

16-Bit Low-Power Stereo Audio ADC With Microphone Bias and Microphone Amplifier

FEATURES

- Analog Front End:
 - Stereo Single End Input With MUX
 - Mono Differential Input
 - Stereo Programmable Gain Amplifier
 - Microphone Boost Amplifier and Bias
- Analog Performances Dynamic Range: 90 dB
- Power-Supply Voltage
 - 1.71 V to 3.6 V for Digital I/O Section
 - 1.71 V to 3.6 V for Digital Core Section
 - 2.4 V to 3.6 V for Analog Section
- Low Power Dissipation:
 - 13 mW in Record, 1.8/2.4 V, 48 kHz, Stereo
 - 5.3 mW in Record, 1.8/2.4 V, 8 kHz, Mono
 - 3.3 μW in All Power Down
- Sampling Frequency: 5 kHz to 50 kHz
- Auto Level Control for Recording
- Operation by Single Clock Input Without PLL
- System Clock: Common Audio Clock (256 f_s/384 f_s), 12/24, 13/26, 13.5/27, 19.2/38.4, 19.68/39.36 MHz
- 2 (I²C) or 3 (SPI) Wire Serial Control
- Programmable Function by Register Control:
 - Digital Soft Mute
 - Power Up/Down Control for Each Module
 - 30-dB to -12-dB Gain for Analog Inputs
 - 0/12/20-dB Boost for Microphone Input
 - Parameter Settings for ALC
 - Three-Band Tone Control and 3D Sound
 - High-Pass Filter and Two-Stage Notch Filter
- Pop Noise Reduction Circuit

- Package:
 - 24-QFN (4 mm imes 5 mm)
 - 24-DSBGA (2.49 mm × 3.49 mm)
- Operation Temperature Range: -40°C to 85°C

APPLICATIONS

- Portable Audio Player, Cellular Phone
- Video Camcorder, Movie Digital Still Camera
- PMP/DMB, Voice Recorder

DESCRIPTION

The PCM1870 is a low-power stereo ADC designed for portable digital audio applications, with line-input amplifier, microphone amplifier, boost bias. programmable gain control, sound effects, and auto level control (ALC). It is available in 24-QFN (4-mm \times 5-mm) and 24-DSBGA (2.49-mm \times 3.49-mm) packages to save footprint. The PCM1870 accepts right-justified, left-justified, l²S, and DSP formats, providing easy interfacing to audio DSP and encoder chips. Sampling rates up to 50 kHz are supported. The user-programmable functions are accessible through a 2- or 3-wire serial control port.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PCM1870

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		PCM1870	UNIT
V_{DD}, V_{IO}, V_{CC}	Supply voltage	-0.3 to 4	V
	Ground voltage differences: DGND, AGND, PGND	±0.1	V
	Input voltage	-0.3 to 4	V
	Input current (any pins except supplies)	±10	mA
	Ambient temperature under bias	-40 to 110	°C
	Storage temperature	-55 to 150	°C
	Junction temperature	150	°C
	Lead temperature (soldering)	260 / 5	°C / s
	Package temperature (reflow, peak)	260	°C

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicted under recommended operating conditions is not impled. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

			MIN	NOM	MAX	UNIT
V _{CC}	Analog supply voltage		2.4	3.3	3.6	V
V _{DD} , V _{IO}	Digital supply voltage		1.71	3.3	3.6	V
	Digital input logic family			CMOS		
		SCKI system clock	3.072		18.432	MHz
	Digital input clock frequency	LRCK sampling clock	8		48	kHz
	Digital output load capacitance				10	pF
T _A	Operating free-air temperature		-40		85	°C



ELECTRICAL CHARACTERISTICS

All specifications at $T_A = 25^{\circ}C$, $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3$ V, $f_S = 48$ kHz, system clock = 256 f_S , and 16-bit data, unless otherwise noted

			PCM1870RHF, PCM1870YZF				
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
	ОАТА						
Data For	rmat						
	Resolution			16		Bits	
	Audio data interface format		I ² S, left-, I	right-justifie	ed, DSP		
	Audio data bit length			16		Bits	
	Audio data format		MSB-firs	t, 2s-compl	ement		
f _S	Sampling frequency		5		50	kHz	
		V _{DD} < 2 V			27		
	System clock	V _{DD} > 2 V			40	MHz	
DIGITAL	. INPUT/OUTPUT						
	Logic family		CMC	S compatil	ole		
VIH			0.7 V _{IO}				
VIL	Input logic level				0.3 V _{IO}	VDC	
IIH		V _{IN} = 3.3 V			10		
IIL	Input logic current	V _{IN} = 0 V			-10	μA	
V _{OH}		$I_{OH} = -2 \text{ mA}$	0.75 V _{IO}				
V _{OL}	Output logic level	$I_{OL} = 2 \text{ mA}$			0.25 V _{IO}	VDC	
	PUT TO DIGITAL OUTPUT THROUGH	ADC (AIN1L/R, AIN2L/R AND PGINL/R-ALC	= OFF, PG	1 = PG2 =	-	= 0 dE	
Dynamic	c Performance	•					
				2.828		Vp-p	
	Full-scale input voltage	0 dB		1		Vrm	
	Dynamic range	EIAJ, A-weighted		90		dB	
SNR	Signal-to-noise ratio	EIAJ, A-weighted	83	90		dB	
	Channel separation			87		dB	
THD+N	Total harmonic distortion + noise	–1 dB		0.009%	0.017%		
Analog I	Input						
	Center voltage			0.5 V _{CC}		V	
	¥	AIN1L, AIN1R, AIN2L, and AIN2R	10	20			
	Input impedance	PGINL and PGINR, PG3 = PG4 = -12 dB	70	142		kΩ	
		PGINL and PGINR, PG3 = PG4 = 30 dB	4.7	9.5			
ANALOO	G OUTPUTS (AOL AND AOR)	1					
	Center voltage			0.5 V _{CC}		V	
			10	00		kΩ	
	Load resistance						
	Load resistance				20	pF	
MICROP	Load capacitance	2 = PG3 = PG4 = 0 dB			20	pF	
MICROP	Load capacitance PHONE BIAS—ALC = OFF, PG1 = PG2	2 = PG3 = PG4 = 0 dB		0.75 Vcc	20	pF V	
MICROP	Load capacitance	2 = PG3 = PG4 = 0 dB		0.75 V _{CC}	20	pF V mA	



ELECTRICAL CHARACTERISTICS (continued)

All specifications at $T_A = 25^{\circ}$ C, $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3$ V, $f_S = 48$ kHz, system clock = 256 f_S , and 16-bit data, unless otherwise noted

		TEAT CONDITIONS	PCM1870RH	IF, PCM18	370YZF	
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
FILTER	R CHARACTERISTICS		U			
Decima	ation Filter for ADC					
	Pass band				0.408 f _S	
	Stop band		0.591 f _S			
	Pass-band ripple				±0.02	dB
	Stop-band attenuation	f < 3.268 f _S	-60			dB
	Group delay			17/f _S		S
High-P	ass Filter for ADC				i	
		–3 dB		3.74		
	Frequency response ($f_c = 4 Hz$)	–0.5 dB		10.66	0.408 f _S	Hz
		–0.1 dB		24.20		
		–3 dB		118.77		
	Frequency response (f _c = 120 Hz)	–0.5 dB		321.75		Hz
	$\frac{1}{10000000000000000000000000000000000$	–0.1 dB		605.52		
		–3 dB		235.68		
	Frequency response (f _c = 240 Hz)	–0.5 dB		609.95		Hz
		–0.1 dB		2601.2		
POWE	R SUPPLY AND SUPPLY CURRENT				i	
V _{IO}		V _{IO}	1.71	3.3	3.6	
V _{DD}	Voltage range	V _{DD}	1.71	3.3	3.6	VDC
V _{CC}		V _{cc}	2.4	3.3	3.6	
	Curratu aurrant	BPZ input, all active, no load		8	12	mA
	Supply current	All inputs are held static.		1	10	μA
	Dower dissipation	BPZ input		26.4	39.6	mW
	Power dissipation	All inputs are held static.		3.3	33	μW
TEMPE	RATURE CONDITION				I.	
	Operation temperature		-40		85	°C
θ_{JA}	Thermal resistance			35		°C/W

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PIN ASSIGNMENTS

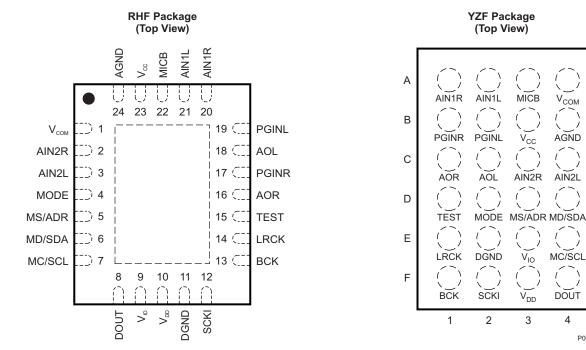
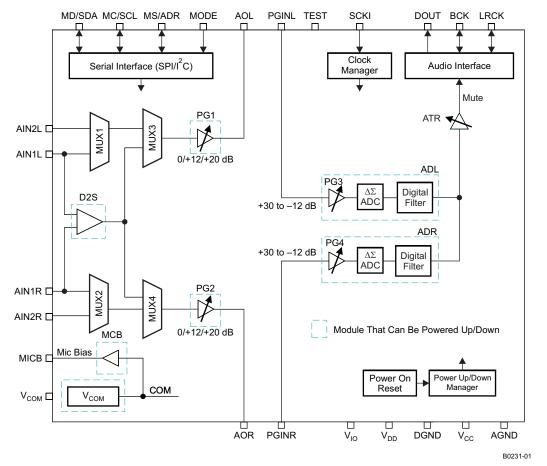


Table 1. TERMINAL FUNCTIONS

TERMINAL					
NAME	N	0.	I/O	DESCRIPTION	
NAME	RHF	YZF			
AGND	24	B4	-	Ground for analog	
AIN1L	21	A2	I	Analog input 1 for L-channel	
AIN1R	20	A1	I	Analog input 1 for R-channel	
AIN2L	3	C4	I	Analog input 2 for L-channel	
AIN2R	2	C3	I	Analog input 2 for R-channel	
AOL	18	C2	0	Microphone amplifier output for L-channel	
AOR	16	C1	0	Microphone amplifier output for R-channel	
BCK	13	F1	I/O	Serial bit clock	
DGND	11	E2	-	Ground for digital	
DOUT	8	F4	0	Serial audio data output	
LRCK	14	E1	I/O	O Left- and right-channel clock	
MC/SCL	7	E4	I	Mode control clock for 3-wire / 2-wire interface	
MD/SDA	6	D4	I/O	Mode control data for 3-wire / 2-wire interface	
MICB	22	A3	0	Microphone bias source output	
MODE	4	D2	I	2- or 3-wire interface selection (LOW: SPI, HIGH: I ² C)	
MS/ADR	5	D3	I	Mode control select for 3-wire / 2-wire interface	
PGINL	19	B2	I	Analog input to gain amplifier for L-channel	
PGINR	17	B1	I	Analog input to gain amplifier for R-channel	
SCKI	12	F2	I	System clock	
TEST	15	D1	I	Test Pin. Should be connected to ground.	
V _{CC}	23	B3	-	Power supply for analog	
V _{COM}	1	A4	-	Common voltage for analog	
V _{DD}	10	F3	-	Power supply for digital core	
V _{IO}	9	E3	-	Power supply for digital I/O	

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FUNCTIONAL BLOCK DIAGRAM





TYPICAL PERFORMANCE CURVES

All specifications at $T_A = 25^{\circ}C$, $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3$ V, $f_S = 8$ kHz to 48 kHz, system clock = 256 f_S and 16-bit data, unless otherwise noted.

