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TPS65182, TPS65182B

SLVSAA2D - MARCH 2010 - REVISED JANUARY 2016

# **TPS65182x PMIC For E Ink® Vizplex™-Enabled Electronic Paper Display**

Technical

Documents

#### 1 Features

- Single Chip Power Management Solution for E Ink<sup>®</sup> Vizplex<sup>™</sup> Electronic Paper Displays
- Generates Positive and Negative Gate and Source Driver Voltages and Back-Plane Bias from a Single, Low-Voltage Input Supply
- 3-V to 6-V Input Voltage Range
- Boost Converter for Positive Rail Base
- Inverting Buck-Boost Converter for Negative Rail Base
- Two Adjustable LDOs for Source Driver Supply
  - LDO1: 15 V, 120 mA (VPOS)
  - LDO2: –15 V, 120 mA (VNEG)
- Accurate Output Voltage Tracking
  - VPOS VNEG = ±50 mV
- Two Charge Pumps for Gate Driver Supply
  - CP1: 22 V, 10 mA (VDDH)
  - CP2: –20 V, 12 mA, (VEE)
- Adjustable VCOM Driver for Accurate Panel-**Backplane Biasing** 
  - -0.3 V to -2.5 V
  - Adjustable Through External Potentiometer
  - 15-mA Max Integrated Switch
- Thermistor Monitoring
  - 10°C to 85°C Temperature Range
  - ±1°C Accuracy from 0°C to 50°C
- I<sup>2</sup>C Serial Interface
  - Slave Address 0x48h (1001000)
- Flexible Power-Up Sequencing

Sleep Mode Support

Tools &

Software

Thermally Enhanced Package for Efficient Heat Management (48-Pin 7 mm × 7 mm × 0.9 mm VQFN)

Support &

Community

**.**...

# 2 Applications

- Power Supply for Active Matrix E Ink Vizplex • Panels
- E-Book Readers
- EPSON® S1D13522 (ISIS) Timing Controller
- EPSON S1D13521 (Broadsheet) Timing Controller
- Application Processors With Integrated or Software Timing Controller (OMAP<sup>™</sup>)

# 3 Description

The TPS65182x device is a single-chip power supplies designed to for E Ink Vizplex displays used in portable e-reader applications and support panel sizes up to 9.7 inches. Two high efficiency DC/DC boost converters generate ±17-V rails which are boosted to 22 V and -20 V by two change pumps to provide the gate driver supply for the Vizplex panel.

### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
TPS65182 <sup>(2)</sup>		7.00 mm 7.00 mm		
TPS65182B	VQFN (48)	7.00 mm × 7.00 mm		

- (1) For all available packages, see the orderable addendum at the end of the data sheet.
- (2) Not recommended for new design.

#### Typical Application Schematic VDDH\_D ⊪ VIN -> From Input ⊣⊦ Supply (3 V to 6 V) Г Positive VDDH DRV Charge 1-v/v Pump VDDH FB TS Temperature Sensor VPOS VIN SW LDO1 VCOM DCDC2 VN • vсом VCOM Panel From Input VB SW Supply (3 V to 6 V) VNEG DCDC2 VR 1 002 VEE D I/O Control VCOM VEE DRV Positive Charge Pump VEE FB



2

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# **4** Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

### Changes from Revision C (October 2010) to Revision D

Added ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and 

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### Page



# **5** Description (continued)

Two tracking LDOs create the  $\pm$ 15-V source driver supplies which support up to 120-mA of output current. All rails are adjustable through the I<sup>2</sup>C interface to accommodate specific panel requirements.

Accurate back-plane biasing is provided by a linear amplifier and can be adjusted either by an external resistor or the I<sup>2</sup>C interface. The VCOM driver can source or sink current depending on panel condition.

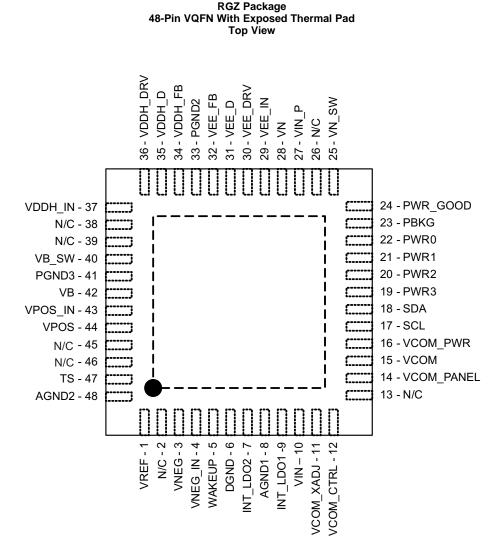
The TPS65182x provides precise temperature measurement function to monitor the panel temperature during operation. The temperature reading is updated every 60 s and can be accessed through the I<sup>2</sup>C interface.

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# 6 Pin Configuration and Functions



**Pin Functions** 

	PIN		DESCRIPTION <sup>(1)</sup>
NO.	NAME	I/O	DESCRIPTION ?
1	VREF	0	Filter pin for 2.25-V internal reference to ADC
2	N/C	—	Not connected
3	VNEG	0	Negative supply output pin for panel source drivers
4	VNEG_IN	I	Input pin for LDO2 (VNEG)
5	WAKEUP	I	Wake up pin (active high). Pull this pin high to wake up from sleep mode.
6	DGND	_	Digital ground
7	INT_LDO2	0	Internal supply (digital circuitry) filter pin
8	AGND1	—	Analog ground for general analog circuitry
9	INT_LDO1	0	Internal supply (analog circuitry) filter pin
10	VIN	I	Input power supply to general circuitry
11	VCOM_XADJ	Ι	Analog input for conventional VCOM setup method. Tie this pin to ground if VCOM is set through $\rm I^2C$ interface.

(1) There will be 0-ns, 93.75-µs, 62.52-µs of deglitch for PWRx, WAKEUP, and VCOM\_CTRL, respectively.



