## Microchip

## PIC18F45J10 Family Data Sheet

## 28/40/44-Pin High-Performance, RISC Microcontrollers with nanoWatt Technology

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## 28/40/44-Pin High-Performance, RISC Microcontrollers with nanoWatt Technology

## Special Microcontroller Features:

- Operating voltage range: 2.0 V to 3.6 V
- 5.5 V tolerant input (digital pins only)
- On-chip 2.5 V regulator
- Low-power, high-speed CMOS Flash technology
- C compiler optimized architecture:
- Optional extended instruction set designed to optimize re-entrant code
- Priority levels for interrupts
- $8 \times 8$ Single-Cycle Hardware Multiplier
- Extended Watchdog Timer (WDT):
- Programmable period from 4 ms to 131 s
- Single-Supply In-Circuit Serial Programming ${ }^{\text {M }}$ (ICSP ${ }^{\text {TM }}$ ) via two pins
- In-Circuit Debug (ICD) with three Break points via two pins
- Power-Managed modes:
- Run: CPU on, peripherals on
- Idle: CPU off, peripherals on
- Sleep: CPU off, peripherals off


## Flexible Oscillator Structure:

- Two Crystal modes, up to 40 MHz
- Two External Clock modes, up to 40 MHz
- Internal 31 kHz oscillator
- Secondary oscillator using Timer1 @ 32 kHz
- Two-Speed Oscillator Start-up
- Fail-Safe Clock Monitor:
- Allows for safe shutdown if peripheral clock stops


## Peripheral Highlights:

- High-current sink/source $25 \mathrm{~mA} / 25 \mathrm{~mA}$ (PORTB and PORTC)
- Three programmable external interrupts
- Four input change interrupts
- One Capture/Compare/PWM (CCP) module
- One Enhanced Capture/Compare/PWM (ECCP) module:
- One, two or four PWM outputs
- Selectable polarity
- Programmable dead time
- Auto-Shutdown and Auto-Restart
- Two Master Synchronous Serial Port (MSSP) modules supporting 3 -wire $\mathrm{SPI}^{\mathrm{TM}}$ (all 4 modes) and $\mathrm{I}^{2} \mathrm{C}^{\text {M }}$ Master and Slave modes
- One Enhanced Addressable USART module:
- Supports RS-485, RS-232 and LIN 1.2
- Auto-Wake-up on Start bit
- Auto-Baud Detect
- 10-bit, up to 13-channel Analog-to-Digital Converter module (A/D):
- Auto-acquisition capability
- Conversion available during Sleep
- Self-calibration feature
- Dual analog comparators with input multiplexing

| Device | Program Memory |  | SRAM Data Memory (bytes) | I/O | 10-bit A/D (ch) | $\begin{aligned} & \text { CCP/ } \\ & \text { ECCP } \\ & \text { (PWM) } \end{aligned}$ | MSSP |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Flash (bytes) | \# Single-Word Instructions |  |  |  |  |  | SPI ${ }^{\text {Tm }}$ | Master $1^{2} C^{\text {tM }}$ |  |  |  |
| PIC18F24J10 | 16K | 8192 | 1024 | 21 | 10 | 2/0 | 1 | Y | Y | 1 | 2 | 1/2 |
| PIC18F25J10 | 32K | 16384 | 1024 | 21 | 10 | 2/0 | 1 | Y | Y | 1 | 2 | 1/2 |
| PIC18F44J10 | 16K | 8192 | 1024 | 32 | 13 | 1/1 | 2 | Y | Y | 1 | 2 | 1/2 |
| PIC18F45J10 | 32K | 16384 | 1024 | 32 | 13 | 1/1 | 2 | Y | Y | 1 | 2 | 1/2 |

## PIC18F45J10 FAMILY

## Pin Diagrams

## 28-Pin SPDIP, SOIC, SSOP (300 MIL)



* Pin feature is dependent on device configuration.


## 28-Pin QFN



* Pin feature is dependent on device configuration.


## Pin Diagrams (Continued)

## 40-Pin PDIP (600 MIL)



* Pin feature is dependent on device configuration.


## 44-Pin QFN



* Pin feature is dependent on device configuration.

Pin Diagrams (Continued)

## 44-Pin TQFP



* Pin feature is dependent on device configuration.


## PIC18F45J10 FAMILY

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## PIC18F45J10 FAMILY

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### 1.0 DEVICE OVERVIEW

This document contains device specific information for the following devices:

- PIC18F24J10
- PIC18LF24J10
- PIC18F25J10
- PIC18LF25J10
- PIC18F44J10
- PIC18LF44J10
- PIC18F45J10
- PIC18LF45J10

This family offers the advantages of all PIC18 microcontrollers - namely, high computational performance at an economical price. The PIC18F45J10 family introduces design enhancements that make these microcontrollers a logical choice for many high-performance, power sensitive applications.

### 1.1 New Core Features

### 1.1.1 nanoWatt TECHNOLOGY

All of the devices in the PIC18F45J10 family incorporate a range of features that can significantly reduce power consumption during operation. Key items include:

- Alternate Run Modes: By clocking the controller from the Timer1 source or the internal oscillator block, power consumption during code execution can be reduced by as much as $90 \%$.
- Multiple Idle Modes: The controller can also run with its CPU core disabled but the peripherals still active. In these states, power consumption can be reduced even further, to as little as $4 \%$ of normal operation requirements.
- On-the-fly Mode Switching: The power-managed modes are invoked by user code during operation, allowing the user to incorporate power-saving ideas into their application's software design.
- Low Consumption in Key Modules: The power requirements for both Timer1 and the Watchdog Timer are minimized. See Section 23.0 "Electrical Characteristics" for values.


### 1.1.2 MULTIPLE OSCILLATOR OPTIONS AND FEATURES

All of the devices in the PIC18F45J10 family offer three different oscillator options. These include:

- One Crystal mode, using crystals or ceramic resonators
- One External Clock mode
- INTRC source (approximately 31 kHz )

Besides its availability as a clock source, the internal oscillator block provides a stable reference source that gives the family additional features for robust operation:

- Fail-Safe Clock Monitor: This option constantly monitors the main clock source against a reference signal provided by the internal oscillator. If a clock failure occurs, the controller is switched to the internal oscillator block, allowing for continued low-speed operation or a safe application shutdown.
- Two-Speed Start-up: This option allows the internal oscillator to serve as the clock source from Power-on Reset, or wake-up from Sleep mode, until the primary clock source is available.


### 1.2 Other Special Features

- Communications: The PIC18F45J10 family incorporates a range of serial communication peripherals, including 1 independent Enhanced USART and 2 Master SSP modules capable of both SPI and $\mathrm{I}^{2} \mathrm{C}$ (Master and Slave) modes of operation. Also, one of the general purpose I/O ports can be reconfigured as an 8-bit Parallel Slave Port for direct processor-to-processor communications.
- Self-programmability: These devices can write to their own program memory spaces under internal software control. By using a bootloader routine, it becomes possible to create an application that can update itself in the field.
- Extended Instruction Set: The PIC18F45J10 family introduces an optional extension to the PIC18 instruction set, which adds 8 new instructions and an Indexed Addressing mode. This extension, enabled as a device configuration option, has been specifically designed to optimize re-entrant application code originally developed in high-level languages, such as C.
- Enhanced CCP module: In PWM mode, this module provides 1,2 or 4 modulated outputs for controlling half-bridge and full-bridge drivers. Other features include Auto-Shutdown, for disabling PWM outputs on interrupt or other select conditions and Auto-Restart, to reactivate outputs once the condition has cleared.
- Enhanced Addressable USART: This serial communication module is capable of standard RS-232 operation and provides support for the LIN bus protocol. Other enhancements include automatic baud rate detection and a 16-bit Baud Rate Generator for improved resolution.
- 10-bit A/D Converter: This module incorporates programmable acquisition time, allowing for a channel to be selected and a conversion to be initiated without waiting for a sampling period and thus, reduce code overhead.
- Extended Watchdog Timer (WDT): This enhanced version incorporates a 16-bit prescaler, allowing an extended time-out range that is stable across operating voltage and temperature. See Section 23.0 "Electrical Characteristics" for time-out periods.


### 1.3 Details on Individual Family Members

Devices in the PIC18F45J10 family are available in 28-pin and 40/44-pin packages. Block diagrams for the two groups are shown in Figure 1-1 and Figure 1-2.
The devices are differentiated from each other in five ways:

1. Flash program memory (16 Kbytes for PIC18F24J10/44J10 devices and 32 Kbytes for PIC18F25J10/45J10).
2. $A / D$ channels ( 10 for 28 -pin devices, 13 for 40/44-pin devices).
3. I/O ports ( 3 bidirectional ports on 28 -pin devices, 5 bidirectional ports on 40/44-pin devices).
4. CCP and Enhanced CCP implementation (28-pin devices have 2 standard CCP modules, 40/44-pin devices have one standard CCP module and one ECCP module).
5. Parallel Slave Port (present only on 40/44-pin devices).
6. One MSSP module for PIC18F24J10/25J10 devices and 2 MSSP modules for PIC18F44J10/45J10 devices
7. Parts designated with an "F" part number (i.e., PIC18F25J10) have a minimum VDD of 2.8 volts, whereas parts designated with an "LF" part number (i.e., PIC18LF25J10) can operate between 2.0-3.6 volts on VDD; however, Vddcore should never exceed Vdd.
All other features for devices in this family are identical. These are summarized in Table 1-1.

The pinouts for all devices are listed in Table 1-2 and Table 1-3.

The PIC18F45J10 family of devices provides an on-chip voltage regulator to supply the correct voltage levels to the core. Parts designated with an "F" part number (such as PIC18F25J10) have the voltage regulator enabled. These parts can run from 2.7-3.6 volts on VDD but should have the VdDCORE pin connected to Vss through a lowESR capacitor. Parts designated with an "LF" part number (such as PIC18LF24J10) do not enable the voltage regulator. An external supply of 2.0-2.7 Volts has to be supplied to the VdDCORE pin while 2.0-3.6 Volts can be supplied to VDD ( VdDCORE should never exceed VDD). See Section 20.3 "On-Chip Voltage Regulator" for more details about the internal voltage regulator.

## PIC18F45J10 FAMILY

TABLE 1-1: DEVICE FEATURES

| Features | PIC18F24J10 | PIC18F25J10 | PIC18F44J10 | PIC18F45J10 |
| :---: | :---: | :---: | :---: | :---: |
| Operating Frequency | DC - 40 MHz | DC - 40 MHz | DC - 40 MHz | DC - 40 MHz |
| Program Memory (Bytes) | 16384 | 32768 | 16384 | 32768 |
| Program Memory (Instructions) | 8192 | 16384 | 8192 | 16384 |
| Data Memory (Bytes) | 768 | 1536 | 768 | 1536 |
| Interrupt Sources | 19 | 19 | 20 | 20 |
| I/O Ports | Ports A, B, C | Ports A, B, C | Ports A, B, C, D, E | Ports A, B, C, D, E |
| Timers | 3 | 3 | 3 | 3 |
| Capture/Compare/PWM Modules | 2 | 2 | 1 | 1 |
| Enhanced Capture/Compare/PWM Modules | 0 | 0 | 1 | 1 |
| Serial Communications | MSSP, <br> Enhanced USART | MSSP, <br> Enhanced USART | MSSP, <br> Enhanced USART | MSSP, <br> Enhanced USART |
| Parallel Communications (PSP) | No | No | Yes | Yes |
| 10-bit Analog-to-Digital Module | 10 Input Channels | 10 Input Channels | 13 Input Channels | 13 Input Channels |
| Resets (and Delays) | POR, BOR ${ }^{(1)}$ RESET Instruction, Stack Full, Stack Underflow (PWRT, OST), MCLR, WDT | POR, BOR ${ }^{(1)}$ RESET Instruction, Stack Full, Stack Underflow (PWRT, OST), MCLR, WDT | POR, BOR ${ }^{(1)}$, RESET Instruction, Stack Full, Stack Underflow (PWRT, OST), MCLR, WDT | POR, $\mathrm{BOR}^{(1)}$ RESET Instruction, Stack Full, Stack Underflow (PWRT, OST), MCLR, WDT |
| Programmable Brown-out Reset | Yes | Yes | Yes | Yes |
| Instruction Set | 75 Instructions; 83 with Extended Instruction Set enabled | 75 Instructions; 83 with Extended Instruction Set enabled | 75 Instructions; 83 with Extended Instruction Set enabled | 75 Instructions; 83 with Extended Instruction Set enabled |
| Packages | 28-pin SPDIP <br> 28-pin SOIC <br> 28-pin SSOP <br> 28-pin QFN | 28-pin SPDIP <br> 28-pin SOIC <br> 28-pin SSOP <br> 28-pin QFN | 40-pin PDIP <br> 44-pin QFN <br> 44-pin TQFP | 40-pin PDIP <br> 44-pin QFN <br> 44-pin TQFP |

Note 1: BOR is not available in PIC18LF2XJ10/4XJ10 devices.

FIGURE 1-1: PIC18F24J10/25J10 (28-PIN) BLOCK DIAGRAM


FIGURE 1-2: PIC18F44J10/45J10 (40/44-PIN) BLOCK DIAGRAM


Note 1: CCP2 is multiplexed with RC1 when configuration bit CCP2MX is set, or RB3 when CCP2MX is not set.
2: BOR is not available in PIC18LF2XJ10/4XJ10 devices.

## PIC18F45J10 FAMILY

TABLE 1-2: PIC18F24J10/25J10 PINOUT I/O DESCRIPTIONS

| Pin Name | Pin Number |  | Pin <br> Type | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | SPDIP, SOIC, SSOP | QFN |  |  |  |
| $\frac{\overline{\mathrm{MCLR}}}{\overline{\mathrm{MCLR}}}$ | 1 | 26 | 1 | ST | Master Clear (input) or programming voltage (input). Master Clear (Reset) input. This pin is an active-low Reset to the device. |
| $\begin{array}{\|c} \hline \text { OSC1/CLKI } \\ \text { OSC1 } \\ \text { CLKI } \end{array}$ | 9 | 6 | I | CMOS | Oscillator crystal or external clock input. Oscillator crystal input or external clock source input. External clock source input. Always associated with pin function OSC1. See related OSC2/CLKO pins. |
| $\begin{array}{\|c} \hline \text { OSC2/CLKO } \\ \text { OSC2 } \\ \text { CLKO } \end{array}$ | 10 | 7 | $\begin{aligned} & 0 \\ & 0 \end{aligned}$ | $\begin{aligned} & - \\ & - \end{aligned}$ | Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode. In EC mode, OSC2 pin outputs CLKO which has $1 / 4$ the frequency of OSC1 and denotes the instruction cycle rate. |
| $\text { Legend: } \begin{aligned} \mathrm{TTL} & = \\ \mathrm{ST} & = \\ \mathrm{O} & = \end{aligned}$ | mpatible | input | with C | MOS leve | CMOS $=$ CMOS compatible input or output |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-2: PIC18F24J10/25J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  | $\begin{gathered} \text { Pin } \\ \text { Type } \end{gathered}$ | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { SPDIP, } \\ & \text { SOIC, } \\ & \text { SSOP } \end{aligned}$ | QFN |  |  |  |
| $\begin{gathered} \text { RAO/ANO } \\ \text { RAO } \\ \text { ANO } \end{gathered}$ | 2 | 27 | $\begin{gathered} \text { I/O } \\ \text { I } \end{gathered}$ | TTL Analog | PORTA is a bidirectional I/O port. <br> Digital I/O. <br> Analog input 0. |
| $\begin{gathered} \text { RA1/AN1 } \\ \text { RA1 } \\ \text { AN1 } \end{gathered}$ | 3 | 28 | I/O |  | Digital I/O. <br> Analog input 1. |
| RA2/AN2/VREF-/CVREF <br> RA2 <br> AN2 <br> Vref- <br> CVref | 4 | 1 | $\begin{gathered} \text { I/O } \\ \text { I } \\ 1 \\ 0 \end{gathered}$ | TTL <br> Analog <br> Analog <br> Analog | Digital I/O. <br> Analog input 2. <br> A/D reference voltage (low) input. Comparator reference voltage output. |
| ```RA3/AN3/VREF+ RA3 AN3 VREF+``` | 5 | 2 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I } \end{gathered}$ | TTL <br> Analog <br> Analog | Digital I/O. <br> Analog input 3. <br> A/D reference voltage (high) input. |
| $\begin{aligned} & \text { RA5/AN4/ } \overline{\text { SS1 }} / \mathrm{C} 2 O U T \\ & \text { RA5 } \\ & \text { AN4 } \\ & \overline{\text { SS1 }} \\ & \text { C2OUT } \end{aligned}$ | 7 | 4 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I } \\ 0 \end{gathered}$ | TTL <br> Analog <br> TTL | Digital I/O. <br> Analog input 4. <br> SPI ${ }^{\text {TM }}$ slave select input. <br> Comparator 2 output. |
| $\begin{aligned} \text { Legend: } \mathrm{TTL} & =\text { TTL compatible input } \\ \text { ST } & =\text { Schmitt Trigger input with CMOS levels } \\ \mathrm{O} & =\text { Output } \end{aligned}$ |  |  |  |  | $\begin{aligned} & \text { CMOS } \end{aligned}=\text { CMOS compatible input or output } \quad \begin{aligned} & \text { I } \\ & \\ & \\ & \text { P Input } \\ & \\ & \end{aligned}$ |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-2: PIC18F24J10/25J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

|  | Pin Number |  | $\begin{aligned} & \text { Pin } \\ & \text { Type } \end{aligned}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Pin Name | SPDIP, SOIC, SSOP | QFN |  | Buffer Type | Description |
|  |  |  |  |  | PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs. |
| RB0/INT0/FLT0/AN12 | 21 | 18 |  |  |  |
| RB0 |  |  | I/O | TTL | Digital I/O. |
| INTO |  |  | 1 | ST | External interrupt 0. |
| FLTO |  |  | 1 | ST | PWM Fault input for CCP1. |
| AN12 |  |  | 1 | Analog | Analog input 12. |
| RB1/INT1/AN10 | 22 | 19 |  |  |  |
| RB1 |  |  | I/O | TTL | Digital I/O. |
| INT1 |  |  | 1 | ST | External interrupt 1. |
| AN10 |  |  | 1 | Analog | Analog input 10. |
| RB2/INT2/AN8 | 23 | 20 |  |  |  |
| RB2 |  |  | I/O | TTL | Digital I/O. |
| INT2 |  |  | 1 | ST | External interrupt 2. |
| AN8 |  |  | 1 | Analog | Analog input 8. |
| RB3/AN9/CCP2 | 24 | 21 |  |  |  |
| RB3 |  |  | I/O | TTL | Digital I/O. |
| AN9 |  |  | 1 | Analog | Analog input 9. |
| CCP2 ${ }^{(1)}$ |  |  | I/O | ST | Capture 2 input/Compare 2 output/PWM 2 output. |
| RB4/KBIO/AN11 | 25 | 22 |  |  |  |
| RB4 |  |  | I/O | TTL | Digital I/O. |
| KBIO |  |  | 1 | TTL | Interrupt-on-change pin. |
| AN11 |  |  | I | Analog | Analog input 11. |
| RB5/KBI1/T0CKI/ | 26 | 23 |  |  |  |
| C1OUT |  |  | I/O | TTL | Digital I/O. |
| RB5 |  |  | 1 | TTL | Interrupt-on-change pin. |
| KBI1 |  |  | 1 | ST | Timer0 external clock input. |
| TOCKI |  |  | 0 | - | Comparator 1 output. |
| C1OUT |  |  |  |  |  |
| RB6/KBI2/PGC | 27 | 24 |  |  |  |
| RB6 |  |  | I/O | TTL | Digital I/O. |
| KBI2 |  |  | 1 | TTL | Interrupt-on-change pin. |
| PGC |  |  | I/O | ST | In-Circuit Debugger and ICSP ${ }^{\text {™ }}$ programming clock pin. |
| RB7/KBI3/PGD | 28 | 25 |  |  |  |
| RB7 |  |  | I/O | TTL | Digital I/O. |
| KBI3 |  |  | 1 | TTL | Interrupt-on-change pin. |
| PGD |  |  | I/O | ST | In-Circuit Debugger and ICSP programming data pin. |

Legend: TTL = TTL compatible input
CMOS = CMOS compatible input or output
ST = Schmitt Trigger input with CMOS levels I = Input
O = Output
$\mathrm{P} \quad=$ Power
Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-2: PIC18F24J10/25J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  | $\begin{gathered} \text { Pin } \\ \text { Type } \end{gathered}$ | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { SPDIP, } \\ & \text { SOIC, } \\ & \text { SSOP } \end{aligned}$ | QFN |  |  |  |
| $\begin{aligned} & \text { RC0/T1OSO/T1CKI } \\ & \text { RC0 } \\ & \text { T1OSO } \\ & \text { T1CKI } \end{aligned}$ | 11 | 8 | $\begin{gathered} \text { I/O } \\ \mathrm{O} \\ \mathrm{I} \end{gathered}$ | $\frac{S T}{-}$ | PORTC is a bidirectional I/O port. <br> Digital I/O. <br> Timer1 oscillator output. <br> Timer1 external clock input. |
| $\begin{gathered} \mathrm{RC} 1 / \mathrm{T} 1 \mathrm{OSI} / \mathrm{CCP} 2 \\ \mathrm{RC} 1 \\ \mathrm{~T} 10 \mathrm{OI} \\ \mathrm{CCP}^{(2)} \end{gathered}$ | 12 | 9 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I/O } \end{gathered}$ | ST Analog ST | Digital I/O. <br> Timer1 oscillator input. Capture 2 input/Compare 2 output/PWM 2 output. |
| $\begin{gathered} \mathrm{RC} 2 / \mathrm{CCP} 1 \\ \mathrm{RC} 2 \\ \mathrm{CCP} 1 \end{gathered}$ | 13 | 10 | $\begin{aligned} & \text { I/O } \\ & \text { I/O } \end{aligned}$ | $\begin{aligned} & \text { ST } \\ & \text { ST } \end{aligned}$ | Digital I/O. <br> Capture 1 input/Compare 1 output/PWM 1 output. |
| $\begin{aligned} & \text { RC3/SCK1/SCL1 } \\ & \text { RC3 } \\ & \text { SCK1 } \\ & \text { SCL1 } \end{aligned}$ | 14 | 11 | I/O <br> I/O <br> I/O | ST ST ST | Digital I/O. <br> Synchronous serial clock input/output for SPI ${ }^{\text {TM }}$ mode. Synchronous serial clock input/output for $\mathrm{I}^{2} \mathrm{C}^{\text {TM }}$ mode. |
| $\begin{aligned} & \text { RC4/SDI1/SDA1 } \\ & \text { RC4 } \\ & \text { SDI1 } \\ & \text { SDA1 } \end{aligned}$ | 15 | 12 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I/O } \end{gathered}$ | $\begin{aligned} & \text { ST } \\ & \text { ST } \\ & \text { ST } \end{aligned}$ | Digital I/O. <br> SPI data in. <br> $\mathrm{I}^{2} \mathrm{C}$ data $\mathrm{I} / \mathrm{O}$. |
| $\begin{gathered} \text { RC5/SDO1 } \\ \text { RC5 } \\ \text { SDO1 } \end{gathered}$ | 16 | 13 | $\begin{gathered} \text { I/O } \\ \mathrm{O} \end{gathered}$ | ST | Digital I/O. <br> SPI data out. |
| $\begin{gathered} \text { RC6/TX/CK } \\ \text { RC6 } \\ \text { TX } \\ \text { CK } \end{gathered}$ | 17 | 14 | $\begin{gathered} \text { I/O } \\ \mathrm{O} \\ \text { I/O } \end{gathered}$ | $\begin{aligned} & S T \\ & \frac{S T}{} \end{aligned}$ | Digital I/O. <br> EUSART asynchronous transmit. EUSART synchronous clock (see related RX/DT). |
| $\begin{aligned} & \text { RC7/RX/DT } \\ & \text { RC7 } \\ & \text { RX } \\ & \text { DT } \end{aligned}$ | 18 | 15 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I/O } \end{gathered}$ | ST <br> ST <br> ST | Digital I/O. <br> EUSART asynchronous receive. EUSART synchronous data (see related TX/CK). |
| Vss | 8,19 | 5,16 | P | - | Ground reference for logic and I/O pins. |
| Vdd | 20 | 17 | P | - | Positive supply for logic and I/O pins. |
| VDDCORE/VCAP Vddcore Vcap | 6 | 3 | $\begin{aligned} & P \\ & P \end{aligned}$ | - | Positive supply for logic and I/O pins. Ground reference for logic and I/O pins. |

Legend: TTL = TTL compatible input
CMOS = CMOS compatible input or output

| $S T=$ Schmitt Trigger input with CMOS levels |  | 1 |
| :--- | :--- | :--- |
|  | $=$ Input |  |
| $O=$ Output | $P$ | $=$ Power |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

## PIC18F45J10 FAMILY

TABLE 1-3: PIC18F44J10/45J10 PINOUT I/O DESCRIPTIONS

| Pin Name | Pin Number |  |  | $\begin{gathered} \text { Pin } \\ \text { Type } \end{gathered}$ | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PDIP | QFN | TQFP |  |  |  |
|  | 1 | 18 | 18 | 1 | ST | Master Clear (input) or programming voltage (input). Master Clear (Reset) input. This pin is an active-low Reset to the device. |
| $\begin{gathered} \text { OSC1/CLKI } \\ \text { OSC1 } \\ \text { CLKI } \end{gathered}$ | 13 | 32 | 30 | I | CMOS | Oscillator crystal or external clock input. Oscillator crystal input or external clock source input. External clock source input. Always associated with pin function OSC1. See related OSC2/CLKO pins. |
| $\begin{gathered} \hline \text { OSC2/CLKO } \\ \text { OSC2 } \\ \text { CLKO } \end{gathered}$ | 14 | 33 | 31 | $0$ $0$ | - | Oscillator crystal or clock output. <br> Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode. In RC mode, OSC2 pin outputs CLKO which has $1 / 4$ the frequency of OSC1 and denotes the instruction cycle rate. |
| Legend: TTL $=$ TTL compatible input  CMOS = CMOS compatible input or output |  |  |  |  |  |  |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-3: PIC18F44J10/45J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  |  | $\begin{gathered} \text { Pin } \\ \text { Type } \end{gathered}$ | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PDIP | QFN | TQFP |  |  |  |
| $\begin{gathered} \text { RAO/ANO } \\ \text { RAO } \\ \text { ANO } \end{gathered}$ | 2 | 19 | 19 | $\begin{gathered} \text { I/O } \\ \text { I } \end{gathered}$ | TTL <br> Analog | PORTA is a bidirectional I/O port. <br> Digital I/O. <br> Analog input 0. |
| $\begin{gathered} \text { RA1/AN1 } \\ \text { RA1 } \\ \text { AN1 } \end{gathered}$ | 3 | 20 | 20 | $\begin{gathered} \text { I/O } \\ \text { I } \end{gathered}$ | TTL Analog | Digital I/O. <br> Analog input 1. |
| RA2/AN2/VREF-/CVREF RA2 <br> AN2 <br> Vref- <br> CVReF | 4 | 21 | 21 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I } \\ 0 \end{gathered}$ | TTL <br> Analog <br> Analog <br> Analog | Digital I/O. <br> Analog input 2. A/D reference voltage (low) input. Comparator reference voltage output. |
| $\begin{aligned} & \text { RA3/AN3/VREF+ } \\ & \text { RA3 } \\ & \text { AN3 } \\ & \text { VREF+ } \end{aligned}$ | 5 | 22 | 22 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \hline \end{gathered}$ | TTL <br> Analog <br> Analog | Digital I/O. <br> Analog input 3. <br> A/D reference voltage (high) input. |
| $\begin{aligned} & \text { RA5/AN4/SS1/C2OUT } \\ & \text { RA5 } \\ & \text { AN4 } \\ & \overline{\text { SS1 }} \\ & \text { C2OUT } \end{aligned}$ | 7 | 24 | 24 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I } \\ 0 \end{gathered}$ | TTL <br> Analog TTL $\qquad$ | Digital I/O. <br> Analog input 4. <br> SPI ${ }^{\text {TM }}$ slave select input. <br> Comparator 2 output. |
| ```Legend: TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels O = Output``` |  |  |  |  |  | $\begin{aligned} \text { CMOS } & =\text { CMOS compatible input or output } \\ & =\text { Input } \\ & =\text { Power } \end{aligned}$ |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-3: PIC18F44J10/45J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  |  | Pin <br> Type | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PDIP | QFN | TQFP |  |  |  |
|  |  |  |  |  |  | PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs. |
| RB0/INT0/FLTO/AN12 | 33 | 9 | 8 |  |  |  |
| RB0 |  |  |  | I/O | TTL | Digital I/O. |
| INT0 |  |  |  | I | ST | External interrupt 0. |
| FLTO |  |  |  | 1 | ST | PWM Fault input for Enhanced CCP1. |
| AN12 |  |  |  | I | Analog | Analog input 12. |
| RB1/INT1/AN10 | 34 | 10 | 9 |  |  |  |
| RB1 |  |  |  | I/O | TTL | Digital I/O. |
| INT1 |  |  |  | I | ST | External interrupt 1. |
| AN10 |  |  |  | I | Analog | Analog input 10. |
| RB2/INT2/AN8 | 35 | 11 | 10 | I/O |  |  |
| RB2 <br> INT2 |  |  |  | I/O | $\begin{aligned} & \text { TTL } \\ & \text { ST } \end{aligned}$ | Digital I/O. <br> External interrupt 2. |
| AN8 |  |  |  | 1 | Analog | Analog input 8. |
| RB3/AN9/CCP2 | 36 | 12 | 11 |  |  |  |
| RB3 |  |  |  | I/O | TTL | Digital I/O. |
| AN9 |  |  |  | 1 | Analog | Analog input 9. |
| CCP2 ${ }^{(1)}$ |  |  |  | I/O | ST | Capture 2 input/Compare 2 output/PWM 2 output. |
| RB4/KBIO/AN11 | 37 | 14 | 14 |  |  |  |
| RB4 |  |  |  | I/O | TTL | Digital I/O. |
| KBIO |  |  |  | I | TTL | Interrupt-on-change pin. |
| AN11 |  |  |  | I | Analog | Analog input 11. |
| RB5/KBI1/C1OUT RB5 | 38 | 15 | 15 |  |  | Digital I/O |
| $\begin{aligned} & \text { RB5 } \\ & \text { KBI1 } \end{aligned}$ |  |  |  | 1/0 | TTL | Digital I/O. <br> Interrupt-on-change pin. |
| C1OUT |  |  |  | 0 | - | Comparator 1 output. |
| RB6/KBI2/PGC | 39 | 16 | 16 |  |  |  |
| RB6 |  |  |  | I/O | TTL | Digital I/O. |
| KBI2 |  |  |  | I | TTL | Interrupt-on-change pin. |
| PGC |  |  |  | I/O | ST | In-Circuit Debugger and ICSPTM programming clock pin. |
| RB7/KBI3/PGD | 40 | 17 | 17 |  |  |  |
| RB7 |  |  |  | I/O | TTL | Digital I/O. |
| KBI3 |  |  |  | 1 | TTL | Interrupt-on-change pin. |
| PGD |  |  |  | I/O | ST | In-Circuit Debugger and ICSP programming data pin. |
| Legend: TTL = TTL compatible input |  |  |  |  |  | MOS = CMOS compatible input or output |
| ST = Schmitt Trigger input with CMOS levels |  |  |  |  |  | = Input |
| $\mathrm{O}=$ Output |  |  |  |  |  | $=$ Power |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-3: PIC18F44J10/45J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  |  | $\begin{gathered} \text { Pin } \\ \text { Type } \\ \hline \end{gathered}$ | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PDIP | QFN | TQFP |  |  |  |
|  |  |  |  |  |  | PORTC is a bidirectional I/O port. |
| RC0/T1OSO/T1CKI | 15 | 34 | 32 |  |  |  |
| RC0 |  |  |  | I/O | ST | Digital I/O. |
| T1OSO |  |  |  | 0 | - | Timer1 oscillator output. |
| T1CKI |  |  |  | 1 | ST | Timer1 external clock input. |
| RC1/T1OSI/CCP2 | 16 | 35 | 35 |  |  |  |
| RC1 |  |  |  | I/O | ST | Digital I/O. |
| T1OSI |  |  |  | 1 | CMOS | Timer1 oscillator input. |
| CCP2 ${ }^{(2)}$ |  |  |  | I/O | ST | Capture 2 input/Compare 2 output/PWM 2 output. |
| RC2/CCP1/P1A | 17 | 36 | 36 |  |  |  |
| RC2 |  |  |  | I/O | ST | Digital I/O. |
| CCP1 |  |  |  | I/O | ST | Capture 1 input/Compare 1 output/PWM 1 output. |
| P1A |  |  |  | 0 | - | Enhanced CCP1 output. |
| RC3/SCK1/SCL1 | 18 | 37 | 37 |  |  |  |
| RC3 |  |  |  | I/O | ST | Digital I/O. |
| SCK1 |  |  |  | I/O | ST | Synchronous serial clock input/output for |
| SCL1 |  |  |  | I/O | ST | Synchronous serial clock input/output for $\mathrm{I}^{2} \mathrm{C}^{\text {TM }}$ mode. |
| RC4/SDI1/SDA1 | 23 | 42 | 42 |  |  |  |
| RC4 |  |  |  | I/O | ST | Digital I/O. |
| SDI1 |  |  |  | 1 | ST | SPI data in. |
| SDA1 |  |  |  | I/O | ST | $\mathrm{I}^{2} \mathrm{C}$ data I/O. |
| RC5/SDO1 | 24 | 43 | 43 |  |  |  |
| RC5 |  |  |  | I/O | ST | Digital I/O. |
| SDO1 |  |  |  | 0 | - | SPI data out. |
| RC6/TX/CK | 25 | 44 | 44 |  |  |  |
| RC6 |  |  |  | I/O | ST | Digital I/O. |
| TX |  |  |  | 0 |  | EUSART asynchronous transmit. |
| CK |  |  |  | I/O | ST | EUSART synchronous clock (see related RX/DT). |
| RC7/RX/DT | 26 | 1 | 1 |  |  |  |
| RC7 |  |  |  | I/O | ST | Digital I/O. |
| RX |  |  |  | 1 | ST | EUSART asynchronous receive. |
| DT |  |  |  | I/O | ST | EUSART synchronous data (see related TX/CK). |

$\begin{aligned} \text { Legend: TTL } & =\text { TTL compatible input } & \text { CMOS } & =\text { CMOS compatible input or output } \\ \text { ST } & =\text { Schmitt Trigger input with CMOS levels } & 1 & =\text { Input }\end{aligned}$ ST $=$ Schmitt Trigger input with CMOS levels $\quad$ I $=$ Input
O = Output
$\mathrm{P}=$ Power

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-3: PIC18F44J10/45J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  |  | $\begin{gathered} \text { Pin } \\ \text { Type } \\ \hline \end{gathered}$ | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PDIP | QFN | TQFP |  |  |  |
|  |  |  |  |  |  | PORTD is a bidirectional I/O port or a Parallel Slave Port (PSP) for interfacing to a microprocessor port. These pins have TTL input buffers when PSP module is enabled. |
| RD0/PSP0/SCK2/ SCL2 | 19 | 38 | 38 |  |  |  |
| RD0 |  |  |  | I/O | ST | Digital I/O. |
| PSP0 |  |  |  | I/O | TTL | Parallel Slave Port data. |
| SCK2 |  |  |  | I/O | ST | Synchronous serial clock input/output for SPITM mode. |
| SCL2 |  |  |  | I/O | ST | Synchronous serial clock input/output for I ${ }^{2} C^{\text {TM }}$ mode. |
| RD1/PSP1/SDI2/SDA2 | 20 | 39 | 39 |  |  |  |
| RD1 |  |  |  | I/O | ST | Digital I/O. |
| PSP1 |  |  |  | I/O | TTL | Parallel Slave Port data. |
| SDI2 |  |  |  | 1 | ST | SPI data in. |
| SDA2 |  |  |  | I/O | ST | $\mathrm{I}^{2} \mathrm{C}$ data I/O. |
| RD2/PSP2/SDO2 | 21 | 40 | 40 |  |  |  |
| RD2 |  |  |  | I/O | ST | Digital I/O. |
| PSP2 |  |  |  | I/O | TTL | Parallel Slave Port data. |
| SDO2 |  |  |  | 0 | - | SPI data out. |
| RD3/PSP3/SS2 | 22 | 41 | 41 |  |  |  |
| RD3 <br> PSP3 |  |  |  | I/O | ST | Digital I/O. |
| $\frac{\mathrm{PSP}}{\mathrm{SS} 2}$ |  |  |  | I/O | TTL | Parallel Slave Port data. |
|  |  |  |  |  |  |  |
| RD4/PSP4 | 27 | 2 | 2 |  |  |  |
| RD4 PSP4 |  |  |  | I/O | ST | Digital I/O. |
| PSP4 |  |  |  | I/O | TTL | Parallel Slave Port data. |
| RD5/PSP5/P1B <br> RD5 | 28 | 3 | 3 |  |  |  |
| RD5 |  |  |  | 1/0 | TTL | Parallel Slave Port data |
| P1B |  |  |  | 0 | - | Enhanced CCP1 output. |
| RD6/PSP6/P1C | 29 | 4 | 4 |  |  |  |
| RD6 |  |  |  | I/O | ST | Digital I/O. |
| PSP6 |  |  |  | I/O | TTL | Parallel Slave Port data. |
| P1C |  |  |  | 0 | - | Enhanced CCP1 output. |
| RD7/PSP7/P1D | 30 | 5 | 5 |  |  |  |
| RD7 |  |  |  | I/O | ST | Digital I/O. |
| PSP7 |  |  |  | I/O | TTL | Parallel Slave Port data. |
| P1D |  |  |  | 0 | - | Enhanced CCP1 output. |
| Legend: TTL $=$ TTL compatible input  CMOS = CMOS compatible input or output |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

TABLE 1-3: PIC18F44J10/45J10 PINOUT I/O DESCRIPTIONS (CONTINUED)

| Pin Name | Pin Number |  |  | $\begin{gathered} \text { Pin } \\ \text { Type } \end{gathered}$ | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PDIP | QFN | TQFP |  |  |  |
| $\begin{gathered} \mathrm{REO} / \overline{\mathrm{RD}} / \mathrm{AN} 5 \\ \frac{\mathrm{RE}}{\mathrm{RD}} \\ \mathrm{AN5} \end{gathered}$ | 8 | 25 | 25 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I } \end{gathered}$ | ST <br> TTL <br> Analog | PORTE is a bidirectional I/O port. <br> Digital I/O. <br> Read control for Parallel Slave Port (see also $\overline{\mathrm{WR}}$ and $\overline{\mathrm{CS}}$ pins). Analog input 5. |
| RE1/ $\overline{W R} /$ AN6 $\frac{R E 1}{\mathrm{WR}}$ <br> AN6 | 9 | 26 | 26 | $\begin{gathered} \text { I/O } \\ \text { I } \\ \text { I } \end{gathered}$ | ST <br> TTL <br> Analog | Digital I/O. <br> Write control for Parallel Slave Port (see $\overline{\mathrm{CS}}$ and $\overline{\mathrm{RD}}$ pins). <br> Analog input 6. |
| $\begin{aligned} & \mathrm{RE} 2 / \overline{\mathrm{CS}} / \mathrm{AN} 7 \\ & \frac{\mathrm{RE} 2}{\mathrm{CS}} \\ & \mathrm{AN7} \\ & \hline \end{aligned}$ | 10 | 27 | 27 | I/O I | ST <br> TTL <br> Analog | Digital I/O. <br> Chip Select control for Parallel Slave Port (see related $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WR}}$ pins). Analog input 7. |
| Vss | 12, 31 | $\begin{gathered} 6,30, \\ 31 \end{gathered}$ | 6,29 | P | - | Ground reference for logic and I/O pins. |
| VDD | 11, 32 | $\begin{gathered} 7,8, \\ 28,29 \end{gathered}$ | 7, 28 | P | - | Positive supply for logic and I/O pins. |
| VDDCore/Vcap <br> Vddcore <br> Vcap | 6 | 23 | 23 | $\begin{aligned} & P \\ & P \end{aligned}$ | - | Positive supply for logic and I/O pins. Ground reference for logic and I/O pins. |
| NC | - | 13 | $\begin{array}{\|l\|} \hline 12,13, \\ 33,34 \end{array}$ | - | - | No connect. |
| Legend: TTL $=$ TTL compatible input  CMOS = CMOS compatible input or output |  |  |  |  |  |  |

Note 1: Default assignment for CCP2 when configuration bit CCP2MX is set.
2: Alternate assignment for CCP2 when configuration bit CCP2MX is cleared.

## PIC18F45J10 FAMILY

NOTES:

### 2.0 OSCILLATOR CONFIGURATIONS

### 2.1 Oscillator Types

The PIC18F45J10 family of devices can be operated in five different oscillator modes:

1. HS High-Speed Crystal/Resonator
2. HSPLL High-Speed Crystal/Resonator with Software PLL Control
3. EC External Clock with Fosc/4 Output
4. ECPLL External Clock with Software PLL Control
5. INTRC Internal 31 kHz Oscillator

Four of these are selected by the user by programming the FOSC2:FOSC0 configuration bits. The fifth mode (INTRC) may be invoked under software control; it can also be configured as the default mode on device Resets.

### 2.2 Crystal Oscillator/Ceramic Resonators (HS Modes)

In HS or HSPLL Oscillator modes, a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation. Figure 2-1 shows the pin connections.

The oscillator design requires the use of a parallel cut crystal.
Note: Use of a series cut crystal may give a frequency out of the crystal manufacturer's specifications.

FIGURE 2-1: CRYSTAL/CERAMIC RESONATOR OPERATION (HS OR HSPLL CONFIGURATION)


Note 1: See Table 2-1 and Table 2-2 for initial values of C1 and C2.
2: A series resistor (Rs) may be required for AT strip cut crystals.
3: RF varies with the oscillator mode chosen.

TABLE 2-1: CAPACITOR SELECTION FOR CERAMIC RESONATORS

| Typical Capacitor Values Used: |  |  |  |
| :---: | :---: | :---: | :---: |
| Mode | Freq. | OSC1 | OSC2 |
| HS | 8.0 MHz | 27 pF | 27 pF |
|  | 16.0 MHz | 22 pF | 22 pF |

Capacitor values are for design guidance only.
These capacitors were tested with the resonators listed below for basic start-up and operation. These values are not optimized.
Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application.
See the notes following Table 2-2 for additional information.

Resonators Used:

| 4.0 MHz |
| :---: |
| 8.0 MHz |
| 16.0 MHz |

TABLE 2-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

| Osc Type | Crystal <br> Freq. | Typical Capacitor Values <br> Tested: |  |
| :---: | :---: | :---: | :---: |
|  |  | $\mathbf{C 1}$ | $\mathbf{C 2}$ |
|  | 4 MHz | 27 pF | 27 pF |
|  | 8 MHz | 22 pF | 22 pF |
|  | 20 MHz | 15 pF | 15 pF |

Capacitor values are for design guidance only.
These capacitors were tested with the crystals listed below for basic start-up and operation. These values are not optimized.
Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application.
See the notes following this table for additional information.

| Crystals Used: |
| :---: |
| 4 MHz |
| 8 MHz |
| 20 MHz |

Note 1: Higher capacitance increases the stability of oscillator but also increases the start-up time.
2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
3: Rs may be required to avoid overdriving crystals with low drive level specification.
4: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

### 2.3 External Clock Input (EC Modes)

The EC and ECPLL Oscillator modes require an external clock source to be connected to the OSC1 pin. There is no oscillator start-up time required after a Power-on Reset or after an exit from Sleep mode.

In the EC Oscillator mode, the oscillator frequency divided by 4 is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure $2-2$ shows the pin connections for the EC Oscillator mode.

FIGURE 2-2: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)


An external clock source may also be connected to the OSC1 pin in the HS mode, as shown in Figure 2-3. In this configuration, the divide-by-4 output on OSC2 is not available.

FIGURE 2-3: EXTERNAL CLOCK INPUT OPERATION (HS OSC CONFIGURATION)


### 2.4 PLL Frequency Multiplier

A Phase Locked Loop (PLL) circuit is provided as an option for users who want to use a lower frequency oscillator circuit, or to clock the device up to its highest rated frequency from a crystal oscillator. This may be useful for customers who are concerned with EMI due to high-frequency crystals, or users who require higher clock speeds from an internal oscillator. For these reasons, the HSPLL and ECPLL modes are available.

The HSPLL and ECPLL modes provide the ability to selectively run the device at 4 times the external oscillating source to produce frequencies up to 40 MHz . The PLL is enabled by setting the PLLEN bit in the OSCTUNE register (Register 2-1).

FIGURE 2-4: PLL BLOCK DIAGRAM


REGISTER 2-1: OSCTUNE: PLL CONTROL REGISTER

bit $7 \quad$ Unimplemented: Read as ' 0 '
bit 6 PLLEN: Frequency Multiplier PLL Enable bit ${ }^{(1)}$
$1=$ PLL enabled
$0=$ PLL disabled
Note 1: Available only for ECPLL and HSPLL oscillator configurations; otherwise, this bit is unavailable and read as ' 0 '.
bit 5-0 Unimplemented: Read as ' 0 '

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 2.5 Internal Oscillator Block

The PIC18F45J10 family of devices includes an internal oscillator source (INTRC) which provides a nominal 31 kHz output. The INTRC is enabled on device power-up and clocks the device during its configuration cycle until it enters operating mode. INTRC is also enabled if it is selected as the device clock source or if any of the following are enabled:

- Fail-Safe Clock Monitor
- Watchdog Timer
- Two-Speed Start-up

These features are discussed in greater detail in Section 20.0 "Special Features of the CPU".
The INTRC can also be optionally configured as the default clock source on device start-up by setting the FOSC2 configuration bit. This is discussed in Section 2.6.1 "Oscillator Control Register".

### 2.6 Clock Sources and Oscillator Switching

The PIC18F45J10 family includes a feature that allows the device clock source to be switched from the main oscillator to an alternate clock source. PIC18F45J10 family devices offer two alternate clock sources. When an alternate clock source is enabled, the various power-managed operating modes are available.
Essentially, there are three clock sources for these devices:

- Primary oscillators
- Secondary oscillators
- Internal oscillator

The primary oscillators include the External Crystal and Resonator modes and the External Clock modes. The particular mode is defined by the FOSC2:FOSC0 configuration bits. The details of these modes are covered earlier in this chapter.
The secondary oscillators are those external sources not connected to the OSC1 or OSC2 pins. These sources may continue to operate even after the controller is placed in a power-managed mode.
PIC18F45J10 family devices offer the Timer1 oscillator as a secondary oscillator. This oscillator, in all power-managed modes, is often the time base for functions such as a real-time clock.

Most often, a 32.768 kHz watch crystal is connected between the RC0/T1OSO/T13CKI and RC1/T1OSI pins. Loading capacitors are also connected from each pin to ground.
The Timer1 oscillator is discussed in greater detail in Section 11.3 "Timer1 Oscillator".

In addition to being a primary clock source, the internal oscillator is available as a power-managed mode clock source. The INTRC source is also used as the clock source for several special features, such as the WDT and Fail-Safe Clock Monitor.
The clock sources for the PIC18F45J10 family devices are shown in Figure 2-5. See Section 20.0 "Special Features of the CPU" for Configuration register details.

FIGURE 2-5: PIC18F45J10 FAMILY CLOCK DIAGRAM


### 2.6.1 OSCILLATOR CONTROL REGISTER

The OSCCON register (Register 2-2) controls several aspects of the device clock's operation, both in full power operation and in power-managed modes.
The System Clock Select bits, SCS1:SCS0, select the clock source. The available clock sources are the primary clock (defined by the FOSC2:FOSC0 configuration bits), the secondary clock (Timer1 oscillator) and the internal oscillator. The clock source changes after one or more of the bits are written to, following a brief clock transition interval.
The OSTS (OSCCON<3>) and T1RUN (T1CON<6>) bits indicate which clock source is currently providing the device clock. The OSTS bit indicates that the Oscillator Start-up Timer (OST) has timed out and the primary clock is providing the device clock in primary clock modes. The T1RUN bit indicates when the Timer1 oscillator is providing the device clock in secondary clock modes. In power-managed modes, only one of these bits will be set at any time. If neither of these bits are set, the INTRC is providing the clock, or the internal oscillator has just started and is not yet stable.
The IDLEN bit determines if the device goes into Sleep mode or one of the Idle modes when the SLEEP instruction is executed.
The use of the flag and control bits in the OSCCON register is discussed in more detail in Section 3.0 "Power-Managed Modes".

Note 1: The Timer1 oscillator must be enabled to select the secondary clock source. The Timer1 oscillator is enabled by setting the T1OSCEN bit in the Timer1 Control register ( $\mathrm{T} 1 \mathrm{CON}<3>$ ). If the Timer1 oscillator is not enabled, then any attempt to select a secondary clock source when executing a SLEEP instruction will be ignored.
2: It is recommended that the Timer1 oscillator be operating and stable before executing the SLEEP instruction or a very long delay may occur while the Timer1 oscillator starts.

### 2.6.1.1 System Clock Selection and the FOSC2 Configuration Bit

The SCS bits are cleared on all forms of Reset. In the device's default configuration, this means the primary oscillator defined by FOSC1:FOSC0 (that is, one of the HC or EC modes) is used as the primary clock source on device Resets.

The default clock configuration on Reset can be changed with the FOSC2 configuration bit. The effect of this bit is to set the clock source selected when SCS1:SCS0 $=00$. When FOSC2 = 1 (default), the oscillator source defined by FOSC1:FOSC0 is selected whenever SCS1:SCS0 $=00$. When FOSC2 $=0$, the INTRC oscillator is selected whenever SCS1:SCS2 $=00$. Because the SCS bits are cleared on Reset, the FOSC2 setting also changes the default oscillator mode on Reset.
Regardless of the setting of FOSC2, INTRC will always be enabled on device power-up. It will serve as the clock source until the device has loaded its configuration values from memory. It is at this point that the FOSC configuration bits are read and the oscillator selection of operational mode is made.
Note that either the primary clock or the internal oscillator will have two bit setting options, at any given time, depending on the setting of FOSC2.

### 2.6.2 OSCILLATOR TRANSITIONS

PIC18F45J10 family devices contain circuitry to prevent clock "glitches" when switching between clock sources. A short pause in the device clock occurs during the clock switch. The length of this pause is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Clock transitions are discussed in greater detail in Section 3.1.2 "Entering Power-Managed Modes".

## REGISTER 2-2: OSCCON: OSCILLATOR CONTROL REGISTER

| R/W-0 | U-0 | U-0 | U-0 | R-q ${ }^{(1)}$ | U-0 | R/W-0 |  | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IDLEN | - | - | - | OSTS | - | SCS1 | SCS0 |  |
| bit 7 |  |  |  |  |  |  |  |  |

$\begin{array}{ll}\text { bit } 7 & \text { IDLEN: Idle Enable bit } \\ & 1=\text { Device enters Idle mode on SLEEP instruction } \\ & 0=\text { Device enters Sleep mode on SLEEP instruction }\end{array}$
bit 6-4 Unimplemented: Read as ' 0 '
bit 3 OSTS: Oscillator Start-up Time-out Status bit ${ }^{(1)}$
1 = Oscillator Start-up Timer time-out has expired; primary oscillator is running
$0=$ Oscillator Start-up Timer time-out is running; primary oscillator is not ready
Note 1: The Reset value is ' 0 ' when HS mode and Two-Speed Start-up are both enabled; otherwise, it is ' 1 '.
bit 2 Unimplemented: Read as ' 0 '
bit 1-0 SCS1:SCS0: System Clock Select bits
11 = Internal oscillator
$10=$ Primary oscillator
01 = Timer1 oscillator
When FOSC2 $=1$ :
$00=$ Primary oscillator
When FOSC2 = 0:
$00=$ Internal oscillator

## Legend:

$\mathrm{U}=$ Unimplemented, read as ' 0 '
$-\mathrm{n}=$ Value at POR $\quad \mathrm{R}=$ Readable bit $\quad ' 0$ ' = Bit is cleared $\quad \mathrm{W}=$ Writable bit

### 2.7 Effects of Power-Managed Modes on the Various Clock Sources

When PRI_IDLE mode is selected, the designated primary oscillator continues to run without interruption. For all other power-managed modes, the oscillator using the OSC1 pin is disabled. The OSC1 pin (and OSC2 pin if used by the oscillator) will stop oscillating.
In Secondary Clock modes (SEC_RUN and SEC_IDLE), the Timer1 oscillator is operating and providing the device clock. The Timer1 oscillator may also run in all power-managed modes if required to clock Timer1 or Timer3.
In RC_RUN and RC_IDLE modes, the internal oscillator provides the device clock source. The 31 kHz INTRC output can be used directly to provide the clock and may be enabled to support various special features, regardless of the power-managed mode (see Section 20.2 "Watchdog Timer (WDT)" through Section 20.5 "Fail-Safe Clock Monitor" for more information on WDT, Fail-Safe Clock Monitor and Two-Speed Start-up).

If the Sleep mode is selected, all clock sources are stopped. Since all the transistor switching currents have been stopped, Sleep mode achieves the lowest current consumption of the device (only leakage currents).
Enabling any on-chip feature that will operate during Sleep will increase the current consumed during Sleep. The INTRC is required to support WDT operation. The Timer1 oscillator may be operating to support a real-time clock. Other features may be operating that do not require a device clock source (i.e., MSSP slave, PSP, INTn pins and others). Peripherals that may add significant current consumption are listed in Section 23.2 "DC Characteristics: Power-Down and Supply Current".

### 2.8 Power-up Delays

Power-up delays are controlled by two timers, so that no external Reset circuitry is required for most applications. The delays ensure that the device is kept in Reset until the device power supply is stable under normal circumstances and the primary clock is operating and stable. For additional information on power-up delays, see Section 4.5 "Power-up Timer (PWRT)".
The first timer is the Power-up Timer (PWRT), which provides a fixed delay on power-up (parameter 33, Table 23-10). It is always enabled.

The second timer is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable (HS modes). The OST does this by counting 1024 oscillator cycles before allowing the oscillator to clock the device.
There is a delay of interval TCSD (parameter 38, Table 23-10), following POR, while the controller becomes ready to execute instructions.

TABLE 2-3: OSC1 AND OSC2 PIN STATES IN SLEEP MODE

| Oscillator Mode | OSC1 Pin | OSC2 Pin |
| :--- | :--- | :--- |
| EC, ECPLL | Floating, pulled by external clock | At logic low (clock/4 output) |
| HS, HSPLL | Feedback inverter disabled at quiescent <br> voltage level | Feedback inverter disabled at quiescent <br> voltage level |

Note: See Table 4-2 in Section 4.0 "Reset" for time-outs due to Sleep and $\overline{\text { MCLR Reset. }}$

## PIC18F45J10 FAMILY

NOTES:

### 3.0 POWER-MANAGED MODES

The PIC18F45J10 family devices provide the ability to manage power consumption by simply managing clocking to the CPU and the peripherals. In general, a lower clock frequency and a reduction in the number of circuits being clocked constitutes lower consumed power. For the sake of managing power in an application, there are three primary modes of operation:

- Run mode
- Idle mode
- Sleep mode

These modes define which portions of the device are clocked and at what speed. The Run and Idle modes may use any of the three available clock sources (primary, secondary or internal oscillator block); the Sleep mode does not use a clock source.
The power-managed modes include several power-saving features offered on previous PICmicro ${ }^{\circledR}$ devices. One is the clock switching feature, offered in other PIC18 devices, allowing the controller to use the Timer1 oscillator in place of the primary oscillator. Also included is the Sleep mode, offered by all PICmicro devices, where all device clocks are stopped.

### 3.1 Selecting Power-Managed Modes

Selecting a power-managed mode requires two decisions: if the CPU is to be clocked or not and which clock source is to be used. The IDLEN bit ( $\mathrm{OSCCON}<7>$ ) controls CPU clocking, while the SCS1:SCS0 bits ( $O S C C O N<1: 0>$ ) select the clock source. The individual modes, bit settings, clock sources and affected modules are summarized in Table 3-1.

### 3.1.1 CLOCK SOURCES

The SCS1:SCS0 bits allow the selection of one of three clock sources for power-managed modes. They are:

- the primary clock, as defined by the FOSC1:FOSC0 configuration bits
- the secondary clock (Timer1 oscillator)
- the internal oscillator


### 3.1.2 ENTERING POWER-MANAGED MODES

Switching from one power-managed mode to another begins by loading the OSCCON register. The SCS1:SCS0 bits select the clock source and determine which Run or Idle mode is to be used. Changing these bits causes an immediate switch to the new clock source, assuming that it is running. The switch may also be subject to clock transition delays. These are discussed in Section 3.1.3 "Clock Transitions and Status Indicators" and subsequent sections.
Entry to the power-managed Idle or Sleep modes is triggered by the execution of a SLEEP instruction. The actual mode that results depends on the status of the IDLEN bit.
Depending on the current mode and the mode being switched to, a change to a power-managed mode does not always require setting all of these bits. Many transitions may be done by changing the oscillator select bits, or changing the IDLEN bit, prior to issuing a SLEEP instruction. If the IDLEN bit is already configured correctly, it may only be necessary to perform a SLEEP instruction to switch to the desired mode.