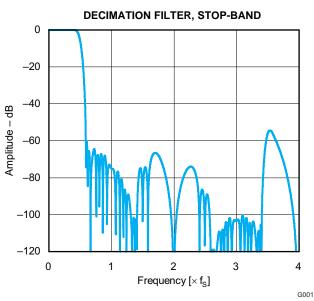
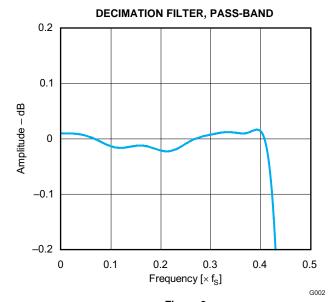


Figure 1.



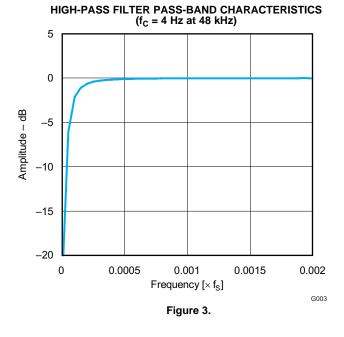
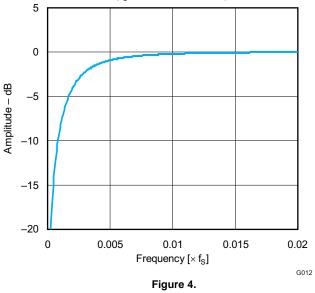


Figure 2.

HIGH-PASS FILTER PASS-BAND CHARACTERISTICS (f_c = 120 Hz at 48 kHz)



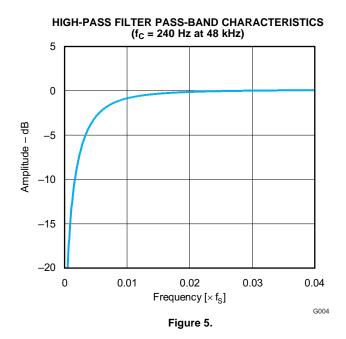
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TYPICAL PERFORMANCE CURVES (continued)

All specifications at $T_A = 25^{\circ}C$, $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3$ V, $f_S = 8$ kHz to 48 kHz, system clock = 256 f_S and 16-bit data, unless otherwise noted.



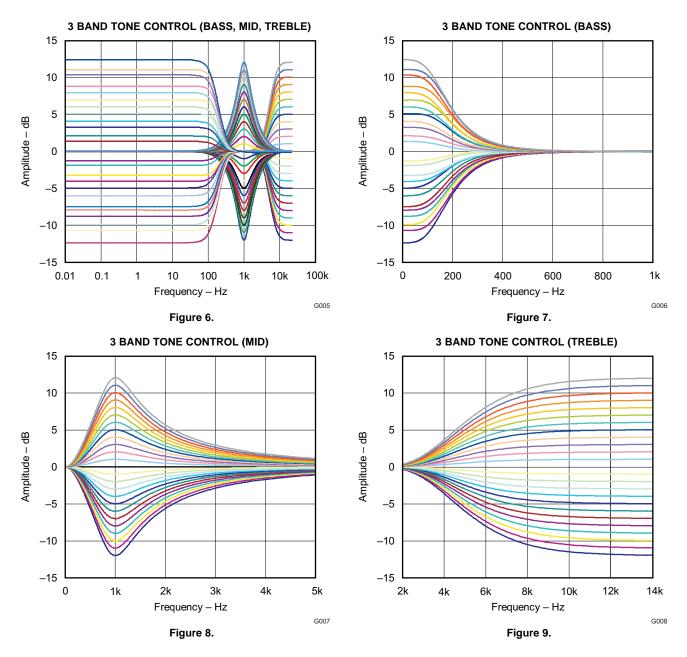
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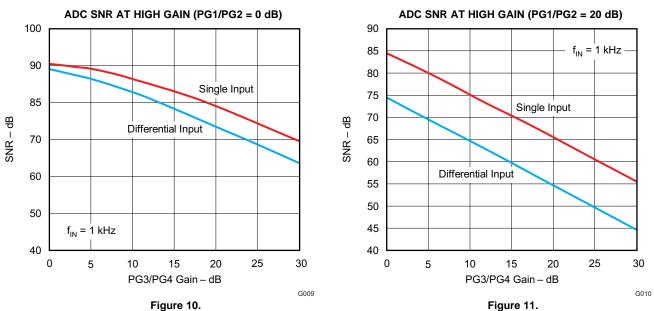
TYPICAL PERFORMANCE CURVES (continued)

All specifications at $T_A = 25^{\circ}$ C, $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3$ V, $f_S = 44.1$ kHz, system clock = 256 f_S and 16-bit data, unless otherwise noted.

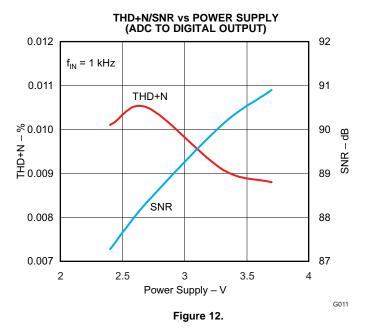


TYPICAL PERFORMANCE CURVES (continued)

All specifications at $T_A = 25^{\circ}C$, $V_{DD} = V_{IO} = V_{CC} = V_{PA} = 3.3$ V, $f_S = 48$ kHz, system clock = 256 f_S and 16-bit data, unless otherwise noted.









PCM1870 DESCRIPTION

Analog Input

The AIN1L, AIN1R, AIN2L and AIN2R pins can be used as microphone or line inputs with selectable 0- or 20-dB boost and 1-Vrms input. All analog inputs have high input impedance (20 k Ω), which is not changed by gain settings. One pair of inputs is selected by register 87 (AIL[1:0], AIR[1:0]). AIN1L and AIN1R can also be used as a monaural differential input.

Gain Setting for Analog Input

Analog signals can be adjusted from 30 dB to -12 dB in 1-dB steps after the 0-, 12- or 20-dB boost amplifier. Gain level can be set for each channel by register 79, 80 (ALV[5:0], ARV[5:0]).

A/D Converter

The ADC includes a multilevel delta-sigma modulator, aliasing filter, decimation filter, high-pass filter and notch filter and can accept 1 Vrms as full-scale input voltage. The decimation filter has a digital soft mute controlled by register 81 (RMUL, RMUR), and the high-pass and notch filters can be disabled by register 81 (HPF[1:0]) and registers 96 through 104 if they are not needed to cancel dc offset or avoid wind noise.

Common Voltage

Unbuffered common voltage. The V_{COM} pin is normally biased to 0.5 V_{CC} , and it provides common voltage to internal circuitry. Connecting a 4.7- μ F capacitor to this pin is recommended to optimize analog performance.

Microphone Bias

The MICB pin is a microphone bias source for an external microphone, which can provide 2 mA (typ) bias current.

Auto Level Control (ALC) for Recording

The sound when microphone recording should be adjusted to a suitable level without saturation. The digitally controlled auto level control (ALC) automatically expands small input signals and compresses large input signals while recording. Expansion level, compression level, attack time, and recovery time can be selected by register 83. See the bit descriptions of register 83 for detailed settings.

3D Sound

A 3D sound effect is provided by mixing L-channel and R-channel data through a band-pass filter with two control parameters, mixing ratio and band-pass filter characteristic. The control parameters are set in register 95 (3DP[3:0], 3FLO). The 3D sound effect is applied to the ADC digital output.

3-Band Tone Control

Tone control has bass, midrange, and treble controls that can be adjusted from 12 dB to -12 dB in 1-dB steps by register 92 to 94 (LGA[4:0], MGA[4:0] and HGA[4:0]). Register 92 (LPAE), which attenuates the digital input signal automatically, can prevent clipping of the output signal at settings higher than 0 dB for bass control. LPAE is not effective for midrange and treble control.

High-Pass Filter and Notch Filter

The high-pass filter cuts dc offset in the analog section of the ADC and can be set to 4 Hz, 120 Hz, or 240 Hz at 48-kHz sampling by register 81 (HPF[1:0]).

Notch filters remove noise at particular frequencies, CCD noise, motor noise and other mechanical noise in an application. The PCM1870 has two notch filters, whose center frequency and frequency bandwidth can be programmed by registers 96 to 104.

Digital Monaural Mixing

The audio data can be mixed to monaural digital data from stereo digital data in the internal audio interface section by register 96 (MXEN).



Zero-Cross Detection

Zero-cross detection minimizes audible zipper noise while changing analog volume and digital attenuation. This function can be applied to digital input or digital output by register 86 (ZCRS).

Power Up/Down for Each Module

Using register 73 (PBIS), register 82 (PAIR, PAIL, PADS, PMCB, PADR, PADL), and register 90 (PCOM), unused modules can be powered down to minimize power consumption (13 mW when recording only).

Digital Interface

All digital I/O pins can interface at various power supply voltages. The V_{IO} pin can be connected to a 1.71-V to 3.6-V power supply.

Power Supply

The V_{CC} pin can be connected to 2.4 V to 3.6 V. The V_{DD} pin and V_{IO} pin can be connected to 1.71 to 3.6 V. A different voltage can be applied to each of these pins (for example, $V_{DD} = 1.8$ V, $V_{IO} = 3.3$ V).

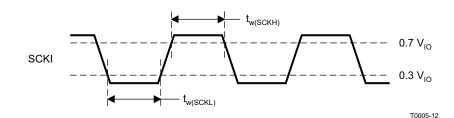
DESCRIPTION OF OPERATION

System Clock Input

The PCM1870 can accept input clocks of various frequencies without a PLL. The clocks are used for clocking of the digital filters, auto level control, and delta-sigma modulators, and classified into common-audio and application-specific clocks. Table 2 shows frequencies of the common audio clock and the application-specific clock. Figure 13 shows timing requirements for system clock inputs. The sampling rate and frequency of the system clock are determined by settings in register 86 (MSR[2:0]) and register 85 (NPR[5:0]). Note that the sampling rate of the application-specific clock has a little sampling error. The details are shown in Table 8.

Table 2. Frequency of Common Audio Clock

	FREQUENCY
Common audio clock	11.2896, 12.288, 16.9344, 18.432 MHz
Application-specific clock	12, 13, 13.5, 24, 26, 27, 19.2, 19.68, 38.4, 39.36 MHz



PARAMETERS	SYMBOL	MIN	UNIT
System clock pulse duration, high	t _{w(SCKH)}	14	ns
System clock pulse duration, low	t _{w(SCKL)}	14	ns

Figure 13. System Clock Timing

Power-On Reset and System Reset

The power-on-reset circuit outputs reset signal, typically at $V_{DD} = 1.2$ V, and this circuit does not depend on the voltage of other power-supplies (V_{CC} , V_{PA} , and V_{IO}). Internal circuits are cleared to default status, then signals are removed from all analog and digital outputs. The PCM1870 does not require any power supply sequencing. Register data must be written after turning all power supplies on.

System reset is enabled by setting register 85 (SRST), and all registers are cleared automatically. All circuits are reset to their default status at once. Note that the PCM1870 has audible pop noise on the analog outputs when enabling SRST.

