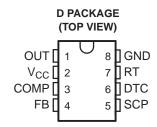


PULSE-WIDTH-MODULATION CONTROL CIRCUITS

FEATURES

- Qualified for Automotive Applications
- Complete PWM Power Control
- 3.6-V to 40-V Operation
- Internal Undervoltage-Lockout Circuit
- Internal Short-Circuit Protection
- Oscillator Frequency: 20 kHz to 500 kHz
- Variable Dead Time Provides Control Over Total Range
- ±3% Tolerance on Reference Voltage
- Available in Q-Temperature Automotive
 - High-Reliability Automotive Applications
 - Configuration Control / Print Support
 - Qualification to Automotive Standards



DESCRIPTION

The TL5001A incorporates on a single monolithic chip all the functions required for a pulse-width-modulation (PWM) control circuit. Designed primarily for power-supply control, the TL5001A contains an error amplifier, a regulator, an oscillator, a PWM comparator with a dead-time-control input, undervoltage lockout (UVLO), short-circuit protection (SCP), and an open-collector output transistor. The TL5001A has a typical reference voltage tolerance of ±3%.

The error-amplifier common-mode voltage ranges from 0 V to 1.5 V. The noninverting input of the error amplifier is connected to a 1-V reference. Dead-time control (DTC) can be set to provide 0% to 100% dead time by connecting an external resistor between DTC and GND. The oscillator frequency is set by terminating RT with an external resistor to GND. During low V_{CC} conditions, the UVLO circuit turns the output off until V_{CC} recovers to its normal operating range.

The TL5001A is characterized for operation from -40°C to 125°C.

AVAILABLE OPTIONS(1)

-	PACKAGED DEVICES ⁽²⁾
T _A	SMALL OUTLINE (D) ⁽³⁾
-40°C to 125°C	TL5001AQDRQ1

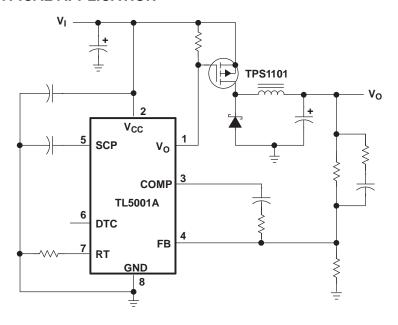
- (1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.
- (2) Package drawings, thermal data, and symbolization are available at www.ti.com/packaging.
- (3) The D package is available taped and reeled. Add the suffix R to the device type (e.g., TL5001ADR).



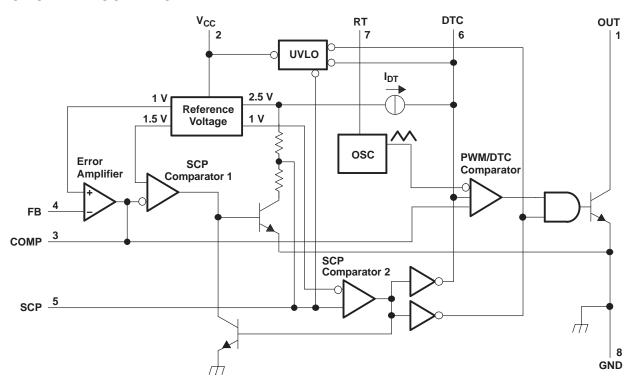
Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



SCHEMATIC FOR TYPICAL APPLICATION



FUNCTIONAL BLOCK DIAGRAM





DETAILED DESCRIPTION

VOLTAGE REFERENCE

A 2.5-V regulator operating from V_{CC} is used to power the internal circuitry of the TL5001A and as a reference for the error amplifier and SCP circuits. A resistive divider provides a 1-V reference for the error amplifier noninverting input which typically is within 2% of nominal over the operating temperature range.

