

# TFA9810

## Stereo full-bridge audio amplifier 2 x 12 W

Rev. 02 — 30 August 2007

Preliminary data sheet

### 1. General description

The TFA9810 is a two-channel power comparator for high-efficiency class D audio amplifier systems. It contains two full-bridge Bridge-Tied Load (BTL) power stages, drive logic, protection control logic and full differential input comparators. By using this power comparator a compact closed-loop self-oscillating digital amplifier system or open-loop system can be built. The TFA9810 does not require a heat sink and operates using an asymmetrical supply voltage.

### 2. Features

- Stereo full-bridge power comparator for class D audio amplifier applications
- No external heat sink required
- Operating voltage range: asymmetrical from 8 V to 20 V
- Thermally protected
- Zero dead-time switching
- Current-limiting (no audible interruptions)

### 3. Applications

- Self-oscillating or open-loop class D audio amplifier applications
- Flat-panel television sets
- Flat-panel monitors
- Multimedia systems
- Wireless speakers
- High-end CRT television sets

### 4. Quick reference data

**Table 1. Quick reference data**

$T_{amb} = 25\text{ }^{\circ}\text{C}$ ;  $V_P = 12\text{ V}$ ;  $f_{osc} = 550\text{ kHz}$ ; [Figure 33](#) unless otherwise specified

Symbol	Parameter	Condition	Min	Typ	Max	Unit
$V_P$	supply voltage	$V_P = V_{DDPx} - V_{SSPx}$	8	12	20	V
$I_{off}$	off-state current	Off mode	-	110	200	$\mu\text{A}$
$I_q$	quiescent current	With load, filter and snubbers connected	-	35	45	mA
$\eta_{po}$	output power efficiency	Output power 2 x 9 W into 8 $\Omega$ ; $P_o = P_{o(nom)}$	87	89	-	%

**Table 1. Quick reference data ...continued** $T_{amb} = 25\text{ }^{\circ}\text{C}$ ;  $V_P = 12\text{ V}$ ;  $f_{osc} = 550\text{ kHz}$ ; [Figure 33](#) unless otherwise specified

Symbol	Parameter	Condition	Min	Typ	Max	Unit
$P_{o(RMS)}$	continuous RMS output power per channel	$R_L = 8\ \Omega$ ; 12 V; THD = 10 %; Two channel driven; no heat sink required.	-	9.5	-	W
$P_o$	output power	$V_P = 12\text{ V}$ ; $R_L = 8\ \Omega$	-	-	-	-
		THD = 10 %	8.5	9.5	-	W
		THD = 1 %	6.5	7.5	-	W
		$V_P = 14\text{ V}$ ; $R_L = 8\ \Omega$ ; THD = 10 %; thermally limited	-	15	-	W
		$V_P = 16\text{ V}$ ; $R_L = 8\ \Omega$ ; THD = 10 %; thermally limited	-	15	-	W
		$V_P = 12\text{ V}$ ; $R_L = 6\ \Omega$ ; THD = 10 %; thermally limited	-	12	-	W
$V_P = 12\text{ V}$ ; $R_L = 4\ \Omega$ ; THD = 10 %; thermally limited	-	12	-	W		

## 5. Ordering information

**Table 2. Ordering information**

Type number	Package		Version
	Name	Description	
TFA9810T	SO32	plastic small outline package; 32 leads; body width 7.5 mm	SOT287-1

6. Block diagram

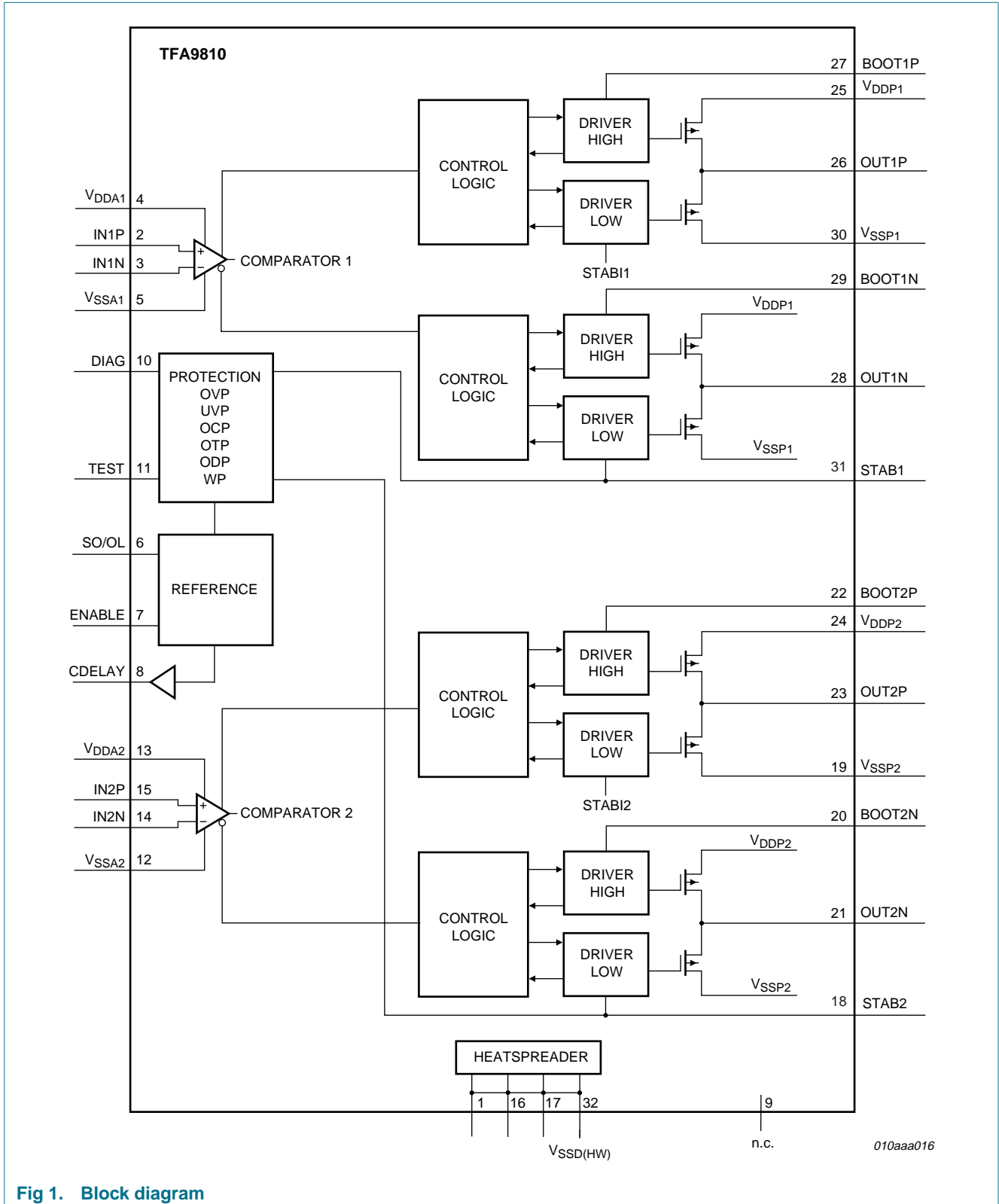


Fig 1. Block diagram

## 7. Pinning information

### 7.1 Pinning

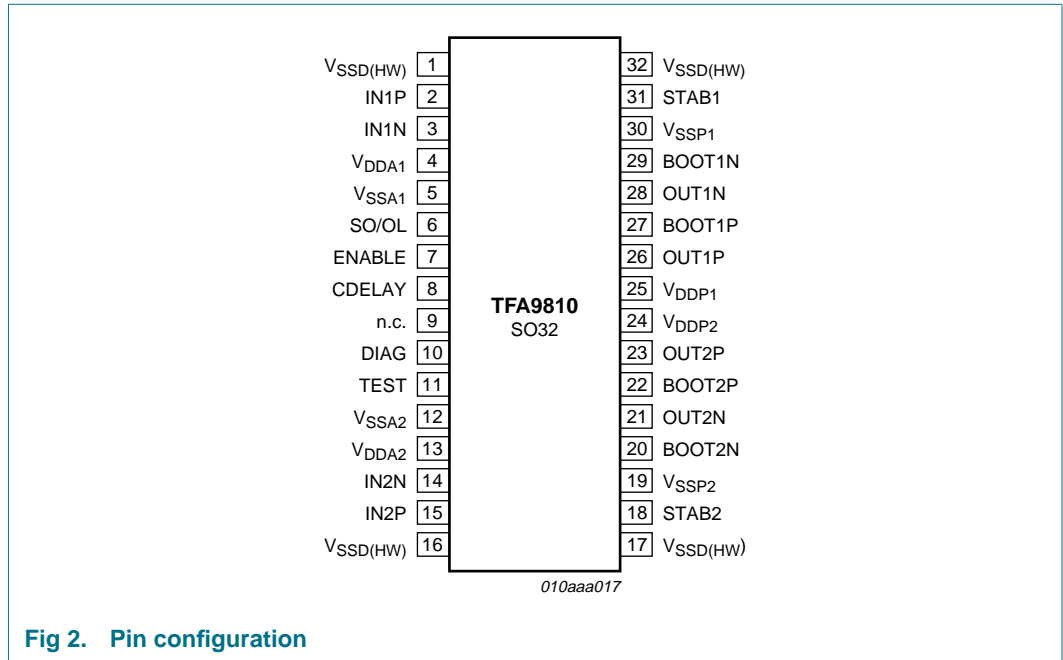


Fig 2. Pin configuration

The SO32 package has four corner leads. These leads (1, 16, 17, and 32) are internally connected to the die pad and must be connected to  $V_{SSA}$  in the application. Together with the applied copper area on the Printed Circuit Board (PCB) these leads determine the ambient temperature, which affects the thermal resistance of the junction.

### 7.2 Pin description

Table 3. Pin description

Symbol	Pin	Description
$V_{SSD(HW)}$	1, 16, 17, 32	Negative digital supply voltage and handle wafer
IN1P	2	Positive input comparator channel 1
IN1N	3	Negative input comparator channel 1
$V_{DDA1}$	4	Positive analog supply voltage channel 1
$V_{SSA1}$	5	Negative analog supply voltage channel 1
SO/OL	6	SO/OL input enables self-oscillating / open-loop configuration
ENABLE	7	Enable input to switch between SLEEP and OPERATING
CDELAY	8	CDELAY input determines the switch on/off timing
n.c.	9	Not connected
DIAG	10	Diagnostic output; open drain
TEST	11	Test signal input; for testing purposes only
$V_{SSA2}$	12	Negative analog supply voltage channel 2
$V_{DDA2}$	13	Positive analog supply voltage channel 2

Table 3. Pin description ...continued

Symbol	Pin	Description
IN2N	14	Negative input comparator channel 2
IN2P	15	Positive input comparator channel 2
STAB2	18	Decoupling of internal 11 V regulator for channel 2 drivers
V <sub>SSP2</sub>	19	Negative power-supply voltage channel 2
BOOT2N	20	Bootstrap high-side driver negative output channel 2
OUT2N	21	Negative output channel 2
BOOT2P	22	Bootstrap high-side driver positive output channel 2
OUT2P	23	Positive output channel 2
V <sub>DDP2</sub>	24	Positive supply voltage power channel 2
V <sub>DDP1</sub>	25	Positive power supply voltage channel 1
OUT1P	26	Positive output channel 1
BOOT1P	27	Bootstrap high-side driver positive output channel 1
OUT1N	28	Negative output channel 1
BOOT1N	29	Bootstrap high-side driver negative output channel 1
V <sub>SSP1</sub>	30	Negative supply voltage power channel 1
STAB1	31	Decoupling of internal 11 V regulator for channel 1 drivers

## 8. Functional description

### 8.1 General

The TFA9810 is a dual-switching power comparator. It is the main building block for a stereo high-efficiency Class D audio power amplifier system. It contains two full-bridge BTL power stages, drive logic, protection-control logic and full differential input comparators and references (see [Figure 1](#)). By using this power comparator a compact closed-loop self-oscillating digital amplifier system or open-loop system can be built. A second-order low-pass filter converts the Pulse Width Modulation (PWM) output signal into an analog audio signal across the speaker.

