

FEATURES

High dc precision

- 150 μV maximum offset voltage
- 1.5 $\mu\text{V}/^\circ\text{C}$ maximum offset voltage drift
- 270 pA maximum input bias current
- 0.3 pA/ $^\circ\text{C}$ typical I_B drift

Low noise: 0.5 μV p-p

Typical noise: 0.1 Hz to 10 Hz

Low power: 600 μA maximum supply current per amplifier

Dual version: [AD706](#)

APPLICATIONS

Industrial/process controls

Weigh scales

ECG/EKG instrumentation

Low frequency active filters

GENERAL DESCRIPTION

The AD704 is a quad, low power bipolar op amp that has the low input bias current of a BiFET amplifier and offers a significantly lower I_B drift over temperature. It uses superbeta bipolar input transistors to achieve picoampere input bias current levels (similar to FET input amplifiers at room temperature), while its I_B typically increases only by $5\times$ at 125°C (unlike a BiFET amp, for which I_B doubles every 10°C , resulting in a $1000\times$ increase at 125°C). In addition, the AD704 achieves 150 μV offset voltage and the low noise characteristics of a precision bipolar input op amp.

Because it has only 1/20 the input bias current of an OP07, the AD704 does not require the commonly used balancing resistor. Furthermore, the current noise is 1/5 that of the OP07, which makes the AD704 usable with much higher source impedances. At 1/6 the supply current (per amplifier) of the OP07, the AD704 is better suited for today's higher density circuit boards and battery-powered applications.

The AD704 is an excellent choice for use in low frequency active filters in 12- and 14-bit data acquisition systems, in precision instrumentation, and as a high quality integrator. The AD704 is internally compensated for unity gain stability. The AD704J is rated over the commercial temperature range of 0°C to 70°C . The AD704A is rated over the industrial temperature of -40°C to $+85^\circ\text{C}$. The AD704S is rated over the military temperature range of -55°C to $+125^\circ\text{C}$, processed to MIL-STD-883B.

CONNECTION DIAGRAMS

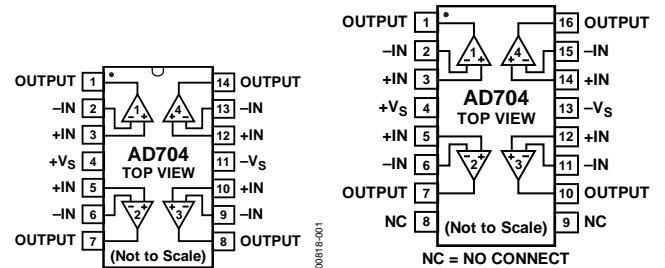


Figure 1. 14-Lead Plastic DIP (N)

Figure 2. 16-Lead SOIC (R) Package

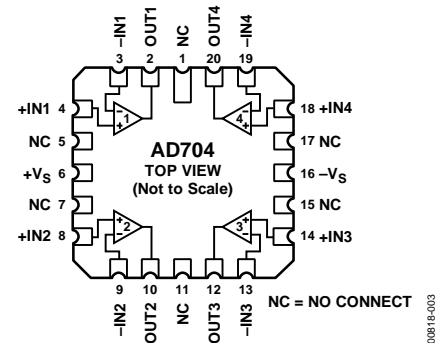


Figure 3. 20-Terminal LCC (E-20-1) Package

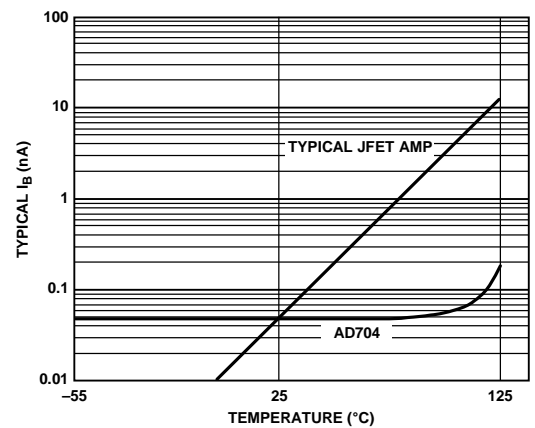


Figure 4. Input Bias Current Over Temperature

Table 1. Low I_B @ 125°C

Model	30V	16V	1.3 to 5V	Next Generation
Single	N/A	AD8663	AD8603	N/A
Dual	AD706	AD8667	AD8607	AD8622
Quad	AD704	AD8669	AD8609	AD8624

Rev. E

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AD704* Product Page Quick Links

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- AD704/883B: Quad Picoampere Input Current, Precision, Bipolar Op Amp Military Data Sheet
- AD704: Picoampere Input Current Quad Bipolar Op Amp Data Sheet

[Tools and Simulations](#)

- AD704 SPICE Macro-Model

[Design Resources](#)

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- PCN-PDN Information
- Quality And Reliability
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REVISION HISTORY

1/10—Rev. D to Rev. E

Updated Format.....	Universal
Changes to Features and General Description Section, Added Table 1, Renumbered Sequentially	1
Changes to Table 2.....	3
Changes to Table 3.....	5
Updated Outline Dimensions	13
Changes to Ordering Guide	14

12/09—Rev. C to Rev. D

Updated Outline Dimensions	10
Changes to Ordering Guide	10

11/01—Rev. B to Rev. C

Edits to Features.....	1
Edits to Product Description	1
Edits to Absolute Maximum Ratings	3
Deleted Metalization Photograph	3
Edits to Ordering Guide	4

SPECIFICATIONS

$T_A = 25^\circ\text{C}$, $V_{CM} = 0\text{ V}$, and $V_S = \pm 15\text{ V}$ dc, unless otherwise noted.

Table 2.

