13-BIT 210 MSPS ANALOG-TO-DIGITAL CONVERTER

FEATURES

- **13-Bit Resolution**
- **210 MSPS Sample Rate**
- **SNR = 69 dBc at 100-MHz IF and 210 MSPS**
- **SFDR = 76 dBc at 100-MHz IF and 210 MSPS**
- **SNR = 68.1 dBc at 230-MHz IF and 210 MSPS**
- **SFDR = 74 dBc at 230-MHz IF and 210 MSPS**
- **2.2 V$_{PP}$ Differential Input Voltage**
- **Fully Buffered Analog Inputs**
- **5 V Analog Supply Voltage**
- **LVDS Compatible Outputs**
- **Total Power Dissipation: 2 W**
- **Offset Binary Output Format**
- **TQFP-80 PowerPAD™ Package**
- **Pin Compatible with the ADS5444**
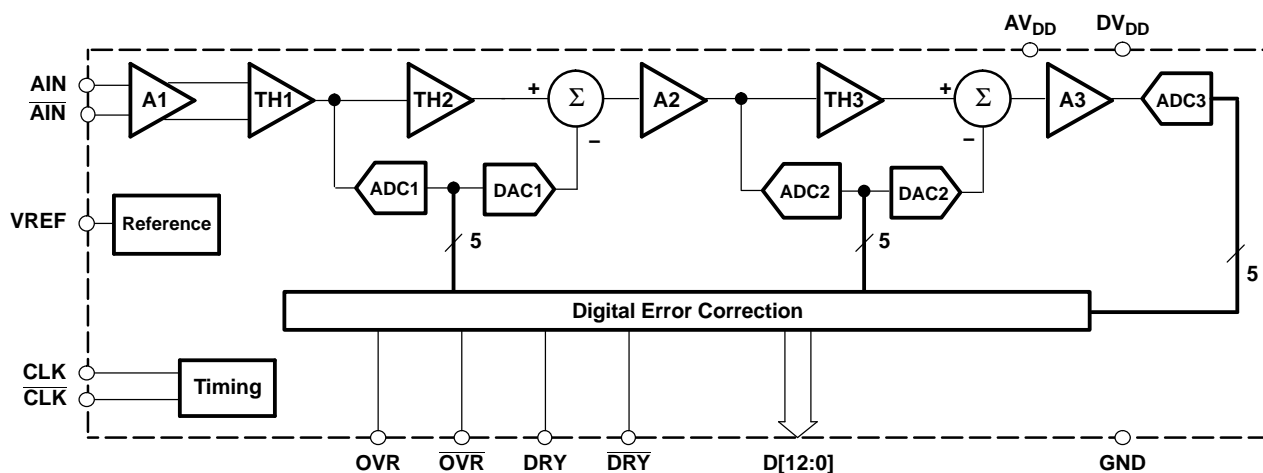- **Industrial Temperature Range = −40°C to 85°C**

APPLICATIONS

- **Test and Measurement**
- **Software-Defined Radio**
- **Multi-channel Basestation Receivers**
- **Basestation Tx Digital Predistortion**
- **Communications Instrumentation**

RELATED PRODUCTS

- **ADS5424 - 14-bit, 105 MSPS ADC**
- **ADS5423 - 14-bit, 80 MSPS ADC**
- **ADS5444 - 13-bit, 250 MSPS ADC**

DESCRIPTION

The ADS5440 is a 13-bit 210 MSPS analog-to-digital converter (ADC) that operates from a 5 V supply, while providing LVDS-compatible digital outputs from a 3.3 V supply. The ADS5440 input buffer isolates the internal switching of the onboard track and hold (T&H) from disturbing the signal source. An internal reference generator is also provided to further simplify the system design. The ADS5440 has outstanding low noise and linearity over input frequency.

B0061-01

The ADS5440 is available in an 80-pin TQFP PowerPAD™ package. The ADS5440 is built on a state of the art Texas Instruments complementary bipolar process (BiCom3X) and is specified over the full industrial temperature range (−40°C to 85°C).

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PACKAGING/ORDERING INFORMATION[1]

Product	Package-Lead	Package Designator [1]	Specified Temperature Range	Package Marking	Ordering Number	Transport Media, Quantity
ADS5440	HTQFP-80[2] PowerPAD	PFP	−40°C to 85°C	ADS5440IPFP	ADS5440IPFP	Tray, 96
					ADS5440IPFPR	Tape and Reel, 1000

(1) For the most current product and ordering information, see the Package Option Addendum located at the end of this data sheet.
(2) Thermal pad size: 7,5 mm x 7,5 mm (typ)

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)[1]

		VALUE / UNIT
Supply voltage	AV_{DD} to GND	6 V
	DRV_{DD} to GND	5 V
Analog input to GND		−0.3 V to AV_{DD}+0.3 V
Clock input to GND		−0.3 V to AV_{DD}+0.3 V
CLK to $\overline{CLK}$		±2.5 V
Digital data output to GND		−0.3 V to DRV_{DD}+0.3 V
Operating temperature range		−40°C to 85°C
Maximum junction temperature		150°C
Storage temperature range		−65°C to 150°C
ESD Human Body Model (HBM)		2.5 kV

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only and functional operation of the device at these or any other conditions beyond those specified is not implied.

THERMAL CHARACTERISTICS[1]

PARAMETER	TEST CONDITIONS	TYP	UNIT
θ_{JA}	Soldered slug, no airflow	21.7	°C/W
	Soldered slug, 250-LFPM airflow	15.4	°C/W
	Unsoldered slug, no airflow	50	°C/W
	Unsoldered slug, 250-LFPM airflow	43.4	°C/W
θ_{JC}	Bottom of package (heatslug)	2.99	°C/W

(1) Using 36 thermal vias (6 x 6 array). See the Application Section.

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RECOMMENDED OPERATING CONDITIONS

		MIN	NOM	MAX	UNIT
SUPPLIES					
AV_{DD}	Analog supply voltage	4.75	5	5.25	V
DRV_{DD}	Output driver supply voltage	3	3.3	3.6	V
ANALOG INPUT					
	Differential input range		2.2		V_{PP}
V_{CM}	Input common mode		2.4		V
CLOCK INPUT					
$1/t_C$	ADCLK input sample rate (sine wave)	10		210	MSPS
	Clock amplitude, differential sine wave		3		Vpp
	Clock duty cycle		50%		
T_A	Open free air temperature	−40		85	°C

ELECTRICAL CHARACTERISTICS

MIN, TYP, and MAX values at T_A = 25°C, full temperature range is T_{MIN} = −40°C to T_{MAX} = 85°C, sampling rate = 210 MSPS, 50% clock duty cycle, AV_{DD} = 5 V, DRV_{DD} = 3.3 V, −1 dBFS differential input, and 3 V_{PP} differential clock, unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Resolution			13		Bits
ANALOG INPUTS						
	Differential input range			2.2		V_{pp}
	Differential input resistance (DC)			1		kΩ
	Differential input capacitance			1.5		pF
	Analog input bandwidth			800		MHz
INTERNAL REFERENCE VOLTAGE						
VREF	Reference voltage			2.4		V
DYNAMIC ACCURACY						
	No missing codes			Assured		
DNL	Differential linearity error	f_{IN} = 10 MHz	-1	±0.4	1	LSB
INL	Integral linearity error	f_{IN} = 10 MHz	-2.2	±0.9	2.2	LSB
	Offset error		-11		11	mV
	Offset temperature coefficient			0.0005		mV/°C
	Gain error		−5		5	%FS
	Gain temperature coefficient			-0.02		Δ%/°C
	PSRR	100-MHz supply frequency		1		mV/V
POWER SUPPLY						
I_{AVDD}	Analog supply current			340	390	mA
I_{DRVDD}	Output buffer supply current	V_{IN} = full scale, f_{IN} = 100 MHz, F_S = 210 MSPS		80	100	mA
	Power dissipation			2	2.28	W

ELECTRICAL CHARACTERISTICS (continued)

MIN, TYP, and MAX values at T_A = 25°C, full temperature range is T_{MIN} = −40°C to T_{MAX} = 85°C, sampling rate = 210 MSPS, 50% clock duty cycle, AV_{DD} = 5 V, DRV_{DD} = 3.3 V, −1 dBFS differential input, and 3 V_{PP} differential clock, unless otherwise noted

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
DYNAMIC AC CHARACTERISTICS							
SNR	Signal-to-noise ratio	f_{IN} = 10 MHz			69.8		dBc
		f_{IN} = 70 MHz			69.2		
		f_{IN} = 100 MHz	T_A = 25°C	67.5	69		
			Full Temp Range	67	69		
		f_{IN} = 170 MHz			68.3		
		f_{IN} = 230 MHz			68		
		f_{IN} = 300 MHz			66.9		
		f_{IN} = 400 MHz			65		
SFDR	Spurious free dynamic range	f_{IN} = 10 MHz			84		dBc
		f_{IN} = 70 MHz			77		
		f_{IN} = 100 MHz	T_A = 25°C	70	76		
			Full Temp Range	68	76		
		f_{IN} = 170 MHz			74		
		f_{IN} = 230 MHz			74		
		f_{IN} = 300 MHz			69		
		f_{IN} = 400 MHz			64		
HD2	Second harmonic	f_{IN} = 10 MHz			96		dBc
		f_{IN} = 70 MHz			83		
		f_{IN} = 100 MHz			87		
		f_{IN} = 170 MHz			76		
		f_{IN} = 230 MHz			76		
		f_{IN} = 300 MHz			69		
		f_{IN} = 400 MHz			64		
HD3	Third harmonic	f_{IN} = 10 MHz			84		dBc
		f_{IN} = 70 MHz			77		
		f_{IN} = 100 MHz			76		
		f_{IN} = 170 MHz			74		
		f_{IN} = 230 MHz			74		
		f_{IN} = 300 MHz			73		
		f_{IN} = 400 MHz			70		
	Worst other harmonic/spur (other than HD2 and HD3)	f_{IN} = 10 MHz			92		dBc
		f_{IN} = 70 MHz			92		
		f_{IN} = 100 MHz			87		
		f_{IN} = 170 MHz			87		
		f_{IN} = 230 MHz			83		
		f_{IN} = 300 MHz			83		
		f_{IN} = 400 MHz			80		

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ELECTRICAL CHARACTERISTICS (continued)

MIN, TYP, and MAX values at T_A = 25°C, full temperature range is T_{MIN} = −40°C to T_{MAX} = 85°C, sampling rate = 210 MSPS, 50% clock duty cycle, AV_{DD} = 5 V, DRV_{DD} = 3.3 V, −1 dBFS differential input, and 3 V_{PP} differential clock, unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	SINAD	f_{IN} = 10 MHz		69.6		dBc
		f_{IN} = 70 MHz		68.8		
		f_{IN} = 100 MHz		68		
		f_{IN} = 170 MHz		66.8		
		f_{IN} = 230 MHz		66		
		f_{IN} = 300 MHz		64		
		f_{IN} = 400 MHz		60		
ENOB	Effective number of bits	f_{IN} = 10 MHz		11.4		Bits
	RMS idle channel noise	Inputs tied to common-mode		0.4		LSB
DIGITAL CHARACTERISTICS − LVDS DIGITAL OUTPUTS						
	Differential output voltage		0.247		0.452	V
	Output offset voltage		1.125	1.25	1.375	V

