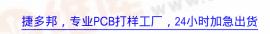
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TVP5022 Data Manual

NTSC/PAL VIDEO DECODER

SLAS274 July 2000







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I

1 Introduction

The TVP5022 is a high-quality single-chip digital video decoder that converts base-band analog NTSC and PAL video signals into digital component video. Both composite and S-video are supported. Sampling is square-pixel or ITU-R BT.601 (13.5 MHz) and is line-locked for correct pixel alignment. The output formats can be 8-bit or 16-bit 4:2:2, 12-bit 4:1:1, or 8-bit ITU-R BT.656. The TVP5022 uses TI patented technology for locking to weak, noisy, or unstable signals. A genlock control output is generated for synchronizing downstream video encoders.

Two-line (1-H delay) comb filtering is available for both the luma and chroma data paths to reduce both cross-luma and cross-chroma artifacts; a chroma trap filter is also available. Video characteristics including hue, contrast, and saturation are programmable using one of five supported host port interfaces. The TVP5022 generates synchronization, blanking, field, lock, and clock signals in addition to digital video outputs.

The TVP5022 includes advanced vertical blanking interval (VBI) data retrieval. The VBI data processor slices, parses, and performs error checking on teletext data in several formats. A built-in FIFO stores up to 14 lines of teletext data and, with proper host port synchronization, full-field teletext retrieval is possible. The VBI data processor also retrieves closed-caption data.

The main blocks of TVP5022 include:

- Analog processors and A/D converters
- Y/C separation
- Chrominance processor
- Luminance processor
- Clock/Timing processor and power-down control
- Output formatter
- Host port interface
- VBI data processor

1.1 Features

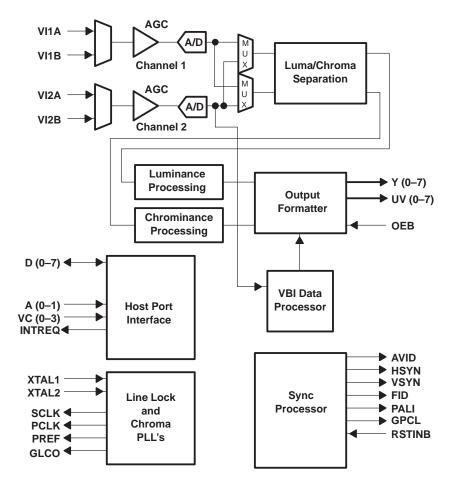
- NTSC (M) and PAL (B, D, G, H, I, M, N) composite video, S-video
- Four analog video inputs for up to 4 composite inputs or 2 S-video inputs
- Two built-in-analog signal processing channels with clamping and AGC
- Dual high speed 8-bit A/D converters
- Patented architecture for locking to weak, noisy, or unstable signals
- Comb filters for both cross-color and cross-luminance noise reductions
- Line locked clock and sampling
- Programmable data rates:
 - 12.2727 MHz square-pixel (NTSC)
 - 14.7500 MHz square-pixel (PAL)
 - 13.5 MHz ITU-R BT.601 (NTSC and PAL)

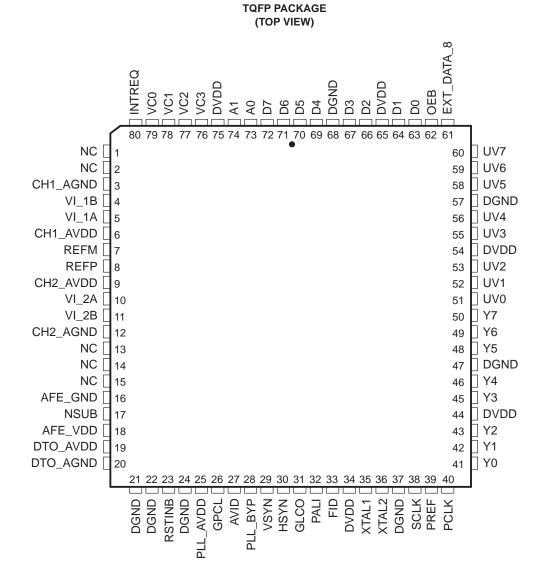
- Programmable output formats: 16-bit or 8-bit 4:2:2 YCbCr, 12-Bit 4:1:1 YCbCr and ITU-R BT.656 with embedded syncs
- Teletext (NABTS, WST) and closed caption decode with FIFO
- ITU-R BT.601 or extended coding range
- Programmable host port options including I²C, VMI (3 modes), and VIP
- 80-terminal TQFP package

1.2 Applications

- Digital image processing
- Video conferencing
- Multimedia
- Digital video
- Desktop video
- Video capture
- Video editing

1.3 Functional Block Diagram





1.4 Terminal Assignments

1–3

TERMINAL										
NAME	NO.	I/O	DESCRIPTIONS							
Analog video										
VI_1A VI_1B VI_2A VI_2B	5 4 10 11	I		Analog video inputs. Up to four composite inputs or two S-video inputs or a combination of the two. The inputs must be AC coupled. The recommended coupling capacitor is 0.1 μ F.						
Clock Signals										
PCLK	40	0	Pixel clock output. The fr 14.75 MHz for square-pixel l		1 1 1					
PREF	39	0	Clock phase reference signa to clock data that is changin		edges when SCLK is used					
SCLK	38	0	System clock output with tw	ice the frequency of the pix	el clock (PCLK).					
XTAL1 XTAL2	35 36	Ι	External clock reference. T oscillator or to one terminal o other terminal of the crysta sampling uses an oscillator f an oscillator frequency of 24	f a crystal oscillator. The use I oscillator or not connect 2 requency of 26.800 MHz. IT	r may connect XTAL2 to the XTAL2 at all. Square pixel					
Digital Video										
EXT_DATA_8	61	Ι	Bit [8] of a 9-or 10-bit digital composite video input							
UV[0:7]51, 52, 53, 55,I/ODigital chrominance outputs. These terminals may be configured to output data fr the channel 2 A/D converter. A vendor modifiable subsystem ID may be initialized configuring the UV [7:0] terminals with pull-up/pull-down resistors.59, 6059, 60										
Y[0:7]	41, 42, 43, 45, 46, 48, 49, 50	0	Digital luminance outputs, or multiplexed luminance and chrominance outputs These terminals may be configured to output data from the channel 1 A/D converter							
HOST PORT-b	ous		VMI	l ² C	VIP					
A[0:1]	73, 74	Ι	VMI address port							
		I/O	VMI data port – bit [7:0].							
INTREQ	80	0	Interrupt request (INTREQ)		Interrupt request (VIRQ)					
VC0	79	I/O	VMI port data ack. or ready signal (DTACK)	Serial clock (SCL)	Hardware address bit-0 (HAD[0])					
VC1	78	I/O	VMI Port Read-Write or Serial data (SDA) Hardware addres Write (RW/WR) HAD[1]							
VC2	77	I/O	VMI port data strobe or read signal (DS/RD) Hardware control (HCTL)							
VC3	76	I	VMI port chip select. (VC)	Slave address select (I ² CA)	VIPCLK					

1.5 Terminal Functions

1.5	Terminal	Functions	(Continued)
-----	----------	-----------	-------------

TERMINAL		1/0	DESCRIPTIONS				
NAME	NO.	1/0	DESCRIPTIONS				
Miscellaneous	s signals						
GPCL	26	I/O	General-purpose control logic. This terminal has three functions:				
			 General-purpose output. In this mode the state of GPCL is directly programmed via the host port. 				
			2. Vertical blank output. In this mode the GPCL terminal is used to indicate the vertical blanking interval of the output video. The beginning and end times of this signal are programmable via the host port control.				
			3. Sync lock control input. In this mode when GPCL is high, the output clock frequencies and sync timing are forced to nominal values.				
GLCO	31	I/O	This serial output carries color PLL information. A slave device can decode the information to allow genlocking to the TVP5022. Data is transmitted at the SCLK rate. Additionally, this terminal, in conjunction with PALI and FID, is used to determine the host port mode configuration during initial power up.				
OEB	62	I	Output enable for Y and UV terminals. Output enable is also controllable via the host port.				
PLL_BYP	28		Connect a 0.1 μ F capacitor between this terminal and PLL_AVDD				
RSTINB	23	I	Reset input, active low				
NC	1, 2, 13, 14, 15		Not connected				
Power Supplie	es						
AFE_GND	16		Analog ground				
AFE_VDD	18		Analog supply, connect to 5 V				
CH1_AGND CH2_AGND	3 12		Analog grounds				
CH1_AVDD CH2_AVDD	6 9		Analog supply, connect to 5 V				
DGND	21, 22, 24, 37, 47, 57, 68		Digital grounds				
DTO_AGND	20		DTO ground, connect to analog ground				
DTO_AVDD	19		DTO supply, connect to 5 V analog				
DVDD	34, 44, 54, 65, 75		Digital supply, connect to 3.3 V				
NSUB	17		Substrate ground, connect to analog ground				
PLL_AVDD	25		PLL supply connect to 3.3 V				
REFP	8		A/D reference supply, connect to 5 V analog				
REFM	7		A/D reference ground, connect to analog ground				

TERMIN	TERMINAL		DECODIDITIONS
NAME	NO.	1/0	DESCRIPTIONS
Sync Signals			
AVID	27	0	Active video indicator. This signal is high during the horizontal active time of the video output on the Y and UV terminals. AVID continues to toggle during vertical blanking intervals.
FID 33		I/O	Odd/even field indicator or vertical lock indicator. For odd/even indicator, a logic 1 indicates the odd field. For vertical lock indicator, a logic 1 indicates the internal vertical PLL is in a locked state. Additionally, this terminal in conjunction with GLCO and PALI is used to determine the host port mode configuration during initial power up and reset.
HSYN	30	0	Horizontal sync signal. The rising edge time is programmable via the host.
PALI	32	I/O	PAL line indicator or horizontal lock indicator. For PAL line indicator, a logic 1 indicates a noninverted line, and a logic 0 indicates an inverted line. For horizontal lock indicator, a logic 1 indicates the internal horizontal PLL is in a locked state. Additionally, this terminal in conjunction with GLCO and FID is used to determine the host port mode configuration during initial power up.
VSYN	29	0	Vertical sync signal

1.5 Terminal Functions (Continued)

2 Detailed Description

2.1 Analog Video Processors and A/D Converters

Figure 2–1 is a functional diagram of the TVP5022 analog video processors and A/D converters. This block accepts up to four inputs and performs analog signal conditioning (i.e., video clamping, video amplifying), and carries out analog-to-digital conversion.

2.1.1 Video Input Selection

Four high-impedance video inputs are sources for two internal analog channels in the TVP5022. Internal multiplexers, controlled by the host port, select the desired input. The user can connect the four analog video inputs in the following combinations:

- Four selectable individual composite video inputs
- 1 S-video input and two composite video inputs
- 2 S-video inputs

2.1.2 Analog Input Clamping and Automatic Gain Control Circuits

The internal clamp circuit restores the ac-coupled video signals to a fixed dc level before A/D conversion. The clamping circuit provides line-by-line restoration of the video sync level to a fixed dc reference voltage. The circuit has two modes of clamping, coarse and fine. In coarse mode, the most negative portion of the signal (typically the sync tip) is clamped to a fixed dc level. The circuit uses fine mode to prevent spurious level shifting caused by noise that is more negative than the sync tip on the input signal. When fine mode is enabled, after the sync position is detected, clamping is only enabled during the sync period. S-video requires fine clamping mode for proper operation.

Input video signal amplitudes may vary significantly from the nominal level of 1 Vpp. An automatic gain control (AGC) circuit adjusts the signal amplitude to use the maximum range of the A/D converters without clipping.

The AGC circuit adjusts the signal amplitude based on the detected amplitude of the sync portion of the input signal. Signal peaks may be present on nonstandard signals that cause clipping at the A/D converters after gain adjustment based only on sync amplitude. In these cases, the signal will be attenuated before A/D conversion.

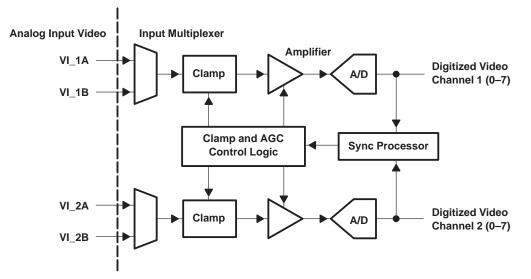


Figure 2–1. Analog Video Processors and A/D Converters

2.1.3 A/D Converters

The TVP5022 contains two 8-bit A/D converters which digitize the analog video signal inputs. To prevent high frequencies which are above half of the sampling rate from entering into the system, video inputs may require external anti-aliasing low pass filters.

2.2 Digital Processing

Figure 2–2 is a block diagram of the TVP5022 digital video processing. This block receives digitized composite or S-video signals from the A/D converters, and performs Y/C separation, chroma demodulation, and Y-signal enhancements. It also generates the horizontal and vertical syncs. The YUV digital output may be programmed into various formats: 16-bit or 8-bit 4:2:2, 12-bit 4:1:1 and ITU-R BT.656 parallel interface standard. The circuit uses comb filters to reduce the cross-chroma and cross-luma noise.

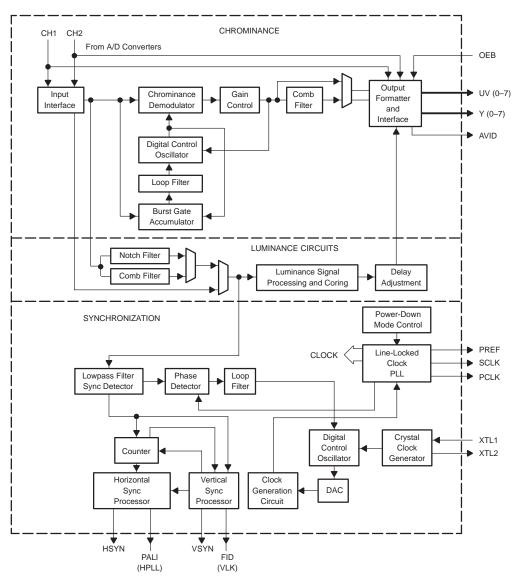


Figure 2–2. Digital Video Signal Processing Block Diagram

2.2.1 Y/C Separation

Luma/chroma separation may be done using either 2-line (1–H delay) comb filtering or a chroma trap filter. Comb filtering is available for both the luminance and the chrominance portion of the data path. The characteristics of the chroma trap filter are shown in Figures 2–3 and 2–4.

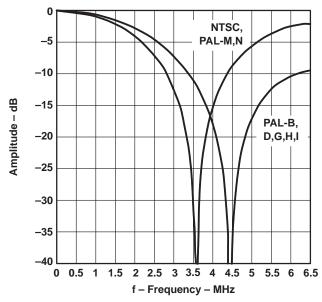


Figure 2–3. Chroma Trap Filter Frequency Response for 13.5 MHz Sampling

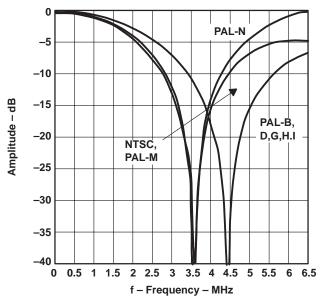


Figure 2–4. Chroma Trap Filter Frequency Response for Square-Pixel Sampling

2.2.2 Luminance Processing

The digitized composite video signal from the output of the A/D converters passes through a luminance comb filter or a chroma trap filter that removes the chrominance signal from the composite signal to generate the luminance signal. The luminance signal is then fed to the input of luminance signal peaking and coring circuits. Figure 2–5 illustrates the functions of the luminance data path. In the case of S-video, the luminance signal will bypass the comb filter or notch filter and be fed to the peaking and coring circuits directly. High frequency components of the luminance signal are enhanced further by the peaking filter (edge enhancer).

Figures 2–6, 2–7, and 2–8 show the characteristic of the peaking filter at maximum gain. The coring circuit reduces low-level, high -frequency noise. Figure 2–9 shows the transfer curve of the coring function. The peaking frequency, peaking gain, and coring threshold are programmable.