Parameters	Conditions	AD704J/A			Unit
		Min	Typ	Max	
INPUT OFFSET VOLTAGE					
Initial Offset			50	150	μV
Offset	$T_{MIN} - T_{MAX}$		100	250	μV
vs. Temp, Average TC			0.2	1.5	$\mu\text{V}/^\circ\text{C}$
vs. Supply (PSRR)	$V_S = \pm 2\text{ V to } \pm 18\text{ V}$	100	132		dB
$T_{MIN} - T_{MAX}$	$V_S = \pm 2.5\text{ V to } \pm 18\text{ V}$	100	126		dB
Long-Term Stability			0.3		$\mu\text{V}/\text{month}$
INPUT BIAS CURRENT ¹	$V_{CM} = 0\text{ V}$		100	270	pA
	$V_{CM} = \pm 13.5\text{ V}$			300	pA
vs. Temp, Average TC			0.3		$\text{pA}/^\circ\text{C}$
$T_{MIN} - T_{MAX}$	$V_{CM} = 0\text{ V}$			300	pA
	$V_{CM} = \pm 13.5\text{ V}$			400	pA
INPUT OFFSET CURRENT	$V_{CM} = 0\text{ V}$		80	250	pA
	$V_{CM} = \pm 13.5\text{ V}$			300	pA
vs. Temp, Average TC			0.6		$\text{pA}/^\circ\text{C}$
$T_{MIN} - T_{MAX}$	$V_{CM} = 0\text{ V}$		100	300	pA
	$V_{CM} = \pm 13.5\text{ V}$		100	400	pA
MATCHING CHARACTERISTICS					
Offset Voltage	$T_{MIN} - T_{MAX}$			250	μV
				400	μV
Input Bias Current ²	$T_{MIN} - T_{MAX}$			500	pA
				600	pA
Common-Mode Rejection ³	$T_{MIN} - T_{MAX}$	94			dB
		94			dB
Power Supply Rejection ⁴	$T_{MIN} - T_{MAX}$	94			dB
		94			dB
Crosstalk ⁵	$T_{MIN} - T_{MAX}$ $f = 10\text{ Hz}$ $R_{LOAD} = 2\text{ k}\Omega$	94			dB
			150		dB
FREQUENCY RESPONSE UNITY GAIN					
Crossover Frequency			0.8		MHz
Slew Rate, Unity Gain	$G = -1$		0.15		$\text{V}/\mu\text{s}$
Slew Rate	$T_{MIN} - T_{MAX}$		0.1		$\text{V}/\mu\text{s}$
INPUT IMPEDANCE					
Differential			40 2		$\text{M}\Omega \text{pF}$
Common-Mode			300 2		$\text{G}\Omega \text{pF}$
INPUT VOLTAGE RANGE					
Common-Mode Voltage		± 13.5	± 14		V
Common-Mode Rejection Ratio	$V_{CM} = \pm 13.5\text{ V}$	100	132		dB
	$T_{MIN} - T_{MAX}$	98	128		dB
INPUT CURRENT NOISE	0.1 Hz to 10 Hz		3		pA p-p
	$f = 10\text{ Hz}$		50		$\text{fA}/\sqrt{\text{Hz}}$
INPUT VOLTAGE NOISE	0.1 Hz to 10 Hz		0.5		μV p-p
	$f = 10\text{ Hz}$		17		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$		15	22	$\text{nV}/\sqrt{\text{Hz}}$

AD704

Parameters	Conditions	AD704J/A			Unit
		Min	Typ	Max	
OPEN-LOOP GAIN	$V_O = \pm 12\text{ V}$				
	$R_{LOAD} = 10\text{ k}\Omega$	200	2000		V/mV
	$T_{MIN} - T_{MAX}$	150	1500		V/mV
	$V_O = \pm 10\text{ V}$				
OUTPUT CHARACTERISTICS	$R_{LOAD} = 2\text{ k}\Omega$	200	1000		V/mV
	$T_{MIN} - T_{MAX}$	150	1000		V/mV
	Voltage Swing				V
	$R_{LOAD} = 10\text{ k}\Omega$				
Current	$T_{MIN} - T_{MAX}$	± 13	± 14		mA
	Short circuit		± 15		mA
CAPACITIVE LOAD					
Drive Capability	Gain = 1		10,000		pF
POWER SUPPLY					
Rated Performance			± 15		V
Operating Range		± 2.0		± 18	V
Quiescent Current			1.5	2.4	mA
	$T_{MIN} - T_{MAX}$		1.6	2.6	mA
TRANSISTOR COUNT	Number of transistors		180		

¹ Bias current specifications are guaranteed maximum at either input.

² Input bias current match is the maximum difference between corresponding inputs of all four amplifiers.

³ CMRR match is the difference of $\Delta V_{OS}/\Delta V_{CM}$ between any two amplifiers, expressed in dB.

⁴ PSRR match is the difference between $\Delta V_{OS}/\Delta V_{SUPPLY}$ for any two amplifiers, expressed in dB.

⁵ See Figure 5 for test circuit.

ABSOLUTE MAXIMUM RATINGS

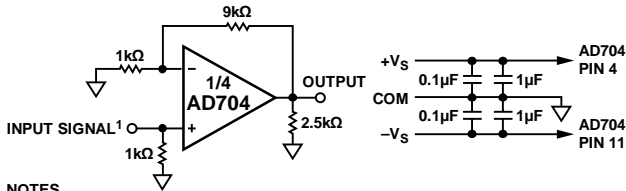
Table 3.

Parameter	Rating
Supply Voltage	±18 V
Internal Power Dissipation (25°C) ¹	
Input Voltage	±V _S
Differential Input Voltage ²	±0.7 V
Output Short-Circuit Duration (Single Input)	Indefinite
Storage Temperature Range	-65°C to +125°C
Operating Temperature Range	
AD704J	0°C to 70°C
AD704A	-40°C to +85°C
Lead Temperature (Soldering, 10 sec)	300°C

¹Specification is for the device in free air:
 14-lead plastic package: θ_{JA} = 150°C/W.
 16-lead SOIC package: θ_{JA} = 100°C/W.
 20-terminal LCC package: θ_{JA} = 150°C/W.

²The input pins of this amplifier are protected by back-to-back diodes. If the differential voltage exceeds ±0.7 volts, external series protection resistors should be added to limit the input current to less than 25 mA.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.



NOTES

1. ALL FOUR AMPLIFIERS ARE CONNECTED AS SHOWN.

¹THE SIGNAL INPUT (SUCH THAT THE AMPLIFIER'S OUTPUT IS AT MAXIMUM AMPLITUDE WITHOUT CLIPPING OR SLEW LIMITING) IS APPLIED TO ONE AMPLIFIER AT A TIME. THE OUTPUTS OF THE OTHER THREE AMPLIFIERS ARE THEN MEASURED FOR CROSSTALK.

