

**Preliminary Datasheet** 

#### **BOOST CONTROLLER**

AP3039

#### **General Description**

AP3039 is a current mode high voltage low-side Nchannel MOSFET controller which is ideal for boost regulators. It contains all the features needed to implement single ended primary topology DC/DC converters.

The input voltage range of AP3039 is from 6V to 27V. Its operation frequency is adjustable from 400K to 1MHz.

The AP3039 has UVLO (Under Voltage Lock Out) circuit. It uses two external resistors to set the UVLO voltage. The AP3039 also has an over output voltage protection to limit the output voltage. The OVP voltage can be set through external resistors. If the output voltage is higher than the OVP high threshold point, it will disable the driver, when the output voltage drops to the OVP low threshold point, it will enable the driver. It also features a soft start to reduce the inrush current when power on, the soft start time can be set through an external capacitor.

AP3039 is available in QFN-3x3-16 package.

#### **Features**

- Input Voltage Range 6V to 27V
- 0.6A Peak MOSFET Gate Driver
- 20ns Quick MOSFET Gate Driver
- Duty Cycle Limit of 90%
- Programmable UVLO
- Programmable Over Voltage Protection
- Cycle by Cycle Current Limit
- Adjustable Soft-Start
- Adjustable Operation Frequency from 400 kHz to 1 MHz

#### **Applications**

- Notebook
- LCD Display Modules



Figure 1. Package Type of AP3039





AP3039

#### **Pin Configuration** FN Package Pin 1 Identification (QFN-3x3-16) OV UVLO SS COMP 1 ΕN NC [12] $\overline{2}$ VIN [11] FΒ ΕP <u>3</u> ] NC 10 SHDN

4

VCC

OUT PGND RT CS
Figure 2. Pin Configuration of AP3039 (Top View)

5 6 7 8

[ 9 ]

**AGND** 

Exposed PAD

# **Pin Description**

Pin Number	Pin Name	Function
1	EN	Enable pin
2	VIN	Input supply pin, must be locally bypassed
3, 12	NC	No connection
4	VCC	$6V$ linear regulator output pin. VCC is used to bias the gate driver for the external MOSFET. If $V_{IN}$ is less than $8.5V$ , the $V_{CC}$ is equal to $V_{IN}$ minus drop voltage across bypass switch. This pin should be bypassed to GND (recommend to connect with AGND pin) with a ceramic capacitor
5	OUT	Connect this pin to the gate of external MOSFET, the gate driver has 0.6A peak current capability
6	PGND	Power ground
7	RT	An external resistor connected from this pin to GND to set the operating frequency
8	CS	Sense switch current pin, which is used for current mode control and for current limit
9	AGND	Reference ground
10	SHDN	This pin can be connected to current matched chip and receives error signal used to shut down the system
11	FB	Voltage Feedback Pin. The reference voltage is 500mV
13	COMP	Compensation Pin. This pin is the output of the internal Error Amplifier



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# **Pin Description (Continued)**

Pin Number	Pin Name	Function
14	SS	An external soft start time capacitor is connected from this pin to ground and is charged by internal $12\mu A$ current source to control regulator soft start time
15	UVLO	Two resistors connected from this pin to ground and the VIN pin respectively to set start up and shutdown level
16	OV	Over output voltage protection pin
	EP	Exposed backside pad. Solder to the circuit board ground plane with sufficient copper connection to ensure low thermal resistance

