

January 1995

#### LM150, LM350A/LM350 3-Amp Adjustable Regulators

#### **General Description**

The LM150 series of adjustable 3-terminal positive voltage regulators is capable of supplying in excess of 3A over a 1.2V to 33V output range. They are exceptionally easy to use and require only 2 external resistors to set the output voltage. Further, both line and load regulation are comparable to discrete designs. Also, the LM150 is packaged in standard transistor packages which are easily mounted and

In addition to higher performance than fixed regulators, the LM150 series offers full overload protection available only in IC's. Included on the chip are current limit, thermal overload protection and safe area protection. All overload protection circuitry remains fully functional even if the adjustment terminal is accidentally disconnected.

Normally, no capacitors are needed unless the device is situated more than 6 inches from the input filter capacitors in which case an input bypass is needed. An output capacitor can be added to improve transient response, while bypassing the adjustment pin will increase the regulator's ripple rejection.

Besides replacing fixed regulators or discrete designs, the LM150 is useful in a wide variety of other applications. Since the regulator is "floating" and sees only the input-to-output differential voltage, supplies of several hundred volts can be regulated as long as the maximum input to output differential is not exceeded, i.e., avoid short-circuiting the output.

By connecting a fixed resistor between the adjustment pin and output, the LM150 can be used as a precision current regulator. Supplies with electronic shutdown can be achieved by clamping the adjustment terminal to ground which programs the output to 1.2V where most loads draw little current.

The part numbers in the LM150 series which have a K suffix are packaged in a standard Steel TO-3 package, while those with a T suffix are packaged in a TO-220 plastic package. The LM150 is rated for  $-55^{\circ}\text{C} \leq T_{J} \leq \, +150^{\circ}\text{C},$  while the LM350A is rated for  $-40^{\circ}$ C  $\leq T_{.1} \leq +125^{\circ}$ C, and the LM350 is rated for  $0^{\circ}$ C  $\leq T_{J} \leq +125^{\circ}$ C.

#### **Features**

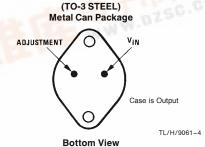
- Adjustable output down to 1.2V
- Guaranteed 3A output current
- Guaranteed thermal regulation
- Output is short circuit protected
- Current limit constant with temperature
- P<sup>+</sup> Product Enhancement tested
- 86 dB ripple rejection
- Guaranteed 1% output voltage tolerance (LM350A)
- Guaranteed max. 0.01%/V line regulation (LM350A)
- Guaranteed max. 0.3% load regulation (LM350A)

#### **Applications**

■ Adjustable power supplies

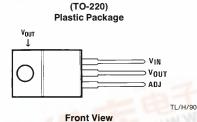
- Constant current regulators
- Battery chargers

#### **Connection Diagrams**



Order Number LM150K STEEL or LM350K STEEL See NS Package Number K02A

Order Number LM150K/883 See NS Package Number K02C



Order Number LM350AT or LM350T See NS Package Number T03B

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#### **Absolute Maximum Ratings**

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Power Dissipation

Input-Output Voltage Differential Storage Temperature

+35V

Internally Limited

-65°C to +150°C

LM150

Operating Temperature Range LM350A

Metal Package (Soldering, 10 sec.)

Plastic Package (Soldering, 4 sec.)

Lead Temperature

ESD Tolerance

 $\begin{array}{l} -55^{\circ}C \leq T_{J} \leq +150^{\circ}C \\ -40^{\circ}C \leq T_{J} \leq +125^{\circ}C \end{array}$ 

300°C 260°C

TBD

LM350  $0^{\circ}C \leq T_{J} \leq \, +\, 125^{\circ}C$ 

#### **Electrical Characteristics**

Specifications with standard type face are for  $T_J=25^{\circ}C$ , and those with **boldface type** apply over **full Operating Temperature Range.** Unless otherwise specified,  $V_{IN}-V_{OUT}=5V$ , and  $I_{OUT}=10$  mA. (Note 2)