Figure 5. Crosstalk Test Circuit

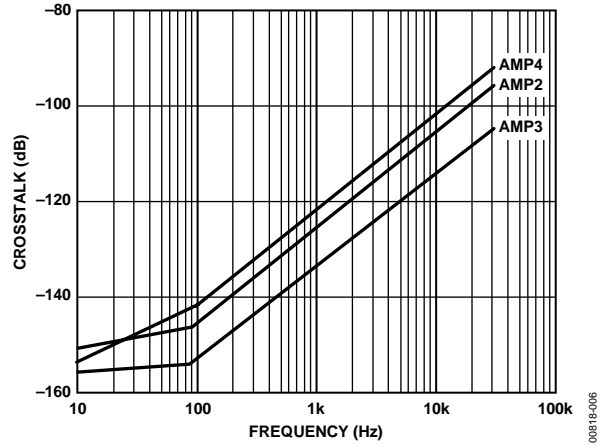


Figure 6. Crosstalk vs. Frequency

ESD CAUTION



ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

TYPICAL PERFORMANCE CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{ V dc}$, unless otherwise noted.

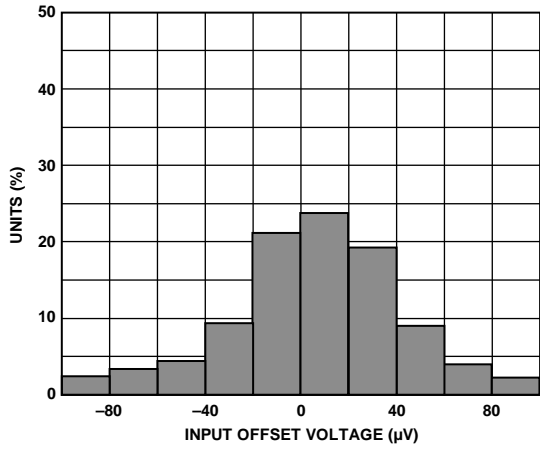


Figure 7. Typical Distribution of Input Offset Voltage

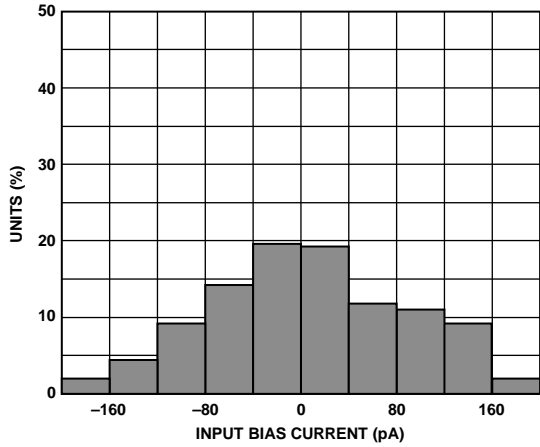


Figure 8. Typical Distribution of Input Bias Current

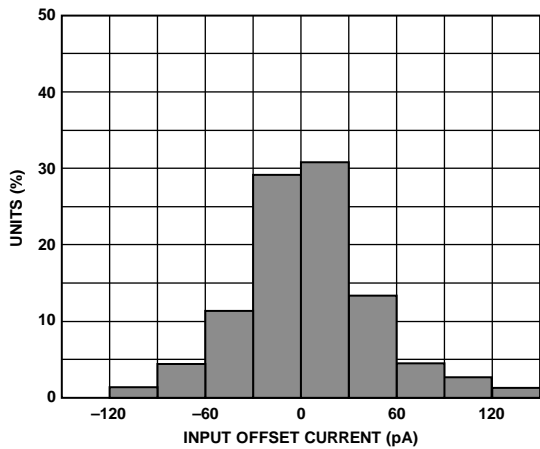


Figure 9. Typical Distribution of Input Offset Current

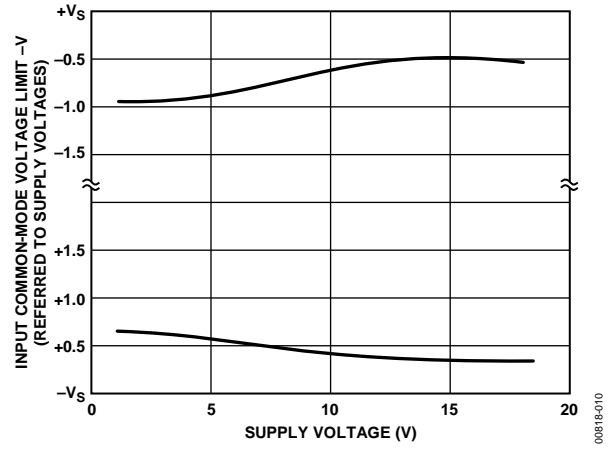


Figure 10. Input Common-Mode Voltage Range vs. Supply Voltage

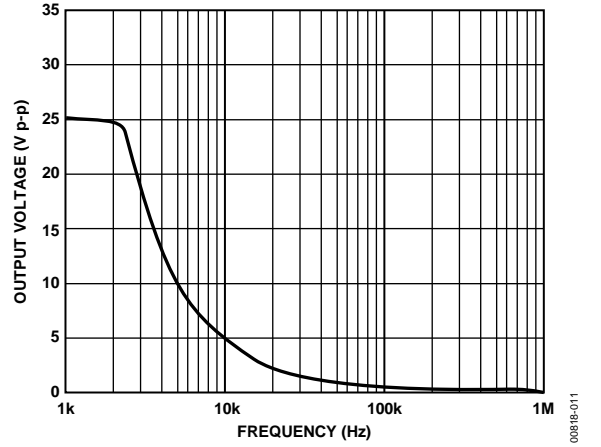


Figure 11. Large Signal Frequency Response

