FAIRCHILD
SEMICONDUCTOR®

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FSEZ1216 — Primary-Side-Regulation PWM Integrated Power MOSFET

Features

- Constant-Voltage (CV) and Constant-Current (CC) Control without Secondary-Feedback Circuitry
- Green-Mode Function: PWM Frequency Linearly Decreasing
- Fixed PWM Frequency at 42kHz with Frequency Hopping to Solve EMI Problem
- Cable Compensation in CV Mode
- Low Startup Current: 10µA
- Low Operating Current: 3.5mA
- Peak-Current-Mode Control in CV Mode
- Cycle-by-Cycle Current Limiting
- V_{DD} Over-Voltage Protection with Auto-Restart
- V_{DD} Under-Voltage Lockout (UVLO)
- Fixed Over-Temperature Protection with Latch
- DIP-8 Package Available

Applications

- Battery chargers for cellular phones, cordless phones, PDA, digital cameras, power tools
- Replaces linear transformer and RCC SMPS

Description

This highly integrated PWM controller, FSEZ1216, provides several features to enhance the performance of low-power flyback converters. The proprietary topology of FSEZ1216 enables simplified circuit design for battery charger applications. A low-cost, smaller, and lighter charger results when compared to a conventional design or a linear transformer. The startup current is only 10µA, which allows use of large startup resistance for further power saving.

To minimize the standby power consumption, the proprietary green-mode function provides off-time modulation to linearly decrease PWM frequency under light-load conditions. This green-mode function assists the power supply in meeting power conservation requirements.

Using FSEZ1216, a charger can be implemented with few external components and minimized cost. A typical output CV/CC characteristic is shown in Figure 1.

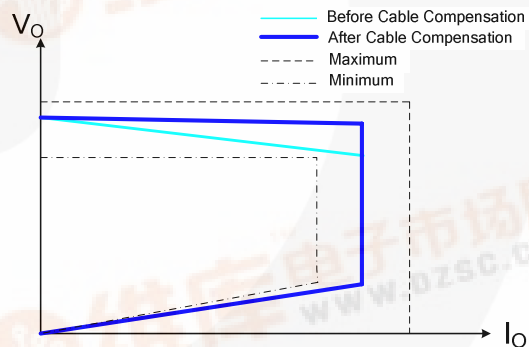FSEZ1216 controller is available in an 8-pin DIP package.

Figure 1.   Typical Output V-I Characteristic

Ordering Information

Part Number	Operating Temperature Range	🌿 Eco Status	Package	Packing Method
FSEZ1216NY	-40°C to +105°C	Green	8-Lead, Dual Inline Package (DIP-8)	Tube

For Fairchild's definition of "green" Eco Status, please visit: http://www.fairchildsemi.com/company/green/rohs_green.html.
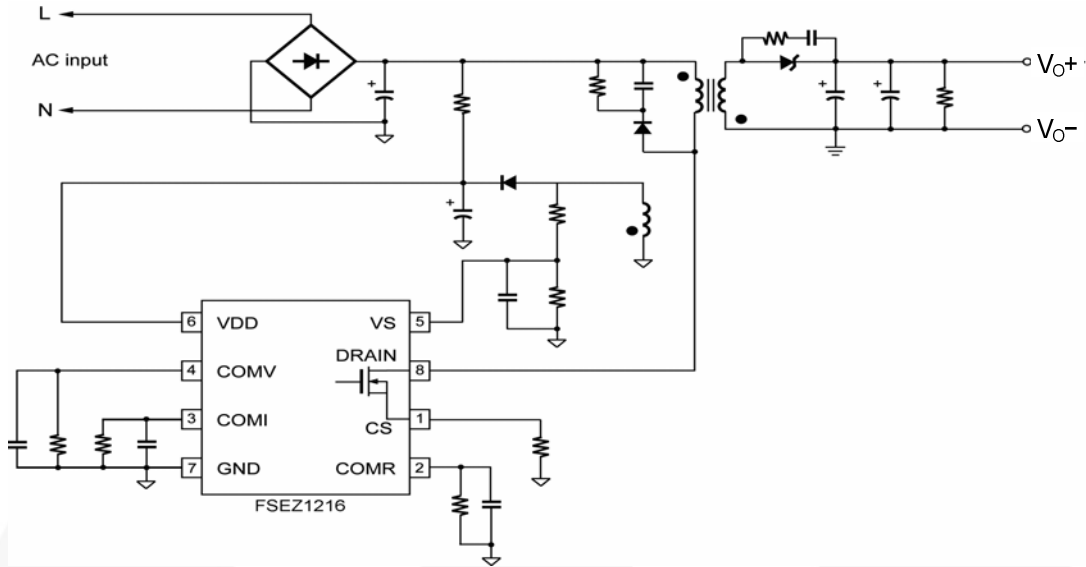
Application Diagram
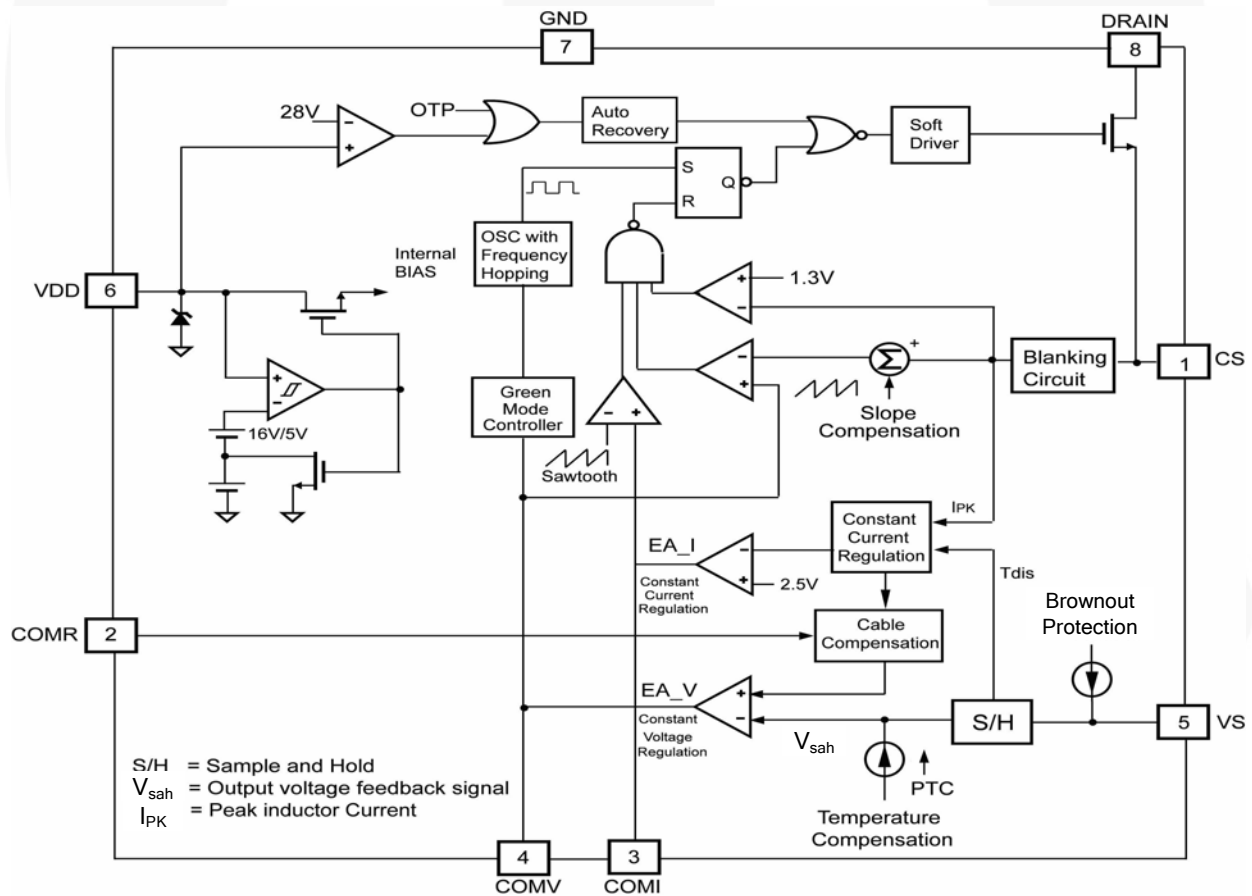

