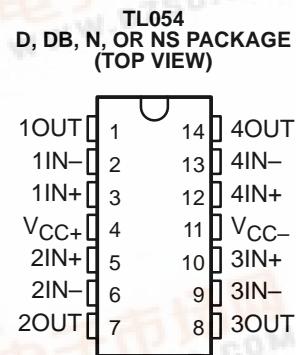
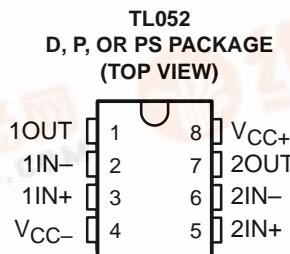
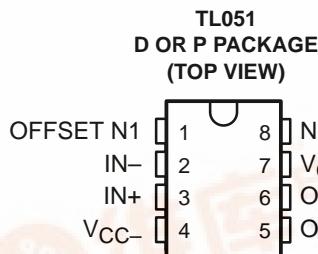


TL05x, TL05xA ENHANCED-JFET LOW-OFFSET OPERATIONAL AMPLIFIERS

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- Direct Upgrades to TL07x and TL08x BiFET Operational Amplifiers
- Faster Slew Rate (20 V/ μ s Typ) Without Increased Power Consumption
- On-Chip Offset-Voltage Trimming for Improved DC Performance and Precision Grades Are Available (1.5 mV, TL051A)



description/ordering information

The TL05x series of JFET-input operational amplifiers offers improved dc and ac characteristics over the TL07x and TL08x families of BiFET operational amplifiers. On-chip Zener trimming of offset voltage yields precision grades as low as 1.5 mV (TL051A) for greater accuracy in dc-coupled applications. Texas Instruments improved BiFET process and optimized designs also yield improved bandwidth and slew rate without increased power consumption. The TL05x devices are pin-compatible with the TL07x and TL08x and can be used to upgrade existing circuits or for optimal performance in new designs.

BiFET operational amplifiers offer the inherently higher input impedance of the JFET-input transistors, without sacrificing the output drive associated with bipolar amplifiers. This makes them better suited for interfacing with high-impedance sensors or very low-level ac signals. They also feature inherently better ac response than bipolar or CMOS devices having comparable power consumption.

The TL05x family was designed to offer higher precision and better ac response than the TL08x, with the low noise floor of the TL07x. Designers requiring significantly faster ac response or ensured lower noise should consider the Excalibur TLE208x and TLE207x families of BiFET operational amplifiers.

Because BiFET operational amplifiers are designed for use with dual power supplies, care must be taken to observe common-mode input voltage limits and output swing when operating from a single supply. DC biasing of the input signal is required, and loads should be terminated to a virtual-ground node at mid-supply. Texas Instruments TLE2426 integrated virtual ground generator is useful when operating BiFET amplifiers from single supplies.

The TL05x are fully specified at ± 15 V and ± 5 V. For operation in low-voltage and/or single-supply systems, Texas Instruments LinCMOS families of operational amplifiers (TLC-prefix) are recommended. When moving from BiFET to CMOS amplifiers, particular attention should be paid to the slew rate and bandwidth requirements, and also the output loading.

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

TL05x, TL05xA
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ORDERING INFORMATION

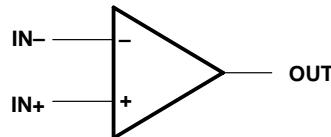
T _A	V _{I0} max AT 25°C	PACKAGE†	ORDERABLE PART NUMBER	TOP-SIDE MARKING
0°C to 70°C	800 µV	PDIP (P)	Tube of 50	TL051ACP
			TL052ACP	TL052ACP
		SOIC (D)	Tube of 75	TL051ACD
			Tube of 75	TL052ACD
			Reel of 2500	TL052ACDR
	1.5 mV	PDIP (P)	Tube of 50	TL051CP
			TL052CP	TL052CP
		PDIP (N)	Tube of 25	TL054ACN
			Tube of 75	TL051CD
		SOIC (D)	Reel of 2500	TL051CDR
			Tube of 75	TL052CD
			Reel of 2500	TL052CDR
			Tube of 50	TL054ACD
		SOP (PS)	Reel of 2500	TL054ACDR
			Reel of 2000	TL052CPSR
			Reel of 2000	TL054CDBR
	4 mV	PDIP (N)	Tube of 25	TL054CN
			Tube of 50	TL054CD
		SOIC (D)	Reel of 2500	TL054CDR
			Reel of 2000	TL054CNSR
–40°C to 85°C	800 µV	PDIP (P)	Tube of 50	TL052AIP
			Tube of 75	TL052AID
		SOIC (D)	Reel of 2500	TL052AIDR
			PDIP (N)	TL054AIN
			Tube of 25	TL054AIN
	1.5 mV	PDIP (P)	Tube of 50	TL051IP
			TL052IP	TL052IP
		SOIC (D)	Tube of 75	TL051ID
			Tube of 75	TL052ID
			Reel of 2500	TL052IDR
			Tube of 50	TL054AID
			Reel of 2500	TL054AIDR
	4 mV	PDIP (N)	Tube of 25	TL054IN
			Tube of 50	TL054ID
		SOIC (D)	Reel of 2500	TL054IDR

† Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

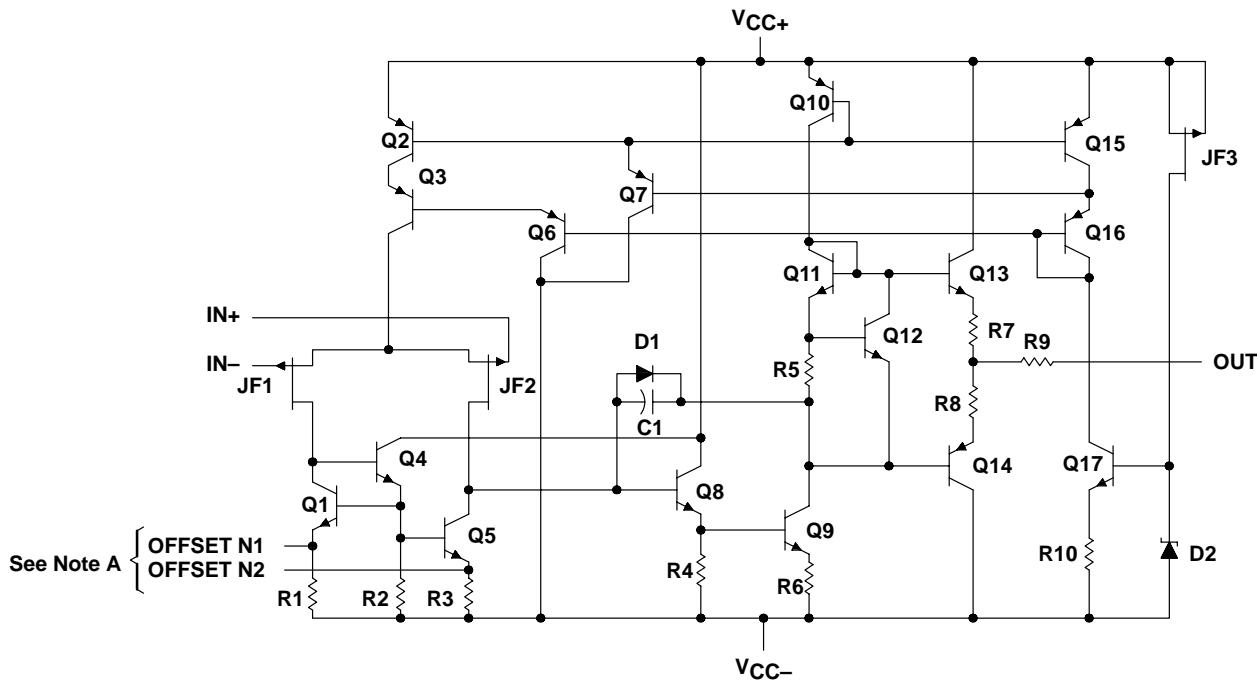
**TL05x, TL05xA
ENHANCED-JFET LOW-OFFSET
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symbol (each amplifier)



equivalent schematic (each amplifier)



NOTE A: OFFSET N1 and OFFSET N2 are available only on the TL051x.

ACTUAL DEVICE COMPONENT COUNT†			
COMPONENT	TL051	TL052	TL054
Transistors	20	34	62
Resistors	10	19	37
Diodes	2	3	5
Capacitors	1	2	4

† These figures include all four amplifiers and all ESD, bias, and trim circuitry.

TL05x, TL05xA ENHANCED-JFET LOW-OFFSET OPERATIONAL AMPLIFIERS

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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V_{CC+} (see Note 1)	18 V
Supply voltage, V_{CC-} (see Note 1)	-18 V
Differential input voltage (see Note 2)	± 30 V
Input voltage range, V_I (any input, see Notes 1 and 3)	± 15 V
Input current, I_I (each input)	± 1 mA
Output current, I_O (each output)	± 80 mA
Total current into V_{CC+}	160 mA
Total current out of V_{CC-}	160 mA
Duration of short-circuit current at (or below) 25°C	Unlimited
Package thermal impedance, θ_{JA} (see Notes 4 and 5):	
D package (8 pin)	97°C/W
D package (14 pin)	86°C/W
DB package (14 pin)	96°C/W
N package (14 pin)	80°C/W
NS package (14 pin)	76°C/W
P package (8 pin)	85°C/W
PS package (8 pin)	95°C/W
Operating virtual junction temperature, T_J	150°C
Lead temperature 1,6 mm (1/16inch) from case for 10 seconds	260°C
Storage temperature range	-65°C to 150°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between V_{CC+} and V_{CC-} .

2. Differential voltages are at IN+ with respect to IN-.
 3. The magnitude of the input voltage must never exceed the magnitude of the supply voltage or 15 V, whichever is less.
 4. Maximum power dissipation is a function of $T_J(\text{max})$, θ_{JA} , and T_A . The maximum allowable power dissipation at any allowable ambient temperature is $P_D = (T_J(\text{max}) - T_A)/\theta_{JA}$. Operating at the absolute maximum T_J of 150°C can impact reliability.
 5. The package thermal impedance is calculated in accordance with JESD 51-7.

recommended operating conditions

		C SUFFIX		I SUFFIX		UNIT	
		MIN	MAX	MIN	MAX		
V _{CC±}	Supply voltage	±5	±15	±5	±15	V	
V _{IC}	Common-mode input voltage	V _{CC±} = ±5 V		-1	4	-1	4
		V _{CC±} = ±15 V		-11	11	-11	11
T _A	Operating free-air temperature	0	70	-40	85	°C	

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TL051C and TL051AC electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T _A [†]	TL051C, TL051AC						UNIT	
			V _{CC} [‡] = ±5 V			V _{CC} [‡] = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051C	25°C	0.75	3.5	0.59	1.5		mV	
			Full range		4.5		2.5			
		TL051AC	25°C	0.55	2.8	0.35	0.8		μV/°C	
			Full range		3.8		1.8			
α _{V_{IO}} Temperature coefficient of input offset voltage [‡]		TL051C	25°C to 70°C		8		8		μV/°C	
		TL051AC	25°C to 70°C		8		8	25		
			25°C		0.04		0.04			
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C		4	100	5	100	pA	pA	
		70°C		0.02	1	0.025	1	nA		
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C		20	200	30	200	pA	pA	
		70°C		0.15	4	0.2	4	nA		
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6		-11 to 11	-12.3 to 15.6		V	
		Full range	-1 to 4			-11 to 11				
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2		13	13.9		V	
		Full range	3			13				
	R _L = 2 kΩ	25°C	2.5	3.8		11.5	12.7			
		Full range	2.5			11.5				
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5		-12	-13.2		V	
		Full range	-2.5			-12				
	R _L = 2 kΩ	25°C	-2.3	-3.2		-11	-12			
		Full range	-2.3			-11				
AVD Large-signal differential voltage amplification [¶]	R _L = 2 kΩ	25°C	25	59		50	105		V/mV	
		0°C	30	65		60	129			
		70°C	20	46		30	85			
r _i Input resistance		25°C		10 ¹²		10 ¹²		Ω		
c _i Input capacitance		25°C		10		12		pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICR} _{min} , V _O = 0, R _S = 50 Ω	25°C	65	85		75	93		dB	
		0°C	65	84		75	92			
		70°C	65	84		75	91			
k _{SVR} Supply-voltage rejection ratio ($\Delta V_{CC} \pm \Delta V_{IO}$)	V _O = 0, R _S = 50 Ω	25°C	75	99		75	99		dB	
		0°C	75	98		75	98			
		70°C	75	97		75	97			
I _{CC} Supply current	V _O = 0, No load	25°C		2.6	3.2	2.7	3.2		mA	
		0°C		2.7	3.2	2.8	3.2			
		70°C		2.6	3.2	2.7	3.2			

[†] Full range is 0°C to 70°C.

[‡] This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

[§] Typical values are based on the input offset-voltage shift observed through 168 hours of operating life test at T_A = 150°C, extrapolated to T_A = 25°C using the Arrhenius equation, and assuming an activation energy of 0.96 eV.

[¶] For V_{CC}[‡] = ±5 V, V_O = ±2.3 V, or for V_{CC}[‡] = ±15 V, V_O = ±10 V.

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TL051C and TL051AC operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T_A^\dagger	TL051C, TL051AC						UNIT	
			$V_{CC\pm} = \pm 5 \text{ V}$			$V_{CC\pm} = \pm 15 \text{ V}$				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	Positive slew rate at unity gain‡	$R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figure 1	25°C	16	20	13	20	22.6	V/ μ s	
			Full range	16.4		11	22.6			
			25°C	15	18	13	18			
			Full range	16		11	19.3			
t _r	Rise time	$V_I(\text{PP}) = \pm 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figures 1 and 2	25°C	55	56	55	56		ns	
			0°C	54		54	55			
			70°C	63		63	63			
			25°C	55	57	55	57			
t _f	Fall time	$V_I(\text{PP}) = \pm 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figures 1 and 2	0°C	54	56	54	56			
			70°C	62		62	64			
			25°C	24	19	24	19			
			0°C	24		24	19			
Overshoot factor			70°C	24		24	19		%	
			25°C	75		75				
			0°C	18		18	30			
			70°C	4		4				
V_n	Equivalent input noise voltage§	$R_S = 20 \Omega$, See Figure 3	$f = 10 \text{ Hz}$	25°C		75			nV/ $\sqrt{\text{Hz}}$	
$V_{N(\text{PP})}$	Peak-to-peak equivalent input noise voltage		$f = 1 \text{ kHz}$	25°C		18				
I_n	Equivalent input noise current		$f = 10 \text{ Hz}$ to 10 kHz	25°C		4				
THD	Total harmonic distortion¶	$R_S = 1 \text{ k}\Omega$, $f = 1 \text{ kHz}$	$R_L = 2 \text{ k}\Omega$	25°C	0.003		0.003		%	
B ₁	Unity-gain bandwidth	$V_I = 10 \text{ mV}$, $C_L = 25 \text{ pF}$, See Figure 4	$R_L = 2 \text{ k}\Omega$	25°C	3	3.1			MHz	
				0°C	3.2	3.3				
				70°C	2.7	2.8				
ϕ_m	Phase margin at unity gain	$V_I = 10 \text{ mV}$, $C_L = 25 \text{ pF}$, See Figure 4	$R_L = 2 \text{ k}\Omega$	25°C	59	62			deg	
				0°C	58	62				
				70°C	59	62				

† Full range is 0°C to 70°C.

‡ For $V_{CC\pm} = \pm 5 \text{ V}$, $V_I(\text{PP}) = \pm 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_I(\text{PP}) = \pm 5 \text{ V}$.

§ This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

¶ For $V_{CC\pm} = \pm 5 \text{ V}$, $V_O(\text{RMS}) = 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_O(\text{RMS}) = 6 \text{ V}$.

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TL051I and TL051AI electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	TA [†]	TL051I, TL051AI						UNIT	
			V _{CC} [‡] = ±5 V			V _{CC} [‡] = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO}	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL051I	25°C	0.75	3.5	0.59	1.5		mV	
			Full range		5.3		3.3			
		TL051AI	25°C	0.55	2.8	0.35	0.8			
			Full range		4.6		2.6			
		TL051I	25°C to 85°C		7		8		μV/°C	
			TL051AI	25°C to 85°C		8		8 25		
α _{V_{IO}}	Temperature coefficient of input offset voltage [‡]		25°C		0.04		0.04		μV/mo	
		V _O = 0, V _{IC} = 0, See Figure 5	25°C	4	100	5	100	pA	pA	
			85°C	0.06	10	0.07	10	nA		
I _{IO}	Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C	20	200	30	200	pA	pA	
			85°C	0.6	20	0.7	20	nA		
V _{ICR}	Common-mode input voltage range		25°C	-1	-2.3	-11	-12.3		V	
				to	to	to	to			
			Full range	4	5.6	11	15.6			
				-1		-11				
V _{OM} +	Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2	13	13.9		V	
			Full range	3		13				
		R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7			
			Full range	2.5		11.5				
V _{OM} -	Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2		V	
			Full range	-2.5		-12				
		R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12			
			Full range	-2.3		-11				
AVD	Large-signal differential voltage amplification [¶]	R _L = 2 kΩ	25°C	25	59	50	105		V/mV	
			-40°C	30	74	60	145			
			85°C	20	43	30	76			
r _i	Input resistance		25°C		10 ¹²		10 ¹²	Ω		
c _i	Input capacitance		25°C		10		12	pF		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin} , V _O = 0, R _S = 50 Ω	25°C	65	85	75	93		dB	
			-40°C	65	83	75	90			
			85°C	65	84	75	93			
k _{SVR}	Supply-voltage rejection ratio (ΔV _{CC} [‡] /ΔV _{IO})	V _O = 0, R _S = 50 Ω	25°C	75	99	75	99		dB	
			-40°C	75	98	75	98			
			85°C	75	99	75	99			
I _{CC}	Supply current	V _O = 0, No load	25°C		2.6 3.2	2.7	3.2		mA	
			-40°C		2.4 3.2	2.6	3.2			
			85°C		2.5 3.2	2.6	3.2			

[†] Full range is -40°C to 85°C

[‡] This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or non-testing of other parameters.

[§] Typical values are based on the input offset-voltage shift observed through 168 hours of operating life test at T_A = 150°C, extrapolated to T_A = 25°C using the Arrhenius equation, and assuming an activation energy of 0.96 eV.

[¶] For V_{CC}[‡] = ±5 V, V_O = ±2.3 V, or for V_{CC}[‡] = ±15 V, V_O = ±10 V.

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TL051I and TL051AI operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T_A^\dagger	TL051I, TL051AI						UNIT	
			$V_{CC\pm} = \pm 5 \text{ V}$			$V_{CC\pm} = \pm 15 \text{ V}$				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	$R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figure 1	25°C	16			13	20		V/ μ s	
		Full range				11				
SR-		25°C	15			13	18			
		Full range				11				
t_r Rise time	$V_I(\text{PP}) = \pm 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figures 1 and 2	25°C	55			56			ns	
		-40°C	52			53				
		85°C	64			65				
t_f Fall time		25°C	55			57				
		-40°C	51			53				
		85°C	64			65				
Overshoot factor		25°C	24			19			%	
		-40°C	24			19				
		85°C	24			19				
V_n Equivalent input noise voltage \S	$R_S = 20 \Omega$, See Figure 3	$f = 10 \text{ Hz}$	25°C	75		75			nV/ $\sqrt{\text{Hz}}$	
		$f = 1 \text{ kHz}$	25°C	18		18	30			
$V_{N(\text{PP})}$ Peak-to-peak equivalent input noise voltage		$f = 10 \text{ Hz}$ to 10 kHz	25°C	4		4			μV	
I_n Equivalent input noise current	$f = 1 \text{ kHz}$		25°C	0.01		0.01			pA/ $\sqrt{\text{Hz}}$	
THD Total harmonic distortion \ddagger	$R_S = 1 \text{ k}\Omega$, $f = 1 \text{ kHz}$	$R_L = 2 \text{ k}\Omega$,	25°C	0.003		0.003			%	
B ₁ Unity-gain bandwidth	$V_I = 10 \text{ mV}$, $C_L = 25 \text{ pF}$, See Figure 4	$R_L = 2 \text{ k}\Omega$,	25°C	3		3.1			MHz	
			-40°C	3.5		3.6				
			85°C	2.6		2.7				
ϕ_m Phase margin at unity gain	$V_I = 10 \text{ mV}$, $C_L = 25 \text{ pF}$, See Figure 4	$R_L = 2 \text{ k}\Omega$,	25°C	59		62			deg	
			-40°C	58		61				
			85°C	59		62				

\dagger Full range is -40°C to 85°C.

\ddagger For $V_{CC\pm} = \pm 5 \text{ V}$, $V_I(\text{PP}) = \pm 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_I(\text{PP}) = \pm 5 \text{ V}$.

\S This parameter is tested on a sample basis for the TL051A. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

\ddagger For $V_{CC\pm} = \pm 5 \text{ V}$, $V_O(\text{RMS}) = 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_O(\text{RMS}) = 6 \text{ V}$.

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TL052C and TL052AC electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	TA [†]	TL052C, TL052AC						UNIT	
			V _{CC} [‡] = ±5 V			V _{CC} [‡] = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL052C	25°C	0.73	3.5	0.65	1.5		mV	
			Full range		4.5		2.5			
		TL052AC	25°C	0.51	2.8	0.4	0.8			
			Full range		3.8		1.8			
α _{V_{IO}} Temperature coefficient of input offset voltage [‡]		TL052C	25°C to 70°C		8		8		μV/°C	
		TL052AC	25°C to 70°C		8		6	25		
Input offset-voltage long-term drift [§]	V _O = 0, R _S = 50 Ω	V _{IC} = 0,	25°C		0.04		0.04		μV/mo	
I _{IO} Input offset current	V _O = 0, See Figure 5	V _{IC} = 0,	25°C	4	100	5	100	pA		
			70°C	0.02	1	0.025	1	nA		
I _{IB} Input bias current	V _O = 0, See Figure 5	V _{IC} = 0,	25°C	20	200	30	200	pA		
			70°C	0.15	4	0.2	4	nA		
V _{ICR} Common-mode input voltage range			25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6		V	
			Full range	-1 to 4		-11 to 11				
		R _L = 10 kΩ	25°C	3	4.2	13	13.9			
			Full range	3		13				
V _{OM+} Maximum positive peak output voltage swing		R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7		V	
			Full range	2.5		11.5				
		R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2			
			Full range	-2.5		-12				
V _{OM-} Maximum negative peak output voltage swing		R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12		V	
			Full range	-2.3		-11				
		R _L = 2 kΩ	25°C	25	59	50	105		V/mV	
			0°C	30	65	60	129			
A _{VD} Large-signal differential voltage amplification [¶]		R _L = 2 kΩ	70°C	20	46	30	85			
			25°C		10 ¹²		10 ¹²	Ω		
r _i Input resistance			25°C		10 ¹²		10 ¹²	Ω		
c _i Input capacitance			25°C		10		12	pF		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin} , V _O = 0, R _S = 50 Ω		25°C	65	85	75	93		dB	
			0°C	65	84	75	92			
			70°C	65	84	75	91			

[†] Full range is 0°C to 70°C.

[‡] This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

[§] Typical values are based on the input offset-voltage shift observed through 168 hours of operating life test at T_A = 150°C, extrapolated to T_A = 25°C using the Arrhenius equation, and assuming an activation energy of 0.96 eV.

[¶] For V_{CC}[‡] = ±5 V, V_O = ±2.3 V; at V_{CC}[‡] = ±15 V, V_O = ±10 V.

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TL052C and TL052AC electrical characteristics at specified free-air temperature (continued)

PARAMETER	TEST CONDITIONS	TA	TL052C, TL052AC						UNIT	
			V _{CC} _± = ±5 V			V _{CC} _± = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
k _{SVR}	Supply-voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	99	dB	
			0°C	75	98	75	98	98		
			70°C	75	97	75	97	97		
I _{CC}	Supply current (two amplifiers)	V _O = 0, No load	25°C	4.6	5.6	4.8	5.6	5.6	mA	
			0°C	4.7	6.4	4.8	6.4	6.4		
			70°C	4.4	6.4	4.6	6.4	6.4		
V _{O1} /V _{O2}	Crosstalk attenuation	A _{VD} = 100	25°C	120	120	120	120	120	dB	

TL052C and TL052AC operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	TA [†]	TL052C, TL052AC						UNIT	
			V _{CC} _± = ±5 V			V _{CC} _± = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	Slew rate at unity gain	R _L = 2 kΩ, C _L = 100 pF, See Figure 1	25°C	17.8	20.7	9	20.7	20.7	V/μs	
			Full range			8		8		
SR-	Negative slew rate at unity gain [‡]		25°C	15.4	17.8	9	17.8	17.8		
			Full range			8		8		
t _r	Rise time	V _I (PP) = ±10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C	55	56	56	56	56	ns	
			0°C	54	55	55	55	55		
			70°C	63	63	63	63	63		
t _f	Fall time		25°C	55	57	57	57	57	ns	
			0°C	54	56	56	56	56		
			70°C	62	64	64	64	64		
Overshoot factor			25°C	24	19	19	19	19	%	
			0°C	24	19	19	19	19		
			70°C	24	19	19	19	19		
V _n	Equivalent input noise voltage [§]	R _S = 20 Ω, See Figure 3	f = 10 Hz	25°C	71	71	71	71	nV/√Hz	
			f = 1 kHz	25°C	19	19	19	30		
V _{N(PP)}	Peak-to-peak equivalent input noise current		f = 10 Hz to 10 kHz	25°C	4	4	4	4	μV	
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.01	0.01	0.01	0.01	pA/√Hz	
THD	Total harmonic distortion [¶]	R _S = 1 kΩ, f = 1 kHz	R _L = 2 kΩ,	25°C	0.003	0.003	0.003	0.003	%	
B ₁	Unity-gain bandwidth	V _I = 10 mV, C _L = 25 pF, See Figure 4	25°C	3	3	3	3	3	MHz	
			0°C	3.2	3.2	3.2	3.2	3.2		
			70°C	2.6	2.7	2.7	2.7	2.7		
φ _m	Phase margin at unity gain	V _I = 10 mV, C _L = 25 pF, See Figure 4	25°C	60	63	63	63	63	deg	
			0°C	59	63	63	63	63		
			70°C	60	63	63	63	63		

[†] Full range is 0°C to 70°C.

[‡] For V_{CC}_± = ±5 V, V_I(PP) = ±1 V; for V_{CC}_± = ±15 V, V_I(PP) = ±5 V.

[§] This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

[¶] For V_{CC}_± = ±5 V, V_O(RMS) = 1 V; for V_{CC}_± = ±15 V, V_O(RMS) = 6 V.

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TL052I and TL052AI electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T_A^\dagger	TL052I, TL052AI						UNIT	
			$V_{CC\pm} = \pm 5 \text{ V}$			$V_{CC\pm} = \pm 15 \text{ V}$				
			MIN	TYP	MAX	MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_O = 0,$ $V_{IC} = 0,$ $R_S = 50 \Omega$	TL052I	25°C	0.73	3.5	0.65	1.5		mV	
			Full range		5.3			3.3		
		TL052AI	25°C	0.51	2.8	0.4	0.8			
			Full range		4.6			2.6		
$\alpha_{V_{IO}}$ Temperature coefficient‡		TL052I	25°C to 85°C		7			6	$\mu\text{V}/^\circ\text{C}$	
		TL052AI	25°C to 85°C		6			6 25		
I_{IO} Input offset current	$V_O = 0,$ $R_S = 50 \Omega$	$V_{IC} = 0,$ See Figure 5	25°C	0.04		0.04			μA	
I_{IB} Input bias current	$V_O = 0,$ See Figure 5	$V_{IC} = 0,$ See Figure 5	25°C	20	200	30	200		pA	
			85°C	0.6	20	0.7	20		nA	
V_{ICR} Common-mode input voltage range			25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6		V	
			Full range	-1 to 4		-11 to 11				
		$R_L = 10 \text{ k}\Omega$	25°C	3	4.2	13	13.9			
			Full range	3		13				
V_{OM+} Maximum positive peak output voltage swing		$R_L = 2 \text{ k}\Omega$	25°C	2.5	3.8	11.5	12.7		V	
			Full range	2.5		11.5				
		$R_L = 10 \text{ k}\Omega$	25°C	-2.5	-3.5	-12	-13.2			
			Full range	-2.5		-12				
V_{OM-} Maximum negative peak output voltage swing		$R_L = 2 \text{ k}\Omega$	25°C	-2.3	-3.2	-11	-12		V	
			Full range	-2.3		-11				
		$R_L = 2 \text{ k}\Omega$	25°C	25	59	50	105		V/mV	
			-40°C	30	74	60	145			
A_{VD} Large-signal differential voltage amplification¶	$R_L = 2 \text{ k}\Omega$	85°C	20	43		30	76			
		25°C		10 ¹²		10 ¹²			Ω	
		25°C		10		12				
$CMRR$ Common-mode rejection ratio	$V_{IC} = V_{ICR\min},$ $V_O = 0,$ $R_S = 50 \Omega$	25°C	65	85		75	93		dB	
		-40°C	65	83		75	90			
		85°C	65	84		75	93			

† Full range is -40°C to 85°C.

