

19-1261: Rev 1: 11/98

EVALUATION KIT MANUAL FOLLOWS DATA SHEET



# **LED Driver with** Programmable Prebias Voltage

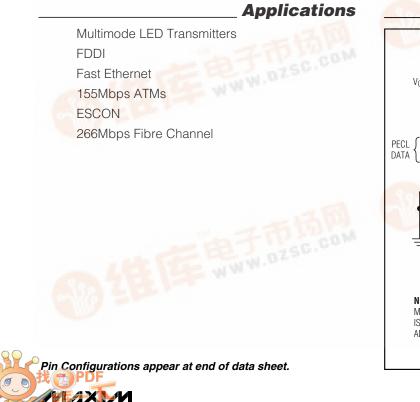
## **General Description**

The MAX3966 is a programmable LED driver for fiber optic transmitters operating at data rates up to 266Mbps. The circuit contains a high-speed current driver with programmable temperature coefficient (tempco), adjustments for LED prebias voltage, and a V<sub>BB</sub> reference voltage generator. The circuit accepts PECL data inputs, and operates from a single +3V to +5.5V power supply.

The MAX3966 can switch up to 100mA into typical high-speed light-emitting diodes. As temperature increases, the device's modulation current increases with a tempco that is programmable from 2500ppm/°C to 12,000ppm/°C. The modulation current is programmed with a single external resistor.

The MAX3966's LED prebias voltage is programmable from 400mV to 925mV. The prebias circuit produces peaking current, which improves the LED switching speed.

Complementary current outputs help to maintain a constant supply current, reducing EMI and supply noise generated by the transmitter module. The MAX3966 is available in die form, or in 16-pin and 24-pin QSOP packages.



## Features

- Programmable LED Prebias Voltage
- Single +3V to +5.5V Power Supply
- Complementary Output Reduces Supply Noise
- ♦ Adjustable Modulation Current
- ♦ Adjustable Temperature Compensation
- Available in 16-Pin/24-Pin QSOP or Die

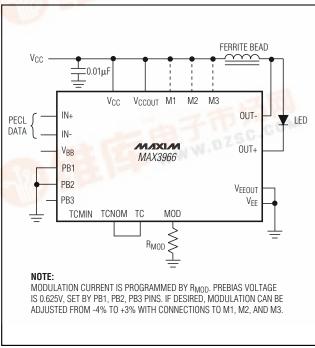
## **Ordering Information**

PART	TEMP. RANGE	PIN-PACKAGE
MAX3966CEE*	0°C to +70°C	16 QSOP
MAX3966CEG	0°C to +70°C	24 QSOP
MAX3966C/D	0°C to +100°C**	Dice

\*Some features are not available in the 16-pin QSOP package. Contact factory for information and availability.

\*\*Dice are designed to operate over a 0°C to +100°C junction temperature (Tj) range, but are tested and guaranteed only at  $T_A = +25$ °C.

## **Typical Operating Circuit**



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## **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage at V<sub>CC</sub>, V<sub>CCOUT</sub> (V<sub>EE</sub> = 0V) .....-0.5V to 7V Current into OUT+, OUT- .....-40mA to 160mA Differential Output Voltage (V<sub>OUT</sub>+ to V<sub>OUT</sub>-) .....-3.3V to 3.3V Voltage at M1, M2, M3, PB1, PB2, PB3,

IN+, IN-, OUT+, OUT-, V<sub>BB</sub>.....-0.5V to (V<sub>CC</sub> + 0.5V) Voltage at TCMIN, TCNOM, TC, MOD.....-0.5V to 2V Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )

24-pin QSOP (derate 10mW/°C above +70°C)770mW/°C
Operating Junction Temperature Range50°C to +150°C
Die Attach Temperature+400°C
Storage Temperature Range50°C to +150°C
Lead Temperature (soldering, 10sec)+300°C

CCZ mN//OC

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## DC ELECTRICAL CHARACTERISTICS

(Load as specified in Figure 1;  $V_{CC} = 3.0V$  to 5.5V,  $V_{EE} = 0V$ ,  $T_A = 0^{\circ}C$  to +70°C, unless otherwise noted. Temperature coefficients are referenced to  $T_A = +25^{\circ}C$ . Typical values are at  $T_A = +25^{\circ}C$ ,  $V_{CC} = 3.3V$ , unless otherwise noted. Dice are tested at  $T_A = +25^{\circ}C$  only.)

