



1 MHz to 2.7 GHz RF Gain Block

AD8353

FEATURES

- Fixed gain of 20 dB
- Operational frequency of 1 MHz to 2.7 GHz
- Linear output power up to 9 dBm
- Input/output internally matched to 50 Ω
- Temperature and power supply stable
- Noise figure: 5.3 dB
- Power supply: 3 V or 5 V

APPLICATIONS

- VCO buffers
- General Tx/Rx amplification
- Power amplifier predrivers
- Low power antenna drivers

GENERAL DESCRIPTION

The AD8353 is a broadband, fixed-gain, linear amplifier that operates at frequencies from 1 MHz up to 2.7 GHz. It is intended for use in a wide variety of wireless devices, including cellular, broadband, CATV, and LMDS/MMDS applications.

By taking advantage of ADI's high performance, complementary Si bipolar process, these gain blocks provide excellent stability over process, temperature, and power supply. This amplifier is single-ended and internally matched to 50 Ω with a return loss of greater than 10 dB over the full operating frequency range.

The AD8353 provides linear output power of 9 dBm with 20 dB of gain at 900 MHz when biased at 3 V and an external RF choke is connected between the power supply and the output pin. The dc supply current is 42 mA. At 900 MHz, the output third-order intercept (OIP3) is greater than 23 dBm and is 19 dBm at 2.7 GHz.

FUNCTIONAL BLOCK DIAGRAM

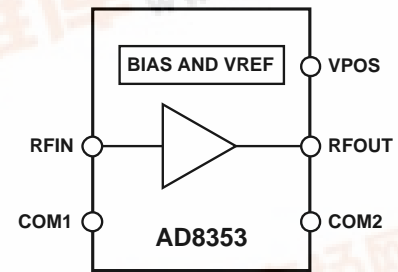


Figure 1.

The noise figure is 5.3 dB at 900 MHz. The reverse isolation (S_{12}) is -36 dB at 900 MHz and -30 dB at 2.7 GHz.

The AD8353 can also operate with a 5 V power supply; in which case, no external inductor is required. Under these conditions, the AD8353 delivers 8 dBm with 20 dB of gain at 900 MHz. The dc supply current is 42 mA. At 900 MHz, the OIP3 is greater than 22 dBm and is 19 dBm at 2.7 GHz. The noise figure is 5.6 dB at 900 MHz. The reverse isolation (S_{12}) is -35 dB.

The AD8353 is fabricated on ADI's proprietary, high performance, 25 GHz, Si complementary, bipolar IC process. The AD8353 is available in a chip scale package that uses an exposed paddle for excellent thermal impedance and low impedance electrical connection to ground. It operates over a -40°C to +85°C temperature range, and an evaluation board is also available.



找PDF
Rev. B

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REVISION HISTORY

12/05—Rev. A to Rev. B

| | |
|---|----|
| Changes to Table 1 | 3 |
| Changes to Table 2 | 4 |
| Changes to Figure 16 | 9 |
| Changes to Figure 32 | 11 |
| Moved Figure 39 to Page 15; Renumbered Sequentially | 15 |
| Changes to Ordering Guide | 16 |

8/05—Rev. 0 to Rev. A

| | |
|--|-----------|
| Updated Format | Universal |
| Changes to Product Title | 1 |
| Changes to Features, Figure 1, and General Description | 1 |
| Changes to Table 1 | 3 |
| Changes to Table 2 | 4 |
| Changes to Figure 2 and Table 4 | 6 |
| Changes to Figure 3 caption and Figure 6 caption | 7 |
| Changes to Figure 17 caption and Figure 20 caption | 9 |
| Changes to Basic Connections Section | 13 |
| Added Low Frequency Applications Below 100 MHz Section | 14 |
| Changes to Table 5 | 15 |
| Changes to Ordering Guide | 16 |
| Updated Outline Dimensions | 16 |

2/02—Revision 0: Initial Version

SPECIFICATIONS

$V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, 100 nH external inductor between RFOUT and VPOS, $Z_O = 50\ \Omega$, unless otherwise noted.

Table 1.

| Parameter | Conditions | Min | Typ | Max | Unit |
|--------------------------------|--|-----|--------|------|------------------------------|
| OVERALL FUNCTION | | | | | |
| Frequency Range | | 1 | | 2700 | MHz |
| Gain | $f = 900\text{ MHz}$ | | 19.8 | | dB |
| | $f = 1.9\text{ GHz}$ | | 17.7 | | dB |
| | $f = 2.7\text{ GHz}$ | | 15.6 | | dB |
| Delta Gain | $f = 900\text{ MHz}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -0.97 | | dB |
| | $f = 1.9\text{ GHz}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.15 | | dB |
| | $f = 2.7\text{ GHz}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.34 | | dB |
| Gain Supply Sensitivity | VPOS $\pm 10\%$, $f = 900\text{ MHz}$ | | 0.04 | | dB/V |
| | $f = 1.9\text{ GHz}$ | | -0.004 | | dB/V |
| | $f = 2.7\text{ GHz}$ | | -0.04 | | dB/V |
| Reverse Isolation (S_{12}) | $f = 900\text{ MHz}$ | | -35.6 | | dB |
| | $f = 1.9\text{ GHz}$ | | -34.9 | | dB |
| | $f = 2.7\text{ GHz}$ | | -30.3 | | dB |
| RF INPUT INTERFACE | Pin RFIN | | | | |
| Input Return Loss | $f = 900\text{ MHz}$ | | 22.3 | | dB |
| | $f = 1.9\text{ GHz}$ | | 20.9 | | dB |
| | $f = 2.7\text{ GHz}$ | | 11.2 | | dB |
| RF OUTPUT INTERFACE | Pin RFOUT | | | | |
| Output Compression Point | $f = 900\text{ MHz}$, 1 dB compression | | 9.1 | | dBm |
| | $f = 1.9\text{ GHz}$ | | 8.4 | | dBm |
| | $f = 2.7\text{ GHz}$ | | 7.6 | | dBm |
| Delta Compression Point | $f = 900\text{ MHz}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.46 | | dB |
| | $f = 1.9\text{ GHz}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.17 | | dB |
| | $f = 2.7\text{ GHz}, -40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1 | | dB |
| Output Return Loss | $f = 900\text{ MHz}$ | | 26.3 | | dB |
| | $f = 1.9\text{ GHz}$ | | 16.9 | | dB |
| | $f = 2.7\text{ GHz}$ | | 13.3 | | dB |
| DISTORTION/NOISE | | | | | |
| Output Third-Order Intercept | $f = 900\text{ MHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 23.6 | | dBm |
| | $f = 1.9\text{ GHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 20.8 | | dBm |
| | $f = 2.7\text{ GHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 19.5 | | dBm |
| Output Second-Order Intercept | $f = 900\text{ MHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 31.6 | | dBm |
| Noise Figure | $f = 900\text{ MHz}$ | | 5.3 | | dB |
| | $f = 1.9\text{ GHz}$ | | 6 | | dB |
| | $f = 2.7\text{ GHz}$ | | 6.8 | | dB |
| POWER INTERFACE | Pin VPOS | | | | |
| Supply Voltage | | 2.7 | 3 | 3.3 | V |
| Total Supply Current | | 35 | 41 | 48 | mA |
| Supply Voltage Sensitivity | | | 15.3 | | mA/V |
| Temperature Sensitivity | $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | 60 | | $\mu\text{A}/^\circ\text{C}$ |

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$V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, no external inductor between RFOUT and VPOS, $Z_O = 50\ \Omega$, unless otherwise noted.

