



# MIC2950/2951

## 150mA Low-Dropout Voltage Regulator

### General Description

The MIC2950 and MIC2951 are “bulletproof” micropower voltage regulators with very low dropout voltage (typically 40mV at light loads and 250mV at 100mA), and very low quiescent current. Like their predecessors, the LP2950 and LP2951, the quiescent current of the MIC2950/MIC2951 increases only slightly in dropout, thus prolonging battery life. The MIC2950/MIC2951 are pin for pin compatible with the LP2950/LP2951, but offer lower dropout, lower quiescent current, reverse battery, and automotive load dump protection.

The key additional features and protection offered include higher output current (150mA), positive transient protection for up to 60V (load dump), and the ability to survive an unregulated input voltage transient of -20V below ground (reverse battery).

The plastic DIP and SOIC versions offer additional system functions such as programmable output voltage and logic controlled shutdown. The 3-pin TO-92 MIC2950 is pin-compatible with the older 5V regulators.

These system functions also include an error flag output that warns of a low output voltage, which is often due to failing batteries on the input. This may also be used as a power-on reset. A logic-compatible shutdown input is also available which enables the regulator to be switched on and off. This part may also be pin-strapped for a 5V output, or programmed from 1.24V to 29V with the use of two external resistors.

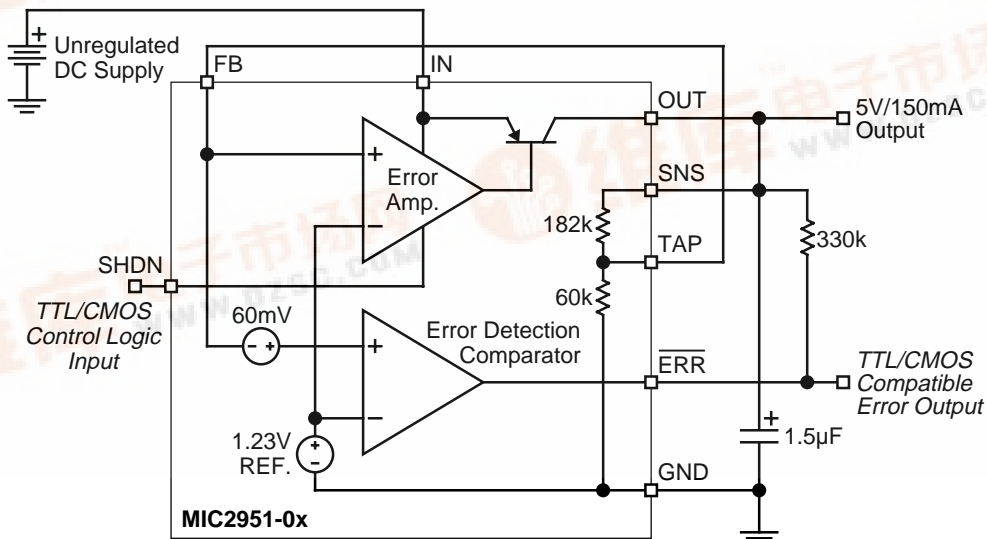
### Features

- High accuracy 3.3, 4.85, or 5V, guaranteed 150mA output
- Extremely low quiescent current
- Low-dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Use as regulator or reference
- Needs only 1.5µF for stability
- Current and thermal limiting
- Unregulated DC input can withstand -20V reverse battery and +60V positive transients
- Error flag warns of output dropout (MIC2951)
- Logic-controlled electronic shutdown (MIC2951)
- Output programmable from 1.24V to 29V (MIC2951)

### Applications

- Automotive Electronics
- Battery Powered Equipment
- Cellular Telephones
- SMPS Post-Regulator
- Voltage Reference
- Avionics
- High Efficiency Linear Power Supplies

### Block Diagram



The MIC2950 is available as either an -05 or -06 version. The -05 and -06 versions are guaranteed for junction temperatures from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ ; the -05 version has a tighter output and reference voltage specification range over temperature. The MIC2951 is available as an -02 or -03 version.

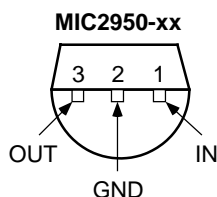
The MIC2950 and MIC2951 have a tight initial tolerance (0.5% typical), a very low output voltage temperature coefficient which allows use as a low-power voltage reference, and extremely good load and line regulation (0.04% typical). This greatly reduces the error in the overall circuit, and is the result of careful design techniques and process control.

## Ordering Information

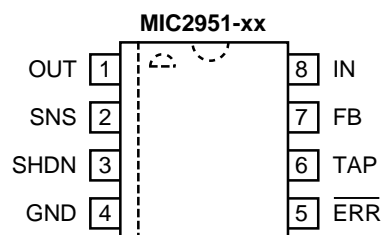
Part Number	Voltage	Accuracy	Temperature Range*	Package
MIC2950-05BZ	5.0V	0.5%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	TO-92
MIC2950-06BZ	5.0V	1.0%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	TO-92
MIC2951-02BM	5.0V	0.5%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-pin SOIC
MIC2951-03BM	5.0V	1.0%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-pin SOIC
MIC2951-02BN	5.0V	0.5%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-pin plastic DIP
MIC2951-03BN	5.0V	1.0%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-pin plastic DIP
MIC2951-03BMM	5.0V	1.0%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-lead MM8™
MIC2951-3.3BM	3.3V	1.0%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-pin SOIC
MIC2951-4.8BM	4.85V	1.0%	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	8-pin SOIC

\* junction temperature

## Pin Configuration



TO-92 (Z)  
(Bottom View)



DIP (N), SOIC (M), MM8™ (MM)  
(Top View)

## Pin Description

Pin # MIC2950	Pin # MIC2951	Pin Name	Pin Function
3	1	OUT	Regulated Output
	2	SNS	Sense (Input): Output-voltage sensing end of internal voltage divider for fixed 5V operation. Not used in adjustable configuration.
	3	SHDN	Shutdown/Enable (Input): TTL compatible input. High = shutdown, low or open = enable.
2	4	GND	Ground
	5	$\overline{\text{ERR}}$	Error Flag (Output): Active low, open-collector output (low = error, floating = normal).
	6	TAP	3.3V/4.85V/5V Tap: Output of internal voltage divider when the regulator is configured for fixed operation. Not used in adjustable configuration.
	7	FB	Feedback (Input): 1.235V feedback from internal voltage divider's TAP (for fixed operation) or external resistor network (adjustable configuration).
1	8	IN	Unregulated Supply Input

**Absolute Maximum Ratings (Note 1)**

Input Supply Voltage ( $V_{IN}$ ) **Note 5** ..... -20V to +60V  
 Feedback Input Voltage ( $V_{FB}$ ) **Note 6, 7** ..... -1.5V to +26V  
 Shutdown Input Voltage ( $V_{SHDN}$ ) **Note 6** ..... -0.3V to +30V  
 Power Dissipation ( $P_D$ ) **Note 4** ..... Internally Limited  
 Storage Temperature ..... -65°C to +150°C  
 Lead Temperature (soldering, 5 sec.) ..... 260°C  
 ESD, **Note 3**

**Operating Ratings (Note 2)**

Input Supply Voltage ( $V_{IN}$ ) ..... +2.0V to +30V  
 Junction Temperature ( $T_J$ ) ..... **Note 4**  
 MIC2950-05/MIC2950-06 ..... -40°C to +125°C  
 MIC2951-02/MIC2950-03 ..... -40°C to +125°C

**Electrical Characteristics**

$V_{IN} = 6V$ ;  $I_L = 100\mu A$ ;  $C_L = 1\mu F$ ;  $T_J = 25^\circ C$ , **bold** values indicate  $-40^\circ C \leq T_J \leq +125^\circ C$ ; **Note 8**; unless noted.