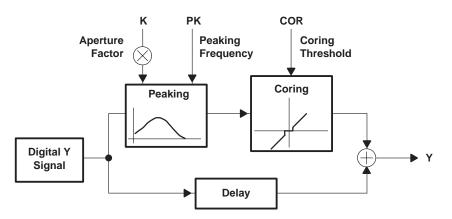


Figure 2–5. Luminance Edge-Enhancer

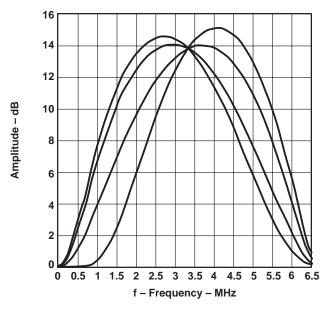


Figure 2–6. Peaking Filter Response, 13.5 MHz Sampling

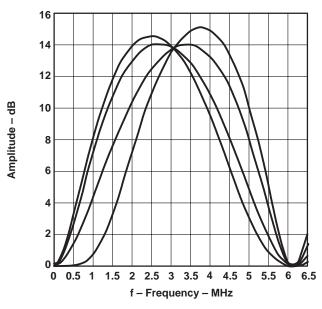


Figure 2–7. Peaking Filter Response, NTSC AND PAL-M SQUARE PIXEL

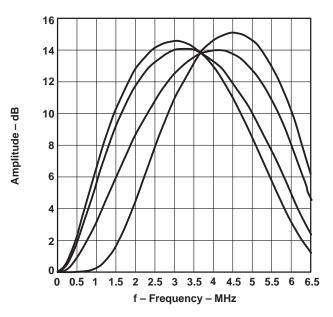


Figure 2–8. Peaking Filter Response, PAL Square Pixel

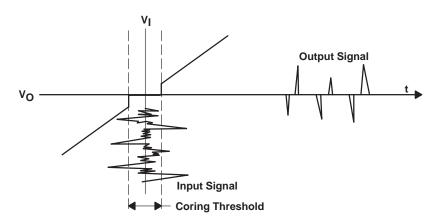


Figure 2–9. Transfer Curve of Coring Circuit

2.2.3 Chrominance Processing

A quadrature demodulator removes the U and V components from the composite signal in composite video mode, or the U and V components from the chroma signal in S-video mode. The U/V signals then pass through the gain control stage for chroma saturation adjustment. The U and V components pass through a comb filter to eliminate cross-chrominance noise. Phase shifting the digitally-controlled oscillator controls hue. The block includes an automatic color killer (ACK) circuit that suppresses the chroma output when the color burst of the video signal is weak or not present.

2.2.4 Clock Circuits

An internal line-locked PLL generates the system and pixel clocks. Figure 2–10 shows a simplified clock circuit diagram. The digital control oscillator (DCO) generates the reference signal for the horizontal PLL. The DCO outputs a signal that is fed to the D/A converter. The D/A converter outputs a line-locked clock signal (LCLK). The DCO requires a 26.8 or a 24.576 MHz clock as an input. The input for the DCO may enter terminal XTAL1 as a TTL level on terminal XTAL1 or a 26.8 or 24.576 MHz crystal connected across terminals XTAL1 and XTAL2. Figure 2–11 shows the various reference clock configurations.

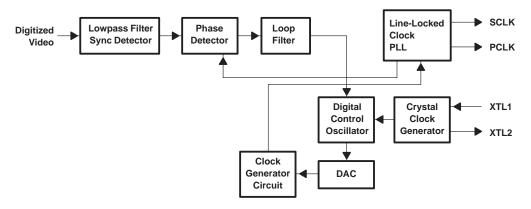


Figure 2–10. Clock Circuit Diagram

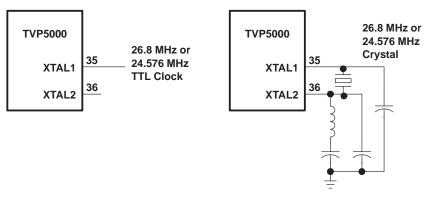


Figure 2–11. Reference Clock Configurations

The TVP5022 generates three signals PCLK, SCLK, and PREF used for clocking data. PCLK is the pixel clock at the sampling frequency. PCLK can be used for clocking data in the 16-bit 4:2:2 and the 12-bit 4:1:1 output formats. SCLK is at twice the sampling frequency and may be used for clocking data in the 8-bit 4:2:2 and ITU-R BT.656 formats. PREF is used as a clock qualifier with SCLK to clock data in the 16-bit 4:2:2 and the 12-bit 4:1:1 formats

2.3 Video Output Format

The TVP5022 supports both square-pixel and ITU-R BT.601 sampling formats and multiple output formats:

- 16-bit 4:2:2
- 12-bit 4:1:1
- 8-bit 4:2:2
- ITU-R BT.656

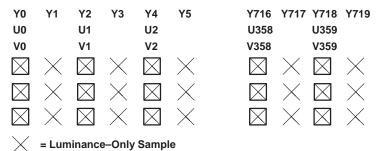
2.3.1 Sampling Frequencies and Patterns

The sampling frequencies that control the number of pixels per line differ depending on the video format and standards. Table 2–1 shows a summary of the sampling frequencies.

 Table 2–1.
 Summary of the Line Frequencies, Data Rates, and Pixel Counts

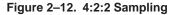
STANDARDS	HORIZONTAL LINE RATE (kHz)	PIXELS PER LINE	ACTIVE PIXELS PER LINE	PCLK (MHz)	SCLK (MHz)
NTSC, square-pixel	15.73426	780	640	12.2727	24.54
NTSC, ITU-R BT.601	15.73426	858	720	13.5	27.0
PAL (B,D,G,H,I), square-pixel	15.625	944	768	14.75	29.5
PAL (B,D,G,H,I), ITU-R BT.601	15.625	864	720	13.5	27.0
PAL(M), square-pixel	15.73426	780	640	12.2727	24.54
PAL(M), ITU-R BT.601	15.73426	858	720	13.5	27.0
PAL(N), square-pixel	15.625	944	768	14.75	29.5
PAL(N), ITU-R BT.601	15.625	864	720	13.5	27.0

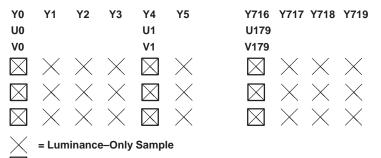
Depending on the output format chosen, The TVP5022 outputs data in the 4:2:2 or 4:1:1 sampling pattern. The patterns differ in the number of chrominance samples derived from the original samples. For the 4:2:2 pattern, every second sample is both a luminance and a chrominance sample; the remainder are luminance-only samples. For the 4:1:1 pattern, every fourth sample is both a luminance and a chrominance sample; the remainder are sample; the remainder are luminance-only samples.



= Luminance and Chrominance Sample

Numbering Shown is for 13.5 MHz Sampling.



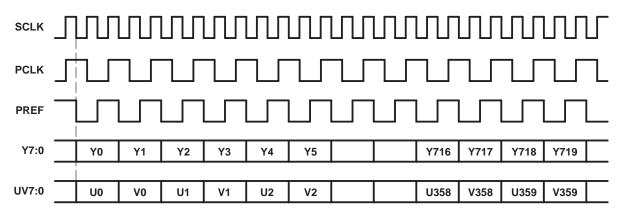


= Luminance and Chrominance Sample

Numbering Shown is for 13.5 MHz Sampling.

Figure 2–13. 4:1:1 Sampling

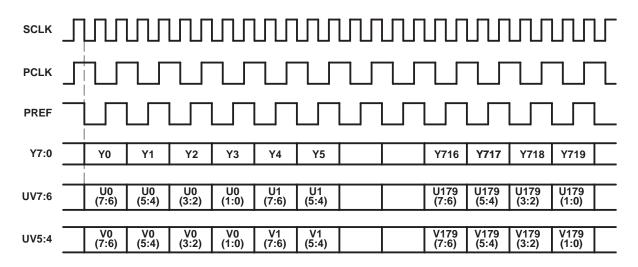
2.3.2 Video Port 16-Bit 4:2:2 Output Format Timing



Numbering Shown is for 13.5 MHz Sampling.

Figure 2–14. 16-Bit 4:2:2 Output Format

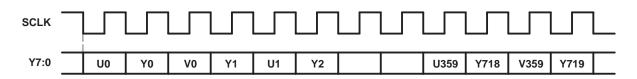
2.3.3 Video Port 12-Bit 4:1:1 Output Format Timing



UV3:0 Logic 0 Numbering Shown is for 13.5 MHz Sampling.

Figure 2–15. 12-bit 4:1:1 Output Format

2.3.4 Video Port 8-Bit 4:2:2 and ITU-R BT.656 Output Format Timing



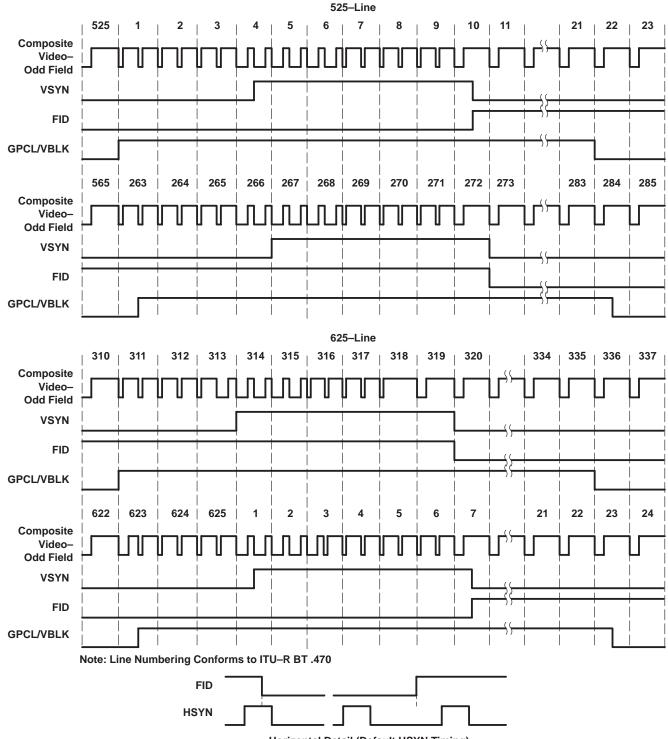
UV7:0 High Impedance

Numbering Shown is for 13.5 MHz Sampling.

Figure 2–16. 8-Bit 4:2:2 and ITU-R BT.656 Output Formats

2.4 Synchronization Signals

The TVP5022 generates synchronization signals to accompany the output video data stream. The following figures show the default timing for the horizontal and vertical sync and associated signals.



Horizontal Detail (Default HSYN Timing) Figure 2–17. Vertical Synchronization Signals

e.
k (SCLK) Referenc
(SCLK)
Cloc
2x Pixel
vith 2x
Timing v
t 4:2:2 Ti
8-Bit 4

_					1	_				1	1	
3	3	≻	3	≻ -		m	≻ -					
2	2	ບັວ	2	సం		7	రం					
-	٢	≻₀	٢	≻0		-	≻ 0					
0	0	о Ср	0	0 CP		•	o Cp					
1715	1727	xx	1559	ХХ		1887	хх					
1714	1726	00	1558	00		1886	00					
1713	1725	8	1557	00		1885	00					
1712	1724	ΕF	1556	ЦЦ		1884	ΕF					
1711	1723	10	 1555	10		1883	10	 				
:		:	:	:		:	:					
1560	1592	80	1452	80		1716	80					
1599	1591	10	 1451	10		1715	10	 				
:	:	:	:	:		:	:					
1472	1464	80	1324	80		1588	80					
1471	1463	10	 1323	10		1587	10	 	_			
:	:	:	:	:		:	:					
1441	1441	00	1281	00		1537	00					
1440	1440	Ξ	1280	ΕF		1536	FF					
1439	1439	۲ 719	1279	Υ 639		1535	۲67 ۲67					
1438	1438	Cr 359	 1278	Cr 319		1534	Cr 383					
1437	1437	۲ 718	1277	Υ 638		1533	۲ 766					
1436	1436	Cb 359	1276	Cb 319		1532	Cb 383					
NTSC 601	PAL 601	ITU 656 Datastream	NTSC sqp	ITU 656 Datastream		PAL sqp	ITU 656 Datastream			HSYN		AVID

16-Bit 4:2:2 Timing with 1x Pixel Clock (PCLK) Reference.

1	1	1	1						
0	0	0	0						
857	863	779	943		_				
856	862	778	942						
855	861	777	941						
		:							
800	796	726	858						
799	795	725	857						
:		:							
736	732	662	794				_		
735	731	661	793						
:		::	:						
720	720	640	768						
719	719	639	767						
718	718	638	766						
NTSC 601	PAL 601	NTSC sqp	PAL sqp			NYSH		AVID	

Figure 2–18. Horizontal Synchronization Signals

2.5 I²C/VIP/VMI Host Interface

Communication with the TVP5022 is via an interface that is configurable at power up to support an I²C, VIP, or VMI bus host. The host interface accesses status and control registers and retrieves sliced VBI data. The host interface also initializes the TVP5022's internal microprocessor.

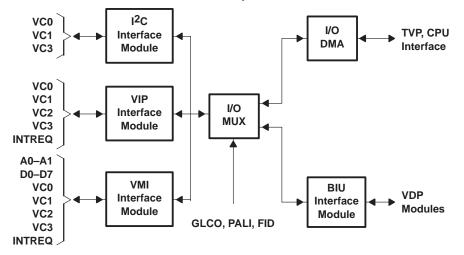


Figure 2–19. Host Interface

The host port mode is selected by attaching external pullup and pulldown resistors to the GLCO, PALI, and FID terminals. The TVP5022 samples the state of the terminals at trailing edge of RSTINB and configures the host port accordingly. Table 2–2 shows the pullup/pulldown combinations required to select each of the host port modes. Figure 2–19 is a block diagram of the host interface showing the mode selection and host interface terminals, as well as interfaces to the internal microprocessor and VBI data processor.

TERMINALS	GLCO	PALI	FID						
TERMINALS	2	1	0						
I ² C Host Port	0	0	1						
VIP Host Port	0	1	0						
VMI Host Port Mode A	1	0	1						
VMI Host Port Mode B	1	1	0						
VMI Host Port Mode C	1	1	1						

Table	2–2.	Host	Port	Select
Table	~ ~.	11031	I OIL	001001

2.6 I²C Interface

The TVP5022 host port interface is configured for $I^{2}C$ operation by attaching external pull-up and pull-down resisters to the GLCO, PALI, and FID terminals. The following is the combination of resisters required to select the $I^{2}C$ host mode. (1 is pullup and 0 is pulldown)

TERMINALS	GLCO	PALI	FID
TERMINALS	2	1	0
I ² C Host Port Enabled	0	0	1

2.6.1 I²C Host Port Select

The I²C standard consists of two signals, serial input/output data (VC1) line and input/output clock line (VC0), that carry information between the devices connected to the bus. A third signal (VC3) is used for slave address selection. Although the I²C system can be multimastered, the TVP5022 will function as a slave device only.

Both SDA and SCL are bidirectional lines that connect to a positive supply voltage via a pullup resistor. When the bus is free, both lines are high.

The slave address select terminal (VC3) enables the use of two TVP5022 devices tied to the same I²C bus.

Table 2–3 summarizes the terminal functions of the I²C mode host interface.

SIGNAL	TYPE	DESCRIPTION
VC3 (I2CA)	I	Slave address selection
VC0 (SCL)	I/O (OD)	Input/output clock line
VC1 (SDA)	I/O (OD)	Input/output data line

Table 2–3. I²C Host Port Terminal Description

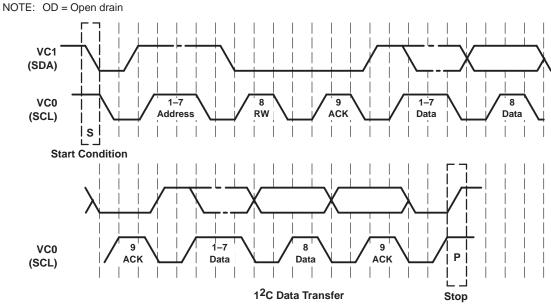


Figure 2–20. I²C Data Transfer Example

Data transfer rate on the bus is up to 400 kbits/s. The number of interfaces connected to the bus is dependent on the bus capacitance limit of 400 pF. The data on the SDA line must be stable during the high period of the clock. The high or low state of the data line can only change with the clock signal on the SCL line being low.