## TABLE 3-1: POWER-MANAGED MODES

| Mode | OSCCON bits |  | Module Clocking |  | Available Clock and Oscillator Source |
| :--- | :---: | :---: | :---: | :---: | :--- |
|  | IDLEN<7> ${ }^{(1)}$ | SCS1:SCS0<1:0> | CPU | Peripherals |  |
| Sleep | 0 | N/A | Off | Off | None - All clocks are disabled |
| PRI_RUN | N/A | 10 | Clocked | Clocked | Primary - HS, EC; <br> this is the normal full power execution mode |
| SEC_RUN | N/A | 01 | Clocked | Clocked | Secondary - Timer1 Oscillator |
| RC_RUN | N/A | 11 | Clocked | Clocked | Internal Oscillator |
| PRI_IDLE | 1 | 10 | Off | Clocked | Primary - HS, EC |
| SEC_IDLE | 1 | 01 | Off | Clocked | Secondary - Timer1 Oscillator |
| RC_IDLE | 1 | 11 | Off | Clocked | Internal Oscillator |

Note 1: IDLEN reflects its value when the SLEEP instruction is executed.

### 3.1.3 CLOCK TRANSITIONS AND STATUS INDICATORS

The length of the transition between clock sources is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Two bits indicate the current clock source and its status: OSTS (OSCCON<3>) and T1RUN ( $\mathrm{T} 1 \mathrm{CON}<6>$ ). In general, only one of these bits will be set while in a given power-managed mode. When the OSTS bit is set, the primary clock is providing the device clock. When the T1RUN bit is set, the Timer1 oscillator is providing the clock. If neither of these bits is set, INTRC is clocking the device.

$$
\begin{array}{ll}
\text { Note: } & \text { Executing a SLEEP instruction does not } \\
\text { necessarily place the device into Sleep } \\
\text { mode. It acts as the trigger to place the } \\
\text { controller into either the Sleep mode or } \\
\text { one of the Idle modes, depending on the } \\
\text { setting of the IDLEN bit. }
\end{array}
$$

### 3.1.4 MULTIPLE SLEEP COMMANDS

The power-managed mode that is invoked with the SLEEP instruction is determined by the setting of the IDLEN bit at the time the instruction is executed. If another SLEEP instruction is executed, the device will enter the power-managed mode specified by IDLEN at that time. If IDLEN has changed, the device will enter the new power-managed mode specified by the new setting.

### 3.2 Run Modes

In the Run modes, clocks to both the core and peripherals are active. The difference between these modes is the clock source.

### 3.2.1 PRI_RUN MODE

The PRI_RUN mode is the normal, full power execution mode of the microcontroller. This is also the default mode upon a device Reset unless Two-Speed Start-up is enabled (see Section 20.4 "Two-Speed Start-up" for details). In this mode, the OSTS bit is set. (see Section 2.6.1 "Oscillator Control Register").

### 3.2.2 SEC_RUN MODE

The SEC_RUN mode is the compatible mode to the "clock switching" feature offered in other PIC18 devices. In this mode, the CPU and peripherals are clocked from the Timer1 oscillator. This gives users the option of lower power consumption while still using a high-accuracy clock source.
SEC_RUN mode is entered by setting the SCS1:SCS0 bits to ' 01 '. The device clock source is switched to the Timer1 oscillator (see Figure 3-1), the primary oscillator is shut down, the T1RUN bit ( $\mathrm{T} 1 \mathrm{CON}<6>$ ) is set and the OSTS bit is cleared.

Note: The Timer1 oscillator should already be running prior to entering SEC_RUN mode. If the T1OSCEN bit is not set when the SCS1:SCS0 bits are set to ' 01 ', entry to SEC_RUN mode will not occur. If the Timer1 oscillator is enabled, but not yet running, device clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result.
On transitions from SEC_RUN mode to PRI_RUN mode, the peripherals and CPU continue to be clocked from the Timer1 oscillator while the primary clock is started. When the primary clock becomes ready, a clock switch back to the primary clock occurs (see Figure 3-2). When the clock switch is complete, the T1RUN bit is cleared, the OSTS bit is set and the primary clock is providing the clock. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run.

FIGURE 3-1: TRANSITION TIMING FOR ENTRY TO SEC_RUN MODE


### 3.2.3 RC_RUN MODE

In RC_RUN mode, the CPU and peripherals are clocked from the internal oscillator; the primary clock is shut down. This mode provides the best power conservation of all the Run modes, while still executing code. It works well for user applications which are not highly timing-sensitive or do not require high-speed clocks at all times.

This mode is entered by setting SCS to ' 11 '. When the clock source is switched to the INTRC (see Figure 3-2), the primary oscillator is shut down and the OSTS bit is cleared.

On transitions from RC_RUN mode to PRI_RUN mode, the device continues to be clocked from the INTRC while the primary clock is started. When the primary clock becomes ready, a clock switch to the primary clock occurs (see Figure 3-3). When the clock switch is complete, the OSTS bit is set and the primary clock is providing the device clock. The IDLEN and SCS bits are not affected by the switch. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

FIGURE 3-2: TRANSITION TIMING TO RC_RUN MODE


FIGURE 3-3: TRANSITION TIMING FROM RC_RUN MODE TO PRI_RUN MODE


Note 1: TOST = 1024 Tosc. These intervals are not shown to scale.

### 3.3 Sleep Mode

The power-managed Sleep mode is identical to the legacy Sleep mode offered in all other PICmicro devices. It is entered by clearing the IDLEN bit (the default state on device Reset) and executing the SLEEP instruction. This shuts down the selected oscillator (Figure 3-4). All clock source status bits are cleared.
Entering the Sleep mode from any other mode does not require a clock switch. This is because no clocks are needed once the controller has entered Sleep. If the WDT is selected, the INTRC source will continue to operate. If the Timer1 oscillator is enabled, it will also continue to run.
When a wake event occurs in Sleep mode (by interrupt, Reset or WDT time-out), the device will not be clocked until the clock source selected by the SCS1:SCS0 bits becomes ready (see Figure 3-5), or it will be clocked from the internal oscillator if either the Two-Speed Start-up or the Fail-Safe Clock Monitor are enabled (see Section 20.0 "Special Features of the CPU"). In either case, the OSTS bit is set when the primary clock is providing the device clocks. The IDLEN and SCS bits are not affected by the wake-up.

### 3.4 Idle Modes

The Idle modes allow the controller's CPU to be selectively shut down while the peripherals continue to operate. Selecting a particular Idle mode allows users to further manage power consumption.

If the IDLEN bit is set to a ' 1 ' when a SLEEP instruction is executed, the peripherals will be clocked from the clock source selected using the SCS1:SCS0 bits; however, the CPU will not be clocked. The clock source status bits are not affected. Setting IDLEN and executing a SLEEP instruction provides a quick method of switching from a given Run mode to its corresponding Idle mode.
If the WDT is selected, the INTRC source will continue to operate. If the Timer1 oscillator is enabled, it will also continue to run.

Since the CPU is not executing instructions, the only exits from any of the Idle modes are by interrupt, WDT time-out or a Reset. When a wake event occurs, CPU execution is delayed by an interval of TCSD (parameter 38, Table 23-10) while it becomes ready to execute code. When the CPU begins executing code, it resumes with the same clock source for the current Idle mode. For example, when waking from RC_IDLE mode, the internal oscillator block will clock the CPU and peripherals (in other words, RC_RUN mode). The IDLEN and SCS bits are not affected by the wake-up.
While in any Idle mode or the Sleep mode, a WDT time-out will result in a WDT wake-up to the Run mode currently specified by the SCS1:SCS0 bits.

FIGURE 3-4: TRANSITION TIMING FOR ENTRY TO SLEEP MODE


FIGURE 3-5: TRANSITION TIMING FOR WAKE FROM SLEEP


Note1: Tost $=1024$ Tosc. These intervals are not shown to scale.

### 3.4.1 PRI_IDLE MODE

This mode is unique among the three low-power Idle modes, in that it does not disable the primary device clock. For timing-sensitive applications, this allows for the fastest resumption of device operation with its more accurate primary clock source, since the clock source does not have to "warm up" or transition from another oscillator.
PRI_IDLE mode is entered from PRI_RUN mode by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then set the SCS bits to ' 10 ' and execute SLEEP. Although the CPU is disabled, the peripherals continue to be clocked from the primary clock source specified by the FOSCO configuration bit. The OSTS bit remains set (see Figure 3-6).
When a wake event occurs, the CPU is clocked from the primary clock source. A delay of interval TCSD is required between the wake event and when code execution starts. This is required to allow the CPU to become ready to execute instructions. After the wake-up, the OSTS bit remains set. The IDLEN and SCS bits are not affected by the wake-up (see Figure 3-7).

### 3.4.2 SEC_IDLE MODE

In SEC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the Timer1 oscillator. This mode is entered from SEC_RUN by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then set SCS1:SCS0 to '01' and execute SLEEP. When the clock source is switched to the Timer1 oscillator, the primary oscillator is shut-down, the OSTS bit is cleared and the T1RUN bit is set.
When a wake event occurs, the peripherals continue to be clocked from the Timer1 oscillator. After an interval of TCSD following the wake event, the CPU begins executing code being clocked by the Timer1 oscillator. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run (see Figure 3-7).

Note: The Timer1 oscillator should already be running prior to entering SEC_IDLE mode. If the T1OSCEN bit is not set when the SLEEP instruction is executed, the SLEEP instruction will be ignored and entry to SEC_IDLE mode will not occur. If the Timer1 oscillator is enabled, but not yet running, peripheral clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result.

FIGURE 3-6: TRANSITION TIMING FOR ENTRY TO IDLE MODE


FIGURE 3-7: TRANSITION TIMING FOR WAKE FROM IDLE TO RUN MODE


### 3.4.3 RC_IDLE MODE

In RC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator. This mode allows for controllable power conservation during Idle periods.
From RC_RUN, this mode is entered by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, first set IDLEN, then clear the SCS bits and execute SLEEP. When the clock source is switched to the INTRC, the primary oscillator is shut down and the OSTS bit is cleared.
When a wake event occurs, the peripherals continue to be clocked from the INTRC. After a delay of TCSD following the wake event, the CPU begins executing code being clocked by the INTRC. The IDLEN and SCS bits are not affected by the wake-up. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

### 3.5 Exiting Idle and Sleep Modes

An exit from Sleep mode, or any of the Idle modes, is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed modes sections (see Section 3.2 "Run Modes", Section 3.3 "Sleep Mode" and Section 3.4 "Idle Modes").

### 3.5.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode, or the Sleep mode, to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCON or PIE registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.
On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/GIEH bit (INTCON $<7>$ ) is set. Otherwise, code execution continues or resumes without branching (see Section 8.0 "Interrupts").
A fixed delay of interval TCSD following the wake event is required when leaving Sleep and Idle modes. This delay is required for the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

### 3.5.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs.
If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see Section 3.2 "Run Modes" and Section 3.3 "Sleep Mode"). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see Section 20.2 "Watchdog Timer (WDT)").
The WDT timer and postscaler are cleared by one of the following events:

- executing a SLEEP or CLRWDT instruction
- the loss of a currently selected clock source (if the Fail-Safe Clock Monitor is enabled)


### 3.5.3 EXIT BY RESET

Exiting an Idle or Sleep mode by Reset automatically forces the device to run from the INTRC.

### 3.5.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. There are two cases:

- PRI_IDLE mode where the primary clock source is not stopped; and
- the primary clock source is the EC mode.

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI_IDLE), or normally does not require an oscillator start-up delay (EC). However, a fixed delay of interval TCSD following the wake event is still required when leaving Sleep and Idle modes to allow the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

### 4.0 RESET

The PIC18F45J10 family of devices differentiate between various kinds of Reset:
a) Power-on Reset (POR)
b) $\overline{M C L R}$ Reset during normal operation
c) $\overline{\mathrm{MCLR}}$ Reset during power-managed modes
d) Watchdog Timer (WDT) Reset (during execution)
e) Brown-out Reset (BOR)
f) RESET Instruction
g) Stack Full Reset
h) Stack Underflow Reset

This section discusses Resets generated by $\overline{M C L R}$, POR and BOR and covers the operation of the various start-up timers. Stack Reset events are covered in Section 5.1.2.4 "Stack Full and Underflow Resets". WDT Resets are covered in Section 20.2 "Watchdog Timer (WDT)".
A simplified block diagram of the On-Chip Reset Circuit is shown in Figure 4-1.

### 4.1 RCON Register

Device Reset events are tracked through the RCON register (Register 4-1). The lower five bits of the register indicate that a specific Reset event has occurred. In most cases, these bits can only be set by the event and must be cleared by the application after the event. The state of these flag bits, taken together, can be read to indicate the type of Reset that just occurred. This is described in more detail in Section 4.6 "Reset State of Registers".
The RCON register also has a control bit for setting interrupt priority (IPEN). Interrupt priority is discussed in Section 8.0 "Interrupts".

FIGURE 4-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT


Note 1: The Brown-out Reset is not available in PIC18LF2XJ10/4XJ10 devices.

REGISTER 4-1: RCON: RESET CONTROL REGISTER
R/W-0

| IPEN | U-0 | U-0 | R/W-1 | R-1 | R-1 | R/W-0 |  | R/W-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| bit 7 | - | $\overline{\mathrm{RI}}$ | $\overline{\mathrm{TO}}$ | $\overline{\mathrm{PD}}$ | $\overline{\mathrm{POR}}$ | $\overline{\mathrm{BOR}}$ |  |  |

bit $7 \quad$ IPEN: Interrupt Priority Enable bit
1 = Enable priority levels on interrupts
$0=$ Disable priority levels on interrupts (PIC16CXXX Compatibility mode)
bit 6-5 Unimplemented: Read as ' 0 '
bit $4 \quad \overline{\mathbf{R I}}$ : RESET Instruction Flag bit
1 = The RESET instruction was not executed (set by firmware only)
$0=$ The RESET instruction was executed causing a device Reset (must be set in software after a Brown-out Reset occurs)
bit $3 \quad \overline{\text { TO }}$ : Watchdog Time-out Flag bit
$1=$ Set by power-up, CLRWDT instruction or SLEEP instruction
$0=$ A WDT time-out occurred
bit $2 \quad \overline{P D}$ : Power-Down Detection Flag bit
1 = Set by power-up or by the CLRWDT instruction
$0=$ Set by execution of the SLEEP instruction
bit $1 \quad \overline{\text { POR: Power-on Reset Status bit }}$
1 = A Power-on Reset has not occurred (set by firmware only)
$0=$ A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)
bit $0 \quad$ BOR: Brown-out Reset Status bit
$1=A$ Brown-out Reset has not occurred (set by firmware only)
$0=$ A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)
Note: BOR is not available in PIC18LF2XJ10/4XJ10 devices.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

Note 1: It is recommended that the $\overline{\text { POR }}$ bit be set after a Power-on Reset has been detected, so that subsequent Power-on Resets may be detected.
2: If the on-chip voltage regulator is disabled, $\overline{B O R}$ remains ' 0 ' at all times. See Section 4.4.1 "Detecting BOR" for more information.
3: Brown-out Reset is said to have occurred when $\overline{B O R}$ is ' 0 ' and POR is ' 1 ' (assuming that $\overline{\text { POR }}$ was set to ' 1 ' by software immediately after POR).

### 4.2 Master Clear (MCLR)

The $\overline{M C L R}$ pin provides a method for triggering a hard external Reset of the device. A Reset is generated by holding the pin low. PIC18 extended microcontroller devices have a noise filter in the $\overline{M C L R}$ Reset path which detects and ignores small pulses.
The $\overline{M C L R}$ pin is not driven low by any internal Resets, including the WDT.

### 4.3 Power-on Reset (POR)

A Power-on Reset condition is generated on-chip whenever VDD rises above a certain threshold. This allows the device to start in the initialized state when VDD is adequate for operation.
To take advantage of the POR circuitry, tie the $\overline{\text { MCLR }}$ pin through a resistor ( $1 \mathrm{k} \Omega$ to $10 \mathrm{k} \Omega$ ) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset delay. A minimum rise rate for VDD is specified (parameter D004). For a slow rise time, see Figure 4-2.
When the device starts normal operation (i.e., exits the Reset condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.
POR events are captured by the $\overline{\mathrm{POR}}$ bit ( $\mathrm{RCON}<1>$ ). The state of the bit is set to ' 0 ' whenever a POR occurs; it does not change for any other Reset event. POR is not reset to ' 1 ' by any hardware event. To capture multiple events, the user manually resets the bit to ' 1 ' in software following any POR.

### 4.4 Brown-out Reset (BOR) (PIC18F2XJ10/4XJ10 Devices Only)

Once a BOR has occurred, the Power-up Timer will keep the chip in Reset for Tpwrt (parameter 33). If Vdd drops below VBor while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be initialized. Once Vdd rises above VBOR, the Power-up Timer will execute the additional time delay.

FIGURE 4-2:
EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)


Note 1: External Power-on Reset circuit is required only if the VDD power-up slope is too slow. The diode $D$ helps discharge the capacitor quickly when VDD powers down.
2: $R<40 \mathrm{k} \Omega$ is recommended to make sure that the voltage drop across $R$ does not violate the device's electrical specification.
3: $\quad \mathrm{R} 1 \geq 1 \mathrm{k} \Omega$ will limit any current flowing into $\overline{\mathrm{MCLR}}$ from external capacitor C , in the event of $\overline{M C L R}$ pin breakdown, due to Electrostatic Discharge (ESD) or Electrical Overstress (EOS).

### 4.4.1 DETECTING BOR

The $\overline{B O R}$ bit always resets to ' 0 ' on any BOR or POR event. This makes it difficult to determine if a BOR event has occurred just by reading the state of $\overline{\mathrm{BOR}}$ alone. A more reliable method is to simultaneously check the state of both $\overline{\mathrm{POR}}$ and $\overline{\mathrm{BOR}}$. This assumes that the $\overline{\mathrm{POR}}$ bit is reset to ' 1 ' in software immediately after any POR event. If $\overline{\mathrm{BOR}}$ is ' 0 ' while $\overline{\mathrm{POR}}$ is ' 1 ', it can be reliably assumed that a BOR event has occurred.
In devices designated with an "LF" part number (such as PIC18LF25J10), Brown-out Reset functionality is disabled. In this case, the $\overline{\mathrm{BOR}}$ bit cannot be used to determine a BOR event. The $\overline{B O R}$ bit is still cleared by a POR event.

## PIC18F45J10 FAMILY

### 4.5 Power-up Timer (PWRT)

PIC18F45J10 family devices incorporate an on-chip Power-up Timer (PWRT) to help regulate the Power-on Reset process. The PWRT is always enabled. The main function is to ensure that the device voltage is stable before code is executed.
The Power-up Timer (PWRT) of the PIC18F45J10 family devices is an 11-bit counter which uses the INTRC source as the clock input. This yields an approximate time interval of $2048 \times 32 \mu \mathrm{~s}=65.6 \mathrm{~ms}$. While the PWRT is counting, the device is held in Reset.

The power-up time delay depends on the INTRC clock and will vary from chip to chip due to temperature and process variation. See DC parameter 33 for details.

### 4.5.1 TIME-OUT SEQUENCE

If enabled, the PWRT time-out is invoked after the POR pulse has cleared. The total time-out will vary based on the status of the PWRT. Figure 4-3, Figure 4-4, Figure 4-5 and Figure 4-6 all depict time-out sequences on power-up with the Power-up Timer enabled.

Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the PWRT will expire. Bringing $\overline{\text { MCLR high will begin execution immediately }}$ (Figure 4-5). This is useful for testing purposes, or to synchronize more than one PIC18F device operating in parallel.

FIGURE 4-3: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO Vdd, Vdd RISE < TPWRT)


FIGURE 4-4: TIME-OUT SEQUENCE ON POWER-UP ( $\overline{M C L R}$ NOT TIED TO Vdd): CASE 1


FIGURE 4-5: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO Vdd): CASE 2


FIGURE 4-6: SLOW RISE TIME ( $\overline{\text { MCLR }}$ TIED TO Vdd, Vdd RISE > TPWRT)


## PIC18F45J10 FAMILY

### 4.6 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a "Reset state" depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. Status bits from the RCON register, $\overline{\mathrm{Rl}}, \overline{\mathrm{TO}}$, $\overline{\mathrm{PD}}, \overline{\mathrm{POR}}$ and $\overline{\mathrm{BOR}}$, are set or cleared differently in different Reset situations, as indicated in Table 4-1. These bits are used in software to determine the nature of the Reset.

Table 4-2 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets and WDT wake-ups.

TABLE 4-1: STATUS BITS, THEIR SIGNIFICANCE AND THE INITIALIZATION CONDITION FOR RCON REGISTER

| Condition | Program Counter ${ }^{(1)}$ | RCON Register |  |  |  |  | STKPTR Register |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\overline{\mathbf{R I}}$ | TO | $\overline{\text { PD }}$ | $\overline{\text { POR }}$ | $\overline{B O R}^{(2)}$ | STKFUL | STKUNF |
| Power-on Reset | 0000h | 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| RESET instruction | 0000h | 0 | u | u | u | u | u | u |
| Brown-out | 0000h | 1 | 1 | 1 | u | 0 | u | u |
| $\overline{\text { MCLR }}$ during power-managed Run modes | 0000h | u | 1 | u | u | u | u | u |
| $\overline{\text { MCLR }}$ during power-managed Idle modes and Sleep mode | 0000h | u | 1 | 0 | u | u | u | u |
| WDT time-out during full power or power-managed Run modes | 0000h | u | 0 | u | u | u | u | u |
| $\overline{\mathrm{MCLR}}$ during full power execution | 0000h | u | u | u | u | u | u | u |
| Stack Full Reset (STVREN = 1) | 0000h | u | u | u | u | u | 1 | u |
| Stack Underflow Reset (STVREN = 1) | 0000h | u | u | u | u | u | u | 1 |
| Stack Underflow Error (not an actual Reset, STVREN = 0) | 0000h | u | u | u | u | u | u | 1 |
| WDT time-out during power-managed Idle or Sleep modes | PC + 2 | u | 0 | 0 | u | u | u | u |
| Interrupt exit from power-managed modes | PC + 2 | u | u | 0 | u | u | u | u |

Legend: $u=$ unchanged
Note 1: When the wake-up is due to an interrupt and the GIEH or GIEL bits are set, the PC is loaded with the interrupt vector (0008h or 0018h).
2: BOR is not available in PIC18LF2XJ10/4XJ10 devices.

## TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS

| Register | Applicable Devices |  | Power-on Reset, Brown-out Reset | $\overline{M C L R}$ Resets WDT Reset RESET Instruction Stack Resets | Wake-up via WDT or Interrupt |
| :---: | :---: | :---: | :---: | :---: | :---: |
| TOSU | PIC18F2XJ10 | PIC18F4XJ10 | ---0 0000 | ---0 0000 | ---0 uuuu ${ }^{(1)}$ |
| TOSH | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu ${ }^{(1)}$ |
| TOSL | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu ${ }^{(1)}$ |
| STKPTR | PIC18F2XJ10 | PIC18F4XJ10 | 00-0 0000 | uu-0 0000 | uu-u uuuu ${ }^{(1)}$ |
| PCLATU | PIC18F2XJ10 | PIC18F4XJ10 | ---0 0000 | ---0 0000 | ---u uuuu |
| PCLATH | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| PCL | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | $\mathrm{PC}+2^{(2)}$ |
| TBLPTRU | PIC18F2XJ10 | PIC18F4XJ10 | --00 0000 | --00 0000 | --uu uuuu |
| TBLPTRH | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| TBLPTRL | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| TABLAT | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| PRODH | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| PRODL | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| INTCON | PIC18F2XJ10 | PIC18F4XJ10 | 0000 000x | 0000 000u | uuuu uuuu ${ }^{(3)}$ |
| INTCON2 | PIC18F2XJ10 | PIC18F4XJ10 | 1111 -1-1 | 1111 -1-1 | uuuu -u-u ${ }^{(3)}$ |
| INTCON3 | PIC18F2XJ10 | PIC18F4XJ10 | 11-0 0-00 | 11-0 0-00 | uu-u u-uu ${ }^{(3)}$ |
| INDF0 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| POSTINC0 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| POSTDEC0 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| PREINC0 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| PLUSW0 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| FSROH | PIC18F2XJ10 | PIC18F4XJ10 | ---- xxxx | ---- uuuu | ---- uuuu |
| FSROL | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| WREG | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| INDF1 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| POSTINC1 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| POSTDEC1 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| PREINC1 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| PLUSW1 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| FSR1H | PIC18F2XJ10 | PIC18F4XJ10 | ---- xxxx | ---- uuuu | ---- uuuu |
| FSR1L | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| BSR | PIC18F2XJ10 | PIC18F4XJ10 | ---- 0000 | ---- 0000 | ---- uuuu |

Legend: $u=u n c h a n g e d, ~ x=u n k n o w n,-=u n i m p l e m e n t e d ~ b i t, ~ r e a d ~ a s ~ ' ~ o ', ~ q ~=~ v a l u e ~ d e p e n d s ~ o n ~ c o n d i t i o n . ~$ Shaded cells indicate conditions do not apply for the designated device.
Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.
2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
4: See Table 4-1 for Reset value for specific condition.

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

| Register | Applicable Devices |  | Power-on Reset, Brown-out Reset | $\overline{M C L R}$ Resets WDT Reset RESET Instruction Stack Resets | Wake-up via WDT or Interrupt |
| :---: | :---: | :---: | :---: | :---: | :---: |
| INDF2 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| POSTINC2 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| POSTDEC2 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| PREINC2 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| PLUSW2 | PIC18F2XJ10 | PIC18F4XJ10 | N/A | N/A | N/A |
| FSR2H | PIC18F2XJ10 | PIC18F4XJ10 | ---- xxxx | ---- uuuu | ---- uuuu |
| FSR2L | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| STATUS | PIC18F2XJ10 | PIC18F4XJ10 | ---x xxxx | ---u uuuu | ---u uuuu |
| TMROH | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| TMROL | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| T0CON | PIC18F2XJ10 | PIC18F4XJ10 | 11111111 | 11111111 | uuuu uuuu |
| OSCCON | PIC18F2XJ10 | PIC18F4XJ10 | 0--- q-00 | 0--- q-00 | u--- q-uu |
| WDTCON | PIC18F2XJ10 | PIC18F4XJ10 | ---- ---0 | ---- ---0 | ---- ---u |
| RCON ${ }^{(4)}$ | PIC18F2XJ10 | PIC18F4XJ10 | 0--1 11q0 | $0--q$ qquu | u--u qquu |
| TMR1H | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| TMR1L | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| T1CON | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | uOuu uuuu | uuuu uuuu |
| TMR2 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| PR2 | PIC18F2XJ10 | PIC18F4XJ10 | 11111111 | 11111111 | 11111111 |
| T2CON | PIC18F2XJ10 | PIC18F4XJ10 | -000 0000 | -000 0000 | -uuu uuuu |
| SSP1BUF | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| SSP1ADD | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SSP1STAT | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SSP1CON1 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SSP1CON2 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| ADRESH | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| ADRESL | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| ADCON0 | PIC18F2XJ10 | PIC18F4XJ10 | 0-00 0000 | 0-00 0000 | u-uu uuuu |
| ADCON1 | PIC18F2XJ10 | PIC18F4XJ10 | --00 0qqq | --00 0qqq | --uu uqqq |
| ADCON2 | PIC18F2XJ10 | PIC18F4XJ10 | 0-00 0000 | 0-00 0000 | u-uu uuuu |

Legend: $u=$ unchanged, $x=$ unknown, - = unimplemented bit, read as ' 0 ', $q=$ value depends on condition. Shaded cells indicate conditions do not apply for the designated device.
Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.
2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
4: See Table 4-1 for Reset value for specific condition.

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

| Register | Applicable Devices |  | Power-on Reset, Brown-out Reset | $\overline{M C L R}$ Resets WDT Reset RESET Instruction Stack Resets | Wake-up via WDT or Interrupt |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CCPR1H | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uauu |
| CCPR1L | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| CCP1CON | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| CCPR2H | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| CCPR2L | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| CCP2CON | PIC18F2XJ10 | PIC18F4XJ10 | $--000000$ | --00 0000 | --uu uuuu |
| BAUDCON | PIC18F2XJ10 | PIC18F4XJ10 | 01-0 0-00 | 01-0 0-00 | uu-u u-uu |
| ECCP1DEL | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| ECCP1AS | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| CVRCON | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| CMCON | PIC18F2XJ10 | PIC18F4XJ10 | 00000111 | 00000111 | uuuu uuuu |
| SPBRGH | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SPBRG | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| RCREG | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| TXREG | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| TXSTA | PIC18F2XJ10 | PIC18F4XJ10 | 00000010 | 00000010 | uuuu uuuu |
| RCSTA | PIC18F2XJ10 | PIC18F4XJ10 | 0000 000x | 0000 000x | uuuu uauu |
| EECON2 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| EECON1 | PIC18F2XJ10 | PIC18F4XJ10 | ---0 x00- | ---0 x00- | ---u uuu- |
| IPR3 | PIC18F2XJ10 | PIC18F4XJ10 | 11-- ---- | 11-- --- | uu-- ---- |
| PIR3 | PIC18F2XJ10 | PIC18F4XJ10 | 00-- ---- | 00-- ---- | uu-- ----(3) |
| PIE3 | PIC18F2XJ10 | PIC18F4XJ10 | 00-- ---- | 00-- ---- | uu-- ---- |
| IPR2 | PIC18F2XJ10 | PIC18F4XJ10 | 11-- 1--1 | 11-- 1--1 | uu-- u--u |
| PIR2 | PIC18F2XJ10 | PIC18F4XJ10 | 00-- 0--0 | 00-- 0--0 | uu-- u--u ${ }^{(3)}$ |
| PIE2 | PIC18F2XJ10 | PIC18F4XJ10 | 00-- 0--0 | 00-- 0--0 | uu-- u--u |
| IPR1 | PIC18F2XJ10 | PIC18F4XJ10 | 11111111 | 11111111 | uuuu uuuu |
| PIR1 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu unuu $^{(3)}$ |
| PIE1 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |

Legend: $u=$ unchanged, $x=u n k n o w n, ~-~=~ u n i m p l e m e n t e d ~ b i t, ~ r e a d ~ a s ~ ' ~ o ', ~ q ~=~ v a l u e ~ d e p e n d s ~ o n ~ c o n d i t i o n . ~$ Shaded cells indicate conditions do not apply for the designated device.
Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.
2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
4: See Table 4-1 for Reset value for specific condition.

TABLE 4-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

| Register | Applicable Devices |  | Power-on Reset, Brown-out Reset | MCLR Resets WDT Reset RESET Instruction Stack Resets | Wake-up via WDT or Interrupt |
| :---: | :---: | :---: | :---: | :---: | :---: |
| TRISE | PIC18F2XJ10 | PIC18F4XJ10 | 0000-111 | 1111 -111 | uuuu -uuu |
| TRISD | PIC18F2XJ10 | PIC18F4XJ10 | 11111111 | 11111111 | uuuu uuuu |
| TRISC | PIC18F2XJ10 | PIC18F4XJ10 | 11111111 | 11111111 | uuuu uuuu |
| TRISB | PIC18F2XJ10 | PIC18F4XJ10 | 11111111 | 11111111 | uuuu uuuu |
| TRISA | PIC18F2XJ10 | PIC18F4XJ10 | --1- 1111 | --1- 1111 | --u- uuuu |
| SSP2BUF | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| LATE | PIC18F2XJ10 | PIC18F4XJ10 | ---- -xxx | ---- -uuu | ---- -uuu |
| LATD | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| LATC | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| LATB | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| LATA | PIC18F2XJ10 | PIC18F4XJ10 | --x- xxxx | --u- uuuu | --u- uuuu |
| SSP2ADD | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SSP2STAT | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SSP2CON1 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| SSP2CON2 | PIC18F2XJ10 | PIC18F4XJ10 | 00000000 | 00000000 | uuuu uuuu |
| PORTE | PIC18F2XJ10 | PIC18F4XJ10 | ---- -xxx | ---- -uuu | ---- -uuu |
| PORTD | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| PORTC | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| PORTB | PIC18F2XJ10 | PIC18F4XJ10 | xxxx xxxx | uuuu uuuu | uuuu uuuu |
| PORTA | PIC18F2XJ10 | PIC18F4XJ10 | --0-0000 | --0-0000 | --u- uuuu |

Legend: $u=$ unchanged, $x=$ unknown, - = unimplemented bit, read as ' 0 ', $q=$ value depends on condition. Shaded cells indicate conditions do not apply for the designated device.
Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.
2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).
3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
4: See Table 4-1 for Reset value for specific condition.

### 5.0 MEMORY ORGANIZATION

There are two types of memory in PIC18 Enhanced microcontroller devices:

- Program Memory
- Data RAM

As Harvard architecture devices, the data and program memories use separate busses; this allows for concurrent access of the two memory spaces.

Additional detailed information on the operation of the Flash program memory is provided in Section 6.0 "Flash Program Memory".

### 5.1 Program Memory Organization

PIC18 microcontrollers implement a 21-bit program counter, which is capable of addressing a 2-Mbyte program memory space. Accessing a location between the upper boundary of the physically implemented memory and the 2-Mbyte address will return all ' 0 's (a NOP instruction).
The PIC18F24J10 and PIC18F44J10 each have 16 Kbytes of Flash memory and can store up to 8,192 single-word instructions. The PIC18F25J10 and PIC18F45J10 each have 32 Kbytes of Flash memory and can store up to 16,384 single-word instructions.

PIC18 devices have two interrupt vectors. The Reset vector address is at 0000h and the interrupt vector addresses are at 0008h and 0018h.

The program memory map for the PIC18F45J10 family devices is shown in Figure 5-1.

FIGURE 5-1: PROGRAM MEMORY MAP AND STACK FOR PIC18F45J10 FAMILY DEVICES


### 5.1.1 PROGRAM COUNTER

The Program Counter (PC) specifies the address of the instruction to fetch for execution. The PC is 21 bits wide and is contained in three separate 8 -bit registers. The low byte, known as the PCL register, is both readable and writable. The high byte, or PCH register, contains the $\mathrm{PC}<15: 8>$ bits; it is not directly readable or writable. Updates to the PCH register are performed through the PCLATH register. The upper byte is called PCU. This register contains the $\mathrm{PC}<20: 16>$ bits; it is also not directly readable or writable. Updates to the PCU register are performed through the PCLATU register.
The contents of PCLATH and PCLATU are transferred to the program counter by any operation that writes PCL. Similarly, the upper two bytes of the program counter are transferred to PCLATH and PCLATU by an operation that reads PCL. This is useful for computed offsets to the PC (see Section 5.1.4.1 "Computed GOTO").
The PC addresses bytes in the program memory. To prevent the PC from becoming misaligned with word instructions, the Least Significant bit of PCL is fixed to a value of ' 0 '. The PC increments by 2 to address sequential instructions in the program memory.
The CALL, RCALL, GOTO and program branch instructions write to the program counter directly. For these instructions, the contents of PCLATH and PCLATU are not transferred to the program counter.

### 5.1.2 RETURN ADDRESS STACK

The return address stack allows any combination of up to 31 program calls and interrupts to occur. The PC is pushed onto the stack when a CALL or RCALL instruction is executed or an interrupt is Acknowledged. The PC value is pulled off the stack on a RETURN, RETLW or RETFIE instruction. PCLATU and PCLATH are not affected by any of the RETURN or CALL instructions.

The stack operates as a 31-word by 21-bit RAM and a 5-bit Stack Pointer, STKPTR. The stack space is not part of either program or data space. The Stack Pointer is readable and writable and the address on the top of the stack is readable and writable through the top-ofstack Special File Registers. Data can also be pushed to, or popped from the stack, using these registers.
A CALL type instruction causes a push onto the stack; the Stack Pointer is first incremented and the location pointed to by the Stack Pointer is written with the contents of the PC (already pointing to the instruction following the CALL). A RETURN type instruction causes a pop from the stack; the contents of the location pointed to by the STKPTR are transferred to the PC and then the Stack Pointer is decremented.

The Stack Pointer is initialized to ' 00000 ' after all Resets. There is no RAM associated with the location corresponding to a Stack Pointer value of ' 00000 '; this is only a Reset value. Status bits indicate if the stack is full or has overflowed or has underflowed.

### 5.1.2.1 Top-of-Stack Access

Only the top of the return address stack (TOS) is readable and writable. A set of three registers, TOSU:TOSH:TOSL, hold the contents of the stack location pointed to by the STKPTR register (Figure 5-2). This allows users to implement a software stack if necessary. After a CALL, RCALL or interrupt, the software can read the pushed value by reading the TOSU:TOSH:TOSL registers. These values can be placed on a user-defined software stack. At return time, the software can return these values to TOSU:TOSH:TOSL and do a return.
The user must disable the global interrupt enable bits while accessing the stack to prevent inadvertent stack corruption.

FIGURE 5-2: RETURN ADDRESS STACK AND ASSOCIATED REGISTERS


### 5.1.2.2 Return Stack Pointer (STKPTR)

The STKPTR register (Register 5-1) contains the Stack Pointer value, the STKFUL (Stack Overflow) status bit and the STKUNF (Stack Underflow) status bits. The value of the Stack Pointer can be 0 through 31. The Stack Pointer increments before values are pushed onto the stack and decrements after values are popped off the stack. On Reset, the Stack Pointer value will be zero. The user may read and write the Stack Pointer value. This feature can be used by a Real-Time Operating System (RTOS) for return stack maintenance.
After the PC is pushed onto the stack 31 times (without popping any values off the stack), the STKFUL bit is set. The STKFUL bit is cleared by software or by a POR.
The action that takes place when the stack becomes full depends on the state of the STVREN (Stack Overflow Reset Enable) configuration bit. (Refer to Section 20.1 "Configuration Bits" for a description of the device configuration bits.) If STVREN is set (default), the 31st push will push the (PC + 2) value onto the stack, set the STKFUL bit and reset the device. The STKFUL bit will remain set and the Stack Pointer will be set to zero.

If STVREN is cleared, the STKFUL bit will be set on the 31st push and the Stack Pointer will increment to 31. Any additional pushes will not overwrite the 31st push and the STKPTR will remain at 31 .

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and sets the STKUNF bit, while the Stack Pointer remains at zero. The STKUNF bit will remain set until cleared by software or until a POR occurs.
Note: Returning a value of zero to the PC on an underflow has the effect of vectoring the program to the Reset vector, where the stack conditions can be verified and appropriate actions can be taken. This is not the same as a Reset, as the contents of the SFRs are not affected.

### 5.1.2.3 PUSH and POP Instructions

Since the Top-of-Stack is readable and writable, the ability to push values onto the stack and pull values off the stack without disturbing normal program execution is a desirable feature. The PIC18 instruction set includes two instructions, PUSH and POP, that permit the TOS to be manipulated under software control. TOSU, TOSH and TOSL can be modified to place data or a return address on the stack.
The PUSH instruction places the current PC value onto the stack. This increments the Stack Pointer and loads the current PC value onto the stack.
The POP instruction discards the current TOS by decrementing the Stack Pointer. The previous value pushed onto the stack then becomes the TOS value.

## REGISTER 5-1: STKPTR: STACK POINTER REGISTER

| R/C-0 | R/C-0 | U-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |  | R/W-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| STKFUL | (1) | STKUNF $^{(\mathbf{1})}$ | - | SP4 | SP3 | SP2 | SP1 | SP0 |
| bit 7 |  |  |  |  |  |  |  |  |


| bit 7 | STKFUL: Stack Overflow Flag bit ${ }^{(1)}$ |
| :---: | :---: |
|  | 1 = Stack became full or overflowed |
|  | 0 = Stack has not become full or overflowed |
| bit 6 | STKUNF: Stack Underflow Flag bit ${ }^{(1)}$ |
|  | 1 = Stack underflow occurred |
|  | 0 = Stack underflow did not occur |
| bit 5 | Unimplemented: Read as '0' |
| bit 4-0 | SP4:SP0: Stack Pointer Location bits |

Note 1: Bit 7 and bit 6 are cleared by user software or by a POR.

| Legend: |  |  |  |
| :--- | :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented | $C=$ Clearable only bit |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0 '=$ Bit is cleared | $x=$ Bit is unknown |

### 5.1.2.4 Stack Full and Underflow Resets

Device Resets on stack overflow and stack underflow conditions are enabled by setting the STVREN bit in Configuration Register 4L. When STVREN is set, a full or underflow will set the appropriate STKFUL or STKUNF bit and then cause a device Reset. When STVREN is cleared, a full or underflow condition will set the appropriate STKFUL or STKUNF bit but not cause a device Reset. The STKFUL or STKUNF bits are cleared by the user software or a Power-on Reset.

### 5.1.3 FAST REGISTER STACK

A fast register stack is provided for the STATUS, WREG and BSR registers, to provide a "fast return" option for interrupts. The stack for each register is only one level deep and is neither readable nor writable. It is loaded with the current value of the corresponding register when the processor vectors for an interrupt. All interrupt sources will push values into the stack registers. The values in the registers are then loaded back into their associated registers if the RETFIE, FAST instruction is used to return from the interrupt.

If both low and high priority interrupts are enabled, the stack registers cannot be used reliably to return from low priority interrupts. If a high priority interrupt occurs while servicing a low priority interrupt, the stack register values stored by the low priority interrupt will be overwritten. In these cases, users must save the key registers in software during a low priority interrupt.
If interrupt priority is not used, all interrupts may use the fast register stack for returns from interrupt. If no interrupts are used, the fast register stack can be used to restore the STATUS, WREG and BSR registers at the end of a subroutine call. To use the fast register stack for a subroutine call, a CALL label, FAST instruction must be executed to save the STATUS, WREG and BSR registers to the fast register stack. A RETURN, FAST instruction is then executed to restore these registers from the fast register stack.
Example 5-1 shows a source code example that uses the fast register stack during a subroutine call and return.

EXAMPLE 5-1: FAST REGISTER STACK CODE EXAMPLE


### 5.1.4 LOOK-UP TABLES IN PROGRAM MEMORY

There may be programming situations that require the creation of data structures, or look-up tables, in program memory. For PIC18 devices, look-up tables can be implemented in two ways:

- Computed Goto
- Table Reads


### 5.1.4.1 Computed GOTO

A computed GOTO is accomplished by adding an offset to the program counter. An example is shown in Example 5-2.
A look-up table can be formed with an ADDWF PCL instruction and a group of RETLW nn instructions. The W register is loaded with an offset into the table before executing a call to that table. The first instruction of the called routine is the ADDWF PCL instruction. The next instruction executed will be one of the RETLW nn instructions that returns the value ' nn ' to the calling function.

The offset value (in WREG) specifies the number of bytes that the program counter should advance and should be multiples of $2(\mathrm{LSb}=0)$.
In this method, only one data byte may be stored in each instruction location and room on the return address stack is required.

## EXAMPLE 5-2: COMPUTED GOTO USING

 AN OFFSET VALUE|  | MOVF | OFFSET, W |
| :--- | :--- | :--- |
| ORG | CALL | TABLE |
| TABLE | ADDWF | PCL |
|  | RETLW | nnh |
|  | RETLW | nnh |
|  | RETLW | nnh |
|  | $\cdot$ |  |
|  |  |  |
|  |  |  |
|  |  |  |
|  |  |  |

### 5.1.4.2 Table Reads and Table Writes

A better method of storing data in program memory allows two bytes of data to be stored in each instruction location.

Look-up table data may be stored two bytes per program word by using table reads and writes. The Table Pointer (TBLPTR) register specifies the byte address and the Table Latch (TABLAT) register contains the data that is read from or written to program memory. Data is transferred to or from program memory one byte at a time.
Table read and table write operations are discussed further in Section 6.1 "Table Reads and Table Writes".

### 5.2 PIC18 Instruction Cycle

### 5.2.1 CLOCKING SCHEME

The microcontroller clock input, whether from an internal or external source, is internally divided by four to generate four non-overlapping quadrature clocks (Q1, Q2, Q3 and Q4). Internally, the program counter is incremented on every Q1; the instruction is fetched from the program memory and latched into the instruction register during Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow are shown in Figure 5-3.

### 5.2.2 INSTRUCTION FLOW/PIPELINING

An "Instruction Cycle" consists of four $Q$ cycles: Q1 through Q4. The instruction fetch and execute are pipelined in such a manner that a fetch takes one instruction cycle, while the decode and execute take another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (e.g., GOTO), then two cycles are required to complete the instruction (Example 5-3).
A fetch cycle begins with the Program Counter (PC) incrementing in Q1.
In the execution cycle, the fetched instruction is latched into the Instruction Register (IR) in cycle Q1. This instruction is then decoded and executed during the Q2, Q3 and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

FIGURE 5-3: CLOCK/INSTRUCTION CYCLE


EXAMPLE 5-3: INSTRUCTION PIPELINE FLOW


All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is "flushed" from the pipeline while the new instruction is being fetched and then executed.

### 5.2.3 INSTRUCTIONS IN PROGRAM MEMORY

The program memory is addressed in bytes. Instructions are stored as two bytes or four bytes in program memory. The Least Significant Byte of an instruction word is always stored in a program memory location with an even address ( $\mathrm{LSb}=0$ ). To maintain alignment with instruction boundaries, the PC increments in steps of 2 and the LSb will always read ' 0 ' (see Section 5.1.1 "Program Counter").
Figure 5-4 shows an example of how instruction words are stored in the program memory.

The CALL and GOTO instructions have the absolute program memory address embedded into the instruction. Since instructions are always stored on word boundaries, the data contained in the instruction is a word address. The word address is written to $\mathrm{PC}<20: 1>$, which accesses the desired byte address in program memory. Instruction \#2 in Figure 5-4 shows how the instruction, GOTO 0006 h , is encoded in the program memory. Program branch instructions, which encode a relative address offset, operate in the same manner. The offset value stored in a branch instruction represents the number of single-word instructions that the PC will be offset by. Section 21.0 "Instruction Set Summary" provides further details of the instruction set.

FIGURE 5-4: INSTRUCTIONS IN PROGRAM MEMORY

|  |  |  | LSB = 1 | LSB $=0$ | ord Addres |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Program | mory |  |  | 000000h |
|  | Byte Loc | s $\rightarrow$ |  |  | 000002h |
|  |  |  |  |  | 000004h |
|  |  |  |  |  | 000006h |
| Instruction 1: | MOVLW | 055h | 0Fh | 55h | 000008h |
| Instruction 2: | Gото | 0006h | EFh | 03h | 00000Ah |
|  |  |  | FOh | 00h | 00000Ch |
| Instruction 3: | MOVFF | 123h, 456h | C1h | 23h | 00000Eh |
|  |  |  | F4h | 56h | 000010h |
|  |  |  |  |  | 000012h |
|  |  |  |  |  | 000014h |

### 5.2.4 TWO-WORD INSTRUCTIONS

The standard PIC18 instruction set has four two-word instructions: CALL, MOVFF, GOTO and LSFR. In all cases, the second word of the instructions always has '1111' as its four Most Significant bits; the other 12 bits are literal data, usually a data memory address.

The use of ' 1111 ' in the 4 MSbs of an instruction specifies a special form of NOP. If the instruction is executed in proper sequence - immediately after the first word the data in the second word is accessed and used by
the instruction sequence. If the first word is skipped for some reason and the second word is executed by itself, a NOP is executed instead. This is necessary for cases when the two-word instruction is preceded by a conditional instruction that changes the PC. Example 5-4 shows how this works.

| Note: | See Section 5.6 "PIC18 Instruction |
| :--- | :--- |
|  | Execution and the Extended Instruc- |
|  | tion Set" for information on two-word |
|  | instructions in the extended instruction set. |

## EXAMPLE 5-4: TWO-WORD INSTRUCTIONS

| CASE 1: |  |  |  |
| :---: | :---: | :---: | :---: |
| Object Code | Source Code |  |  |
| 0110011000000000 | TSTFSZ | REG1 | ; is RAM location 0? |
| 1100000100100011 | MOVFF | REG1, REG2 | ; No, skip this word |
| 1111010001010110 |  |  | ; Execute this word as a NOP |
| 0010010000000000 | ADDWF | REG3 | ; continue code |
| CASE 2: |  |  |  |
| Object Code | Source Code |  |  |
| 0110011000000000 | TSTFSZ | REG1 | ; is RAM location 0? |
| 1100000100100011 | MOVFF | REG1, REG2 | ; Yes, execute this word |
| 1111010001010110 |  |  | ; 2nd word of instruction |
| 0010010000000000 | ADDWF | REG3 | ; continue code |

### 5.3 Data Memory Organization

Note: The operation of some aspects of data memory are changed when the PIC18 extended instruction set is enabled. See Section 5.5 "Data Memory and the Extended Instruction Set" for more information.

The data memory in PIC18 devices is implemented as static RAM. Each register in the data memory has a 12-bit address, allowing up to 4096 bytes of data memory. The memory space is divided into as many as 16 banks that contain 256 bytes each; PIC18F45J10 family devices implement all 16 banks. Figure 5-5 shows the data memory organization for the PIC18F45J10 family devices.

The data memory contains Special Function Registers (SFRs) and General Purpose Registers (GPRs). The SFRs are used for control and status of the controller and peripheral functions, while GPRs are used for data storage and scratchpad operations in the user's application. Any read of an unimplemented location will read as '0's.
The instruction set and architecture allow operations across all banks. The entire data memory may be accessed by Direct, Indirect or Indexed Addressing modes. Addressing modes are discussed later in this subsection.
To ensure that commonly used registers (SFRs and select GPRs) can be accessed in a single cycle, PIC18 devices implement an Access Bank. This is a 256-byte memory space that provides fast access to SFRs and the lower portion of GPR Bank 0 without using the BSR. Section 5.3.2 "Access Bank" provides a detailed description of the Access RAM.

### 5.3.1 BANK SELECT REGISTER (BSR)

Large areas of data memory require an efficient addressing scheme to make rapid access to any address possible. Ideally, this means that an entire address does not need to be provided for each read or write operation. For PIC18 devices, this is accomplished with a RAM banking scheme. This divides the memory space into 16 contiguous banks of 256 bytes. Depending on the instruction, each location can be addressed directly by its full 12-bit address, or an 8-bit low-order address and a 4-bit bank pointer.

Most instructions in the PIC18 instruction set make use of the bank pointer, known as the Bank Select Register (BSR). This SFR holds the 4 Most Significant bits of a location's address; the instruction itself includes the 8 Least Significant bits. Only the four lower bits of the BSR are implemented (BSR3:BSR0). The upper four bits are unused; they will always read ' 0 ' and cannot be written to. The BSR can be loaded directly by using the MOVLB instruction.
The value of the BSR indicates the bank in data memory. The 8 bits in the instruction show the location in the bank and can be thought of as an offset from the bank's lower boundary. The relationship between the BSR's value and the bank division in data memory is shown in Figure 5-6.
Since up to 16 registers may share the same low-order address, the user must always be careful to ensure that the proper bank is selected before performing a data read or write. For example, writing what should be program data to an 8 -bit address of F9h while the BSR is $0 F h$ will end up resetting the program counter.
While any bank can be selected, only those banks that are actually implemented can be read or written to. Writes to unimplemented banks are ignored, while reads from unimplemented banks will return 'o's. Even so, the STATUS register will still be affected as if the operation was successful. The data memory map in Figure 5-5 indicates which banks are implemented.

In the core PIC18 instruction set, only the MOVFF instruction fully specifies the 12-bit address of the source and target registers. This instruction ignores the BSR completely when it executes. All other instructions include only the low-order address as an operand and must use either the BSR or the Access Bank to locate their target registers.

## PIC18F45J10 FAMILY

FIGURE 5-5: DATA MEMORY MAP FOR PIC18F24J10/44J10 DEVICES


FIGURE 5-6: USE OF THE BANK SELECT REGISTER (DIRECT ADDRESSING)


Note 1: The Access RAM bit of the instruction can be used to force an override of the selected bank (BSR<3:0>) to the registers of the Access Bank.
2: The MOVFF instruction embeds the entire 12-bit address in the instruction.

### 5.3.2 ACCESS BANK

While the use of the BSR with an embedded 8-bit address allows users to address the entire range of data memory, it also means that the user must always ensure that the correct bank is selected. Otherwise, data may be read from or written to the wrong location. This can be disastrous if a GPR is the intended target of an operation but an SFR is written to instead. Verifying and/or changing the BSR for each read or write to data memory can become very inefficient.
To streamline access for the most commonly used data memory locations, the data memory is configured with an Access Bank, which allows users to access a mapped block of memory without specifying a BSR. The Access Bank consists of the first 128 bytes of memory (00h-7Fh) in Bank 0 and the last 128 bytes of memory ( $80 \mathrm{~h}-\mathrm{FFh}$ ) in Block 15. The lower half is known as the "Access RAM" and is composed of GPRs. This upper half is also where the device's SFRs are mapped. These two areas are mapped contiguously in the Access Bank and can be addressed in a linear fashion by an 8-bit address (Figure 5-5).
The Access Bank is used by core PIC18 instructions that include the Access RAM bit (the ' $a$ ' parameter in the instruction). When ' $a$ ' is equal to ' 1 ', the instruction uses the BSR and the 8-bit address included in the opcode for the data memory address. When ' $a$ ' is ' 0 ',
however, the instruction is forced to use the Access Bank address map; the current value of the BSR is ignored entirely.
Using this "forced" addressing allows the instruction to operate on a data address in a single cycle without updating the BSR first. For 8-bit addresses of 80h and above, this means that users can evaluate and operate on SFRs more efficiently. The Access RAM below 80h is a good place for data values that the user might need to access rapidly, such as immediate computational results or common program variables. Access RAM also allows for faster and more code efficient context saving and switching of variables.
The mapping of the Access Bank is slightly different when the extended instruction set is enabled (XINST configuration bit =1). This is discussed in more detail in Section 5.5.3 "Mapping the Access Bank in Indexed Literal Offset Mode".

### 5.3.3 GENERAL PURPOSE REGISTER FILE

PIC18 devices may have banked memory in the GPR area. This is data RAM which is available for use by all instructions. GPRs start at the bottom of Bank 0 (address 000h) and grow upwards towards the bottom of the SFR area. GPRs are not initialized by a Power-on Reset and are unchanged on all other Resets.

### 5.3.4 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy the top half of Bank 15 (F80h to FFFh). A list of these registers is given in Table 5-1 and Table 5-2.

The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The reset and interrupt registers are described in their respective chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of a peripheral feature are described in the chapter for that peripheral.
The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as ' 0 's.