Table 2.

| Parameter | Conditions | Min | Typ | Max | Unit |
|--------------------------------|--|-----|-------|------|------------------------------|
| OVERALL FUNCTION | | | | | |
| Frequency Range | | 1 | | 2700 | MHz |
| Gain | $f = 900\text{ MHz}$ | | 19.5 | | dB |
| | $f = 1.9\text{ GHz}$ | | 17.6 | | dB |
| | $f = 2.7\text{ GHz}$ | | 15.7 | | dB |
| Delta Gain | $f = 900\text{ MHz}$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -0.96 | | dB |
| | $f = 1.9\text{ GHz}$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.18 | | dB |
| | $f = 2.7\text{ GHz}$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.38 | | dB |
| Gain Supply Sensitivity | VPOS $\pm 10\%$, $f = 900\text{ MHz}$ | | 0.09 | | dB/V |
| | $f = 1.9\text{ GHz}$ | | -0.01 | | dB/V |
| | $f = 2.7\text{ GHz}$ | | -0.09 | | dB/V |
| Reverse Isolation (S_{12}) | $f = 900\text{ MHz}$ | | -35.4 | | dB |
| | $f = 1.9\text{ GHz}$ | | -34.6 | | dB |
| | $f = 2.7\text{ GHz}$ | | -30.2 | | dB |
| RF INPUT INTERFACE | Pin RFIN | | | | |
| Input Return Loss | $f = 900\text{ MHz}$ | | 22.9 | | dB |
| | $f = 1.9\text{ GHz}$ | | 21.7 | | dB |
| | $f = 2.7\text{ GHz}$ | | 11.5 | | dB |
| RF OUTPUT INTERFACE | Pin RFOUT | | | | |
| Output Compression Point | $f = 900\text{ MHz}$ | | 8.3 | | dBm |
| | $f = 1.9\text{ GHz}$ | | 8.1 | | dBm |
| | $f = 2.7\text{ GHz}$ | | 7.5 | | dBm |
| Delta Compression Point | $f = 900\text{ MHz}$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.05 | | dB |
| | $f = 1.9\text{ GHz}$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.49 | | dB |
| | $f = 2.7\text{ GHz}$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | -1.33 | | dB |
| Output Return Loss | $f = 900\text{ MHz}$ | | 27 | | dB |
| | $f = 1.9\text{ GHz}$ | | 22 | | dB |
| | $f = 2.7\text{ GHz}$ | | 14.3 | | dB |
| DISTORTION/NOISE | | | | | |
| Output Third-Order Intercept | $f = 900\text{ MHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 22.8 | | dBm |
| | $f = 1.9\text{ GHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 20.6 | | dBm |
| | $f = 2.7\text{ GHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 19.5 | | dBm |
| Output Second-Order Intercept | $f = 900\text{ MHz}$, $\Delta f = 1\text{ MHz}$, $P_{IN} = -28\text{ dBm}$ | | 30.3 | | dBm |
| Noise Figure | $f = 900\text{ MHz}$ | | 5.6 | | dB |
| | $f = 1.9\text{ GHz}$ | | 6.3 | | dB |
| | $f = 2.7\text{ GHz}$ | | 7.1 | | dB |
| POWER INTERFACE | Pin VPOS | | | | |
| Supply Voltage | | 4.5 | 5 | 5.5 | V |
| Total Supply Current | | 35 | 42 | 52 | mA |
| Supply Voltage Sensitivity | | | 4.3 | | mA/V |
| Temperature Sensitivity | $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ | | 45.7 | | $\mu\text{A}/^\circ\text{C}$ |

ABSOLUTE MAXIMUM RATINGS

Table 3.

| Parameter | Rating |
|-------------------------------------|-----------------|
| Supply Voltage, VPOS | 5.5 V |
| Input Power (re: 50 Ω) | 10 dBm |
| Equivalent Voltage | 700 mV rms |
| Internal Power Dissipation | |
| Paddle Not Soldered | 325 mW |
| Paddle Soldered | 812 mW |
| θ_{JA} (Paddle Soldered) | 80°C/W |
| θ_{JA} (Paddle Not Soldered) | 200°C/W |
| Maximum Junction Temperature | 150°C |
| Operating Temperature Range | –40°C to +85°C |
| Storage Temperature Range | –65°C to +150°C |
| Lead Temperature (Soldering 60 sec) | 240°C |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



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PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

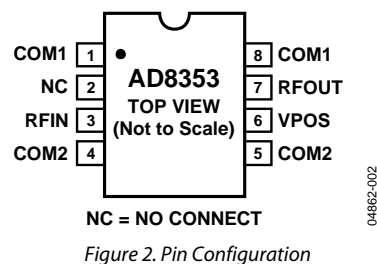


Table 4. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
|---------|----------|---|
| 1, 8 | COM1 | Device Common. Connect to low impedance ground. |
| 2 | NC | No Connection. |
| 3 | RFIN | RF Input Connection. Must be ac-coupled. |
| 4, 5 | COM2 | Device Common. Connect to low impedance ground. |
| 6 | VPOS | Positive Supply Voltage. |
| 7 | RFOUT | RF Output Connection. Must be ac-coupled. |

TYPICAL PERFORMANCE CHARACTERISTICS

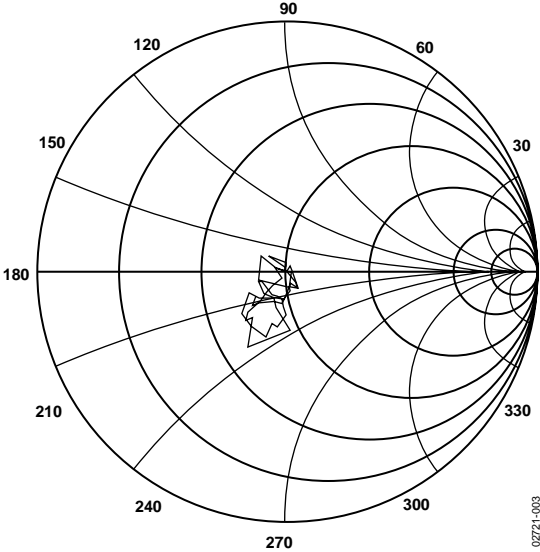


Figure 3. S_{11} vs. Frequency, $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $dc \leq f \leq 3\text{ GHz}$

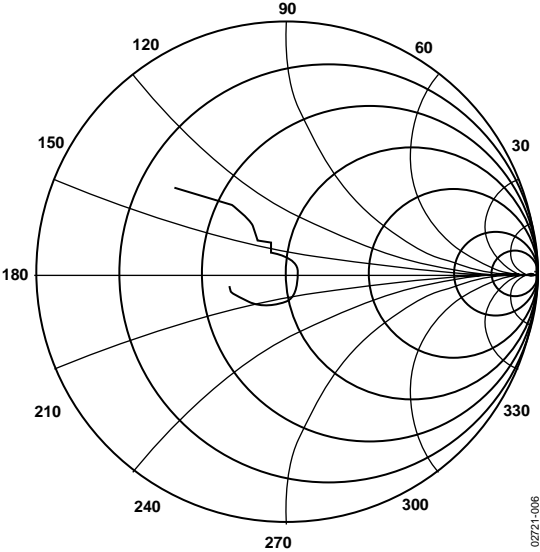


Figure 6. S_{22} vs. Frequency, $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $dc \leq f \leq 3\text{ GHz}$

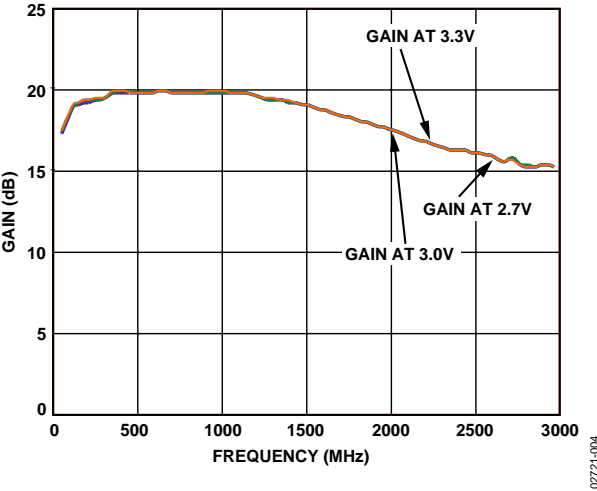


Figure 4. Gain vs. Frequency, $V_S = 2.7\text{ V}$, 3 V , and 3.3 V , $T_A = 25^\circ\text{C}$