- When transferring multiple bytes during one read or write operation, the internal subaddress is not automatically incremented.
- A high to low transition on the SDA line while the SCL is high indicates a start condition.
- A low to high transition on the SDA line while the SCL is high indicates a stop condition
- Acknowledge is signalled by SDA low during the ninth SCL high.
- Not-acknowledge is signalled by SDA high during the ninth SCL high.

Every byte placed on the SDA line must be 8 bits long. The number of bytes that can be transferred is unrestricted. An acknowledge bit follows each byte. If the slave can not receive another complete byte of data until it has performed another function, it holds the clock line (SCL) low. An SCL low forces the master into a wait state. Data transfer continues when the slave is ready for another byte of data and releases the clock line (SCL).

Data transfer with acknowledge is necessary. The master generates an acknowledge related clock pulse. The master releases the SDA line high during the acknowledge clock pulse. The slave pulls down the SDA line during the acknowledge clock pulse so that it remains stable low during the high period of this clock pulse.

When a slave does not acknowledge the slave address, the data line is left high. The master then generates a stop condition to abort the transfer.

If a slave acknowledges the slave address, but some time later in the transfer cannot receive any more data bytes, the master again aborts the transfer. The slave indicates a not ready condition by generating the not acknowledge. The slave leaves the data line high and the master generates the stop condition.

If a master-receiver is involved in a transfer, it indicates the end of data to the slave-transmitter by not generating an acknowledge on the last byte that was clocked out of the slave. The slave-transmitter must release the data line to allow the master to generate a stop or repeated start condition.

2.6.2 I²C Write Operation

Data transfers occur using the following illustrated formats.

The I²C master initiates a write operation to the TVP5022 by generating a start condition followed by the TVP5022's I²C address (101110X). The address is in MSB first bit order followed by a 0 to indicate a write cycle. After receiving a TVP5022 acknowledge, the I²C master sends a subaddress of the register or the block of registers where it will write. Following the subaddress is one or more bytes of data, with MSB first. The TVP5022 acknowledges the receipt of each byte upon completion of each transfer. The I²C master ends a write operation by generating a stop condition.

The X in the address of the TVP5022 is 0 when the I^2CA terminal is low and the X is 1 when the I^2CA is high. If the read or write cycle contains more than one byte, the internal subaddress does not increment automatically.

		-						
	0							
I ² C Start (Master)	S]						
	7	6	5	4	3	2	1	0
I ² C General address (Master)	1	0	1	1	1	0	Х	0
	9]						
I ² C Acknowledge (Slave)	A]						
	7	6	5	4	3	2	1	0
I ² C Write register address (Master)	Addr	Addr	Addr	Addr	Addr	 Addr	Addr	Addr
	9]						
I ² C Acknowledge (Slave)	А	1						
	7	6	5	4	3	2	1	0
I ² C Write data (Master)	Data	Data	Data	Data	Data	Data	Data	Data
	9	1						
I ² C Acknowledge (Slave)	А]						
	0							
I ² C Stop (Master)	Р	1						

2.6.3 I²C Read Operation

The read operation has two phases, the address phase and the data phase. In the address phase, the I²C master initiates a write operation to the TVP5022 by generating a start condition followed by the TVP5022's I²C address (101110X). The address is in MSB first bit order followed by a 0 to indicate a write cycle. After receiving a TVP5022 acknowledge, the I²C master sends a subaddress of the register or the block of registers where it will read. The TVP5022 acknowledges the receipt of the address. The I²C master ends the address phase by generating a stop condition. During the data phase, the I²C address (101110X). The address is in MSB first bit order followed by the TVP5022's I²C address (101110X). The address is in MSB first bit order followed by a 1 to indicate a read cycle. After receiving a TVP5022 acknowledge, the TVP5022 transfers a data byte to the master. The I²C master acknowledges the receipt of each byte upon completion of each transfer. After the TVP5022 transfers the last byte, the I²C master ends the read operation by generating a not acknowledge followed by a stop condition.

Read Address Phase

	0							
I ² C Start (Master)	S]						
	7	6	5	4	3	2	1	0
I ² C General address (Master)	1	0	1	1	1	0	Х	0
	9	1						
I ² C Acknowledge (Slave)	A	1						
		J T		1			1	
	7	6	5	4	3	2	1	0
I ² C Write register address (Master)	Addr	Addr	Addr	Addr	Addr	Addr	Addr	Addr
	9]						
I ² C Acknowledge (Slave)	A	1						
		1						
I ² C Stop (Master)	0	1						
	1 '	1						
Read Data Phase								
	0]						
I ² C Start (Master)	S]						
	7	6	5	4	3	2	1	0
I ² C General address (Master)	1	0	1	1	1	0	X	1
		1						
	9	-						
I ² C Acknowledge (Slave)	A]						
	7	6	5	4	3	2	1	0
I ² C Read data (Slave)	Data	Data	Data	Data	Data	Data	Data	Data
	9	1						
I ² C Acknowledge (Master)	/A	1						
		- 1						
12C Stop (Master)	0	-						
I ² C Stop (Master)		l						

2.6.4 I²C Microcode Write Operation

Data written during the microcode write operation will be written to the TVP5022 program RAM. During the write cycle the microprocessor resets and points to location zero in the program and remains reset. Upon completion of the write operation, the microprocessor requires a restart operation. To perform a clear-reset requires writing into the 7F register to clear reset and resume microprocessor function. (The 7F register requires no specific data written into the register, any data will resume microprocessor function).

		-						
	0							
I ² C Start (Master)	S]						
								•
	7	6	5	4	3	2	1	0
I ² C General address (Master)	1	0	1	1	1	0	Х	0
		1						
	9	4						
I ² C Acknowledge (Slave)	A]						
	7	6	5	4	3	2	1	0
I ² C Write register address (Master)	0	1	1	1	1	1	1	0
		_					-	
	9							
I ² C Acknowledge (Slave)	Α							
	7	6	5	4	3	2	1	0
I ² C Write data (Master)	Data							
		1						
	9	4						
I ² C Acknowledge (Slave)	A]						
	0	1						
I ² C Stop (Master)	P	1						
	Г							

Microprocessor CLEAR Reset

	0]						
I ² C Start (Master)	S							
	7	6	5	4	3	2	1	0
	1	0	5	4	3	2	'	0
I ² C General address (Master)	1	0	1	1	1	0	Х	0
		1						
	9							
I ² C Acknowledge (Slave)	А	1						
	7	6	5	4	3	2	1	0
I ² C Write register address (Master)	0	1	1	1	1	1	1	1
	· ·	1		-			-	
	9							
I ² C Acknowledge (Slave)	А							
		-					1	
	7	6	5	4	3	2	1	0
I ² C Write data (Master)								

Any Data Written to 7F Will Start the Microprocessor

	9
I ² C Acknowledge (Slave)	А
	0
I ² C Stop (Master)	Р

2.6.5 I²C Microcode Read Operation

Data read during the microcode read operation will be read from the TVP5022 Program RAM. During the read cycle the microprocessor resets and points to location zero in the program and remains reset. Upon completion of the read operation, the microprocessor requires a restart operation. To perform a clear–reset requires writing into the 7F register to clear reset and resume microprocessor function. (The 7F register requires no specific data written into the register, any data will resume microprocessor function).

Read Address Phase

	0]					
I ² C Start (Master)	S]					
	7	6	5	4	3	2	1
I ² C General address (Master)	1	0	1	1	1	0	Х
	9]					
I ² C Acknowledge (Slave)	А]					
	7	6	5	4	3	2	1
I ² C Read register address (Master)	1	0	0	0	1	1	1
	9	1					
I ² C Acknowledge (Slave)	A	1					

	0
I ² C Stop (Master)	Р

Read Data Phase

	0
I ² C Start (Master)	S

	7	6	5	4	3	2	1	0
I ² C General address (Master)	1	0	1	1	1	0	Х	1

		7						
I ² C Acknowledge (Slave)		A						
	7	6	5	4	3	2	1	0
I ² C Read data (Slave)	Data							
		7						
I ² C Acknowledge (Master)		A						

NOTE: Repeat byte read sequence for N–1 bytes where N is the total number of bytes to be read

	7	6	5	4	3	2	1	0
I ² C read data (Slave)	Data							
	0	1						

	9
I ² C Acknowledge (Master)	/A

	0
I ² C Stop (Master)	Р

Microprocessor CLEAR Reset

	0]						
I ² C Start (Master)	S]						
	7	6	5	4	3	2	1	0
I ² C General address (Master)	1	0	1	1	1	0	Х	0
		_						
	9							
120 1 1 (01)		1						
I ² C Acknowledge (Slave)	A							
I2C Acknowledge (Slave)	A] 						
I2C Acknowledge (Slave)	7	6	5	4	3	2	1	0

Write to FIFO address=7F

	9
I ² C Acknowledge (Slave)	А

	7	6	5	4	3	2	1	0
I ² C Write data (Master)	Data							
NOTE: Any data written to 7F will start the microprocessor.								

	9
I ² C Acknowledge (Slave)	А
	0
I ² C Stop (Master)	Р

2.7 VIP Host Interface Port

The TVP5022 host interface is configured for video interface port (VIP) by attaching external pullup and pulldown resisters to the GLCO, PALI, and FID terminals. The following is the combination of resisters required to select the VIP host mode. (1 is pullup and 0 is pulldown)

TERMINALS	GLCO	PALI	FID
TERMINALS	2	1	0
VIP host port enabled	0	1	0

The VIP is a standard interface, conforming to the video interface port (VIP) specification Version 1.1. The interface is between a video enabled graphics device and one or more video devices. The video port of VIP transports various types of real-time signal streams. Signal names in parentheses denote the signal name referenced in VIP specification.Host port transfers require five terminals, VC3, VC0, VC1, VC2, INTREQ.

2.7.1 VIP Host Port Terminal Description

Table 2–4.	VIP Host Port	Terminal	Description
			Dooonpuon

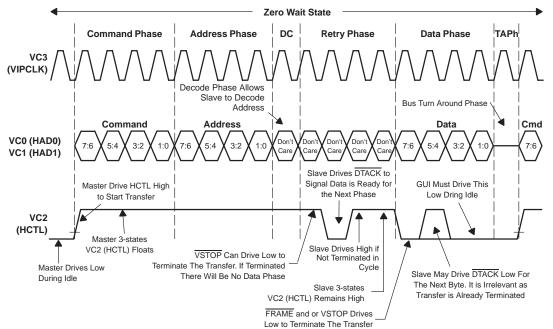
SIGNAL	TYPE	DESCRIPTION
VC3 (VIPCLK)	I	VIP Host clock (25 MHz – 33 MHz)
VC0 VC1 (HAD[0:1])	I/O	Host address/data bus VC0 = (HAD_0) VC1 = (HAD_1)
VC2 (HCTL)	I/O (OD)	Host control: This includes the symbolic signals of VFRAME, DTACK, and VSTOP
INTREQ (VIRQ#)	O (OD)	Interrupt request. Shared signal with VDP

VC3 (VIPCLK) is the host port clock, operating from 25 MHz–33 MHz. VIPCLK can come from any source.

VC0 and VC1 (HAD[0:1]) is a two-wire bus that transfers commands, addresses, and data between master and slave devices.

VC2 (HCTL) is a shared control terminal. The master drives it to initiate and terminate data transfers. The slave drives it to terminate and add wait states to data transfers. Because VC2 is a shared control signal, special attention must be given to its generation to avoid bus conflicts.

INTREQ is a normally open drain pin that signals interrupts to the host controller. Using the interrupt configuration register at subaddress C2, this terminal can be configured as a conventional CMOS I/O buffer (non-open drain). Conflict is possible if multiple devices are connected to the INTREQ signal when it is configured in non-open drain mode.



2.7.2 VIP Phases

Figure 2–21. VIP Transfer Example

PHASE	EXPLANATION
Command	All host port transfers start with a command phase. The 8-bit command/address byte is multiplexed onto VC0 and VC1 (HAD[1:0]) during the command phase. The command byte selects between devices, read, and write cycles, register or FIFO transfers, and contains the most significant four bits of the register address.
Address	During register transfers the command phase is followed by the address extension phase. The least significant 8-bits of the VIP register address are multiplexed onto VC0 and VC1 (HAD[1:0]) during the address extension phase. This phase is not present during FIFO transfers.
Decode	Following the command or command/address phase(s), the bus requires one clock delay to allow slave devices to decode the address and determine if they are able to respond within the 1 wait phase requirement for active operation.
Retry	The four clock cycles immediately following the decode phase constitute the retry phase. During the retry phase, the slave indicates its desire to terminate the operation without transferring any data (retry), add a wait phase or transfer the first byte of data. When the slave asserts VSTOP, the transfer ends with the retry phase. When the slave neither terminates the transfer nor accepts the byte, the retry phase is followed by a wait phase.
Wait	During the second cycle of a decode, retry or wait phase, the slave indicates its ability to transfer the next byte of data by driving VC2 (HCTL) low. When the slave does not drive VC2 (HCTL) low and the transfer is not terminated, the current phase is followed by a wait phase. During wait phases, the current owner (master for writes, slave for reads) continues to drive the HAD bus, but no data is transferred. The slave is allowed to add one wait phase per byte to register accesses without compromising system timing. Additional wait phases are not prevented but overall system reliability may be compromised.
Data	When VC2 (HCTL) is removed during cycle 1 of a retry, wait or data phase, a data phase follows the current phase. The bus transfers data between master and slave devices during data phases. The data is multiplexed onto VC0 and VC1 (HAD[1:0]).
T _A	Immediately following the last transfer phase of a read transfer, the slave requires a one cycle delay giving time to 3-state the VC0 and VC1 (HAD) bus. The master is free to begin a new bus transfer, driving VC0 and VC1 (HAD) and VC2 (HCTL) immediately following the T_A phase.

Table 2–5. VIP Host Port Phase Description

Table 2–6. Condensed Table Command/Address

CO	MMAN	ID	Cmd/Addr	REGISTER ADDRESS	DATA	COMMENT
[7:4]			[3:0]	[7:0]	[7:0]	
01	0/1	0	0000	00000000 thru 11111111	ddddddd	VIP configuration registers
01	0/1	0	0001	00000000 thru 11111111	ddddddd	General TVP registers
01	1	0	0010	00000000 thru 11111111	XXXXXXXX	No latency read access 1 phase
01	1	0	0011	addr as prev. written	dddddd	No latency read access 2 phase
01	1	1	0000	No addr phase	xxx0/1xxx0/1	FIFO status 0 read
01	1	1	0001	No addr phase	xxxxxx11	FIFO status 1 read
01	1	1	0100	No addr phase	ddddddd	FIFO VBI data read

2.7.3 Command Byte

During the command byte phase, the hardware control line (VC2) will transition high and the hardware address lines (VC0 and VC1) transmits the command byte from the host to the TVP5022. The command byte determines the nature of the data transfer and the affected TVP5022 address space.

	7	6	5	4	3	2	1	0
Command	DEVSEL1 (0)	DEVSEL0 (1)	R/W	F/R	A11	A10	A9	A8

NAME	DESCRIPTION
DEVSEL1:0	Device select. Always 01 for TVP5022
R/W	1=Read 0=Write
F/R	1=FIFO 0=Register access
A11:8	Address bus upper 4 bits For register accesses: 0000=VIP specific configuration registers 0001=General TVP5022 registers 0010=No latency read access phase 1 0011=No latency read access phase 2 For FIFO accesses: 0000=FIFO status 0 0001=FIFO status 1 0100=VBI FIFO

Command Byte Bit Description

2.7.3.1 Access Latency and Wait States

VIP accesses to registers or the VBI FIFO require the TVP5022 to insert one or more wait states into the access sequence. For register accesses, the wait states may total as much as $64 \,\mu$ s. All writes will release the host port immediately, but internal wait states will continue to be generated until the operation completes. Any access attempts to the host port while the write operation is not completed will cause the slave to terminate the operator. Reads (except for no-latency reads) will hold the host port until completion.

