

Features

- 300mA Low Dropout Regulator with EN
- Very low I_Q: 40μA
- Wide input voltage range: 2V 6V
- Adjustable output: 1V 5V
- High PSRR: 65dB at 1kHz
- Ultra-fast start-up time: 25µs
- Stable with low ESR, 1µF ceramic output capacitor
- Excellent Load/Line Transient Response
- Low dropout: 350mV at 300mA
- Current limit protection
- Short circuit protection
- Thermal shutdown protection
- Ambient temperature range: -40°C to 85°C
- SOT25 and DFN2020-6 (Preview): Available in "Green" Molding Compound (No Br, Sb)

Description

The AP133 is a 300mA, adjustable output voltage, low dropout linear regulator. The device included pass element, error amplifier, band-gap, current limit and thermal shutdown circuitry. The device is turned on when EN pin is set to logic high level.

The characteristics of low dropout voltage and less quiescent current make it good for low power applications, for example, battery powered devices. The typical quiescent current is approximately $40\mu A$ from zero to maximum load.

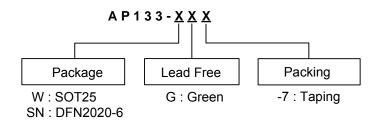
Built-in current-limit and thermal-shutdown functions prevent IC from damage in fault conditions.

The AP133 is available in SOT25 and DFN2020-6 (preview) package.

Applications

- Cellular Phones
- Smart Phones, PDAs
- MP3/MP4
- Bluetooth head set
- Low power application

Ordering Information



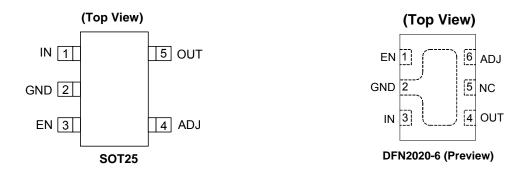
Note: 1. RoHS revision 13.2.2003. Glass and High Temperature Solder Exemptions Applied, see EU Directive Annex Notes 5 and 7.

	Device	Package Code	Packaging	7" Tape and Reel		
	Device	rackage code	(Note 2)	Quantity	Part Number Suffix	
@ ,	AP133-W	W	SOT25	3000/Tape & Reel	-7	
9	AP133-SN	SN	DFN2020-6	3000/Tape & Reel	-7	

Note: 2. Pad layout as shown on Diodes Inc. suggested pad layout document AP02001, which can be found on our website at http://www.diodes.com/datasheets/ap02001.pdf.



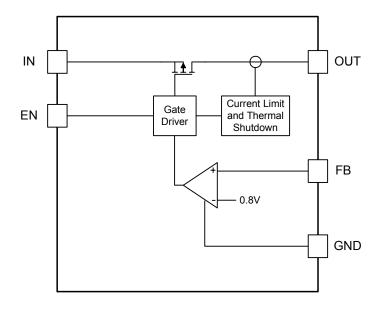
Pin Assignment



Pin Descriptions

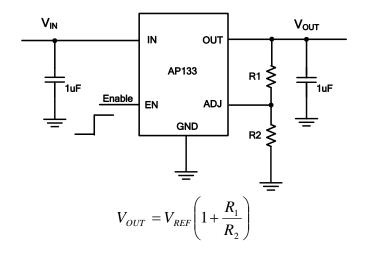
Name	Pin Number		Description			
Name	SOT25 DFN2020-6		Description			
IN	1	3	Voltage input pin. Bypass to ground through at least 0.1µF capacitor			
GND	2	2	Ground			
EN	3	1	Enable input, active high			
ADJ	4	6	Output feedback pin			
NC		5	No connection			
OUT	5	4	Voltage output pin. Bypass to ground through 1µF ceramic capacitor			

Functional Block Diagram





Typical Application Circuit



Absolute Maximum Ratings

Symbol	Parameter		Ratings	Units
V_{IN}	Input Voltage		7	V
	OUT, ADJ, EN Voltage		$V_{IN} + 0.3$	V
	Continuous Load Current		Internal Limited	
T_OP	Operating Junction Tempera	ature Range	-40 ~ 125	°C
T_{ST}	Storage Temperature Range		-65 ~150	°C
P _D	Power Dissipation (Note 3)	SOT25	610	mW
i D		DFN2020-6	880	mW
T_J	Maximum Junction Temperature		150	°C

Note: 3. Ratings apply to ambient temperature at 25°C.