## TABLE 5-1: SPECIAL FUNCTION REGISTER MAP FOR PIC18F45J10 FAMILY DEVICES

| Address | Name | Address | Name | Address | Name | Address | Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FFFh | TOSU | FDFh | INDF2 ${ }^{(1)}$ | FBFh | CCPR1H | F9Fh | IPR1 |
| FFEh | TOSH | FDEh | POSTINC2 ${ }^{(1)}$ | FBEh | CCPR1L | F9Eh | PIR1 |
| FFDh | TOSL | FDDh | POSTDEC2 ${ }^{(1)}$ | FBDh | CCP1CON | F9Dh | PIE1 |
| FFCh | STKPTR | FDCh | PREINC2 ${ }^{(1)}$ | FBCh | CCPR2H | F9Ch | -(2) |
| FFBh | PCLATU | FDBh | PLUSW2 ${ }^{(1)}$ | FBBh | CCPR2L | F9Bh | - ${ }^{(2)}$ |
| FFAh | PCLATH | FDAh | FSR2H | FBAh | CCP2CON | F9Ah | $-^{(2)}$ |
| FF9h | PCL | FD9h | FSR2L | FB9h | - ${ }^{(2)}$ | F99h | - ${ }^{(2)}$ |
| FF8h | TBLPTRU | FD8h | STATUS | FB8h | BAUDCON | F98h | - ${ }^{(2)}$ |
| FF7h | TBLPTRH | FD7h | TMROH | FB7h | ECCP1DEL ${ }^{(3)}$ | F97h | - ${ }^{(2)}$ |
| FF6h | TBLPTRL | FD6h | TMROL | FB6h | ECCP1AS ${ }^{(3)}$ | F96h | TRISE ${ }^{(3)}$ |
| FF5h | TABLAT | FD5h | TOCON | FB5h | CVRCON | F95h | TRISD ${ }^{(3)}$ |
| FF4h | PRODH | FD4h | -(2) | FB4h | CMCON | F94h | TRISC |
| FF3h | PRODL | FD3h | OSCCON | FB3h | $\square^{(2)}$ | F93h | TRISB |
| FF2h | INTCON | FD2h | - ${ }^{(2)}$ | FB2h | $-^{(2)}$ | F92h | TRISA |
| FF1h | INTCON2 | FD1h | WDTCON | FB1h | - ${ }^{(2)}$ | F91h | $-^{(2)}$ |
| FFOh | INTCON3 | FDOh | RCON | FB0h | SPBRGH | F90h | - ${ }^{(2)}$ |
| FEFh | INDFO ${ }^{(1)}$ | FCFh | TMR1H | FAFh | SPBRG | F8Fh | - ${ }^{(2)}$ |
| FEEh | POSTINC0 ${ }^{(1)}$ | FCEh | TMR1L | FAEh | RCREG | F8Eh | SSP2BUF |
| FEDh | POSTDEC0 ${ }^{(1)}$ | FCDh | T1CON | FADh | TXREG | F8Dh | LATE ${ }^{(3)}$ |
| FECh | PREINC0 ${ }^{(1)}$ | FCCh | TMR2 | FACh | TXSTA | F8Ch | LATD ${ }^{(3)}$ |
| FEBh | PLUSW0 ${ }^{(1)}$ | FCBh | PR2 | FABh | RCSTA | F8Bh | LATC |
| FEAh | FSROH | FCAh | T2CON | FAAh | - ${ }^{(2)}$ | F8Ah | LATB |
| FE9h | FSROL | FC9h | SSP1BUF | FA9h | (2) | F89h | LATA |
| FE8h | WREG | FC8h | SSP1ADD | FA8h | - ${ }^{(2)}$ | F88h | SSP2ADD ${ }^{(3)}$ |
| FE7h | INDF1 ${ }^{(1)}$ | FC7h | SSP1STAT | FA7h | EECON2 ${ }^{(1)}$ | F87h | SSP2STAT ${ }^{(3)}$ |
| FE6h | POSTINC1 ${ }^{(1)}$ | FC6h | SSP1CON1 | FA6h | EECON1 | F86h | SSP2CON1 ${ }^{(3)}$ |
| FE5h | POSTDEC $1^{(1)}$ | FC5h | SSP1CON2 | FA5h | IPR3 | F85h | SSP2CON2 ${ }^{(3)}$ |
| FE4h | PREINC1 ${ }^{(1)}$ | FC4h | ADRESH | FA4h | PIR3 | F84h | PORTE ${ }^{(3)}$ |
| FE3h | PLUSW ${ }^{(1)}$ | FC3h | ADRESL | FA3h | PIE3 | F83h | PORTD ${ }^{(3)}$ |
| FE2h | FSR1H | FC2h | ADCON0 | FA2h | IPR2 | F82h | PORTC |
| FE1h | FSR1L | FC1h | ADCON1 | FA1h | PIR2 | F81h | PORTB |
| FEOh | BSR | FCOh | ADCON2 | FAOh | PIE2 | F80h | PORTA |

Note 1: This is not a physical register.
2: Unimplemented registers are read as ' 0 '.
3: This register is not available in 28 -pin devices.

TABLE 5-2: REGISTER FILE SUMMARY (PIC18F24J10/25J10/44J10/45J10)

| File Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on POR, BOR | Details on page: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TOSU | - | - | - | Top-of-Stack | Upper Byte | <20:16>) |  |  | ---0 0000 | 43, 48 |
| TOSH | Top-of-Stack High Byte (TOS<15:8>) |  |  |  |  |  |  |  | 00000000 | 43, 48 |
| TOSL | Top-of-Stack Low Byte (TOS<7:0>) |  |  |  |  |  |  |  | 00000000 | 43, 48 |
| STKPTR | STKFUL | STKUNF | - | Return Stack Pointer |  |  |  |  | 00-0 0000 | 43, 49 |
| PCLATU | - | - | - | Holding Register for PC<20:16> |  |  |  |  | ---0 0000 | 43, 48 |
| PCLATH | Holding Register for $\mathrm{PC}<15: 8>$ |  |  |  |  |  |  |  | 00000000 | 43, 48 |
| PCL | PC Low Byte (PC<7:0>) |  |  |  |  |  |  |  | 00000000 | 43, 48 |
| TBLPTRU | - | - | bit 21 | Program Memory Table Pointer Upper Byte (TBLPTR<20:16>) |  |  |  |  | --00 0000 | 43, 70 |
| TBLPTRH | Program Memory Table Pointer High Byte (TBLPTR<15:8>) |  |  |  |  |  |  |  | 00000000 | 43, 70 |
| TBLPTRL | Program Memory Table Pointer Low Byte (TBLPTR<7:0>) |  |  |  |  |  |  |  | 00000000 | 43, 70 |
| TABLAT | Program Memory Table Latch |  |  |  |  |  |  |  | 00000000 | 43, 70 |
| PRODH | Product Register High Byte |  |  |  |  |  |  |  | xxxx xxxx | 43, 77 |
| PRODL | Product Register Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 43, 77 |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 0000 000x | 43, 81 |
| INTCON2 | $\overline{\text { RBPU }}$ | INTEDGO | INTEDG1 | INTEDG2 | - | TMROIP | - | RBIP | 1111 -1-1 | 43, 82 |
| INTCON3 | INT2IP | INT1IP | - | INT2IE | INT1IE | - | INT2IF | INT1IF | 11-0 0-00 | 43, 83 |
| INDF0 | Uses contents of FSR0 to address data memory - value of FSR0 not changed (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| POSTINC0 | Uses contents of FSR0 to address data memory - value of FSR0 post-incremented (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| POSTDEC0 | Uses contents of FSR0 to address data memory - value of FSR0 post-decremented (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| PREINC0 | Uses contents of FSR0 to address data memory - value of FSR0 pre-incremented (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| PLUSW0 | Uses contents of FSRO to address data memory - value of FSRO pre-incremented (not a physical register) value of FSRO offset by W |  |  |  |  |  |  |  | N/A | 43, 62 |
| FSROH | - | - | - | - | Indirect Da | mory Add | Pointer | Byte | - xxxx | 43, 62 |
| FSROL | Indirect Data Memory Address Pointer 0 Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 43, 62 |
| WREG | Working Register |  |  |  |  |  |  |  | xxxx xxxx | 43 |
| INDF1 | Uses contents of FSR1 to address data memory - value of FSR1 not changed (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| POSTINC1 | Uses contents of FSR1 to address data memory - value of FSR1 post-incremented (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| POSTDEC1 | Uses contents of FSR1 to address data memory - value of FSR1 post-decremented (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| PREINC1 | Uses contents of FSR1 to address data memory - value of FSR1 pre-incremented (not a physical register) |  |  |  |  |  |  |  | N/A | 43, 62 |
| PLUSW1 | Uses contents of FSR1 to address data memory - value of FSR1 pre-incremented (not a physical register) value of FSR1 offset by W |  |  |  |  |  |  |  | N/A | 43, 62 |
| FSR1H | - | - | - | - | Indirect Da | mory Add | Pointer | Byte | - xxxx | 43, 62 |
| FSR1L | Indirect Data Memory Address Pointer 1 Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 43, 62 |
| BSR | - | - | - | - | Bank Sele | gister |  |  | ---- 0000 | 43, 53 |
| INDF2 | Uses contents of FSR2 to address data memory - value of FSR2 not changed (not a physical register) |  |  |  |  |  |  |  | N/A | 44, 62 |
| POSTINC2 | Uses contents of FSR2 to address data memory - value of FSR2 post-incremented (not a physical register) |  |  |  |  |  |  |  | N/A | 44, 62 |
| POSTDEC2 | Uses contents of FSR2 to address data memory - value of FSR2 post-decremented (not a physical register) |  |  |  |  |  |  |  | N/A | 44, 62 |
| PREINC2 | Uses contents of FSR2 to address data memory - value of FSR2 pre-incremented (not a physical register) |  |  |  |  |  |  |  | N/A | 44, 62 |
| PLUSW2 | Uses contents of FSR2 to address data memory - value of FSR2 pre-incremented (not a physical register) value of FSR2 offset by W |  |  |  |  |  |  |  | N/A | 44, 62 |
| FSR2H | - | - | - | - | Indirect Da | mory Add | Pointer | Byte | ---- xxxx | 44, 62 |
| FSR2L | Indirect Data Memory Address Pointer 2 Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 44, 62 |
| STATUS | - | - | - | N | OV | Z | DC | C | ---x xxxx | 44, 60 |

Legend: $\quad \mathrm{x}=$ unknown, $\mathrm{u}=$ unchanged, $-=$ unimplemented, $\mathrm{q}=$ value depends on condition
Note 1: See Section 4.4 "Brown-out Reset (BOR) (PIC18F2XJ10/4XJ10 Devices Only)".
2: These registers and/or bits are not implemented on 28-pin devices and are read as ' 0 '. Reset values are shown for 40/44-pin devices; individual unimplemented bits should be interpreted as '-'.

## PIC18F45J10 FAMILY

TABLE 5-2: REGISTER FILE SUMMARY (PIC18F24J10/25J10/44J10/45J10) (CONTINUED)

| File Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on POR, BOR | Details on page: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMROH | Timer0 Register High Byte |  |  |  |  |  |  |  | 00000000 | 44, 113 |
| TMROL | Timer0 Register Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 44, 113 |
| TOCON | TMR0ON | T08BIT | TOCS | TOSE | PSA | TOPS2 | TOPS1 | TOPSO | 11111111 | 44, 111 |
| OSCCON | IDLEN | - | - | - | OSTS | - | SCS1 | SCSO | 0--- q-00 | 28, 44 |
| WDTCON | - | - | - | - | - | - | - | SWDTEN | --- ---0 | 44, 235 |
| RCON | IPEN | - | - | $\overline{\mathrm{RI}}$ | TO | $\overline{P D}$ | $\overline{\text { POR }}$ | $\overline{\mathrm{BOR}}{ }^{(1)}$ | 0--1 11q0 | 38, 42, 90 |
| TMR1H | Timer1 Register High Byte |  |  |  |  |  |  |  | xxxx xxxx | 44, 119 |
| TMR1L | Timer1 Register Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 44, 119 |
| T1CON | RD16 | T1RUN | T1CKPS1 | T1CKPS0 | T1OSCEN | T1SYNC | TMR1CS | TMR1ON | 00000000 | 44, 115 |
| TMR2 | Timer2 Register |  |  |  |  |  |  |  | 00000000 | 44, 122 |
| PR2 | Timer2 Period Register |  |  |  |  |  |  |  | 11111111 | 44,122 |
| T2CON | - | T2OUTPS3 | T2OUTPS2 | T2OUTPS1 | T2OUTPS0 | TMR2ON | T2CKPS1 | T2CKPS0 | -000 0000 | 44, 121 |
| SSP1BUF | MSSP1 Receive Buffer/Transmit Register |  |  |  |  |  |  |  | xxxx xxxx | 44, 154 |
| SSP1ADD | MSSP1 Address Register in $\mathrm{I}^{2} \mathrm{C}^{\text {M }}$ Slave mode. MSSP1 Baud Rate Reload Register in $\mathrm{I}^{2} \mathrm{C}$ Master mode. |  |  |  |  |  |  |  | 00000000 | 44, 155 |
| SSP1STAT | SMP | CKE | D/ $\bar{A}$ | P | S | $\mathrm{R} / \bar{W}$ | UA | BF | 00000000 | $\begin{gathered} 44,146 \\ 156 \end{gathered}$ |
| SSP1CON1 | WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 | 00000000 | $\begin{gathered} 44,147, \\ 157 \end{gathered}$ |
| SSP1CON2 | GCEN | ACKSTAT | ACKDT | ACKEN | RCEN | PEN | RSEN | SEN | 00000000 | 44, 158 |
| ADRESH | A/D Result Register High Byte |  |  |  |  |  |  |  | xxxx xxxx | 44,218 |
| ADRESL | A/D Result Register Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 44,218 |
| ADCON0 | ADCAL | - | CHS3 | CHS2 | CHS1 | CHSO | GO/DONE | ADON | 0-00 0000 | 44,209 |
| ADCON1 | - | - | VCFG1 | VCFG0 | PCFG3 | PCFG2 | PCFG1 | PCFG0 | --00 0qqq | 44, 210 |
| ADCON2 | ADFM | - | ACQT2 | ACQT1 | ACQT0 | ADCS2 | ADCS1 | ADCSO | 0-00 0000 | 44, 211 |
| CCPR1H | Capture/Compare/PWM Register 1 High Byte |  |  |  |  |  |  |  | xxxx xxxx | 45, 124 |
| CCPR1L | Capture/Compare/PWM Register 1 Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 45, 124 |
| CCP1CON | P1M1 ${ }^{(2)}$ | P1M0 ${ }^{(2)}$ | DC1B1 | DC1B0 | CCP1M3 | CCP1M2 | CCP1M1 | CCP1M0 | 00000000 | 45, 123, |
| CCPR2H | Capture/Compare/PWM Register 2 High Byte |  |  |  |  |  |  |  | xxxx xxxx | 45, 124 |
| CCPR2L | Capture/Compare/PWM Register 2 Low Byte |  |  |  |  |  |  |  | xxxx xxxx | 45, 124 |
| CCP2CON | - | - | DC2B1 | DC2B0 | CCP2M3 | CCP2M2 | CCP2M1 | CCP2M0 | --00 0000 | 45, 123 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 01-0 0-00 | 45, 190 |
| ECCP1DEL | PRSEN | PDC6 ${ }^{(2)}$ | PDC5 ${ }^{(2)}$ | PDC4 ${ }^{(2)}$ | PDC3 ${ }^{(2)}$ | PDC2 ${ }^{(2)}$ | PDC1 ${ }^{(2)}$ | PDC0 ${ }^{(2)}$ | 00000000 | 45, 140 |
| ECCP1AS | ECCPASE | ECCPAS2 | ECCPAS1 | ECCPAS0 | PSSAC1 | PSSAC0 | PSSBD1 ${ }^{(2)}$ | PSSBD0 ${ }^{(2)}$ | 00000000 | 45, 141 |
| CVRCON | CVREN | CVROE | CVRR | CVRSS | CVR3 | CVR2 | CVR1 | CVR0 | 00000000 | 45, 225 |
| CMCON | C2OUT | C1OUT | C2INV | C1INV | CIS | CM2 | CM1 | CM0 | 00000111 | 45, 219 |

Legend: $\quad \mathrm{x}=$ unknown, $\mathrm{u}=$ unchanged, $-=$ unimplemented, $\mathrm{q}=$ value depends on condition
Note 1: See Section 4.4 "Brown-out Reset (BOR) (PIC18F2XJ10/4XJ10 Devices Only)".
2: These registers and/or bits are not implemented on 28 -pin devices and are read as ' 0 '. Reset values are shown for 40/44-pin devices; individual unimplemented bits should be interpreted as '-'.

TABLE 5-2: REGISTER FILE SUMMARY (PIC18F24J10/25J10/44J10/45J10) (CONTINUED)

| File Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Value on POR, BOR | Details on page: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 00000000 | 45, 192 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 00000000 | 45, 192 |
| RCREG | EUSART Receive Register |  |  |  |  |  |  |  | 00000000 | 45, 199 |
| TXREG | EUSART Transmit Register |  |  |  |  |  |  |  | xxxx xxxx | 45, 197 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 00000010 | 45, 188 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 0000 000x | 45, 189 |
| EECON2 | EEPROM Control Register 2 (not a physical register) |  |  |  |  |  |  |  | 00000000 | 45, 68 |
| EECON1 | - | - | - | FREE | WRERR | WREN | WR | - | ---0 x00- | 45, 69 |
| IPR3 | SSP2IP | BCL2IP | - | - | - | - | - | - | 11-- ---- | 45, 89 |
| PIR3 | SSP2IF | BCL2IF | - | - | - | - | - | - | 00-- ---- | 45, 85 |
| PIE3 | SSP2IE | BCL2IE | - | - | - | - | - | - | 00-- ---- | 45, 87 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 11-- 1--1 | 45, 89 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 00-- 0--0 | 45, 85 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 00-- 0--0 | 45, 87 |
| IPR1 | PSPIP ${ }^{(2)}$ | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 11111111 | 45, 88 |
| PIR1 | PSPIF ${ }^{(2)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 00000000 | 45, 84 |
| PIE1 | PSPIE ${ }^{(2)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 00000000 | 45, 86 |
| TRISE ${ }^{(2)}$ | IBF | OBF | IBOV | PSPMODE | - | TRISE2 | TRISE1 | TRISE0 | $1111-111$ | 46, 107 |
| TRISD ${ }^{(2)}$ | PORTD Data Direction Control Register |  |  |  |  |  |  |  | 11111111 | 46, 103 |
| TRISC | PORTC Data Direction Control Register |  |  |  |  |  |  |  | 11111111 | 46, 100 |
| TRISB | PORTB Data Direction Control Register |  |  |  |  |  |  |  | 11111111 | 46, 97 |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISAO | --1- 1111 | 46, 94 |
| SSP2BUF | MSSP2 Receive Buffer/Transmit Register |  |  |  |  |  |  |  | xxxx xxxx | 46, 154 |
| LATE ${ }^{(2)}$ | - | - | - | - | - | PORTE Da <br> (Read and | atch Regis e to Data |  | ---- -xxx | 46, 106 |
| LATD ${ }^{(2)}$ | PORTD Data Latch Register (Read and Write to Data Latch) |  |  |  |  |  |  |  | xxxx xxxx | 46, 103 |
| LATC | PORTC Data Latch Register (Read and Write to Data Latch) |  |  |  |  |  |  |  | xxxx xxxx | 46, 100 |
| LATB | PORTB Data Latch Register (Read and Write to Data Latch) |  |  |  |  |  |  |  | xxxx $x x x x$ | 46, 97 |
| LATA | - | - | PORTA Data | Latch Registe | Read and | Write to Data |  |  | --xx xxxx | 46, 94 |
| SSP2ADD | MSSP2 Address Register in $I^{2} \mathrm{C}^{\text {TM }}$ Slave mode. MSSP2 Baud Rate Reload Register in $\mathrm{I}^{2} \mathrm{C}$ Master mode. |  |  |  |  |  |  |  | 00000000 | 46, 154 |
| SSP2STAT | SMP | CKE | D/ $\bar{A}$ | P | S | R/W | UA | BF | 00000000 | $\begin{gathered} 46,146, \\ 156 \end{gathered}$ |
| SSP2CON1 | WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 | 00000000 | $\begin{gathered} 46,147, \\ 157 \end{gathered}$ |
| SSP2CON2 | GCEN | ACKSTAT | ACKDT | ACKEN | RCEN | PEN | RSEN | SEN | 00000000 | 46, 158 |
| PORTE ${ }^{(2)}$ | - | - | - | - | - | RE2 ${ }^{(2)}$ | $R E 1^{(2)}$ | RE0 ${ }^{(2)}$ | ---- -xxx | 46, 106 |
| PORTD ${ }^{(2)}$ | RD7 | RD6 | RD5 | RD4 | RD3 | RD2 | RD1 | RD0 | xxxx xxxx | 46, 103 |
| PORTC | RC7 | RC6 | RC5 | RC4 | RC3 | RC2 | RC1 | RC0 | xxxx xxxx | 46, 100 |
| PORTB | RB7 | RB6 | RB5 | RB4 | RB3 | RB2 | RB1 | RB0 | xxxx xxxx | 46, 97 |
| PORTA | - | - | RA5 | - | RA3 | RA2 | RA1 | RAO | --0-0000 | 46, 94 |

Legend: $\quad x=$ unknown, $u=$ unchanged, $-=$ unimplemented, $q=$ value depends on condition
Note 1: See Section 4.4 "Brown-out Reset (BOR) (PIC18F2XJ10/4XJ10 Devices Only)".
2: These registers and/or bits are not implemented on 28 -pin devices and are read as ' 0 '. Reset values are shown for 40/44-pin devices; individual unimplemented bits should be interpreted as '-'.

### 5.3.5 STATUS REGISTER

The STATUS register, shown in Register 5-2, contains the arithmetic status of the ALU. As with any other SFR, it can be the operand for any instruction.
If the STATUS register is the destination for an instruction that affects the $\mathrm{Z}, \mathrm{DC}, \mathrm{C}, \mathrm{OV}$ or N bits, the results of the instruction are not written; instead, the STATUS register is updated according to the instruction performed. Therefore, the result of an instruction with the STATUS register as its destination may be different than intended. As an example, CLRF STATUS will set the $Z$ bit and leave the remaining status bits unchanged ('000u uluu').

It is recommended that only BCF, BSF, SWAPF, MOVFF and MOVWF instructions are used to alter the STATUS register, because these instructions do not affect the $Z$, $\mathrm{C}, \mathrm{DC}, \mathrm{OV}$ or N bits in the STATUS register.
For other instructions that do not affect Status bits, see the instruction set summaries in Table 21-2 and Table 21-3.

Note: The C and DC bits operate as the $\overline{\text { borrow }}$ and digit borrow bits, respectively, in subtraction.

## REGISTER 5-2: STATUS REGISTER

| U-0 | U-0 | U-0 | R/W-x | R/W-x | R/W-x | R/W-x | R/W-x |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | N | OV | Z | DC | C |
| 7 bit |  |  |  |  |  |  |  |

bit 7-5 Unimplemented: Read as ' 0 '
bit $4 \quad N$ : Negative bit
This bit is used for signed arithmetic (2's complement). It indicates whether the result was negative (ALU MSB = 1).
1 = Result was negative
$0=$ Result was positive
bit $3 \quad$ OV: Overflow bit
This bit is used for signed arithmetic (2's complement). It indicates an overflow of the 7-bit magnitude which causes the sign bit (bit 7 of the result) to change state.
1 = Overflow occurred for signed arithmetic (in this arithmetic operation)
0 = No overflow occurred
bit $2 \quad$ Z: Zero bit
$1=$ The result of an arithmetic or logic operation is zero
$0=$ The result of an arithmetic or logic operation is not zero
bit 1 DC: Digit Carry/borrow bit
For ADDWF, ADDLW, SUBLW and SUBWF instructions:
$1=$ A carry-out from the 4th low-order bit of the result occurred
$0=$ No carry-out from the 4th low-order bit of the result
Note: For $\overline{b o r r o w}$, the polarity is reversed. A subtraction is executed by adding the 2's complement of the second operand. For rotate (RRF, RLF) instructions, this bit is loaded with either bit 4 or bit 3 of the source register.
bit $0 \quad$ C: Carry/borrow bit
For ADDWF, ADDLW, SUBLW and SUBWF instructions:
1 = A carry-out from the Most Significant bit of the result occurred
$0=$ No carry-out from the Most Significant bit of the result occurred
Note: For borrow, the polarity is reversed. A subtraction is executed by adding the 2's complement of the second operand. For rotate (RRF, RLF) instructions, this bit is loaded with either the high or low-order bit of the source register.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 5.4 Data Addressing Modes

Note: The execution of some instructions in the core PIC18 instruction set are changed when the PIC18 extended instruction set is enabled. See Section 5.5 "Data Memory and the Extended Instruction Set" for more information.

While the program memory can be addressed in only one way - through the program counter - information in the data memory space can be addressed in several ways. For most instructions, the addressing mode is fixed. Other instructions may use up to three modes, depending on which operands are used and whether or not the extended instruction set is enabled.
The addressing modes are:

- Inherent
- Literal
- Direct
- Indirect

An additional addressing mode, Indexed Literal Offset, is available when the extended instruction set is enabled (XINST configuration bit $=1$ ). Its operation is discussed in greater detail in Section 5.5.1 "Indexed Addressing with Literal Offset".

### 5.4.1 INHERENT AND LITERAL ADDRESSING

Many PIC18 control instructions do not need any argument at all; they either perform an operation that globally affects the device or they operate implicitly on one register. This addressing mode is known as Inherent Addressing. Examples include SLEEP, RESET and DAW.
Other instructions work in a similar way but require an additional explicit argument in the opcode. This is known as Literal Addressing mode because they require some literal value as an argument. Examples include ADDLW and MOVLW, which respectively, add or move a literal value to the W register. Other examples include CALL and GOTO, which include a 20-bit program memory address.

### 5.4.2 DIRECT ADDRESSING

Direct addressing specifies all or part of the source and/or destination address of the operation within the opcode itself. The options are specified by the arguments accompanying the instruction.
In the core PIC18 instruction set, bit-oriented and byteoriented instructions use some version of direct addressing by default. All of these instructions include some 8-bit literal address as their Least Significant Byte. This address specifies either a register address in one of the banks of data RAM (Section 5.3.3 "General Purpose Register File") or a location in the Access Bank (Section 5.3.2 "Access Bank") as the data source for the instruction.

The Access RAM bit ' $a$ ' determines how the address is interpreted. When 'a' is ' 1 ', the contents of the BSR (Section 5.3.1 "Bank Select Register (BSR)") are used with the address to determine the complete 12-bit address of the register. When ' $a$ ' is ' 0 ', the address is interpreted as being a register in the Access Bank. Addressing that uses the Access RAM is sometimes also known as Direct Forced Addressing mode.

A few instructions, such as MOVFF, include the entire 12-bit address (either source or destination) in their opcodes. In these cases, the BSR is ignored entirely.
The destination of the operation's results is determined by the destination bit ' $d$ '. When ' $d$ ' is ' 1 ', the results are stored back in the source register, overwriting its original contents. When ' $d$ ' is ' 0 ', the results are stored in the W register. Instructions without the 'd' argument have a destination that is implicit in the instruction; their destination is either the target register being operated on or the W register.

### 5.4.3 INDIRECT ADDRESSING

Indirect addressing allows the user to access a location in data memory without giving a fixed address in the instruction. This is done by using File Select Registers (FSRs) as pointers to the locations to be read or written to. Since the FSRs are themselves located in RAM as Special File Registers, they can also be directly manipulated under program control. This makes FSRs very useful in implementing data structures, such as tables and arrays in data memory.
The registers for indirect addressing are also implemented with Indirect File Operands (INDFs) that permit automatic manipulation of the pointer value with auto-incrementing, auto-decrementing or offsetting with another value. This allows for efficient code, using loops, such as the example of clearing an entire RAM bank in Example 5-5.

| EXAMPLE 5-5: | HOW TO CLEAR RAM (BANK 1) USING INDIRECT ADDRESSING |  |
| :---: | :---: | :---: |
| LFSR | FSRO, 100h | ; |
| NEXT CLRF | POSTINC0 | ; Clear INDF <br> ; register then <br> ; inc pointer |
| BTFSS | FSROH, 1 | ; All done with <br> ; Bank1? |
| BRA | NEXT | ; NO, clear next |
| CONTINUE |  | ; YES, continue |

### 5.4.3.1 FSR Registers and the INDF Operand

At the core of indirect addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8 -bit registers, FSRnH and FSRnL. The four upper bits of the FSRnH register are not used, so each FSR pair holds a 12 -bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.
Indirect addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as "virtual" registers; they are mapped in the SFR space but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L. Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction's target. The INDF operand is just a convenient way of using the pointer.
Because indirect addressing uses a full 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.

### 5.4.3.2 FSR Registers and POSTINC, POSTDEC, PREINC and PLUSW

In addition to the INDF operand, each FSR register pair also has four additional indirect operands. Like INDF, these are "virtual" registers that cannot be indirectly read or written to. Accessing these registers actually accesses the associated FSR register pair, but also performs a specific action on it stored value. They are:

- POSTDEC: accesses the FSR value, then automatically decrements it by 1 afterwards
- POSTINC: accesses the FSR value, then automatically increments it by 1 afterwards
- PREINC: increments the FSR value by 1 , then uses it in the operation
- PLUSW: adds the signed value of the W register (range of -127 to 128) to that of the FSR and uses the new value in the operation.
In this context, accessing an INDF register uses the value in the FSR registers without changing them. Similarly, accessing a PLUSW register gives the FSR value offset by that in the W register; neither value is actually changed in the operation. Accessing the other virtual registers changes the value of the FSR registers.
Operations on the FSRs with POSTDEC, POSTINC and PREINC affect the entire register pair; that is, rollovers of the FSRnL register, from FFh to 00h, carry over to the FSRnH register. On the other hand, results of these operations do not change the value of any flags in the STATUS register (e.g., Z, N, OV, etc.).

FIGURE 5-7: INDIRECT ADDRESSING


The PLUSW register can be used to implement a form of indexed addressing in the data memory space. By manipulating the value in the W register, users can reach addresses that are fixed offsets from pointer addresses. In some applications, this can be used to implement some powerful program control structure, such as software stacks, inside of data memory.

### 5.4.3.3 Operations by FSRs on FSRs

Indirect addressing operations that target other FSRs or virtual registers represent special cases. For example, using an FSR to point to one of the virtual registers will not result in successful operations. As a specific case, assume that FSROH:FSROL contains FE7h, the address of INDF1. Attempts to read the value of the INDF1 using INDFO as an operand will return 00h. Attempts to write to INDF1 using INDF0 as the operand will result in a NOP.
On the other hand, using the virtual registers to write to an FSR pair may not occur as planned. In these cases, the value will be written to the FSR pair but without any incrementing or decrementing. Thus, writing to INDF2 or POSTDEC2 will write the same value to the FSR2H:FSR2L.
Since the FSRs are physical registers mapped in the SFR space, they can be manipulated through all direct operations. Users should proceed cautiously when working on these registers, particularly if their code uses indirect addressing.
Similarly, operations by indirect addressing are generally permitted on all other SFRs. Users should exercise the appropriate caution that they do not inadvertently change settings that might affect the operation of the device.

### 5.5 Data Memory and the Extended Instruction Set

Enabling the PIC18 extended instruction set (XINST configuration bit $=1$ ) significantly changes certain aspects of data memory and its addressing. Specifically, the use of the Access Bank for many of the core PIC18 instructions is different; this is due to the introduction of a new addressing mode for the data memory space.
What does not change is just as important. The size of the data memory space is unchanged, as well as its linear addressing. The SFR map remains the same. Core PIC18 instructions can still operate in both Direct and Indirect Addressing mode; inherent and literal instructions do not change at all. Indirect addressing with FSR0 and FSR1 also remain unchanged.

### 5.5.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of indirect addressing using the FSR2 register pair within Access RAM. Under the proper conditions, instructions that use the Access Bank - that is, most bit-oriented and byte-oriented instructions can invoke a form of indexed addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset, or Indexed Literal Offset mode.
When using the extended instruction set, this addressing mode requires the following:

- The use of the Access Bank is forced ('a' = 0) and
- The file address argument is less than or equal to 5Fh.
Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in direct addressing), or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an address pointer, specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.


### 5.5.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use direct addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte-oriented and bit-oriented instructions, or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.
Additionally, byte-oriented and bit-oriented instructions are not affected if they do not use the Access Bank (Access RAM bit is ' 1 '), or include a file address of 60 h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled in shown in Figure 5-8.
Those who desire to use byte-oriented or bit-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in Section 21.2.1 "Extended Instruction Syntax".

FIGURE 5-8: COMPARING ADDRESSING OPTIONS FOR BIT-ORIENTED AND BYTE-ORIENTED INSTRUCTIONS (EXTENDED INSTRUCTION SET ENABLED)

Example Instruction: ADDWF, f, d, a (Opcode: 0010 01da ffff ffff)

When ' $a$ ' $=0$ and $\mathbf{f} \geq 60 \mathrm{~h}$ :
The instruction executes in Direct Forced mode. ' $f$ ' is interpreted as a location in the Access RAM between 060h and OFFh. This is the same as locations 060h to 07Fh (Bank 0) and F80h to FFFh (Bank 15) of data memory.
Locations below 60h are not available in this addressing mode.

When ' $a$ ' $=0$ and $\mathrm{f} \leq 5 \mathrm{Fh}$ :
The instruction executes in Indexed Literal Offset mode. 'f' is interpreted as an offset to the address value in FSR2. The two are added together to obtain the address of the target register for the instruction. The address can be anywhere in the data memory space.

Note that in this mode, the correct syntax is now:
ADDWF [k], d
where ' $k$ ' is the same as ' $f$ '.

When ' $a$ ' = 1 (all values of $f$ ): The instruction executes in Direct mode (also known as Direct Long mode). ' $f$ ' is interpreted as a location in one of the 16 banks of the data memory space. The bank is designated by the Bank Select Register (BSR). The address can be in any implemented bank in the data memory space.


### 5.5.3 MAPPING THE ACCESS BANK IN INDEXED LITERAL OFFSET MODE

The use of Indexed Literal Offset Addressing mode effectively changes how the first 96 locations of Access RAM ( 00 h to 5 Fh ) are mapped. Rather than containing just the contents of the bottom half of Bank 0, this mode maps the contents from Bank 0 and a user-defined "window" that can be located anywhere in the data memory space. The value of FSR2 establishes the lower boundary of the addresses mapped into the window, while the upper boundary is defined by FSR2 plus 95 (5Fh). Addresses in the Access RAM above 5Fh are mapped as previously described (see Section 5.3.2 "Access Bank"). An example of Access Bank remapping in this addressing mode is shown in Figure 5-9.

Remapping of the Access Bank applies only to operations using the Indexed Literal Offset mode. Operations that use the BSR (Access RAM bit is ' 1 ') will continue to use direct addressing as before.

### 5.6 PIC18 Instruction Execution and the Extended Instruction Set

Enabling the extended instruction set adds eight additional commands to the existing PIC18 instruction set. These instructions are executed as described in Section 21.2 "Extended Instruction Set".

FIGURE 5-9: REMAPPING THE ACCESS BANK WITH INDEXED LITERAL OFFSET ADDRESSING

## Example Situation:

ADDWF f, d, a
FSR2H:FSR2L $=120 \mathrm{~h}$
Locations in the region from the FSR2 pointer (120h) to the pointer plus 05Fh (17Fh) are mapped to the bottom of the Access RAM (000h-05Fh).
Locations in Bank 0 from 060h to 07Fh are mapped, as usual, to the middle half of the Access Bank.
Special File Registers at F80h through FFFh are mapped to 80h through FFh, as usual.
Bank 0 addresses below 5Fh can still be addressed by using the BSR.


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NOTES:

### 6.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.
A read from program memory is executed on one byte at a time. A write to program memory is executed on blocks of 64 bytes at a time. Program memory is erased in blocks of 1024 bytes at a time. A bulk erase operation may not be issued from user code.
Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase; therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

### 6.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).
Table read operations retrieve data from program memory and place it into the data RAM space. Figure 6-1 shows the operation of a table read with program memory and data RAM.
Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in Section 6.5 "Writing to Flash Program Memory". Figure 6-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word aligned.

## FIGURE 6-1: TABLE READ OPERATION



Note 1: Table Pointer register points to a byte in program memory.

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## FIGURE 6-2: TABLE WRITE OPERATION



### 6.2 Control Registers

Several control registers are used in conjunction with the TBLRD and TBLWT instructions. These include the:

- EECON1 register
- EECON2 register
- TABLAT register
- TBLPTR registers


### 6.2.1 EECON1 AND EECON2 REGISTERS

The EECON1 register (Register 6-1) is the control register for memory accesses. The EECON2 register is not a physical register; it is used exclusively in the memory write and erase sequences. Reading EECON2 will read all 'o's.
The FREE bit, when set, will allow a program memory erase operation. When FREE is set, the erase operation is initiated on the next WR command. When FREE is clear, only writes are enabled.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WR bit is set and cleared when the internal programming timer expires and the write operation is complete.

Note: During normal operation, the WRERR is read as ' 1 '. This can indicate that a write operation was prematurely terminated by a Reset, or a write operation was attempted improperly.

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software; it is cleared in hardware at the completion of the write operation.

REGISTER 6-1: EECON1: DATA EEPROM CONTROL REGISTER 1

| U-0 | U-0 | U-0 | R/W-0 | R/W-x | R/W-0 |  | R/S-0 |  | U-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | FREE | WRERR | WREN | WR | - |  |  |
| bit 7 |  |  |  |  |  |  |  |  |  |

bit 7-5 Unimplemented: Read as ' 0 '
bit 4 FREE: Flash Row Erase Enable bit
1 = Erase the program memory row addressed by TBLPTR on the next WR command (cleared by completion of erase operation)
$0=$ Perform write only
bit 3 WRERR: Flash Program/Data EEPROM Error Flag bit
$1=A$ write operation is prematurely terminated (any Reset during self-timed programming in normal operation, or an improper write attempt)
$0=$ The write operation completed
bit 2 WREN: Flash Program/Data EEPROM Write Enable bit 1 = Allows write cycles to Flash program/data EEPROM
$0=$ Inhibits write cycles to Flash program/data EEPROM
bit $1 \quad$ WR: Write Control bit
1 = Initiates a data EEPROM erase/write cycle or a program memory erase cycle or write cycle.
(The operation is self-timed and the bit is cleared by hardware once write is complete.
The WR bit can only be set (not cleared) in software.)
$0=$ Write cycle to the EEPROM is complete
bit $0 \quad$ Unimplemented: Read as ' 0 '

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit |  |
| $S=$ Bit can be set by software, but not cleared | $U=$ Unimplemented bit, read as ' 0 ' |  |
| $-n=$ Value at POR | $' 1$ ' = Bit is set | $\prime 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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### 6.2.2 TABLAT - TABLE LATCH REGISTER

The Table Latch (TABLAT) is an 8-bit register mapped into the SFR space. The Table Latch register is used to hold 8-bit data during data transfers between program memory and data RAM.

### 6.2.3 TBLPTR - TABLE POINTER REGISTER

The Table Pointer (TBLPTR) register addresses a byte within the program memory. The TBLPTR is comprised of three SFR registers: Table Pointer Upper Byte, Table Pointer High Byte and Table Pointer Low Byte (TBLPTRU:TBLPTRH:TBLPTRL). These three registers join to form a 22-bit wide pointer. The low-order 21 bits allow the device to address up to 2 Mbytes of program memory space. The 22nd bit allows access to the device ID, the user ID and the configuration bits.
The Table Pointer register, TBLPTR, is used by the TBLRD and TBLWT instructions. These instructions can update the TBLPTR in one of four ways based on the table operation. These operations are shown in Table 6-1. These operations on the TBLPTR only affect the low-order 21 bits.

### 6.2.4 TABLE POINTER BOUNDARIES

TBLPTR is used in reads, writes and erases of the Flash program memory.

When a TBLRD is executed, all 22 bits of the TBLPTR determine which byte is read from program memory into TABLAT.
When a TBLWT is executed, the six LSbs of the Table Pointer register (TBLPTR<5:0>) determine which of the 64 program memory holding registers is written to. When the timed write to program memory begins (via the WR bit), the 16 MSbs of the TBLPTR (TBLPTR<21:6>) determine which program memory block of 64 bytes is written to. For more detail, see Section 6.5 "Writing to Flash Program Memory".
When an erase of program memory is executed, the 7 MSbs of the Table Pointer register (TBLPTR<21:10>) point to the 1024-byte block that will be erased. The Least Significant bits (TBLPTR<9:0>) are ignored.
Figure 6-3 describes the relevant boundaries of TBLPTR based on Flash program memory operations.

## TABLE 6-1: TABLE POINTER OPERATIONS WITH TBLRD AND TBLWT INSTRUCTIONS

| Example | Operation on Table Pointer |
| :--- | :---: |
| TBLRD* <br> TBLWT* | TBLPTR is not modified |
| TBLRD* + <br> TBLWT* + | TBLPTR is incremented after the read/write |
| TBLRD* <br> TBLWT* | TBLPTR is decremented after the read/write |
| TBLRD+* <br> TBLWT+* | TBLPTR is incremented before the read/write |

FIGURE 6-3: TABLE POINTER BOUNDARIES BASED ON OPERATION


### 6.3 Reading the Flash Program Memory

The TBLRD instruction is used to retrieve data from program memory and places it into data RAM. Table reads from program memory are performed one byte at a time.

TBLPTR points to a byte address in program space. Executing TBLRD places the byte pointed to into TABLAT. In addition, TBLPTR can be modified automatically for the next table read operation.

The internal program memory is typically organized by words. The Least Significant bit of the address selects between the high and low bytes of the word. Figure 6-4 shows the interface between the internal program memory and the TABLAT.

FIGURE 6-4: READS FROM FLASH PROGRAM MEMORY


EXAMPLE 6-1: READING A FLASH PROGRAM MEMORY WORD


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### 6.4 Erasing Flash Program Memory

The minimum erase block is 1024 bytes. Only through the use of an external programmer, or through ICSP control, can larger blocks of program memory be bulk erased. Word erase in the Flash array is not supported.

When initiating an erase sequence from the microcontroller itself, a block of 1024 bytes of program memory is erased. The Most Significant 7 bits of the TBLPTR<21:10> point to the block being erased. TBLPTR<9:0> are ignored.
The EECON1 register commands the erase operation. The WREN bit must be set to enable write operations. The FREE bit is set to select an erase operation.
For protection, the write initiate sequence for EECON2 must be used.
A long write is necessary for erasing the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

### 6.4.1 <br> FLASH PROGRAM MEMORY ERASE SEQUENCE

The sequence of events for erasing a block of internal program memory location is:

1. Load Table Pointer register with address of row being erased.
2. Set the EECON1 register for the erase operation:

- set WREN bit to enable writes;
- set FREE bit to enable the erase.

3. Disable interrupts.
4. Write 55h to EECON2.
5. Write OAAh to EECON2.
6. Set the WR bit. This will begin the row erase cycle.
7. The CPU will stall for duration of the erase (about 2 ms using internal timer).
8. Re-enable interrupts.

## EXAMPLE 6-2: ERASING A FLASH PROGRAM MEMORY ROW

|  | MOVLW | CODE_ADDR_UPPER | ; load TBLPTR with the base |
| :---: | :---: | :---: | :---: |
|  | MOVWF | TBLPTRU | ; address of the memory block |
|  | MOVLW | CODE_ADDR_HIGH |  |
|  | MOVWF | TBLPTRH |  |
|  | MOVLW | CODE_ADDR_LOW |  |
|  | MOVWF | TBLPTRL |  |
| ERASE_ROW |  |  |  |
|  | BSF | EECON1, WREN | ; enable write to memory |
|  | BSF | EECON1, FREE | ; enable Row Erase operation |
|  | BCF | INTCON, GIE | ; disable interrupts |
| Required | MOVLW | 55h |  |
| Sequence | MOVWF | EECON2 | ; write 55h |
|  | MOVLW | OAAh |  |
|  | MOVWF | EECON2 | ; write OAAh |
|  | BSF | EECON1, WR | ; start erase (CPU stall) |
|  | BSF | INTCON, GIE | ; re-enable interrupts |

### 6.5 Writing to Flash Program Memory

The minimum programming block is 32 words or 64 bytes. Word or byte programming is not supported.
Table writes are used internally to load the holding registers needed to program the Flash memory. There are 64 holding registers used by the table writes for programming.
Since the Table Latch (TABLAT) is only a single byte, the TBLWT instruction may need to be executed 64 times for each programming operation. All of the table write operations will essentially be short writes because only the holding registers are written. At the end of updating the 64 holding registers, the EECON1 register must be written to in order to start the programming operation with a long write.
The long write is necessary for programming the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

The EEPROM on-chip timer controls the write time. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

Note: Unlike previous devices, the PIC18F45J10 family of devices does not reset the holding registers after a write occurs. The holding registers must be cleared or overwritten before a programming sequence.
In order to maintain the endurance of the cells, each Flash byte should not be programmed more then twice between erase operations. Either a bulk or row erase of the target row is required before attempting to modify the contents a third time.

FIGURE 6-5: TABLE WRITES TO FLASH PROGRAM MEMORY


### 6.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

1. If the section of program memory to be written to has been programmed previously, then the memory will need to be erased before the write occurs (see 6.4.1 "Flash Program Memory Erase Sequence").
2. Write the 64 bytes into the holding registers with auto-increment.
3. Set the EECON1 register for the write operation: - set WREN to enable byte writes.
4. Disable interrupts.
5. Write 55h to EECON2.
6. Write OAAh to EECON2.
7. Set the WR bit. This will begin the write cycle.
8. The CPU will stall for duration of the write (about 2 ms using internal timer).
9. Re-enable interrupts.
10. Verify the memory (table read).

An example of the required code is shown in Example 6-3.

Note: Before setting the WR bit, the Table Pointer address needs to be within the intended address range of the 64 bytes in the holding register.

## EXAMPLE 6-3: WRITING TO FLASH PROGRAM MEMORY

|  | MOVLW | CODE_ADDR_UPPER | ; Load TBLPTR with the base |
| :---: | :---: | :---: | :---: |
|  | MOVWF | TBLPTRU | ; address of the memory block |
|  | MOVLW | CODE_ADDR_HIGH |  |
|  | MOVWF | TBLPTRH |  |
|  | MOVLW | CODE_ADDR_LOW |  |
|  | MOVWF | TBLPTRL |  |
| ERASE_BLOCK |  |  |  |
|  | BSF | EECON1, WREN | ; enable write to memory |
|  | BSF | EECON1, FREE | ; enable Row Erase operation |
|  | BCF | INTCON, GIE | ; disable interrupts |
|  | MOVLW | 55h |  |
|  | MOVWF | EECON2 | ; write 55h |
|  | MOVLW | OAAh |  |
|  | MOVWF | EECON2 | ; write OAAh |
|  | BSF | EECON1, WR | ; start erase (CPU stall) |
|  | BSF | INTCON, GIE | ; re-enable interrupts |
|  | MOVLW | D'16' |  |
|  | MOVWF | WRITE_COUNTER | ; Need to write 16 blocks of 64 to write <br> ; one erase block of 1024 |
| RESTART_BUFFER |  |  |  |
|  | MOVLW | D'64' |  |
|  | MOVWF | COUNTER |  |
|  | MOVLW | BUFFER_ADDR_HIGH | ; point to buffer |
|  | MOVWF | FSROH |  |
|  | MOVLW | BUFFER_ADDR_LOW |  |
|  | MOVWF | FSROL |  |
| FILL_BUFFER |  |  |  |
|  | $\cdots$ |  | ; read the new data from I2C, SPI, <br> ; PSP, USART, etc. |
| WRITE_BUFFER |  |  |  |
|  | MOVLW | D'64 | ; number of bytes in holding register |
|  | MOVWF | COUNTER |  |
| WRITE_BYTE_TO_HREGS |  |  |  |
|  | MOVFF | POSTINCO, WREG | ; get low byte of buffer data |
|  | MOVWF | TABLAT | ; present data to table latch |
|  | TBLWT+* |  | ; write data, perform a short write <br> ; to internal TBLWT holding register. |
|  | DECFSZ | COUNTER | ; loop until buffers are full |
|  | BRA | WRITE_WORD_TO_HREGS |  |
| PROGRAM_MEMORY |  |  |  |
|  | BSF | EECON1, WREN | ; enable write to memory |
|  | BCF | INTCON, GIE | ; disable interrupts |
|  | MOVLW | 55h |  |
| Required | MOVWF | EECON2 | ; write 55h |
| Sequence | MOVLW | 0 AAh |  |
|  | MOVWF | EECON2 | ; write OAAh |
|  | BSF | EECON1, WR | ; start program (CPU stall) |
|  | BSF | INTCON, GIE | ; re-enable interrupts |
|  | $B C F$ | EECON1, WREN | ; disable write to memory |
|  | DECFSZ | WRITE_COUNTER | ; done with one write cycle |
|  | BRA | RESTART_BUFFER | ; if not done replacing the erase block |

### 6.5.2 WRITE VERIFY

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

### 6.5.3 UNEXPECTED TERMINATION OF WRITE OPERATION

If a write is terminated by an unplanned event, such as loss of power or an unexpected Reset, the memory location just programmed should be verified and reprogrammed if needed. If the write operation is interrupted by a $\overline{M C L R}$ Reset or a WDT Time-out Reset during normal operation, the user can check the WRERR bit and rewrite the location(s) as needed.

### 6.5.4 <br> PROTECTION AGAINST SPURIOUS WRITES

To protect against spurious writes to Flash program memory, the write initiate sequence must also be followed. See Section 20.0 "Special Features of the CPU" for more detail.

### 6.6 Flash Program Operation During Code Protection

See Section 20.6 "Program Verification and Code Protection" for details on code protection of Flash program memory.

TABLE 6-2: REGISTERS ASSOCIATED WITH PROGRAM FLASH MEMORY

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TBLPTRU | - | - | bit 21 | Program Memory Table Pointer Upper Byte (TBLPTR<20:16>) |  |  |  |  | 43 |
| TBPLTRH | Program Memory Table Pointer High Byte (TBLPTR<15:8>) |  |  |  |  |  |  |  | 43 |
| TBLPTRL | Program Memory Table Pointer Low Byte (TBLPTR<7:0>) |  |  |  |  |  |  |  | 43 |
| TABLAT | Program Memory Table Latch |  |  |  |  |  |  |  | 43 |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| EECON2 | EEPROM Control Register 2 (not a physical register) |  |  |  |  |  |  |  | 45 |
| EECON1 | - | - | - | FREE | WRERR | WREN | WR | - | 45 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 45 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 45 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used during Flash/EEPROM access.

## PIC18F45J10 FAMILY

NOTES:

## $7.08 \times 8$ HARDWARE MULTIPLIER

### 7.1 Introduction

All PIC18 devices include an $8 \times 8$ hardware multiplier as part of the ALU. The multiplier performs an unsigned operation and yields a 16-bit result that is stored in the product register pair, PRODH:PRODL. The multiplier's operation does not affect any flags in the STATUS register.
Making multiplication a hardware operation allows it to be completed in a single instruction cycle. This has the advantages of higher computational throughput and reduced code size for multiplication algorithms and allows the PIC18 devices to be used in many applications previously reserved for digital signal processors. A comparison of various hardware and software multiply operations, along with the savings in memory and execution time, is shown in Table 7-1.

### 7.2 Operation

Example 7-1 shows the instruction sequence for an $8 \times 8$ unsigned multiplication. Only one instruction is required when one of the arguments is already loaded in the WREG register.
Example 7-2 shows the sequence to do an $8 \times 8$ signed multiplication. To account for the sign bits of the arguments, each argument's Most Significant bit (MSb) is tested and the appropriate subtractions are done.

EXAMPLE 7-1: $8 \times 8$ UNSIGNED
MULTIPLY ROUTINE

| MOVF | ARG1, W | $;$ |
| :--- | :--- | :--- |
| MULWF | ARG2 | $;$ ARG1 * ARG2 -> <br>  |
|  |  |  |

EXAMPLE 7-2: $8 \times 8$ SIGNED MULTIPLY ROUTINE

| MOVF | ARG1, W |  |
| :---: | :---: | :---: |
| MULWF | ARG2 | ; ARG1 * ARG2 -> |
|  |  | ; PRODH:PRODL |
| BTFSC | ARG2, SB | ; Test Sign Bit |
| SUBWF | PRODH, F | ; PRODH $=$ PRODH |
|  |  | ; - ARG1 |
| MOVF | ARG2, W |  |
| BTFSC | ARG1, SB | ; Test Sign Bit |
| SUBWF | PRODH, F | ; PRODH $=$ PRODH |
|  |  | ; - ARG2 |

TABLE 7-1: PERFORMANCE COMPARISON FOR VARIOUS MULTIPLY OPERATIONS

| Routine | Multiply Method | Program Memory (Words) | Cycles (Max) | Time |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | @ 40 MHz | @ 10 MHz | @ 4 MHz |
| $8 \times 8$ unsigned | Without hardware multiply | 13 | 69 | $6.9 \mu \mathrm{~s}$ | 27.6 us | $69 \mu \mathrm{~s}$ |
|  | Hardware multiply | 1 | 1 | 100 ns | 400 ns | $1 \mu \mathrm{~s}$ |
| $8 \times 8$ signed | Without hardware multiply | 33 | 91 | $9.1 \mu \mathrm{~s}$ | 36.4 ¢ | $91 \mu \mathrm{~s}$ |
|  | Hardware multiply | 6 | 6 | 600 ns | $2.4 \mu \mathrm{~s}$ | $6 \mu \mathrm{~s}$ |
| $16 \times 16$ unsigned | Without hardware multiply | 21 | 242 | 24.2 ¢ | 96.8 ¢ | $242 \mu \mathrm{~s}$ |
|  | Hardware multiply | 28 | 28 | $2.8 \mu \mathrm{~s}$ | $11.2 \mu \mathrm{~s}$ | $28 \mu \mathrm{~s}$ |
| $16 \times 16$ signed | Without hardware multiply | 52 | 254 | $25.4 \mu \mathrm{~s}$ | $102.6 \mu \mathrm{~s}$ | $254 \mu \mathrm{~s}$ |
|  | Hardware multiply | 35 | 40 | $4.0 \mu \mathrm{~s}$ | 16.0 ¢s | $40 \mu \mathrm{~s}$ |

Example 7-3 shows the sequence to do a $16 \times 16$ unsigned multiplication. Equation 7-1 shows the algorithm that is used. The 32-bit result is stored in four registers (RES3:RESO).

## EQUATION 7-1: $16 \times 16$ UNSIGNED MULTIPLICATION ALGORITHM

| RES3:RES0 $=$ | ARG1H:ARG1L $\bullet$ ARG2H:ARG2L |
| :--- | :--- |
| $=$ | $\left(\right.$ ARG1H $\bullet$ ARG2H $\left.\bullet 2^{16}\right)+$ |
|  | $\left(\right.$ ARG1H $\bullet$ ARG2L $\left.\bullet 2^{8}\right)+$ |
|  | $\left(\right.$ ARG1L $\bullet$ ARG2H $\left.\bullet 2^{8}\right)+$ |
|  | $($ ARG1L $\bullet$ ARG2L $)$ |

EXAMPLE 7-3: $\quad 16 \times 16$ UNSIGNED

|  | MOVF MULWF | ARG1L, W ARG2L | ; ARG1L * ARG2L-> <br> ; PRODH:PRODL |
| :---: | :---: | :---: | :---: |
|  | MOVFF | PRODH, RES1 | ; |
|  | MOVFF | PRODL, RESO | ; |
| ; | MOVF | ARG1H, W |  |
|  | MULWF | ARG2H | ; ARG1H * ARG2H-> <br> ; PRODH:PRODL |
|  | MOVFF | PRODH, RES3 | ; |
|  | MOVFF | PRODL, RES2 | ; |
| ; |  |  |  |
|  | MOVF | ARG1L, W |  |
|  | MULWF | ARG2H | ; ARG1L * ARG2H-> <br> ; PRODH:PRODL |
|  | MOVF | PRODL, W | ; |
|  | ADDWF | RES1, F | ; Add cross |
|  | MOVF | PRODH, W | ; products |
|  | ADDWFC | RES2, F | ; |
|  | CLRF | WREG | ; |
|  | ADDWFC | RES3, F | ; |
| ; |  |  |  |
|  | MOVF | ARG1H, W | ; |
|  | MULWF | ARG2L | ; ARG1H * ARG2L-> <br> ; PRODH:PRODL |
|  | MOVF | PRODL, W | ; |
|  | ADDWF | RES1, F | ; Add cross |
|  | MOVF | PRODH, W | ; products |
|  | ADDWFC | RES2, F | ; |
|  | CLRF | WREG | ; |
|  | ADDWFC | RES3, F | ; |

Example 7-4 shows the sequence to do a $16 \times 16$ signed multiply. Equation 7-2 shows the algorithm used. The 32-bit result is stored in four registers (RES3:RES0). To account for the sign bits of the arguments, the MSb for each argument pair is tested and the appropriate subtractions are done.

EQUATION 7-2: $16 \times 16$ SIGNED MULTIPLICATION ALGORITHM
RES3:RES0 = ARG1H:ARG1L •ARG2H:ARG2L
$=\left(\right.$ ARG1H $\cdot$ ARG2H $\left.\bullet 2^{16}\right)+$
$\left(\right.$ ARG1H $\bullet$ ARG2L $\left.\bullet 2^{8}\right)+$ (ARG1L•ARG2H • $2^{8}$ ) + (ARG1L•ARG2L) + $\left(-1 \bullet\right.$ ARG2H<7> • ARG1H:ARG1L $\left.\bullet 2^{16}\right)+$ $\left(-1 \bullet\right.$ ARG1H〈7> • ARG2H:ARG2L • $2^{16}$ )

EXAMPLE 7-4: $16 \times 16$ SIGNED MULTIPLY ROUTINE


### 8.0 INTERRUPTS

Members of the PIC18F45J10 family of devices have multiple interrupt sources and an interrupt priority feature that allows most interrupt sources to be assigned a high priority level or a low priority level. The high priority interrupt vector is at 0008 h and the low priority interrupt vector is at 0018 h . High priority interrupt events will interrupt any low priority interrupts that may be in progress.
There are thirteen registers which are used to control interrupt operation. These registers are:

- RCON
- INTCON
- INTCON2
- INTCON3
- PIR1, PIR2, PIR3
- PIE1, PIE2, PIE3
- IPR1, IPR2, IPR3

It is recommended that the Microchip header files supplied with MPLAB ${ }^{\circledR}$ IDE be used for the symbolic bit names in these registers. This allows the assembler/compiler to automatically take care of the placement of these bits within the specified register.
In general, interrupt sources have three bits to control their operation. They are:

- Flag bit to indicate that an interrupt event occurred
- Enable bit that allows program execution to branch to the interrupt vector address when the flag bit is set
- Priority bit to select high priority or low priority

The interrupt priority feature is enabled by setting the IPEN bit (RCON $<7>$ ). When interrupt priority is enabled, there are two bits which enable interrupts globally. Setting the GIEH bit (INTCON $<7>$ ) enables all interrupts that have the priority bit set (high priority). Setting the GIEL bit (INTCON<6>) enables all interrupts that have the priority bit cleared (low priority). When the interrupt flag, enable bit and appropriate global interrupt enable bit are set, the interrupt will vector immediately to address 0008h or 0018h, depending on the priority bit setting. Individual interrupts can be disabled through their corresponding enable bits.