TIMING CHARACTERISTICS
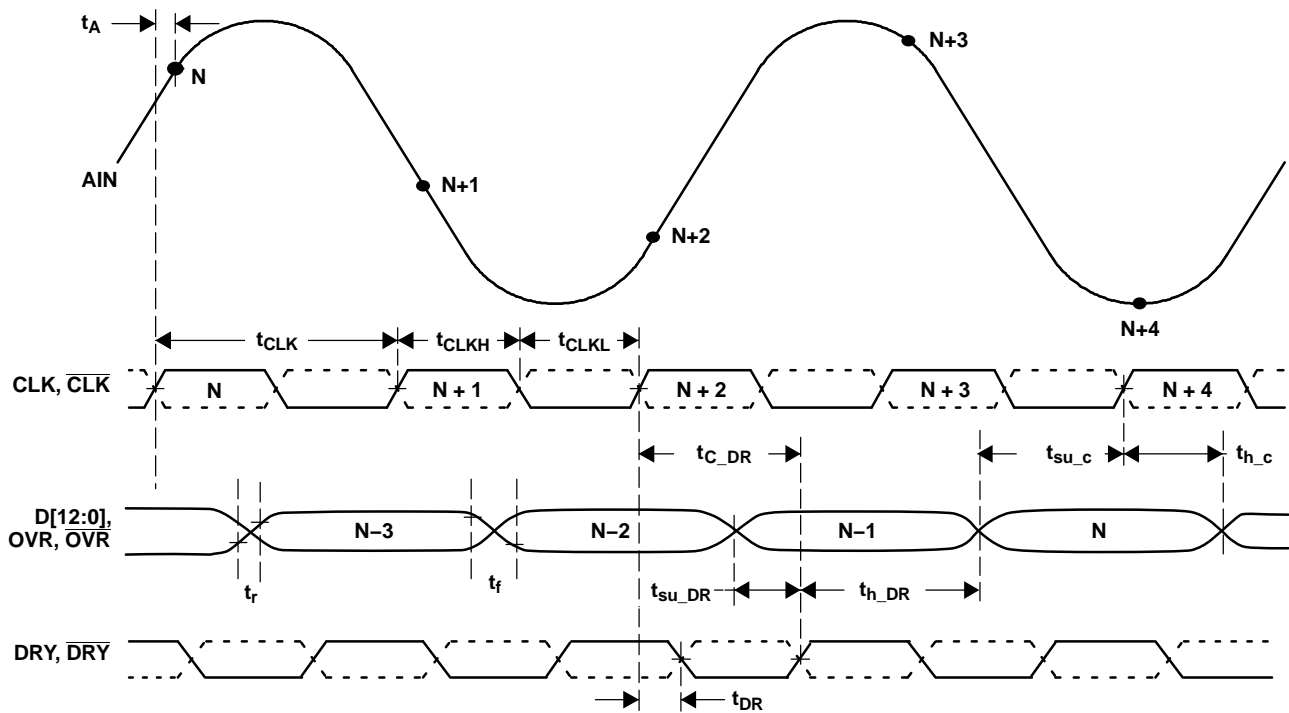
Figure 1. Timing Diagram

TIMING CHARACTERISTICS

Min, Typ, Max over full temperature range, 50% clock duty cycle, sampling rate = 210 MSPS, AV_{DD} = 5 V, DRV_{DD} = 3.3 V

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_A	Aperature delay			500		ps
t_J	Clock slope independent aperture uncertainty (jitter)			150		fs RMS
	Latency			4		cycles
Clock Input						
t_{CLK}	Clock period			4.76		ns
t_{CLKH}	Clock pulsewidth high			2.38		ns
t_{CLKL}	Clock pulsewidth low			2.38		ns
Clock to DataReady (DRY)						
t_{DR}	Clock rising to DataReady falling			1.1		ns
t_{C_DR}	Clock rising to DataReady rising	Clock duty cycle = 50% [1]	3.1	3.5	3.9	ns
Clock to DATA, OVR [2]						
t_r	Data rise time (20% to 80%)			0.6		ns
t_f	Data fall time (80% to 20%)			0.6		ns
t_{su_c}	Data valid to clock (setup time)			3.5		ns
t_{h_c}	Clock to invalid Data (hold time)			0.2		ns
DataReady (DRY)/DATA, OVR [2]						
$t_{su(DR)}$	Data valid to DRY		2.1	2.4		ns
$t_{h(DR)}$	DRY to invalid Data		0.9	1.3		ns

(1) $t_{C\_DR} = t_{DR} + t_{CLKH}$ for clock duty cycles other than 50%.
(2) Data is updated with clock falling edge or DRY rising edge.

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DEVICE INFORMATION

PFP PACKAGE
(TOP VIEW)
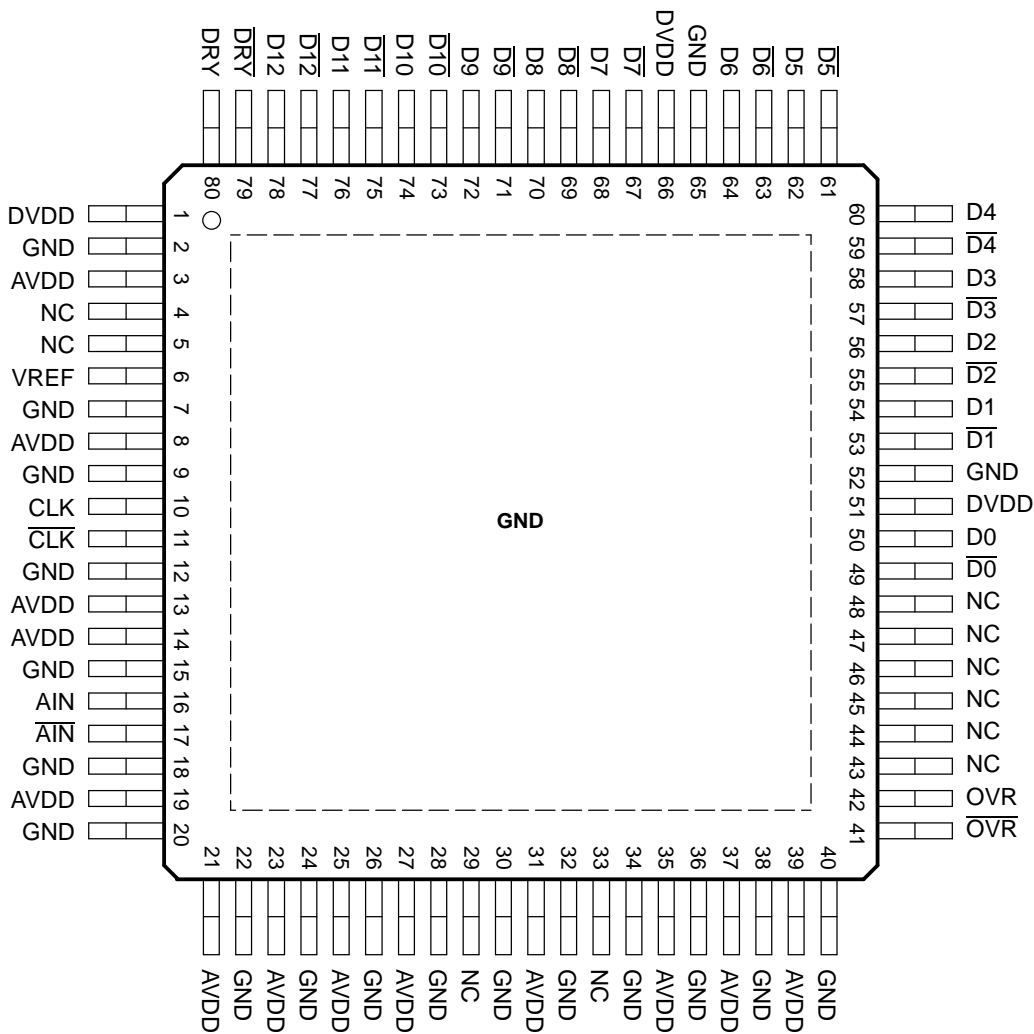
P0027-01

TERMINAL FUNCTIONS

TERMINAL		DESCRIPTION
NAME	NO.	
AVDD	3, 8, 13, 14, 19, 21, 23, 25, 27, 31, 35, 37, 39	Analog power supply
DVDD	1, 51, 66	Output driver power supply
GND	2, 7, 9, 12, 15, 18, 20, 22, 24, 26, 28, 30, 32, 34, 36, 38, 40, 52, 65	Ground
VREF	6	Reference voltage
CLK	10	Differential input clock (positive). Conversion initiated on rising edge
$\overline{CLK}$	11	Differential input clock (negative)
AIN	16	Differential input signal (positive)
$\overline{AIN}$	17	Differential input signal (negative)

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DEVICE INFORMATION (continued)

TERMINAL FUNCTIONS (continued)

TERMINAL		DESCRIPTION
NAME	NO.	
OVR, $\overline{\text{OVR}}$	42, 41	Over range indicator LVDS output. A logic high signals an analog input in excess of the full-scale range.
D0, $\overline{\text{D0}}$	50, 49	LVDS digital output pair, least-significant bit (LSB)
D1−D6, $\overline{\text{D1}}$−$\overline{\text{D6}}$	53−64	LVDS digital output pairs
D7−D11, $\overline{\text{D7}}$−$\overline{\text{D11}}$	67−76	LVDS digital output pairs
D12, $\overline{\text{D12}}$	78, 77	LVDS digital output pair, most-significant bit (MSB)
DRY, $\overline{\text{DRY}}$	80, 79	Data ready LVDS output pair
NC	4, 5, 29, 33, 43−48	No connect

DEFINITION OF SPECIFICATIONS

Analog Bandwidth The analog input frequency at which the power of the fundamental is reduced by 3 dB with respect to the low frequency value.

Aperture Delay The delay in time between the rising edge of the input sampling clock and the actual time at which the sampling occurs.

Aperture Uncertainty (Jitter) The sample-to-sample variation in aperture delay.

Clock Pulse Width/Duty Cycle The duty cycle of a clock signal is the ratio of the time the clock signal remains at a logic high (clock pulse width) to the period of the clock signal. Duty cycle is typically expressed as a percentage. A perfect differential sine wave clock results in a 50% duty cycle.

Maximum Conversion Rate The maximum sampling rate at which certified operation is given. All parametric testing is performed at this sampling rate unless otherwise noted.

Minimum Conversion Rate The minimum sampling rate at which the ADC functions.

Differential Nonlinearity (DNL) An ideal ADC exhibits code transitions at analog input values spaced exactly 1 LSB apart. The DNL is the deviation of any single step from this ideal value, measured in units of LSB.

Integral Nonlinearity (INL) The INL is the deviation of the ADCs transfer function from a best fit line determined by a least squares curve fit of that transfer function. The INL at each analog input value is the difference between the actual transfer function and this best fit line, measured in units of LSB.

Gain Error The gain error is the deviation of the ADCs actual input full-scale range from its ideal value. The gain error is given as a percentage of the ideal input full-scale range.

Offset Error Offset error is the deviation of output code from mid-code when both inputs are tied to common-mode.

Temperature Drift Temperature drift (with respect to gain error and offset error) specifies the change from the value at the nominal temperature to the value at T_{MIN} or T_{MAX}. It is computed as the maximum variation the parameters over the whole temperature range divided by T_{MIN}− T_{MAX}.

Signal-to-Noise Ratio (SNR) SNR is the ratio of the power of the fundamental (P_S) to the noise floor power (P_N), excluding the power at dc and the first five harmonics.

$$\text{SNR} = 10\log_{10} \frac{P_S}{P_N}$$

(1)

SNR is either given in units of dBc (dB to carrier) when the absolute power of the fundamental is used as the reference, or dBFS (dB to full scale) when the power of the fundamental is extrapolated to the converter's full-scale range.

