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CMOS LSI



LC75366, 75366M

Two-Channel Electronic Volume Control



Overview

The LC75366 (DIP20) and the LC75366M (MFP20) are electronic volume controls that can be controlled by serial input data and provide volume, balance and loudness functions.

Features

• Silicon gate CMOS process for low switching noise

Functions

• Volume: 0 dB to -68 dB (in 2 dB steps) and $-\infty$; 36 positions.

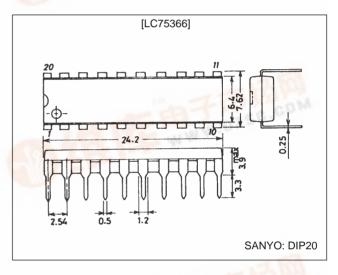
A balance function can be implemented by controlling the left and right channel volume settings independently.

- Loudness: Taps are provided at the -20 dB positions in the 10 dB step resistor ladder used by the volume control function. A loudness function can be implemented by attaching external RC circuits at these tap points.
- An address selection pin (the S pin) allows two LC75366 chips to be used on the same bus.
- Serial data input: Supports CCB* format communication with the system controller.
 - CCB is a trademark of SANYO ELECTRIC CO., LTD.
 - CCB is SANYO's original bus format and all the bus addresses are controlled by SANYO.

Package Dimensions

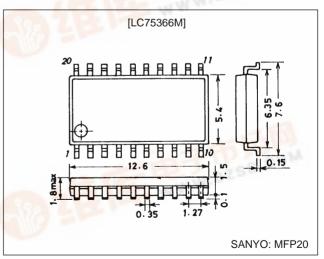
unit: mm

3021B-DIP20



unit: mm

3036B-MFP20



Specifications

Absolute Maximum Ratings at $Ta = 25^{\circ}C$, $V_{SS} = 0 V$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V _{DD} max V _{DD}		12	V
Maximum input volto go	V _{IN} max1	CL, DI, CE, S	V_{SS} – 0.3 to V_{DD} + 0.3	V
Maximum input voltage	V _{IN} max2	L10dBIN, L2dBIN, R10dBIN, R2dBIN	V_{SS} – 0.3 to V_{DD} + 0.3	V
Allowable power dissipation	Pd max	Ta ≤ 85°C	140	mW
Operating temperature	Topr		-40 to +85	°C
Storage temperature	Tstg		-50 to +125	°C



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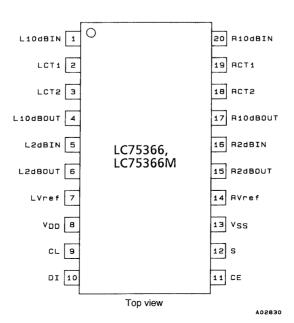
Allowable Operating Ranges at $Ta=25^{\circ}C,\,V_{SS}$ = 0 V

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V _{DD}	V _{DD}	4.0 to 11.0	V
Input high level voltage	V _{IH} (1)	CL, DI, CE	0.3 V _{DD} + 1 to V _{DD}	V
input high level voltage	V _{IH} (2)	S	0.8 V _{DD} to V _{DD}	V
Input low level voltage	V _{IL} (1)	CL, DI, CE	V_{SS} to 0.2 V_{DD}	V
	V _{IL} (2)	S	V_{SS} to 0.2 V_{DD}	V
Input voltage amplitude	V _{IN}	L10dBIN, L2dBIN, R10dBIN, R2dBIN	V _{SS} to V _{DD}	Vp-p
Input pulse width	t _{øW}	CL	1 or longer	μs
Setup time	t _{set up}	CL, DI, CE	1 or longer	μs
Hold time	t _{hold}	CL, DI, CE	1 or longer	μs
Operating frequency	fop	CL	Up to 500	kHz

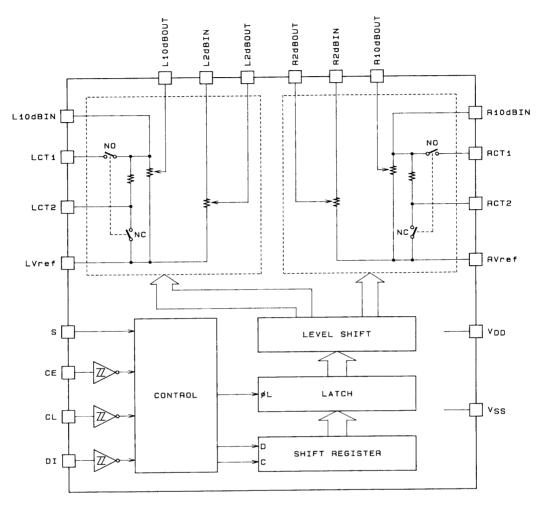
Electrical Characteristics at $Ta=25^{\circ}C,\,V_{SS}$ = 0 V