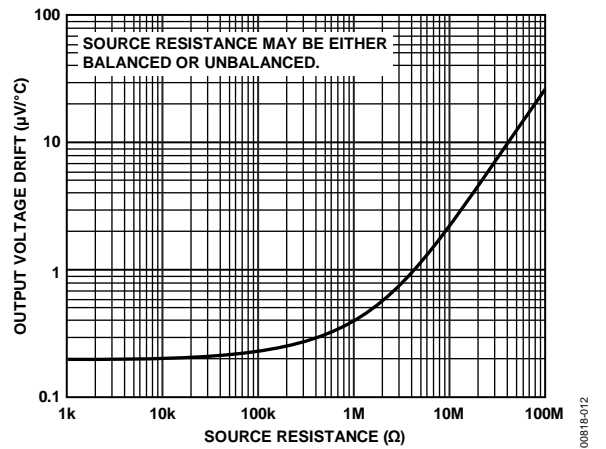


Figure 12. Offset Voltage Drift vs. Source Resistance

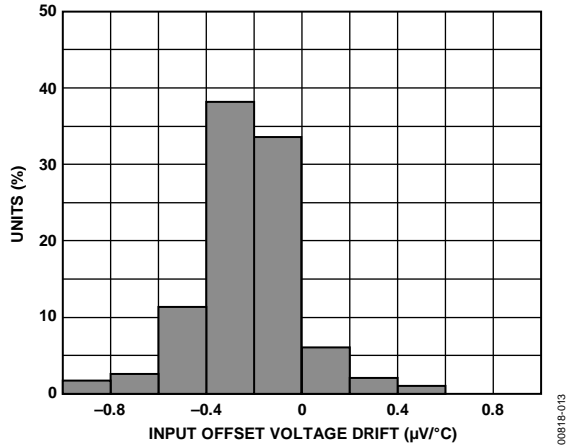


Figure 13. Typical Distribution of Input Offset Voltage Drift

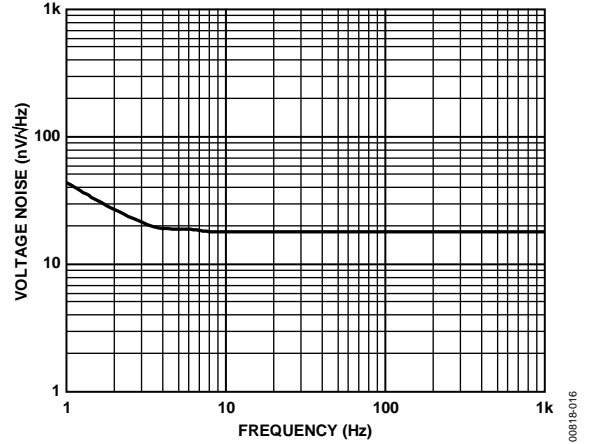


Figure 16. Input Noise Voltage Spectral Density

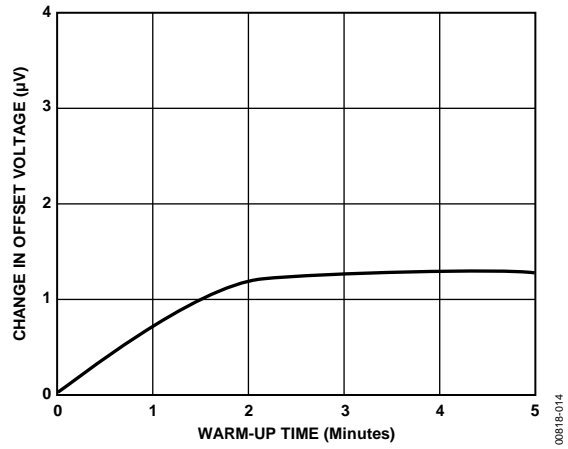


Figure 14. Change in Input Offset Voltage vs. Warm-Up Time

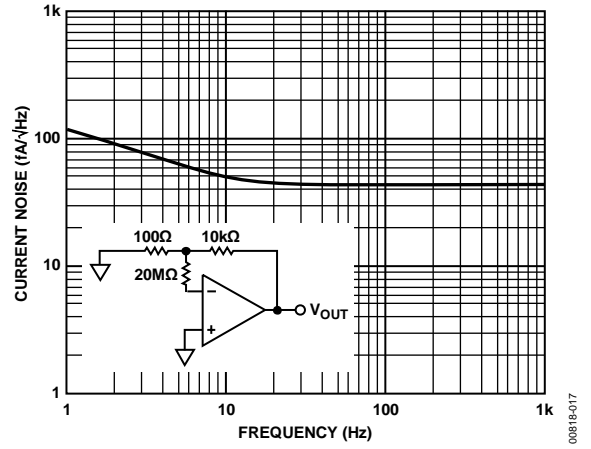


Figure 17. Input Noise Current Spectral Density

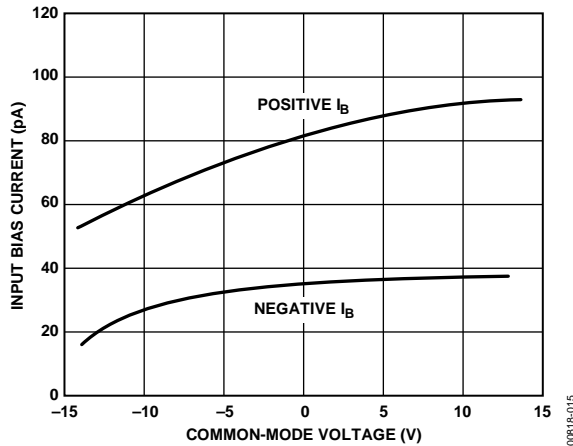


Figure 15. Input Bias Current vs. Common-Mode Voltage

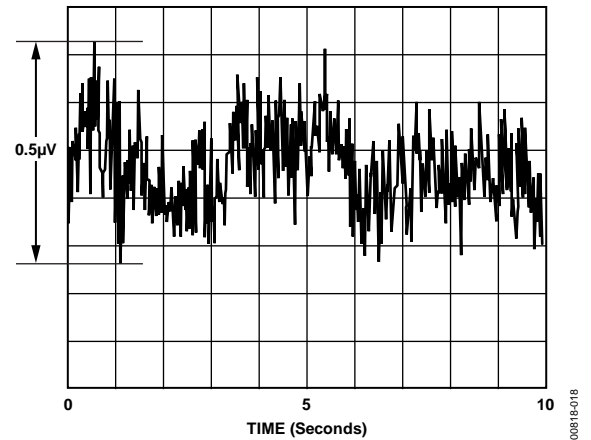


Figure 18. 0.1 Hz to 10 Hz Noise Voltage

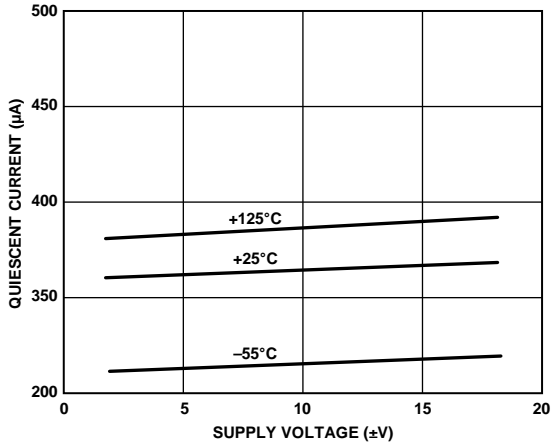


Figure 19. Quiescent Supply Current vs. Supply Voltage (per Amplifier)