Signal-to-Noise and Distortion (SINAD) SINAD is the ratio of the power of the fundamental (P_S) to the power of all the other spectral components including noise (P_N) and distortion (P_D), but excluding dc.

$$\text{SINAD} = 10\log_{10} \frac{P_S}{P_N + P_D}$$

(2)

SINAD is either given in units of dBc (dB to carrier) when the absolute power of the fundamental is used as the reference, or dBFS (dB to full scale) when the power of the fundamental is extrapolated to the converter's full-scale range.

DEFINITION OF SPECIFICATIONS (continued)

Effective Resolution Bandwidth The highest input frequency where the SNR (dB) is dropped by 3 dB for a full-scale input amplitude.

Total Harmonic Distortion (THD) THD is the ratio of the power of the fundamental (P_S) to the power of the first five harmonics (P_D).
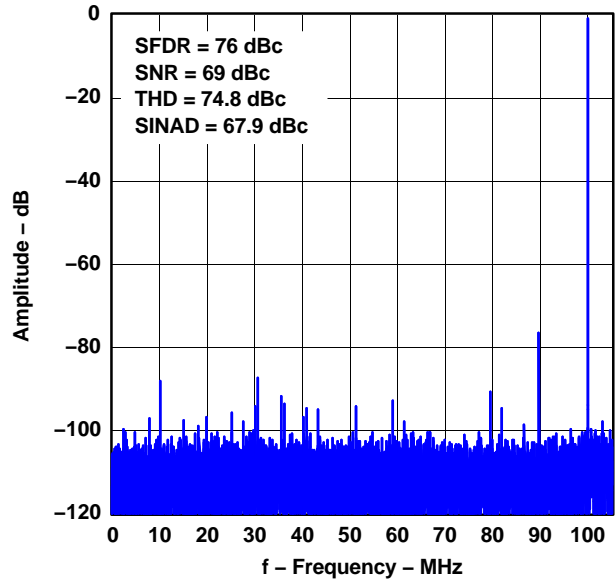
$$THD = 10\log_{10}\frac{P_S}{P_D}$$

(3)

THD is typically given in units of dBc (dB to carrier).

Two-Tone Intermodulation Distortion IMD3 is the ratio of the power of the fundamental (at frequencies f_1, f_2) to the power of the worst spectral component at either frequency $2f_1- f_2$ or $2f_2- f_1$). IMD3 is either given in units of dBc (dB to carrier) when the absolute power of the fundamental is used as the reference or dBFS (dB to full scale) when the power of the fundamental is extrapolated to the converter's full-scale range.

TYPICAL CHARACTERISTICS

**Spectral Performance
(FFT For 10-MHz Input Signal)**

SFDR = 83 dBc
SNR = 69.9 dBc
THD = 81.4 dBc
SINAD = 69.6 dBc
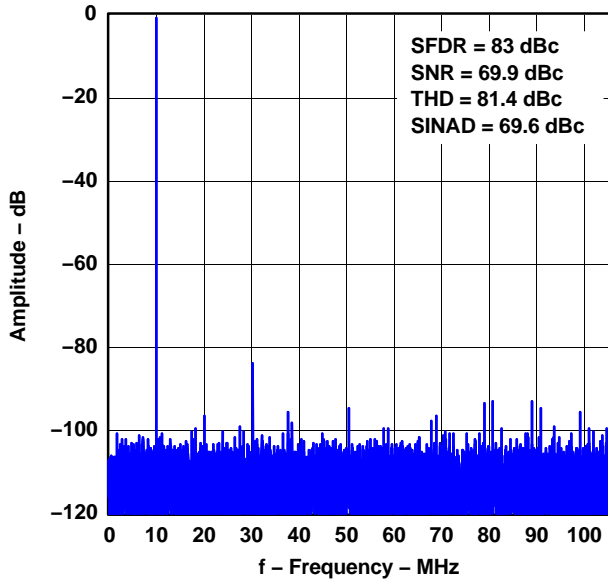
Figure 2.

**Spectral Performance
(FFT For 100-MHz Input Signal)**

SFDR = 76 dBc
SNR = 69 dBc
THD = 74.8 dBc
SINAD = 67.9 dBc

Figure 3.

**Spectral Performance
(FFT For 170-MHz Input Signal)**

SFDR = 74 dBc
SNR = 68.3 dBc
THD = 71.5 dBc
SINAD = 66.8 dBc
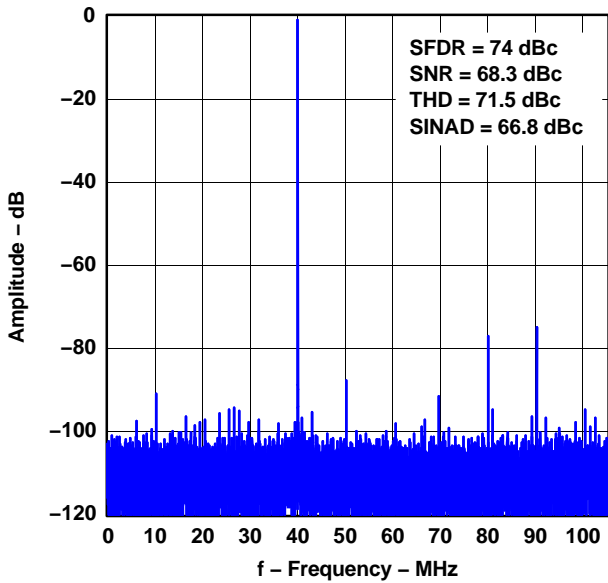
Figure 4.

**Spectral Performance
(FFT For 230-MHz Input Signal)**

SFDR = 74 dBc
SNR = 67.9 dBc
THD = 70.3 dBc
SINAD = 65.9 dBc
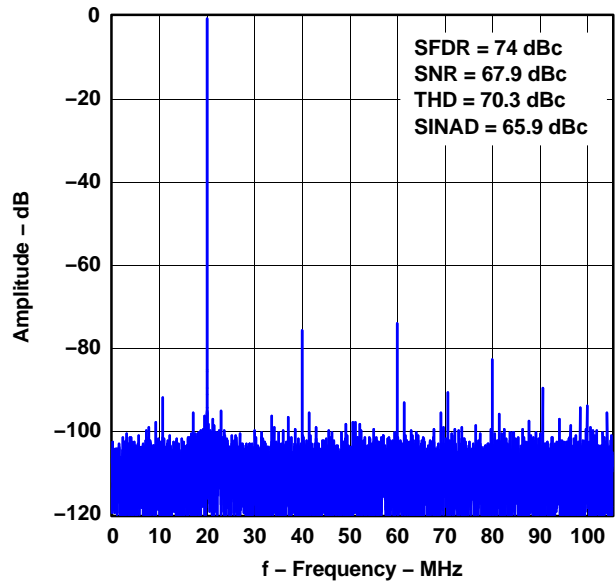
Figure 5.

TYPICAL CHARACTERISTICS (continued)

**Spectral Performance
(FFT For 300-MHz Input Signal)**

SFDR = 69 dBc
SNR = 66.9 dBc
THD = 67.1 dBc
SINAD = 63.9 dBc

Amplitude – dB

f – Frequency – MHz

G005

Figure 6.

**Spectral Performance
(FFT For 400-MHz Input Signal)**

SFDR = 63.3 dBc
SNR = 64.9 dBc
THD = 62.1 dBc
SINAD = 60.1 dBc

Amplitude – dB

f – Frequency – MHz

G006

Figure 7.

**Two-Tone Intermodulation Distortion
(FFT For 51.5-MHz and 52.5 MHz Input Signals)**

F_{IN1} = 51.5 MHz, −7 dBFS
F_{IN2} = 52.5 MHz, −7 dBFS
IMD3 = 89.3 dBFS

Amplitude – dB

f – Frequency – MHz

G007

Figure 8.

**Two-Tone Intermodulation Distortion
(FFT For 51.5-MHz and 52.5 MHz Input Signals)**

F_{IN1} = 51.5 MHz, −16 dBFS
F_{IN2} = 52.5 MHz, −16 dBFS
IMD3 = 97.4 dBFS

Amplitude – dB

f – Frequency – MHz

G008

Figure 9.

TYPICAL CHARACTERISTICS (continued)

**Two-Tone Intermodulation Distortion
(FFT For 151-MHz and 152 MHz Input Signals)**

F_{IN1} = 151 MHz, −7 dBFS
F_{IN2} = 152 MHz, −7 dBFS
IMD3 = 87.9 dBFS

G009

Figure 10.

**Two-Tone Intermodulation Distortion
(FFT For 151-MHz and 152 MHz Input Signals)**

F_{IN1} = 151 MHz, −16 dBFS
F_{IN2} = 152 MHz, −16 dBFS
IMD3 = 99.4 dBFS

G010

Figure 11.

**Two-Tone Intermodulation Distortion
(FFT For 254-MHz and 255 MHz Input Signals)**

F_{IN1} = 254 MHz, −7 dBFS
F_{IN2} = 255 MHz, −7 dBFS
IMD3 = 80.5 dBFS

G011

Figure 12.

**Two-Tone Intermodulation Distortion
(FFT For 254-MHz and 255 MHz Input Signals)**

F_{IN1} = 254 MHz, −16 dBFS
F_{IN2} = 255 MHz, −16 dBFS
IMD3 = 93.9 dBFS

G012

Figure 13.

TYPICAL CHARACTERISTICS (continued)

Two-Tone Intermodulation Distortion
(FFT For 355-MHz and 356 MHz Input Signals)

F_{IN1} = 355 MHz, −7 dBFS
F_{IN2} = 356 MHz, −7 dBFS
IMD3 = 79.7 dBFS
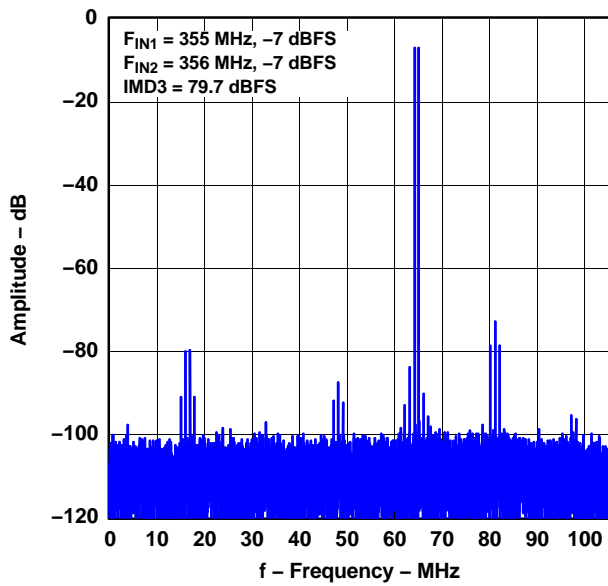
G013

Figure 14.

