

# **KBA2338**

## 3W Filterless Class-D Audio Power Amplifier

## **General Description**

The kB2338A is a high efficiency, 3W mono class-D audio power amplifier. A low noise, filterless PWM architecture eliminates the output filter, reducing external component count, system cost, and simplifying design.

Operating in a single 5V supply, kB2338A is capable of driving  $4\Omega$  speaker load at a continuous average output of 3W/10% THD+N or 2W/1% THD+N. The kB2338A has high efficiency with speaker load compared to a typical class AB amplifier. With a 3.6V supply driving an  $8\Omega$  speaker , the efficiency for a 400mW power level is 88%.

In cellular handsets, the earpiece, speaker phone, and melody ringer can each be driven by the kB2338A. The gain of kB2338A is externally configurable which allows independent gain control from multiple sources by summing signals from seperate sources.

The kB2338A is available in space-saving WCSP and DFN packages.

## **Features**

- Unique Modulation Scheme Reduces EMI Emissions
- Efficiency at 3.6V With an 8- $\Omega$  Speaker:
  - 88% at 400 mW
  - 80% at 100 mW
- Low 2.38-mA Quiescent Current and 0.5-μA Shutdown Current
- 2.5V to 6.0V Wide Supply Voltage
- Optimized PWM Output Stage Eliminates LC Output Filter
- Improved PSRR (-72 dB) Eliminates Need for a Voltage Regulator
- Fully Differential Design Reduces RF Rectification and Eliminates Bypass Capacitor
- Improved CMRR Eliminates Two Input Coupling Capacitors
- Internally Generated 250-kHz Switching Frequency
- Integrated Pop and Click Suppression Circuitry
- 1.5mm × 1.5mm Wafer Chip Scale Package (WCSP) and 3mm × 3mm DFN-8 package
- RoHS compliant and 100% lead(Pb)-free

# 9-PIN FLIP-CHIP WCSP MARKING DIAGRAMS A3 C3 A20 © 5 A1 C1



## **Applications**

- Cellular Phone
- Portable Electronic Devices
- PDAs and Smart Phones
- Portable Computer

# **Ordering Information**

Order Number	Package Type	Marking	Operating Temperature range
-kBA2338B-UDFN	<del>UDFN-8</del>		
kBA2338-WCSP	WCSP-9	A20 5	-40 °C to 85°C

# **Pin Configuration**

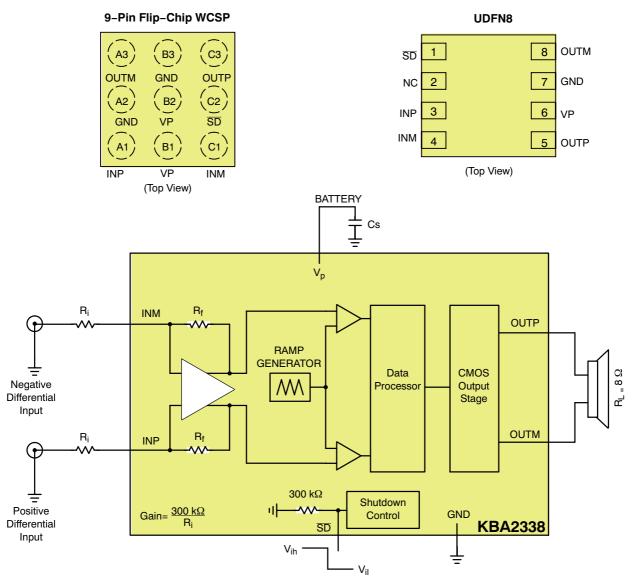


Figure 1. Typical Application

# **Pin Description**

Pin No.				
WCSP	WCSP UDFN8 Symbol Type		Туре	Description
A1	3	INP	1	Positive Differential Input.
A2	7	GND	I	Analog Ground.
A3	8	OUTM	0	Negative BTL Output.
B1		V <sub>p</sub>	I	Analog Positive Supply. Range: 2.5 V – 6.0 V.
B2	6	V <sub>p</sub>	1	Power Analog Positive Supply. Range: 2.5 V – 6.0 V.
В3	7	GND	I	Analog Ground.
C1	4	INM	I	Negative Differential Input.
C2	1	SD	I	The device enters in Shutdown Mode when a low level is applied on this pin. An internal 300 k $\Omega$ resistor will force the device in shutdown mode if no signal is applied to this pin. It also helps to save space and cost.
С3	5	OUTP	0	Positive BTL Output.