ERROR AMPLIFIER

The error amplifier compares a sample of the dc-to-dc converter output voltage to the 1-V reference and generates an error signal for the PWM comparator. The dc-to-dc converter output voltage is set by selecting the error-amplifier gain (see Figure 1), using the following expression:

$$V_0 = (1 + R1/R2) (1 V)$$

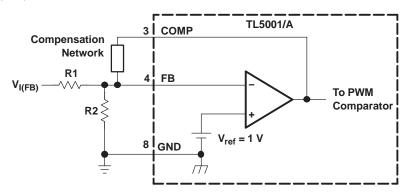


Figure 1. Error-Amplifier Gain Setting

The error-amplifier output is brought out as COMP for use in compensating the dc-to-dc converter control loop for stability. Because the amplifier can only source 45 μ A, the total dc-load resistance should be 100 k Ω or more.

OSCILLATOR/PWM

The oscillator frequency (f_{osc}) can be set between 20 kHz and 500 kHz by connecting a resistor between RT and GND. Acceptable resistor values range from 15 k Ω to 250 k Ω . The oscillator frequency can be determined by using the graph shown in Figure 5.

The oscillator output is a triangular wave with a minimum value of approximately 0.7 V and a maximum value of approximately 1.3 V. The PWM comparator compares the error-amplifier output voltage and the DTC input voltage to the triangular wave and turns the output transistor off whenever the triangular wave is greater than the lesser of the two inputs.

DEAD-TIME CONTROL (DTC)

DTC provides a means of limiting the output-switch duty cycle to a value less than 100%, which is critical for boost and flyback converters. A current source generates a reference current (I_{DT}) at DTC that is nominally equal to the current at the oscillator timing terminal (RT). Connecting a resistor between DTC and GND generates a dead-time reference voltage (V_{DT}), which the PWM/DTC comparator compares to the oscillator triangle wave as described in the previous section. Nominally, the maximum duty cycle is 0% when VDT is 0.7 V or less and 100% when V_{DT} is 1.3 V or greater. Because the triangle wave amplitude is a function of frequency and the source impedance of RT is relatively high (1250 Ω), choosing R_{DT} for a specific maximum duty cycle (D) is accomplished using the following equation and the voltage limits for the frequency in question as found in Figure 11 (V_{osc} max and V_{osc} min are the maximum and minimum oscillator levels):

$$\frac{R_{DT} = (R_t + 1250) \left[D(V_{osc} max - V_{osc} min) + V_{osc} min \right]}{V_{RT}} \quad ; \quad V_{RT} = 1 \text{ V}$$
(1)

Where

 \boldsymbol{R}_{DT} and \boldsymbol{R}_{t} are in $\Omega,$ D is in decimal



Soft start can be implemented by paralleling the DTC resistor with a capacitor (C_{DT}) as shown in Figure 2. During soft start, the voltage at DTC is derived by the following equation:

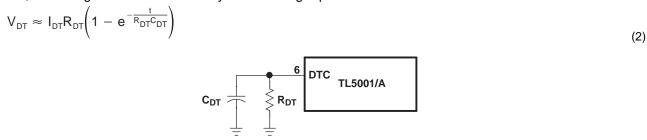


Figure 2. Soft-Start Circuit

If the dc-to-dc converter must be in regulation within a specified period of time, the time constant ($R_{DT}C_{DT}$) should be $t_0/3$ to $t_0/5$. The TL5001A remains off until $V_{DT}\approx 0.7$ V, the minimum ramp value. C_{DT} is discharged every time UVLO or SCP becomes active.

UNDERVOLTAGE-LOCKOUT (UVLO) PROTECTION

The undervoltage-lockout circuit turns the output transistor off and resets the SCP latch whenever the supply voltage drops too low (approximately 3 V at 25°C) for proper operation. A hysteresis voltage of 200 mV eliminates false triggering on noise and chattering.