# Pin Functions (continued)

PIN			
NO.	NAME	I/O	DESCRIPTION <sup>(1)</sup>
12	VCOM_CTRL	I	VCOM_PANEL gate driver enable (active high)
13	N/C		Not connected
14	VCOM_PANEL	0	Panel common-voltage output pin
15	VCOM	0	Filter pin for panel common-voltage driver
16	VCOM_PWR	I	Internal supply input pin to VCOM buffer. Connect to the output of DCDC2.
17	SCL	I	Serial interface (I <sup>2</sup> C) clock input
18	SDA	I/O	Serial interface (I <sup>2</sup> C) data input/output
19	PWR3	I	Enable pin for CP1 (VDDH) (active high)
20	PWR2	I	Enable pin for LDO1 (VPOS) (active high)
21	PWR1	I	Enable pin for CP2 (VEE) (active high)
22	PWR0	I	Enable pin for LDO2 (VNEG) and VCOM (active high)
24	PWR_GOOD	0	Open drain power good output pin (active low)
25	VN_SW	0	Inverting buck-boost converter switch out (DCDC2)
26	N/C	_	Not connected
27	VIN_P	I	Input power supply to inverting buck-boost converter (DCDC2)
28	VN	I	Feedback pin for inverting buck-boost converter (DCDC2)
29	VEE_IN	I	Input supply pin for CP1 (VEE)
30	VEE_DRV	0	Driver output pin for negative charge pump (CP2)
31	VEE_D	0	Base voltage output pin for negative charge pump (CP2)
32	VEE_FB	I	Feedback pin for negative charge pump (CP2)
33	PGND2	_	Power ground for CP1 (VDDH) and CP2 (VEE) charge pumps
34	VDDH_FB	I	Feedback pin for positive charge pump (CP1)
35	VDDH_D	0	Base voltage output pin for positive charge pump (CP1)
36	VDDH_DRV	0	Driver output pin for positive charge pump (CP1)
37	VDDH_IN	I	Input supply pin for positive charge pump (CP1)
38	N/C	_	Not connected
39	N/C	_	Not connected
40	VB_SW	0	Boost converter switch out (DCDC1)
41	PGND3	_	Power ground for DCDC1
42	VB	I	Feedback pin for boost converter (DCDC1)
43	VPOS_IN	I	Input pin for LDO1 (VPOS)
44	VPOS	0	Positive supply output pin for panel source drivers
45	N/C	_	Not connected
46	N/C	_	Not connected
47	TS	I	Thermistor input pin. Connect a 10k NTC thermistor and a 43k linearization resistor between this pin and AGND2.
48	AGND2		Reference point to external thermistor and linearization resistor
23	PowerPad (PBKG)	_	Die substrate/thermal pad. Connect to VN with short, wide trace. Wide copper trace will improve heat dissipation. PowerPad must not be connected to ground.

# 7 Specifications

# 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)(2)

	MIN	MAX	UNIT
Input voltage range at VIN, VINP	-0.3	7	V
Ground pins to system ground	-0.3	0.3	V
Voltage range at SDA, SCL, WAKEUP, PWR3, PWR2, PWR1, PWR0, VCOM_CTRL, VDDH_FB, VEE_FB, PWR_GOOD	-0.3	3.6	V
VCOM_XADJ	-3.6	0.3	V
Voltage on VB, VB_SW, VPOS_IN, VDDH_IN	-0.3	20	V
Voltage on VN, VNEG_IN, VEE_IN, VCOM_PWR	-20	0.3	V
Voltage from VINP to VN_SW	-0.3	30	V
Peak output current		Internally limited	mA
Continuous total power dissipation		2	W
T <sub>J</sub> Operating junction temperature	-10	125	°C
T <sub>A</sub> Operating ambient temperature <sup>(3)</sup>	-10	85	°C
T <sub>stg</sub> Storage temperature	-65	150	°C

Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute maximum rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to network ground terminal.

(3) It is recommended that copper plane in proper size on board be in contact with die thermal pad to dissipate heat efficiently. Thermal pad is electrically connected to PBKG, which is supposed to be tied to the output of buck-boost converter. Thus wide copper trace in the buck-boost output will help heat dissipated efficiently.

# 7.2 ESD Ratings

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 $^{\left(2\right)}$	±500	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

# 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
	Input voltage range at VIN, VINP	3	3.7	6	V
	Voltage range at SDA, SCL, WAKEUP, PWR3, PWR2, PWR1, PWR0, VCOM_CTRL, VDDH_FB, VEE_FB, VCOM_XADJ, PWR_GOOD	0		3.6	V
T <sub>A</sub>	Operating ambient temperature	-10		85	°C
TJ	Operating junction temperature	-10		125	°C



# 7.4 Thermal Information

		TPS65182x	
	THERMAL METRIC <sup>(1)</sup>	RGZ (VQFN)	UNIT
		48 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance (2)	30.5	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	16.2	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	7.1	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	0.2	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	7.1	°C/W
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	_	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

(2) Estimated when mounted on high K JEDEC board per JESD 51-7 with thickness of 1.6 mm, 4 layers, size of 76.2 mm X 114.3 mm, and 2 oz. copper for top and bottom plane. Actual thermal impedance will depend on PCB used in the application.

# 7.5 Electrical Characteristics

 $V_{IN} = 3.7 \text{ V}, T_A = -10^{\circ}\text{C}$  to 85°C, Typical values are at  $T_A = 25^{\circ}\text{C}$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INPUT VO	LTAGE					
V <sub>IN</sub>	Input voltage range		3	3.7	6	V
V <sub>UVLO</sub>	Undervoltage lockout threshold	V <sub>IN</sub> falling		2.9		V
V <sub>HYS</sub>	Undervoltage lockout hysteresis	V <sub>IN</sub> rising		400		mV
INPUT CU	RRENT					
l <sub>Q</sub>	Operating quiescent current into VIN	Device switching, no load		5.5		mA
I <sub>STD</sub>	Operating quiescent current into VIN	Device in standby mode		130		μA
I <sub>SLEEP</sub>	Shutdown current	Device in sleep mode		2.8	10	μA
INTERNAL	. SUPPLIES					
VI <sub>NT_LDO1</sub>	Internal supply			2.7		V
V <sub>INT_LDO2</sub>	Internal supply			2.7		V
V <sub>REF</sub>	Internal supply			2.25		V
DCDC1 (P	OSITIVE BOOST REGULATOR)					
V <sub>IN</sub>	Input voltage range		3	3.7	6	V
N/	Output voltage range			17		V
V <sub>OUT</sub>	DC set tolerance		-5%		5%	
I <sub>OUT</sub>	Output current				160	mA
R <sub>DS(ON)</sub>	MOSFET on resistance	V <sub>IN</sub> = 3.7 V		350		mΩ
	Switch current limit			1.5		А
I <sub>LIMIT</sub>	Switch current accuracy		-30%		30%	
f <sub>SW</sub>	Switching frequency			1		MHz
L	Inductor			2.2		μH
С	Capacitor			2x4.7		μF
ESR	Capacitor ESR			20		mΩ
DCDC2 (IN	IVERTING BUCK-BOOST REGULATO	R)				
V <sub>IN</sub>	Input voltage range		3	3.7	6	V
V	Output voltage range			-17		V
V <sub>OUT</sub>	DC set tolerance		-5%		5%	
I <sub>OUT</sub>	Output current				160	mA
R <sub>DS(ON)</sub>	MOSFET on resistance	V <sub>IN</sub> = 3.7 V		350		mΩ
	Switch current limit			1.5		А
ILIMIT	Switch current accuracy		-30%		30%	

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# **Electrical Characteristics (continued)**