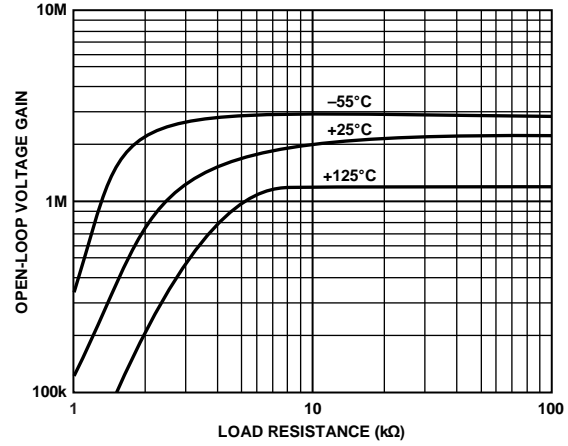


Figure 22. Open-Loop Gain vs. Load Resistance Over Temperature

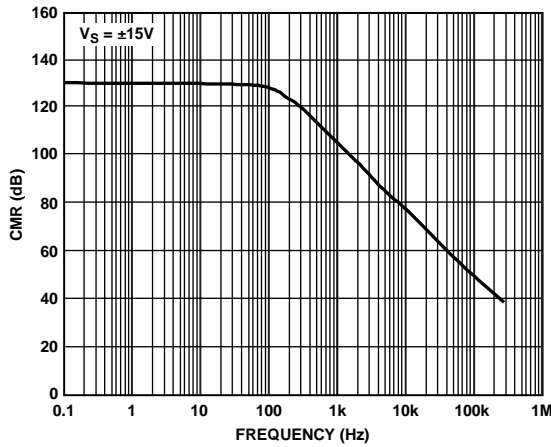


Figure 20. Common-Mode Rejection vs. Frequency

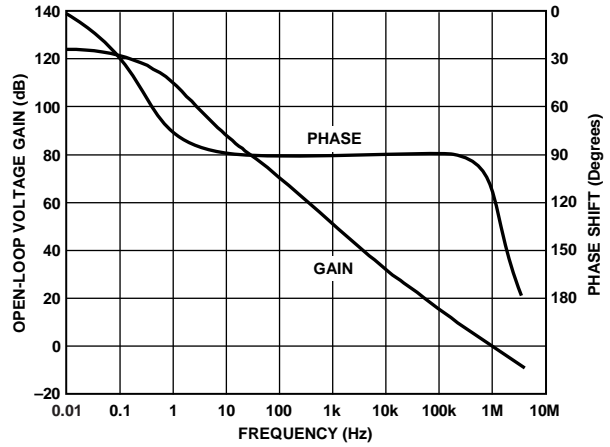


Figure 23. Open-Loop Gain and Phase vs. Frequency

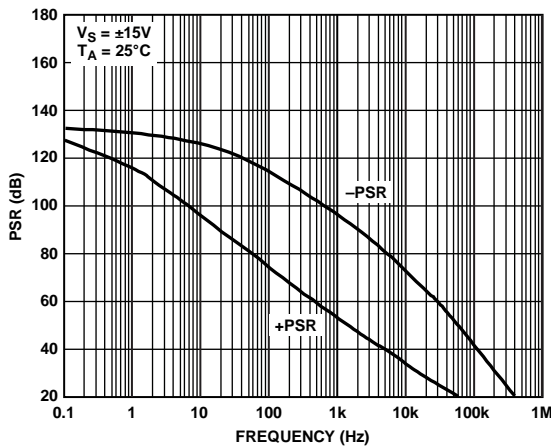


Figure 21. Power Supply Rejection vs. Frequency

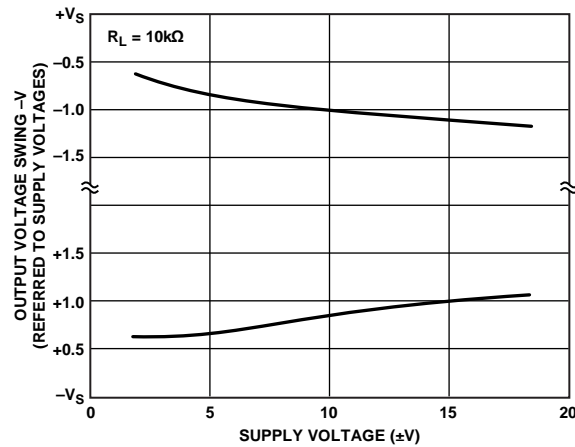


Figure 24. Output Voltage Swing vs. Supply Voltage

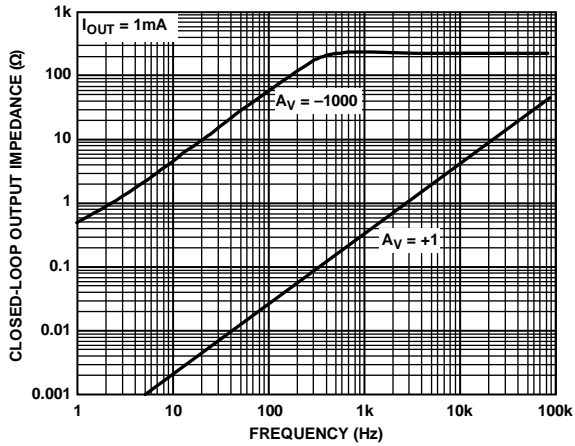


Figure 25. Closed-Loop Output Impedance vs. Frequency

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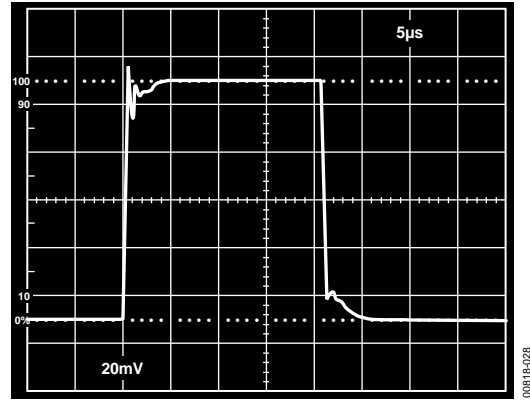


