捷多邦,专业PCB打样工厂**T企V2对**中**第世V2**72, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT

SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

- Rail-To-Rail Output
- Wide Bandwidth . . . 3 MHz
- High Slew Rate . . . 2 .4 V/μs
- Supply Voltage Range . . . 2.7 V to 16 V
- Supply Current . . . 550 μA/Channel
- Input Noise Voltage . . . 39 nV/√Hz
- Input Bias Current . . . 1 pA
- Specified Temperature Range
 0°C to 70°C...Commercial Grade
 -40°C to 125°C...Industrial Grade
- Ultrasmall Packaging
 - 5 Pin SOT-23 (TLV271)
 - 8 Pin MSOP (TLV272)
- Ideal Upgrade for TLC27x Family

description

The TLV27x takes the minimum operating supply voltage down to 2.7 V over the extended industrial temperature range while adding the rail-to-rail output swing feature. This makes it an ideal alternative to the TLC27x family for applications where rail-to-rail output swings are essential. The TLV27x also provides 3-MHz bandwidth from only $550~\mu A$.

Like the TLC27x, the TLV27x is fully specified for 5-V and \pm 5-V supplies. The maximum recommended supply voltage is 16 V, which allows the devices to be operated from a variety of rechargeable cells (\pm 8 V supplies down to \pm 1.35 V).

The CMOS inputs enable use in high-impedance sensor interfaces, with the lower voltage operation making an attractive alternative for the TLC27x in battery-powered applications.

All members are available in PDIP and SOIC with the singles in the small SOT-23 package, duals in the MSOP, and quads in the TSSOP package.

The 2.7-V operation makes it compatible with Li-lon powered systems and the operating supply voltage range of many micropower microcontrollers available today including TI's MSP430.

SELECTION OF SIGNAL AMPLIFIER PRODUCTS†

DEVICE	V _{DD} (V)	V _{IO} (μV)	lq/Ch (μA)	I _{IB} (pA)	GBW (MHz)	SR (V/μs)	KAIL		SINGLES/DUALS/QUADS	
TLV27x	2.7–16	500	550	1	3	2.4	_	0	S/D/Q	
TLC27x	3–16	1100	675	1	1.7	3.6	_	25-7	S/D/Q	
TLV237x	2.7–16	500	550	1	3	2.4	Yes	I/O	S/D/Q	
TLC227x	4–16	300	1100	1	2.2	3.6		0	D/Q	
TLV246x	2.7–6	150	550	1300	6.4	1.6	Yes	I/O	S/D/Q	
TLV247x	2.7–6	250	600	2	2.8	1.5	Yes	I/O	S/D/Q	
TLV244x	2.7–10	300	725	1	1.8	1.4	_	0	D/Q	

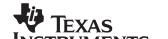
† Typical values measured at 5 V, 25°C



Operational Amplifier

Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of



TLV271, TLV272, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS

SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

FAMILY PACKAGE TABLE

DEVICE	NUMBER OF		PAC	KAGE TY	PES		OLULTDOWN.	UNIVERSAL	
DEVICE	CHANNELS	PDIP	SOIC	SOT-23	TSSOP	MSOP	SHUTDOWN	EVM BOARD	
TLV271	1	8	8	5	_			Refer to the EVM	
TLV272	2	8	8	_	_	8	_	Selection Guide	
TLV274	4	14	14	_	14	_	_	(Lit# SLOU060)	

TLV271 AVAILABLE OPTIONS

ĺ				PACKAGED DEVICES						
	T_A	V _{IO} MAX AT 25°C	SMALL OUTLINE	SOT-23	PLASTIC DIP					
		23 0	(D) [†]	(DBV)‡	SYMBOL	(P)				
	0°C to 70°C	F\/	TLV271CD	TLV271CDBV	VBHC	_				
ĺ	-40°C to 125°C	5 mV	TLV271ID	TLV271IDBV	VBHI	TLV271IP				

[†] This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLV271IDR).