Power On/Off Sequence

To reduce audible pop noise, a sequence of register settings is required after turning all power supplies on when powering up, or before turning the power supplies off when powering down. If some modules are not required for a particular application or operation, they should be placed in the power-down state after performing the power-on sequence. The recommended power-on and power-off sequences are shown in Table 3 and Table 4, respectively.



Table 3. Recommended Power-On Sequence

STEP	REGISTER SETTINGS	NOTE
1	—	Turn ON all power supplies ⁽¹⁾
2	5102 _H	ADC audio interface format (left-justified) (2)
3	5A00 _H	PG1, PG2 gain control (0 dB)
4	4980 _H	Analog bias power up
5	5601 _H	Zero-cross detection enable
6	4A01 _H	V _{COM} power up
7	523F _H	Analog front end (ADL, ADR, D2S, MCB, PG1, 2, 5, 6) power up
8	5711 _H	Analog input (MUX3, MUX4) select. Analog input (MUX1, MUX2) select
9	4F0C _H	Analog input L-ch (PG3) volume (0 dB) ⁽³⁾
10	500C _H	Analog input R-ch (PG4) volume (0 dB) (3)

(1) V_{DD} should be turned on first or at the same time with other power supplies. It is recommended to set the register data after turning on all power supplies and while the system clock is running.

(2) The audio interface format should be set to match the DSP or decoder being used.

(3) Any level is acceptable for volume or attenuation. The level should return to that recorded in the register data when system was last powered off.

Table 4. Recommended Power-Off Sequence

STEP	REGISTER SETTINGS	NOTE
1	5132 _H	ADC L-ch/R-ch digital soft mute enable, ADC audio interface format (left-justified) ⁽¹⁾
2	5200 _H	Analog front end (ADL, ADR, D2S, MCB, PG1, 2, 5, 6) power down
3	4A00 _H	V _{COM} power down
4	4900 _H	Analog bias power down
5	—	Turn OFF all power supplies. ⁽²⁾

(1) The audio interface format should be set to match the DSP or decoder being used.

(2) Power-supply sequencing is not required. It is recommended to make the required register settings while the system clock is running, then turn off all power supplies.

Power-Supply Current

The current consumption of the PCM1870 depends on the power-up/down status of each circuit module. In order to save power consumption, disabling each module is recommended when it is not used in an application or operation. Table 5 shows current consumption in some states.

	POW	ER SUPPL		NT [mA]	PD	[mW]	
OPERATION MODE	V _{DD} (1.8 V)	V _{DD} (3.3 V)	V _{DD} V _{CC} (3.3 V) (3.3 V)		TOTAL (V _{DD} = 1.8 V)	TOTAL (V _{DD} = 3.3 V)	
ALL POWER DOWN	0.000	0.000	0.001	0.000	0.003	0.003	
Recording (f _S = 48 kHz)							
Line input (AIN2L/AIN2R)	1.78	3.71	4.58	0.10	18.3	27.7	
Mic input (AIN1L/AIN1R, 20 dB)	1.79	3.71	5.06	0.10	19.9	29.3	
Mic input (AIN1L/AIN1R, 20 dB) with ALC	2.73	5.59	5.06	0.10	21.6	35.5	
Mono mic input (AIN1L, 20 dB)	1.33	2.80	3.56	0.10	14.1	21.3	
Mono mic input (AIN1L, 20 dB) with ALC	2.21	4.60	3.56	0.10	15.7	27.3	
Mono diff mic input (AIN1L/AIN1R, 20 dB)	1.33	2.80	3.88	0.10	15.2	22.4	
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	2.21	4.60	3.88	0.10	16.8	28.3	
Recording (f _S = 22.05 kHz)		••		· · ·		•	
Line input (AIN2L/AIN2R)	0.82	1.66	3.71	0.10	13.7	18.1	
Mic input (AIN1L/AIN1R, 20 dB)	0.82	1.66	4.20	0.10	15.3	19.7	
Mic input (AIN1L/AIN1R, 20 dB) with ALC	1.26	2.55	4.20	0.10	16.1	22.6	
Mono mic input (AIN1L, 20 dB)	0.61	1.23	2.74	0.10	10.1	13.4	
Mono mic input (AIN1L, 20 dB) with ALC	1.03	2.10	2.74	0.10	10.9	1.63	
Mono diff mic input (AIN1L/AIN1R, 20 dB)	0.61	1.23	3.06	0.10	11.2	14.5	
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	1.02	2.08	3.06	0.10	11.9	17.3	
Recording (f _S = 16 kHz)							
Line input (AIN2L/AIN2R)	0.59	1.18	3.51	0.10	12.7	15.8	
Mic input (AIN1L/AIN1R, 20 dB)	0.59	1.18	3.99	0.10	14.2	17.4	
Mic input (AIN1L/AIN1R, 20 dB) with ALC	0.91	1.85	3.99	0.10	14.8	19.6	
Mono mic input (AIN1L, 20 dB)	0.44	0.87	2.55	0.10	9.2	11.6	
Mono mic input (AIN1L, 20 dB) with ALC	0.75	1.52	2.55	0.10	9.8	13.8	
Mono diff mic input (AIN1L/AIN1R, 20 dB)	0.44	0.87	2.87	0.10	10.3	12.7	
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC	0.74	1.50	2.87	0.10	10.8	14.8	
Recording (f _S = 8 kHz)							
Line input (AIN2L/AIN2R)	0.29	0.54	3.23	0.10	11.2	12.8	
Mic input (AIN1L/AIN1R, 20 dB)	0.29	0.54	3.72	0.10	12.8	14.4	
Mic input (AIN1L/AIN1R, 20 dB) with ALC	0.46	0.88	3.72	0.10	13.1	15.5	
Mono mic input (AIN1L, 20 dB)	0.22	0.39	2.29	0.10	8.0	9.2	
Mono mic input (AIN1L, 20 dB) with ALC	0.37	0.70	2.29	0.10	8.2	10.2	
Mono diff mic input (AIN1L/AIN1R, 20 dB)	0.22	0.39	2.61	0.10	9.0	10.2	
Mono diff mic input (AIN1L/AIN1R, 20 dB) with ALC)	0.37	0.70	2.61	0.10	9.3	11.3	
Condition: 256 f _S , 16 bits, slave mode, zero data input,	no load						

Table 5. Power Consumption Table



Audio Serial Interface

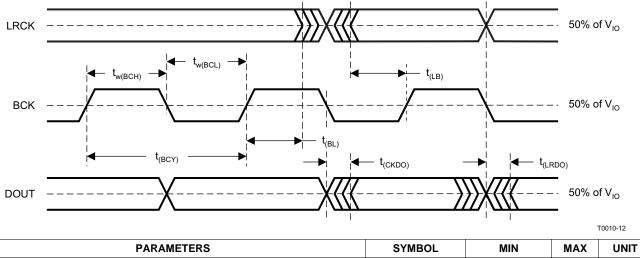
The audio serial interface of the PCM1870 consists of LRCK, BCK and DOUT. Sampling rate (f_s), left and right channel are present on LRCK. DOUT transmits the serial data from the decimation filter for the ADC. BCK is used to transmit the serial audio data on DOUT at its high-to-low transition. BCK and LRCK should be synchronized with audio system clock. Ideally, it is recommended that they are derived from it.

The PCM1870 requires LRCK to be synchronized with the system clock. The PCM1870 do not need a specific phase relationship between LRCK and the system clock.

The PCM1870 has both master mode and slave mode interface formats, which can be selected by register 84 (MSTR). LRCK and BCK are generated from the system clock in master mode.

Audio Data Formats and Timing

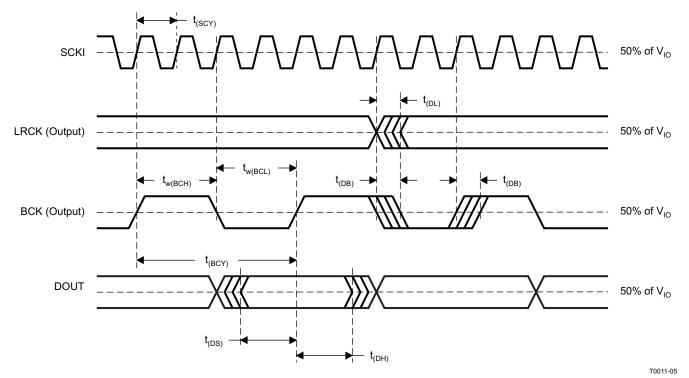
The PCM1870 supports I²S, right-justified, left-justified, and DSP formats. The data formats are shown in Figure 16, and they are selected using resister 70 (RFM[1:0], PFM[1:0]). All formats require binary 2s-complement, MSB-first audio data. The default format is I²S. Figure 14 shows a detailed timing diagram.



PARAMETERS	SYMBOL	MIN	MAX	UNIT
BCK pulse cycle time (I ² S, left- and right-justified formats)	t _(BCY)	1/(64f _S) ⁽¹⁾		
BCK pulse cycle time (DSP format)	t _(BCY)	1/(256f _S) ⁽¹⁾		
BCK high-level time	t _{w(BCH)}	35		ns
BCK low-level time	t _{w(BCL)}	35		ns
BCK rising edge to LRCK edge	t _(BL)	10		ns
LRCK edge to BCK rising edge	t _(LB)	10		ns
DOUT delay time from BCK falling edge	t _(CKDO)		40	ns

(1) f_S is the sampling frequency.

Figure 14. Audio Interface Timing (Slave Mode)



PARAMETERS	SYMBOL	MIN	MAX	UNIT
SCKI pulse cycle time	t _(SCY)	1/(256f _S) ⁽¹⁾		
LRCK edge from SCKI rising edge	t _(DL)	5	140	ns
BCK edge from SCKI rising edge	t _(DB)	5	140	ns
BCK pulse cycle time	t _(BCY)	1/(64f _S) ⁽¹⁾		
BCK high-level time	t _{w(BCH)}	146		ns
BCK low-level time	t _{w(BCL)}	146		ns
DOUT setup time	t _(DS)	10		ns
DOUT hold time	t _(DH)	10		ns

(1) f_S is up to 48 kHz. f_S is the sampling frequency Figure 15. Audio Interface Timing (Master Mode)

PCM1870



(a) Right-Justified Data Format; L-Channel = HIGH, R-Channel = LOW LRCK R-Channe L-Channe (= 32 f_S, 48 f_S, or 64 f_S 16-Bit Right-Justified DOUT 15 16 15 11 12 16 MSB I SB MSB I SB (b) I²S Data Format; L-Channel = LOW, R-Channel = HIGH L-Channe R-Channel LRCK <u>ਗ਼</u> BCK ∱L∱I∱I **fififi** (= 32 f_S, 48 f_S, or 64 f_s)⁽¹⁾ DOUT LSB LSB (c) Left-Justified Data Format; L-Channel = HIGH, R-Channel = LOW LRCK L-Channel R-Channel ┟┫┫┫┫┫┫┫┫┫ BCK I ♠ (= 32 f_s, 48 f_s, or 64 f_s) DOUT 2 3 4 15 16 1 2 15 1 2 1 5 6 11 12 3 16 MSB I SB MSB LSB (d) Burst BCK Interface Format in Master Mode; L-Channel = HIGH, R-Channel = LOW LRCK I -Channel R-Channel በብብብብብብብ **fififi** ſſ BCK (= 32 f_S, 48 f_S DOUT 2 1 2 LSB LSB MSB (e) DSP Format LRCK <u>ללת המתומת ה</u> BCK (= 32 f_s, 48 128 fo or 256 fo DOUT 5 6 12 16 1 3 15 2 R MSB LSB MSB LSB T0009-09 (1) All audio interface formats support BCK = 64 f_S in master mode (register 69, MSTR = 1). When set to multisampling

rate, f_S of BCK is set to half the rate of the DSM operation frequency.