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS	
Data Input High Voltage	Referenced to V <sub>CC</sub>	-1.165		-0.880	V	
Data Input Low Voltage	Referenced to V <sub>CC</sub>	-1.810		-1.475	V	
Supply Current	(Note 1)		30	35	mA	
Input Current at IN+ or IN-		-50		50	μA	
Maximum Programmable Modulation Current	(Note 2)	100	130		mA	
Minimum Programmable Modulation Current				40	mA	
Modulation-Current Accuracy	(Note 3)	-10		10	%	
	PB1, PB2, PB3 = (open, open, open)	0.380	0.400	0.428		
Prebias Voltage	PB1, PB2, PB3 = (V <sub>EE</sub> , V <sub>EE</sub> , open)	0.594	0.625	0.663	V	
	$PB1, PB2, PB3 = (V_{EE}, V_{EE}, V_{EE})$	0.879	0.925	0.981		
Towns and we Or affinished of	Maximum tempco (TC open)		12,000			
Temperature Coefficient of Modulation Current	Nominal tempco (TC, TCNOM shorted)		3600	3600 ppm/°C		
	Minimum tempco (TC, TCMIN shorted)		2500			
V <sub>BB</sub> Voltage	Referenced to V <sub>CC</sub>	-1.38	-1.33	-1.26	V	
Prebias Resistor (RPREBIAS)		66.3	78	89.7	Ω	

Note 1:  $R_{MOD} = 1k\Omega$ . Excludes  $I_{OUT+}$  and  $I_{OUT-}$ .

Note 2:  $V_{CC}$  = 3.0V.  $R_{MOD}$  = 698 $\Omega$ , TC connected to TCMIN.

Note 3:  $V_{CC} = 3.3V$ ,  $V_{LED} = 1.55V$ , prebias voltage programmed at 0.625V (nominal),  $T_A = +25^{\circ}C$ .  $R_{MOD} = 1k\Omega$ , and programs approximately 80mA current. M1 =  $V_{CC}$ , M2 = open, M3 = open, TC connected to TCNOM.

## **AC ELECTRICAL CHARACTERISTICS**

(Load as specified in Figure 1, unless otherwise noted.  $V_{CC} = 3.0V$  to 5.5V,  $R_{MOD} = 1k\Omega$ ,  $T_A = 0^{\circ}C$  to +70°C. Input data-edge speed = 1ns (typ),  $V_{PREBIAS} = 0.625V$ . Typical values are at  $T_A = +25^{\circ}C$ ,  $V_{CC} = 3.3V$ , nominal tempco.)

PARAMETER	PARAMETER CONDITIONS		ТҮР	MAX	UNITS
Output Current Edge Speed	20% to 80%, input is a 12.5MHz square wave		615	1000	ps
Output Current Pulse-Width Correction (PWC)	Input is a 12.5MHz square wave (Note 4)		-210		ps
Output Current Data-Dependent	266Mbps (Note 5)		150		20
Jitter	155Mbps (Note 6)		215	330	ps
Random Jitter	RMS		1.1		ps

**Note 4:** PWC = (widthCURRENT ON - widthCURRENT OFF) / 2.

Note 5: Test pattern is a 2 - 1PRBS transmitted at 266Mbps (3.75ns per bit).

Note 6: Test pattern is a 2<sup>13</sup> - 1PRBS containing 72 consecutive zeros or 72 consecutive ones.