Recommended Operating Conditions

Symbol	Parameter	Min	Max	Unit
V _{IN}	Input voltage	2	6	V
I _{OUT}	Output Current	0	300	mA
T _A	Operating Ambient Temperature	-40	85	°C



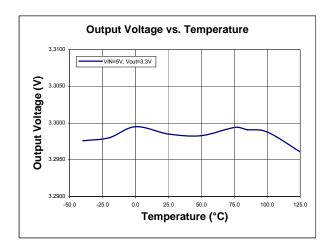
Electrical Characteristics

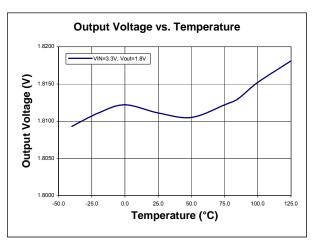
 $(T_A = 25^{\circ}C, V_{IN} = V_{OUT} + 1V, C_{IN} = 1 \mu F, C_{OUT} = 1 \mu F, V_{EN} = 2V, unless otherwise stated)$

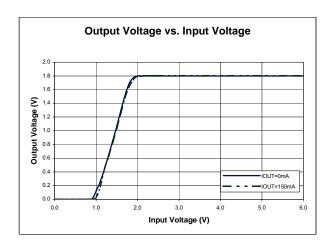
Symbol	Parameter		Test Conditions	Min	Тур	Max	Unit
IQ	Input Quiescent Current		I _{OUT} = 0~300mA	_	40	60	μΑ
I _{SHDN}	Input Shutdown Current		$V_{EN} = 0V$, $I_{OUT} = 0$	_		1	μА
I _{LEAK}	Input Leakage Cu	rrent	V_{EN} = 0V, OUT grounded	_		1	μΑ
$V_{Dropout}$	Dropout Voltage		$V_{OUT} \ge 1.5V$, $I_{OUT} = 300$ mA		350	450	mV
V_{REF}	ADJ reference vol	ltage	I _{OUT} = 0		8.0		V
I_{ADJ}	ADJ leakage			_		1	μΑ
V_{OUT}	Output Voltage Accuracy			-2		2	%
	Line Regulation		$V_{IN} = V_{OUT} + 1V$ to 5.5V, $I_{OUT} = 1$ mA		0.05		%/V
$\Delta V_{OUT} / V_{OUT}$	Load Regulation		I _{OUT} from 1mA to 300mA	-1		1	%
t _{ST}	Start-up Time		$V_{EN} = 0V \text{ to } 2.0V, I_{OUT} = 300\text{mA}$		25		μS
PSRR	PSRR		1kHz, I _{OUT} = 0mA		65		dB
I _{SHORT}	Short-circuit Current		$V_{IN} = 5.0V, V_{OUT} < 0.2V$		120		mA
I _{LIMIT}	Current limit		V_{OUT} = 3V, R_{OUT} = 3 Ω	400	600		mA
V_{IL}	EN Input Logic Low Voltage					0.4	V
V_{IH}	EN Input Logic High Voltage			1.4			V
I _{EN}	EN Input leakage		$V_{EN} = 0V \text{ or } 5.5V$	-1		1	μΑ
T _{SHDN}	Thermal shutdown threshold				145		°C
T _{HYS}	Thermal shutdown hysteresis				20		°C
	Thermal	SOT25			197		
$ heta_{JA}$	Resistance Junction-to- Ambient	DFN2020-6	Device mounted on FR-4 substrate, 1"x1" 1oz copper,		136		°C/W
	Thermal	SOT25	single-sided PCB board		45		°C/W
$ heta_{ extsf{JC}}$	Resistance Junction-to-Case	DFN2020-6			102		

