offset Binary) Operation - AD7533

The circuit configuration for operating the AD7533 in the bipolar mode is given in Figure 3. Using offset binary digital input codes and positive and negative reference voltage values, 4-Quadrant multiplication can be realized. The "Digital Input Code/Analog Output Value" table for bipolar mode is given in Table 4.

A "Logic 1" input at any digital input forces the corresponding ladder switch to steer the bit current to  $I_{OUT1}$  bus. A "Logic 0" input forces the bit current to  $I_{OUT2}$  bus. For any code the  $I_{OUT1}$  and  $I_{OUT2}$  bus currents are complements of one

another. The current amplifier at  $I_{OUT2}$  changes the polarity of  $I_{OUT2}$  current and the transconductance amplifier at  $I_{OUT1}$  output sums the two currents. This configuration doubles the output range. The difference current resulting at zero offset binary code, (MSB = "Logic 1", all other bits = "Logic 0"), is corrected by using an external resistor, ( $10M\Omega$ ), from  $V_{REF}$  to  $I_{OUT2}$ .

**TABLE 4. UNIPOLAR BINARY CODE - AD7533**

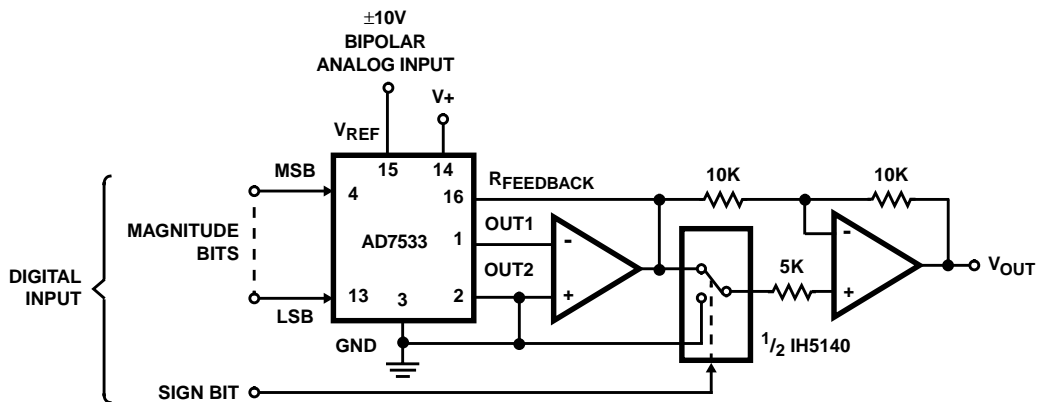
DIGITAL INPUT MSB    LSB	(NOTE 1) NOMINAL ANALOG OUTPUT
1111111111	$-V_{REF} \left( \frac{511}{512} \right)$
1000000001	$-V_{REF} \left( \frac{1}{512} \right)$
1000000000	0
0111111111	$+V_{REF} \left( \frac{1}{512} \right)$
0000000001	$+V_{REF} \left( \frac{511}{512} \right)$
0000000000	$+V_{REF} \left( \frac{512}{512} \right)$

**NOTES:**

1.  $V_{OUT}$  as shown in the Functional Diagram.
2. Nominal Full Scale for the circuit of Figure 6 is given by:  

$$FSR = V_{REF} \left( \frac{1023}{512} \right).$$
3. Nominal LSB magnitude for the circuit of Figure 3 is given by:

$$LSB = V_{REF} \left( \frac{1}{512} \right).$$



**FIGURE 4. 10-BIT AND SIGN MULTIPLYING DAC**

## AD7523, AD7533

### Offset Adjustment

1. Adjust  $V_{REF}$  to approximately +10V.
2. Connect all digital inputs to "Logic 1".
3. Adjust  $I_{OUT2}$  amplifier offset adjust trimpot for  $0V \pm 1mV$  at  $I_{OUT2}$  amplifier output.
4. Connect MSB (Bit 1) to "Logic 1" and all other bits to "Logic 0".
5. Adjust  $I_{OUT1}$  amplifier offset adjust trimpot for  $0V \pm 1mV$  at  $V_{OUT}$ .

### Gain Adjustment

1. Connect all digital inputs to  $V+$ .
2. Monitor  $V_{OUT}$  for a  $-V_{REF} (1 - 2^{-9})$  volts reading.
3. To increase  $V_{OUT}$ , connect a series resistor of up to  $250\Omega$  between  $V_{OUT}$  and  $R_{FEEDBACK}$ .
4. To decrease  $V_{OUT}$ , connect a series resistor of up to  $250\Omega$  between the reference voltage and the  $V_{REF}$  terminal.