Parameter	Conditions		Units		
		Min	Тур	Max	Ullits
Reference Voltage	$ 3V \leq (V_{IN} - V_{OUT}) \leq 35V, $ $ 10 \text{ mA} \leq I_{OUT} \leq 3A, P \leq 30W $	1.20	1.25	1.30	V
Line Regulation	$3V \le (V_{IN} - V_{OUT}) \le 35V \text{ (Note 3)}$		0.005	0.01	%/V
			0.02	0.05	%/V
Load Regulation	10 mA $\leq$ I <sub>OUT</sub> $\leq$ 3A (Note 3)		0.1	0.3	%
			0.3	1	%
Thermal Regulation	20 ms Pulse		0.002	0.01	%/W
Adjustment Pin Current			50	100	μΑ
Adjustment Pin Current Change	10 mA $\leq$ I <sub>OUT</sub> $\leq$ 3A, 3V $\leq$ (V <sub>IN</sub> $-$ V <sub>OUT</sub> ) $\leq$ 35V		0.2	5	μΑ
Temperature Stability	$T_{MIN} \le T_{J} \le T_{MAX}$		1		%
Minimum Load Current	$V_{IN} - V_{OUT} = 35V$		3.5	5	mA
Current Limit	$ \begin{vmatrix} V_{IN} - V_{OUT} \le 10V \\ V_{IN} - V_{OUT} = 30V \end{vmatrix} $	<b>3.0</b> 0.3	<b>4.5</b>		A A
RMS Output Noise, % of V <sub>OUT</sub>	$10 \text{ Hz} \le \text{f} \le 10 \text{ kHz}$		0.001		%
Ripple Rejection Ratio	$V_{OUT} = 10V, f = 120 \text{ Hz}, C_{ADJ} = 0 \mu F$		65		dB
	$V_{OUT} = 10V, f = 120 \text{ Hz}, C_{ADJ} = 10 \mu\text{F}$	66	86		dB
Long-Term Stability	T <sub>J</sub> = 125°C, 1000 hrs		0.3	1	%
Thermal Resistance, Junction to Case	K Package		1.2	1.5	°C/W
Thermal Resistance, Junction to Ambient (No Heat Sink)	K Package		35		°C/W

#### **Electrical Characteristics** (Continued)

Specifications with standard type face are for  $T_J=25^{\circ}\text{C}$ , and those with **boldface type** apply over **full Operating Temperature Range.** Unless otherwise specified,  $V_{\text{IN}}-V_{\text{OUT}}=5\text{V}$ , and  $I_{\text{OUT}}=10\text{ mA}$ . (Note 2) (Continued)