When the IPEN bit is cleared (default state), the interrupt priority feature is disabled and interrupts are compatible with $\mathrm{PICmicro}^{\circledR}$ mid-range devices. In Compatibility mode, the interrupt priority bits for each source have no effect. INTCON<6> is the PEIE bit which enables/disables all peripheral interrupt sources. INTCON<7> is the GIE bit which enables/disables all interrupt sources. All interrupts branch to address 0008h in Compatibility mode.
When an interrupt is responded to, the global interrupt enable bit is cleared to disable further interrupts. If the IPEN bit is cleared, this is the GIE bit. If interrupt priority levels are used, this will be either the GIEH or GIEL bit. High priority interrupt sources can interrupt a low priority interrupt. Low priority interrupts are not processed while high priority interrupts are in progress.
The return address is pushed onto the stack and the PC is loaded with the interrupt vector address (0008h or 0018h). Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bits must be cleared in software before re-enabling interrupts to avoid recursive interrupts.
The "return from interrupt" instruction, RETFIE, exits the interrupt routine and sets the GIE bit (GIEH or GIEL if priority levels are used) which re-enables interrupts.
For external interrupt events, such as the INT pins or the PORTB input change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding enable bit or the GIE bit.

Note: Do not use the MOVFF instruction to modify any of the interrupt control registers while any interrupt is enabled. Doing so may cause erratic microcontroller behavior.

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FIGURE 8-1: PIC18F24J10/25J10/44J10/45J10 INTERRUPT LOGIC


### 8.1 INTCON Registers

The INTCON registers are readable and writable registers which contain various enable, priority and flag bits.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

## REGISTER 8-1: INTCON: INTERRUPT CONTROL REGISTER

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-x |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF |

bit 7
bit 0
bit 7 GIE/GIEH: Global Interrupt Enable bit
When IPEN = 0:
1 = Enables all unmasked interrupts
$0=$ Disables all interrupts
When IPEN = 1:
1 = Enables all high priority interrupts
$0=$ Disables all interrupts
bit 6 PEIE/GIEL: Peripheral Interrupt Enable bit
When IPEN $=0$ :
1 = Enables all unmasked peripheral interrupts
$0=$ Disables all peripheral interrupts
When IPEN = 1:
1 = Enables all low priority peripheral interrupts
$0=$ Disables all low priority peripheral interrupts
bit 5 TMROIE: TMR0 Overflow Interrupt Enable bit
1 = Enables the TMRO overflow interrupt
$0=$ Disables the TMR0 overflow interrupt
bit 4 INTOIE: INTO External Interrupt Enable bit
1 = Enables the INT0 external interrupt
$0=$ Disables the INT0 external interrupt
bit 3 RBIE: RB Port Change Interrupt Enable bit
1 = Enables the RB port change interrupt
$0=$ Disables the RB port change interrupt
bit 2 TMROIF: TMRO Overflow Interrupt Flag bit
$1=$ TMR0 register has overflowed (must be cleared in software)
$0=$ TMRO register did not overflow
bit 1 INTOIF: INTO External Interrupt Flag bit
1 = The INTO external interrupt occurred (must be cleared in software)
$0=$ The INT0 external interrupt did not occur
bit $0 \quad$ RBIF: RB Port Change Interrupt Flag bit
1 = At least one of the RB7:RB4 pins changed state (must be cleared in software)
$0=$ None of the RB7:RB4 pins have changed state
Note: A mismatch condition will continue to set this bit. Reading PORTB will end the mismatch condition and allow the bit to be cleared.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

## PIC18F45J10 FAMILY

REGISTER 8-2: INTCON2: INTERRUPT CONTROL REGISTER 2

| R/W-1 | R/W-1 | R/W-1 | R/W-1 | U-0 | R/W-1 | U-0 | R/W-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { RBPU }}$ | INTEDG0 | INTEDG1 | INTEDG2 | - | TMR0IP | - | RBIP |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad$ RBPU: PORTB Pull-up Enable bit
$1=$ All PORTB pull-ups are disabled
$0=$ PORTB pull-ups are enabled by individual port latch values
bit 6 INTEDGO: External Interrupt 0 Edge Select bit
1 = Interrupt on rising edge
$0=$ Interrupt on falling edge
bit 5 INTEDG1: External Interrupt 1 Edge Select bit
1 = Interrupt on rising edge
$0=$ Interrupt on falling edge
bit 4 INTEDG2: External Interrupt 2 Edge Select bit
1 = Interrupt on rising edge
$0=$ Interrupt on falling edge
bit 3 Unimplemented: Read as ' 0 '
bit 2 TMROIP: TMRO Overflow Interrupt Priority bit
1 = High priority
0 = Low priority
bit 1 Unimplemented: Read as ' 0 '
bit $0 \quad$ RBIP: RB Port Change Interrupt Priority bit
$1=$ High priority
$0=$ Low priority

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 8-3: INTCON3: INTERRUPT CONTROL REGISTER 3

| R/W-1 | R/W-1 | U-0 | R/W-0 | R/W-0 | U-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INT2IP | INT1IP | - | INT2IE | INT1IE | - | INT2IF | INT1IF |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad$ INT2IP: INT2 External Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit $6 \quad$ INT1IP: INT1 External Interrupt Priority bit
1 = High priority
$0=$ Low priority
bit 5 Unimplemented: Read as ' 0 '
bit 4 INT2IE: INT2 External Interrupt Enable bit
1 = Enables the INT2 external interrupt
$0=$ Disables the INT2 external interrupt
bit 3 INT1IE: INT1 External Interrupt Enable bit
1 = Enables the INT1 external interrupt
$0=$ Disables the INT1 external interrupt
bit 2 Unimplemented: Read as ' 0 '
bit 1 INT2IF: INT2 External Interrupt Flag bit
$1=$ The INT2 external interrupt occurred (must be cleared in software)
$0=$ The INT2 external interrupt did not occur
bit $0 \quad$ INT1IF: INT1 External Interrupt Flag bit
1 = The INT1 external interrupt occurred (must be cleared in software)
$0=$ The INT1 external interrupt did not occur

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0 '=$ Bit is cleared $\quad x=$ Bit is unknown |

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global interrupt enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

### 8.2 PIR Registers

The PIR registers contain the individual flag bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Request (Flag) registers (PIR1, PIR2, PIR3).

Note 1: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit, GIE (INTCON<7>).
2: User software should ensure the appropriate interrupt flag bits are cleared prior to enabling an interrupt and after servicing that interrupt.

REGISTER 8-4: PIR1: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 1

| R/W-0 | R/W-0 | R-0 | R-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad$ PSPIF: Parallel Slave Port Read/Write Interrupt Flag bit ${ }^{(1)}$
$1=$ A read or a write operation has taken place (must be cleared in software)
$0=$ No read or write has occurred
Note: This bit is not implemented on 28-pin devices and should be read as ' 0 '.
bit 6 ADIF: A/D Converter Interrupt Flag bit
1 = An A/D conversion completed (must be cleared in software)
$0=$ The A/D conversion is not complete
bit 5 RCIF: EUSART Receive Interrupt Flag bit
1 = The EUSART receive buffer, RCREG, is full (cleared when RCREG is read)
$0=$ The EUSART receive buffer is empty
bit 4 TXIF: EUSART Transmit Interrupt Flag bit
$1=$ The EUSART transmit buffer, TXREG, is empty (cleared when TXREG is written)
$0=$ The EUSART transmit buffer is full
bit 3 SSP1IF: Master Synchronous Serial Port 1 Interrupt Flag bit
$1=$ The transmission/reception is complete (must be cleared in software)
$0=$ Waiting to transmit/receive
bit 2 CCP1IF: ECCP1/CCP1 Interrupt Flag bit
Capture mode:
$1=\mathrm{A}$ TMR1 register capture occurred (must be cleared in software)
$0=$ No TMR1 register capture occurred
Compare mode:
1 = A TMR1 register compare match occurred (must be cleared in software)
$0=$ No TMR1 register compare match occurred
PWM mode:
Unused in this mode.
bit 1 TMR2IF: TMR2 to PR2 Match Interrupt Flag bit
$1=$ TMR2 to PR2 match occurred (must be cleared in software)
$0=$ No TMR2 to PR2 match occurred
bit $0 \quad$ TMR1IF: TMR1 Overflow Interrupt Flag bit
$1=$ TMR1 register overflowed (must be cleared in software)
$0=$ TMR1 register did not overflow

Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 8-5: PIR2: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 2

| R/W-0 | R/W-0 | U-0 | U-0 | R/W-0 | U-0 | U-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF |
| bit 7 |  |  |  |  |  |  |  |

bit 7 OSCFIF: Oscillator Fail Interrupt Flag bit
1 = Device oscillator failed, clock input has changed to INTOSC (must be cleared in software)
$0=$ Device clock operating
bit 6 CMIF: Comparator Interrupt Flag bit
$1=$ Comparator input has changed (must be cleared in software)
$0=$ Comparator input has not changed
bit 5-4 Unimplemented: Read as ' 0 '
bit 3 BCL1IF: Bus Collision Interrupt Flag bit (MSSP1 module)
$1=\mathrm{A}$ bus collision occurred (must be cleared in software)
$0=$ No bus collision occurred
bit 2-1 Unimplemented: Read as ' 0 '
bit $0 \quad$ CCP2IF: CCP2 Interrupt Flag bit
Capture mode:
$1=\mathrm{A}$ TMR1 register capture occurred (must be cleared in software)
$0=$ No TMR1 register capture occurred
Compare mode:
$1=$ A TMR1 register compare match occurred (must be cleared in software)
$0=$ No TMR1 register compare match occurred
PWM mode:
Unused in this mode.

## Legend:

| $R=$ Readable bit | W = Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1$ ' $=$ Bit is set | ' 0 ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 8-6: PIR3: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 3

| R/W-0 | R/W-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SSP2IF | BCL2IF | - | - | - | - | - | - |
| bit 7 |  |  |  |  |  |  |  |

bit 7 SSP2IF: Master Synchronous Serial Port 2 Interrupt Flag bit
$1=$ The transmission/reception is complete (must be cleared in software)
$0=$ Waiting to transmit/receive
bit 6 BCL2IF: Bus Collision Interrupt Flag bit (MSSP2 module)
$1=\mathrm{A}$ bus collision occurred (must be cleared in software)
$0=$ No bus collision occurred
bit 5-0 Unimplemented: Read as ' 0 '

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

## PIC18F45J10 FAMILY

### 8.3 PIE Registers

The PIE registers contain the individual enable bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Enable registers (PIE1, PIE2, PIE3). When IPEN $=0$, the PEIE bit must be set to enable any of these peripheral interrupts.

## REGISTER 8-7: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |  | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE |

bit 7
bit 0
bit $7 \quad$ PSPIE: Parallel Slave Port Read/Write Interrupt Enable bit ${ }^{(1)}$
1 = Enables the PSP read/write interrupt
$0=$ Disables the PSP read/write interrupt
Note: This bit is not implemented on 28-pin devices and should be read as ' 0 '.
bit 6 ADIE: A/D Converter Interrupt Enable bit
1 = Enables the A/D interrupt
$0=$ Disables the A/D interrupt
bit 5 RCIE: EUSART Receive Interrupt Enable bit
1 = Enables the EUSART receive interrupt
0 = Disables the EUSART receive interrupt
bit 4 TXIE: EUSART Transmit Interrupt Enable bit
1 = Enables the EUSART transmit interrupt
0 = Disables the EUSART transmit interrupt
bit 3 SSP1IE: Master Synchronous Serial Port 1 Interrupt Enable bit
1 = Enables the MSSP1 interrupt
$0=$ Disables the MSSP1 interrupt
bit 2 CCP1IE: ECCP1/CCP1 Interrupt Enable bit
1 = Enables the ECCP1/CCP1 interrupt
$0=$ Disables the ECCP1/CCP1 interrupt
bit 1 TMR2IE: TMR2 to PR2 Match Interrupt Enable bit
1 = Enables the TMR2 to PR2 match interrupt
$0=$ Disables the TMR2 to PR2 match interrupt
bit 0 TMR1IE: TMR1 Overflow Interrupt Enable bit
1 = Enables the TMR1 overflow interrupt
$0=$ Disables the TMR1 overflow interrupt

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0 '=$ Bit is cleared $\quad x=$ Bit is unknown |

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REGISTER 8-8: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

| R/W-0 | R/W-0 | U-0 | U-0 | R/W-0 | U-0 | U-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE |
| bit 7 |  |  |  |  |  |  |  |

bit 7 OSCFIE: Oscillator Fail Interrupt Enable bit
1 = Enabled
$0=$ Disabled
bit 6 CMIE: Comparator Interrupt Enable bit
1 = Enabled
$0=$ Disabled
bit 5-4 Unimplemented: Read as ' 0 '
bit 3 BCL1IE: Bus Collision Interrupt Enable bit (MSSP1 module)
1 = Enabled
$0=$ Disabled
bit 2-1 Unimplemented: Read as ' 0 '
bit $0 \quad$ CCP2IE: CCP2 Interrupt Enable bit
1 = Enabled
$0=$ Disabled

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1$ ' $=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 8-9: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

| R/W-0 | R/W-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SSP2IE | BCL2IE | - | - | - | - | - | - |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad$ SSP2IE: Master Synchronous Serial Port 2 Interrupt Enable bit
1 = Enabled
$0=$ Disabled
bit 6 BCL2IE: Bus Collision Interrupt Enable bit (MSSP2 module)
1 = Enabled
$0=$ Disabled
bit 5-0 Unimplemented: Read as ' 0 '

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1$ ' $=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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### 8.4 IPR Registers

The IPR registers contain the individual priority bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are three Peripheral Interrupt Priority registers (IPR1, IPR2, IPR3). Using the priority bits requires that the Interrupt Priority Enable (IPEN) bit be set.

## REGISTER 8-10: IPR1: PERIPHERAL INTERRUPT PRIORITY REGISTER 1

| R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP |

bit 7
bit 0
bit $7 \quad$ PSPIP: Parallel Slave Port Read/Write Interrupt Priority bit ${ }^{(1)}$
$1=$ High priority
0 = Low priority
Note: This bit is not implemented on 28-pin devices and should be read as ' 0 '.
bit 6 ADIP: A/D Converter Interrupt Priority bit
$1=$ High priority
0 = Low priority
bit 5 RCIP: EUSART Receive Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit 4 TXIP: EUSART Transmit Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit 3 SSP1IP: Master Synchronous Serial Port 1 Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit 2 CCP1IP: ECCP1/CCP1 Interrupt Priority bit
$1=$ High priority
0 = Low priority
bit 1 TMR2IP: TMR2 to PR2 Match Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit $0 \quad$ TMR1IP: TMR1 Overflow Interrupt Priority bit
$1=$ High priority
$0=$ Low priority

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 8-11: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

| R/W-1 | R/W1 | U-0 | U-0 | R/W-1 | U-0 | U-0 | R/W-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad$ OSCFIP: Oscillator Fail Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit $6 \quad$ CMIP: Comparator Interrupt Priority bit
$1=$ High priority
$0=$ Low priority
bit 5-4 Unimplemented: Read as ' 0 '
bit 3 BCL1IP: Bus Collision Interrupt Priority bit (MSSP1 module)
1 = High priority
0 = Low priority
bit 2-1 Unimplemented: Read as ' 0 '
bit $0 \quad$ CCP2IP: CCP2 Interrupt Priority bit
$1=$ High priority
$0=$ Low priority

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 8-12: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

| R/W-1 | R/W-1 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SSP2IP | BCL2IP | - | - | - | - | - | - |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad$ SSP2IP: Master Synchronous Serial Port 2 Interrupt Priority bit
1 = High priority
$0=$ Low priority
bit 6 BCL2IP: Bus Collision Interrupt Priority bit (MSSP2 module)
$1=$ High priority
$0=$ Low priority
bit 5-0 Unimplemented: Read as ' 0 '

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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### 8.5 RCON Register

The RCON register contains bits used to determine the cause of the last Reset or wake-up from Idle or Sleep modes. RCON also contains the bit that enables interrupt priorities (IPEN).

REGISTER 8-13: RCON: RESET CONTROL REGISTER

| R/W-0 | U-0 | U-0 | R/W-1 | R-1 | R-1 | R/W-0 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IPEN | - | - | $\overline{\mathrm{RI}}$ | $\overline{\mathrm{TO}}$ | $\overline{\mathrm{PD}}$ | $\overline{\mathrm{POR}}$ | $\overline{\mathrm{BOR}}$ |
| bit 7 |  |  |  |  |  |  |  |

bit 7 IPEN: Interrupt Priority Enable bit
1 = Enable priority levels on interrupts
$0=$ Disable priority levels on interrupts (PIC16CXXX Compatibility mode)
bit 6-5 Unimplemented: Read as ' 0 '
bit $4 \quad$ RI: RESET Instruction Flag bit
For details of bit operation, see Register 4-1.
bit $3 \quad \overline{\mathbf{T O}}$ : Watchdog Timer Time-out Flag bit
For details of bit operation, see Register 4-1.
bit $2 \quad \overline{\mathbf{P D}}$ : Power-Down Detection Flag bit
For details of bit operation, see Register 4-1.
bit $1 \quad \overline{\text { POR: }}$ : Power-on Reset Status bit
For details of bit operation, see Register 4-1.
bit $0 \quad \overline{\text { BOR: Brown-out Reset Status bit }}$
For details of bit operation, see Register 4-1.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 8.6 INTn Pin Interrupts

External interrupts on the RB0/INT0, RB1/INT1 and RB2/INT2 pins are edge-triggered. If the corresponding INTEDGx bit in the INTCON2 register is set (= 1), the interrupt is triggered by a rising edge; if the bit is clear, the trigger is on the falling edge. When a valid edge appears on the RBx/INTx pin, the corresponding flag bit, INTxIF, is set. This interrupt can be disabled by clearing the corresponding enable bit, INTxIE. Flag bit, INTxIF, must be cleared in software in the Interrupt Service Routine before re-enabling the interrupt.
All external interrupts (INT0, INT1 and INT2) can wake-up the processor from the power-managed modes if bit INTxIE was set prior to going into the power-managed modes. If the Global Interrupt Enable bit, GIE, is set, the processor will branch to the interrupt vector following wake-up.
Interrupt priority for INT1 and INT2 is determined by the value contained in the interrupt priority bits, INT1IP (INTCON3<6>) and INT2IP (INTCON3<7>). There is no priority bit associated with INTO. It is always a high priority interrupt source.

### 8.7 TMRO Interrupt

In 8-bit mode (which is the default), an overflow in the TMR0 register (FFh $\rightarrow 00 \mathrm{~h}$ ) will set flag bit, TMROIF. In 16-bit mode, an overflow in the TMROH:TMROL register pair (FFFFh $\rightarrow 0000 \mathrm{~h}$ ) will set TMROIF. The interrupt can be enabled/disabled by setting/clearing enable bit, TMROIE (INTCON $<5>$ ). Interrupt priority for TimerO is determined by the value contained in the interrupt priority bit, TMROIP (INTCON2<2>). See Section 10.0 "Timer0 Module" for further details on the Timer0 module.

### 8.8 PORTB Interrupt-on-Change

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON $<3>$ ). Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

### 8.9 Context Saving During Interrupts

During interrupts, the return PC address is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the fast return stack. If a fast return from interrupt is not used (see Section 5.3 "Data Memory Organization"), the user may need to save the WREG, STATUS and BSR registers on entry to the Interrupt Service Routine. Depending on the user's application, other registers may also need to be saved. Example 8-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

## EXAMPLE 8-1: SAVING STATUS, WREG AND BSR REGISTERS IN RAM

```
MOVWF W_TEMP ; W_TEMP is in virtual bank
MOVFF STATUS, STATUS_TEMP ; STATUS_TEMP located anywhere
MOVFF BSR, BSR_TEMP ; BSR_TMEP located anywhere
; USER ISR CODE
MOVFF BSR_TEMP, BSR ; Restore BSR
MOVF W_TEMP, W ; Restore WREG
MOVFF STATUS_TEMP, STATUS ; Restore STATUS
```


## PIC18F45J10 FAMILY

NOTES:

### 9.0 I/O PORTS

Depending on the device selected and features enabled, there are up to five ports available. Some pins of the I/O ports are multiplexed with an alternate function from the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.
Each port has three registers for its operation. These registers are:

- TRIS register (data direction register)
- Port register (reads the levels on the pins of the device)
- LAT register (output latch)

The Data Latch (LAT register) is useful for read-modifywrite operations on the value that the I/O pins are driving.
A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 9-1.

FIGURE 9-1: GENERIC I/O PORT OPERATION


Note 1: I/O pins have diode protection to VDD and Vss.

### 9.1 I/O Port Pin Capabilities

When developing an application, the capabilities of the port pins must be considered. Outputs on some pins have higher output drive strength than others. Similarly, some pins can tolerate higher than VDD input levels.

### 9.1.1 PIN OUTPUT DRIVE

The output pin drive strengths vary for groups of pins intended to meet the needs for a variety of applications. PORTB and PORTC are designed to drive higher loads, such as LEDs. All other ports are designed for small loads, typically indication only. Table 9-1 summarizes the output capabilities. Refer to Section 23.0 "Electrical Characteristics" for more details.

## TABLE 9-1: OUTPUT DRIVE LEVELS

| Port | Drive | Description |
| :--- | :--- | :--- |
| PORTA |  |  |
| PORTD | Minimum | Intended for indication. |
| PORTE |  | High | | Suitable for direct LED drive |
| :--- |
| levels. |

### 9.1.2 INPUT PINS AND VOLTAGE CONSIDERATIONS

The voltage tolerance of pins used as device inputs is dependent on the pin's input function. Pins that are used as digital only inputs are able to handle DC voltages up to 5.5 V ; a level typical for digital logic circuits. In contrast, pins that also have analog input functions of any kind can only tolerate voltages up to VdD. Voltage excursions beyond VDD on these pins should be avoided. Table 9-2 summarizes the input capabilities. Refer to Section 23.0 "Electrical Characteristics" for more details.

TABLE 9-2: INPUT VOLTAGE LEVELS

| Port or Pin | Tolerated Input | Description |
| :---: | :---: | :---: |
| PORTA<5:0> | VDD | Only Vdd input levels tolerated. |
| PORTB<5:0> |  |  |
| PORTC<1:0> |  |  |
| PORTE<2:0> |  |  |
| PORTB<7:6> | 5.5V | Tolerates input levels above VDD, useful for most standard logic. |
| PORTC<7:2> |  |  |
| PORTD<7:0> |  |  |

### 9.1.3 INTERFACING TO A 5V SYSTEM

Though the Vddmax of the PIC18F45J10 family is 3.6 V , these devices are still capable of interfacing with 5 V systems, even if the VIH of the target system is above 3.6 V . This is accomplished by adding a pull-up resistor to the port pin (Figure 9-2), clearing the LAT bit for that pin and manipulating the corresponding TRIS bit (Figure 9-1) to either allow the line to be pulled high or to drive the pin low. Only port pins that are tolerant of voltages up to 5.5 V can be used for this type of interface (refer to Section 9.1.2 "Input Pins and Voltage Considerations").

FIGURE 9-2: +5V SYSTEM HARDWARE INTERFACE


EXAMPLE 9-1: COMMUNICATING WITH THE +5 V SYSTEM

```
BCF LATD, 7 ; set up LAT register so
    ; changing TRIS bit will
    ; drive line low
BCF TRISD, 7 ; send a 0 to the 5V system
BCF TRISD, 7 ; send a l to the 5V system
```


### 9.2 PORTA, TRISA and LATA Registers

PORTA is a 5 -bit wide, bidirectional port. The corresponding data direction register is TRISA. Setting a TRISA bit (=1) will make the corresponding PORTA pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISA bit $(=0)$ will make the corresponding PORTA pin an output (i.e., put the contents of the output latch on the selected pin).
Reading the PORTA register reads the status of the pins, whereas writing to it, will write to the port latch.
The Data Latch (LATA) register is also memory mapped. Read-modify-write operations on the LATA register read and write the latched output value for PORTA.
The other PORTA pins are multiplexed with analog inputs, the analog Vref+ and Vref- inputs and the comparator voltage reference output. The operation of pins RA3:RA0 and RA5 as A/D converter inputs is selected by clearing or setting the control bits in the ADCON1 register (A/D Control Register 1).
Pins RAO and RA3 may also be used as comparator inputs and RA5 may be used as the C2 comparator output by setting the appropriate bits in the CMCON register. To use RA3:RAO as digital inputs, it is also necessary to turn off the comparators.

Note: On a Power-on Reset, RA5 and RA3:RA0 are configured as analog inputs and read as ' 0 '.
All PORTA pins have TTL input levels and full CMOS output drivers.
The TRISA register controls the direction of the PORTA pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

EXAMPLE 9-2: INITIALIZING PORTA

| CLRF | PORTA | ; Initialize PORTA by |
| :--- | :--- | :--- |
|  |  | ; clearing output |
| CLRF | LATA | ; Altata latches |
|  |  | ; to clear output |
| MOVLW | $07 h$ | ; Configure A/D |
| MOVWF | ADCON1 | ; for digital inputs |
| MOVWF | $07 h$ | ; Configure comparators |
| MOVWF | CMCON | ; for digital input |
| MOVLW | OCFh | ; Value used to |
|  |  | ; initialize data |
|  |  | direction |
| MOVWF | TRISA | Set RA<3:0> as inputs |
|  |  | RA<5:4> as outputs |

TABLE 9-3: PORTA I/O SUMMARY

| Pin | Function | TRIS Setting | I/O | $\begin{aligned} & \text { I/O } \\ & \text { Type } \end{aligned}$ | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RAO/ANO | RAO | 0 | 0 | DIG | LATA <0> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTA<0> data input; disabled when analog input enabled. |
|  | ANO | 1 | 1 | ANA | A/D input channel 0 and Comparator C1- input. Default input configuration on POR; does not affect digital output. |
| RA1/AN1 | RA1 | 0 | 0 | DIG | LATA<1> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTA<1> data input; disabled when analog input enabled. |
|  | AN1 | 1 | 1 | ANA | A/D input channel 1 and Comparator C2- input. Default input configuration on POR; does not affect digital output. |
| RA2/AN2/ Vref-/CVref | RA2 | 0 | 0 | DIG | LATA<2> data output; not affected by analog input. Disabled when CVref output enabled. |
|  |  | 1 | 1 | TTL | PORTA<2> data input. Disabled when analog functions enabled; disabled when CVREF output enabled. |
|  | AN2 | 1 | 1 | ANA | A/D input channel 2 and Comparator C2+ input. Default input configuration on POR; not affected by analog output. |
|  | Vref- | 1 | 1 | ANA | A/D and comparator voltage reference low input. |
|  | CVref | x | 0 | ANA | Comparator voltage reference output. Enabling this feature disables digital I/O. |
| RA3/AN3/VREF+ | RA3 | 0 | 0 | DIG | LATA <3> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTA<3> data input; disabled when analog input enabled. |
|  | AN3 | 1 | 1 | ANA | A/D input channel 3 and Comparator $\mathrm{C} 1+$ input. Default input configuration on POR. |
|  | VREF+ | 1 | 1 | ANA | A/D and comparator voltage reference high input. |
| $\begin{aligned} & \hline \text { RA5/AN4/ } \overline{\text { SS1/ } / ~} \\ & \text { C2OUT } \end{aligned}$ | RA5 | 0 | 0 | DIG | LATA < 5 > data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTA<5> data input; disabled when analog input enabled. |
|  | AN4 | 1 | 1 | ANA | A/D input channel 4. Default configuration on POR. |
|  | SS1 | 1 | 1 | TTL | Slave select input for MSSP1 (MSSP1 module). |
|  | C2OUT | 0 | 0 | DIG | Comparator 2 output; takes priority over port data. |
| OSC2/CLKO | OSC2 | x | 0 | ANA | Main oscillator feedback output connection (HS mode). |
|  | CLKO | x | 0 | DIG | System cycle clock output (FOSC/4) in RC and EC Oscillator modes. |
| OSC1/CLKI | OSC1 | x | 1 | ANA | Main oscillator input connection. |
|  | CLKI | x | 1 | ANA | Main clock input connection. |

Legend: DIG = Digital level output; TTL = TTL input buffer; ST = Schmitt Trigger input buffer; ANA = Analog level input/output; $\mathrm{x}=$ Don't care (TRIS bit does not affect port direction or is overridden for this option).

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TABLE 9-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PORTA | - | - | RA5 | - | RA3 | RA2 | RA1 | RA0 | 46 |
| LATA | - | - | PORTA Data Latch Register (Read and Write to Data Latch) | 46 |  |  |  |  |  |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISA0 | 46 |
| ADCON1 | - | - | VCFG1 | VCFG0 | PCFG3 | PCFG2 | PCFG1 | PCFG0 | 44 |
| CMCON | C2OUT | C1OUT | C2INV | C1INV | CIS | CM2 | CM1 | CM0 | 45 |
| CVRCON | CVREN | CVROE | CVRR | CVRSS | CVR3 | CVR2 | CVR1 | CVR0 | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by PORTA.

### 9.3 PORTB, TRISB and LATB Registers

PORTB is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISB. Setting a TRISB bit (= 1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., put the contents of the output latch on the selected pin).
The Data Latch register (LATB) is also memory mapped. Read-modify-write operations on the LATB register read and write the latched output value for PORTB.

## EXAMPLE 9-3: INITIALIZING PORTB

| CLRF | PORTB | ; Initialize PORTB by <br> ; clearing output <br> ; data latches |
| :---: | :---: | :---: |
| CLRF | LATB | ; Alternate method <br> ; to clear output <br> ; data latches |
| MOVLW | OFh | ; Set RB<4:0> as |
| MOVWF | ADCON1 | ; digital I/O pins <br> ; (required if config bit <br> ; PBADEN is set) |
| MOVLW | OCFh | ; Value used to <br> ; initialize data <br> ; direction |
| MOVWF | TRISB | ; Set $\mathrm{RB}<3: 0>$ as inputs <br> ; RB<5:4> as outputs <br> ; RB<7:6> as inputs |

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit, $\overline{\mathrm{RBPU}}$ (INTCON2<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Note: On a Power-on Reset, RB4:RB0 are configured as analog inputs by default and read as ' 0 '; RB7:RB5 are configured as digital inputs.
By programming the configuration bit, PBADEN, RB4:RB0 will alternatively be configured as digital inputs on POR.

Four of the PORTB pins (RB7:RB4) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupt-on-change comparison). The input pins (of RB7:RB4) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB7:RB4 are ORed together to generate the RB Port Change Interrupt with Flag bit, RBIF (INTCON<0>).
This interrupt can wake the device from Sleep mode or any of the Idle modes. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:
a) Any read or write of PORTB (except with the MOVFF (ANY), PORTB instruction).
b) Clear flag bit, RBIF.

A mismatch condition will continue to set flag bit, RBIF. Reading PORTB will end the mismatch condition and allow flag bit, RBIF, to be cleared.
The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.
RB3 can be configured by the configuration bit, CCP2MX, as the alternate peripheral pin for the CCP2 module (CCP2MX = 0).
The RB5 pin is multiplexed with the Timer0 module clock input and one of the comparator outputs to become the RB5/KBI1/T0CKI/C1OUT pin.

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## TABLE 9-5: PORTB I/O SUMMARY

| Pin | Function | TRIS Setting | I/O | $\begin{aligned} & \text { I/O } \\ & \text { Type } \end{aligned}$ | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RB0/INT0/FLT0/AN12 | RBO | 0 | 0 | DIG | LATB<0> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTB<0> data input; weak pull-up when $\overline{\text { RBPU }}$ bit is cleared. Disabled when analog input enabled. ${ }^{(1)}$ |
|  | INTO | 1 | 1 | ST | External interrupt 0 input. |
|  | FLTO | 1 | 1 | ST | PWM Fault input (ECCP1/CCP1 module); enabled in software. |
|  | AN12 | 1 | 1 | ANA | A/D input channel 12. ${ }^{(1)}$ |
| RB1/INT1/AN10 | RB1 | 0 | 0 | DIG | LATB<1> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTB<1> data input; weak pull-up when $\overline{\text { RBPU }}$ bit is cleared. Disabled when analog input enabled. ${ }^{(1)}$ |
|  | INT1 | 1 | 1 | ST | External interrupt 1 input. |
|  | AN10 | 1 | 1 | ANA | A/D input channel 10. ${ }^{(1)}$ |
| RB2/INT2/AN8 | RB2 | 0 | 0 | DIG | LATB<2> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTB<2> data input; weak pull-up when $\overline{\text { RBPU }}$ bit is cleared. Disabled when analog input enabled. ${ }^{(1)}$ |
|  | INT2 | 1 | 1 | ST | External interrupt 2 input. |
|  | AN8 | 1 | 1 | ANA | A/D input channel 8. ${ }^{(1)}$ |
| RB3/AN9/CCP2 | RB3 | 0 | 0 | DIG | LATB<3> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTB<3> data input; weak pull-up when $\overline{\text { RBPU }}$ bit is cleared. Disabled when analog input enabled. ${ }^{(1)}$ |
|  | AN9 | 1 | 1 | ANA | A/D input channel 9. ${ }^{(1)}$ |
|  | $\mathrm{CCP2}^{(2)}$ | 0 | 0 | DIG | CCP2 compare and PWM output. |
|  |  | 1 | 1 | ST | CCP2 capture input |
| RB4/KBIO/AN11 | RB4 | 0 | 0 | DIG | LATB<4> data output; not affected by analog input. |
|  |  | 1 | 1 | TTL | PORTB<4> data input; weak pull-up when $\overline{\text { RBPU }}$ bit is cleared. Disabled when analog input enabled. ${ }^{(1)}$ |
|  | KBIO | 1 | 1 | TTL | Interrupt-on-change pin. |
|  | AN11 | 1 | 1 | ANA | A/D input channel 11. ${ }^{(1)}$ |
| $\begin{aligned} & \hline \text { RB5/KBI1/TOCKI/ } \\ & \text { C1OUT } \end{aligned}$ | RB5 | 0 | 0 | DIG | LATB<5> data output. |
|  |  | 1 | 1 | TTL | PORTB<5> data input; weak pull-up when $\overline{\mathrm{RBPU}}$ bit is cleared. |
|  | KBI1 | 1 | 1 | TTL | Interrupt-on-change pin. |
|  | TOCKI | 1 | 1 | ST | Timer0 clock input. |
|  | C10UT | 0 | 0 | DIG | Comparator 1 output; takes priority over port data. |
| RB6/KBI2/PGC | RB6 | 0 | 0 | DIG | LATB<6> data output. |
|  |  | 1 | 1 | TTL | PORTB<6> data input; weak pull-up when $\overline{\mathrm{RBPU}}$ bit is cleared. |
|  | KBI2 | 1 | 1 | TTL | Interrupt-on-change pin. |
|  | PGC | x | 1 | ST | Serial execution (ICSPP ${ }^{\text {TM }}$ ) clock input for ICSP and ICD operation. ${ }^{(3)}$ |
| RB7/KBI3/PGD | RB7 | 0 | 0 | DIG | LATB<7> data output. |
|  |  | 1 | 1 | TTL | PORTB<7> data input; weak pull-up when $\overline{\mathrm{RBPU}}$ bit is cleared. |
|  | KBI3 | 1 | 1 | TTL | Interrupt-on-change pin. |
|  | PGD | x | 0 | DIG | Serial execution data output for ICSP and ICD operation. ${ }^{(3)}$ |
|  |  | x | 1 | ST | Serial execution data input for ICSP and ICD operation. ${ }^{(3)}$ |

Legend: DIG = Digital level output; TTL = TTL input buffer; ST = Schmitt Trigger input buffer; ANA = Analog level input/output; $\mathrm{x}=$ Don't care (TRIS bit does not affect port direction or is overridden for this option).
Note 1: Configuration on POR is determined by the PBADEN configuration bit. Pins are configured as analog inputs by default when PBADEN is set and digital inputs when PBADEN is cleared.
2: Alternate assignment for CCP2 when the CCP2MX configuration bit is ' 0 '. Default assignment is RC1.
3: All other pin functions are disabled when ICSP or ICD are enabled.

## PIC18F45J10 FAMILY

TABLE 9-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PORTB | RB7 |  | RB6 | RB5 | RB4 | RB3 | RB2 | RB1 | RB0 |
| 46 |  |  |  |  |  |  |  |  |  |
| LATB | PORTB Data Latch Register (Read and Write to Data Latch) |  |  | 46 |  |  |  |  |  |
| TRISB | PORTB Data Direction Control Register |  |  |  | 46 |  |  |  |  |
| INTCON | GIE/GIEH | PEIE/GIEL | TMR0IE | INT0IE | RBIE | TMR0IF | INTOIF | RBIF | 43 |
| INTCON2 | $\overline{R B P U}$ | INTEDG0 | INTEDG1 | INTEDG2 | - | TMR0IP | - | RBIP | 43 |
| INTCON3 | INT2IP | INT1IP | - | INT2IE | INT1IE | - | INT2IF | INT1IF | 43 |
| ADCON1 | - | - | VCFG1 | VCFG0 | PCFG3 | PCFG2 | PCFG1 | PCFG0 | 44 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by PORTB.

## PIC18F45J10 FAMILY

### 9.4 PORTC, TRISC and LATC Registers

PORTC is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISC. Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., put the contents of the output latch on the selected pin).
The Data Latch register (LATC) is also memory mapped. Read-modify-write operations on the LATC register read and write the latched output value for PORTC.

PORTC is multiplexed with several peripheral functions (Table 9-7). The pins have Schmitt Trigger input buffers. RC1 is normally configured by configuration bit, CCP2MX, as the default peripheral pin of the CCP2 module (default/erased state, CCP2MX = 1).
When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for additional information.

## Note: On a Power-on Reset, these pins are configured as digital inputs.

The contents of the TRISC register are affected by peripheral overrides. Reading TRISC always returns the current contents, even though a peripheral device may be overriding one or more of the pins.

## EXAMPLE 9-4: INITIALIZING PORTC

| CLRF | PORTC | ; Initialize PORTC by <br> ; clearing output |
| :---: | :---: | :---: |
|  |  | ; data latches |
| CLRF | LATC | ; Alternate method |
|  |  | ; to clear output |
|  |  | ; data latches |
| MOVLW | OCFh | ; Value used to |
|  |  | ; initialize data |
|  |  | ; direction |
| MOVWF | TRISC | ; Set RC<3:0> as inputs |
|  |  | ; RC<5:4> as outputs |
|  |  | ; RC<7:6> as inputs |

TABLE 9-7: PORTC I/O SUMMARY

| Pin | Function | TRIS Setting | I/O | $\begin{aligned} & \text { I/O } \\ & \text { Type } \end{aligned}$ | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { RC0/T1OSO/ } \\ & \text { T1CKI } \end{aligned}$ | RC0 | 0 | 0 | DIG | LATC<0> data output. |
|  |  | 1 | 1 | ST | PORTC<0> data input. |
|  | T10SO | x | 0 | ANA | Timer1 oscillator output; enabled when Timer1 oscillator enabled. Disables digital I/O. |
|  | T1CKI | 1 | 1 | ST | Timer1 counter input. |
| RC1/T1OSI/CCP2 | RC1 | 0 | 0 | DIG | LATC<1> data output. |
|  |  | 1 | 1 | ST | PORTC<1> data input. |
|  | T10SI | x | 1 | ANA | Timer1 oscillator input; enabled when Timer1 oscillator enabled. Disables digital I/O. |
|  | $\mathrm{CCP} 2^{(1)}$ | 0 | 0 | DIG | CCP2 compare and PWM output; takes priority over port data. |
|  |  | 1 | 1 | ST | CCP2 capture input. |
| RC2/CCP1/P1A | RC2 | 0 | 0 | DIG | LATC<2> data output. |
|  |  | 1 | 1 | ST | PORTC<2> data input. |
|  | CCP1 | 0 | 0 | DIG | ECCP1/CCP1 compare or PWM output; takes priority over port data. |
|  |  | 1 | 1 | ST | ECCP1/CCP1 capture input. |
|  | $\mathrm{P} 1 \mathrm{~A}^{(2)}$ | 0 | 0 | DIG | ECCP1 Enhanced PWM output, channel A. May be configured for tri-state during Enhanced PWM shutdown events. Takes priority over port data. |
| RC3/SCK1/SCL1 | RC3 | 0 | 0 | DIG | LATC<3> data output. |
|  |  | 1 | 1 | ST | PORTC<3> data input. |
|  | SCK1 | 0 | 0 | DIG | SPITM clock output (MSSP1 module); takes priority over port data. |
|  |  | 1 | 1 | ST | SPI clock input (MSSP1 module). |
|  | SCL1 | 0 | 0 | DIG | $1^{2} \mathrm{C}^{\text {TM }}$ clock output (MSSP1 module); takes priority over port data. |
|  |  | 1 | 1 | $\mathrm{I}^{2} \mathrm{C} / \mathrm{SMB}$ | $\mathrm{I}^{2} \mathrm{C}$ clock input (MSSP1 module); input type depends on module setting. |
| RC4/SDI1/SDA1 | RC4 | 0 | 0 | DIG | LATC<4> data output. |
|  |  | 1 | I | ST | PORTC<4> data input. |
|  | SDI1 | 1 | 1 | ST | SPI data input (MSSP1 module). |
|  | SDA1 | 1 | 0 | DIG | $I^{2} \mathrm{C}$ data output (MSSP1 module); takes priority over port data. |
|  |  | 1 | 1 | $\mathrm{I}^{2} \mathrm{C} / \mathrm{SMB}$ | $I^{2} \mathrm{C}$ data input (MSSP1 module); input type depends on module setting. |
| RC5/SDO1 | RC5 | 0 | 0 | DIG | LATC<5> data output. |
|  |  | 1 | 1 | ST | PORTC<5> data input. |
|  | SDO1 | 0 | 0 | DIG | SPI data output (MSSP1 module); takes priority over port data. |
| RC6/TX/CK | RC6 | 0 | 0 | DIG | LATC<6> data output. |
|  |  | 1 | 1 | ST | PORTC<6> data input. |
|  | TX | 1 | 0 | DIG | Asynchronous serial transmit data output (EUSART module); takes priority over port data. User must configure as output. |
|  | CK | 1 | 0 | DIG | Synchronous serial clock output (EUSART module); takes priority over port data. |
|  |  | 1 | 1 | ST | Synchronous serial clock input (EUSART module). |
| RC7/RX/DT | RC7 | 0 | 0 | DIG | LATC<7> data output. |
|  |  | 1 | 1 | ST | PORTC<7> data input. |
|  | RX | 1 | 1 | ST | Asynchronous serial receive data input (EUSART module). |
|  | DT | 1 | 0 | DIG | Synchronous serial data output (EUSART module); takes priority over port data. |
|  |  | 1 | 1 | ST | Synchronous serial data input (EUSART module). User must configure as an input. |

Legend: $\quad$ DIG = Digital level output; TTL = TTL input buffer; ST = Schmitt Trigger input buffer; ANA = Analog level input/output; $I^{2} \mathrm{C} / \mathrm{SMB}=I^{2} \mathrm{C} /$ SMBus input buffer; $\mathrm{x}=$ Don't care (TRIS bit does not affect port direction or is overridden for this option).
Note 1: Default assignment for CCP2 when the CCP2MX configuration bit is set. Alternate assignment is RB3.
2: Enhanced PWM output is available only on PIC18F44J10/45J10 devices.

## PIC18F45J10 FAMILY

TABLE 9-8: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PORTC | RC7 | RC6 | RC5 | RC4 | RC3 | RC2 | RC1 | RC0 | 46 |
| LATC | PORTC Data Latch Register (Read and Write to Data Latch) |  |  | 46 |  |  |  |  |  |
| TRISC | PORTC Data Direction Control Register |  |  |  |  |  |  |  |  |

### 9.5 PORTD, TRISD and LATD Registers

PORTD is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISD. Setting a TRISD bit (= 1) will make the corresponding PORTD pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISD bit (= 0) will make the corresponding PORTD pin an output (i.e., put the contents of the output latch on the selected pin).
The Data Latch register (LATD) is also memory mapped. Read-modify-write operations on the LATD register read and write the latched output value for PORTD.
All pins on PORTD are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.
Three of the PORTD pins are multiplexed with outputs P1B, P1C and P1D of the Enhanced CCP module. The operation of these additional PWM output pins is covered in greater detail in Section 14.0 "Enhanced Capture/Compare/PWM (ECCP) Module".

Note: On a Power-on Reset, these pins are configured as digital inputs.

PORTD can also be configured as an 8-bit wide microprocessor port (Parallel Slave Port) by setting control bit, PSPMODE (TRISE<4>). In this mode, the input buffers are TTL. See Section 9.7 "Parallel Slave Port" for additional information on the Parallel Slave Port (PSP).

Note: When the Enhanced PWM mode is used with either dual or quad outputs, the PSP functions of PORTD are automatically disabled.

## EXAMPLE 9-5: INITIALIZING PORTD

| CLRF | PORTD | $\begin{aligned} & \text {; Initialize PORTD by } \\ & \text {; clearing output } \\ & \text {; data latches } \end{aligned}$ |
| :---: | :---: | :---: |
| CLRF | LATD | ; Alternate method <br> ; to clear output <br> ; data latches |
| MOVLW | OCFh | ; Value used to <br> ; initialize data <br> ; direction |
| MOVWF | TRISD | ; Set $\mathrm{RD}<3: 0>$ as inputs <br> ; RD<5:4> as outputs <br> ; RD<7:6> as inputs |

## PIC18F45J10 FAMILY

TABLE 9-9: PORTD I/O SUMMARY

| Pin | Function | TRIS Setting | I/O | $\begin{aligned} & \text { l/O } \\ & \text { Type } \end{aligned}$ | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RD0/PSP0/SCK2/ SCL2 | RD0 | 0 | 0 | DIG | LATD<0> data output. |
|  |  | 1 | I | ST | PORTD<0> data input. |
|  | PSP0 | x | 0 | DIG | PSP read data output (LATD<0>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | SCK2 | 0 | 0 | DIG | SPI ${ }^{\text {TM }}$ clock output (MSSP2 module); takes priority over port data. |
|  |  | 1 | 1 | ST | SPI clock input (MSSP2 module). |
|  | SCL2 | 0 | 0 | DIG | $1^{2} \mathrm{C}^{\text {TM }}$ clock output (MSSP2 module); takes priority over port data. |
|  |  | 1 | 1 | $\mathrm{I}^{2} \mathrm{C} / \mathrm{SMB}$ | $\mathrm{I}^{2} \mathrm{C}$ clock input (MSSP2 module); input type depends on module setting. |
| RD1/PSP1/SDI2/ SDA2 | RD1 | 0 | 0 | DIG | LATD<1> data output. |
|  |  | 1 | 1 | ST | PORTD<1> data input. |
|  | PSP1 | x | 0 | DIG | PSP read data output (LATD<1>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | SDI2 | 1 | 1 | ST | SPI data input (MSSP2 module). |
|  | SDA2 | 1 | 0 | DIG | $1^{2} \mathrm{C}$ data output (MSSP2 module); takes priority over port data. |
|  |  | 1 | 1 | $\mathrm{I}^{2} \mathrm{C} / \mathrm{SMB}$ | $1^{2} \mathrm{C}$ data input (MSSP2 module); input type depends on module setting. |
| RD2/PSP2/SDO2 | RD2 | 0 | 0 | DIG | LATD<2> data output. |
|  |  | 1 | 1 | ST | PORTD<2> data input. |
|  | PSP2 | x | 0 | DIG | PSP read data output (LATD<2>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | SDO2 | 0 | 0 | DIG | SPI data output (MSSP2 module); takes priority over port data. |
| RD3/PSP3/SS2 | RD3 | 0 | 0 | DIG | LATD<3> data output. |
|  |  | 1 | 1 | ST | PORTD<3> data input. |
|  | PSP3 | x | O | DIG | PSP read data output (LATD<3>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | $\overline{\mathrm{SS} 2}$ | 1 | 1 | TTL | Slave select input for MSSP2 (MSSP2 module). |
| RD4/PSP4 | RD4 | 0 | 0 | DIG | LATD<4> data output. |
|  |  | 1 | 1 | ST | PORTD<4> data input. |
|  | PSP4 | x | 0 | DIG | PSP read data output (LATD<4>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
| RD5/PSP5/P1B | RD5 | 0 | 0 | DIG | LATD<5> data output. |
|  |  | 1 | 1 | ST | PORTD<5> data input. |
|  | PSP5 | x | 0 | DIG | PSP read data output (LATD<5>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | P1B | 0 | O | DIG | ECCP1 Enhanced PWM output, channel B; takes priority over port and PSP data. May be configured for tri-state during Enhanced PWM shutdown events. |
| RD6/PSP6/P1C | RD6 | 0 | 0 | DIG | LATD<6> data output. |
|  |  | 1 | 1 | ST | PORTD<6> data input. |
|  | PSP6 | x | O | DIG | PSP read data output (LATD<6>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | P1C | 0 | 0 | DIG | ECCP1 Enhanced PWM output, channel C; takes priority over port and PSP data. May be configured for tri-state during Enhanced PWM shutdown events. |
| RD7/PSP7/P1D | RD7 | 0 | 0 | DIG | LATD<7> data output. |
|  |  | 1 | 1 | ST | PORTD<7> data input. |
|  | PSP7 | x | O | DIG | PSP read data output (LATD<7>); takes priority over port data. |
|  |  | x | 1 | TTL | PSP write data input. |
|  | P1D | 0 | 0 | DIG | ECCP1 Enhanced PWM output, channel D; takes priority over port and PSP data. May be configured for tri-state during Enhanced PWM shutdown events. |
|  $\mathrm{x}=$ Don't care (TRIS bit does not affect port direction or is overridden for this option). |  |  |  |  |  |

TABLE 9-10: $\quad$ SUMMARY OF REGISTERS ASSOCIATED WITH PORTD

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PORTD $^{(\mathbf{1})}$ | RD7 | RD6 | RD5 | RD4 | RD3 | RD2 | RD1 | RD0 | 46 |
| LATD $^{(\mathbf{1})}$ | PORTD Data Latch Register (Read and Write to Data Latch) |  |  | 46 |  |  |  |  |  |
| TRISD $^{(\mathbf{1})}$ | PORTD Data Direction Control Register |  |  | 46 |  |  |  |  |  |
| TRISE $^{(\mathbf{1})}$ | IBF | OBF | IBOV | PSPMODE | - | TRISE2 | TRISE1 | TRISE0 | 46 |
| CCP1CON | P1M1 | P1M0 | DC1B1 | DC1B0 | CCP1M3 | CCP1M2 | CCP1M1 | CCP1M0 | 45 |

Legend: $-=$ unimplemented, read as ' 0 '. Shaded cells are not used by PORTD.
Note 1: These registers are not available in 28 -pin devices.

### 9.6 PORTE, TRISE and LATE Registers

## Note: PORTE is only available in 40/44-pin devices.

Depending on the particular PIC18F45J10 family device selected, PORTE is implemented in two different ways.
For 40/44-pin devices, PORTE is a 4-bit wide port. Three pins (RE0/RD/AN5, RE1/ $\overline{\mathrm{WR}} / \mathrm{AN6}$ and RE2/ $\overline{\mathrm{CS}} /$ AN7) are individually configurable as inputs or outputs. These pins have Schmitt Trigger input buffers. When selected as analog inputs, these pins will read as 'o's.
The corresponding data direction register is TRISE. Setting a TRISE bit (=1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISE bit (=0) will make the corresponding PORTE pin an output (i.e., put the contents of the output latch on the selected pin).
TRISE controls the direction of the RE pins, even when they are being used as analog inputs. The user must make sure to keep the pins configured as inputs when using them as analog inputs.
Note: On a Power-on Reset, RE2:REO are configured as analog inputs.

The upper four bits of the TRISE register also control the operation of the Parallel Slave Port. Their operation is explained in Register 9-1.

The Data Latch register (LATE) is also memory mapped. Read-modify-write operations on the LATE register read and write the latched output value for PORTE.

EXAMPLE 9-6: INITIALIZING PORTE

| CLRF | PORTE | ; Initialize PORTE by <br> ; clearing output <br> ; data latches |
| :---: | :---: | :---: |
| CLRF | LATE | ; Alternate method <br> ; to clear output <br> ; data latches |
| MOVLW | 0Ah | ; Configure A/D |
| MOVWF | ADCON1 | ; for digital inputs |
| MOVLW | 03h | ; Value used to <br> ; initialize data <br> ; direction |
| MOVWF | TRISE | ; Set RE<0> as inputs <br> ; RE<l> as outputs <br> ; $\mathrm{RE}<2>$ as inputs |

REGISTER 9-1: TRISE REGISTER (40/44-PIN DEVICES ONLY)

| R-0 | R-0 | R/W-0 | R/W-0 | U-0 | R/W-1 | R/W-1 | R/W-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IBF | OBF | IBOV | PSPMODE | - | TRISE2 | TRISE1 | TRISE0 |
| bit 7 bit |  |  |  |  |  |  |  |

bit 4 PSPMODE: Parallel Slave Port Mode Select bit
bit 2 TRISE2: RE2 Direction Control bit
bit 1 TRISE1: RE1 Direction Control bit
bit 7
bit 6
bit 5
bit 3
bit 0

IBF: Input Buffer Full Status bit
$1=A$ word has been received and waiting to be read by the CPU
$0=$ No word has been received
OBF: Output Buffer Full Status bit
1 = The output buffer still holds a previously written word $0=$ The output buffer has been read
IBOV: Input Buffer Overflow Detect bit (in Microprocessor mode)
1 = A write occurred when a previously input word has not been read (must be cleared in software)
0 = No overflow occurred

1 = Parallel Slave Port mode
$0=$ General purpose I/O mode
t 3 Unimplemented: Read as ' 0 '

1 = Input
$0=$ Output

$$
1=\text { Input }
$$

$0=$ Output
TRISEO: REO Direction Control bit
1 = Input
$0=$ Output

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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TABLE 9-11: PORTE I/O SUMMARY

| Pin | Function | TRIS Setting | I/O | $\begin{aligned} & \text { I/O } \\ & \text { Type } \end{aligned}$ | Description |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RE0/ $\overline{\mathrm{RD}} / \mathrm{AN} 5$ | RE0 | 0 | 0 | DIG | LATE<0> data output; not affected by analog input. |
|  |  | 1 | 1 | ST | PORTE<0> data input; disabled when analog input enabled. |
|  | $\overline{\mathrm{RD}}$ | 1 | 1 | TTL | PSP read enable input (PSP enabled). |
|  | AN5 | 1 | 1 | ANA | A/D input channel 5; default input configuration on POR. |
| RE1/VR/AN6 | RE1 | 0 | 0 | DIG | LATE<1> data output; not affected by analog input. |
|  |  | 1 | 1 | ST | PORTE<1> data input; disabled when analog input enabled. |
|  | $\overline{\mathrm{WR}}$ | 1 | 1 | TTL | PSP write enable input (PSP enabled). |
|  | AN6 | 1 | 1 | ANA | A/D input channel 6; default input configuration on POR. |
| RE2/CS/AN7 | RE2 | 0 | 0 | DIG | LATE<2> data output; not affected by analog input. |
|  |  | 1 | I | ST | PORTE<2> data input; disabled when analog input enabled. |
|  | $\overline{\mathrm{CS}}$ | 1 | 1 | TTL | PSP write enable input (PSP enabled). |
|  | AN7 | 1 | 1 | ANA | A/D input channel 7; default input configuration on POR. |

Legend: DIG = Digital level output; TTL = TTL input buffer; ST = Schmitt Trigger input buffer; ANA = Analog level input/output; $\mathrm{x}=$ Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 9-12: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PORTE $^{(\mathbf{1 )}}$ | - | - | - | - | - | RE2 | RE1 | RE0 | 46 |
| LATE $^{(\mathbf{1})}$ | - | - | - | - | - | PORTE Data Latch Register <br> (Read and Write to Data Latch) | 46 |  |  |
| TRISE $^{(\mathbf{1})}$ | IBF | OBF | IBOV | PSPMODE | - | TRISE2 | TRISE1 | TRISE0 | 46 |
| ADCON1 | - | - | VCFG1 | VCFG0 | PCFG3 | PCFG2 | PCFG1 | PCFG0 | 44 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by PORTE.
Note 1: These registers are not available in 28-pin devices.