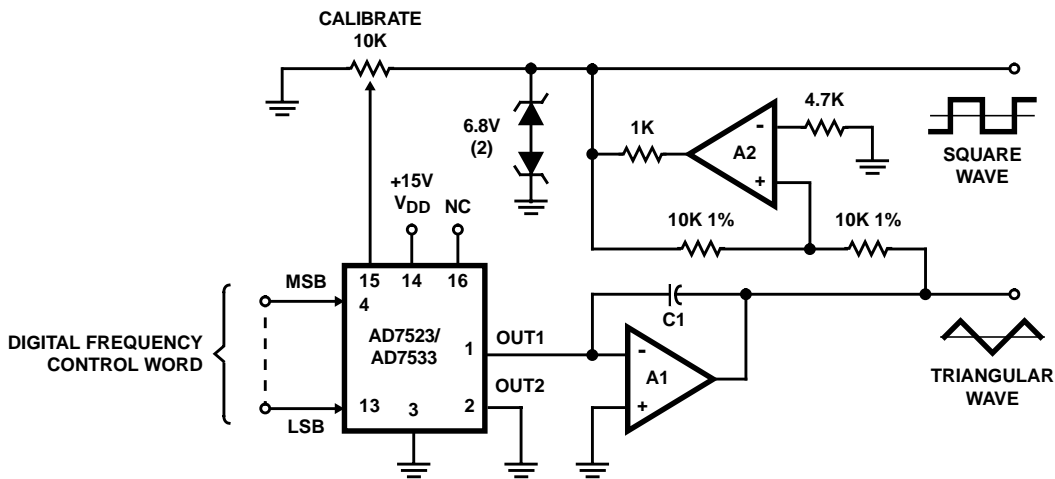
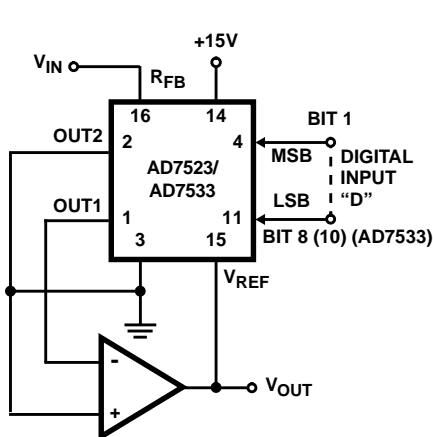


FIGURE 5. PROGRAMMABLE FUNCTION GENERATOR



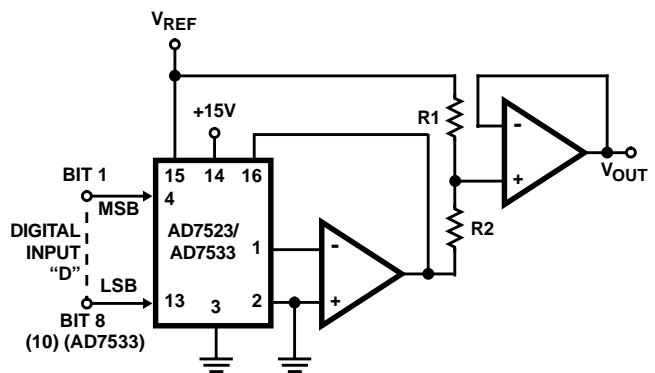
$$V_{OUT} = -V_{IN}/D$$

Where:

$$D = \frac{\text{Bit 1}}{2^1} + \frac{\text{Bit 2}}{2^2} + \dots + \frac{\text{Bit 8}}{2^8}$$

$$\left(0 \leq D \leq \frac{255}{256}\right)$$

FIGURE 6. DIVIDER (DIGITALLY CONTROLLED GAIN)



$$V_{OUT} = V_{REF} \left[ \left( \frac{R_2}{R_1 + R_2} \right) - \left( \frac{R_1 D}{R_1 + R_2} \right) \right]$$

Where  $D = \frac{\text{Bit 1}}{2^1} + \frac{\text{Bit 2}}{2^2} + \dots + \frac{\text{Bit 8}}{2^8}$

$$\left(0 \leq D \leq \frac{255}{256}\right)$$

FIGURE 7. MODIFIED SCALE FACTOR AND OFFSET

## AD7523, AD7533

### Die Characteristics

**DIE DIMENSIONS:**

101 mils x 103 mils (2565micrms x 2616micrms)