‡ This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

§ Typical values are based on the input offset-voltage shift observed through 168 hours of operating life test at $T_A = 150^\circ\text{C}$, extrapolated to $T_A = 25^\circ\text{C}$ using the Arrhenius equation, and assuming an activation energy of 0.96 eV.

¶ At $V_{CC\pm} = \pm 5 \text{ V}$, $V_O = \pm 2.3 \text{ V}$; at $V_{CC\pm} = \pm 15 \text{ V}$, $V_O = \pm 10 \text{ V}$.

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TL052I and TL052AI electrical characteristics at specified free-air temperature (continued)

PARAMETER	TEST CONDITIONS	TA	TL052I, TL052AI						UNIT	
			V _{CC} ± = ±5 V			V _{CC} ± = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
k _{SVR}	Supply-voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	99	dB	
			-40°C	75	98	75	98	98		
			85°C	75	99	75	99	99		
I _{CC}	Supply current (two amplifiers)	V _O = 0, No load	25°C	4.6	5.6	4.8	5.6	5.6	mA	
			-40°C	4.5	6.4	4.7	6.4	6.4		
			85°C	4.4	6.4	4.6	6.4	6.4		
V _{O1} /V _{O2}	Crosstalk attenuation	A _{VD} = 100	25°C	120	120	120	120	120	dB	

TL052I and TL052AI operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	TA†	TL052I, TL052AI						UNIT	
			V _{CC} ± = ±5 V			V _{CC} ± = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	Slew rate at unity gain‡	R _L = 2 kΩ, C _L = 100 pF, See Figure 1	25°C	17.8	20.7	9	20.7	20.7	V/μs	
			Full range			8				
SR-	Negative slew rate at unity gain‡		25°C	15.4	17.8	9	17.8	17.8		
			Full range			8				
t _r	Rise time	V _I (PP) = ±10 mV, R _L = 2 kΩ, C _L = 100 pF, See Figures 1 and 2	25°C	55	56	56	56	56	ns	
			-40°C	52	53	53	53	53		
			85°C	64	65	65	65	65		
t _f	Fall time		25°C	55	57	57	57	57		
			-40°C	51	53	53	53	53		
			85°C	64	65	65	65	65		
Overshoot factor			25°C	24%	19%	19%	19%	19%	%	
			-40°C	24%	19%	19%	19%	19%		
			85°C	24%	19%	19%	19%	19%		
V _n	Equivalent input noise voltage§	R _S = 20 Ω, See Figure 3	f = 10 Hz	25°C	71	71	71	71	nV/√Hz	
			f = 1 kHz	25°C	19	19	19	30		
V _{N(PP)}	Peak-to-peak equivalent input noise current		f = 10 Hz to 10 kHz	25°C	4	4	4	4	μV	
I _n	Equivalent input noise current	f = 1 kHz		25°C	0.01	0.01	0.01	0.01	pA/√Hz	
THD	Total harmonic distortion¶	R _S = 1 kΩ, R _L = 2 kΩ, f = 1 kHz		25°C	0.003	0.003	0.003	0.003	%	
B ₁	Unity-gain bandwidth	V _I = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C	3	3	3	3	3	MHz	
			-40°C	3.5	3.6	3.6	3.6	3.6		
			85°C	2.5	2.6	2.6	2.6	2.6		
φ _m	Phase margin at unity gain	V _I = 10 mV, R _L = 2 kΩ, C _L = 25 pF, See Figure 4	25°C	60	63	63	63	63	deg	
			-40°C	58	61	61	61	61		
			85°C	60	63	63	63	63		

† Full range is -40°C to 85°C.

‡ For V_{CC}± = ±5 V, V_I(PP) = ±1 V; for V_{CC}± = ±15 V, V_I(PP) = ±5 V.

§ This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

¶ For V_{CC}± = ±5 V, V_O(RMS) = 1 V; for V_{CC}± = ±15 V, V_O(RMS) = 6 V.

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TL054C and TL054AC electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T _A [†]	TL054C, TL054AC						UNIT	
			V _{CC±} = ±5 V			V _{CC±} = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054C	25°C	0.64	5.5	0.56	4	4	mV	
			Full range		7.7			6.2		
		TL054AC	25°C	0.57	3.5	0.5	1.5	1.5		
			Full range		5.7			3.7		
		TL054C	25°C to 70°C		25		23	23	μV/°C	
			TL054AC	25°C to 70°C		24		23		
α _{V_{IO}} Temperature coefficient of input offset voltage			25°C		0.04		0.04	0.04	μV/mo	
		V _O = 0, V _{IC} = 0, See Figure 5	25°C	4	100	5	100	100	pA	
			70°C	0.02	1	0.025	1	1	nA	
		V _O = 0, V _{IC} = 0, See Figure 5	25°C	20	200	30	200	200	pA	
			70°C	0.15	4	0.2	4	4	nA	
			25°C	-1 to 4	-2.3 to 5.6	-11 to 11	-12.3 to 15.6	-12.3	V	
V _{ICR} Common-mode input voltage range			Full range	-1 to 4	-1 to 4	-11 to 11	-11 to 11	-11		
		R _L = 10 kΩ	25°C	3	4.2	13	13.9	13.9	V	
			Full range	3		13				
		R _L = 2 kΩ	25°C	2.5	3.8	11.5	12.7	12.7		
			Full range	2.5		11.5				
V _{OM+} Maximum positive peak output voltage swing		R _L = 10 kΩ	25°C	-2.5	-3.5	-12	-13.2	-13.2	V	
			Full range	-2.5		-12				
		R _L = 2 kΩ	25°C	-2.3	-3.2	-11	-12	-12		
			Full range	-2.3		-11				
		R _L = 2 kΩ	25°C	25	72	50	133	133	V/mV	
			0°C	30	88	60	173	173		
			70°C	20	57	30	85	85		
r _i Input resistance			25°C		10 ¹²		10 ¹²	10 ¹²	Ω	
c _i Input capacitance			25°C		10		12	12	pF	
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin} , V _O = 0, R _S = 50 Ω	25°C	65	84	75	92	92	92	dB	
		0°C	65	84	75	92	92	92		
		70°C	65	84	75	93	93	93		
k _{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	V _{CC±} = ±5 V to ±15 V, V _O = 0, R _S = 50 Ω	25°C	75	99	75	99	99	99	dB	
		0°C	75	99	75	99	99	99		
		70°C	75	99	75	99	99	99		
I _{CC} Supply current (four amplifiers)	V _O = 0, No load	25°C		8.1	11.2	8.4	11.2	11.2	mA	
		0°C		8.2	12.8	8.5	12.8	12.8		
		70°C		7.9	11.2	8.2	11.2	11.2		
V _{O1} /V _{O2} Crosstalk attenuation	A _{VD} = 100	25°C		120		120		120	dB	

[†] Full range is 0°C to 70°C.

[‡] Typical values are based on the input offset-voltage shift observed through 168 hours of operating life test at T_A = 150°C, extrapolated to T_A = 25°C using the Arrhenius equation, and assuming an activation energy of 0.96 eV.

[§] For V_{CC±} = ±5 V, V_O = ±2.3 V, at V_{CC±} = ±15 V, V_O = ±10 V.B

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TL054C and TL054AC operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T_A^\dagger	TL054C, TL054C						UNIT	
			$V_{CC\pm} = \pm 5 \text{ V}$			$V_{CC\pm} = \pm 15 \text{ V}$				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	$R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figure 1 and Note 7	25°C	15.4			10	17.8		V/ μ s	
		0°C	15.7			8	17.9			
		70°C	14.4			8	17.5			
		25°C	13.9			10	15.9			
		0°C	14.3			8	16.1			
		70°C	13.3			8	15.5			
t_r		25°C	55			56			ns	
		0°C	54			55				
		70°C	63			63				
		25°C	55			57				
		0°C	54			56				
		70°C	62			64				
Overshoot factor	$V_I(\text{PP}) = \pm 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 100 \text{ pF}$, See Figures 1 and 2	25°C	24%			19%			%	
		0°C	24%			19%				
		70°C	24%			19				
		25°C	75			75				
		25°C	21			21	45			
		25°C	4			4				
I_n	$f = 10 \text{ Hz}$	25°C	0.01			0.01			$\text{pA}/\sqrt{\text{Hz}}$	
		25°C	0.003			0.003				
		25°C	2.7			2.7				
		0°C	3			3				
		70°C	2.4			2.4				
		25°C	61			64				
B_1	$V_I = 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 25 \text{ pF}$, See Figure 4	25°C	60			64			MHz	
		0°C	61			63				
		70°C	61			63				
		25°C	61			64				
		0°C	60			64				
		70°C	61			63				
ϕ_m	$V_I = 10 \text{ mV}$, $R_L = 2 \text{ k}\Omega$, $C_L = 25 \text{ pF}$, See Figure 4	25°C	61			64			deg	
		0°C	60			64				
		70°C	61			63				

[†] Full range is 0°C to 70°C.

[‡] For $V_{CC\pm} = \pm 5 \text{ V}$, $V_I(\text{PP}) = \pm 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_I(\text{PP}) = \pm 5 \text{ V}$.

[§] This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

[¶] For $V_{CC\pm} = \pm 5 \text{ V}$, $V_O(\text{RMS}) = 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_O(\text{RMS}) = 6 \text{ V}$.

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TL054I and TL054AI electrical characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T _A †	TL054I, TL054AI						UNIT	
			V _{CC} ± = ±5 V			V _{CC} ± = ±15 V				
			MIN	TYP	MAX	MIN	TYP	MAX		
V _{IO} Input offset voltage	V _O = 0, V _{IC} = 0, R _S = 50 Ω	TL054I	25°C	0.64	5.5	0.56	4	4	mV	
			Full range		8.8		7.3			
		TL054AI	25°C	0.57	3.5	0.5	1.5	1.5		
			Full range		6.8		4.8			
α _{V_{IO}} Temperature coefficient of input offset voltage		TL054I	25°C to 85°C		25		24	24	μV/°C	
		TL054AI	25°C to 85°C		25		23	23		
			25°C		0.04		0.04	0.04		
I _{IO} Input offset current	V _O = 0, V _{IC} = 0, See Figure 5	25°C		4	100		5	100	pA	
		85°C		0.06	10		0.07	10	nA	
I _{IB} Input bias current	V _O = 0, V _{IC} = 0, See Figure 5	25°C		20	200		30	200	pA	
		85°C		0.6	20		0.7	20	nA	
V _{ICR} Common-mode input voltage range		25°C	-1 to 4	-2.3 to 5.6		-11 to 11	-12.3 to 15.6	-12.3	V	
			Full range	-1 to 4		-11 to 11	-12.3 to 15.6	-12.3		
V _{OM+} Maximum positive peak output voltage swing	R _L = 10 kΩ	25°C	3	4.2		13	13.9	13.9	V	
		Full range	3			13		13		
	R _L = 2 kΩ	25°C	2.5	3.8		11.5	12.7	12.7		
		Full range	2.5			11.5		11.5		
V _{OM-} Maximum negative peak output voltage swing	R _L = 10 kΩ	25°C	-2.5	-3.5		-12	-13.2	-13.2	V	
		Full range	-2.5			-12		-12		
	R _L = 2 kΩ	25°C	-2.3	-3.2		-11	-12	-12		
		Full range	-2.3			-11		-11		
AVD Large-signal differential voltage amplification§	R _L = 2 kΩ	25°C	25	72		50	133	133	V/mV	
		-40°C	30	101		60	212	212		
		85°C	20	50		30	70	70		
r _i Input resistance		25°C		10 ¹²			10 ¹²	10 ¹²	Ω	
c _i Input capacitance		25°C		10			12	12	pF	
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin} , V _O = 0, R _S = 50 Ω	25°C	65	84		75	92	92	dB	
		-40°C	65	83		75	92	92		
		85°C	65	84		75	93	93		
k _{SVR} Supply-voltage rejection ratio ($\Delta V_{CC\pm}/\Delta V_{IO}$)	V _{CC} ± = ±5 V to ±15 V, V _O = 0, R _S = 50 Ω	25°C	75	99		75	99	99	dB	
		-40°C	75	98		75	99	99		
		85°C	75	99		75	99	99		
I _{CC} Supply current (four amplifiers)	V _O = 0, No load	25°C		8.1	11.2		8.4	11.2	mA	
		-40°C		7.9	12.8		8.2	12.8		
		85°C		7.6	11.2		7.9	11.2		
V _{O1} /V _{O2} Crosstalk attenuation	AVD = 100	25°C		120			120	120	dB	

† Full range is -40°C to 85°C.

‡ Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T_A = 150°C, extrapolated to T_A = 25°C using the Arrhenius equation, and assuming an activation energy of 0.96 eV.

§ For V_{CC}± = ±5 V, V_O = ±2.3 V, at V_{CC}± = ±15 V, V_O = ±10 V.

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TL054I and TL054AI operating characteristics at specified free-air temperature

PARAMETER	TEST CONDITIONS	T_A^\dagger	TL054I, TL054AI						UNIT	
			$V_{CC\pm} = \pm 5 \text{ V}$			$V_{CC\pm} = \pm 15 \text{ V}$				
			MIN	TYP	MAX	MIN	TYP	MAX		
SR+	$R_L = 2 \text{ k}\Omega, C_L = 100 \text{ pF}$, See Figure 1	25°C	15.4		10	17.8			$\text{V}/\mu\text{s}$	
		-40°C	16.4		8	18				
		85°C	14		8	17.3				
		25°C	13.9		10	15.9				
		-40°C	14.7		8	16.1				
		85°C	13		8	15.3				
t_r	Rise time	25°C	55		56				ns	
		-40°C	52		53					
		85°C	64		65					
t_f	$V_I(\text{PP}) = \pm 10 \text{ mV}, R_L = 2 \text{ k}\Omega, C_L = 100 \text{ pF}$, See Figures 1 and 2	25°C	55		57					
		-40°C	51		53					
		85°C	64		65					
Overshoot factor		25°C	24		19				%	
		-40°C	24		19					
		85°C	24		19					
V_n	$R_S = 20 \Omega$, See Figure 3	$f = 10 \text{ Hz}$	25°C	75		75			$\text{nV}/\sqrt{\text{Hz}}$	
		$f = 1 \text{ kHz}$	25°C	21		21	45			
$V_{N(\text{PP})}$	Peak-to-peak equivalent input noise voltage	$f = 10 \text{ Hz to } 10 \text{ kHz}$	25°C	4		4			μV	
I_n	Equivalent input noise current	$f = 1 \text{ kHz}$	25°C	0.01		0.01			$\text{pA}/\sqrt{\text{Hz}}$	
THD	Total harmonic distortion \ddagger	$R_S = 1 \text{ k}\Omega, R_L = 2 \text{ k}\Omega, f = 1 \text{ kHz}$	25°C	0.003%		0.003%			%	
B_1	Unity-gain bandwidth	$V_I = 10 \text{ mV}, R_L = 2 \text{ k}\Omega, C_L = 25 \text{ pF}$, See Figure 4	25°C	2.7		2.7			MHz	
			-40°C	3.3		3.3				
			85°C	2.3		2.4				
ϕ_m	Phase margin at unity gain	$V_I = 10 \text{ mV}, R_L = 2 \text{ k}\Omega, C_L = 25 \text{ pF}$, See Figure 4	25°C	61		64			deg	
			-40°C	59		62				
			85°C	61		64				

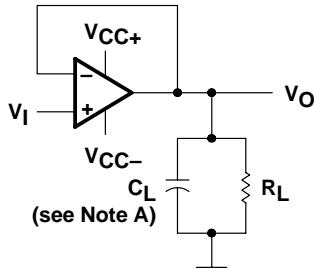
\dagger Full range is -40°C to 85°C.

\ddagger For $V_{CC\pm} = \pm 5 \text{ V}$, $V_I(\text{PP}) = \pm 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_I(\text{PP}) = \pm 5 \text{ V}$.

\S This parameter is tested on a sample basis. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

\P For $V_{CC\pm} = \pm 5 \text{ V}$, $V_O(\text{RMS}) = 1 \text{ V}$; for $V_{CC\pm} = \pm 15 \text{ V}$, $V_O(\text{RMS}) = 6 \text{ V}$.

PARAMETER MEASUREMENT INFORMATION



NOTE A: C_L includes fixture capacitance.

Figure 1. Slew Rate, Rise/Fall Time, and Overshoot Test Circuit

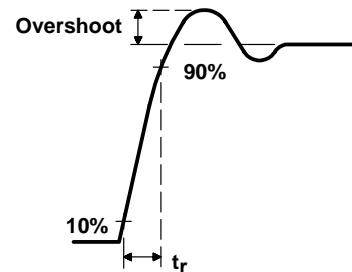


Figure 2. Rise-Time and Overshoot Waveform

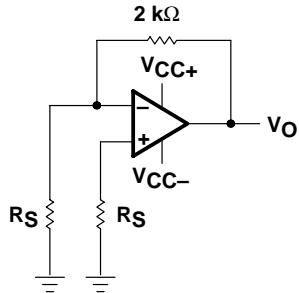
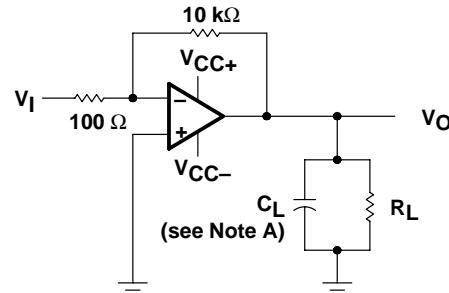


Figure 3. Noise-Voltage Test Circuit



NOTE A: C_L includes fixture capacitance.
Figure 4. Unity-Gain Bandwidth and Phase-Margin Test Circuit

typical values

Typical values, as presented in this data sheet represent the median (50% point) of device parametric performance.

input bias and offset current

At the picoamp-bias-current level typical of the TL05x and TL05xA, accurate measurement of the bias current becomes difficult. Not only does this measurement require a picoammeter, but test-socket leakages easily can exceed the actual device bias currents. To accurately measure these small currents, Texas Instruments uses a two-step process. The socket leakage is measured using picoammeters with bias voltages applied, but with no device in the socket. The device then is inserted in the socket, and a second test that measures both the socket leakage and the device input bias current is performed. The two measurements then are subtracted algebraically to determine the bias current of the device.

noise

Because of the increasing emphasis on low noise levels in many of today's applications, the input noise voltage density is sample tested at $f = 1$ kHz. Texas Instruments also has additional noise-testing capability to meet specific application requirements. Please contact the factory for details.

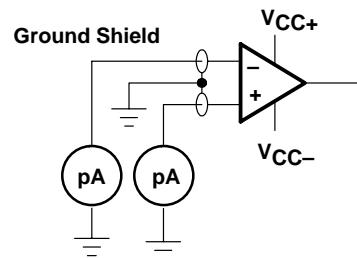


Figure 5. Input-Bias and Offset-Current Test Circuit

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TYPICAL CHARACTERISTICS

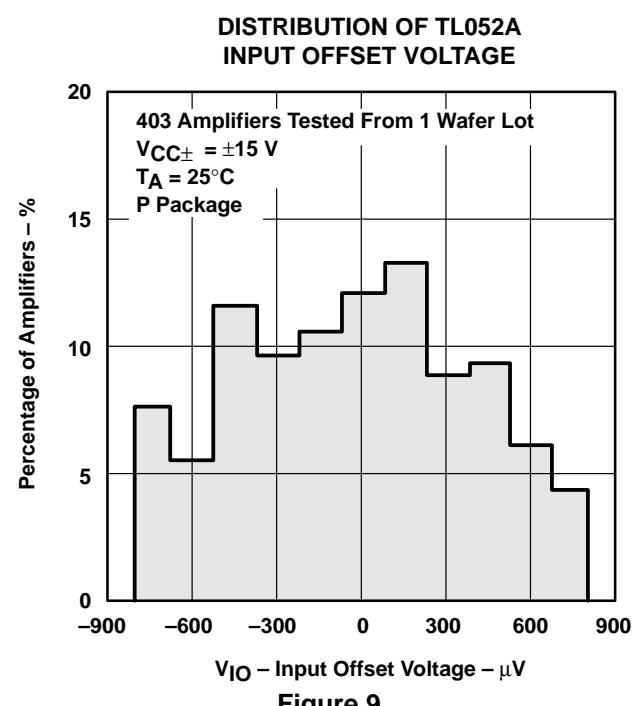
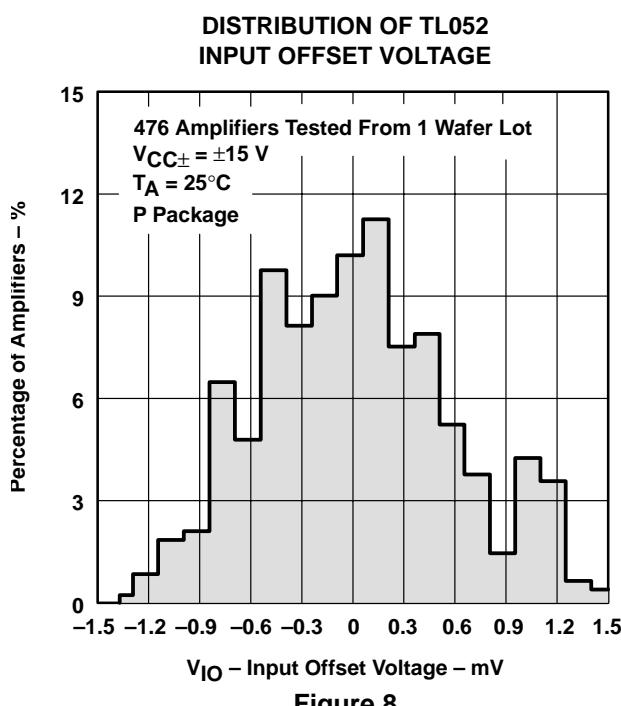
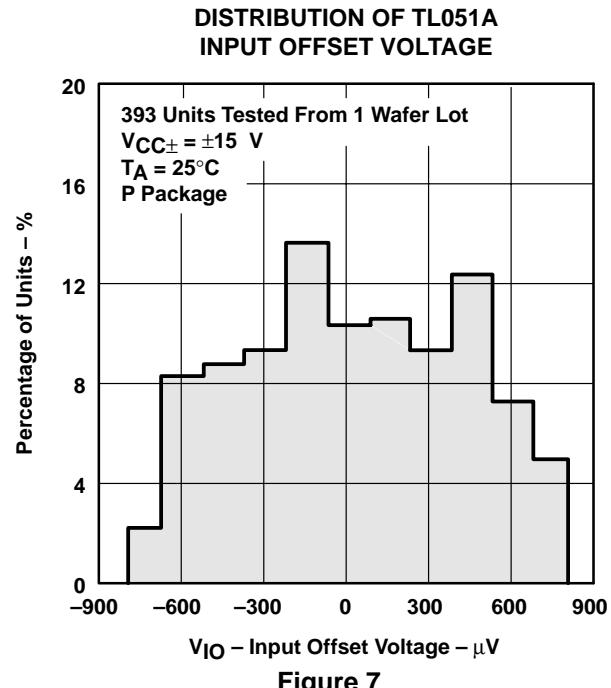
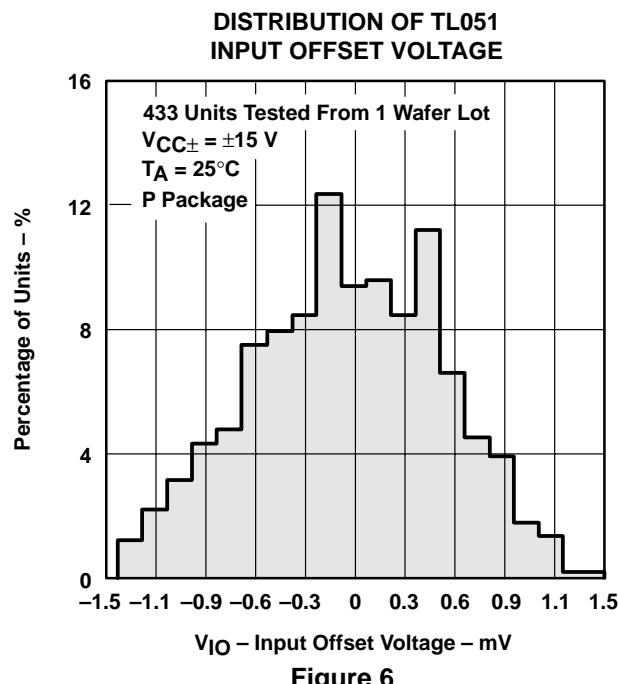
Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	6–11
$\alpha_{V_{IO}}$	Temperature coefficient of input offset voltage	Distribution	12, 13, 14
I_{IB}	Input bias current	vs Common-mode input voltage vs Free-air temperature	15 16
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V_{IC}	Common-mode input voltage range limits	vs Supply voltage vs Free-air temperature	17 18
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V_n	Equivalent input noise voltage	vs Frequency	61, 62
THD	Total harmonic distortion	vs Frequency	63
B_1	Unity-gain bandwidth	vs Supply voltage vs Free-air temperature	64, 65, 66 67, 68, 69
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TYPICAL CHARACTERISTICS



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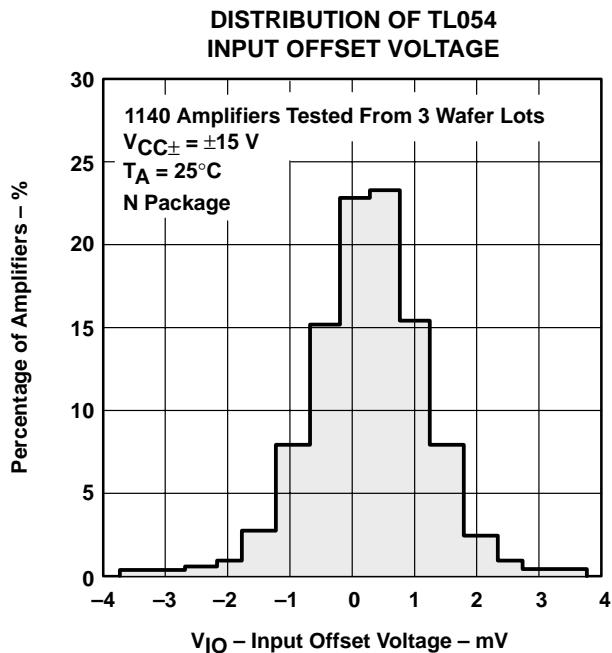