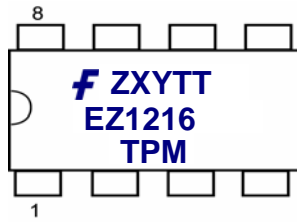**Figure 2.   Typical Application**

Internal Block Diagram

Figure 3.   Functional Block Diagram

Marking Information

ZXYTT
EZ1216
TPM

8

1

F- Fairchild logo
Z- Plant Code
X- 1 digit year code
Y- 1 digit week code
TT: 2 digits die run code
T: Package type (N=DIP)
P: Z: Pb free, Y: Green package
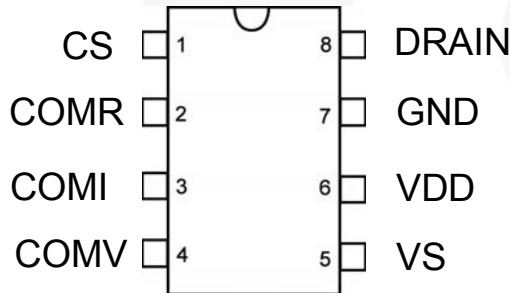M: Manufacture flow code

Figure 4.   Top Mark

Pin Configuration

CS — 1   8 — DRAIN
COMR — 2   7 — GND
COMI — 3   6 — VDD
COMV — 4   5 — VS

Figure 5.   Pin Configuration

Pin Definitions

Pin #	Name	Description
1	CS	Analog input, current sense. Connected to a current-sense resistor for peak-current-mode control in CV mode. The current-sense signal is also provided for output-current regulation in CC mode.
2	COMR	Analog output, cable compensation. Connect a resistor between COMR and GND for cable loss compensation in CV mode.
3	COMI	Analog output, current compensation. Output of the current error amplifier. Connect a capacitor between COMI pin and GND for frequency compensation.
4	COMV	Analog output, voltage compensation. Output of the voltage error amplifier. Connect a capacitor between COMV pin and GND for frequency compensation.
5	VS	Analog input, voltage sense. Output-voltage-sense input for output-voltage regulation.
6	VDD	Supply, power supply.
7	GND	Voltage reference, ground.
8	DRAIN	Driver output, power MOSFET drain. This pin is the high-voltage power MOSFET drain.

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter		Min.	Max.	Unit
V_{VDD}	DC Supply Voltage[1]			30	V
V_{VS}	VS Pin Input Voltage		-0.3	7.0	V
V_{CS}	CS Pin Input Voltage		-0.3	7.0	V
V_{COMV}	Voltage Error Amplifier Output Voltage		-0.3	7.0	V
V_{COMI}	Voltage Error Amplifier Output Voltage		-0.3	7.0	V
V_{DS}	Drain-Source Voltage			600	V
I_D	Continuous Drain Current	T_C=25°C		1	A
		T_C=100°C		0.6	A
I_{DM}	Pulsed Drain Current			4	A
E_{AS}	Single Pulse Avalanche Energy			33	mJ
I_{AR}	Avalanche Current			1	A
P_D	Power Dissipation ($T_A < 50$°C)			800	mW
Θ_{JA}	Thermal Resistance (Junction-to-Air)			113	°C /W
Θ_{JC}	Thermal Resistance (Junction-to-Case)			67	°C /W
T_J	Operating Junction Temperature			+150	°C
T_{STG}	Storage Temperature Range		-55	+150	°C
T_L	Lead Temperature (Wave Soldering or IR, 10 seconds)			+260	°C
ESD	Electrostatic Discharge Capability, Human Body Model, JEDEC: JESD22-A114			2.5	KV
	Electrostatic Discharge Capability, Charged Device Model, JEDEC: JESD22-C101			1250	V

Note:

1. All voltage values, except differential voltages, are given with respect to GND pin.

Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
T_A	Operating Ambient Temperature		-40		+105	°C

Electrical Characteristics

V_{DD}=15V and T_A=25°C unless otherwise specified.