**METALLIZATION:**

Type: Pure Aluminum  
 Thickness:  $10 \pm 1 \text{k}\text{\AA}$

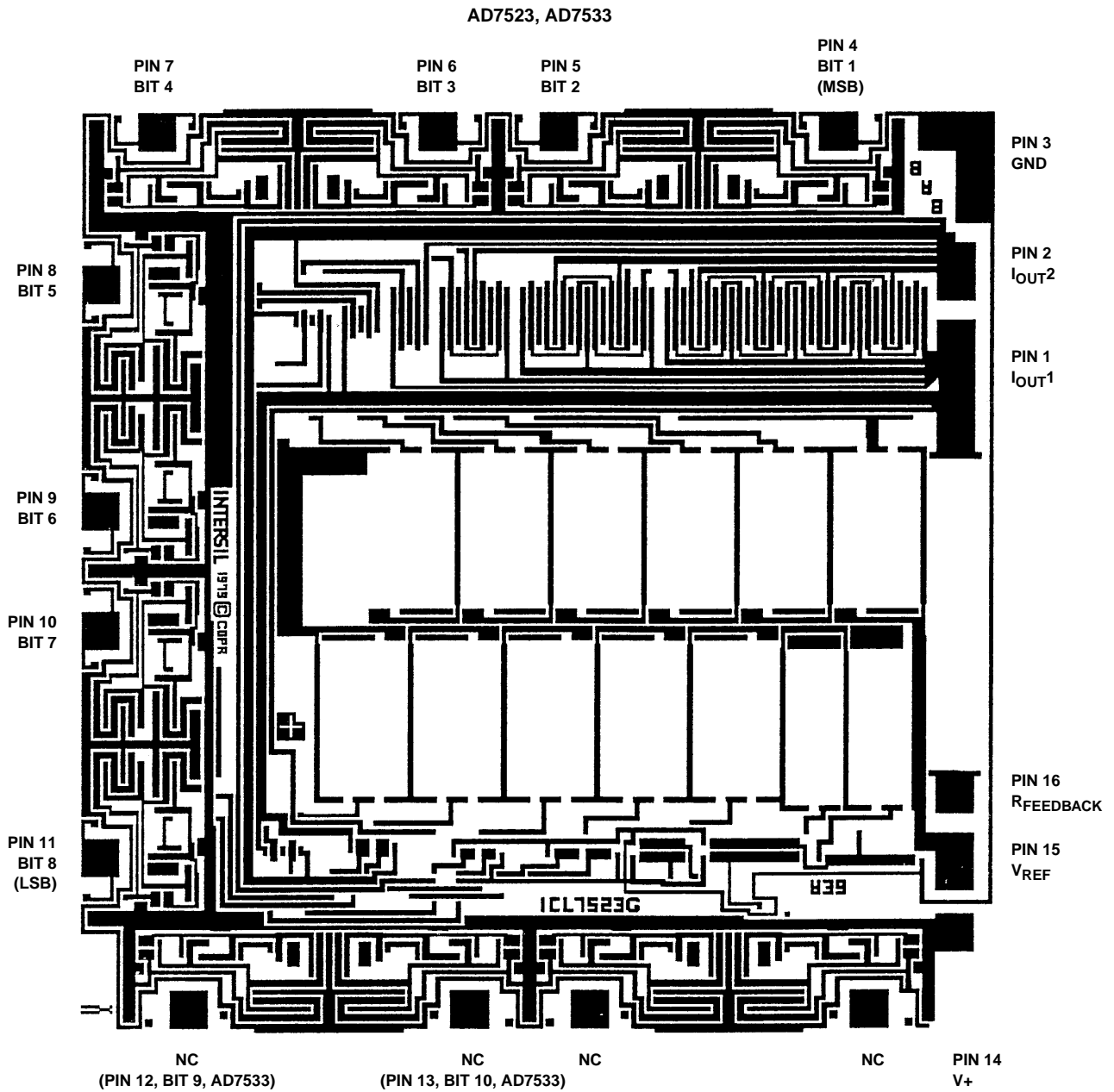
**PASSIVATION:**

Type: PSG/Nitride  
 PSG:  $7 \pm 1.4 \text{k}\text{\AA}$   
 Nitride:  $8 \pm 1.2 \text{k}\text{\AA}$

**PROCESS:**

CMOS Metal Gate

### Metallization Mask Layout





## **AD7523, AD7533**

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