Parameter	Condition	Min	Typ	Max	Units
Output Voltage	MIC295x-02/-05 ( $\pm 0.5\%$ ), $T_J = 25^\circ C$	4.975	5.000	5.025	V
	MIC295x-03/-06 ( $\pm 1\%$ ), $T_J = 25^\circ C$	4.950	5.000	5.050	V
	MIC2951-3.3 ( $\pm 1\%$ ), $T_J = 25^\circ C$	3.267	3.300	3.333	V
	MIC2951-4.8 ( $\pm 1\%$ ), $T_J = 25^\circ C$	4.802	4.850	4.899	V
Output Voltage	MIC295x-02/-05 ( $\pm 0.5\%$ ), $-25^\circ C \leq T_J \leq +85^\circ C$	4.950		5.050	V
	MIC295x-03/-06 ( $\pm 1\%$ ), $-25^\circ C \leq T_J \leq +85^\circ C$	4.925		5.075	V
	MIC2951-3.3 ( $\pm 1\%$ ), $-25^\circ C \leq T_J \leq +85^\circ C$	3.251		3.350	V
	MIC2951-4.8 ( $\pm 1\%$ ), $-25^\circ C \leq T_J \leq +85^\circ C$	4.777		4.872	V
Output Voltage Over Full Temperature Range	MIC295x-02/-05 ( $\pm 0.5\%$ ), $-40^\circ C$ to $+125^\circ C$	<b>4.940</b>		<b>5.060</b>	V
	MIC295x-03/-06 ( $\pm 1\%$ ), $-40^\circ C$ to $+125^\circ C$	<b>4.900</b>		<b>5.100</b>	V
	MIC2951-3.3 ( $\pm 1\%$ ), $-40^\circ C$ to $+125^\circ C$	<b>3.234</b>		<b>3.366</b>	V
	MIC2951-4.8 ( $\pm 1\%$ ), $-40^\circ C$ to $+125^\circ C$	<b>4.753</b>		<b>4.947</b>	V
Output Voltage Over Load Variation	MIC295x-02/-05 ( $\pm 0.5\%$ ), $100\mu A \leq I_L \leq 150mA$ , $T_J \leq T_{J(max)}$	<b>4.930</b>		<b>5.070</b>	V
	MIC295x-03/-06 ( $\pm 1\%$ ), $100\mu A \leq I_L \leq 150mA$ , $T_J \leq T_{J(max)}$	<b>4.880</b>		<b>5.120</b>	V
	MIC2951-3.3 ( $\pm 1\%$ ), $100\mu A \leq I_L \leq 150mA$ , $T_J \leq T_{J(max)}$	<b>3.221</b>		<b>3.379</b>	V
	MIC2951-4.8 ( $\pm 1\%$ ), $100\mu A \leq I_L \leq 150mA$ , $T_J \leq T_{J(max)}$	<b>4.733</b>		<b>4.967</b>	V
Output Voltage Temperature Coefficient	MIC295x-02/-05 ( $\pm 0.5\%$ ), <b>Note 9</b>		<b>20</b>	<b>100</b>	ppm/ $^\circ C$
	MIC295x-03/-06 ( $\pm 1\%$ ), <b>Note 9</b>		<b>50</b>	<b>150</b>	ppm/ $^\circ C$
	MIC2951-3.3 ( $\pm 1\%$ ), <b>Note 9</b>		<b>50</b>	<b>150</b>	ppm/ $^\circ C$
	MIC2951-4.8 ( $\pm 1\%$ ), <b>Note 9</b>		<b>50</b>	<b>150</b>	ppm/ $^\circ C$
Line Regulation	MIC295x-02/-05 ( $\pm 0.5\%$ ), <b>Note 10, 11</b>		0.03	0.10 <b>0.20</b>	% %
	MIC295x-03/-06 ( $\pm 1\%$ ), <b>Note 10, 11</b>		0.04	0.20 <b>0.40</b>	% %
	MIC2951-3.3 ( $\pm 1\%$ ), <b>Note 10, 11</b>		0.04	0.20 <b>0.40</b>	% %
	MIC2951-4.8 ( $\pm 1\%$ ), <b>Note 10, 11</b>		0.04	0.20 <b>0.40</b>	% %

Parameter	Condition	Min	Typ	Max	Units
Load Regulation	MIC295x-02/-05 ( $\pm 0.5\%$ ), $100\mu\text{A} \leq I_L \leq 150\text{mA}$ , <b>Note 10</b>		0.04	0.10 <b>0.20</b>	% %
	MIC295x-03/-06 ( $\pm 1\%$ ), $100\mu\text{A} \leq I_L \leq 150\text{mA}$ , <b>Note 10</b>		0.10	0.20 <b>0.30</b>	% %
	MIC2951-3.3 ( $\pm 1\%$ ), $100\mu\text{A} \leq I_L \leq 150\text{mA}$ , <b>Note 10</b>		0.10	0.20 <b>0.30</b>	% %
	MIC2951-4.8 ( $\pm 1\%$ ), $100\mu\text{A} \leq I_L \leq 150\text{mA}$ , <b>Note 10</b>		0.10	0.20 <b>0.30</b>	% %
Dropout Voltage	MIC295x-02/-03/-05/-06, $I_L = 100\mu\text{A}$ , <b>Note 12</b>		40	80 <b>140</b>	mV mV
	MIC295x-02/-03/-05/-06, $I_L = 100\text{mA}$ , <b>Note 12</b>		250	300	mV
	MIC295x-02/-03/-05/-06, $I_L = 150\text{mA}$ , <b>Note 12</b>		300	450 <b>600</b>	mV mV
	MIC2951-3.3 ( $\pm 1\%$ ), $I_L = 100\mu\text{A}$ , <b>Note 12</b>		40	80 <b>150</b>	mV mV
	MIC2951-3.3 ( $\pm 1\%$ ), $I_L = 100\text{mA}$ , <b>Note 12</b>		250	350	mV
	MIC2951-3.3 ( $\pm 1\%$ ), $I_L = 150\text{mA}$ , <b>Note 12</b>		320	450 <b>600</b>	mV mV
	MIC2951-4.8 ( $\pm 1\%$ ), $I_L = 100\mu\text{A}$ , <b>Note 12</b>		40	80 <b>140</b>	mV mV
	MIC2951-4.8 ( $\pm 1\%$ ), $I_L = 100\text{mA}$ , <b>Note 12</b>		250	300	mV
	MIC2951-4.8 ( $\pm 1\%$ ), $I_L = 150\text{mA}$ , <b>Note 12</b>		300	450 <b>600</b>	mV mV
	Ground Current	MIC295x-02/-03/-05/-06, $I_L = 100\mu\text{A}$		120	180 <b>300</b>
MIC295x-02/-03/-05/-06, $I_L = 100\text{mA}$			1.7	2.5 <b>3.5</b>	mA mA
MIC295x-02/-03/-05/-06, $I_L = 150\text{mA}$			4	6 <b>8</b>	mA mA
MIC2951-3.3 ( $\pm 1\%$ ), $I_L = 100\mu\text{A}$			100	180 <b>300</b>	$\mu\text{A}$ $\mu\text{A}$
MIC2951-3.3 ( $\pm 1\%$ ), $I_L = 100\text{mA}$			1.7	2.5	mA
MIC2951-3.3 ( $\pm 1\%$ ), $I_L = 150\text{mA}$			4	6 <b>10</b>	mA mA
MIC2951-4.8 ( $\pm 1\%$ ), $I_L = 100\mu\text{A}$			120	180 <b>300</b>	$\mu\text{A}$ $\mu\text{A}$
MIC2951-4.8 ( $\pm 1\%$ ), $I_L = 100\text{mA}$			1.7	2.5 <b>3.5</b>	mA mA
MIC2951-4.8 ( $\pm 1\%$ ), $I_L = 150\text{mA}$			4	6 <b>8</b>	mA mA
Dropout Ground Current	MIC295x-02/-03/-05/-06 ( $\pm 0.5\%$ ), $V_{\text{IN}} = 4.5\text{V}$ , $I_L = 100\mu\text{A}$		280	350 <b>400</b>	$\mu\text{A}$ $\mu\text{A}$
	MIC2951-3.3 ( $\pm 1\%$ ), $V_{\text{IN}} = 3.0\text{V}$ , $I_L = 100\mu\text{A}$		150	350 <b>400</b>	$\mu\text{A}$ $\mu\text{A}$
	MIC2951-4.8 ( $\pm 1\%$ ), $V_{\text{IN}} = 4.3\text{V}$ , $I_L = 100\mu\text{A}$		280	350 <b>400</b>	$\mu\text{A}$ $\mu\text{A}$