Figure 10

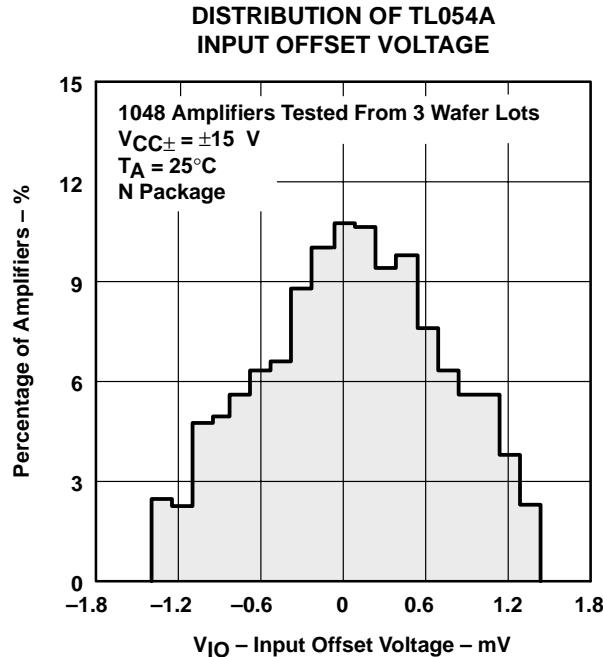


Figure 11

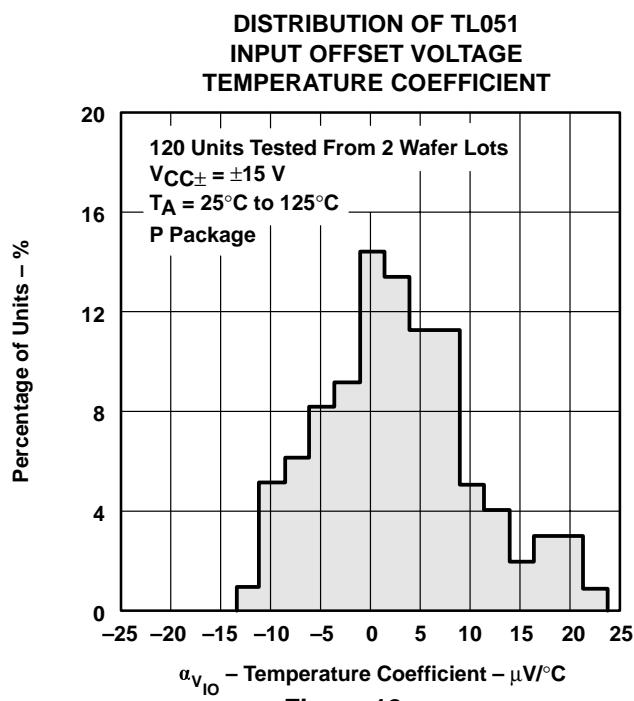


Figure 12

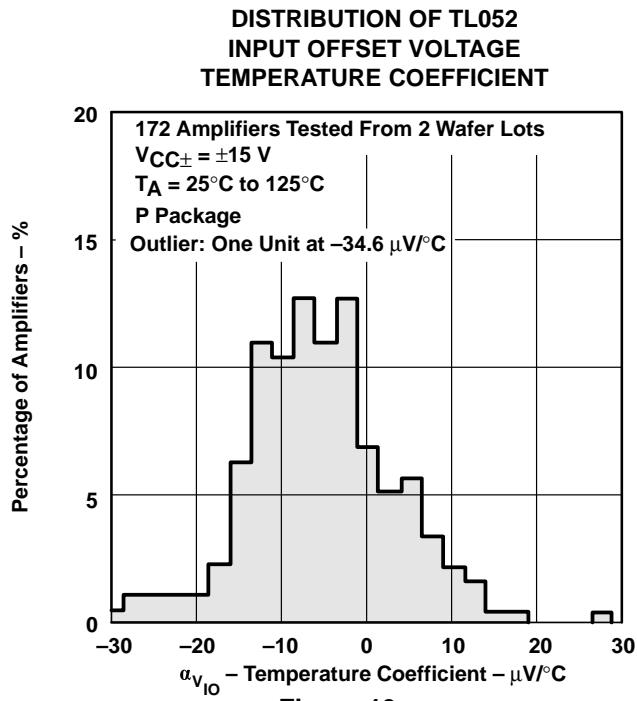
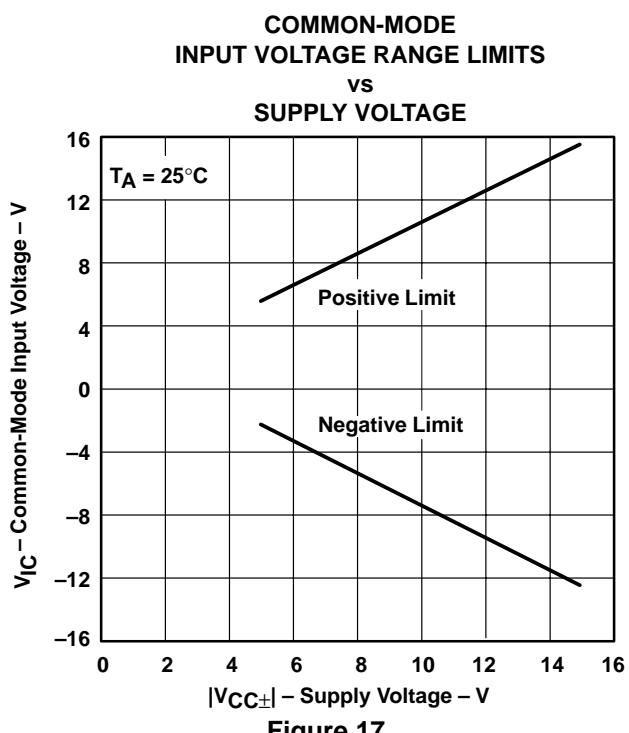
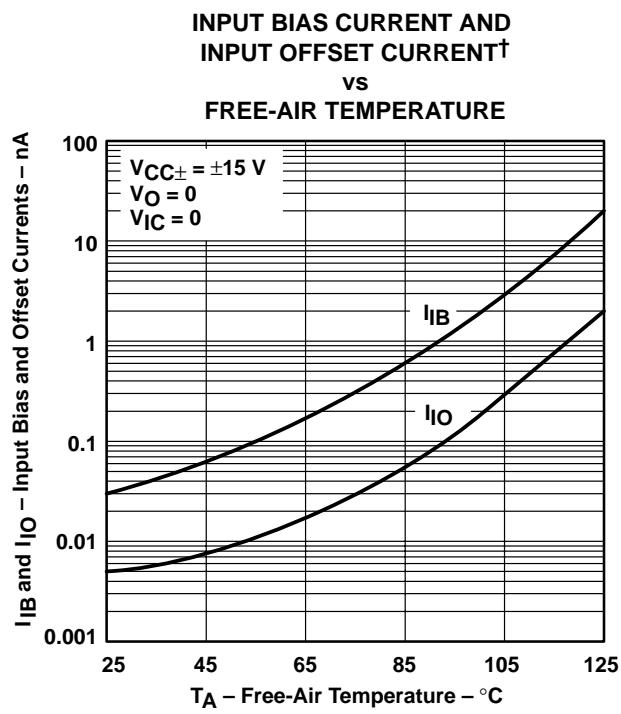
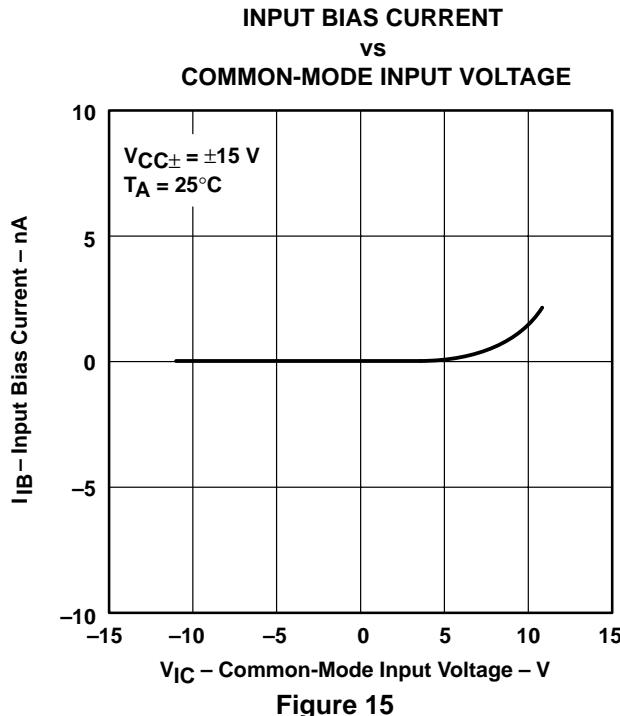
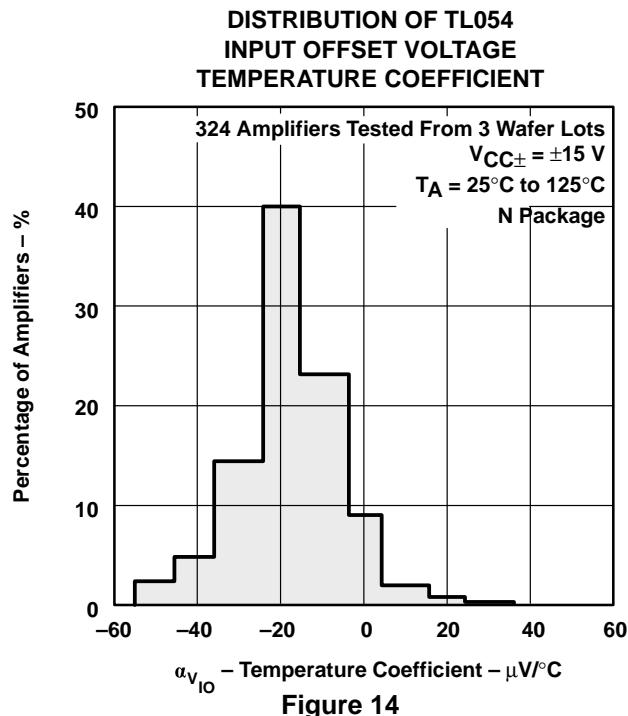


Figure 13

TL05x, TL05xA
ENHANCED-JFET LOW-OFFSET
OPERATIONAL AMPLIFIERS

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TYPICAL CHARACTERISTICS



[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

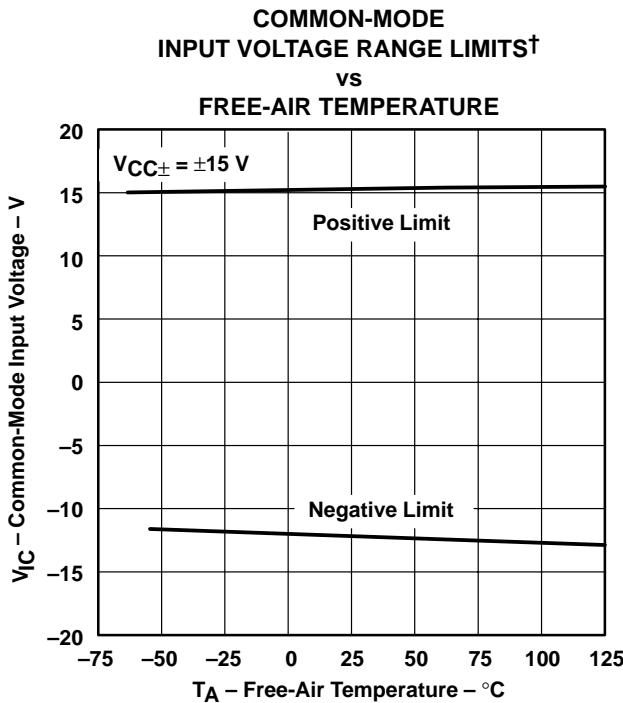


Figure 18

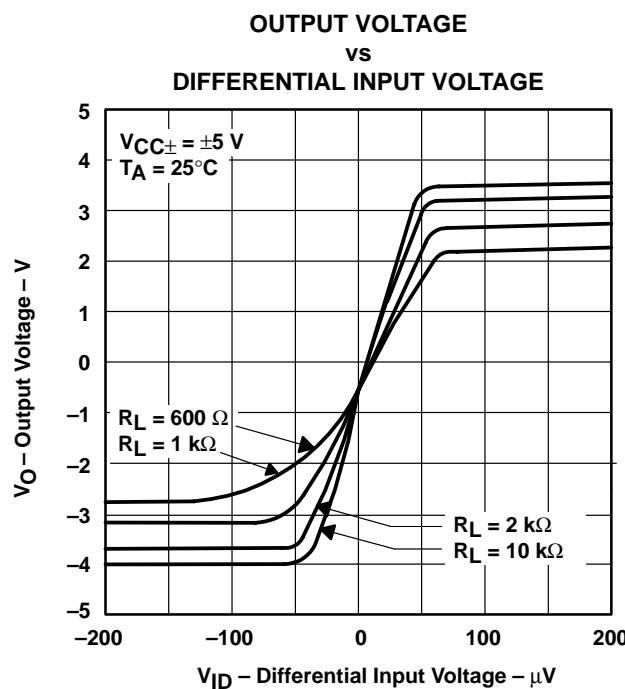


Figure 19

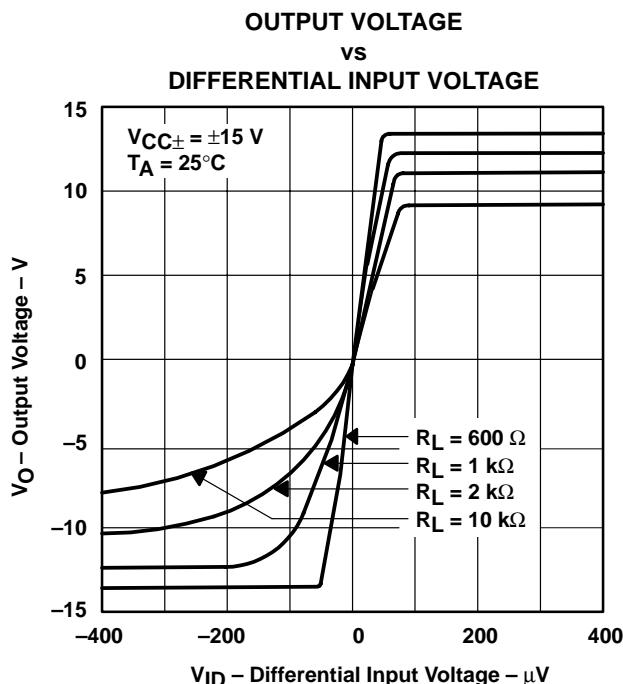


Figure 20

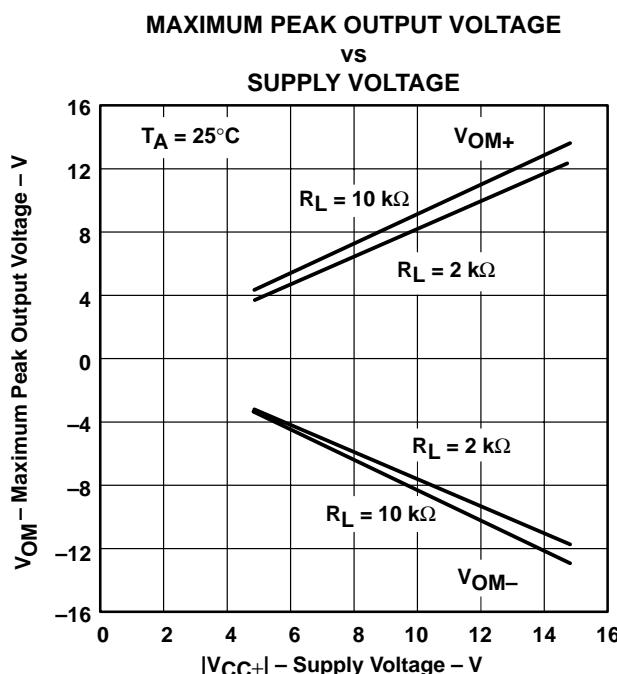


Figure 21

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

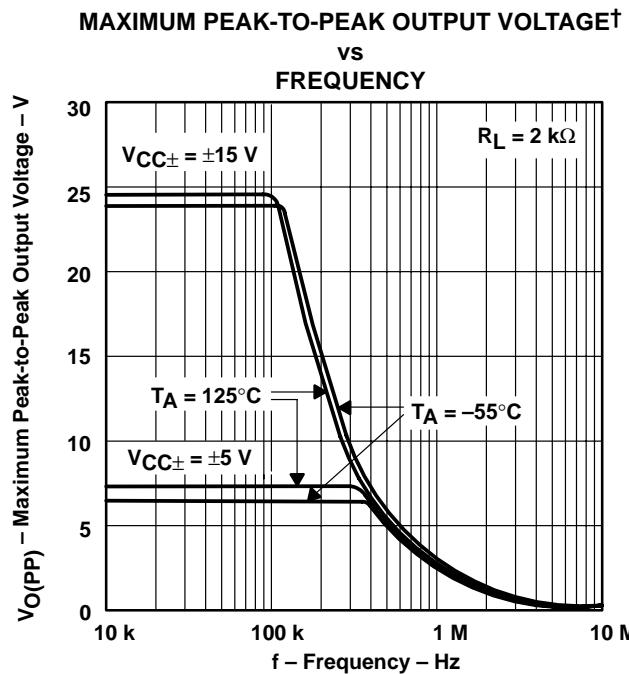


Figure 22

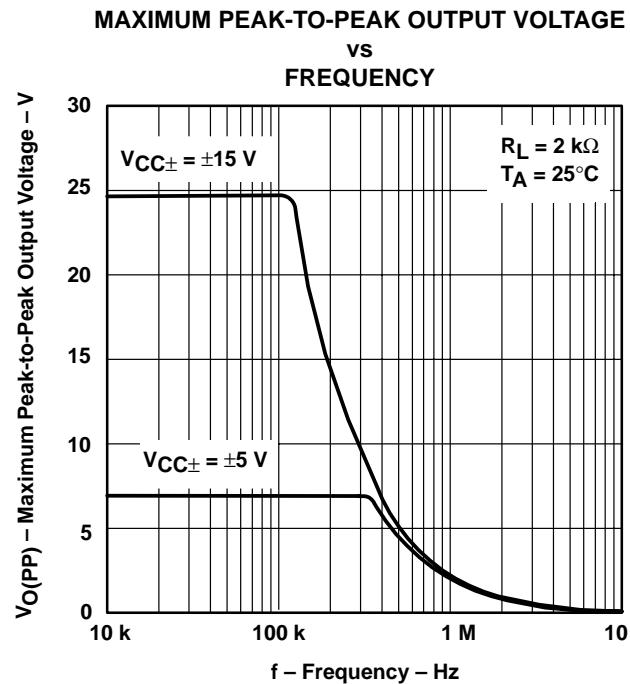


Figure 23

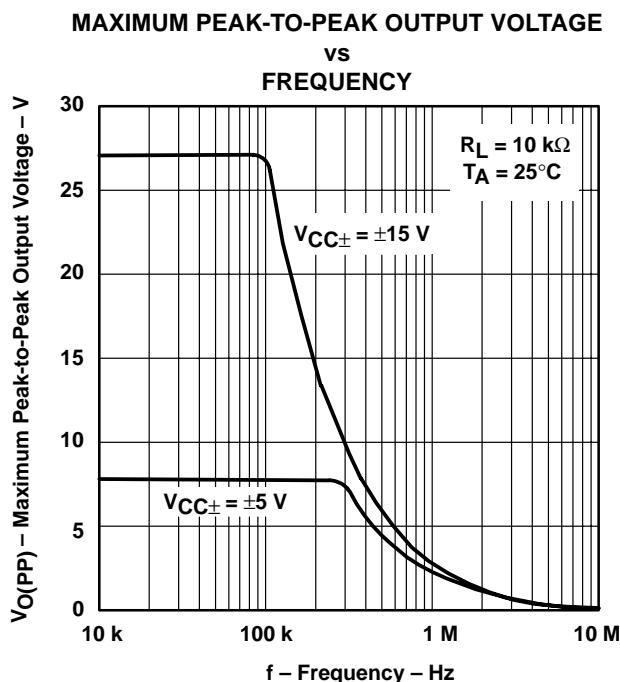


Figure 24

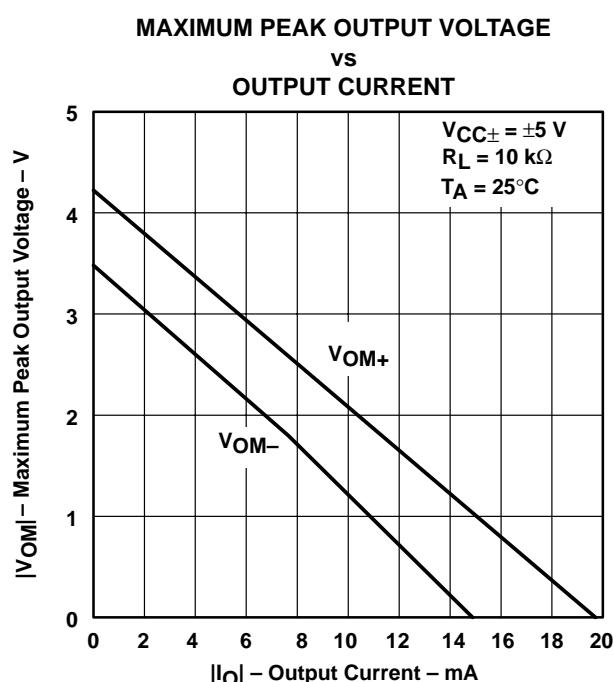


Figure 25

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

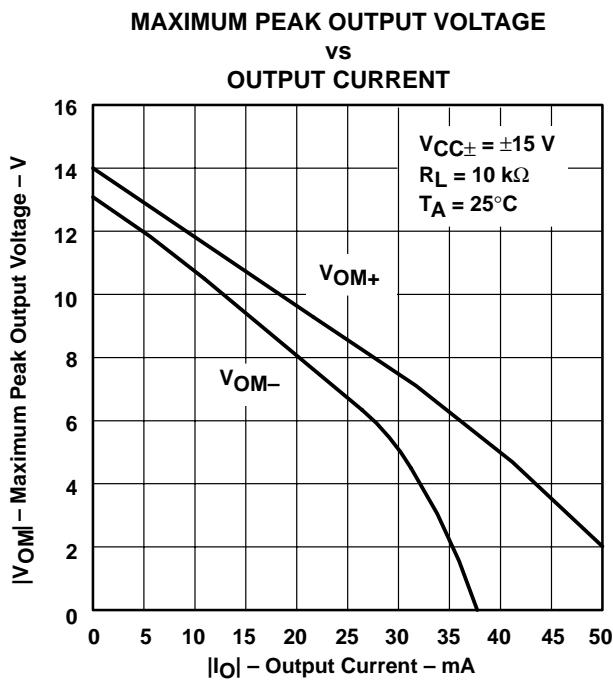


Figure 26

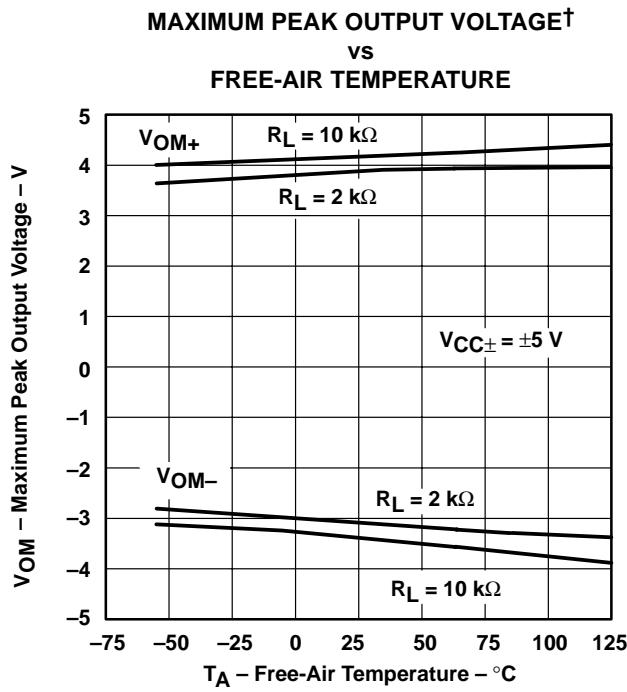


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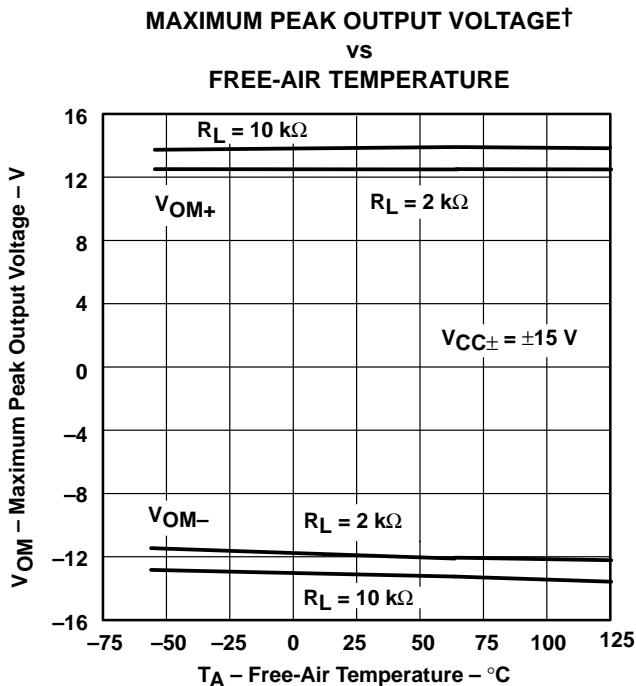


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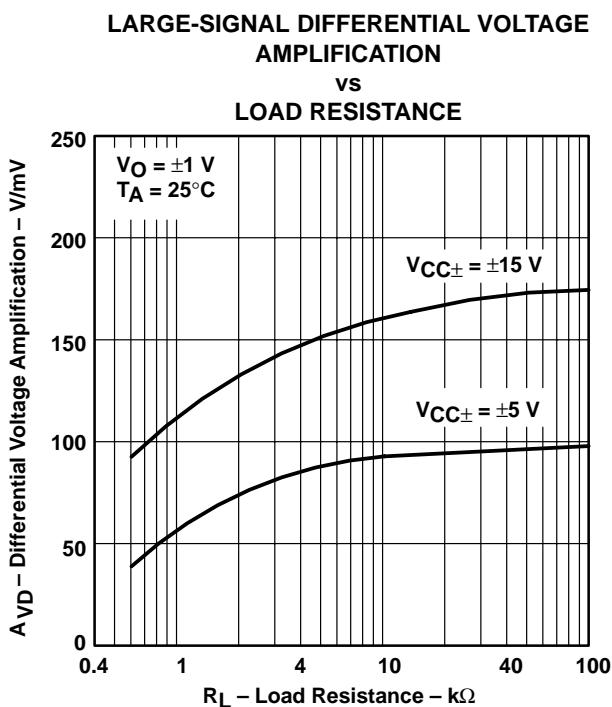


Figure 29

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS
LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT

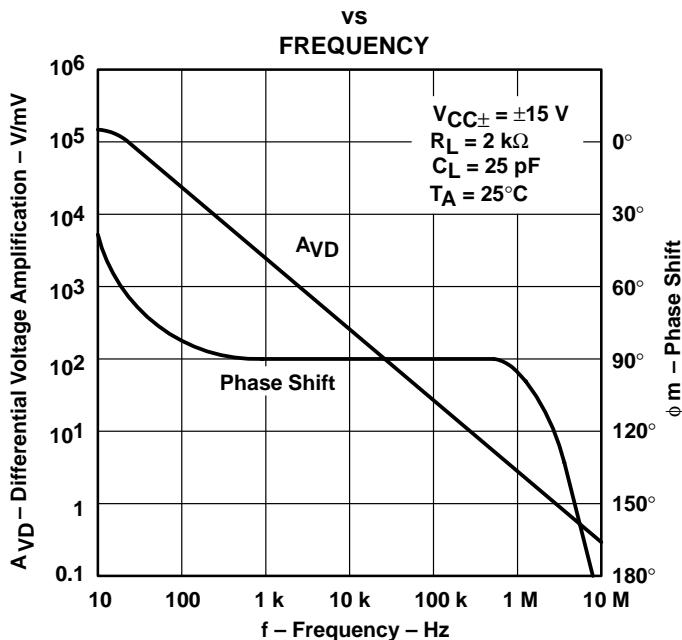


Figure 30

TL051 AND TL052
LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION†
vs
FREE-AIR TEMPERATURE