TLV272 AVAILABLE OPTIONS

		PACKAGED DEVICES						
TA	V _{IO} MAX AT 25°C	SMALL OUTLINE	MSOP	PLASTIC DIP				
	20 0	(D)§	(DGK)§	SYMBOL	(P)			
0°C to 70°C	E m\/	TLV272CD	TLV272CDGK	AVF	_			
-40°C to 125°C	5 mV	TLV272ID	TLV272IDGK	AVG	TLV272IP			

[§] This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLV272IDR).

TLV274 AVAILABLE OPTIONS

		PACKAGED DEVICES				
TA	V _{IO} MAX AT 25°C	SMALL OUTLINE (D)¶	PLASTIC DIP (N)	TSSOP (PW)¶		
0°C to 70°C	5 mV	TLV274CD	_	TLV274CPW		
-40°C to 125°C	51117	TLV274ID	TLV274IN	TLV274IPW		

[¶] This package is available taped and reeled. To order this packaging option, add an R suffix to the part number (e.g., TLV274IDR).

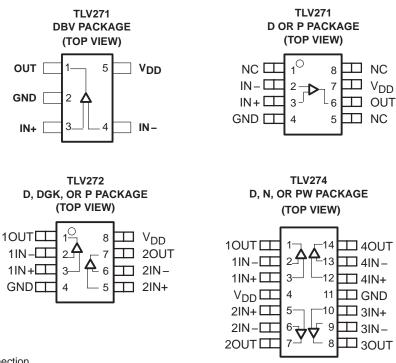


[‡] This package is only available taped and reeled. For standard quantities (3,000 pieces per reel), add an R suffix (e.g., TLV270IDBVR). For smaller quantities (250 pieces per mini-reel), add a T suffix to the part number (e.g., TLV270IDBVT).

TLV271, TLV272, TLV274 FAMILY OF 550-µA/Ch 3-MHz RAIL-TO-RAIL OUTPUT

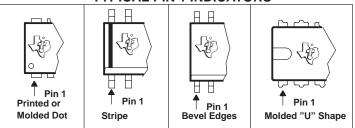
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TLV27x PACKAGE PINOUTS(1)



NC - No internal connection (1) SOT-23 may or may not be indicated

TYPICAL PIN 1 INDICATORS





TLV271, TLV272, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS

SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V _{DD} (see Note 1)	
Differential input voltage, V _{ID}	±V _{DD}
Input voltage range, V _I (see Note 1)	\dots -0.2 V to V _{DD} + 0.2 V
Input current range, I _I	±10 mA
Output current range, IO	±100 mA
Continuous total power dissipation	
Operating free-air temperature range, T _A : C suffix	0°C to 70°C
I suffix	–40°C to 125°C
Maximum junction temperature, T _J	150°C
Storage temperature range, T _{stq}	
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values, except differential voltages, are with respect to GND.

DISSIPATION RATING TABLE

PACKAGE	(₀C\M) _θ ገC	θJA (°C/W)	$T_{\mbox{\scriptsize A}} \le 25^{\circ}\mbox{\scriptsize C}$ POWER RATING	T _A = 25°C POWER RATING
D (8)	38.3	176	710 mW	396 mW
D (14)	26.9	122.3	1022 mW	531 mW
D (16)	25.7	114.7	1090 mW	567 mW
DBV (5)	55	324.1	385 mW	201 mW
DBV (6)	55	294.3	425 mW	221 mW
DGK (8)	54.23	259.96	481 mW	250 mW
DGS (10)	54.1	257.71	485 mW	252 mW
N (14, 16)	32	78	1600 mW	833 mW
P (8)	41	104	1200 mW	625 mW
PW (14)	29.3	173.6	720 mW	374 mW
PW (16)	28.7	161.4	774 mW	403 mW

recommended operating conditions

			MIN	MAX	UNIT
Complements as No.	Single supply		2.7	16	\/
Supply voltage, V _{DD}	Split supply		±1.35	±8	V
Common-mode input voltage range, V _{ICR}			0	V _{DD} -1.35	V
Operation from air temperature T.	C-suffix		0	70	°C
Operating free-air temperature, T _A	I-suffix		-40	125	, o