### 9.7 Parallel Slave Port

Note: The Parallel Slave Port is only available in 40/44-pin devices.

In addition to its function as a general I/O port, PORTD can also operate as an 8-bit wide Parallel Slave Port (PSP) or microprocessor port. PSP operation is controlled by the 4 upper bits of the TRISE register (Register 9-1). Setting control bit, PSPMODE (TRISE<4>), enables PSP operation as long as the Enhanced CCP module is not operating in dual output or quad output PWM mode. In Slave mode, the port is asynchronously readable and writable by the external world.
The PSP can directly interface to an 8-bit microprocessor data bus. The external microprocessor can read or write the PORTD latch as an 8 -bit latch. Setting the control bit, PSPMODE, enables the PORTE I/O pins to become control inputs for the microprocessor port. When set, port pin RE0 is the $\overline{\mathrm{RD}}$ input, RE1 is the $\overline{\mathrm{WR}}$ input and RE2 is the $\overline{\mathrm{CS}}$ (Chip Select) input. For this functionality, the corresponding data direction bits of the TRISE register (TRISE<2:0>) must be configured as inputs (set). The A/D port configuration bits, PFCG3:PFCG0 (ADCON1<3:0>), must also be set to a value in the range of '1010' through '1111'.
A write to the PSP occurs when both the $\overline{\mathrm{CS}}$ and $\overline{W R}$ lines are first detected low and ends when either are detected high. The PSPIF and IBF flag bits are both set when the write ends.
A read from the PSP occurs when both the $\overline{C S}$ and $\overline{R D}$ lines are first detected low. The data in PORTD is read out and the OBF bit is clear. If the user writes new data to PORTD to set OBF, the data is immediately read out; however, the OBF bit is not set.
When either the $\overline{\mathrm{CS}}$ or $\overline{\mathrm{RD}}$ lines are detected high, the PORTD pins return to the input state and the PSPIF bit is set. User applications should wait for PSPIF to be set before servicing the PSP; when this happens, the IBF and OBF bits can be polled and the appropriate action taken.

The timing for the control signals in Write and Read modes is shown in Figure 9-4 and Figure 9-5, respectively.

FIGURE 9-3: PORTD AND PORTE BLOCK DIAGRAM (PARALLEL SLAVE PORT)


Note: I/O pins have diode protection to VDD and Vss.

## PIC18F45J10 FAMILY

FIGURE 9-4: PARALLEL SLAVE PORT WRITE WAVEFORMS


FIGURE 9-5:

## PARALLEL SLAVE PORT READ WAVEFORMS



TABLE 9-13: REGISTERS ASSOCIATED WITH PARALLEL SLAVE PORT

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PORTD ${ }^{(1)}$ | RD7 | RD6 | RD5 | RD4 | RD3 | RD2 | RD1 | RD0 | 46 |
| LATD $^{(1)}$ | PORTD Data Latch Register (Read and Write to Data Latch) |  |  |  |  |  |  |  | 46 |
| TRISD ${ }^{(1)}$ | PORTD Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| PORTE ${ }^{(1)}$ | - | - | - | - | - | RE2 | RE1 | RE0 | 46 |
| LATE ${ }^{(1)}$ | - | - | - | - | - | PORTE Data Latch Register (Read and Write to Data Latch) |  |  | 46 |
| TRISE ${ }^{(1)}$ | IBF | OBF | IBOV | PSPMODE | - | TRISE2 | TRISE1 | TRISE0 | 46 |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| ADCON1 | - | - | VCFG1 | VCFGO | PCFG3 | PCFG2 | PCFG1 | PCFG0 | 44 |

[^1]Note 1: These registers and/or bits are not implemented on 28 -pin devices and should be read as ' 0 '.

### 10.0 TIMERO MODULE

The Timer0 module incorporates the following features:

- Software selectable operation as a timer or counter in both 8-bit or 16-bit modes
- Readable and writable registers
- Dedicated 8-bit, software programmable prescaler
- Selectable clock source (internal or external)
- Edge select for external clock
- Interrupt-on-overflow

The TOCON register (Register 10-1) controls all aspects of the module's operation, including the prescale selection. It is both readable and writable.
A simplified block diagram of the Timer0 module in 8-bit mode is shown in Figure 10-1. Figure $10-2$ shows a simplified block diagram of the TimerO module in 16-bit mode.

## REGISTER 10-1: TOCON: TIMERO CONTROL REGISTER

| R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 | R/W-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMR0ON | T08BIT | T0CS | T0SE | PSA | TOPS2 | TOPS1 | TOPS0 |

bit 7 TMR0ON: Timer0 On/Off Control bit
1 = Enables Timer0
$0=$ Stops Timer0
bit 6 T08BIT: Timer0 8-Bit/16-Bit Control bit
$1=$ Timer0 is configured as an 8-bit timer/counter
$0=$ Timer0 is configured as a 16-bit timer/counter
bit 5 TOCS: Timer0 Clock Source Select bit
1 = Transition on TOCKI pin
$0=$ Internal instruction cycle clock (CLKO)
bit 4 TOSE: TimerO Source Edge Select bit
1 = Increment on high-to-low transition on TOCKI pin
$0=$ Increment on low-to-high transition on TOCKI pin
bit $3 \quad$ PSA: Timer0 Prescaler Assignment bit
1 = TImer0 prescaler is not assigned. Timer0 clock input bypasses prescaler.
$0=$ Timer0 prescaler is assigned. Timer0 clock input comes from prescaler output.
bit 2-0 TOPS2:TOPSO: Timer0 Prescaler Select bits
$111=1: 256$ Prescale value
$110=1: 128$ Prescale value
$101=1: 64$ Prescale value
$100=1: 32$ Prescale value
$011=1: 16$ Prescale value
$010=1: 8 \quad$ Prescale value
$001=1: 4 \quad$ Prescale value
$000=1: 2 \quad$ Prescale value

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 10.1 Timer0 Operation

Timer0 can operate as either a timer or a counter; the mode is selected with the TOCS bit (TOCON $<5>$ ). In Timer mode (TOCS = 0), the module increments on every clock by default unless a different prescaler value is selected (see Section 10.3 "Prescaler"). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMRO register.

The Counter mode is selected by setting the TOCS bit (=1). In this mode, Timer0 increments either on every rising or falling edge of pin RB5/T0CKI. The incrementing edge is determined by the Timer0 Source Edge Select bit, TOSE (TOCON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.
An external clock source can be used to drive Timer0; however, it must meet certain requirements to ensure that the external clock can be synchronized with the
internal phase clock (TosC). There is a delay between synchronization and the onset of incrementing the timer/counter.

### 10.2 TimerO Reads and Writes in 16-Bit Mode

TMROH is not the actual high byte of Timer0 in 16-bit mode. It is actually a buffered version of the real high byte of Timer0 which is not directly readable nor writable (refer to Figure 10-2). TMROH is updated with the contents of the high byte of Timer0 during a read of TMROL. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte were valid, due to a rollover between successive reads of the high and low byte.
Similarly, a write to the high byte of Timer0 must also take place through the TMROH Buffer register. The high byte is updated with the contents of TMROH when a write occurs to TMROL. This allows all 16 bits of Timer0 to be updated at once.

FIGURE 10-1: TIMERO BLOCK DIAGRAM (8-BIT MODE)


Note: Upon Reset, Timer0 is enabled in 8-bit mode with clock input from TOCKI max. prescale.

FIGURE 10-2: TIMERO BLOCK DIAGRAM (16-BIT MODE)


[^2]
### 10.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module. The prescaler is not directly readable or writable. Its value is set by the PSA and TOPS2:TOPSO bits ( $\mathrm{TOCON}<3: 0>$ ) which determine the prescaler assignment and prescale ratio.
Clearing the PSA bit assigns the prescaler to the Timer0 module. When it is assigned, prescale values from 1:2 through 1:256 in power-of-2 increments are selectable.

When assigned to the Timer0 module, all instructions writing to the TMRO register (e.g., CLRF TMRO, MOVWF TMRO, BSF TMRO, etc.) clear the prescaler count.
Note: Writing to TMR0 when the prescaler is assigned to Timer0 will clear the prescaler count but will not change the prescaler assignment.

### 10.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control and can be changed "on-the-fly" during program execution.

### 10.4 TimerO Interrupt

The TMRO interrupt is generated when the TMRO register overflows from FFh to 00h in 8-bit mode, or from FFFFh to 0000h in 16-bit mode. This overflow sets the TMROIF flag bit. The interrupt can be masked by clearing the TMROIE bit (INTCON<5>). Before re-enabling the interrupt, the TMROIF bit must be cleared in software by the Interrupt Service Routine.

Since Timer0 is shut down in Sleep mode, the TMR0 interrupt cannot awaken the processor from Sleep.

TABLE 10-1: REGISTERS ASSOCIATED WITH TIMERO

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMROL | Timer0 Register Low Byte |  |  |  |  |  |  |  |  |
| TMROH | Timer0 Register High Byte |  |  |  |  |  |  |  |  |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| TOCON | TMROON | TO8BIT | TOCS | TOSE | PSA | TOPS2 | TOPS1 | TOPS0 | 44 |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISA0 | 46 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by Timer0.

## PIC18F45J10 FAMILY

NOTES:

### 11.0 TIMER1 MODULE

The Timer1 timer/counter module incorporates these features:

- Software selectable operation as a 16 -bit timer or counter
- Readable and writable 8-bit registers (TMR1H and TMR1L)
- Selectable clock source (internal or external) with device clock or Timer1 oscillator internal options
- Interrupt-on-overflow
- Reset on CCP Special Event Trigger
- Device clock status flag (T1RUN)

A simplified block diagram of the Timer1 module is shown in Figure 11-1. A block diagram of the module's operation in Read/Write mode is shown in Figure 11-2.
The module incorporates its own low-power oscillator to provide an additional clocking option. The Timer1 oscillator can also be used as a low-power clock source for the microcontroller in power-managed operation.
Timer1 can also be used to provide Real-Time Clock (RTC) functionality to applications with only a minimal addition of external components and code overhead.
Timer1 is controlled through the T1CON Control register (Register 11-1). It also contains the Timer1 Oscillator Enable bit (T1OSCEN). Timer1 can be enabled or disabled by setting or clearing control bit, TMR1ON (T1CON<0>).

## REGISTER 11-1: T1CON: TIMER1 CONTROL REGISTER

| R/W-0 | R-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RD16 | T1RUN | T1CKPS1 | T1CKPS0 | T1OSCEN | T1SYNC | TMR1CS | TMR1ON |

bit 7
bit 0
bit 7 RD16: 16-Bit Read/Write Mode Enable bit
1 = Enables register read/write of TImer1 in one 16-bit operation
$0=$ Enables register read/write of Timer1 in two 8-bit operations
bit 6 T1RUN: Timer1 System Clock Status bit
1 = Device clock is derived from Timer1 oscillator
$0=$ Device clock is derived from another source
bit 5-4 T1CKPS1:T1CKPS0: Timer1 Input Clock Prescale Select bits
$11=1: 8$ Prescale value
$10=1: 4$ Prescale value
$01=1: 2$ Prescale value
$00=1: 1$ Prescale value
bit 3 T1OSCEN: Timer1 Oscillator Enable bit
1 = Timer1 oscillator is enabled
$0=$ Timer1 oscillator is shut off
The oscillator inverter and feedback resistor are turned off to eliminate power drain.
bit $2 \quad \overline{\text { T1SYNC: Timer1 External Clock Input Synchronization Select bit }}$
When TMR1CS = 1:
1 = Do not synchronize external clock input
0 = Synchronize external clock input
When TMR1CS = 0:
This bit is ignored. Timer1 uses the internal clock when TMR1CS $=0$.
bit 1 TMR1CS: Timer1 Clock Source Select bit
1 = External clock from pin RC0/T1OSO/T1CKI (on the rising edge)
$0=$ Internal clock (Fosc/4)
bit $0 \quad$ TMR1ON: Timer1 On bit
1 = Enables Timer1
$0=$ Stops Timer1

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

## PIC18F45J10 FAMILY

### 11.1 Timer1 Operation

Timer1 can operate in one of these modes:

- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR1CS (T1CON<1>). When TMR1CS is cleared (= 0), Timer1 increments on every internal instruction
cycle (Fosc/4). When the bit is set, Timer1 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

When Timer1 is enabled, the RC1/T1OSI and RC0/T1OSO/T1CKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as ' 0 '.

FIGURE 11-1: TIMER1 BLOCK DIAGRAM


Note 1: When enable bit, T1OSCEN, is cleared, the inverter and feedback resistor are turned off to eliminate power drain.

FIGURE 11-2: TIMER1 BLOCK DIAGRAM (16-BIT READ/WRITE MODE)


Note 1: When enable bit, T1OSCEN, is cleared, the inverter and feedback resistor are turned off to eliminate power drain.

### 11.2 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes (see Figure 11-2). When the RD16 control bit ( $\mathrm{T} 1 \mathrm{CON}<7>$ ) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L will load the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.
A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits to both the high and low bytes of Timer1 at once.
The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler. The prescaler is only cleared on writes to TMR1L.

### 11.3 Timer1 Oscillator

An on-chip crystal oscillator circuit is incorporated between pins T1OSI (input) and T1OSO (amplifier output). It is enabled by setting the Timer1 Oscillator Enable bit, T1OSCEN (T1CON<3>). The oscillator is a low-power circuit rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical oscillator is shown in Figure 11-3. Table 11-1 shows the capacitor selection for the Timer1 oscillator.

The user must provide a software time delay to ensure proper start-up of the Timer1 oscillator.

FIGURE 11-3: EXTERNAL COMPONENTS FOR THE TIMER1 OSCILLATOR


TABLE 11-1: CAPACITOR SELECTION FOR THETIMEROSCILLATOR ${ }^{(2,3,4)}$

| Oscillator <br> Type | Freq. | C1 | C2 |
| :---: | :---: | :---: | :---: |
| LP | 32 kHz | $27 \mathrm{pF}^{(\mathbf{1})}$ | $27 \mathrm{pF}^{(\mathbf{1})}$ |

Note 1: Microchip suggests these values as a starting point in validating the oscillator circuit.
2: Higher capacitance increases the stability of the oscillator but also increases the start-up time.
3: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
4: Capacitor values are for design guidance only.

### 11.3.1 USING TIMER1 AS A CLOCK SOURCE

The Timer1 oscillator is also available as a clock source in power-managed modes. By setting the clock select bits, SCS1:SCS0 (OSCCON<1:0>), to ' 01 ', the device switches to SEC_RUN mode; both the CPU and peripherals are clocked from the Timer1 oscillator. If the IDLEN bit ( $O S C C O N<7>$ ) is cleared and a SLEEP instruction is executed, the device enters SEC_IDLE mode. Additional details are available in Section 3.0 "Power-Managed Modes".
Whenever the Timer1 oscillator is providing the clock source, the Timer1 system clock status flag, T1RUN ( $\mathrm{T} 1 \mathrm{CON}<6>$ ), is set. This can be used to determine the controller's current clocking mode. It can also indicate the clock source being currently used by the Fail-Safe Clock Monitor. If the Clock Monitor is enabled and the Timer1 oscillator fails while providing the clock, polling the T1RUN bit will indicate whether the clock is being provided by the Timer1 oscillator or another source.

### 11.3.2 TIMER1 OSCILLATOR LAYOUT CONSIDERATIONS

The Timer1 oscillator circuit draws very little power during operation. Due to the low-power nature of the oscillator, it may also be sensitive to rapidly changing signals in close proximity.
The oscillator circuit, shown in Figure 11-3, should be located as close as possible to the microcontroller. There should be no circuits passing within the oscillator circuit boundaries other than Vss or Vdd.
If a high-speed circuit must be located near the oscillator (such as the CCP1 pin in Output Compare or PWM mode, or the primary oscillator using the OSC2 pin), a grounded guard ring around the oscillator circuit, as shown in Figure 11-4, may be helpful when used on a single-sided PCB or in addition to a ground plane.

FIGURE 11-4: OSCILLATOR CIRCUIT WITH GROUNDED GUARD RING


Note: Not drawn to scale.

### 11.4 Timer1 Interrupt

The TMR1 register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The Timer1 interrupt, if enabled, is generated on overflow which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled or disabled by setting or clearing the Timer1 Interrupt Enable bit, TMR1IE (PIE1<0>).

### 11.5 Resetting Timer1 Using the ECCP/CCP Special Event Trigger

If ECCP1/CCP1 or CCP2 are configured to generate a Special Event Trigger in Compare mode (CCPxM3:CCPxM0 = 1011), this signal will reset Timer1. The trigger from CCP2 will also start an A/D conversion if the $A / D$ module is enabled (see Section 14.2.1 "Special Event Trigger" for more information).
The module must be configured as either a timer or a synchronous counter to take advantage of this feature. When used this way, the CCPRH:CCPRL register pair effectively becomes a period register for Timer1.

If Timer1 is running in Asynchronous Counter mode, this Reset operation may not work.
In the event that a write to Timer1 coincides with a Special Event Trigger, the write operation will take precedence.

```
Note: The Special Event Triggers from the
        ECCP1/CCPx module will not set the
        TMR1IF interrupt flag bit (PIR1<0>).
```


### 11.6 Using Timer1 as a Real-Time Clock

Adding an external LP oscillator to Timer1 (such as the one described in Section 11.3 "Timer1 Oscillator" above) gives users the option to include RTC functionality to their applications. This is accomplished with an inexpensive watch crystal to provide an accurate time base and several lines of application code to calculate the time. When operating in Sleep mode and using a battery or supercapacitor as a power source, it can completely eliminate the need for a separate RTC device and battery backup.
The application code routine, RTCisr, shown in Example 11-1, demonstrates a simple method to increment a counter at one-second intervals using an Interrupt Service Routine. Incrementing the TMR1 register pair to overflow triggers the interrupt and calls the routine which increments the seconds counter by one. Additional counters for minutes and hours are incremented as the previous counter overflows.
Since the register pair is 16 bits wide, counting up to overflow the register directly from a 32.768 kHz clock would take 2 seconds. To force the overflow at the required one-second intervals, it is necessary to preload it. The simplest method is to set the MSb of TMR1H with a BSF instruction. Note that the TMR1L register is never preloaded or altered; doing so may introduce cumulative error over many cycles.
For this method to be accurate, Timer1 must operate in Asynchronous mode and the Timer1 overflow interrupt must be enabled (PIE1<0> = 1) as shown in the routine, RTCinit. The Timer1 oscillator must also be enabled and running at all times.

EXAMPLE 11-1: IMPLEMENTING A REAL-TIME CLOCK USING A TIMER1 INTERRUPT SERVICE

| RTCinit |  |  |  |
| :---: | :---: | :---: | :---: |
|  | MOVLW | 80h | ; Preload TMR1 register pair |
|  | MOVWF | TMR1H | ; for 1 second overflow |
|  | CLRF | TMR1L |  |
|  | MOVLW | $\mathrm{b}^{\prime} 00001111^{\prime}$ | ; Configure for external clock, |
|  | MOVWF | T1CON | ; Asynchronous operation, external oscillator |
|  | CLRF | secs | ; Initialize timekeeping registers |
|  | CLRF | mins | ; |
|  | MOVLW | . 12 |  |
|  | MOVWF | hours |  |
|  | BSF | PIE1, TMR1IE | ; Enable Timerl interrupt |
|  | RETURN |  |  |
| RTCiss |  |  |  |
|  | BSF | TMR1H, 7 | ; Preload for 1 sec overflow |
|  | BCF | PIR1, TMR1IF | ; Clear interrupt flag |
|  | INCF | secs, F | ; Increment seconds |
|  | MOVLW | . 59 | ; 60 seconds elapsed? |
|  | CPFSGT | secs |  |
|  | RETURN |  | ; No, done |
|  | CLRF | secs | ; Clear seconds |
|  | INCF | mins, F | ; Increment minutes |
|  | MOVLW | . 59 | ; 60 minutes elapsed? |
|  | CPFSGT | mins |  |
|  | RETURN |  | ; No, done |
|  | CLRF | mins | ; clear minutes |
|  | INCF | hours, F | ; Increment hours |
|  | MOVLW | . 23 | ; 24 hours elapsed? |
|  | CPFSGT | hours |  |
|  | RETURN |  | ; No, done |
|  | CLRF | hours | ; Reset hours |
|  | RETURN |  | ; Done |

TABLE 11-2: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP ${ }^{(1)}$ | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| TMR1L | Timer1 Register Low Byte |  |  |  |  |  |  |  | 44 |
| TMR1H | Timer1 Register High Byte |  |  |  |  |  |  |  | 44 |
| T1CON | RD16 | T1RUN | T1CKPS1 | T1CKPS0 | T1OSCEN | T1SYNC | TMR1CS | TMR1ON | 44 |

Legend: Shaded cells are not used by the Timer1 module.
Note 1: These bits are not implemented on 28 -pin devices and should be read as ' 0 '.

## PIC18F45J10 FAMILY

NOTES:

### 12.0 TIMER2 MODULE

The Timer2 timer module incorporates the following features:

- 8-bit timer and period registers (TMR2 and PR2, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4 and 1:16)
- Software programmable postscaler
(1:1 through 1:16)
- Interrupt on TMR2-to-PR2 match
- Optional use as the shift clock for the MSSP module
The module is controlled through the T2CON register (Register 12-1) which enables or disables the timer and configures the prescaler and postscaler. Timer2 can be shut off by clearing control bit, TMR2ON (T2CON<2>), to minimize power consumption.

A simplified block diagram of the module is shown in Figure 12-1.

### 12.1 Timer2 Operation

In normal operation, TMR2 is incremented from 00h on each clock (Fosc/4). A 4-bit counter/prescaler on the clock input gives direct input, divide-by-4 and divide-by-16 prescale options; these are selected by the prescaler control bits, T2CKPS1:T2CKPS0 ( $\mathrm{T} 2 \mathrm{CON}<1: 0>$ ). The value of TMR2 is compared to that of the period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/postscaler (see Section 12.2 "Timer2 Interrupt").
The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, while the PR2 register initializes at FFh. Both the prescaler and postscaler counters are cleared on the following events:

- a write to the TMR2 register
- a write to the T2CON register
- any device Reset (Power-on Reset, $\overline{\text { MCLR }}$ Reset, Watchdog Timer Reset or Brown-out Reset)

TMR2 is not cleared when T2CON is written.

## REGISTER 12-1: T2CON: TIMER2 CONTROL REGISTER

| U-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | T2OUTPS3 | T2OUTPS2 | T2OUTPS1 | T2OUTPS0 | TMR2ON | T2CKPS1 | T2CKPS0 |
| bit 7 |  |  |  |  |  |  |  |

bit 7 Unimplemented: Read as ' 0 '
bit 6-3 T2OUTPS3:T2OUTPS0: Timer2 Output Postscale Select bits
$0000=1: 1$ Postscale
0001 = 1:2 Postscale
-
-
-
$1111=1: 16$ Postscale
bit 2 TMR2ON: Timer2 On bit
$1=$ Timer2 is on
$0=$ Timer2 is off
bit 1-0 T2CKPS1:T2CKPS0: Timer2 Clock Prescale Select bits
$00=$ Prescaler is 1
$01=$ Prescaler is 4
$1 \mathrm{x}=$ Prescaler is 16

## Legend:

| $\mathrm{R}=$ Readable bit | $\mathrm{W}=$ Writable bit | $\mathrm{U}=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' $=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 12.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2-to-PR2 match) provides the input for the 4-bit output counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF (PIR1<1>). The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE (PIE1<1>).
A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS3:T2OUTPS0 (T2CON<6:3>).

### 12.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the CCP modules, where it is used as a time base for operations in PWM mode.
Timer2 can be optionally used as the shift clock source for the MSSP module operating in SPI mode. Additional information is provided in Section 15.0 "Master Synchronous Serial Port (MSSP) Module".

FIGURE 12-1: TIMER2 BLOCK DIAGRAM


TABLE 12-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INT0IE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF $^{(\mathbf{1})}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE $^{(\mathbf{1})}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| TMR2 | Timer2 Register |  |  |  |  |  |  |  |  |
| T2CON | - | T2OUTPS3 | T2OUTPS2 | T2OUTPS1 | T2OUTPS0 | TMR2ON | T2CKPS1 | T2CKPS0 | 44 |
| PR2 | Timer2 Period Register |  |  |  |  |  |  |  |  |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by the Timer2 module.
Note 1: These bits are not implemented on 28-pin devices and should be read as ' 0 '.

### 13.0 CAPTURE/COMPARE/PWM (CCP) MODULES

PIC18F45J10 family devices all have two CCP (Capture/Compare/PWM) modules. Each module contains a 16-bit register which can operate as a 16-bit Capture register, a 16-bit Compare register or a PWM Master/Slave Duty Cycle register.
In 28-pin devices, the two standard CCP modules (CCP1 and CCP2) operate as described in this chapter. In 40/44-pin devices, CCP1 is implemented as an Enhanced CCP module (ECCP1) with standard Capture and Compare modes and Enhanced PWM modes. The Enhanced CCP implementation is discussed in Section 14.0 "Enhanced Capture/Compare/PWM (ECCP) Module".

The Capture and Compare operations described in this chapter apply to all standard and Enhanced CCP modules.

Note: Throughout this section and Section 14.0 "Enhanced Capture/Compare/PWM (ECCP) Module", references to the register and bit names for CCP modules are referred to generically by the use of ' $x$ ' or ' $y$ ' in place of the specific module number. Thus, "CCPxCON" might refer to the control register for CCP1, CCP2 or ECCP1. "CCPxCON" is used throughout these sections to refer to the module control register regardless of whether the CCP module is a standard or Enhanced implementation.

## REGISTER 13-1: CCPxCON REGISTER (CCP1, CCP2 MODULES IN 28-PIN DEVICES)

| U-0 | U-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |  | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | ---: |
| - | - | RCxB1 | DCxB0 | CCPxM3 | CCPxM2 | CCPxM1 | CCPxM0 |
| bit 7 |  |  |  |  |  |  |  |

bit 7-6 Unimplemented: Read as ' 0 '
bit 5-4 DCxB1:DCxB0: PWM Duty Cycle bit 1 and bit 0
Capture mode:
Unused.
Compare mode:
Unused.
PWM mode:
These bits are the two LSbs (bit 1 and bit 0 ) of the 10-bit PWM duty cycle. The eight MSbs (DCx9:DCx2) of the duty cycle are found in CCPRxL.
bit 3-0 CCPxM3:CCPxM0: CCPx Mode Select bits
0000 = Capture/Compare/PWM disabled (resets CCPx module)
0001 = Reserved
0010 = Compare mode, toggle output on match (CCPxIF bit is set)
0011 = Reserved
0100 = Capture mode, every falling edge
0101 = Capture mode, every rising edge
$0110=$ Capture mode, every 4th rising edge
0111 = Capture mode, every 16th rising edge
$1000=$ Compare mode: initialize CCPx pin low; on compare match, force CCPx pin high (CCPxIF bit is set)
1001 = Compare mode: initialize CCPx pin high; on compare match, force CCPx pin low (CCPxIF bit is set)
$1010=$ Compare mode: generate software interrupt on compare match (CCPxIF bit is set, CCPx pin reflects I/O state)
1011 = Compare mode: trigger special event, reset timer, start A/D conversion on CCPx match (CCPxIF bit is set)
$11 x x=P W M$ mode

## Legend:

| $\mathrm{R}=$ Readable bit | $\mathrm{W}=$ Writable bit | $\mathrm{U}=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 13.1 CCP Module Configuration

Each Capture/Compare/PWM module is associated with a control register (generically, CCPxCON) and a data register (CCPRx). The data register, in turn, is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte). All registers are both readable and writable.

### 13.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize Timers 1 or 2, depending on the mode selected. Timer1 is available to modules in Capture or Compare modes, while Timer2 is available for modules in PWM mode.

## TABLE 13-1: ECCP/CCP MODE - TIMER RESOURCE

| ECCP/CCP Mode | Timer Resource |
| :---: | :---: |
| Capture | Timer1 |
| Compare | Timer1 |
| PWM | Timer2 |

Both modules may be active at any given time and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time. The interactions between the two modules are summarized in Figure 13-1 and Figure 13-2. In Timer1 in Asynchronous Counter mode, the capture operation will not work.

### 13.1.2 CCP2 PIN ASSIGNMENT

The pin assignment for CCP2 (Capture input, Compare and PWM output) can change, based on device configuration. The CCP2MX configuration bit determines which pin CCP2 is multiplexed to. By default, it is assigned to RC1 (CCP2MX = 1). If the configuration bit is cleared, CCP2 is multiplexed with RB3.
Changing the pin assignment of CCP2 does not automatically change any requirements for configuring the port pin. Users must always verify that the appropriate TRIS register is configured correctly for CCP2 operation regardless of where it is located.

TABLE 13-2: INTERACTIONS BETWEEN ECCP1/CCP1 AND CCP2 FOR TIMER RESOURCES

| CCP1 Mode | CCP2 Mode | Interaction |
| :---: | :---: | :--- |
| Capture | Capture | Each module uses TMR1 as the time base. |
| Capture | Compare | CCP2 can be configured for the Special Event Trigger to reset TMR1. Automatic A/D <br> conversions on the trigger event can also be done. Operation of ECCP1/CCP1 will be <br> affected. |
| Compare | Capture | ECCP1/CCP1 can be configured for the Special Event Trigger to reset TMR1. Operation <br> of CCP2 will be affected. |
| Compare | Compare | Either module can be configured for the Special Event Trigger to reset TMR1. Automatic <br> A/D conversions on the CCP2 trigger event can be done. |
| Capture | PWM $^{(\mathbf{1})}$ | None |
| Compare | PWM $^{(\mathbf{1})}$ | None |
| PWM $^{(\mathbf{1})}$ | Capture | None |
| $\mathrm{PWM}^{(\mathbf{1})}$ | Compare | None |
| $\mathrm{PWM}^{(\mathbf{1 )}}$ | PWM | Both PWMs will have the same frequency and update rate (TMR2 interrupt). |

Note 1: Includes standard and Enhanced PWM operation.

### 13.2 Capture Mode

In Capture mode, the CCPRxH:CCPRxL register pair captures the 16-bit value of the TMR1 register when an event occurs on the corresponding CCPx pin. An event is defined as one of the following:

- every falling edge
- every rising edge
- every 4th rising edge
- every 16 th rising edge

The event is selected by the mode select bits, CCPxM3:ССРxM0 (CCPxCON $<3: 0>$ ). When a capture is made, the interrupt request flag bit, CCPxIF, is set; it must be cleared in software. If another capture occurs before the value in register CCPRx is read, the old captured value is overwritten by the new captured value.

### 13.2.1 CCP PIN CONFIGURATION

In Capture mode, the appropriate CCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

## Note: If RB3/CCP2 or RC1/CCP2 is configured

 as an output, a write to the port can cause a capture condition.
### 13.2.2 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit clear to avoid false interrupts. The interrupt flag bit, CCPxIF, should also be cleared following any such change in operating mode.

### 13.2.3 CCP PRESCALER

There are four prescaler settings in Capture mode; they are specified as part of the operating mode selected by the mode select bits (CCPxM3:CCPxM0). Whenever the CCP module is turned off or Capture mode is disabled, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.
Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared; therefore, the first capture may be from a non-zero prescaler. Example 13-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAMPLE 13-1: CHANGING BETWEEN CAPTURE PRESCALERS (CCP2 SHOWN)

| CLRF | CCP2CON | $;$ Turn CCP module off |
| :--- | :--- | :--- |
| MOVLW | NEW_CAPT_PS | $;$ Load WREG with the |
|  |  | $;$ new prescaler mode |
| MOVWF | CCP2CON | ; Lalue and CCP ON <br>  |
|  |  | Load CCP2CON with |

FIGURE 13-1: CAPTURE MODE OPERATION BLOCK DIAGRAM


### 13.3 Compare Mode

In Compare mode, the 16-bit CCPRx register value is constantly compared against the TMR1 register value. When a match occurs, the CCPx pin can be:

- driven high
- driven low
- toggled (high-to-low or low-to-high)
- remain unchanged (that is, reflects the state of the I/O latch)
The action on the pin is based on the value of the mode select bits (CCPxM3:CCPxM0). At the same time, the interrupt flag bit, CCPxIF, is set.


### 13.3.1 CCP PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the appropriate TRIS bit.

Note: Clearing the CCP2CON register will force the RB3 or RC1 compare output latch (depending on device configuration) to the default low level. This is not the PORTB or PORTC I/O data latch.

### 13.3.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode or Synchronized Counter mode if the CCP module is using the compare feature. In Asynchronous Counter mode, the compare operation may not work.

### 13.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen ( $С С Р x \mathrm{M} 3: С С Р x M 0=1010$ ), the corresponding CCPx pin is not affected. Only a CCP interrupt is generated, if enabled and the CCPxIE bit is set.

### 13.3.4 SPECIAL EVENT TRIGGER

Both CCP modules are equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode $(C C P x M 3: С С Р x M 0=1011)$.
For either CCP module, the Special Event Trigger resets the Timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable period register for either timer.
The Special Event Trigger for CCP2 can also start an A/D conversion. In order to do this, the A/D converter must already be enabled.

FIGURE 13-2: COMPARE MODE OPERATION BLOCK DIAGRAM


TABLE 13-3: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE AND TIMER1

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMR0IF | INTOIF | RBIF | 43 |
| RCON | IPEN | - | - | $\overline{\mathrm{RI}}$ | $\overline{\mathrm{TO}}$ | $\overline{\mathrm{PD}}$ | $\overline{\text { POR }}$ | $\overline{\mathrm{BOR}}$ | 42 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 45 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 45 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 45 |
| TRISB | PORTB Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TRISC | PORTC Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TMR1L | Timer1 Register Low Byte |  |  |  |  |  |  |  | 44 |
| TMR1H | Timer1 Register High Byte |  |  |  |  |  |  |  | 44 |
| T1CON | RD16 | T1RUN | T1CKPS1 | T1CKPS0 | T1OSCEN | T1SYNC | TMR1CS | TMR1ON | 44 |
| CCPR1L | Capture/Compare/PWM Register 1 Low Byte |  |  |  |  |  |  |  | 45 |
| CCPR1H | Capture/Compare/PWM Register 1 High Byte |  |  |  |  |  |  |  | 45 |
| CCP1CON | P1M1 ${ }^{(1)}$ | P1M0 ${ }^{(1)}$ | DC1B1 | DC1B0 | CCP1M3 | CCP1M2 | CCP1M1 | CCP1M0 | 45 |
| CCPR2L | Capture/Compare/PWM Register 2 Low Byte |  |  |  |  |  |  |  | 45 |
| CCPR2H | Capture/Compare/PWM Register 2 High Byte |  |  |  |  |  |  |  | 45 |
| CCP2CON | - | - | DC2B1 | DC2B0 | CCP2M3 | CCP2M2 | CCP2M1 | CCP2M0 | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by Capture/Compare or Timer1.
Note 1: These bits are not implemented on 28 -pin devices and should be read as ' 0 '.

### 13.4 PWM Mode

In Pulse-Width Modulation (PWM) mode, the CCPx pin produces up to a 10-bit resolution PWM output. Since the CCP2 pin is multiplexed with a PORTB or PORTC data latch, the appropriate TRIS bit must be cleared to make the CCP2 pin an output.

Note: Clearing the CCP2CON register will force the RB3 or RC1 output latch (depending on device configuration) to the default low level. This is not the PORTB or PORTC I/O data latch.

Figure 13-3 shows a simplified block diagram of the CCP module in PWM mode.

For a step-by-step procedure on how to set up the CCP module for PWM operation, see Section 13.4.4 "Setup for PWM Operation".

FIGURE 13-3: SIMPLIFIED PWM BLOCK DIAGRAM


Note 1: The 8-bit TMR2 value is concatenated with the 2-bit internal Q clock, or 2 bits of the prescaler, to create the 10-bit time base.

A PWM output (Figure 13-4) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

FIGURE 13-4: PWM OUTPUT


### 13.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

## EQUATION 13-1:

$$
\begin{aligned}
\text { PWM Period }= & {[(\mathrm{PR} 2)+1] \bullet 4 \bullet \mathrm{TosC} \bullet } \\
& (\mathrm{TMR} 2 \text { Prescale Value })
\end{aligned}
$$

PWM frequency is defined as $1 /[\mathrm{PWM}$ period].
When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCPx pin is set (exception: if PWM duty cycle $=0 \%$, the CCPx pin will not be set)
- The PWM duty cycle is latched from CCPRxL into CCPRxH
Note: $\quad$ The Timer2 postscalers (see Section 12.0 "Timer2 Module") are not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.


### 13.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPRxL register and to the CCPxCON<5:4> bits. Up to 10-bit resolution is available. The CCPRxL contains the eight MSbs and the CCPxCON<5:4> contains the two LSbs. This 10 -bit value is represented by CCPRxL:CCPxCON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

## EQUATION 13-2:

```
PWM Duty Cycle = (CCPRxL:CCPxCON<5:4>) \bullet
    TOSC • (TMR2 Prescale Value)
```

CCPRxL and CCPxCON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPRxH until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPRxH is a read-only register.

The CCPRxH register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation.

When the CCPRxH and 2-bit latch match TMR2, concatenated with an internal 2-bit $Q$ clock or 2 bits of the TMR2 prescaler, the CCPx pin is cleared.
The maximum PWM resolution (bits) for a given PWM frequency is given by the equation:

EQUATION 13-3:
PWM Resolution (max) $=\frac{\log \left(\frac{\text { FOSC }}{\text { FPWM }}\right)}{\log (2)}$ bits

Note: If the PWM duty cycle value is longer than the PWM period, the CCP2 pin will not be cleared.

TABLE 13-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 40 MHz

| PWM Frequency | $\mathbf{2 . 4 4} \mathbf{~ k H z}$ | $\mathbf{9 . 7 7} \mathbf{~ k H z}$ | $\mathbf{3 9 . 0 6} \mathbf{~ k H z}$ | $\mathbf{1 5 6 . 2 5} \mathbf{~ k H z}$ | $\mathbf{3 1 2 . 5 0} \mathbf{~ k H z}$ | $\mathbf{4 1 6 . 6 7} \mathbf{~ k H z}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Timer Prescaler (1, 4, 16) | 16 | 4 | 1 | 1 | 1 | 1 |
| PR2 Value | FFh | FFh | FFh | $3 F h$ | $1 F h$ | 17 h |
| Maximum Resolution (bits) | 10 | 10 | 10 | 8 | 7 | 6.58 |

### 13.4.3 PWM AUTO-SHUTDOWN (CCP1 ONLY)

The PWM auto-shutdown features of the Enhanced CCP module are also available to CCP1 in 28-pin devices. The operation of this feature is discussed in detail in Section 14.4.7 "Enhanced PWM Auto-Shutdown".
Auto-shutdown features are not available for CCP2.

### 13.4.4 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

1. Set the PWM period by writing to the PR2 register.
2. Set the PWM duty cycle by writing to the CCPRxL register and CCPxCON $<5: 4>$ bits.
3. Make the CCPx pin an output by clearing the appropriate TRIS bit.
4. Set the TMR2 prescale value, then enable Timer2 by writing to T2CON.
5. Configure the CCPx module for PWM operation.

## PIC18F45J10 FAMILY

TABLE 13-5: REGISTERS ASSOCIATED WITH PWM AND TIMER2

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| RCON | IPEN | - | - | $\overline{\mathrm{RI}}$ | TO | $\overline{\mathrm{PD}}$ | $\overline{\text { POR }}$ | BOR | 42 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| TRISB | PORTB Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TRISC | PORTC Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TMR2 | Timer2 Register |  |  |  |  |  |  |  | 44 |
| PR2 | Timer2 Period Register |  |  |  |  |  |  |  | 44 |
| T2CON | - | T2OUTPS3 | T2OUTPS2 | T2OUTPS1 | T2OUTPS0 | TMR2ON | T2CKPS1 | T2CKPS0 | 44 |
| CCPR1L | Capture/Compare/PWM Register 1 Low Byte |  |  |  |  |  |  |  | 45 |
| CCPR1H | Capture/Compare/PWM Register 1 High Byte |  |  |  |  |  |  |  | 45 |
| CCP1CON | P1M1 ${ }^{(1)}$ | P1M0 ${ }^{(1)}$ | DC1B1 | DC1B0 | CCP1M3 | CCP1M2 | CCP1M1 | CCP1M0 | 45 |
| CCPR2L | Capture/Compare/PWM Register 2 Low Byte |  |  |  |  |  |  |  | 45 |
| CCPR2H | Capture/Compare/PWM Register 2 High Byte |  |  |  |  |  |  |  | 45 |
| CCP2CON | - | - | DC2B1 | DC2B0 | CCP2M3 | CCP2M2 | CCP2M1 | CCP2M0 | 45 |
| ECCP1AS | ECCPASE | ECCPAS2 | ECCPAS1 | ECCPAS0 | PSSAC1 | PSSAC0 | PSSBD1 ${ }^{(1)}$ | PSSBD0 ${ }^{(1)}$ | 45 |
| ECCP1DEL | PRSEN | PDC6 ${ }^{(1)}$ | PDC5 ${ }^{(1)}$ | PDC4 ${ }^{(1)}$ | PDC3 ${ }^{(1)}$ | PDC2 ${ }^{(1)}$ | PDC1 ${ }^{(1)}$ | PDC0 ${ }^{(1)}$ | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by PWM or Timer2.
Note 1: These bits are not implemented on 28-pin devices and should be read as ' 0 '.

### 14.0 ENHANCED CAPTURE/ COMPARE/PWM (ECCP) MODULE

In PIC18F44J10/45J10 devices, ECCP1 is implemented as a standard CCP module with Enhanced PWM capabilities. These include the provisions for 2 or 4 output channels, user-selectable polarity, dead-band control and automatic shutdown
and restart. The Enhanced features are discussed in detail in Section 14.4 "Enhanced PWM Mode". Capture, Compare and single output PWM functions of the ECCP module are the same as described for the standard CCP module.
The control register for the Enhanced CCP module is shown in Register 14-1. It differs from the CCP1CON register in PIC18F24J10/25J10 devices in that the two Most Significant bits are implemented to control PWM functionality.

## REGISTER 14-1: CCP1CON REGISTER (ECCP1 MODULE, 40/44-PIN DEVICES)

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| P1M1 | P1M0 | DC1B1 | DC1B0 | CCP1M3 | CCP1M2 | CCP1M1 | CCP1M0 |
| bit 7 |  |  |  |  |  |  |  |

bit 7-6 P1M1:P1M0: Enhanced PWM Output Configuration bits
If CCP1M3:CCP1M2 $=00,01,10$ :
$\mathrm{xx}=\mathrm{P} 1 \mathrm{~A}$ assigned as Capture/Compare input/output; P1B, P1C, P1D assigned as port pins
If CCP1M3: $\mathrm{CCP} 1 \mathrm{M} 2=11$ :
$00=$ Single output: P1A modulated; P1B, P1C, P1D assigned as port pins
01 = Full-bridge output forward: P1D modulated; P1A active; P1B, P1C inactive
$10=$ Half-bridge output: P1A, P1B modulated with dead-band control; P1C, P1D assigned as port pins
11 = Full-bridge output reverse: P1B modulated; P1C active; P1A, P1D inactive
bit 5-4 DC1B1:DC1B0: PWM Duty Cycle bit 1 and bit 0
Capture mode:
Unused.
Compare mode:
Unused.
PWM mode:
These bits are the two LSbs of the 10-bit PWM duty cycle. The eight MSbs of the duty cycle are found in CCPR1L.
bit 3-0 CCP1M3:CCP1M0: Enhanced CCP Mode Select bits
0000 = Capture/Compare/PWM off (resets ECCP module)
0001 = Reserved
$0010=$ Compare mode, toggle output on match
0011 = Capture mode
0100 = Capture mode, every falling edge
0101 = Capture mode, every rising edge
$0110=$ Capture mode, every 4th rising edge
0111 = Capture mode, every 16th rising edge
$1000=$ Compare mode, initialize CCP1 pin low, set output on compare match (set CCP1IF)
1001 = Compare mode, initialize CCP1 pin high, clear output on compare match (set CCP1IF)
1010 = Compare mode, generate software interrupt only, CCP1 pin reverts to I/O state
1011 = Compare mode, trigger special event (ECCP resets TMR1, sets CCP1IF bit)
$1100=$ PWM mode; P1A, P1C active-high; P1B, P1D active-high
1101 = PWM mode; P1A, P1C active-high; P1B, P1D active-low
$1110=$ PWM mode; P1A, P1C active-low; P1B, P1D active-high
$1111=$ PWM mode; P1A, P1C active-low; P1B, P1D active-low

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' $=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

In addition to the expanded range of modes available through the CCP1CON register and ECCP1AS register, the ECCP module has an additional register associated with Enhanced PWM operation and auto-shutdown features. It is:

- ECCP1DEL (Dead-Band Delay)


### 14.1 ECCP Outputs and Configuration

The Enhanced CCP module may have up to four PWM outputs, depending on the selected operating mode. These outputs, designated P1A through P1D, are multiplexed with I/O pins on PORTC and PORTD. The outputs that are active depend on the ECCP operating mode selected. The pin assignments are summarized in Table 14-1.
To configure the I/O pins as PWM outputs, the proper PWM mode must be selected by setting the P1M1:P1M0 and CCP1M3:CCP1M0 bits. The appropriate TRISC and TRISD direction bits for the port pins must also be set as outputs.

### 14.1.1 ECCP MODULES AND TIMER RESOURCES

Like the standard CCP modules, the ECCP module can utilize Timers 1 or 2 , depending on the mode selected. Timer 1 is available for modules in Capture or Compare modes, while Timer2 is available for modules in PWM mode. Interactions between the standard and Enhanced CCP modules are identical to those described for standard CCP modules. Additional details on timer resources are provided in Section 13.1.1 "CCP Modules and Timer Resources".

### 14.2 Capture and Compare Modes

Except for the operation of the Special Event Trigger discussed below, the Capture and Compare modes of the ECCP module are identical in operation to that of CCP2. These are discussed in detail in Section 13.2 "Capture Mode" and Section 13.3 "Compare Mode". No changes are required when moving between 28-pin and 40/44-pin devices.

### 14.2.1 SPECIAL EVENT TRIGGER

The Special Event Trigger output of ECCP1 resets the TMR1 register pair. This allows the CCPR1 register to effectively be a 16-bit programmable period register for Timer1.

### 14.3 Standard PWM Mode

When configured in Single Output mode, the ECCP module functions identically to the standard CCP module in PWM mode, as described in Section 13.4 "PWM Mode". This is also sometimes referred to as "Compatible CCP" mode, as in Table 14-1.

Note: When setting up single output PWM operations, users are free to use either of the processes described in Section 13.4.4 "Setup for PWM Operation" or Section 14.4.9 "Setup for PWM Operation". The latter is more generic and will work for either single or multi-output PWM.

TABLE 14-1: PIN ASSIGNMENTS FOR VARIOUS ECCP1 MODES

| ECCP Mode | CCP1CON <br> Configuration | RC2 | RD5 | RD6 | RD7 |
| :--- | :---: | :---: | :---: | :---: | :---: |
| All 40/44-pin Devices: |  |  |  |  |  |
| Compatible CCP | $00 x x$ 11xx | CCP1 | RD5/PSP5 | RD6/PSP6 | RD7/PSP7 |
| Dual PWM | $10 x x ~ 11 x x$ | P1A | P1B | RD6/PSP6 | RD7/PSP7 |
| Quad PWM | x1xx 11xx | P1A | P1B | P1C | P1D |

Legend: $\mathrm{x}=$ Don't care. Shaded cells indicate pin assignments not used by ECCP1 in a given mode.

### 14.4 Enhanced PWM Mode

The Enhanced PWM mode provides additional PWM output options for a broader range of control applications. The module is a backward compatible version of the standard CCP module and offers up to four outputs, designated P1A through P1D. Users are also able to select the polarity of the signal (either active-high or active-low). The module's output mode and polarity are configured by setting the P1M1:P1M0 and CCP1M3:CCP1M0 bits of the CCP1CON register.
Figure 14-1 shows a simplified block diagram of PWM operation. All control registers are double-buffered and are loaded at the beginning of a new PWM cycle (the period boundary when Timer2 resets) in order to prevent glitches on any of the outputs. The exception is the PWM Delay register, ECCP1DEL, which is loaded at either the duty cycle boundary or the period boundary (whichever comes first). Because of the buffering, the module waits until the assigned timer resets instead of starting immediately. This means that Enhanced PWM waveforms do not exactly match the standard PWM waveforms, but are instead offset by one full instruction cycle (4 Tosc).
As before, the user must manually configure the appropriate TRIS bits for output.

### 14.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following equation.

## EQUATION 14-1:

$$
\begin{aligned}
\text { PWM Period }= & {[(\mathrm{PR} 2)+1] \bullet 4 \bullet \text { TosC } \bullet } \\
& (\mathrm{TMR2} \text { Prescale Value })
\end{aligned}
$$

PWM frequency is defined as $1 /[P W M$ period]. When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set (if PWM duty cycle $=0 \%$, the CCP1 pin will not be set)
- The PWM duty cycle is copied from CCPR1L into CCPR1H

Note: The Timer2 postscaler (see Section 12.0 "Timer2 Module") is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

## FIGURE 14-1: SIMPLIFIED BLOCK DIAGRAM OF THE ENHANCED PWM MODULE



Note: The 8-bit TMR2 register is concatenated with the 2-bit internal Q clock, or 2 bits of the prescaler, to create the 10-bit time base.

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### 14.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR1L register and to the CCP1CON $<5: 4>$ bits. Up to 10 -bit resolution is available. The CCPR1L register contains the eight MSbs and the CCP1CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR1L:CCP1CON $<5: 4>$. The PWM duty cycle is calculated by the following equation:

## EQUATION 14-2:

$$
\begin{aligned}
\text { PWM Duty Cycle }= & (\mathrm{CCPR} 1 \mathrm{~L}: \mathrm{CCP1CON}\langle 5: 4>) \bullet \\
& \operatorname{TosC} \bullet(\mathrm{TMR2} \text { Prescale Value })
\end{aligned}
$$

CCPR1L and CCP1CON<5:4> can be written to at any time, but the duty cycle value is not copied into CCPR1H until a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR1H is a read-only register.
The CCPR1H register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation. When the CCPR1H and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock or two bits of the TMR2 prescaler, the CCP1 pin is cleared. The maximum PWM resolution (bits) for a given PWM frequency is given by the following equation:

EQUATION 14-3:
PWM Resolution (max $)=\frac{\log \left(\frac{\text { FOSC }}{\text { FPWM }}\right)}{\log (2)}$ bits

Note: If the PWM duty cycle value is longer than the PWM period, the CCP1 pin will not be cleared.

### 14.4.3 PWM OUTPUT CONFIGURATIONS

The P1M1:P1M0 bits in the CCP1CON register allow one of four configurations:

- Single Output
- Half-Bridge Output
- Full-Bridge Output, Forward mode
- Full-Bridge Output, Reverse mode

The Single Output mode is the standard PWM mode discussed in Section 14.4 "Enhanced PWM Mode". The Half-Bridge and Full-Bridge Output modes are covered in detail in the sections that follow.

The general relationship of the outputs in all configurations is summarized in Figure 14-2.

TABLE 14-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 40 MHz

| PWM Frequency | $\mathbf{2 . 4 4} \mathbf{~ k H z}$ | $\mathbf{9 . 7 7} \mathbf{~ k H z}$ | $\mathbf{3 9 . 0 6} \mathbf{~ k H z}$ | $\mathbf{1 5 6 . 2 5} \mathbf{~ k H z}$ | $\mathbf{3 1 2 . 5 0} \mathbf{~ k H z}$ | $\mathbf{4 1 6 . 6 7} \mathbf{~ k H z}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Timer Prescaler (1, 4, 16) | 16 | 4 | 1 | 1 | 1 | 1 |
| PR2 Value | FFh | FFh | FFh | $3 F \mathrm{~h}$ | 1Fh | 17 h |
| Maximum Resolution (bits) | 10 | 10 | 10 | 8 | 7 | 6.58 |

FIGURE 14-2: PWM OUTPUT RELATIONSHIPS (ACTIVE-HIGH STATE)


FIGURE 14-3: PWM OUTPUT RELATIONSHIPS (ACTIVE-LOW STATE)


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### 14.4.4 HALF-BRIDGE MODE

In the Half-Bridge Output mode, two pins are used as outputs to drive push-pull loads. The PWM output signal is output on the P1A pin, while the complementary PWM output signal is output on the P1B pin (Figure 14-4). This mode can be used for half-bridge applications, as shown in Figure 14-5, or for full-bridge applications where four power switches are being modulated with two PWM signals.
In Half-Bridge Output mode, the programmable deadband delay can be used to prevent shoot-through current in half-bridge power devices. The value of bits, PDC6:PDC0, sets the number of instruction cycles before the output is driven active. If the value is greater than the duty cycle, the corresponding output remains inactive during the entire cycle. See Section 14.4.6 "Programmable Dead-Band Delay" for more details of the dead-band delay operations.

Since the P1A and P1B outputs are multiplexed with the PORTC<2> and PORTD<5> data latches, the TRISC<2> and TRISD<5> bits must be cleared to configure P 1 A and P 1 B as outputs.

FIGURE 14-4: HALF-BRIDGE PWM OUTPUT

td = Dead-Band Delay
Note 1: At this time, the TMR2 register is equal to the PR2 register.
2: Output signals are shown as active-high.

FIGURE 14-5: EXAMPLES OF HALF-BRIDGE OUTPUT MODE APPLICATIONS


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### 14.4.5 FULL-BRIDGE MODE

In Full-Bridge Output mode, four pins are used as outputs; however, only two outputs are active at a time. In the Forward mode, pin P1A is continuously active and pin P1D is modulated. In the Reverse mode, pin P1C is continuously active and pin P1B is modulated. These are illustrated in Figure 14-6.

P1A, P1B, P1C and P1D outputs are multiplexed with the PORTC<2> and PORTD<7:5> data latches. The TRISC<2> and TRISD<7:5> bits must be cleared to make the P1A, P1B, P1C and P1D pins outputs.

FIGURE 14-6: FULL-BRIDGE PWM OUTPUT


Reverse Mode


Note 1: At this time, the TMR2 register is equal to the PR2 register.
2: Output signal is shown as active-high.

FIGURE 14-7: EXAMPLE OF FULL-BRIDGE APPLICATION


### 14.4.5.1 Direction Change in Full-Bridge Mode

In the Full-Bridge Output mode, the P1M1 bit in the CCP1CON register allows the user to control the forward/reverse direction. When the application firmware changes this direction control bit, the module will assume the new direction on the next PWM cycle.
Just before the end of the current PWM period, the modulated outputs (P1B and P1D) are placed in their inactive state, while the unmodulated outputs (P1A and P1C) are switched to drive in the opposite direction. This occurs in the time interval, 4 Tosc * (Timer2 Prescale Value), before the next PWM period begins. The Timer2 prescaler will be either 1, 4 or 16, depending on the value of the T2CKPS1:T2CKPS0 bits ( $\mathrm{T} 2 \mathrm{CON}<1: 0>$ ). During the interval from the switch of the unmodulated outputs to the beginning of the next period, the modulated outputs (P1B and P1D) remain inactive. This relationship is shown in Figure 14-8.
Note that in the Full-Bridge Output mode, the ECCP1 module does not provide any dead-band delay. In general, since only one output is modulated at all times, dead-band delay is not required. However, there is a situation where a dead-band delay might be required. This situation occurs when both of the following conditions are true:

1. The direction of the PWM output changes when the duty cycle of the output is at or near $100 \%$.
2. The turn-off time of the power switch, including the power device and driver circuit, is greater than the turn-on time.

Figure 14-9 shows an example where the PWM direction changes from forward to reverse at a near $100 \%$ duty cycle. At time t1, the outputs P1A and P1D become inactive while output P1C becomes active. In this example, since the turn-off time of the power devices is longer than the turn-on time, a shoot-through current may flow through power devices, QC and QD (see Figure 14-7), for the duration of ' t '. The same phenomenon will occur to power devices, QA and QB, for PWM direction change from reverse to forward.
If changing PWM direction at high duty cycle is required for an application, one of the following requirements must be met:

1. Reduce PWM for a PWM period before changing directions.
2. Use switch drivers that can drive the switches off faster than they can drive them on.
Other options to prevent shoot-through current may exist.

FIGURE 14-8: PWM DIRECTION CHANGE


Note 1: The direction bit in the ECCP1 Control register (CCP1CON $<7>$ ) is written any time during the PWM cycle.
2: When changing directions, the P1A and P1C signals switch before the end of the current PWM cycle at intervals of 4 Tosc, 16 Tosc or 64 Tosc, depending on the Timer2 prescaler value. The modulated P1B and P1D signals are inactive at this time.