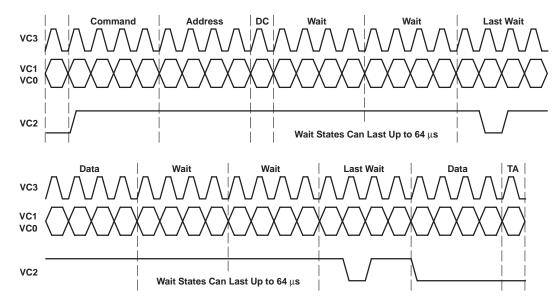


Figure 2–22. Functional Timing Reading From TVP (Example)

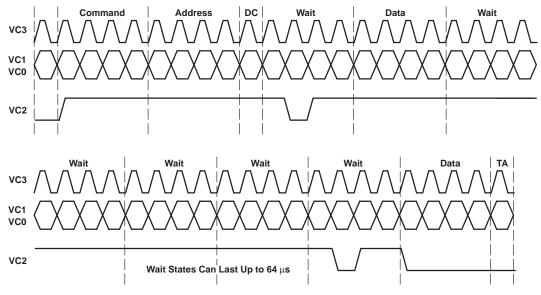
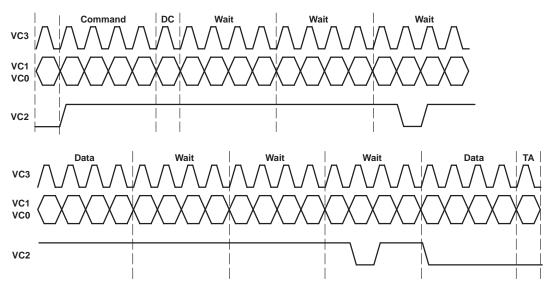


Figure 2–23. Functional Timing Writing to TVP (Example)





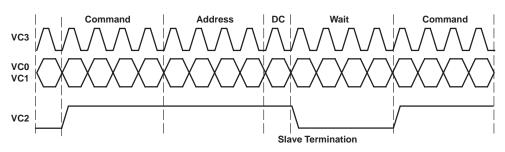


Figure 2–25. Functional Timing Example of Slave Termination

VIP Configuration Register:

The TVP5022 supports VIP configuration registers that are accessible only in the VIP host mode.

		со	MM	AN	ID I	РΗΑ	SE			AD	DR	ES	S P	HA	SE				DAT froi					
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
VIP Config. Register read	0	1	1	0	0	0	0	0	А	А	А	А	А	А	А	А	D	D	D	D	D	D	D	D

General TVP5022 Registers:

The bulk of the TVP5022 register space consists of status and control registers that are available to the I^2C , VIP, and VMI host modes.

		со	MN	IAN	ID I	РΗΑ	SE			AD	DR	ES	S P	HA	SE							SE 020		
	7 6 5 4 3 2 1 0 7 6 5 4 3 2 1 0 7 6 5 4 3 2 1 0 7 6 5 4 3 2 1											1	0											
General TVP register read	0 1 1 0 0 0 1 A A A A A A A A D D D D D D											D	D											
		со	MN	IAN	ID I	PH/	SE			AD	DR	ES	S P	HA	SE			[A F TV		SE 20)		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
General TVP register write	0 1 0 0 0 0 0 1 A A A A A A A A D D D D D D									D														

2.7.3.2 No-latency Read

No-latency read eliminates the extended wait states of normal read operations. No latency read consists of two zero-wait state phases separated by an idle period during which the host may perform other operations. The first phase identifies the register address to be read. This will be stored in an intermediate data buffer. The current data in the intermediate data buffer is immediately read out. This information is to be ignored because it is invalid data.

Following the completion of the first phase, the host must wait for at least 64 μ s to ensure that the data requested in the first phase is stored in the intermediate data buffer. The host initiates the second phase of the read by reading the data from the intermediate data buffer.

No Latency Read Phase 1:

		COMMAND PHASE									DR	ES	S P	HA	SE					NTA F om T				
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
No latency read phase 1	0	1	1	0	0	0	1	0	А	А	A	А	А	А	А	А	du	du	du	du	du	du	du	du

No Latency Read Phase 2:

		COMMAND PHASE									DR	ES	S P	HA	SE					A F TV		SE 20)		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
No Latency Read Phase 2	0	1	1	0	0	0	1	1	А	А	А	А	А	А	А	А	D	D	D	D	D	D	D	D

2.7.3.3 FIFO Status 0 Register

FIFO status 0 register returns two bits that report status and six unused bits.

	7	6	5	4	3	2	1	0
FIFO Status 0	Undefined	Undefined	Undefined	DREQA	Undefined	Undefined	Undefined	VIRQ

DREQA: DMA request for FIFO A. This bit is the same as the teletext threshold bit (bit 1 of the interrupt status register at VIP address1C0).

VIRQ: this bit returns the status of the INTREQ terminal. Reading this register does not clear this terminal.

		0	СОМ	MAN	D Pł	IASI	=				D	ATA F	PHAS	SE	_	
	7 6 5 4 3 2 1 0 7 6 5 4 3 2											2	1	0		
FIFO Status 0	0	1	1	1	0	0	0	0	d	d	d	0/1	d	d	d	0/1

There is no address associated with reading the FIFO status 0 register.

2.7.3.4 FIFO Status 1 Register

The FIFO status 1 register returns two bits that report status and six unused bits.

	7	6	5	4	3	2	1	0
FIFO Status 1	Undefined	Undefined	Undefined	Undefined	Undefined	Undefined	R/W	PRESENT

R/W This bit identifies the VBI FIFO is read–only and is always set to 1. PRESENT This bit indicates the VBI FIFO is present and is always set to 1.

		C	СОМ	MAN	D Pł	IASE	Ξ				DA		PHA	SE		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
FIFO Status 1	0	1	1	1	0	0	0	1	dc	dc	dc	dc	dc	dc	1	1

There is no address associated with reading the FIFO status 1 register.

2.7.3.5 VBI FIFO

The VBI FIFO stores sliced VBI data. The data may be read from the FIFO at an average rate of one data byte per three cycles (1 data cycle and 2 wait cycles) when the VIPCLK is at maximum speed.

		C	OM	MAN	ND P	HAS	E				DA	TA F	PHAS	SE		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
FIFO VBI data read	0	1	1	1	0	1	0	0	D	D	D	D	D	D	D	D

There is no address associated with reading the VBI FIFO.

2.7.4 **VIP Microcode Write Operation**

Data is written to the TVP5022 program RAM during the microcode write operation. During the write cycle, the microprocessor resets and points to location zero in the program and remains reset. Following the first data phase, the data phase is repeated until all microcode is written. The microprocessor requires a clear-reset operation upon completion of the write operation. The host performs the reset by writing into the 7F register to clear reset and resume microprocessor function. (There is no specific data to be written into the 7F register; any data will resume microprocessor function).

		со	MM	AN	ID F	РΗΑ	SE			AD	DR	ES	S P	HA	SE					TA F		-		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Microcode write	0	1	0	0	0	0	0	1	0	1	1	1	1	1	1	0	D	D	D	D	D	D	D	D

The data phase is repeated to the end of the microcode.

		со	MM	AN	DF	РНА	SE			AD	DR	ES	S P	HA	SE					A F		SE 20)		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Clear reset	0	1	0	0	0	0	0	1	0	1	1	1	1	1	1	1	Х	Х	Х	Х	Х	Х	Х	Х

2.7.5 **VIP Microcode Read Operation**

Data is read from the TVP5022 program RAM during the microcode read operation. During the read cycle, the microprocessor resets and points to location zero in the program and remains reset. Following the first data phase, the data phase is repeated until all microcode is read. The microprocessor requires a clear-reset operation upon completion of the read operation. The host performs the reset by writing into the 7F register to clear reset and resume microprocessor function. (There is no specific data to be written into the 7F register; any data will resume microprocessor function).

		со	MM	AN	ID F	РΗΑ	SE			AD	DR	ES	S P	HA	SE							SE 020		
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Microcode read	0	1	1	0	0	0	0	1	1	0	0	0	1	1	1	0	D	D	D	D	D	D	D	D

The data phase is repeated to the end of the microcode.

		COMMAND PHASE						AD	DR	ES	S P	HA	SE			[DAT (to			-				
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Clear reset	0	1	0	0	0	0	0	1	0	1	1	1	1	1	1	1	Х	Х	Х	Х	Х	Х	Х	Х

2.8 Video Module Interface (VMI) Host Interface

The VMI host port interface is configurable for three possible modes of operation. The mode of operation is determined by attaching pullup or pulldown resistors to the GLCO, PALI, and FID terminals. Table 2-7 shows the various modes. (0=pulldown and 1=pullup)

Table 2–7. VMI Host F	Port Sele	ect	
TERMINALS	GLCO	PALI	FID
TERMINALS	2	1	0
VMI Host Port Mode A	1	0	1
VMI Host Port Mode B	1	1	0
VMI Host Port Mode C	1	1	1

Table 2–7.	VMI Host Port Select

VMI modes A and B are from the Video Electronics Standards Association (VESA) video module interface (VMI) proposal version 1.4. Mode C is designed to conform to the interface requirements of the IBM PowerPC 403GC. Table 2–8 summarizes the terminal functions of the VMI mode host interface.

SIGNAL	VMI SIGNAL NAME	ТҮРЕ	DESCRIPTION
VC3	CS	I	Active low chip select - mode B
VCO	DTACK – mode A READY – mode B	O (see below)	Data acknowledge – mode A Data ready – modes B and C
VC1	R/W – mode A WR – mode B	I	Read/write – modes A and C Write strobe – mode B
VC2	DS – mode A RD – mode B	I	Data strobe – mode A and C Read strobe – mode B
A1:A0	HA[1:0]	I	Address bus from host
D7:D0	HD[7:0]	I/O	Input/output data bus from host
INTREQ	INTREQ	O (nominal open drain)	Interrupt request

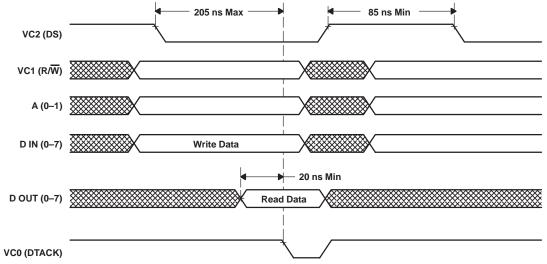
Table 2-8.	VMI Host	Port	Terminal	Definitions
------------	----------	------	----------	-------------

INTREQ is a nominally open drain terminal that signals interrupts to the host controller. The interrupt configuration register at subaddress C2 can configure this terminal as a conventional CMOS I/O buffer (non-open drain). Conflict is possible if INTREQ is connected to multiple devices and INTREQ is not configured in the open drain mode.

VC0 (DTACK/READY) is in the high impedance state when VC3 is not asserted.

2.9 Host Port – Mode A Timing

Host port mode A has a bus interface that accommodates the Motorola type of control signals. The diagram below shows the timing of the mode A signals. The host initiates the cycle when VC2 transitions low. The target responds by pulling VC0 low to indicate it is receiving the data or that the requested data is present on the bus. The host completes its cycle by pulling VC2 high. Once the host completes its cycle, the target pulls VC0 high. The host may change VC1, A[1:0], and Din[7:0] as soon as it receives VC0.





2.10 Host Port – Mode B Timing

Host port mode B has a bus interface that accommodates the Intel type of control signals. The diagram below shows the timing of the mode B signals. The host presents the address and chip select when the command (VC1 or VC2) transitions low. The VMI module pulls VC0 low until the read or write operation is complete. VC1 and VC2 cannot both be low while VC3 is low.

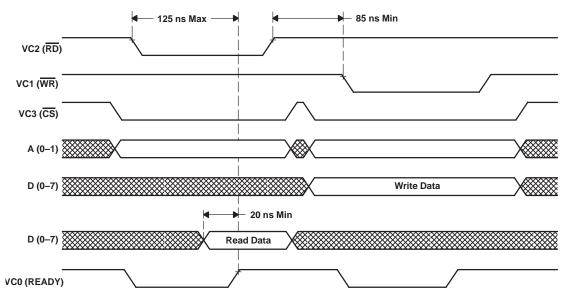


Figure 2–27. Host Port Mode B Timing

2.11 Host Port – Mode C Timing (PPC)

Host port mode C has a bus interface that accommodates the IBM PowerPC 403GC type of control signals. The Figure 2–21 shows the timing of the mode C signals. Mode C is similar to mode A (Motorola) except that a ready signal is used instead of VC0.

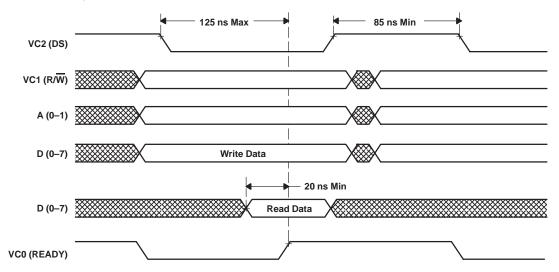


Figure 2–28. Host Port Mode C Timing

2.11.1 VMI Register Mapping

The VMI module contains only four registers that are accessible by the host. The address register holds an indirect address for internal register access. When the host accesses the data register, the VMI module reads or writes the internal register selected by the indirect address register.

Two other registers are provided for direct access. The FIFO register provides direct access to the VBI FIFO and the second direct access register is the status/interrupt register. This register contains the state of the interrupt sources.

A[1:0]	
00	Address register
01	Data register
10	FIFO
11	Status register

Figure 2–29. VMI Address Register Map

Normally, read or write operations require two accesses. To read the FIFO register, set A[1:0] to 10(binary) and perform a read cycle. The FIFO read data will be placed on the D[7:0] bus. To read/write the status/interrupt register, set A[1:0] to 11(binary) and perform a read/write cycle. The read/write will be muxed appropriately to/from the external data bus.

Indirect Register Read/Write:

All VMI accesses except for the VBI FIFO and the status/interrupt register require a two-step operation. To access an indirect register, write the internal address to the VMI address register. The first step requires setting A[1:0] to 00 and performing a write cycle with D[7:0] equaling the indirect address. To write to an indirect register, the second step is to write the data to VMI address 01. To read an indirect register, the second step is to read the the requested data from address 01.

STEP 1	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write register address	0	0	Register address							
STEP 2	A1	A0	D7 D6 D5 D4 D3 D2 D1 D0							
Read register data	0	1	Data from register							

Read the indirect register

Write to the indirect register

STEP 1	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write register address	0	0	Register address							
STEP 2	A1	A0	D7 D6 D5 D4 D3 D2 D1 D0							
SIEF 2	AI	AU	Di	00	D5	D4	D3	D2	וט	D0
Write register data	0	1	Data to register							

Latency:

VMI access to indirect addresses 00–8F requires special consideration due to response latencies of up to 64 μ s for these addresses. Latency occurs between steps one and two for a read operation and following step two of a write operation. To avoid violating the VMI cycle time requirements, the host can poll the cycle complete bit in the VMI status register following step one for a read or step two for a write. Alternatively, the cycle complete enable bit in the interrupt enable register (indirect address C1) can be set to generate an interrupt for the host when access is complete.

VMI access to indirect addresses 90-CF occur with minimal latency. Interrupts will not be generated for the completion of access cycles to these addresses.

VBI FIFO:

The VBI FIFO containing sliced VBI data is directly readable by the VMI host.

	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Read VBI FIFO	1	0	Data from FIFO							

Status/Interrupt Register:

The status/interrupt register provides the host with information containing the source of an interrupt. After an interrupt condition is set, the condition can be reset by writing a one to the appropriate bit in the status/interrupt register.

		A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Ac	cess status/interrupt register	1	1	Data to/from status/interrupt register				er			

2.11.2 VMI Microcode Write Operation

Data written to indirect register 7E will be written to the TVP5022 program RAM. During the address write cycle the microprocessor resets and points to location zero in the program and remains reset. The microprocessor requires a clear-reset operation upon completion of the write operation. The host performs the reset by writing into the 7F register to clear reset and resume microprocessor function. (There is no specific data to be written into the 7F register; any data will resume microprocessor function).

To avoid violating VMI cycle time requirements during microcode write operation, the host can poll the cycle complete bit in the VMI status register after writing each byte of data to the VMI data register. Alternatively, the cycle complete enable bit in the interrupt enable register (indirect address C1) can be set to generate an interrupt for the host when a write operation is complete.