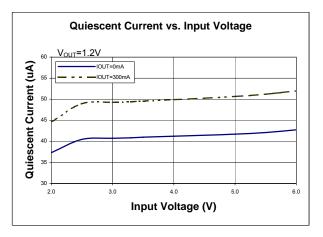


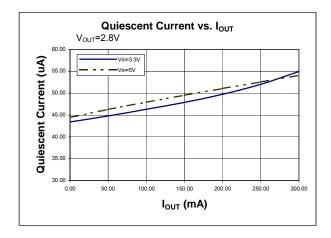
Typical Performance Characteristics

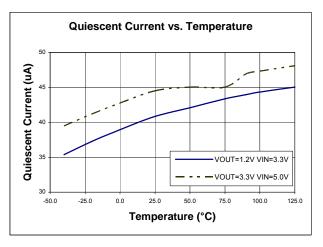




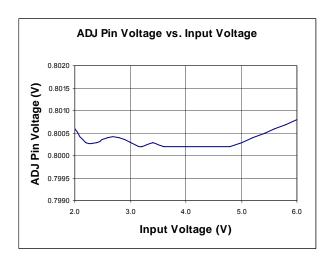


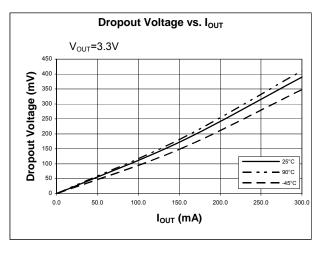


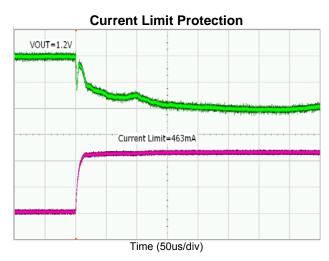


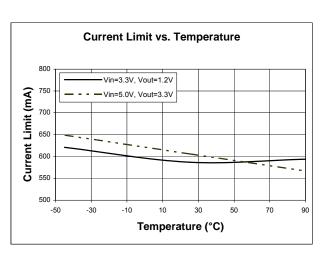


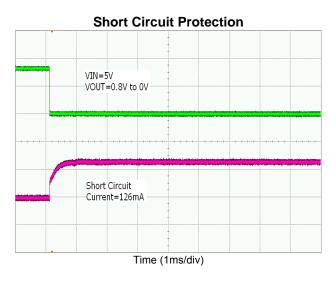


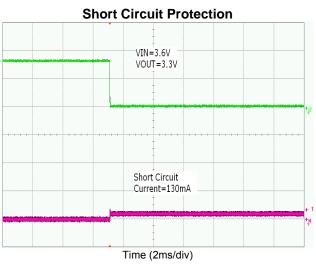




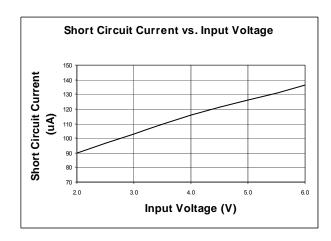


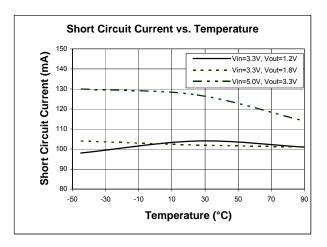


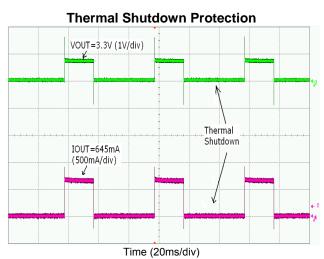


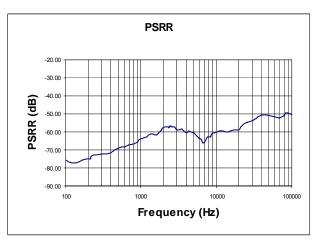


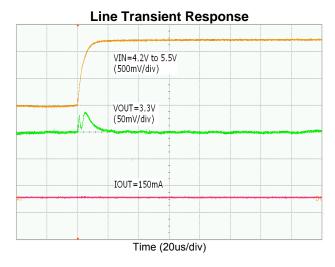


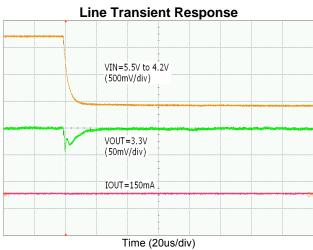




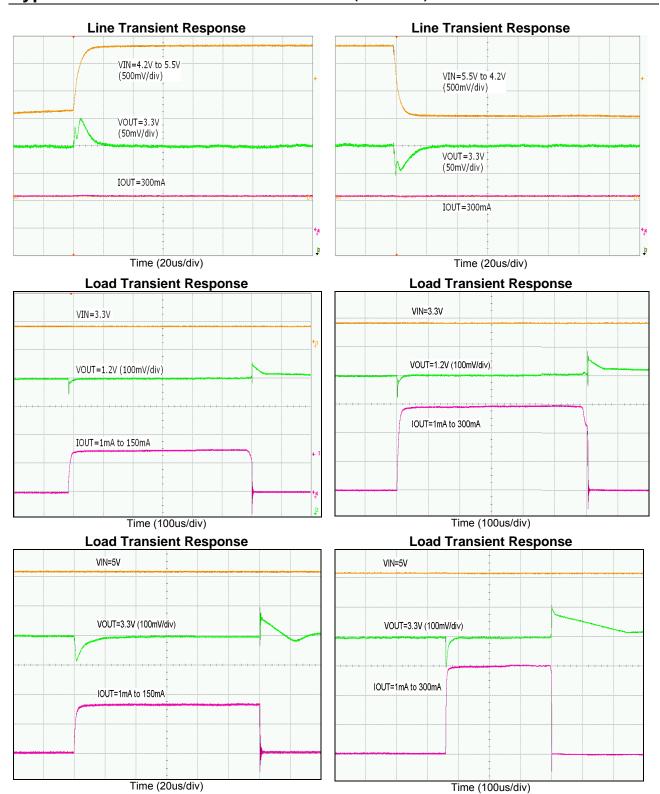