# **Functional Block Diagram**

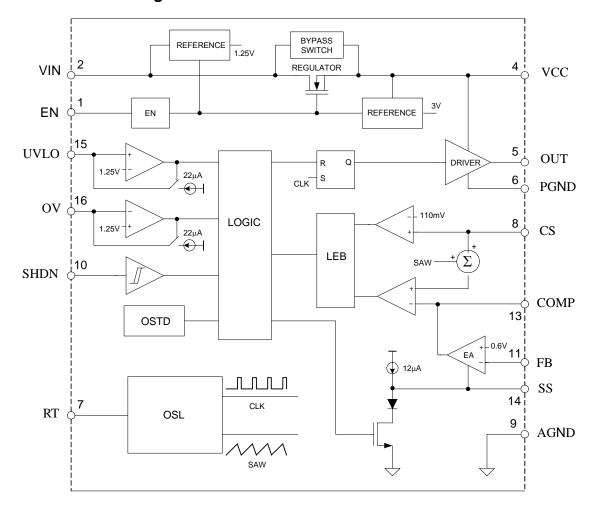
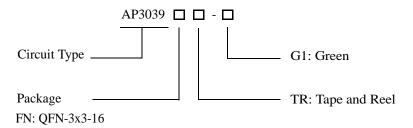


Figure 3. Functional Block Diagram of AP3039



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# **Ordering Information**



Package	Temperature Range	Part Number	Marking ID	Packing Type
QFN-3x3-16	-40 to 85°C	AP3039FNTR-G1	B2A	Tape & Reel

BCD Semiconductor's products, as designated with "G1" suffix in the part number, are RoHS compliant and Green.



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# **Absolute Maximum Ratings (Note 1)**

Parameter	Symbol	Value	Unit
Input Voltage	V <sub>IN</sub>	30	V
VCC Pin Voltage	V <sub>CC</sub>	10	V
OUT Pin Voltage	V <sub>OUT</sub>	10	V
Feedback Pin Voltage	$V_{\mathrm{FB}}$	7	V
UVLO Pin Voltage	V <sub>UVLO</sub>	7	V
CS Pin Voltage	V <sub>CS</sub>	7	V
SHDN Pin Voltage	V <sub>SHDN</sub>	7	V
Enable Pin Voltage	V <sub>EN</sub>	V <sub>IN</sub>	V
OV Pin Voltage	V <sub>OV</sub>	7	V
Thermal Resistance (Junction to Ambient, no Heat sink)	$R_{ heta JA}$	60	°C/W
Operating Junction Temperature	$T_{J}$	150	°C
Storage Temperature Range	$T_{STG}$	-65 to 150	°C
Lead Temperature (Soldering, 10sec)	$T_{LEAD}$	260	°C
ESD (Machine Model)		200	V
ESD (Human Body Model)		2000	V

Note 1: Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "Recommended Operating Conditions" is not implied. Exposure to "Absolute Maximum Ratings" for extended periods may affect device reliability.

#### **Recommended Operating Conditions**

Parameter	Symbol	Min	Max	Unit
Input Voltage	V <sub>IN</sub>	6	27	V
Operating Frequency	f	400	1000	kHz
Operating Temperature	T <sub>A</sub>	-40	85	°C



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# **Electrical Characteristics**