Parameter	Condition	Min	Typ	Max	Units
Current Limit	$V_{OUT} = 0V$		300	400 <b>450</b>	mA mA
Thermal Regulation	<b>Note 13</b>		0.05	0.20	%/W
Output Noise	10Hz to 100kHz, $C_L = 1.5\mu F$		430		$\mu V_{RMS}$
	10Hz to 100kHz, $C_L = 200\mu F$		160		$\mu V_{RMS}$
	10Hz to 100kHz, $C_L = 3.3\mu F$ , 0.01 $\mu F$ bypass Feedback to Output		100		$\mu V_{RMS}$
Reference Voltage	MIC295x-02/-05 ( $\pm 0.5\%$ )	1.220 <b>1.200</b>	1.235	1.250 <b>1.260</b>	V V
	MIC295x-03/-06 ( $\pm 1\%$ )	1.210 <b>1.200</b>	1.235	1.260 <b>1.270</b>	V V
	MIC2951-3.3 ( $\pm 1\%$ )	1.210 <b>1.200</b>	1.235	1.260 <b>1.270</b>	V V
	MIC2951-4.8 ( $\pm 1\%$ )	1.210 <b>1.200</b>	1.235	1.260 <b>1.270</b>	V V
Reference Voltage	MIC295x-02/-05 ( $\pm 0.5\%$ ), <b>Note 14</b>	<b>1.190</b>		<b>1.270</b>	V
	MIC295x-03/-06 ( $\pm 1\%$ ), <b>Note 14</b>	<b>1.185</b>		<b>1.285</b>	V
	MIC2951-3.3 ( $\pm 1\%$ ), <b>Note 14</b>	<b>1.185</b>		<b>1.285</b>	V
	MIC2951-4.8 ( $\pm 1\%$ ), <b>Note 14</b>	<b>1.185</b>		<b>1.285</b>	V
Feedback Bias Current			20	40 <b>60</b>	nA nA
Reference Voltage Temperature Coefficient	MIC295x-02/-05 ( $\pm 0.5\%$ ), <b>Note 9</b>		20		ppm/ $^{\circ}C$
	MIC295x-03/-06 ( $\pm 1\%$ ), <b>Note 9</b>		50		ppm/ $^{\circ}C$
	MIC2951-3.3 ( $\pm 1\%$ ), <b>Note 9</b>		50		ppm/ $^{\circ}C$
	MIC2951-4.8 ( $\pm 1\%$ ), <b>Note 9</b>		50		ppm/ $^{\circ}C$
Feedback Bias Current Temperature Coefficient			0.1		nA/ $^{\circ}C$
Error Comparator (Flag) Output Leakage Current	$V_{OH} = 30V$		0.01	1.00 <b>2.00</b>	$\mu A$ $\mu A$
Error Comparator (Flag) Output Low Voltage	$V_{IN} = 4.5V$ , $I_{OL} = 200\mu A$		150	250 <b>400</b>	mV mV
Error Comparator Upper Threshold Voltage	<b>Note 15</b>	40 <b>25</b>	60		mV mV
Error Comparator Lower Threshold Voltage	<b>Note 15</b>		75	95 <b>140</b>	mV mV
Error Comparator Hysteresis	<b>Note 15</b>		15		mV

Parameter	Condition	Min	Typ	Max	Units
Shutdown Input Logic Voltage	MIC295x-02/-05 ( $\pm 0.5\%$ )		1.3		V
		Low		<b>0.7</b>	V
		High	<b>2.0</b>		V
	MIC295x-03/-06 ( $\pm 1\%$ )		1.3		V
		Low		<b>0.7</b>	V
		High	<b>2.0</b>		V
	MIC2951-3.3 ( $\pm 1\%$ )		1.3		V
		Low		<b>0.7</b>	V
High		<b>2.0</b>		V	
MIC2951-4.8 ( $\pm 1\%$ )		1.3		V	
	Low		<b>0.7</b>	V	
	High	<b>2.0</b>		V	
Shutdown Input Current	$V_{\text{SHUTDOWN}} = 2.4\text{V}$		30	50 <b>100</b>	$\mu\text{A}$ $\mu\text{A}$
	$V_{\text{SHUTDOWN}} = 30\text{V}$		450	600 <b>750</b>	$\mu\text{A}$ $\mu\text{A}$
Regulator Output Current in Shutdown	<b>Note 7</b>		3	10 <b>20</b>	$\mu\text{A}$ $\mu\text{A}$

**Note 1.** Exceeding the absolute maximum rating may damage the device.

**Note 2.** The device is not guaranteed to function outside its operating rating.

**Note 3.** Devices are ESD sensitive. Handling precautions are recommended.

**Note 4.** The junction-to-ambient thermal resistance of the TO-92 package is  $180^{\circ}\text{C}/\text{W}$  with 0.4" leads and  $160^{\circ}\text{C}/\text{W}$  with 0.25" leads to a PC board. The thermal resistance of the 8-pin DIP package is  $105^{\circ}\text{C}/\text{W}$  junction-to-ambient when soldered directly to a PC board. Junction-to-ambient thermal resistance for the SOIC (M) package is  $160^{\circ}\text{C}/\text{W}$ . Junction-to-ambient thermal resistance for the MM8™ (MM) is  $250^{\circ}\text{C}/\text{W}$ .

**Note 5.** The maximum positive supply voltage of 60V must be of limited duration ( $\leq 100\text{ms}$ ) and duty cycle ( $\leq 1\%$ ). The maximum continuous supply voltage is 30V.

**Note 6.** When used in dual-supply systems where the output terminal sees loads returned to a negative supply, the output voltage should be diode-clamped to ground.

**Note 7.**  $V_{\text{SHDN}} \geq 2\text{V}$ ,  $V_{\text{IN}} \leq 30\text{V}$ ,  $V_{\text{OUT}} = 0$ , with the FB pin connected to TAP.

**Note 8.** Additional conditions for 8-pin devices are  $V_{\text{FB}} = 5\text{V}$ , TAP and OUT connected to SNS ( $V_{\text{OUT}} = 5\text{V}$ ) and  $V_{\text{SHDN}} \leq 0.8\text{V}$ .

**Note 9.** Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

**Note 10.** Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered in the specification for thermal regulation.

**Note 11.** Line regulation for the MIC2951 is tested at  $150^{\circ}\text{C}$  for  $I_{\text{L}} = 1\text{mA}$ . For  $I_{\text{L}} = 100\mu\text{A}$  and  $T_{\text{J}} = 125^{\circ}\text{C}$ , line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.