# **Absolute Maximum Ratings**

Symbol	Rating	Max	Unit		
V <sub>p</sub>	Supply Voltage	Active Mode Shutdown Mode	6.0 7.0	V	
V <sub>in</sub>	Input Voltage		-0.3 to V <sub>CC</sub> +0.3	V	
I <sub>out</sub>	Max Output Current (Note 1)		1.5	Α	
$P_d$	Power Dissipation (Note 2)	Power Dissipation (Note 2)			
T <sub>A</sub>	Operating Ambient Temperature	-40 to +85	°C		
TJ	Max Junction Temperature	150	°C		
T <sub>stg</sub>	Storage Temperature Range		-65 to +150	°C	
$R_{ hetaJA}$	Thermal Resistance Junction-to-Air	9-Pin Flip-Chip UDFN8	90 (Note 3) 50	°C/W	
-	ESD Protection Human Body Model (HBM) (Note 4) Machine Model (MM) (Note 5)		> 2000 > 200	V	
-	Latchup Current @ T <sub>A</sub> = 85°C (Note 6)	9-Pin Flip-Chip UDFN8	±70 ±100	mA	
MSL	Moisture Sensitivity (Note 7)		Level 1		

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

- 1. The device is protected by a current breaker structure. See "Current Breaker Circuit" in the Description Information section for more
- 2. The thermal shutdown is set to 160°C (typical) avoiding irreversible damage to the device due to power dissipation.
- 3. For the 9–Pin Flip–Chip CSP package, the  $R_{\theta JA}$  is highly dependent of the PCB Heatsink area. For example,  $R_{\theta JA}$  can equal 195°C/W with 50 mm<sup>2</sup> total area and also 135°C/W with 500 mm<sup>2</sup>. When using ground and power planes, the value is around 90°C/W, as specified in table.
- 4. Human Body Model: 100 pF discharged through a 1.5 kΩ resistor following specification JESD22/A114. On 9-Pin Flip-Chip, B2 Pin (V<sub>P</sub>) is qualified at 1500 V.
- 5. Machine Model: 200 pF discharged through all pins following specification JESD22/A115.
  6. Latchup Testing per JEDEC Standard JESD78.
- 7. Moisture Sensitivity Level (MSL): 1 per IPC/JEDEC standard: J-STD-020A.

# **Electrical Characteristics**

(Limits apply for  $T_A = +25^{\circ}C$  unless otherwise noted) (WCSP)

Characteristic	Symbol	Conditions	Min	Тур	Max	Unit
Operating Supply Voltage	V <sub>p</sub>	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	2.5	-	6.0	V
Supply Quiescent Current	I <sub>dd</sub>	$V_p$ = 3.6 V, $R_L$ = 8.0 $\Omega$ $V_p$ = 5.5 V, No Load $V_p$ from 2.5 V to 5.5 V, No Load	-	2.15 2.61	-	mA
		$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$	_	_	4.6	
Shutdown Current	I <sub>sd</sub>	$V_p = 4.2 \text{ V}$ $T_A = +25^{\circ}\text{C}$ $T_A = +85^{\circ}\text{C}$	_	0.42 0.45	0.8	μΑ
		V <sub>p</sub> = 5.5 V T <sub>A</sub> = +25°C T <sub>A</sub> = +85°C	_ 	0.8	1.5	μΑ
Shutdown Voltage High	V <sub>sdih</sub>	- -	1.2	-	_	V
Shutdown Voltage Low	V <sub>sdil</sub>		-	_	0.4	V
Switching Frequency	F <sub>sw</sub>	$V_p$ from 2.5 V to 5.5 V $T_A = -40^{\circ}\text{C}$ to +85°C	190	250	310	kHz
Gain	G	$R_L = 8.0 \Omega$	285 kΩ R <sub>i</sub>	300 kΩ R <sub>i</sub>	<u>315 kΩ</u> R <sub>i</sub>	V V
Output Impedance in Shutdown Mode	Z <sub>SD</sub>	-	_	300	-	Ω
Resistance from SD to GND	Rs	-	_	300	-	kΩ
Output Offset Voltage	Vos	V <sub>p</sub> = 5.5 V	_	6.0	-	mV
Turn On Time	Ton	V <sub>p</sub> from 2.5 V to 5.5 V	-	9.0	-	ms
Turn Off Time	Toff	V <sub>p</sub> from 2.5 V to 5.5 V	-	5.0	-	ms
Thermal Shutdown Temperature	Tsd	-	-	160	-	°C
Output Noise Voltage	Vn	V <sub>p</sub> = 3.6 V, f = 20 Hz to 20 kHz no weighting filter with A weighting filter	_ _	65 42	- -	μVrms
RMS Output Power	Po	$\begin{aligned} R_L = 8.0 \ \Omega, \ f = 1.0 \ \text{kHz}, \ THD + N < 1\% \\ V_p = 2.5 \ \text{V} \\ V_p = 3.0 \ \text{V} \\ V_p = 3.6 \ \text{V} \\ V_p = 4.2 \ \text{V} \\ V_p = 5.0 \ \text{V} \end{aligned}$	- - - -	0.32 0.48 0.7 0.97 1.38	- - - -	W
		$\begin{array}{c} R_L = 8.0~\Omega,  f = 1.0~\text{kHz},  \text{THD+N} < 10\% \\ V_p = 2.5~\text{V} \\ V_p = 3.0~\text{V} \\ V_p = 3.6~\text{V} \\ V_p = 4.2~\text{V} \\ V_p = 5.0~\text{V} \end{array}$	- - - -	0.4 0.59 0.87 1.19 1.7	- - - -	W
		$R_L = 4.0 \ \Omega, \ f = 1.0 \ kHz, \ THD+N < 1\%$ $V_p = 2.5 \ V$ $V_p = 3.0 \ V$ $V_p = 3.6 \ V$ $V_p = 4.2 \ V$ $V_p = 5.0 \ V$	- - - -	0.49 0.72 1.06 1.62 2.12	- - - -	W
		$R_L = 4.0 \ \Omega, \ f = 1.0 \ \text{kHz}, \ \text{THD+N} < 10\%$ $V_p = 2.5 \ \text{V}$ $V_p = 3.0 \ \text{V}$ $V_p = 3.6 \ \text{V}$ $V_p = 4.2 \ \text{V}$ $V_p = 5.0 \ \text{V}$	- - - -	0.6 0.9 1.33 2.0 2.63	- - - -	W