# $V_{IN}$ = 3.7 V, $T_A$ = -10°C to 85°C, Typical values are at $T_A$ = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
L	Inductor			4.7		μH
С	Capacitor			2x4.7		μF
ESR	Capacitor ESR			20		mΩ
LDO1 (VPC	DS)					
V <sub>POS_IN</sub>	Input voltage range		16.15	17	17.85	V
V <sub>SET</sub>	Output voltage set value	$V_{IN} = 17 V$	14.25	15	15.75	V
VINTERVAL	Output voltage set resolution	$V_{IN} = 17 V$		250		mV
V <sub>POS_OUT</sub>	Output voltage range	$V_{SET}$ = 15 V, $I_{LOAD}$ = 20 mA	14.85	15	15.15	V
V <sub>OUTTOL</sub>	Output tolerance	$V_{SET}$ = 15 V, $I_{LOAD}$ = 20 mA	-1%		1%	
V <sub>DROPOUT</sub>	Dropout voltage	$I_{LOAD} = 120 \text{ mA}$			250	mV
V <sub>LOADREG</sub>	Load regulation – DC	I <sub>LOAD</sub> = 10% to 90%			1%	
I <sub>LOAD</sub>	Load current range			120		mA
I <sub>LIMIT</sub>	Output current limit		200			mA
T <sub>SS</sub>	Soft start time			1		ms
С	Recommended output capacitor			4.7		μF
LDO2 (VN	EG)	·			1	
V <sub>NEG_IN</sub>	Input voltage range		-17.85	-17	-16.15	V
V <sub>SET</sub>	Output voltage set value	V <sub>IN</sub> = -17 V	-15.75	-15	-14.25	V
VINTERVAL	Output voltage set resolution	$V_{IN} = -17 \text{ V}$		250		mV
V <sub>NEG_OUT</sub>	Output voltage range	V <sub>SET</sub> = -15 V, I <sub>LOAD</sub> = -20 mA	-15.15	-15	-14.85	V
V <sub>OUTTOL</sub>	Output tolerance	$V_{SET} = -15 \text{ V}, \text{ I}_{LOAD} = -20 \text{ mA}$	-1%		1%	
VDROPOUT	Dropout voltage	$I_{LOAD} = 120 \text{ mA}$			250	mV
VLOADREG	Load regulation – DC	$I_{LOAD} = 10\%$ to 90%			1%	
ILOAD	Load current range			120		mA
ILIMIT	Output current limit		200			mA
T <sub>SS</sub>	Soft start time			1		ms
C	Recommended output capacitor			4.7		μF
LD01 (POS	6) AND LDO2 (VNEG) TRACKING	ł	1		I	
V <sub>DIFF</sub>	Difference between VPOS and VNEG	$V_{SET} = \pm 15 \text{ V},$ $I_{LOAD} = \pm 20 \text{ mA}, 0^{\circ}\text{C} \text{ to } 60^{\circ}\text{C}$	-50		50	mV
VCOM DRI						
V <sub>COM</sub>	Output voltage range		-2.5		-0.3	V
G	V <sub>COM</sub> gain (V <sub>COM_XADJ</sub> /V <sub>COM</sub> )	$V_{COM_{ADJ}} = 0 V$		1		V/V
VCOM SW	ІТСН	I	1			
T <sub>ON</sub>	Switch ON time				1	ms
R <sub>DS(ON)</sub>	MOSFET ON resistance	$V_{COM} = -1.25 \text{ V}, I_{COM} = 30 \text{ mA}$		20	35	Ω
I <sub>LIMIT</sub>	MOSFET current limit	Not tested in production		200		mA
I <sub>SWLEAK</sub>	Switch leakage current	$V_{COM} = 0 V,$ $V_{COM_{PANEL}} = -2.5 V$			8.3	nA
CP1 (VDD	H) CHARGE PUMP					
V <sub>DDH_IN</sub>	Input voltage range		16.15	17	17.85	V
	Feedback voltage			1		V
V <sub>FB</sub>	Accuracy		-3%		3%	
V <sub>DDH_OUT</sub>	Output voltage range	$V_{SET} = 22 \text{ V}, \text{ I}_{LOAD} = 2 \text{ mA}$	21	22	23	V
I <sub>LOAD</sub>	Load current range				10	mA
		1	1			

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# **Electrical Characteristics (continued)**

$V_{IN} = 3.7 \text{ V}, T_A = -10^{\circ}\text{C} \text{ to } 85^{\circ}\text{C}$	Typical values are at T 25°C	(unless otherwise noted)
$v_{\rm N} = 5.7 v$ , $r_{\rm A} = -10 \ 0.000 \ 0$	Typical values are at $T_A = 25$ C	

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
CD	Recommended driver capacitor			10		nF
Co	Recommended output capacitor			4.7		μF
CP2 (VEE)	NEGATIVE CHARGE PUMP	•				
V <sub>EE_IN</sub>	Input voltage range		-17.75	-17	-16.15	V
M	Feedback voltage			-1		V
V <sub>FB</sub>	Accuracy		-3%		3%	
V <sub>EE_OUT</sub>	Output voltage range	$V_{SET} = -20 \text{ V}, \text{ I}_{LOAD} = 3 \text{ mA}$	-21	-20	-19	V
I <sub>LOAD</sub>	Load current range				12	mA
f <sub>SW</sub>	Switching frequency			560		KHz
CD	Recommended driver capacitor			10		nF
Co	Recommended output capacitor			4.7		μF
THERMIST	OR MONITOR <sup>(1)</sup>	·				
A <sub>TMS</sub>	Temperature to voltage ratio	Not tested in production		-0.0158		V/°C
Offset <sub>TMS</sub>	Offset	Temperature = 0°C		1.575		V
V <sub>TMS_HOT</sub>	Temp hot trip voltage (T = 50°C)	TEMP_HOT_SET = 0x8C		0.768		V
V <sub>TMS_COOL</sub>	Temp hot escape voltage (T = $45^{\circ}$ C)	TEMP_COOL_SET = 0x82		0.845		V
V <sub>TMS_MAX</sub>	Maximum input level			2.25		V
R <sub>NTC_PU</sub>	Internal pull up resistor			7.307		ΚΩ
R <sub>LINEAR</sub>	External linearization resistor			43		KΩ
ADC <sub>RES</sub>	ADC resolution	Not tested in production, 1 bit		16.1		mV
ADC <sub>DEL</sub>	ADC conversion time	Not tested in production		19		μs
TMST <sub>TOL</sub>	Accuracy	Not tested in production	-1		1	LSB
LOGIC LEV	ELS AND TIMING CHARTERISTICS (	SCL, SDA, PWR_GOOD, PWRx, WAKEU	IP)			
V <sub>OL</sub>	Output low threshold level	I <sub>O</sub> = 3 mA, sink current (SDA, PWR_GOOD)			0.4	V
V <sub>IL</sub>	Input low threshold level				0.4	V
V <sub>IH</sub>	Input high threshold level		1.2			V
I <sub>(bias)</sub>	Input bias current	V <sub>IO</sub> = 1.8 V			1	μA
t <sub>low,WAKEUP</sub>	WAKEUP low time	minimum low time for WAKEUP pin	150			ms
f <sub>SCL</sub>	SCL clock frequency				400	KHz
OSCILLAT	DR		<u> </u>			
f <sub>OSC</sub>	Oscillator frequency			9		MHz
	Frequency accuracy	$T_A = -40^{\circ}C$ to $85^{\circ}C$	-10%		10%	
THERMAL	SHUTDOWN	•	<u> </u>			
T <sub>SHTDWN</sub>	Thermal trip point			150		°C
	Thermal hysteresis			20		°C

 10-KΩ Murata NCP18XH103F03RB thermistor (1%) in parallel with a linearization resistor (43 KΩ, 1%) are used at TS pin for panel temperature measurement.

