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**NJM2904**

## SINGLE-SUPPLY DUAL OPERATIONAL AMPLIFIER

### ■ GENERAL DESCRIPTION

The NJM2904 consists of two independent, high gain, internally frequency compensated operation amplifiers which were designed specifically to operate from a single power supply over a wide range of voltages. Operation from split power supplies is also possible and the low power supply current drain is independent of the magnitude of the power supply voltage.

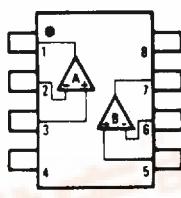
Application areas include transducer amplifiers, DC gain blocks, and all the conventional op amp circuits which now can be more easily implemented in single power supply systems. For example, the NJM2904 can be directly operated off of the standard +5V power supply voltage which is used in digital systems and will easily provide the required interface electronics without requiring the additional  $\pm 15$ V power supplies.

### ■ FEATURES

- Single Supply
- Operating Voltage ( $+3V \sim +32V$ )
- Low Operating Current (0.7mA typ.)
- Slew Rate (0.5V/ $\mu s$  typ.)
- Bipolar Technology
- Package Outline DIP8, DMP8, SIP8, SSOP8

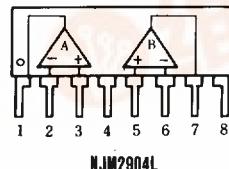
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### ■ PIN CONFIGURATION

NJM2904D  
NJM2904M  
NJM2904V

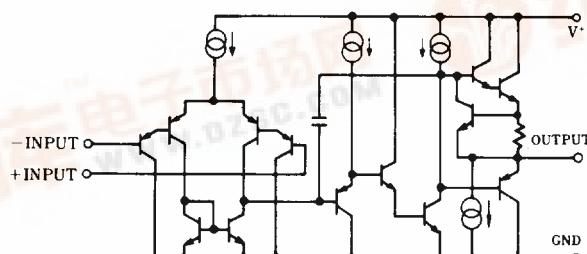
PIN FUNCTION

1.	A OUTPUT
2.	A-INPUT
3.	A+INPUT
4.	GND
5.	B+INPUT
6.	B-INPUT
7.	B OUTPUT
8.	V+



NJM2904L

### ■ EQUIVALENT CIRCUIT (1/2 Shown)





**NJM2904**

■ ABSOLUTE MAXIMUM RATINGS

(Ta=25°C)

PARAMETER	SYMBOL	RATINGS	UNIT
Supply Voltage	V <sup>+</sup> (V <sup>+</sup> /V <sup>-</sup> )	32(or ±16)	V
Differential Input Voltage	V <sub>ID</sub>	32	V
Input Voltage	V <sub>I</sub>	-0.3~+32	V
Power Dissipation	P <sub>D</sub>	(DIP8) 500	mW
		(DMP8) 300	mW
		(SSOP8) 300	mW
		(SIP8) 800	mW
Operating Temperature Range	T <sub>opr</sub>	-40~+85	°C
Storage Temperature Range	T <sub>stg</sub>	-50~+120	°C

■ ELECTRICAL CHARACTERISTICS

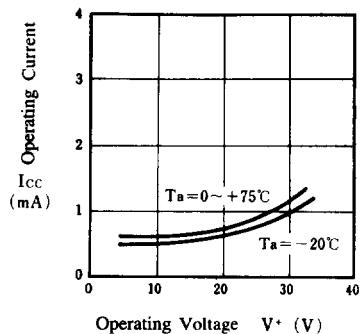
(Ta=25°C V<sup>+</sup>=5V)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Input Offset Voltage	V <sub>IO</sub>	R <sub>S</sub> =0Ω	—	2	7	mV
Input Offset Current	I <sub>IO</sub>		—	5	50	nA
Input Bias Current	I <sub>B</sub>		—	25	250	nA
Large Signal Voltage Gain	A <sub>V</sub>	R <sub>L</sub> ≥2kΩ	—	100	—	dB
Maximum Output Voltage Swing	V <sub>OPP</sub>	R <sub>L</sub> =2kΩ	3.5	—	—	V
Input Common Mode Voltage Range	V <sub>ICM</sub>		0~3.5	—	—	V
Common Mode Rejection Ratio	CMR		—	85	—	dB
Supply Voltage Rejection Ratio	SVR		—	100	—	dB
Output Source Current	I <sub>SOURCE</sub>	V <sub>IN</sub> <sup>+</sup> =1V, V <sub>IN</sub> <sup>-</sup> =0V	20	30	—	mA
Output Sink Current	I <sub>SINK</sub>	V <sub>IN</sub> <sup>+</sup> =0V, V <sub>IN</sub> <sup>-</sup> =1V	8	20	—	mA
Channel Separation	CS	f=1k~20kHz, Input Referred	—	120	—	dB
Operating Current	I <sub>cc</sub>	R <sub>L</sub> =∞	—	0.7	1.2	mA
Slew Rate	SR		—	0.5	—	V/μs
Unity Gain Bandwidth	f <sub>T</sub>		—	0.2	—	MHz