Two-Tone Intermodulation Distortion
(FFT For 355-MHz and 356 MHz Input Signals)

F_{IN1} = 355 MHz, −16 dBFS
F_{IN2} = 356 MHz, −16 dBFS
IMD3 = 96.8 dBFS
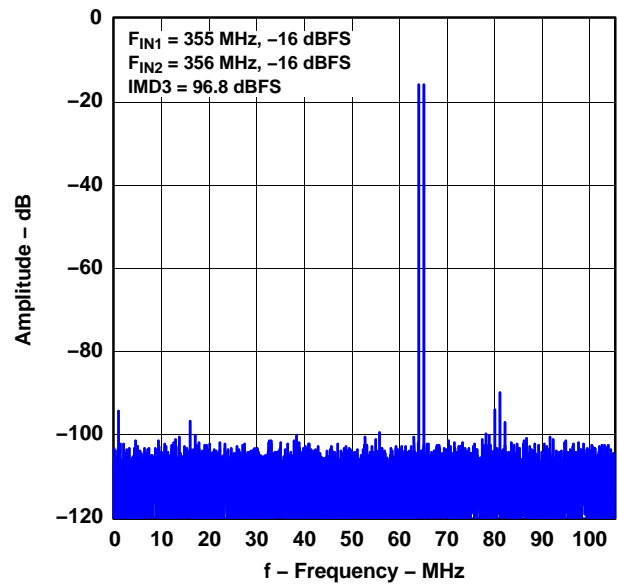
G014

Figure 15.

WCDMA Carrier
(f_{IN} = 70 MHz)

f_S = 184.32 MSPS
f_{IN} = 70 MHz
ACPR Adj Top = 73 dB
ACPR Adj Low = 72.7 dB
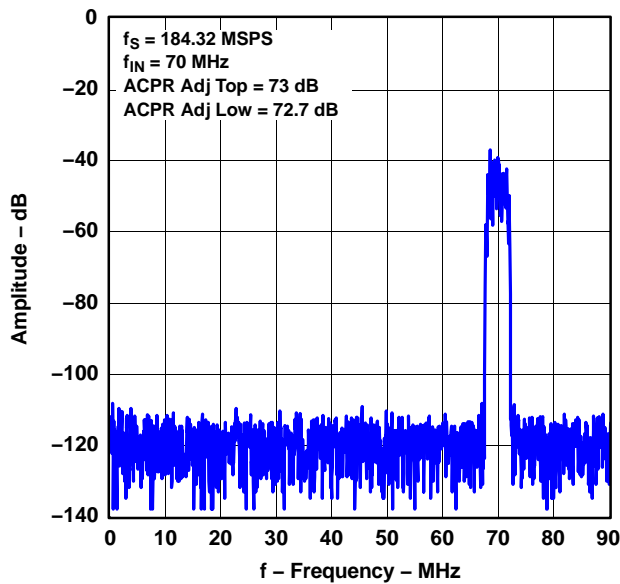
G015

Figure 16.

WCDMA Carrier
(f_{IN} = 140 MHz)

f_S = 184.32 MSPS
f_{IN} = 140 MHz
ACPR Adj Top = 72.1 dB
ACPR Adj Low = 72.1 dB
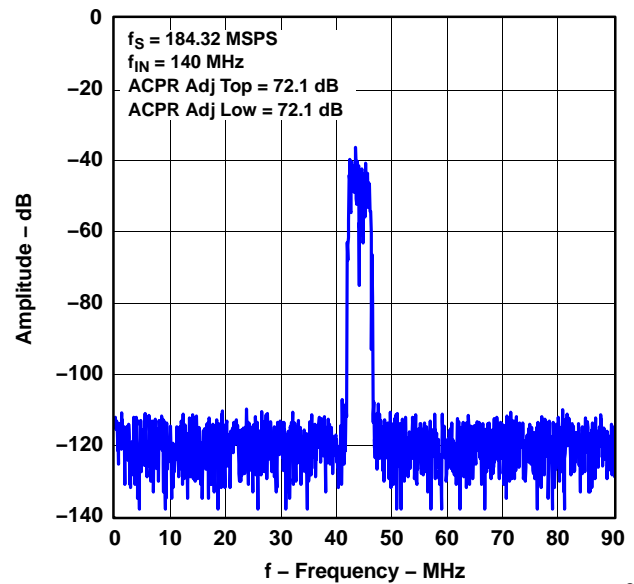
G016

Figure 17.

TYPICAL CHARACTERISTICS (continued)
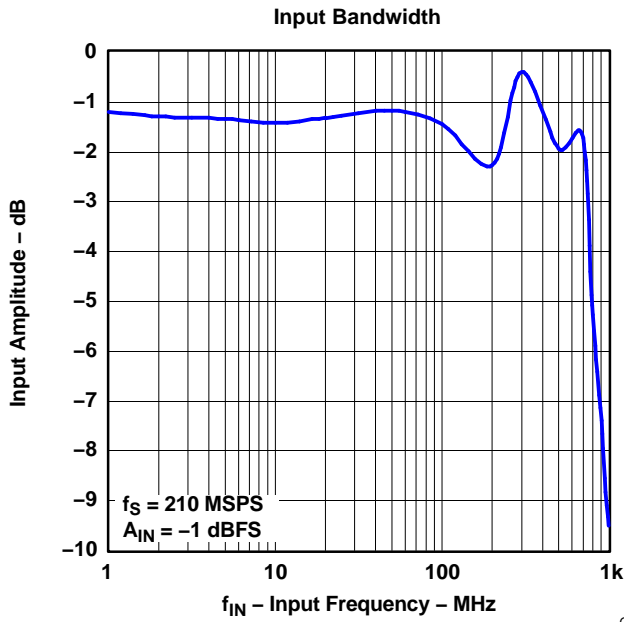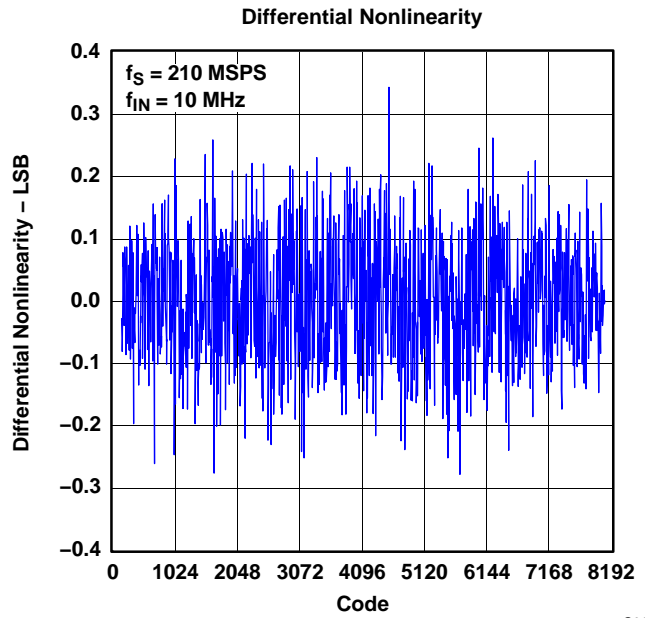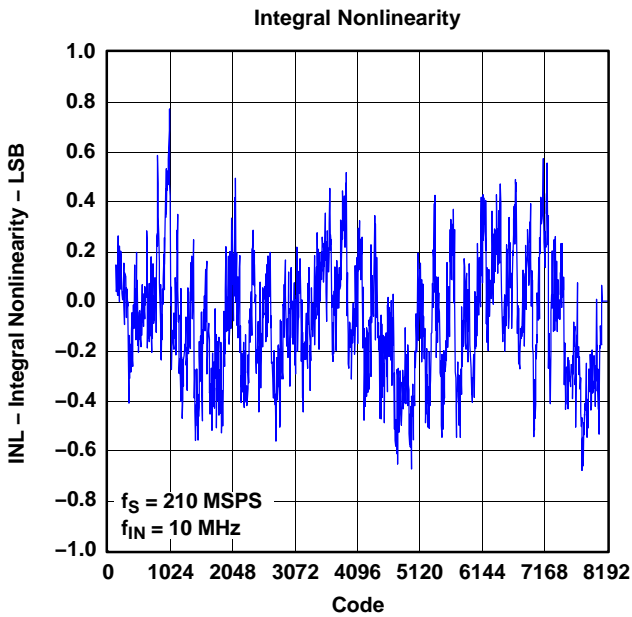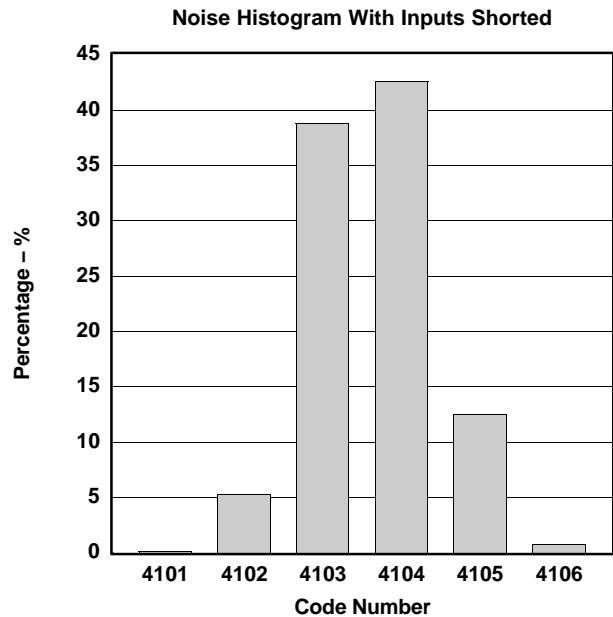
Input Bandwidth

Figure 18.

Differential Nonlinearity

Figure 19.

Integral Nonlinearity

Figure 20.

Noise Histogram With Inputs Shorted

Figure 21.

TYPICAL CHARACTERISTICS (continued)

**AC Performance
vs
Input Amplitude**

Figure 22.

**AC Performance
vs
Input Amplitude**

Figure 23.

**Two-Tone Spurious Free Dynamic Range
vs
Input Amplitude**

Figure 24.

**Spurious Free Dynamic Range
vs
Clock Duty Cycle**

Figure 25.

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TYPICAL CHARACTERISTICS (continued)

Spurious Free Dynamic Range
vs
Clock Level

Figure 26.

Signal-to-Noise Ratio
vs
Clock Level

Figure 27.

Performance
vs
Clock Common Mode Level

Figure 28.

Spurious Free Dynamic Range
vs
AVDD Across Temperature

Figure 29.

TYPICAL CHARACTERISTICS (continued)

Signal-to-Noise Ratio
vs
AVDD Across Temperature

Figure 30.

Spurious Free Dynamic Range
vs
DRVDD Across Temperature

Figure 31.

Signal-to-Noise Ratio
vs
DRVDD Across Temperature

Figure 32.

TYPICAL CHARACTERISTICS (continued)

**SNR
vs
Input Frequency and Sampling Frequency**

Figure 33.