### 8.2 Interfacing

The pins ENABLE and SO/OL control the operating mode of the TFA9810. Both the ENABLE and the SO/OL pins refer to V<sub>SSD(HW)</sub>.

When the SO/OL pin is connected to V<sub>SSA</sub> the TFA9810 is in self-oscillating mode: when the SO/OL pin is floating the TFA9810 is in open-loop mode.

The TEST pin needs to be connected to ground in both situations.

Table 4. SO/OL connections

Interfacing	CONFIGURATION
SO/OL connected to	
V <sub>SSD(HW)</sub>	Self-oscillating
Open	Open-loop

The device has two modes: SLEEP and OPERATING.

In SLEEP mode the TFA9810 is not biased and has a very low supply current.

When the TFA9810 is set into OPERATING mode the device is started via the start-up sequence, which provides a pop-free start-up behavior. After start-up the reference voltages STAB are present and the outputs start switching.

**Table 5. Start-up**

Interfacing	
ENABLE [V]	MODE
ENABLE < 0.8 V	SLEEP
ENABLE > 3 V	OPERATING

### 8.3 Input comparators

The input stages have a differential input and are optimized for low noise and low offset. This results in maximum flexibility in the application.

#### 8.3.1 Operating in self-oscillating configuration

The inputs (IN1P, IN1N, IN2P, IN2N) of the comparators are internally set to a voltage level of  $0.5V_P$ , but only during the start-up sequence. In operating mode the inputs are high-ohmic.

#### 8.3.2 Operating in open-loop configuration

No internal voltages are applied to the inputs. The input pins (IN1P, IN1N, IN2P, IN2N) are pulled down to  $V_{SSA}$  level by internal resistors.

### 8.4 Diagnostic

The DIAG output is an open-drain output. The maximum current is 2 mA. Whenever one of the protections is triggered the DIAG output is activated low. The DIAG output refers to  $V_{SSD}$ .

### 8.5 Protections

Overtemperature, overcurrent, overvoltage, undervoltage, over-dissipation sensors, and window protection are included in the TFA9810. When one of these sensors exceeds its threshold level either the output power stage is switched off and the outputs (OUT1N, OUT1P, OUT2N, OUT2P) become floating, or the TFA9810 shuts down and starts up immediately.

- **OverTemperature Protection (OTP)**

If the junction temperature ( $T_j$ ) exceeds a threshold level of about 150 °C then the outputs become floating. The device will start switching again after 5  $\mu$ s and when the temperature is below 150 °C. This is thermal limitation without audible interruptions.

- **OverCurrent Protection (OCP)**

If the output current exceeds the maximum output current threshold level the output becomes floating. The device will start switching again after 5  $\mu$ s. This is current limitation without audible interruptions.

- **OverVoltage Protection (OVP)**

When the supply voltage applied to the TFA9810 exceeds the maximum supply voltage threshold level the device will shut down. The device will restart when the supply voltage is within the operating range.

- **UnderVoltage Protection (UVP)**

When the supply voltage applied to the TFA9810 falls below the minimum supply voltage threshold level the device will shut down. The device will restart when the supply voltage is within the operating range.

- **Over-Dissipation Protection (ODP)**

The ODP in the TFA9810 is a combination of two protections. Exceeding a temperature threshold level of 135 °C an internal pre-warning is generated. When an overcurrent is detected during the pre-warning the device will shut down. When the ENABLE pin is high the TFA9810 will restart automatically. The restart sequence (switch-off → switch-on) will take 200 ms to 500 ms.

- **Window Protection (WP)**

During start-up, if one of the outputs is shorted to  $V_{SS}$  or  $V_{DD}$  the device will not start. This is an effective measure to protect the device against shorts between the outputs (before the filter) and the ground or supply lines. The supply must be switched off prior to removing any short. The WP protects the device against failure during board assembly.

**Table 6. Overview protections**

Protections				
Symbol	Condition	DIAG	Outputs	Recovering
OTP	$T_j > 150\text{ °C}$	LOW	Floating	Automatic, after 5 $\mu\text{s}$ and $T_j < 150\text{ °C}$
OCP	$I_O > I_{ORM}$	LOW	Floating	Automatic, after 5 $\mu\text{s}$ and $I_O < I_{ORM}$
OVP	$V_P > 20\text{ V}$	LOW	Floating	Restart (switch-off → switch-on when $V_P < 20\text{ V}$ )
UVP	$V_P < 8\text{ V}$	LOW	Floating	Restart (switch-off → switch-on when $V_P > 8\text{ V}$ )
ODP	$T_j > 135\text{ °C}$ and $I_O > I_{ORM}$	LOW	Floating	Restart (switch-off → switch-on when $T_j < 135\text{ °C}$ or $I_O < I_{ORM}$ )
WP	$OUTX > V_{DDA} - 1\text{ V}$ or $OUTX < V_{SSA} + 1\text{ V}$	LOW		