# **Electrical Characteristics**

(Limits apply for  $T_A = +25^{\circ}C$  unless otherwise noted) (WCSP)

Characteristic	Symbol	Conditions	Min	Тур	Max	Unit
Efficiency	-	$R_L = 8.0 \Omega$ , f = 1.0 kHz $V_p = 5.0 V$ , $P_{out} = 1.2 W$ $V_p = 3.6 V$ , $P_{out} = 0.6 W$	- -	91 90	- -	%
		$R_L = 4.0 \Omega$ , f = 1.0 kHz $V_p = 5.0 V$ , $P_{out} = 2.0 W$ $V_p = 3.6 V$ , $P_{out} = 1.0 W$	- -	82 81	- -	
Total Harmonic Distortion + Noise	THD+N	$V_p$ = 5.0 V, $R_L$ = 8.0 $\Omega$ , f = 1.0 kHz, $P_{out}$ = 0.25 W $V_p$ = 3.6 V, $R_L$ = 8.0 $\Omega$ ,	-	0.05	-	%
		f = 1.0 kHz, P <sub>out</sub> = 0.25 W	-	0.09	-	
Common Mode Rejection Ratio	CMRR	$V_p$ from 2.5 V to 5.5 V $V_{ic} = 0.5$ V to $V_p - 0.8$ V $V_p = 3.6$ V, $V_{ic} = 1.0$ V $V_{pp}$	-	-62	-	dB
		f = 217 Hz f = 1.0 kHz	-	-56 -57	- -	
Power Supply Rejection Ratio	PSRR	$V_{p\_ripple\_pk-pk}$ = 200 mV, $R_L$ = 8.0 $\Omega$ , Inputs AC Grounded $V_p$ = 3.6 V				dB
		f = 217 kHz f = 1.0 kHz	_ _	-62 -65	-   -	

# **Electrical Characteristics**

(Limits apply for  $T_A = +25^{\circ}C$  unless otherwise noted) (UDFN)

Characteristic	Symbol	Conditions	Min	Тур	Max	Unit
Operating Supply Voltage	V <sub>p</sub>	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	2.5	-	6.0	V
Supply Quiescent Current	l <sub>dd</sub>	$V_p = 3.6 \text{ V}, R_L = 8.0 \Omega$ $V_p = 5.5 \text{ V}, \text{ No Load}$ $V_p \text{ from } 2.5 \text{ V to } 5.5 \text{ V}, \text{ No Load}$		2.15 2.61	- -	mA
		$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	-	-	3.8	
Shutdown Current	I <sub>sd</sub>	$V_p = 4.2 \text{ V}$ $T_A = +25^{\circ}\text{C}$ $T_A = +85^{\circ}\text{C}$		0.42 0.45	0.8 2.0	μА
		$V_p = 5.5 \text{ V}$ $T_A = +25^{\circ}\text{C}$ $T_A = +85^{\circ}\text{C}$		0.8 0.9	1.5 -	μΑ
Shutdown Voltage High	V <sub>sdih</sub>	_	1.2	_	-	V
Shutdown Voltage Low	V <sub>sdil</sub>	-	-	-	0.4	V
Switching Frequency	F <sub>sw</sub>	$V_p$ from 2.5 V to 5.5 V $T_A = -40^{\circ}\text{C}$ to +85°C	180	240	300	kHz
Gain	G	$R_L = 8.0 \Omega$	285 kΩ R <sub>i</sub>	300 kΩ R <sub>i</sub>	<u>315 kΩ</u> R <sub>i</sub>	V V
Output Impedance in Shutdown Mode	Z <sub>SD</sub>	_	-	20	-	kΩ
Resistance from SD to GND	Rs	-	-	300	-	kΩ
Output Offset Voltage	Vos	V <sub>p</sub> = 5.5 V	-	6.0	-	mV
Turn On Time	Ton	V <sub>p</sub> from 2.5 V to 5.5 V	-	1.0	-	μs
Turn Off Time	Toff	V <sub>p</sub> from 2.5 V to 5.5 V	-	1.0	_	μS
Thermal Shutdown Temperature	Tsd	-	-	160	-	°C
Output Noise Voltage	Vn	$V_p$ = 3.6 V, f = 20 Hz to 20 kHz no weighting filter with A weighting filter		65 42	- -	μVrms