FIGURE 14-9: PWM DIRECTION CHANGE AT NEAR 100\% DUTY CYCLE


### 14.4.6 PROGRAMMABLE DEAD-BAND DELAY

| Note: | Programmable dead-band delay is not <br> implemented in 28-pin devices with <br> standard CCP modules. |
| :--- | :--- |

In half-bridge applications, where all power switches are modulated at the PWM frequency at all times, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on and the other turned off), both switches may be on for a short period of time until one switch completely turns off. During this brief interval, a very high current (shootthrough current) may flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shoot-through current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.
In the Half-Bridge Output mode, a digitally programmable dead-band delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the nonactive state to the active state. See Figure 14-4 for an illustration. Bits PDC6:PDC0 of the ECCP1DEL register (Register 14-2) set the delay period in terms of microcontroller instruction cycles (Tcy or 4 Tosc). These bits are not available in 28 -pin devices as the standard CCP module does not support half-bridge operation.

### 14.4.7 ENHANCED PWM AUTO-SHUTDOWN

When the ECCP1 is programmed for any of the Enhanced PWM modes, the active output pins may be configured for auto-shutdown. Auto-shutdown immediately places the Enhanced PWM output pins into a defined shutdown state when a shutdown event occurs.

A shutdown event can be caused by either of the comparator modules, a low level on the Fault input pin (FLTO) or any combination of these three sources. The comparators may be used to monitor a voltage input proportional to a current being monitored in the bridge circuit. If the voltage exceeds a threshold, the comparator switches state and triggers a shutdown. Alternatively, a low digital signal on FLTO can also trigger a shutdown. The auto-shutdown feature can be disabled by not selecting any auto-shutdown sources. The autoshutdown sources to be used are selected using the ECCPAS2:ECCPAS0 bits (bits<6:4> of the ECCP1AS register).
When a shutdown occurs, the output pins are asynchronously placed in their shutdown states, specified by the PSSAC1:PSSAC0 and PSSBD1:PSSBD0 bits (ECCPAS3:ECCPAS0). Each pin pair (P1A/P1C and P1B/P1D) may be set to drive high, drive low or be tri-stated (not driving). The ECCPASE bit (ECCP1AS<7>) is also set to hold the Enhanced PWM outputs in their shutdown states.
The ECCPASE bit is set by hardware when a shutdown event occurs. If automatic restarts are not enabled, the ECCPASE bit is cleared by firmware when the cause of the shutdown clears. If automatic restarts are enabled, the ECCPASE bit is automatically cleared when the cause of the auto-shutdown has cleared.
If the ECCPASE bit is set when a PWM period begins, the PWM outputs remain in their shutdown state for that entire PWM period. When the ECCPASE bit is cleared, the PWM outputs will return to normal operation at the beginning of the next PWM period.

Note: Writing to the ECCPASE bit is disabled while a shutdown condition is active.

## REGISTER 14-2: ECCP1DEL: PWM DEAD-BAND DELAY REGISTER

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PRSEN | PDC6 $^{(1)}$ | PDC5 $^{(1)}$ | $\mathrm{PDC4}^{(1)}$ | $\mathrm{PDC3}^{(1)}$ | $\mathrm{PDC2}^{(1)}$ | $\mathrm{PDC1}^{(1)}$ | $\mathrm{PDC0}^{(1)}$ | bit 7 bit 0

bit 7 PRSEN: PWM Restart Enable bit 1 = Upon auto-shutdown, the ECCPASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically
$0=$ Upon auto-shutdown, ECCPASE must be cleared in software to restart the PWM
bit 6-0
PDC6:PDC0: PWM Delay Count bits ${ }^{(1)}$
Delay time, in number of Fosc/4 (4 * TOSC) cycles, between the scheduled and actual time for a PWM signal to transition to active.
Note 1: Reserved on 28-pin devices; maintain these bits clear.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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## REGISTER 14-3: ECCP1AS: ENHANCED CAPTURE/COMPARE/PWM AUTO-SHUTDOWN

 CONTROL REGISTER| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | ---: | ---: | ---: |
| ECCPASE | ECCPAS2 | ECCPAS1 | ECCPAS0 | PSSAC1 | PSSAC0 | PSSBD1 ${ }^{(1)}$ | PSSBD0 |
| bit 7 |  |  |  |  |  |  |  |
| bit 7 |  |  |  |  |  |  |  |

bit 7 ECCPASE: ECCP Auto-Shutdown Event Status bit
1 = A shutdown event has occurred; ECCP outputs are in shutdown state
$0=$ ECCP outputs are operating
bit 6-4 ECCPAS2:ECCPAS0: ECCP Auto-Shutdown Source Select bits
111 = FLTO, Comparator 1 or Comparator 2
$110=$ FLTO or Comparator 2
$101=$ FLTO or Comparator 1
$100=$ FLTO
011 = Either Comparator 1 or 2
010 = Comparator 2 output
001 = Comparator 1 output
$000=$ Auto-shutdown is disabled
bit 3-2 PSSAC1:PSSAC0: Pins A and C Shutdown State Control bits
$1 \mathrm{x}=$ Pins A and C are tri-state (40/44-pin devices);
PWM output is tri-state (28-pin devices)
01 = Drive Pins A and C to ' 1 '
$00=$ Drive Pins A and C to ' 0 '
bit 1-0 PSSBD1:PSSBD0: Pins B and D Shutdown State Control bits ${ }^{(1)}$
1x = Pins B and D tri-state
$01=$ Drive Pins B and D to ' 1 '
$00=$ Drive Pins B and D to ' 0 '
Note 1: Reserved on 28-pin devices; maintain these bits clear.

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 14.4.7.1 Auto-Shutdown and Automatic Restart

The auto-shutdown feature can be configured to allow automatic restarts of the module following a shutdown event. This is enabled by setting the PRSEN bit of the ECCP1DEL register (ECCP1DEL<7>).
In Shutdown mode with PRSEN = 1 (Figure 14-10), the ECCPASE bit will remain set for as long as the cause of the shutdown continues. When the shutdown condition clears, the ECCPASE bit is cleared. If PRSEN $=0$ (Figure 14-11), once a shutdown condition occurs, the ECCPASE bit will remain set until it is cleared by firmware. Once ECCPASE is cleared, the Enhanced PWM will resume at the beginning of the next PWM period.

> | Note: | Writing to the ECCPASE bit is disabled |
| :--- | :--- |
| while a shutdown condition is active. |  |

Independent of the PRSEN bit setting, if the autoshutdown source is one of the comparators, the shutdown condition is a level. The ECCPASE bit cannot be cleared as long as the cause of the shutdown persists.
The Auto-Shutdown mode can be forced by writing a ' 1 ' to the ECCPASE bit.

### 14.4.8 START-UP CONSIDERATIONS

When the ECCP module is used in the PWM mode, the application hardware must use the proper external pullup and/or pull-down resistors on the PWM output pins. When the microcontroller is released from Reset, all of the I/O pins are in the high-impedance state. The external circuits must keep the power switch devices in the OFF state until the microcontroller drives the I/O pins with the proper signal levels, or activates the PWM output(s).
The CCP1M1:CCP1M0 bits (CCP1CON<1:0>) allow the user to choose whether the PWM output signals are active-high or active-low for each pair of PWM output pins (P1A/P1C and P1B/P1D). The PWM output polarities must be selected before the PWM pins are configured as outputs. Changing the polarity configuration while the PWM pins are configured as outputs is not recommended, since it may result in damage to the application circuits.
The P1A, P1B, P1C and P1D output latches may not be in the proper states when the PWM module is initialized. Enabling the PWM pins for output at the same time as the ECCP module may cause damage to the application circuit. The ECCP module must be enabled in the proper output mode and complete a full PWM cycle before configuring the PWM pins as outputs. The completion of a full PWM cycle is indicated by the TMR2IF bit being set as the second PWM period begins.

FIGURE 14-10: PWM AUTO-SHUTDOWN (PRSEN = 1, AUTO-RESTART ENABLED)


FIGURE 14-11: PWM AUTO-SHUTDOWN (PRSEN = 0, AUTO-RESTART DISABLED)


### 14.4.9 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the ECCP module for PWM operation:

1. Configure the PWM pins, P1A and P1B (and P1C and P1D, if used), as inputs by setting the corresponding TRIS bits.
2. Set the PWM period by loading the PR2 register.
3. If auto-shutdown is required:

- Disable auto-shutdown $(E C C P A S E=0)$
- Configure source (FLTO, Comparator 1 or Comparator 2)
- Wait for non-shutdown condition

4. Configure the ECCP module for the desired PWM mode and configuration by loading the CCP1CON register with the appropriate values:

- Select one of the available output configurations and direction with the P1M1:P1M0 bits.
- Select the polarities of the PWM output signals with the CCP1M3:CCP1M0 bits.

5. Set the PWM duty cycle by loading the CCPR1L register and CCP1CON<5:4> bits.
6. For Half-Bridge Output mode, set the deadband delay by loading ECCP1DEL<6:0> with the appropriate value.
7. If auto-shutdown operation is required, load the ECCP1AS register:

- Select the auto-shutdown sources using the ECCPAS2:ECCPAS0 bits.
- Select the shutdown states of the PWM output pins using the PSSAC1:PSSAC0 and PSSBD1:PSSBD0 bits.
- Set the ECCPASE bit (ECCP1AS<7>).
- Configure the comparators using the CMCON register.
- Configure the comparator inputs as analog inputs.

8. If auto-restart operation is required, set the PRSEN bit (ECCP1DEL<7>).
9. Configure and start TMR2:

- Clear the TMR2 interrupt flag bit by clearing the TMR2IF bit (PIR1<1>).
- Set the TMR2 prescale value by loading the T2CKPS bits (T2CON<1:0>).
- Enable Timer2 by setting the TMR2ON bit ( $\mathrm{T} 2 \mathrm{CON}<2>$ ).

10. Enable PWM outputs after a new PWM cycle has started:

- Wait until TMRn overflows (TMRnIF bit is set).
- Enable the CCP1/P1A, P1B, P1C and/or P1D pin outputs by clearing the respective TRIS bits.
- Clear the ECCPASE bit (ECCP1AS<7>).


### 14.4.10 OPERATION IN POWER-MANAGED MODES

In Sleep mode, all clock sources are disabled. Timer2 will not increment and the state of the module will not change. If the CCP1 pin is driving a value, it will continue to drive that value. When the device wakes up, it will continue from this state. If Two-Speed Start-ups are enabled, the initial start-up frequency from INTOSC and the postscaler may not be stable immediately.
In PRI_IDLE mode, the primary clock will continue to clock the ECCP module without change. In all other power-managed modes, the selected power-managed mode clock will clock Timer2. Other power-managed mode clocks will most likely be different than the primary clock frequency.

### 14.4.10.1 Operation with Fail-Safe Clock Monitor

If the Fail-Safe Clock Monitor is enabled, a clock failure will force the device into the power-managed RC_RUN mode and the OSCFIF bit (PIR2<7>) will be set. The ECCP will then be clocked from the internal oscillator clock source, which may have a different clock frequency than the primary clock.
See the previous section for additional details.

### 14.4.11 EFFECTS OF A RESET

Both Power-on Reset and subsequent Resets will force all ports to Input mode and the CCP registers to their Reset states.
This forces the Enhanced CCP module to reset to a state compatible with the standard CCP module.

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TABLE 14-3: REGISTERS ASSOCIATED WITH ECCP1 MODULE AND TIMER1

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| RCON | IPEN | - | - | $\overline{\mathrm{RI}}$ | TO | $\overline{\mathrm{PD}}$ | $\overline{\text { POR }}$ | $\overline{\text { BOR }}$ | 42 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP ${ }^{(1)}$ | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 45 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 45 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 45 |
| TRISB | PORTB Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TRISC | PORTC Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TRISD ${ }^{(1)}$ | PORTD Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| TMR1L | Timer1 Register Low Byte |  |  |  |  |  |  |  | 44 |
| TMR1H | Timer1 Register High Byte |  |  |  |  |  |  |  | 44 |
| T1CON | RD16 | T1RUN | T1CKPS1 | T1CKPS0 | T1OSCEN | T1SYNC | TMR1CS | TMR1ON | 44 |
| TMR2 | Timer2 Register |  |  |  |  |  |  |  | 44 |
| T2CON | - | T2OUTPS3 | T2OUTPS2 | T2OUTPS1 | T2OUTPS0 | TMR2ON | T2CKPS1 | T2CKPS0 | 44 |
| PR2 | Timer2 Period Register |  |  |  |  |  |  |  | 44 |
| CCPR1L | Capture/Compare/PWM Register 1 Low Byte |  |  |  |  |  |  |  | 45 |
| CCPR1H | Capture/Compare/PWM Register 1 High Byte |  |  |  |  |  |  |  | 45 |
| CCP1CON | P1M1 ${ }^{(1)}$ | P1M0 ${ }^{(1)}$ | DC1B1 | DC1B0 | CCP1M3 | CCP1M2 | CCP1M1 | CCP1M0 | 45 |
| ECCP1AS | ECCPASE | ECCPAS2 | ECCPAS1 | ECCPASO | PSSAC1 | PSSAC0 | PSSBD1 ${ }^{(1)}$ | PSSBD0 ${ }^{(1)}$ | 45 |
| ECCP1DEL | PRSEN | PDC6 ${ }^{(1)}$ | PDC5 ${ }^{(1)}$ | PDC4 ${ }^{(1)}$ | PDC3 ${ }^{(1)}$ | PDC2 ${ }^{(1)}$ | PDC1 ${ }^{(1)}$ | PDC0 ${ }^{(1)}$ | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used during ECCP operation.
Note 1: These registers and/or bits are not implemented on 28 -pin devices and should be read as ' 0 '.

### 15.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

### 15.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface, useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPITM)
- Inter-Integrated Circuit $\left(I^{2} \mathrm{C}^{\top M}\right)$
- Full Master mode
- Slave mode (with general address call)

The $I^{2} \mathrm{C}$ interface supports the following modes in hardware:

- Master mode
- Multi-Master mode
- Slave mode

PIC18F24J10/25J10 (28-pin) devices have one MSSP module designated as MSSP1. PIC18F44J10/45J10 (40/44-pin) devices have two MSSP modules, designated as MSSP1 and MSSP2. Each module operates independently of the other.

Note: Throughout this section, generic references to an MSSP module in any of its operating modes may be interpreted as being equally applicable to MSSP1 or MSSP2. Register names and module I/O signals use the generic designator ' $x$ ' to indicate the use of a numeral to distinguish a particular module, when required. Control bit names are not individuated.

### 15.2 Control Registers

Each MSSP module has three associated control registers. These include a status register (SSPxSTAT) and two control registers (SSPxCON1 and SSPxCON2). The use of these registers and their individual configuration bits differ significantly depending on whether the MSSP module is operated in SPI or ${ }^{2} \mathrm{C}$ mode.
Additional details are provided under the individual sections.

Note: In devices with more than one MSSP module, it is very important to pay close attention to SSPCON register names. SSP1CON1 and SSP1CON2 control different operational aspects of the same module, while SSP1CON1 and SSP2CON1 control the same features for two different modules.

### 15.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

- Serial Data Out (SDOx) - RC5/SDO1 or RD2/PSP2/SDO2
- Serial Data In (SDIx) - RC4/SDI1/SDA1 or RD1/PSP1/SDI2/SDA2
- Serial Clock (SCKx) - RC3/SCK1/SCL1 or RD0/PSP0/SCK2/SCL2
Additionally, a fourth pin may be used when in a Slave mode of operation:
- Slave Select $(\overline{\mathrm{SSx}})$ - RA5/AN4/ $\overline{\mathrm{SS} 1} / \mathrm{C} 2 O U T$ or RD3/PSP3/SS2
Figure 15-1 shows the block diagram of the MSSP module when operating in SPI mode.

FIGURE 15-1: MSSP BLOCK DIAGRAM (SPITM MODE)


Note: Only port I/O names are used in this diagram for the sake of brevity. Refer to the text for a full list of multiplexed functions.

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### 15.3.1 REGISTERS

Each MSSP module has four registers for SPI mode operation. These are:

- MSSP Control Register 1 (SSPxCON1)
- MSSP Status Register (SSPxSTAT)
- Serial Receive/Transmit Buffer Register (SSPxBUF)
- MSSP Shift Register (SSPxSR) - Not directly accessible

SSPxCON1 and SSPxSTAT are the control and status registers in SPI mode operation. The SSPxCON1 register is readable and writable. The lower 6 bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

SSPxSR is the shift register used for shifting data in or out. SSPxBUF is the buffer register to which data bytes are written to or read from.
In receive operations, SSPxSR and SSPxBUF together create a double-buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not double-buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

## REGISTER 15-1: SSPxSTAT: MSSPx STATUS REGISTER (SPITM MODE)

| R/W-0 | R/W-0 | R-0 | R-0 | R-0 | R-0 | R-0 | R-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMP | CKE | D/ $\bar{A}$ | P | S | R/ $\bar{W}$ | UA | BF |
| bit 7 |  |  |  |  |  |  |  |

bit 7 SMP: Sample bit
SPI Master mode:
1 = Input data sampled at end of data output time
$0=$ Input data sampled at middle of data output time
SPI Slave mode:
SMP must be cleared when SPI is used in Slave mode.
bit 6 CKE: SPI Clock Select bit
$1=$ Transmit occurs on transition from active to Idle clock state
$0=$ Transmit occurs on transition from Idle to active clock state
Note: Polarity of clock state is set by the CKP bit (SSPxCON1<4>).
bit 5 D/A: Data/Address bit
Used in $I^{2} \mathrm{C}$ mode only.
bit $4 \quad \mathbf{P}$ : Stop bit
Used in $I^{2} \mathrm{C}$ mode only. This bit is cleared when the MSSP module is disabled, SSPEN is cleared.
bit $3 \quad$ S: Start bit Used in $I^{2} \mathrm{C}$ mode only.
bit $2 \quad \mathbf{R} / \overline{\mathbf{W}}$ : Read/ $\overline{\text { Write }}$ Information bit Used in $I^{2} \mathrm{C}$ mode only.
bit 1 UA: Update Address bit
Used in $I^{2} \mathrm{C}$ mode only.
bit $0 \quad$ BF: Buffer Full Status bit (Receive mode only)
1 = Receive complete, SSPxBUF is full
$0=$ Receive not complete, SSPxBUF is empty

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $\mathrm{W}=$ Writable bit | $\mathrm{U}=$ = Unimplemented bit, read as ' 0 ' |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 15-2: SSPxCON1: MSSPx CONTROL REGISTER 1 (SPITM MODE)

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 |

bit 7
bit 0
bit 7 WCOL: Write Collision Detect bit (Transmit mode only)
$1=$ The SSPxBUF register is written while it is still transmitting the previous word (must be cleared in software)
$0=$ No collision
bit 6 SSPOV: Receive Overflow Indicator bit
SPI Slave mode:
1 = A new byte is received while the SSPxBUF register is still holding the previous data. In case of overflow, the data in SSPxSR is lost. Overflow can only occur in Slave mode. The user must read the SSPxBUF, even if only transmitting data, to avoid setting overflow (must be cleared in software).
$0=$ No overflow
Note: In Master mode, the overflow bit is not set since each new reception (and transmission) is initiated by writing to the SSPxBUF register.
bit 5 SSPEN: Master Synchronous Serial Port Enable bit
1 = Enables serial port and configures SCKx, SDOx, SDIx and $\overline{\text { SSx }}$ as serial port pins
$0=$ Disables serial port and configures these pins as I/O port pins
Note: When enabled, these pins must be properly configured as input or output.
bit 4 CKP: Clock Polarity Select bit
1 = Idle state for clock is a high level
0 = Idle state for clock is a low level
bit 3-0 SSPM3:SSPM0: Master Synchronous Serial Port Mode Select bits
0101 = SPI Slave mode, clock = SCKx pin, $\overline{\text { SSx }}$ pin control disabled, $\overline{\text { SSx }}$ can be used as I/O pin
$0100=$ SPI Slave mode, clock $=$ SCKx pin, SSx pin control enabled
0011 = SPI Master mode, clock = TMR2 output/2
$0010=$ SPI Master mode, clock $=$ Fosc/64
0001 = SPI Master mode, clock $=$ Fosc/16
$0000=$ SPI Master mode, clock $=$ Fosc/4
Note: Bit combinations not specifically listed here are either reserved or implemented in $\mathrm{I}^{2} \mathrm{C}$ mode only.

## Legend:

| $\mathrm{R}=$ Readable bit | $\mathrm{W}=$ Writable bit | $\mathrm{U}=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' $=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 15.3.2 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPxCON1<5:0> and SSPxSTAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCKx is the clock output)
- Slave mode (SCKx is the clock input)
- Clock Polarity (Idle state of SCKx)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCKx)
- Clock Rate (Master mode only)
- Slave Select mode (Slave mode only)

Each MSSP consists of a transmit/receive shift register (SSPxSR) and a buffer register (SSPxBUF). The SSPxSR shifts the data in and out of the device, MSb first. The SSPxBUF holds the data that was written to the SSPxSR until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSPxBUF register. Then, the Buffer Full detect bit BF (SSPxSTAT<0>) and the interrupt flag bit SSPxIF are set. This double-buffering of the received data (SSPxBUF) allows the next byte to start reception before reading the data that was just received. Any write to the

SSPxBUF register during transmission/reception of data will be ignored and the Write Collision detect bit, WCOL (SSPxCON1<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPxBUF register completed successfully.
When the application software is expecting to receive valid data, the SSPxBUF should be read before the next byte of data to transfer is written to the SSPxBUF. The Buffer Full bit, BF (SSPxSTAT<0>), indicates when SSPxBUF has been loaded with the received data (transmission is complete). When the SSPxBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. The SSPxBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 15-1 shows the loading of the SSP1BUF (SSP1SR) for data transmission.
The SSPxSR is not directly readable or writable and can only be accessed by addressing the SSPxBUF register. Additionally, the SSPxSTAT register indicates the various status conditions.

## EXAMPLE 15-1: LOADING THE SSP1BUF (SSP1SR) REGISTER

| LOOP | BTFSS | SSP1STAT, BF | ; Has data been received (transmit complete)? |
| :--- | :--- | :--- | :--- |
|  | BRA | LOOP | ;No |
|  | MOVF | SSP1BUF, W | ;WREG reg = contents of SSP1BUF |
|  | MOVWF | RXDATA | ;Save in user RAM, if data is meaningful |
|  | MOVF | TXDATA, W | ;W reg = contents of TXDATA |
|  | MOVWF New data to xmit |  |  |

### 15.3.3 ENABLING SPI I/O

To enable the serial port, MSSP Enable bit, SSPEN (SSPxCON1<5>), must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, reinitialize the SSPxCON registers and then set the SSPEN bit. This configures the SDIx, SDOx, SCKx and $\overline{S S x}$ pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDIx is automatically controlled by the SPI module
- SDOx must have TRISC<5> (or TRISD<2>) bit cleared
- SCKx (Master mode) must have TRISC<3> (or TRISD<0>) bit cleared
- SCKx (Slave mode) must have TRISC<3> (or TRISD<0>) bit set
- $\overline{\mathrm{SSx}}$ must have TRISA<5> (or TRISD<3>) bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

### 15.3.4 TYPICAL CONNECTION

Figure 15-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCKx signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- Master sends data - Slave sends dummy data
- Master sends data - Slave sends data
- Master sends dummy data - Slave sends data

FIGURE 15-2: SPITM MASTER/SLAVE CONNECTION


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### 15.3.5 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCKx. The master determines when the slave (Processor 2, Figure 15-2) will broadcast data by the software protocol.
In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDOx output could be disabled (programmed as an input). The SSPxSR register will continue to shift in the signal present on the SDIx pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode.

The clock polarity is selected by appropriately programming the CKP bit (SSPxCON1<4>). This then, would give waveforms for SPI communication as shown in Figure 15-3, Figure 15-5 and Figure 15-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user-programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 • Tcy)
- Fosc/64 (or 16 • Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz ) of 10.00 Mbps .

Figure 15-3 shows the waveforms for Master mode. When the CKE bit is set, the SDOx data is valid before there is a clock edge on SCKx. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.

FIGURE 15-3:

## SPI ${ }^{\text {™ }}$ MODE WAVEFORM (MASTER MODE)



### 15.3.6 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCKx. When the last bit is latched, the SSPxIF interrupt flag bit is set.
Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCKx pin. The Idle state is determined by the CKP bit (SSPxCON1<4>).
While in Slave mode, the external clock is supplied by the external clock source on the SCKx pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.
While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from Sleep.

### 15.3.7 SLAVE SELECT SYNCHRONIZATION

The $\overline{\mathrm{SSx}}$ pin allows a Synchronous Slave mode. The SPI must be in Slave mode with $\overline{S S x}$ pin control enabled (SSPxCON1<3:0> = 04h). When the SSx pin is low, transmission and reception are enabled and the

SDOx pin is driven. When the $\overline{S S x}$ pin goes high, the SDOx pin is no longer driven, even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

Note 1: When the SPI is in Slave mode with $\overline{\mathrm{SSx}}$ pin control enabled (SSPxCON1<3:0> = 0100), the SPI module will reset if the $\overline{\mathrm{SSx}}$ pin is set to VDD.

2: If the SPI is used in Slave mode with CKE set, then the $\overline{\mathrm{SSx}}$ pin control must be enabled.

When the SPI module resets, the bit counter is forced to ' 0 '. This can be done by either forcing the $\overline{\mathrm{SSx}}$ pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDOx pin can be connected to the SDIx pin. When the SPI needs to operate as a receiver, the SDOx pin can be configured as an input. This disables transmissions from the SDOx. The SDIx can always be left as an input (SDIx function) since it cannot create a bus conflict.

FIGURE 15-4: SLAVE SYNCHRONIZATION WAVEFORM


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FIGURE 15-5: $\quad$ SPITM MODE WAVEFORM (SLAVE MODE WITH CKE = 0)


FIGURE 15-6: $\quad$ SPITM MODE WAVEFORM (SLAVE MODE WITH CKE = 1)


### 15.3.8 OPERATION IN POWER-MANAGED MODES

In SPI Master mode, module clocks may be operating at a different speed than when in full power mode; in the case of Sleep mode, all clocks are halted.

In Idle modes, a clock is provided to the peripherals. That clock should be from the primary clock source, the secondary clock (Timer1 oscillator at 32.768 kHz ) or the INTOSC source. See Section 2.6 "Clock Sources and Oscillator Switching" for additional information.
In most cases, the speed that the master clocks SPI data is not important; however, this should be evaluated for each system.
If MSSP interrupts are enabled, they can wake the controller from Sleep mode, or one of the Idle modes, when the master completes sending data. If an exit from Sleep or Idle mode is not desired, MSSP interrupts should be disabled.
If the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the devices wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.
In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in any power-managed mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device.

### 15.3.9 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

### 15.3.10 BUS MODE COMPATIBILITY

Table 15-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

## TABLE 15-1: $\quad$ SPITM ${ }^{\text {TM }}$ BUS MODES

| Standard SPITM Mode <br> Terminology | Control Bits State |  |
| :---: | :---: | :---: |
|  | CKP | CKE |
| 0,0 | 0 | 1 |
| 0,1 | 0 | 0 |
| 1,0 | 1 | 1 |
| 1,1 | 1 | 0 |

There is also an SMP bit which controls when the data is sampled.

### 15.3.11 SPI CLOCK SPEED AND MODULE INTERACTIONS

Because MSSP1 and MSSP2 are independent modules, they can operate simultaneously at different data rates. Setting the SSPM3:SSPM0 bits of the SSPxCON1 register determines the rate for the corresponding module.
An exception is when both modules use Timer2 as a time base in Master mode. In this instance, any changes to the Timer2 operation will affect both MSSP modules equally. If different bit rates are required for each module, the user should select one of the other three time base options for one of the modules.

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TABLE 15-2: REGISTERS ASSOCIATED WITH SPITM OPERATION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| PIR3 | SSP2IF | BCL2IF | - | - | - | - | - | - | 45 |
| PIE3 | SSP2IE | BCL2IE | - | - | - | - | - | - | 45 |
| IPR3 | SSP2IP | BCL2IP | - | - | - | - | - | - | 45 |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISAO | 46 |
| TRISC | TRISC7 | TRISC6 | TRISC5 | TRISC4 | TRISC3 | TRISC2 | TRISC1 | TRISC0 | 46 |
| TRISD ${ }^{(1)}$ | TRISD7 | TRISD6 | TRISD5 | TRISD4 | TRISD3 | TRISD2 | TRISD1 | TRISD0 | 46 |
| SSP1BUF | MSSP1 Receive Buffer/Transmit Register |  |  |  |  |  |  |  | 44 |
| SSP1CON1 | WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 | 44 |
| SSP1STAT | SMP | CKE | D/ $\bar{A}$ | P | S | R/W | UA | BF | 44 |
| SSP2BUF | MSSP2 Receive Buffer/Transmit Register |  |  |  |  |  |  |  | 46 |
| SSP2CON1 | WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 | 46 |
| SSP2STAT | SMP | CKE | D/ $\bar{A}$ | P | S | R/W | UA | BF | 46 |

Legend: Shaded cells are not used by the MSSP module in SPI ${ }^{T M}$ mode.
Note 1: These registers and/or bits are not implemented on 28 -pin devices and should be read as ' 0 '.

## $15.4 \quad I^{2} \mathrm{C}$ Mode

The MSSP module in $I^{2} \mathrm{C}$ mode fully implements all master and slave functions (including general call support) and provides interrupts on Start and Stop bits in hardware to determine a free bus (multi-master function). The MSSP module implements the standard mode specifications as well as 7-bit and 10-bit addressing.
Two pins are used for data transfer:

- Serial clock (SCLx) - RC3/SCK1/SCL1 or RD0/PSP0/SCK2/SCL2
- Serial data (SDAx) - RC4/SDI1/SDA1 or RD1/PSP1/SDI2/SDA2

The user must configure these pins as inputs by setting the TRISC<4:3> or TRISD<1:0> bits.

FIGURE 15-7: MSSP BLOCK DIAGRAM ( $1^{2} \mathrm{C}^{\text {TM }}$ MODE)


Note: Only port I/O names are used in this diagram for the sake of brevity. Refer to the text for a full list of multiplexed functions.

### 15.4.1 REGISTERS

The MSSP module has six registers for $\mathrm{I}^{2} \mathrm{C}$ operation. These are:

- MSSP Control Register 1 (SSPxCON1)
- MSSP Control Register 2 (SSPxCON2)
- MSSP Status Register (SSPxSTAT)
- Serial Receive/Transmit Buffer Register (SSPxBUF)
- MSSP Shift Register (SSPxSR) - Not directly accessible
- MSSP Address Register (SSPxADD)

SSPxCON1, SSPxCON2 and SSPxSTAT are the control and status registers in $\mathrm{I}^{2} \mathrm{C}$ mode operation. The SSPxCON1 and SSPxCON2 registers are readable and writable. The lower 6 bits of the SSPxSTAT are read-only. The upper 2 bits of the SSPxSTAT are read/write.
SSPxSR is the shift register used for shifting data in or out. SSPxBUF is the buffer register to which data bytes are written to or read from.
SSPxADD register holds the slave device address when the MSSP is configured in $I^{2} \mathrm{C}$ Slave mode. When the MSSP is configured in Master mode, the lower seven bits of SSPxADD act as the Baud Rate Generator reload value.
In receive operations, SSPxSR and SSPxBUF together create a double-buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.
During transmission, the SSPxBUF is not double-buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

## REGISTER 15-3: SSPxSTAT: MSSPx STATUS REGISTER ( $\mathbf{I}^{2} \mathrm{C}^{\text {TM }}$ MODE)

| R/W-0 | R/W-0 | R-0 | R-0 | R-0 | R-0 | R-0 | R-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMP | CKE | D/ $/ \bar{A}$ | P | S | $R / \bar{W}$ | UA | BF |
| bit 7 |  |  |  |  |  |  |  |

bit 7 SMP: Slew Rate Control bit
In Master or Slave mode:
1 = Slew rate control disabled for Standard Speed mode ( 100 kHz and 1 MHz )
0 = Slew rate control enabled for High-Speed mode ( 400 kHz )
bit 6 CKE: SMBus Select bit
In Master or Slave mode:
1 = Enable SMBus specific inputs
$0=$ Disable SMBus specific inputs
bit 5 D/A: Data/Address bit
In Master mode:
Reserved.
In Slave mode:
1 = Indicates that the last byte received or transmitted was data
$0=$ Indicates that the last byte received or transmitted was address
bit $4 \quad$ P: Stop bit
1 = Indicates that a Stop bit has been detected last
$0=$ Stop bit was not detected last
Note: This bit is cleared on Reset and when SSPEN is cleared.
bit $3 \quad$ S: Start bit
1 = Indicates that a Start bit has been detected last
$0=$ Start bit was not detected last
Note: This bit is cleared on Reset and when SSPEN is cleared.
bit $2 \quad \mathbf{R} / \overline{\mathbf{W}}$ : Read $/ \overline{\text { Write }}$ Information bit ( ${ }^{2} \mathrm{C}$ mode only)
In Slave mode:
1 = Read
$0=$ Write
Note: This bit holds the $R / \bar{W}$ bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit or not $\overline{\text { ACK }}$ bit.
In Master mode:
1 = Transmit is in progress
$0=$ Transmit is not in progress
Note: ORing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSP is in Active mode.
bit 1 UA: Update Address bit (10-bit Slave mode only)
1 = Indicates that the user needs to update the address in the SSPxADD register
$0=$ Address does not need to be updated
bit $0 \quad$ BF: Buffer Full Status bit
In Transmit mode:
1 = SSPxBUF is full
$0=$ SSPxBUF is empty
In Receive mode:
1 = SSPxBUF is full (does not include the $\overline{\mathrm{ACK}}$ and Stop bits)
$0=$ SSPxBUF is empty (does not include the $\overline{A C K}$ and Stop bits)

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 15-4: SSPxCON1: MSSPx CONTROL REGISTER 1 ( $\mathbf{I}^{2} C^{\text {TM }}$ MODE)

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 |

bit 7
bit 0
bit 7 WCOL: Write Collision Detect bit
In Master Transmit mode:
1 = A write to the SSPxBUF register was attempted while the $I^{2} \mathrm{C}$ conditions were not valid for a transmission to be started (must be cleared in software)
$0=$ No collision
In Slave Transmit mode:
1 = The SSPxBUF register is written while it is still transmitting the previous word (must be cleared in software)
$0=$ No collision
In Receive mode (Master or Slave modes):
This is a "don't care" bit.
bit 6 SSPOV: Receive Overflow Indicator bit
In Receive mode:
$1=\mathrm{A}$ byte is received while the SSPxBUF register is still holding the previous byte (must be cleared in software)
$0=$ No overflow
In Transmit mode:
This is a "don't care" bit in Transmit mode.
bit 5 SSPEN: Master Synchronous Serial Port Enable bit
1 = Enables the serial port and configures the SDAx and SCLx pins as the serial port pins
$0=$ Disables serial port and configures these pins as I/O port pins
Note: When enabled, the SDAx and SCLx pins must be properly configured as input or output.
bit 4 CKP: SCKx Release Control bit
In Slave mode:
1 = Release clock
$0=$ Holds clock low (clock stretch), used to ensure data setup time
In Master mode:
Unused in this mode.
bit 3-0 SSPM3:SSPM0: Master Synchronous Serial Port Mode Select bits
$1111=I^{2} \mathrm{C}$ Slave mode, 10 -bit address with Start and Stop bit interrupts enabled
$1110=1^{2} \mathrm{C}$ Slave mode, 7 -bit address with Start and Stop bit interrupts enabled
$1011=I^{2} \mathrm{C}$ Firmware Controlled Master mode (Slave Idle)
$1000=I^{2} \mathrm{C}$ Master mode, clock $=\mathrm{Fosc} /(4$ * (SSPxADD +1$)$ )
$0111=I^{2} \mathrm{C}$ Slave mode, 10 -bit address
$0110=I^{2} \mathrm{C}$ Slave mode, 7 -bit address
Bit combinations not specifically listed here are either reserved or implemented in SPI mode only.

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

## REGISTER 15-5: SSPxCON2: MSSPx CONTROL REGISTER 2 ( $\mathbf{I}^{2} \mathbf{C}^{\text {TM }}$ MODE)

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| GCEN | ACKSTAT | ACKDT | ACKEN $^{(1)}$ | RCEN $^{(1)}$ | PEN $^{(1)}$ | RSEN $^{(1)}$ | SEN $^{(1)}$ |
| bit 7 |  |  |  |  |  |  |  |

bit 7 GCEN: General Call Enable bit (Slave mode only)
1 = Enable interrupt when a general call address (0000h) is received in the SSPxSR
$0=$ General call address disabled
bit 6 ACKSTAT: Acknowledge Status bit (Master Transmit mode only)
1 = Acknowledge was not received from slave
$0=$ Acknowledge was received from slave
bit 5 ACKDT: Acknowledge Data bit (Master Receive mode only)
1 = Not Acknowledge
0 = Acknowledge
Note: Value that will be transmitted when the user initiates an Acknowledge sequence at the end of a receive.
bit 4 ACKEN: Acknowledge Sequence Enable bit (Master Receive mode only) ${ }^{(1)}$
$1=$ Initiate Acknowledge sequence on SDAx and SCLx pins and transmit ACKDT data bit. Automatically cleared by hardware.
$0=$ Acknowledge sequence Idle
bit $3 \quad$ RCEN: Receive Enable bit (Master mode only) ${ }^{(\mathbf{1})}$
1 = Enables Receive mode for $I^{2} \mathrm{C}$
0 = Receive Idle
bit 2 PEN: Stop Condition Enable bit (Master mode only) ${ }^{(\mathbf{1 )}}$
1 = Initiate Stop condition on SDAx and SCLx pins. Automatically cleared by hardware.
0 = Stop condition Idle
bit 1 RSEN: Repeated Start Condition Enable bit (Master mode only) ${ }^{(1)}$
$1=$ Initiate Repeated Start condition on SDAx and SCLx pins. Automatically cleared by hardware.
$0=$ Repeated Start condition Idle
bit 0 SEN: Start Condition Enable/Stretch Enable bit ${ }^{(1)}$
In Master mode:
1 = Initiate Start condition on SDAx and SCLx pins. Automatically cleared by hardware.
$0=$ Start condition Idle
In Slave mode:
1 = Clock stretching is enabled for both slave transmit and slave receive (stretch enabled)
$0=$ Clock stretching is disabled
Note 1: For bits ACKEN, RCEN, PEN, RSEN, SEN: If the ${ }^{2}$ C module is not in the Idle mode, these bits may not be set (no spooling) and the SSPxBUF may not be written (or writes to the SSPxBUF are disabled).

## Legend:

| $\mathrm{R}=$ Readable bit | $\mathrm{W}=$ Writable bit | $\mathrm{U}=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

### 15.4.2 OPERATION

The MSSP module functions are enabled by setting the MSSP Enable bit, SSPEN (SSPxCON1<5>).
The SSPxCON1 register allows control of the $I^{2} C$ operation. Four mode selection bits (SSPxCON1<3:0>) allow one of the following $I^{2} C$ modes to be selected:

- $\mathrm{I}^{2} \mathrm{C}$ Master mode,
clock $=($ Fosc/4) $\times($ SSPxADD +1$)$
- $I^{2} \mathrm{C}$ Slave mode (7-bit address)
- $I^{2} \mathrm{C}$ Slave mode (10-bit address)
- $\left.\right|^{2} \mathrm{C}$ Slave mode (7-bit address) with Start and Stop bit interrupts enabled
- $\mathrm{I}^{2} \mathrm{C}$ Slave mode (10-bit address) with Start and Stop bit interrupts enabled
- $I^{2} \mathrm{C}$ Firmware Controlled Master mode, slave is Idle
Selection of any $\mathrm{I}^{2} \mathrm{C}$ mode, with the SSPEN bit set, forces the SCLx and SDAx pins to be open-drain, provided these pins are programmed to inputs by setting the appropriate TRISC or TRISD bits. To ensure proper operation of the module, pull-up resistors must be provided externally to the SCLx and SDAx pins.


### 15.4.3 SLAVE MODE

In Slave mode, the SCLx and SDAx pins must be configured as inputs (TRISC $<4: 3>$ or TRISD $<1: 0>$ set). The MSSP module will override the input state with the output data when required (slave-transmitter).
The $I^{2} \mathrm{C}$ Slave mode hardware will always generate an interrupt on an address match. Through the mode select bits, the user can also choose to interrupt on Start and Stop bits
When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge ( $\overline{\mathrm{ACK}}$ ) pulse and load the SSPxBUF register with the received value currently in the SSPxSR register.
Any combination of the following conditions will cause the MSSP module not to give this $\overline{\mathrm{ACK}}$ pulse:

- The Buffer Full bit, BF (SSPxSTAT<0>), was set before the transfer was received.
- The overflow bit, SSPOV (SSPxCON1<6>), was set before the transfer was received.
In this case, the SSPxSR register value is not loaded into the SSPxBUF, but bit SSPxIF is set. The BF bit is cleared by reading the SSPxBUF register, while bit SSPOV is cleared through software.
The SCLx clock input must have a minimum high and low for proper operation. The high and low times of the $I^{2} \mathrm{C}$ specification, as well as the requirement of the MSSP module, are shown in timing parameter 100 and parameter 101.


### 15.4.3.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPxSR register. All incoming bits are sampled with the rising edge of the clock (SCLx) line. The value of register SSPxSR<7:1> is compared to the value of the SSPxADD register. The address is compared on the falling edge of the eighth clock (SCLx) pulse. If the addresses match and the BF and SSPOV bits are clear, the following events occur:

1. The SSPxSR register value is loaded into the SSPxBUF register.
2. The Buffer Full bit, BF, is set.
3. An $\overline{\mathrm{ACK}}$ pulse is generated.
4. The MSSP Interrupt Flag bit, SSPxIF, is set (and interrupt is generated, if enabled) on the falling edge of the ninth SCLx pulse.
In 10-bit Address mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. Bit R/W (SSPxSTAT<2>) must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '11110 A9 A8 0', where ' $A 9$ ' and 'A8' are the two MSbs of the address. The sequence of events for 10 -bit address is as follows, with steps 7 through 9 for the slave-transmitter:
5. Receive first (high) byte of address (bits SSPxIF, BF and UA (SSPxSTAT<1>) are set).
6. Update the SSPxADD register with second (low) byte of address (clears bit UA and releases the SCLx line).
7. Read the SSPxBUF register (clears bit BF) and clear flag bit SSPxIF.
8. Receive second (low) byte of address (bits SSPxIF, BF and UA are set).
9. Update the SSPxADD register with the first (high) byte of address. If match releases SCLx line, this will clear bit UA.
10. Read the SSPxBUF register (clears bit BF) and clear flag bit SSPxIF.
11. Receive Repeated Start condition.
12. Receive first (high) byte of address (bits SSPxIF and $B F$ are set).
13. Read the SSPxBUF register (clears bit BF) and clear flag bit SSPxIF.

### 15.4.3.2 Reception

When the $\mathrm{R} / \overline{\mathrm{W}}$ bit of the address byte is clear and an address match occurs, the R/W bit of the SSPxSTAT register is cleared. The received address is loaded into the SSPxBUF register and the SDAx line is held low ( $\overline{\mathrm{ACK}}$ ).
When the address byte overflow condition exists, then the no Acknowledge ( $\overline{\mathrm{ACK}}$ ) pulse is given. An overflow condition is defined as either bit BF (SSPxSTAT $<0>$ ) is set, or bit SSPOV (SSPxCON1<6>) is set.
An MSSP interrupt is generated for each data transfer byte. The interrupt flag bit, SSPxIF, must be cleared in software. The SSPxSTAT register is used to determine the status of the byte.
If SEN is enabled (SSPxCON2<0> = 1), SCKx/SCLx (RC3 or RD0) will be held low (clock stretch) following each data transfer. The clock must be released by setting bit, CKP (SSPxCON1<4>). See Section 15.4.4 "Clock Stretching" for more details.

### 15.4.3.3 Transmission

When the $\mathrm{R} / \overline{\mathrm{W}}$ bit of the incoming address byte is set and an address match occurs, the $R / \bar{W}$ bit of the SSPxSTAT register is set. The received address is loaded into the SSPxBUF register. The ACK pulse will be sent on the ninth bit and pin RC3 or RD6 is held low, regardless of SEN (see Section 15.4.4 "Clock Stretching" for more details). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data. The transmit data must be loaded into the SSPxBUF register which also loads the SSPxSR register. Then pin RC3 or RDO should be enabled by setting bit, CKP (SSPxCON1<4>). The eight data bits are shifted out on the falling edge of the SCLx input. This ensures that the SDAx signal is valid during the SCLx high time (Figure 15-9).
The $\overline{\text { ACK }}$ pulse from the master-receiver is latched on the rising edge of the ninth SCLx input pulse. If the SDAx line is high (not $\overline{A C K}$ ), then the data transfer is complete. In this case, when the $\overline{\mathrm{ACK}}$ is latched by the slave, the slave logic is reset (resets SSPxSTAT register) and the slave monitors for another occurrence of the Start bit. If the SDAx line was low ( $\overline{\mathrm{ACK}}$ ), the next transmit data must be loaded into the SSPxBUF register. Again, pin RC3 or RD0 must be enabled by setting bit CKP.
An MSSP interrupt is generated for each data transfer byte. The SSPxIF bit must be cleared in software and the SSPxSTAT register is used to determine the status of the byte. The SSPxIF bit is set on the falling edge of the ninth clock pulse.

FIGURE 15-8: $\quad I^{2} C^{T M}$ SLAVE MODE TIMING WITH SEN $=0$ (RECEPTION, 7-BIT ADDRESS)


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FIGURE 15-9: $\quad I^{2} C^{\text {TM }}$ SLAVE MODE TIMING (TRANSMISSION, 7-BIT ADDRESS)


FIGURE 15-10: $\quad I^{2} C^{\text {TM }}$ SLAVE MODE TIMING WITH SEN $=0$ (RECEPTION, 10-BIT ADDRESS)

|  |  |  |
| :---: | :---: | :---: |
|  |  |  |
|  |  |  |
|  |  |  |
|  |  |  |
|  |  |  |

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FIGURE 15-11: $\quad I^{2} C^{\text {TM }}$ SLAVE MODE TIMING (TRANSMISSION, 10-BIT ADDRESS)


### 15.4.4 CLOCK STRETCHING

Both 7 -bit and 10 -bit Slave modes implement automatic clock stretching during a transmit sequence.
The SEN bit (SSPxCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCLx pin to be held low at the end of each data receive sequence.

### 15.4.4.1 Clock Stretching for 7-bit Slave Receive Mode (SEN = 1)

In 7-bit Slave Receive mode, on the falling edge of the ninth clock at the end of the $\overline{A C K}$ sequence, if the BF bit is set, the CKP bit in the SSPxCON1 register is automatically cleared, forcing the SCLx output to be held low. The CKP being cleared to ' 0 ' will assert the SCLx line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and read the contents of the SSPxBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 15-13).

Note 1: If the user reads the contents of the SSPxBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

### 15.4.4.2 Clock Stretching for 10-bit Slave Receive Mode (SEN = 1)

In 10-bit Slave Receive mode during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10 -bit address and following the receive of the second byte of the 10 -bit address with the $\mathrm{R} / \overline{\mathrm{W}}$ bit cleared to ' 0 '. The release of the clock line occurs upon updating SSPxADD. Clock stretching will occur on each data receive sequence as described in 7 -bit mode.
Note: If the user polls the UA bit and clears it by updating the SSPxADD register before the falling edge of the ninth clock occurs and if the user hasn't cleared the BF bit by reading the SSPxBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching on the basis of the state of the BF bit only occurs during a data sequence, not an address sequence.

### 15.4.4.3 Clock Stretching for 7-bit Slave Transmit Mode

The 7 -bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock, if the BF bit is clear. This occurs regardless of the state of the SEN bit.
The user's ISR must set the CKP bit before transmission is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and load the contents of the SSPxBUF before the master device can initiate another transmit sequence (see Figure 15-9).

Note 1: If the user loads the contents of SSPxBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
2: The CKP bit can be set in software regardless of the state of the $B F$ bit.

### 15.4.4.4 Clock Stretching for 10-bit Slave Transmit Mode

In 10-bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10 -bit Slave Receive mode. The first two addresses are followed by a third address sequence which contains the high-order bits of the 10 -bit address and the $R / \bar{W}$ bit set to ' 1 '. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7 -bit Slave Transmit mode (see Figure 15-11).

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### 15.4.4.5 Clock Synchronization and the CKP bit

When the CKP bit is cleared, the SCLx output is forced to ' 0 '. However, clearing the CKP bit will not assert the SCLx output low until the SCLx output is already sampled low. Therefore, the CKP bit will not assert the SCLx line until an external $\mathrm{I}^{2} \mathrm{C}$ master device has
already asserted the SCLx line. The SCLx output will remain low until the CKP bit is set and all other devices on the $I^{2} \mathrm{C}$ bus have deasserted SCLx. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCLx (see Figure 15-12).

FIGURE 15-12: CLOCK SYNCHRONIZATION TIMING


FIGURE 15-13: $\quad I^{2} \mathbf{C}^{\text {TM }}$ SLAVE MODE TIMING WITH SEN $=1$ (RECEPTION, 7-BIT ADDRESS)


FIGURE 15-14: $\quad I^{2} C^{\text {TM }}$ SLAVE MODE TIMING WITH SEN $=1$ (RECEPTION, 10-BIT ADDRESS)


### 15.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the $\mathrm{I}^{2} \mathrm{C}$ bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.
The general call address is one of eight addresses reserved for specific purposes by the $\mathrm{I}^{2} \mathrm{C}$ protocol. It consists of all ' 0 's with $\mathrm{R} / \overline{\mathrm{W}}=0$.
The general call address is recognized when the General Call Enable bit, GCEN, is enabled (SSPxCON2<7> set). Following a Start bit detect, 8 bits are shifted into the SSPxSR and the address is compared against the SSPxADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPxSR is transferred to the SSPxBUF, the BF flag bit is set (eighth bit) and on the falling edge of the ninth bit ( $\overline{\mathrm{ACK}}$ bit), the SSPxIF interrupt flag bit is set.
When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPxBUF. The value can be used to determine if the address was device specific or a general call address.
In 10-bit mode, the SSPxADD is required to be updated for the second half of the address to match and the UA bit is set (SSPxSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-bit Address mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 15-15).

FIGURE 15-15: SLAVE MODE GENERAL CALL ADDRESS SEQUENCE (7 OR 10-BIT ADDRESS MODE)


### 15.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPxCON1 and by setting the SSPEN bit. In Master mode, the SCLx and SDAx lines are manipulated by the MSSP hardware.
Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop ( P ) and Start ( S ) bits are cleared from a Reset or when the MSSP module is disabled. Control of the $I^{2} \mathrm{C}$ bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.
In Firmware Controlled Master mode, user code conducts all $\mathrm{I}^{2} \mathrm{C}$ bus operations based on Start and Stop bit conditions.
Once Master mode is enabled, the user has six options.

1. Assert a Start condition on SDAx and SCLx.
2. Assert a Repeated Start condition on SDAx and SCLx.
3. Write to the SSPxBUF register initiating transmission of data/address.
4. Configure the $\mathrm{I}^{2} \mathrm{C}$ port to receive data.
5. Generate an Acknowledge condition at the end of a received byte of data.
6. Generate a Stop condition on SDAx and SCLx.

Note: The MSSP module, when configured in $I^{2} \mathrm{C}$ Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPxBUF register to initiate transmission before the Start condition is complete. In this case, the SSPxBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPxBUF did not occur.

The following events will cause the MSSP Interrupt Flag bit, SSPxIF, to be set (and MSSP interrupt, if enabled):

- Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmit
- Repeated Start

FIGURE 15-16: MSSP BLOCK DIAGRAM ( ${ }^{2} \mathbf{C}^{\text {TM }}$ MASTER MODE)


### 15.4.6.1 $\quad I^{2} C$ Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the $\mathrm{I}^{2} \mathrm{C}$ bus will not be released.
In Master Transmitter mode, serial data is output through SDAx, while SCLx outputs the serial clock. The first byte transmitted contains the slave address of the receiving device ( 7 bits) and the Read/Write $(R / \bar{W})$ bit. In this case, the R/W bit will be logic ' 0 '. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.
In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the $\mathrm{R} / \overline{\mathrm{W}}$ bit will be logic ' 1 '. Thus, the first byte transmitted is a 7 -bit slave address followed by a ' 1 ' to indicate the receive bit. Serial data is received via SDAx, while SCLx outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.
The Baud Rate Generator used for the SPI mode operation is used to set the SCLx clock frequency for either $100 \mathrm{kHz}, 400 \mathrm{kHz}$ or $1 \mathrm{MHz} \mathrm{I}^{2} \mathrm{C}$ operation. See Section 15.4.7 "Baud Rate" for more detail.

A typical transmit sequence would go as follows:

1. The user generates a Start condition by setting the Start Enable bit, SEN (SSPxCON2<0>).
2. SSPxIF is set. The MSSP module will wait the required start time before any other operation takes place.
3. The user loads the SSPxBUF with the slave address to transmit.
4. Address is shifted out the SDAx pin until all 8 bits are transmitted.
5. The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPxCON2 register (SSPxCON2<6>).
6. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
7. The user loads the SSPxBUF with eight bits of data.
8. Data is shifted out the SDAx pin until all 8 bits are transmitted.
9. The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPxCON2 register (SSPxCON2<6>).
10. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
11. The user generates a Stop condition by setting the Stop Enable bit, PEN (SSPxCON2<2>).
12. Interrupt is generated once the Stop condition is complete.

### 15.4.7 BAUD RATE

In $I^{2} \mathrm{C}$ Master mode, the Baud Rate Generator (BRG) reload value is placed in the lower 7 bits of the SSPxADD register (Figure 15-17). When a write occurs to SSPxBUF, the Baud Rate Generator will automatically begin counting. The BRG counts down to ' 0 ' and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (TCY) on the Q2 and Q4 clocks. In ${ }^{2}$ C Master mode, the BRG is reloaded automatically.
Once the given operation is complete (i.e., transmission of the last data bit is followed by $\overline{\mathrm{ACK}})$, the internal clock will automatically stop counting and the SCLx pin will remain in its last state.

Table 15-3 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPxADD.

### 15.4.7.1 Baud Rate and Module Interdependence

Because MSSP1 and MSSP2 are independent, they can operate simultaneously in $\mathrm{I}^{2} \mathrm{C}$ Master mode at different baud rates. This is done by using different $B R G$ reload values for each module.

Because this mode derives its basic clock source from the system clock, any changes to the clock will affect both modules in the same proportion. It may be possible to change one or both baud rates back to a previous value by changing the BRG reload value.

FIGURE 15-17: BAUD RATE GENERATOR BLOCK DIAGRAM


TABLE 15-3: $\quad \mathrm{I}^{2} \mathrm{C}^{\text {тм }}$ CLOCK RATE w/BRG

| Fcy | FCY *2 | BRG Value | FscL <br> (2 Rollovers of BRG) |
| :---: | :---: | :---: | :---: |
| 10 MHz | 20 MHz | 18 h | $400 \mathrm{kHz}{ }^{(\mathbf{1})}$ |
| 10 MHz | 20 MHz | 1 Fh | 312.5 kHz |
| 10 MHz | 20 MHz | 63 h | 100 kHz |
| 4 MHz | 8 MHz | 09 h | $400 \mathrm{kHz}{ }^{(\mathbf{1 )}}$ |
| 4 MHz | 8 MHz | 0 h | 308 kHz |
| 4 MHz | 8 MHz | 27 h | 100 kHz |
| 1 MHz | 2 MHz | 02 h | $333 \mathrm{kHz}{ }^{(\mathbf{1 )}}$ |
| 1 MHz | 2 MHz | 09 h | 100 kHz |
| 1 MHz | 2 MHz | 00 h | $1 \mathrm{MHz}{ }^{(\mathbf{1})}$ |

Note 1: The $\mathrm{I}^{2} \mathrm{C}^{\top M}$ interface does not conform to the $400 \mathrm{kHz} \mathrm{I}^{2} \mathrm{C}$ specification (which applies to rates greater than 100 kHz ) in all details, but may be used with care where higher rates are required by the application.

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### 15.4.7.2 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCLx pin (SCLx allowed to float high). When the SCLx pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCLx pin is actually sampled high. When the

SCLx pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and begins counting. This ensures that the SCLx high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 15-18).

FIGURE 15-18: BAUD RATE GENERATOR TIMING WITH CLOCK ARBITRATION


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### 15.4.8 $\quad \mathrm{I}^{2} \mathrm{C}$ MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start Enable bit, SEN (SSPxCON2<0>). If the SDAx and SCLx pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and starts its count. If SCLx and SDAx are both sampled high when the Baud Rate Generator times out (TBRG), the SDAx pin is driven low. The action of the SDAx being driven low while SCLx is high is the Start condition and causes the S bit (SSPxSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and resumes its count. When the Baud Rate Generator times out (TbRG), the SEN bit (SSPxCON2<0>) will be automatically cleared by hardware. The Baud Rate Generator is suspended, leaving the SDAx line held low and the Start condition is complete.