Parameter	Symbol	Conditions	min	typ	max	Unit
Total harmonic distortion	THD (1)	V_{IN} = 1 Vrms, f = 1 kHz, all settings flat overall, V _{DD} = 9 V		0.006		%
rotal narmonic distortion	THD (2)	V_{IN} = 1 Vrms, f = 20 kHz, all settings flat overall, V _{DD} = 9 V		0.015		%
Crosstalk	СТ	V_{IN} = 1 Vrms, f = 20 kHz, all settings flat overall, Rg = 1 kΩ		85		dB
Output at maximum attenuation	V _O min	V_{IN} = 1 Vrms, f = 20 kHz, volume setting: -∞, with a 470 µF capacitor between L/R Vref and V _{SS}		-80		dB
Total resistance	R _{VOL} (1)	10 dB steps	28.2	47	65.8	kΩ
	R _{VOL} (2)	2 dB steps	12	20	28	kΩ
Output off leakage current	loff	L10dBIN, R10dBIN, LCT1, L2dBIN, R2dBIN, RCT1, L10dBOUT, R10dBOUT, LCT2, L2dBOUT, R2dBOUT, RCT2, LVref, RVref	-10		+10	μΑ
Input high level current	IIH	VI = V _{DD} (CL, CE and DI pins)			10	μA
Input low level current	Ι _{ΙL}	VI = V _{SS} (CL, CE and DI pins)	-10			μA
Output noise voltage	V _N	All settings flat overall (IHF-A), V_{DD} = 9 V, Rg = 1 k Ω		2	10	μV
Current drain	I _{DD}	$V_{DD} - V_{SS} = 11 V$			1	mA
Analog switch on resistance (Design target value)	R _{ON}	CT1	180	300	420	Ω
		For use between CT2 and Vref	90	150	210	Ω
		0 dB, -∞	0.6	1.0	1.4	kΩ
		Other than the above	6.0	10.0	14.0	kΩ

Pin Assignment



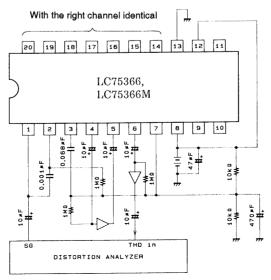
Equivalent Circuit Block Diagram



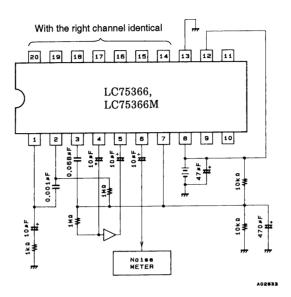
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Test Circuits

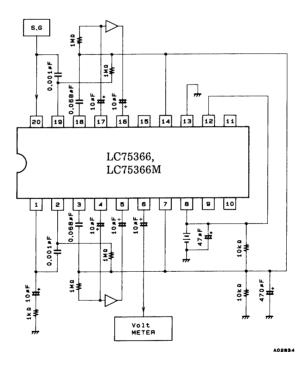
1. Total harmonic distortion



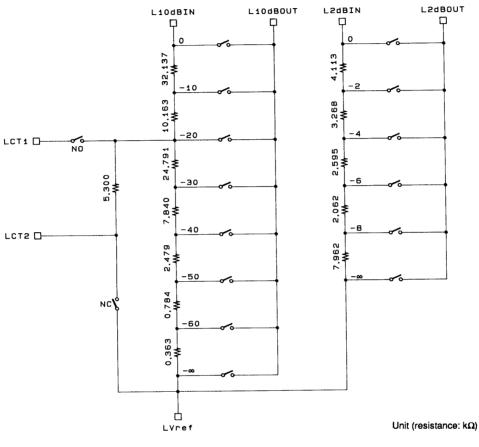
2. Output noise voltage



3. Crosstalk



Volume Block Equivalent Circuit



The right channel is identical.