TLV271, TLV272, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

electrical characteristics at specified free-air temperature, V_{DD} = 2.7 V, 5 V, and ± 5 V (unless otherwise noted)

dc performance

	PARAMETER	TEST CONDIT	IONS	T _A †	MIN	TYP	MAX	UNIT
,,				25°C		0.5	5	.,
VIO	Input offset voltage	$V_{IC} = V_{DD}/2,$ $R_{I} = 10 \text{ k}\Omega,$	$V_O = V_{DD}/2$, RS = 50 Ω	Full range			7	mV
ανιο	Offset voltage drift			25°C		2		μV/°C
		$V_{IC} = 0 \text{ to } V_{DD} - 1.35V,$	\/ 0.7\/	25°C	58	70		
	Common-mode rejection ratio	$R_S = 50 \Omega$	$V_{DD} = 2.7 \text{ V}$	Full range	55			dB
CMDD		$V_{IC} = 0$ to V_{DD} -1.35V, R _S = 50 Ω ,	V _{DD} = 5 V	25°C	65	80		
CMRR				Full range	62			
		$V_{IC} = -5 \text{ to } V_{DD} - 1.35V,$ $R_S = 50 \Omega,$	V _{DD} = ±5 V	25°C	69	85		
				Full range	66			
			., 07.,	25°C	97	106		
			$V_{DD} = 2.7 \text{ V}$	Full range	76			1
Δ	Large-signal differential voltage	$V_{O(PP)} = V_{DD}/2,$., 5.,	25°C	100	110		٩D
AVD	amplification	$R_L = 10 \text{ k}\Omega$	$V_{DD} = 5 V$	Full range	86			dB
			.5.4	25°C	100	115		
			$V_{DD} = \pm 5 \text{ V}$	Full range	90			

[†] Full range is 0°C to 70°C for C suffix and full range is -40°C to 125°C for I suffix. If not specified, full range is -40°C to 125°C.

input characteristics

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
			25°C		1	60	
liO	Input offset current		70°C			100	pА
		$V_{DD} = 5 \text{ V}, V_{IC} = V_{DD}/2,$	125°C			1000	
		$V_O = V_{DD}/2$, $R_S = 50 \Omega$	25°C		1	60	
I _{IB}	Input bias current		70°C			100	рΑ
			125°C			1000	
r _{i(d)}	Differential input resistance		25°C		1000		GΩ
C _{IC}	Common-mode input capacitance	f = 21 kHz	25°C		8		pF

TLV271, TLV272, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

electrical characteristics at specified free-air temperature, $V_{\mbox{DD}}$ = 2.7 V, 5 V, and ± 5 V (unless otherwise noted)