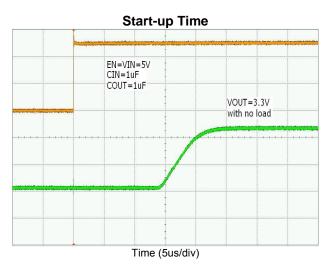


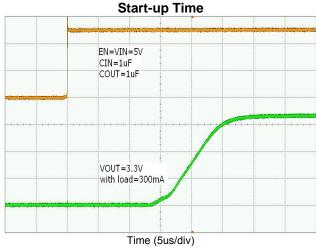


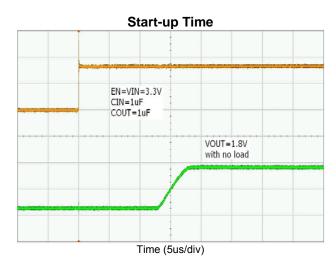


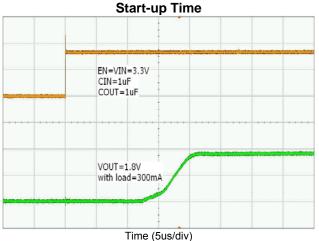














Application Note

Input Capacitor

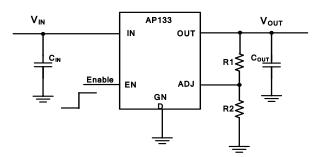
 $A1\mu F$ ceramic capacitor is recommended to connect between V_{IN} and GND pins to decouple input power supply glitch and noise. The amount of the capacitance may be increased without limit. This input capacitor must be located as close as possible to the device to assure input stability and less noise. For PCB layout, a wide copper trace is required for both V_{IN} and GND. A lower ESR capacitor allows the use of less capacitance, while higher ESR type require more capacitance.