# **Electrical Characteristics**

(Limits apply for  $T_A = +25^{\circ}C$  unless otherwise noted) (UDFN)

Characteristic	Symbol	Conditions	Min	Тур	Max	Unit
RMS Output Power	Ро	$\begin{array}{c} R_L = 8.0 \; \Omega,  f = 1.0 \; \text{kHz},  \text{THD+N} < 1\% \\ V_p = 2.5 \; \text{V} \\ V_p = 3.0 \; \text{V} \\ V_p = 3.6 \; \text{V} \\ V_p = 4.2 \; \text{V} \\ V_p = 5.0 \; \text{V} \end{array}$	- - - -	0.22 0.33 0.45 0.67 0.92	- - - -	W
		$\begin{array}{c} R_L = 8.0~\Omega,~f = 1.0~\text{kHz},~\text{THD+N} < 10\% \\ V_p = 2.5~\text{V} \\ V_p = 3.0~\text{V} \\ V_p = 3.6~\text{V} \\ V_p = 4.2~\text{V} \\ V_p = 5.0~\text{V} \end{array}$	- - - -	0.36 0.53 0.76 1.07 1.49	- - - -	W
		$\begin{array}{c} R_L = 4.0 \; \Omega,  f = 1.0 \; \text{kHz},  \text{THD+N} < 1\% \\ V_p = 2.5 \; \text{V} \\ V_p = 3.0 \; \text{V} \\ V_p = 3.6 \; \text{V} \\ V_p = 4.2 \; \text{V} \\ V_p = 5.0 \; \text{V} \end{array}$	- - - -	0.24 0.38 0.57 0.83 1.2	- - - -	W
		$\begin{array}{c} R_L = 4.0 \; \Omega, \; f = 1.0 \; \text{kHz}, \; \text{THD+N} < 10\% \\ V_p = 2.5 \; \text{V} \\ V_p = 3.0 \; \text{V} \\ V_p = 3.6 \; \text{V} \\ V_p = 4.2 \; \text{V} \\ V_p = 5.0 \; \text{V} \end{array}$	- - - -	0.52 0.8 1.125 1.58 2.19	- - - -	W
Efficiency	-	$R_L = 8.0 \ \Omega, \ f = 1.0 \ \text{kHz}$ $V_p = 5.0 \ \text{V}, \ P_{out} = 1.2 \ \text{W}$ $V_p = 3.6 \ \text{V}, \ P_{out} = 0.6 \ \text{W}$	- -	87 87	- -	%
		$R_L = 4.0 \ \Omega, \ f = 1.0 \ \text{kHz}$ $V_p = 5.0 \ \text{V}, \ P_{out} = 2.0 \ \text{W}$ $V_p = 3.6 \ \text{V}, \ P_{out} = 1.0 \ \text{W}$	- -	79 78	- -	
Total Harmonic Distortion + Noise	THD+N	$\begin{aligned} V_p &= 5.0 \text{ V, R}_L = 8.0 \ \Omega, \\ f &= 1.0 \text{ kHz, P}_{out} = 0.25 \text{ W} \\ V_p &= 3.6 \text{ V, R}_L = 8.0 \ \Omega, \end{aligned}$	-	0.05	-	%
		f = 1.0 kHz, P <sub>out</sub> = 0.25 W	-	0.06	-	
Common Mode Rejection Ratio	CMRR	$V_p$ from 2.5 V to 5.5 V $V_{ic} = 0.5$ V to $V_p - 0.8$ V $V_p = 3.6$ V, $V_{ic} = 1.0$ V $V_p = 0.0$ V $V_p = $	_	-62	-	dB
		f = 217 Hz f = 1.0 kHz	-	-56 -57	-	
Power Supply Rejection Ratio	PSRR	$V_{p\_ripple\_pk-pk}$ = 200 mV, $R_L$ = 8.0 $\Omega$ , Inputs AC Grounded $V_p$ = 3.6 V				dB
		f = 217 kHz f = 1.0 kHz	<u> </u>	-62 -65	- -	

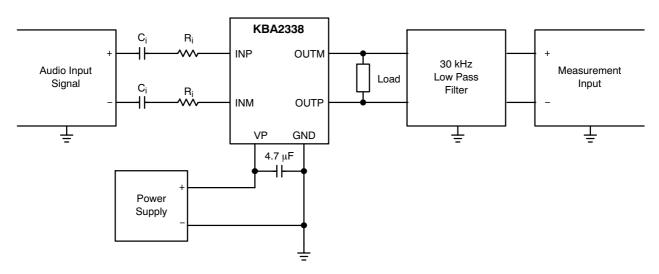


Figure 2. Test Setup for Graphs

#### NOTES:

- 1. Unless otherwise noted,  $C_i$  = 100 nF and  $R_i$ = 150 k $\Omega$ . Thus, the gain setting is 2 V/V and the cutoff frequency of the input high pass filter is set to 10 Hz. Input capacitors are shorted for CMRR measurements.
- 2. To closely reproduce a real application case, all measurements are performed using the following loads:

 $R_L = 8 \Omega$  means Load = 15  $\mu$ H + 8  $\Omega$  + 15  $\mu$ H

 $R_L$  = 4  $\Omega$  means Load = 15  $\mu$ H + 4  $\Omega$  + 15  $\mu$ H

Very low DCR 15  $\mu$ H inductors (50 m $\Omega$ ) have been used for the following graphs. Thus, the electrical load measurements are performed on the resistor (8  $\Omega$  or 4  $\Omega$ ) in differential mode.

3. For Efficiency measurements, the optional 30 kHz filter is used. An RC low-pass filter is selected with (100  $\Omega$ , 47 nF) on each PWM output.

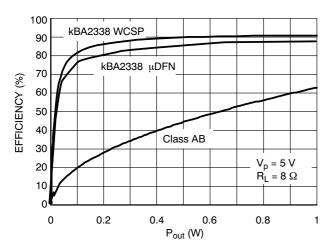


Figure 3. Efficiency vs.  $P_{out}$  V<sub>p</sub> = 5 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz

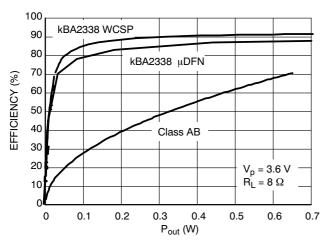


Figure 5. Efficiency vs. P  $_{out}$  V  $_{p}$  = 3.6 V, R  $_{L}$  = 8  $\Omega,$  f = 1 kHz

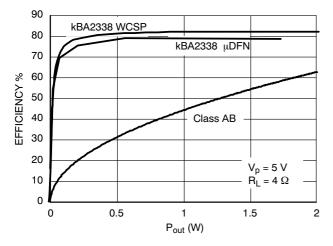


Figure 6. Efficiency vs.  $P_{out}$ V<sub>p</sub> = 5 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz

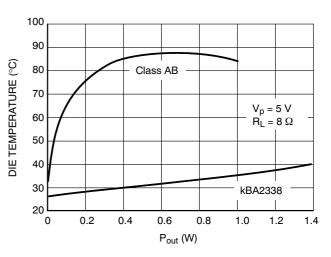


Figure 4. Die Temperature vs.  $P_{out}$   $V_p$  = 5 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz @  $T_A$  = +25°C

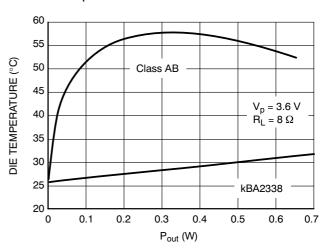


Figure 8. Die Temperature vs. P  $_{out}$  V $_{p}$  = 3.6 V, R $_{L}$  = 8  $\Omega$ , f = 1 kHz @ T $_{A}$  = +25°C

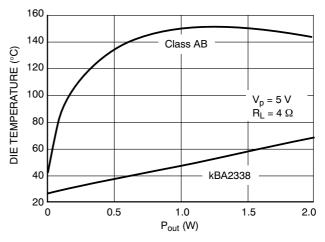


Figure 7. Die Temperature vs.  $P_{out}$   $V_p$  = 5 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz @  $T_A$  = +25°C

I----I- 0007

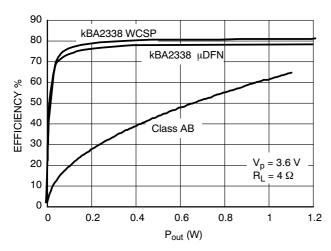


Figure 9. Efficiency vs.  $P_{out}$  V<sub>p</sub> = 3.6 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz

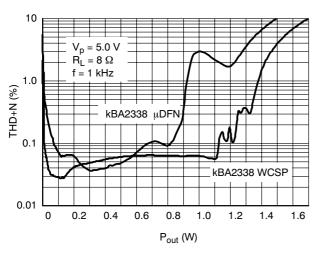


Figure 11. THD+N vs.  $P_{out}$  V<sub>p</sub> = 5 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz

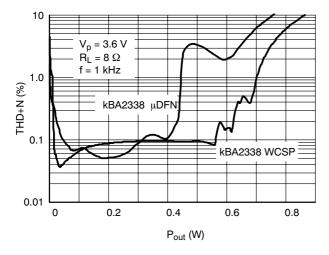


Figure 13. THD+N vs.  $P_{out}$  V<sub>p</sub> = 3.6 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz

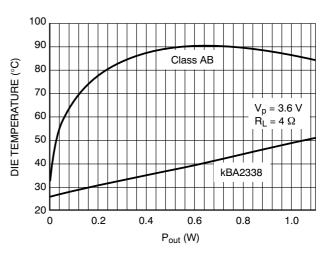


Figure 10. Die Temperature vs.  $P_{out}$   $V_p$  = 3.6 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz @  $T_A$  = +25°C