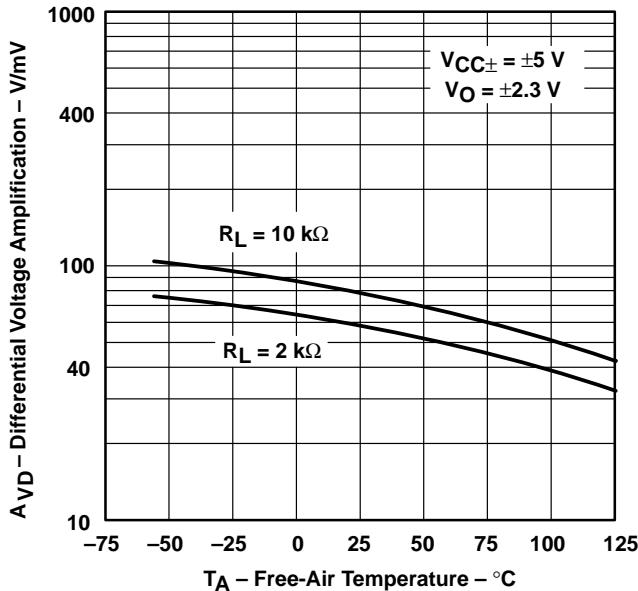


Figure 31

TL054
LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION†
vs
FREE-AIR TEMPERATURE

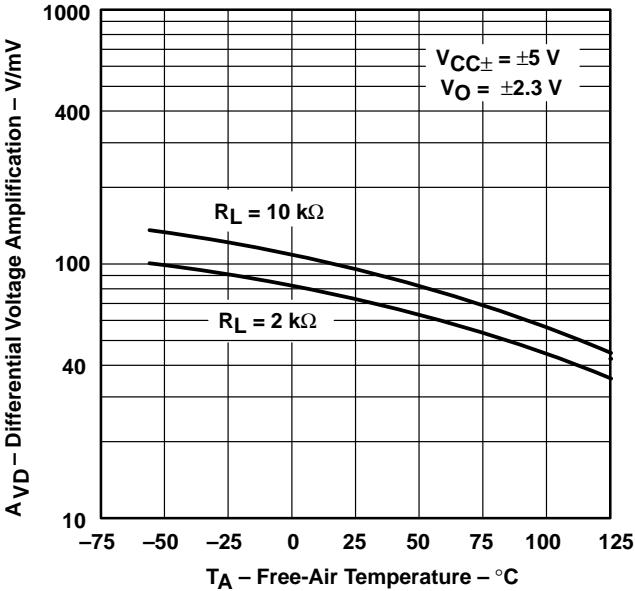


Figure 32

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION†
VS
FREE-AIR TEMPERATURE

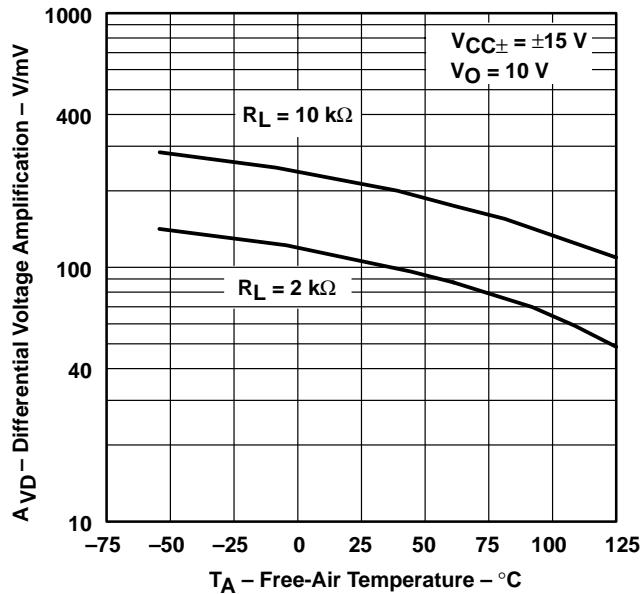


Figure 33

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

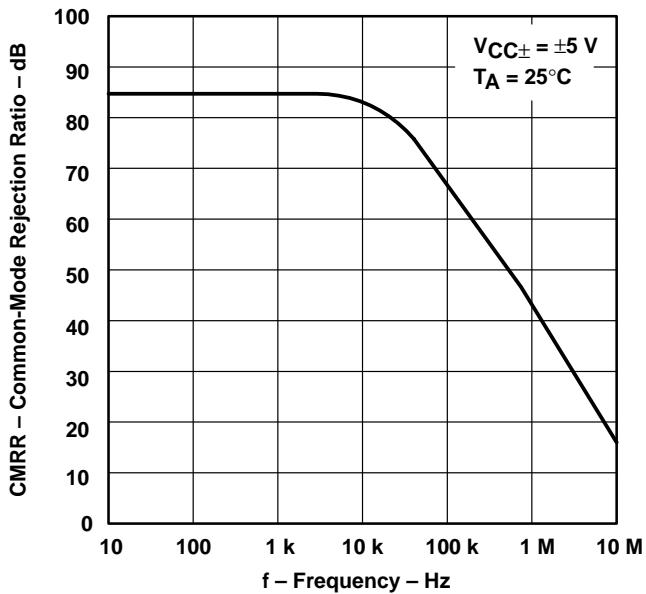


Figure 34

COMMON-MODE REJECTION RATIO
VS
FREQUENCY

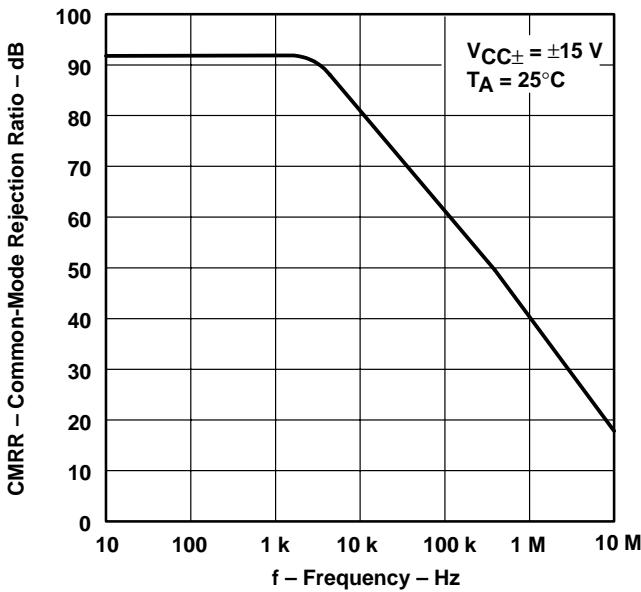


Figure 35

COMMON-MODE REJECTION RATIO†
VS
FREE-AIR TEMPERATURE

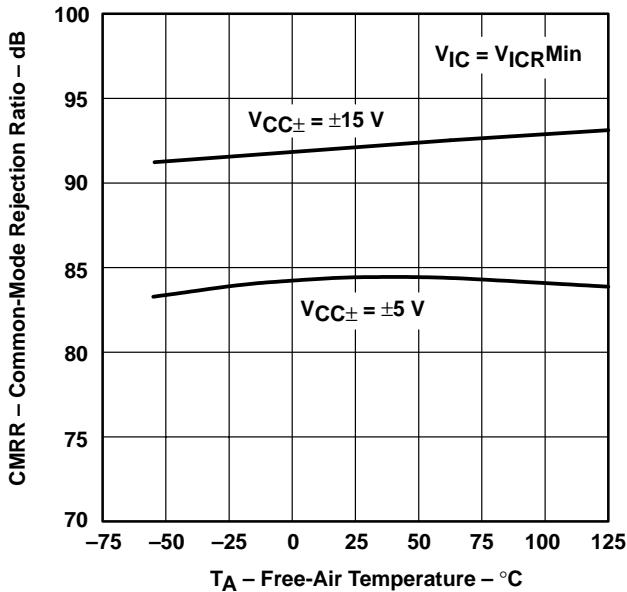


Figure 36

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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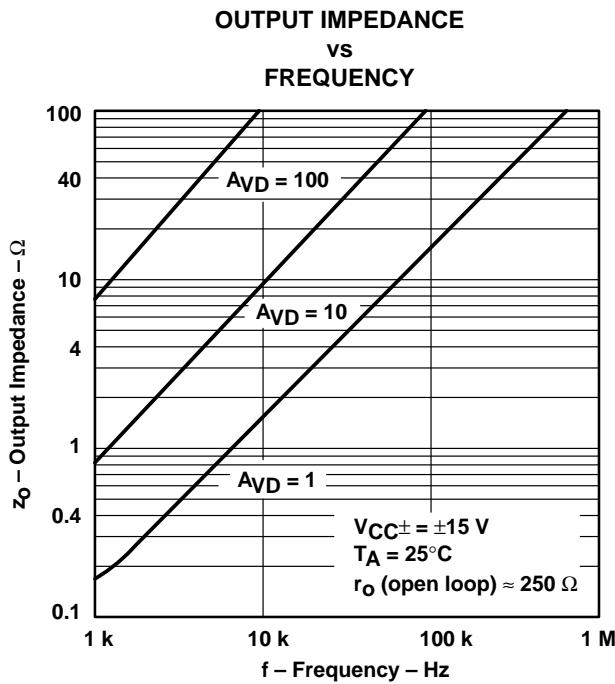


Figure 37

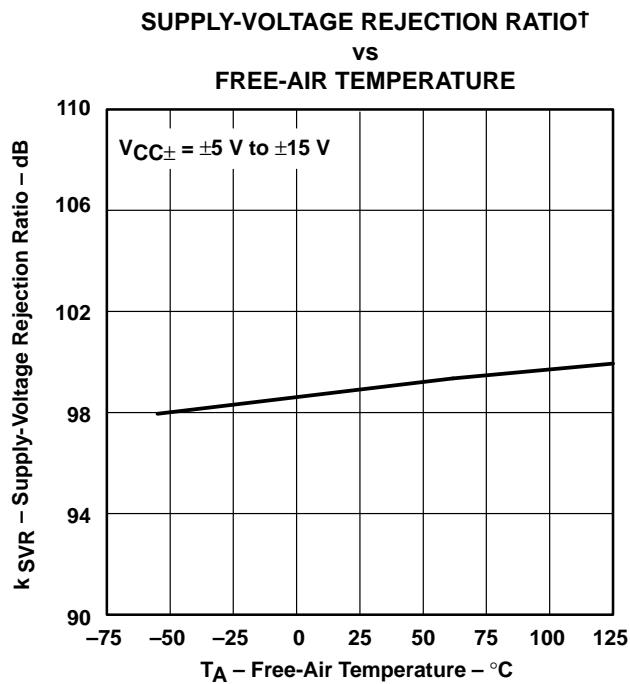


Figure 38

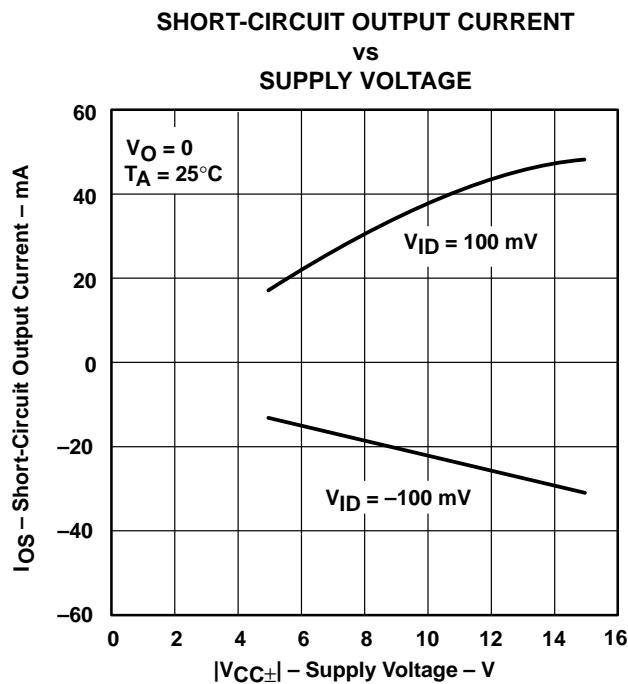


Figure 39

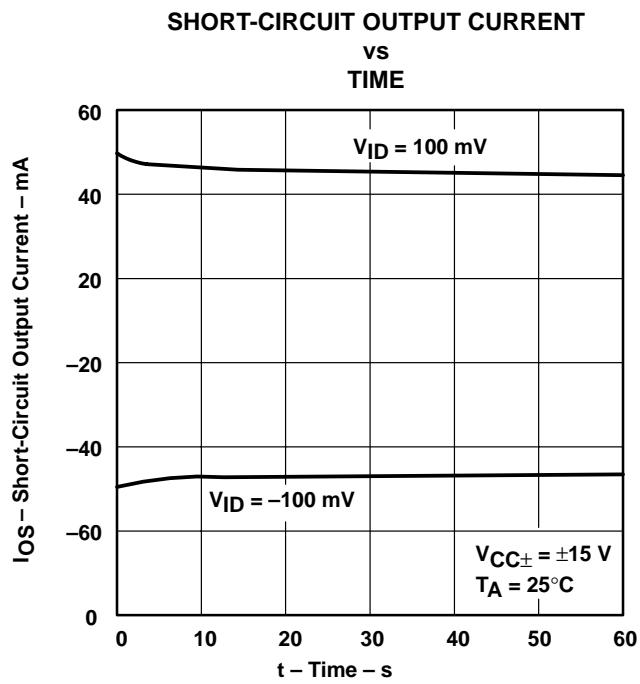


Figure 40

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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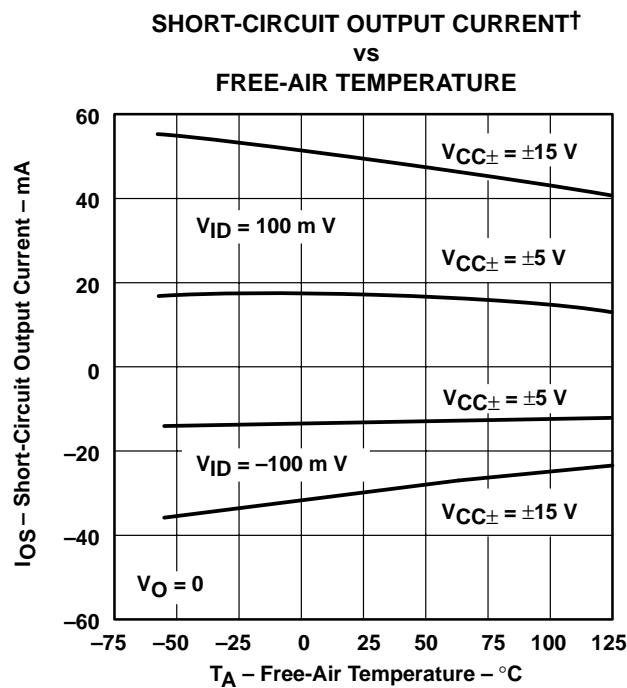


Figure 41

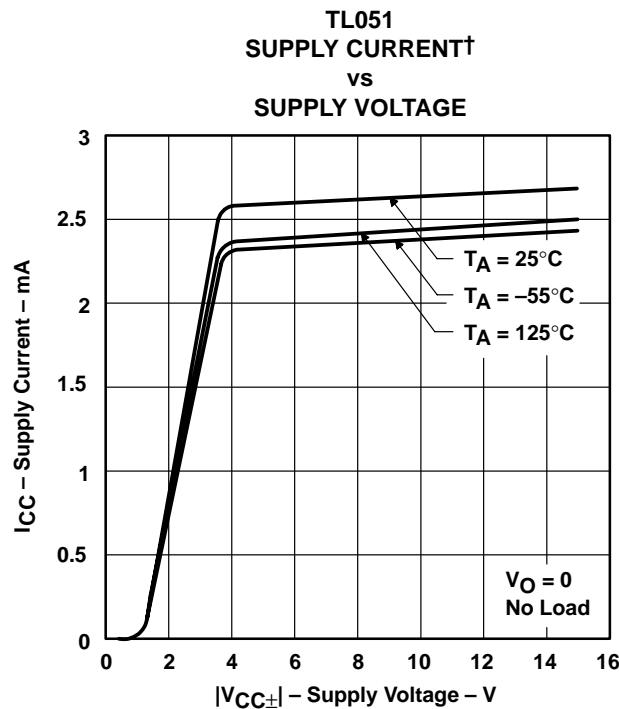


Figure 42

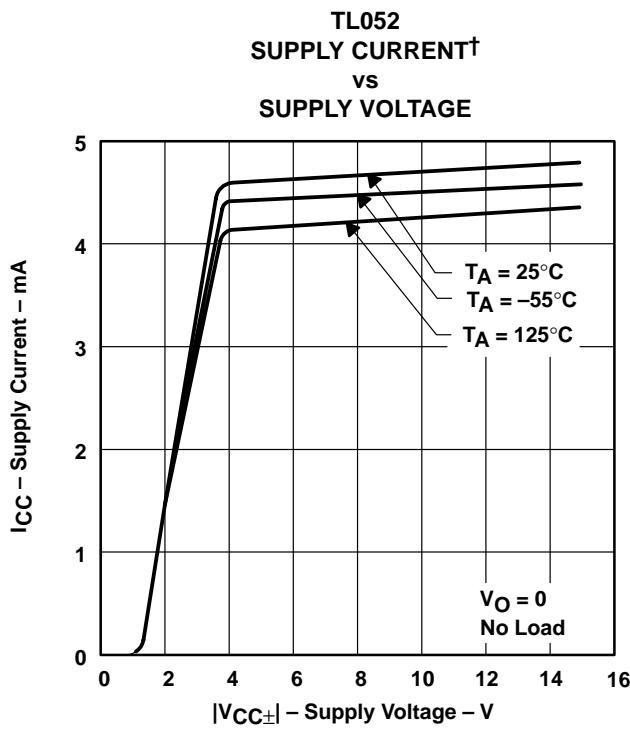


Figure 43

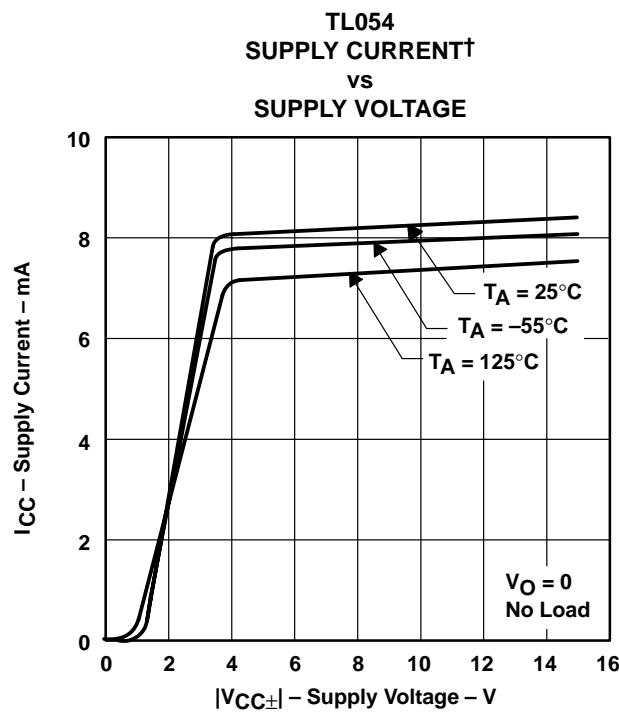


Figure 44

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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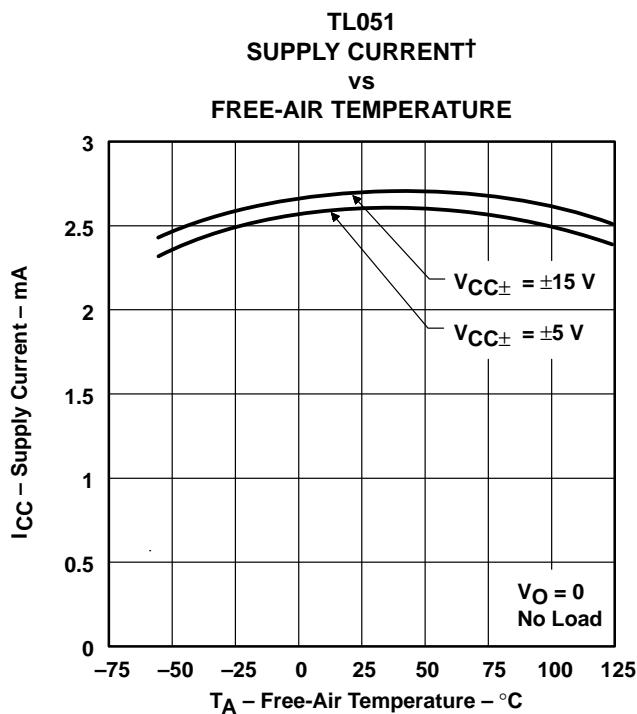


Figure 45

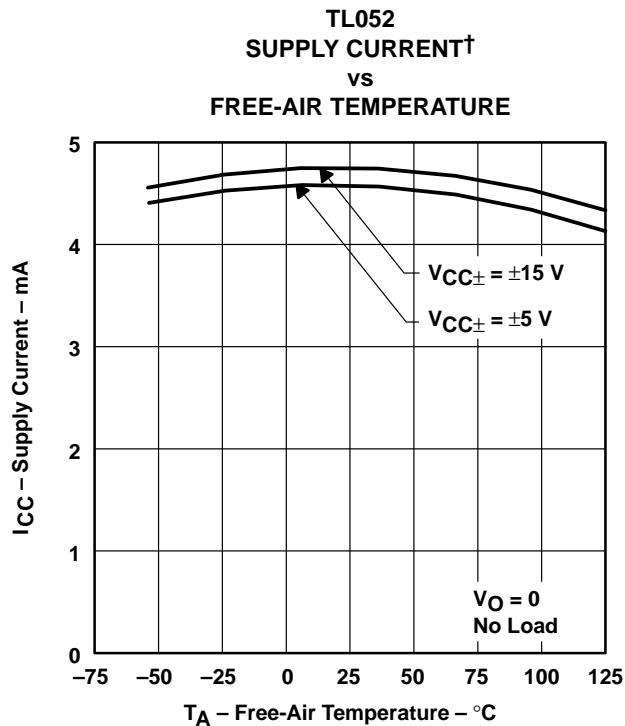


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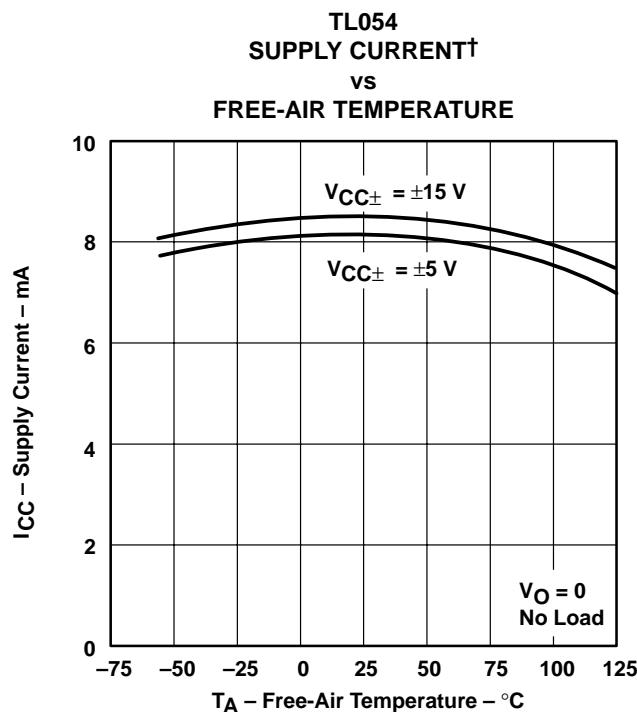


Figure 47

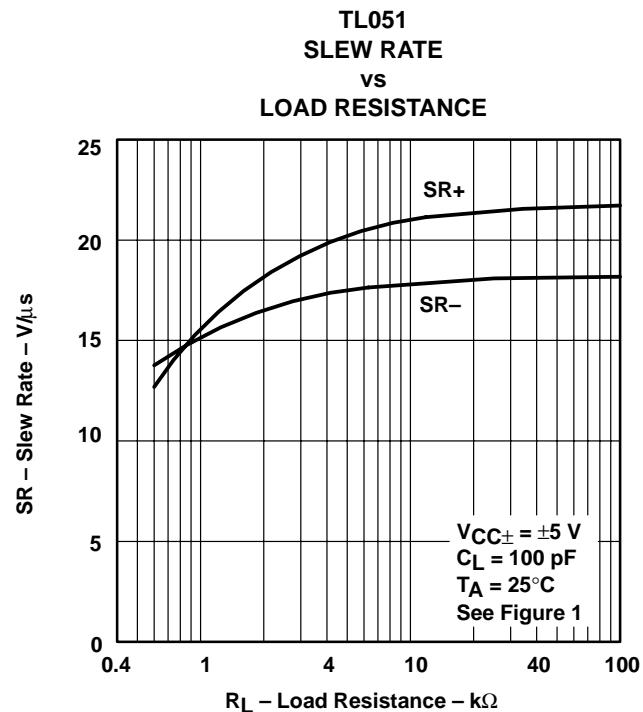


Figure 48

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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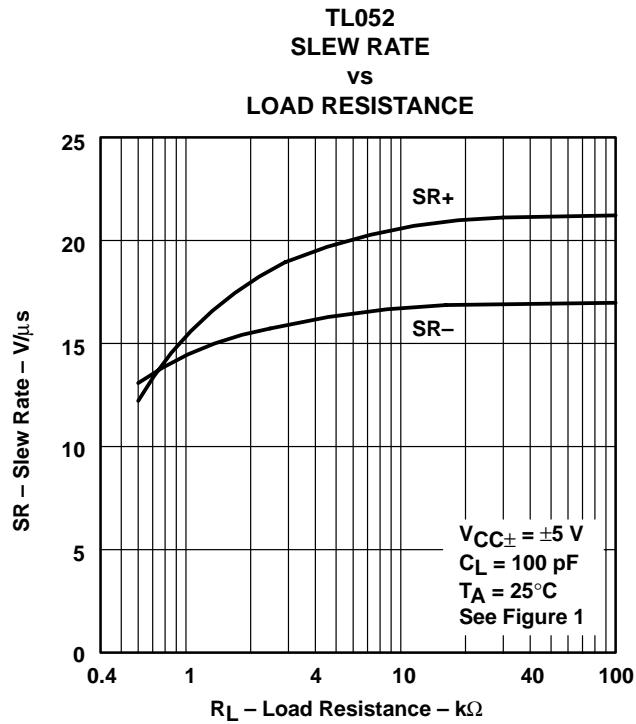