Figure 16. Audio Data Output Formats



THREE-WIRE INTERFACE (SPI, MODE (PIN 28) = LOW)

All write operations for the serial control port use 16-bit data words. Figure 17 shows the control data word format. The most significant bit must be 0. There are seven bits, labeled IDX[6:0], that set the register address for the write operation. The least-significant eight bits, D[7:0], contain the data to be written to the register specified by IDX[6:0].

Figure 18 shows the functional timing diagram for writing to the serial control port. To write the data into the mode register, the data is clocked into an internal shift register on the rising edge of the MC clock. The serial data should change on the falling edge of the MC clock and should be LOW during write mode. The rising edge of MS should be aligned with the falling edge of the last MC clock pulse in the 16-bit frame. The MC can run continuously between transactions while MS is in the LOW state.

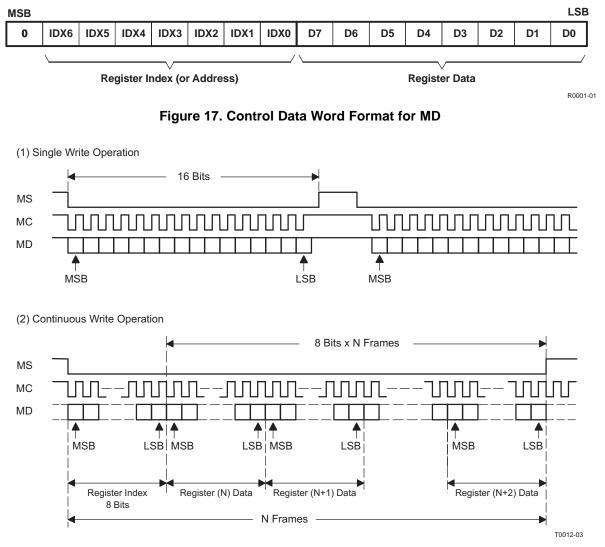
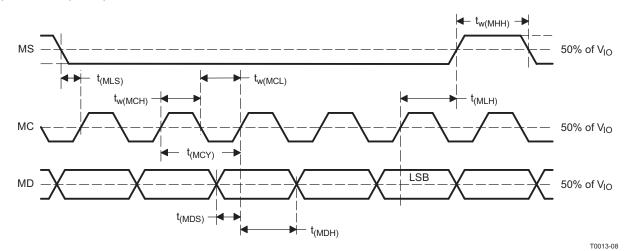


Figure 18. Register Write Operation



Three-Wire Interface (SPI) Timing Requirements

Figure 19 shows a detailed timing diagram for the serial control interface. These timing parameters are critical for proper control port operation.



PARAMETERS	SYMBOL	MIN	TYP	MAX	UNIT
MC pulse cycle time	t _(MCY)	500 ⁽¹⁾			ns
MC low-level time	t _{w(MCL)}	50			ns
MC high-level time	t _{w(MCH)}	50			ns
MS high-level time	t _{w(MHH)}	See ⁽¹⁾			ns
MS falling edge to MC rising edge	t _(MLS)	50			ns
MS hold time	t _(MLH)	20			ns
MD hold time	t _(MDH)	15			ns
MD setup time	t _(MDS)	20			ns

(1) $3/(128 f_S)$ s (min), where f_S is the sampling frequency

Figure 19. SPI Interface Timing

TWO-WIRE INTERFACE [I²C, MODE (PIN 28) = HIGH]

The PCM1870 supports the I^2C serial bus and the data transmission protocol for the I^2C standard as a slave device. This protocol is explained in I^2C specification 2.0.

In I²C mode, the control terminals are changed as follows.

TERMINAL NAME	PROPERTY	DESCRIPTION
MS/ADR	Input	I ² C address
MD/SDA	Input/output	I ² C data
MC/SCL	Input	I ² C clock

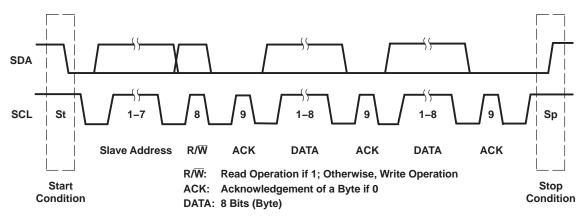
Slave Address

MSB							LSB	
1	0	0	0	1	1	ADR	R/W	

The PCM1870 has its 7-bit slave address. The first six bits (MSBs) of the slave address are factory preset to 1000 11. The next bit of the address byte is the device select bit, which can be user-defined by ADR terminal. A maximum of two PCM1870s can be connected on the same bus at one time. Each PCM1870 responds when it receives its own slave address.

Packet Protocol

A master device must control packet protocol, which is start condition, slave address with read/write bit, data if write or acknowledgement if read, and stop condition. The PCM1870 supports only slave-receiver and slave-transmitter.



Write Operation

Transmitter	М	М	М	S	М	S	М	S	М
Data Type	St	Slave Address	R/W	ACK	DATA	ACK	DATA	ACK	Sp

Read Operation

Transmitter	М	М	М	S	S	М	S	М	М
Data Type	St	Slave Address	R/W	ACK	DATA	ACK	DATA	NACK	Sp

M: Master Device St: Start Condition S: Slave Device Sp: Stop Condition

T0049-03

Figure 20. Basic I²C Framework



Write Operation

A master can write any PCM1870 registers using single access. The master sends a PCM1870 slave address with a write bit, a register address, and the data. When undefined registers are accessed, the PCM1870 does not send an acknowledgement. The Figure 21 shows a diagram of the write operation.

Transmitter	М	М	М	S	М	S	М	S	М
Data Type	St	Slave Address	W	ACK	Reg Address	ACK	Write Data	ACK	Sp

M: Master Device S: Slave Device

St: Start Condition W: Write ACK: Acknowledge Sp: Stop Condition

R0002-01

Figure 21. Framework for Write Operation

Read Operation

A master can read the PCM1870 register. The value of the register address is stored in an indirect index register in advance. The master sends a PCM1870 slave address with a read bit after storing the register address. Then the PCM1870 transfers the data which the index register points to. Figure 22 shows a diagram of the read operation.

Transmitter	М	М	М	S	М	S	М	М	М	S	S	М	М
Data Type	St	Slave Address	W	ACK	Reg Address	ACK	Sr	Slave Address	R	ACK	Read Data	NACK	Sp

M: Master Device S: Slave Device St: Start Condition

Sr: Repeated Start Condition ACK: Acknowledge Sp: Stop Condition NACK: Not Acknowledge

W: Write R: Read

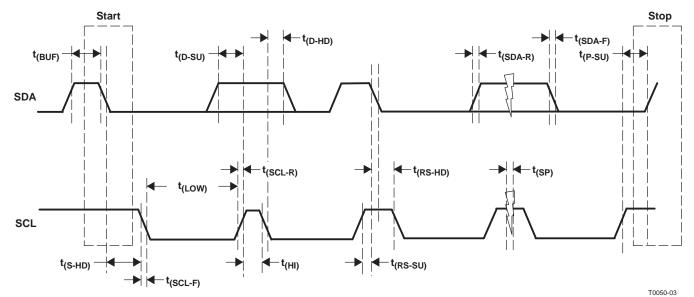
R0002-02

NOTE: The slave address after the repeated start condition must be the same as the previous slave address.

Figure 22. Read Operation



Timing Diagram



TIMING CH	IARACTERISTICS				
SYMBOL	PARAMETER	CONDITIONS	MIN	MAX	UNIT
f _{SCL}	SCL clock frequency	Standard		100	kHz
t _(BUF)	Bus free time between a STOP and START condition	Standard	4.7		μs
t _(LOW)	Low period of the SCL clock	Standard	4.7		μs
t _(HI)	High period of the SCL clock	Standard	4		μs
t _(RS-SU)	Setup time for START condition	Standard	4.7		μs
t _(S-HD)	Hold time for START condition	Standard	4		μs
t _(D-SU)	Data setup time	Standard	250		ns
t _(D-HD)	Data hold time	Standard	0	900	ns
t _(SCL-R)	Rise time of SCL signal	Standard	20 + 0.1 C _B	1000	ns
t _(SCL-F)	Fall time of SCL signal	Standard	20 + 0.1 C _B	1000	ns
t _(SDA-R)	Rise time of SDA signal	Standard	20 + 0.1 C _B	1000	ns
t _(SDA-F)	Fall time of SDA signal	Standard	20 + 0.1 C _B	1000	ns
t _(P-SU)	Setup time for STOP condition	Standard	4		μs
C _B	Capacitive load for SDA and SCL line			400	pF
t _(SP)	Pulse duration of spike suppressed			25	ns

Figure 23. I²C Interface Timing



USER-PROGRAMMABLE MODE CONTROLS

Register Map

The mode control register map is shown in Table 6. Each register includes an index (or address) indicated by the IDX[6:0] bits.

REGISTER	IDX [6:0] (B14–B8)	DESCRIPTION	B7	B6	В5	B4	B3	B2	B1	В0
Register 73	49h	Analog bias power up/down	PBIS	RSV						
Register 74	4Ah	V _{COM} power up/down	RSV	PCOM						
Register 79	4Fh	Volume for ADC input (L-ch)	RSV	RSV	ALV5	ALV4	ALV3	ALV2	ALV1	ALV0
Register 80	50h	Volume for ADC input (R-ch)	RSV	RSV	ARV5	ARV4	ARV3	ARV2	ARV1	ARV0
Register 81	51h	ADC high pass-filter, soft mute, audio interface	HPF1	HPF0	RMUL	RMUR	RSV	DSMC	RFM1	RFM0
Register 82	52h	ADC, MCB, PG1, 2, 5, 6, D2S power up/down	RSV	RSV	PAIR	PAIL	PADS	PMCB	PADR	PADL
Register 83	53h	Auto level control for recording	RALC	RSV	RRTC	RATC	RCP1	RCP0	RLV1	RLV0
Register 84	54h	Master mode	RSV	RSV	RSV	RSV	RSV	MSTR	RSV	BIT0
Register 85	55h	System reset, sampling rate control	SRST	RSV	NPR5	NPR4	NPR3	NPR2	NPR1	NPR0
Register 86	56h	BCK config, master mode, zero cross	MBST	MSR2	MSR1	MSR0	RSV	RSV	RSV	ZCRS
Register 87	57h	Analog input select (MUX1, 2, 3, 4)	AD2S	RSV	AIR1	AIR0	RSV	RSV	AIL1	AIL0
Register 90	5Ah	V _{COM} power up/down, ramp up/down time, boost	RSV	RSV	RSV	RSV	RSV	RSV	G20R	G20L
Register 92	5Ch	Bass boost gain level (200 Hz)	LPAE	RSV	RSV	LGA4	LGA3	LGA2	LGA1	LGA0
Register 93	5Dh	Middle boost gain level (1 kHz)	RSV	RSV	RSV	MGA4	MGA3	MGA2	MGA1	MGA0
Register 94	5Eh	Treble boost gain level (5 kHz)	RSV	RSV	RSV	HGA4	HGA3	HGA2	HGA1	HGA0
Register 95	5Fh	Sound effect source select, 3D sound	RSV	3DEN	RSV	3FL0	3DP3	3DP2	3DP1	3DP0
Register 96	60h	2-stage notch filter, digital monaural mixing	NEN2	NEN1	NUP2	NUP1	RSV	RSV	RSV	MXEN
Register 97	61h	1st-stage notch filter lower coefficient (a1)	F107	F106	F105	F104	F103	F102	F101	F100
Register 98	62h	1st-stage notch filter upper coefficient (a1)	F115	F114	F113	F112	F111	F110	F109	F108
Register 99	63h	1st-stage notch filter lower coefficient (a2)	F207	F206	F205	F204	F203	F202	F201	F200
Register 100	64h	1st-stage notch filter upper coefficient (a2)	F215	F214	F213	F212	F211	F210	F209	F208
Register 101	65h	2nd-stage notch filter lower coefficient (a1)	S107	S106	S105	S104	S103	S102	S101	S100
Register 102	66h	2nd-stage notch filter upper coefficient (a1)	S115	S114	S113	S112	S111	S110	S109	S108
Register 103	67h	2nd-stage notch filter lower coefficient (a2)	S207	S206	S205	S204	S203	S202	S201	S200
Register 104	68h	2nd-stage notch filter upper coefficient (a2)	S215	S214	S213	S212	S211	S210	S209	S208
Register 124	7Ch	Mic boost	RSV	RSV	RSV	RSV	RSV	RSV	G12R	G12L
ADC: A/D conv	erter MCB	: Microphone bias								
PGx: Analog in	put buffer	D2S: Differential to single-ended amplifier								