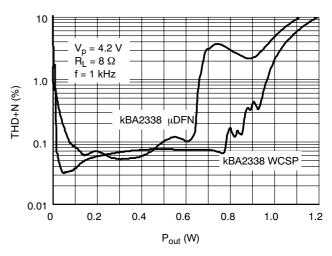


Figure 12. THD+N vs.  $P_{out}$  V<sub>D</sub> = 4.2 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz

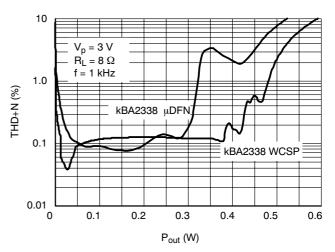
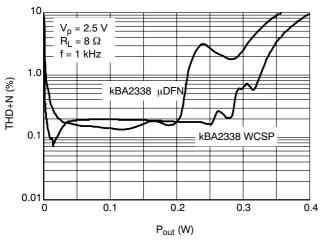


Figure 14. THD+N vs.  $P_{out}$  V<sub>p</sub> = 3 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz



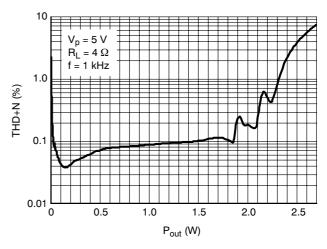
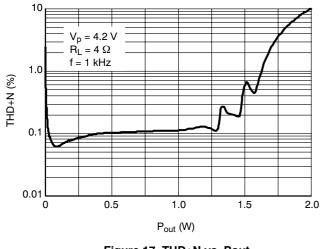


Figure 15. THD+N vs. Pout  $V_p$  = 2.5 V,  $R_L$  = 8  $\Omega$ , f = 1 kHz

Figure 16. THD+N vs. Pout  $V_p = 5 \text{ V}$ ,  $R_L = 4 \Omega$ , f = 1 kHz



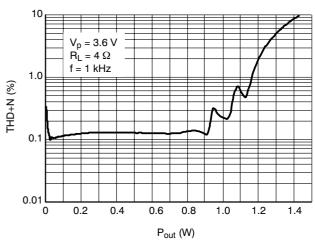
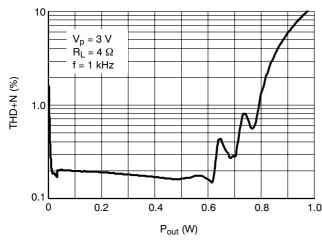


Figure 17. THD+N vs. Pout  $V_p = 4.2 \text{ V}$ ,  $R_L = 4 \Omega$ , f = 1 kHz

Figure 18. THD+N vs. Pout  $V_p$  = 3.6 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz



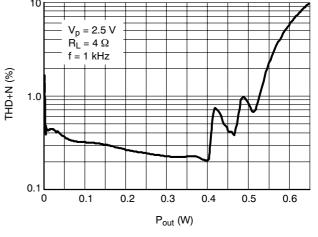
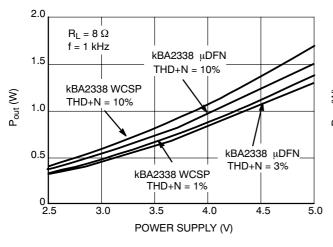


Figure 19. THD+N vs. Power Out  $V_p$  = 3 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz

Figure 20. THD+N vs. Power Out  $V_p$  = 2.5 V,  $R_L$  = 4  $\Omega$ , f = 1 kHz



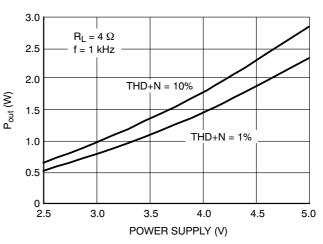
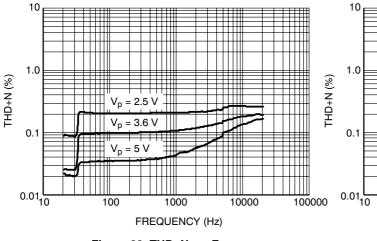


Figure 21. Output Power vs. Power Supply  $R_L = 8 \Omega @ f = 1 \text{ kHz}$ 

Figure 22. Output Power vs. Power Supply  $\mathbf{R_L} = \mathbf{4} \ \Omega \ \textcircled{0} \ \mathbf{f} = \mathbf{1} \ \mathbf{kHz}$ 



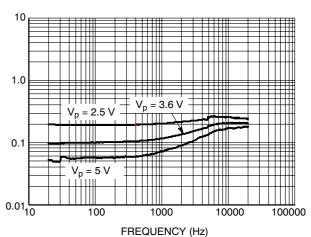
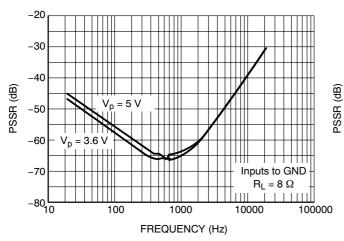