Note: If at the beginning of the Start condition, the SDAx and SCLx pins are already sampled low, or if during the Start condition, the SCLx line is sampled low before the SDAx line is driven low, a bus collision occurs. The Bus Collision Interrupt Flag, BCLxIF, is set, the Start condition is aborted and the $I^{2} \mathrm{C}$ module is reset into its Idle state.

### 15.4.8.1 WCOL Status Flag

If the user writes the SSPxBUF when a Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPxCON2 is disabled until the Start condition is complete.

FIGURE 15-19: FIRST START BIT TIMING


### 15.4.9 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPxCON2<1>) is programmed high and the $\mathrm{I}^{2} \mathrm{C}$ logic module is in the Idle state. When the RSEN bit is set, the SCLx pin is asserted low. When the SCLx pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPxADD<6:0> and begins counting. The SDAx pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDAx is sampled high, the SCLx pin will be deasserted (brought high). When SCLx is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and begins counting. SDAx and SCLx must be sampled high for one Tbrg. This action is then followed by assertion of the SDAx pin (SDAx $=0$ ) for one TbRG while SCLx is high. Following this, the RSEN bit (SSPxCON2<1>) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDAx pin held low. As soon as a Start condition is detected on the SDAx and SCLx pins, the $S$ bit (SSPxSTAT<3>) will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.
2: A bus collision during the Repeated Start condition occurs if:

- SDAx is sampled low when SCLx goes from low-to-high.
- SCLx goes low before SDAx is asserted low. This may indicate that another master is attempting to transmit a data ' 1 '.

Immediately following the SSPxIF bit getting set, the user may write the SSPxBUF with the 7-bit address in 7 -bit mode or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode) or eight bits of data (7-bit mode).

### 15.4.9.1 WCOL Status Flag

If the user writes the SSPxBUF when a Repeated Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).
Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPxCON2 is disabled until the Repeated Start condition is complete.

FIGURE 15-20: REPEATED START CONDITION WAVEFORM
Write to SSPxCON2 occurs here: SDAx=1, $\begin{gathered}\text { SDAx }=1, \\ \text { SCLx (no change) } \\ \begin{array}{c}\text { RSEN bit set by hardware }\end{array} \\ \text { on falling edge of ninth clock, } \\ \text { end of Xmit }\end{gathered}$

### 15.4.10 $\quad \mathrm{I}^{2} \mathrm{C}$ MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSPxBUF register. This action will set the Buffer Full flag bit, BF and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDAx pin after the falling edge of SCLx is asserted (see data hold time specification parameter 106). SCLx is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCLx is released high (see data setup time specification parameter 107). When the SCLx pin is released high, it is held that way for TbRG. The data on the SDAx pin must remain stable for that duration and some hold time after the next falling edge of SCLx. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDAx. This allows the slave device being addressed to respond with an $\overline{\mathrm{ACK}}$ bit during the ninth bit time if an address match occurred, or if data was received properly. The status of $\overline{A C K}$ is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared; if not, the bit is set. After the ninth clock, the SSPxIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPxBUF, leaving SCLx low and SDAx unchanged (Figure 15-21).
After the write to the SSPxBUF, each bit of the address will be shifted out on the falling edge of SCLx until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDAx pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDAx pin to see if the address was recognized by a slave. The status of the $\overline{\text { ACK }}$ bit is loaded into the ACKSTAT status bit (SSPxCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPxIF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPxBUF takes place, holding SCLx low and allowing SDAx to float.

### 15.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPxSTAT<0>) is set when the CPU writes to SSPxBUF and is cleared when all 8 bits are shifted out.

### 15.4.10.2 WCOL Status Flag

If the user writes to the SSPxBUF when a transmit is already in progress (i.e., SSPxSR is still shifting out a data byte), the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur) after 2 TCY after the SSPxBUF write. If SSPxBUF is rewritten within 2 TCY, the WCOL bit is set and SSPxBUF is updated. This may result in a corrupted transfer.

The user should verify that the WCOL is clear after each write to SSPxBUF to ensure the transfer is correct. In all cases, WCOL must be cleared in software.

### 15.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPxCON2<6>) is cleared when the slave has sent an Acknowledge $(\overline{\mathrm{ACK}}=0)$ and is set when the slave does not Acknowledge ( $\overline{\mathrm{ACK}}=1$ ). A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

### 15.4.11 $\quad I^{2} \mathrm{C}$ MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPxCON2<3>).

Note: The MSSP module must be in an Idle state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCLx pin changes (high-to-low/low-to-high) and data is shifted into the SSPxSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPxSR are loaded into the SSPxBUF, the BF flag bit is set, the SSPxIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCLx low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPxCON2<4>).

### 15.4.11.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSPxBUF from SSPxSR. It is cleared when the SSPxBUF register is read.

### 15.4.11.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when 8 bits are received into the SSPxSR and the BF flag bit is already set from a previous reception.

### 15.4.11.3 WCOL Status Flag

If the user writes the SSPxBUF when a receive is already in progress (i.e., SSPxSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

FIGURE 15-21: $\quad I^{2} C^{\text {TM }}$ MASTER MODE WAVEFORM (TRANSMISSION, 7 OR 10-BIT ADDRESS)


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FIGURE 15-22: $\quad 1^{2} \mathbf{C}^{\text {TM }}$ MASTER MODE WAVEFORM (RECEPTION, 7-BIT ADDRESS)


### 15.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit, ACKEN (SSPxCON2<4>). When this bit is set, the SCLx pin is pulled low and the contents of the Acknowledge data bit are presented on the SDAx pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TbRG) and the SCLx pin is deasserted (pulled high). When the SCLx pin is sampled high (clock arbitration), the Baud Rate Generator counts for TbRG. The SCLx pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into Idle mode (Figure 15-23).

### 15.4.12.1 WCOL Status Flag

If the user writes the SSPxBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

### 15.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDAx pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPxCON2<2>). At the end of a receive/transmit, the SCLx line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDAx line low. When the SDAx line is sampled low, the Baud Rate Generator is reloaded and counts down to ' 0 '. When the Baud Rate Generator times out, the SCLx pin will be brought high and one Tbrg (Baud Rate Generator rollover count) later, the SDAx pin will be deasserted. When the SDAx pin is sampled high while SCLx is high, the $P$ bit (SSPxSTAT<4>) is set. A TbRG later, the PEN bit is cleared and the SSPxIF bit is set (Figure 15-24).

### 15.4.13.1 WCOL Status Flag

If the user writes the SSPxBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

FIGURE 15-23: ACKNOWLEDGE SEQUENCE WAVEFORM


FIGURE 15-24: STOP CONDITION RECEIVE OR TRANSMIT MODE


### 15.4.14 SLEEP OPERATION

While in Sleep mode, the $I^{2} C$ module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

### 15.4.15 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

### 15.4.16 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop $(P)$ and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the $I^{2} \mathrm{C}$ bus may be taken when the $P$ bit (SSPxSTAT<4>) is set, or the bus is Idle, with both the $S$ and $P$ bits clear. When the bus is busy, enabling the MSSP interrupt will generate the interrupt when the Stop condition occurs.
In multi-master operation, the SDAx line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed in hardware with the result placed in the BCLxIF bit.
The states where arbitration can be lost are:

- Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition


### 15.4.17 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDAx pin, arbitration takes place when the master outputs a ' 1 ' on SDAx, by letting SDAx float high and another master asserts a ' 0 '. When the SCLx pin floats high, data should be stable. If the expected data on SDAx is a ' 1 ' and the data sampled on the SDAx pin $=0$, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLxIF and reset the $\mathrm{I}^{2} \mathrm{C}$ port to its Idle state (Figure 15-25).
If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDAx and SCLx lines are deasserted and the SSPxBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the $I^{2} \mathrm{C}$ bus is free, the user can resume communication by asserting a Start condition.
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDAx and SCLx lines are deasserted and the respective control bits in the SSPxCON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the $\mathrm{I}^{2} \mathrm{C}$ bus is free, the user can resume communication by asserting a Start condition.
The master will continue to monitor the SDAx and SCLx pins. If a Stop condition occurs, the SSPxIF bit will be set.
A write to the SSPxBUF will start the transmission of data at the first data bit regardless of where the transmitter left off when the bus collision occurred.
In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the $\mathrm{I}^{2} \mathrm{C}$ bus can be taken when the P bit is set in the SSPxSTAT register, or the bus is Idle and the $S$ and $P$ bits are cleared.

FIGURE 15-25: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE


### 15.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:
a) SDAx or SCLx are sampled low at the beginning of the Start condition (Figure 15-26).
b) SCLx is sampled low before SDAx is asserted low (Figure 15-27).
During a Start condition, both the SDAx and the SCLx pins are monitored.
If the SDAx pin is already low, or the SCLx pin is already low, then all of the following occur:

- the Start condition is aborted;
- the BCLxIF flag is set; and
- the MSSP module is reset to its Idle state (Figure 15-26).
The Start condition begins with the SDAx and SCLx pins deasserted. When the SDAx pin is sampled high, the Baud Rate Generator is loaded from SSPxADD<6:0> and counts down to ' 0 '. If the SCLx pin is sampled low while SDAx is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data ' 1 ' during the Start condition.

If the SDAx pin is sampled low during this count, the BRG is reset and the SDAx line is asserted early (Figure 15-28). If, however, a ' 1 ' is sampled on the SDAx pin, the SDAx pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to ' 0 '. If the SCLx pin is sampled as ' 0 ' during this time, a bus collision does not occur. At the end of the BRG count, the SCLx pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDAx before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

FIGURE 15-26: BUS COLLISION DURING START CONDITION (SDAx ONLY)


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FIGURE 15-27: BUS COLLISION DURING START CONDITION (SCLx = 0)


FIGURE 15-28: BRG RESET DUE TO SDAx ARBITRATION DURING START CONDITION


### 15.4.17.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:
a) A low level is sampled on SDAx when SCLx goes from low level to high level.
b) SCLx goes low before SDAx is asserted low, indicating that another master is attempting to transmit a data ' 1 '.
When the user deasserts SDAx and the pin is allowed to float high, the BRG is loaded with SSPxADD<6:0> and counts down to ' 0 '. The SCLx pin is then deasserted and when sampled high, the SDAx pin is sampled.

If SDAx is low, a bus collision has occurred (i.e., another master is attempting to transmit a data ' 0 ', see Figure 15-29). If SDAx is sampled high, the BRG is reloaded and begins counting. If SDAx goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDAx at exactly the same time.
If SCLx goes from high-to-low before the BRG times out and SDAx has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data ' 1 ' during the Repeated Start condition (see Figure 15-30).
If, at the end of the BRG time-out, both SCLx and SDAx are still high, the SDAx pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCLx pin, the SCLx pin is driven low and the Repeated Start condition is complete.

FIGURE 15-29: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)


FIGURE 15-30:
BUS COLLISION DURING REPEATED START CONDITION (CASE 2)


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### 15.4.17.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:
a) After the SDAx pin has been deasserted and allowed to float high, SDAx is sampled low after the BRG has timed out.
b) After the SCLx pin is deasserted, SCLx is sampled low before SDAx goes high.

The Stop condition begins with SDAx asserted low. When SDAx is sampled low, the SCLx pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPxADD<6:0> and counts down to ' 0 '. After the BRG times out, SDAx is sampled. If SDAx is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data ' 0 ' (Figure 15-31). If the SCLx pin is sampled low before SDAx is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data ' 0 ' (Figure 15-32).

FIGURE 15-31: BUS COLLISION DURING A STOP CONDITION (CASE 1)


FIGURE 15-32: BUS COLLISION DURING A STOP CONDITION (CASE 2)


TABLE 15-4: REGISTERS ASSOCIATED WITH I ${ }^{2} C^{\text {TM }}$ OPERATION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on Page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 45 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 45 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 45 |
| PIR3 | SSP2IF | BCL2IF | - | - | - | - | - | - | 45 |
| PIE3 | SSP2IE | BCL2IE | - | - | - | - | - | - | 45 |
| IPR3 | SSP2IP | BCL2IP | - | - | - | - | - | - | 45 |
| TRISC | TRISC7 | TRISC6 | TRISC5 | TRISC4 | TRISC3 | TRISC2 | TRISC1 | TRISC0 | 46 |
| TRISD ${ }^{(1)}$ | TRISD7 | TRISD6 | TRISD5 | TRISD4 | TRISD3 | TRISD2 | TRISD1 | TRISD0 | 46 |
| SSP1BUF | MSSP1 Receive Buffer/Transmit Register |  |  |  |  |  |  |  | 44 |
| SSP1ADD | MSSP1 Address Register ( $I^{2} C^{\text {TM }}$ Slave mode). MSSP1 Baud Rate Reload Register ( ${ }^{2} \mathrm{C}$ Master mode). |  |  |  |  |  |  |  | 44 |
| SSP1CON1 | WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 | 44 |
| SSP1CON2 | GCEN | ACKSTAT | ACKDT | ACKEN | RCEN | PEN | RSEN | SEN | 44 |
| SSP1STAT | SMP | CKE | D/ $\bar{A}$ | P | S | R/W | UA | BF | 44 |
| SSP2BUF | MSSP2 Receive Buffer/Transmit Register |  |  |  |  |  |  |  | 46 |
| SSP2ADD | MSSP2 Address Register ( ${ }^{2} \mathrm{C}$ Slave mode). MSSP2 Baud Rate Reload Register ( ${ }^{2} \mathrm{C}$ Master mode). |  |  |  |  |  |  |  | 46 |
| SSP2CON1 | WCOL | SSPOV | SSPEN | CKP | SSPM3 | SSPM2 | SSPM1 | SSPM0 | 46 |
| SSP2CON2 | GCEN | ACKSTAT | ACKDT | ACKEN | RCEN | PEN | RSEN | SEN | 46 |
| SSP2STAT | SMP | CKE | D/ $\bar{A}$ | P | S | R/W | UA | BF | 46 |

Legend: $-=$ unimplemented, read as ' 0 '. Shaded cells are not used by the MSSP module in $I^{2} C^{\text {TM }}$ mode.
Note 1: These registers and/or bits are not implemented on 28 -pin devices and should be read as ' 0 '.

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NOTES:

### 16.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is one of the two serial I/O modules. (Generically, the USART is also known as a Serial Communications Interface or SCI.) The EUSART can be configured as a full-duplex asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers. It can also be configured as a halfduplex synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc.
The Enhanced USART module implements additional features, including automatic baud rate detection and calibration, automatic wake-up on Sync Break reception and 12-bit Break character transmit. These make it ideally suited for use in Local Interconnect Network bus (LIN bus) systems.
The EUSART can be configured in the following modes:

- Asynchronous (full duplex) with:
- Auto-wake-up on character reception
- Auto-baud calibration
- 12-bit Break character transmission
- Synchronous - Master (half duplex) with selectable clock polarity
- Synchronous - Slave (half duplex) with selectable clock polarity

The pins of the Enhanced USART are multiplexed with PORTC. In order to configure RC6/TX/CK and RC7/RX/DT as an EUSART:

- bit SPEN (RCSTA<7>) must be set (=1)
- bit TRISC<7> must be set (= 1 )
- bit TRISC<6> must be set (=1)

Note: The EUSART control will automatically reconfigure the pin from input to output as needed.
The operation of the Enhanced USART module is controlled through three registers:

- Transmit Status and Control (TXSTA)
- Receive Status and Control (RCSTA)
- Baud Rate Control (BAUDCON)

These are detailed on the following pages in Register 16-1, Register 16-2 and Register 16-3, respectively.

## REGISTER 16-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R-1 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D |
| bit 7 |  |  |  |  |  |  |  |

bit 7 CSRC: Clock Source Select bit
Asynchronous mode:
Don't care.
Synchronous mode:
1 = Master mode (clock generated internally from BRG)
0 = Slave mode (clock from external source)
bit 6 TX9: 9-bit Transmit Enable bit
1 = Selects 9-bit transmission
$0=$ Selects 8-bit transmission
bit 5 TXEN: Transmit Enable bit
1 = Transmit enabled
$0=$ Transmit disabled
Note: SREN/CREN overrides TXEN in Sync mode.
bit 4 SYNC: EUSART Mode Select bit
1 = Synchronous mode
0 = Asynchronous mode
bit 3 SENDB: Send Break Character bit
Asynchronous mode:
1 = Send Sync Break on next transmission (cleared by hardware upon completion)
$0=$ Sync Break transmission completed
Synchronous mode:
Don't care.
bit 2 BRGH: High Baud Rate Select bit
Asynchronous mode:
1 = High speed
0 = Low speed
Synchronous mode:
Unused in this mode.
bit 1 TRMT: Transmit Shift Register Status bit
1 = TSR empty
$0=$ TSR full
bit $0 \quad$ TX9D: 9th bit of Transmit Data
Can be address/data bit or a parity bit.

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 16-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R-0 | R-0 | R-x |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D |
| bit 7 |  |  |  |  |  |  |  |

bit 7 SPEN: Serial Port Enable bit
1 = Serial port enabled (configures RX/DT and TX/CK pins as serial port pins)
$0=$ Serial port disabled (held in Reset)
bit 6 RX9: 9-bit Receive Enable bit
1 = Selects 9-bit reception
$0=$ Selects 8-bit reception
bit 5 SREN: Single Receive Enable bit
Asynchronous mode:
Don't care.
Synchronous mode - Master:
1 = Enables single receive
$0=$ Disables single receive
This bit is cleared after reception is complete.
Synchronous mode - Slave:
Don't care.
bit 4 CREN: Continuous Receive Enable bit
Asynchronous mode:
1 = Enables receiver
0 = Disables receiver
Synchronous mode:
1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)
$0=$ Disables continuous receive
bit 3 ADDEN: Address Detect Enable bit
Asynchronous mode 9-bit (RX9 = 1):
1 = Enables address detection, enables interrupt and loads the receive buffer when RSR<8> is set
$0=$ Disables address detection, all bytes are received and ninth bit can be used as parity bit
Asynchronous mode 9-bit (RX9 = 0):
Don't care.
bit 2 FERR: Framing Error bit
1 = Framing error (can be updated by reading RCREG register and receiving next valid byte)
$0=$ No framing error
bit 1 OERR: Overrun Error bit
1 = Overrun error (can be cleared by clearing bit CREN)
0 = No overrun error
bit $0 \quad$ RX9D: 9th bit of Received Data
This can be address/data bit or a parity bit and must be calculated by user firmware.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 16-3: BAUDCON: BAUD RATE CONTROL REGISTER

| R/W-0 | R-1 | U-0 | R/W-0 | R/W-0 | U-0 | R/W-0 |  | R/W-0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN |  |
| bit 7 |  |  |  |  |  |  |  |  |


| bit 7 | ABDOVF: Auto-Baud Acquisition Rollover Status bit |
| :---: | :---: |
|  | ```\(1=\) A BRG rollover has occurred during Auto-Baud Rate Detect mode (must be cleared in software) \(0=\) No BRG rollover has occurred``` |
| bit 6 | RCIDL: Receive Operation Idle Status bit |
|  | 1 = Receive operation is Idle |
|  | $0=$ Receive operation is active |
| bit 5 | Unimplemented: Read as '0' |
| bit 4 | SCKP: Synchronous Clock Polarity Select bit |
|  | Asynchronous mode: |
|  | Unused in this mode. |
|  | Synchronous mode: |
|  | 1 = Idle state for clock (CK) is a high level |
|  | 0 = Idle state for clock (CK) is a low level |
| bit 3 | BRG16: 16-bit Baud Rate Register Enable bit |
|  | $1=16$-bit Baud Rate Generator - SPBRGH and SPBRG <br> $0=8$-bit Baud Rate Generator - SPBRG only (Compatible mode), SPBRGH value ignored |
| bit 2 | Unimplemented: Read as '0' |
| bit 1 | WUE: Wake-up Enable bit |
|  | Asynchronous mode: |
|  | 1 = EUSART will continue to sample the RX pin - interrupt generated on falling edge; bit cleared in hardware on following rising edge <br> $0=R X$ pin not monitored or rising edge detected |
|  | Synchronous mode: |
|  | Unused in this mode. |
| bit 0 | ABDEN: Auto-Baud Detect Enable bit |
|  | Asynchronous mode: |
|  | ```1 = Enable baud rate measurement on the next character. Requires reception of a Sync field (55h); cleared in hardware upon completion \(0=\) Baud rate measurement disabled or completed``` |
|  | Synchronous mode: |
|  | Unused in this mode. |

## Legend:

| $\mathrm{R}=$ Readable bit | $\mathrm{W}=$ Writable bit | $\mathrm{U}=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | ' 1 ' $=$ Bit is set | ' 0 ' = Bit is cleared $\quad x=$ Bit is unknown |

### 16.1 Baud Rate Generator (BRG)

The BRG is a dedicated 8 -bit or 16 -bit generator that supports both the Asynchronous and Synchronous modes of the EUSART. By default, the BRG operates in 8 -bit mode; setting the BRG16 bit (BAUDCON<3>) selects 16-bit mode.
The SPBRGH:SPBRG register pair controls the period of a free running timer. In Asynchronous mode, bits BRGH (TXSTA<2>) and BRG16 (BAUDCON $<3>$ ) also control the baud rate. In Synchronous mode, BRGH is ignored. Table 16-1 shows the formula for computation of the baud rate for different EUSART modes which only apply in Master mode (internally generated clock).
Given the desired baud rate and Fosc, the nearest integer value for the SPBRGH:SPBRG registers can be calculated using the formulas in Table 16-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 16-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 16-2. It may be advan-
tageous to use the high baud rate $(\mathrm{BRGH}=1)$ or the 16-bit BRG to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRGH:SPBRG registers causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

### 16.1.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRG register pair.

### 16.1.2 SAMPLING

The data on the RX pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX pin.

## TABLE 16-1: BAUD RATE FORMULAS

| Configuration Bits |  |  | BRG/EUSART Mode | Baud Rate Formula |
| :---: | :---: | :---: | :---: | :---: |
| SYNC | BRG16 | BRGH |  |  |
| 0 | 0 | 0 | 8-bit/Asynchronous | Fosc/[64 ( $\mathrm{n}+1$ ] |
| 0 | 0 | 1 | 8-bit/Asynchronous | Fosc/[16 ( $n+1$ ] |
| 0 | 1 | 0 | 16-bit/Asynchronous | FosC/[16( $n+1$ ] |
| 0 | 1 | 1 | 16-bit/Asynchronous |  |
| 1 | 0 | x | 8-bit/Synchronous | Fosc/[4 ( $\mathrm{n}+1$ )] |
| 1 | 1 | x | 16-bit/Synchronous |  |

Legend: $\mathrm{x}=$ Don't care, $\mathrm{n}=$ value of SPBRGH:SPBRG register pair

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EXAMPLE 16-1: CALCULATING BAUD RATE ERROR
For a device with FOSC of 16 MHz , desired baud rate of 9600, Asynchronous mode, 8-bit BRG:
Desired Baud Rate $=$ Fosc/(64 ([SPBRGH:SPBRG] + 1))
Solving for SPBRGH:SPBRG:
$\mathrm{X}=(($ Fosc/Desired Baud Rate $) / 64)-1$
$=((16000000 / 9600) / 64)-1$
$=[25.042]=25$
Calculated Baud Rate $=16000000 /(64(25+1))$
$=9615$
Error $\quad=$ (Calculated Baud Rate - Desired Baud Rate)/Desired Baud Rate
$=(9615-9600) / 9600=0.16 \%$

TABLE 16-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  | 45 |  |  |  |  |  |
| SPBRG | EUSART Baud Rate Generator Register Low Byte | 45 |  |  |  |  |  |  |  |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by the BRG.

TABLE 16-3: BAUD RATES FOR ASYNCHRONOUS MODES

| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=0, \mathrm{BRG1} 16=0$ |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=40.000 \mathrm{MHz}$ |  |  | Fosc $=\mathbf{2 0 . 0 0 0 ~ M H z ~}$ |  |  | Fosc $=10.000 \mathrm{MHz}$ |  |  | Fosc $=8.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | $\begin{gathered} \text { \% } \\ \text { Error } \end{gathered}$ | SPBRG value (decimal) |
| 0.3 | - | - | - | - | - | - | - | - | - | - | - | - |
| 1.2 | - | - | - | 1.221 | 1.73 | 255 | 1.202 | 0.16 | 129 | 1201 | -0.16 | 103 |
| 2.4 | 2.441 | 1.73 | 255 | 2.404 | 0.16 | 129 | 2.404 | 0.16 | 64 | 2403 | -0.16 | 51 |
| 9.6 | 9.615 | 0.16 | 64 | 9.766 | 1.73 | 31 | 9.766 | 1.73 | 15 | 9615 | -0.16 | 12 |
| 19.2 | 19.531 | 1.73 | 31 | 19.531 | 1.73 | 15 | 19.531 | 1.73 | 7 | - | - | - |
| 57.6 | 56.818 | -1.36 | 10 | 62.500 | 8.51 | 4 | 52.083 | -9.58 | 2 | - | - | - |
| 115.2 | 125.000 | 8.51 | 4 | 104.167 | -9.58 | 2 | 78.125 | -32.18 | 1 | - | - | - |


| BAUD RATE (K) | SYNC $=0, B R G H=0, B R G 16=0$ |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=4.000 \mathrm{MHz}$ |  |  | Fosc $=2.000 \mathrm{MHz}$ |  |  | Fosc $=1.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% <br> Error | SPBRG value (decimal) | Actual Rate (K) | $\begin{gathered} \text { \% } \\ \text { Error } \end{gathered}$ | SPBRG value (decimal) |
| 0.3 | 0.300 | 0.16 | 207 | 300 | -0.16 | 103 | 300 | -0.16 | 51 |
| 1.2 | 1.202 | 0.16 | 51 | 1201 | -0.16 | 25 | 1201 | -0.16 | 12 |
| 2.4 | 2.404 | 0.16 | 25 | 2403 | -0.16 | 12 | - | - | - |
| 9.6 | 8.929 | -6.99 | 6 | - | - | - | - | - | - |
| 19.2 | 20.833 | 8.51 | 2 | - | - | - | - | - | - |
| 57.6 | 62.500 | 8.51 | 0 | - | - | - | - | - | - |
| 115.2 | 62.500 | -45.75 | 0 | - | - | - | - | - | - |


| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=1, \mathrm{BRG16}=0$ |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=40.000 \mathrm{MHz}$ |  |  | Fosc $=\mathbf{2 0 . 0 0 0 ~ M H z ~}$ |  |  | Fosc $=10.000 \mathrm{MHz}$ |  |  | Fosc $=8.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | $\%$ <br> Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) |
| 0.3 | - | - | - | - | - | - | - | - | - | - | - | - |
| 1.2 | - | - | - | - | - | - | - | - | - | - | - | - |
| 2.4 | - | - | - | - | - | - | 2.441 | 1.73 | 255 | 2403 | -0.16 | 207 |
| 9.6 | 9.766 | 1.73 | 255 | 9.615 | 0.16 | 129 | 9.615 | 0.16 | 64 | 9615 | -0.16 | 51 |
| 19.2 | 19.231 | 0.16 | 129 | 19.231 | 0.16 | 64 | 19.531 | 1.73 | 31 | 19230 | -0.16 | 25 |
| 57.6 | 58.140 | 0.94 | 42 | 56.818 | -1.36 | 21 | 56.818 | -1.36 | 10 | 55555 | 3.55 | 8 |
| 115.2 | 113.636 | -1.36 | 21 | 113.636 | -1.36 | 10 | 125.000 | 8.51 | 4 | - | - | - |


| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=1, \mathrm{BRG16}=0$ |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=4.000 \mathrm{MHz}$ |  |  | Fosc $=2.000 \mathrm{MHz}$ |  |  | Fosc $=1.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | $\%$ <br> Error | SPBRG value <br> (decimal) | Actual Rate (K) | $\begin{gathered} \text { \% } \\ \text { Error } \end{gathered}$ | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) |
| 0.3 | - | - | - | - | - | - | 300 | -0.16 | 207 |
| 1.2 | 1.202 | 0.16 | 207 | 1201 | -0.16 | 103 | 1201 | -0.16 | 51 |
| 2.4 | 2.404 | 0.16 | 103 | 2403 | -0.16 | 51 | 2403 | -0.16 | 25 |
| 9.6 | 9.615 | 0.16 | 25 | 9615 | -0.16 | 12 | - | - | - |
| 19.2 | 19.231 | 0.16 | 12 | - | - | - | - | - | - |
| 57.6 | 62.500 | 8.51 | 3 | - | - | - | - | - | - |
| 115.2 | 125.000 | 8.51 | 1 | - | - | - | - | - | - |

TABLE 16-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=0, \mathrm{BRG16}=1$ |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=40.000 \mathrm{MHz}$ |  |  | Fosc $=20.000 \mathrm{MHz}$ |  |  | Fosc $=10.000 \mathrm{MHz}$ |  |  | Fosc $=8.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | \% <br> Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | $\begin{gathered} \text { \% } \\ \text { Error } \end{gathered}$ | SPBRG value (decimal) |
| 0.3 | 0.300 | 0.00 | 8332 | 0.300 | 0.02 | 4165 | 0.300 | 0.02 | 2082 | 300 | -0.04 | 1665 |
| 1.2 | 1.200 | 0.02 | 2082 | 1.200 | -0.03 | 1041 | 1.200 | -0.03 | 520 | 1201 | -0.16 | 415 |
| 2.4 | 2.402 | 0.06 | 1040 | 2.399 | -0.03 | 520 | 2.404 | 0.16 | 259 | 2403 | -0.16 | 207 |
| 9.6 | 9.615 | 0.16 | 259 | 9.615 | 0.16 | 129 | 9.615 | 0.16 | 64 | 9615 | -0.16 | 51 |
| 19.2 | 19.231 | 0.16 | 129 | 19.231 | 0.16 | 64 | 19.531 | 1.73 | 31 | 19230 | -0.16 | 25 |
| 57.6 | 58.140 | 0.94 | 42 | 56.818 | -1.36 | 21 | 56.818 | -1.36 | 10 | 55555 | 3.55 | 8 |
| 115.2 | 113.636 | -1.36 | 21 | 113.636 | -1.36 | 10 | 125.000 | 8.51 | 4 | - | - | - |


| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=0, \mathrm{BRG16}=1$ |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=4.000 \mathrm{MHz}$ |  |  | Fosc $=2.000 \mathrm{MHz}$ |  |  | Fosc $=1.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) |
| 0.3 | 0.300 | 0.04 | 832 | 300 | -0.16 | 415 | 300 | -0.16 | 207 |
| 1.2 | 1.202 | 0.16 | 207 | 1201 | -0.16 | 103 | 1201 | -0.16 | 51 |
| 2.4 | 2.404 | 0.16 | 103 | 2403 | -0.16 | 51 | 2403 | -0.16 | 25 |
| 9.6 | 9.615 | 0.16 | 25 | 9615 | -0.16 | 12 | - | - | - |
| 19.2 | 19.231 | 0.16 | 12 | - | - | - | - | - | - |
| 57.6 | 62.500 | 8.51 | 3 | - | - | - | - | - | - |
| 115.2 | 125.000 | 8.51 | 1 | - | - | - | - | - | - |


| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=1, \mathrm{BRG16}=1$ or SYNC $=1, \mathrm{BRG16}=1$ |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=40.000 \mathrm{MHz}$ |  |  | Fosc $=20.000 \mathrm{MHz}$ |  |  | Fosc $=10.000 \mathrm{MHz}$ |  |  | Fosc $=8.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | $\begin{gathered} \text { \% } \\ \text { Error } \end{gathered}$ | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) |
| 0.3 | 0.300 | 0.00 | 33332 | 0.300 | 0.00 | 16665 | 0.300 | 0.00 | 8332 | 300 | -0.01 | 6665 |
| 1.2 | 1.200 | 0.00 | 8332 | 1.200 | 0.02 | 4165 | 1.200 | 0.02 | 2082 | 1200 | -0.04 | 1665 |
| 2.4 | 2.400 | 0.02 | 4165 | 2.400 | 0.02 | 2082 | 2.402 | 0.06 | 1040 | 2400 | -0.04 | 832 |
| 9.6 | 9.606 | 0.06 | 1040 | 9.596 | -0.03 | 520 | 9.615 | 0.16 | 259 | 9615 | -0.16 | 207 |
| 19.2 | 19.193 | -0.03 | 520 | 19.231 | 0.16 | 259 | 19.231 | 0.16 | 129 | 19230 | -0.16 | 103 |
| 57.6 | 57.803 | 0.35 | 172 | 57.471 | -0.22 | 86 | 58.140 | 0.94 | 42 | 57142 | 0.79 | 34 |
| 115.2 | 114.943 | -0.22 | 86 | 116.279 | 0.94 | 42 | 113.636 | -1.36 | 21 | 117647 | -2.12 | 16 |


| BAUD RATE (K) | SYNC $=0, \mathrm{BRGH}=1, \mathrm{BRG16}=1$ or SYNC $=1, \mathrm{BRG16}=1$ |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Fosc $=4.000 \mathrm{MHz}$ |  |  | Fosc $=2.000 \mathrm{MHz}$ |  |  | Fosc $=1.000 \mathrm{MHz}$ |  |  |
|  | Actual Rate (K) | \% Error | SPBRG value (decimal) | Actual Rate (K) | $\begin{gathered} \text { \% } \\ \text { Error } \end{gathered}$ | SPBRG value (decimal) | Actual Rate (K) | \% Error | SPBRG value (decimal) |
| 0.3 | 0.300 | 0.01 | 3332 | 300 | -0.04 | 1665 | 300 | -0.04 | 832 |
| 1.2 | 1.200 | 0.04 | 832 | 1201 | -0.16 | 415 | 1201 | -0.16 | 207 |
| 2.4 | 2.404 | 0.16 | 415 | 2403 | -0.16 | 207 | 2403 | -0.16 | 103 |
| 9.6 | 9.615 | 0.16 | 103 | 9615 | -0.16 | 51 | 9615 | -0.16 | 25 |
| 19.2 | 19.231 | 0.16 | 51 | 19230 | -0.16 | 25 | 19230 | -0.16 | 12 |
| 57.6 | 58.824 | 2.12 | 16 | 55555 | 3.55 | 8 | - | - | - |
| 115.2 | 111.111 | -3.55 | 8 | - | - | - | - | - | - |

### 16.1.3 AUTO-BAUD RATE DETECT

The Enhanced USART module supports the automatic detection and calibration of baud rate. This feature is active only in Asynchronous mode and while the WUE bit is clear.
The automatic baud rate measurement sequence (Figure 16-1) begins whenever a Start bit is received and the ABDEN bit is set. The calculation is self-averaging.
In the Auto-Baud Rate Detect (ABD) mode, the clock to the $B R G$ is reversed. Rather than the BRG clocking the incoming $R X$ signal, the $R X$ signal is timing the $B R G$. In ABD mode, the internal Baud Rate Generator is used as a counter to time the bit period of the incoming serial byte stream.
Once the ABDEN bit is set, the state machine will clear the BRG and look for a Start bit. The Auto-Baud Rate Detect must receive a byte with the value 55h (ASCII "U", which is also the LIN bus Sync character) in order to calculate the proper bit rate. The measurement is taken over both a low and a high bit time in order to minimize any effects caused by asymmetry of the incoming signal. After a Start bit, the SPBRG begins counting up, using the preselected clock source on the first rising edge of $R X$. After eight bits on the RX pin or the fifth rising edge, an accumulated value totalling the proper BRG period is left in the SPBRGH:SPBRG register pair. Once the 5th edge is seen (this should correspond to the Stop bit), the ABDEN bit is automatically cleared.
If a rollover of the BRG occurs (an overflow from FFFFh to 0000h), the event is trapped by the ABDOVF status bit (BAUDCON<7>). It is set in hardware by BRG rollovers and can be set or cleared by the user in software. ABD mode remains active after rollover events and the ABDEN bit remains set (Figure 16-2).

While calibrating the baud rate period, the BRG registers are clocked at $1 / 8$ th the preconfigured clock rate. Note that the BRG clock will be configured by the BRG16 and BRGH bits. Independent of the BRG16 bit setting, both the SPBRG and SPBRGH will be used as a 16-bit counter. This allows the user to verify that no carry occurred for 8 -bit modes by checking for 00 h in the SPBRGH register. Refer to Table 16-4 for counter clock rates to the BRG.
While the ABD sequence takes place, the EUSART state machine is held in Idle. The RCIF interrupt is set once the fifth rising edge on RX is detected. The value in the RCREG needs to be read to clear the RCIF interrupt. The contents of RCREG should be discarded.

Note 1: If the WUE bit is set with the ABDEN bit, Auto-Baud Rate Detection will occur on the byte following the Break character.

2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible due to bit error rates. Overall system timing and communication baud rates must be taken into consideration when using the Auto-Baud Rate Detection feature.

## TABLE 16-4: BRG COUNTER <br> CLOCK RATES

| BRG16 | BRGH | BRG Counter Clock |
| :---: | :---: | :---: |
| 0 | 0 | Fosc/512 |
| 0 | 1 | Fosc/128 |
| 1 | 0 | Fosc/128 |
| 1 | 1 | Fosc/32 |

Note: During the ABD sequence, SPBRG and SPBRGH are both used as a 16-bit counter, independent of BRG16 setting.

### 16.1.3.1 ABD and EUSART Transmission

Since the BRG clock is reversed during ABD acquisition, the EUSART transmitter cannot be used during $A B D$. This means that whenever the ABDEN bit is set, TXREG cannot be written to. Users should also ensure that ABDEN does not become set during a transmit sequence. Failing to do this may result in unpredictable EUSART operation.

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FIGURE 16-1: AUTOMATIC BAUD RATE CALCULATION


Note: The ABD sequence requires the EUSART module to be configured in Asynchronous mode and WUE $=0$.

FIGURE 16-2: BRG OVERFLOW SEQUENCE


### 16.2 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA<4>). In this mode, the EUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.
The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x 16 or x 64 of the bit shift rate depending on the BRGH and BRG16 bits (TXSTA<2> and BAUDCON $<3>$ ). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.
When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver
- Auto-Wake-up on Sync Break Character
- 12-bit Break Character Transmit
- Auto-Baud Rate Detection


### 16.2.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 16-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG register (if available).

Once the TXREG register transfers the data to the TSR register (occurs in one TcY), the TXREG register is empty and the TXIF flag bit (PIR1<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXIE (PIE1<4>). TXIF will be set regardless of the state of TXIE; it cannot be cleared in software. TXIF is also not cleared immediately upon loading TXREG, but becomes valid in the second instruction cycle following the load instruction. Polling TXIF immediately following a load of TXREG will return invalid results.
While TXIF indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1: The TSR register is not mapped in data memory so it is not available to the user.
2: Flag bit TXIF is set when enable bit TXEN is set.

To set up an Asynchronous Transmission:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
3. If interrupts are desired, set enable bit TXIE.
4. If 9-bit transmission is desired, set transmit bit TX9. Can be used as address/data bit.
5. Enable the transmission by setting bit TXEN which will also set bit TXIF.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Load data to the TXREG register (starts transmission).
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 16-3: EUSART TRANSMIT BLOCK DIAGRAM


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FIGURE 16-4: ASYNCHRONOUS TRANSMISSION


FIGURE 16-5: ASYNCHRONOUS TRANSMISSION (BACK TO BACK)


Note: This timing diagram shows two consecutive transmissions.

## TABLE 16-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| TXREG | EUSART Transmit Register |  |  |  |  |  |  |  | 45 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 45 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 45 |

Legend: - = unimplemented locations read as ' 0 '. Shaded cells are not used for asynchronous transmission.
Note 1: These bits are not implemented on 28-pin devices and should be read as ' 0 '.

### 16.2.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 16-6. The data is received on the RX pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at $x 16$ times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.
To set up an Asynchronous Reception:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
3. If interrupts are desired, set enable bit RCIE.
4. If 9-bit reception is desired, set bit RX9.
5. Enable the reception by setting bit CREN.
6. Flag bit, RCIF, will be set when reception is complete and an interrupt will be generated if enable bit, RCIE, was set.
7. Read the RCSTA register to get the 9th bit (if enabled) and determine if any error occurred during reception.
8. Read the 8-bit received data by reading the RCREG register.
9. If any error occurred, clear the error by clearing enable bit CREN.
10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

### 16.2.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
3. If interrupts are required, set the RCEN bit and select the desired priority level with the RCIP bit.
4. Set the RX9 bit to enable 9-bit reception.
5. Set the ADDEN bit to enable address detect.
6. Enable reception by setting the CREN bit.
7. The RCIF bit will be set when reception is complete. The interrupt will be Acknowledged if the RCIE and GIE bits are set.
8. Read the RCSTA register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
9. Read RCREG to determine if the device is being addressed.
10. If any error occurred, clear the CREN bit.
11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.

## FIGURE 16-6: EUSART RECEIVE BLOCK DIAGRAM



## FIGURE 16-7: ASYNCHRONOUS RECEPTION



Note: This timing diagram shows three words appearing on the RX input. The RCREG (receive buffer) is read after the third word causing the OERR (Overrun) bit to be set.

## TABLE 16-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP ${ }^{(1)}$ | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| RCREG | EUSART Receive Register |  |  |  |  |  |  |  | 45 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 45 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 45 |

Legend: $-=$ unimplemented locations read as ' 0 '. Shaded cells are not used for asynchronous reception.
Note 1: These bits are not implemented on 28 -pin devices and should be read as ' 0 '.

### 16.2.4 AUTO-WAKE-UP ON SYNC BREAK CHARACTER

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper byte reception cannot be performed. The auto-wake-up feature allows the controller to wake-up due to activity on the RX/DT line while the EUSART is operating in Asynchronous mode.
The auto-wake-up feature is enabled by setting the WUE bit (BAUDCON<1>). Once set, the typical receive sequence on RX/DT is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a Wake-up Signal character for the LIN protocol.)

Following a wake-up event, the module generates an RCIF interrupt. The interrupt is generated synchronously to the $Q$ clocks in normal operating modes (Figure 16-8) and asynchronously, if the device is in Sleep mode (Figure 16-9). The interrupt condition is cleared by reading the RCREG register.
The WUE bit is automatically cleared once a low-tohigh transition is observed on the $R X$ line following the wake-up event. At this point, the EUSART module is in Idle mode and returns to normal operation. This signals to the user that the Sync Break event is over.

### 16.2.4.1 Special Considerations Using Auto-Wake-up

Since auto-wake-up functions by sensing rising edge transitions on RX/DT, information with any state changes before the Stop bit may signal a false end-ofcharacter and cause data or framing errors. To work properly, therefore, the initial character in the transmission must be all ' 0 's. This can be 00 h ( 8 bytes) for standard RS-232 devices or 000h ( 12 bits) for LIN bus.
Oscillator start-up time must also be considered, especially in applications using oscillators with longer start-up intervals (i.e., HS mode). The Sync Break (or Wake-up Signal) character must be of sufficient length and be followed by a sufficient interval to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

### 16.2.4.2 Special Considerations Using the WUE Bit

The timing of WUE and RCIF events may cause some confusion when it comes to determining the validity of received data. As noted, setting the WUE bit places the EUSART in an Idle mode. The wake-up event causes a receive interrupt by setting the RCIF bit. The WUE bit is cleared after this when a rising edge is seen on RX/DT. The interrupt condition is then cleared by reading the RCREG register. Ordinarily, the data in RCREG will be dummy data and should be discarded.
The fact that the WUE bit has been cleared (or is still set) and the RCIF flag is set should not be used as an indicator of the integrity of the data in RCREG. Users should consider implementing a parallel method in firmware to verify received data integrity.
To assure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

FIGURE 16-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING NORMAL OPERATION


Note 1: The EUSART remains in Idle while the WUE bit is set.

FIGURE 16-9: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP


Note 1: If the wake-up event requires long oscillator warm-up time, the auto-clear of the WUE bit can occur before the oscillator is ready. This sequence should not depend on the presence of Q clocks.
2: The EUSART remains in Idle while the WUE bit is set.

### 16.2.5 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. The Break character transmit consists of a Start bit, followed by twelve ' 0 ' bits and a Stop bit. The frame Break character is sent whenever the SENDB and TXEN bits (TXSTA<3> and TXSTA $<5>$ ) are set while the Transmit Shift register is loaded with data. Note that the value of data written to TXREG will be ignored and all ' 0 's will be transmitted.
The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).
Note that the data value written to the TXREG for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.
The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 16-10 for the timing of the Break character sequence.

### 16.2.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to set up the Break character.
3. Load the TXREG with a dummy character to initiate transmission (the value is ignored).
4. Write '55h' to TXREG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.
When the TXREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXREG.

### 16.2.6 RECEIVING A BREAK CHARACTER

The Enhanced USART module can receive a Break character in two ways.
The first method forces configuration of the baud rate at a frequency of $9 / 13$ the typical speed. This allows for the Stop bit transition to be at the correct sampling location ( 13 bits for Break versus Start bit and 8 data bits for typical data).
The second method uses the auto-wake-up feature described in Section 16.2.4 "Auto-Wake-up on Sync Break Character". By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCIF interrupt and receive the next data byte followed by another interrupt.
Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABD bit once the TXIF interrupt is observed.

FIGURE 16-10: SEND BREAK CHARACTER SEQUENCE


### 16.3 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA $<7>$ ). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit SYNC (TXSTA<4>). In addition, enable bit SPEN ( $\mathrm{RCSTA}<7>$ ) is set in order to configure the TX and RX pins to CK (clock) and DT (data) lines, respectively.
The Master mode indicates that the processor transmits the master clock on the CK line. Clock polarity is selected with the SCKP bit (BAUDCON<4>). Setting SCKP sets the Idle state on CK as high, while clearing the bit sets the Idle state as low. This option is provided to support Microwire devices with this module.

### 16.3.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

The EUSART transmitter block diagram is shown in Figure 16-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available).

Once the TXREG register transfers the data to the TSR register (occurs in one TCY), the TXREG is empty and the TXIF flag bit (PIR1<4>) is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXIE (PIE1<4>). TXIF is set regardless of the state of enable bit TXIE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREG register.
While flag bit TXIF indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.
To set up a Synchronous Master Transmission:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. If interrupts are desired, set enable bit TXIE.
4. If 9-bit transmission is desired, set bit TX9.
5. Enable the transmission by setting bit TXEN.
6. If 9 -bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Start transmission by loading data to the TXREG register.
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON $<7: 6>$ ) are set.

FIGURE 16-11: SYNCHRONOUS TRANSMISSION


Note: Sync Master mode, SPBRG $=0$, continuous transmission of two 8 -bit words.

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FIGURE 16-12: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)


TABLE 16-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP ${ }^{(1)}$ | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| TXREG | EUSART Transmit Register |  |  |  |  |  |  |  | 45 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 45 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used for synchronous master transmission.
Note 1: These bits are not implemented on 28-pin devices and should be read as ' 0 '.

### 16.3.2 EUSART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTA $<5>$ ), or the Continuous Receive Enable bit, CREN (RCSTA<4>). Data is sampled on the $R X$ pin on the falling edge of the clock.
If enable bit SREN is set, only a single word is received. If enable bit CREN is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.
To set up a Synchronous Master Reception:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
3. Ensure bits CREN and SREN are clear.
4. If interrupts are desired, set enable bit RCIE.
5. If 9-bit reception is desired, set bit RX9.
6. If a single reception is required, set bit SREN. For continuous reception, set bit CREN.
7. Interrupt flag bit, RCIF, will be set when reception is complete and an interrupt will be generated if the enable bit, RCIE, was set.
8. Read the RCSTA register to get the 9th bit (if enabled) and determine if any error occurred during reception.
9. Read the 8 -bit received data by reading the RCREG register.
10. If any error occurred, clear the error by clearing bit CREN.
11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON $<7: 6>$ ) are set.

FIGURE 16-13: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)


TABLE 16-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| RCREG | EUSART Receive Register |  |  |  |  |  |  |  | 45 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 45 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used for synchronous master reception.
Note 1: These bits are not implemented on 28-pin devices and should be read as ' 0 '.

### 16.4 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTA<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CK pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

### 16.4.1 EUSART SYNCHRONOUS SLAVE TRANSMISSION

The operation of the Synchronous Master and Slave modes is identical, except in the case of the Sleep mode.
If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:
a) The first word will immediately transfer to the TSR register and transmit.
b) The second word will remain in the TXREG register.
c) Flag bit, TXIF, will not be set.
d) When the first word has been shifted out of TSR, the TXREG register will transfer the second word to the TSR and flag bit, TXIF, will now be set.
e) If enable bit TXIE is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting bits SYNC and SPEN and clearing bit CSRC.
2. Clear bits CREN and SREN.
3. If interrupts are desired, set enable bit TXIE.
4. If 9-bit transmission is desired, set bit TX9.
5. Enable the transmission by setting enable bit TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
7. Start transmission by loading data to the TXREG register.
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON $<7: 6>$ ) are set.

TABLE 16-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| TXREG | EUSART Transmit Register |  |  |  |  |  |  |  | 45 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 45 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used for synchronous slave transmission.
Note 1: These bits are not implemented on 28-pin devices and should be read as ' 0 '.

### 16.4.2 EUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical, except in the case of Sleep, or any Idle mode and bit SREN, which is a "don't care" in Slave mode.
If receive is enabled by setting the CREN bit prior to entering Sleep or any Idle mode, then a word may be received while in this low-power mode. Once the word is received, the RSR register will transfer the data to the RCREG register; if the RCIE enable bit is set, the interrupt generated will wake the chip from the low-power mode. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Reception:

1. Enable the synchronous master serial port by setting bits SYNC and SPEN and clearing bit CSRC.
2. If interrupts are desired, set enable bit RCIE.
3. If 9-bit reception is desired, set bit RX9.
4. To enable reception, set enable bit CREN.
5. Flag bit, RCIF, will be set when reception is complete. An interrupt will be generated if enable bit, RCIE, was set.
6. Read the RCSTA register to get the 9th bit (if enabled) and determine if any error occurred during reception.
7. Read the 8-bit received data by reading the RCREG register.
8. If any error occurred, clear the error by clearing bit CREN.
9. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

TABLE 16-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| RCSTA | SPEN | RX9 | SREN | CREN | ADDEN | FERR | OERR | RX9D | 45 |
| RCREG | EUSART Receive Register |  |  |  |  |  |  |  | 45 |
| TXSTA | CSRC | TX9 | TXEN | SYNC | SENDB | BRGH | TRMT | TX9D | 45 |
| BAUDCON | ABDOVF | RCIDL | - | SCKP | BRG16 | - | WUE | ABDEN | 45 |
| SPBRGH | EUSART Baud Rate Generator Register High Byte |  |  |  |  |  |  |  | 45 |
| SPBRG | EUSART Baud Rate Generator Register Low Byte |  |  |  |  |  |  |  | 45 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used for synchronous slave reception.
Note 1: These bits are not implemented on 28 -pin devices and should be read as ' 0 '.

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NOTES:

### 17.0 10-BIT ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) converter module has 10 inputs for the 28 -pin devices and 13 for the $40 / 44$-pin devices. This module allows conversion of an analog input signal to a corresponding 10-bit digital number.
The module has five registers:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register 0 (ADCONO)
- A/D Control Register 1 (ADCON1)
- A/D Control Register 2 (ADCON2)

The ADCONO register, shown in Register 17-1, controls the operation of the $A / D$ module. The ADCON1 register, shown in Register 17-2, configures the functions of the port pins. The ADCON2 register, shown in Register 17-3, configures the A/D clock source, programmed acquisition time and justification.

## REGISTER 17-1: ADCONO: A/D CONTROL REGISTER 0

| R/W-0 | U-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADCAL | - | CHS3 | CHS2 | CHS1 | CHS0 | GO/ $\overline{\text { DONE }}$ | ADON |

bit 7
bit 0
bit 7 ADCAL: A/D Calibration bit
$1=$ Calibration is performed on next $A / D$ conversion
$0=$ Normal A/D converter operation
bit 6 Unimplemented: Read as ' 0 '
bit 5-2 CHS3:CHSO: Analog Channel Select bits
0000 = Channel 0 (ANO)
0001 = Channel 1 (AN1)
$0010=$ Channel 2 (AN2)
0011 = Channel 3 (AN3)
0100 = Channel 4 (AN4)
$0101=$ Channel 5 (AN5) ${ }^{(1,2)}$
$0110=$ Channel 6 (AN6) ${ }^{(1,2)}$
$0111=$ Channel 7 (AN7) ${ }^{(\mathbf{1 , 2})}$
$1000=$ Channel 8 (AN8)
1001 = Channel 9 (AN9)
$1010=$ Channel 10 (AN10)
1011 = Channel 11 (AN11)
$1100=$ Channel 12 (AN12
$1101=$ Unimplemented ${ }^{(2)}$
$1110=$ Unimplemented ${ }^{(2)}$
$1111=$ Unimplemented ${ }^{(\mathbf{2})}$
Note 1: These channels are not implemented on 28-pin devices.
2: Performing a conversion on unimplemented channels will return a floating input measurement.
bit 1 GO/DONE: A/D Conversion Status bit
When ADON = 1:
$1=A / D$ conversion in progress
$0=A / D$ Idle
bit $0 \quad$ ADON: A/D On bit
$1=A / D$ converter module is enabled
$0=A / D$ converter module is disabled

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' $=$ Bit is set | $' 0$ ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 17-2: ADCON1: A/D CONTROL REGISTER 1

| U-0 | U-0 | R/W-0 | R/W-0 | R/W-0 ${ }^{(1)}$ | $\mathrm{R} / \mathrm{W}^{(1)}$ | R/W ${ }^{(1)}$ | $\mathrm{R} / \mathrm{W}^{(1)}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | VCFG1 | VCFG0 | PCFG3 | PCFG2 | PCFG1 | PCFG0 |
|  |  |  |  |  |  |  | bit 0 |

bit 7-6
bit 5
bit 4
bit 3-0

Unimplemented: Read as ' 0 '
VCFG1: Voltage Reference Configuration bit (Vref- source)
1 = VREF- (AN2)
$0=\mathrm{Vss}$
VCFGO: Voltage Reference Configuration bit (VREF+ source)
1 = VREF+ (AN3)
$0=$ VDD
PCFG3:PCFG0: A/D Port Configuration Control bits:

| PCFG3: <br> PCFG0 | $\underset{\mathbb{K}}{\underset{\gtrless}{N}}$ | $\underset{\mathbb{Z}}{\underset{Z}{2}}$ | $\frac{0}{2}$ | $\underset{<}{2}$ | $\sum_{<}^{\infty}$ | $\frac{\widehat{N}}{\underset{\sim}{2}}$ | $\frac{\widehat{N}}{\substack{0}}$ | $\frac{\widehat{y}}{\frac{10}{2}}$ | $\underset{4}{\mathbb{Z}}$ | $\sum_{\lll}^{\infty}$ | $\underset{\mathbb{Z}}{\sim}$ | $\underset{\ll}{\Sigma}$ | $\underset{4}{2}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $0000^{(1)}$ | A | A | A | A | A | A | A | A | A | A | A | A | A |
| 0001 | A | A | A | A | A | A | A | A | A | A | A | A | A |
| 0010 | A | A | A | A | A | A | A | A | A | A | A | A | A |
| 0011 | D | A | A | A | A | A | A | A | A | A | A | A | A |
| 0100 | D | D | A | A | A | A | A | A | A | A | A | A | A |
| 0101 | D | D | D | A | A | A | A | A | A | A | A | A | A |
| 0110 | D | D | D | D | A | A | A | A | A | A | A | A | A |
| $0111^{(1)}$ | D | D | D | D | D | A | A | A | A | A | A | A | A |
| 1000 | D | D | D | D | D | D | A | A | A | A | A | A | A |
| 1001 | D | D | D | D | D | D | D | A | A | A | A | A | A |
| 1010 | D | D | D | D | D | D | D | D | A | A | A | A | A |
| 1011 | D | D | D | D | D | D | D | D | D | A | A | A | A |
| 1100 | D | D | D | D | D | D | D | D | D | D | A | A | A |
| 1101 | D | D | D | D | D | D | D | D | D | D | D | A | A |
| 1110 | D | D | D | D | D | D | D | D | D | D | D | D | A |
| 1111 | D | D | D | D | D | D | D | D | D | D | D | D | D |

A = Analog input
D = Digital I/O
Note 1: The POR value of the PCFG bits depends on the value of the PBADEN configuration bit. When PBADEN $=1$, PCFG<3:0> $=0000$; when $\operatorname{PBADEN}=0$, PCFG<3:0> = 0111 .
2: AN5 through AN7 are available only on 40/44-pin devices.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

REGISTER 17-3: ADCON2: A/D CONTROL REGISTER 2

| R/W-0 | U-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADFM | - | ACQT2 | ACQT1 | ACQT0 | ADCS2 | ADCS1 | ADCS0 |
| bit 7 |  |  |  |  |  |  |  |

bit 7 ADFM: A/D Result Format Select bit
1 = Right justified
$0=$ Left justified
bit $6 \quad$ Unimplemented: Read as ' 0 '
bit 5-3 ACQT2:ACQTO: A/D Acquisition Time Select bits
$111=20$ TAD
$110=16$ TAD
$101=12$ TAD
$100=8$ TAD
$011=6$ TAD
$010=4$ TAD
$001=2$ TAD
$000=0$ TAD $^{(1)}$
bit 2-0 ADCS2:ADCSO: A/D Conversion Clock Select bits
$111=$ FRC (clock derived from A/D RC oscillator) ${ }^{(1)}$
$110=\mathrm{FOSC} / 64$
$101=$ Fosc/16
100 = Fosc/4
$011=$ FRC (clock derived from A/D RC oscillator) ${ }^{(1)}$
$010=$ FOSC/32
$001=\mathrm{Fosc} / 8$
$000=\mathrm{Fosc} / 2$
Note 1: If the A/D FRC clock source is selected, a delay of one Tcy (instruction cycle) is added before the A/D clock starts. This allows the SLEEP instruction to be executed before starting a conversion.