### 8.6 Start-up sequence

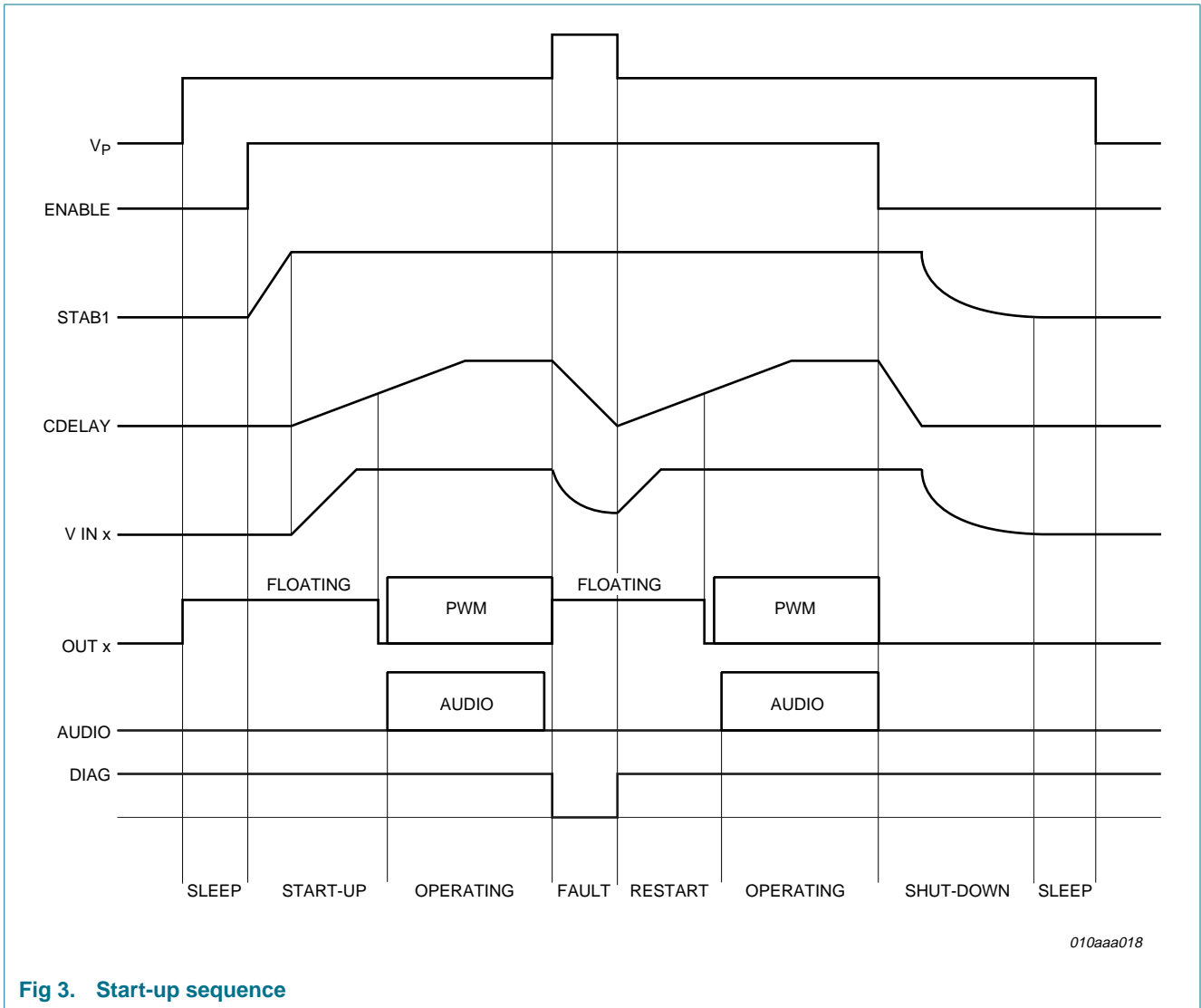


Fig 3. Start-up sequence

## 9. Internal circuitry

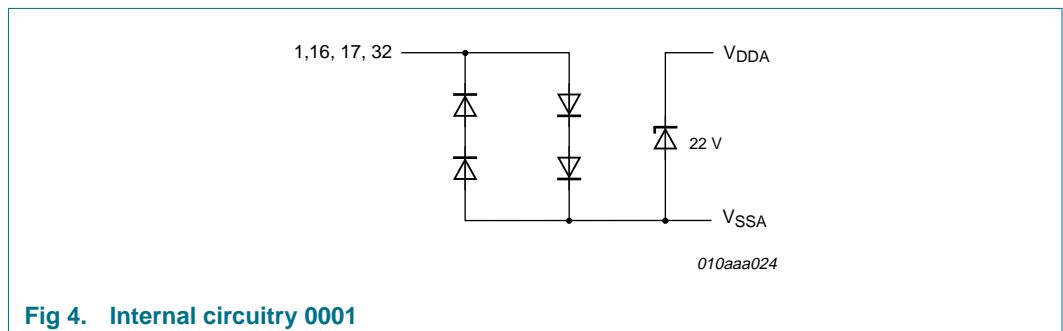


Fig 4. Internal circuitry 0001



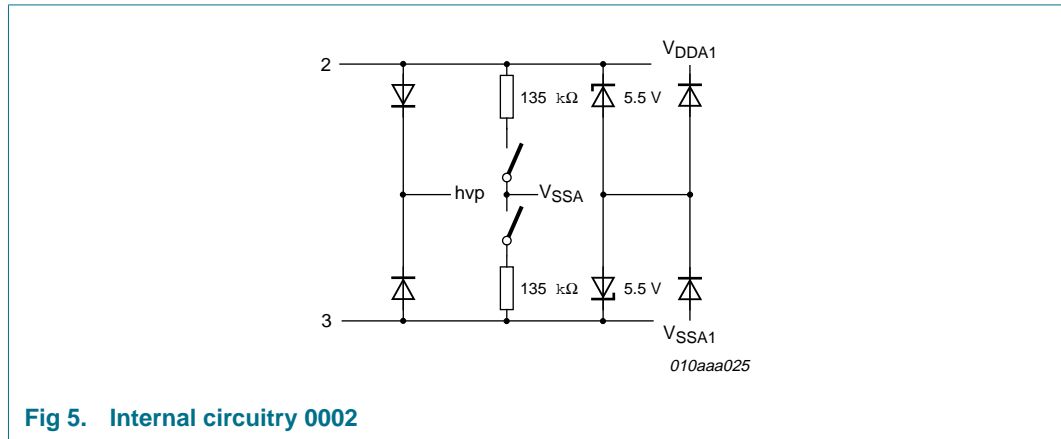


Fig 5. Internal circuitry 0002

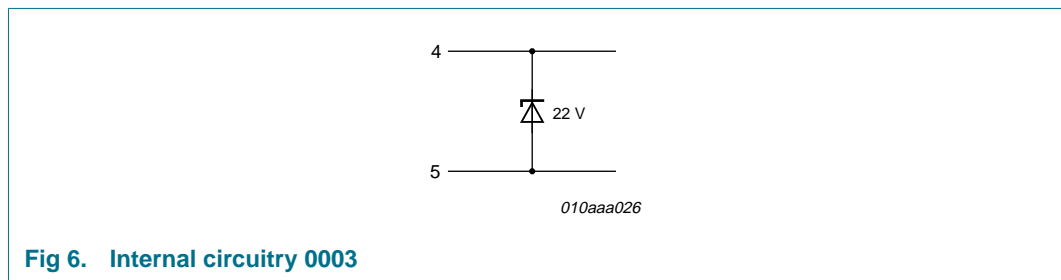


Fig 6. Internal circuitry 0003

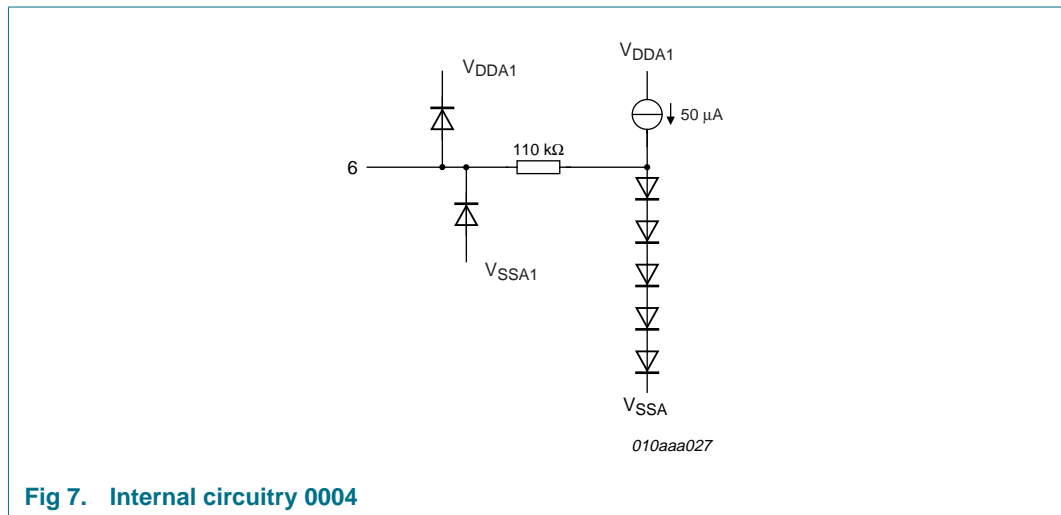