	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write microcode Register address	0	0	0	1	1	1	1	1	1	0

	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0	
Write microcode	0	1		First byte of microcode data							
Register data Write microcode Register Data	0	1		(Wait for cycle complete status or interrupt) Second Byte of microcode data (Wait for cycle complete status or interrupt)							
Write microcode Register data	0	1		Last byte of microcode data (Wait for cycle complete status or interrupt)							
	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0	
Write clear reset Register address	0	0	0 1 1 1 1 1 1 1						1		
	A1	A0	D7 D6 D5 D4 D3 D2 D1 D0								
Write clear reset Dummy data	0	1	Х	Х	Х	Х	Х	Х	Х	Х	

2.11.3 VMI Microcode Read Operation

Data read from indirect register 8E will be read from the TVP5022 program RAM. During the read cycle, the microprocessor resets, points to location zero in the program, and remains reset. The microprocessor requires a clear-reset operation upon completion of the read operation. The host performs the reset by writing into the 7F register to clear reset and resume microprocessor function. (There is no specific data to be written into the 7F register; any data will resume microprocessor function).

To avoid violating VMI cycle time requirements during the microcode read operation, the host can poll the cycle complete bit in the VMI status register. This polling shoud be done following the address load cycle (for indirect register 8Eh). Alternatively, the cycle complete enable bit in the interrupt enable register (indirect address C1) can be set to generate an interrupt for the host when data from a read cycle is unavailable.

STEP 1	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write microcode Register address	0	0	1	0	0	0	1	1	1	0
STEP 2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Read register data	0	1	Data							

STEP 1	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write microcode Register address	0	0	1	0	0	0	1	1	1	0
STEP 2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Read register data	0	1	Data	dData						
	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write clear reset Register address	0	0	0	1	1	1	1	1	1	1
	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
Write clear reset Dummy data	0	1	Data							

Note: Must readdress for each read of incremental data

2.12 Genlock Control

The frequency control word of the internal color subcarrier digital control oscillator (DCO) and the subcarrier phase reset bit are transmitted via the GLCO terminal. The frequency control word is a 23-bit binary number. The frequency of the DCO can be calculated from the following equation:

$$F_{dco} = \frac{F_{ctrl}}{2^{23}} \times F_{sclk}$$

Where F_{dco} is the frequency of the DCO, F_{ctrl} is the 23-bit DCO frequency control, and F_{sclk} is the frequency of the SCLK.

The last bit (bit 0) of the DCO frequency control is always 0.

A write of 1 to bit 4 of the chrominance control register at host port subaddress 1Ah causes the subcarrier DCO phase reset bit to be sent on the next scan line on GLCO. The active low reset bit occurs 8 SCLKs after the transmission of the last bit of DCO frequency control. Upon the transmission of the reset bit, the phase of the TVP5022 internal subcarrier DCO is reset to zero.

A genlocking slave device connected to the GLCO terminal can use the information on GLCO to synchronize its internal color DCO to achieve clean line and color lock.

Figure 2–30 shows the timing of GLCO.

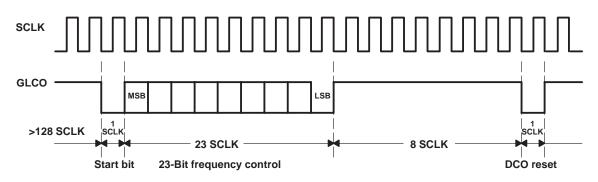


Figure 2–30. GLCO Timing

2.13 VBI Data Processor

The TVP5022 VBI data processor slices, parses, and performs error checking on teletext data contained in the vertical blanking interval or during active lines. Teletext formats supported are North American Basic Teletext Specification (NABTS) equivalent to ITU-R BT.653 system C, and World System Teletext (WST) equivalent to ITU-R BT.653 system B. Data is stored in an internal FIFO and may be read via the host port or transmitted as ancillary data in the digital video stream in BT.656 mode. The VBI data FIFO holds up to 14 lines of NABTS or 11 lines of WST data. Interrupts generated by the VBI data processor are configurable to enable host synchronization for retrieval of full-field teletext data.

VBI data processor may also slice closed caption data and store it in two registers accessible via the host port.

2.13.1 Teletext Data Byte Order

Table 2–9 shows the order in which teletext data is read from the FIFO.

BYTE NUMBER	NABTS – 525 OR 625 LINE SYSTEM	WST – 525 LINE SYSTEM	WST – 625 LINE SYSTEM							
1	Video line [7:0]	Video Line [7:0]	Video Line [7:0]							
2	00, Hamming error, parity error, LPC error, match #2, match #1, video line [8]	00, Hamming error, parity error, LPC error, match #2, match #1, video line [8]	00, Hamming error, parity error, LPC error, match #2, match #1, video line [8]							
3	Packet address 1	Magazine	Magazine							
4	Packet address 2	Row address	Row address							
5	Packet address 3	Data byte 1	Data byte 1							
6	Continuity index	Data byte 2	Data byte 2							
7	Packet structure	Data byte 3	Data byte 3							
8	Data block 1	Data byte 4	Data byte 4							
9	Data block 2	Data byte 5	Data byte 5							
10	Data block 3	Data byte 6	Data byte 6							
11	Data block 4	Data byte 7	Data byte 7							
12	Data block 5	Data byte 8	Data byte 8							

Table	2_9	Teletext	Byte	Order
Iable	Z-J.	ICICICAL	Dyte	Oldel

BYTE NUMBER	NABTS – 525 OR 625 LINE SYSTEM	WST – 525 LINE SYSTEM	WST – 625 LINE SYSTEM
13	Data block 6	Data byte 9	Data byte 9
14	Data block 7	Data byte 10	Data byte 10
15	Data block 8	Data byte 11	Data byte 11
16	Data block 9	Data byte 12	Data byte 12
17	Data block 10	Data byte 13	Data byte 13
18	Data block 11	Data byte 14	Data byte 14
19	Data block 12	Data byte 15	Data byte 15
20	Data block 13	Data byte 16	Data byte 16
21	Data block 14	Data byte 17	Data byte 17
22	Data block 15	Data byte 18	Data byte 18
23	Data block 16	Data byte 19	Data byte 19
24	Data block 17	Data byte 20	Data byte 20
25	Data block 18	Data byte 21	Data byte 21
26	Data block 19	Data byte 22	Data byte 22
27	Data block 20	Data byte 23	Data byte 23
28	Data block 21	Data byte 24	Data byte 24
29	Data block 22	Data byte 25	Data byte 25
30	Data block 23	Data byte 26	Data byte 26
31	Data block 24	Data byte 27	Data byte 27
32	Data block 25	Data byte 28	Data byte 28
33	Data block 26	Data byte 29	Data byte 29
34	Data block 27/suffix	Data byte 30	Data byte 30
35	Data block 28/suffix	Data byte 31	Data byte 31
36	Padding byte [†]	Data byte 32	Data byte 32
37			Data byte 33
38			Data byte 34
39			Data byte 35
40			Data byte 36
41			Data byte 37
42			Data byte 38
43			Data byte 39
44			Data byte 40

Table 2–9. Teletext Byte Order (Continued)

[†] The padding byte is used to ensure an even number of writes. This byte does not contain any useful information. The read pointer automatically advances past this byte so the user does not have to read the padding byte.

2.13.2 Teletext as Ancillary Data in Video Stream

Sliced teletext data can be output as ancillary data in the video stream in ITU-R BT.656 mode. Teletext data is output on the Y7:0 pins during the horizontal blanking period following the line that the data was retrieved. Dummy ancillary data blocks with special timing header information are inserted during certain horizontal blanking periods to provide data synchronization information. Tables 2–10 through 2–13 show the format and sequence of the ancillary data inserted into the video stream.

BYTE NO.	MSB 7	6	5	4	3	2	1	LSB 0	DESCRIPTION
1	0	0	0	0	0	0	0	0	
2	1	1	1	1	1	1	1	1	Ancillary data preamble
3	1	1	1	1	1	1	1	1	
4	NEP	EP	0	1	0	DID2	DID1	DID0	Data ID
5	1	0	0	0	0	0	0	0	Secondary data ID
6	1	0	0	0	1	0	0	1	Number of 32-bit data words
7	Video I	ine nur	mber 7:0						
8	0	0	Hamming error	Parity error	LPC error	Match #2	Match #1	Video line 8	Internal data ID
9	Packet	addres	ss 1						Data byte
10	Packet	addres	ss 2						Data byte
11	Packet	addres	ss 3						Data byte
12	Contin	uity ind	ex						Data byte
13	Packet	structu	ure						Data byte
14	Teletex	t data	1						Data byte
15	Teletex	t data 2	2						Data byte
39	Teletex	t data 2	26						Data byte
40	Teletex	t data 2	27/suffix						Data byte
41	Teletex	t data 2	27/suffix						Data byte
42	Checks	sum							Checksum
43	1	0	0	0	0	0	0	0	Fill byte
44	1	0	0	0	0	0	0	0	Fill byte

Table 2–10. NABTS 525/625-Line Ancillary Data Sequence

BYTE NO.	MSB 7	6	5	4	3	2	1	LSB 0	DESCRIPTION	
1	0	0	0	0	0	0	0	0		
2	1	1	1	1	1	1	1	1	Ancillary data preamble	
3	1	1	1	1	1	1	1	1		
4	NEP	EP	0	1	0	DID2	DID1	DID0	Data ID	
5	1	0	0	0	0	0	0	0	Secondary data ID	
6	1	0	0	0	1	0	0	1	Number of 32-bit data words	
7	Video line number 7:0									
8	0	0	Hamming error	Parity error	LPC error	Match #2	Match #1	Video line 8	Internal data ID	
9	Magaz	ine		Data byte						
10	Row ad	ddress							Data byte	
11	Teletex	t data	1						Data byte	
12	Teletex	t data	2						Data byte	
									Data byte	
42	Teletex	t data	32						Data byte	
43	NEP	EP	Checksum						Checksum	
44	1	0	0	0	0	0	0	0	Fill byte	

Table 2–11. WST 525-Line Ancillary Data Sequence

Table 2–12. WST 625-Line Ancillary Data Sequence

BYTE NO.	MSB 7	6	5	4	3	2	1	LSB 0	DESCRIPTION	
1	0	0	0	0	0	0	0	0		
2	1	1	1	1	1	1	1	1	Ancillary data preamble	
3	1	1	1	1	1	1	1	1		
4	NEP	EP	0	1	0	DID2	DID1	DID0	Data ID	
5	1	0	0	0	0	0	0	0	Secondary data ID	
6	0	1	0	0	1	0	1	1	Number of 32-bit data words	
7	Video I	/ideo line number 7:0								
8	0	0	Hamming error	Parity error	LPC error	Match #2	Match #1	Video line 8	Internal data ID	
9	Magaz	ine		Data byte						
10	Row ad	ddress							Data byte	
11	Teletex	t data	1						Data byte	
12	Teletex	t data	2						Data byte	
									Data byte	
50	Teletex	t data	40						Data byte	
51	NEP	EP	Checksum						Checksum	
52	1	0	0	0	0	0	0	0	Fill byte	

BYTE NO.	MSB 7	6	5	4	3	2	1	LSB 0	DESCRIPTION	
1	0	0	0	0	0	0	0	0		
2	1	1	1	1	1	1	1	1	Ancillary data preamble	
3	1	1	1	1	1	1	1	1		
4	NEP	EP	0	1	0	DID2	DID1	DID0	Data ID	
5	1	0	0	0	0	0	0	0	Secondary data ID	
6	1	0	0	0	0	0	0	0	Number of 32-bit data words	

Table 2–13. Dummy Timing Ancillary Data Sequence

In the tables above, EP is even parity on the lower 6 bits and NEP is negated even parity. The checksum for teletext data blocks is the 8 LSB's of the sum of the data bytes. The data ID byte provides timing information. Table 2–14 shows the possible values of the data ID byte and their meanings.

DATA ID	EVENT IN SOURCE STREAM	DATA TYPE				
50	Start of first, odd field	Dummy timing block				
91	Sliced data of lines 1-23 of first field	VBI data				
92	End of nominal VBI of first field, line 23	Dummy timing block				
53	Sliced data of line 24 to end of first field	Full field teletext data				
94	Start of second, even field	Dummy timing block				
55	Sliced data of lines 1-23 of second field	VBI data				
56	End of nominal VBI of second field, line 23	Dummy timing block				
97	Sliced data of line 24 to end of second field	Full field teletext data				

Table 2–14. Ancillary Data ID

A dummy timing block will be inserted into the video stream during the horizontal blanking period following line 23 of each field. If teletext data is available from line 23 it will be inserted into the video stream prior to the dummy timing block.

2.14 Reset

Reset is initiated at power up or any time the RSTINB terminal is brought low. Table 2–15 describes the status of the terminals on the TVP5022 during and immediately after reset. Following a powerup reset, the host downloads microcode to the program memory of the TVP5022 for use by the internal microprocessor.

Table 2–15.	Power-Up	Reset Sequ	ence
-------------	----------	-------------------	------

· _ · _ ·		
SIGNAL NAMES	DURING RESET	AFTER RESET
RSTINB, OEB, EXT_DATA_8	In	In
GLCO, PALI, FID, GPCL	In	Out
UV[7:0]	In	High-impedance
SCLK, PCLK, PREF	High-impedance	Out
Y[7:0], HSYN, VSYN, AVID	High-impedance	High-impedance
Host interface terminals: VC0, VC1, VC2, VC3, INTREQ, A1:0, D7:0	High-impedance	Active

Signals take their reset values during power up and anytime the RSTINB terminal is low.

The active state of the host interface terminals may be input, output, or input/output depending on the host mode selected during reset.

2.15 Internal Control Registers

Table 2–16 shows the summary of the TVP5022 registers. The detailed programming information of each register is described in the following sections. For writable registers, the reserved bits must be written with a 0.

REGISTER FUNCTION	VIP	VMI	l ² C	R/W
Vendor ID	000 – 001	NA	NA	R
Device ID	002 - 003	NA	NA	R
Subsystem vendor ID	004 – 005	NA	NA	R
Subsystem device ID	006 – 007	NA	NA	R
Reserved	008 – 00B	NA	NA	
Revision ID	00C - 00D	NA	NA	R
Reserved	00E – 0FF	NA	NA	
Analog input source selection	100	00 → 00	00	W
Analog channel controls	101	00 ← 01	01	W
Operation mode controls	102	00 ← 02	02	W
Miscellaneous controls	103	00 ← 03	03	W
Reserved	104 – 105	$00 \leftarrow 04 - 05$	04 – 5	
Color killer threshold control	106	00 ← 06	06	W
Luminance processing controls-#1	107	00 ← 07	07	W
Luminance processing controls-#2	108	80 → 00	08	W
Brightness control	109	00 ← 09	09	W
Color saturation control	10A	A0 ightarrow 00	0A	W
Hue control	10B	00 ← 0B	0B	W
Contrast control	10C	m O0 ightarrow m O0	0C	W
Outputs and data rate select	10D	$00 \leftarrow 0D$	0D	W
Reserved	10E – 115	$00 \leftarrow 0E - 15$	0E – 15	
Horizontal sync start NTSC	116	00 ← 16	16	W
Horizontal sync start PAL	117	00 ← 17	17	W
Vertical blanking start	118	00 ← 18	18	W
Vertical blanking stop	119	00 ← 19	19	W
Chrominance control	11A	00 ← 1A	1A	W
Reserved	11B – 11F	$00 \leftarrow 1B - 1F$	1B – 1F	
Digital channel selection	120	00 ← 20	20	W
Reserved	121 – 17D	$00 \leftarrow 21 - 7D$	21 – 7D	
Program RAM write	17E	$00 \leftarrow 7E$	7E	W
Microprocessor reset clear	17F	$00 \leftarrow 7F$	7F	W
Firmware version	180	00 → 00	80	R
Status #1	181	00 ← 81	81	R
Status #2	182	00 ← 82	82	R
AGC gain	183	00 ← 83	83	R

Table 2–16	Registers	Summary
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REGISTER FUNCTION	VIP	VMI	l ² C	R/W
Reserved	184 – 18D	$00 \leftarrow 84 - 8D$	84 – 8D	
Program RAM read	18E	00 ← 8E	8E	R
Reserved	18F	00 ← 8F	8F	
TXF filter #1 parameters (R/W)	190 – 194	$00 \leftarrow 90 - 94$	90 - 94	R/W
TXF filter #2 parameters (R/W)	195 – 199	$00 \leftarrow 95 - 99$	95 – 99	R/W
TXF error filtering enable	19A	00 ← 9A	9A	R/W
TXF transaction processing enables	19B	00 ← 9B	9B	R/W
Reserved	19C – 19F	$00 \gets 9C - 9F$	9C – 9F	
TTX control register	1A0	$00 \leftrightarrow A0$	A0	R/W
Line enable register	1A1 – 1A2	$00 \leftarrow A1 - A2$	A1 – A2	R/W
Sync pattern register	1A3	$00 \leftarrow A3$	A3	R/W
Reserved	1A4 – AF	$00 \gets A4 - AF$	A4 – AF	
Teletext FIFO	1B0	00 ← B0	B0	R
Closed caption data	1B1 – 1B2	$00 \leftarrow B1 - B2$	B1 – B2	R
Buffer status	1B3	00 ← B3	B3	R
Interrupt threshold	1B4	00 ← B4	B4	R/W
Interrupt line number	1B5	00 ← B5	B5	R/W
FIFO control	1B6	00 ← B6	B6	R/W
Reserved	1B7 – 1BF	$00 \leftarrow B7 - BF$	B7 – BF	R/W
Interrupt status register	1C0	00 ← C0	C0	R/W
Interrupt enable register	1C1	00 ← C1	C1	R/W
Interrupt configuration register	1C2	00 ← C2	C2	R/W
Reserved	1C3 – 1FF	$00 \gets \text{C3} - \text{FF}$	C3 – FF	
No-Latency read access 1	2xx			R
No-Latency read access 2	Зхх			R
FIFO		10		R
Interrupt status register		11		R/W

Table 2–16. Registers Summary (Continued)

2.15.1 Register Definitions

2.15.2 Vendor ID

		_						
VIP address	VIP address 000–001							
VMI address		Ν	IA	1				
I ² C address		Ν	IA					
Address	7	6	5	4	3	2	1	0
000	0	1	0	0	1	1	0	0
001	0	0	0	1	0	0	0	0

This field identifies the manufacturer of the device. Address 001 is the MSB. This field is a constant of 104C.