Symbol	Parameter		Conditions	Min.	Typ.	Max.	Units
V_{DD} Section							
V_{OP}	Continuously Operating Voltage					25	V
V_{DD-ON}	Turn-On Threshold Voltage			15	16	17	V
V_{DD-OFF}	Turn-Off Threshold Voltage			4.5	5.0	5.5	V
I_{DD-ST}	Startup Current		0< V_{DD} < V_{DD-ON}-0.16V		10	20	µA
I_{DD-OP}	Operating Current		V_{DD}=20V, f_S=f_{OSC}, V_{VS}=2V, V_{CS}=3V, C_L=1nF		3.5	5.0	mA
$I_{DD-GREEN}$	Green Mode Operating Supply Current		V_{DD}=20V, V_{VS}=2.7V, f_S=$f_{OSC-N-MIN}$, V_{CS}=0V, C_L=1nF, V_{COMV}=0V		1	2	mA
V_{DD-OVP}	V_{DD} Over-Voltage-Protection Level		V_{CS}=3V, V_{VS}=2.3V	27	28	29	V
$t_{D-VDDOVP}$	V_{DD} Over-Voltage-Protection Debounce Time		fs= f_{OSC}, V_{VS}=2.3V	100	250	400	µs
Oscillator Section							
f_{OSC}	Frequency	Center Frequency	T_A=25°C	39	42	45	KHz
		Frequency Hopping Range	T_A=25°C	±1.8	±2.6	±3.6	
t_{FHR}	Frequency Hopping Period		T_A=25°C		3		ms
$f_{OSC-N-MIN}$	Minimum Frequency at No Load		V_{VS}=2.7V, V_{COMV}=0V		550		Hz
$f_{OSC-CM-MIN}$	Minimum Frequency at CCM		V_{VS}=2.3V, V_{CS}=0.5V		20		KHz
f_{DV}	Frequency Variation vs. V_{DD} Deviation		V_{DD}=10V to 25V			5	%
f_{DT}	Frequency Variation vs. Temperature Deviation		T_A=-40°C to +85°C			15	%
Voltage-Sense Section							
I_{VS-UVP}	Sink Current for Brownout Protection		R_{VS}=20KΩ		125		µA
I_{tc}	IC Compensation Bias Current				9.5		µA
$V_{BIAS-COMV}$	Adaptive Bias Voltage Dominated by V_{COMV}		V_{COMV}=0V, T_A=25°C, R_{VS}=20KΩ		1.4		V
Current-Sense Section							
t_{PD}	Propagation Delay to GATE Output				100	200	ns
t_{MIN-N}	Minimum On Time at No Load		V_{VS}=-0.8V, R_S=2KΩ, V_{COMV}=1V		1100		ns
t_{MINCC}	Minimum On Time in CC Mode		V_{VS}=0V, V_{COMV}=2V		400		ns
D_{SAW}	Duty Cycle of SAW Limiter				40		%
V_{TH}	Threshold Voltage for Current Limit				1.3		V

Electrical Characteristics

V_{DD}=15V and T_A=25°C unless otherwise specified.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
Voltage-Error-Amplifier Section						
V_{VR}	Reference Voltage		2.475	2.500	2.525	V
V_N	Green Mode Starting Voltage on COMV Pin	f_S=f_{OSC}-2KHz, V_{VS}=2.3V		2.8		V
V_G	Green Mode Ending Voltage on COMV Pin	f_S=1KHz		0.8		V
I_{V-SINK}	Output Sink Current	V_{VS}=3V, V_{COMV}=2.5V		90		µA
$I_{V-SOURCE}$	Output Source Current	V_{VS}=2V, V_{COMV}=2.5V		90		µA
V_{V-HGH}	Output High Voltage	V_{VS}=2.3V	4.5			V
Current-Error-Amplifier Section						
V_{IR}	Reference Voltage		2.475	2.500	2.525	V
I_{I-SINK}	Output Sink Current	V_{CS}=3V, V_{COMI}=2.5V		55		µA
$I_{I-SOURCE}$	Output Source Current	V_{CS}=0V, V_{COMI}=2.5V		55		µA
V_{I-HGH}	Output High Voltage	V_{CS}=0V	4.5			V
Cable Compensation Section						
V_{COMR}	Variation Test Voltage on COMR Pin for Cable Compensation	R_{COMR}=100k		0.735		V
Internal MOSFET Section						
DCY_{MAX}	Maximum Duty Cycle			75		%
BV_{DSS}	Drain-Source Breakdown Voltage	I_D=250µA, V_{GS}=0V	600			V
$\Delta BV_{DSS}/\Delta T_J$	Breakdown Voltage Temperature Coefficient	I_D=250µA, Referenced to 25°C		0.6		V/°C
I_S	Maximum Continuous Drain-Source Diode Forward Current				1	A
I_{SM}	Maximum Pulsed Drain-Source Diode Forward Current				4	A
$R_{DS(ON)}$	Static Drain-Source On-Resistance	I_D=0.5A, V_{GS}=10V		9.3	11.5	Ω
I_{DSS}	Drain-Source Leakage Current	V_{DS}=600V, V_{GS}=0V, T_C=25°C			1	µA
		V_{DS}=480V, V_{GS}=0V, T_C=100°C			10	µA
t_{D-ON}	Turn-On Delay Time[2,3]	V_{DS}=300V, I_D=1.1A, R_G=25Ω		7	24	ns
t_r	Rise Time			21	52	ns
t_{D-OFF}	Turn-Off Delay Time			13	36	ns
t_f	Fall Time			27	64	ns
C_{ISS}	Input Capacitance	V_{GS}=0V, V_{DS}=25V, f_S=1MHz		130	170	pF
C_{OSS}	Output Capacitance			19	25	pF
Over-Temperature-Protection Section						
T_{OTP}	Threshold Temperature for OTP[4]			+140		°C