Fig 7. Internal circuitry 0004

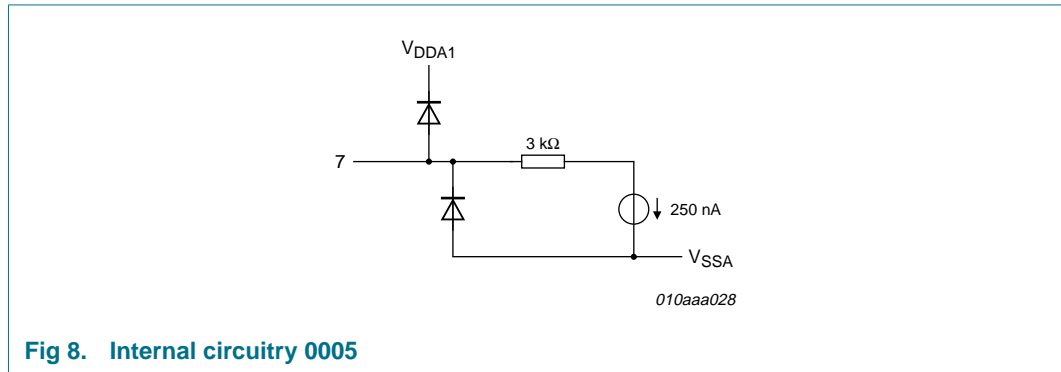


Fig 8. Internal circuitry 0005

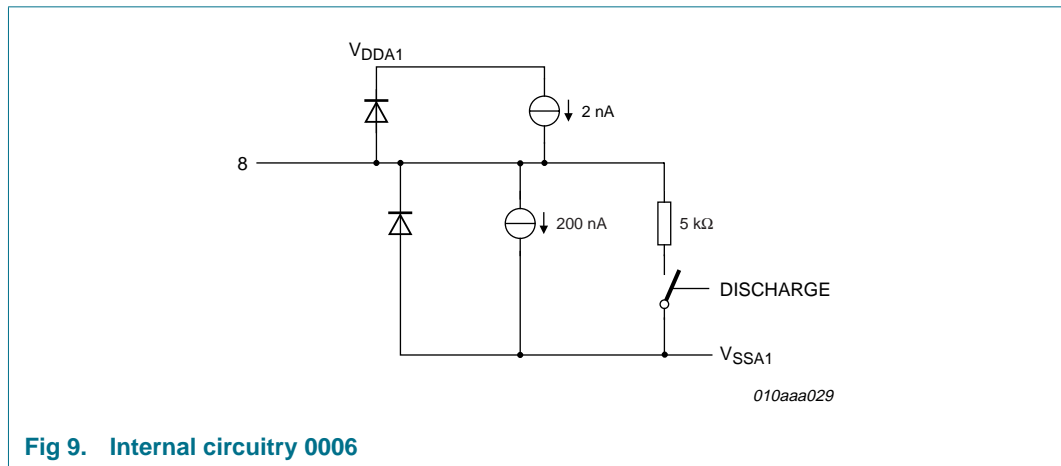


Fig 9. Internal circuitry 0006

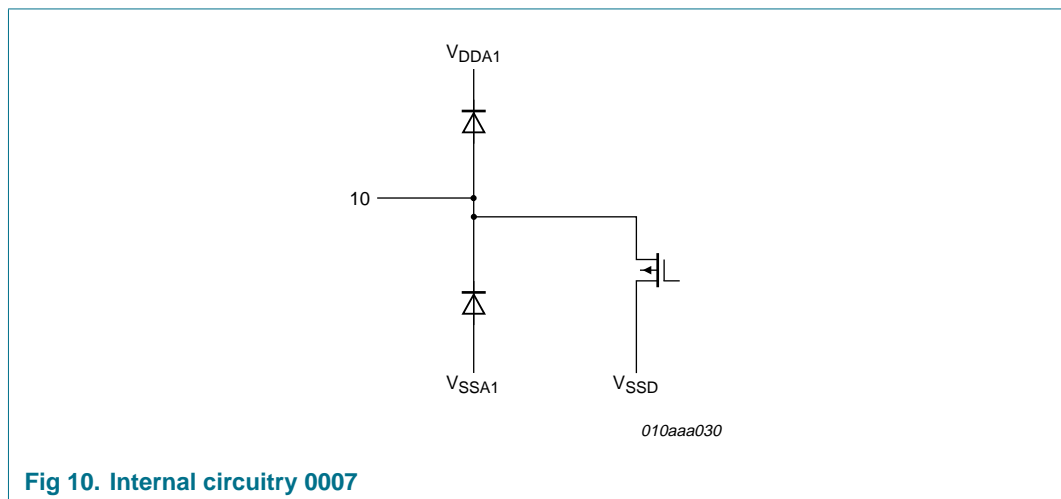
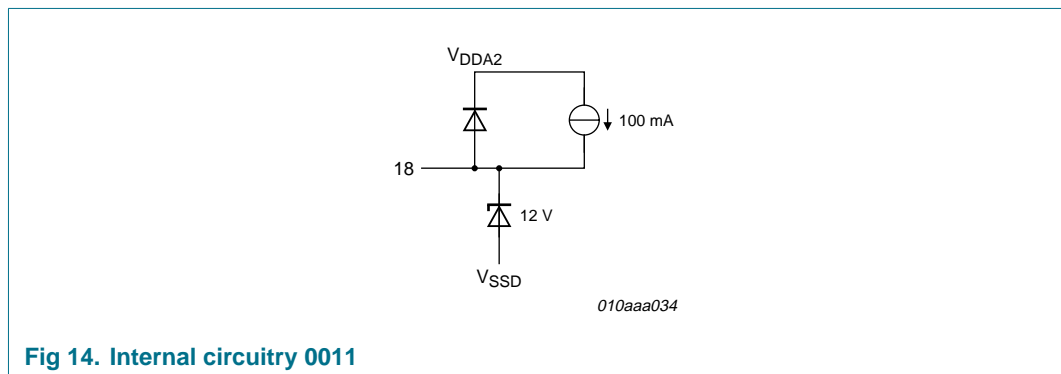
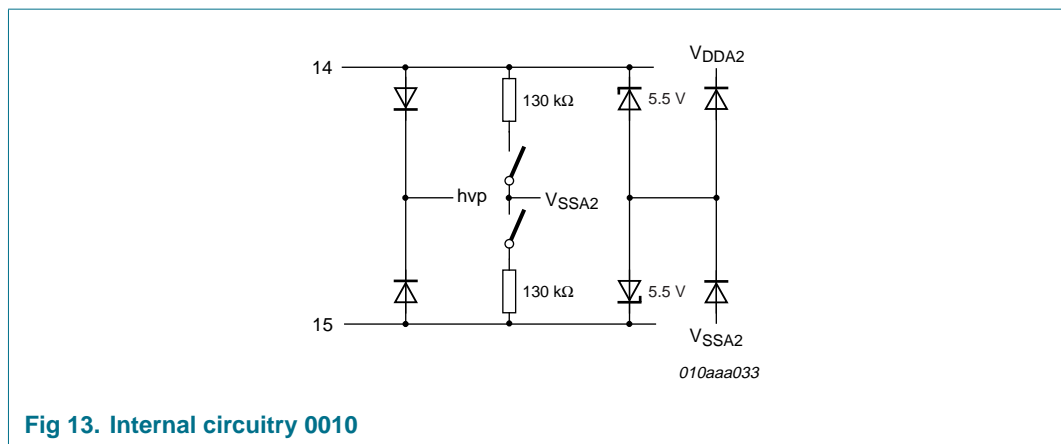
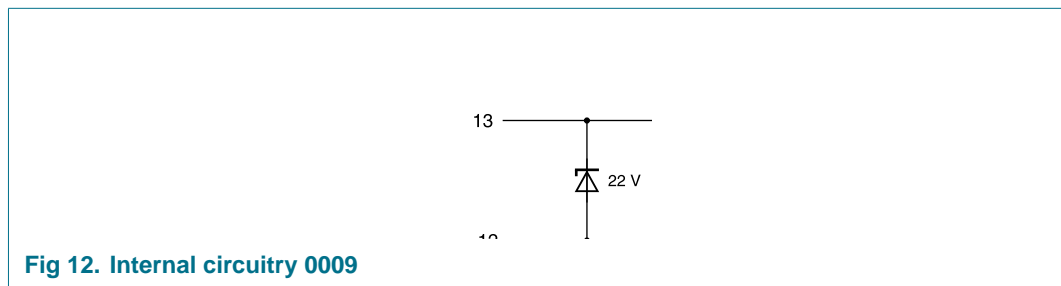
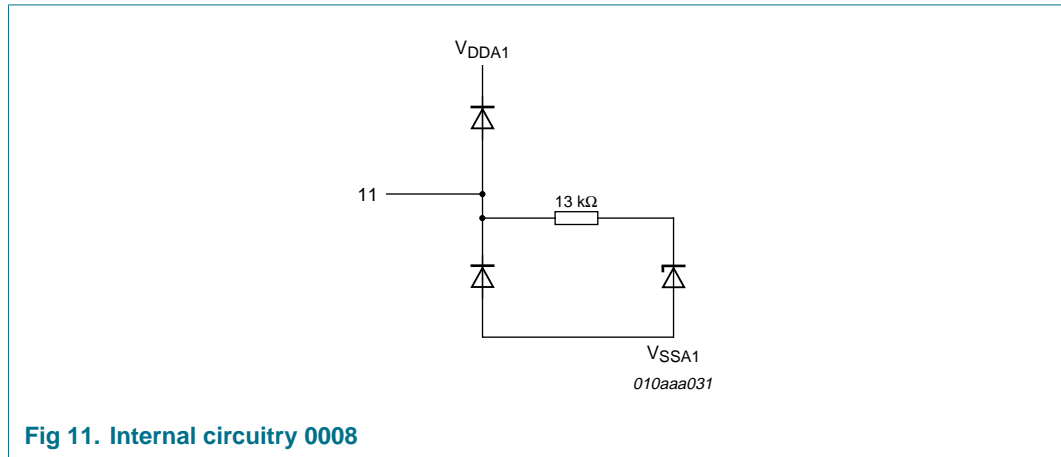