2.15.3 Device ID

-	002–003
VMI address	NA
I ² C address	NA

Address	7	6	5	4	3	2	1	0
002	0	0	1	0	0	0	0	0
003	0	1	0	1	0	0	0	1

This field identifies the particular device. Address 003 is the MSB. This field is a constant of 5120

2.15.4 Subsystem Vendor ID

VIP address		004	-005]			
VMI address		N	IA				
I ² C address		N	IA]			
Address	7	6	5	4	3	2	1
004		•	d from	UV pir	÷		reset
005	1	1	1	1	1	1	1

This field identifies the subsystem manufacturer (e.g., the board manufacturer). Address 005 is the MSB. This field is a constant of FF. The value of the LSB is set at the device power up or reset by sampling the state of UV[7:0] terminals. The UV[7:0] terminals may be tied to pullup or pulldown resistors to determine a fixed value for the subsystem vendor ID.

0

1

2.15.5 Subsystem Device ID

VIP address	006–007
VMI address	NA
I ² C address	NA

Address	7	6	5	4	3	2	1	0
006	1	1	1	1	1	1	1	1
007	1	1	1	1	1	1	1	1

This field identifies the subsystem device. Address 007 is the MSB. This field is a constant of FFFF.

2.15.6 Subsystem Revision ID

VIP address	00C-00D
VMI address	NA
I ² C address	NA

Address	7	6	5	4	3	2	1	0
00C	0	0	0	0	0	0	0	0
00D	0	0	0	0	0	0	0	1

This identifies the device hardware revision. Address 00D is the MSB. This field is a constant of 0100.

2.15.7 Analog Input Source Select	ion
-----------------------------------	-----

7	6	5	4	3	2
I ² C a	ddress		C	00	
VMI a	ddress		C	00	
VIP a	ddress		1	00	

7	6	5	4	3	2	1	0
		Rese	erved			Channel 1 source selection	Channel 2 source selection

Channel 1 source selection:

- 0 = VI1A selected (default)
- 1 = VI1B selected

Channel 2 source selection:

- 0 = VI2A selected (default)
- 1 = VI2B selected

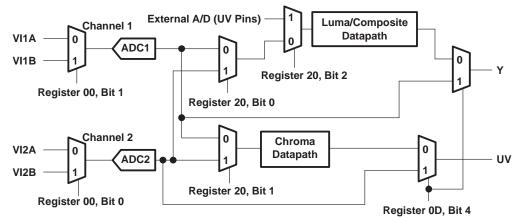


Figure 2–31. Video Input Source Selection

Table 2–17. Analog Ch	hannel and Video Mode Selection
-----------------------	---------------------------------

		ADDRE	ESS 00	ADDRE	ESS 20
	INPUT(s) SELECTED	BIT 1	BIT 0	BIT 1	BIT 0
Composite	1A	0	х	0	0
	1B	1	х	0	0
	2A	х	0	1	1
	2B	х	1	1	1
S-video	1A luma, 2A chroma	0	0	1	0
	1A luma, 2B chroma	0	1	1	0
	1B luma, 2A chroma	1	0	1	0
	1B luma, 2B chroma	1	1	1	0
	2A luma, 1A chroma	0	0	0	1
	2A luma, 1B chroma	1	0	0	1
	2B luma, 1A chroma	0	1	0	1
	2B luma, 1B chroma	1	1	0	1

01
01

7	6	5	4	3 2		1 0	
Reserved		Automatic clar Chan		Automatic clar Chan		Automatic g	gain control

Automatic clamping control, channel 2:

- 00 = Reserved
- 01 = Automatic clamping enabled (default)
- 10 = Reserved
- 11 = Clamping level frozen

Automatic clamping control, channel 1:

- 00 = Reserved
- 01 = Automatic clamping enabled (default)
- 10 = Reserved

11 = Clamping level frozen

- Automatic gain control:
- 00 = Reserved
- 01 = AGC enabled using luma input as the reference (default)
- 10 = Reserved
- 11 = AGC frozen

VIP address	102
VMI address	02
I ² C address	02

[7	6	5	4	3	2	1	0
	External A/D width		TV/VCF	R mode	Reserved	Disable color PLL	Reserved	Power down mode

External A/D width:

- 00 = 8-bit external A/D
- 01 = 9-bit external A/D terminal OEB is the LSB of the 9-bit input data
- 10 = 10-bit external A/D terminal GPCL is the LSB of the 10-bit input data, and terminal OEB is the next-to-LSB (default)
- 11 = Reserved

An external A/D converter can be used to provide up to 10-bit data directly to the digital data path. These bits set the width of the external data path. External A/D access is enabled by register 20, bit 2. See Figure 2–31 and register 20.

TV/VCR mode:

- 00 = Automatic mode determined by the internal detection circuit (default)
- 01 = Reserved
- 10 = VCR (nonstandard video) mode
- 11 = TV (standard video) mode

With automatic detection enabled, unstable or nonstandard syncs on input video will force the device into VCR (nonstandard) mode. This turns off the luminance and chrominance comb filters and turns on the chroma trap filter.

Disable color PLL:

- 0 = Normal operation (default)
- 1 = Force color PLL increment and genlock (GLCO) serial output to zero

Power down mode:

- 0 = Normal operation (default)
- 1 = Power down mode

2.15.10 Miscellaneous Controls

VIP address	103
VMI address	03
I ² C address	03

7	6	5	4	3	2	1	0
Terminal GPCL function select		Terminals PALI and FID function select	YUV out- put enable	HSYN, VSYN, AVID enable	Reserved	Vertical blanking on/off	Clock enable

Terminal GPCL function select:

- 00 = Terminal GPCL is logic 0 output (default)
- 01 = Terminal GPCL is logic 1 output
- 10 = Terminal GPCL is vertical blank output
- 11 = Terminal GPCL is external sync lock control input

When GPCL is configured as a vertical blank output, the vertical blanking on/off bit is used to activate the output.

When GPCL is configured as a sync lock control, it can be used to force the internal PLLs to their normal settings. This causes all clocks and synchronization signals to assume nominal values. The sync lock control input is active high.

Terminals PALI and FID function select:

- 0 = Terminal PALI outputs PAL indicator signal and terminal FID outputs field ID signal (default)
- 1 = Terminal PALI outputs horizontal lock indicator (HLK) and terminal FID outputs vertical lock indicator (VLK)

YUV output enable:

- 0 = YUV high impedance (default)
- 1 = YUV active

Y[7:0] and UV[7:0] terminals are controlled by this bit, the YUV enable bit in register C2, and the OEB terminal.

OEB Pin	Reg 03, Bit 4	Rec C2, Bit 2	YUV OUTPUT
0	х	0	High impedance
0	0	х	High impedance
0	1	1	Active
1	х	х	High impedance

Table 2–18. YUV Output Controls

Horizontal sync (HSYN), Vertical sync (VSYN), and Active video indicator (AVID) outputs enable:

- 0 = HSYN, VSYN, and AVID disabled, (high impedance state) (default)
- 1 = HSYN, VSYN, and AVID active

Vertical blanking on/off control:

- 0 = Vertical blanking off. Output is logic zero (default)
- 1 = Vertical blanking on

This bit is functional if GCPL is configured to output vertical blank.

Clock enable:

- 0 = SCLK and PCLK outputs are high impedance
- 1 = SCLK and PCLK outputs are enabled (default)

2.15.11 Color Killer Threshold Control

VIP address	106
VMI address	06
I ² C address	06

7	6	5	4	3	2	1	0		
Reserved	Automatic color killer		erved Automatic color killer Color killer threshold						

Automatic color killer:

- 00 = Automatic mode (default)
- 01 = Reserved
- 10 = Color killer enabled. The UV terminals are forced to a zero color state
- 11 = Color killer disabled

Color killer threshold (ref. 0 dB = nominal burst amplitude):

- 1 1 1 1 1 = 30 dB
- 1 0 0 0 0 = -24 dB (default)
- $0 \ 0 \ 0 \ 0 \ 0 = -18 \text{ dB}$

2.15.12 Luminance Processing Control 1

VIP address	107
VMI address	07
I ² C address	07
L	

7	6	5	4	3	2	1	0
Reserved	Pedestal	Reserved	Luma bypass during vertical blank			delay with ance signa	

Pedestal:

- 0 = 7.5 IRE pedestal is present on the analog video input (default)
- 1 = Pedestal is not present on the analog video input

Luminance bypass mode during vertical blanking:

- 0 = No (default)
- 1 = Yes

When the luminance bypass is enabled, the luminance comb and notch filters are turned off and the chrominance components of the output video are sent to a zero color state. Luminance bypass will occur for the duration of the vertical blanking as defined by registers 18 and 19. This feature may be used to prevent distortion of test and data signals present during the vertical blanking interval.

Luma signal delay with respect to chroma signal in pixel clock increments (range -8 to 7 pixel clocks):

- 1 1 1 1 = -8 pixel clocks delay
- 1 0 1 1 = -4 pixel clocks delay
- 1 0 0 0 = -1 pixel clocks delay
- $0 \quad 0 \quad 0 \quad 0 = 0$ pixel clocks delay (default)
- $0 \quad 0 \quad 1 \quad 1 = 3$ pixel clocks delay
- 0 1 1 1 = 7 pixel clocks delay

2.15.13 Luminance Processing Control 2

VIP address	108
VMI address	08
I ² C address	08

7	6	5	4	3	2	1	0
Luma filt	er select	Coring th	hreshold	Peakir	ng gain	Peaking f	requency

Luminance filter select:

- 00 = Automatic select (default)
- 01 = Reserved
- 10 = Chroma trap (notch filter)
- 11 = Comb filter

With automatic select enabled, unstable or nonstandard syncs on input video will force selection of the chroma trap (notch) filter. For PAL imputs, automatic select will always selct the chroma trap filter.

Coring threshold:

- 00 = Coring off (default)
- $01 = \pm 1 \text{ LSB}$
- $10 = \pm 2 \text{ LSB}$
- $11 = \pm 3 \text{ LSB}$

Peaking gain:

- 00 = Peaking disabled (default)
- $01 = 6 \, dB$
- $10 = 12 \, dB$
- 11 = 18 dB

Peaking frequency:

Square-pixel sampling rate:

	NTSC	PAL	PAL M	PAL N	
00 =	3.8 MHz	4.5 MHz	3.8 MHz	4.5 MHz	(default)
01 =	3.4 MHz	4.1 MHz	3.4 MHz	4.1 MHz	
10 =	2.5 MHz	3.0 MHz	2.5 MHz	3.0 MHz	
11 =	2.7 MHz	3.2 MHz	2.7 MHz	3.2 MHz	
	ITU-R BT.601	sampling rate:			

ALL STANDARDS

- 00 = 4.1 MHz (default)
- 01 = 3.7 MHz
- 10 = 2.8 MHz
- 11 = 3.0 MHz

2.15.14 Brightness Control

	<u> </u>								
VIP address	10	9							
VMI address	09)							
I ² C address 09									
7 6 5 4 3 2 1 0									
	Brightness control								

Brightness:

1 1 1 1 1 1 1 1 1 = 255 (bright)

1 0 0 0 0 0 0 0 = 128 (default)

 $0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ = 0 \ (dark)$

2.15.15 Color Saturation Control

VIP address		10A						
VMI address	0A							
I ² C address		0A						
7 6 5 4 3 2 1 0								
	Saturation control							

Saturation:

1 1 1 1 1 1 1 1 1 1 = 255 (maximum) 1 0 0 0 0 0 0 0 0 = 128 (default)

 $0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 = 0$ (no color)

2.15.16 Hue Control

This register (Sub-address = 0Bh) sets the hue of the color signal.

····e ·egietei (
VIP address	10B							
VMI address	0B							
I ² C address	0B							
7 6 5 4 3 2 1 0								
	Hue control							

Hue:

0 1 1 1 1 1 1 1 = 180 degrees

0 0 0 0 0 0 0 0 0 = 0 degrees (default)

1 0 0 0 0 0 0 0 0 = -180 degrees

2.15.17 Contrast Control

VIP address	10C	;				
VMI address	0C					
I ² C address	0C					
7	6	5	4	3	2	2 1
			Contras	t control		

Contrast:

1 1 1 1 1 1 1 1 1 = 255 (maximum contrast)

1 0 0 0 0 0 0 0 = 128 (default)

 $0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ = 0$ (minimum contrast)

2.15.18 Outputs and Data Rates Select

VIP address	10D
VMI address	0D
I ² C address	0D

7	6	5	4	3	2	1	0
Reserved	YUV output code range	UV code format	YUV data path bypass	Reserved	YUV o	output f	ormat

YUV output code range:

- 0 = ITU-R BT.601 coding range (Y ranges from 16 to 235, UV components range from 16 to 240) (default)
- 1 = Extended coding range (Y, and UV components range from 1 to 254)

UV code format:

- 0 = Offset binary code (2's complement + 128) (default)
- 1 = Straight binary code (2's complement)

YUV data path bypass:

- 0 = Normal operation (default)
- 1 = Y terminals connected to channel 1 A/D output, UV terminals connected to channel 2 A/D output

YUV output format:

- 000 = 16-bit 4:2:2 (default)
- 001 = Reserved
- 010 = 12-bit 4:1:1
- 011 = Reserved
- 100 = 8-bit 4:2:2
- 101 = Reserved
- 110 = Reserved
- 111 = 8-bit ITU-R BT. 656 with embedded syncs

2.15.19 Horizontal Sync (HSYN) Start

VIP address	116–117
VMI address	16–17
I ² C address	16–17

Address	7	6 5 4 3 2 1									
16		HSYN start for NTSC									
17		HSYN start for PAL									

HSYN start:

1	1	1	1	1	1	1	1	= -127×4 pixel clocks
1	1	1	1	1	1	1	0	= -126×4 pixel clocks
1	1	1	1	1	1	0	1	= -125×4 pixel clocks
1	0	0	0	0	0	0	0	= 0 pixel clocks (default)
0	1	1	1	1	1	1	1	= 1×4 pixel clocks
0	1	1	1	1	1	1	0	= 2×4 pixel clocks
0	0	0	0	0	0	0	0	= 128×4 pixel clocks

2.15.20 Vertical Blanking Start

VIP address	118
VMI address	18
I ² C address	18

7	6	5	4	3	2	1	0	
VBLK start								

VBLK start:

0	1	1	1	1	1	1	1	= 127 lines after start of vertical blanking interval
0	0	0	0	0	0	0	1	= 1 line after start of vertical blanking interval
0	0	0	0	0	0	0	0	= Same time as start of vertical blanking interval (default)
0	1	1	1	1	1	1	1	= 1 line before start of vertical blanking interval
1	0	0	0	0	0	0	0	= 128 lines before start of vertical blanking interval

Vertical blanking is adjustable with respect to the standard vertical blanking intervals as shown in Figure 2–17. The setting in this register determines the timing of the GPCL signal when it is configured to output vertical blank (see register 03). The setting in this register is also used to determine the duration of the luma bypass function (see register 07).