output characteristics

	PARAMETER	TEST CONDITIONS	3	T _A †	MIN	TYP	MAX	UNIT	
			.,	25°C	2.55	2.58			
			$V_{DD} = 2.7 V$	Full range	2.48				
		V V (0 1 4 A	., 5.,	25°C	4.9	4.93			
		$V_{IC} = V_{DD}/2$, $I_{OH} = -1 \text{ mA}$	$V_{DD} = 5 V$	Full range	4.85				
			\/ \ \F \/	25°C	4.92	4.96			
V	LPak lavel extend veltage		$V_{DD} = \pm 5 V$	Full range	4.9			V	
VOH	High-level output voltage		\/ 27\/	25°C	1.9	2.1		V	
			V _{DD} = 2.7 V	Full range	1.5				
		$V_{IC} = V_{DD}/2$, $I_{OH} = -5 \text{ mA}$	\/ F \/	25°C	4.6	4.68			
		VIC = VDD/2, $IOH = -5 IIIA$	V _{DD} = 5 V	Full range	4.5				
			\/n= - + E \/	25°C	4.7	4.84			
			$V_{DD} = \pm 5 V$	Full range	4.65				
			\/nn - 2.7.\/	25°C		0.1	0.15	V	
			V _{DD} = 2.7 V	Full range			0.22		
		$V_{IC} = V_{DD}/2$, $I_{OL} = 1 \text{ mA}$	V _{DD} = 5 V	25°C		0.05	0.1		
			100 = 0 1	Full range			0.15		
			V _{DD} = ±5 V	25°C		-4.95	-4.92		
\/a.	Low-level output voltage			Full range			-4.9		
VOL	Low-level output voltage		V _{DD} = 2.7 V	25°C		0.5	0.7		
			VDD = 2.7 V	Full range			1.1		
		$V_{IC} = V_{DD}/2$, $I_{OL} = 5 \text{ mA}$	V _{DD} = 5 V	25°C		0.28	0.4		
		V C = VDD/2, $V C = 3 IIIA$	VDD = 3 √	Full range			0.5		
			V _{DD} = ±5 V	25°C		-4.84	-4.7		
			VDD = ±3 V	Full range			-4.65		
		$V_O = 0.5 \text{ V from rail, } V_{DD} = 2.7 \text{ V}$	Positive rail	25°C		4			
		VO = 5.5 V HOITHAII, VDD = 2.7 V	Negative rail	25°C		5			
ا	Output ourrant	V 05V/ 11V -V	Positive rail	25°C		7]	
10	Output current	$V_O = 0.5 \text{ V from rail}, V_{DD} = 5 \text{ V}$	Negative rail	25°C		8		mA	
		V 05 V (mass mall V 10 V	Positive rail	25°C		13		1	
		$V_O = 0.5 \text{ V from rail}, V_{DD} = 10 \text{ V}$	Negative rail	25°C		12			

[†] Full range is 0°C to 70°C for C suffix and full range is -40°C to 125°C for I suffix. If not specified, full range is -40°C to 125°C.



[‡] Depending on package dissipation rating

TLV271, TLV272, TLV274 FAMILY OF 550- μ A/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

electrical characteristics at specified free-air temperature, $\rm V_{DD}$ = 2.7 V, 5 V, and ± 5 V (unless otherwise noted) (continued)

power supply

	PARAMETER	TEST COND	T _A †	MIN	TYP	MAX	UNIT	
		$V_O = V_{DD}/2$	$V_{DD} = 2.7 \text{ V}$	25°C		470	560	μΑ
I _{DD}	Supply current (per channel)		V _{DD} = 5 V	25°C		550	660	
			V 40 V	25°C		625	800	
			V _{DD} = 10 V	Full range			1000	
DODD	Supply voltage rejection ratio	$V_{DD} = 2.7 \text{ V to } 16 \text{ V},$	$V_{IC} = V_{DD}/2$,	25°C	70	80		-ID
PSRR	$(\Delta V_{DD} / \Delta V_{IO})$	No load		Full range	65			dB

[†] Full range is 0°C to 70°C for C suffix and full range is -40°C to 125°C for I suffix. If not specified, full range is -40°C to 125°C.

dynamic performance

PARAMETER		TEST CONDITIONS		T _A †	MIN	TYP	MAX	UNIT
UGBW	Unity gain bandwidth	$R_L = 2 \text{ k}\Omega$, $C_L = 10 \text{ pF}$	V _{DD} = 2.7 V	25°C		2.4		MHz
			V _{DD} = 5 V to 10 V	25°C		3		
	Slew rate at unity gain	$V_{O(PP)} = V_{DD}/2$, $C_{L} = 50 \text{ pF}$, $R_{L} = 10 \text{ k}Ω$,	V _{DD} = 2.7 V	25°C	1.35	2.1		V/μs
SR				Full range	1			
			V _{DD} = 5 V	25°C	1.45	2.4		V/μs
				Full range	1.2			
			V _{DD} = ±5 V	25°C	1.8	2.6		V/µs
				Full range	1.3			
φm	Phase margin	$R_L = 2 k\Omega$	C _L = 10 pF	25°C		65		0
	Gain margin	$R_L = 2 k\Omega$	C _L = 10 pF	25°C		18		dB
t _s	Settling time	$V_{DD} = 2.7 \text{ V},$ $V_{(STEP)PP} = 1 \text{ V}, A_{V} = -1,$ $C_{L} = 10 \text{ pF}, R_{L} = 2 \text{ k}\Omega$	0.1%	0500		2.9		_
		$V_{DD} = 5 \text{ V}, \pm 5 \text{ V},$ $V_{(STEP)PP} = 1 \text{ V}, A_{V} = -1,$ $C_{L} = 47 \text{ pF}, R_{L} = 2 \text{ k}\Omega$	0.1%	25°C		2		μs