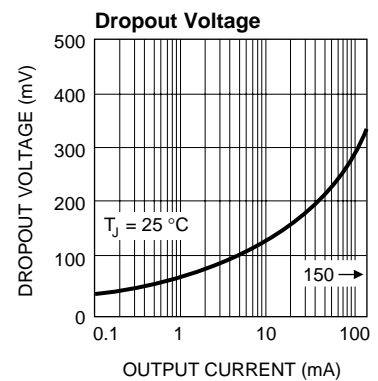
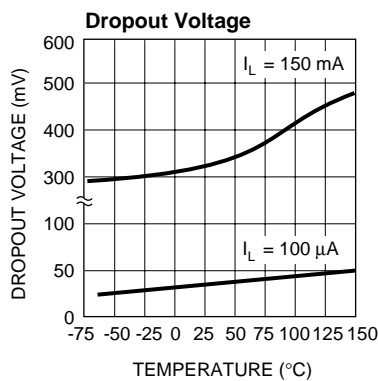
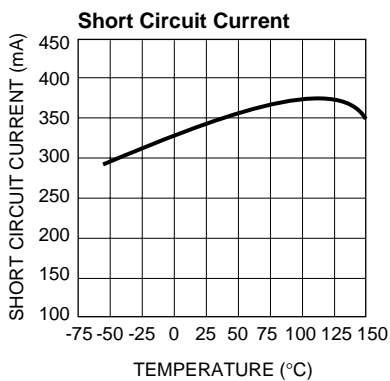
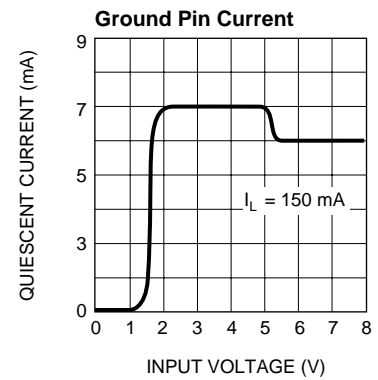
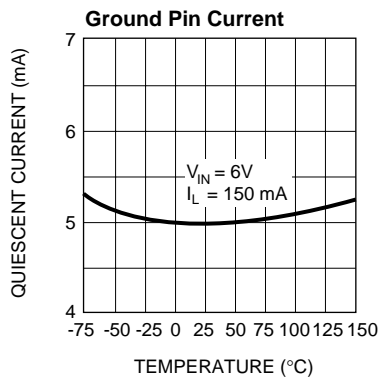
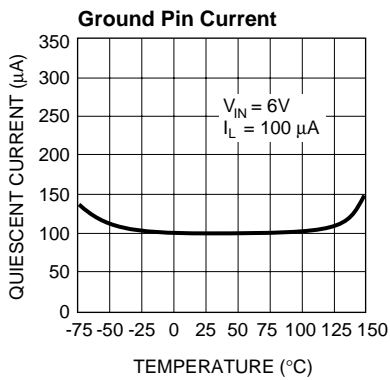
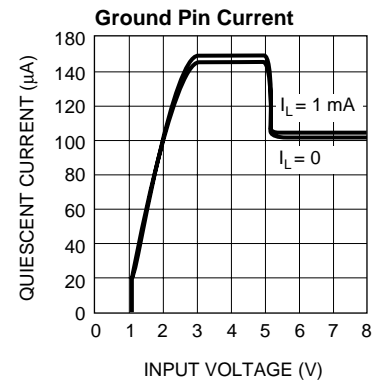
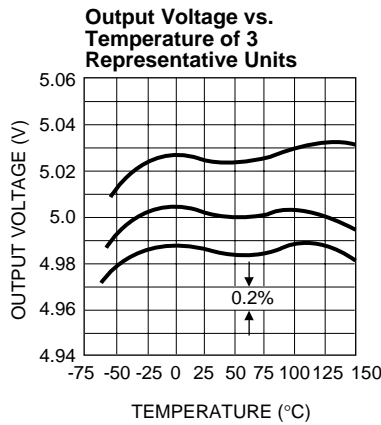
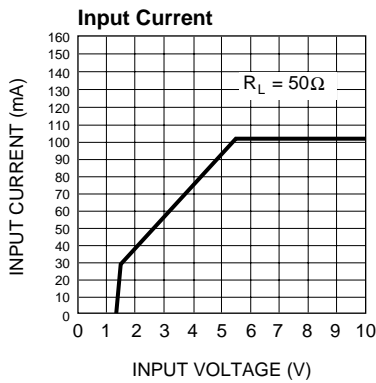
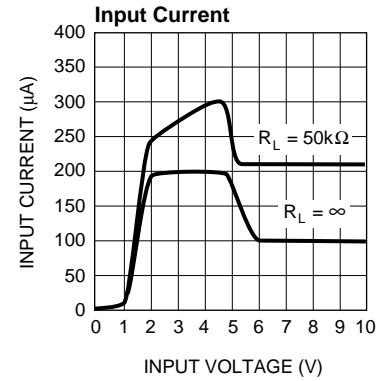
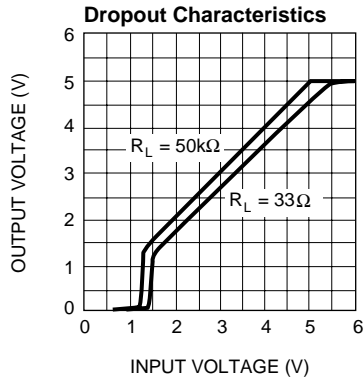
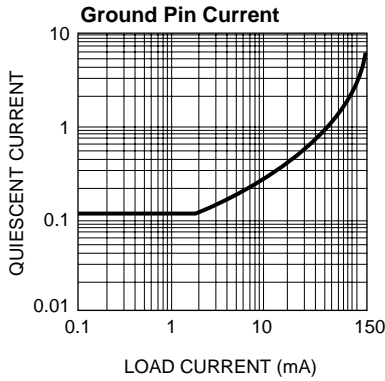
**Note 12.** Dropout voltage is defined as the input to output differential at which the output voltage drops 100mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.