Notes:
2. Pulse test: pulse width $\leqq$ 300µs, duty cycle $\leqq$ 2%.
3. Essentially independent of operating temperature.
4. When over-temperature protection is activated, the power system enters latch mode and output is disabled.
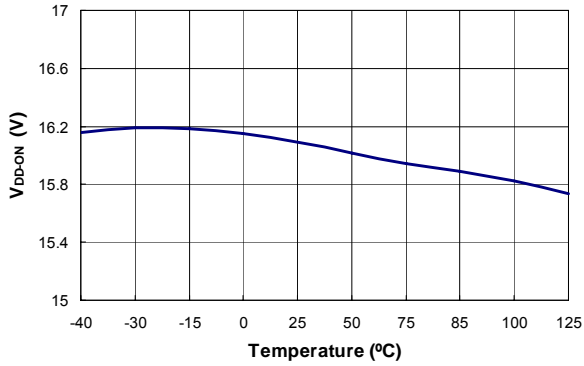
Typical Performance Characteristics

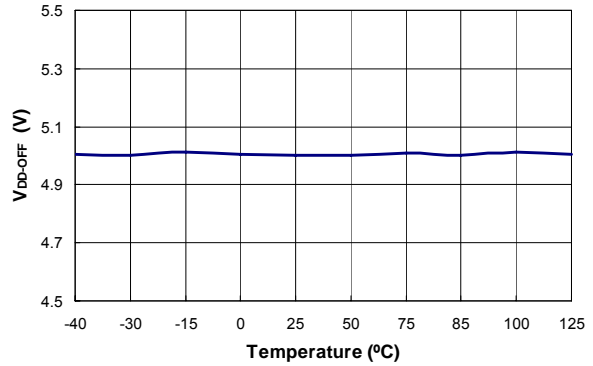**Figure 6.** Turn-on Threshold Voltage (V_{DD-ON})
vs. Temperature

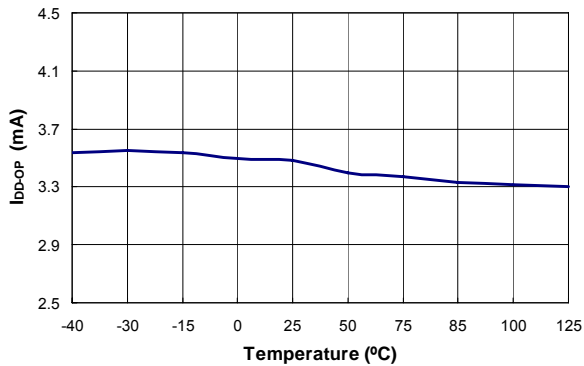**Figure 7.** Turn-off Threshold Voltage (V_{DD-OFF})
vs. Temperature

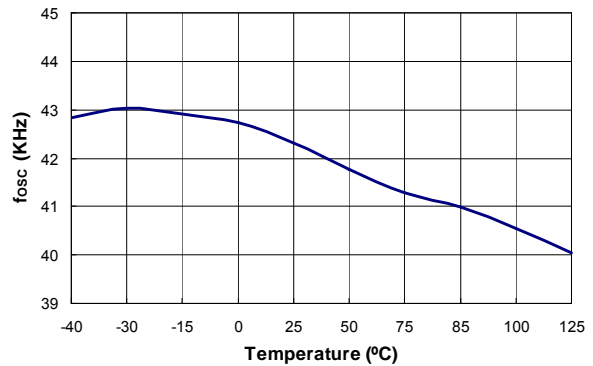**Figure 8.** Operating Current (I_{DD-OP})
vs. Temperature

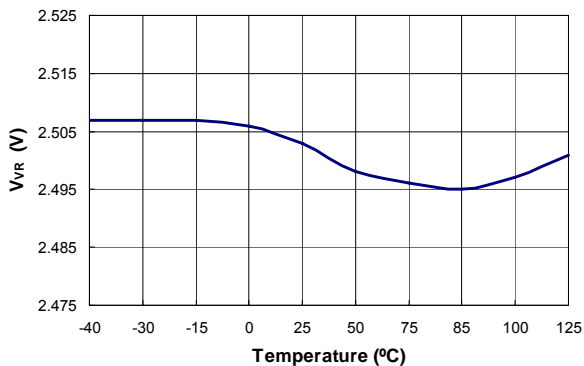**Figure 9.** Center Frequency (f_{OSC}) vs. Temperature

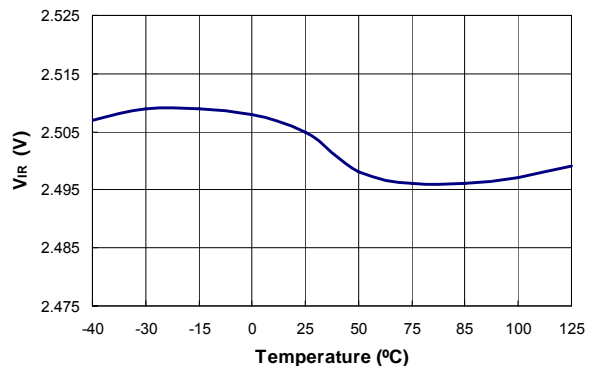**Figure 10.** Reference Voltage (V_{VR}) vs. Temperature

Figure 11. Reference Voltage (V_{IR}) vs. Temperature
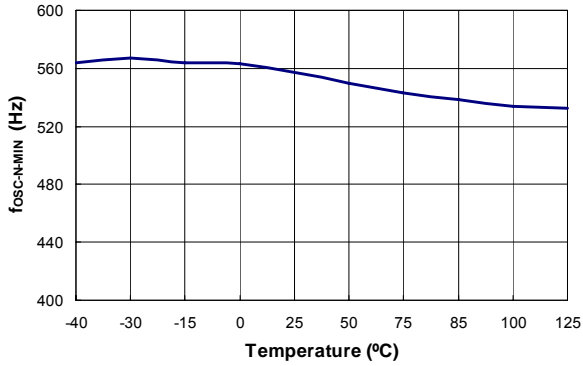
Typical Performance Characteristics (Continued)