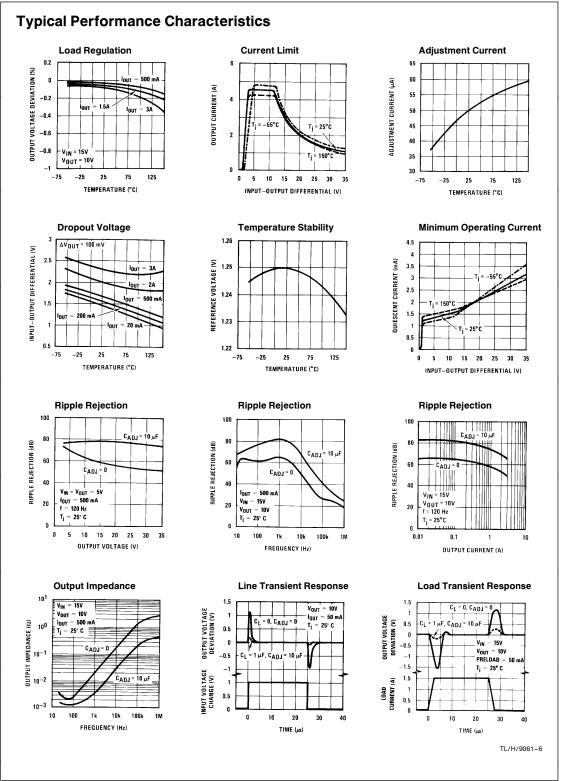
Parameter	Conditions	LM350A			LM350			Units
		Min	Тур	Max	Min	Тур	Max	Oilles
Reference Voltage	$I_{OUT} = 10 \text{ mA}, T_{J} = 25^{\circ}\text{C}$	1.238	1.250	1.262				٧
	$\begin{aligned} 3V &\leq (V_{IN} - V_{OUT}) \leq 35V, \\ 10 \text{ mA} &\leq I_{OUT} \leq 3A, P \leq 30W \end{aligned}$	1.225	1.250	1.270	1.20	1.25	1.30	V
Line Regulation	$3V \le (V_{IN} - V_{OUT}) \le 35V \text{ (Note 3)}$		0.005	0.01		0.005	0.03	%/V
			0.02	0.05		0.02	0.07	%/V
Load Regulation	$10 \text{ mA} \le I_{OUT} \le 3A \text{ (Note 3)}$		0.1	0.3		0.1	0.5	%
			0.3	1		0.3	1.5	%
Thermal Regulation	20 ms Pulse		0.002	0.01		0.002	0.03	%/W
Adjustment Pin Current			50	100		50	100	μΑ
Adjustment Pin Current Change	$10~\text{mA} \leq I_{OUT} \leq 3\text{A, }3\text{V} \leq (\text{V}_{IN} - \text{V}_{OUT}) \leq 35\text{V}$		0.2	5		0.2	5	μΑ
Temperature Stability	$T_{MIN} \le T_{J} \le T_{MAX}$		1			1		%
Minimum Load Current	$V_{IN} - V_{OUT} = 35V$		3.5	10		3.5	10	mA
Current Limit	$\begin{aligned} &V_{IN} - V_{OUT} \leq 10V \\ &V_{IN} - V_{OUT} = 30V \end{aligned}$	<b>3.0</b> 0.3	<b>4.5</b>		<b>3.0</b> 0.25	<b>4.5</b>		A A
RMS Output Noise, % of V <sub>OUT</sub>	$10 \text{ Hz} \leq f \leq 10 \text{ kHz}$		0.001			0.001		%
Ripple Rejection Ratio	$V_{OUT} = 10V, f = 120 \text{ Hz}, C_{ADJ} = 0 \mu\text{F}$		65			65		dB
	$V_{OUT} = 10V, f = 120 \text{ Hz}, C_{ADJ} = 10 \mu\text{F}$	66	86		66	86		dB
Long-Term Stability	T <sub>J</sub> = 125°C, 1000 hrs		0.25	1		0.25	1	%
Thermal Resistance, Junction to Case	K Package T Package		3	4		1.2 3	1.5 4	°C/W
Thermal Resistance, Junction to Ambient (No Heat Sink)	K Package T Package		50			35 50		°C/W

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics.

Note 2: These specifications are applicable for power dissipations up to 30W for the TO-3 (K) package and 25W for the TO-220 (T) package. Power dissipation is guaranteed at these values up to 15V input-output differential. Above 15V differential, power dissipation will be limited by internal protection circuitry. All limits (i.e., the numbers in the Min. and Max. columns) are guaranteed to National's AOQL (Average Outgoing Quality Level).

Note 3: Regulation is measured at a constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specifications for thermal regulation.

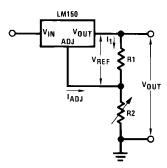
Note 4: Refer to RETS150K drawing for military specifications of the LM150K.



#### **Application Hints**

In operation, the LM150 develops a nominal 1.25V reference voltage,  $V_{\rm REF}$ , between the output and adjustment terminal. The reference voltage is impressed across program resistor R1 and, since the voltage is constant, a constant current  $I_1$  then flows through the output set resistor R2, giving an output voltage of

$$V_{OUT} = V_{REF} \left( 1 + \frac{R2}{R1} \right) + I_{ADJ} R2.$$



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Since the 50  $\mu$ A current from the adjustment terminal represents an error term, the LM150 was designed to minimize I<sub>ADJ</sub> and make it very constant with line and load changes. To do this, all quiescent operating current is returned to the output establishing a minimum load current requirement. If there is insufficient load on the output, the output will rise.