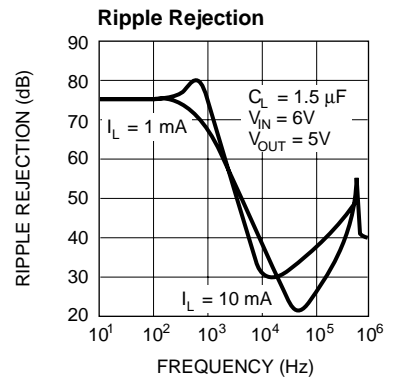
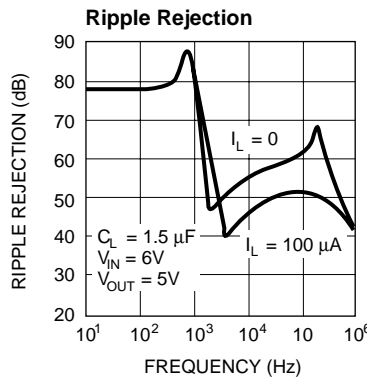
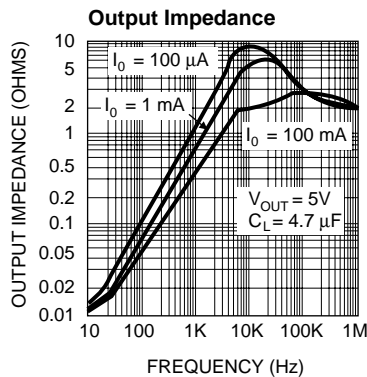
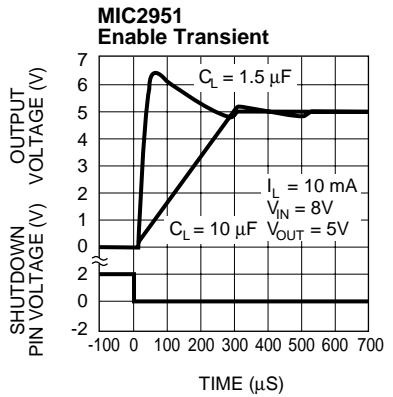
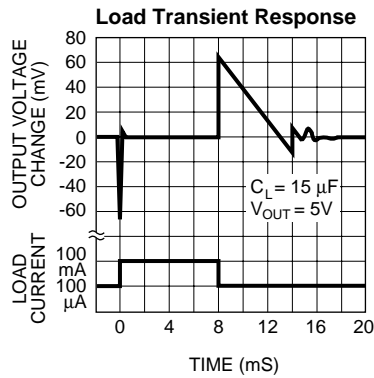
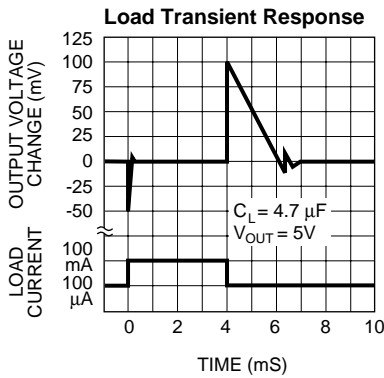
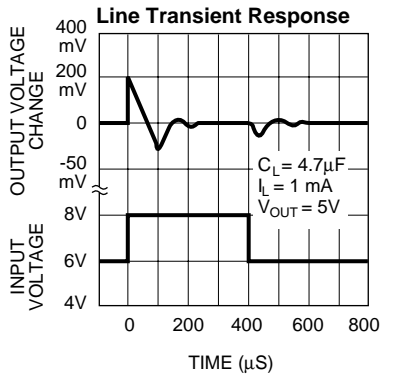
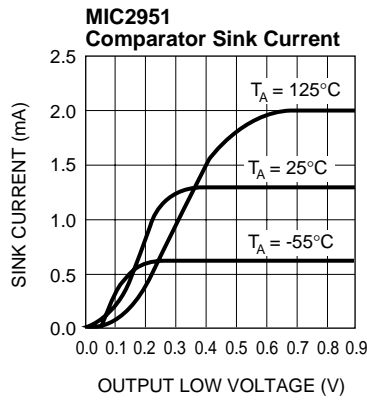
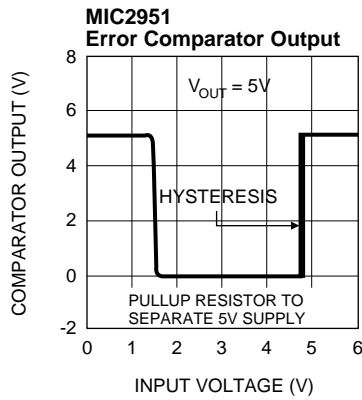
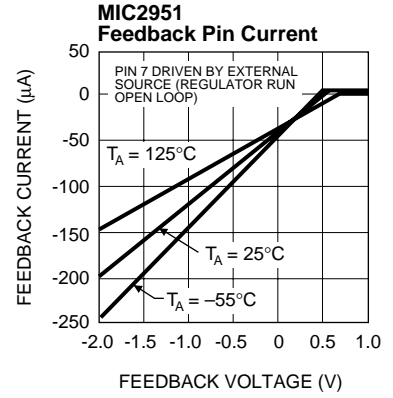
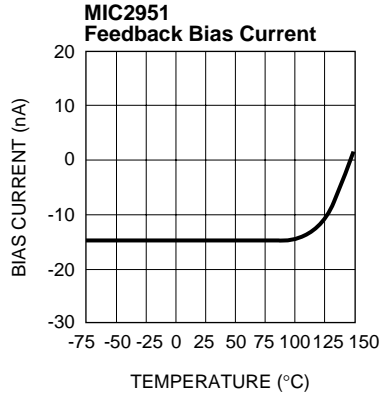
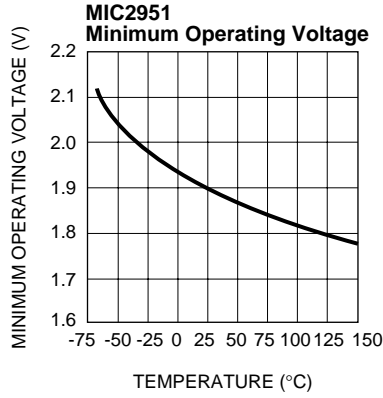
**Note 13.** Thermal regulation is defined as the change in output voltage at a time "t" after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50mA load pulse at  $V_{\text{IN}} = 30\text{V}$  (1.25W pulse) for  $t = 10\text{ms}$ .

**Note 14.**  $V_{\text{REF}} \leq V_{\text{OUT}} \leq (V_{\text{IN}} - 1\text{V})$ ,  $2.3\text{V} \leq V_{\text{IN}} \leq 30\text{V}$ ,  $100\mu\text{A} < I_{\text{L}} \leq 150\text{mA}$ ,  $T_{\text{J}} \leq T_{\text{JMAX}}$ .

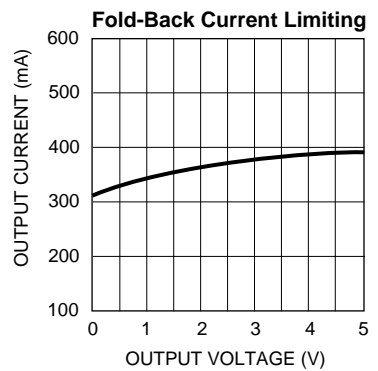
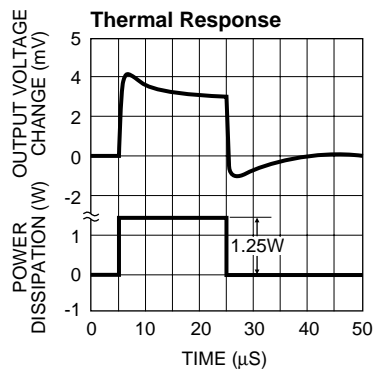
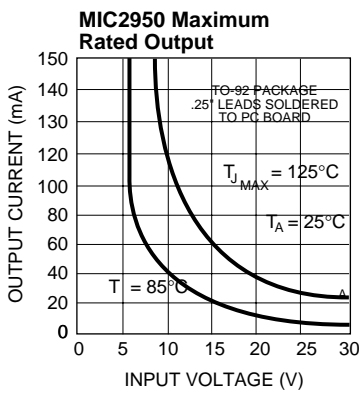
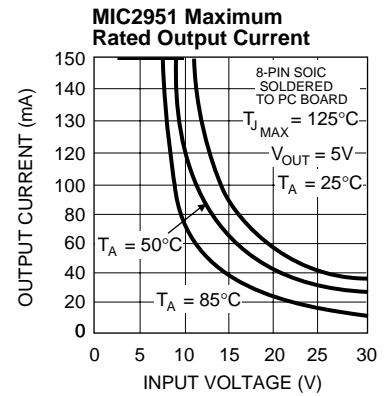
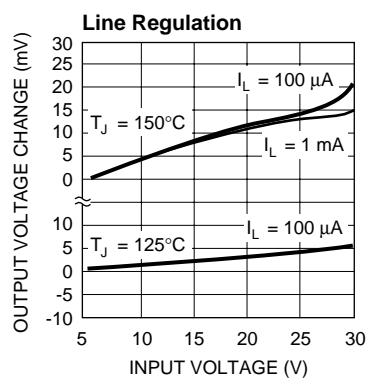
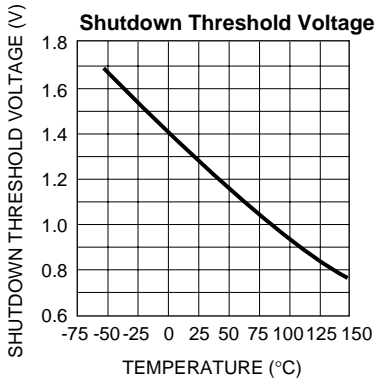
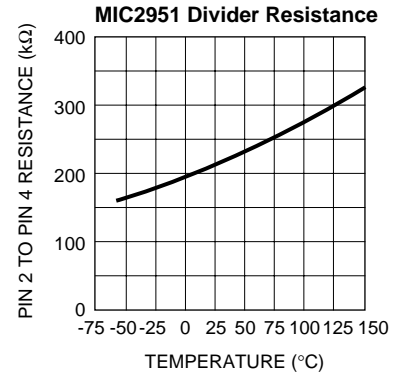
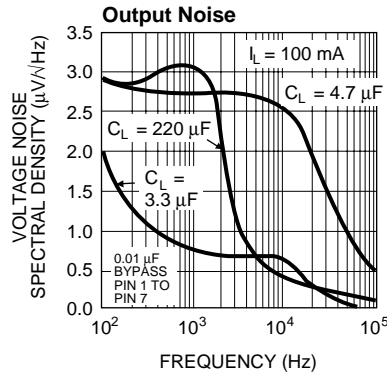
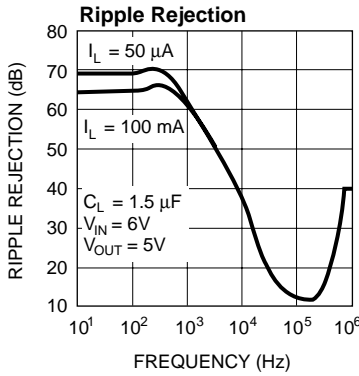
**Note 15.** Comparator thresholds are expressed in terms of a voltage differential at the FB terminal below the nominal reference voltage measured at 6V input. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain =  $V_{\text{OUT}}/V_{\text{REF}} = (R1 + R2)/R2$ . For example, at a programmed output voltage of 5V, the error output is guaranteed to go low when the output drops by  $95\text{mV} \times 5\text{V}/1.235\text{V} = 384\text{mV}$ . Thresholds remain constant as a percent of  $V_{\text{OUT}}$  as  $V_{\text{OUT}}$  is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.