Figure 12. Minimum Frequency at No Load ($f_{OSC-N-MIN}$) vs. Temperature
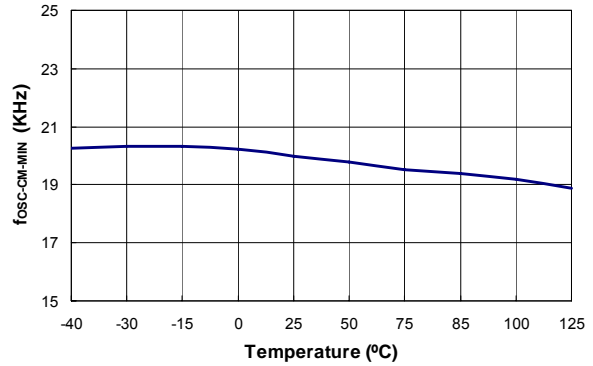
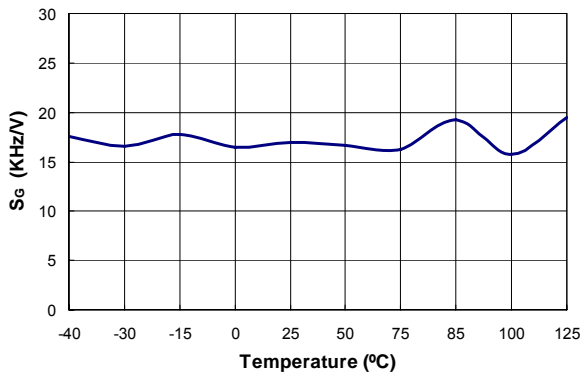**Figure 13. Minimum Frequency at CCM ($f_{OSC-CM-MIN}$) vs. Temperature**

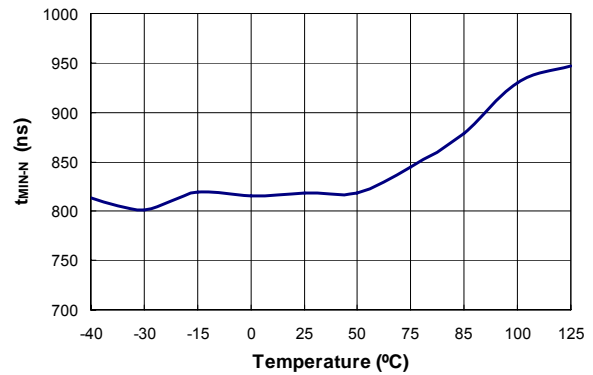**Figure 14. Green Mode Frequency Decreasing Rate (S_G) vs. Temperature**

Figure 15. Minimum On Time at No Load (t_{MIN-N}) vs. Temperature
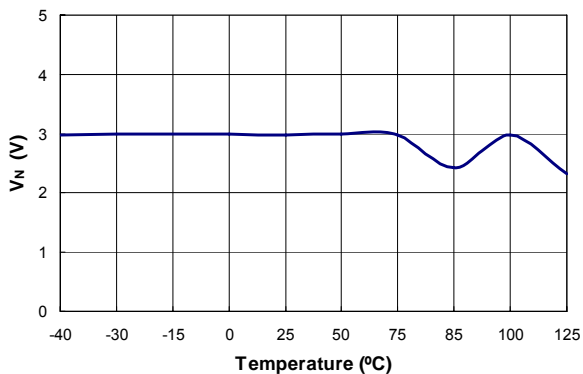
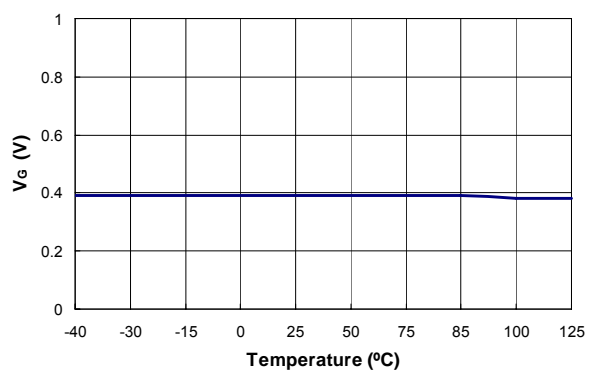**Figure 16. Green Mode Starting Voltage on COMV Pin (V_N) vs. Temperature**

Figure 17. Green Mode Ending Voltage on COMV Pin (V_G) vs. Temperature

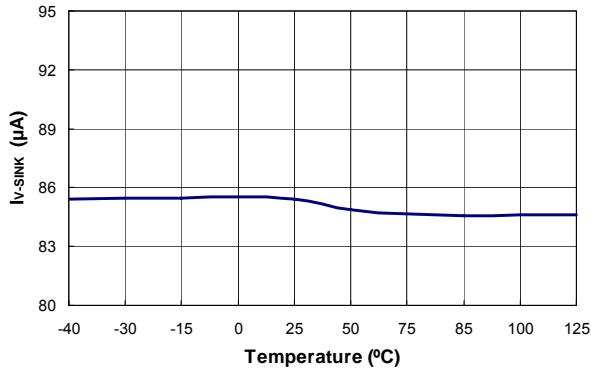# Typical Performance Characteristics (Continued)

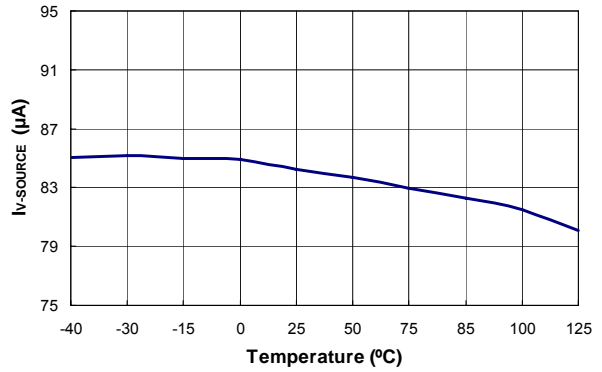**Figure 18. Output Sink Current (I_{V-SINK}) vs. Temperature**