Figure 23. THD+N vs. Frequency R<sub>L</sub> = 8  $\Omega$ , P<sub>out</sub> = 250 mW @ f = 1 kHz

Figure 24. THD+N vs. Frequency  $R_L$  = 4  $\Omega$ ,  $P_{out}$  = 250 mW @ f = 1 kHz



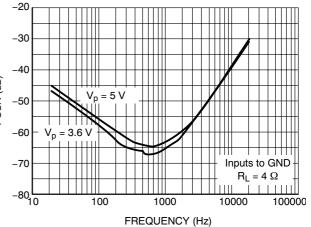
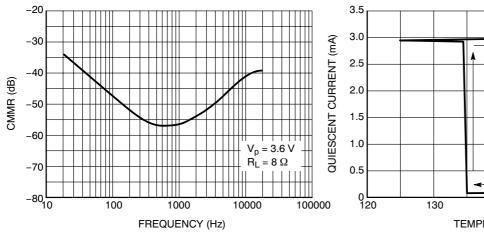


Figure 25. PSRR vs. Frequency Inputs Grounded, R<sub>L</sub> = 8  $\Omega$ , Vripple = 200 mvpkpk

Figure 26. PSRR vs. Frequency Inputs grounded, R<sub>L</sub> = 4  $\Omega$ , Vripple = 200 mVpkpk

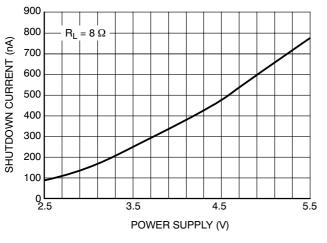
A---- 0007



3.5
3.0
2.5
2.0
1.5
1.0
0.5
0.1
120
130
140
150
160
TEMPERATURE (°C)

Figure 27. PSRR vs. Frequency  $\mbox{V}_{\mbox{\scriptsize p}}$  = 3.6 V,  $\mbox{R}_{\mbox{\scriptsize L}}$  = 8  $\Omega,$  Vic = 200 mvpkpk

Figure 28. Thermal Shutdown vs. Temperature  $\mbox{\sc V}_{\mbox{\sc p}}$  = 5 V,  $\mbox{\sc R}_{\mbox{\sc L}}$  = 8  $\Omega,$ 



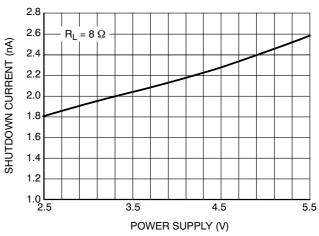
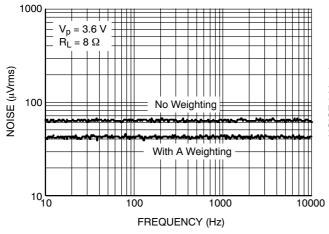


Figure 29. Shutdown Current vs. Power Supply  $\mbox{\bf R}_{\mbox{\bf L}}$  = 8  $\Omega$ 

Figure 30. Quiescent Current vs. Power Supply  $\mbox{\bf R}_{\mbox{\bf L}}$  = 8  $\Omega$ 



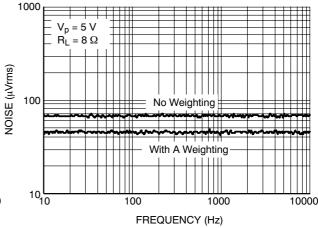
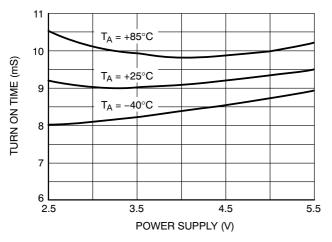


Figure 31. Noise Floor, Inputs AC Grounded with 1  $\mu$ F V<sub>p</sub> = 3.6 V

Figure 32. Noise Floor, Inputs AC Grounded with 1  $\mu\text{F V}_{p}$  = 5 V



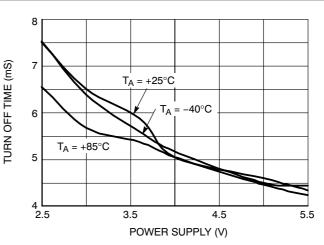


Figure 33. Turn on Time

Figure 34. Turn off Time

## **Description Information**

#### **Detailed Description**

The basic structure of the Kba2338 is composed of one analog pre–amplifier, a pulse width modulator and an H–bridge CMOS power stage. The first stage is externally configurable with gain–setting resistor  $R_i$  and the internal fixed feedback resistor  $R_f$  (the closed–loop gain is fixed by the ratios of these resistors) and the other stage is fixed. The load is driven differentially through two output stages.