Figure 28. Unity Gain Follower Small Signal Pulse Response $R_F = 0 \Omega$, $C_L = 100 \text{ pF}$

00818-028

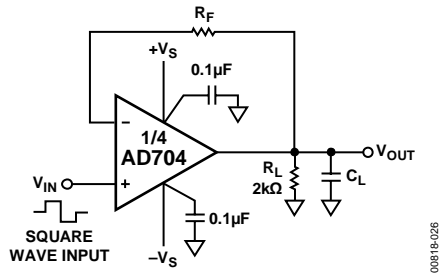


Figure 26. Unity Gain Follower (for Large Signal Applications, Resistor R_F Limits the Current Through the Input Protection Diodes)

00818-026

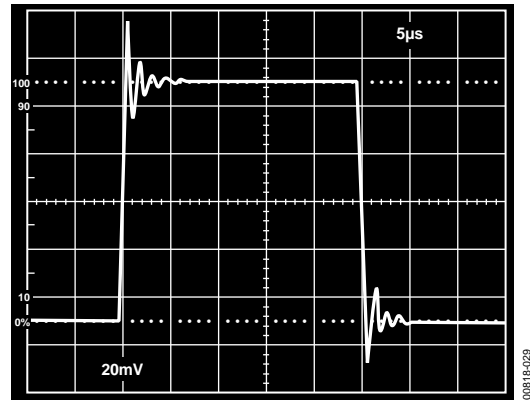


Figure 29. Unity Gain Follower Small Signal Pulse Response $R_F = 0 \Omega$, $C_L = 1000 \text{ pF}$

00818-029

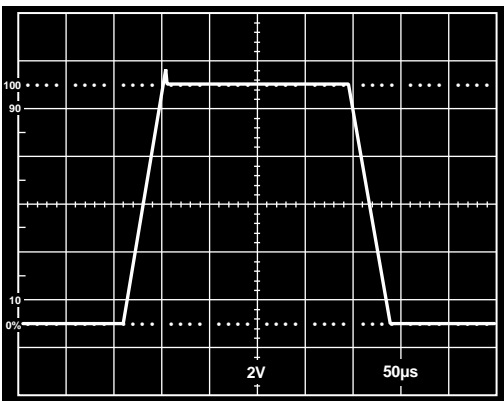


Figure 27. Unity Gain Follower Large Signal Pulse Response $R_F = 10 \text{ k}\Omega$, $C_L = 1000 \text{ pF}$

00818-027

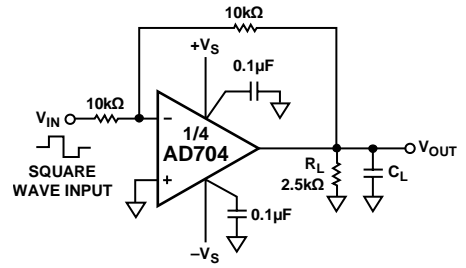
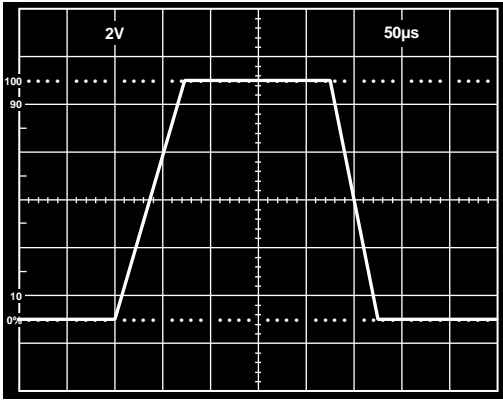


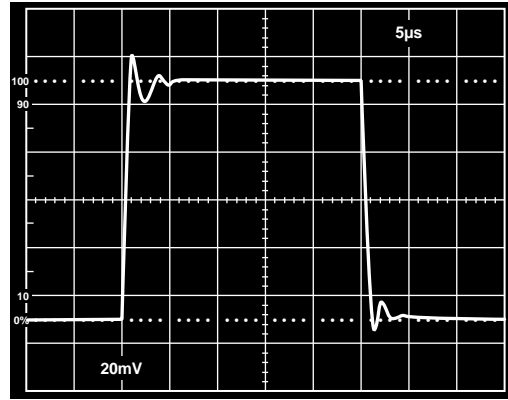
Figure 30. Unity Gain Inverter Connection

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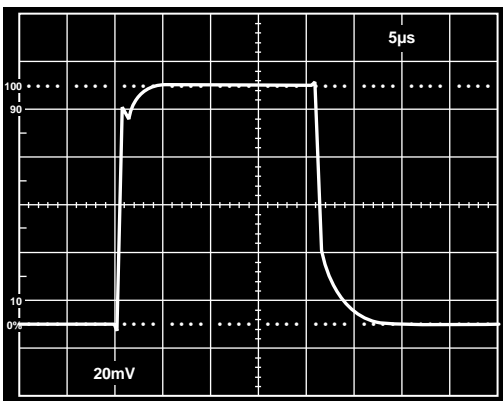
00816-031

Figure 31. Unity Gain Inverter Large Signal Pulse Response, $C_L = 1000\text{ pF}$