Output Capacitor

The output capacitor is required to stabilize and help transient response for LDO. The AP133 is stable with very small ceramic output capacitors. The recommended capacitance is from $1\mu F$ to $4.7\mu F$, Equivalent Series Resistance (ESR) is from $10m\Omega$ to $100m\Omega$, and temperature characteristics is X7R or X5R. Higher capacitance values help to improve load/line transient response. The output capacitance may be increased to keep low undershoot/overshoot. Place output capacitor as close as possible to OUT and GND pins, and keep the leads as short as possible.

Adjustable Operation

The AP133 provides output voltage from 1.0V to 5.0V through external resistor divider as shown below.



The output voltage is calculated by:

$$V_{OUT} = V_{REF} \left(1 + \frac{R_1}{R_2} \right)$$

Where V_{REF}=0.8V (the internal reference voltage)

Rearranging the equation will give the following that is used for adjusting the output to a particular voltage:

$$R_1 = R_2 \left(\frac{V_{OUT}}{V_{REF}} - 1 \right)$$

To maintain the stability of the internal reference voltage, R_2 need to be kept smaller than 250k Ω .

No Load Stability

Other than external resistor divider, no minimum load is required to keep the device stable. The device will remain stable and regulated in no load condition.

ON/OFF Input Operation

The AP133 is turned on by setting the EN pin high, and is turned off by pulling it low. If this feature is not used, the EN pin should be tied to IN pin to keep the regulator output on at all time. To ensure proper operation, the signal source used to drive the EN pin must be able to swing above and below the specified turn-on/off voltage thresholds listed in the Electrical Characteristics section under $V_{\rm IL}$ and $V_{\rm IH}$.

Current Limit Protection

When output current at OUT pin is higher than current limit threshold, the current limit protection will be triggered and clamp the output current to approximately 600mA to prevent over-current and to protect the regulator from damage due to overheating.

Short Circuit Protection

When OUT pin is short-circuit to GND or OUT pin voltage is less than 200mV, short circuit protection will be triggered and clamp the output current to approximately 120mA. This feature protects the regulator from over-current and damage due to overheating.

Thermal Shutdown Protection

Thermal protection disables the output when the junction temperature rises to approximately +145°C, allowing the device to cool down. When the junction temperature reduces to approximately +125°C the output circuitry is enabled again. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits the heat dissipation of the regulator, protecting it from damage due to overheating.

Ultra Fast Start-up

After enabled, the AP133 is able to provide full power in as little as tens of microseconds, typically 25µs, without sacrificing low ground current. This feature will help load circuitry move in and out of standby mode in real time, eventually extend battery life for mobile phones and other portable devices.

Fast Transient Response

Fast transient response LDOs can also extend battery life. TDMA-based cell phone protocols such as Global System for Mobile Communications (GSM) have a transmit/receive duty factor of only 12.5 percent, enabling power savings by putting much of the baseband circuitry into standby mode in between transmit cycles. In baseband circuits, the load often transitions virtually instantaneously from 100µA to 100mA. To meet this load requirement, the LDO must react very quickly without a large voltage droop or overshoot — a requirement that cannot be met with conventional, general-purpose LDOs.

The AP133's fast transient response from 0 to 300mA provides stable voltage supply for fast DSP and GSM chipset with fast changing load.