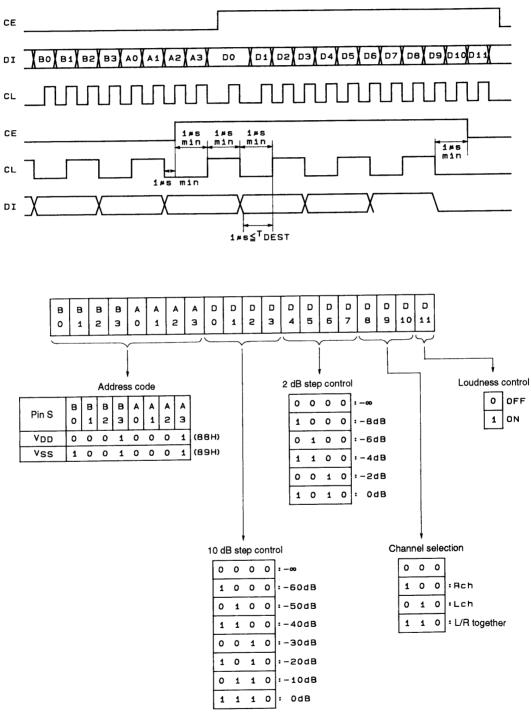
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Pin Functions

Pin No.	Symbol	Function	Note
1	L10dBIN	Input pins for the 10 dB step attenuator. Must be driven from a low impedance.	
20	R10dBIN		VSS # # A02836
2	LCT1		
3	LCT2	Loudness connections. Connect a high band compensation capacitor between	
19	RCT1	 CT1 and 10dBIN and connect a low band compensation capacitor between CT2 and Vref. 	
18	RCT2		
4	L10dBOUT	Output pins for the 10 dB step attenuator. These outputs must be received by a	
17	R10dBOUT	load of about 1 MΩ.	₩ ₩ VSS 402837
5	L2dBIN		
16	R2dBIN	Input pins for the 2 dB step attenuator. Must be driven from a low impedance.	□
6	L2dBOUT	Output pins for the 2 dB step attenuator. These outputs must be received by a	
15	R2dBOUT	load of about 1 MΩ.	₩ ₩ ₩ VSS
7	LVref	Volume circuit common pins. The impedance of the pattern connected to these pins should be kept as low as possible. Since the capacitors between Vref and V _{SS} form the residual resistance	
14	RVref	components when the volume is cut, adequate care must be taken in determining the value of these capacitors.	Vref ZZ A02838
12	S	Pin that selects the address code in the data format. Data will be accepted for an address code of 88 when this pin is tied to V_{DD} , and for an address code of 89 when tied to V_{SS} .	VDD
9	CL		- VDD
10	DI	Serial data and clock inputs for control	× ×
11	CE	Chip enable. The internally latched data is written and the analog switches operate when this pin goes from high to low. Data transfer is enabled when this pin is at the high level.	VSS A02840
8	V _{DD}		
13	V _{SS}	 These pins must be connected to the power supply. 	

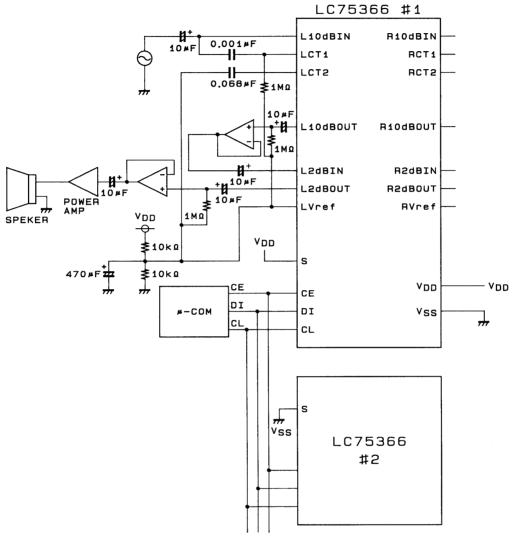
Control System Timing and Data Format

Apply the stipulated serial data to the CE, CL and DI pins to control the LC75366 and LC75366M. The data consists of 20 bits, of which 8 bits are the address and 12 bits are control data.