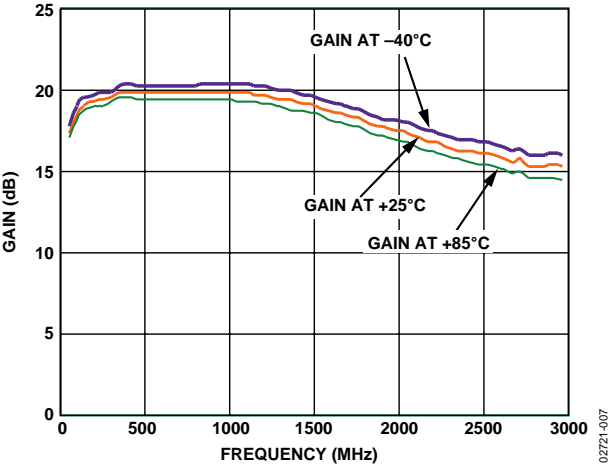


Figure 7. Gain vs. Frequency, $V_S = 3\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

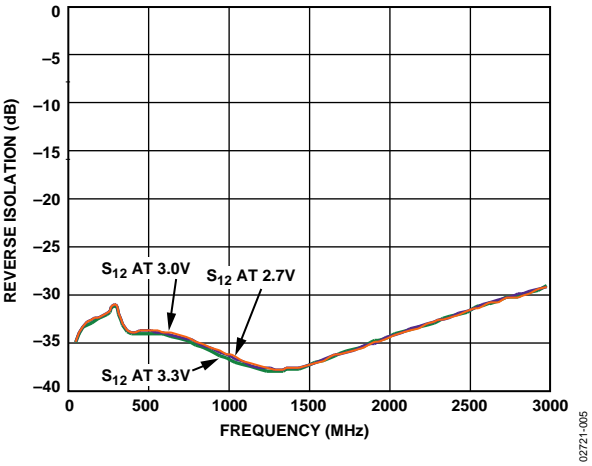


Figure 5. Reverse Isolation vs. Frequency, $V_S = 2.7\text{ V}$, 3 V , and 3.3 V , $T_A = 25^\circ\text{C}$

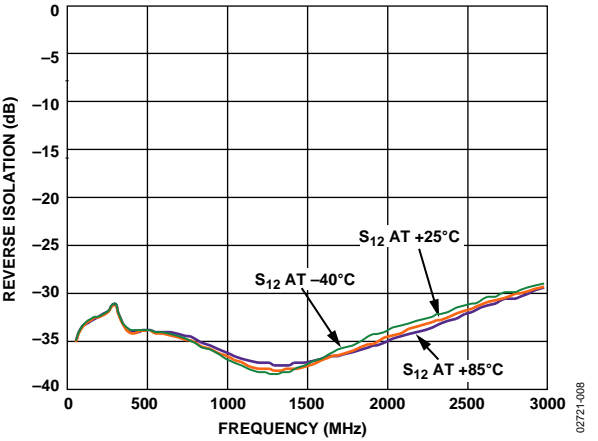


Figure 8. Reverse Isolation vs. Frequency, $V_S = 3\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

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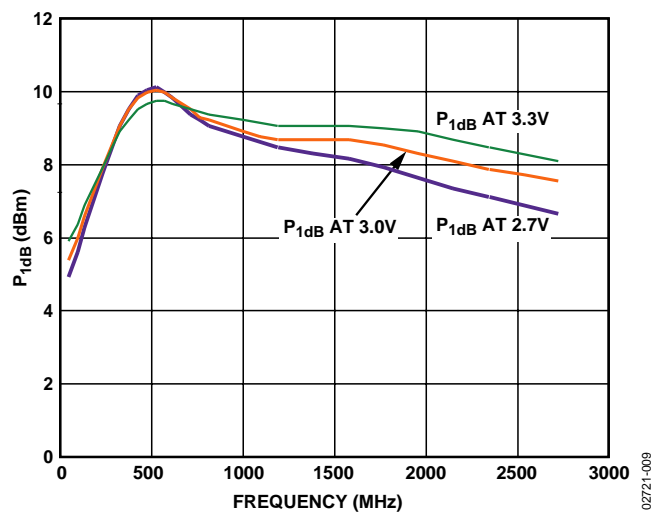


Figure 9. P_{1dB} vs. Frequency, $V_S = 2.7\text{ V}$, 3 V , and 3.3 V , $T_A = 25^\circ\text{C}$

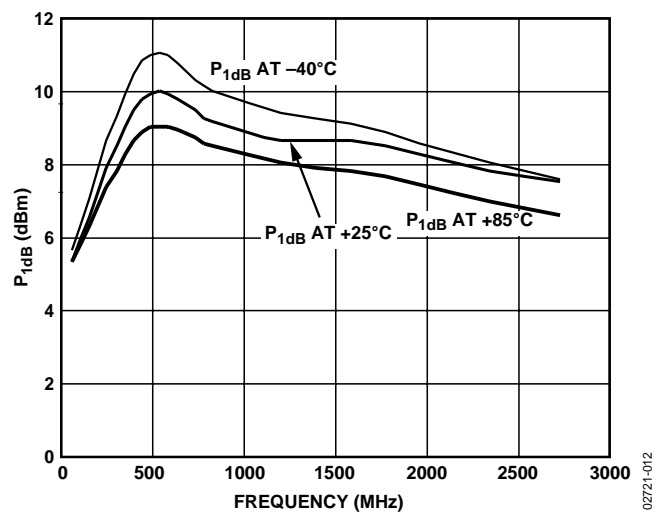


Figure 12. P_{1dB} vs. Frequency, $V_S = 3\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

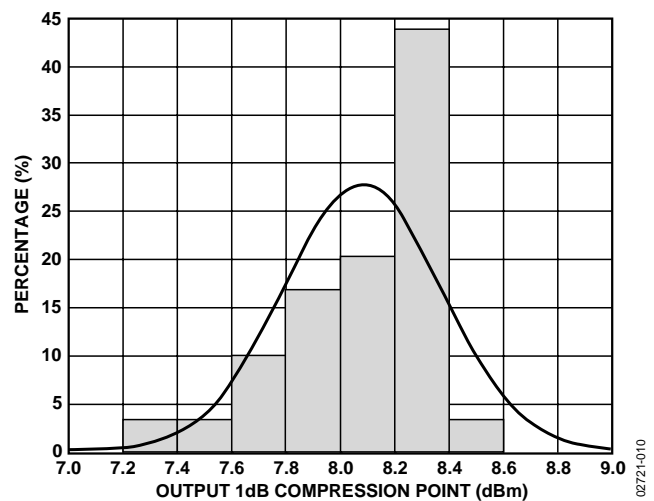


Figure 10. Distribution of P_{1dB} , $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 2.2\text{ GHz}$

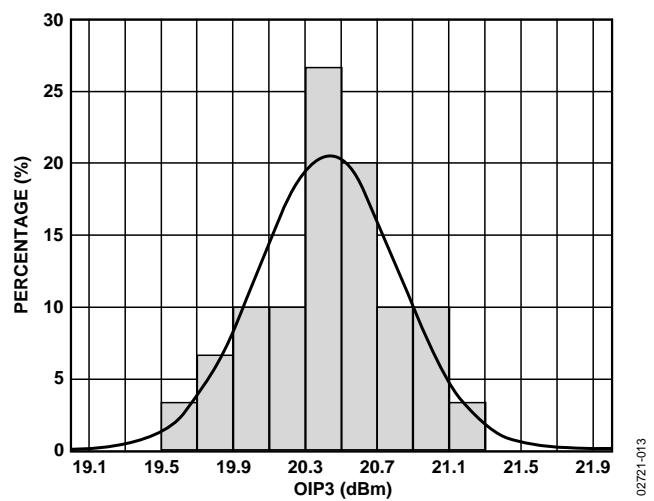


Figure 13. Distribution of OIP_3 , $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 2.2\text{ GHz}$