(V  $_{\mbox{\footnotesize{IN}}}\!=\!12\mbox{\footnotesize{V}}\!,\, V_{\mbox{\footnotesize{EN}}}\!=\!V_{\mbox{\footnotesize{IN}}},\, T_{\mbox{\footnotesize{A}}}\!=\!25^{\mbox{\footnotesize{o}}}\!C,$  unless otherwise specified.)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input Voltage	$V_{\rm IN}$		6		27	V
Feedback Voltage	$V_{\mathrm{FB}}$		470	500	530	mV
FB Pin Bias Current	$I_{\mathrm{FB}}$			35	100	nA
Supply Current	$I_{CC}$	Not Switching		3	5	mA
Shutdown Supply Current	$I_Q$	$V_{EN}=0V$		1	2	μΑ
VCC Voltage	V <sub>CC</sub>	9V≤V <sub>IN</sub> ≤27V	5.5	6	6.5	V
Vec voltage	, ((	6V≤V <sub>IN</sub> <9V	5			_ v
VCC Current Limit	I <sub>CC-LIM</sub>			50		mA
Drop Voltage Across Bypass Switch	V <sub>IN</sub> -V <sub>CC</sub>	I <sub>CC</sub> =0mA, f <sub>OSC</sub> ≤400kHz, 6V≤V <sub>IN</sub> <8.5V		300		mV
Bypass Switch Turn-off Threshold	V <sub>BYP-HI</sub>	V <sub>IN</sub> increasing		8.7		V
Bypass Switch Threshold Hysteresis	V <sub>BYP-HYS</sub>	V <sub>IN</sub> decreasing		260		mV
VCC Pin UVLO Rising Threshold	V <sub>CC-HI</sub>			4.7		V
VCC Pin UVLO Falling Hysteresis	V <sub>CC-HYS</sub>			300		mV
Oscillator Frequency	f <sub>OSC</sub>	Adjustable, $R_T$ =51k $\Omega$ to 150k $\Omega$	400		1000	kHz
UVLO Threshold	V <sub>UVLO</sub>		1.22	1.25	1.28	V
UVLO Hysteresis Current Source	I <sub>HYS</sub>			22		μΑ
Current Limit Threshold Voltage	V <sub>CS</sub>		90	110	130	mV
RT Voltage	V <sub>RT</sub>		1.20	1.25	1.30	V
Error Amplifier Transconductance	G <sub>S</sub>			470		μA/V
EN Pin Threshold Voltage	V <sub>EH</sub>		2.0			V
LIVE III THE SHORT VORTAGE	V <sub>EL</sub>				0.5	•
SHDN Pin Threshold Voltage	V <sub>IH</sub>		2.0			V
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# **Electrical Characteristics (Continued)**

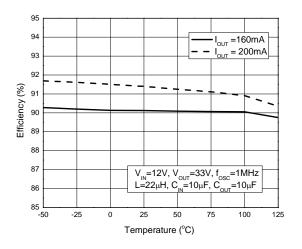
(V  $_{IN}\!\!=\!\!12V\!,\,V_{EN}\!=\!\!V_{IN},\,T_{A}\!\!=\!\!25^{o}C\!,$  unless otherwise specified.)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
OV Threshold	V <sub>OV</sub>			1.25		V
OV Hysteresis Current Source	I <sub>OV-HYS</sub>			22		μΑ
Maximum Duty Cycle	D <sub>MAX</sub>			90	93	%
Soft Start Current Source	I <sub>SS</sub>			12		μΑ
Out Pin Rise Time	t <sub>RISE</sub>	Out Pin Load =1nF		20		ns
Out Pin Fall Time	t <sub>FALL</sub>	Out Pin Load =1nF		20		ns
OUT Dropout Voltage (V <sub>CC</sub> -V <sub>OUT</sub> )	V <sub>OUT-H</sub>	I <sub>OUT</sub> =50mA		0.25	0.75	V
OUT Low Voltage Level (V <sub>OUT</sub> )	V <sub>OUT-L</sub>	I <sub>OUT</sub> =100mA		0.25	0.75	V
Thermal Shutdown Temperature	T <sub>OTSD</sub>			160		°C
Thermal Shutdown Hysteresis	T <sub>HYS</sub>			20		°C



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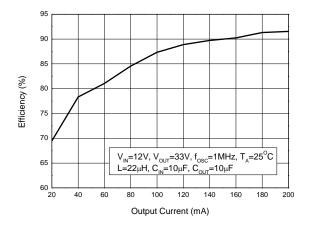
# **Typical Performance Characteristics**



f<sub>osc</sub> =400kHz 94 osc= 1MHz 93 92 Efficiency (%) 91 90 89 87  $I_{OUT}$ =160mA,  $V_{OUT}$ =33V,  $T_{A}$ =25 $^{\circ}$ C L=22 $\mu$ H,  $C_{IN}$ =10 $\mu$ F;  $C_{OUT}$ =10 $\mu$ F 86 85 12 18 24 27 Input Voltage (V)