### TPS65182, TPS65182B

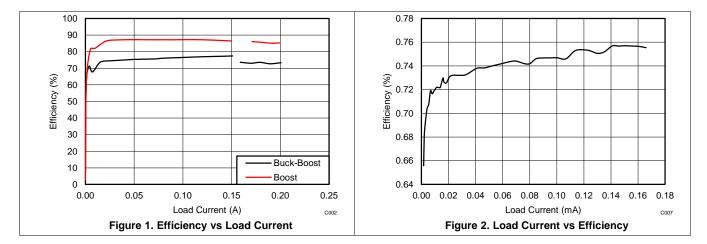
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# 7.6 Data Transmission Timing

 $V_{BAT}$  = 3.6 V ±5%,  $T_A$  = 25°C,  $C_L$  = 100 pF (unless otherwise noted)

			MIN	NOM MAX	UNI
f <sub>(SCL)</sub>	Serial clock frequency		100	4	00 KH2
	Hold time (repeated) START condition. After this	SCL = 100 KHz	4		μs
t <sub>HD;STA</sub>	period, the first clock pulse is generated.	SCL = 400 KHz	600		ns
	LOW pariod of the SCL clock	SCL = 100 KHz	4.7		
t <sub>LOW</sub>	LOW period of the SCL clock	SCL = 400 KHz	1.3		μs
•	HIGH period of the SCL clock	SCL = 100 KHz	4		μs
t <sub>HIGH</sub>		SCL = 400 KHz	600		ns
	Set up time for a repeated STADT condition	SCL = 100 KHz	4.7		μs
t <sub>SU;STA</sub>	Set-up time for a repeated START condition	SCL = 400 KHz	600		ns
	Data hold time	SCL = 100 KHz	0	3.	45 µs
t <sub>HD;DAT</sub>		SCL = 400 KHz	0	9	00 ns
	Data aat un time	SCL = 100 KHz	250		
t <sub>SU;DAT</sub>	Data set-up time	SCL = 400 KHz	100		ns
	Disa time of both CDA and CCL signals	SCL = 100 KHz		10	
t <sub>r</sub>	Rise time of both SDA and SCL signals	SCL = 400 KHz		3	ns 00
	Fall time of both SDA and SCL signals	SCL = 100 KHz		3	00
t <sub>f</sub>	Fall time of both SDA and SCL signals	SCL = 400 KHz		3	ns 00
	Cat up time for CTOD and lities	SCL = 100 KHz	4		μs
t <sub>SU;STO</sub>	Set-up time for STOP condition	SCL = 400 KHz	600		ns
	Due Free Time Deturger Stee and Start Condition	SCL = 100 KHz	4.7		
t <sub>BUF</sub>	Bus Free Time Between Stop and Start Condition	SCL = 400 KHz	1.3		μs
	Pulse width of spikes which mst be suppressed	SCL = 100 KHz	n/a	r	/a
t <sub>SP</sub>	by the input filter	SCL = 400 KHz	0		50 ns
<u> </u>	Conceptive lead for each bus line	SCL = 100 KHz		4	00 pr
C <sub>b</sub>	Capacitive load for each bus line	SCL = 400 KHz		4	00 pF

# 7.7 Typical Characteristics



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# 8 Detailed Description

## 8.1 Overview

The TPS65182x family of devices provides two adjustable LDOs, inverting buck-boost converter, boost converter, thermistor monitoring, and flexible power-up and power-down sequencing. The system can be supplied by a regulated input voltage ranging from 3 V to 6 V. The device is characterized across a  $-10^{\circ}$ C to 85°C temperature range, best suited for personal electronic applications.

The  $I_2C$  interface provides comprehensive features for using the TPS65182x family of devices. All rails can be enabled or disabled. Power-up and power-down sequences can also be programmed through the  $I_2C$  interface, as well as thermistor and interrupt configurations. Voltage adjustment can also be controlled through the  $I_2C$  interface.

The adjustable LDOs can supply up to 120 mA of current. The default output voltages for each LDO can be adjusted through the  $I_2C$  interface. LDO1 (VPOS) and LDO2 (VNEG) track each other in a way that they are of opposite sign, but same magnitude. The sum of VLDO1 and VLOD2 is guaranteed to be less than 50 mv.

There are two charge pumps: VDDH and VEE 10 mA and 12 mA respectively. These charge pumps boost the DC-DC boost converters ±16-V rails to provide a gate channel supply. The power good functionality is open-drain output, if any of the four power rails (CP1, CP2, LDO1, LDO2) are not in regulation, encounters a fault, or is disabled, the pin is pulled low. PWR\_GOOD remains low if one of the rails is not enabled by the host, and only after all rails are in regulation, PWR\_GOOD is released to HiZ state (pulled up by external resistor).

The TPS65182x family of devices provides circuitry to bias and measure an external NTC to monitor the display panel temperature in a range from  $-10^{\circ}$ C to  $85^{\circ}$ C with an accuracy of  $\pm 1^{\circ}$ C from  $0^{\circ}$ C to  $50^{\circ}$ C. Temperature measurements are triggered by the controlling host and the last temperature reading is always stored in the TMST\_VALUE register. Interrupts are issued when the temperature exceeds the programmable HOT, or drops below the programmable COLD threshold, or when the temperature has changed by more than a user-defined threshold from the baseline value.

# TPS65182, TPS65182B

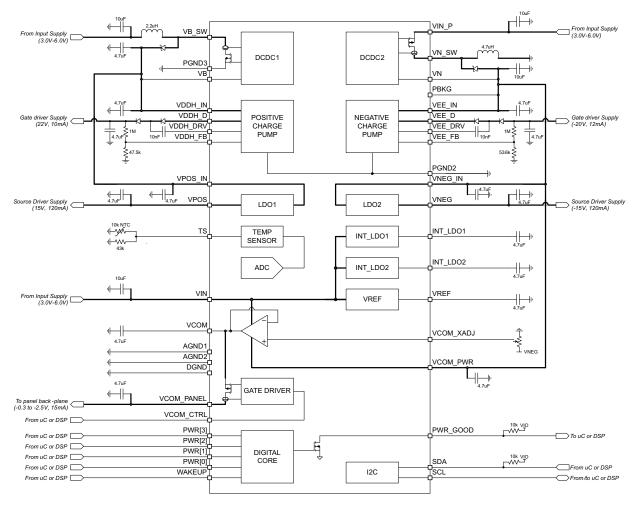
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INSTRUMENTS

Texas

# 8.2 Functional Block Diagram



# 8.3 Feature Description

## 8.3.1 Modes of Operation

The TPS65182x has three modes of operation, SLEEP, STANDBY, and ACTIVE. SLEEP mode is the lowestpower mode in which all internal circuitry is turned off. In STANDBY, all power rails are shut down but the device is ready to accept commands through PWR[3:0] pins and/or I<sup>2</sup>C interface. In ACTIVE mode one or more power rails are enabled.