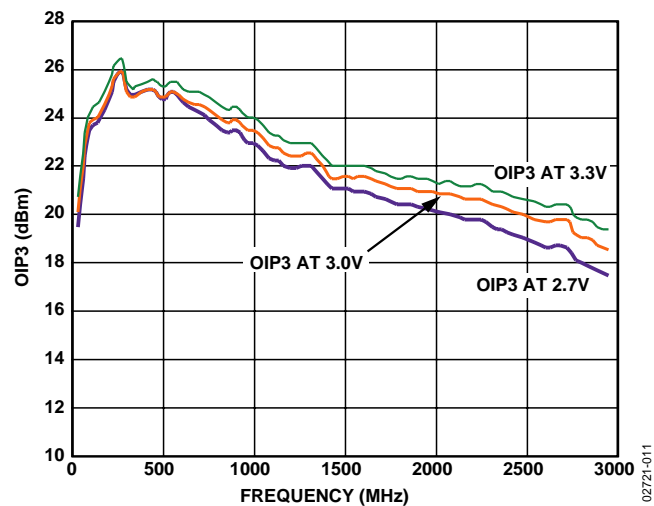


Figure 11. OIP_3 vs. Frequency, $V_S = 2.7\text{ V}$, 3 V , and 3.3 V , $T_A = 25^\circ\text{C}$

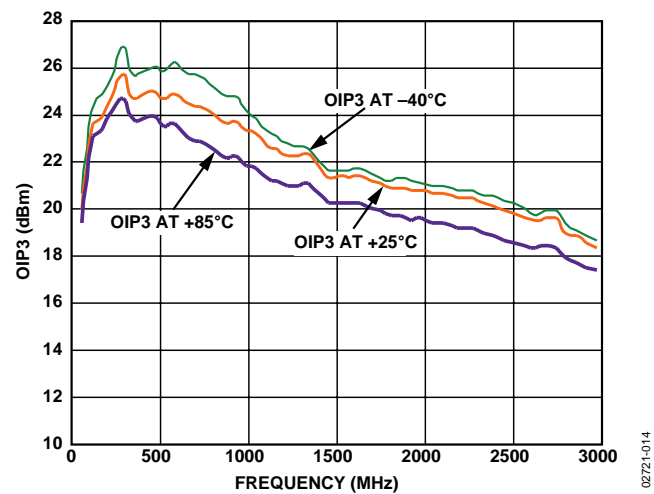


Figure 14. OIP_3 vs. Frequency, $V_S = 3\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

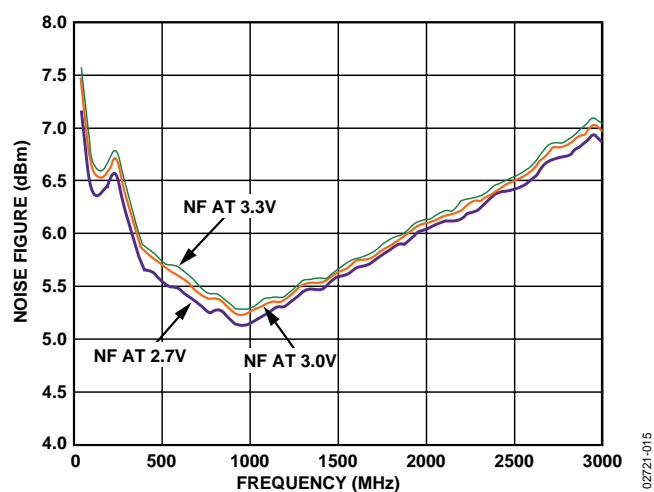


Figure 15. Noise Figure vs. Frequency, $V_S = 2.7\text{ V}$, 3 V , and 3.3 V , $T_A = 25^\circ\text{C}$

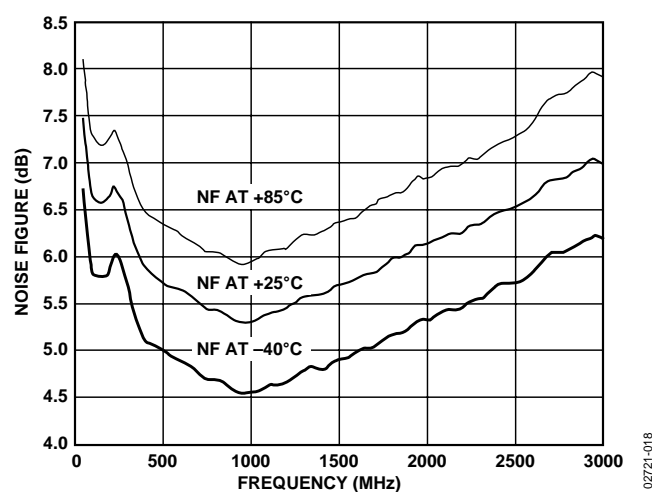


Figure 18. Noise Figure vs. Frequency, $V_S = 3\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

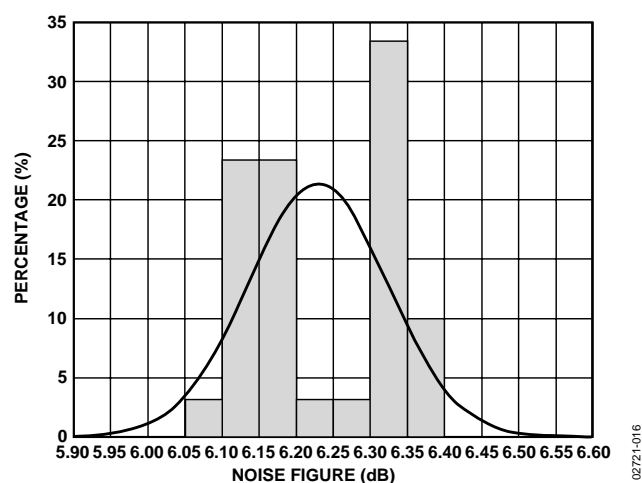


Figure 16. Distribution of Noise Figure, $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 2.2\text{ GHz}$

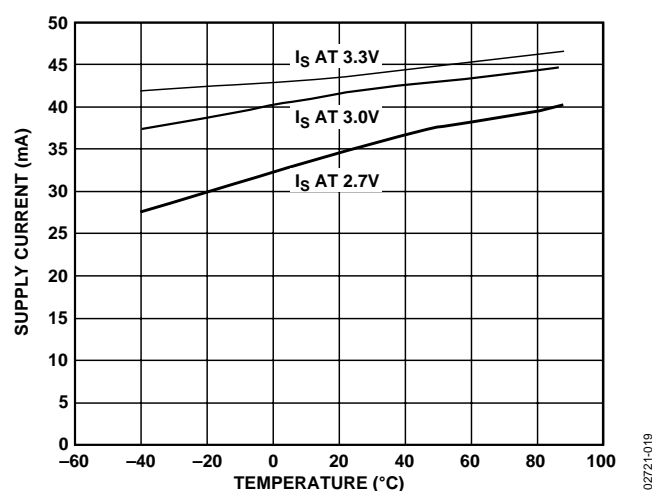


Figure 19. Supply Current vs. Temperature, $V_S = 2.7\text{ V}$, 3 V , and 3.3 V

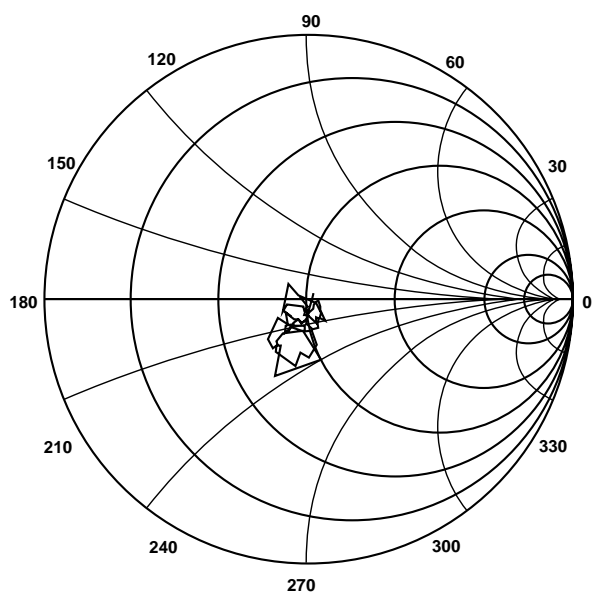


Figure 17. S_{11} vs. Frequency, $V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $dc \leq f \leq 3\text{ GHz}$