The differential PWM output signal is a digital image of the analog audio input signal. The human ear is a band pass filter regarding acoustic waveforms, the typical values of which are 20 Hz and 20 kHz. Thus, the user will hear only the amplified audio input signal within the frequency range. The switching frequency and its harmonics are fully filtered. The inductive parasitic element of the loudspeaker helps to guarantee a superior distortion value.

#### **Power Amplifier**

The output PMOS and NMOS transistors of the amplifier have been designed to deliver the output power of the specifications without clipping. The channel resistance ( $R_{on}$ ) of the NMOS and PMOS transistors is typically 0.4  $\Omega$ .

# Turn On and Turn Off Transitions in Case of 9 Pin Flip-Chip Package

In order to eliminate "pop and click" noises during transition, the output power in the load must not be established or cutoff suddenly. When a logic high is applied to the shutdown pin, the internal biasing voltage rises quickly and, 4 ms later, once the output DC level is around the common mode voltage, the gain is established slowly (5.0 ms). This method to turn on the device is optimized in terms of rejection of "pop and click" noises. Thus, the total turn on time to get full power to the load is 9 ms (typical).

The device has the same behavior when it is turned—off by a logic low on the shutdown pin. No power is delivered to the load 5 ms after a falling edge on the shutdown pin. Due to the fast turn on and off times, the shutdown signal can be used as a mute signal as well.

#### Turn On and Turn Off Transitions in Case of UDFN8

In case of UDFN8 package, the audio signal is established instantaneously after the rising edge on the shutdown pin. The audio is also suddenly cut once a low level is sent to the amplifier. This way to turn on and off the device in a very fast way also prevents from "pop & click" noise.

#### **Shutdown Function**

The device enters shutdown mode when the shutdown signal is low. During the shutdown mode, the DC quiescent current of the circuit does not exceed  $1.5~\mu A$ .

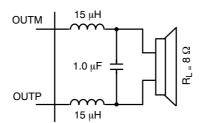
#### **Current Breaker Circuit**

The maximum output power of the circuit corresponds to an average current in the load of 820 mA.

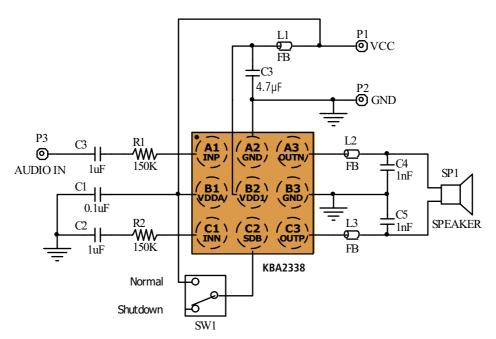
In order to limit the excessive power dissipation in the load if a short-circuit occurs, a current breaker cell shuts down the output stage. The current in the four output MOS transistors are real-time controlled, and if one current exceeds the threshold set to 1.5 A, the MOS transistor is opened and the current is reduced to zero. As soon as the short-circuit is removed, the circuit is able to deliver the expected output power.

This patented structure protects the Kba2338. Since it completely turns off the load, it minimizes the risk of the chip overheating which could occur if a soft current limiting circuit was used.

Manala 0007



**Optional Audio Output Filter** 



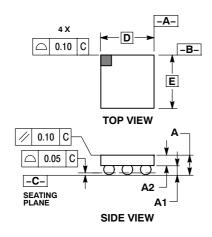
Ferrite Bead specification: Z=120  $\Omega$  @100MHz , Current rating=3A

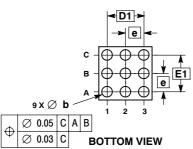
**kBA2338 WCSP Application Schematic** 

Manala 0007

# **PACAGE DESCRIPTION**

#### 9 PIN WCSP





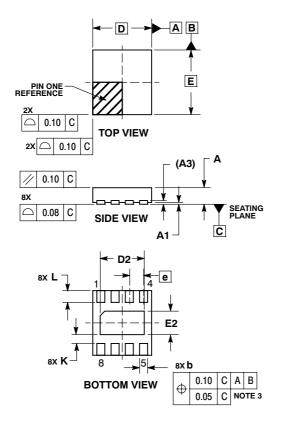
	MILLIMETERS					
DIM	MIN	MAX				
Α	0.540	0.660				
A1	0.210	0.270				
A2	0.330	0.390				
D	1.450	BSC				
E	1.450 BSC					
b	0.290	0.340				
е	0.500 BSC					
D1	1.000 BSC					
E1	1 000 BSC					

Manak 0007

# **PACAGE DESCRIPTION**

#### 8 PIN UDFN, 2x2.2, 0.5P

16



	MILLIMETERS						
DIM	MIN	NOM	MAX				
Α	0.45	0.50	0.55				
A1	0.00	0.03	0.05				
А3		0.127 R	EF				
b	0.20	0.25	0.30				
D		2.00 BS	C				
D2	1.40	1.50	1.60				
Е		2.20 BS	SC				
E2	0.70	0.80	0.90				
е		0.50 BSC					
K	0.20						
L	0.35	0.40	0.45				