Figure 4. Efficiency vs. Case Temperature

Figure 5. Efficiency vs. Input Voltage



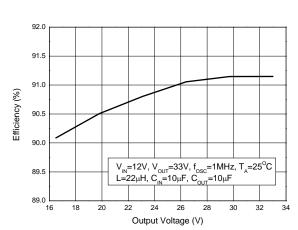


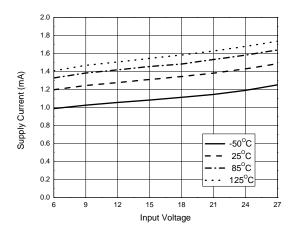
Figure 6. Efficiency vs. Output Current

Figure 7. Efficiency vs. Output Voltage



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# **Typical Performance Characteristics (Continued)**



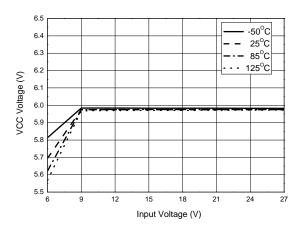
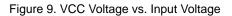
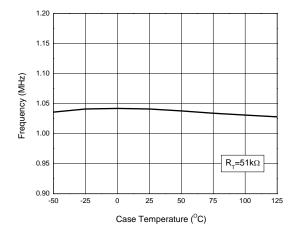


Figure 8. Supply Current vs. Input Voltage





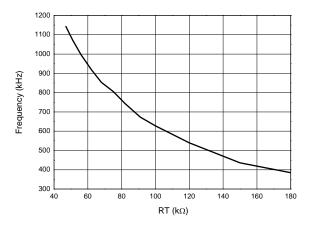


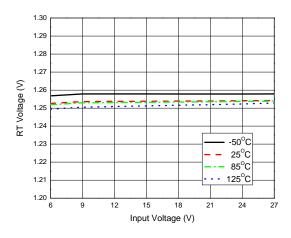
Figure 10. Switching Frequency vs. Case Temperature

Figure 11. Switching Frequency vs. RT Value



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# **Typical Performance Characteristics (Continued)**



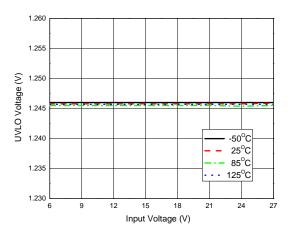
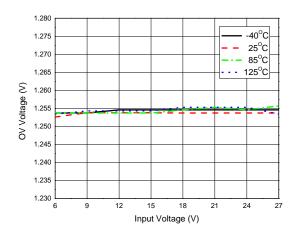


Figure 12. RT Voltage vs. Input Voltage

Figure 13. UVLO Voltage vs. Input Voltage



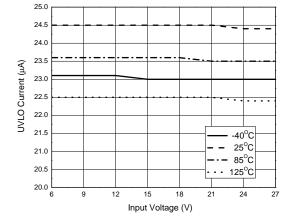


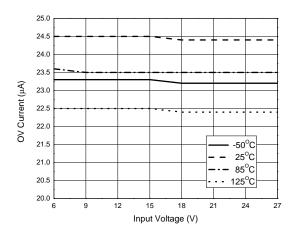
Figure 14. OV Voltage vs. Input Voltage

Figure 15. UVLO Current vs. Input Voltage



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# **Typical Performance Characteristics (Continued)**



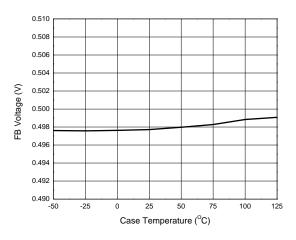
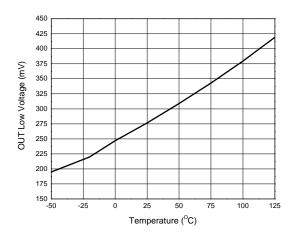