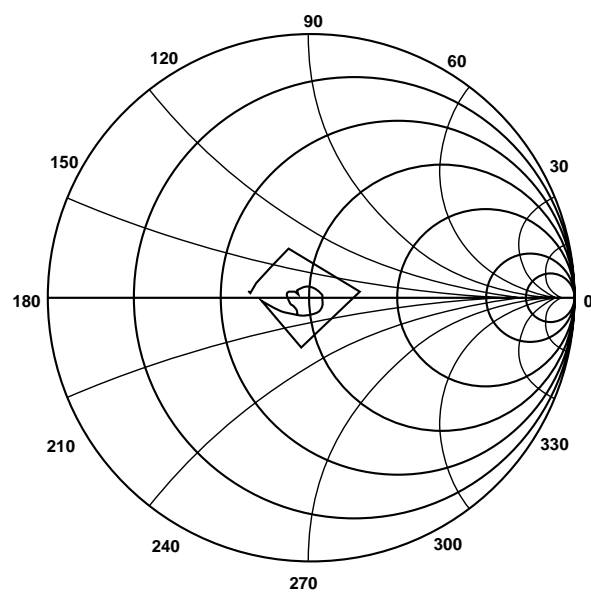


Figure 20. S_{22} vs. Frequency, $V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $dc \leq f \leq 3\text{ GHz}$

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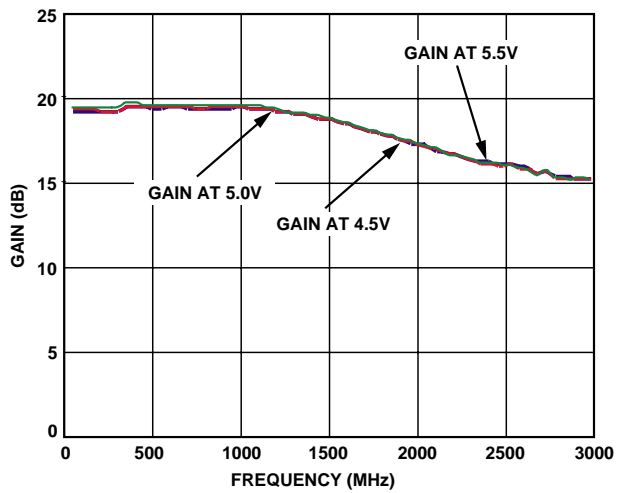


Figure 21. Gain vs. Frequency, $V_S = 4.5\text{ V}$, 5 V , and 5.5 V , $T_A = 25^\circ\text{C}$

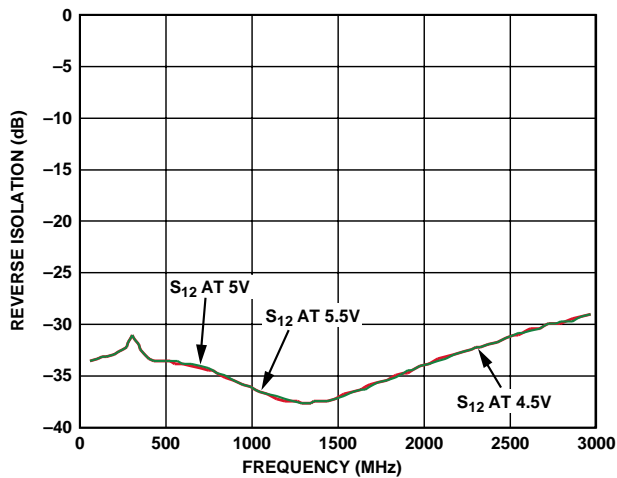


Figure 22. Reverse Isolation vs. Frequency, $V_S = 4.5\text{ V}$, 5 V , and 5.5 V , $T_A = 25^\circ\text{C}$

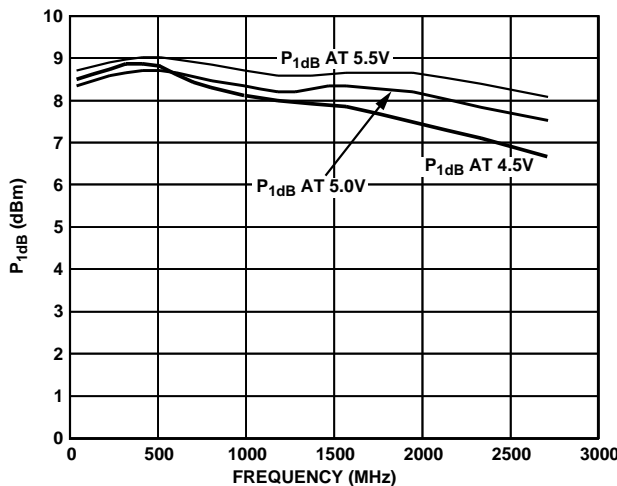


Figure 23. $P_{1\text{dB}}$ vs. Frequency, $V_S = 4.5\text{ V}$, 5 V , and 5.5 V , $T_A = 25^\circ\text{C}$

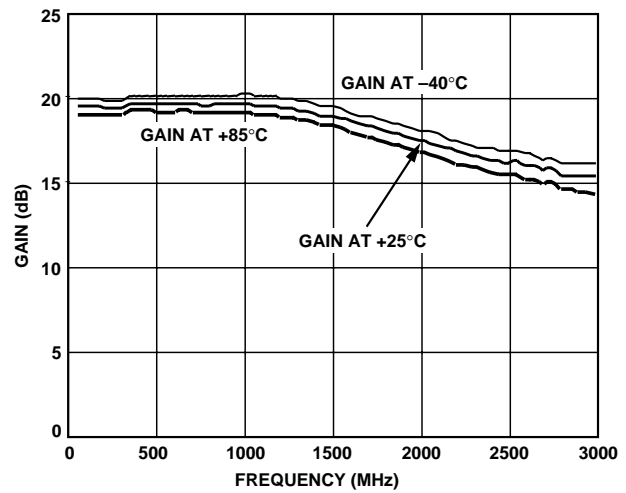


Figure 24. Gain vs. Frequency, $V_S = 5\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

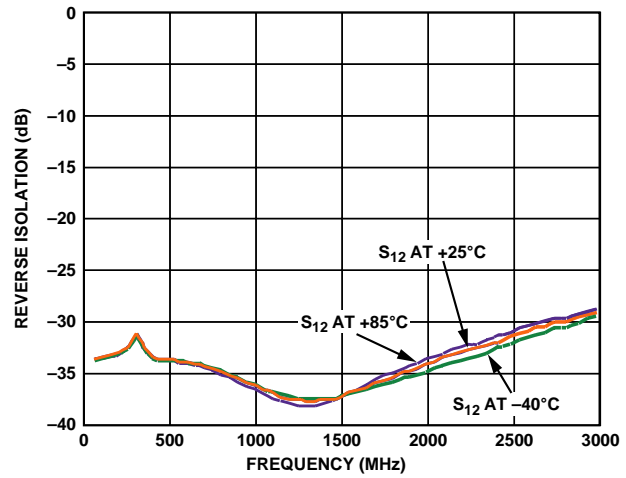


Figure 25. Reverse Isolation vs. Frequency, $V_S = 5\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