Fig 10. Internal circuitry 0007



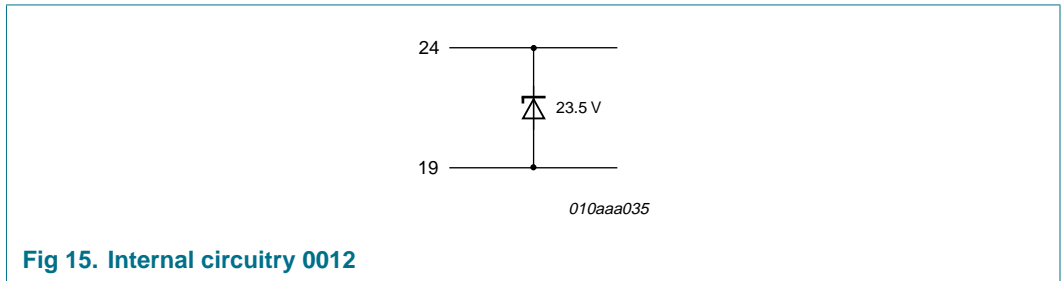


Fig 15. Internal circuitry 0012

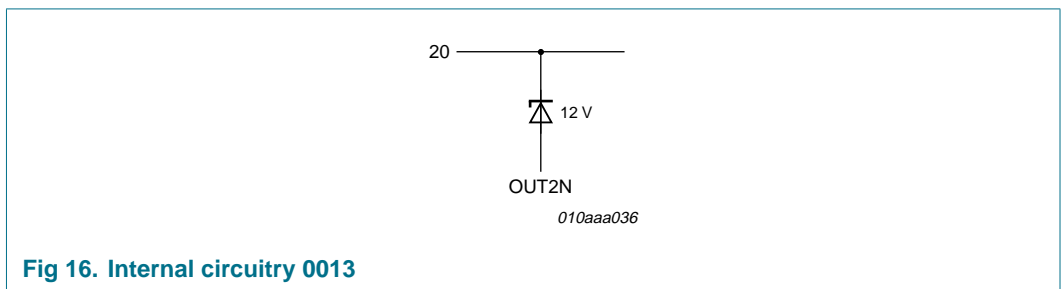


Fig 16. Internal circuitry 0013

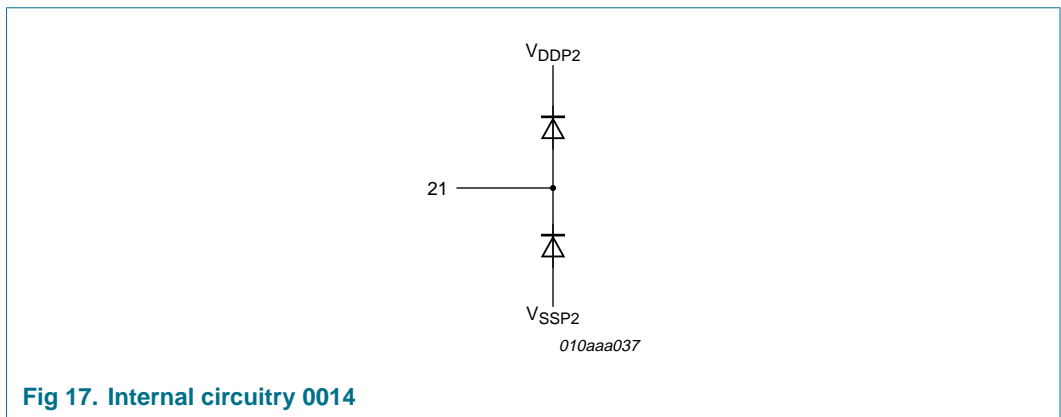


Fig 17. Internal circuitry 0014

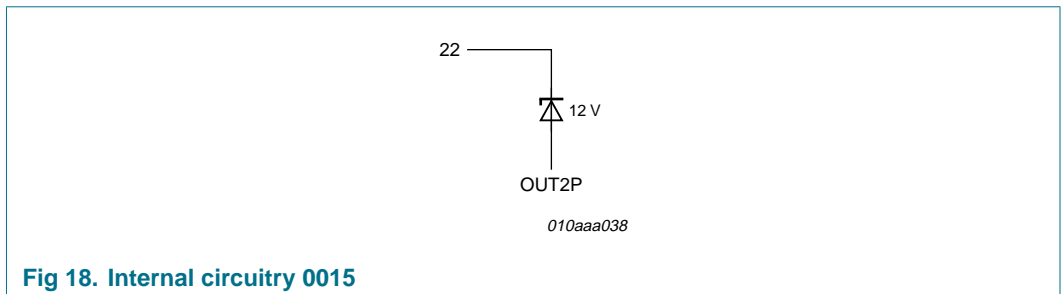
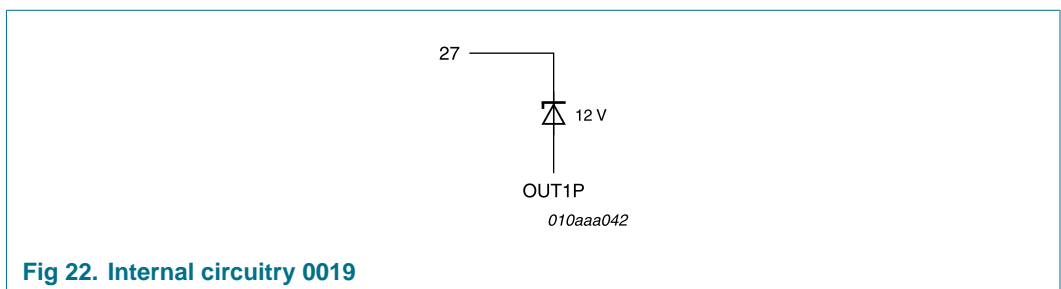
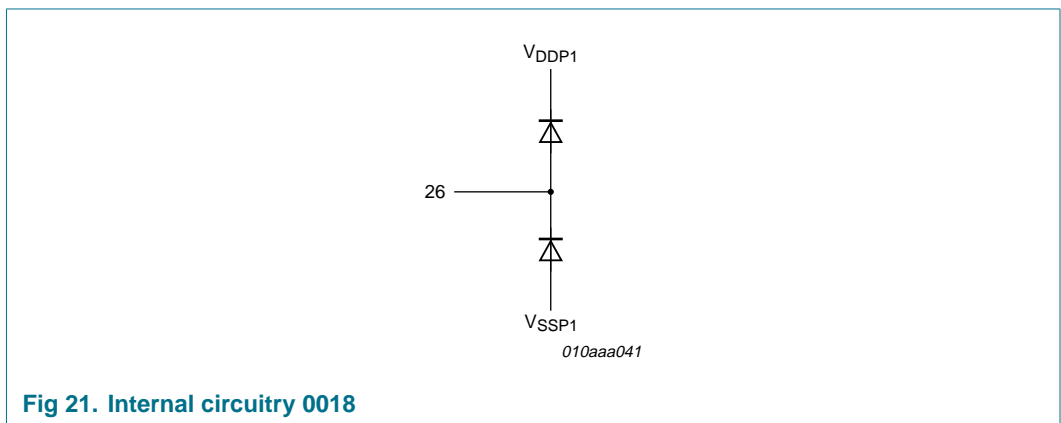
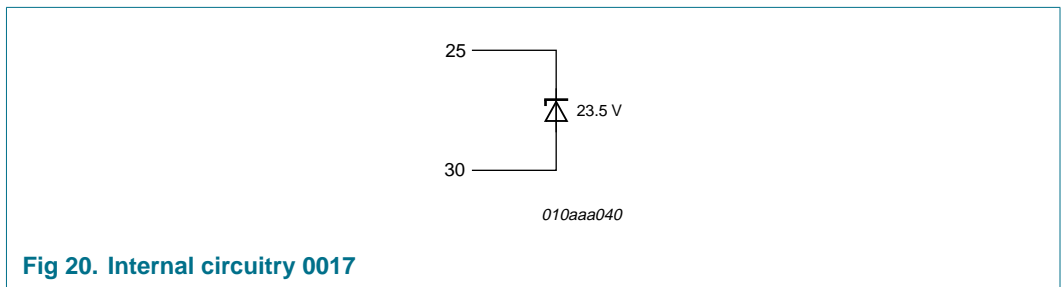
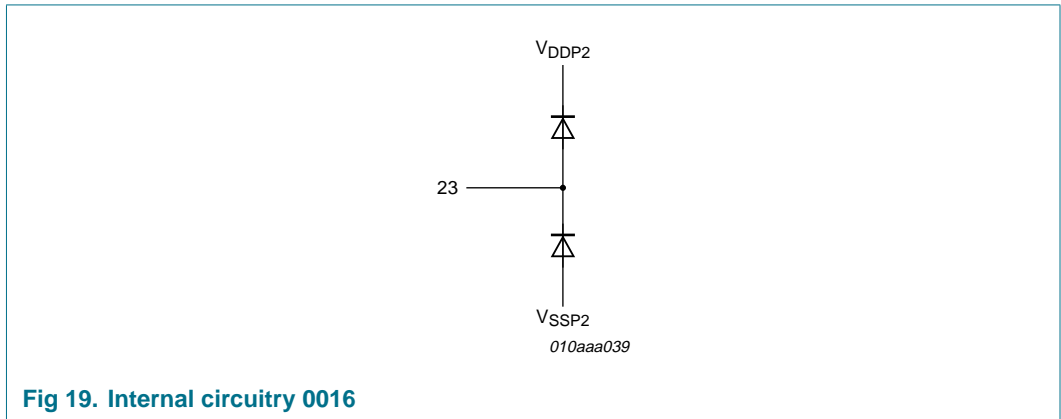
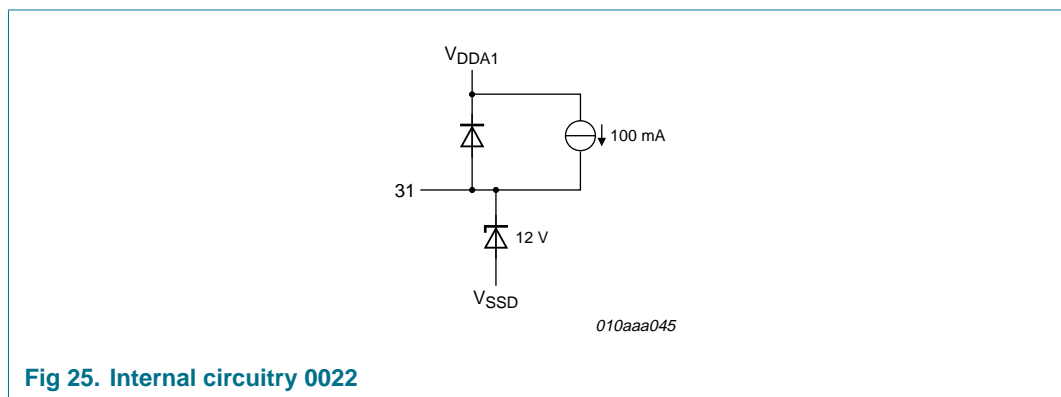
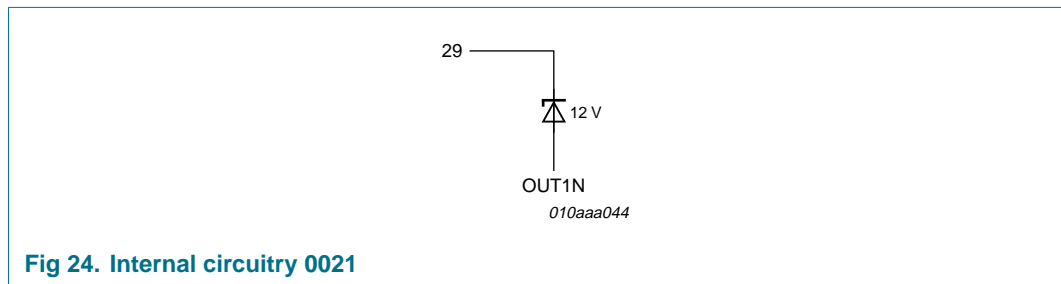
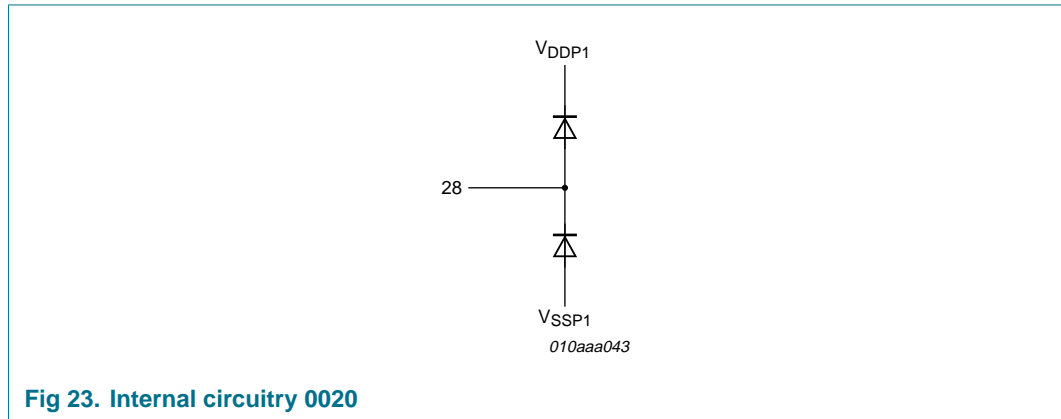


Fig 18. Internal circuitry 0015





## 10. Limiting values

**Table 7. Limiting values**

*In accordance with the Absolute Maximum Rating System (IEC 60134).*

Symbol	Parameter	Conditions	Min.	Max.	Unit
$V_P$	supply voltage	asymmetrical	-0.3	+23	V
$I_{ORM}$	repetitive peak output current		3	-	A
$T_j$	junction temperature		-	+150	°C
$T_{stg}$	storage temperature		-55	+150	°C
$T_{amb}$	ambient temperature		-40	+85	°C
$P_{max}$	maximum power dissipation		-	2.5	W

**Table 7. Limiting values ...continued**  
 In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min.	Max.	Unit
V <sub>x</sub>	voltage on pin x	DIAG	V <sub>SS</sub> – 0.3	+12	V
		IN1P - IN1N	–12	+12	V
		IN2P - IN2N	–12	+12	V
		all other pins	V <sub>SS</sub> – 0.3	V <sub>DD</sub> + 0.3	V
V <sub>esd</sub>	electrostatic discharge voltage	VINX with respect to other pins	–1500	+1500	V
		all other pins	–2000	+2000	V

## 11. Thermal characteristics

**Table 8. Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
R <sub>th(j-a)</sub>	thermal resistance from junction to ambient	SO32. JEDEC test board	[1] -	41	44	K/W
		SO32. Two-layer application board	-	44	-	K/W
Ψ <sub>j-lead</sub>	thermal characterization parameter from junction to lead	SO32	[1]		30	K/W
Ψ <sub>j-top</sub>	thermal characterization parameter from junction to top of package	SO32	[1] [2]	4	8	K/W

[1] Measured in a JEDEC high K-factor test board (standard EIA/JESD 51-7) in free air with natural convection.

[2] Strongly depends on where the measurement is taken on the case.

## 12. Characteristics

### 12.1 Static characteristics

**Table 9. Static characteristics**  
*T<sub>amb</sub> = 25 °C; V<sub>P</sub> = 12 V; f<sub>osc</sub> = 550 kHz; Figure 33 unless otherwise specified*

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
<b>Supply voltage</b>						
V <sub>P</sub>	supply voltage	V <sub>P</sub> = V <sub>DDPx</sub> – V <sub>SSPx</sub>	8	12	20	V
I <sub>off</sub>	off-state current	off mode	-	110	200	μA
I <sub>q</sub>	quiescent current	with load, filter, and snubbers connected	-	35	45	mA
<b>ENABLE input</b>						
V <sub>IL</sub>	LOW-level input voltage	with respect to V <sub>SSD</sub>	–0.3	-	+0.8	V
V <sub>IH</sub>	HIGH-level input voltage	with respect to V <sub>SSD</sub>	3	-	V <sub>P</sub>	V

**Table 9. Static characteristics ...continued** $T_{amb} = 25\text{ }^{\circ}\text{C}$ ;  $V_P = 12\text{ V}$ ;  $f_{osc} = 550\text{ kHz}$ ; [Figure 33](#) unless otherwise specified

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_I$	input current	$V_I = 5\text{ V}$	-	1	20	$\mu\text{A}$
<b>SO/OL input</b>						
$V_{IL}$	LOW-level input voltage	with respect to $V_{SSD}$	-	-	0.4	V
$V_{IH}$	HIGH-level input voltage	with respect to $V_{SSD}$	3	4	4.5	V
<b>STABI</b>						
$V_{STABI}$	voltage on pin STABI	with respect to $V_{SS}$	10	11	12	V
<b>Comparator full-differential input stage</b>						
$V_{\text{offset}(i)(\text{eq})}$	equivalent input offset voltage		-	-	1	mV
		$20\text{ Hz} < f < 20\text{ kHz}$	-	-	15	$\mu\text{V}$
$V_{\text{cm}}$	common mode voltage		$V_{SSA} + 3$	-	$V_{DDA} - 1$	V
$I_{IB}$	input bias current		-	-	1	$\mu\text{A}$
<b>OverTemperature Protection (OTP)</b>						
$T_{\text{prot}}$	protection temperature		150	-	-	$^{\circ}\text{C}$
<b>OverVoltage Protection (OVP)</b>						
$V_{\text{th}(\text{ovp})}$	overvoltage protection threshold voltage	level internal fixed	20	21.5	23	V
<b>UnderVoltage Protection (UVP)</b>						
$V_{P(\text{uvp})}$	undervoltage protection supply voltage	level internal fixed	7	7.5	8	V
<b>OverCurrent Protection (OCP)</b>						
$I_{O(\text{ocp})}$	overcurrent protection output current		<a href="#">[1]</a> 3	3.5	-	A
<b>Window Protection (WP)</b>						
$V_O$	output voltage	high level	-	$V_{DDA} - 1$	-	V
		low level	-	$V_{SSA} + 1$	-	V

[1] Current limiting concept: in overcurrent condition no interruption of the audio signal in case of impedance drop.