Figure 16. OV Current vs. Input Voltage

Figure 17. Feedback Voltage vs. Case Temperature



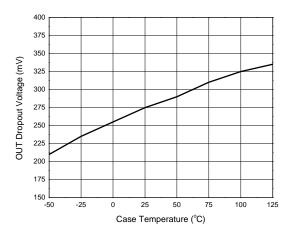


Figure 18. OUT Low Voltage vs. Case Temperature

Figure 19. OUT Dropout Voltage vs. Case Temperature



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# **Application Information Operation**

AP3039 is a boost DC-DC controller with adjustable operation frequency. Current mode control scheme provides excellent line and load regulation. Operation can be best understood by referring to Figure 3.

At the start of each oscillator cycle, the SR latch is set and external power switch Q1 (see Figure 4) turns on and the switch current will increase linearly. The voltage on external sense resistor R<sub>CS</sub> (see Figure 4), connected from CS pin to GND, is proportional to the switch current. This voltage is added to a stabilizing ramp and the result is fed into the non-inversion input of the PWM comparator. When this non-inversion input voltage exceeds inversion input voltage of PWM comparator which is the output voltage of the error amplifier EA, the SR latch is reset and the external power switch turns off. The voltage level at inversion input of PWM comparator sets the peak current level to keep the output voltage in regulation. This voltage level is the amplified signal of the voltage difference between feedback voltage and reference voltage of 0.5V. So, a constant output current can be provided by this operation mode.

#### **Input Under-Voltage Detector**

AP3039 contains an Under Voltage Lock Out (UVLO) circuit. Two resistors R1 and R2 are connected from UVLO pin to ground and VIN pin respectively (see Figure 4). The resistor divider must be designed such that the voltage on the UVLO pin is higher than 1.25V when VIN is in the desired operating range. If the voltage on the pin is below under voltage threshold, all functions of AP3039 are disabled, but the system will remain in a low power standby state. UVLO hysteresis is accomplished through an internal 22µA current source which switched on or off 22µA current into the impedance of the set-point divider. When the UVLO threshold is exceeded, the current source is activated to instantly raise the voltage on the UVLO pin. When the UVLO pin voltage falls below the threshold the current source is turned off, causing the voltage on the UVLO pin to fall. The formula for UVLO can be expresses as

For Input Threshold Voltage  $V_{IN\_THRESHOLD} \!\!=\! 1.25V^*(R1 \!\!+\! R2)\!/R2$ 

For Input Hysteresis Voltage V<sub>IN-HYSTERESIS</sub>=22µA\*R1

#### **Over Voltage Protection**

AP3039 has an over voltage protection (OVP) circuit. The OV Pin is connected to the center tap of R3 and R4 resistor voltage-divider from the high voltage output to GND. (see Figure 21). When the loop is open or the output voltage becomes excessive in any case, result the voltage on OV pin exceeds 1.25V, all functions of AP3039 will be disabled, and the output voltage will fall. OVP hysteresis is accomplished with an internal  $22\mu A$  current source and the operation mode is the same as UVLO. The formula for OVP can be expresses as blow:

For OVP Voltage  $V_{OVP}{=}1.25V*(R3{+}R4)/R4$  For OVP Hysteresis Voltage  $V_{OVP-HYSTERESIS}{=}22\mu A*R3$ 

#### **Frequency Selection**

An external resistor  $R_T$ , connected from RT pin to GND, is used to set the operating frequency (see Figure 4). Operation frequency range is from 400kHz to 1MHz (see Table 1). High frequency operation optimizes the regulator for the smallest component size, while low frequency operation can reduce the switch losses.