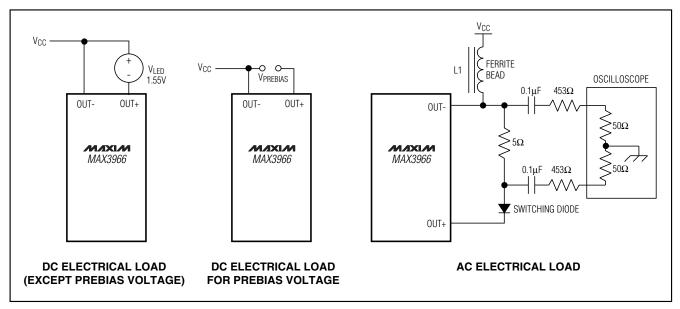
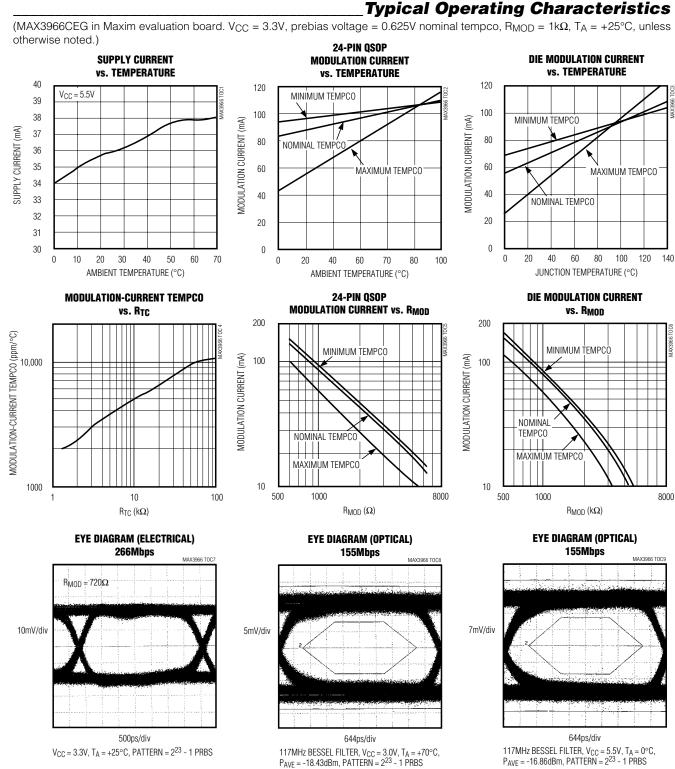


Figure 1. MAX3966 Output Test Loads



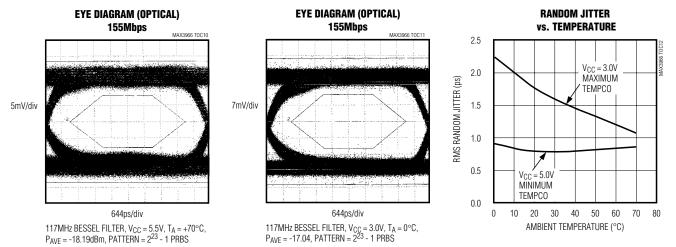
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**MAX3966** 

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## Typical Operating Characteristics (continued)

(MAX3966CEG in Maxim evaluation board.  $V_{CC}$  = 3.3V, prebias voltage = 0.625V nominal tempco,  $R_{MOD}$  = 1k $\Omega$ , T<sub>A</sub> = +25°C, unless otherwise noted.)



## Pin Description

PIN		NAME	FUNCTION	
24-PIN	16-PIN	NAME	FUNCTION	
1, 2, 24		M2, M3, M1	Pins M1, M2, and M3 provide fine adjustment of the modulation current (Table 2).	
3	2	MOD	A resistor from MOD to $V_{\mbox{\scriptsize EE}}$ programs the LED modulation current.	
4, 23	1	Vcc	Provides current to the internal amplifiers.	
5	3	V <sub>BB</sub>	Input Reference Voltage PECL Threshold Voltage Output. Connect $V_{BB}$ to IN- for single-ended input.	
6	4	IN-	Inverting Data Input	
7	5	IN+	Noninverting Data Input	
8	6	V <sub>EE</sub>	Ground for Internal Amplifiers	
9	7	TCMIN	Shorting TC to TCMIN provides the minimum modulation-current tempco.	
10		TCNOM	Shorting TC to TCNOM provides a modulation tempco of approximately 3600ppm/°C.	
11	8	TC	A resistor connected between the TC and TCMIN pins sets the tempco of the modulation current. Leaving TC unconnected provides the maximum tempco.	
12, 13, 14	_	PB1, PB2, PB3	Programs the voltage prebias at the OUT+ pin (Table 1).	
15, 16	9, 10	VEEOUT	Ground for the Output Current Drivers	
17, 18	11, 12	OUT+	Current-Output Pins	
19, 20	13, 14	OUT-	Complementary Current-Output Pins	
21, 22	15, 16	VCCOUT	Supply Connection for the Output Current Drivers	