## 12.2 Dynamic characteristics

**Table 10. Dynamic characteristics**

$T_{amb} = 25\text{ }^{\circ}\text{C}$ ;  $V_P = 12\text{ V}$ ;  $R_L = 8\text{ }\Omega$ ; [Figure 33](#) unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
<b>PWM Output</b>						
$t_r$	rise time		-	10	-	ns
$t_f$	fall time		-	10	-	ns
$t_{resp}$	response time	transition PWM output from LOW to HIGH	-	60	-	ns
		$V_I = 70\text{ mV}$	-	50	-	ns
		$V_I = 3.3\text{ V}$	-	60	-	ns
		transition PWM output from HIGH to LOW	-	50	-	ns
		$V_I = 70\text{ mV}$	-		-	
		$V_I = 3.3\text{ V}$	-		-	
$t_{w(min)}$	minimum pulse width	PWM output	-	60	-	ns
$R_{DSon}$	drain-source on-state resistance		[1]	0.28	0.35	$\Omega$
$\eta_{po}$	output power efficiency	output power 2 x 9 W into 8 $\Omega$ . $P_o = P_{o(nom)}$	[2]	87	89	%

[1] High-side and low-side power switch have the same series resistance.

[2] Output power measured across the loudspeaker load. Output power is measured indirectly via  $R_{DSon}$ .

## 12.3 AC characteristics measured in a typical application

**Table 11. AC characteristics measured in typical application**

$T_{amb} = 25\text{ }^{\circ}\text{C}$ ;  $V_P = 12\text{ V}$ ;  $R_L = 8\text{ }\Omega$ ;  $f_{osc} = 550\text{ kHz}$ ; [Figure 33](#) unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_P$	operating supply voltage	$V_P = V_{DDPX} - V_{SSPX}$	8	12	20	V
$P_{o(RMS)}$	continuous RMS output power per channel.	$R_L = 8\text{ }\Omega$ ; $V_P =$ 12 V THD = 10 %; Two channel driven; no heat sink required.	-	9.5	-	W
$P_o$	output power	$V_P = 12\text{ V}$ ; $R_L = 8\text{ }\Omega$	-	-	-	W
		THD = 10 %	8.5	9.5	-	W
		THD = 1 %	6.5	7.5	-	W
		$V_P = 14\text{ V}$ ; $R_L = 8\text{ }\Omega$ ; THD = 10 %; thermally limited	-	12	-	W
		$V_P = 16\text{ V}$ ; $R_L = 8\text{ }\Omega$ ; THD = 10 %; thermally limited	-	15	-	W
		$V_P = 12\text{ V}$ ; $R_L = 6\text{ }\Omega$ ; THD = 10 %; thermally limited	-	12	-	W
		$R_L = 4\text{ }\Omega$ ; $V_P = 12\text{ V}$ ; THD = 10 %; thermally limited	-	15	-	W

**Table 11. AC characteristics measured in typical application ...continued**

$T_{amb} = 25\text{ }^{\circ}\text{C}$ ;  $V_P = 12\text{ V}$ ;  $R_L = 8\text{ }\Omega$ ;  $f_{osc} = 550\text{ kHz}$ ; [Figure 33](#) unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
THD+N	total harmonic distortion-plus-noise	$P_o$ is 1 W; $f = 1\text{ kHz}$ ; AES17 brick-wall filter	-	0.04	0.1	%
$\eta_{po}$	output power efficiency	$P_o = 9\text{ W}$	87	89	-	%
$G_{v(cl)}$	closed-loop voltage gain	$V_I = 100\text{ mV (RMS)}$ ; $f_i = 1\text{ kHz}$	19	19.7	21	dB
$V_{n(o)}$	noise output voltage	Inputs shorted; AES17 brick-wall filter	-	150	-	$\mu\text{V}$
S/N	signal-to-noise ratio	$V_o = 10\text{ V(RMS)}$ ; gain 20 dB	-	96	-	dB
SVRR	supply voltage ripple rejection	$V_{ripple} = 2\text{ Vpp}$ ; $f_i = 1\text{ kHz}$ <a href="#">[1]</a>	34	45	-	dB
$\alpha_{cs}$	channel separation	$P_o = 1\text{ W}$ ; $f_i = 1\text{ kHz}$	55	70	-	dB

[1] Minimum value determined by R5, R10, R17, R22 equalling +1 % and R7, R14, R18, R20 equalling -1 %.

### 13. Quality specification

*In accordance with SNW-FQ-611-E, 'if this type is used as an audio amplifier'. The number of the quality specification can be found in the Quality Reference Handbook. The handbook can be ordered using the code 9398 510 63011.*

### 14. Application information

#### 14.1 Output power estimation

For BTL configuration the output power just before clipping can be estimated using [Equation 1](#):

$$BTL : P_{o0.5\%} = \frac{\left[ \left[ \frac{R_L}{R_L + 2 \times (R_{DSon} + R_s)} \right] \times V_P \right]^2}{2 \times R_L} \tag{1}$$

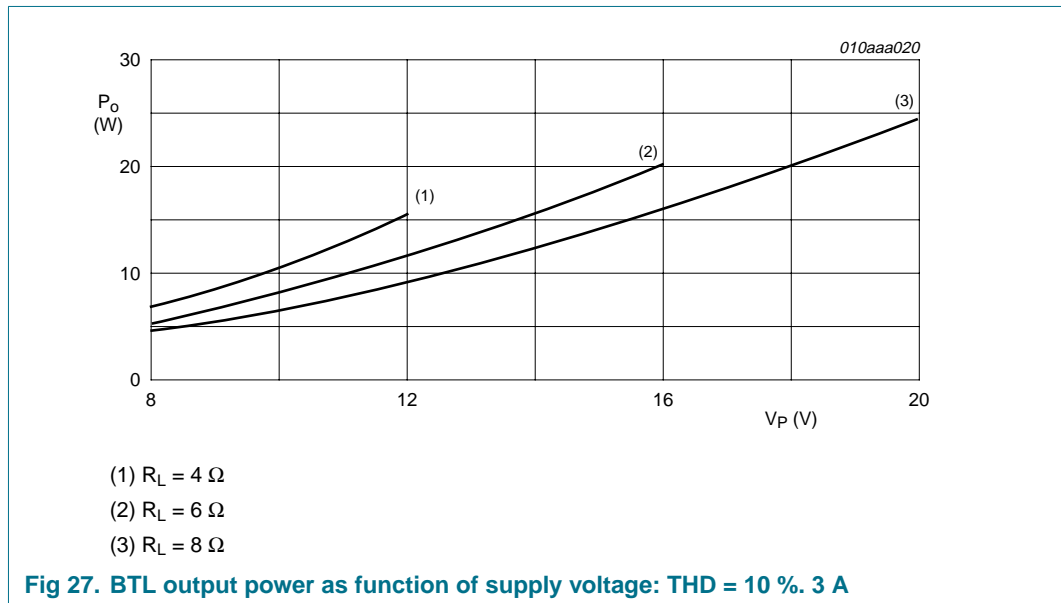
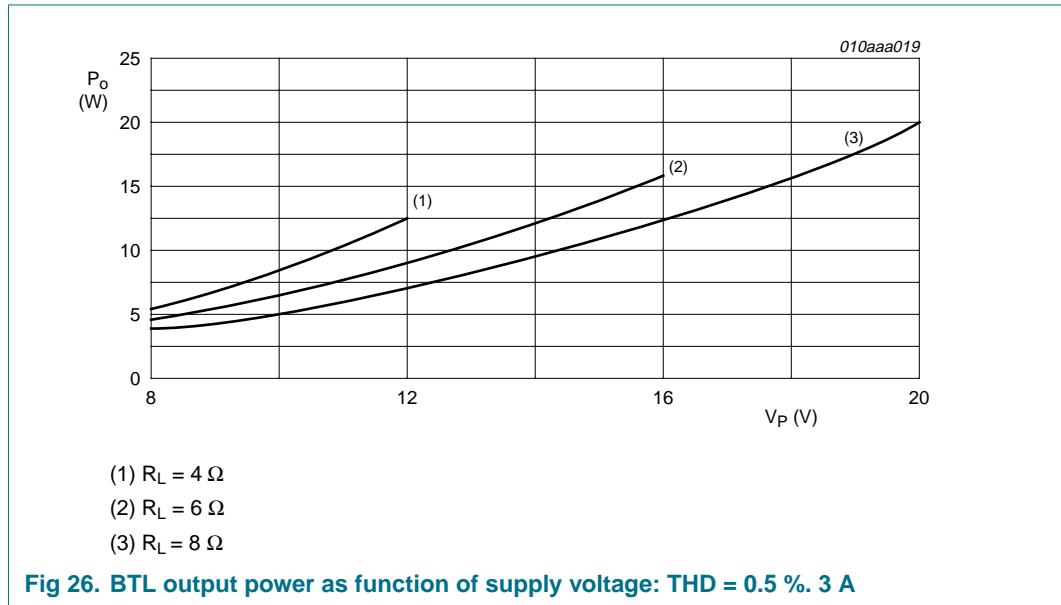
Where,

- $V_P$  = supply voltage ( $V_{DDPx} - V_{SSPx}$ ) [V]
- $R_L$  = load resistance [ $\Omega$ ]
- $R_{DSon}$  = drain-source on-state resistance [ $\Omega$ ]
- $R_s$  = series resistance [ $\Omega$ ]
- $P_{o0.5\%}$  = output power at the THD level of 0.5 % [W]

The output power at 10 % THD can be estimated by using [Equation 2](#):

$$P_{o10\%} = 1.25 \times P_{o0.5\%} \tag{2}$$

[Figure 26](#) and [Figure 27](#) below show the estimated output power at THD = 0.5 % and THD = 10 % as a function of the BLT supply voltage for different load impedances.



## 14.2 Output current limiting

The peak output current is internally limited above a level of 3 A minimum. During normal operation the output current should not exceed this threshold level of 3 A, otherwise the output signal will be distorted. The peak output current in BTL can be estimated using [Equation 3](#):

$$I_{Omax} \leq \frac{V_P}{R_L + 2 \times (R_{DSon} + R_s)} \leq 3 \text{ A} \quad (3)$$

Where:

- $V_P$  = supply voltage ( $V_{DDPX} - V_{SSPX}$ ) [V].
- $R_L$  = load resistance [ $\Omega$ ].
- $R_{DSon}$  = drain-source on-state resistance [ $\Omega$ ].
- $R_s$  = series resistance [ $\Omega$ ].

Example:

A 4  $\Omega$  speaker in BTL configuration can be used up to a supply voltage of 12 V without running into current limiting. Current limiting (clipping) will avoid audio holes, but it causes a sound distortion similar to voltage clipping.

## 14.3 Speaker configuration and impedance

For a flat-frequency response (second-order Butterworth filter) it is necessary to change the low-pass filter components  $L_{LC}$  and  $C_{LC}$  according to the speaker configuration and impedance. [Table 12](#) shows the practical required values:

**Table 12. Filter component values**

Configuration	Impedance [ $\Omega$ ]	$L_{LC}$ [ $\mu\text{F}$ ]	$C_{LC}$ [nF]
BTL	4	10	1500
	6	16	1000
	8	22	680

## 14.4 Differential input

For a high common-mode rejection ratio and a maximum of flexibility in the application, the audio inputs of the application are fully differential.