TYPICAL CHARACTERISTICS (continued)

SFDR
vs
Input Frequency and Sampling Frequency
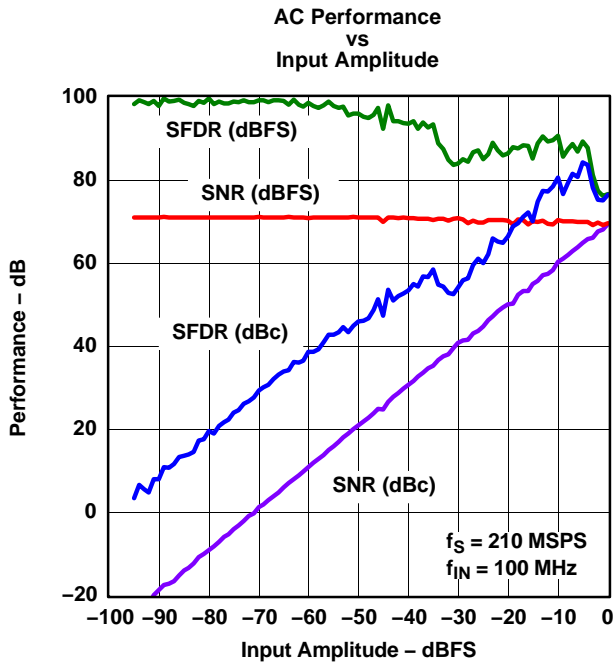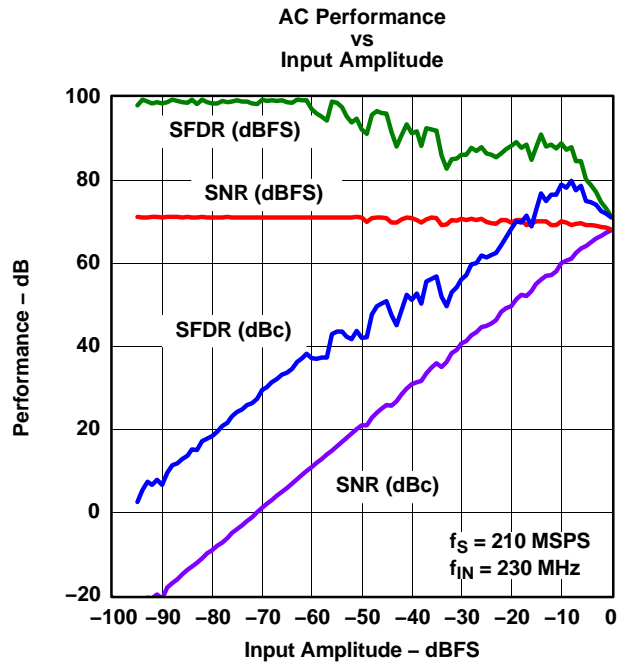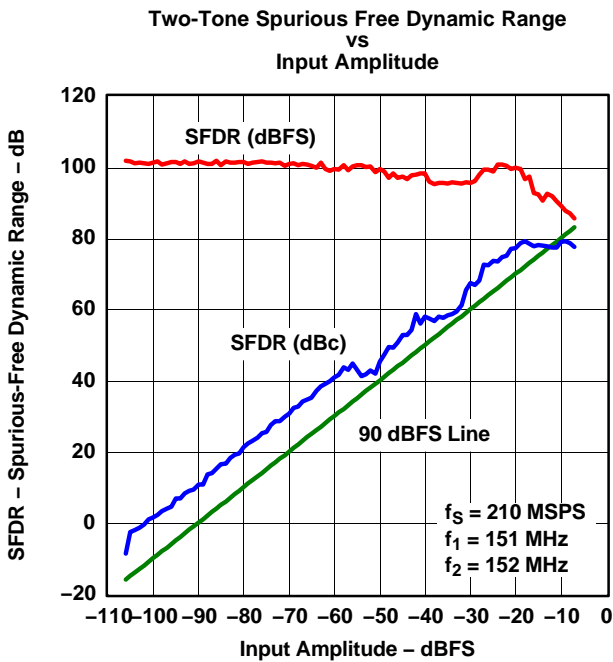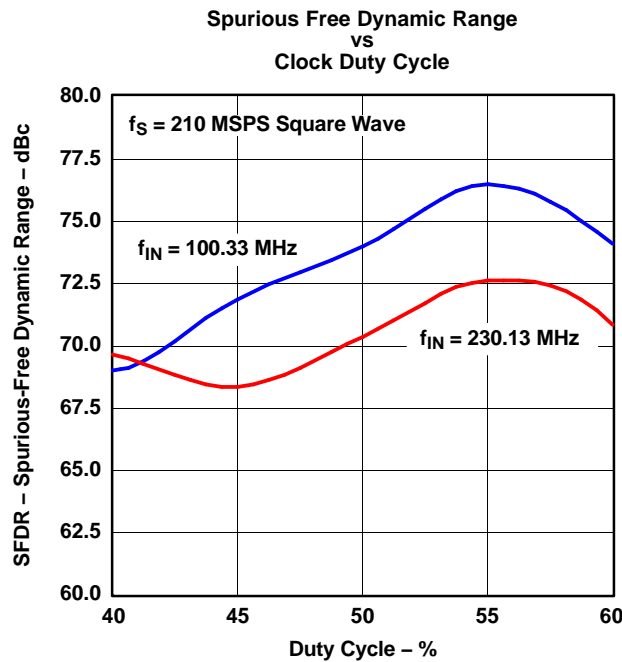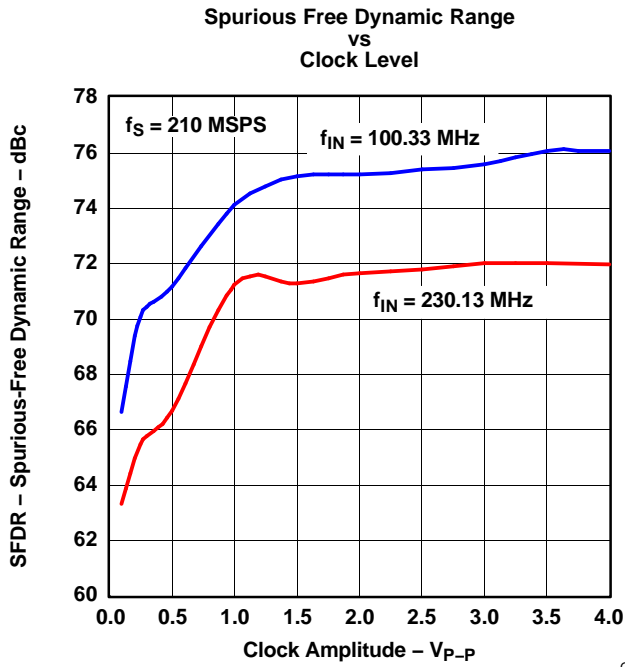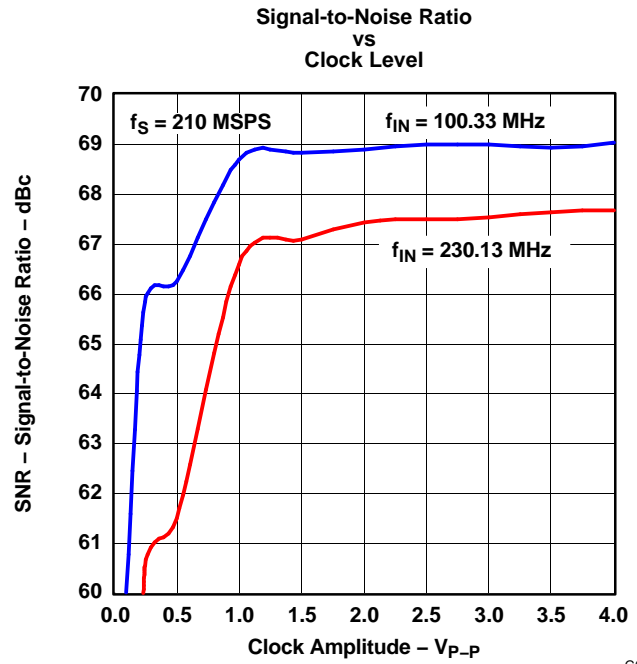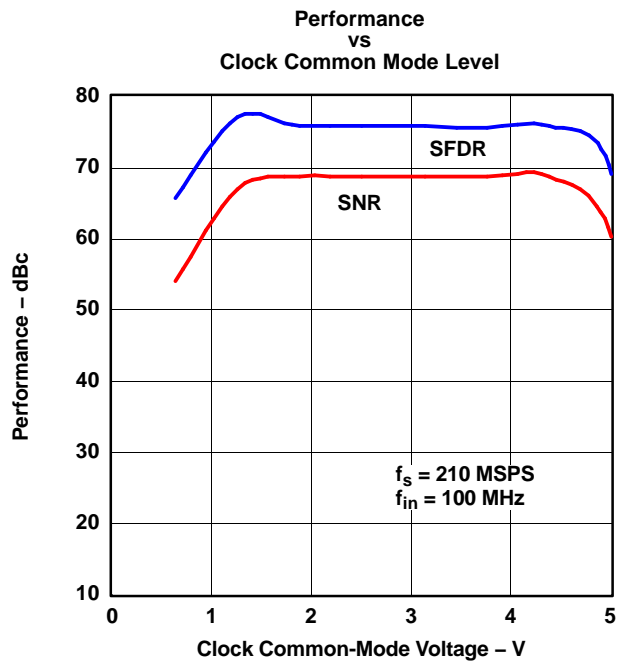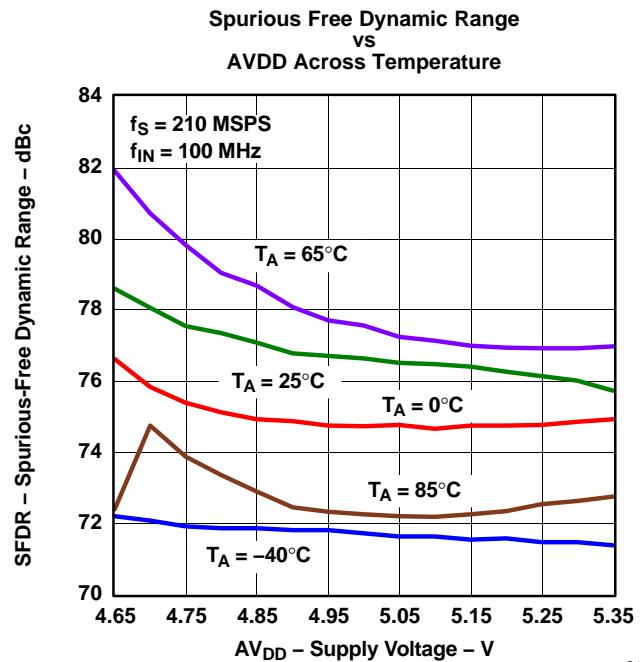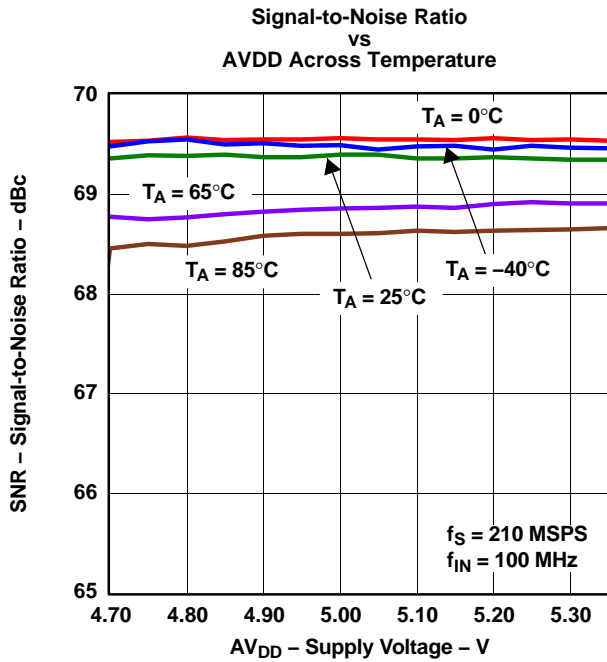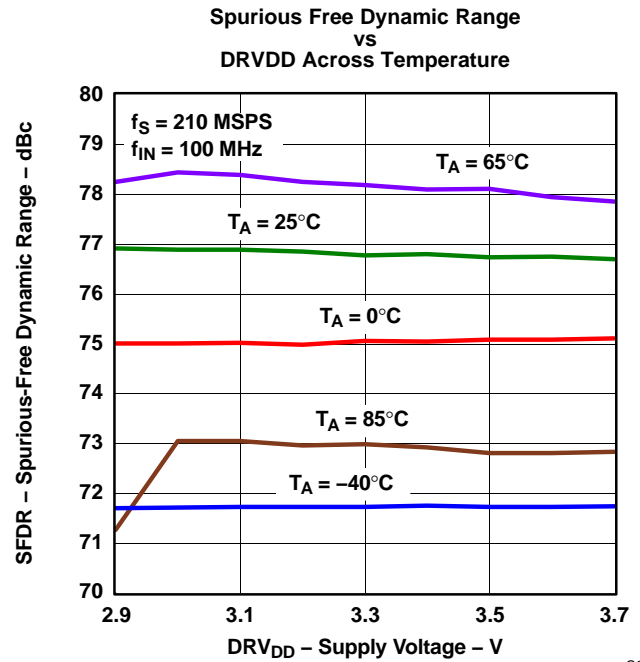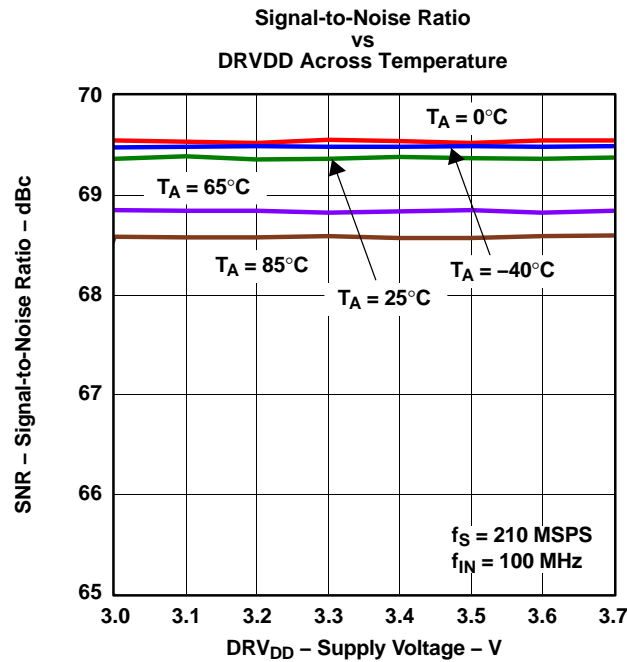
Figure 34.
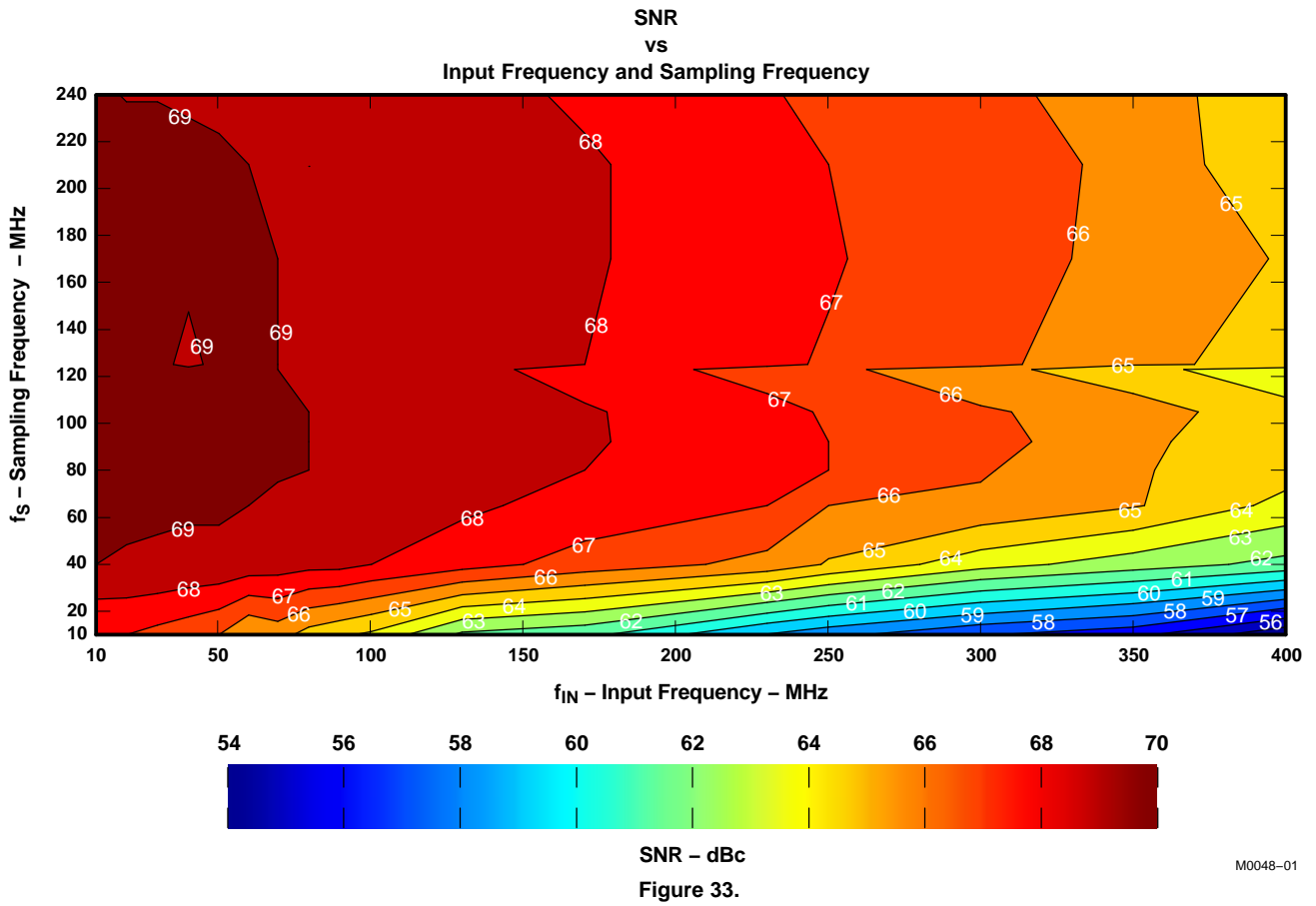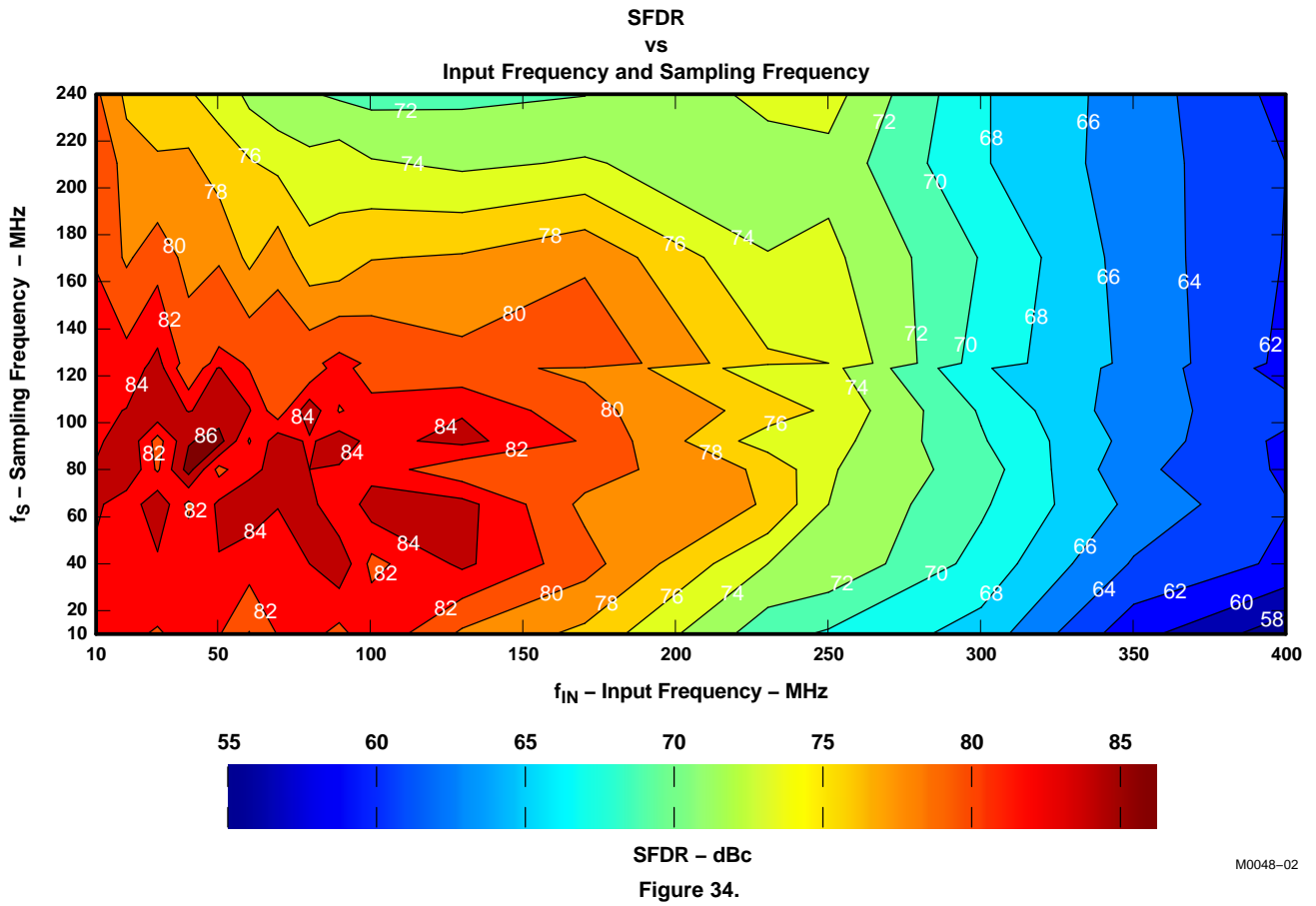
APPLICATION INFORMATION