SHORT-CIRCUIT PROTECTION (SCP)

The TL5001A includes short-circuit protection (see Figure 3), which turns the power switch off to prevent damage when the converter output is shorted. When activated, the SCP prevents the switch from being turned on until the internal latching circuit is reset. The circuit is reset by reducing the input voltage until UVLO becomes active or until the SCP terminal is pulled to ground externally.

When a short circuit occurs, the error-amplifier output at COMP rises to increase the power-switch duty cycle in an attempt to maintain the output voltage. SCP comparator 1 starts an RC timing circuit when COMP exceeds 1.5 V. If the short is removed and the error-amplifier output drops below 1.5 V before time out, normal converter operation continues. If the fault is still present at the end of the time-out period, the timer sets the latching circuit and turns off the TL5001/A output transistor.

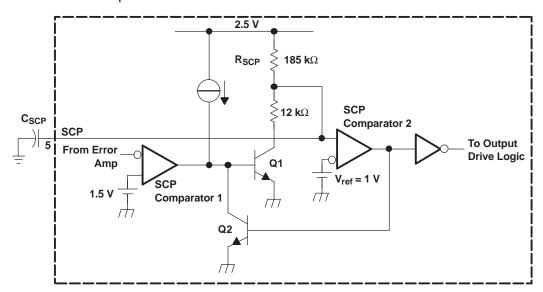


Figure 3. SCP Circuit



The timer operates by charging an external capacitor (C_{SCP}) connected between the SCP terminal and ground, towards 2.5 V through a 185-k Ω resistor (R_{SCP}). The circuit begins charging from an initial voltage of approximately 185 mV and times out when the capacitor voltage reaches 1 V. The output of SCP comparator 2 then goes high, turns on Q2, and latches the timer circuit. The expression for setting the SCP time period is derived from Equation 3:

$$V_{SCP} = (2.5 - 0.185) (1 - e^{-t/\tau}) + 0.185$$
 (3)

Where

 $\tau = R_{SCP}C_{SCP}$

The end of the time-out period (t_{SCP}) occurs when $V_{SCP} = 1$ V. Solving for C_{SCP} yields Equation 4:

$$C_{SCP} = 12.46 \times t_{SCP} \tag{4}$$

Where

t is in seconds, C is in μF

 t_{SCP} must be much longer (generally 10 to 15 times) than the converter start-up period, or the converter will not start.

OUTPUT TRANSISTOR

The output of the TL5001A is an open-collector transistor with a maximum collector current rating of 21 mA and a voltage rating of 51 V. The output is turned on under the following conditions: the oscillator triangle wave is lower than both the DTC voltage and the error-amplifier output voltage, the UVLO circuit is inactive, and the short-circuit protection circuit is inactive.

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ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) (1)

V _{CC}	Supply voltage ⁽²⁾	41 V
V _{I(FB)}	Amplifier input voltage	20 V
Vo	Output voltage, OUT	51 V
Io	Output current, OUT	21 mA
I _{O(peak)}	Output peak current, OUT	100 mA
	Continuous total power dissipation	See Dissipation Rating
T _A	Operating ambient temperature range, TL5001AQDRQ1	−40°C to 125°C
T _{stg}	Storage temperature range	−65°C to 150°C
	Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

⁽¹⁾ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATINGS

PACKA	E T _A ≤ 25°C	DERATING FACTOR	T _A = 70°C	T _A = 85°C	T _A = 125°C
	POWER RATING	ABOVE T _A = 25°C	POWER RATING	POWER RATING	POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW

RECOMMENDED OPERATING CONDITIONS

			MIN	MAX	UNIT
V _{CC}	Supply voltage	3.6	40	V	
V _{I(FB)}	Amplifier input voltage	0	1.5	V	
Vo	Output voltage, OUT			50	V
Io	Output current, OUT		20	mA	
	COMP source current		45	μΑ	
	COMP dc load resistance	100		kΩ	
R _t	Oscillator timing resistor	15	250	kΩ	
fosc	Oscillator frequency	20	500	kHz	
T _A	Operating ambient temperature	TL5001AQDRQ1	-40	125	°C

Product Folder Link(s): TL5001A-Q1

John Documentation Feedback

⁽²⁾ All voltage values are with respect to network ground terminal.



ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 100 kHz (unless otherwise noted)

DADAMETED	PARAMETER TEST CONDITIONS					UNIT
PARAMETER	IES	or CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
REFERENCE						
Output valta an	T _A = 25°C	COMP assessed to ED	0.97	1	1.03	V
Output voltage	$T_A = MIN \text{ to } MAX$	COMP connected to FB	0.94	0.98	1.06	V
Input regulation	$T_A = MIN \text{ to MAX},$	VCC = 3.6 V to 40 V		2	12.5	mV
Output voltage change with temperature	$T_A = MIN \text{ to } MAX$		-6% ⁽²⁾	2%	6% ⁽²⁾	
UNDERVOLTAGE LOCKOUT						
Henry therebold will an	T _A = MIN, 25°C			3		
Upper threshold voltage	$T_A = MAX$			2.55		V
Lawrenth and hald walks are	T _A = MIN, 25°C			2.8		
Lower threshold voltage	$T_A = MAX$			2.0		V
Hysteresis	$T_A = MIN \text{ to } MAX$		80	200		mV
	T _A = MIN, 25°C		2.1	2.55		.,
Reset threshold voltage	$T_A = MAX$		0.35	0.63		V
SHORT-CIRCUIT PROTECTION					1.	
200 11 11 11	T _A = MIN, 25°C		0.97	1	1.03	
SCP threshold voltage	$T_A = MAX$		0.94	0.98	1.06	V
SCP voltage, latched	$T_A = MIN \text{ to } MAX$	No pullup	140	185	230	mV
SCP voltage, UVLO standby	$T_A = MIN \text{ to } MAX$	No pullup		60	120	mV
Equivalent timing resistance	$T_A = MIN \text{ to } MAX$			185		kΩ
SCP comparator 1 threshold voltage	$T_A = MIN \text{ to } MAX$			1.5		V
OSCILLATOR					1	
Frequency	$T_A = MIN \text{ to } MAX$	$R_t = 100 \text{ k}\Omega$		100		kHz
Standard deviation of frequency	$T_A = MIN \text{ to } MAX$			2		kHz
Frequency change with voltage	$T_A = MIN \text{ to } MAX$	V _{CC} = 3.6 V to 40 V		1		kHz
	T 14111 / 1411	Q suffix	-9 ⁽²⁾	5	9(2)	
Frequency change with temperature	$T_A = MIN \text{ to } MAX$	M suffix	-9 ⁽²⁾	5	9(2)	kHz
Voltage at RT	T _A = MIN to MAX			1		V
DEAD-TIME CONTROL		+			1.	
Output (source) current	$T_A = MIN \text{ to } MAX$	V _(DT) = 1.5 V	0.9 x I _{RT} ⁽³⁾		1.1 x I _{RT} ⁽³⁾	μΑ
		Duty cycle 0%	0.5	0.7		
	T _A = 25°C	Duty cycle 100%		1.3	1.5	\ /
Input threshold voltage		Duty cycle 0%	0.4	0.7		V
	$T_A = MIN \text{ to } MAX$	Duty cycle 100%		1.3	1.7	

⁽¹⁾ All typical values are at T_A = 25°C.
(2) Not production tested.
(3) Output source current at RT



ELECTRICAL CHARACTERISTICS

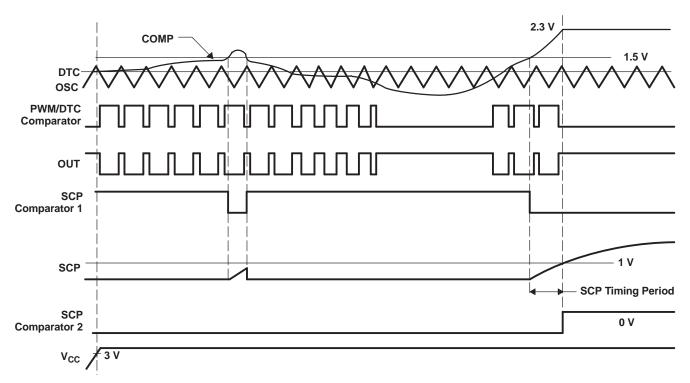
over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 100 kHz (unless otherwise noted)