The input configuration for a differential-input application is illustrated in [Figure 28](#).

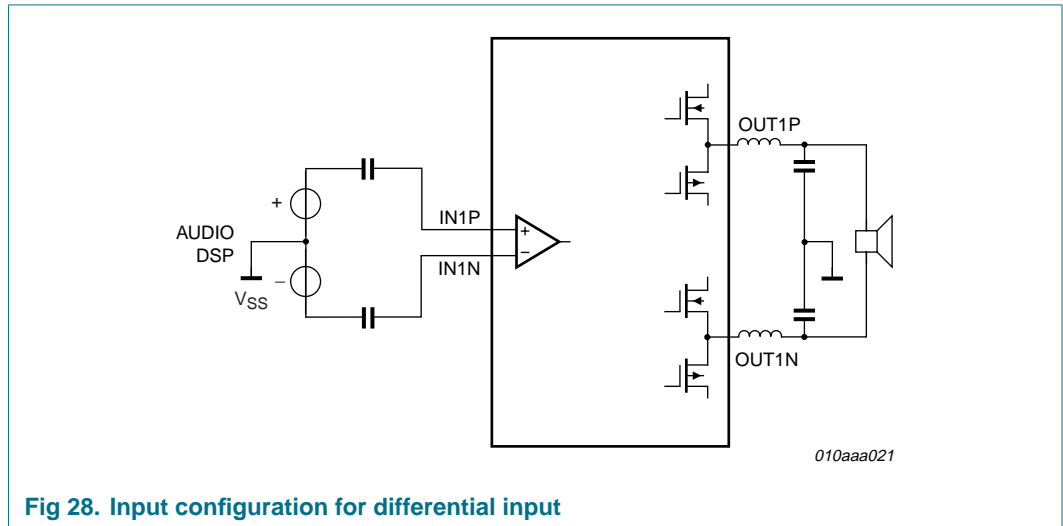


Fig 28. Input configuration for differential input

### 14.5 Single-ended input

When using an audio source with a single-ended 'out', it is important to connect the IN1N from the application board to the VSS/GND of the audio source (e.g. Audio DSP).

The input configuration for single-ended 'in' application is illustrated in [Figure 29](#).

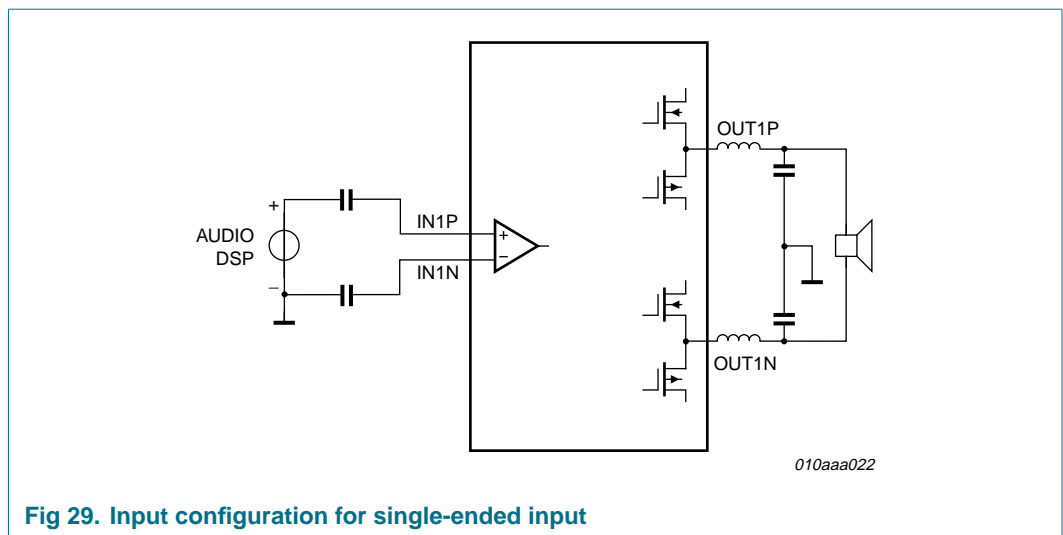


Fig 29. Input configuration for single-ended input

14.6 Curves measured in a typical application

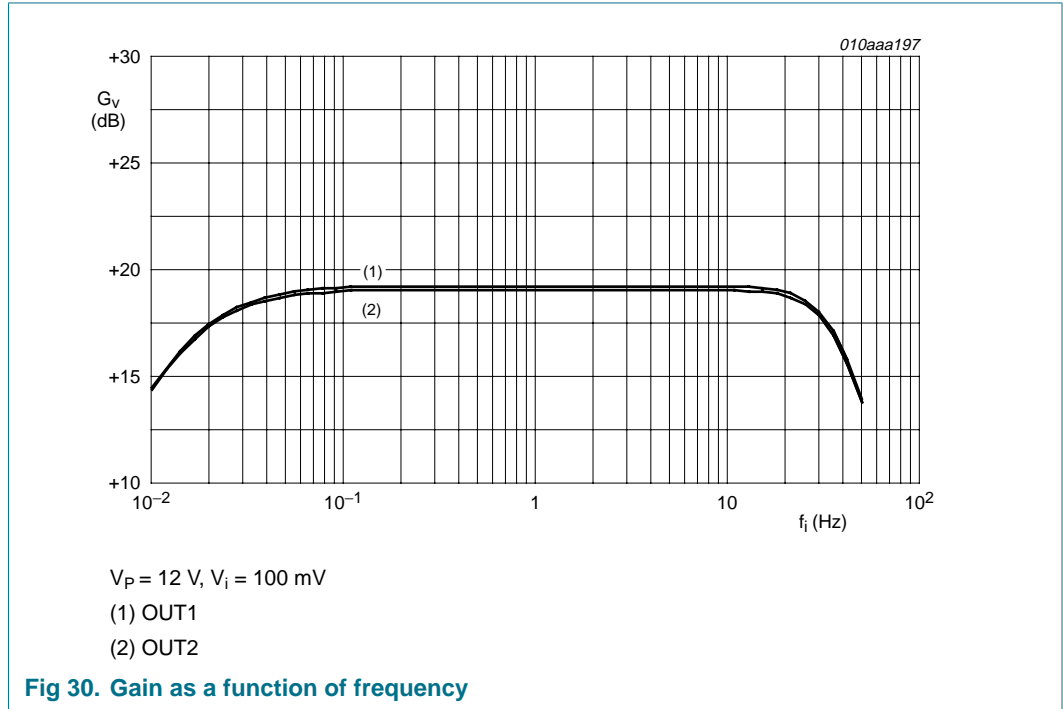


Fig 30. Gain as a function of frequency

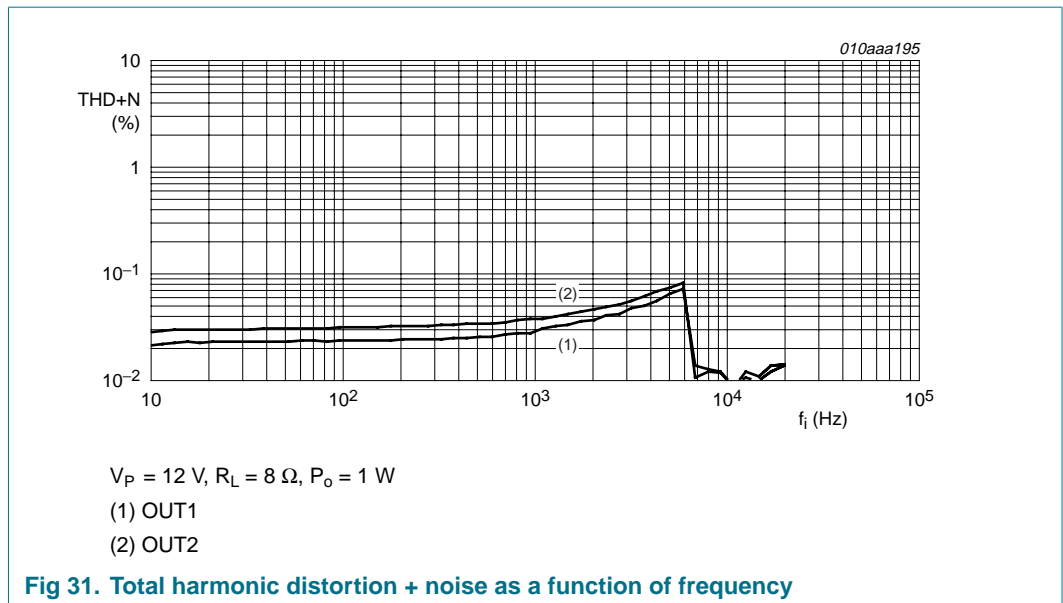
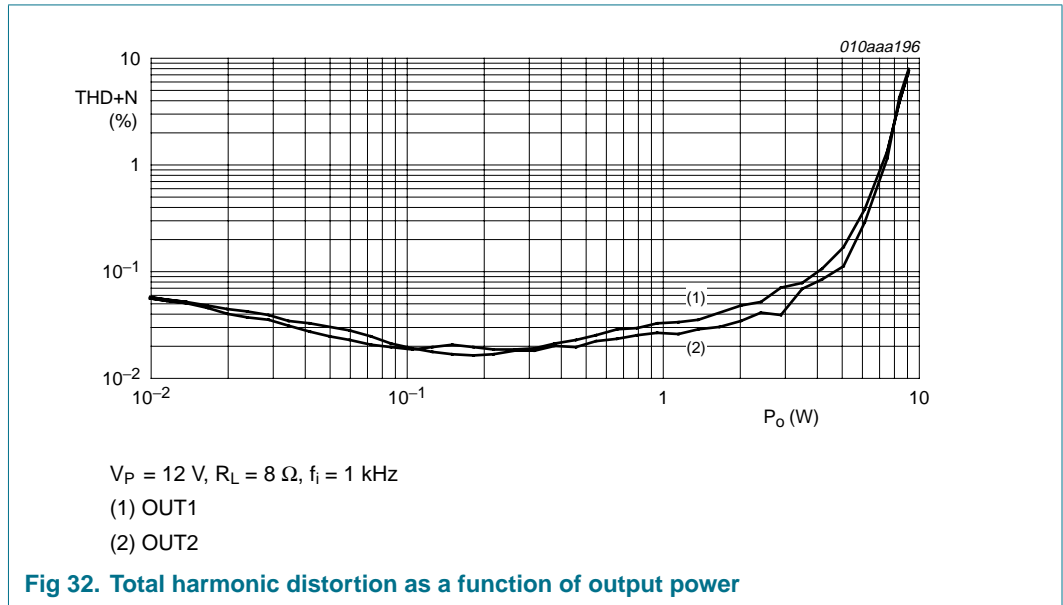
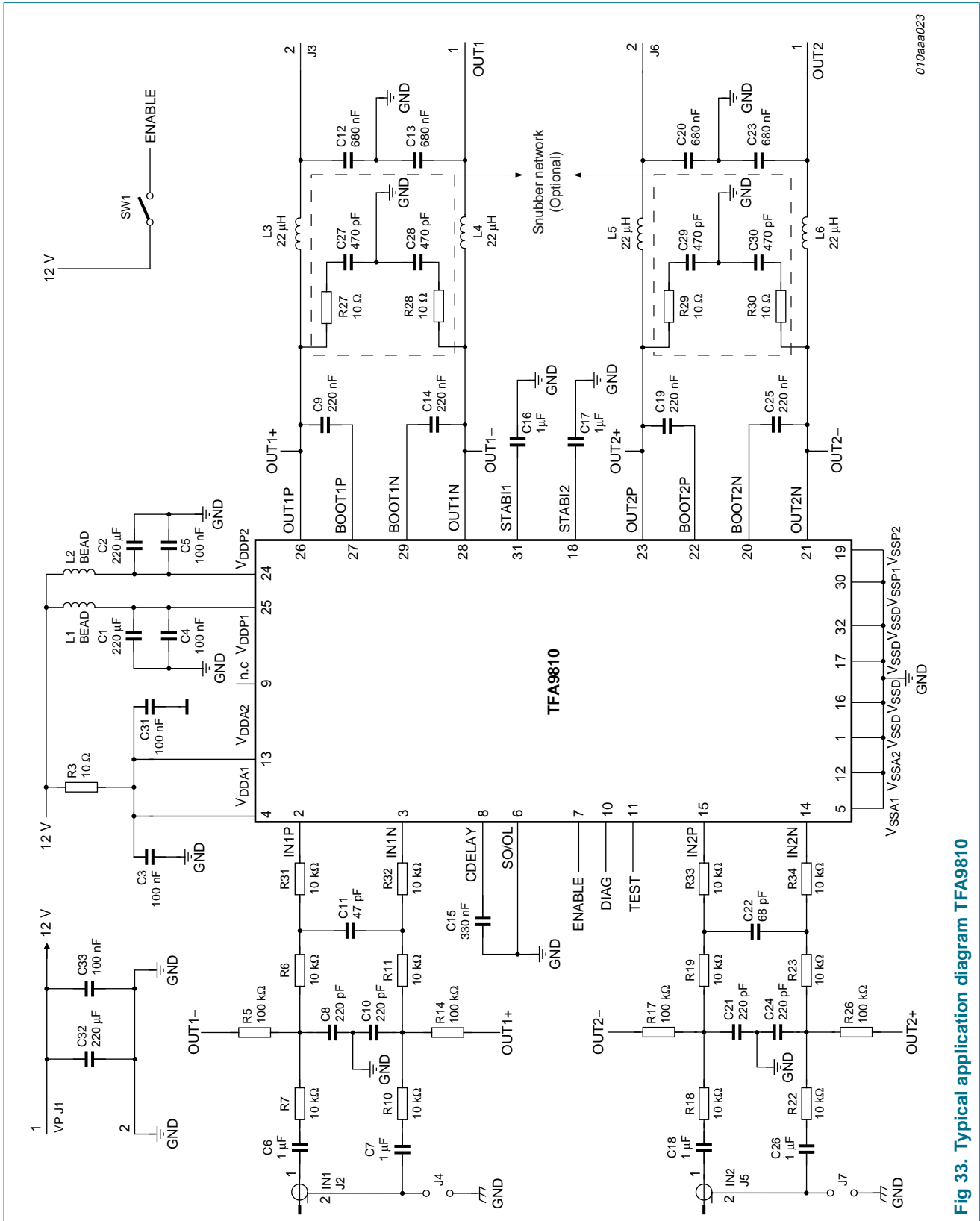