Figure 49

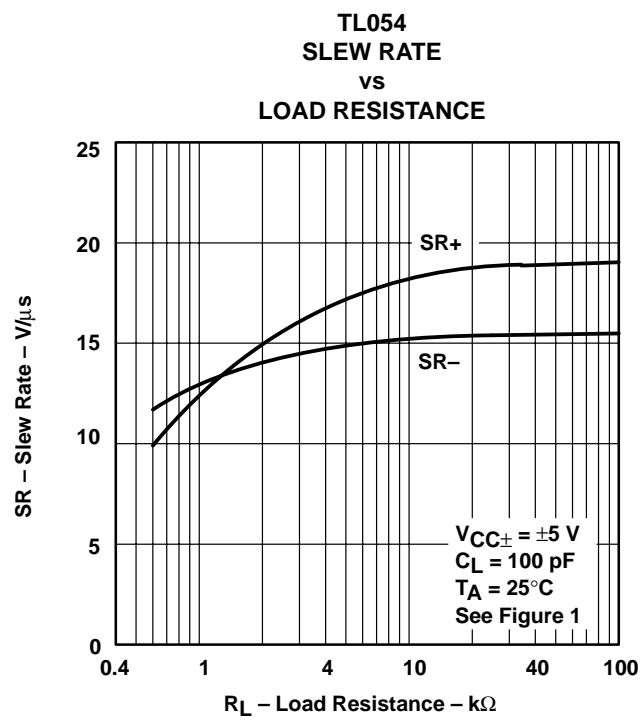


Figure 50

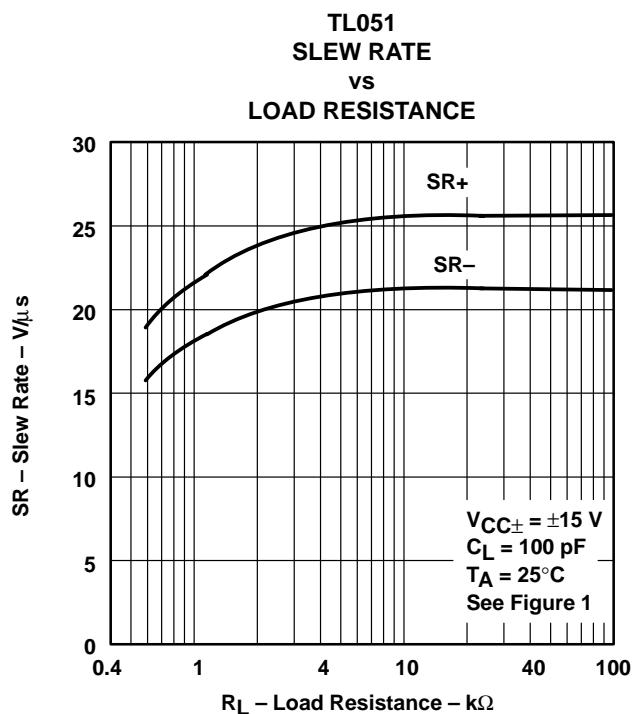


Figure 51

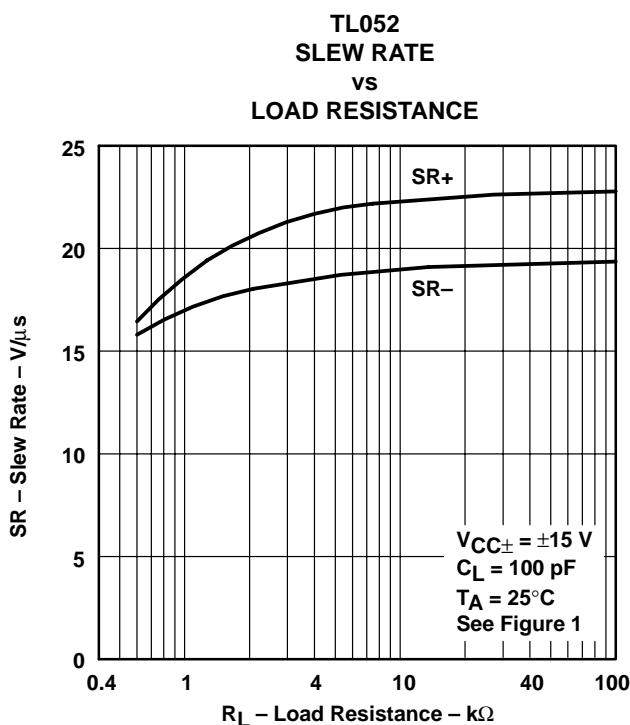


Figure 52

TL05x, TL05xA
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TYPICAL CHARACTERISTICS

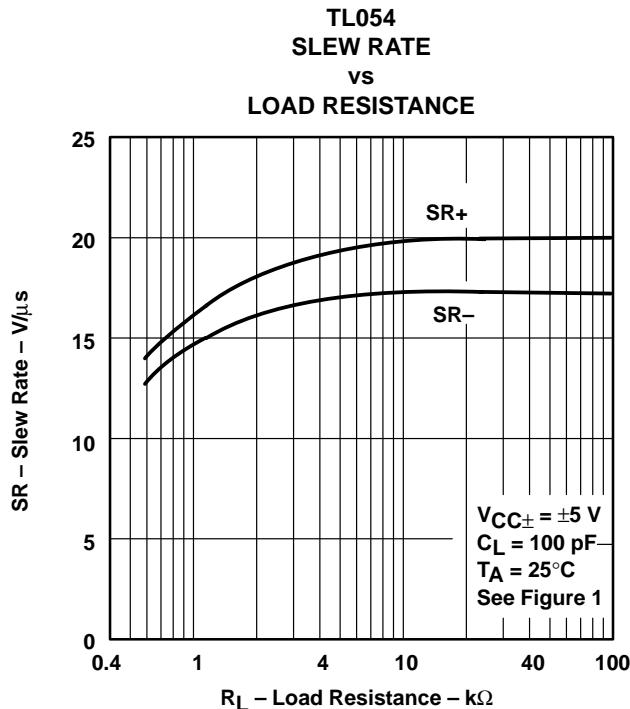


Figure 53

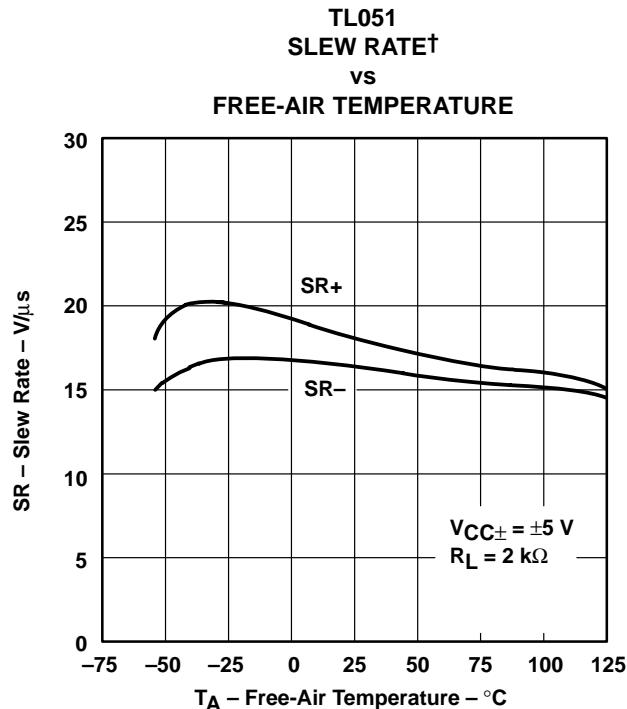


Figure 54

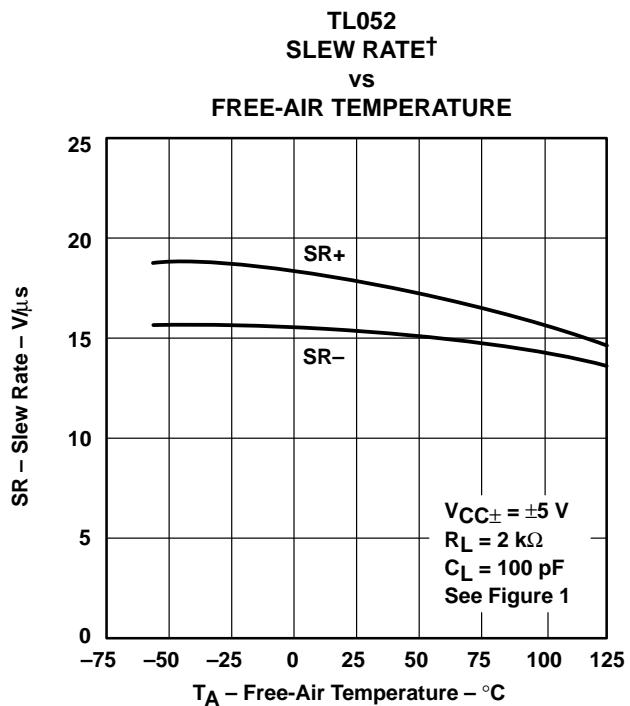


Figure 55

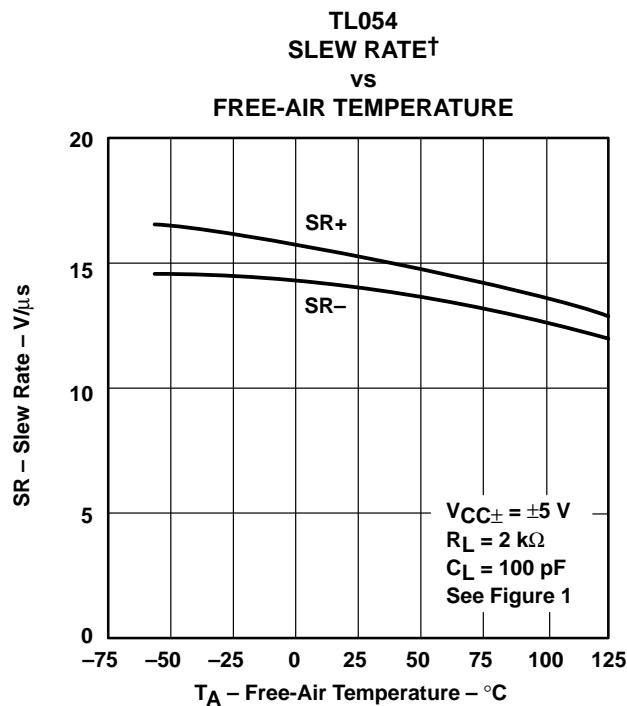


Figure 56

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

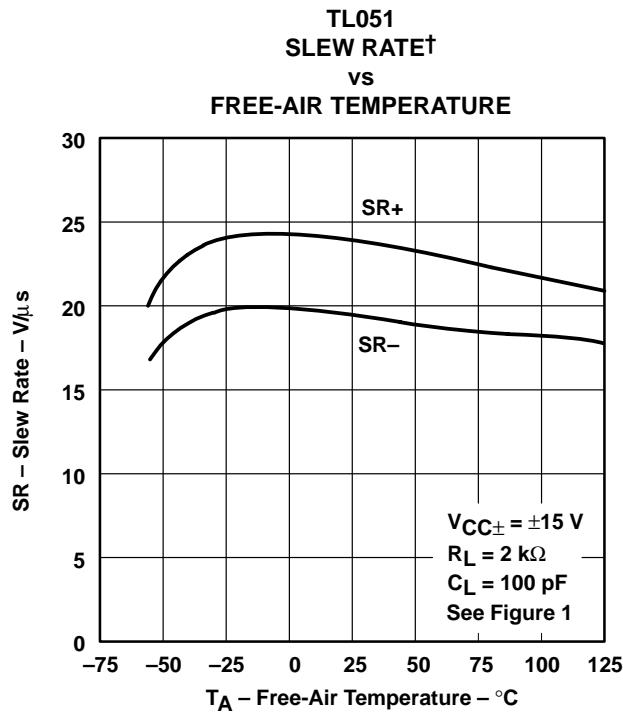


Figure 57

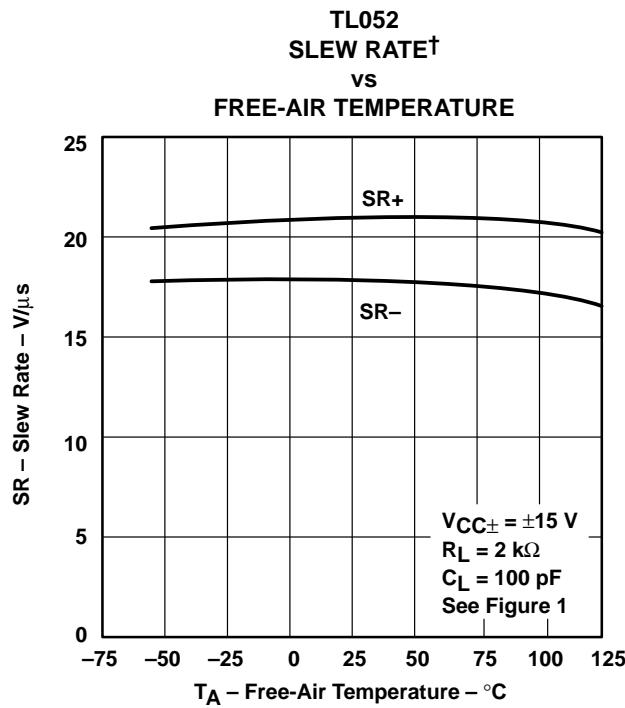


Figure 58

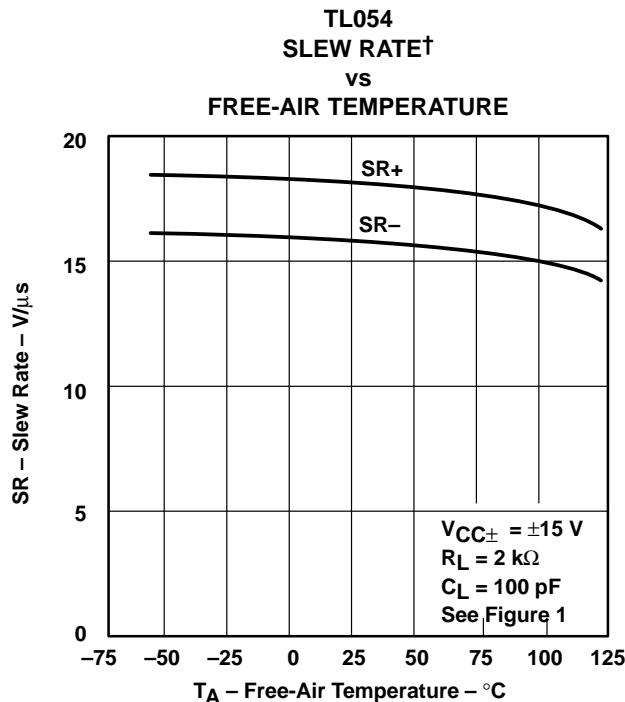


Figure 59

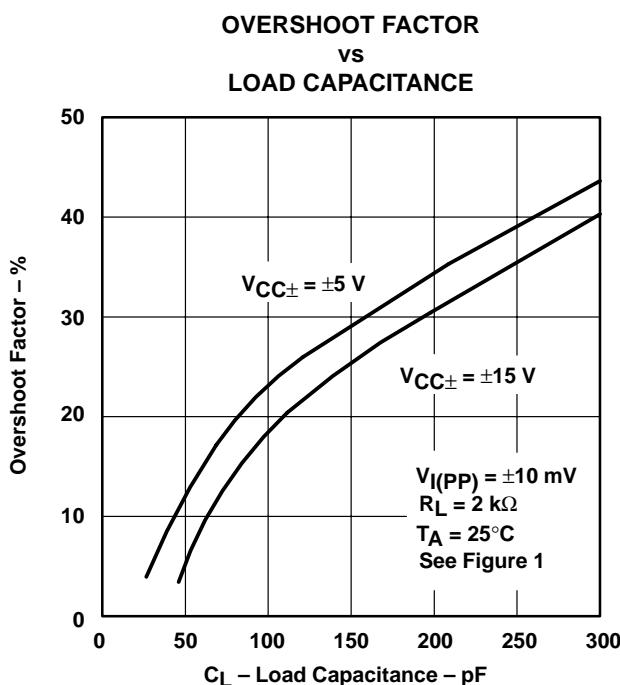


Figure 60

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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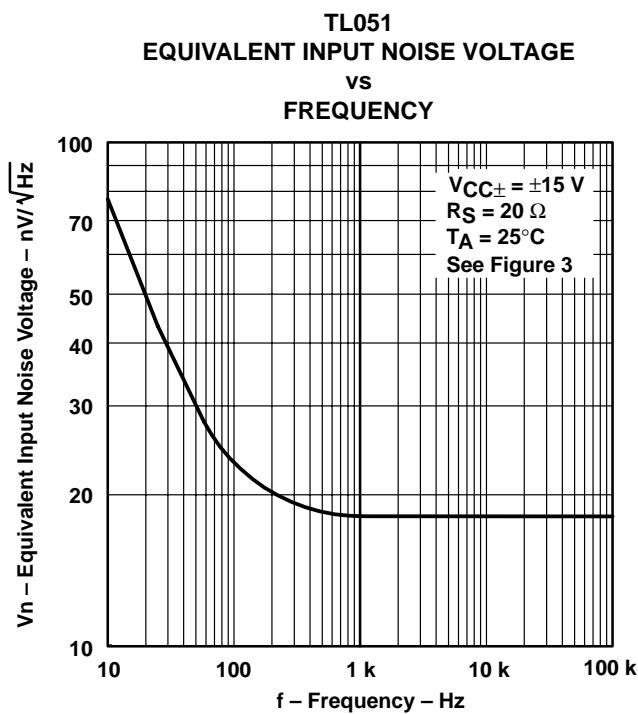


Figure 61

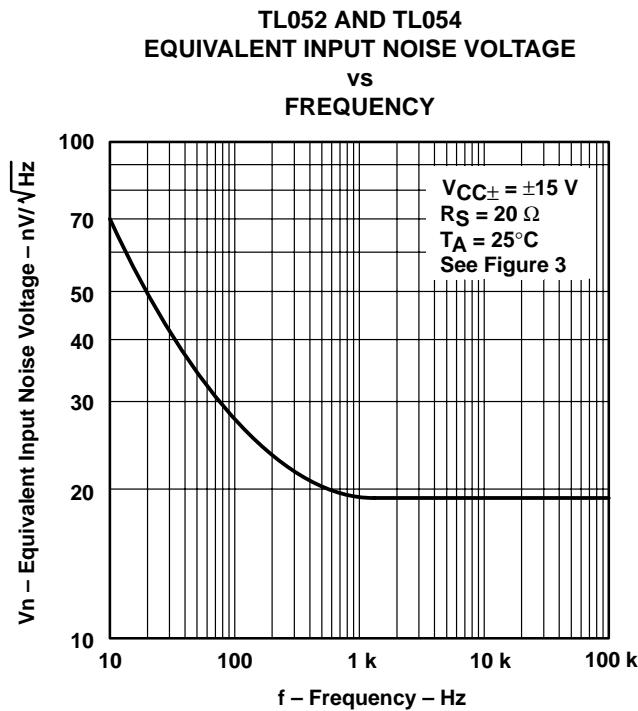


Figure 62

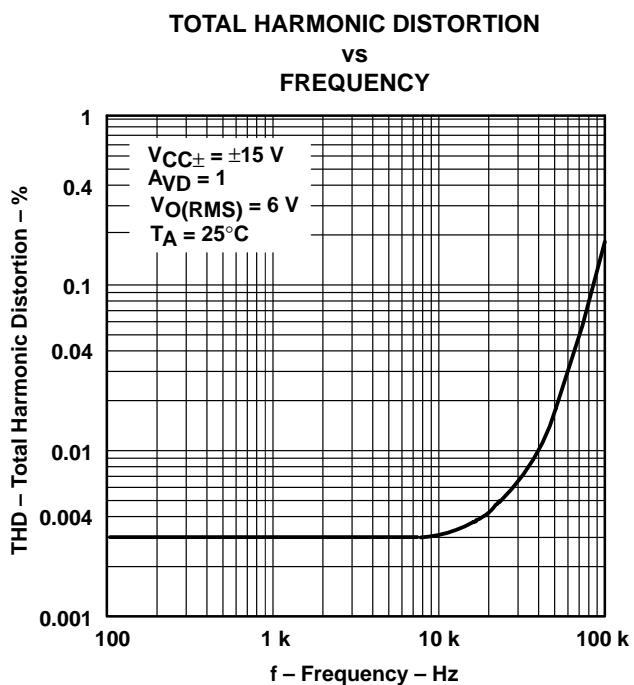


Figure 63

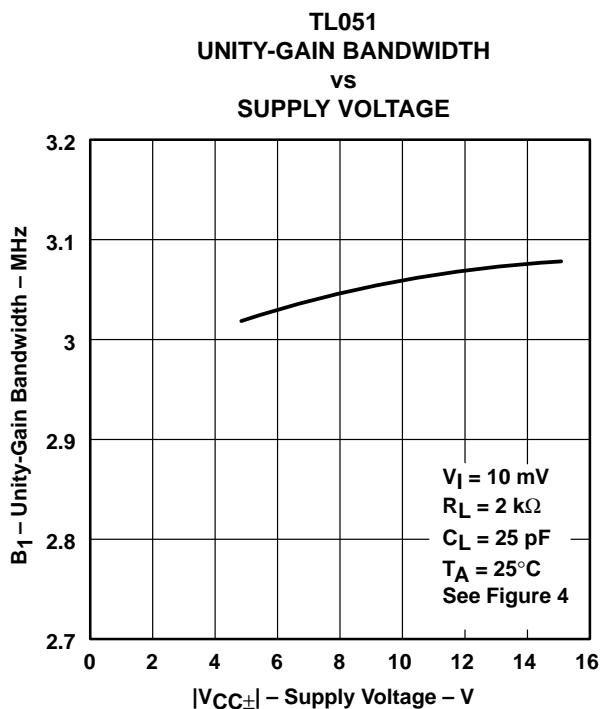


Figure 64

TL05x, TL05xA ENHANCED-JFET LOW-OFFSET OPERATIONAL AMPLIFIERS

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TYPICAL CHARACTERISTICS

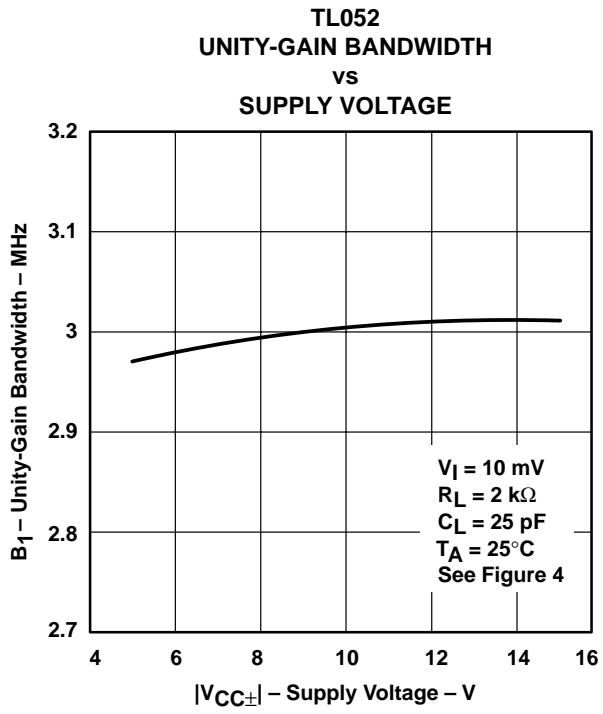


Figure 65

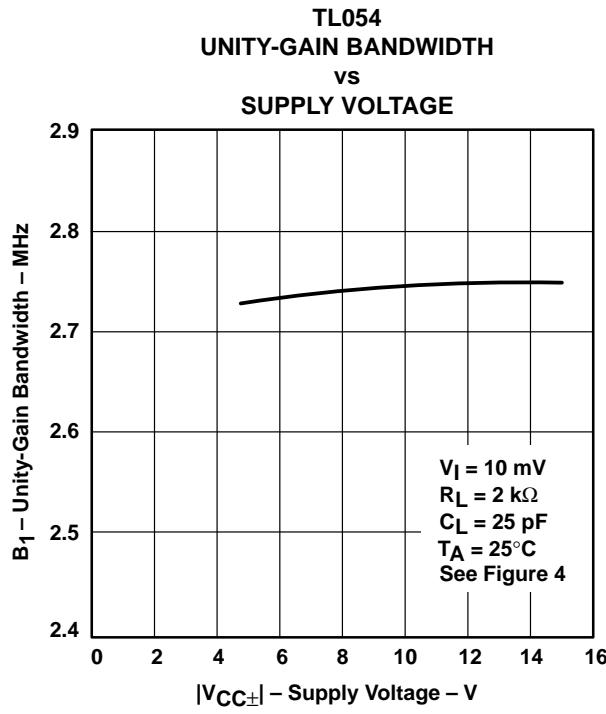


Figure 66

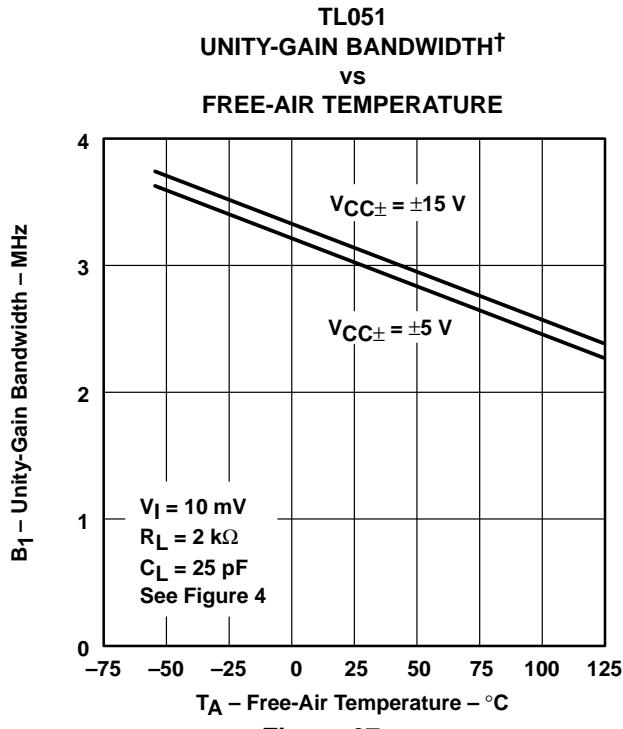


Figure 67

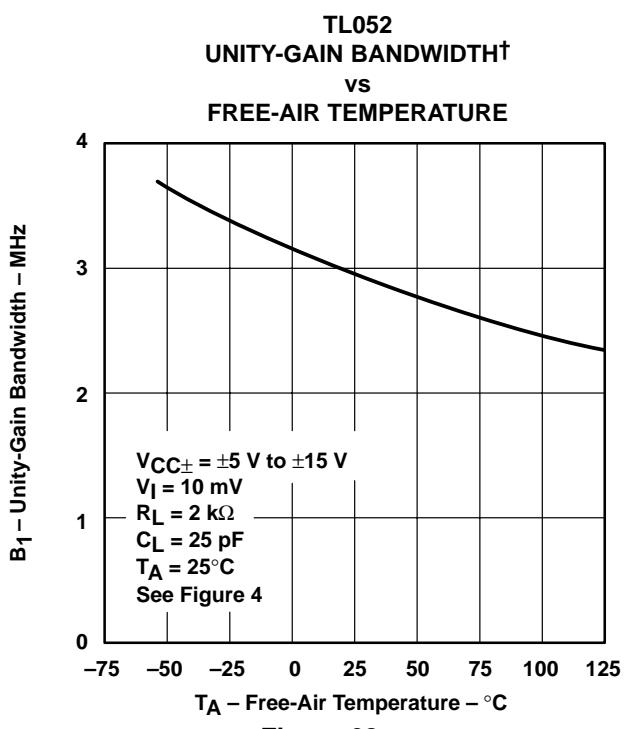


Figure 68

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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TYPICAL CHARACTERISTICS