- **SLEEP** This is the lowest power mode of operation. All internal circuitry is turned off and the device does not respond to I<sup>2</sup>C communications. TPS65182x enters SLEEP mode whenever WAKEUP pin is pulled low.
- **STANDBY** In STANDBY all internal support circuitry is powered up and the device is ready to accept commands either through GPIO or I<sup>2</sup>C control but none of the power rails are enabled. To enter STANDBY mode the WAKEUP pin must be pulled high and all PWRx pins must be pulled low. The device also enters STANDBY mode if input under voltage lock out (UVLO), positive boost under voltage (VB\_UV), or inverting buck-boost under voltage (VN\_UV) is detected, or thermal shutdown occurs.
- ACTIVE The device is in ACTIVE mode when any of the output rails are enabled and no fault condition is present. This is the normal mode of operation while the device is powered up. In ACTIVE mode, a falling edge on any PWRx pin shuts down and a rising edge powers up the corresponding rail.



### Feature Description (continued)

### 8.3.2 Mode Transistions

- SLEEP → ACTIVE WAKEUP pin is pulled high (rising edge) with any PWRx pin high. Rails come up in a predefined power-up sequence.
- SLEEP → STANDBY WAKEUP pin is pulled high (rising edge) with all PWRx pins low. Rails will remain down until one or more PWRx pin is pulled high.
- ACTIVE → SLEEP WAKEUP pin is pulled low (falling edge). Rails are shut down following the pre-defined power-down sequence.
- ACTIVE → STANDBY WAKEUP pin is high. All PWRx pins are pulled low (falling edge). Rails shut down in the order in which PWRx pins are pulled low. In the event of thermal shut down (TSD), under voltage lock out (UVLO), positive boost or inverting buck-boost under voltage (UV), the device shuts down all rails in a pre-defined power-down sequence.
- **STANDBY** → **ACTIVE** WAKEUP pin is high and any PWRx pin is pulled high (rising edge). Rails come up in the same order as PWRx pins are pulled high.
- **STANDBY**  $\rightarrow$  **SLEEP** WAKEUP pin is pulled low (falling edge) while none of the output rails are enabled.

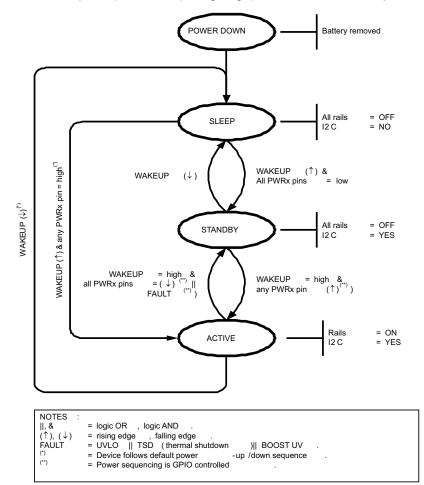


Figure 3. Global State Diagram



## Feature Description (continued)

### 8.3.3 Wake-Up and Power Up Sequencing

The TPS65182x supports a default power-up sequence supporting E Ink Vizplex displays. It also offers full user control of the power-up sequence through GPIO control using the PWR3, 2, 1, 0 pins. Using GPIO control, the output rails are enabled/disabled in the order in which the PWRx pins are asserted/de-asserted, respectively, and the power-up timing is controlled by the host only. Rails are in regulation 2 ms after their respective PWRx pin has been asserted with the exception of the first rail, which takes 6 ms to power up. The additional time is needed to power up the positive and inverting buck-boost regulator which need to be turned on before any other rail can be enabled. Once all rails are enabled and in regulation the PWR\_GOOD pin is released (pin status = HiZ and power good line is pulled high by external pull-up resistor). The PWRx pins are assigned to the rails as follows:

- PWR0: LDO2 (VNEG) and VCOM
- PWR1: CP2 (VEE)
- PWR3: LDO2 (VPOS)
- PWR4: CP1 (VDDH)

Rails are powered down whenever the host de-asserts the respective PWRx pin, and once all rails are disabled the device enters STANDBY mode. The next step is then to de-assert the WAKEUP pin to enter SLEEP mode which is the lowest-power mode of operation.

It is possible for the host to force the TPS65182x directly into SLEEP mode from ACTIVE mode by de-asserting the WAKEUP pin in which case the device follows the pre-defined power-down sequence before entering SLEEP mode.

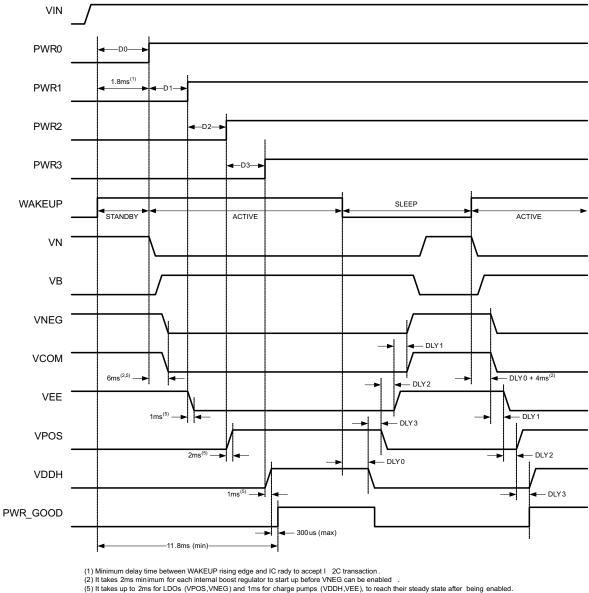
### 8.3.4 Dependencies Between Rails

Charge pumps, LDOs, and VCOM driver are dependent on the positive and inverting buck-boost converters and several dependencies exist that affect the power-up sequencing. These dependencies are listed below.