# MAX3966



### **Detailed Description**

The MAX3966 provides a flexible current drive for the modulation of fiber optic light-emitting diodes (LEDs). The circuit is designed to be used with +3.3V or +5V power supplies. The IC provides up to 100mA of modulation current. An adjustable prebias current source sets the LED prebias voltage. An integrated resistor provides passive peaking and optical pulse-width compensation.

Figure 2 shows a block diagram of the MAX3966, which comprises a reference-voltage generator, modulationcurrent generator, input buffer, prebias current generator, main output driver, complementary output driver, and LED-compensation network.

#### Reference-Voltage Generator and Temperature Compensation

The reference-voltage generator circuit provides the supply-referenced  $V_{BB}$  reference voltage and a reference voltage for the modulation-current generator.

A PECL-compatible V<sub>BB</sub> generator is provided to simplify operation with a single-ended data input. For single-ended operation, connect the input signal to IN+, and connect V<sub>BB</sub> directly to IN-. The V<sub>BB</sub> circuit is not designed to drive other external circuitry. The V<sub>BB</sub> output is implemented with an NPN emitter follower.

The reference-voltage generator circuit provides two voltage sources that create modulation-current temperature compensation. A positive modulation-current

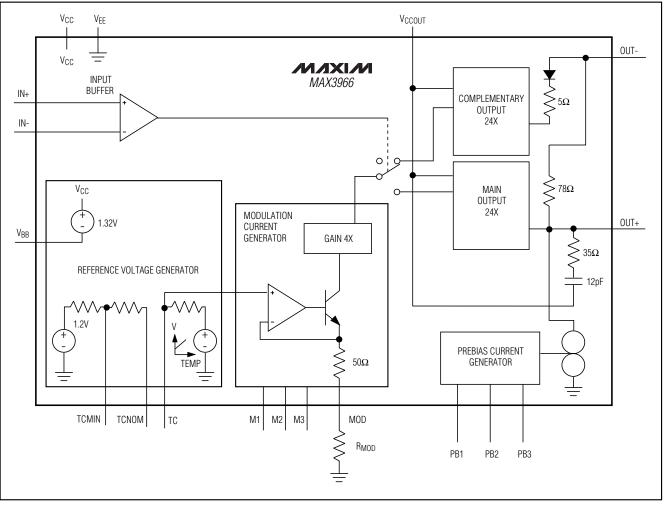


Figure 2. Functional Diagram

temperature coefficient (tempco) is useful to compensate for the temperature characteristics of typical fiber optic LEDs. The first source has a temperaturestable output. The second source has a temperatureincreasing output with a tempco of approximately 12,000ppm/°C (relative to +25°C). A resistor divider between the two reference generators programs the modulation-current tempco. For maximum modulationcurrent tempco, leave the TC pin disconnected. For a tempco of approximately 3600ppm/°C, connect TC to TCNOM. To obtain the minimum tempco, connect TCMIN to TC. Intermediate tempco values can be programmed by connecting an external resistor (RTC) between TCMIN and TC.

Input Buffer

The inputs are connected to the PECL-compatible differential input buffer. If left unconnected, IN+ is internally pulled to a PECL low and IN- is pulled to a PECL high, causing low light output. The input impedance of IN+ and IN- is approximately  $50k\Omega$ .