DADAMETE	n		TL5001AQ			LINUT	
PARAMETE	K	16	ST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
ERROR AMPLIFIER		<u>'</u>		1		'	
Input bias current		$T_A = MIN \text{ to } MAX$			-160	-500	nA
Output valtage ewing	Positive	T MINI to MAY		1.5	2.3		V
Output voltage swing	Negative	$T_A = MIN \text{ to MAX}$			0.3	0.4	V
Open-loop voltage amplifica	ntion	T _A = MIN to MAX	AX 80				dB
Unity gain bandwidth		$T_A = MIN \text{ to } MAX$			1.5		MHz
Output (sink) current		$T_A = MIN \text{ to } MAX$	V _{I(FB)} = 1.2 V, COMP = 1 V	100	600		μΑ
2		$T_A = MIN, 25^{\circ}C$		-45	-70		
Output (source) current		$T_A = MAX$	$V_{I(FB)} = 0 \text{ V, COMP} = 1 \text{ V}$	-30	-45		μΑ
OUTPUT							
Output saturation voltage		$T_A = MIN \text{ to } MAX$	I _O = 10 mA		1.5	2	V
Off state summent		$T_A = MIN \text{ to } MAX$	V _O = 50 V, V _{CC} = 0			10	^
Off-state current			V _O = 50 V			10	μΑ
Short-circuit output current		T _A = MIN to MAX	V _O = 6 V		40		mA
TOTAL DEVICE						,	
Standby supply current	Off state	$T_A = MIN \text{ to } MAX$			1	1.5	mA
Average supply current		$T_A = MIN \text{ to } MAX$	$R_L = 100 \text{ k}\Omega$		1.4	2.1	mA

⁽¹⁾ All typical values are at $T_A = 25$ °C.



PARAMETER MEASUREMENT INFORMATION

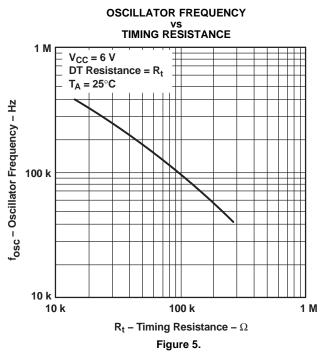


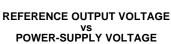
A. The waveforms show timing characteristics for an intermittent short circuit and a longer short circuit that is sufficient to activate SCP.