Theory of Operation

The ADS5440 is a 13 bit, 210 MSPS, monolithic pipeline analog to digital converter. Its bipolar analog core operates from a 5 V supply, while the output uses a 3.3 V supply to provide LVDS compatible outputs. The conversion process is initiated by the rising edge of the external input clock. At that instant, the differential input signal is captured by the input track and hold (T&H) and the input sample is sequentially converted by a series of small resolution stages, with the outputs combined in a digital correction logic block. Both the rising and the falling clock edges are used to propagate the sample through the pipeline every half clock cycle. This process results in a data latency of four clock cycles, after which the output data is available as a 13 bit parallel word, coded in offset binary format.

Input Configuration

The analog input for the ADS5440 consists of an analog differential buffer followed by a bipolar track-and-hold. The analog buffer isolates the source driving the input of the ADC from any internal switching. The input common mode is set internally through a 500 Ω resistor connected from 2.4 V to each of the inputs. This results in a differential input impedance of 1 kΩ.

For a full-scale differential input, each of the differential lines of the input signal (pins 16 and 17) swings symmetrically between 2.4 +0.55 V and 2.4 -0.55 V. This means that each input has a maximum signal swing of 1.1 V_{PP} for a total differential input signal swing of 2.2 V_{PP}. The maximum swing is determined by the internal reference voltage generator eliminating the need for any external circuitry for this purpose.