2.15.21	Vertical	Blanking	VBL	K Stop
---------	----------	----------	-----	--------

		0					
VIP address	119						
VMI address	19						
I ² C address	19						
7	6	5	4	3	2	1	0
			VBL	K end			

VBLK start:

0	1	1	1	1	1	1	1	= 127 lines after end of vertical blanking interval
0	0	0	0	0	0	0	1	= 1 line after end of vertical blanking interval
0	0	0	0	0	0	0	0	= Same time as end of vertical blanking interval (default)
0	1	1	1	1	1	1	1	= 1 line before end of vertical blanking interval
1	0	0	0	0	0	0	0	= 128 lines before end of vertical blanking interval

Vertical blanking is adjustable with respect to the standard vertical blanking intervals as shown in Figure 2–17. The setting in this register determines the timing of the GPCL signal when it is configured to output vertical blank (see register 03). The setting in this register is also used to determine the duration of the luma bypass function (see register 07).

2.15.22 Chrominance Control

VIP address	11A
VMI address	1A
I ² C address	1A

7	6	5	4	3	2	1	0
Reserved			Color reset	Chrominance co	omb filter control	Automatic colo	or gain control

Color reset:

0 = Color not reset (default)

1 = Color reset

When this bit is set the color PLL phase is reset to zero and the phase reset bit is sent in the next transmission of the genlock control signal GCLO. The color reset control bit is then automatically reset to zero. Refer to Figure 2–30 for genlock timing.

Chrominance comb filter control:

- 00 = Automatic select (default)
- 01 = Reserved
- 10 = Comb filter on
- 11 = Comb filter bypassed

With automatic select enabled, unstable or nonstandard syncs on input video will force bypass of the chrominance comb filter. For PAL inputs, automatic select will always bypass the chrominance comb filter.

Automatic color gain control:

- 00 = ACC enabled (default)
- 01 = Reserved
- 10 = ACC disabled
- 11 = ACC frozen

2.15.23 Digital Channel Selection

VIP address	120			
VMI address	20			
I ² C address	20			
	1 1			

7	6		5	4	3	2	1	0
	Reserved Exte					External A/D select	Chroma channel select	Luma/composite channel select

NOTE: Also see analog input source selection register 00

External A/D select

- 0 = Use internal A/D converters
- 1 = Use external A/D converter (default)

Chroma channel select

- 0 = ADC1 selected (default)
- 1 = ADC2 selected

Luma/Composite channel select:

- 0 = ADC1 selected (default)
- 1 = ADC2 selected

2.15.24 Firmware Version

VIP address	180		
VMI address	80		
I ² C address	80		

7	6	5	4	3	2	1	0		
	Firmware version								

2.	1	5.25	Status	Register	1
----	---	------	--------	----------	---

VIP address	181
VMI address	81
I ² C address	81

7	6	5	4	3	2	1	0
Peak white detect	Reserved	Field rate	Lost lock detect	Color lock	Vertical lock	Horizontal lock	TV/VCR

Peak white detect:

0 = Peak white is not detected

1 = Peak white is detected

Field rate:

0 = 60 Hz

1 = 50 Hz

Lost lock detect:

0 = No lost horizontal lock since status was last read

1 = Lost horizontal lock since status was last read

Color lock:

- 0 = Color subcarrier is not locked
- 1 = Color subcarrier is locked

Vertical lock:

- 0 = Vertical sync is not locked
- 1 = Vertical sync is locked

Horizontal lock:

- 0 = Horizontal sync is not locked
- 1 = Horizontal sync is locked

TV/VCR:

- 0 = TV
- 1 = VCR

This bit indicates whether or not nonstandard syncs are detected on the video input.

2.15.26 Status Register 2

7	6					
I ² C a	ddress	s 82				
VMI a	addres	s 82				
VIP a	ddres	5 182				

7	6	5	4	3	2	1	0
Rese	erved	PAL switch polarity	Field sequence	AGC and clamping lock	F	Reserved	

PAL switch polarity of first line of odd field:

- 0 = PAL switch is zero (color burst phase = 135 degree)
- 1 = PAL switch is one (color burst phase = 225 degree)

Field sequence:

- 0 = Even field
- 1 = Odd field

Automatic gain and clamping lock status:

- 0 = Automatic gain and clamping is not locked
- 1 = Automatic gain and clamping is locked

2.15.27 AGC Gain

VIP address	183
VMI address	83
I ² C address	83

7	6	5	4	3	2	1	0
AGC gain							

AGC gain (step size = 0.831%):

0 0 0 0 0 0 0 0 0 0 0 = 70.7%(-3 dB) 0 1 0 0 0 0 0 0 0 = 100% (0 dB)

1 0 0 1 0 0 0 0 = 141% (3 dB)

1 1 0 0 0 0 0 0 = 200% (6 dB)

1 1 1 1 1 1 1 1 1 = 28.3% (9 dB)

VIP address	190 – 194
VMI address	90 - 94
I ² C address	90 - 94

2.15.28 TXF Filter #1 Parameters

Address	7	6	5	4	3	2	1	0	
90	Filter #1 mask_1[3:0]				Filter #1 pattern_1[3:0]				
91	Filter #1 mask_2[3:0]				Filter #1 pattern_2[3:0]				
92	Filter #1 mask_3[3:0]				Filter #1 pattern_3[3:0]				
93	Filter #1 mask_4[3:0]				Filter #1 pattern_4[3:0]				
94	Filter #1 mask_5[3:0]					Filter #1 pa	attern_5[3:0]		

These registers hold the search parameters for filter #1. The parameters parse the first five bytes of NABTS teletext transactions or the first two bytes of WST transactions. These bytes of teletext always contain four data bits interlaced with four hamming protection bits. The filter ignores the protection bits.

For an NABTS system, the packet prefix consists of five bytes: P1, P2, P3, CI, and PS. Each byte contains 4 data bits interlaced with 4 hamming protection bits.

Pattern_1[3:0] corresponds to P1[7], P1[5], P1[3], P1[1] (Packet address) Pattern_2[3:0] corresponds to P2[7], P2[5], P2[3], P2[1] Pattern_3[3:0] corresponds to P3[7], P3[5], P3[3], P3[1] Pattern_4[3:0] corresponds to CI[7], CI[5], CI[3], CI[1] Pattern_5[3:0] corresponds to PS[7], PS[5], PS[3], PS[1]

(Packet address) (Packet address) (Continuity index) (Packet structure)

For a WST system (PAL or NTSC), the magazine and row address group consists of two bytes. The two bytes contain three bits of magazine number (M[2:0]) and 5 bits of row address (R[4:0]), interlaced with eight hamming protection bits.

Pattern_1[3:0] corresponds to R[0], M[2], M[1], M[0] Pattern_2[3:0] corresponds to R[4], R[3], R2], R[1] Pattern 3[3:0] is ignored Pattern_4[3:0] is ignored Pattern_5[3:0] is ignored

(Magazine and row LSBit) (Upper bits of row address)

The mask bits enable filtering using the corresponding bit in the pattern register. For example, A 1 in the LSB of Mask 1 means that the VBI data processor (VDP) should compare the LSB of Nybble 1 in the pattern register to the first data bit of the transaction. A 0 in the LSB of Mask 1 means that the VDP should ignore the first data bit of the transaction.

NOTE:

The TXF filter #1 parameters can only be written and read when both the filter #1 enable and the filter #2 enable bits are 0. When reading the values, the values must be read consecutively, starting with the first value.

VIP address	195 – 199
VMI address	95 – 99
I ² C address	99 – 99

2.15.29 TXF Filter #2 Parameters

Address	7	6	5	4	3	2	1	0			
95	Filter #2 mask_1[3:0] Filter #2 pattern_1[3:0]										
96	Filter #2 mask_2[3:0] Filter #2 pattern_2[3:0]										
97	Filter #2 mask_3[3:0] Filter #2 pattern_3[3:0]										
98	Filter #2 mask_4[3:0] Filter #2 pattern_4[3:0]										
99		Filter #2 m	ask_5[3:0]		Filter #2 mask_5[3:0] Filter #2 pattern_5[3:0]						

These registers hold the search parameters for Filter #2. The parameters parse the first five bytes of NABTS teletext transactions or the first two bytes of WST transactions. These bytes of teletext are expected to always contain four data bits interlaced with four hamming protection bits. The filter ignores the protection bits.

For an NABTS system, the packet prefix consists of five bytes: P1, P2, P3, CI, and PS. Each byte contains 4 data bits interlaced with 4 hamming protection bits.

 Pattern_1[3:0] corresponds to P1[7], P1[5], P1[3], P1[1]
 (p

 Pattern_2[3:0] corresponds to P2[7], P2[5], P2[3], P2[1]
 (p

 Pattern_3[3:0] corresponds to P3[7], P3[5], P3[3], P3[1]
 (p

 Pattern_4[3:0] corresponds to CI[7], CI[5], CI[3], CI[1]
 (d

 Pattern_5[3:0] corresponds to PS[7], PS[5], PS[3], PS[1]
 (d

(packet address) (packet address) (packet address) (continuity index) (packet structure)

For a WST system (PAL or NTSC), the magazine and row address group consists of two bytes. The two bytes contain three bits of magazine number (M[2:0]) and 5 bits of row address (R[4:0]), interlaced with eight hamming protection bits.

Pattern_1[3:0] corresponds to R[0], M[2], M[1], M[0] Pattern_2[3:0] corresponds to R[4], R[3], R2], R[1] Pattern_3[3:0] is ignored Pattern_4[3:0] is ignored Pattern_5[3:0] is ignored (magazine and row LSB) (upper bits of row address)

The mask bits enable filtering using the corresponding bit in the pattern register. For example, A 1 in the LSB of Mask_1 means that the VBI data processor (VDP) should compare the LSB of Nybble_1 in the pattern register to the first data bit of the transaction. A 0 in the LSB of Mask_1 means that the VDP should ignore the first data bit of the transaction.

NOTE:

The TXF filter #2 parameters can only be written and read when both the filter #1 enable and the filter #2 enable bits are 0. When reading the values, the values must be read consecutively, starting with the TXF filter #1 parameters values.

19A
9A
9A

2.15.30 TXF Error Filtering Enables

7	6	5	4	3	2	1	0
Reserved	Reserved	Reserved	Reserved	LPC error enable	CCD parity error enable	Teletext parity error enable	Hamming error enable
	Hamming Error Enable		ble C) = disable	1 = enable	default = 0	
	Teletext Parity Error Enable		Enable 0) = disable	1 = enable	default = 0	

1 = enable

1 = enable

default = 0

default = 0

These bits allow the VDP module to discard transactions based on bit errors. The hamming error enable allows error correction and detection of hamming encoded bytes. The teletext parity error enable allows the VDP to discard teletext transactions with parity errors. The closed caption data (CCD) parity error enable allows the VDP to discard closed caption transactions with parity errors. The LPC error enable allows the VDP to discard teletext transactions with longitudinal parity errors.

0 = disable

0 = disable

2.15.31 TXF Transaction Processing Enables

CCD Parity Error Enable

LPC Error Enable

VIP address	19B		
VMI address	9B		
I ² C address	9B		

7	6	5	4	3	2	1	0
Reserved	Reserved	Reserved	Filter #2 enable	Filter #1 enable	CCD odd field enable	CCD even field enable	Teletext enable
	Taladaud		0	-lin - la la	4		
	Teletext enable		0 =	disable	1 = enable	default = 0	
	CCD even field enable 0 =		disable	1 = enable	default = 0		
	CCD odd field enable 0 =		e 0=	disable	1 = enable	default = 0	
	Filter #1	enable	0 =	= disable 1 = enable		default = 0	
	Filter #2	Filter #2 enable 0 =		disable	1 = enable	default = 0	

These bits enable or disable certain features. The teletext enable allows the VBI data processor (VDP) to receive teletext data. If this bit is 0, all outputs from the VDP remain idle while teletext data is present. The CCD even field enable and CCD odd field enable allow the VDP to receive closed caption data. The filter #1 enable allows the VDP to parse data based on the values in the filter #1 parameters register. The filter #2 enable allows the VDP to parse data based on the values in the filter #2 parameters register.

2.15.32 TTX Control Regi	ister
--------------------------	-------

VIP address	1A0		
VMI address	A0		
I ² C address	A0		

7	6	5	4	3	2	1	0
Reserved	Reserved	Reserved	Reserved	Full- field enable	Custom framing code	CCD en- able	TTX mode

Teletext mode	0 = NABTS	1 = WST
CCD enable	0 = Closed caption is disabled	1 = Closed caption is enabled
Custom sync	0 = Use default teletext sync pattern	1 = Use sync pattern register
Full field enable	0 = No teletext search after VBI area	1 = Teletext search all lines after VBI

The TTX control register allows operating parameters of the VBI data processor to be controlled. The TTX mode selection is independent of PAL/NTSC mode, which is selected by the TVP5022. This effectively controls the default framing code and data rate. Closed caption data is affected by 525 lines versus 625 lines (but not NABTS/WST). For NTSC and PAL M, the CCD data search is on Line 21; for PAL B,G,H,I,N it is on Line 22. Custom framing code affects teletext data ONLY – closed caption data always uses the default sync pattern.

2.15.33 Line Enable Registers

VIP address	1A1
VMI address	A1
I ² C address	A1

7	6	5	4	3	2	1	0
Enable	Enable	Enable	Enable	Enable	Enable	Enable	Enable
line	line	line	line	line	line	line	line
17/280	16/279	15/278	14/277	13/276	12/275	11/274	10/273
(14/327)	(13/326)	(12/325)	(11/324)	(10/323)	(9/322)	(8/321)	(7/320)

VIP address	1A2
VMI address	A2
I ² C address	A2

7	6	5	4	3	2	1	0
Enable							
line							
25/288	24/287	23/286	22/285	21/284	20/283	19/282	18/281
(22/335)	(21/334)	(20/333)	(19/332)	(18/331)	(17/330)	(16/329)	(15/328)

Line enable XX 0 = No teletext search on line XX 1 = Search line XX for teletext data NOTE: Line numbers in parentheses refer to 625 Line systems

In both VBI only and full field modes, the vertical interval lines can be individually enabled or disabled. Only lines that are enabled are searched for the selected type of teletext data. This allows some amount of filtering on a physical location basis. If closed caption data is enabled, this overrides the enable/disable bit for line 21 (22). If full field mode is enabled, all lines after the vertical interval are searched for the selected type of teletext data. The registers are initalized to 0x00 on reset.

2.15.34 Sync Pattern Register

	,							
VIP address	S	1A3						
VMI addres	s	A3						
I ² C address	S	A3						
7 6 5 4 3 2 1 0								
	Framing code[7:0]							

NOTE:

The custom sync option is only valid for NABTS or WST messages; closed caption always uses the EIA standard start bit pattern.

If the custom sync bit is set in the control register, the sync comparator uses the contents of the sync pattern register as the bit pattern for the teletext framing code. Otherwise, the default sync patterns are used. Relative to the sync pattern register, incoming bits are shifted in MSB first. To illustrate; the default WST framing code would be specified as 0xE4 and the default NABTS framing code would be specified as 0xE7 (although the MSB vs LSB is ambiguous for the latter).

2.15.35 Teletext FIFO

2.10.00	CICLOX	 0					
VIP address	S	1B0					
VMI addres	S	B0					
I ² C address	S	B0					
7	6	5	4	3	2	1	0
			Teletext data	FIFO [7:0]			

Reading this location returns 1 byte from the FIFO that stores teletext transactions. If the FIFO is empty, a read will return the same value as the previous read. The micro must know the number of bytes per transaction. The transaction length depends on whether the data is NABTS, WST-NTSC, or WST-PAL.

2.15.36 Closed Caption Data

There are two registers reserved for closed-caption data. Since closed-caption data always contains two bytes per transaction, each register must be read twice. The first read returns the first byte of the message, the second read returns the second byte. Further reads return the first byte until new data is received.