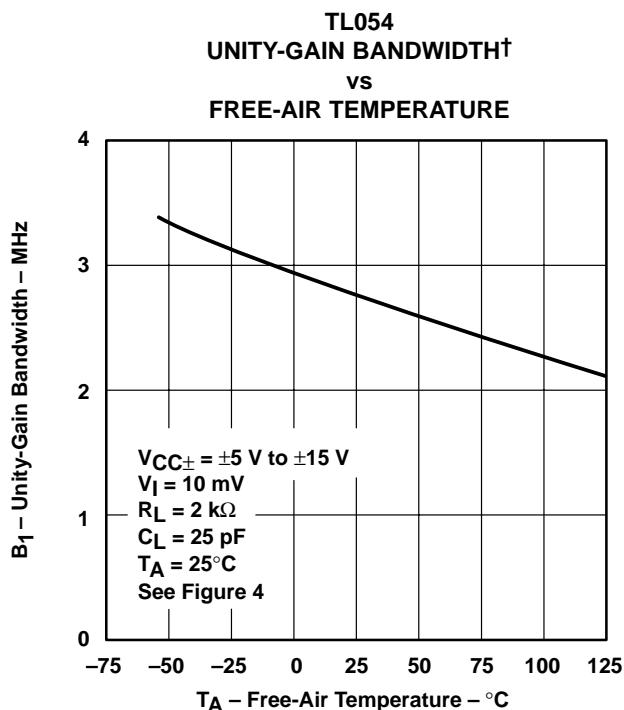


Figure 69

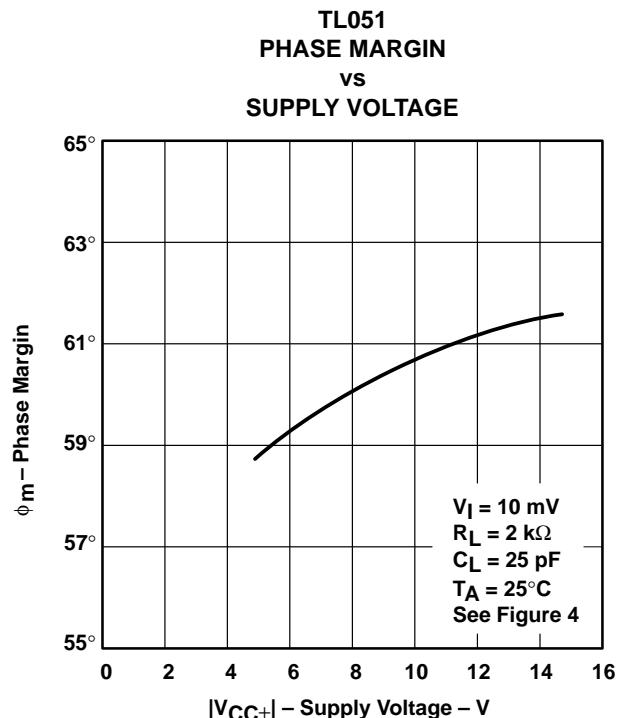


Figure 70

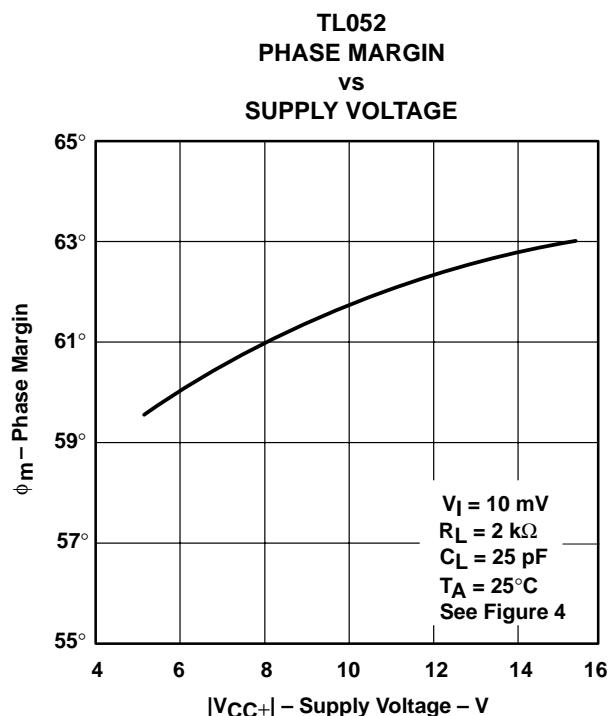


Figure 71

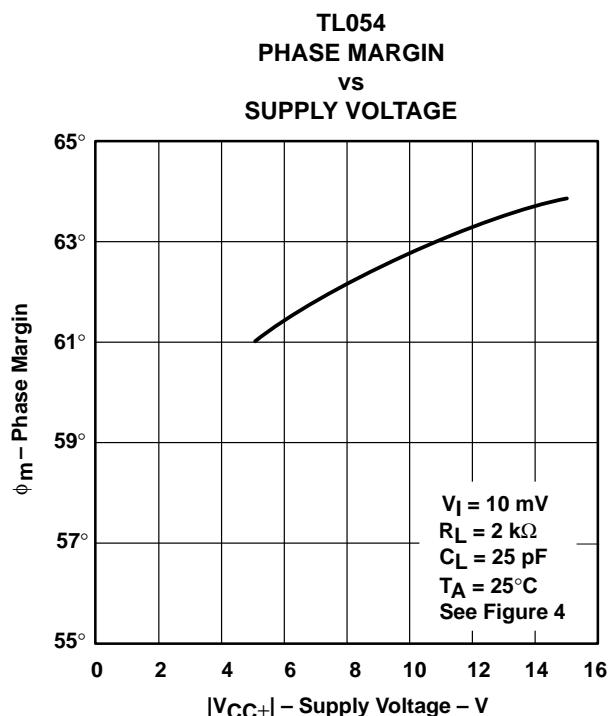


Figure 72

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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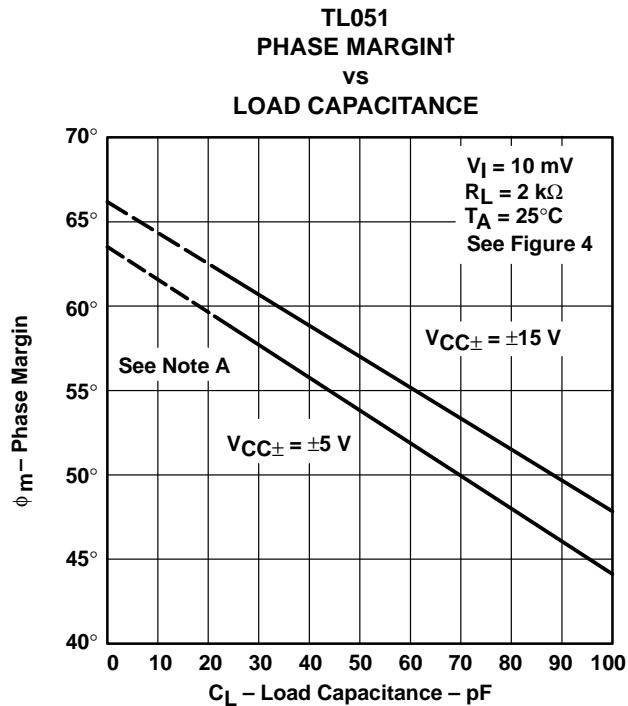


Figure 73

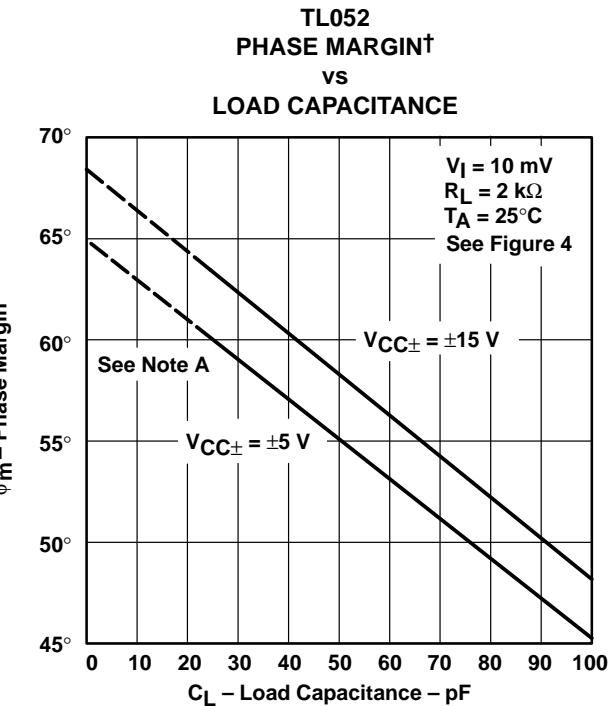


Figure 74

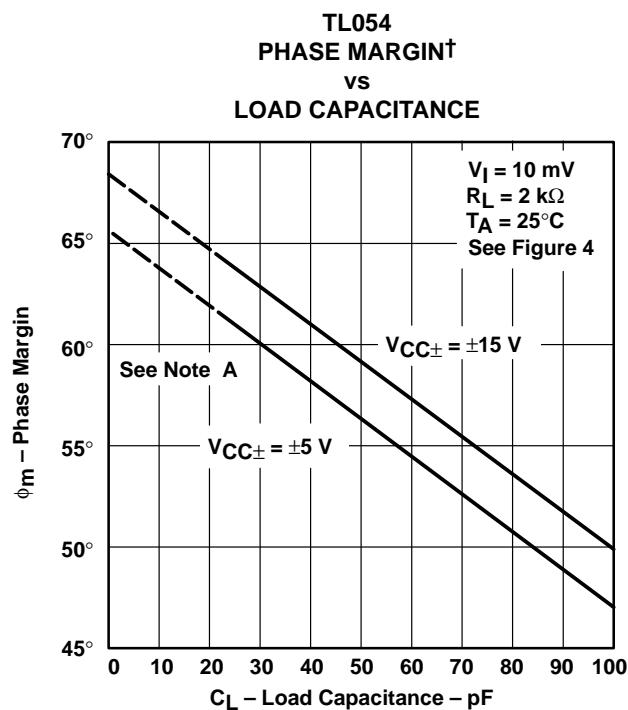


Figure 75

† Values of phase margin below a load capacitance of 25 pF were estimated.

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TYPICAL CHARACTERISTICS

**TL051
PHASE MARGIN†
vs
FREE-AIR TEMPERATURE**

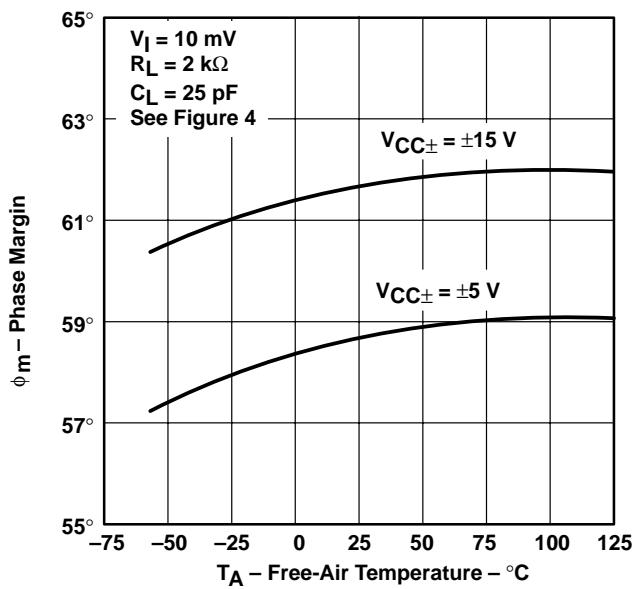


Figure 76

**TL052
PHASE MARGIN†
vs
FREE-AIR TEMPERATURE**

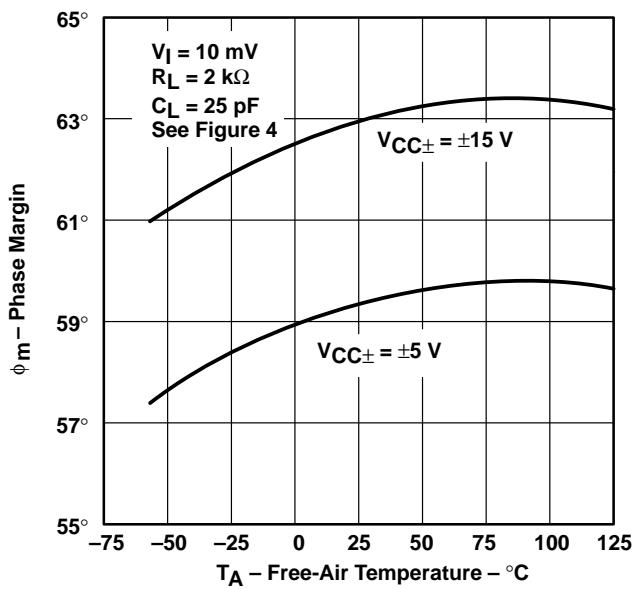


Figure 77

**TL054
PHASE MARGIN†
vs
FREE-AIR TEMPERATURE**

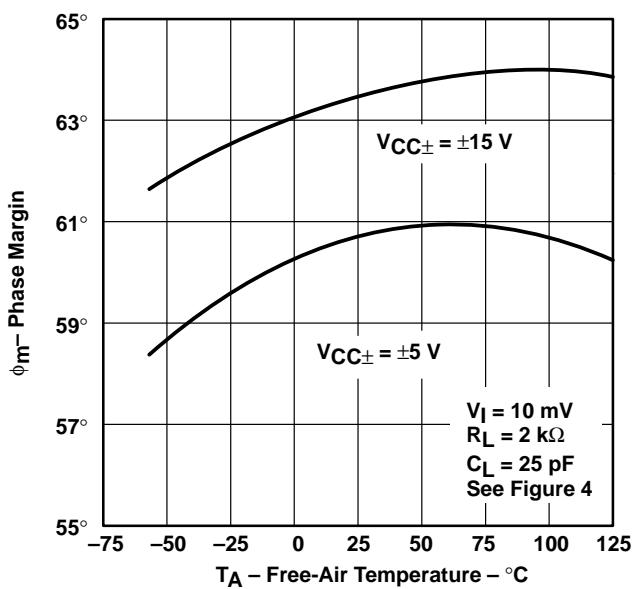


Figure 78

† Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

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ENHANCED-JFET LOW-OFFSET
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TYPICAL CHARACTERISTICS

**VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE**

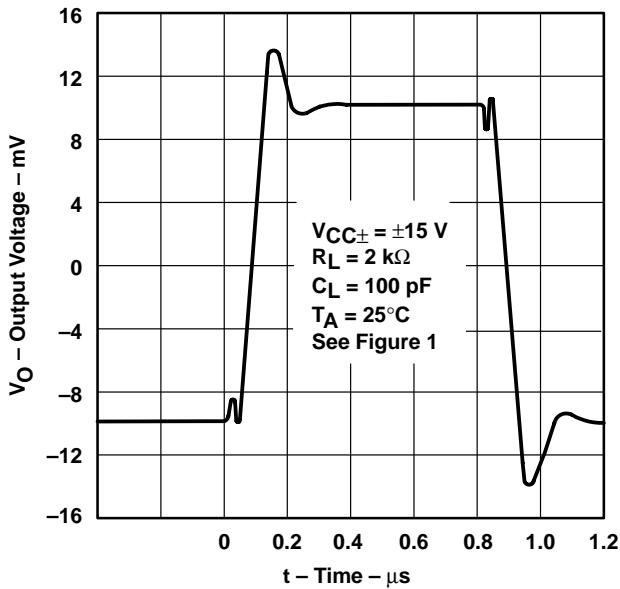


Figure 79

**VOLTAGE-FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE**

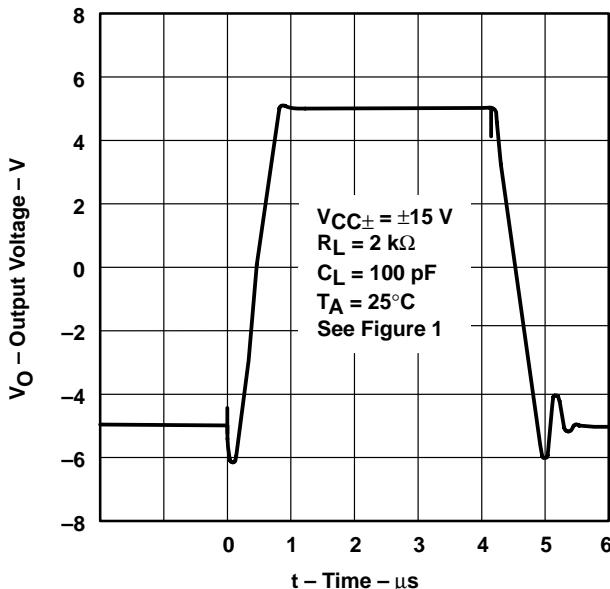


Figure 80

TL05x, TL05xA
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APPLICATION INFORMATION

output characteristics

All operating characteristics (except bandwidth and phase margin) are specified with 100-pF load capacitance. The TL05x and TL05xA drive higher capacitive loads; however, as the load capacitance increases, the resulting response pole occurs at lower frequencies, causing ringing, peaking, or even oscillation. The value of the load capacitance at which oscillation occurs varies with production lots. If an application appears to be sensitive to oscillation due to load capacitance, adding a small resistance in series with the load should alleviate the problem. Capacitive loads of 1000 pF, and larger, may be driven if enough resistance is added in series with the output (see Figure 81 and Figure 82).

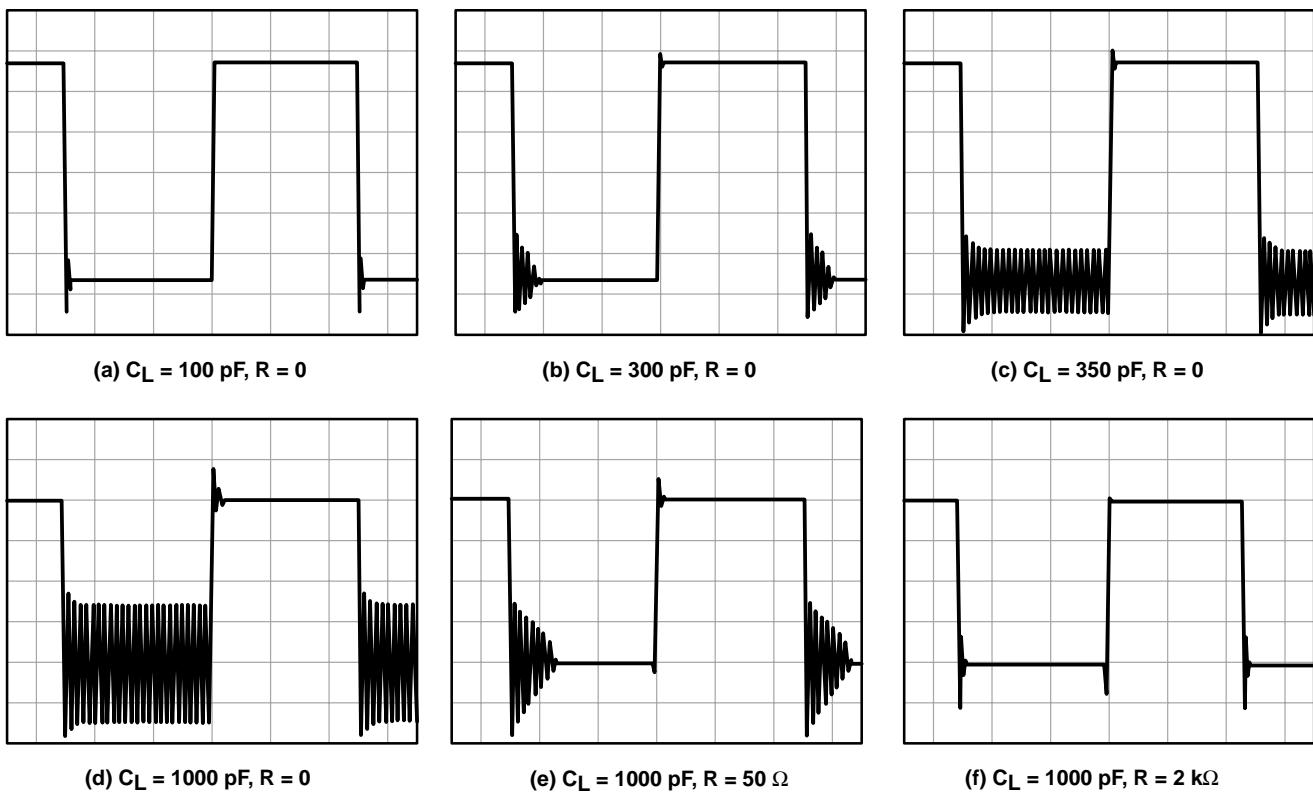


Figure 81. Effect of Capacitive Loads

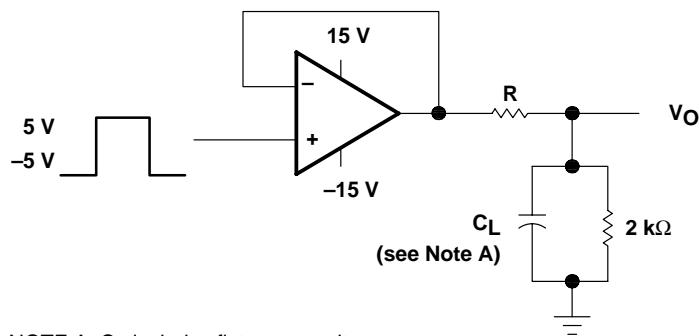


Figure 82. Test Circuit for Output Characteristics

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APPLICATION INFORMATION

input characteristics

The TL05x and TL05xA are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction.

Because of the extremely high input impedance and resulting low-bias current requirements, the TL05x and TL05xA are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets easily can exceed bias current requirements and cause degradation in system performance. It is good practice to include guard rings around inputs (see Figure 83). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input.

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

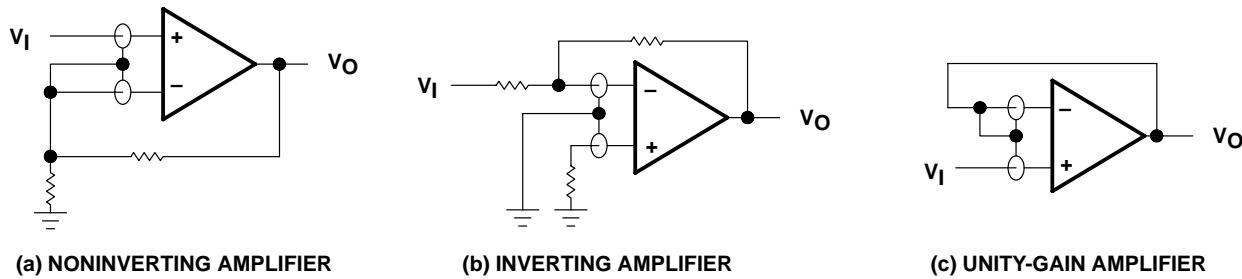


Figure 83. Use of Guard Rings

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input-bias current requirements of the TL05x and TL05xA result in a very low current noise. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω .

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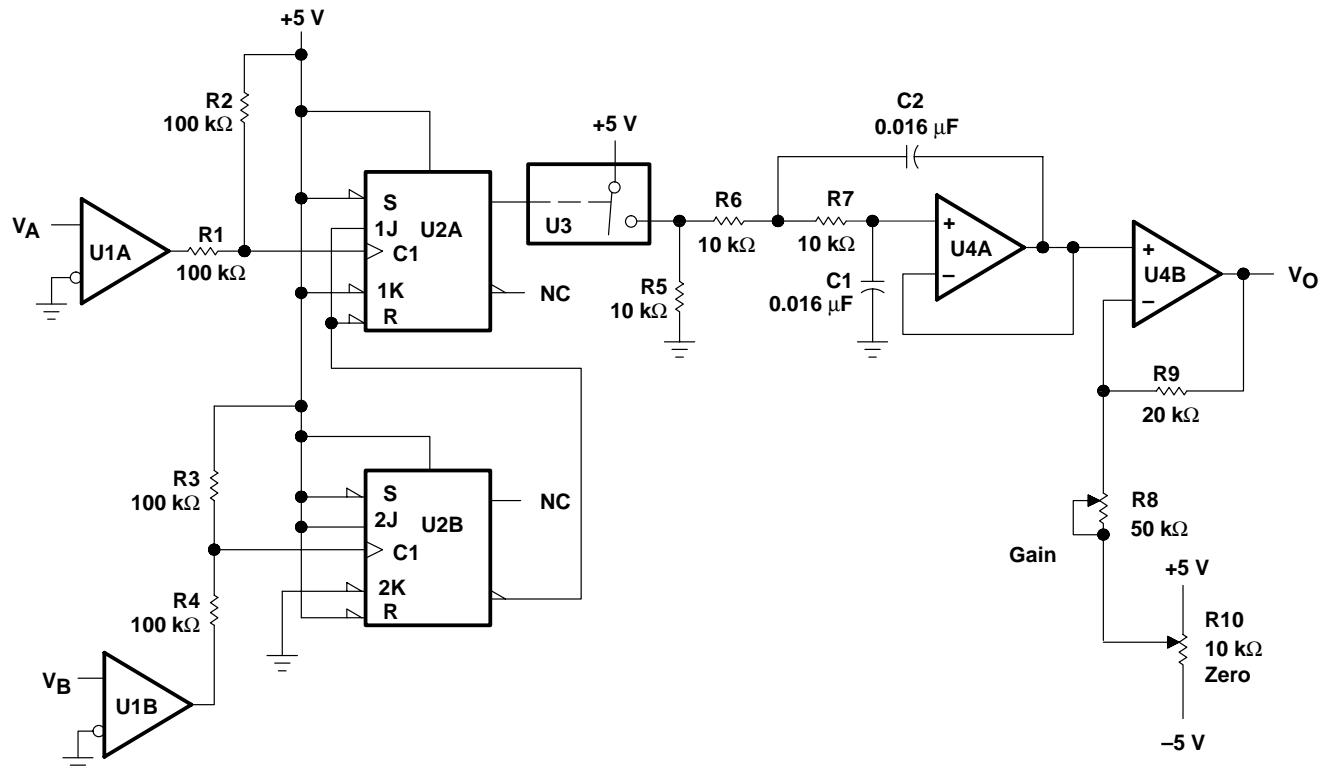
phase meter

The phase meter in Figure 84 produces an output voltage of 10 mV per degree of phase delay between the two input signals V_A and V_B . The reference signal V_A must be the same frequency as V_B . The TLC3702 comparators (U1) convert these two input sine waves into ± 5 -V square waves. Then, R1 and R4 provide level shifting prior to the SN74HC109 dual J-K flip flops.

Flip-flop U2B is connected as a toggle flip-flop and generates a square wave at one-half the frequency of V_B . Flip-flop U2A also produces a square wave at one-half the input frequency. The pulse duration of U2A varies from zero to one-half the period, where zero corresponds to zero phase delay between V_A and V_B and one-half the period corresponds to V_B lagging V_A by 360 degrees.

The output pulse from U2A causes the TLC4066 (U3) switch to charge the TL05x (U4) integrator capacitors C1 and C2. As the phase delay approaches 360 degrees, the output of U4A approximates a square wave, and U2A has an output of almost 2.5 V. U4B acts as a noninverting amplifier with a gain of 1.44 in order to scale the 0- to 2.5-V integrator output to a 0- to 3.6-V output range.

R8 and R10 provide output gain and zero-level calibration. This circuit operates over a 100-Hz to 10-kHz frequency range.



NOTE A: U1 = TLC3702; $V_{CC\pm} = \pm 5$ V
 U2 = SN74HC109
 U3 = TLC4066
 U4, U5 = TL05x; $V_{CC\pm} = \pm 5$ V

Figure 84. Phase Meter

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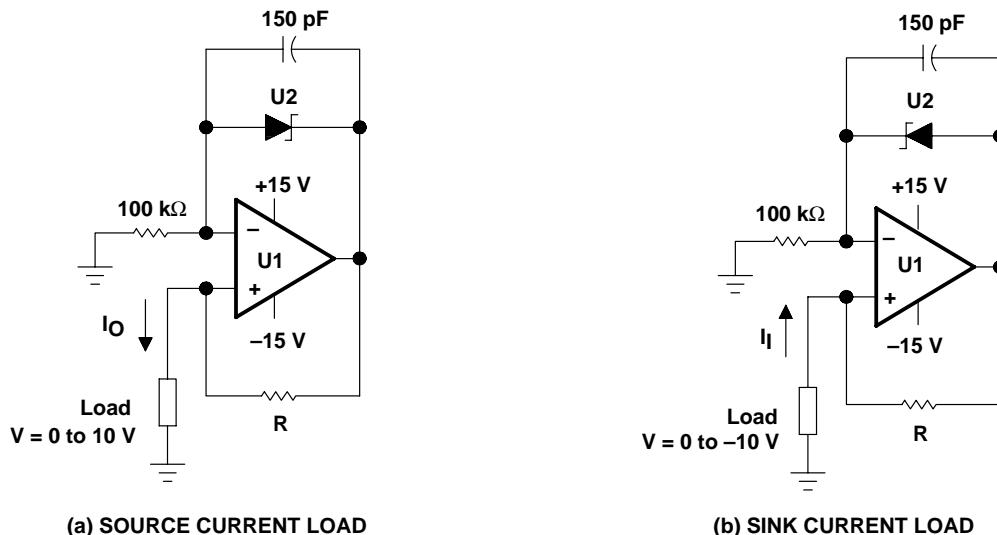
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APPLICATION INFORMATION

precision constant-current source over temperature

A precision current source (see Figure 85) benefits from the high input impedance and stability of Texas Instruments enhanced-JFET process. A low-current shunt regulator maintains 2.5 V between the inverting input and the output of the TL05x. The negative feedback then forces 2.5 V across the current-setting resistor R; therefore, the current to the load simply is 2.5 V divided by R.