#### **Modulation-Current Generator**

The modulation-current generator circuit provides control of the modulation-current amplitude. This amplitude is determined by the voltage at the MOD pin and external resistor  $R_{MOD}$ . Pins M1, M2, and M3 can be selectively connected to the positive supply to provide fine adjustment of the modulation current. Table 2 describes the functions of the M1, M2, and M3 pins.

Do not connect bypass capacitors at the MOD pin. Capacitance at this pin increases high-frequency output noise.

#### **Prebias Current Generator**

A prebias voltage (V<sub>PREBIAS</sub>) can be applied to the LED to improve switching speed. The prebias current generator creates a current that flows through the  $78\Omega$  prebias resistor in the output stage, creating a prebias

Table 1	I. LEC	Prebias	Voltage
---------	--------	---------	---------

PB1	PB2	PB3	PREBIAS (V)
Open	Open	Open	0.400
VEE	Open	Open	0.475
Open	VEE	Open	0.550
VEE	VEE	Open	0.625
Open	Open	VEE	0.700
V <sub>EE</sub>	Open	V <sub>EE</sub>	0.775
Open	VEE	VEE	0.850
VEE	VEE	VEE	0.925

voltage. The prebias voltage can be adjusted by selectively connecting pins PB1, PB2, and PB3 to V<sub>EE</sub>. Table 1 describes the functions of PB1, PB2, and PB3.

#### **Output Current Drivers**

The modulation-current reference is switched and amplified by the output stages, which are implemented with NPN current mirrors.

LED package lead inductance causes ringing and overshoot, which can be compensated with an RC filter network. The MAX3966 includes  $35\Omega$  and 12pF of internal compensation. The compensation network can be optimized by adding additional components between VCCOUT and OUT+.

The MAX3966 includes a complementary output driver, which is switched 180° out of phase with the main output. This configuration helps to maintain constant current flow from the voltage supply, reducing noise and EMI. A large diode and a 5 $\Omega$  resistor are connected in series with the negative output (OUT-) to emulate the LED load at OUT+.

#### **Peaking Current**

The prebias resistor provides peaking current to improve the LED switching speed. The peaking magnitude is given by the following equation:

$$\mathsf{PEAKING}_{\mathsf{CURRENT}} \cong \frac{\mathsf{V}_{\mathsf{LED}} - \mathsf{V}_{\mathsf{PREBIAS}}}{78\Omega}$$

The peaking amplitude is equal for rising and falling data transitions.

#### **Design Procedure**

#### Select an LED

For best performance, select a high-efficiency, lowinductance LED. LED inductance causes large voltage swings and ringing.

#### Table 2. Modulation-Current Adjustment

M1	M2	МЗ	ADJUSTMENT (%)
Open	Open	Open	-4.0
Open	Open	V <sub>CC</sub>	-3.0
Open	V <sub>CC</sub>	Open	-2.0
Open	Vcc	V <sub>CC</sub>	-1.0
Vcc	Open	Open	0
V <sub>CC</sub>	Open	V <sub>CC</sub>	1.0
Vcc	Vcc	Open	2.0
Vcc	V <sub>CC</sub>	V <sub>CC</sub>	3.0



**Program the Modulation-Current Tempco** Select a modulation-current tempco that provides nearly constant LED output power as temperature varies. For the minimum tempco, connect TCMIN to the TC pin. For a tempco of approximately 3600ppm/°C, connect TC to TCNOM and leave TCMIN unconnected. For the maximum tempco, leave TCMIN, TCNOM, and TC unconnected.

Refer to the Modulation-Current Temperature Coefficient vs. RTC Resistor graph in the *Typical Operating Characteristics* to program a custom tempco. From the graph, determine the appropriate resistor and connect it between TCMIN and TC.

For example, if an LED requires a 5000 ppm/°C tempco, choose RTC of  $10\Omega.$ 

#### **Program the Modulation Current**

Determine the required modulation current at  $T_A = +25^{\circ}$ C. Then select the appropriate value of  $R_{MOD}$  from the Modulation Current vs.  $R_{MOD}$  graph in the *Typical Operating Characteristics*, which provides the required current.