Full range is 0°C to 70°C for C suffix and full range is -40°C to 125°C for I suffix. If not specified, full range is -40°C to 125°C.

noise/distortion performance

	PARAMETER	TEST CONDITIONS		TA	MIN	TYP	MAX	UNIT	
	Total harmonic distortion plus noise	$V_{DD} = 2.7 \text{ V},$ $V_{O(PP)} = V_{DD}/2 \text{ V},$ $R_{L} = 2 \text{ k}\Omega, \text{ f} = 10 \text{ kHz}$	A _V = 1	25°C		0.02%			
			A _V = 10			0.05%			
			A _V = 100			0.18%			
THD + N		$V_{DD} = 5 \text{ V}, \pm 5 \text{ V},$ $V_{O}(PP) = V_{DD}/2 \text{ V},$ $R_{L} = 2 \text{ k}\Omega, \text{ f} = 10 \text{ K}$	A _V = 1	25°C		0.02%			
			A _V = 10			0.09%			
			A _V = 100			0.50%			
.,	Employed and beauty as in a continue	f = 1 kHz		25°C		39		nV/√ Hz	
V _n	Equivalent input noise voltage	f = 10 kHz				35			
In	Equivalent input noise current	f = 1 kHz		25°C		0.6		fA/√ Hz	



TLV271, TLV272, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
CMRR	Common-mode rejection ratio	vs Frequency	1
	Input bias and offset current	vs Free-air temperature	2
VOL	Low-level output voltage	vs Low-level output current	3, 5, 7
Vон	High-level output voltage	vs High-level output current	4, 6, 8
V _{O(PP)}	Peak-to-peak output voltage	vs Frequency	9
l _{DD}	Supply current	vs Supply voltage	10
PSRR	Power supply rejection ratio	vs Frequency	11
AVD	Differential voltage gain & phase	vs Frequency	12
	Gain-bandwidth product	vs Free-air temperature	13
SR	Observation	vs Supply voltage	14
	Slew rate	vs Free-air temperature	15
φm	Phase margin	vs Capacitive load	16
V _n	Equivalent input noise voltage	vs Frequency	17
	Voltage-follower large-signal pulse response		18, 19
	Voltage-follower small-signal pulse response		20
	Inverting large-signal response		21, 22
	Inverting small-signal response		23
	Crosstalk	vs Frequency	24



TYPICAL CHARACTERISTICS

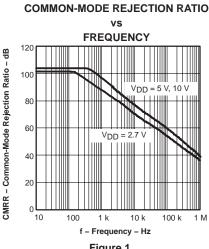
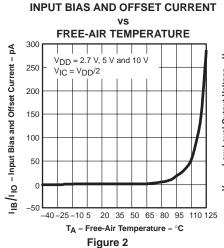


Figure 1



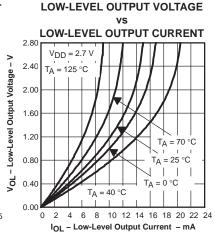


Figure 3

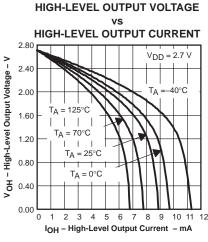


Figure 4

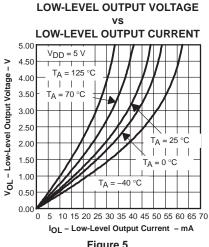


Figure 5

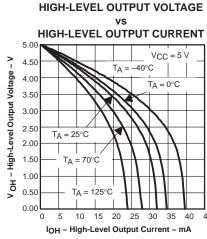
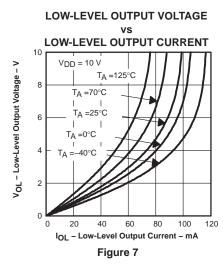
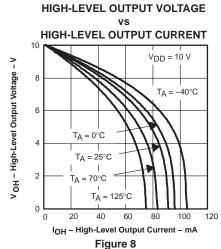