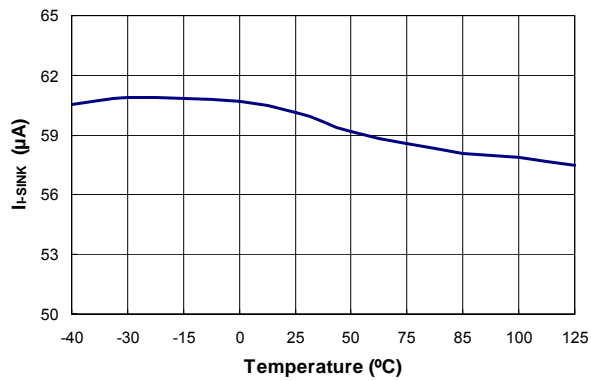**Figure 19. Output Source Current ($I_{V-SOURCE}$) vs. Temperature**

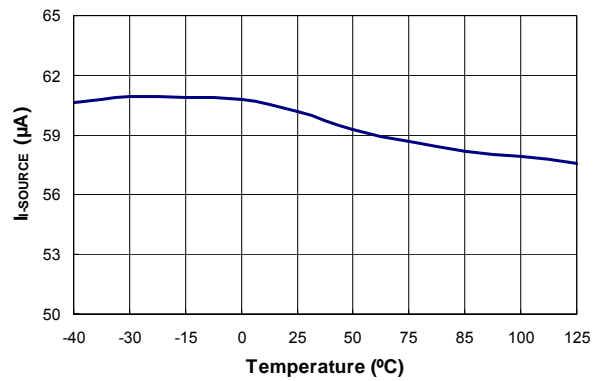**Figure 20. Output Sink Current (I_{I-SINK}) vs. Temperature**

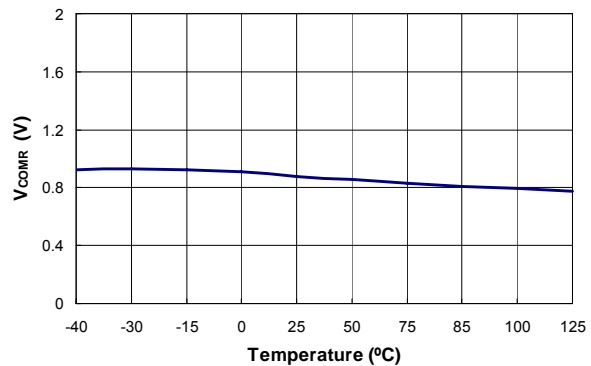**Figure 21. Output Source Current ($I_{I-SOURCE}$) vs. Temperature**

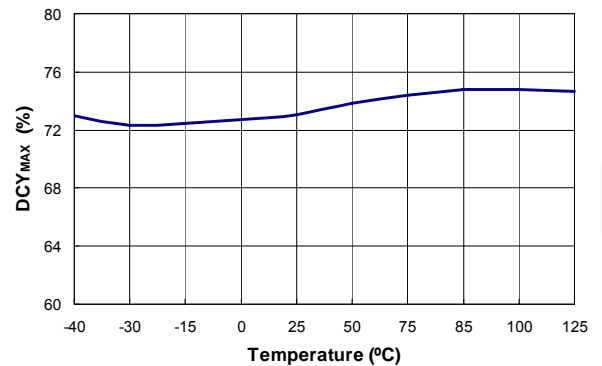**Figure 22. Variation Test Voltage on COMR Pin for Cable Compensation (V_{COMR}) vs. Temperature**

Figure 23. Maximum Duty Cycle (DCY_{MAX}) vs. Temperature

Functional Description

The proprietary topology of FSEZ1216 enables simplified circuit design for battery charger applications. Without secondary feedback circuitry, the CV and CC control can be achieved accurately. As shown in Figure 24, with the frequency-hopping and PWM operation, EMI problems can be solved by using minimized filter components. FSEZ1216 also provides many protection functions. The VDD pin is equipped with over-voltage protection, and under-voltage lockout. Pulse-by-pulse current limiting and CC control ensure over-current protection at heavy loads. The internal over-temperature-protection function shuts down the controller with latch when over heated.
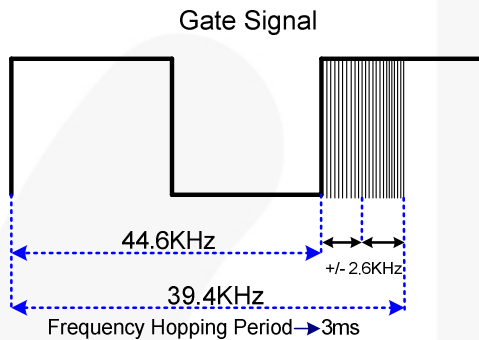
Gate Signal

Figure 24. Frequency Hopping

Startup Current

The startup current is only 10µA, which allows a startup resistor with a high resistance and a low-wattage to supply the startup power for the controller. A 1.5MΩ, 0.25W, startup resistor and a 10µF/25V V_{DD} hold-up capacitor are sufficient for an AC-to-DC power adapter with a wide input range ($100V_{AC}$ to $240V_{AC}$).

Operating Current

The operating current has been reduced to 3.5mA, which results in higher efficiency and reduces the V_{DD} hold-up capacitance requirement. Once FSEZ1216 enter "deep" green mode, the operating current is reduced to 1.2mA, assisting the power supply in meeting the power conservation requirements.

Green-Mode Operation

Figure 25 shows the characteristics of the PWM frequency vs. the output voltage of the error amplifier (V_{COMV}). The FSEZ1216 uses the positive, proportional, output load parameter (V_{COMV}) as an indication of the output load for modulating the PWM frequency. In heavy load conditions, the PWM frequency is fixed at 42KHz. Once V_{COMV} is lower than V_N, the PWM frequency starts to linearly decrease from 42KHz to 550Hz, providing further power savings and meeting international power conservation requirements.
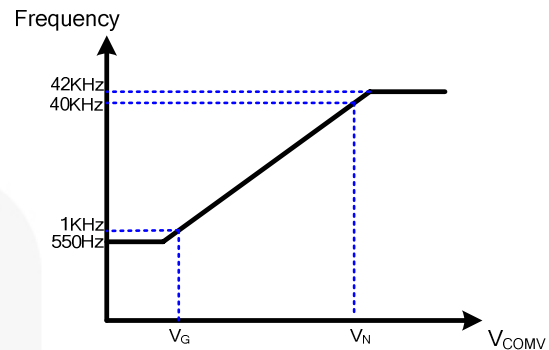