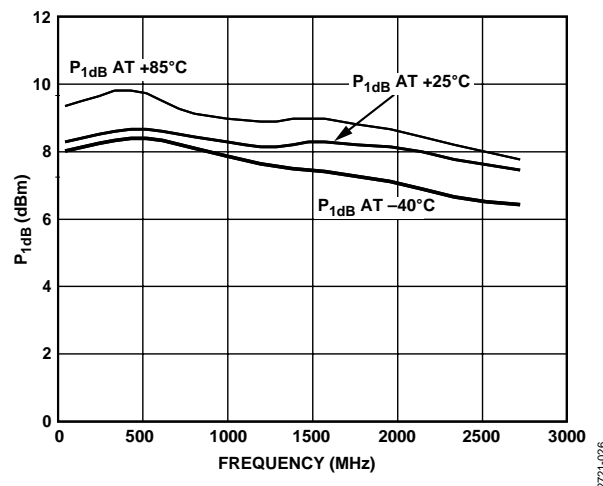


Figure 26. $P_{1\text{dB}}$ vs. Frequency, $V_S = 5\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

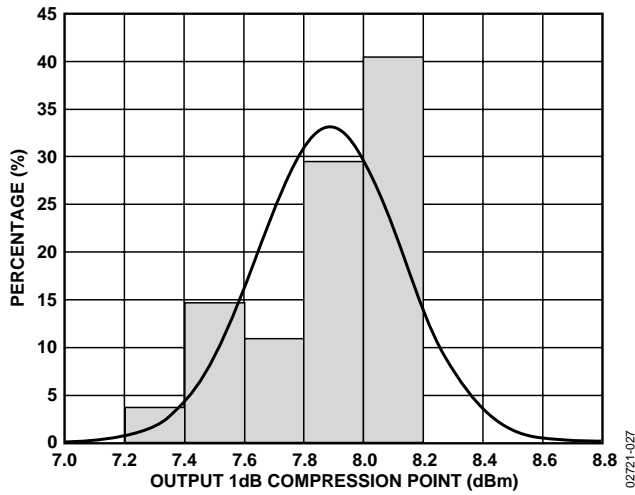


Figure 27. Distribution of P_{1dB} , $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 2.2\text{ GHz}$

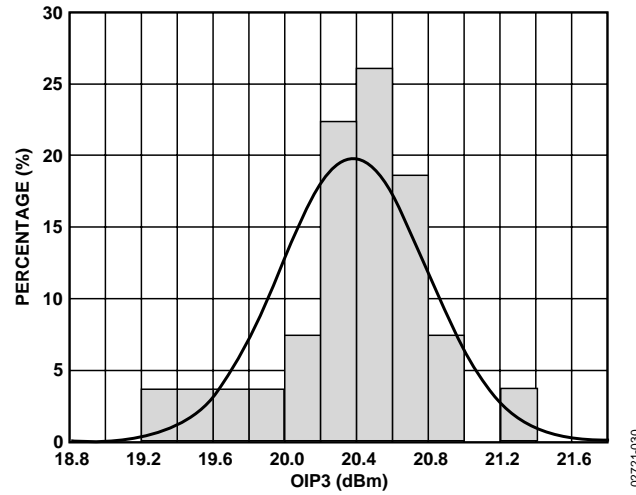


Figure 30. Distribution of OIP3, $V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 2.2\text{ GHz}$

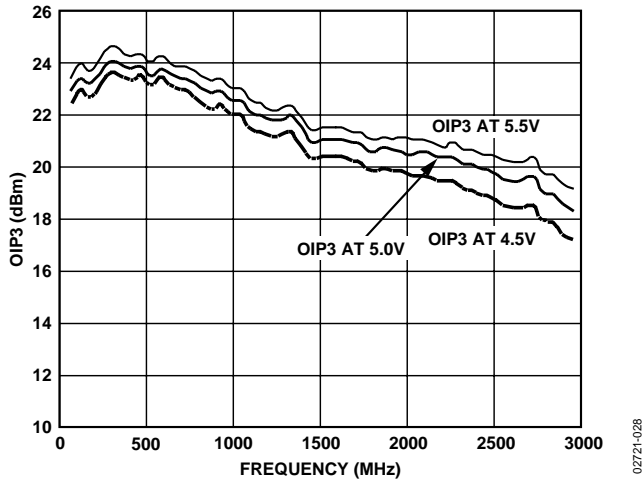


Figure 28. OIP3 vs. Frequency, $V_S = 4.5\text{ V}$, 5 V , and 5.5 V , $T_A = 27^\circ\text{C}$

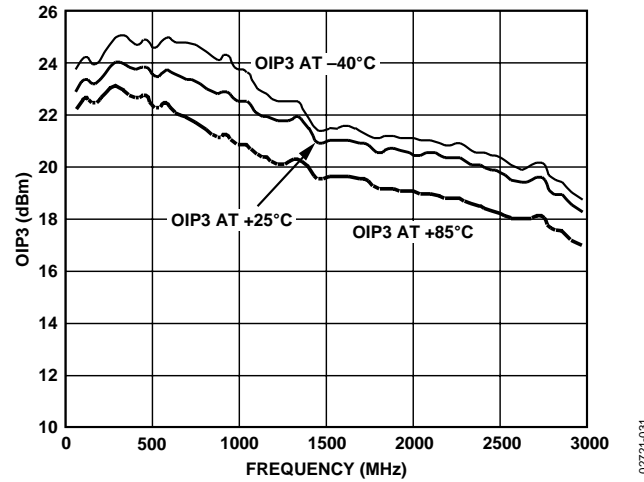


Figure 31. OIP3 vs. Frequency, $V_S = 5\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

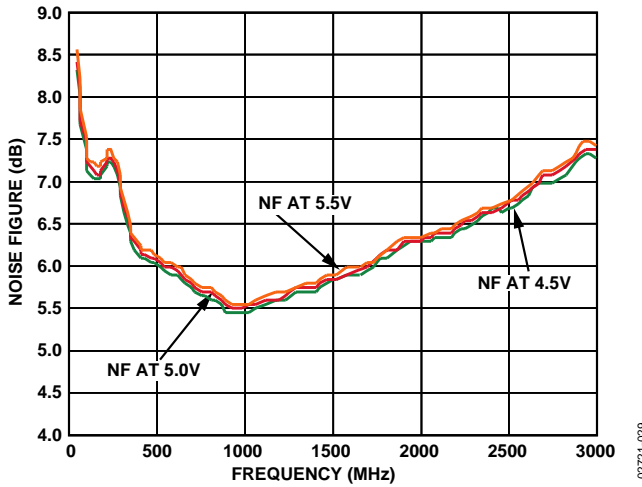


Figure 29. Noise Figure vs. Frequency, $V_S = 4.5\text{ V}$, 5 V , and 5.5 V , $T_A = 25^\circ\text{C}$

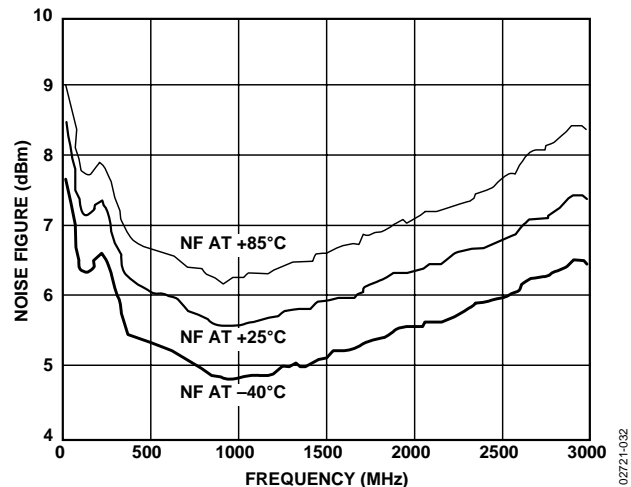


Figure 32. Noise Figure vs. Frequency, $V_S = 5\text{ V}$, $T_A = -40^\circ\text{C}$, $+25^\circ\text{C}$, and $+85^\circ\text{C}$