FIGURE 1

#### **EXTERNAL CAPACITORS**

An input bypass capacitor is recommended. A 0.1  $\mu$ F disc or 1  $\mu$ F solid tantalum on the input is suitable input bypassing for almost all applications. The device is more sensitive to the absence of input bypassing when adjustment or output capacitors are used but the above values will eliminate the possibility of problems.

The adjustment terminal can be bypassed to ground on the LM150 to improve ripple rejection. This bypass capacitor prevents ripple from being amplified as the output voltage is increased. With a 10  $\mu\text{F}$  bypass capacitor 86 dB ripple rejection is obtainable at any output level. Increases over 10  $\mu\text{F}$  do not appreciably improve the ripple rejection at frequencies above 120 Hz. If the bypass capacitor is used, it is sometimes necessary to include protection diodes to prevent the capacitor from discharging through internal low current paths and damaging the device.

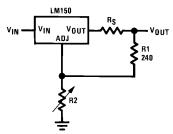
In general, the best type of capacitors to use is solid tantalum. Solid tantalum capacitors have low impedance even at high frequencies. Depending upon capacitor construction, it takes about 25  $\mu\text{F}$  in aluminum electrolytic to equal 1  $\mu\text{F}$  solid tantalum at high frequencies. Ceramic capacitors are also good at high frequencies, but some types have a large decrease in capacitance at frequencies around 0.5 MHz. For this reason, 0.01  $\mu\text{F}$  disc may seem to work better than a 0.1  $\mu\text{F}$  disc as a bypass.

Although the LM150 is stable with no output capacitors, like any feedback circuit, certain values of external capacitance can cause excessive ringing. This occurs with values between 500 pF and 5000 pF. A 1  $\mu\text{F}$  solid tantalum (or 25  $\mu\text{F}$  aluminum electrolytic) on the output swamps this effect and insures stability.

#### LOAD REGULATION

The LM150 is capable of providing extremely good load regulation but a few precautions are needed to obtain maximum performance. The current set resistor connected between the adjustment terminal and the output terminal (usually 240 $\Omega$ ) should be tied directly to the output (case) of the regulator rather than near the load. This eliminates line drops from appearing effectively in series with the reference and degrading regulation. For example, a 15V regulator with  $0.05\Omega$  resistance between the regulator and load will have a load regulation due to line resistance of  $0.05\Omega \times I_{OUT}$ . If the set resistor is connected near the load the effective line resistance will be  $0.05\Omega$  (1 + R2/R1) or in this case, 11.5 times worse

Figure 2 shows the effect of resistance between the regulator and 240  $\!\Omega$  set resistor.



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FIGURE 2. Regulator with Line Resistance in Output Lead

With the TO-3 package, it is easy to minimize the resistance from the case to the set resistor, by using two separate leads to the case. The ground of R2 can be returned near the ground of the load to provide remote ground sensing and improve load regulation.

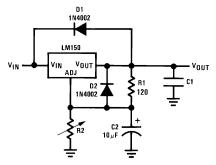
#### **PROTECTION DIODES**

When external capacitors are used with any IC regulator it is sometimes necessary to add protection diodes to prevent the capacitors from discharging through low current points into the regulator. Most 10  $\mu \rm F$  capacitors have low enough internal series resistance to deliver 20A spikes when shorted. Although the surge is short, there is enough energy to damage parts of the IC.

When an output capacitor is connected to a regulator and the input is shorted, the output capacitor will discharge into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage of the regulator, and the rate of decrease of  $V_{\text{IN}}.$  In the LM150, this discharge path is through a large junction that is able to sustain 25A surge with no problem. This is not true of other types of positive regulators. For output capacitors of 25  $\mu\text{F}$  or less, there is no need to use diodes.