Figure 25. Green-Mode Operation (Frequency vs. V_{COMV})

Constant Voltage (CV) and Constant Current (CC) Operation

An innovative technique of the FSEZ1216 can accurately achieve CV/CC characteristic output without secondary-side voltage or current-feedback circuitry. A feedback signal for CV/CC operation from the reflected voltage across the primary auxiliary winding is proportional to secondary winding, so provides controller the feedback signal from secondary side and achieves constant voltage output property. In constant-current output operation, this voltage signal is detected and examined by the precise constant current regulation controller, which then determines the on-time of the MOSFET to control input power and provides constant current output property. With feedback voltage V_{CS} across current sense resistor, the controller can obtain input power of power supply. Therefore, the region of constant current output operation can be adjusted by current sense resistor.

Temperature Compensation

Built-in temperature compensation provides better constant voltage regulation at different ambient temperatures. This internal compensation current is a positive temperature coefficient (PTC) current that can compensate the forward-voltage drop of the secondary diode of varying with temperature. This variation caused output voltage to rise at high temperature.

Leading-Edge Blanking (LEB)

Each time the power MOSFET is switched on, a turn-on spike occurs at the sense-resistor. To avoid premature termination of the switching pulse, a leading-edge blanking time is built in. Conventional RC filtering can therefore be omitted. During this blanking period, the current-limit comparator is disabled and cannot switch off gate driver.

Functional Description (Continued)

Under-Voltage Lockout (UVLO)

The turn-on and turn-off thresholds of the FSEZ1216 are fixed internally at 16V and 5V. During startup, the hold-up capacitor must be charged to 16V through the startup resistor to enable the FSEZ1216. The hold-up capacitor continues to supply V_{DD} until power can be delivered from the auxiliary winding of the main transformer. V_{DD} must not drop below 5V during startup. The UVLO hysteresis window ensures the hold-up capacitor is adequate to supply V_{DD} during startup.

V_{DD} Over-Voltage Protection (OVP)

V_{DD} over-voltage protection prevents damage due to over-voltage conditions. When the V_{DD} voltage exceeds 28V due to abnormal conditions, PWM pulses are disabled until the V_{DD} voltage drops below the UVLO, then starts again. Over-voltage conditions are usually caused by open feedback loops.

Over-Temperature Protection (OTP)

The FSEZ1216 has a built-in temperature sensing circuit to shut down PWM output once the junction temperature exceeds 140°C. While PWM output is shut down, the V_{DD} voltage gradually drops to the UVLO voltage. Some of the internal circuits are shut down and V_{DD} gradually starts increasing again. When V_{DD} reaches 16V, all the internal circuits, including the temperature-sensing circuit, start operating normally. If the junction temperature is still higher than 140°C, the PWM controller shuts down immediately. This situation continues until the temperature drop below 110°C.

Gate Output

The BiCMOS output stage is a fast totem-pole gate driver. Cross conduction has been avoided to minimize heat dissipation, increase efficiency, and enhance reliability. The output driver is clamped by an internal 15V Zener diode to protect the internal power MOSFET transistors against undesired over-voltage gate signals.

Built-in Slope Compensation

The sensed voltage across the current-sense resistor is used for current-mode control and pulse-by-pulse current limiting. Built-in slope compensation improves stability and prevents sub-harmonic oscillations due to peak-current-mode control. A synchronized, positively sloped ramp is built-in at each switching cycle.

Noise Immunity

Noise from the current sense or the control signal can cause significant pulse-width hopping, particularly in continuous-conduction mode. While slope compensation helps alleviate these problems, further precautions should still be taken. Good placement and layout practices should be followed. Avoiding long PCB traces and component leads, locating compensation and filter components near the FSEZ1216.
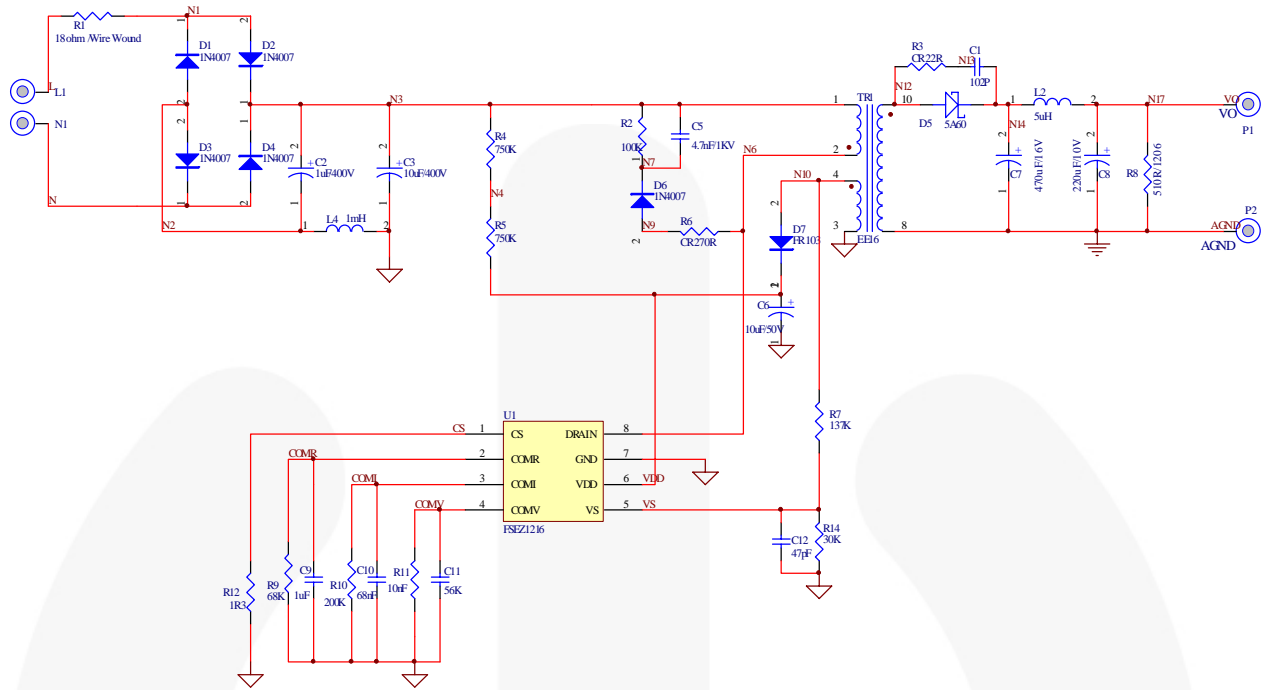
Applications Information