00816-033

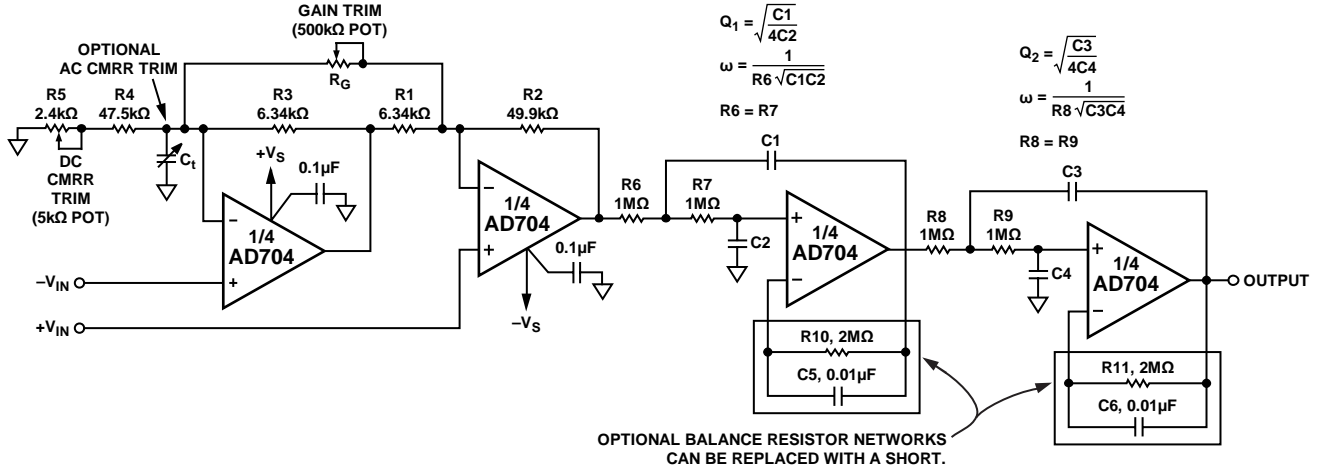
Figure 33. Unity Gain Inverter Small Signal Pulse Response, $C_L = 1000\text{ pF}$



00816-032

Figure 32. Unity Gain Inverter Small Signal Pulse Response, $C_L = 100\text{ pF}$

THEORY OF OPERATION



- NOTES**
- INSTRUMENTATION AMPLIFIER GAIN = $1 + \frac{R_2}{R_1} + \frac{2R_2}{R_G}$ (FOR $R_1 = R_3$, $R_2 = R_4 + R_5$).
 - CAPACITORS C2 AND C4 ARE SOUTHERN ELECTRONICS MPCC, POLYCARBONATE, ±5%, 50V.
 - ALL RESISTORS METAL FILM, 1%.

Figure 34. Gain-of-10 Instrumentation Amplifier with Post Filtering

The instrumentation amplifier with post filtering (see Figure 34) combines two applications that benefit greatly from the AD704. This circuit achieves low power and dc precision over temperature with a minimum of components.

The instrumentation amplifier circuit offers many performance benefits, including BiFET level input bias currents, low input offset voltage drift, and only 1.2 mA quiescent current. It operates for gains that are $G \geq 2$ and, at lower gains, it benefits from no output amplifier offset and no noise contribution as encountered in a 3-op-amp design. Good low frequency CMRR is achieved even without the optional ac CMRR trim (see Figure 35). Table 4 provides resistance values for three common circuit gains. For other gains, use the following equations:

$$R_2 = R_4 + R_5 = 49.9 \text{ k}\Omega$$

$$R_1 = R_3 = \frac{49.9 \text{ k}\Omega}{0.9 G - 1}$$

$$\text{Max Value of } R_G = \frac{99.8 \text{ k}\Omega}{0.06 G}$$

$$C_i \approx \frac{1}{2\pi (R_3) 5 \times 10^5}$$

Table 4. Resistance Values for Various Gains

Circuit Gain (G)	R1 and R3	RG (Max Value of Trim Potentiometer)	Bandwidth (-3 dB), Hz
10	6.34 kΩ	166 kΩ	50 k
100	526 Ω	16.6 kΩ	5k
1000	56.2 Ω	1.66 kΩ	0.5 k

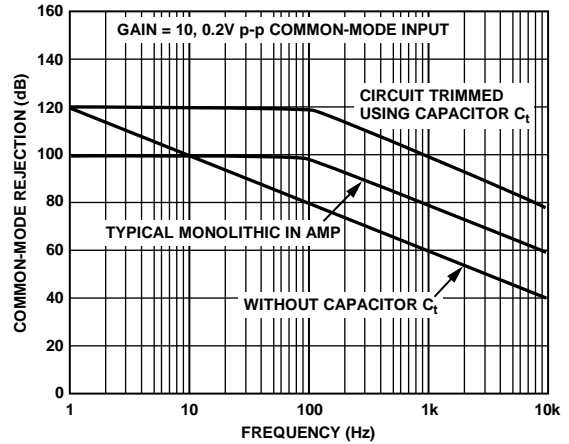


Figure 35. Common-Mode Rejection vs. Frequency with and Without Capacitor C_i

AD704

The 1 Hz, four-pole active filter offers dc precision with a minimum of components and cost. The low current noise, I_{OS} , and I_B allow the use of 1 M Ω resistors without sacrificing the 1 μ V/ $^{\circ}$ C drift of the AD704. This means that lower capacitor values can be used, reducing cost and space. Furthermore, because the AD704's I_B is as low as its I_{OS} , over most of the MIL temperature range, most applications do not require the use of the normal balancing resistor (with its stability capacitor). Adding the optional balancing resistor enhances performance at high temperatures, as shown in Figure 36. Table 5 gives capacitor values for several common low pass responses.