# Typical Characteristics









## Applications Information

### Automotive Applications

The MIC2950/2951 are ideally suited for automotive applications for a variety of reasons. They will operate over a wide range of input voltages, have very low dropout voltages (40mV at light loads), and very low quiescent currents. These features are necessary for use in battery powered systems, such as automobiles. They are also “bulletproof” devices; with the ability to survive both reverse battery (negative transients up to 20V below ground), and load dump (positive transients up to 60V) conditions. A wide operating temperature range with low temperature coefficients is yet another reason to use these versatile regulators in automotive designs.

### External Capacitors

A 1.5  $\mu\text{F}$  (or greater) capacitor is required between the MIC2950/MIC2951 output and ground to prevent oscillations due to instability. Most types of tantalum or aluminum electrolytics will be adequate; film types will work, but are costly and therefore not recommended. Many aluminum electrolytics have electrolytes that freeze at about  $-30^{\circ}\text{C}$ , so solid tantalums are recommended for operation below  $-25^{\circ}\text{C}$ . The important parameters of the capacitor are an effective series resistance of about  $5\Omega$  or less and a resonant frequency above 500kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.5 $\mu\text{F}$  for current below 10mA or 0.15 $\mu\text{F}$  for currents below 1 mA. Using the 8-pin versions at voltages below 5V runs the error amplifier at lower gains so that more output capacitance is needed. For the worst-case situation of a 150mA load at 1.23V output (Output shorted to Feedback) a 5 $\mu\text{F}$  (or greater) capacitor should be used.

The MIC2950 will remain stable and in regulation with no load in addition to the internal voltage divider, unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the MIC2951 version with external resistors, a minimum load of 1 $\mu\text{A}$  is recommended.

A 0.1 $\mu\text{F}$  capacitor should be placed from the MIC2950/MIC2951 input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

Stray capacitance to the MIC2951 Feedback terminal (pin 7) can cause instability. This may especially be a problem when using high value external resistors to set the output voltage. Adding a 100pF capacitor between Output and Feedback and increasing the output capacitor to at least 3.3 $\mu\text{F}$  will remedy this.

### Error Detection Comparator Output

A logic low output will be produced by the comparator whenever the MIC2951 output falls out of regulation by more than approximately 5%. This figure is the comparator’s built-in

offset of about 60mV divided by the 1.235V reference voltage. (Refer to the block diagram on Page 1). This trip level remains “5% below normal” regardless of the programmed output voltage of the MIC2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, thermal limiting, or overvoltage on input (over  $\cong 40\text{V}$ ).

Figure 1 is a timing diagram depicting the  $\overline{\text{ERROR}}$  signal and the regulated output voltage as the MIC2951 input is ramped up and down. The  $\overline{\text{ERROR}}$  signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which  $V_{\text{OUT}} = 4.75$ —for 5.0V applications). Since the MIC2951’s dropout voltage is load-dependent (see curve in Typical Performance Characteristics), the input voltage trip point (about 5V) will vary with the load current. The output voltage trip point does not vary with load.

The error comparator has an open-collector output which requires an external pull-up resistor. Depending on system requirements, this resistor may be returned to the output or some other supply voltage. In determining a value for this resistor, note that while the output is rated to sink 200 $\mu\text{A}$ , this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1M $\Omega$ . The resistor is not required if this output is unused.

### Programming the Output Voltage (MIC2951)

The MIC2951 may be pin-strapped for 5V (or 3.3V or 4.85V) using its internal voltage divider by tying Pin 1 (output) to Pin 2 (sense) and Pin 7 (feedback) to Pin 6 (5V Tap). Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. An external pair of resistors is required, as shown in Figure 2.

The complete equation for the output voltage is

$$V_{\text{OUT}} = V_{\text{REF}} \times \{ 1 + R_1/R_2 \} + I_{\text{FB}} R_1$$

where  $V_{\text{REF}}$  is the nominal 1.235 reference voltage and  $I_{\text{FB}}$  is the feedback pin bias current, nominally  $-20\text{nA}$ . The minimum recommended load current of 1  $\mu\text{A}$  forces an upper limit of 1.2M $\Omega$  on the value of  $R_2$ , if the regulator must work with no load (a condition often found in CMOS in standby),  $I_{\text{FB}}$  will produce a 2% typical error in  $V_{\text{OUT}}$  which may be eliminated at room temperature by trimming  $R_1$ . For better accuracy, choosing  $R_2 = 100\text{k}$  reduces this error to 0.17% while increasing the resistor program current to 12  $\mu\text{A}$ .

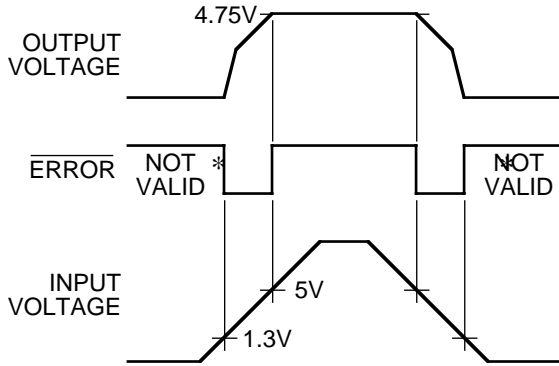
### Reducing Output Noise

In some applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is the only method by which noise can be reduced on the 3 lead MIC2950 and is relatively inefficient, as increasing the capacitor from 1 $\mu\text{F}$  to 220 $\mu\text{F}$  only decreases the noise from 430 $\mu\text{V}$  to 160 $\mu\text{V}$  rms for a 100kHz bandwidth at 5V output.