Figure 26. 5W (5V/1A) Application Circuit

BOM

Designator	Part Type	Designator	Part Type
D1, D2, D3, D4, D6	1N4007	R2	R 100KΩ
D5	SB560	R3	R 22Ω
D7	FR103	R4, R5	R 750KΩ
C1	1nF	R6	R 270Ω
C2	EC 1μF/400V	R7	R 137KΩ
C3	EC 10μF/400V	R8	R 510Ω
C5	4.7nF/1KV	R9	R 68KΩ
C6	EC 10μF/50V	R10	R 200KΩ
C7	EC 470μF/16V	R11	R 56KΩ
C8	EC 220μF/10	R12	R 1.3Ω
C9	1μF	R14	R 30Ω
C10	68nF	L2	5μH
C11	10nF	L4	1mH
C12	47pF	T1	EE16 (1.5mH)
R1	R 18Ω	U1	IC FSEZ1216

Physical Dimensions
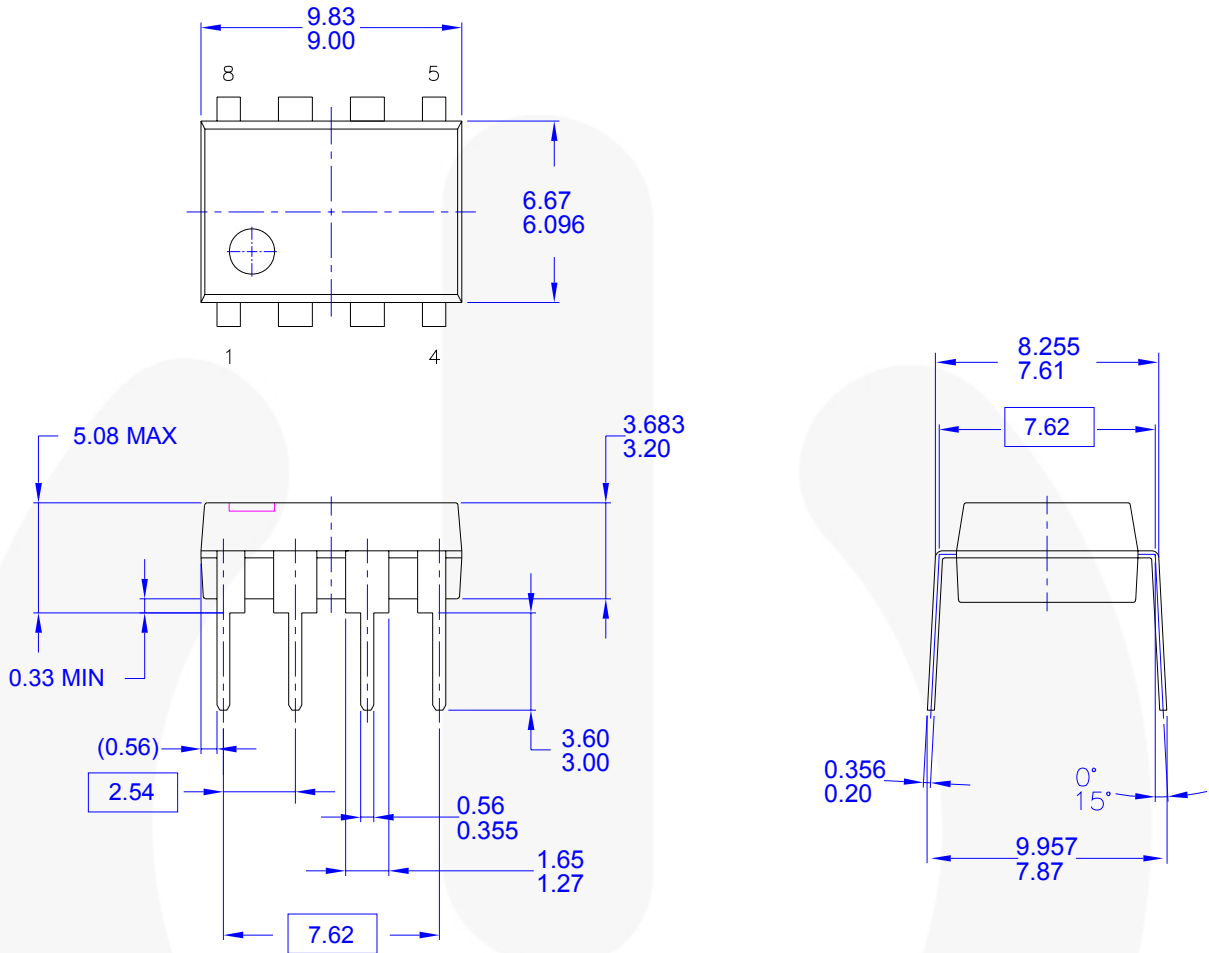
NOTES: UNLESS OTHERWISE SPECIFIED
A) THIS PACKAGE CONFORMS TO
   JEDEC MS-001 VARIATION BA
B) ALL DIMENSIONS ARE IN MILLIMETERS.
C) DIMENSIONS ARE EXCLUSIVE OF BURRS,
   MOLD FLASH, AND TIE BAR EXTRUSIONS.
D) DIMENSIONS AND TOLERANCES PER
   ASME Y14.5M-1994
E) DRAWING FILENAME AND REVSION: MKT-N08FREV2.

Figure 27. 8-Lead, Dual Inline Package (DIP-8)

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Definition of Terms

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