- 1. Inverting buck-boost (DCDC2) must be in regulation before positive boost (DCDC1) can be enabled. Internally, DCDC1 enable is gated by DCDC2 power good.
- 2. Positive boost (DCDC1) must be in regulation before LDO2 (VNEG) can be enabled. Internally LDO2 enable is gated DCDC1 power-good.
- 3. Positive boost (DCDC1) must be in regulation before VCOM can be enabled; Internally VCOM enable is gated by DCDC1 power good.
- 4. Positive boost (DCDC1) must be in regulation before negative charge pump (CP2) can be enabled. Internally CP2 enable is gated by DCDC1 power good.
- 5. Positive boost (DCDC1) must be in regulation before positive charge pump (CP1) can be enabled. Internally CP1 enable is gated by DCDC1 power good.
- 6. LDO2 must be in regulation before LDO1 can be enabled. Internally LDO1 enable is gated by LDO2 power good.
- 7. The minimum delay time between any two PWRx pins must be > 62.5 µs in order to follow the power up sequence defined by GPIO control. If any two PWRx pins are pulled up together (< 62.5 µs apart) rails will be staggered in a manner that a subsequent rail's enable is gated by PG of a preceding rail. In this case, the default order of power-up is LDO2 (VNEG), CP2 (VEE), LDO1 (VPOS), and CP1(VDDH). If any two PWRx pins are pulled low then all rails will go down at the same time.</p>



### Feature Description (continued)



DLY 0-DLY 3 are power up /down delays are factory-set to 2ms.

Figure 4. Power-Up and Power-Down Timing Diagram

### 8.3.5 Soft-Start

Softstart for DCDC1, DCDC2, LDO1, and LDO2 is accomplished by lowering the current limits during start-up. If DCDC1 or DCDC2 are unable to reach power-good status within 10 ms, the device enters STANDBY mode.

### 8.3.6 VCOM Adjustment

VCOM can be adjusted by an external potentiometer by connecting a potentiometer to the VCOM\_XADJ pin. The potentiometer must be connected between ground and a negative supply. The gain from VCOM\_XADJ to VCOM is 1 and therefore the voltage applied to VCOM\_XADJ pin should range from -0.3 to -2.5V.

### 8.3.7 VPOS and VNEG Supply Tracking

LDO1 (VPOS) and LDO2 (VNEG) track each other in a way that they are of opposite sign but same magnitude. The sum of VLDO1 and VLOD2 is guaranteed to be < 50 mV.

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## Feature Description (continued)

### 8.3.8 Fault Handling and Recovery

The TPS65182x monitors input and output voltages and die temperature and will take action if operating conditions are outside normal limits. Whenever the TPS65182x encounters:

- Thermal Shutdown (TSD)
- Positive Boost Under Voltage (VB\_UV)
- Inverting Buck-Boost Under Voltage (VN\_UV)
- Input Under Voltage Lock Out (UVLO)

it will shut down all power rails and enter STANDBY mode. Shut down follows the pre-defined power-down sequence and once a fault is detected, the PWR\_GOOD pin is pulled low.

Whenver the TPS65182x encounters under voltage on VNEG (VNEG\_UV), VPOS (VPOS\_UV), VEE (VEE\_UV) or VDDH (VDDH\_UV) it will shut down the corresponding rail (plus any dependent rail) only and remain in ACTIVE mode, allowing the DCDC converters to remain up. Again, the PWR\_GOOD pin will be pulled low.

As the PWRx inputs are edge sensitive, the host must toggle the PWRx pins to re-enable the rails through GPIO control, i.e. it must bring the PWRx pins low before asserting them again.

### 8.3.9 Power Good Pin

The power good pin (PWR\_GOOD) is an open drain output that is pulled high when all four power rails (CP1, CP2, LDO1, LDO2) are in regulation and is pulled low if any of the rails encounters a fault. PWR\_GOOD remains low if one of the rails is not enabled by the host and only after all rails are in regulation PWR\_GOOD is released to HiZ state (pulled up by external resistor).

### 8.3.10 Panel Temperature Monitoring

The TPS65182x provides circuitry to bias and measure an external negative temperature coefficient resistor (NTC) to monitor device temperature in a range from  $-10^{\circ}$ C to  $85^{\circ}$ C with and accuracy of  $\pm 1^{\circ}$ C from  $0^{\circ}$ C to  $50^{\circ}$ C. Temperature reading is automatically updated every 60 s.

### 8.3.11 NTC Bias Circuit

Figure 5 below shows the block diagram of the NTC bias and measurement circuit. The NTC is biased from an internally generated 2.25-V reference voltage through an integrated 7.307-K $\Omega$  bias resistor. A 43-K $\Omega$  resistor is connected parallel to the NTC to linearize the temperature response curve. The circuit is designed to work with a nominal 10-K $\Omega$  NTC and achieves accuracy of ±1°C from 0°C to 50°C. The voltage drop across the NTC is digitized by a 10-bit SAR ADC and translated into an 8-bit two's complement by digital per Table 1.



## Feature Description (continued)

TEMPERATURE	TMST_VALUE[7:0]
< -10°C	1111 0110
-10°C	1111 0110
-9°C	1111 0111
-2°C	1111 1110
-1°C	1111 1111
0°C	0000 0000
1°C	0000 0001
2°C	0000 0010
25°C	0001 1001
85°C	0101 0101
> 85°C	0101 0101

### Table 1. ADC Output Value vs Termperature

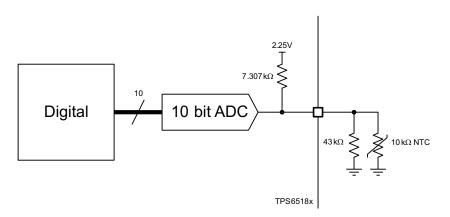


Figure 5. NTC Bias and Measurement Circuit

## 8.4 Device Functional Modes

## 8.4.1 I<sup>2</sup>C Bus Operation

The TPS65182x supports a special I<sup>2</sup>C mode making it compatible with the EPSON<sup>®</sup> Broadsheet S1D13521 timing controller. Standard I<sup>2</sup>C protocol requires the following steps to read data from a register:

- 1. Send device slave address, R/nW bit set low (write command)
- 2. Send register address
- 3. Send device slave address, R/nW set high (read command)
- 4. The slave will respond with data from the specified register address.end device slave address, R/nW set high (read command).

The EPSON<sup>®</sup> Broadsheet S1D13521 controller does not support I<sup>2</sup>C writes nor I<sup>2</sup>C reads from addressed registers, therefore the TPS65182x I<sup>2</sup>C interface has been modified and the reading the temperature data is reduced to two steps:

- 1. Send device address, R/nW set high (read command)
- 2. Read the data from the slave. The slave will respond with data from TMST\_VALUE register address.