Possible choices for the shunt regulator include the LT1004, LT1009, and LM385. If the regulator's cathode connects to the operational amplifier output, this circuit sources load current. Similarly, if the cathode connects to the inverting input, the circuit sinks current from the load. To minimize output current change with temperature, R should be a metal film resistor with a low temperature coefficient. Also, this circuit must be operated with split-voltage supplies.



NOTE A: U1 = 1/2 TL05x
 U2 = LM385, LT1004, or LT1009 voltage reference
 $I = \frac{2.5}{R}$, R = Low-temperature-coefficient metal-film resistor

Figure 85. Precision Constant-Current Source

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APPLICATION INFORMATION

instrumentation amplifier with adjustable gain/null

The instrumentation amplifier in Figure 86 benefits greatly from the high input impedance and stable input offset voltage of the TL05xA. Amplifiers U1A, U1B, and U2A form the actual instrumentation amplifier, while U2B provides offset null. Potentiometer R1 provides gain adjustment. With $R1 = 2\text{ k}\Omega$, the circuit gain equals 100, while with $R1 = 200\text{ k}\Omega$, the circuit gain equals two. The following equation shows the instrumentation amplifier gain as a function of R1:

$$A_V = 1 + \left(\frac{R2 + R3}{R1} \right)$$

Readjusting the offset null is necessary when the circuit gain is changed. If U2B is needed for another application, R7 can be terminated at ground. The low input offset voltage of the TL05xA minimizes the dc error of the circuit. For best matching, all resistors should be one-percent tolerance. The matching between R4, R5, R6, and R7 controls the CMRR of this application.

The following equation shows the output voltages when the input voltage equals zero. This dc error can be nulled by adjusting the offset null potentiometer; however, any change in offset voltage over time or temperature also creates an error. To calculate the error from changes in offset, consider the three offset components in the equation as delta offsets, rather than initial offsets. The improved stability of Texas Instruments enhanced JFETs minimizes the error resulting from change in input offset voltage with time. Assuming V_I equals zero, V_O can be shown as a function of the offset voltage:

$$V_O = V_{IO2} \left[\left(1 + \frac{R3}{R1} \right) \left(\frac{R7}{R5 + R7} \right) \left(1 + \frac{R6}{R4} \right) + \frac{R2(R6)}{R1(R4)} \right] - V_{IO1} \left[\frac{R3}{R1} \left(\frac{R7}{R5 + R7} \right) \left(1 + \frac{R6}{R4} \right) + \frac{R6}{R4} \left(1 + \frac{R2}{R1} \right) \right] + V_{IO3} \left(1 + \frac{R6}{R4} \right)$$

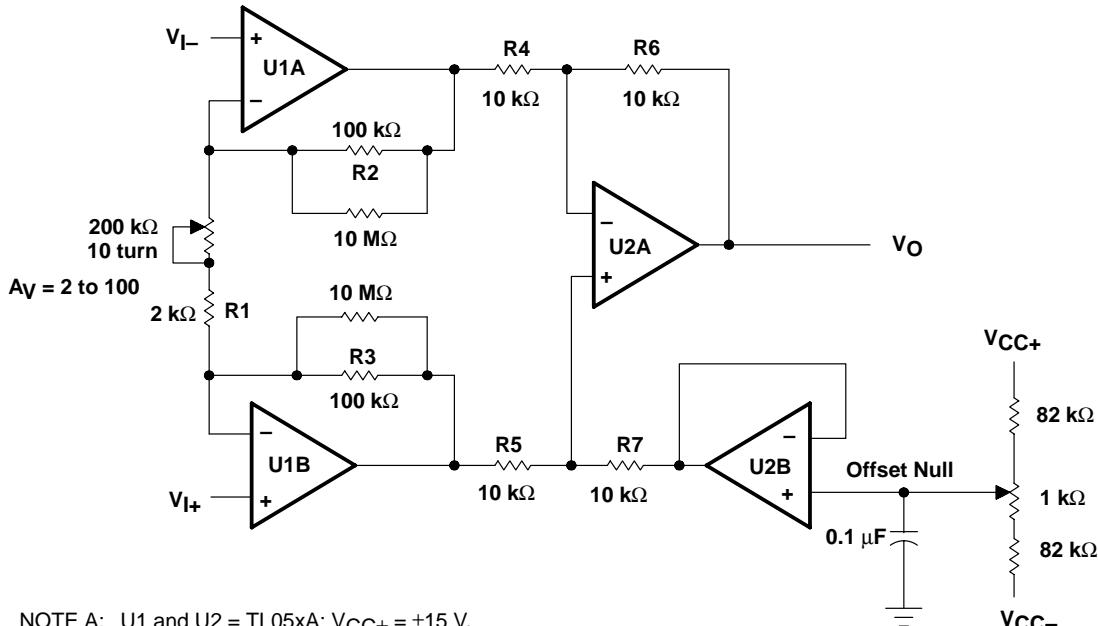


Figure 86. Instrumentation Amplifier

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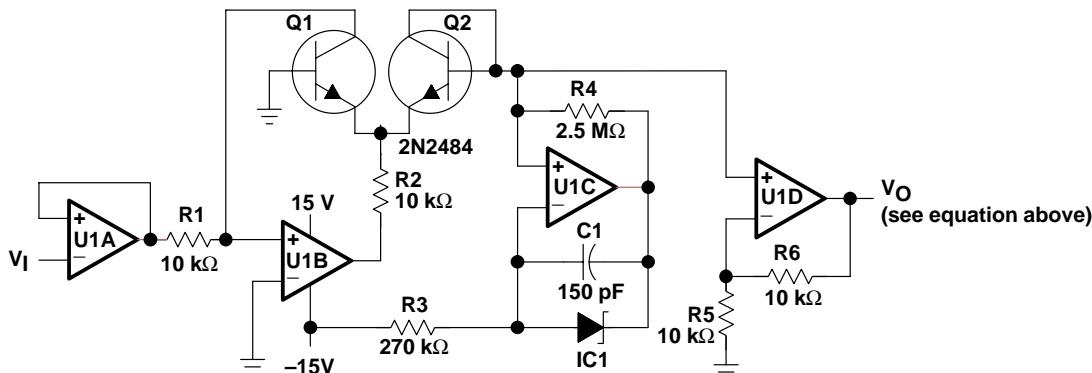
APPLICATION INFORMATION

high input impedance log amplifier

The low input offset voltage and high input impedance of the TL05xA creates a precision log amplifier (see Figure 87). IC1 is a 2.5-V, low-current precision, shunt regulator. Transistors Q1 and Q2 must be a closely matched npn pair. For best performance over temperature, R4 should be a metal-film resistor with a low temperature coefficient.

In this circuit, U1A serves as a high-impedance unity-gain buffer. Amplifier U1B converts the input voltage to a current through R1 and Q1. Amplifier U1C, IC1, and R4 form a 1- μ A temperature-stable current source that sets the base-emitter voltage of Q2. U1D amplifies the difference between the base-emitter voltage of Q1 and Q2 (see Figure 88). The output voltage is given by the following equation:

$$V_O = -\left[1 + \frac{R_6}{R_5}\right] \frac{kT}{q} \left[\ln \frac{V_I}{(R_1 \times 1 \times 10^{-6})} \right] \text{ where } k = 1.38 \times 10^{-23}, q = 1.602 \times 10^{-19}, \text{ and } T \text{ is Kelvin temperature}$$



NOTE A: U1A through U1D = TL05xA. IC1 = LM385, LT1004, or LT1009 voltage reference

Figure 87. Log Amplifier

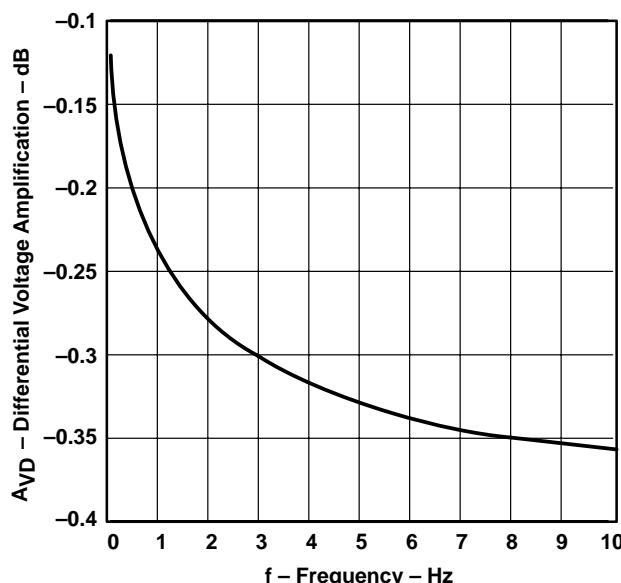


Figure 88. Output Voltage vs Input Voltage for Log Amplifier

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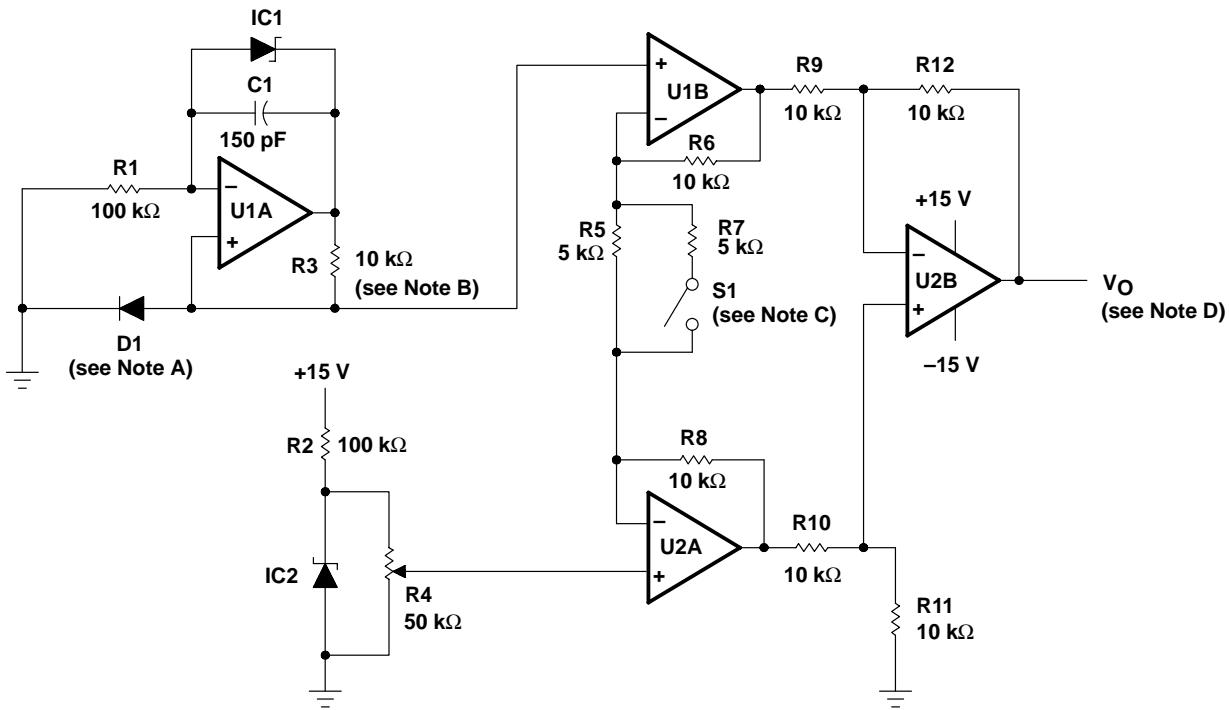
APPLICATION INFORMATION

analog thermometer

By combining a current source that does not vary over temperature with an instrumentation amplifier, a precise analog thermometer can be built (see Figure 89). Amplifier U1A and IC1 establish a constant current through the temperature-sensing diode D1. For this section of the circuit to operate correctly, the TL05x must use split supplies, and R3 must be a metal-film resistor with a low temperature coefficient.

The temperature-sensitive voltage from the diode is compared to a temperature-stable voltage reference set by IC2. R4 should be adjusted to provide the correct output voltage when the diode is at a known temperature. Although this potentiometer resistance varies with temperature, the divider ratio of the potentiometer remains constant.

Amplifiers U1B, U2A, and U2B form the instrumentation amplifier that converts the difference between the diode and reference voltage to a voltage proportional to the temperature. With switch S1 closed, the amplifier gain equals 5 and the output voltage is proportional to temperature in degrees Celsius. With S1 open, the amplifier gain is 9 and the output is proportional to temperature in degrees Fahrenheit. Every time S1 is changed, R4 must be recalibrated. By setting S1 correctly, the output voltage equals 10 mV per degree (C or F).



- NOTES:
- A. Temperature-sensing diode $\approx (-2 \text{ mV}/^\circ\text{C})$
 - B. Metal-film resistor (low temperature coefficient)
 - C. Switch open for ${}^\circ\text{F}$ and closed for ${}^\circ\text{C}$
 - D. $V_O \propto$ temperature; $10 \text{ mV}/^\circ\text{C}$ or $10 \text{ mV}/^\circ\text{F}$
 - E. U1, U2 = TL05x. IC1, IC2 = LM385, LT1004, or LT1009 voltage reference

Figure 89. Analog Thermometer

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APPLICATION INFORMATION

voltage-ratio-to-dB converter

The application in Figure 90 measures the amplitude ratio of two signals, then converts the ratio to decibels (see Figure 91). The output voltage provides a resolution of 100 mV/dB. The two inputs can be either dc or sinusoidal ac signals. When using ac signals, both signals should be the same frequency or output glitches will occur. For measuring two input signals of different frequencies, extra filtering should be added after the rectifiers.

The circuit contains three low-offset TL05xA devices. Two of these devices provide the rectification and logarithmic conversion of the inputs. The third TL05xA forms an instrumentation amplifier. The stage performing the logarithmic conversion also requires two well-matched npn transistors.

The input signal first passes through a high-impedance unity-gain buffer U1A (U2A). Then U1B (U2B) rectifies the input signal at a gain of 0.5, and U1C (U2C) provides a noninverting gain of 2, so that the system gain is still one. U1D (U2D), R6 (R13), and Q1 (Q2) perform the logarithmic conversion of the rectified input signal. The instrumentation amplifier formed by U3A, U3B, U3D scales the difference of the two logarithmic voltages by a gain of 33.6. As a result, the output voltage equals 100 mV/dB. The 1-k Ω potentiometer on the input of U3C calibrates the zero-dB reference level. The following equations are used to derive the relationship between the input voltage ratio, expressed in decibels, and the output voltage.

$$X \text{ dB} = 20 \log \left[\frac{V_A}{V_B} \right] = 20 \left[\frac{\ln(V_A) - \ln(V_B)}{\ln(10)} \right]$$
$$X \text{ dB} = 8.686 \left[\ln(V_A) - \ln(V_B) \right]$$
$$V_{BE(Q1)} = \frac{kT}{q} \ln \left[\frac{V_A}{R \times I_S} \right] \quad V_{BE(Q2)} = \frac{kT}{q} \ln \left[\frac{V_B}{R \times I_S} \right]$$
$$\Delta V_{BE} = V_{BE(Q1)} - V_{BE(Q2)} = \frac{kT}{q} \left[\ln(V_A) - \ln(V_B) \right]$$
$$X \text{ dB} = \frac{8.686}{kT/q} \left[V_{BE(Q1)} - V_{BE(Q2)} \right] = 336 \left[V_{BE(Q1)} - V_{BE(Q2)} \right] \text{ at } 25^\circ\text{C}$$

where

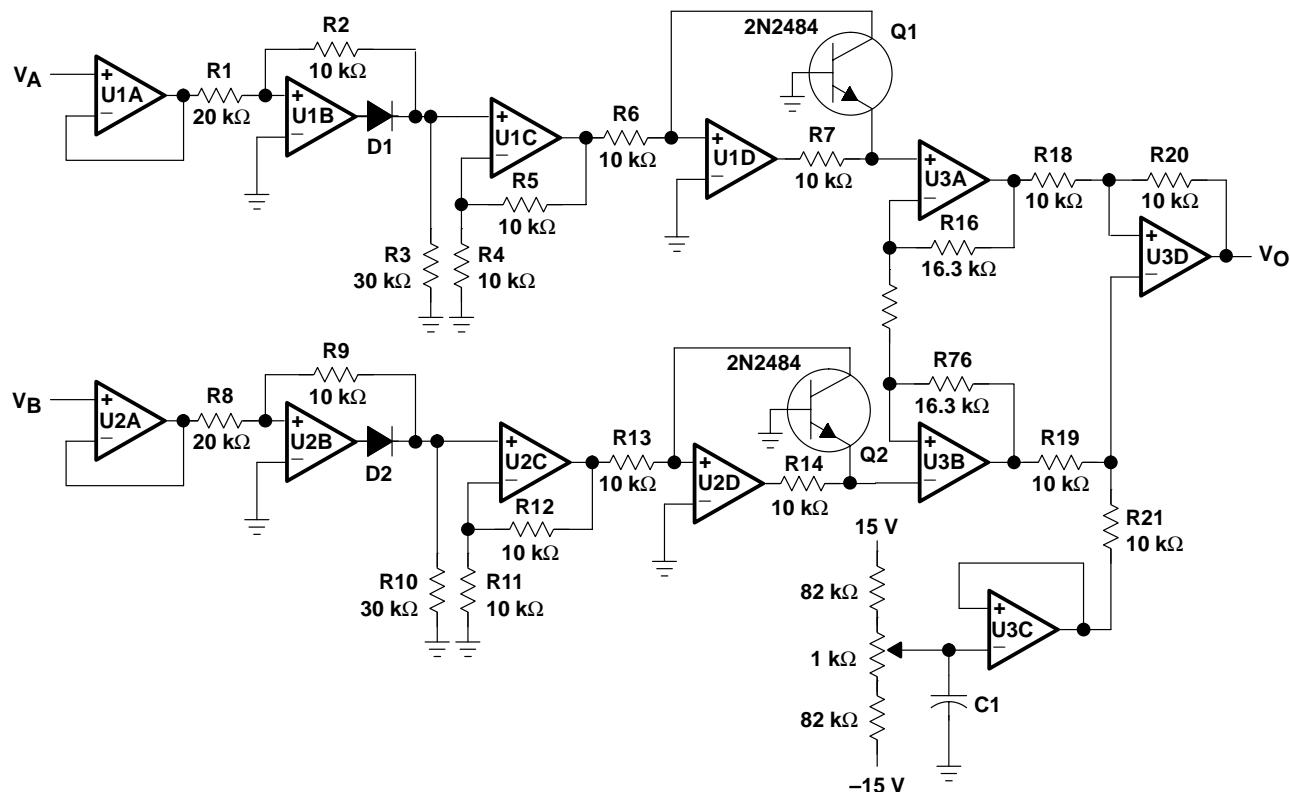
$$k = 1.38 \times 10^{-23}, \quad q = 1.602 \times 10^{-19}, \quad \text{and T is Kelvin temperature}$$

This gives a resolution of 1 V/dB. Therefore, the gain of the instrumentation amplifier is set at 33.6 to obtain 100 mV/dB.

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NOTE A: U1A through U3D = TL05xA, $V_{CC\pm} = \pm 15$ V. D1 and D2 = 1N914.

Figure 90. Voltage Ratio-to-dB Converter

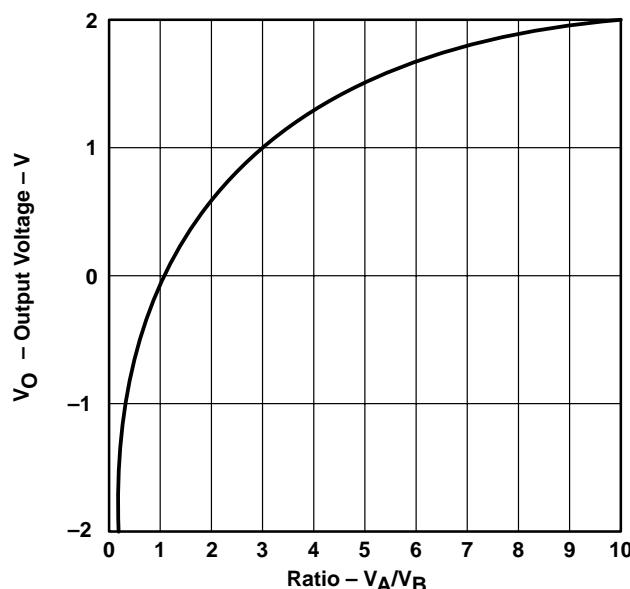


Figure 91. Output Voltage vs the Ratio of the Input Voltages for Voltage-to-dB Converter

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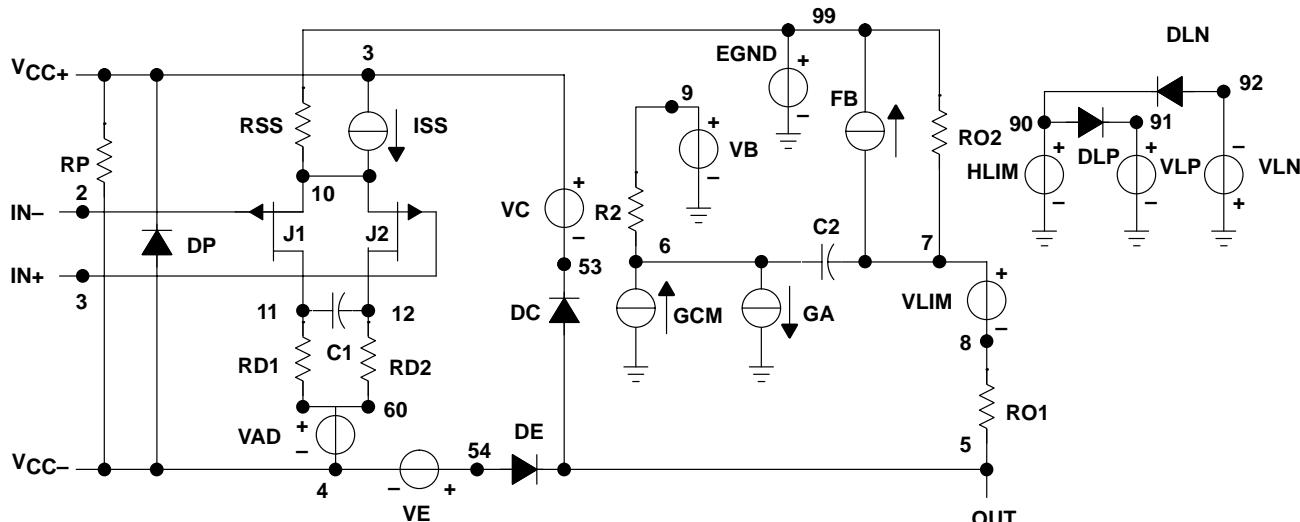
APPLICATION INFORMATION

macromodel information

Macromodel information provided was derived using Microsim *Parts*TM, the model-generation software used with Microsim *PSpice*TM. The Boyle macromodel (see Note 6 and subcircuit Figure 92) are generated using the TL05x typical electrical and operating characteristics at $T_A = 25^\circ\text{C}$. Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 6: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers", *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).



.SUBCKT TL05x 1 2 3 4 5				
C1	11	12	3.988E-12	
C2	6	7	15.00E-12	
DC	5	53	DX	
DE	54	5	DX	
DLP	90	91	DX	
DLN	92	90	DX	
DP	4	3	DX	
EGND99	0	POLY (2) (3,0) (4,0) 0 .5 .5		
FB	7	99	POLY (5) VB VC VE VLP	
+ VLN	0	2.875E6 -3E6 3E6 3E6 -3E6		
GA	6	0	11 12 292.2E-6	
GCM	0	6	10 99 6.542E-9	
ISS	3	10	DC 300.0E-6	
HLIM	90	0	VLM 1K	
J1	11	2	10 JX	
J2	12	1	10 JX	
R2	6	9	100.0E3	

RD1	4	11	3.422E3
RD2	4	12	3.422E3
R01	8	5	125
R02	7	99	125
RP	3	4	11.11E3
RSS	10	99	666.7E6
VB	9	0	DC 0
VC	3	53	DC 3
VE	54	4	DC 3.7
VLM	7	8	DC 0
VLP	91	0	DC 28
VLN	0	92	DC 28
.MODEL	DX D (IS=800.0E-18)		
.MODEL	JX PJF (IS=15.00E-12 BETA=185.2E-6		
+ VTO	=-.1)		
.ENDS			

Figure 92. Boyle Macromodel and Subcircuit

PSpice and Parts are trademarks of MicroSim Corporation.

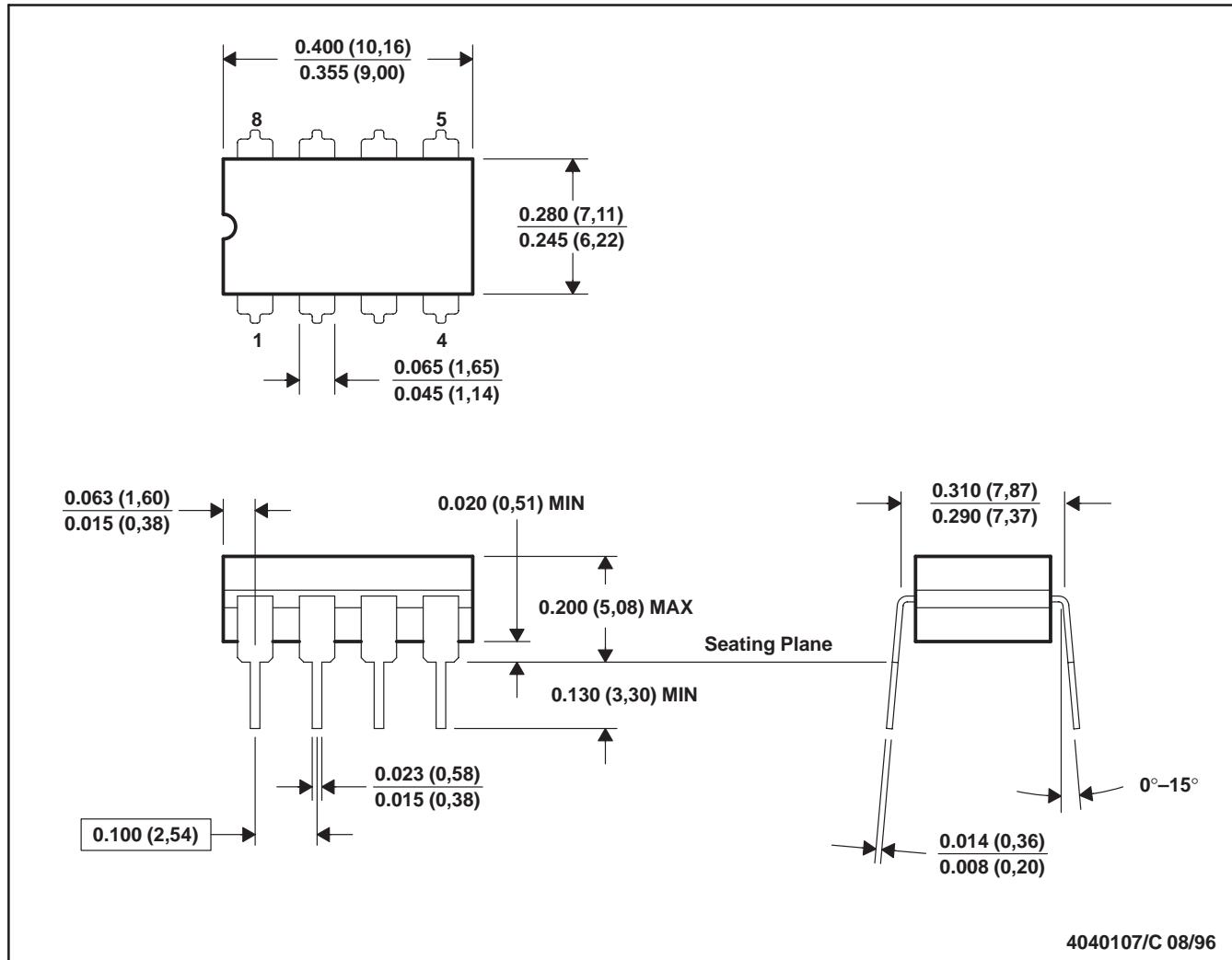
Macromodels, simulation models, or other models provided by TI, directly or indirectly, are not warranted by TI as fully representing all of the specification and operating characteristics of the semiconductor product to which the model relates.