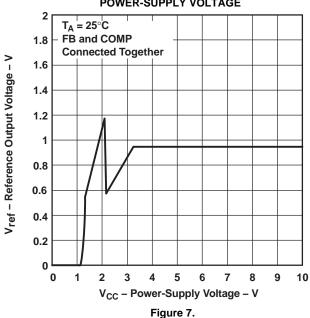
Figure 4. PWM Timing Diagram



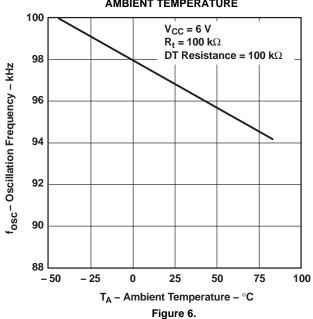
TYPICAL CHARACTERISTICS



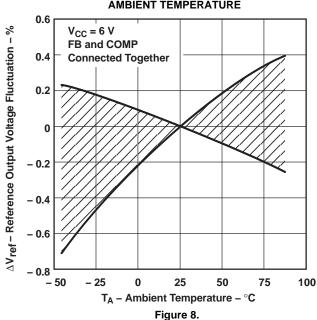




OSCILLATION FREQUENCY VS AMBIENT TEMPERATURE

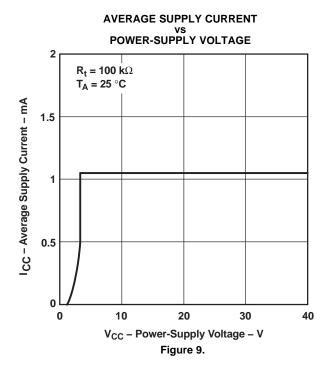


REFERENCE OUTPUT VOLTAGE FLUCTUATION vs AMBIENT TEMPERATURE





TYPICAL CHARACTERISTICS (continued)



PWM TRIANGLE WAVE AMPLITUDE VOLTAGE VS OSCILLATOR FREQUENCY

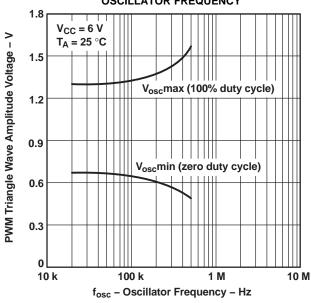
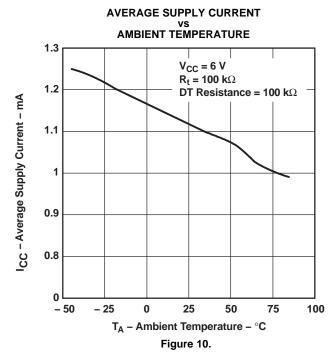
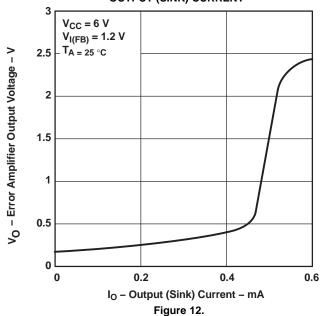


Figure 11.



ERROR AMPLIFIER OUTPUT VOLTAGE vs OUTPUT (SINK) CURRENT

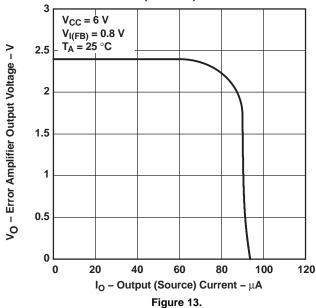


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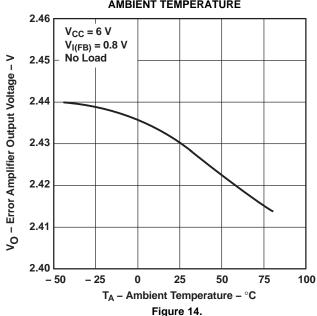
TYPICAL CHARACTERISTICS (continued)