The ADS5440 obtains optimum performance when the analog inputs are driven differentially. The circuit in Figure 35 shows one possible configuration using an RF transformer with termination either on the primary or on the secondary of the transformer. If voltage gain is required, a step up transformer can be used. For voltage gains that would require an impractical transformer turn ratio, a single-ended amplifier driving the transformer is shown in Figure 36).
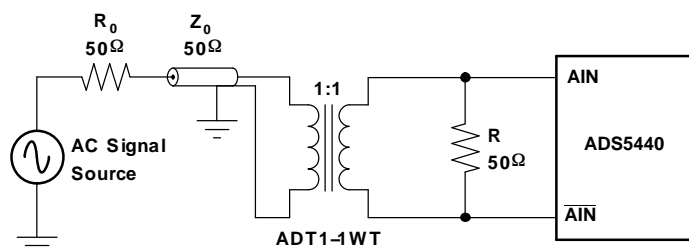
Figure 35. Converting a Single-Ended Input to a Differential Signal Using RF Transformers
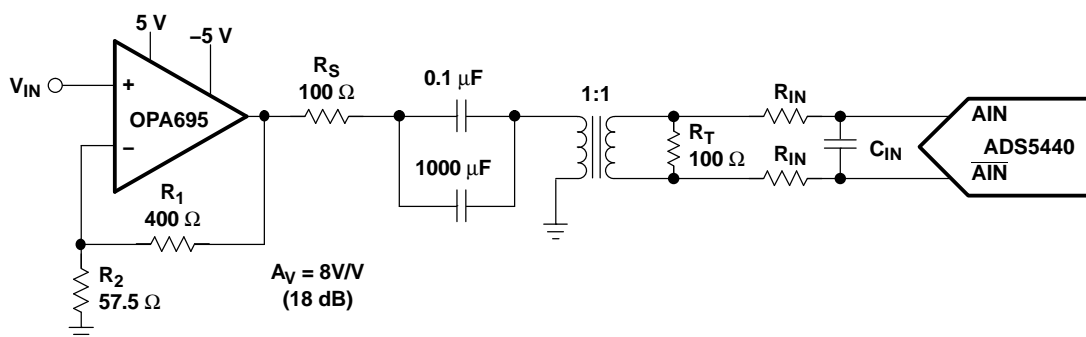
Figure 36. Using the OPA695 With the ADS5440
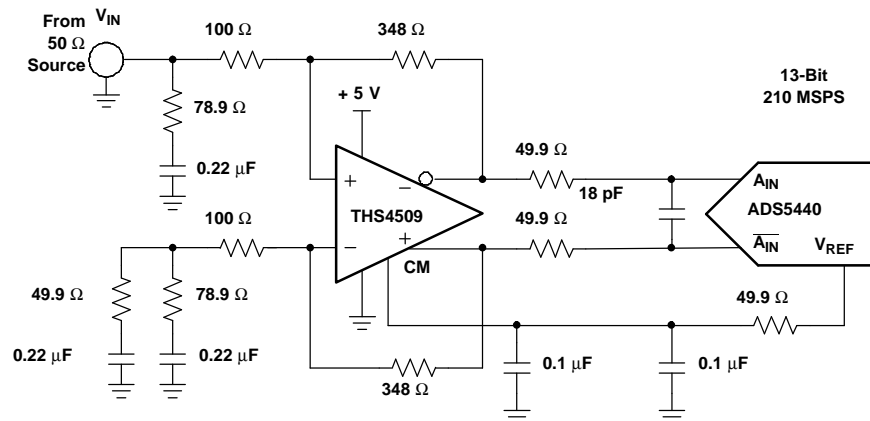
Application Information (continued)

Figure 37. Using the THS4509 With the ADS5440

Besides the OPA695, Texas Instruments offers a wide selection of single-ended operational amplifiers that can be selected depending on the application. An RF gain block amplifier, such as Texas Instrument's THS9001, can also be used with an RF transformer for high input frequency applications. For applications requiring dc-coupling with the signal source, a differential input/differential output amplifier like the THS4509 (see Figure 37) is a good solution as it minimizes board space and reduces the number of components.

In this configuration, the THS4509 amplifier circuit provides 10 dB of gain, converts the single-ended input to differential, and sets the proper input common-mode voltage to the ADS5440.

The 50 Ω resistors and 18 pF capacitor between the THS4509 outputs and ADS5440 inputs (along with the input capacitance of the ADC) limit the bandwidth of the signal to about 70 MHz (-3 dB).

Input termination is accomplished via the 78.9 Ω resistor and 0.22 μF capacitor to ground in conjunction with the input impedance of the amplifier circuit. A 0.22 μF capacitor and 49.9 Ω resistor is inserted to ground across the 78.9 Ω resistor and 0.22 μF capacitor on the alternate input to balance the circuit.

Gain is a function of the source impedance, termination, and 348 Ω feedback resistor. See the THS4509 data sheet for further component values to set proper 50 Ω termination for other common gains.

Since the ADS5440 recommended input common-mode voltage is +2.4 V, the THS4509 is operated from a single power supply input with V_{S+} = +5 V and V_{S-} = 0 V (ground). This maintains maximum headroom on the internal transistors of the THS4509.

Clock Inputs

The ADS5440 clock input can be driven with either a differential clock signal or a single-ended clock input, with little or no difference in performance between both configurations. In low input frequency applications, where jitter may not be a big concern, the use of single-ended clock (see Figure 38) could save some cost and board space without any trade-off in performance. When driven on this configuration, it is best to connect $\overline{CLK}$ to ground with a 0.01 μF capacitor, while CLK is ac-coupled with a 0.01 μF capacitor to the clock source, as shown in Figure 38.
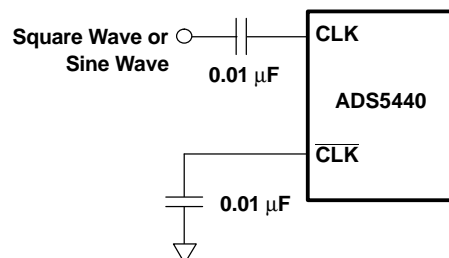

Figure 38. Single-Ended Clock
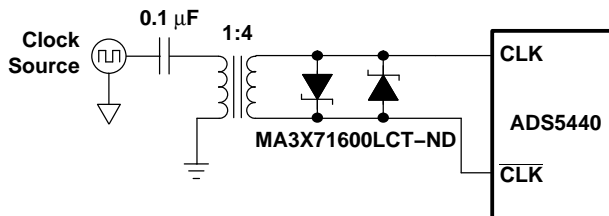
Application Information (continued)

Figure 39. Differential Clock

Nevertheless, for jitter sensitive applications, the use of a differential clock has some advantages (as with any other ADC) at the system level. The first advantage is that it allows for common-mode noise rejection at the PCB level.

A differential clock also allows for the use of bigger clock amplitudes without exceeding the absolute maximum ratings. In the case of a sinusoidal clock, this results in higher slew rates and reduces the impact of clock noise on jitter. See Clocking High Speed Data Converters (SLYT075) for more detail.

Figure 39 shows this approach. The back-to-back Schottky diodes can be added to limit the clock amplitude in cases where this would exceed the absolute maximum ratings, even when using a differential clock.
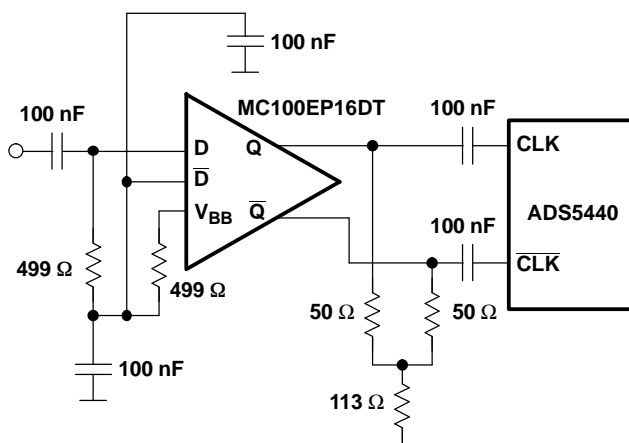
Figure 40. Differential Clock Using PECL Logic

Another possibility is the use of a logic based clock, such as PECL. In this case, the slew rate of the edges will most likely be much higher than the one obtained for the same clock amplitude based on a sinusoidal clock. This solution would minimize the effect of the slope dependent ADC jitter. Using logic gates to square a sinusoidal clock may not produce the best results as logic gates may not have been optimized to act as comparators, adding too much jitter while squaring the inputs.

The common-mode voltage of the clock inputs is set internally to 2.4 V using internal 1 kΩ resistors. It is recommended to use ac coupling, but if this scheme is not possible due to, for instance, asynchronous clocking, the ADS5440 features good tolerance to clock common-mode variation.

Additionally, the internal ADC core uses both edges of the clock for the conversion process. Ideally, a 50% duty cycle clock signal should be provided.

Digital Outputs

The ADC provides 13 data outputs (D12 to D0, with D12 being the MSB and D0 the LSB), a data-ready signal (DRY), and an over-range indicator (OVR) that equals a logic high when the output reaches the full-scale limits. The output format is offset binary. It is recommended to use the DRY signal to capture the output data of the ADS5440.

Application Information (continued)

The ADS5440 digital outputs are LVDS compatible.

Power Supplies

The use of low noise power supplies with adequate decoupling is recommended. Linear supplies are the preferred choice versus switched ones, which tend to generate more noise components that can be coupled to the ADS5440.

The ADS5440 uses two power supplies. For the analog portion of the design, a 5 V AVDD is used, while for the digital outputs supply (DRVDD) we recommend the use of 3.3 V. All the ground pins are marked as GND, although AGND pins and DRGND pins are not tied together inside the package.

Layout Information

The evaluation board represents a good guideline of how to layout the board to obtain the maximum performance out of the ADS5440. General design rules as the use of multilayer boards, single ground plane for ADC ground connections and local decoupling ceramic chip capacitors should be applied. The input traces should be isolated from any external source of interference or noise, including the digital outputs as well as the clock traces. The clock signal traces should also be isolated from other signals, especially in applications where low jitter is required as high IF sampling.

Besides performance oriented rules, care has to be taken when considering the heat dissipation out of the device. The thermal heat sink should be soldered to the board as described in the *PowerPad Package* section.

PowerPAD PACKAGE

The PowerPAD package is a thermally enhanced standard size IC package designed to eliminate the use of bulky heatsinks and slugs traditionally used in thermal packages. This package can be easily mounted using standard printed circuit board (PCB) assembly techniques, and can be removed and replaced using standard repair procedures.

The PowerPAD package is designed so that the leadframe die pad (or thermal pad) is exposed on the bottom of the IC. This provides an extremely low thermal resistance path between the die and the exterior of the package. The thermal pad on the bottom of the IC can then be soldered directly to the printed circuit board (PCB), using the PCB as a heatsink.

Assembly Process

1. Prepare the PCB top-side etch pattern including etch for the leads as well as the thermal pad as illustrated in the Mechanical Data section.
2. Place a 6-by-6 array of thermal vias in the thermal pad area. These holes should be 13 mils in diameter. The small size prevents wicking of the solder through the holes.
3. It is recommended to place a small number of 25 mil diameter holes under the package, but outside the thermal pad area to provide an additional heat path.
4. Connect all holes (both those inside and outside the thermal pad area) to an internal copper plane (such as a ground plane).
5. Do not use the typical web or spoke via connection pattern when connecting the thermal vias to the ground plane. The spoke pattern increases the thermal resistance to the ground plane.
6. The top-side solder mask should leave exposed the terminals of the package and the thermal pad area.
7. Cover the entire bottom side of the PowerPAD vias to prevent solder wicking.
8. Apply solder paste to the exposed thermal pad area and all of the package terminals.