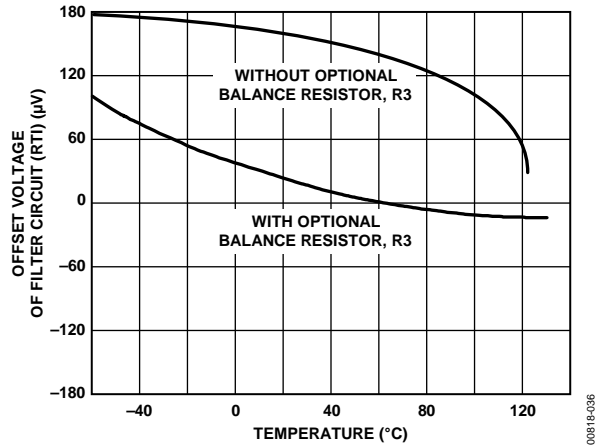


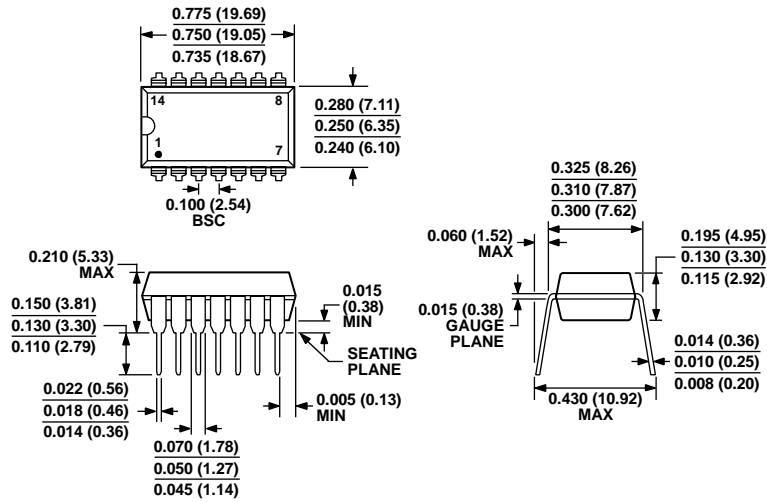
Figure 36. V_{OS} vs. Temperature Performance of the 1 Hz Filter Circuit

Table 5. 1 Hz, Four-Pole Low-Pass Filter Recommended Component Values¹

Desired Low Pass Response	Section 1 Frequency (Hz)	Q	Section 2 Frequency (Hz)	Q	C1 (μ F)	C2 (μ F)	C3 (μ F)	C4 (μ F)
Bessel	1.43	0.522	1.60	0.806	0.116	0.107	0.160	0.0616
Butterworth	1.00	0.541	1.00	1.31	0.172	0.147	0.416	0.0609
0.1 dB Chebychev	0.648	0.619	0.948	2.18	0.304	0.198	0.733	0.0385
0.2 dB Chebychev	0.603	0.646	0.941	2.44	0.341	0.204	0.823	0.0347
0.5 dB Chebychev	0.540	0.705	0.932	2.94	0.416	0.209	1.00	0.0290
1.0 dB Chebychev	0.492	0.785	0.925	3.56	0.508	0.206	1.23	0.0242

¹ Specified values are for a -3 dB point of 1.0 Hz. For other frequencies, simply scale the C1 through C4 capacitors directly; that is, for a 3 Hz Bessel response, C1 = 0.0387 μ F, C2 = 0.0357 μ F, C3 = 0.0533 μ F, and C4 = 0.0205 μ F.

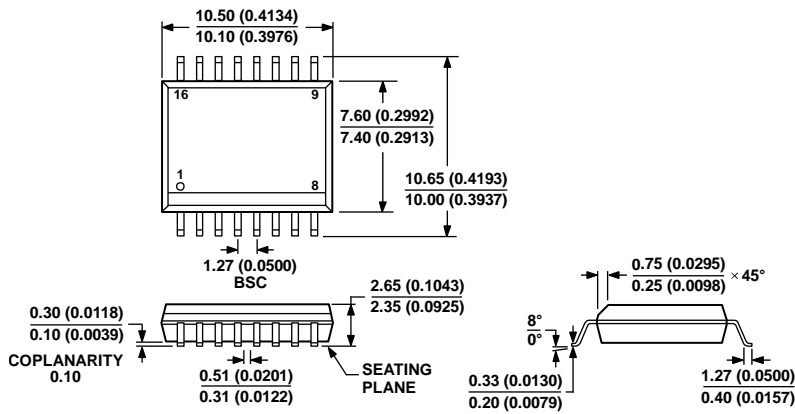
OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-001
 CONTROLLING DIMENSIONS ARE IN INCHES; MILLIMETER DIMENSIONS (IN PARENTHESES) ARE ROUNDED-OFF INCH EQUIVALENTS FOR REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN. CORNER LEADS MAY BE CONFIGURED AS WHOLE OR HALF LEADS.

Figure 37. 14-Lead Plastic Dual In-Line Package [PDIP] Narrow Body (N-14)
 Dimensions shown in inches and (millimeters)

070606-A

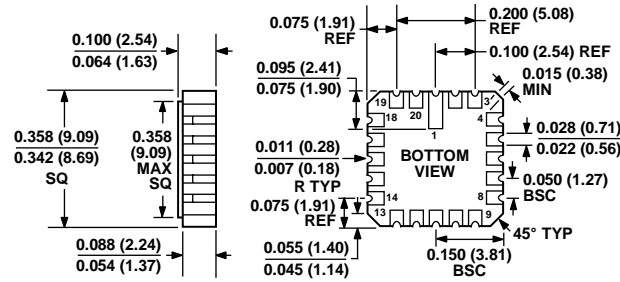


COMPLIANT TO JEDEC STANDARDS MS-013-AA
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

Figure 38. 16-Lead Standard Small Outline Package [SOIC_W] Wide Body (RW-16)
 Dimensions shown in millimeters and (inches)

0312707-B

AD704



CONTROLLING DIMENSIONS ARE IN INCHES; MILLIMETER DIMENSIONS (IN PARENTHESES) ARE ROUNDED-OFF INCH EQUIVALENTS FOR REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

0021 06-A

Figure 39. 20-Terminal Ceramic Leadless Chip Carrier [LCC] (E-20-1)

Dimensions shown in inches and (millimeters)

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option
AD704AR-16	-40°C to +85°C	16-Lead SOIC_W	RW -16
AD704AR-16-REEL	-40°C to +85°C	16-Lead SOIC_W	RW -16
AD704ARZ-16	-40°C to +85°C	16-Lead SOIC_W	RW -16
AD704ARZ-16-REEL	-40°C to +85°C	16-Lead SOIC_W	RW -16
AD704JN	0°C to 70°C	14-Lead PDIP	N-14
AD704JNZ	0°C to 70°C	14-Lead PDIP	N-14
AD704JR-16	0°C to 70°C	16-Lead SOIC_W	RW -16
AD704JR-16-REEL	0°C to 70°C	16-Lead SOIC_W	RW -16
AD704JRZ-16	0°C to 70°C	16-Lead SOIC_W	RW -16
AD704JRZ-16-REEL	0°C to 70°C	16-Lead SOIC_W	RW -16
AD704SE/883B	-55°C to +125°C	20-Terminal LCC	E-20-1

¹ Z = RoHS Compliant Part.

NOTES

AD704

NOTES