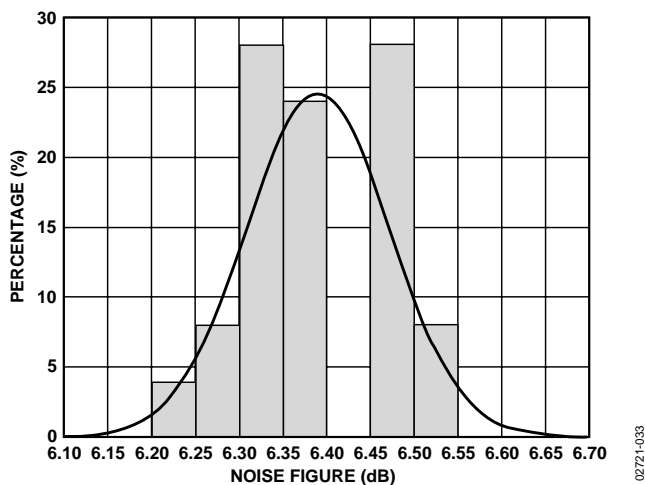


Figure 33. Distribution of Noise Figure, $V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 2.2\text{ GHz}$

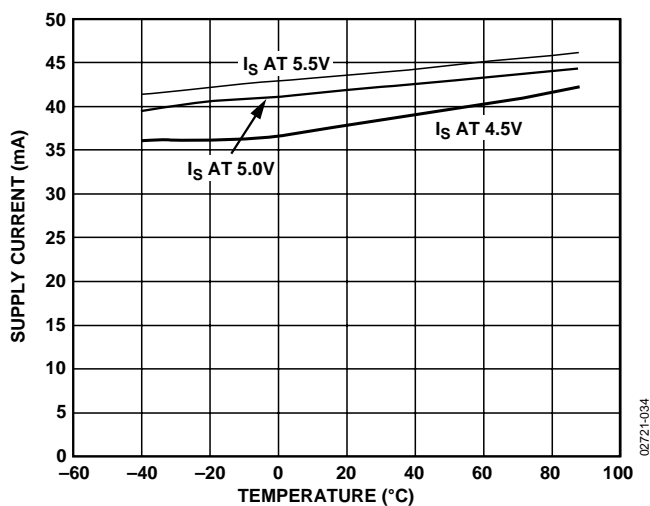


Figure 34. Supply Current vs. Temperature, $V_S = 4.5\text{ V}$, 5 V , and 5.5 V

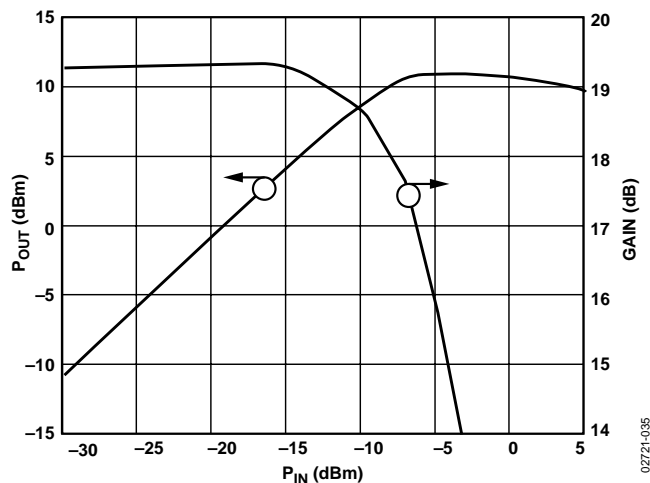


Figure 35. Output Power and Gain vs. Input Power, $V_S = 3\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 900\text{ MHz}$

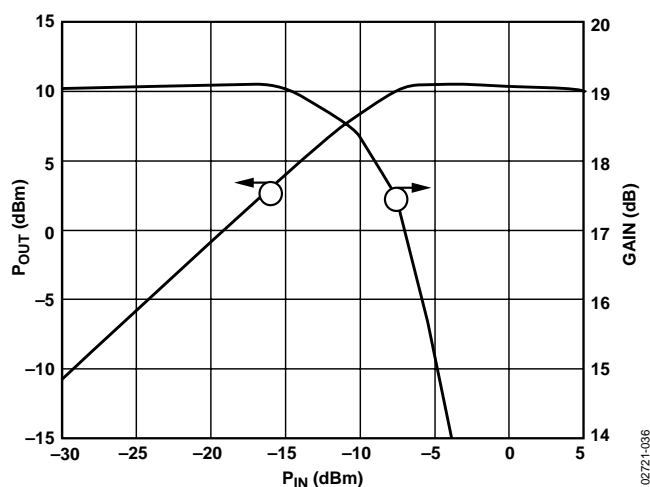


Figure 36. Output Power and Gain vs. Input Power, $V_S = 5\text{ V}$, $T_A = 25^\circ\text{C}$, $f = 900\text{ MHz}$

THEORY OF OPERATION

The AD8353 is a 2-stage, feedback amplifier employing both shunt-series and shunt-shunt feedback. The first stage is degenerated and resistively loaded and provides approximately 10 dB of gain. The second stage is a PNP-NPN Darlington output stage, which provides another 10 dB of gain. Series-shunt feedback from the emitter of the output transistor sets the input impedance to 50 Ω over a broad frequency range. Shunt-shunt feedback from the amplifier output to the input of the Darlington stage helps to set the output impedance to 50 Ω . The amplifier can be operated from a 3 V supply by adding a choke inductor from the amplifier output to VPOS. Without this choke inductor, operation from a 5 V supply is also possible.

BASIC CONNECTIONS

The AD8353 RF gain block is a fixed gain amplifier with single-ended input and output ports whose impedances are nominally equal to 50 Ω over the frequency range 1 MHz to 2.7 GHz. Consequently, it can be directly inserted into a 50 Ω system with no impedance matching circuitry required. The input and output impedances are sufficiently stable vs. variations in temperature and supply voltage that no impedance matching compensation is required. A complete set of scattering parameters is available at www.analog.com.

The input pin (RFIN) is connected directly to the base of the first amplifier stage, which is internally biased to approximately 1 V; therefore, a dc blocking capacitor should be connected between the source that drives the AD8353 and the input pin, RFIN.

It is critical to supply very low inductance ground connections to the ground pins (Pin 1, Pin 4, Pin 5, and Pin 8) as well as to the backside exposed paddle. This ensures stable operation.

The AD8353 is designed to operate over a wide supply voltage range, from 2.7 V to 5.5 V. The output of the part, RFOUT, is taken directly from the collector of the output amplifier stage. This node is internally biased to approximately 2.2 V when the supply voltage is 5 V. Consequently, a dc blocking capacitor should be connected between the output pin, RFOUT, and the load that it drives. The value of this capacitor is not critical, but it should be 100 pF or larger.

When the supply voltage is 3 V, it is recommended that an external RF choke be connected between the supply voltage and the output pin, RFOUT. This increases the dc voltage applied to the collector of the output amplifier stage, which improves performance of the AD8353 to be very similar to the performance produced when 5 V is used for the supply voltage. The inductance of the RF choke should be approximately 100 nH, and care should be taken to ensure that the lowest series self-resonant frequency of this choke is well above the maximum frequency of operation for the AD8353. For lower frequency operation, use a higher value inductor.

Bypass the supply voltage input, VPOS, using a large value capacitance (approximately 0.47 μ F or larger) and a smaller, high frequency bypass capacitor (approximately 100 pF) physically located close to the VPOS pin.

The recommended connections and components are shown in Figure 40.

APPLICATIONS

The AD8353 RF gain block can be used as a general-purpose, fixed gain amplifier in a wide variety of applications, such as a driver for a transmitter power amplifier (see Figure 37). Its excellent reverse isolation also makes this amplifier suitable for use as a local oscillator buffer amplifier that would drive the local oscillator port of an upconverter or downconverter mixer (see Figure 38).