The bypass capacitor on the adjustment terminal can discharge through a low current junction. Discharge occurs when either the input or output is shorted. Internal to the LM150 is a  $50\Omega$  resistor which limits the peak discharge current. No protection is needed for output voltages of 25V or less and 10  $\mu\text{F}$  capacitance. Figure 3 shows an LM150 with protection diodes included for use with outputs greater than 25V and high values of output capacitance.

#### **Application Hints** (Continued)



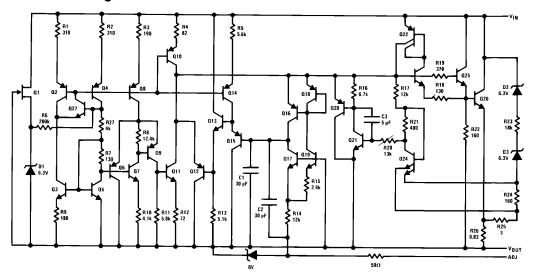
D1 protects against C1 D2 protects against C2

$$V_{OUT} = 1.25 V \left(1 + \frac{R2}{R1}\right) + I_{ADJ}R2$$

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FIGURE 3. Regulator with Protection Diodes

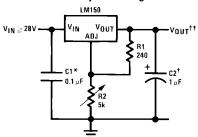
#### **Schematic Diagram**



TL/H/9061-10

#### **Typical Applications**

#### 1.2V—25V Adjustable Regulator



TL/H/9061-1

Full output current not available at high input-output voltages.

†Optional—improves transient response. Output capacitors in the range of 1  $\mu F$  to 1000  $\mu F$  of aluminum or tantalum electrolytic are commonly used to provide improved output impedance and rejection of transients.

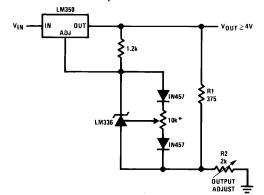
\*Needed if device is more than 6 inches from filter capacitors.

$$\dagger\dagger V_{\mbox{OUT}} = 1.25 V \left(1 + \frac{\mbox{R2}}{\mbox{R1}}\right) + I_{\mbox{ADJ}} \left(\mbox{R2}\right)$$

Note: Usually R1 = 240 $\Omega$  for LM150 and R1 = 120 $\Omega$  for LM350.

#### Typical Applications (Continued)

## Precision Power Regulator with Low Temperature Coefficient

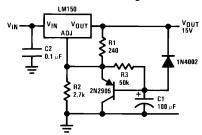


\*Adjust for 3.75V across R1

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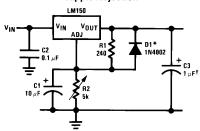
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#### Slow Turn-ON 15V Regulator



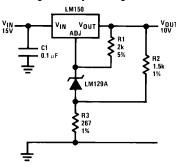
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## Adjustable Regulator with Improved Ripple Rejection



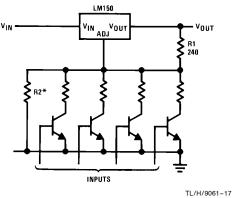
†Solid tantalum \*Discharges C1 if output is shorted to ground