For more detailed information regarding the PowerPAD package and its thermal properties, see either the SLMA004 Application Brief *PowerPAD Made Easy* or the SLMA002 Technical Brief *PowerPAD Thermally Enhanced Package*.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
ADS5440IPFP	ACTIVE	HTQFP	PFP	80	96	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-4-260C-72 HR	-40 to 85	ADS5440IPFP	Samples
ADS5440IPFPR	OBSOLETE	HTQFP	PFP	80		TBD	Call TI	Call TI	-40 to 85	ADS5440IPFP	
ADS5440IPFPRG4	OBSOLETE	HTQFP	PFP	80		TBD	Call TI	Call TI	-40 to 85		

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

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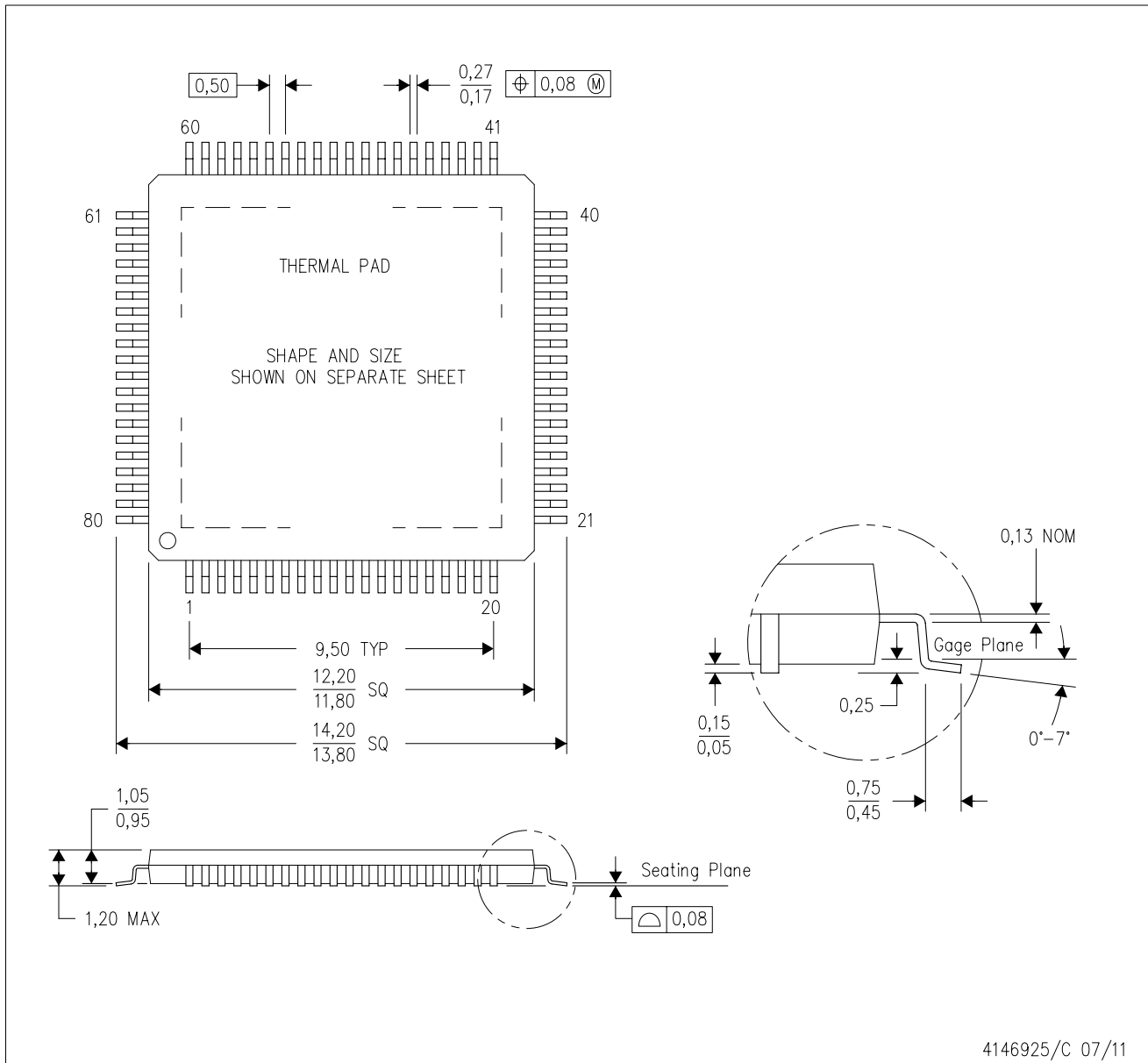**OTHER QUALIFIED VERSIONS OF ADS5440 :**

● Enhanced Product: ADS5440-EP

NOTE: Qualified Version Definitions:

   ● Enhanced Product - Supports Defense, Aerospace and Medical Applications

PFP (S−PQFP−G80)　　　　　　PowerPAD™ PLASTIC QUAD FLATPACK

THERMAL PAD

SHAPE AND SIZE
SHOWN ON SEPARATE SHEET

0,50

$\frac{0,27}{0,17}$

⊕ 0,08 Ⓜ

60　41
61　40
80　21
1　20

9,50 TYP

$\frac{12,20}{11,80}$ SQ

$\frac{14,20}{13,80}$ SQ

$\frac{1,05}{0,95}$

1,20 MAX

Seating Plane

⌓ 0,08

0,13 NOM

Gage Plane

$\frac{0,15}{0,05}$

0,25

$\frac{0,75}{0,45}$

0°−7°

4146925/C 07/11

NOTES:　A.　All linear dimensions are in millimeters.
　　　　　B.　This drawing is subject to change without notice.
　　　　　C.　Body dimensions do not include mold flash or protrusion
　　　　　D.　This package is designed to be soldered to a thermal pad on the board.　Refer to Technical Brief, PowerPad
　　　　　　　Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding
　　　　　　　recommended board layout.　This document is available at www.ti.com <http://www.ti.com>.
　　　　　E.　See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
　　　　　F.　Falls within JEDEC MS−026

PowerPAD is a trademark of Texas Instruments.

PFP (S–PQFP–G80)   PowerPAD™ PLASTIC QUAD FLATPACK

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

Exposed Thermal Pad

$\dfrac{7,50}{6,15}$

$\dfrac{7,50}{6,15}$

Top View

Exposed Thermal Pad Dimensions
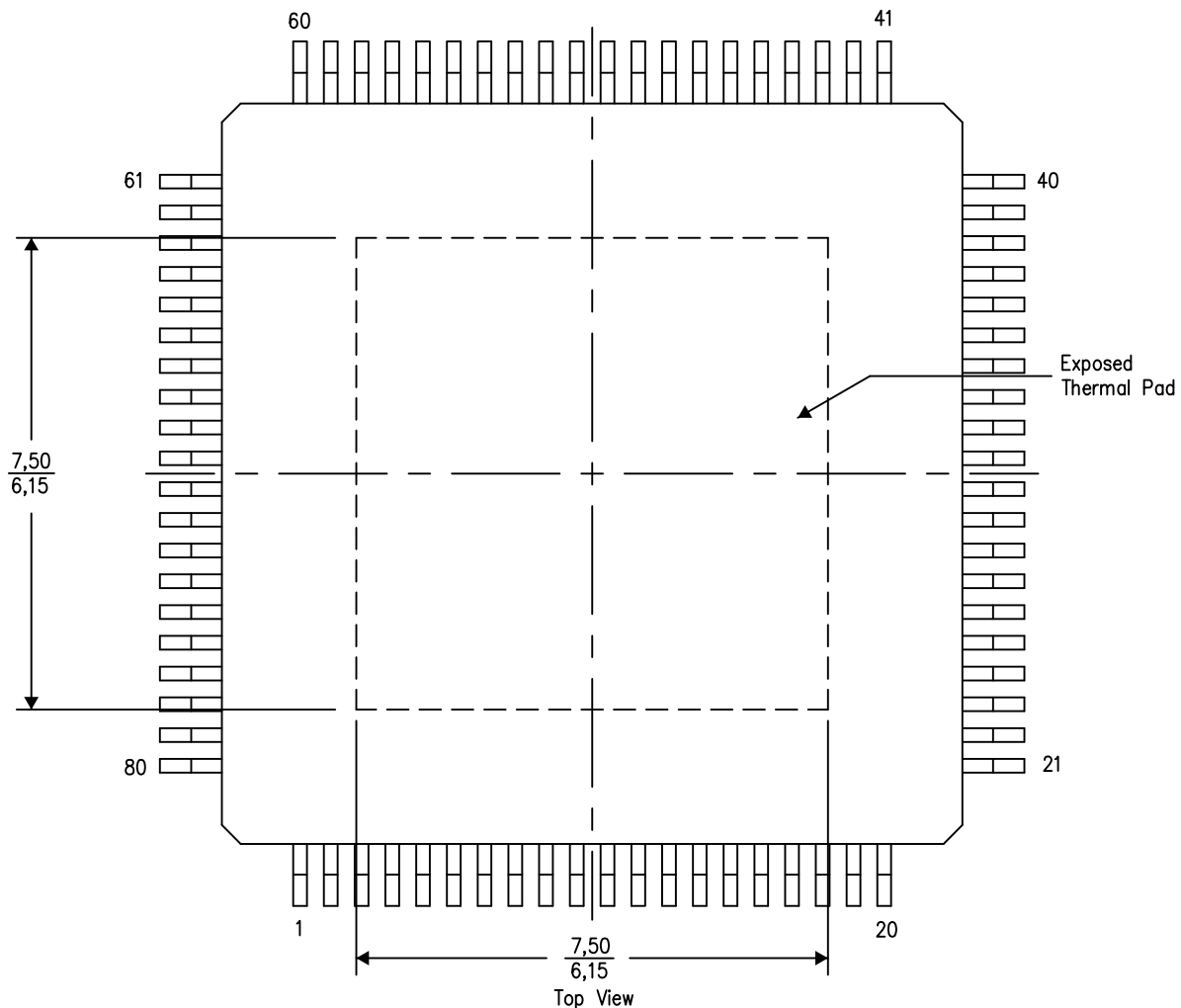
4206327–2/P 05/14

NOTE: A.   All linear dimensions are in millimeters

PowerPAD is a trademark of Texas Instruments

PFP (S−PQFP−G80)   PowerPAD™ PLASTIC QUAD FLATPACK

Example Board Layout
Via pattern and copper area under solder mask
may vary depending on layout constraints

Stencil Openings
Based on a stencil thickness
of .125mm (.005inch).
Reference table below for other
solder stencil thicknesses

49xø0,2
Centered
Array on Pad

76x0,5

80x1,5   76x0,5

80x0,3

1,3

1,3

7,5 10,0 13,4

Y

7,5   13,4

X

7,5

Solder Mask
Over Copper   Pin 1

7,5
10,0
13,4

Solder Mask
Defined Pad
(See Note C, D)

Pin 1   13,4

(See Note E)

0,3

Example
Solder Mask Opening
(See Note F)

1,5

Example
Pad Geometry

0,05
All Around

Center Power Pad Solder Stencil Opening		
Stencil Thickness	X	Y
0.1mm	8.39	8.39
0.125mm	7.5	7.5
0.15mm	6.85	6.85
0.175mm	6.34	6.34

4208544−2/G 02/16

PowerPAD is a trademark of Texas Instruments.

NOTES: A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad
Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets
for specific thermal information, via requirements, and recommended board layout. These documents are available at
www.ti.com <http://www.ti.com>. Publication IPC−7351 is recommended for alternate designs.
E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should
contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric
metal load solder paste. Refer to IPC−7525 for other stencil recommendations.
F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

IMPORTANT NOTICE

Products		Applications	
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Wireless Connectivity	www.ti.com/wirelessconnectivity		