For example, to program 80mA modulation current, the graph indicates an R<sub>MOD</sub> value of 750 $\Omega$  for maximum tempco (12,000ppm/°C) and 1.1k $\Omega$  for nominal tempco (3600ppm/°C). By interpolation, choose an R<sub>MOD</sub> of 1.05k $\Omega$  for a tempco of 5000ppm/°C.

#### **Program Prebias Voltage**

Determine the LED prebias voltage that produces an acceptable trade-off between peaking current and extinction ratio. Refer to Table 1 for PB1, PB2, and PB3 settings.

#### **Layout Considerations**

For optimum performance, total load inductance should not exceed 10nH. Load inductance includes LED inductance, LED package lead inductance, and circuitboard traces. Keep the connections between the MAX3966 OUT pins and the LED as short as possible to minimize inductance.

Chip-and-wire (hybrid) technology reduces package inductance significantly, and provides the best possible performance.

Use good high-frequency layout techniques. Use a multilayer board with an uninterrupted ground plane. Power supplies should be capacitively bypassed to the ground plane with surface-mount capacitors located near the power-supply pins.

### **Applications Information**

#### **Input Terminations**

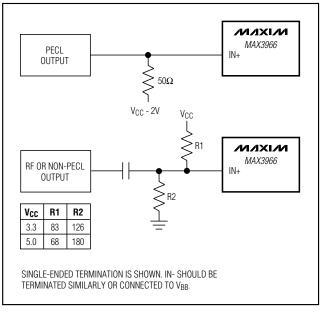
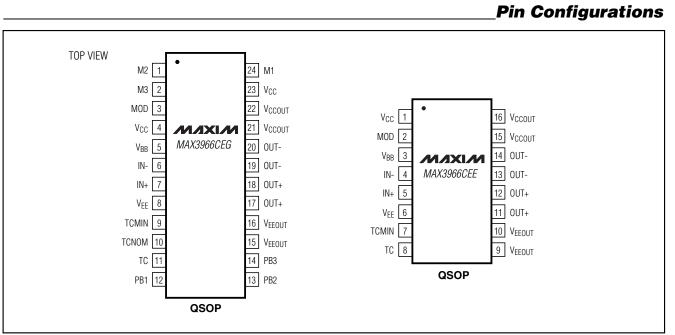


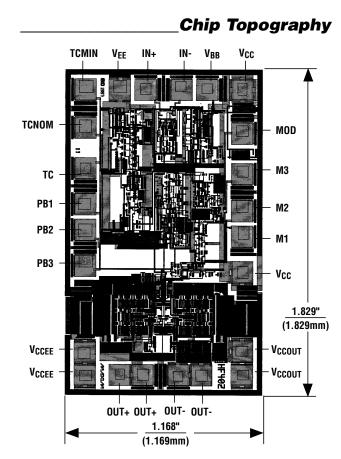
Figure 3. Input Terminations

#### Wire-Bonding Die

The MAX3966 utilizes gold metalization, which provides high reliability. Make connections to the die with gold wire only, using ball-bonding techniques. Wedge bonding is not recommended. Pad size is 4mil square (100µm). Die thickness is typically 15mil (375µm).

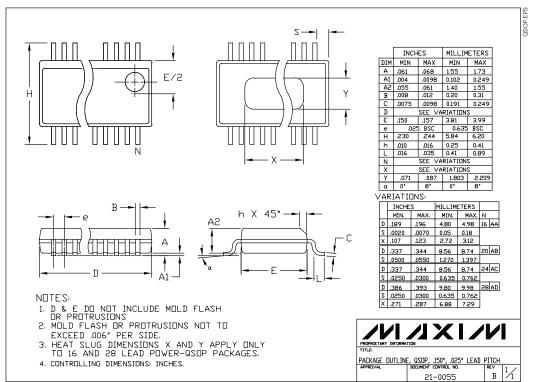


MAX3966



TRANSISTOR COUNT: 304 SUBSTRATE CONNECTED TO  $\mathsf{V}_{\mathsf{EE}}$ 





# **MAX3966**

NOTES