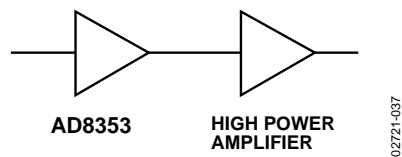


Figure 37. AD8353 as a Driver Amplifier

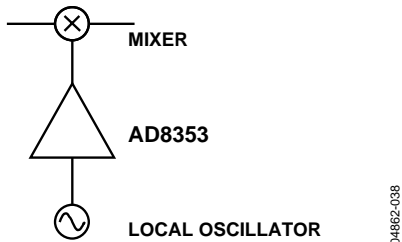


Figure 38. AD8353 as a LO Driver Amplifier

LOW FREQUENCY APPLICATIONS BELOW 100 MHz

The AD8353 RF gain block can be used below 100 MHz. To accomplish this, the series dc blocking capacitors, C1 and C2, need to be changed to a higher value that is appropriate for the desired frequency. C1 and C2 were changed to 0.1 μ F to accomplish the sweep in Figure 39.

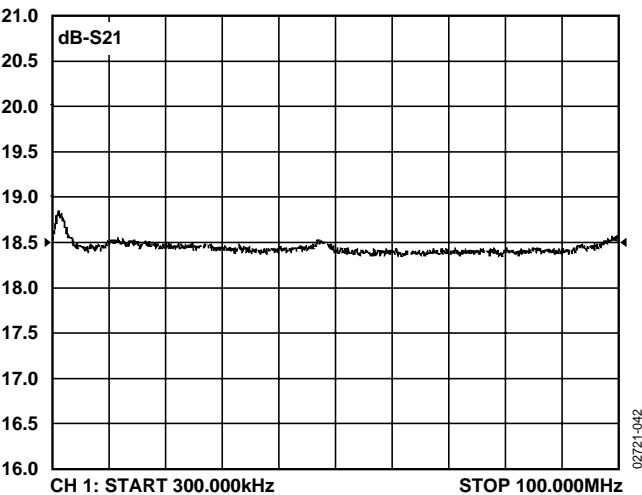


Figure 39. Low Frequency Application from 300 kHz to 100 MHz at 5 V VPOS, -12 dBm Input Power

EVALUATION BOARD

Figure 40 shows the schematic of the AD8353 evaluation board. Note that L1 is shown as an optional component that is used to obtain maximum gain only when $V_P = 3\text{ V}$. The board is powered by a single supply in the 2.7 V to 5.5 V range. The power supply is decoupled by a 0.47 μF and a 100 pF capacitor.

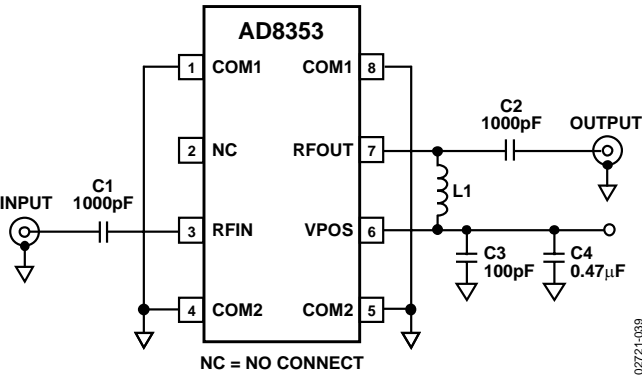


Figure 40. Evaluation Board Schematic

Table 5. Evaluation Board Configuration Options

| Component | Function | Default Value |
|-----------|--|---------------------------|
| C1, C2 | AC coupling capacitors. | 1000 pF, 0603 |
| C3 | High frequency bypass capacitor. | 100 pF 0603 |
| C4 | Low frequency bypass capacitor. | 0.47 μF , 0603 |
| L1 | Optional RF choke, used to increase current through output stage when $V_P = 3\text{ V}$. Not recommended for use when $V_P = 5\text{ V}$. | 100 nH, 0603 |

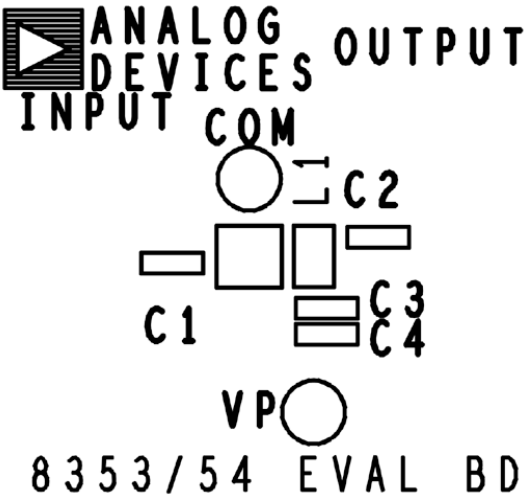


Figure 41. Silkscreen Top

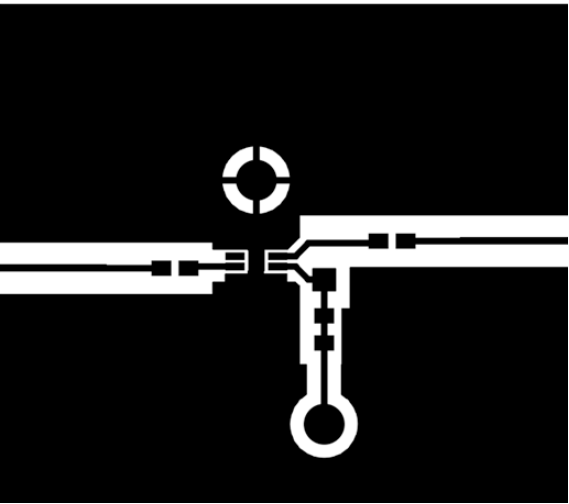


Figure 42. Component Side

AD8353

OUTLINE DIMENSIONS

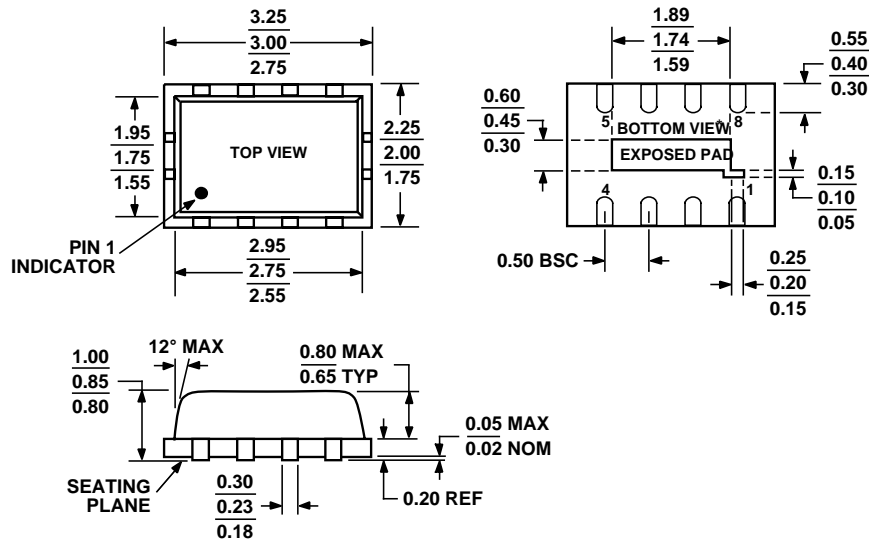


Figure 43. 8-Lead Lead Frame Chip Scale Package [LFCSP_VD]
2 mm × 3 mm Body, Very Thin, Dual Lead
CP-8-1
Dimensions shown in millimeters

ORDERING GUIDE

| Model | Temperature Range | Package Description | Package Option | Branding |
|-------------------------------|-------------------|-----------------------------------|----------------|----------|
| AD8353ACP-R2 | −40°C to +85°C | 8-Lead LFCSP_VD | CP-8-1 | JB |
| AD8353ACP-REEL7 | −40°C to +85°C | 8-Lead LFCSP_VD, 7" Tape and Reel | CP-8-1 | JB |
| AD8353ACPZ-REEL7 ¹ | −40°C to +85°C | 8-Lead LFCSP_VD, 7" Tape and Reel | CP-8-1 | OE |
| AD8353-EVAL | | Evaluation Board | | |

¹ Z = Pb-free part.