Register 73				-								-				
	B15	B14	B13	B12	B11	B10	B9	B8	B7	B 6	B5	B4	B 3	B2	B1	B0
Register 73	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	PBIS	RSV	RSV	RSV	RSV	RSV	RSV	RSV
IDX[6:0]: 1	00 10	01b (4	19h) R	egiste	er 73											
PBIS: Pow	er Up	/Dow	n Cor	trol f	or Bia	IS										
Default valu	ue: 0															
This bit is u	ised to	o cont	rol po	wer u	o/dow	n for t	he an	alog bi	as circ	uit.						
PBIS = 0	I	Power of	down (d	lefault)												
PBIS = 1	l	Power u	qu													
Register 74	4															
j	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 74	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	RSV	PCOM
IDX[6:0]: 1	00 10	10b (4	IAh) F	Reaiste	er 74											
PCOM: Po		``	,	•		сом										
Default valu		•				00.11										
This bit is u	ised to	o cont	rol po	wer u	o/dow	n for \	V _{сом} .									
PCOM = 0		Power of	down (d	lefault)												

PCOM = 0	Power down (default)
PCOM = 1	Power up



Register 79) and	80														
-	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 79	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	ALV5	ALV4	ALV3	ALV2	ALV1	ALV0
Register 80	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	ARV5	ARV4	ARV3	AR2	ARV1	ARV0

IDX[6:0]: 100 1111b (4Fh) Register 79

IDX[6:0]: 101 0000b (50h) Register 80

ALV[5:0]: Gain Control for PG3 (ADC Analog Input R-Channel)

ARV[5:0]: Gain Control for PG4 (ADC Analog Input L-Channel)

Default value: 00

PG3 and PG4 can be independently controlled for ADC input from 30 dB to -12 dB in 1-dB steps. ADC output may have zipper noise when changing levels. In the PCM1870, the noise can be reduced when making the change by using zero-cross detection (Register 85, ZCRS).

ALV[5:0], A	RV[5:0]		ALV[5:0], A	RV[5:0]	
BINARY	HEX	GAIN LEVEL SETTING	BINARY	HEX	GAIN LEVEL SETTING
10 1010	2A	30 dB	01 0100	14	8 dB
10 1001	29	29 dB	01 0011	13	7 dB
10 1000	28	28 dB	01 0010	12	6 dB
10 0111	27	27 dB	01 0001	11	5 dB
10 0110	26	26 dB	01 0000	10	4 dB
10 0101	25	25 dB	00 1111	0F	3 dB
10 0100	24	24 dB	00 1110	0E	2 dB
10 0011	23	23 dB	00 1101	0D	1 dB
10 0010	22	22 dB	00 1100	0C	0 dB
10 0001	21	21 dB	00 1011	0B	–1 dB
10 0000	20	20 dB	00 1010	0A	-2 dB
01 1111	1F	19 dB	00 1001	09	–3 dB
01 1110	1E	18 dB	00 1000	08	-4 dB
01 1101	1D	17 dB	00 0111	07	–5 dB
01 1100	1C	16 dB	00 0110	06	-6 dB
01 1011	1B	15 dB	00 0101	05	-7 dB
01 1010	1A	14 dB	00 0100	04	8 dB
01 1001	19	13 dB	00 0011	03	–9 dB
01 1000	18	12 dB	00 0010	02	–10 dB
01 0111	17	11 dB	00 0001	01	–11 dB
01 0110	16	10 dB	00 0000	00	-12 dB (default)
01 0101	15	9 dB			

Table 7. Gain Level Setting



Register 8 ⁻	1															
	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 81	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	HPF1	HPF0	RMUL	RMUR	RSV	DSMC	RFM1	RFM0

IDX[6:0]: 101 0001b (51h) Register 81

HPF[1:0]: High-Pass Filter Selection

Default value: 00

The PCM1870 has a digital high-pass filter to remove dc voltage at the input of the ADC. The cutoff frequency of the high-pass filter can be selected.

HPF[1:0]	High Pass Filter Selection
0 0	f _C = 4 Hz at 48 kHz (default)
0 1	f _C = 240 Hz at 48 kHz
10	f _C = 120 Hz at 48 kHz
11	High-pass filter disabled

RMUL: Digital Soft Mute Control for ADC L-Channel

RMUR: Digital Soft Mute Control for ADC R-Channel

Default value: 1

The digital output of the ADC can be independently muted or unmuted. The transition from the current volume level to mute, or the return to the previous volume setting from mute, occurs at the rate of one 1-dB step for each $8/f_S$ time period. When RMUL and RMUR = 0, the digital data is increased from mute to the previous attenuation level, and when RMUL and RMUR = 1, the digital data is decreased from the current attenuation level to mute. In the PCM1870, audible zipper noise can be reduced by using zero-cross detection (register 85, ZCRS).

RMUL, RMUR = 0 Mute disabled RMUL, RMUR = 1 Mute enabled (default)

DSMC: Waiting Time Turn ADC Mute Off at Power Up

Default value: 0

ADC digital output has waiting time at power up when DSMC = 0. It is recommended to set DSMC = 0.

DSMC = 0	10 ms at 48 kHz (default)
DSMC = 1	No wait

RFM[1:0]: Audio Interface Selection for ADC (Digital Output)

Default value: 00

The audio interface format for ADC digital output has I²S, right-justified, left-justified, and DSP formats.

RFM[1:0]	Audio Interface Selection for ADC Digital Output
0 0	I ² S (default)
0 1	Right-justified
10	Left-justified
11	DSP mode



Register 82	2 B15	B14	B13	B12	B11	B10	В9	B8	B7	B6	В5	В4	B3	B2	B1	В0	
Register 82	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	PAIR	PAIL	PADS	PMCB	PADR	PADL	

IDX[6:0]: 101 0010b (52h) Register 82

PAIR: Power Up/Down for PG2 and PG6 (Gain Amplifier for Analog Input R-Channel)

PAIR: Power Up/Down for PG1 and PG5 (Gain Amplifier for Analog Input L-Channel)

Default value: 0

This bit is used to control power up/down for PG1, -2 and PG5, -6 (gain amplifier for analog input).

PAIR, PAIL = 0 Power down (default) PAIR, PAIL = 1 Power up

PADS: Power Up/Down for D2S (Differential Amplifier) of AIN1L and AIN1R

Default value: 0

This bit is used to control power up/down for D2S (differential-to-single amplifier).

PADS = 0	Power down (default)
PADS = 1	Power up

PMCB: Power Up/Down Control for Microphone Bias Source

Default value: 0

This bit is used to control power up/down for the microphone bias source.

PMCB = 0	Power down (default)
PMCB = 1	Power up

PADR: Power Up/Down Control for ADR (ADC and Digital Filter R-Channel)

PADL: Power Up/Down Control for ADL (ADC and Digital Filter L-Channel)

Default value: 0

This bit is used to control power up/down for the ADC and decimation filter.

PADR, PADL = 0 Power down (default) PADR, PADL = 1 Power up

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Register 83

•	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 83	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RALC	RSV	RRTC	RATC	RCP1	RCP0	RLV1	RLV0

IDX[6:0]: 1010011b (53h) Register 83

RALC: Automatic Level Control (ALC) Enable for Recording

Default value: 0

Auto level control can be enabled with some parameters for microphone input or lower-level analog source.

RALC = 0	Disabled (default)
RALC = 1	Enabled

RRTC: ALC Recovery Time Control for Recording

Default value: 0

This bit selects the time during which a gain level change completes to compress the signal when the input to the ADC increases in amplitude.

RRTC = 0	3.4 s (default)
RRTC = 1	13.6 s

RATC: ALC Attack Time Control for Recording

Default value: 0

This bit selects the time during which a gain level change completes to expand the signal when the input to the ADC decreases in amplitude.

RATC = 0	1 ms (default)
RATC = 1	2 ms

RCP[1:0]: ALC Compression Level Control for Recording

Default value: 00

Auto level control can set the compression level to -2, -6, or -12 dB. Higher-level signals should be compressed to avoid saturation or degradation of sound quality.

RCP[1:0]	ALC Compression Level Control for Recording
0 0	-2 dB (default)
0 1	-6 dB
10	–12 dB
11	Reserved

RLV[1:0]: ALC Expansion Level Control for Recording

Default value: 00

Auto level control can set the expansion level to 0, 6, 14, or 24 dB. Lower-level signals should be expanded to make a small signal easy to hear. If set to 0 dB, the ALC can be operated only as a limiter.

RLV[1:0]	ALC Expansion Level Control for Recording
0 0	0 dB (default)
0 1	6 dB
10	14 dB
11	24 dB



Register 84–86																
-	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 84	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	MSTR	RSV	BIT0
Register 85	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	SRST	RSV	NPR5	NPR4	NPR3	NPR2	NPR1	NPR0
Register 86	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	MBST	MSR2	MSR1	MSR0	RSV	RSV	RSV	ZCRS

IDX[6:0]: 101 0100b (54h) Register 84

IDX[6:0]: 101 0101b (55h) Register 85

IDX[6:0]: 101 0110b (56h) Register 86

MSTR: Master or Slave Selection for Audio Interface

Default value: 0

This bit is used to select either master or slave mode for the audio interface. In master mode, PCM1870 generates LRCK and BCK from the system clock. In slave mode, it receives LRCK and BCK from another device.

MSTR = 0	Slave interface (default)
MSTR = 1	Master interface

BIT0: Bit Length Selection for Audio Interface

Default value: 1

This bit is used to select data bit length for the ADC output.

BIT0 = 0	Reserved
BIT0 = 1	16 bits (default)

SRST: System Reset

Default value: 0

This bit is used to enable system reset. All circuits are reset by setting SRST = 1. After completing the reset sequence, SRST resets to 0 automatically.

SRST = 0	Reset disabled (default)
SRST = 1	Reset enabled

NPR[5:0]: System Clock Rate Selection

Default value: 00 0000

These bits are used to select the system clock rate. See Table 8 for details.