#### High Stability 10V Regulator



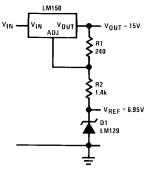
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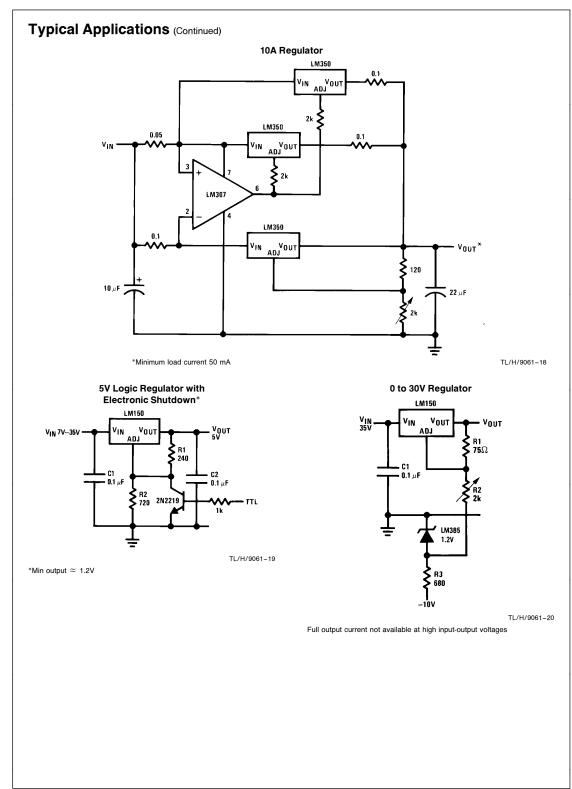
#### **Digitally Selected Outputs**



\*Sets maximum V<sub>OUT</sub>

#### Regulator and Voltage Reference





## Typical Applications (Continued) 5A Constant Voltage/Constant Current Regulator MJ4502 ₹ R3 0.2 5W CURRENT ADJUST R2 250k 100 pF LM350 0UTPUT 1.2V-30V V<sub>OUT</sub> D2 1 1N457 **₹** R4 680 LM301A D1 1N457 D3 LED\* **₹** R6 240 C5 75 pF R5 330k R8 5k Voltage Adjust $\dagger$ Solid tantalum TSolid tantalum -6V T0 -15V \*Lights in constant current mode TL/H/9061-21 12V Battery Charger LM350 **₹** R3 230 TO 12V BATTERY Q1 2N2905 LM301A 1000 pF

START

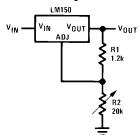
# 

# Precision Current Limiter $v_{IN} = v_{IN} v_{0UT} = v_{REF},$ $v_{IN} = v_{0UT} = v_{REF},$ $v_{0UT} = v_{REF},$ $v_{0UT} = v_{REF},$ $v_{0UT} = v_{0UT} = v_{0UT}$ $v_{0UT} = v_{0UT}$

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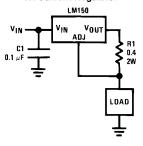
#### 1.2V-20V Regulator with Minimum Program Current

-5V TO −10V



TL/H/9061-25

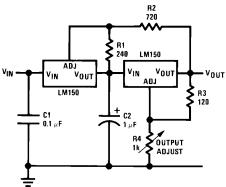
#### 3A Current Regulator



TL/H/9061-26

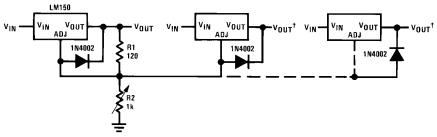
\*Minimum output current  $\approx$  4 mA

#### **Tracking Preregulator**



# Typical Applications (Continued) Adju VIN VOUT VOUT VOUT

# Adjusting Multiple On-Card Regulators with Single Control\*



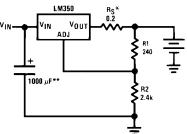
†Minimum load—10 mA

\*All outputs within  $\pm\,100$  mV

# AC Voltage Regulator LM350 VIN VOUT 480 480 VIN VOUT 12 Vp-p 3A ADJ VOUT

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#### Simple 12V Battery Charger



TL/H/9061-30

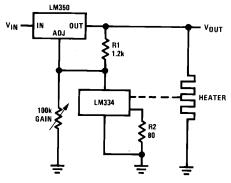
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\*RS—sets output impedance of charger: Z\_{OUT} = R\_{S} \left(1 + \frac{R2}{R1}\right)

Use of  $\ensuremath{\mathsf{R}}_S$  allows low charging rates with fully charged battery.