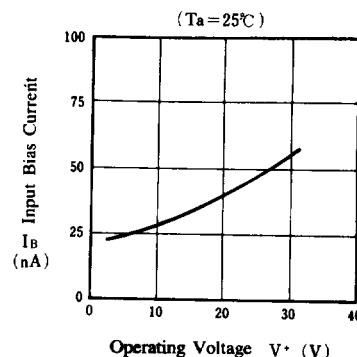


## ■ TYPICAL CHARACTERISTICS

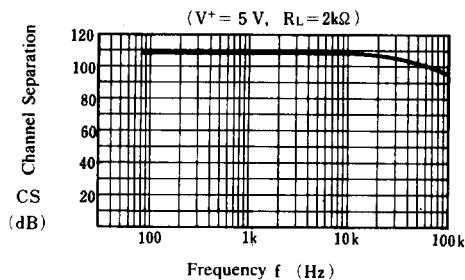
**Operating Current  
vs. Operating Voltage**



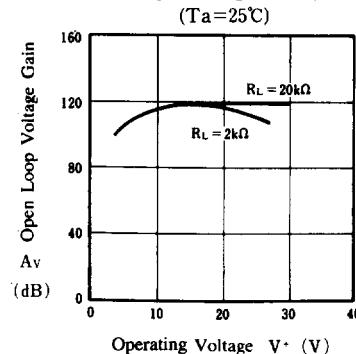
**Input Bias Current  
vs. Operating Voltage**



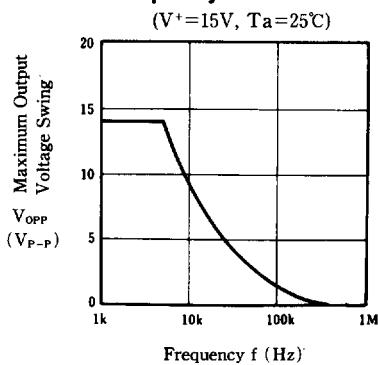
**Channel Separation vs. Frequency**



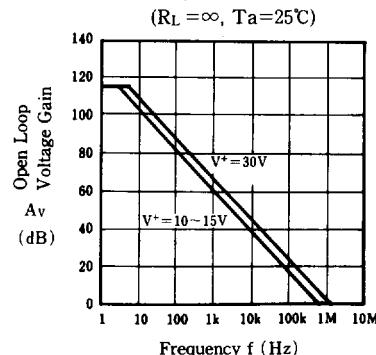
**Voltage Gain  
vs. Operating Voltage**



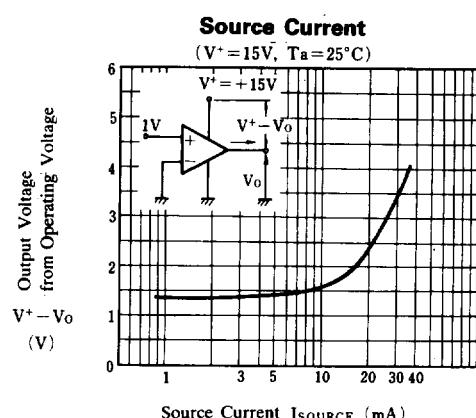
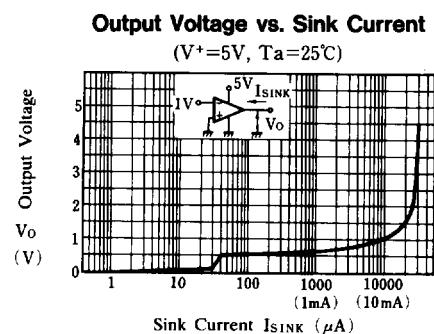
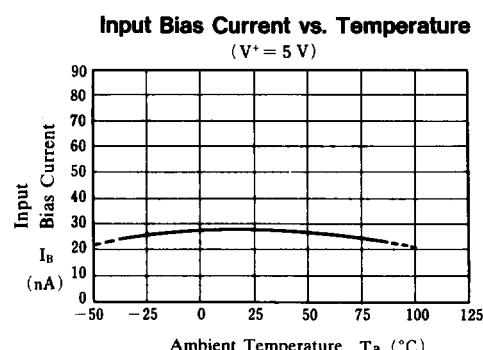
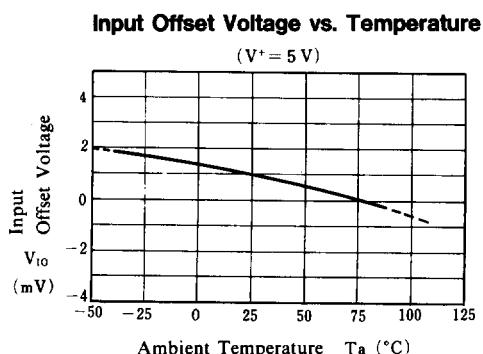
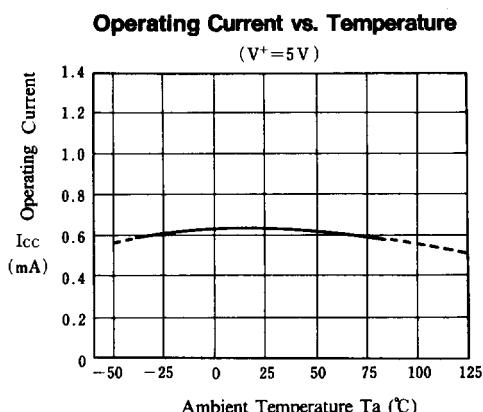
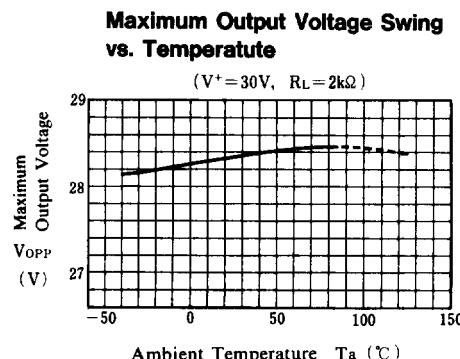
**Maximum Output Voltage Swing  
vs. Frequency**