**Table 1. Frequency Selection** 

Resistance of $R_T$ ( $k\Omega$ )	Operating Frequency (kHz)
147	400
95	600
68	800
51	1000



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# **Application Information (Continued)**

#### **Soft Start**

AP3039 has a soft start circuit to limit the inrush current during startup. The time of soft start is controlled by an internal 12 $\mu$ A current source and an external soft start capacitor  $C_{SS}$  connected from SS pin to GND (see Figure 21). The effective  $C_{SS}$  voltage for Soft Start is from 0 to 2.3V, the time of Soft Start is:

$$t_{SS} = C_{SS} * 2.3 V / 12 \mu A$$

#### **Typical Application**

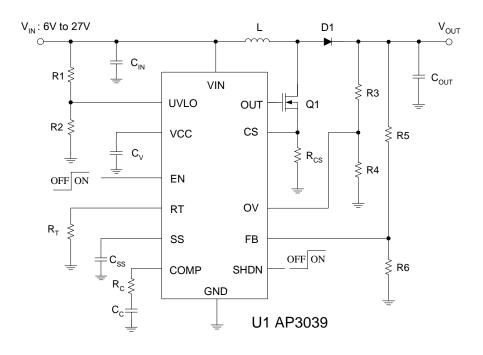
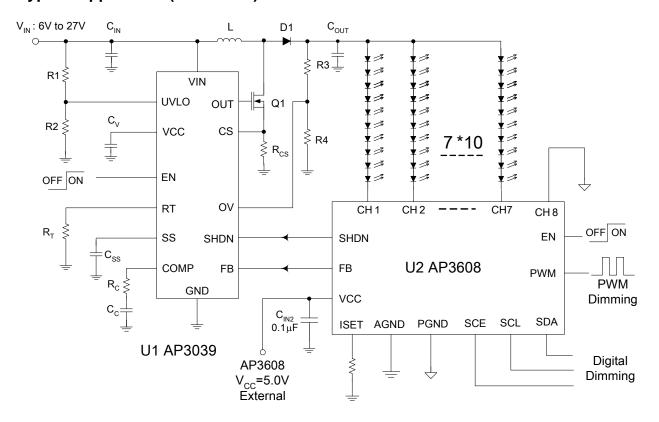


Figure 21. AP3039 Application Circuit



**AP3039** 

#### **Typical Application (Continued)**



C: X5R or X7R Dielectric

L: SUMIDA CDR6D28MN-220NC

Q1: FAIRCHILD FDC5612 D1: FAIRCHILD SS36

Figure 22. AP3039 + AP3608 (Eight Channels Current Sink) Application Circuit

#### **Application Circuit Simple Introduction**

Figure 22 is an application circuit in which the AP3039 works with AP3608 to drive LED array. The AP3608 acts as an eight-channel constant current sink with current match to drive the LEDs.

The SHDN pin and FB pin of AP3608 are the interface terminals to coordinate with the AP3039. The FB pin of AP3608 samples voltage of each channel, and outputs the lowest voltage of all the strings to AP3039.

When there is a shutdown signal on EN pin of AP3608 or all LED channels are inactive, the SHDN pin of AP3608 outputs a low logic signal to turn off AP3039.

If AP3608 is on PWM dimming mode, the SHDN pin of AP3608 outputs a signal to AP3039, which is synchronous with PWM.

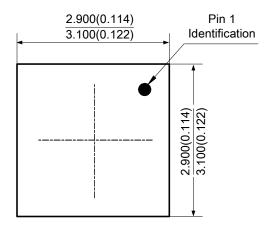


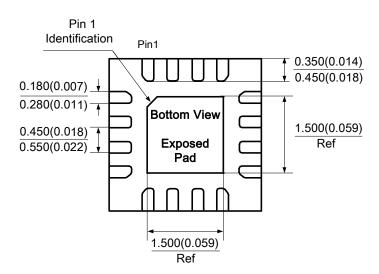
AP3039

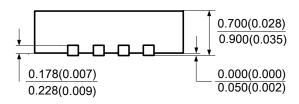
#### **Mechanical Dimensions**

QFN-3x3-16













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