MBST: BCK Output Configuration in Master Mode

Default value: 0

This bit is used to control the BCK output configuration in master mode. V_{IO} (I/O cell power supply) power consumption can be reduced by adjusting BCK edge to bit number when setting MBST = 1. This is effective in master mode (register 69 MSTR = 1).

MBST = 0	Normal output (default)	ĺ
MBST = 1	Burst output	





MSR[2:0]: System Clock Dividing Rate Selection in Master Mode (Register 70)

Default value: 000

These bits are used to set the dividing rate of the input system clock. See Table 8 for details.

SYSTEM CLOCK	ADC SAMPLING RATE	DAC SAMPLING RATE	REGISTER	R SETTING	BIT CLOCK
SCK (MHz)	ADC f _S (kHz)	DAC f _S (kHz)	MSR[2:0]	NPR[5:0]	BCK (f _S)
	24 (SC	CK/256)	010	00 0000	64
	16 (SC	CK/384)	011	00 0000	64
6 1 1 1	12 (SC	CK/512)	100	00 0000	64
6.144	8 (SC	:K/768)	101	00 0000	64
-	6 (SC	K/1024)	110	00 0000	64
-	4 (SC	K/1536)	111	00 0000	64
	32 (SC	CK/256)	010	00 0000	64
8.192	16 (SC	CK/512)	100	00 0000	64
-	8 (SC	K/1024)	110	00 0000 00 0000 00 0000 00 0000 00 0000 00 0000 00 0000 00 0000 00 0000 00 0000 00 0000 00 0000	64
	48 (SC	CK/256)	010	00 0000	64
-	32 (SC	CK/384)	011	00 0000	64
10.000	24 (S0	CK/512)	100	00 0000	64
12.288	16 (SC	CK/768)	101	00 0000	64
-	12 (SC	:K/1024)	110	00 0000	64
-	8 (SC	K/1536)	111	00 0000	64
	48 (SC	CK/384)	011	00 0000	64
18.432	24 (S0	CK/768)	101	00 0000	64
-	12 (SC	:K/1536)	111	101 00 0000 110 00 0000 111 00 0000 010 00 0000 100 00 0000 100 00 0000 110 00 0000 110 00 0000 010 00 0000 011 00 0000 101 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 100 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000 111 00 0000	64
	22.05 (\$	SCK/256)	010	00 0000	64
-	14.7 (S	CK/384)	011	00 0000	64
5 0 4 4 0	11.025 (SCK/512)	100	00 0000	64
5.6448	7.35 (S	CK/768)	101	00 0000	64
-	5.5125 (\$	SCK/1024)	110	00 0000	64
-	3.675 (S	CK/1536)	111	00 0000	64
	44.1 (S	CK/256)	010	00 0000	64
=	29.4 (S	CK/384)	011	00 0000	64
44,0000	22.05 (\$	SCK/512)	100	00 0000	64
11.2896	14.7 (S	CK/768)	101	00 0000	64
F	11.025 (\$	SCK/1024)	110	00 0000	64
	7.35 (S	CK/1536)	111	00 0000	64