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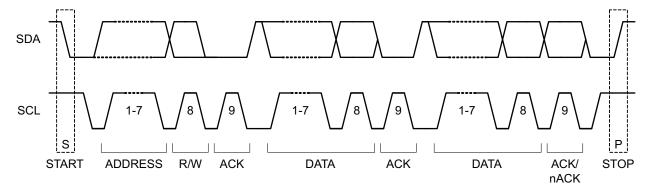
## **Device Functional Modes (continued)**

S	SLAVE ADDRESS	R	А	DATA 0		А	Ρ
	From master to slave		R	Read	S Sta	art	
	From slave to master		А	Acknowlege	P Sto	р	

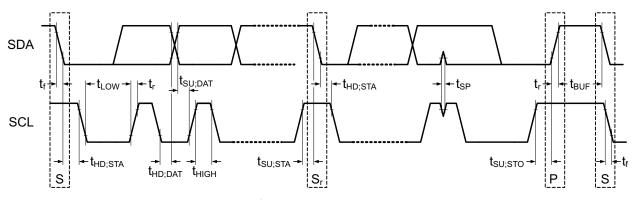
Figure 6. Subaddress in I<sup>2</sup>C Transmission

The I<sup>2</sup>C Bus is a communications link between a controller and a series of slave terminals. The link is established using a two-wired bus consisting of a serial clock signal (SCL) and a serial data signal (SDA). The serial clock is sourced from the controller in all cases where the serial data line is bi-directional for data communication between the controller and the slave terminals. Each device has an open Drain output to transmit data on the serial data line. An external pull-up resistor must be placed on the serial data line to pull the drain output high during data transmission.

Data transmission is initiated with a start bit from the controller as shown in Figure 7. The start condition is recognized when the SDA line transitions from high to low during the high portion of the SCL signal. Upon reception of a start bit, the device will receive serial data on the SDA input and check for valid address and control information. If the appropriate group and address bits are set for the device, then the device will issue an acknowledge pulse and prepare the receive subaddress data. Subaddress data is decoded and responded to as per the Register Map section of this document. Data transmission is completed by either the reception of a stop condition or the reception of the data word sent to the device. A stop condition is recognized as a low to high transition of the SDA input during the high portion of the SCL signal. All other transitions of the SDA line must occur during the low portion of the SCL signal. An acknowledge is issued after the reception of valid address, sub-address and data words. Reference Figure 7.











# 8.5 Register Maps

# Table 2. Register Address Map

REGISTER	ADDRESS (HEX)	NAME	DEFAULT VALUE	DESCRIPTION
0	0x00	TMST_VALUE	N/A	Thermistor value read by ADC

## 8.5.1 Thermistor Readout (TMST\_VALUE) Register (offset = 0x00h)

DATA BIT	D7	D6	D5	D4	D3	D2	D1	D0		
FIELD NAME		TMST_VALUE[7:0]								
READ/WRITE	R	R	R	R	R	R	R	R		
RESET VALUE	N/A	N/A	N/A	N/A	N/A	N/A	N/A	N/A		

FIELD NAME	BIT DEFINITION
	Temperature read-out
	1111 0110 – < -10°C
	1111 0110 – -10°C
	1111 0111 – -9°C
	1111 1110 – -2°C
	1111 1111 – -1 °C
TMST_VALUE[7:0]	0000 0000 – 0 °C
	0000 0001 – 1°C
	0000 0010 – 2°C
	0001 1001 – 25°C
	0101 0101 – 85°C
	0101 0101 – > 85°C



# 9 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 9.1 Application Information

The TPS65185 device is used to power display screens in E-book applications, specifically E-lnk Vizplex display, by connecting the screen to the positive and negative charge pump, LDOs 1 and 2, and VCOM rails. The display screens size that can be supported up to 9.7 inches.

# 9.2 Typical Application

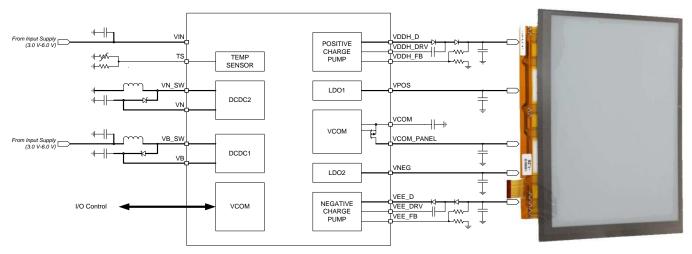


Figure 9. Typical Application Schematic

## 9.2.1 Design Requirements

For this design example, use the parameters listed in Table 3 as the input parameters.

### **Table 3. Design Parameters**

	VOLTAGE	SEQUENCE (STROBE)
VNEG (LDO2)	–15 V	1
VEE (Charge pump 2)	−20 V	2
VPOS (LDO1)	15 V	3
VDDH (Charge pump 1)	22 V	4

### 9.2.2 Detailed Design Procedure

For the positive boost regulator (DCDC1) a 10- $\mu$ F capacitor can be used as the input capacitor value; two 4.7- $\mu$ F capacitor are used as output capacitors to reduce ESR along with a 2.2- $\mu$ H inductor. For the inverting buck-boost regulator (DCDC2), an 10- $\mu$ F capacitor can be used at the input capacitor value; A 10- $\mu$ F and 4.7- $\mu$ F capacitor are used as output capacitors to reduce ESR, with a 4.7- $\mu$ H inductor. Capacitor ESR for all capacitors should be around 20 m $\Omega$ , and ceramic X5R material. These are the typical the values used, additional inductor and capacitor values can be used for improved functionality, but the parts should be rated the same as the recommended external components listed in Table 4.

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TPS65182, TPS65182B

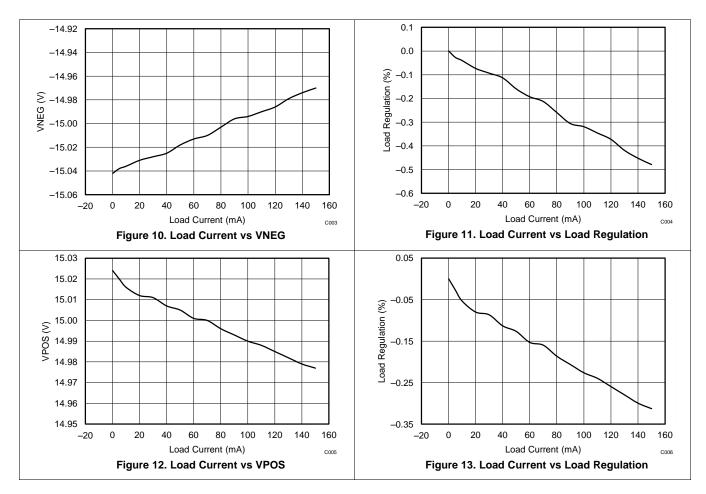
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### Table 4. Recommended External Components

PART NUMBER	VALUE	SIZE	MANUFACTURER
INDUCTORS			
LQH44PN4R7MP0	4.7 µH	4 mm × 4 mm × 1.65 mm	Murata
NR4018T4R7M	4.7 µH	4 mm × 4 mm × 1.8 mm	Taiyo Yuden
VLS252015ET-2R2M	2.2 µH	2 mm × 2.5 mm × 1.5 mm	TDK
NR4012T2R2M	2.2 µH	4 mm × 4 mm × 1.2 mm	Taiyo Yuden
CAPACITORS			•
GRM21BC81E475KA12L	4.7 µF, 25 V, X6S	805	Murata
GRM32ER71H475KA88L	4.7 µF, 50 V, X7R	1210	Murata
All other capacitors	X5R or better	—	—
DIODES			·
BAS3010	—	SOD-323	Infineon
MBR130T1	_	SOD-123	ON-Semi
BAV99	_	SOT-23	Fairchild
THERMISTOR			
NCP18XH103F03RB	10 kΩ	603	Murata

### 9.2.3 Application Curves





# **10 Power Supply Recommendations**

The device is designed to operate with an input voltage supply range from 3 V to 6 V. This input supply can be from a externally regulated supply. If the input supply is located more than a few inches from the TPS65185, additional bulk capacitance may be required in addition to the ceramic bypass capacitors. An electrolytic capacitor with a value of 10  $\mu$ F is a typical choice.