Noise can be reduced fourfold by a bypass capacitor across  $R_1$ , since it reduces the high frequency gain from 4 to unity. Pick:

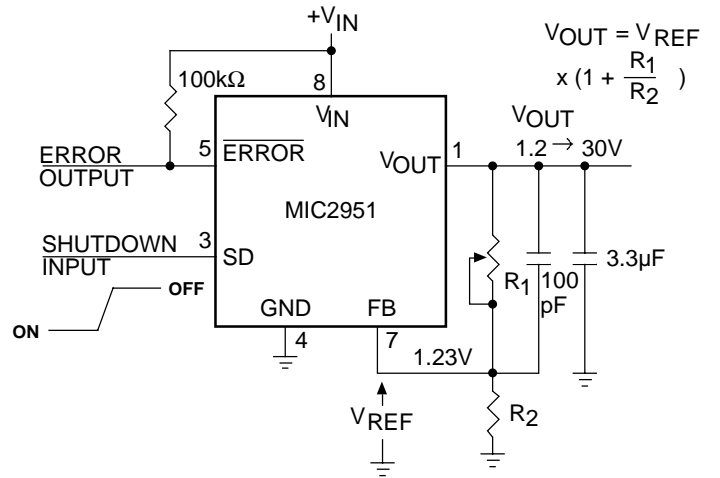
$$C_{\text{BYPASS}} \cong \frac{1}{2\pi R_1 \cdot 200 \text{ Hz}}$$

or about  $0.01 \mu\text{F}$ . When doing this, the output capacitor must be increased to  $3.3 \mu\text{F}$  to maintain stability. These changes reduce the output noise from  $430 \mu\text{V}$  to  $100 \mu\text{V rms}$  for a  $100 \text{ kHz}$  bandwidth at  $5\text{V}$  output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.



\* SEE APPLICATIONS INFORMATION

Figure 1.  $\overline{\text{ERROR}}$  Output Timing

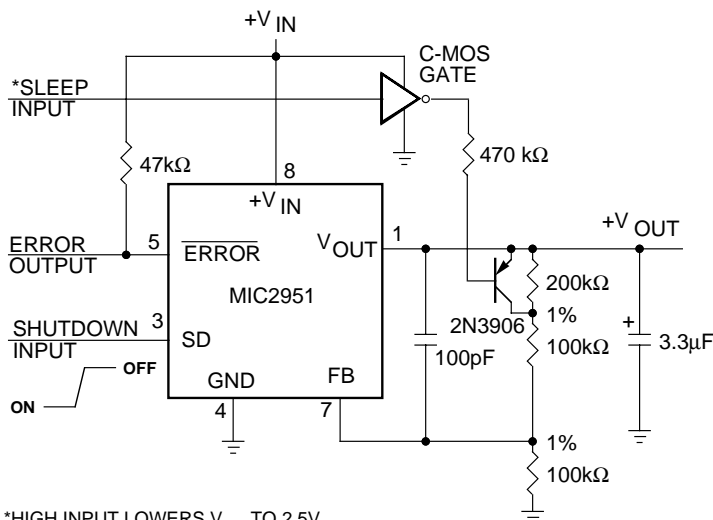


NOTE: PINS 2 AND 6 ARE LEFT OPEN

\*SEE APPLICATIONS INFORMATION

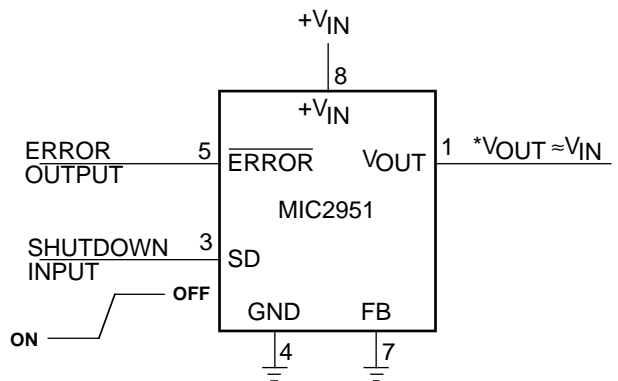
Figure 2. Adjustable Regulator

## Typical Applications



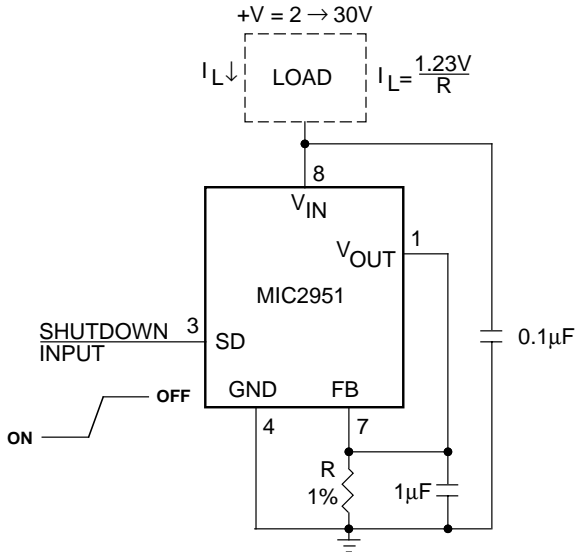
\*HIGH INPUT LOWERS  $V_{\text{OUT}}$  TO 2.5V

5V Regulator with 2.5V Sleep Function

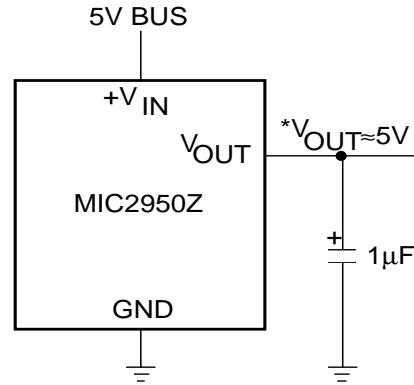


\*MINIMUM INPUT-OUTPUT VOLTAGE RANGES FROM 40mV TO 400mV, DEPENDING ON LOAD CURRENT.

Wide Input Voltage Range Current Limiter

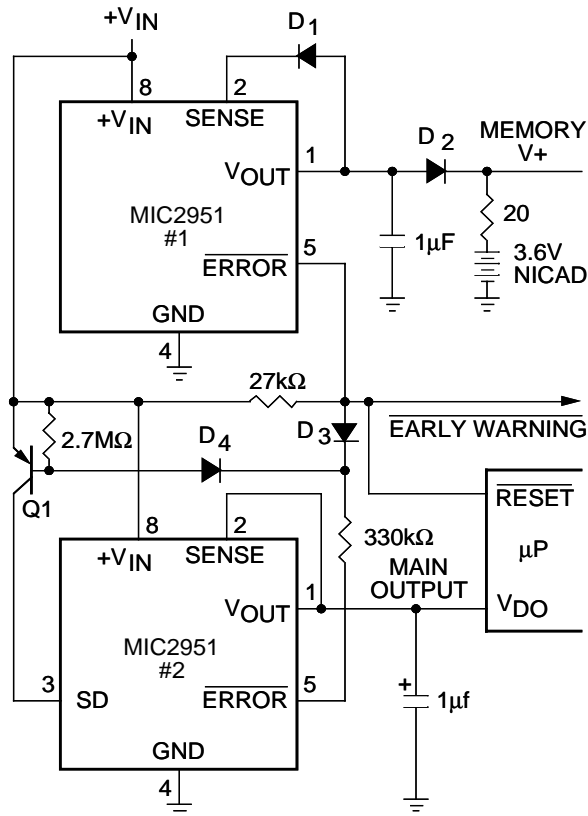


**Low Drift Current Source**



**5-Volt Current Limiter**

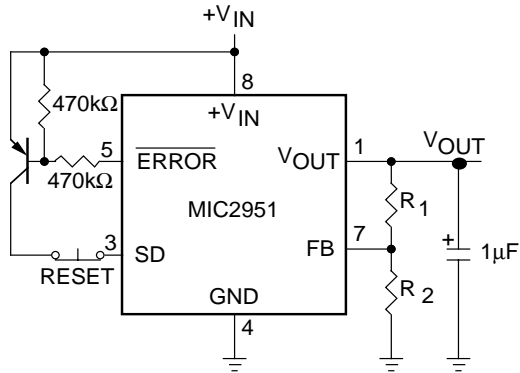
\* MINIMUM INPUT-OUTPUT VOLTAGE RANGES FROM 40mV TO 400mV, DEPENDING ON LOAD CURRENT.