MECHANICAL DATA

MCER001A – JANUARY 1995 – REVISED JANUARY 1997

JG (R-GDIP-T8)

CERAMIC DUAL-IN-LINE

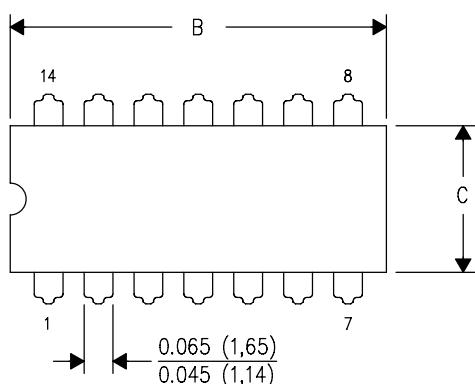


- NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. This package can be hermetically sealed with a ceramic lid using glass frit.
D. Index point is provided on cap for terminal identification.
E. Falls within MIL STD 1835 GDIP1-T8

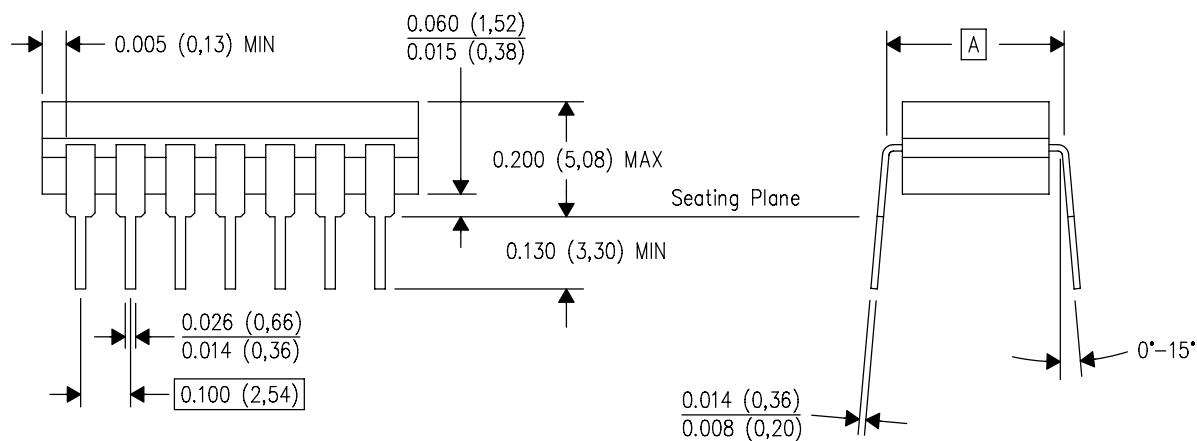
J (R-GDIP-T**)

14 LEADS SHOWN

CERAMIC DUAL IN-LINE PACKAGE



DIM \ PINS **	14	16	18	20
A	0.300 (7,62) BSC	0.300 (7,62) BSC	0.300 (7,62) BSC	0.300 (7,62) BSC
B MAX	0.785 (19,94)	.840 (21,34)	0.960 (24,38)	1.060 (26,92)
B MIN	—	—	—	—
C MAX	0.300 (7,62)	0.300 (7,62)	0.310 (7,87)	0.300 (7,62)
C MIN	0.245 (6,22)	0.245 (6,22)	0.220 (5,59)	0.245 (6,22)



4040083/F 03/03

- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. This package is hermetically sealed with a ceramic lid using glass frit.
 - D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.
 - E. Falls within MIL STD 1835 GDIP1-T14, GDIP1-T16, GDIP1-T18 and GDIP1-T20.

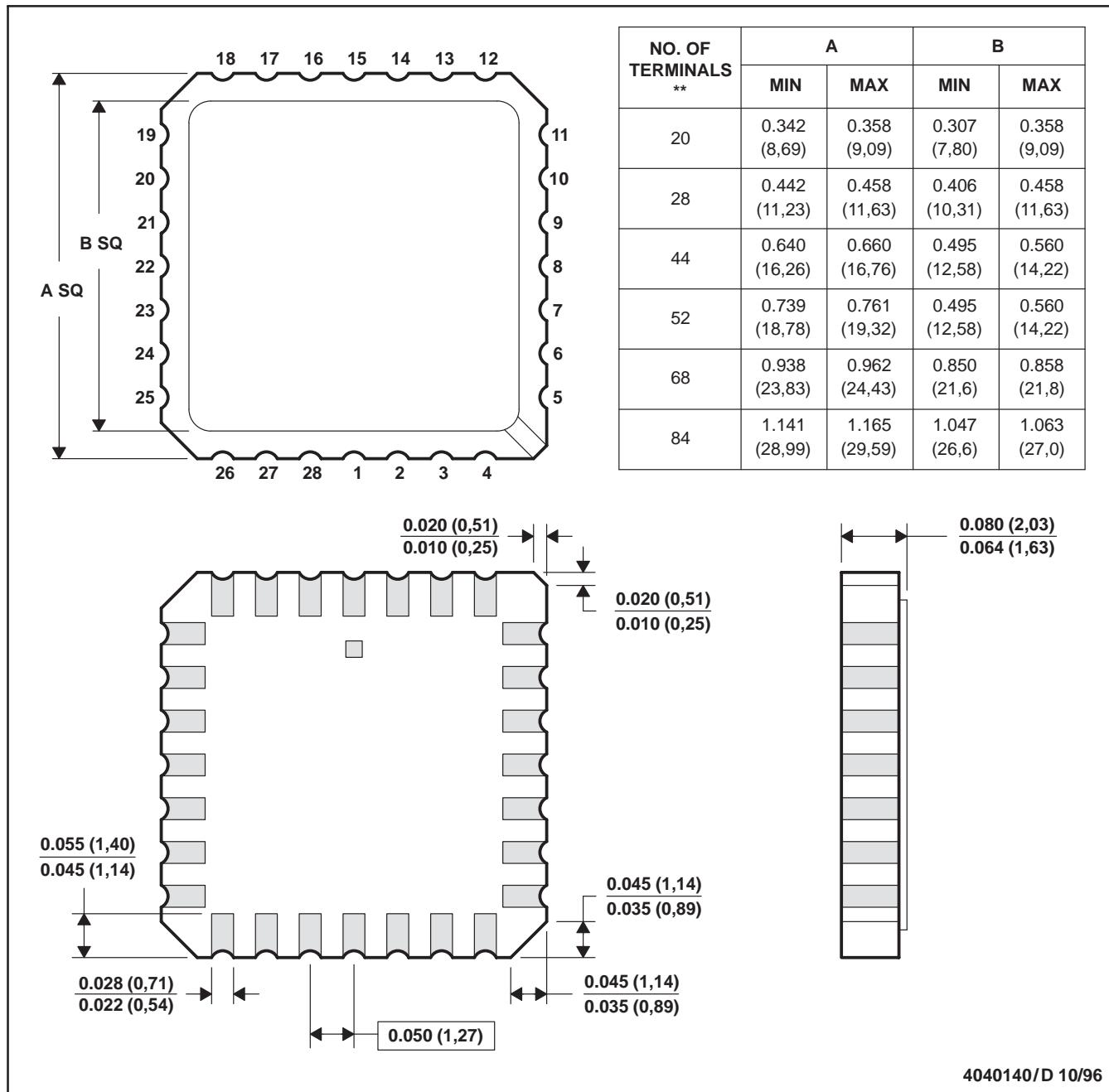
MECHANICAL DATA

MLCC006B – OCTOBER 1996

FK (S-CQCC-N)**

28 TERMINAL SHOWN

LEADLESS CERAMIC CHIP CARRIER



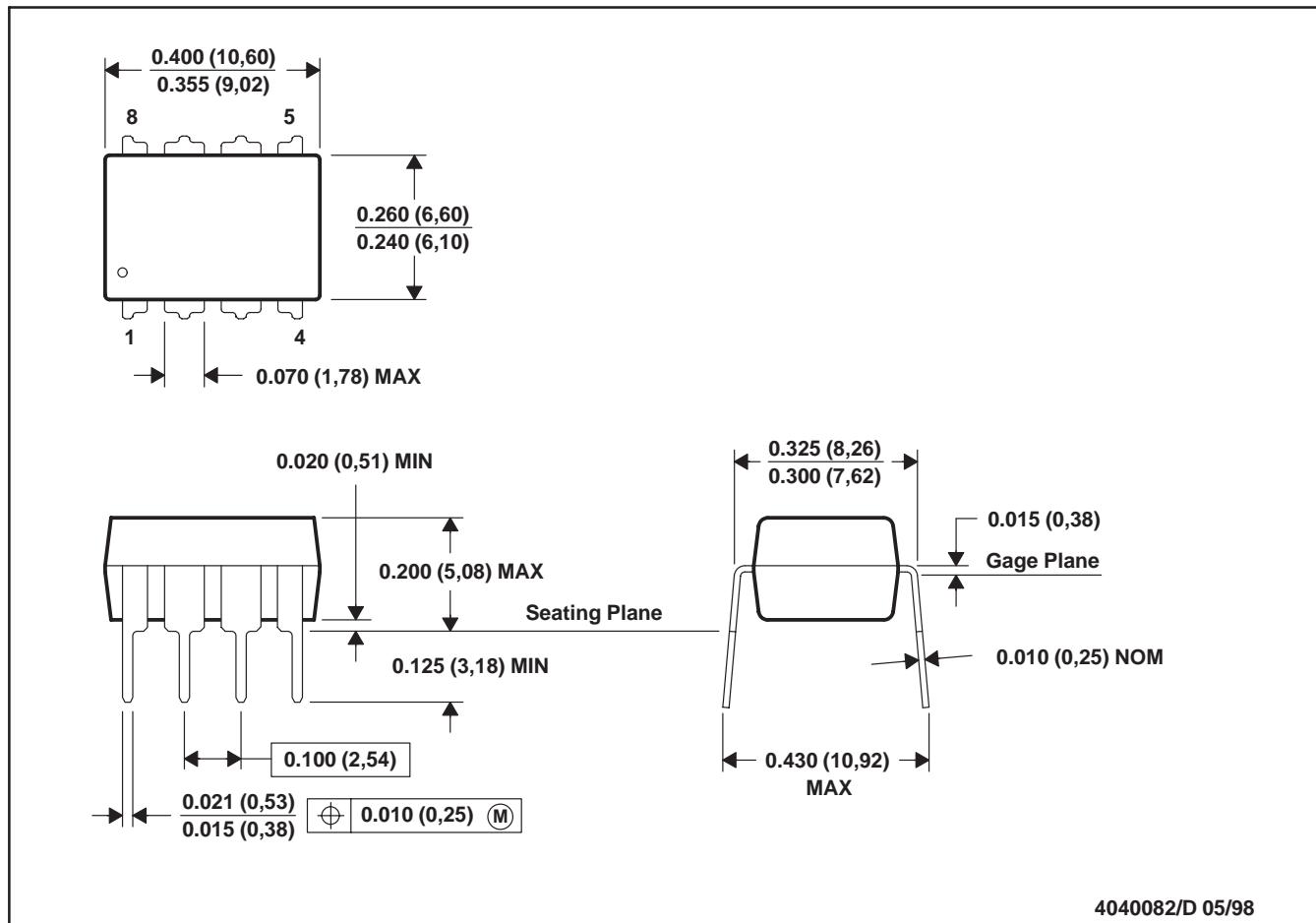
- NOTES:
- All linear dimensions are in inches (millimeters).
 - This drawing is subject to change without notice.
 - This package can be hermetically sealed with a metal lid.
 - The terminals are gold plated.
 - Falls within JEDEC MS-004

MECHANICAL DATA

MPDI001A – JANUARY 1995 – REVISED JUNE 1999

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE



- NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. Falls within JEDEC MS-001

For the latest package information, go to http://www.ti.com/sc/docs/package/pkg_info.htm

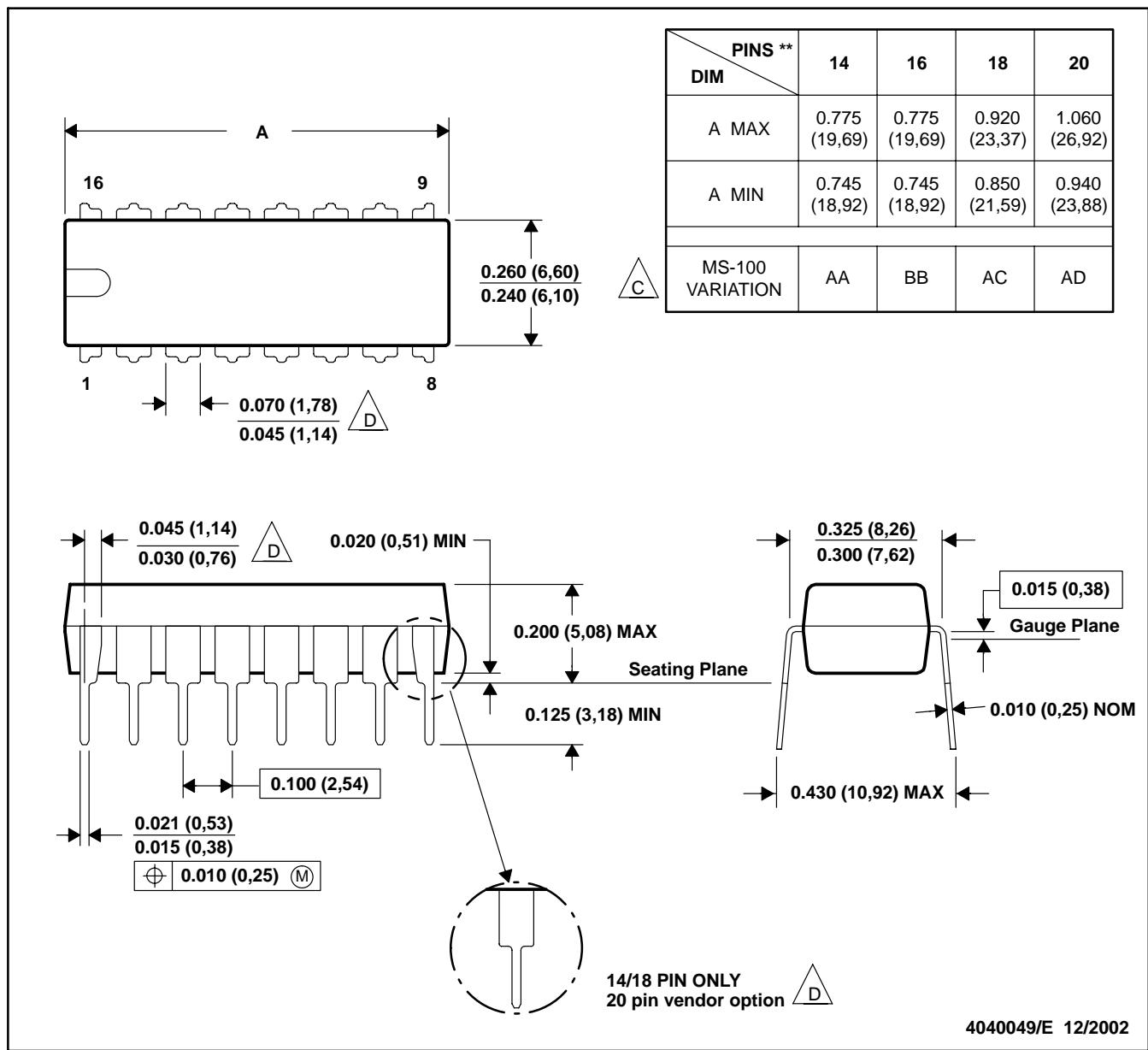
MECHANICAL

MPDI002C – JANUARY 1995 – REVISED DECEMBER 20002

N (R-PDIP-T**)

16 PINS SHOWN

PLASTIC DUAL-IN-LINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).

The 20 pin end lead shoulder width is a vendor option, either half or full width.

4040049/E 12/2002

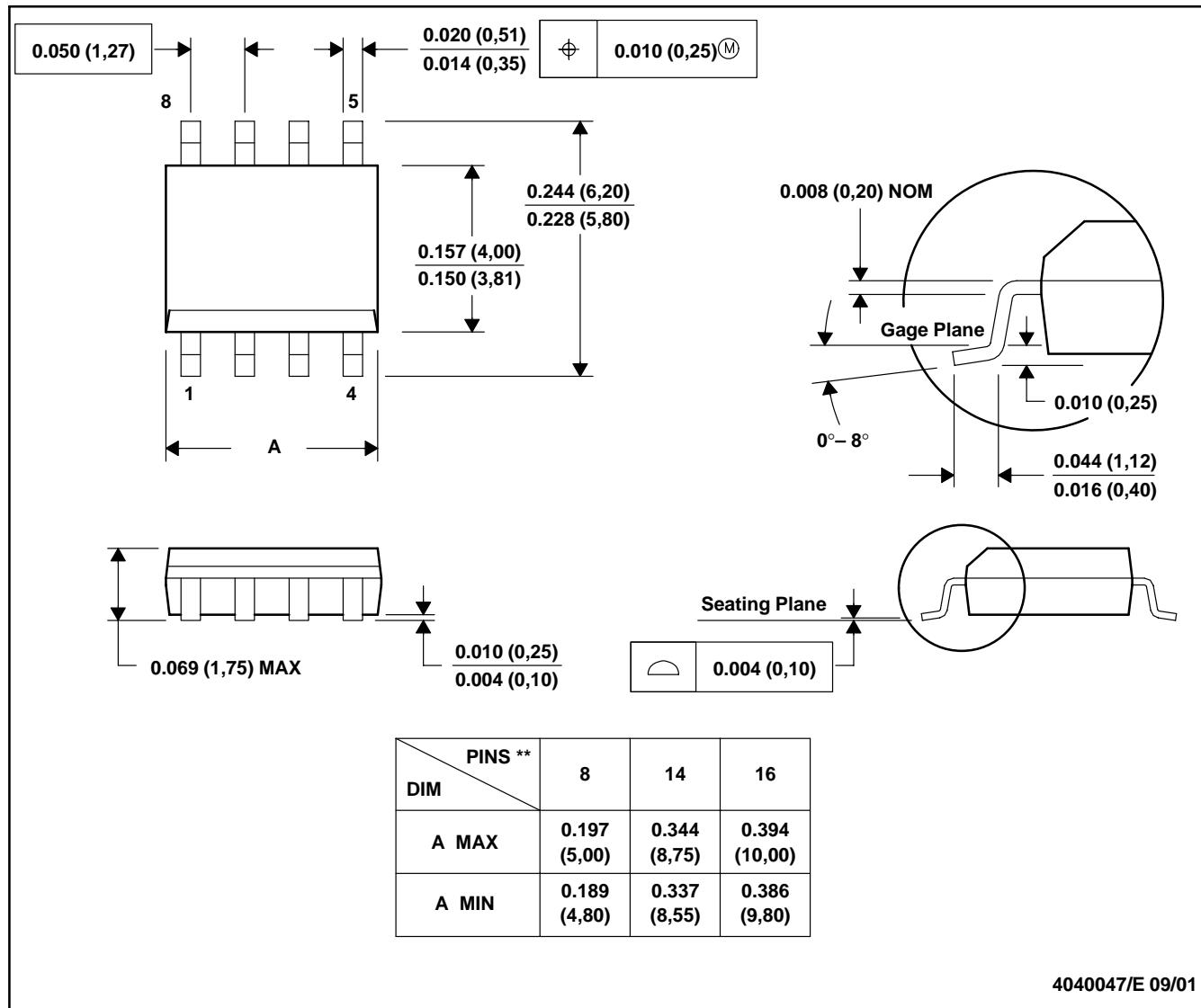
MECHANICAL DATA

MSOI002B – JANUARY 1995 – REVISED SEPTEMBER 2001

D (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

8 PINS SHOWN



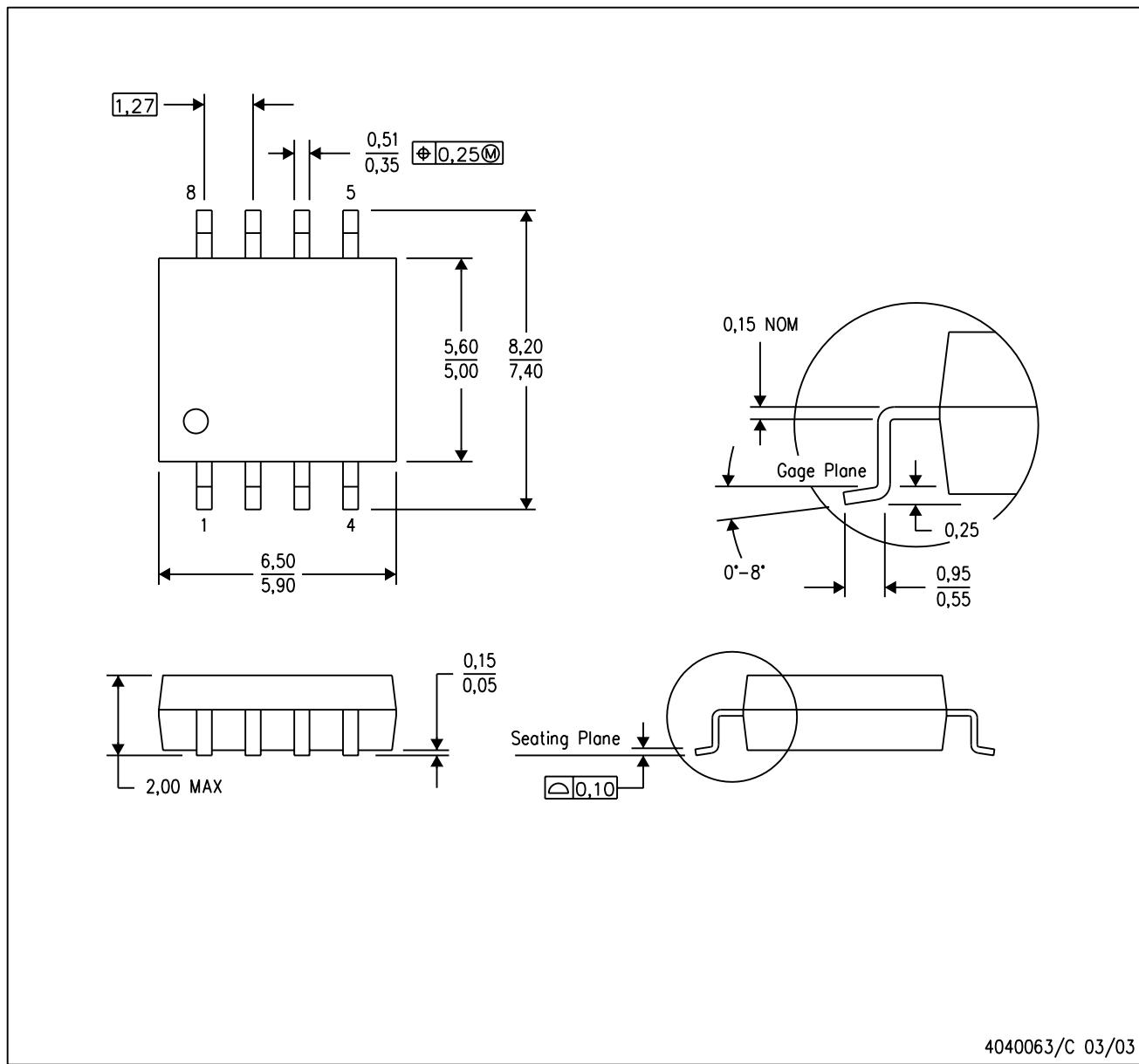
4040047/E 09/01

- NOTES:
- All linear dimensions are in inches (millimeters).
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0.15).
 - Falls within JEDEC MS-012

MECHANICAL DATA

PS (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



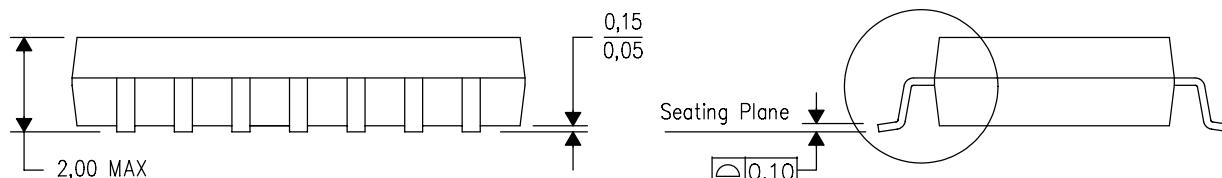
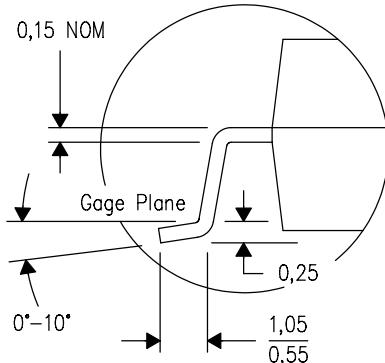
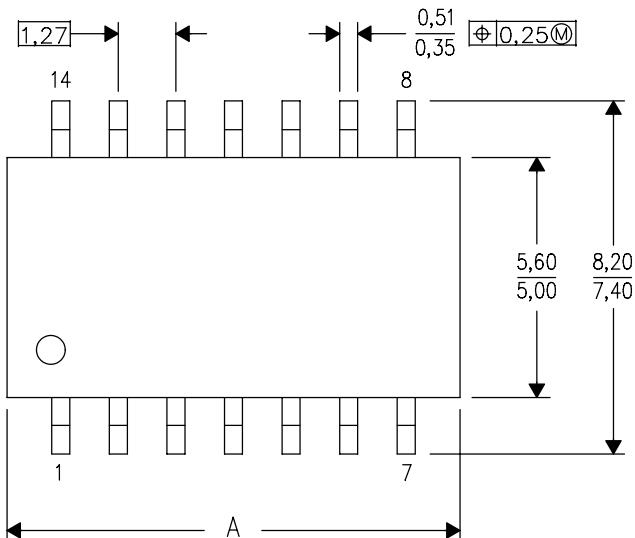
- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0.15.

MECHANICAL DATA

NS (R-PDSO-G)**

14-PINS SHOWN

PLASTIC SMALL-OUTLINE PACKAGE



PINS ** DIM	14	16	20	24
A MAX	10,50	10,50	12,90	15,30
A MIN	9,90	9,90	12,30	14,70

4040062/C 03/03

- NOTES: A. All linear dimensions are in millimeters.
 B. This drawing is subject to change without notice.
 C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

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