Figure 6

PEAK-TO-PEAK OUTPUT VOLTAGE





FREQUENCY Peak-to-Peak Output Voltage – V V_{DD} = 10 V $A_V = -10$ $R_L = 2 k\Omega$ $C_{L} = 10 \text{ pF}$ $T_A = 25^{\circ}C$ THD = 5% $V_{DD} = 5 V$ VO(PP) 10 k 10 M 100 100 k 1 M f - Frequency - Hz

Figure 9



TYPICAL CHARACTERISTICS

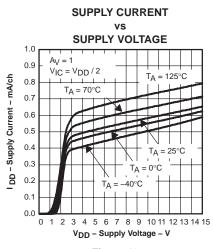


Figure 10

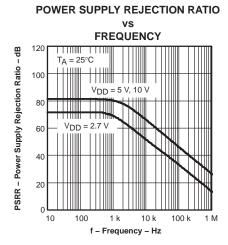


Figure 11

DIFFERENTIAL VOLTAGE GAIN AND PHASE

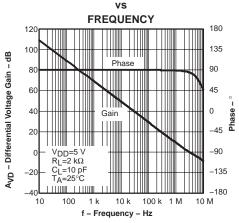


Figure 12

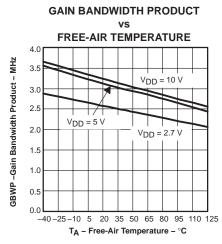


Figure 13

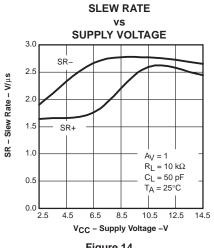
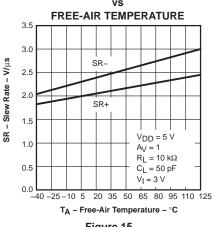
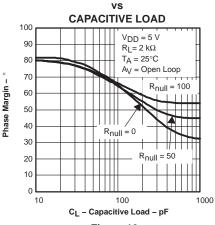


Figure 14



SLEW RATE

Figure 15



PHASE MARGIN

Figure 16



TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE

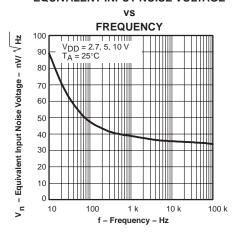


Figure 17

VOLTAGE-FOLLOWER LARGE-SIGNAL PULSE RESPONSE V₁ - Input Voltage - V $V_{DD} = 5 V$ $A_V = 1$ $R_L = 2 k\Omega$ C_L = 10 pF V_O - Output Voltage - V V_I = 3 V_{PP} T_A = 25°C 3 2 ۷o 0

t - Time - μs Figure 18

6 8 10 12 14 16 18

0 2 4

VOLTAGE-FOLLOWER LARGE-SIGNAL PULSE RESPONSE

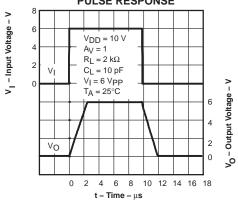
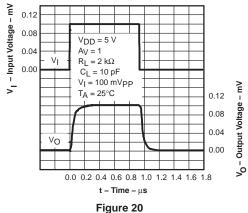