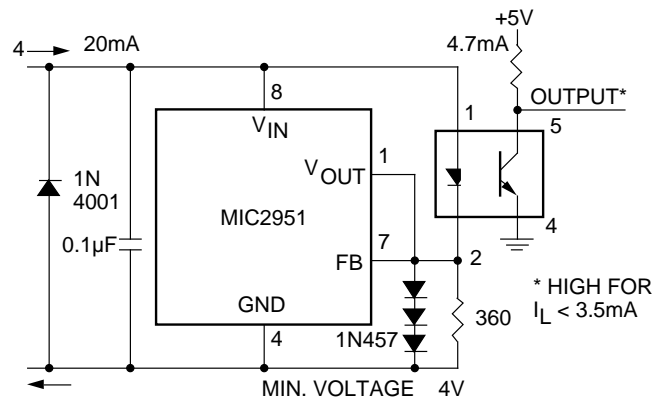
**Regulator with Early Warning and Auxiliary Output**

- EARLY WARNING FLAG ON LOW INPUT VOLTAGE
- MAIN OUTPUT LATCHES OFF AT LOWER INPUT VOLTAGES
- BATTERY BACKUP ON AUXILIARY OUTPUT

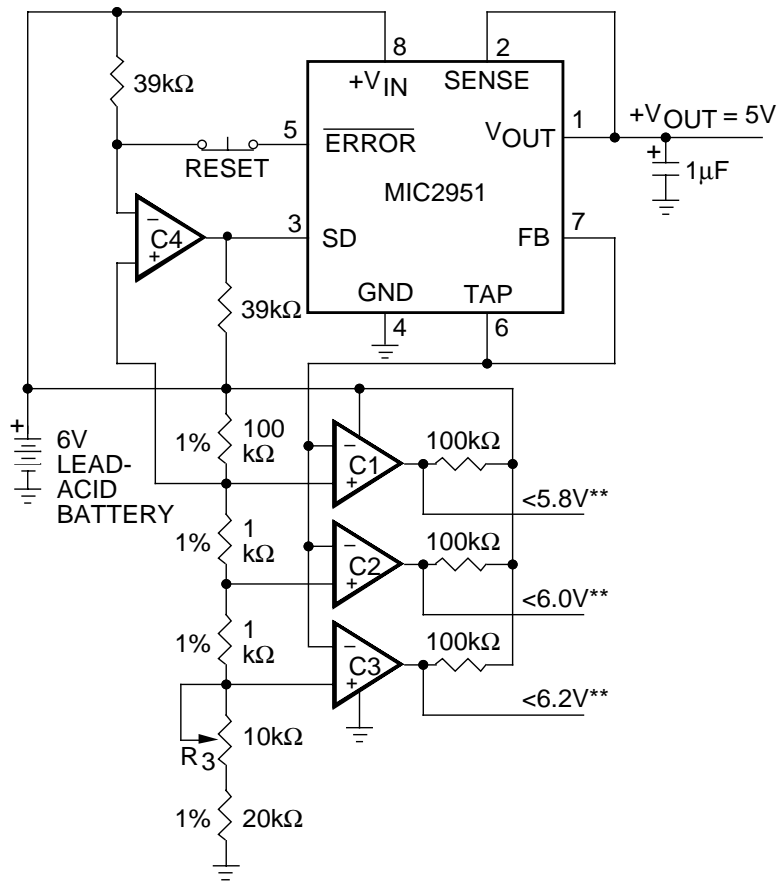
OPERATION: REG. #1'S  $V_{OUT}$  IS PROGRAMMED ONE DIODE DROP ABOVE 5 V. ITS ERROR FLAG BECOMES ACTIVE WHEN  $V_{IN} \leq 5.7$  V. WHEN  $V_{IN}$  DROPS BELOW 5.3 V, THE ERROR FLAG OF REG. #2 BECOMES ACTIVE AND VIA Q1 LATCHES THE MAIN OUTPUT OFF. WHEN  $V_{IN}$  AGAIN EXCEEDS 5.7 V REG. #1 IS BACK IN REGULATION AND THE EARLY WARNING SIGNAL RISES, UNLATCHING REG. #2 VIA D3.



Latch Off When Error Flag Occurs

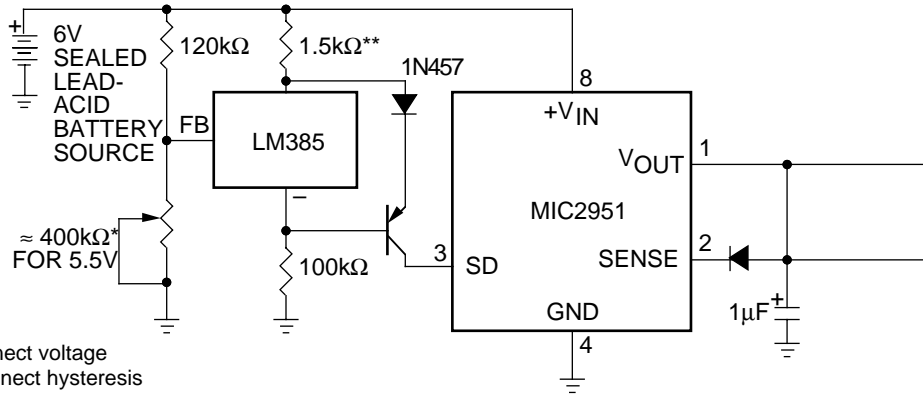


Open Circuit Detector for 4mA to 20mA Current Loop



C1 TO C4 ARE COMPARATORS (LP339 OR EQUIVALENT)  
 \*OPTIONAL LATCH OFF WHEN DROP OUT OCCURS. ADJUST R3 FOR C2 SWITCHING WHEN  $V_{IN}$  IS 6.0V  
 \*\*OUTPUTS GO LOW WHEN  $V_{IN}$  DROPS BELOW DESIGNATED THRESHOLDS.

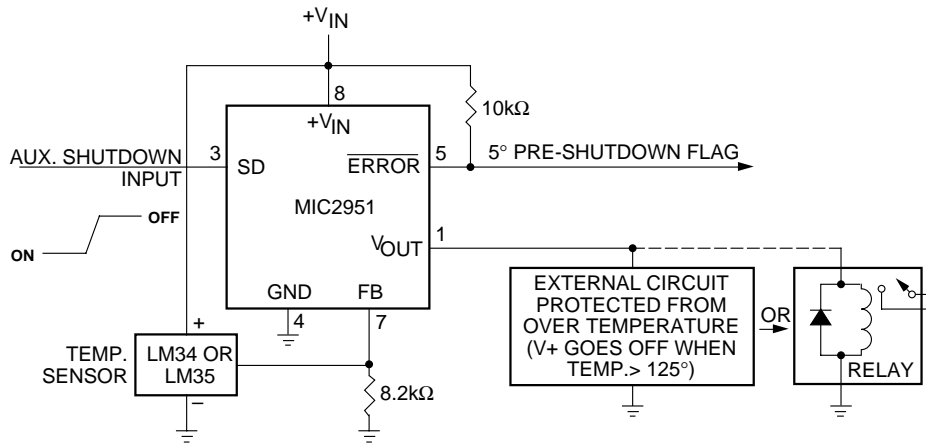
Regulator with State-of-Charge Indicator



\* Sets disconnect voltage  
 \*\* Sets disconnect hysteresis

**Low-Battery Disconnect**

For values shown, Regulator shuts down when  $V_{IN} < 5.5V$  and turns on again at  $6.0V$ . Current drain in disconnected mode is  $150\mu A$ .



LM34 for 125°F Shutdown  
 LM35 for 125°C Shutdown

**System Overtemperature Protection Circuit**

**Schematic Diagram**