**Open Loop Voltage Gain  
vs. Frequency**

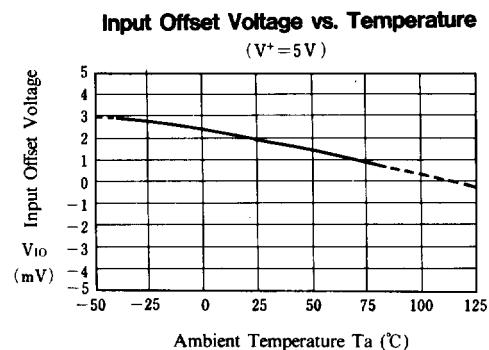
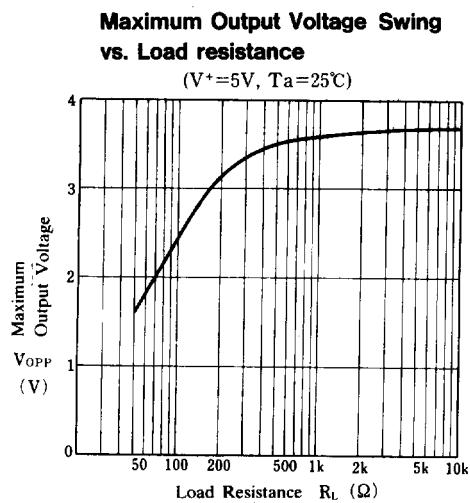
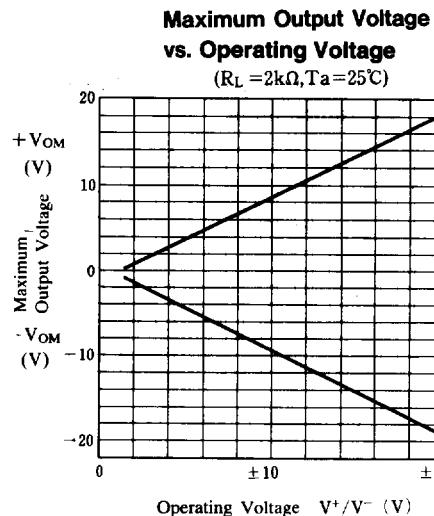
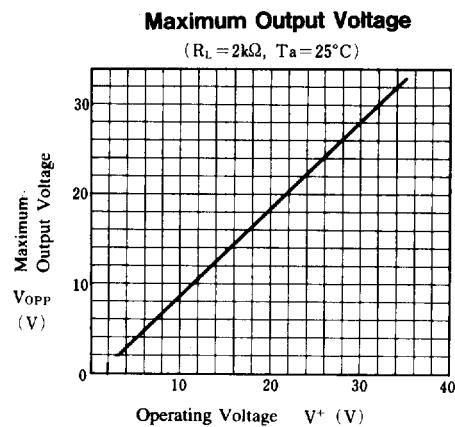


## ■ TYPICAL CHARACTERISTICS





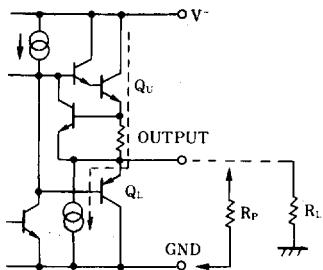
## ■ TYPICAL CHARACTERISTICS





## ■ APPLICATION

- Improvement of Cross-over Distortion  
Equivalent circuit at the output stage



NJM2904, in its static state (No in and output condition) when design,  $Q_U$  being biassed by constant current (breakdown beam) yet,  $Q_L$  stays OFF.

While using with both power source mode, the cross-over distortion might occur instantly when  $Q_L$  ON.

There might be cases when application for amplifier of audio signals, not only distortion but also the apparent frequency bandwidth being narrowed remarkably.

It is advisable especially when using both power source mode, constantly to use with higher current on  $Q_U$  than the load current (including feedback current), and then connect the pull-down resistor  $R_P$  at the part between output and GND pins.