ERROR AMPLIFIER OUTPUT VOLTAGE

OUTPUT (SOURCE) CURRENT

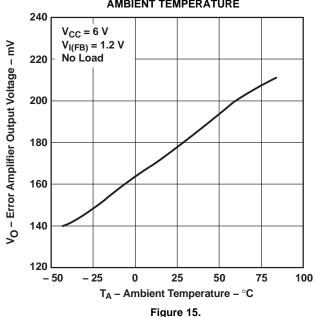


ERROR AMPLIFIER OUTPUT VOLTAGE vs AMBIENT TEMPERATURE



ERROR AMPLIFIER OUTPUT VOLTAGE

vs AMBIENT TEMPERATURE



ERROR AMPLIFIER OPEN-LOOP GAIN AND PHASE SHIFT

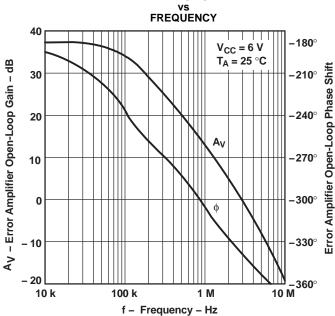
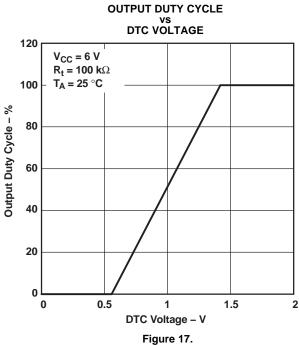
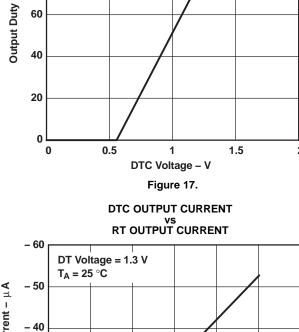


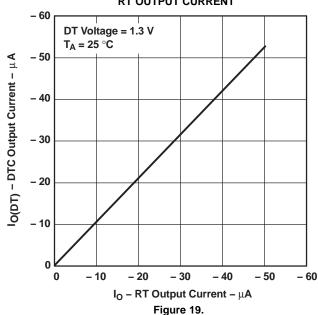
Figure 16.

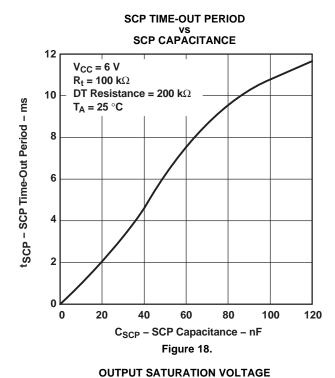


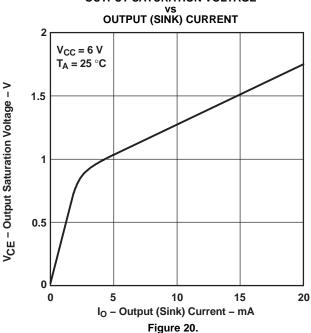
TYPICAL CHARACTERISTICS (continued)





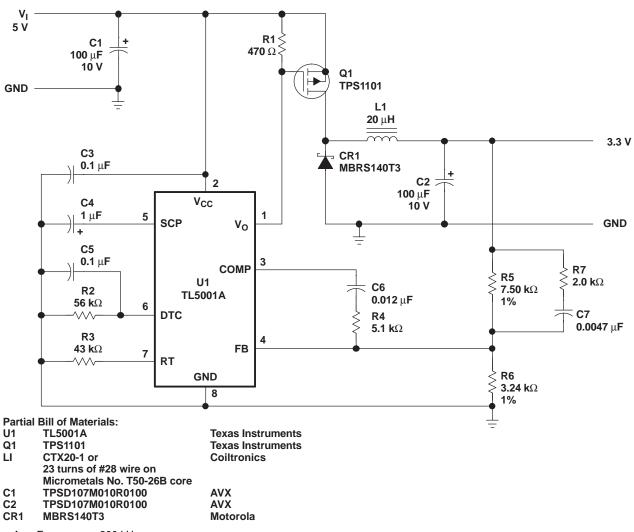








APPLICATION INFORMATION



- A. Frequency = 200 kHz
- B. Duty cycle = 90% max
- C. Soft-start time constant (TC) = 5.6 ms
- D. SCP TC = 70 ms

Figure 21. Step Down Converter





11-Apr-2013

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
TL5001AQDRG4Q1	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	501AQ1	Samples
TL5001AQDRQ1	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	501AQ1	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

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OTHER QUALIFIED VERSIONS OF TL5001A-Q1:

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.



PACKAGE OPTION ADDENDUM

11-Apr-2013

Military: TL5001AM

NOTE: Qualified Version Definitions:

• Catalog - TI's standard catalog product

• Military - QML certified for Military and Defense Applications

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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