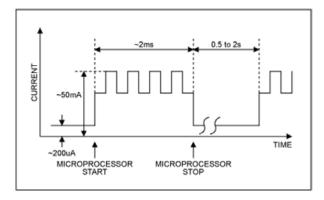
Small Overshoot and Undershoot

The AP133 has small and controlled overshoot and undershoot in load and line transitions. This helps to protect supplied circuit from damage and operation error caused by glitches. This feather also permits the usage of small value output decoupling capacitor with AP133.



Low Quiescent Current

Cellular phone baseband internal digital circuits typically operate from 1.8V to 2.6V. When the Li+ battery voltage falls to 3.2-3.3V, most phones shut off, giving at least 500-600mV of headroom for the baseband digital LDO, so dropout is not critical. Output noise and the PSRR are not critical specs for the digital circuits. Nonetheless, this supply requires low quiescent current at light loads because this LDO stays on at all times. Figure below shows how the digital supply current of a representative GSM chipset core varies as a function of time.



In the standby mode, the microprocessor consumes only around 200 μ A. Since the phone stays in standby for the longest percentage of time, using a 40 μ A quiescent current LDO, instead of 140 μ A, saves 100 μ A and extends the standby time by 340 μ A/240 μ A, or 1.417 times.

The baseband internal analog circuit is typically 2.4V-3.0V, and it requires 200-600mV dropout. This LDO is on all the time, so it requires low quiescent current as well. The cellular phone real-time clock LDO needs a very low quiescent current, since this LDO is on all the time even though the handset is powered off.

The AP133, consuming only around $40\mu A$ for all input range and output loading, provides great power saving in portable and low power applications.

Wide Output Range

The AP133, with a wide output range of 1.0V to 5.0V, provides a versatile LDO solution for many portable applications.

High PSRR

The RF circuit consists of receive and transmit sections, which typically require 2.6V-3.0V supply voltage. The RF circuits such as LNA (low-noise amplifier), up/down-converter, mixer, PLL, VCO, and IF stage, require low noise and high PSRR LDOs. The temperature-compensated crystal oscillator circuit requires very high PSRR at RF power amplifier burst frequency. For instance, minimum 65dB PSRR at 217Hz is recommended for the GSM handsets.

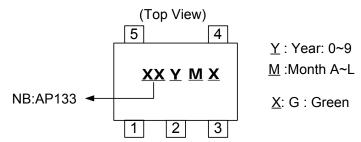
In order to provide good audio quality, the audio power supply for hand-free, game, MP3, and multimedia applications in cellular phones, require low-noise and high PSRR at audio frequency range (20Hz-20kHz).

The AP133, with PSRR of 70dB at 1kHz in best case, is suitable for some of these applications that require high PSRR.



Marking Information

(1) SOT25



(2) DFN2020-6 (Preview)

(Top View)

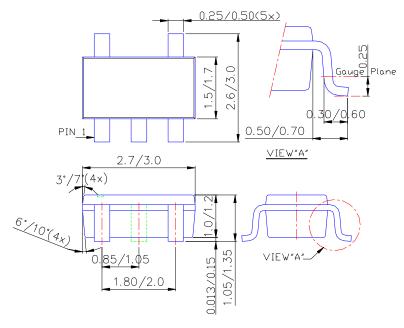
X: Internal code A~Z: Green



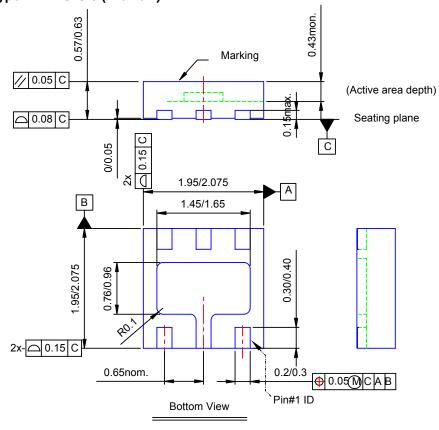
Package Information

(All Dimensions in mm)

(1) Package Type: SOT25



(2) Package Type: DFN2020-6 (Preview)





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