Figure 19

VOLTAGE-FOLLOWER SMALL-SIGNAL PULSE RESPONSE



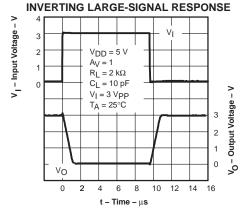


Figure 21

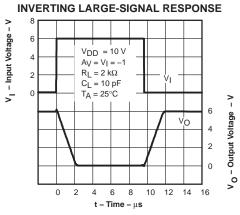
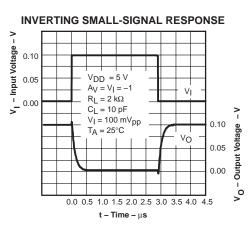
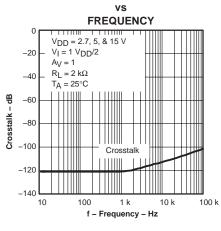


Figure 22

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TYPICAL CHARACTERISTICS





CROSSTALK

Figure 23

Figure 24

APPLICATION INFORMATION

driving a capacitive load

When the amplifier is configured in this manner, capacitive loading directly on the output decreases the device's phase margin leading to high frequency ringing or oscillations. Therefore, for capacitive loads of greater than 10 pF, it is recommended that a resistor be placed in series (R_{NULL}) with the output of the amplifier, as shown in Figure 25. A minimum value of 20 Ω should work well for most applications.

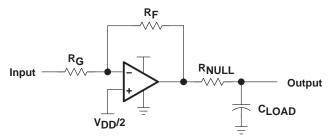


Figure 25. Driving a Capacitive Load



SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

APPLICATION INFORMATION

offset voltage

The output offset voltage, (V_{OO}) is the sum of the input offset voltage (V_{IO}) and both input bias currents (I_{IB}) times the corresponding gains. The following schematic and formula can be used to calculate the output offset voltage:

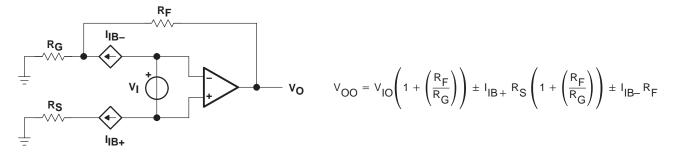


Figure 26. Output Offset Voltage Model

general configurations

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to accomplish this is to place an RC filter at the noninverting terminal of the amplifier (see Figure 27).

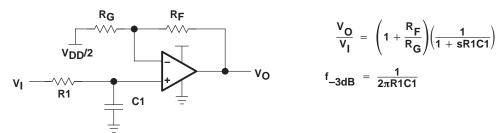


Figure 27. Single-Pole Low-Pass Filter

If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to do this can result in phase shift of the amplifier.

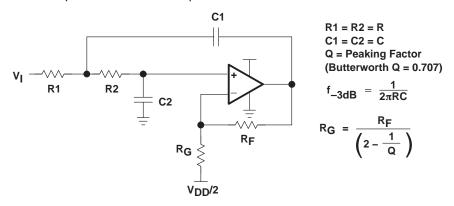


Figure 28. 2-Pole Low-Pass Sallen-Key Filter

TLV271, TLV272, TLV274 FAMILY OF 550-µA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS

SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

APPLICATION INFORMATION

circuit layout considerations

To achieve the levels of high performance of the TLV27x, follow proper printed-circuit board design techniques. A general set of guidelines is given in the following.