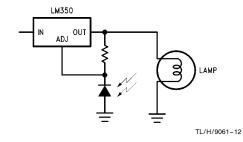
\*\*1000  $\mu\text{F}$  is recommended to filter out any input transients

#### Temperature Controller

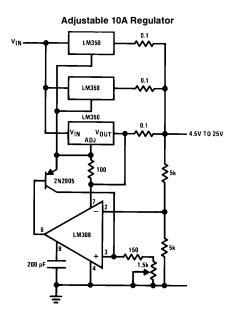


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#### **Light Controller**

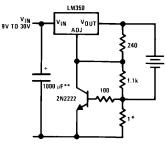


#### Typical Applications (Continued)



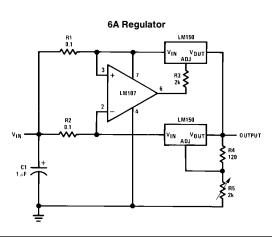
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#### **Current Limited 6V Charger**

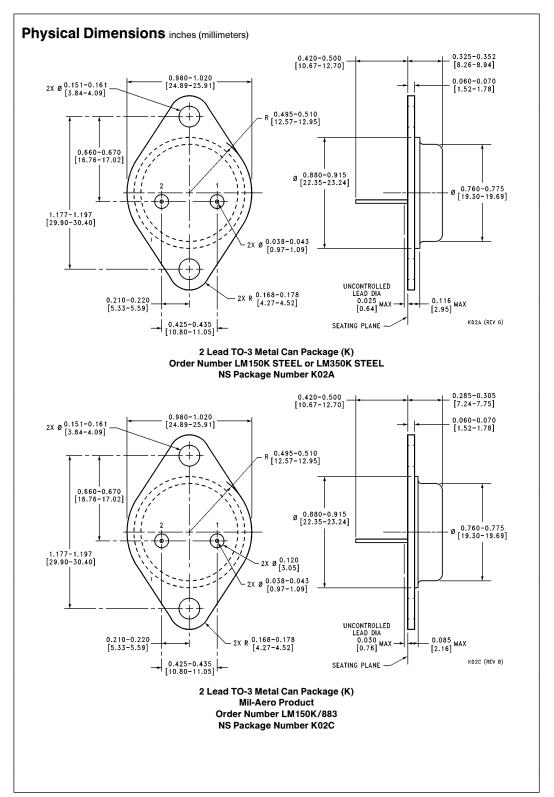


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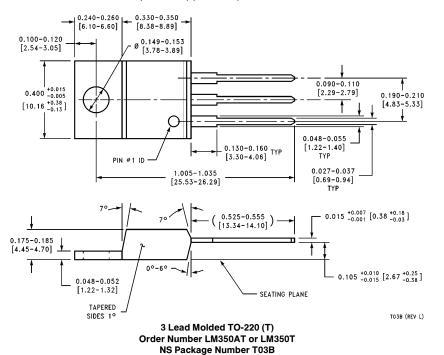
<sup>\*\*1000</sup>  $\mu {\rm F}$  is recommended to filter out any input transients.



<sup>\*</sup>Sets peak current (2A for 0.3 $\Omega$ )



#### Physical Dimensions inches (millimeters) (Continued)



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National Semiconductor Corporation 1111 West Bardin Road Arlington, TX 76017 Tel: 1(800) 272-9959 Fax: 1(800) 737-7018 National Semiconductor Europe

Fax: (+49) 0-180-530 85 86
Email: cnjwge@tevm2.nsc.com
Deutsch Tel: (+49) 0-180-530 85 85
English Tel: (+49) 0-180-532 78 32
Français Tel: (+49) 0-180-532 78 61
Italiano Tel: (+49) 0-180-534 16 80

National Semiconductor Hong Kong Ltd. 13th Floor, Straight Block, Ocean Centre, 5 Canton Rd. Tsimshatsui, Kowloon Hong Kong Tel: (852) 2737-1600 Fax: (852) 2736-9960 National Semiconductor Japan Ltd. Tel: 81-043-299-2309 Fax: 81-043-299-2408