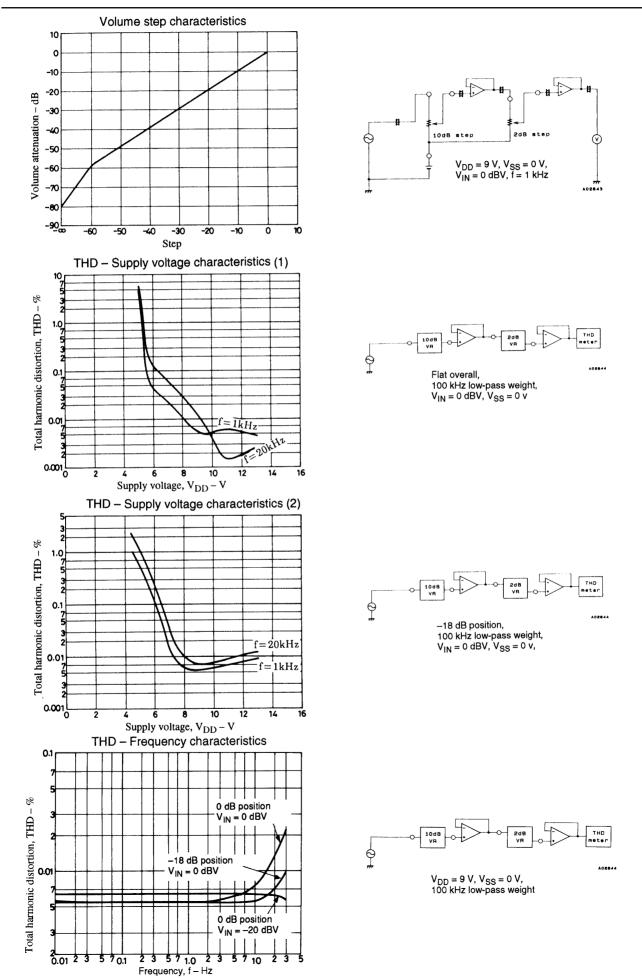


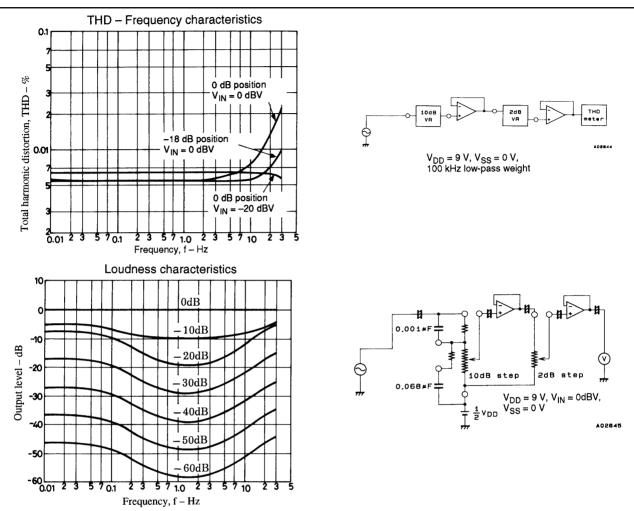
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Sample Application Circuit



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Loudness Function External Circuit Constant Calculation Example

First, refer to the LC75366 and LC75366M 10 dB step internal equivalent circuit shown on page 5. Figure 1 below shows this circuit simplified with the external components used for the loudness function connected for this calculation. The sample calculation below uses this diagram to acquire a 5 dB boost at f = 100 Hz.

(f = 100 Hz, 5 dB boost) Let R and C in Figure 1 be: R1 = R2 = 10 k Ω R3 = 1 k Ω C1 = Z1, C2 = Z2 Then:

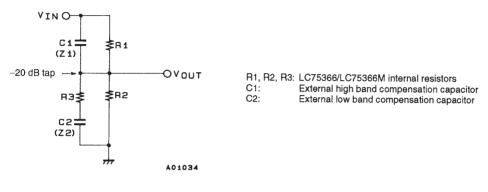
$$V_{OUT} = \frac{\frac{R2 (R3 + Z2)}{R2 + R3 + Z2}}{(at = 1 \text{ kHz})} = -20 \text{ dB}$$
$$(at = 1 \text{ kHz}) \frac{\frac{R1 \cdot Z1}{R1 + Z1} + \frac{R2 (R3 + Z2)}{R2 + R3 + Z2}}{\frac{R2 (R3 + 10 \cdot Z2)}{R2 + R3 + 10 \cdot Z2}} = -15 \text{ dB}$$
$$(at = 100 \text{ Hz}) \frac{\frac{R1 \cdot 10 \cdot Z1}{R1 + 10 \cdot Z1} + \frac{R2 (R3 + 10 \cdot Z2)}{R2 + R3 + 10 \cdot Z2}}{R2 + R3 + 10 \cdot Z2} = -15 \text{ dB}$$

Solving the above equations gives:

 $Z1 \neq 178.3 \text{ k}\Omega$ and $Z2 = 176 \Omega$

Therefore, under such conditions where f = 1 kHz, specifications may be satisfied if C (capacitor) having these impedances is supplied externally.

The result is that C1 = 893 pF and $C2 = 0.9 \mu$ F.





Usage Notes

- 1. The states of the internal analog switches will be indeterminate when power is first applied. Muting should be applied externally until control data has been transferred and stored.
- 2. The signal lines for the CL, DI and CE pins should either be covered by the pattern ground or be formed from shielded cable to prevent the high-frequency digital signals transmitted over these lines from entering the analog system.

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