# 11 Layout

# 11.1 Layout Guidelines

The layout guidelines for TPS65182x are as follows:

- PBKG (Die substrate must connect to VN (–16 V) with short, wide trace. Wide copper trace will improve heat dissipation.
- Power pad is internally connected to PBKG and must be connected to ground, but connected to VN with a short wide copper trace.
- Inductor traces must be kept on the PCB top layer free of any vias.
- Feedback traces must be routed away from any potential noise source to avoid coupling.
- Output caps must be placed immediately at output pin.
- VIN pins must be bypassed to ground with low ESR ceramic bypass capacitors.



# 11.2 Layout Example

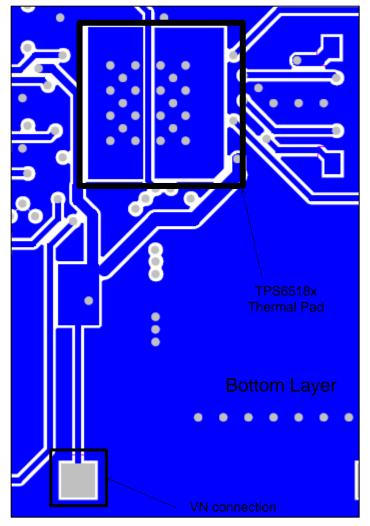


Figure 14. Typical Layout of TPS6518x



# **12 Device and Documentation Support**

### 12.1 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
TPS65182	Click here	Click here	Click here	Click here	Click here
TPS65182B	Click here	Click here	Click here	Click here	Click here

### Table 5. Related Links

### **12.2 Community Resources**

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E<sup>™</sup> Online Community *TI's Engineer-to-Engineer (E2E) Community.* Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support TI's Design Support** Quickly find helpful E2E forums along with design support tools and contact information for technical support.

### 12.3 Trademarks

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E lnk is a registered trademark of E lnk Corporation.

EPSON is a registered trademark of Seiko Epson Corporation.

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### 12.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

# 12.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



23-Oct-2015

# PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TPS65182BRGZR	ACTIVE	VQFN	RGZ	48	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-10 to 85	E INK TPS65182B	Samples
TPS65182BRGZT	ACTIVE	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-10 to 85	E INK TPS65182B	Samples
TPS65182RGZR	NRND	VQFN	RGZ	48	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-10 to 85	E INK TPS65182	
TPS65182RGZT	NRND	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-10 to 85	E INK TPS65182	

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.



23-Oct-2015

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# **TAPE AND REEL INFORMATION**





# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal Device	1	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS65182BRGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
TPS65182BRGZT	VQFN	RGZ	48	250	180.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
TPS65182RGZR	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
TPS65182RGZT	VQFN	RGZ	48	250	180.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2

TEXAS INSTRUMENTS

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# PACKAGE MATERIALS INFORMATION

23-Oct-2015



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS65182BRGZR	VQFN	RGZ	48	2500	367.0	367.0	38.0
TPS65182BRGZT	VQFN	RGZ	48	250	210.0	185.0	35.0
TPS65182RGZR	VQFN	RGZ	48	2500	367.0	367.0	38.0
TPS65182RGZT	VQFN	RGZ	48	250	210.0	185.0	35.0

# **MECHANICAL DATA**



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-leads (QFN) package configuration.

D. The package thermal pad must be soldered to the board for thermal and mechanical performance.

E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.

F. Falls within JEDEC MO-220.



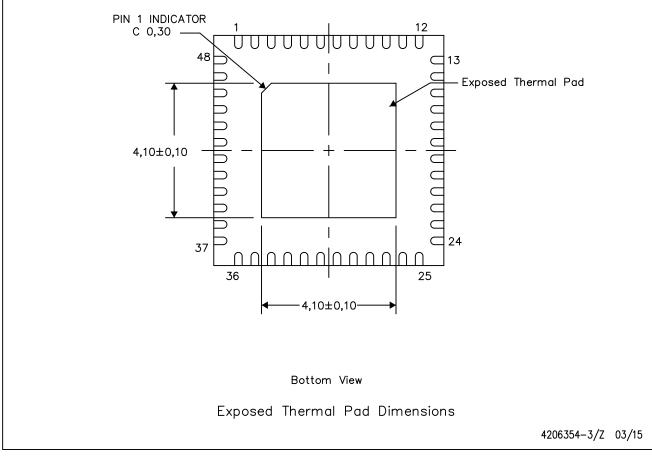
# RGZ (S-PVQFN-N48) PLASTIC QUAD FLATPACK NO-LEAD

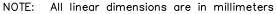
## THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

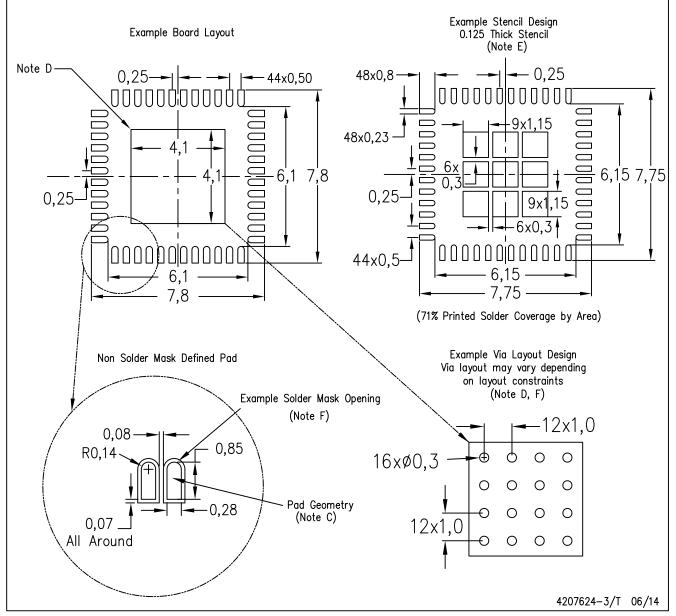






RGZ (S-PVQFN-N48)

# PLASTIC QUAD FLATPACK NO-LEAD



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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