Fig 31. Total harmonic distortion + noise as a function of frequency



### 14.7 Typical application diagram TFA9810

The typical application diagram with the TFA9810 supplied from an asymmetrical supply is shown in [Figure 33](#).



010aaa023

Fig 33. Typical application diagram TFA9810



## 14.8 Typical application: bill of materials

Table 13. Typical application: bill of materials

Item	Quantity	Reference	Part	Description
1	2	C1, C2.	220 $\mu$ F/35 V	General purpose 85 °C, diameter 8 mm
2	5	C3, C4, C5, C31, C33.	100 nF/50 V	SMD 0805 X7R
3	2	C16, C17.	1 $\mu$ F/50 V	SMD 1206 X7R
4	4	C6, C7, C18, C26	1 $\mu$ F/25 V	MKT
5	4	C8, C10, C21, C24.	220 pF/25 V	SMD 0402 NP0
6	4	C9, C14, C19, C25.	220 nF/25 V	SMD 0805 X7R
7	1	C11	47 pF/25 V	SMD 0402 NP0
8	4	C12, C13, C20, C23.	680 nF/25 V	MKT
9	1	C15	330 nF/25 V	SMD 0805 X7R
10	1	C22.	68 pF/25 V	SMD 0402 NP0
11	1	C32	1000 $\mu$ F/25 V	CE12-02R
12	3	J1, J3, J6.	Screw terminal	Two pins
13	2	J2, J5.	CINCH	CINCH
14	2	J4, J7	Jumper	Closed on demo board only
15	2	L1, L2.	BEAD	SMD 1206 Würth Elektronik DC < 0.5 $\Omega$ 10 MHz > 80 $\Omega$
16	4	L3, L4, L5, L6.	22 $\mu$ H	8RDY TOKO A7040HN-220M, 11RHBP TOKO A7503CY-220M or Sagami 7311NA-220M
17	5	R3	10 / 0.25 W / 5 %	SMD 1206
18	4	R5, R14, R17, R26.	100 k / 0.1 W / 1 % for 20 dB 200 k / 0.1 W / 1 % for 26 dB	SMD 0603
19	12	R6, R7, R10, R11, R18, R19, R22, R23, R31, R32, R33, R34.	10 k / 0.1 W / 1 %	SMD 0603
20	1	SW1	PCB switch	Secme 090320901
21	1	U1	TFA9810T	SOT287-1 (SO32) NXP Semiconductors

## 14.9 Snubber network

Table 14. Snubber network: bill of materials

Item	Quantity	Reference	Part	Footprint
1	4	C27, C28, C29, C30	470 pF, 25 V	SMD 0805 X7R
2	4	R27, R28, R29, R30	10 / 0.25 W / 5 %	SMD 1206

## 15. Test information

General Quality Specification for General Applications, Power management and RF Power. Document SNW-FQ-611 refers.

16. Package outline

SO32: plastic small outline package; 32 leads; body width 7.5 mm

SOT287-1

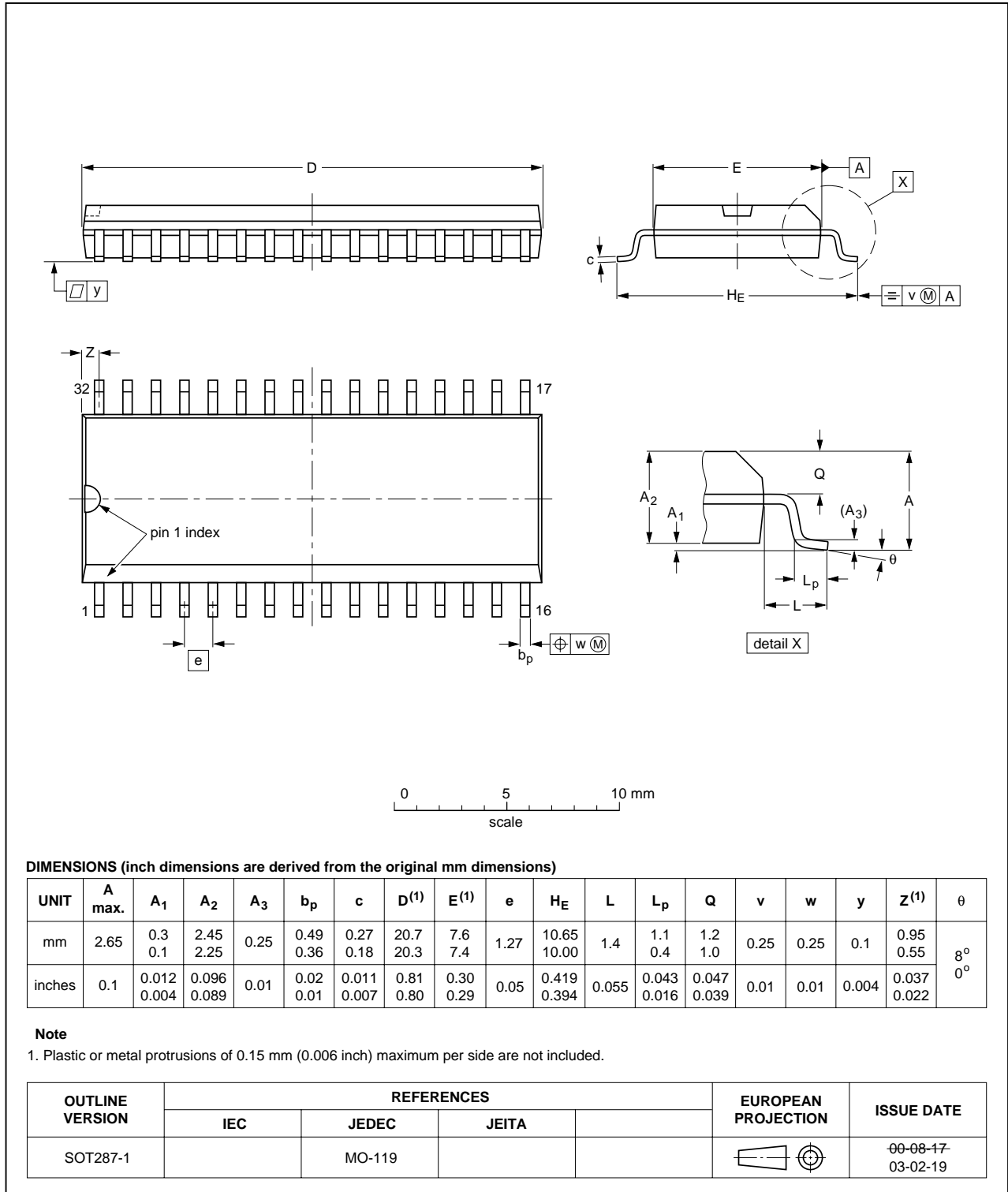


Fig 34. Package outline

## 17. Revision history

Table 15. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
TFA9810_2	20070830	Preliminary data sheet	-	TFA9810_1
Modifications:	• Added legal statement on “Right to make changes” in <a href="#">Section 18.3</a> .			
TFA9810_1	20070815	Preliminary data sheet	-	-

## 18. Legal information

### 18.1 Data sheet status

Document status <sup>[1][2]</sup>	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL <http://www.nxp.com>.

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Date of release: 30 August 2007

Document identifier: TFA9810\_2