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SYSTEM CLOCK	ADC SAMPLING RATE	DAC SAMPLING RATE	REGISTER	R SETTING	BIT CLOCK	
SCK (MHz)	ADC SAMPLING RATE ADC f _S (kHz)	DAC SAMPLING RATE	MSR[2:0]	NPR[5:0]	BCK (f _S)	
	48.214 (SCK/280)	010	00 0010	70	
	44.407 (\$	010	00 0001	76		
		, SCK/420)	010	10 0010	70	
		SCK/560)	100	00 0010	70	
13.5	,	SCK/608)	100	00 0001	76	
	```	, SCK/840)	100	10 0010	70	
		SCK/1120)	110	00 0010	70	
	8.035 (S	CK/1680)	110	10 0010	70	
	48.214 (	SCK/560)	010	01 0010	70	
	44.407 (	SCK/608)	010	01 0001	76	
	32.142 (	SCK/840)	010	11 0010	70	
	24.107 (S	SCK/1120)	100	01 0010	70	
27	,	SCK/1216)	100	01 0001	76	
	16.071 (S	SCK/1680)	100	11 0010	70	
	12.053 (S	SCK/2240)	110	01 0010	70	
	8.035 (S	CK/3360)	110	11 0010	70	
	48.387 (	SCK/248)	010	00 0100	62	
	44.117 (\$	SCK/272)	010	00 0011	68	
	32.258 (	SCK/372)	010	10 0100	62	
10	24.193 (\$	SCK/496)	100	00 0100	62	
12	22.058 (	SCK/544)	100	00 0011	68	
	16.129 (	SCK/744)	100	10 0100	62	
	12.096 (	SCK/992)	110	00 0100	62	
	8.064 (S	CK/1488)	110	10 0100	62	
	48.387 (	SCK/496)	010	01 0100	62	
	44.117 (\$	SCK/544)	010	01 0011	68	
	32.258 (	SCK/744)	010	11 0100	62	
0.4	24.193 (\$	SCK/992)	100	01 0100	62	
24	22.058 (S	SCK/1088)	100	01 0011	68	
	16.129 (S	SCK/1488)	100	11 0100	62	
	12.096 (S	SCK/1984)	110	01 0100	62	
	8.064 (S	CK/2796)	110	11 0100	62	
	48.484 (\$	SCK/396)	011	00 0110	66	
	44.444 (\$	SCK/432)	011	00 0101	72	
	32.323 (	SCK/594)	011	10 0110	66	
10.2	24.242 (\$	SCK/792)	101	00 0110	66	
19.2	22.222 (	SCK/864)	101	00 0101	72	
	16.161 (S	SCK/1188)	101	10 0110	66	
	12.121 (S	SCK/1584)	111	00 0110	66	
	8.080 (S	CK/2376)	111	10 0110	66	

# Table 9. System Clock Frequency for Application-Specific Audio Clock





SYSTEM CLOCK	ADC SAMPLING RATE	DAC SAMPLING RATE	REGISTER	REGISTER SETTING			
SCK (MHz)	ADC f _S (kHz)	DAC f _S (kHz)	MSR[2:0]	NPR[5:0]	BIT CLOCK BCK (f _S )		
	48.484 (	SCK/792)	011	01 0110	66		
-	44.444 (	SCK/864)	011	01 0101	72		
-	32.323 (\$	SCK/1188)	011	11 0110	66		
	24.242 (\$	SCK/1584)	101	01 0110	66		
38.4	22.222 (\$	SCK/1728)	101	01 0101	72		
-	16.161 (5	SCK/2376)	101	11 0110	66		
-	12.121 (8	SCK/3168)	111	01 0110	66		
-	8.080 (S	CK/4752)	111	11 0110	66		
	47.794 (	SCK/272)	010	00 1000	68		
-	43.918 (	SCK/296)	010	00 0111	74		
-	31.862 (	SCK/408)	010	10 1000	68		
40	23.897 (	SCK/544)	100	00 1000	68		
13	21.959 (	SCK/592)	100	00 0111	74		
-	15.931 (	SCK/816)	100	10 1000	68		
	11.948 (\$	SCK/1088)	110	00 1000	68		
-	7.965 (S	CK/1632)	110	10 1000	68		
	47.794 (	SCK/544)	010	01 1000	68		
-	43.918 (	SCK/592)	010	01 0111	74		
-	31.862 (	SCK/816)	010	11 1000	68		
	23.897 (\$	SCK/1088)	100	01 1000	68		
26	21.959 (\$	SCK/1184)	100	01 0111	74		
-	15.931 (\$	SCK/1632)	100	11 1000	68		
-	11.948 (\$	SCK/2176)	110	01 1000	68		
-	7.965 (S	CK/3264)	110	11 1000	68		
	48.235 (	SCK/408)	011	00 1010	68		
-	44.324 (	SCK/444)	011	00 1001	74		
-	32.156 (	SCK/612)	011	10 1010	68		
40.00	24.117 (	SCK/816)	101	00 1010	68		
19.68	22.162 (	SCK/888)	101	00 1001	74		
-	16.078 (\$	SCK/1224)	101	10 1010	68		
-	12.058 (\$	SCK/1632)	111	00 1010	68		
-	8.039 (S	CK/2448)	111	10 1010	68		
	48.235 (	SCK/816)	011	01 1010	68		
F	44.324 (	SCK/888)	011	01 1001	74		
F	32.156 (\$	SCK/1224)	011	11 1010	68		
20.20	24.117 (\$	SCK/1632)	101	01 1010	68		
39.36	22.162 (\$	SCK/1776)	101	01 1001	74		
-	16.078 (\$	SCK/2448)	101	11 1010	68		
F	12.058 (\$	SCK/3264)	111	01 1010	68		
	8.039 (S	CK/4896)	111	11 1010	68		

## Table 9. System Clock Frequency for Application-Specific Audio Clock (continued)



### ZCRS: Zero Cross for Digital Attenuation/Mute and Analog Gain Setting

Default value: 0

This bit is used to enable the zero-cross detector, which reduces zipper noise while the digital soft mute or analog gain setting is being changed. If no zero-cross data is input for a  $512/f_S$  period (10.6 ms at a 48-kHz sampling rate), then a time-out occurs and the PCM1870 starts changing the attenuation, gain, or volume level. The zero-cross detector cannot be used with continuous-zero and dc data.

ZCRS = 0	Zero cross disabled (default)
ZCRS = 1	Zero cross enabled
De vieter 07	

#### Register 87

Register of	B15	B14	B13	B12	B11	B10	B9	B8	B7	<b>B6</b>	B5	B4	B3	B2	B1	B0
Register 87	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	AD2S	RSV	AIR1	AIR0	RSV	RSV	AIL1	AIL0

#### **IDX[6:0]:** 101 0111b (57h) Register 87

#### AD2S: Differential Amplifier Selector (MUX3 and MUX4)

Default value: 0

This bit is used to select whether a single-ended amplifier or differential amplifier (D2S) is used as the input for the ADC. MUX3 and MUX4 use the single-ended input when AD2S = 0. MUX3 and MUX4 use the monaural differential input when AD2S = 1.

AD2S = 0	Single-ended amplifier (default)
AD2S = 1	Differential amplifier

### AIL[1:0]: AIN1L and AIN2L Selector (MUX1)

#### Default value: 00

MUX1 selects the analog input, AIN1L or AIN2L.

AIL[1:0]	AIN L-channel Select
0 0	Disconnect (default)
0 1	AIN1L
10	AIN2L
11	Reserved

#### AIR[1:0]: AIN1R and AIN2R Selector (MUX2)

Default value: 00

MUX2 selects the analog input, AIN1R or AIN2R.

AIR[1:0]	AIN R-channel Select
0 0	Disconnect (default)
0 1	AIN1R
10	AIN2R
11	Reserved



### **Register 90**

	B15	B14	B13	B12	B11	B10	B9	B8	B7	<b>B6</b>	B5	B4	B3	B2	B1	B0
Register 90	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	G20R	G20L

IDX[6:0]: 101 1010b (5Ah) Register 90

### G20R: 20-dB Boost for PG2 (Gain Amplifier for AIN1R and AIN2R)

Default value: 0

This bit is used to boost the microphone signal when the analog input is small.

G12R (REGISTER 124)	G20R (REGISTER 90)	PG2 GAIN
0	0	0 dB (default)
0	1	20 dB
1	0	12 dB
1	1	Reserved

### G20L: 20 dB Boost for PG1 (Gain Amplifier for AIN1L and AIN2L)

Default value: 0

This bit is used to boost the microphone signal when the analog input is small.

G12L (REGISTER 124)	G20L (REGISTER 90)	PG1 GAIN
0	0	0 dB (default)
0	1	20 dB
1	0	12 dB
1	1	Reserved



Register 92	2 B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	В5	B4	В3	B2	B1	В0	
Register 92	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	LPAE	RSV	RSV	LGA4	LGA3	LGA2	LGA1	LGA0	l

### **IDX[6:0]:** 101 1100b (5Ch) Register 92

### LPAE: Automatic Attenuation Setting for Bass Boost Gain Control

Default value: 0

A gain setting for bass boost may cause digital data saturation, depending on the input data level. Where this could occur, LPAE can be used to set the same attenuation level as the bass boost gain level for the digital input data.

LPAE = 1 Enable	LPAE = 0	Disble (default)
	LPAE = 1	Enable

### LGA[4:0]: Bass Boost Gain Control

Default value: 0 0000

These bits are used to set bass boost gain level for the digital data. The center frequency for boost is 200 Hz at 44.1 kHz.

LGA[4:0]	TONE CONTROL GAIN (BASS)	LGA[4:0]	TONE CONTROL GAIN (BASS)
0 0000	0 dB (default)	0 1111	0 dB
0 0011	12 dB	1 0000	-1 dB
0 0100	11 dB	1 0001	-2 dB
0 0101	10 dB	1 0010	–3 dB
0 0110	9 dB	1 0011	-4 dB
0 0111	8 dB	1 0100	–5 dB
0 1000	7 dB	1 0101	-6 dB
0 1001	6 dB	1 0110	-7 dB
0 1010	5 dB	1 0111	-8 dB
0 1011	4 dB	1 1000	–9 dB
0 1100	3 dB	1 1001	-10 dB
0 1101	2 dB	1 1010	–11 dB
0 1110	1 dB	1 1011	–12 dB



Register 93		B14	B13	B12	B11	B10	В9	B8	B7	B6	В5	B4	B3	B2	B1	В0	
Register 93	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	MGA4	MGA3	MGA2	MGA1	MGA0	

IDX[6:0]: 101 1101b (5Dh) Register 93

# MGA[4:0]: Middle Boost Gain Control

Default value: 0 0000

These bits are used to set midrange boost gain level for the digital data. The center frequency for boost is 1 kHz.

MGA[4:0]	TONE CONTROL GAIN (MID)	MGA[4:0]	TONE CONTROL GAIN (MID)
0 0000	0 dB (default)	0 1111	0 dB
0 0011	12 dB	1 0000	–1 dB
0 0100	11 dB	1 0001	–2 dB
0 0101	10 dB	1 0010	–3 dB
0 0110	9 dB	1 0011	4 dB
0 0111	8 dB	1 0100	–5 dB
0 1000	7 dB	1 0101	6 dB
0 1001	6 dB	1 0110	7 dB
0 1010	5 dB	1 0111	–8 dB
0 1011	4 dB	1 1000	–9 dB
0 1100	3 dB	1 1001	-10 dB
0 1101	2 dB	1 1010	–11 dB
0 1110	1 dB	1 1011	-12 dB

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Register 94		B14	B13	B12	B11	B10	В9	B8	B7	B6	B5	B4	В3	B2	B1	В0	
Register 94	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	HGA4	HGA3	HGA2	HGA1	HGA0	

# IDX[6:0]: 101 1110b (5Eh) Register 94

# HGA[4:0]: Treble Boost Gain Control

Default value: 0 0000

These bits are used to set treble boost gain level for the digital data. The center frequency for boost is 5 kHz.

HGA[4:0]	TONE CONTROL GAIN (TREBLE)	HGA[4:0]	TONE CONTROL GAIN (TREBLE)
0 0000	0 dB (default)	0 1111	0 dB
0 0011	12 dB	1 0000	-1 dB
0 0100	11 dB	1 0001	–2 dB
0 0101	10 dB	1 0010	–3 dB
0 0110	9 dB	1 0011	-4 dB
0 0111	8 dB	1 0100	–5 dB
0 1000	7 dB	1 0101	6 dB
0 1001	6 dB	1 0110	–7 dB
0 1010	5 dB	1 0111	-8 dB
0 1011	4 dB	1 1000	–9 dB
0 1100	3 dB	1 1001	-10 dB
0 1101	2 dB	1 1010	–11 dB
0 1110	1 dB	1 1011	–12 dB

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Register 95		B14	B13	B12	B11	B10	В9	B8	B7	B6	В5	B4	B3	B2	B1	B0	
Register 95	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	3DEN	RSV	3FL0	3DP3	3DP2	3DP1	3DP0	

#### **IDX[6:0]:** 1011111b (5Fh) Register 95

### **3DEN: 3D Sound Effect Enable**

Default value: 0

This bit is used for enabling the 3-D sound effect filter. This filter has two independently controlled parameters.

3DEN = 0	Disable (default)
3DEN = 1	Enable

### 3FL0: Filter Selection for 3D Sound

Default value: 0

This bit is used for selecting from two kinds of filter type, narrow and wide. These filters produce different 3-D effects.

3FL0 = 0	Narrow (default)
3FL0 = 1	Wide

### 3DP[3:0]: Efficiency for 3D Sound Effect

Default value: 0000

These bits are used for adjusting the 3-D sound efficiency. Higher percentages have greater efficiency.

3DP[3:0]	3D Sound Effect Efficiency
0000	0% (default)
0001	10%
0010	20%
0011	30%
0100	40%
0101	50%
0110	60%
0111	70%
1000	80%
1001	90%
1010	100%
1011	Reserved
: 1111	: Reserved



Register 96	5															
•	B15	B14	B13	B12	B11	B10	<b>B</b> 9	B8	B7	<b>B</b> 6	B5	B4	B3	B2	B1	B0
Register 96	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	NEN2	NEN1	NUP2	NUP1	RSV	RSV	RSV	MXEN

IDX[6:0]: 110 0000b (60h) Register 96

#### NEN2: Second-Stage Notch Filter Enable

Default value: 0