Even field closed caption data [7:0]

B2

2.15.37	Buffer	Status
---------	--------	---------------

VIP address	6	1	B3						
VMI addres	s	E	B3						
I ² C address	3	I	B3						
7	6		5	4	3	2	1	0	
CCD field1 avail	CCD field2 avail	2		Tx cou	nt [3:0]		FIFO full	Teletext data avail	
Teletext ava	ail	_		it indicates t it is cleared				t transactio	n is in the FIFO.
Tx count		_	- This v	alue represe	ents the nur	nber of com	nplete telete	ext transacti	ions in the FIFO.
CCD field 2	avail	-	- This status bit indicates that closed-caption data has been received in the ev field. The status bit is cleared when both of the two bytes have been read.						
CCD field 1	avail	_		tatus bit ind Γhe status b					eived in the odd d.

2.15.38 Interrupt Threshold

VIP address	S	1B4					
VMI addres	s	B4					
I ² C address	S	B4					
7	6	5	4	3	2	1	0
					Threshold	value [3:0]	

Threshold value

 This value determines how many teletext transactions must be received before the teletext threshold status bit is set in the interrupt status register. The default value is 5.

2.15.39 Interrupt Line Number

VIP addres	s	1B5					
VMI addres	s	B5					
I ² C addres	S	B5					
7	6	5	4	3	2	1	0
		Data required	ired Interrupt line number [4:0]			er [4:0]	

Interrupt line number – This value determines which video line number will be used to generate the teletext data, closed caption even field, and closed caption odd field bits in the interrupt status register. The register value is examined at the start of the line. Since there is no line 0, a value of all zeros in this register disables the three interrupt signals that use this condition. The default value is 24 (decimal).

Data required
 If this bit is set high, teletext data bit will only be set if there is data in the FIFO. This bit does not affect the closed caption even field, and closed caption odd field bits. The default value for this bit is 1.

2.15.40 FIFO Control

VIP address	6	1E	36						
VMI address	s	В	6						
I ² C address	3	В	6						
7	6		5	4	3	2	1	0	
					Read in progress	Reserved	TTX VMI output enable	FIFO reset	
				ng the read	and write p s flags. The	pointers to z	zero, clearing th gs for the close	e Tx count	to zero, and
TTX VMI ou	utput er	nable -	VMI p	oort and dis	ables acces	ss from the	e teletext data output formatte the output form	er. A 0 disa	bles access
CCD Reset	:	-				0	the closed capt is automaticall	0	
Read in pro	gress	-	the la		not been rea		teletext transac can be used to v		

VIP address	S		1C0							
VMI addres	s		C0							
I ² C address C0		C0								
7	6		5	4	3	2	1	0		
tvpLock state	tvpLc interr		Cycle complete	Bus error	CC odd field	CC Even field	Teletext threshold	Teletext data		
Teletext data0 = Teletext data buffer empty or we have not reached the video line number that equals the interrupt line number register.1 = Teletext data bu complete trans video line num line number register.		lete transacti line number	on and the = interrupt	the video	bit can be occur whenever line number = number register t the data.					
Teletext threshold		0 = T	hreshold no	t reached	d 1 = Teletext data in buffer has reached configurable threshold.					
CC even fie	eld	0 = B	uffer empty			ield closed c contains dat				
CC odd field	d	0 = B	Suffer empty			eld closed ca contains dat				
Bus error		0 = No bus error		0 = No bus erro			1 = VMI interface detected an illegal access.			
Cycle comp	olete	0 = Read or write cycle in progress.				1 = Read or write cycle complete				
tvpLock interrupt		(transition h occurred on horizontal lo	the		1 = A transition has occurred on the horizontal lock signal.		generated or	nterrupt will be any transition of al lock signal.	
tvpLock sta	te		TVP5022 n video source	ot locked to	1 = TVP50 sourc)22 locked to e	video	Reflects the the tvpLock.	present state of	

2.15.41 Interrupt Status Register

The interrupt status register is polled by the external processor to determine the interrupt source. After an interrupt condition is set, it can be reset by writing to this register with a 1 in the appropriate bit.

VIP address	1C1
VMI address	C1
I ² C address	C1

2.15.42 Interrupt Enable Register

7	6	5	4	3	2	1	0
Reserved	tvpLock Interrupt Enable	Cycle Complete Enable	Bus Error Enable	CC Odd Field En- able	CC Even Field En- able	Teletext Threshold Enable	Teletext Data Enable

The interrupt enable register is used by the external processor to mask unnecessary interrupt sources. Bits loaded with a 1 allow the corresponding interrupt condition to generate an interrupt on the external terminal. Conversely, bits loaded with a 0 mask the corresponding interrupt condition from generating an interrupt on the external terminal. This register only affects the external terminal, it does not affect the bits in the interrupt status register. A given condition can set the appropriate bit in the status register and not cause an interrupt on the external terminal. To determine if this device is driving the interrupt terminal, check either the interrupt status register with the interrupt enable register or check the state of the interrupt bit in the interrupt configuration register.

0

Interrupt

polarity

1

Interrupt

2.15.43 Interrupt Configuration Register

VIP addres	S		1C2			
VMI addres	S		C2			
I ² C addres	S		C2			
7	6		5	4	3	2
			Reserved			YUV output enable
Interrupt Polarity 0 – Interrup						

Interrupt Polarity	0 = Interrupt is active low.	1 = Interrupt is active high.	
Interrupt	0 = Interrupt terminal is not active.	1 = Interrupt terminal is active.	Reflects state of interrupt terminal.
YUV output enable	0 = YUV terminals are high impedance.	1 = YUV terminals are active.	

The interrupt configuration register configures the polarity of the external interrupt terminal.

2.15.44 Interrupt Status Register

VIP address	NA			
VMI address	11 (direct)			
I2C address	NA			

The status register is a duplicate of the interrupt status register at subaddress C0h.

3 Electrical Specifications

3.1 Absolute Maximum Ratings[†]

Supply voltage , AV _{DD}
Supply voltage, DV _{DD} 3.6 V
Input voltage range, A _{VI}
Input voltage range, D _{VI}
Storage temperature range
Operating free-air temperature
Total power dissipation (Watts) 2.5 W

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

	MIN	NOM	MAX	UNIT
Supply voltage, digital, DV _{DD}	3	3.3	3.6	V
Supply voltage, analog, AV _{DD}	4.75	5	5.25	V
Input voltage, analog (ac coupling necessary), VI (p-p)	0.5	1	1.26	V
Input voltage high, digital, VIH	2			V
Input voltage low, digital, VIL			0.8	V
Input voltage high, VCO and VC1 in I ² C mode, V _{IH} I ² C	2.3			V
Input voltage low, VCO and VC1 in I ² C mode, V _{IL} I ² C			1.0	V
Output current, Vout = 2.4V, I _{OH}	-4	-8		mA
Output current, Vout = 0.4V, I _{OL}	6	8		mA
Operating free-air temperature, T _A	0		70	°C
Crystal Specifications				
Frequency (ITU.601 sampling – 13.5 MHz)		24.576		MHz
Frequency (square pixel sampling)		26.800		MHz
Frequency tolerance			±40	ppm

3.2 Recommended Operating Conditions

3.3 Electrical Characteristics

3.3.1 Analog Processing and Analog-to-Digital Converters

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Zi	Input impedance, analog video inputs	By design	200			kΩ
Ci	Input capacitance, analog video inputs	By design			10	pF
ΔG	Gain control range		-2		6	dB
DNL	DC differential nonlinearity	A/D only			1	LSB
INL	DC integral nonlinearity	A/D only			1.2	LSB
Fr	Frequency response	6 MHz		-0.9	-3	dB
XTALK	Crosstalk	1 MHz			-50	dB
SNR	Signal-to-noise ratio	1 MHz, 1 Vpp		42		dB
NS	Noise spectrum	50% flat field		53		dB
DP	Differential phase			0.5		deg
DG	Differential gain			1.5%		

NOTE 1: Test Conditions: $DV_{DD} = 3.3V$, $AV_{DD} = 5.0V$, $T_A = 70^{\circ}C$ unless otherwise specified

3.3.2 DC Electrical Characteristics

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
IIN(DIG)	DIG) Digital supply current				225	300	mA
IIN(AN)	Analog supply current				85	105	mA
1 li	Input leakage current				10	μA	
Ci	Input capacitance digital inputs	Input capacitance digital inputs				8	pF
VOL	Output voltage	Low				0.4	V
VOH	Output voltage	High		2.4			V

NOTE 2: Measured with a load of 10 k Ω in parallel to 15 pF.

3.4 Timing

3.4.1 Clocks, Video Data, Sync Timing

	PARAMETER	TEST CONDITIONS (see NOTE 3)	MIN	TYP	МАХ	UNIT
δCLK	Duty cycle PCLK, SCLK		40%	50%	60%	
^t r(SCLK)	Rise time SCLK	10% to 90%		3		ns
^t f(SCLK)	Fall time SCLK	90% to 10%		2		ns
^t r(PCLK)	Rise time PCLK	10% to 90%		3		ns
^t f(PCLK)	Fall time PCLK	90% to 10%		2		ns
td(PCLK)	Delay time, SCLK rising edge to PCLK				5	ns
^t d(PREF)	Delay time, SCLK falling edge to PREF	See Note 3			3	ns
^t d(Y:UV)	Delay time, SCLK falling edge to Y, UV	See Note 3			4	ns
^t d(OUT)	Delay time, SCLK falling edge to digital outputs except PCLK, PREF, Y, UV				6	ns
^t su(UV)	Setup time, UV pins (in input mode) to SCLK falling edge, when PREF high		10			ns
^t h((UV)	Hold time, UV pins (in input mode) from SCLK falling edge, when PREF high				2	ns
f(I2C)	I ² C clock frequency				400	kHz

NOTES: 3. C_L = 50 pF

4. SCLK falling edge may occur up to 2 ns after PREF, Y, UV output transitions.

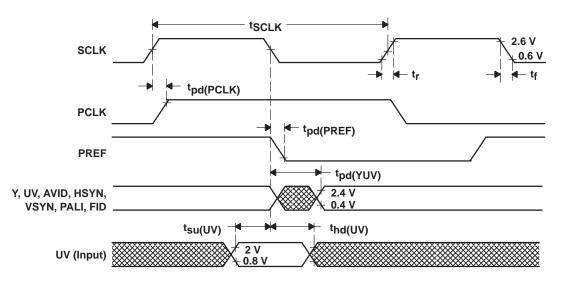


Figure 3–1. Clock, Video Cat, Sync Timing

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
^t BUF	Bus free time between STOP and START		1.3			μS
^t SU:STA	Setup time for a (repeated) START condition		0.6			μS
^t HD:STA	Hold time (repeated) START condition		0.6			μS
^t SU:STO	Setup time for a STOP condition		0.6			μS
^t SU:DAT	Data set-up time		100			nS
^t HD:DAT	Data hold time		0		0.9	μS
^t R	Rise time VC1 (SDA) and VC0 (SCL) signal				250	nS
t _F	Fall time VC1 (SDA) and VC0 (SCL) signal				250	nS
Cb	Capacitive load for each bus line				400	pF
fl2C	I ² C clock frequency				400	kHz

3.4.2 I²C Host Bus Timing

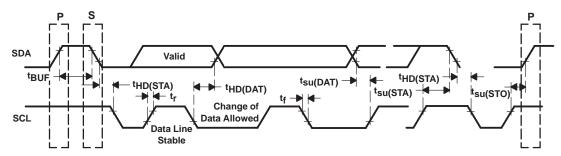


Figure 3–2. I²C Bus Timing

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{su}	VC0, VC1, VC2 setup to VC3 (VIPCLK)		5			ns
thd	(VIPCLK) to VC0, VC1, VC2 hold time		0			ns
^t pd	VC3 (VIPCLK) to VC0, VC1, VC2, INTREQ propagation delay time				11	ns

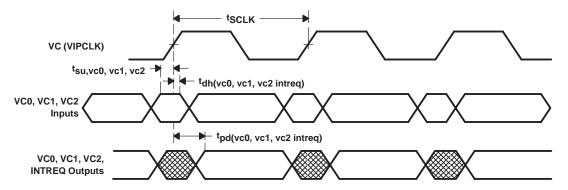


Figure 3–3. VIP Host Port Timing

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
t ₁	A[1:0], D[0:7], VC1 setup until VC2 LOW		5			ns		
t ₂	Delay VC0 LOW after VC2 LOW		0			ns		
t ₃	A[1:0], D[0:7], VC1 hold after VC0 LOW		5			ns		
t5	Delay VC0 HIGH after VC2 HIGH		0			ns		
t ₆	Delay VC2 LOW (next cycle) after VC0 HIGH		5			ns		
t ₈	(Read Cycle) D[7:0] setup until VC0 LOW		10			ns		
t9	(Read Cycle) D[7:0] hold after VC2 HIGH		0			ns		

3.4.4 VMI Host Port Timing (Mode A)

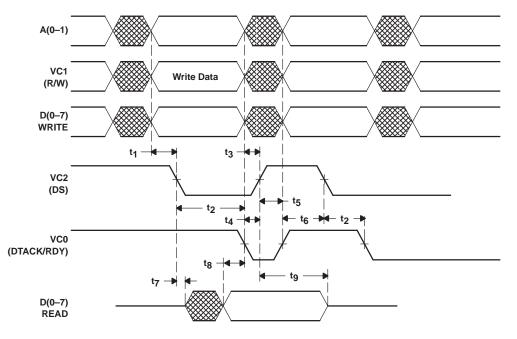


Figure 3–4. VMI Host Port Timing (Mode A)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t ₁	Delay VC1 or VC2 active after valid A[1:0]		10			ns
t ₂	A[1:0] hold after VC1 or VC2 inactive		10			ns
t3	Delay VC0 LOW after VC1 or VC2 active				28	ns
t4	D[7:0] setup until VC1 active		5			ns
t5	D[7:0] hold after VC1 inactive		10			ns
t ₆	VC0 inactive pulse width		10			ns
t ₈	(Read Cycle) VC1 LOW until D[7:0] non-3-state		5			ns
tg	(Read Cycle) D[7:0] setup until VC0 inactive		0			ns
t ₁₀	(Read Cycle) D[7:0] hold after VC0 inactive		0		15	ns
t ₁₁	Hold VC1 active after VC0 active		0			ns

3.4.5 VMI Host Port Timing (Mode B)

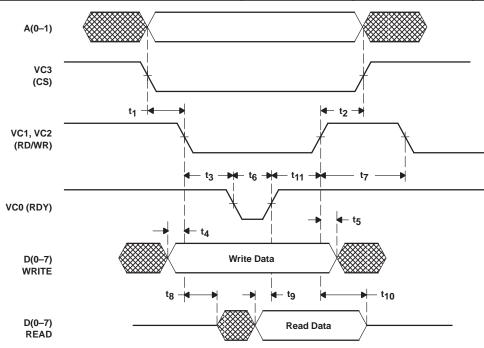


Figure 3–5. VMI Host Port Timing (Mode B)

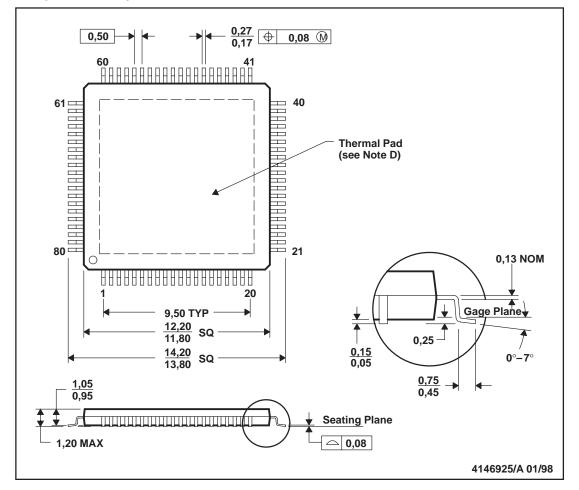
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4 Mechanical Data

PFP (S-PQFP-G80)

PowerPAD[™] PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions include mold flash or protrusions.
- D. The package thermal performance may be enhanced by bonding the thermal pad to an external thermal plane. This pad is electrically and thermally connected to the backside of the die and possibly selected leads.
- E. Falls within JEDEC MS-026

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