- Ground planes—It is highly recommended that a ground plane be used on the board to provide all
 components with a low inductive ground connection. However, in the areas of the amplifier inputs and
 output, the ground plane can be removed to minimize the stray capacitance.
- Proper power supply decoupling—Use a 6.8-μF tantalum capacitor in parallel with a 0.1-μF ceramic capacitor on each supply terminal. It may be possible to share the tantalum among several amplifiers depending on the application, but a 0.1-μF ceramic capacitor should always be used on the supply terminal of every amplifier. In addition, the 0.1-μF capacitor should be placed as close as possible to the supply terminal. As this distance increases, the inductance in the connecting trace makes the capacitor less effective. The designer should strive for distances of less than 0.1 inches between the device power terminals and the ceramic capacitors.
- Sockets—Sockets can be used but are not recommended. The additional lead inductance in the socket pins
 will often lead to stability problems. Surface-mount packages soldered directly to the printed-circuit board
 is the best implementation.
- Short trace runs/compact part placements—Optimum high performance is achieved when stray series
 inductance has been minimized. To realize this, the circuit layout should be made as compact as possible,
 thereby minimizing the length of all trace runs. Particular attention should be paid to the inverting input of
 the amplifier. Its length should be kept as short as possible. This helps to minimize stray capacitance at the
 input of the amplifier.
- Surface-mount passive components—Using surface-mount passive components is recommended for high
 performance amplifier circuits for several reasons. First, because of the extremely low lead inductance of
 surface-mount components, the problem with stray series inductance is greatly reduced. Second, the small
 size of surface-mount components naturally leads to a more compact layout thereby minimizing both stray
 inductance and capacitance. If leaded components are used, it is recommended that the lead lengths be
 kept as short as possible.



SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

APPLICATION INFORMATION

general power dissipation considerations

For a given θ_{JA} , the maximum power dissipation is shown in Figure 29 and is calculated by the following formula:

$$P_{D} = \left(\frac{T_{MAX}^{-T}A}{\theta_{JA}}\right)$$

Where:

P_D = Maximum power dissipation of TLV27x IC (watts)

T_{MAX} = Absolute maximum junction temperature (150°C)

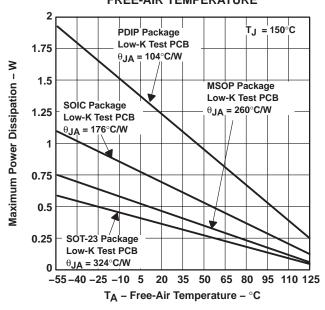
 T_A = Free-ambient air temperature (°C)

 $\theta_{JA} = \theta_{JC} + \theta_{CA}$

 θ_{JC} = Thermal coefficient from junction to case

 θ_{CA} = Thermal coefficient from case to ambient air (°C/W)

MAXIMUM POWER DISSIPATION vs FREE-AIR TEMPERATURE



NOTE A: Results are with no air flow and using JEDEC Standard Low-K test PCB.

Figure 29. Maximum Power Dissipation vs Free-Air Temperature

TLV271, TLV272, TLV274 FAMILY OF 550-μA/Ch 3-MHz RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS

SLOS351D - MARCH 2001 - REVISED FEBRUARY 2004

APPLICATION INFORMATION

macromodel information

Macromodel information provided was derived using Microsim $Parts^{TM}$ Release 9.1, the model generation software used with Microsim $PSpice^{TM}$. The Boyle macromodel (see Note 4) and subcircuit in Figure 30 are generated using TLV27x typical electrical and operating characteristics at $T_A = 25$ °C. Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate

 V_{DD}

- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification

- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

99

NOTE 2: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers," *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).

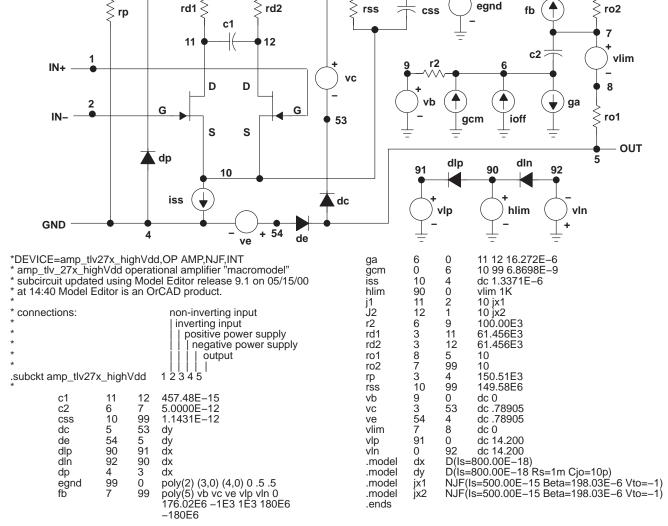


Figure 30. Boyle Macromodel and Subcircuit

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