The PCM1870 has a two-stage notch filter. The two stages can separately set filter characteristics. This bit is used to enable the second stage.

NEN2 = 0	Disabled (default)
NEN2 = 1	Enabled

#### **NEN1: First-Stage Notch Filter Enable**

Default value: 0

The PCM1870 has a two-stage notch filter. The two stages can separately set filter characteristics. This bit is used to enable the first stage.

NEN1 = 0	Disabled (default)	
NEN1 = 1	Enabled	

#### NUP2: Second-Stage Notch Filter Coefficients Update

Default value: 0

This bit is used to update the coefficients for the second-stage notch filter. The coefficients written to registers 101, 102, 103, 104 are updated when NUP2 = 1.

NUP2 = 0	No update (default)
NUP2 = 1	Update

#### NUP1: First-Stage Notch Filter Coefficients Update

Default value: 0

This bit is used to update the coefficients for the first-stage notch filter. The coefficients written to registers 97, 98, 99, 100 are updated when NUP1 = 1.

NUP1 = 0	No update (default)
NUP1 = 1	Update

#### **MXEN: Digital Monaural Mixing**

Default value: 0

This bit is used to enable or disable monaural mixing in the section that combines L-ch and R-ch digital data.

MXEN = 0	Disabled (stereo, default)
MXEN = 1	Enabled (monaural mixing)



#### Registers 97–100

	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	<b>B0</b>
Register 97	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F107	F106	F105	F104	F103	F102	F101	F100
Register 98	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F115	F114	F113	F112	F111	F110	F109	F108
Register 99	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F207	F206	F205	F204	F203	F202	F201	F200
Register 100	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	F215	F214	F213	F212	F211	F210	F209	F208

IDX[6:0]: 110 0001b (61h) Register 97 IDX[6:0]: 110 0010b (62h) Register 98 IDX[6:0]: 110 0011b (63h) Register 99 IDX[6:0]: 110 0100b (64h) Register 100

F[107:100]: Lower 8 Bits of Coefficient  $a_1$  for First-Stage Notch Filter F[115:108]: Upper 8 Bits of Coefficient  $a_1$  for First-Stage Notch Filter F[207:200]: Lower 8 Bits of Coefficient  $a_2$  for First-Stage Notch Filter F[215:208]: Upper 8 Bits of Coefficient  $a_2$  for First-Stage Notch Filter

Default value: 0000 0000

These bits are used to change the characteristics of the first-stage notch filter. See *Calculating Filter Coefficients* for details.

#### Registers 101–104 B15 B14 B13 B12 B11 B10 **B**9 **B8 B7 B6 B5 B4 B**3 B2 **B1 B0** S107 S100 Register 101 0 IDX6 IDX5 IDX4 IDX3 IDX2 IDX1 IDX0 S106 S105 S104 S103 S102 S101 Register 102 0 IDX6 IDX5 IDX4 IDX3 IDX2 IDX1 IDX0 S115 S114 S113 S112 S111 S110 S109 S108 Register 103 0 IDX6 IDX5 IDX4 IDX3 IDX2 IDX1 IDX0 S207 S206 S205 S204 S203 S202 S201 S200 Register 104 0 IDX6 IDX5 IDX4 IDX3 IDX2 IDX1 IDX0 S215 S214 S213 S212 S211 S210 S209 S208

**IDX[6:0]:** 110 0101b (65h) Register 101 **IDX[6:0]:** 110 0110b (66h) Register 102 **IDX[6:0]:** 110 0111b (67h) Register 103 **IDX[6:0]:** 110 1000b (68h) Register 104

S[107:100]: Lower 8 Bits of Coefficient  $a_1$  for Second-Stage Notch Filter S[115:108]: Upper 8 Bts of Coefficient  $a_1$  for Second-Stage Notch Filter S[207:200]: Lower 8 Bits of Coefficient  $a_2$  for Second-Stage Notch Filter S[215:208]: Upper 8 Bits of Coefficient  $a_2$  for Second-Stage Notch Filter

Default value: 0000 0000

These bits are used to change the characteristics of the second-stage notch filter. See *Calculating Filter Coefficients* for details.



#### Calculating Filter Coefficients

The PCM1870 provides a dual-stage notch filter at the digital output of the ADC. The filter characteristics of each filter stage can be programmed. The characteristics are determined by calculating coefficients for three parameters, sampling frequency, center frequency and bandwidth, as shown in the following equations. All coefficients must be written as 2s-complement binary data into registers 97, 98, 99, 100, 101, 102, 103, and 104.  $f_S = Sampling frequency [Hz]$ 

 $f_{C}$  = Center frequency [Hz]

$$f_b = Bandwidth [Hz]$$

$$a_1 = -(1 + a_2) cos \left(\frac{2\pi f_C}{f_S}\right)$$

$$a_{2} = \frac{1 - \tan\left(\frac{2\pi f_{b}^{}/f_{S}^{}}{2}\right)}{1 + \tan\left(\frac{2\pi f_{b}^{}/f_{S}^{}}{2}\right)}$$

#### **Register 124**

	og.o.o		B14	B13	B12	B11	B10	B9	B8	B7	<b>B6</b>	B5	B4	B3	B2	B1	B0	
F	Register 124	0	IDX6	IDX5	IDX4	IDX3	IDX2	IDX1	IDX0	RSV	RSV	RSV	RSV	RSV	RSV	G12R	G12L	1

#### IDX[6:0]: 111 1100b (7Ch) Register 124

#### G12R: 12-dB Boost for PG2 (Gain Amplifier for AIN1R and AIN2R) G12L: 12-dB Boost for PG1 (Gain Amplifier for AIN1L and AIN2L)

Default value: 0

These bits are used to boost the microphone signal when the analog input is small. See Register 90 for the detailed settings.



### CONNECTION DIAGRAM

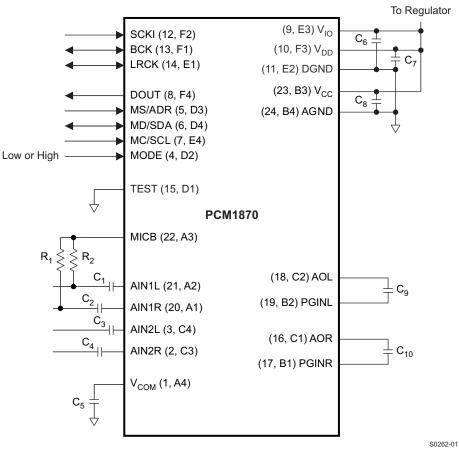


Figure 24. Connection Diagram

Table 10	Recommended	External	Parts
----------	-------------	----------	-------

C ₁ –C ₄	1 μF–10 μF	C ₉ , C ₁₀	1 μF–10 μF
C ₅	1 μF–4.7 μF	R ₁ , R ₂	2.2 kΩ
C ₆	0.1 µF		
C ₇	1 µF		
C ₈	1 μF–4.7 μF		



### BOARD DESIGN AND LAYOUT CONSIDERATIONS

#### $V_{CC}$ , $V_{DD}$ , and $V_{IO}$ Pins

The digital and analog power supply lines to the PCM1870 should be bypassed to the corresponding ground pins with 0.1- to 4.7-µF ceramic capacitors or electrolytic capacitors, placed as close to the pins as possible to maximize the dynamic performance of ADC.

#### **AGND and DGND Pins**

To maximize the dynamic performance of the PCM1870, the analog and digital grounds are not connected internally. These grounds should have very low impedance to avoid digital noise feeding back into the analog ground. So, they should be connected directly to each other under the part to reduce the potential of noise problems.

#### AIN1L, AIN1R, AIN2L, and AIN2R Pins

AIN1L, AIN1R, AIN2L, and AIN2R are single-ended inputs. AIN1L and AIN1R can also be used as a monaural differential input. The anti-aliasing low-pass filters are integrated on these inputs to remove the out-of-band noise from the audio. If the performance of these filters is not good enough for an application, appropriate external anti-aliasing filters are needed. The passive RC filter (100  $\Omega$  and 0.01  $\mu$ F to 1 k $\Omega$  and 1000 pF) is used in general. Any pins that are not used in an application should be left open. Do not select open pins through register settings.

#### AOL, AOR, PGINL, and PGINR Pins

When AIN1L, AIN1R, AIN2L, and AIN2R pins are used as microphone inputs with high gain, AOL and AOR may have a large dc offset. It is recommended to locate a dc-blocking capacitor (1- to 10-µF capacitor) between AOL/AOR and PGINL/PGINR. If an application is not affected by dc offset, the PCM1870 does not need the capacitors.

#### V_{COM} Pin

 $1-\mu F$  to  $4.7-\mu F$  capacitor is recommended between V_{COM} and AGND to ensure low source impedance for the ADC common voltage. This capacitor should be located as close as possible to the V_{COM} pin to reduce dynamic errors on the ADC common voltage.

#### BCK (Master Mode) and DOUT Pins

BCK in the master mode and DOUT have adequate load drive capability, but if the BCK and DOUT lines are long, locating a buffer near the PCM1870 and minimizing load capacitance is recommended in order to minimize crosstalk between digital and analog circuits, maximize the dynamic performance of the ADC, and reduce power consumption.

Cł	hanges from Original (May 2007) to Revision A Pag							
•	Added new package in FEATURES section	1						
•	Added new package in DESCRIPTION section	1						
•	Added pinout for YZF package	5						
	Inserted column in TERMINAL FUNCTIONS table for terminal numbers of YZF package.							
•	Added pin numbers for YZF package to connection diagram	43						



24-Apr-2015

# PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
PCM1870RHFR	NRND	VQFN	RHF	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1870	
PCM1870RHFT	NRND	VQFN	RHF	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1870	
PCM1870RHFTG4	NRND	VQFN	RHF	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1870	
PCM1870YZFT	NRND	DSBGA	YZF	24	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	PCM1870	

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.



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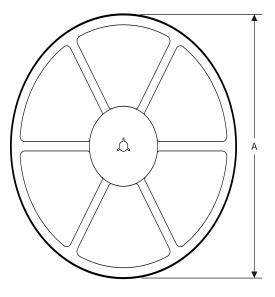
# PACKAGE MATERIALS INFORMATION

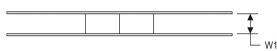
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### TAPE AND REEL INFORMATION

#### REEL DIMENSIONS

Texas Instruments





#### TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

# TAPE AND REEL INFORMATION

*All dimensions are nominal

Device		Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
PCM1870RHFR	VQFN	RHF	24	3000	330.0	12.4	4.3	5.3	1.3	8.0	12.0	Q1
PCM1870RHFT	VQFN	RHF	24	250	180.0	12.4	4.3	5.3	1.3	8.0	12.0	Q1

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# PACKAGE MATERIALS INFORMATION

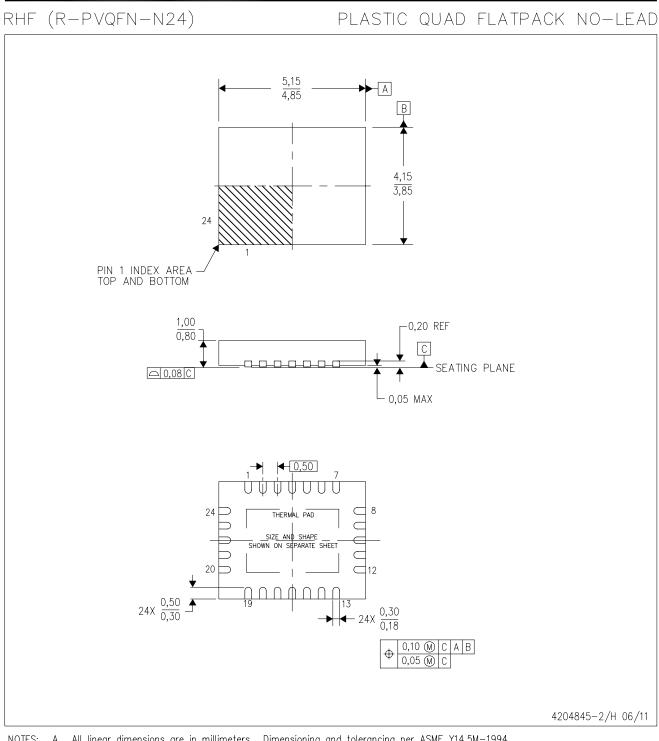
14-Jul-2012



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
PCM1870RHFR	VQFN	RHF	24	3000	367.0	367.0	35.0
PCM1870RHFT	VQFN	RHF	24	250	210.0	185.0	35.0

# **MECHANICAL DATA**



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. QFN (Quad Flatpack No-Lead) Package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.

E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.

F. Falls within JEDEC MO-220.



# RHF (R-PVQFN-N24)

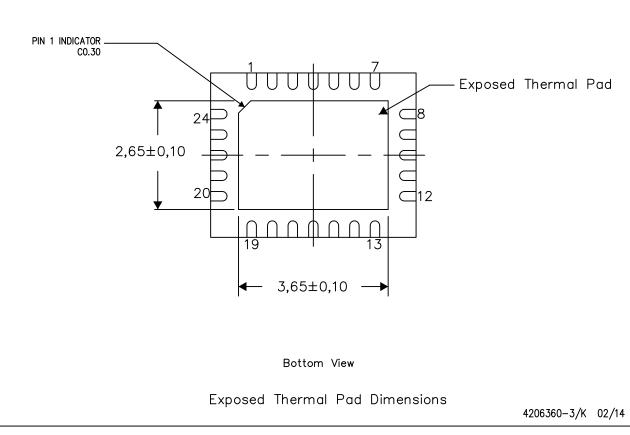
# PLASTIC QUAD FLATPACK NO-LEAD

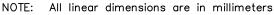
### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

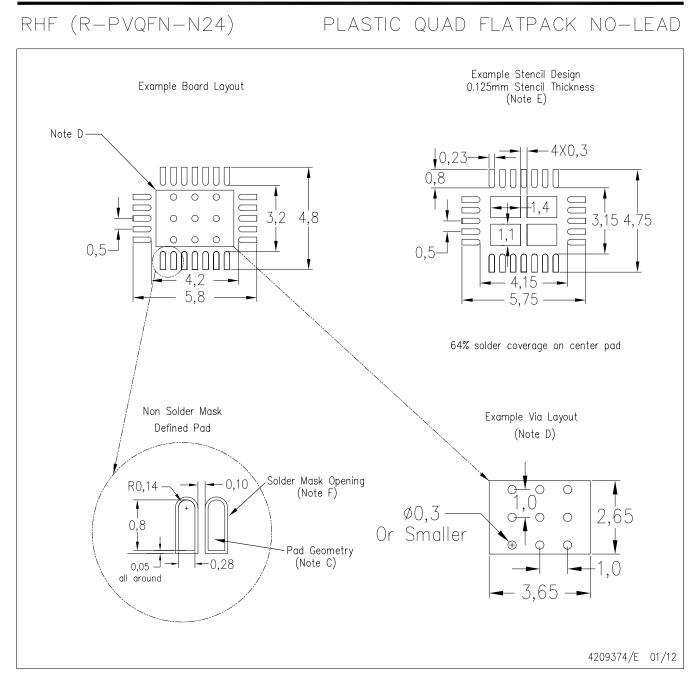
For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.







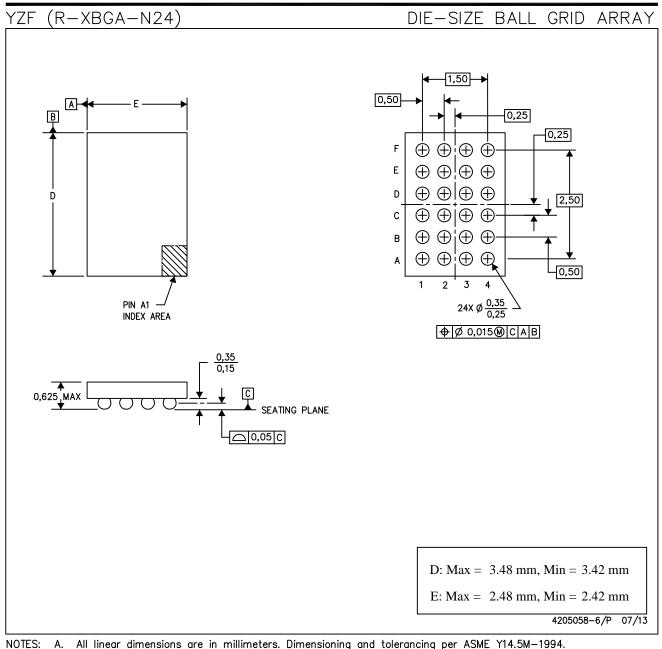


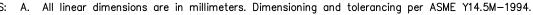
NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in thermal pad.



# **MECHANICAL DATA**





B. This drawing is subject to change without notice.

C. NanoFree™ package configuration.

NanoFree is a trademark of Texas Instruments.



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