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-\mathrm{n}=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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The analog reference voltage is software selectable to either the device's positive and negative supply voltage (VDD and Vss), or the voltage level on the RA3/AN3/ Vref+ and RA2/AN2/Vref-/CVREF pins.
The A/D converter has a unique feature of being able to operate while the device is in Sleep mode. To operate in Sleep, the A/D conversion clock must be derived from the A/D's internal RC oscillator.
The output of the sample and hold is the input into the converter, which generates the result via successive approximation.

A device Reset forces all registers to their Reset state. This forces the A/D module to be turned off and any conversion in progress is aborted.
Each port pin associated with the A/D converter can be configured as an analog input, or as a digital I/O. The ADRESH and ADRESL registers contain the result of the $A / D$ conversion. When the $A / D$ conversion is complete, the result is loaded into the ADRESH:ADRESL register pair, the GO/DONE bit (ADCONO register) is cleared and A/D Interrupt Flag bit, ADIF, is set. The block diagram of the $A / D$ module is shown in Figure 17-1.

## FIGURE 17-1: A/D BLOCK DIAGRAM



Note 1: Channels AN5 through AN7 are not available in 28-pin devices.
2: I/O pins have diode protection to VDD and Vss.

The value in the ADRESH:ADRESL registers is not modified for a Power-on Reset. The ADRESH:ADRESL registers will contain unknown data after a Power-on Reset.
After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as an input. To determine acquisition time, see Section 17.1 "A/D Acquisition Requirements". After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time can be programmed to occur between setting the GO/DONE bit and the actual start of the conversion.
The following steps should be followed to perform an A/D conversion:

1. Configure the A/D module:

- Configure analog pins, voltage reference and digital I/O (ADCON1)
- Select A/D input channel (ADCONO)
- Select A/D acquisition time (ADCON2)
- Select A/D conversion clock (ADCON2)
- Turn on A/D module (ADCONO)

2. Configure $A / D$ interrupt (if desired):

- Clear ADIF bit
- Set ADIE bit
- Set GIE bit

3. Wait the required acquisition time (if required).
4. Start conversion:

- Set GO/ $\overline{\mathrm{DONE}}$ bit (ADCONO register)

5. Wait for $A / D$ conversion to complete, by either:

- Polling for the GO/DONE bit to be cleared

OR

- Waiting for the $A / D$ interrupt

6. Read A/D Result registers (ADRESH:ADRESL); clear bit ADIF, if required.
7. For next conversion, go to step 1 or step 2, as required. The $A / D$ conversion time per bit is defined as TAD. A minimum wait of 2 TAD is required before the next acquisition starts.

FIGURE 17-2: A/D TRANSFER FUNCTION


FIGURE 17-3: ANALOG INPUT MODEL


Legend: CPIN = Input Capacitance
VT $\quad=$ Threshold Voltage
ILEAKAGE = Leakage Current at the pin due to various junctions
RIC $\quad=$ Interconnect Resistance
SS = Sampling Switch
CHOLD = Sample/Hold Capacitance (from DAC)
Rss = Sampling Switch Resistance

VDD | 6 V |  |  |
| :--- | :--- | :--- |
| 5 V |  |  |
| 4 V |  |  |
| 3 V |  |  |
| 2 V |  |  |
|  |  |  |
| Sampling Switch $(\mathrm{k} \Omega$ |  |  |

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### 17.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 17-3. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD). The source impedance affects the offset voltage at the analog input (due to pin leakage current). The maximum recommended impedance for analog sources is $2.5 \mathrm{k} \Omega$. After the analog input channel is selected (changed), the channel must be sampled for at least the minimum acquisition time before starting a conversion.

Note: When the conversion is started, the holding capacitor is disconnected from the input pin.

To calculate the minimum acquisition time, Equation 17-1 may be used. This equation assumes that $1 / 2 \mathrm{LSb}$ error is used ( 1024 steps for the A/D). The $1 / 2$ LSb error is the maximum error allowed for the A/D to meet its specified resolution.
Example 17-3 shows the calculation of the minimum required acquisition time TACQ. This calculation is based on the following application system assumptions:

| ChoLd | $=25 \mathrm{pF}$ |
| :--- | :--- |
| Rs | $=2.5 \mathrm{k} \Omega$ |
| Conversion Error | $\leq 1 / 2 \mathrm{LSb}$ |
| VdD | $=5 \mathrm{~V} \rightarrow$ Rss $=2 \mathrm{k} \Omega$ |
| Temperature | $=85^{\circ} \mathrm{C}$ (system max. |

## EQUATION 17-1: ACQUISITION TIME

TACQ $=$ Amplifier Settling Time + Holding Capacitor Charging Time + Temperature Coefficient

$$
=\mathrm{TAMP}+\mathrm{TC}+\mathrm{TCOFF}
$$

EQUATION 17-2: A/D MINIMUM CHARGING TIME

```
VHOLD = (VREF - (VREF/2048)) • (1- e (-TC/CHOLD(RIC + RSS + RS))}
or
Tc = -(CHOLD)(RIC + RSS + Rs) }\operatorname{ln}(1/2048
```


## EQUATION 17-3: CALCULATING THE MINIMUM REQUIRED ACQUISITION TIME

| TACQ | $=$ TAMP $+\mathrm{TC}+$ TCOFF |
| ---: | :--- |
| TAMP | $=0.2 \mu \mathrm{~s}$ |
| TCOFF | $=\left(\mathrm{Temp}-25^{\circ} \mathrm{C}\right)\left(0.02 \mu \mathrm{~s} /{ }^{\circ} \mathrm{C}\right)$ |
|  | $\left(85^{\circ} \mathrm{C}-25^{\circ} \mathrm{C}\right)\left(0.02 \mu \mathrm{~s} /{ }^{\circ} \mathrm{C}\right)$ |
|  | $1.2 \mu \mathrm{~s}$ |

Temperature coefficient is only required for temperatures $>25^{\circ} \mathrm{C}$. Below $25^{\circ} \mathrm{C}$, $\mathrm{TCOFF}=0 \mathrm{~ms}$.
TC $\quad=\quad-($ CHOLD $)($ RIC + RSS + RS $) \ln (1 / 2047) \mu \mathrm{s}$
$-(25 \mathrm{pF})(1 \mathrm{k} \Omega+2 \mathrm{k} \Omega+2.5 \mathrm{k} \Omega) \ln (0.0004883) \mu \mathrm{s}$
$1.05 \mu \mathrm{~s}$
TACQ $=0.2 \mu \mathrm{~s}+1 \mu \mathrm{~s}+1.2 \mu \mathrm{~s}$
$2.4 \mu \mathrm{~s}$

### 17.2 Selecting and Configuring Acquisition Time

The ADCON2 register allows the user to select an acquisition time that occurs each time the GO/DONE bit is set. It also gives users the option to use an automatically determined acquisition time.

Acquisition time may be set with the ACQT2:ACQT0 bits (ADCON2<5:3>), which provides a range of 2 to 20 TAD. When the GO/DONE bit is set, the A/D module continues to sample the input for the selected acquisition time, then automatically begins a conversion. Since the acquisition time is programmed, there may be no need to wait for an acquisition time between selecting a channel and setting the GO/DONE bit.
Manual acquisition is selected when ACQT2:ACQT0 $=000$. When the GO/DONE bit is set, sampling is stopped and a conversion begins. The user is responsible for ensuring the required acquisition time has passed between selecting the desired input channel and setting the GO/DONE bit. This option is also the default Reset state of the ACQT2:ACQT0 bits and is compatible with devices that do not offer programmable acquisition times.
In either case, when the conversion is completed, the GO/DONE bit is cleared, the ADIF flag is set and the A/D begins sampling the currently selected channel again. If an acquisition time is programmed, there is nothing to indicate if the acquisition time has ended or if the conversion has begun.

### 17.3 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires 11 TAD per 10-bit conversion. The source of the A/D conversion clock is software selectable. There are seven possible options for TAD:

- 2 Tosc
- 4 Tosc
- 8 Tosc
- 16 Tosc
- 32 Tosc
- 64 Tosc
- Internal RC Oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be as short as possible, but greater than the minimum TAD (see parameter 130 for more information).
Table 17-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

TABLE 17-1: TAD vs. DEVICE OPERATING FREQUENCIES

| AD Clock Source (TAD) |  | Maximum Device Frequency |  |
| :---: | :---: | :---: | :---: |
| Operation | ADCS2:ADCS0 | PIC18F2XJ10/4XJ10 | PIC18LF2XJ10/4XJ10(4) |
| 2 Tosc | 000 | 2.86 MHz | 1.43 MHz |
| 4 Tosc | 100 | 5.71 MHz | 2.86 MHz |
| 8 Tosc | 001 | 11.43 MHz | 5.72 MHz |
| 16 Tosc | 101 | 22.86 MHz | 11.43 MHz |
| 32 Tosc | 010 | 40.0 MHz | 22.86 MHz |
| 64 Tosc | 110 | 40.0 MHz | 40.0 MHz |
| RC $^{(\mathbf{3})}$ | $x 11$ | $1.00 \mathrm{MHz}{ }^{(\mathbf{1 )}}$ | 1.00 MHz |

Note 1: The RC source has a typical TAD time of $1.2 \mu \mathrm{~s}$.
2: The RC source has a typical TAD time of $2.5 \mu \mathrm{~s}$.
3: For device frequencies above 1 MHz , the device must be in Sleep for the entire conversion or the $A / D$ accuracy may be out of specification.
4: Low-power (PIC18LF2XJ10/4XJ10) devices only.

### 17.4 Operation in Power-Managed Modes

The selection of the automatic acquisition time and $A / D$ conversion clock is determined in part by the clock source and frequency while in a power-managed mode.
If the $A / D$ is expected to operate while the device is in a power-managed mode, the ACQT2:ACQT0 and ADCS2:ADCS0 bits in ADCON2 should be updated in accordance with the clock source to be used in that mode. After entering the mode, an A/D acquisition or conversion may be started. Once started, the device should continue to be clocked by the same clock source until the conversion has been completed.

If desired, the device may be placed into the corresponding Idle mode during the conversion. If the device clock frequency is less than 1 MHz , the $A / D$ RC clock source should be selected.
Operation in the Sleep mode requires the A/D FRC clock to be selected. If bits ACQT2:ACQT0 are set to ' 000 ' and a conversion is started, the conversion will be delayed one instruction cycle to allow execution of the SLEEP instruction and entry to Sleep mode. The IDLEN bit ( $\mathrm{OSCCON}<7>$ ) must have already been cleared prior to starting the conversion.

### 17.5 A/D Converter Calibration

The A/D converter in the PIC18F45J10 family of devices includes a self-calibration feature which compensates for any offset generated within the module. The calibration process is automated and is initiated by setting the ADCAL bit (ADCONO<7>). The next time the GO/ $\overline{\mathrm{DONE}}$ bit is set, the module will perform a "dummy" conversion (that is, with reading none of the input channels) and store the resulting value internally to compensate for offset. Thus, subsequent offsets will be compensated.
The calibration process assumes that the device is in a relatively steady-state operating condition. If $A / D$ calibration is used, it should be performed after each device Reset, or if there are other major changes in operating conditions.

### 17.6 Configuring Analog Port Pins

The ADCON1, TRISA, TRISB and TRISE registers all configure the A/D port pins. The port pins needed as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VoH or Vol) will be converted.
The $A / D$ operation is independent of the state of the CHS3:CHSO bits and the TRIS bits.

Note 1: When reading the Port register, all pins configured as analog input channels will read as cleared (a low level). Pins configured as digital inputs will convert as analog inputs. Analog levels on a digitally configured input will be accurately converted.

2: Analog levels on any pin defined as a digital input may cause the digital input buffer to consume current out of the device's specification limits.

3: The PBADEN bit in Configuration Register 3H configures PORTB pins to reset as analog or digital pins by controlling how the PCFG0 bits in ADCON1 are reset.

### 17.7 A/D Conversions

Figure 17-4 shows the operation of the A/D converter after the GO/DONE bit has been set and the ACQT2:ACQT0 bits are cleared. A conversion is started after the following instruction to allow entry into Sleep mode before the conversion begins.
Figure 17-5 shows the operation of the A/D converter after the GO/DONE bit has been set and the ACQT2:ACQT0 bits are set to ' 010 ' and selecting a 4 TAD acquisition time before the conversion starts.

Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. This means the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers).
After the A/D conversion is completed or aborted, a 2 TAD wait is required before the next acquisition can be started. After this wait, acquisition on the selected channel is automatically started.

Note: The GO/DONE bit should NOT be set in the same instruction that turns on the A/D.

FIGURE 17-4: A/D CONVERSION TAD CYCLES (ACQT<2:0> $=000$, TACQ $=0$ )


FIGURE 17-5: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 010, TACQ = 4 TAD)


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### 17.8 Use of the CCP2 Trigger

An A/D conversion can be started by the Special Event Trigger of the CCP2 module. This requires that the CCP2M3:ССР2M0 bits (CCP2CON<3:0>) be programmed as ' 1011 ' and that the $A / D$ module is enabled (ADON bit is set). When the trigger occurs, the GO/DONE bit will be set, starting the A/D acquisition and conversion and the Timer1 counter will be reset to zero. Timer1 is reset to automatically repeat the A/D acquisition period with minimal software overhead
(moving ADRESH:ADRESL to the desired location). The appropriate analog input channel must be selected and the minimum acquisition period is either timed by the user, or an appropriate TACQ time selected before the Special Event Trigger sets the GO/DONE bit (starts a conversion).
If the $A / D$ module is not enabled (ADON is cleared), the Special Event Trigger will be ignored by the A/D module, but will still reset the Timer1 counter.

TABLE 17-2: REGISTERS ASSOCIATED WITH A/D OPERATION

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 43 |
| PIR1 | PSPIF ${ }^{(1)}$ | ADIF | RCIF | TXIF | SSP1IF | CCP1IF | TMR2IF | TMR1IF | 45 |
| PIE1 | PSPIE ${ }^{(1)}$ | ADIE | RCIE | TXIE | SSP1IE | CCP1IE | TMR2IE | TMR1IE | 45 |
| IPR1 | PSPIP(1) | ADIP | RCIP | TXIP | SSP1IP | CCP1IP | TMR2IP | TMR1IP | 45 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 45 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 45 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 45 |
| ADRESH | A/D Result Register High Byte |  |  |  |  |  |  |  | 44 |
| ADRESL | A/D Result Register Low Byte |  |  |  |  |  |  |  | 44 |
| ADCON0 | ADCAL | - | CHS3 | CHS2 | CHS1 | CHSO | GO/DONE | ADON | 44 |
| ADCON1 | - | - | VCFG1 | VCFG0 | PCFG3 | PCFG2 | PCFG1 | PCFG0 | 44 |
| ADCON2 | ADFM | - | ACQT2 | ACQT1 | ACQT0 | ADCS2 | ADCS1 | ADCS0 | 44 |
| PORTA | - | - | RA5 | - | RA3 | RA2 | RA1 | RAO | 46 |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISA0 | 46 |
| PORTB | RB7 | RB6 | RB5 | RB4 | RB3 | RB2 | RB1 | RB0 | 46 |
| TRISB | PORTB Data Direction Control Register |  |  |  |  |  |  |  | 46 |
| LATB | PORTB Data Latch Register (Read and Write to Data Latch) |  |  |  |  |  |  |  | 46 |
| PORTE ${ }^{(1)}$ | - | - | - | - | - | RE2 | RE1 | RE0 | 46 |
| TRISE ${ }^{(1)}$ | IBF | OBF | IBOV | PSPMODE | - | TRISE2 | TRISE1 | TRISE0 | 46 |
| LATE ${ }^{(1)}$ | - | - | - | - | - | PORTE D <br> (Read and | ta Latch Reg Write to Dat | ister Latch) | 46 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used for A/D conversion.
Note 1: These registers and/or bits are not implemented on 28 -pin devices and should be read as ' 0 '.

### 18.0 COMPARATOR MODULE

The analog comparator module contains two comparators that can be configured in a variety of ways. The inputs can be selected from the analog inputs multiplexed with pins RA0 through RA5, as well as the on-chip voltage reference (see Section 19.0 "Comparator Voltage Reference Module"). The digital outputs (normal or inverted) are available at the pin level and can also be read through the control register.

The CMCON register (Register 18-1) selects the comparator input and output configuration. Block diagrams of the various comparator configurations are shown in Figure 18-1.

## REGISTER 18-1: CMCON: COMPARATOR CONTROL REGISTER

| R-0 | R-0 | R/W-0 | R/W-0 | R/W-0 | R/W-1 | R/W-1 | R/W-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| C2OUT | C1OUT | C2INV | C1INV | CIS | CM2 | CM1 | CM0 |
| bit 7 |  |  |  |  |  |  |  |


| bit 7 | C2OUT: Comparator 2 Output bit |
| :---: | :---: |
|  | When C2INV $=0$ : |
|  | 1 = C2 VIN+ > C2 Vin- |
|  | $0=\mathrm{C} 2 \mathrm{VIN}+<\mathrm{C} 2 \mathrm{VIN}-$ |
|  | When C2INV = 1: |
|  | 1 = C2 Vin+ < C2 Vin- |
|  | $0=\mathrm{C} 2 \mathrm{VIN}+>\mathrm{C} 2 \mathrm{VIN}-$ |
| bit 6 | C1OUT: Comparator 1 Output bit |
|  | When C1INV = 0: |
|  | 1 = C1 Vin+ > C1 Vin- |
|  | $0=\mathrm{C} 1 \mathrm{VIN}+<\mathrm{C} 1 \mathrm{VIN}-$ |
|  | When C1INV = 1: |
|  | 1 = C1 VIN+ < C1 Vin- |
|  | $0=$ C1 VIN+ > C1 VIN- |

bit $5 \quad$ C2INV: Comparator 2 Output Inversion bit
1 = C2 output inverted
$0=\mathrm{C} 2$ output not inverted
bit $4 \quad$ C1INV: Comparator 1 Output Inversion bit
1 = C1 output inverted
$0=$ C1 output not inverted
bit 3 CIS: Comparator Input Switch bit
When CM2:CMO = 110:
1 = C1 VIN- connects to RA3/AN3/VREF+
C2 VIN- connects to RA2/AN2/VREF-/CVREF
$0=$ C1 VIN- connects to RAO/ANO
C2 Vin- connects to RA1/AN1
bit 2-0 CM2:CMO: Comparator Mode bits
Figure 18-1 shows the Comparator modes and the CM2:CM0 bit settings.

| Legend: |  |  |
| :--- | :--- | :--- |
| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value at POR | $' 1$ ' = Bit is set | $' 0$ ' = Bit is cleared $\quad x=$ Bit is unknown |

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### 18.1 Comparator Configuration

There are eight modes of operation for the comparators, shown in Figure 18-1. Bits CM2:CM0 of the CMCON register are used to select these modes. The TRISA register controls the data direction of the comparator pins for each mode. If the Comparator mode is
changed, the comparator output level may not be valid for the specified mode change delay shown in Section 23.0 "Electrical Characteristics".

Note: Comparator interrupts should be disabled during a Comparator mode change; otherwise, a false interrupt may occur.

FIGURE 18-1: COMPARATOR I/O OPERATING MODES

| Comparators Reset $\text { CM2:CM0 = } 000$ | Comparators Off (POR Default Value) $\text { CM2:CM0 = } 111$ |
| :---: | :---: |
| RAO/ANO A $\qquad$ VIn- | $\text { RAO/ANO } \quad \mathrm{D} \quad \text { VIN- }$ |
|  | RA3/AN3/ $\begin{aligned} & \text { VREF }+\end{aligned} \quad$ VIN+ $+C 1$$\quad$ Off (Read as '0') |
|  | RA1/AN1 $\quad \mathrm{D}$ VIN- |
| $\begin{aligned} & \text { RA2/AN2/ } \frac{\mathrm{A}}{\mathrm{~A}} \mathrm{VIN}+_{+}+\mathrm{C} 2> \\ & \text { VREF-/CVREF } \end{aligned}$ | $\underset{\text { RREF-/CVREF }}{\text { RA2/AN2/ }} \stackrel{\mathrm{D}}{=} \mathrm{VIN+}+\mathrm{C} 2+$ Off (Read as '0') |
| Two Independent Comparators CM2:CM0 = 010 | Two Independent Comparators with Outputs CM2:CM0 = 011 |
| RAO/ANO A VIN- |  |
| $\begin{aligned} & \text { RA3/AN3/ A } \begin{array}{l} \text { VIN }+ \\ \text { VREF+ } \end{array}+\mathrm{C} 1>-\mathrm{C} 1 \mathrm{OUT} \end{aligned}$ | $\begin{aligned} & \text { RA3/AN3/ } \\ & \text { VREF }+\end{aligned}$ RIN $+\infty \quad$ C1OUT |
| VIN- | RB5/KBl1/ |
|  | RA1/AN1 |
| VREF-/CVREF | RA2/AN2/ A$\qquad$ Vin+ Vref-/CVref RA5/AN4/ $\overline{\text { SS1 }} /$ C2OUT* $^{*}$ |
|  |  |
| Two Common Reference Comparators CM2:CMO = 100 | Two Common Reference Comparators with Outputs CM2:CMO = 101 |
| RAO/ANO A <br> VIN- | RAO/ANO |
|  | RA3/AN3/ VREF+ A $\quad$ Vin $+\infty$ C1 |
| I | RB5/KBI1/ TOCKI/C1OUT* |
| RA1/AN1 A | $\begin{array}{ll\|l} \text { RA1/AN1 } & \text { A } & \text { VIN- }^{-} \\ \hline \end{array}$ |
| Vref-/CVref |  |
|  |  |
| One Independent Comparator with Output CM2:CMO = 001 | Four Inputs Multiplexed to Two Comparators $\text { CM2:CMO = } 110$ |
| RAO/ANO A VIN- | $\text { RAO/ANO } \xrightarrow{\mathrm{A}} \mathrm{CIS}=0 \quad \mathrm{VIN}-$ |
|  |  |
| RB5/KBI1/T0CKI/C1OUT* | $\text { /AN1 } \quad \mathrm{A} 0 \quad \text { Ic }$ |
| $\text { RA1/AN1 } \xrightarrow{D}$ |  |
|  |  |
| $A=$ Analog Input, port reads zeros always $\quad D=$ Digital Input $\quad$ CIS (CMCON $<3>$ ) is the Comparator Input Switch <br> * Setting the TRISA<5> bit will disable the comparator outputs by configuring the pins as inputs. |  |

### 18.2 Comparator Operation

A single comparator is shown in Figure 18-2, along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 18-2 represent the uncertainty, due to input offsets and response time.

### 18.3 Comparator Reference

Depending on the comparator operating mode, either an external or internal voltage reference may be used. The analog signal present at VIN - is compared to the signal at VIN+ and the digital output of the comparator is adjusted accordingly (Figure 18-2).

FIGURE 18-2: SINGLE COMPARATOR


### 18.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSs and VDD and can be applied to either pin of the comparator(s).

### 18.3.2 INTERNAL REFERENCE SIGNAL

The comparator module also allows the selection of an internally generated voltage reference from the comparator voltage reference module. This module is described in more detail in Section 19.0 "Comparator Voltage Reference Module".
The internal reference is only available in the mode where four inputs are multiplexed to two comparators (CM2:CM0 = 110). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

### 18.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise, the maximum delay of the comparators should be used (see Section 23.0 "Electrical Characteristics").

### 18.5 Comparator Outputs

The comparator outputs are read through the CMCON register. These bits are read-only. The comparator outputs may also be directly output to the RB5 and RA5 I/O pins. When enabled, multiplexors in the output path of the RB5 and RA5 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 18-3 shows the comparator output block diagram.
The TRISA bits will still function as an output enable/ disable for the RB5 and RA5 pins while in this mode.
The polarity of the comparator outputs can be changed using the C2INV and C1INV bits (CMCON<5:4>).

Note 1: When reading the Port register, all pins configured as analog inputs will read as a ' 0 '. Pins configured as digital inputs will convert an analog input according to the Schmitt Trigger input specification.
2: Analog levels on any pin defined as a digital input may cause the input buffer to consume more current than is specified.

FIGURE 18-3: COMPARATOR OUTPUT BLOCK DIAGRAM


### 18.6 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from $\mathrm{CMCON}<7$ :6>, to determine the actual change that occurred. The CMIF bit (PIR2<6>) is the Comparator Interrupt Flag. The CMIF bit must be reset by clearing it. Since it is also possible to write a ' 1 ' to this register, a simulated interrupt may be initiated.
Both the CMIE bit (PIE2<6>) and the PEIE bit (INTCON $<6>$ ) must be set to enable the interrupt. In addition, the GIE bit (INTCON<7>) must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

> | Note: | If a change in the CMCON register |
| :--- | :--- |
| (C1OUT or C2OUT) should occur when a |  |
| read operation is being executed (start of |  |
| the Q2 cycle), then the CMIF (PIR2 |  |
|  | register) interrupt flag may not get set. |

The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:
a) Any read or write of CMCON will end the mismatch condition.
b) Clear flag bit CMIF.

A mismatch condition will continue to set flag bit CMIF. Reading CMCON will end the mismatch condition and allow flag bit CMIF to be cleared.

### 18.7 Comparator Operation During Sleep

When a comparator is active and the device is placed in Sleep mode, the comparator remains active and the interrupt is functional if enabled. This interrupt will wake-up the device from Sleep mode, when enabled. Each operational comparator will consume additional current, as shown in the comparator specifications. To minimize power consumption while in Sleep mode, turn off the comparators $(\mathrm{CM2}: \mathrm{CM0}=111)$ before entering Sleep. If the device wakes up from Sleep, the contents of the CMCON register are not affected.

### 18.8 Effects of a Reset

A device Reset forces the CMCON register to its Reset state, causing the comparator modules to be turned off (CM2:CM0 = 111). However, the input pins (RAO through RA3) are configured as analog inputs by default on device Reset. The I/O configuration for these pins is determined by the setting of the PCFG3:PCFG0 bits (ADCON1<3:0>). Therefore, device current is minimized when analog inputs are present at Reset time.

### 18.9 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 18-4. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and Vss. The analog input, therefore, must be between Vss and VDD. If the input voltage deviates from this
range by more than 0.6 V in either direction, one of the diodes is forward biased and a latch-up condition may occur. A maximum source impedance of $10 \mathrm{k} \Omega$ is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

FIGURE 18-4: COMPARATOR ANALOG INPUT MODEL


TABLE 18-1: REGISTERS ASSOCIATED WITH COMPARATOR MODULE

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CMCON | C2OUT | C1OUT | C2INV | C1INV | CIS | CM2 | CM1 | CMO | 45 |
| CVRCON | CVREN | CVROE | CVRR | CVRSS | CVR3 | CVR2 | CVR1 | CVR0 | 45 |
| INTCON | GIE/GIEH | PEIE/GIEL | TMROIE | INTOIE | RBIE | TMROIF | INTOIF | RBIF | 46 |
| PIR2 | OSCFIF | CMIF | - | - | BCL1IF | - | - | CCP2IF | 45 |
| PIE2 | OSCFIE | CMIE | - | - | BCL1IE | - | - | CCP2IE | 45 |
| IPR2 | OSCFIP | CMIP | - | - | BCL1IP | - | - | CCP2IP | 45 |
| PORTA | - | - | RA5 | - | RA3 | RA2 | RA1 | RAO | 46 |
| LATA | - | - | PORTA Data Latch Register (Read and Write to Data Latch) |  |  |  |  |  | 46 |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISAO | 46 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are unused by the comparator module.

## PIC18F45J10 FAMILY

NOTES:

### 19.0 COMPARATOR VOLTAGE REFERENCE MODULE

The comparator voltage reference is a 16-tap resistor ladder network that provides a selectable reference voltage. Although its primary purpose is to provide a reference for the analog comparators, it may also be used independently of them.
A block diagram of the module is shown in Figure 19-1. The resistor ladder is segmented to provide two ranges of CVREF values and has a power-down function to conserve power when the reference is not being used. The module's supply reference can be provided from either device VDD/VsS or an external voltage reference.

### 19.1 Configuring the Comparator Voltage Reference

The voltage reference module is controlled through the CVRCON register (Register 19-1). The comparator voltage reference provides two ranges of output voltage, each with 16 distinct levels. The range to be
used is selected by the CVRR bit (CVRCON<5>). The primary difference between the ranges is the size of the steps selected by the CVreF Selection bits (CVR3:CVR0), with one range offering finer resolution. The equations used to calculate the output of the comparator voltage reference are as follows:

```
If CVRR = 1:
CVREF = ((CVR3:CVRO)/24) x CVRSRC
If CVRR = 0:
CVREF = (CVRSRC x 1/4) + (((CVR3:CVR0)/32) x
CVRSRC)
```

The comparator reference supply voltage can come from either VDD and Vss, or the external Vref+ and Vref- that are multiplexed with RA2 and RA3. The voltage source is selected by the CVRSS bit (CVRCON<4>).
The settling time of the comparator voltage reference must be considered when changing the CVREF output (see Table 23-3 in Section 23.0 "Electrical Characteristics").

## REGISTER 19-1: CVRCON: COMPARATOR VOLTAGE REFERENCE CONTROL REGISTER

| R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CVREN | CVROE ${ }^{(1)}$ | CVRR | CVRSS | CVR3 | CVR2 | CVR1 | CVR0 |

$$
\text { bit } 7
$$

| bit 7 | CVREN: Comparator Voltage Reference Enable bit <br> $1=$ CVREF circuit powered on <br> $0=$ CVREF circuit powered down |
| :---: | :---: |
| bit 6 | CVROE: Comparator VREF Output Enable bit ${ }^{(1)}$ <br> $1=$ CVREF voltage level is also output on the RA2/AN2/VREF-/CVREF pin <br> $0=$ CVREF voltage is disconnected from the RA2/AN2/VREF-/CVREF pin <br> Note 1: CVROE overrides the TRISA<2> bit setting. |
| bit 5 | CVRR: Comparator Vref Range Selection bit $\begin{aligned} & 1=0 \text { to } 0.667 \text { CVRSRC, with CVRSRC/24 step size (low range) } \\ & 0=0.25 \text { CVRSRC to } 0.75 \text { CVRSRC, with CVRSRC/32 step size (high range) } \end{aligned}$ |
| bit 4 | CVRSS: Comparator Vref Source Selection bit $\begin{aligned} & 1=\text { Comparator reference source, CVRSRC }=(\text { VREF }+)-(\text { VREF- }) \\ & 0=\text { Comparator reference source, CVRSRC }=\text { VDD }- \text { VSS } \end{aligned}$ |
| bit 3-0 | CVR3:CVR0: Comparator Vref Value Selection bits $(0 \leq(C V R 3: C V R 0) \leq 15)$ When CVRR = 1: <br> CVREF = ((CVR3:CVRO)/24) • (CVRSRC) <br> When CVRR = 0: <br> CVREF $=($ CVRSRC/4 $)+((C V R 3: C V R 0) / 32) \bullet(C V R S R C)$ |

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-\mathrm{n}=$ Value at POR | $' 1 '=$ Bit is set | $' 0 '=$ Bit is cleared $\quad x=$ Bit is unknown |

FIGURE 19-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM
VREF + CDC CVRSS = 1

### 19.2 Voltage Reference Accuracy/Error

The full range of voltage reference cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 19-1) keep CVREF from approaching the reference source rails. The voltage reference is derived from the reference source; therefore, the CVREF output changes with fluctuations in that source. The tested absolute accuracy of the voltage reference can be found in Section 23.0 "Electrical Characteristics".

### 19.3 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the CVRCON register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

### 19.4 Effects of a Reset

A device Reset disables the voltage reference by clearing bit, CVREN (CVRCON<7>). This Reset also disconnects the reference from the RA2 pin by clearing bit, CVROE (CVRCON<6>) and selects the high-voltage range by clearing bit, CVRR (CVRCON $<5>$ ). The CVR value select bits are also cleared.

### 19.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RA2 pin if the CVROE bit is set. Enabling the voltage reference output onto RA2 when it is configured as a digital input will increase current consumption. Connecting RA2 as a digital output with CVRSS enabled will also increase current consumption.
The RA2 pin can be used as a simple D/A output with limited drive capability. Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to Vref. Figure 19-2 shows an example buffering technique.

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FIGURE 19-2: COMPARATOR VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE


Note 1: R is dependent upon the voltage reference configuration bits, $C V R C O N<3: 0>$ and $C V R C O N<5>$.

TABLE 19-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset <br> Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CVRCON | CVREN | CVROE | CVRR | CVRSS | CVR3 | CVR2 | CVR1 | CVR0 | 45 |
| CMCON | C2OUT | C1OUT | C2INV | C1INV | CIS | CM2 | CM1 | CM0 | 45 |
| TRISA | - | - | TRISA5 | - | TRISA3 | TRISA2 | TRISA1 | TRISA0 | 46 |

Legend: Shaded cells are not used with the comparator voltage reference.

## PIC18F45J10 FAMILY

NOTES:

### 20.0 SPECIAL FEATURES OF THE CPU

PIC18F45J10 family devices include several features intended to maximize reliability and minimize cost through elimination of external components. These are:

- Oscillator Selection
- Resets:
- Power-on Reset (POR)
- Power-up Timer (PWRT)
- Oscillator Start-up Timer (OST)
- Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- Fail-Safe Clock Monitor
- Two-Speed Start-up
- Code Protection
- In-Circuit Serial Programming

The oscillator can be configured for the application depending on frequency, power, accuracy and cost. All of the options are discussed in detail in Section 2.0 "Oscillator Configurations".
A complete discussion of device Resets and interrupts is available in previous sections of this data sheet.

In addition to their Power-up and Oscillator Start-up Timers provided for Resets, the PIC18F45J10 family of devices have a configurable Watchdog Timer which is controlled in software.
The inclusion of an internal RC oscillator also provides the additional benefits of a Fail-Safe Clock Monitor (FSCM) and Two-Speed Start-up. FSCM provides for background monitoring of the peripheral clock and automatic switchover in the event of its failure. Two-Speed Start-up enables code to be executed almost immediately on start-up, while the primary clock source completes its start-up delays.
All of these features are enabled and configured by setting the appropriate Configuration register bits.

### 20.1 Configuration Bits

The configuration bits can be programmed (read as ' 0 ') or left unprogrammed (read as ' 1 ') to select various device configurations. These bits are mapped starting at program memory location 300000h. A complete list is shown in Table 20-1. A detailed explanation of the various bit functions is provided in Register 20-1 through Register 20-6.
Note that address 300000h is beyond the user program memory space. In fact, it belongs to the configuration memory space (300000h-3FFFFFh) which can only be accessed using table reads and table writes.

### 20.1.1 CONSIDERATIONS FOR CONFIGURING THE PIC18F45J10 FAMILY DEVICES

Unlike most PIC18 microcontrollers, devices of the PIC18F45J10 family do not use persistent memory registers to store configuration information. The configuration bytes are implemented as volatile memory which means that configuration data must be programmed each time the device is powered up.
Configuration data is stored in the four words at the top of the on-chip program memory space, known as the Flash Configuration Words. It is stored in program memory in the same order shown in Table 20-1, with CONFIG1L at the lowest address and CONFIG3H at the highest. The data is automatically loaded in the proper Configuration registers during device power-up.
When creating applications for these devices, users should always specifically allocate the location of the Flash Configuration Word for configuration data; this is to make certain that program code is not stored in this address when the code is compiled.
The volatile memory cells used for the configuration bits always reset to ' 1 ' on Power-on Resets. For all other type of Reset events, the previously programmed values are maintained and used without reloading from program memory.
The four Most Significant bits of CONFIG1H, CONFIG2H and CONFIG3H in program memory should also be '1111'. This makes these Configuration Words appear to be NOP instructions in the remote event that their locations are ever executed by accident. Since configuration bits are not implemented in the corresponding locations, writing ' 1 's to these locations has no effect on device operation.
To prevent inadvertent configuration changes during code execution, all programmable configuration bits are write-once. After a bit is initially programmed during a power cycle, it cannot be written to again. Changing a device configuration requires a device Reset.

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TABLE 20-1: CONFIGURATION BITS AND DEVICE IDs

| File Name |  | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Default/ Unprogrammed |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 300000h | CONFIG1L | $\overline{\text { DEBUG }}$ | XINST | STVREN | - | - | - | - | WDTEN | 111- ---1 |
| 300001h | CONFIG1H | - ${ }^{(2)}$ | - ${ }^{(2)}$ | $\sim^{(2)}$ | $\sim^{(2)}$ | - ${ }^{(3)}$ | CP0 | - | - | -- x1- |
| 300002h | CONFIG2L | IESO | FCMEN | - | - | - | FOSC2 | FOSC1 | FOSC0 | 11-- -111 |
| 300003h | CONFIG2H | - $^{(2)}$ | $\sim^{(2)}$ | $\sim^{(2)}$ | $-^{(2)}$ | WDTPS3 | WDTPS2 | WDTPS1 | WDTPS0 | -- 1111 |
| 300004h | CONFIG3L | - | - | - | - | - | - | - | - | - -- |
| 300005h | CONFIG3H | - ${ }^{(2)}$ | $\sim^{(2)}$ | - ${ }^{(2)}$ | $-^{(2)}$ | - | - | - | CCP2MX | ---- ---1 |
| 3FFFFEh | DEVID1 | DEV2 | DEV1 | DEV0 | REV4 | REV3 | REV2 | REV1 | REVo | xxxx xxxx ${ }^{(4)}$ |
| 3FFFFFh | DEVID2 | DEV10 | DEV9 | DEV8 | DEV7 | DEV6 | DEV5 | DEV4 | DEV3 | 0001 110x ${ }^{(4)}$ |

Legend: $\mathrm{x}=$ unknown, $\mathrm{u}=$ unchanged, $-=$ unimplemented. Shaded cells are unimplemented, read as ' 0 '.
Note 1: Values reflect the unprogrammed state as received from the factory and following Power-on Resets. In all other Reset states, the configuration bytes maintain their previously programmed states.
2: The value of these bits in program memory should always be ' 1 '. This ensures that the location is executed as a NOP if it is accidentally executed.
3: $\quad$ This bit should always be maintained as ' 0 '.
4: See Register 20-7 and Register 20-8 for DEVID values. These registers are read-only and cannot be programmed by the user.

## REGISTER 20-1: CONFIG1L: CONFIGURATION REGISTER 1 LOW (BYTE ADDRESS 300000h)

| R/WO-1 | R/WO-1 | R/WO-1 | U-0 | U-0 | U-0 | U-0 | R/WO-1 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { DEBUG }}$ | XINST | STVREN | - | - | - | - | WDTEN |
| bit 7 |  |  |  |  |  |  |  |

bit $7 \quad \overline{\text { DEBUG: }}$ : Background Debugger Enable bit
1 = Background debugger disabled; RB6 and RB7 configured as general purpose I/O pins $0=$ Background debugger enabled; RB6 and RB7 are dedicated to In-Circuit Debug
bit $6 \quad$ XINST: Extended Instruction Set Enable bit
1 = Instruction set extension and Indexed Addressing mode enabled
$0=$ Instruction set extension and Indexed Addressing mode disabled (Legacy mode)
bit 5 STVREN: Stack Overflow/Underflow Reset Enable bit
1 = Reset on stack overflow/underflow enabled
$0=$ Reset on stack overflow/underflow disabled
bit 4-1 Unimplemented: Read as ' 0 '
bit $0 \quad$ WDTEN: Watchdog Timer Enable bit
$1=$ WDT enabled
$0=$ WDT disabled (control is placed on SWDTEN bit)

| Legend: |  |
| :--- | :--- | :--- |
| $R=$ Readable bit $\quad$ WO = Write-once bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value when device is unprogrammed | $' 1$ ' = Bit is set $\quad$ ' 0 ' = Bit is cleared |

REGISTER 20-2: CONFIG1H: CONFIGURATION REGISTER 1 HIGH (BYTE ADDRESS 300001h)

bit 7-3 Unimplemented: Read as ' 0 '
bit 2 CPO: Code Protection bit
$1=$ Program memory is not code-protected
$0=$ Program memory is code-protected
bit 1-0 Unimplemented: Read as ' 0 '
Note 1: This bit should always be maintained as ' 0 '.

| Legend: |  |
| :--- | :--- | :--- |
| $R=$ Readable bit $\quad$ WO = Write-once bit | $U=$ Unimplemented bit, read as ' 0 ' |
| $-n=$ Value when device is unprogrammed | $' 1$ ' = Bit is set $\quad$ ' 0 ' = Bit is cleared |

REGISTER 20-3: CONFIG2L: CONFIGURATION REGISTER 2 LOW (BYTE ADDRESS 300002h)

| R/WO-1 | R/WO-1 | U-0 | U-0 | U-0 | R/WO-1 | R/WO-1 | R/WO-1 |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: | ---: |
| IESO | FCMEN | - | - | - | FOSC2 | FOSC1 | FOSC0 |
| bit 7 |  |  |  |  |  |  |  |

bit 7 IESO: Two-Speed Start-up (Internal/External Oscillator Switchover) Control bit
1 = Two-Speed Start-up enabled
0 = Two-Speed Start-up disabled
bit 6 FCMEN: Fail-Safe Clock Monitor Enable bit
1 = Fail-Safe Clock Monitor enabled
$0=$ Fail-Safe Clock Monitor disabled
bit 5-3 Unimplemented: Read as ' 0 '
bit 2 FOSC2: Default/Reset System Clock Select bit
1 = Clock selected by FOSC1:FOSC0 as system clock is enabled when OSCCON<1:0> = 00
$0=$ INTRC enabled as system clock when OSCCON<1:0> $=00$
bit 1-0 FOSC1:FOSC0: Oscillator Selection bits
$11=$ EC oscillator, PLL enabled and under software control, CLKO function on OSC2
$10=\mathrm{EC}$ oscillator, CLKO function on OSC2
$01=$ HS oscillator, PLL enabled and under software control
$00=$ HS oscillator

## Legend:

| $R=$ Readable bit | WO = Write-once bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- | :--- |
| $-n=$ Value when device is unprogrammed | $' 1$ ' = Bit is set | ' 0 ' = Bit is cleared |

REGISTER 20-4: CONFIG2H: CONFIGURATION REGISTER 2 HIGH (BYTE ADDRESS 300003h)

| $\mathrm{U}-0$ | U-0 | U-0 | U-0 | R/WO-1 | R/WO-1 | R/WO-1 | R/WO-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | WDTPS3 | WDTPS2 | WDTPS1 | WDTPS0 |

bit 7-4 Unimplemented: Read as ' 0 '
bit 3-0 WDTPS3:WDTPS0: Watchdog Timer Postscale Select bits

$$
\begin{aligned}
& 1111=1: 32,768 \\
& 1110=1: 16,384 \\
& 1101=1: 8,192 \\
& 1100=1: 4,096 \\
& 1011=1: 2,048 \\
& 1010=1: 1,024 \\
& 1001=1: 512 \\
& 1000=1: 256 \\
& 0111=1: 128 \\
& 0110=1: 64 \\
& 0101=1: 32 \\
& 0100=1: 16 \\
& 0011=1: 8 \\
& 0010=1: 4 \\
& 0001=1: 2 \\
& 0000=1: 1
\end{aligned}
$$

## Legend:

| $R=$ Readable bit | $W O=$ Write-once bit |
| :--- | :--- |
| $-n=$ Value when device is unprogrammed | $U=$ Unimplemented bit, read as ' 0 ' |
|  | 1 ' = Bit is set $\quad ' 0$ ' = Bit is cleared |

REGISTER 20-5: CONFIG3L: CONFIGURATION REGISTER 3 LOW (BYTE ADDRESS 300004h)

bit 7-0 Unimplemented: Read as ' 0 '

Legend:
$R=$ Readable bit $\quad U=$ Unimplemented bit, read as ' 0 '
$-\mathrm{n}=$ Value when device is unprogrammed $\quad$ ' 1 ' = Bit is set $\quad$ ' 0 ' = Bit is cleared

REGISTER 20-6: CONFIG3H: CONFIGURATION REGISTER 3 HIGH (BYTE ADDRESS 300005h)

| U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | R/WO-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | - | - | CCP2MX |
| bit 7 |  |  |  |  |  |  |  |$\quad$| bit 0 |
| :--- | :--- |

bit 7-1 Unimplemented: Read as ' 0 '
bit $0 \quad$ CCP2MX: CCP2 Mux bit
$1=$ CCP2 is multiplexed with RC1
$0=\mathrm{CCP} 2$ is multiplexed with RB3

## Legend:

| $R=$ Readable bit | WO = Write-once bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value when device is unprogrammed | $' 1$ ' $=$ Bit is set | ' 0 ' = Bit is cleared |

REGISTER 20-7: DEVICE ID REGISTER 1 FOR PIC18F45J10 FAMILY DEVICES

| R | R | R | R | R | R | R | R |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DEV2 | DEV1 | DEV0 | REV4 | REV3 | REV2 | REV1 | REV0 |
| bit 7 |  |  |  |  |  |  |  |

bit 7-5 DEV2:DEV0: Device ID bits
011 = PIC18F4XJ10
$010=$ PIC18F2XJ10
001 = PIC18F4XJ10
$000=$ PIC18F2XJ10
Note: Where values for DEV2:DEV0 are shared by more than one device number, the specific device is always identified by using the entire DEV10:DEV0 bit sequence.
bit 4-0 REV4:REV0: Revision ID bits
These bits are used to indicate the device revision.

## Legend:

| $R=$ Read-only bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- |
| $-n=$ Value when device is unprogrammed | $U=$ Unchanged from programmed state |

REGISTER 20-8: DEVICE ID REGISTER 2 FOR PIC18F45J10 FAMILY DEVICES

| R | R | R | R | R | R | R |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DEV10 | DEV9 | DEV8 | DEV7 | DEV6 | DEV5 | DEV4 | DEV3 |
| bit 7 |  |  |  |  |  |  |  |

bit 7-0 DEV10:DEV3: Device ID bits
These bits are used with the DEV2:DEV0 bits in the Device ID Register 1 to identify the part number.
0001 1100 $=$ PIC18FX5J10 devices
0001 1101 $=$ PIC18FX4J10 devices
Note: The values for DEV10:DEV3 may be shared with other device families. The specific device is always identified by using the entire DEV10:DEV0 bit sequence.

## Legend:

| $R=$ Read-only bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- |
| $-n=$ Value when device is unprogrammed | $U=$ Unchanged from programmed state |

### 20.2 Watchdog Timer (WDT)

For PIC18F45J10 family devices, the WDT is driven by the INTRC oscillator. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the INTRC oscillator.
The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexor, controlled by the WDTPS bits in Configuration Register 2H. Available periods range from 4 ms to 131.072 seconds ( 2.18 minutes). The WDT and postscaler are cleared whenever a SLEEP or CLRWDT instruction is executed, or a clock failure (primary or Timer1 oscillator) has occurred.

Note 1: The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
2: When a CLRWDT instruction is executed, the postscaler count will be cleared.

### 20.2.1 CONTROL REGISTER

The WDTCON register (Register 20-9) is a readable and writable register. The SWDTEN bit enables or disables WDT operation.

FIGURE 20-1: WDT BLOCK DIAGRAM


## REGISTER 20-9: WDTCON: WATCHDOG TIMER CONTROL REGISTER

| U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | R/W-0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | - | - | SWDTEN $^{(1)}$ |
| bit 7 |  |  |  |  |  |  |  |

bit 7-1 Unimplemented: Read as ' 0 '
bit 0 SWDTEN: Software Controlled Watchdog Timer Enable bit ${ }^{(1)}$
1 = Watchdog Timer is on
$0=$ Watchdog Timer is off
Note 1: This bit has no effect if the configuration bit, WDTEN, is enabled.

## Legend:

| $R=$ Readable bit | $W=$ Writable bit | $U=$ Unimplemented bit, read as ' 0 ' |
| :--- | :--- | :--- |
| $-n=$ Value at POR | $' 1 '=$ Bit is set | $' 0$ ' $=$ Bit is cleared $\quad x=$ Bit is unknown |

TABLE 20-2: SUMMARY OF WATCHDOG TIMER REGISTERS

| Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset Values <br> on page |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RCON | IPEN | - | - | $\overline{\mathrm{RI}}$ | $\overline{\mathrm{TO}}$ | $\overline{\mathrm{PD}}$ | $\overline{\mathrm{POR}}$ | $\overline{\mathrm{BOR}}$ | 44 |
| WDTCON | - | - | - | - | - | - | - | SWDTEN | 44 |

Legend: - = unimplemented, read as ' 0 '. Shaded cells are not used by the Watchdog Timer.

### 20.3 On-Chip Voltage Regulator

Note: The on-chip voltage regulator is only available in parts designated with an " $F$ ", such as PIC18F45J10.

In parts designated "LF", the microcontroller core can be powered from an external source that is separate from VDD, or it can be powered from an on-chip regulator which derives power from VDD. Both sources use the common Vddcore/Vcap pin.
In "F" devices, a low ESR capacitor must be connected to the VdDCORE/VCAP pin for proper device operation. In parts designated with an "LF" part number (i.e., PIC18LF45J10), power to the core must be supplied on VdDCore/Vcap. It is always good design practice to have sufficient capacitance on all supply pins. Examples are shown in Figure 20-2.

$$
\begin{array}{ll}
\text { Note: } & \text { In parts designated with an "LF", such as } \\
& \text { PIC18LF45J10, VDDCORE must never } \\
& \text { exceed VDD. }
\end{array}
$$

The specifications for core voltage and capacitance are listed in Section 23.4 "AC (Timing) Characteristics".

FIGURE 20-2: CONNECTIONS FOR THE ON-CHIP REGULATOR
PIC18FXXJ10 Devices (Regulator Enabled):


PIC18LFXXJ10 Devices (Regulator Disabled):


OR


### 20.3.1 ON-CHIP REGULATOR AND BOR

PIC18F45J10 family devices (designated with an "F" part number) also have a simple brown-out capability. If the voltage supplied to the regulator is inadequate to maintain a regulated level, the regulator Reset circuitry will generate a BOR Reset. This event is captured by the $\overline{B O R}$ flag bit ( $\mathrm{RCON}<0>$ ).
The operation of the BOR is described in more detail in Section 4.4 "Brown-out Reset (BOR) (PIC18F2XJ10/4XJ10 Devices Only)" and Section 4.4.1 "Detecting BOR". The brown-out voltage levels are specific in Section 23.1 "DC Characteristics".

### 20.3.2 POWER-UP REQUIREMENTS

The on-chip regulator is designed to meet the power-up requirements for the device. While powering up, VDDCORE must never exceed VdD by 0.3 volts.

### 20.4 Two-Speed Start-up

The Two-Speed Start-up feature helps to minimize the latency period, from oscillator start-up to code execution, by allowing the microcontroller to use the INTRC oscillator as a clock source until the primary clock source is available. It is enabled by setting the IESO configuration bit.
Two-Speed Start-up should be enabled only if the primary oscillator mode is HS (Crystal-based) modes. Since the EC mode does not require an OST start-up delay, Two-Speed Start-up should be disabled.
When enabled, Resets and wake-ups from Sleep mode cause the device to configure itself to run from the internal oscillator block as the clock source, following the time-out of the Power-up Timer after a POR Reset is enabled. This allows almost immediate code execution while the primary oscillator starts and the OST is running. Once the OST times out, the device automatically switches to PRI_RUN mode.

In all other power-managed modes, Two-Speed Start-up is not used. The device will be clocked by the currently selected clock source until the primary clock source becomes available. The setting of the IESO bit is ignored.

### 20.4.1 SPECIAL CONSIDERATIONS FOR USING TWO-SPEED START-UP

While using the INTRC oscillator in Two-Speed Start-up, the device still obeys the normal command sequences for entering power-managed modes, including serial SLEEP instructions (refer to Section 3.1.4 "Multiple Sleep Commands"). In practice, this means that user code can change the SCS1:SCS0 bit settings or issue SLEEP instructions before the OST times out. This would allow an application to briefly wake-up, perform routine "housekeeping" tasks and return to Sleep before the device starts to operate from the primary oscillator.

User code can also check if the primary clock source is currently providing the device clocking by checking the status of the OSTS bit ( $O S C C O N<3>$ ). If the bit is set, the primary oscillator is providing the clock. Otherwise, the internal oscillator block is providing the clock during wake-up from Reset or Sleep mode.

FIGURE 20-3: TIMING TRANSITION FOR TWO-SPEED START-UP (INTRC)


Note 1: TOST = 1024 Tosc. These intervals are not shown to scale.

### 20.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the microcontroller to continue operation in the event of an external oscillator failure by automatically switching the device clock to the internal oscillator block. The FSCM function is enabled by setting the FCMEN configuration bit.
When FSCM is enabled, the INTRC oscillator runs at all times to monitor clocks to peripherals and provide a backup clock in the event of a clock failure. Clock monitoring (shown in Figure 20-4) is accomplished by creating a sample clock signal which is the INTRC output divided by 64. This allows ample time between FSCM sample clocks for a peripheral clock edge to occur. The peripheral device clock and the sample clock are presented as inputs to the Clock Monitor latch (CM). The CM is set on the falling edge of the device clock source but cleared on the rising edge of the sample clock.

FIGURE 20-4: FSCM BLOCK DIAGRAM


Clock failure is tested for on the falling edge of the sample clock. If a sample clock falling edge occurs while CM is still set, a clock failure has been detected (Figure 20-5). This causes the following:

- the FSCM generates an oscillator fail interrupt by setting bit OSCFIF (PIR2<7>);
- the device clock source is switched to the internal oscillator block (OSCCON is not updated to show the current clock source - this is the fail-safe condition); and
- the WDT is reset.

During switchover, the postscaler frequency from the internal oscillator block may not be sufficiently stable for timing sensitive applications. In these cases, it may be desirable to select another clock configuration and enter an alternate power-managed mode. This can be done to attempt a partial recovery or execute a controlled shutdown. See Section 3.1.4 "Multiple Sleep Commands" and Section 20.4.1 "Special Considerations for Using Two-Speed Start-up" for more details.

To use a higher clock speed on wake-up, the INTOSC or postscaler clock sources can be selected to provide a higher clock speed by setting bits IRCF2:IRCF0 immediately after Reset. For wake-ups from Sleep, the INTOSC or postscaler clock sources can be selected by setting IRCF2:IRCF0 prior to entering Sleep mode.
The FSCM will detect failures of the primary or secondary clock sources only. If the internal oscillator block fails, no failure would be detected, nor would any action be possible.

### 20.5.1 FSCM AND THE WATCHDOG TIMER

Both the FSCM and the WDT are clocked by the INTRC oscillator. Since the WDT operates with a separate divider and counter, disabling the WDT has no effect on the operation of the INTRC oscillator when the FSCM is enabled.

As already noted, the clock source is switched to the INTRC clock when a clock failure is detected; this may mean a substantial change in the speed of code execution. If the WDT is enabled with a small prescale value, a decrease in clock speed allows a WDT time-out to occur and a subsequent device Reset. For this reason, Fail-Safe Clock Monitor events also reset the WDT and postscaler, allowing it to start timing from when execution speed was changed and decreasing the likelihood of an erroneous time-out.

### 20.5.2 EXITING FAIL-SAFE OPERATION

The fail-safe condition is terminated by either a device Reset or by entering a power-managed mode. On Reset, the controller starts the primary clock source specified in Configuration Register 2H (with the OST oscillator, start-up delays if running in HS mode). The INTRC oscillator provides the device clock until the primary clock source becomes ready (similar to a Two-Speed Start-up). The clock source is then switched to the primary clock (indicated by the OSTS bit in the OSCCON register becoming set). The Fail-Safe Clock Monitor then resumes monitoring the peripheral clock.
The primary clock source may never become ready during start-up. In this case, operation is clocked by the INTRC oscillator. The OSCCON register will remain in its Reset state until a power-managed mode is entered.

FIGURE 20-5: FSCM TIMING DIAGRAM


### 20.5.3 FSCM INTERRUPTS IN POWER-MANAGED MODES

By entering a power-managed mode, the clock multiplexor selects the clock source selected by the OSCCON register. Fail-Safe Monitoring of the power-managed clock source resumes in the power-managed mode.
If an oscillator failure occurs during power-managed operation, the subsequent events depend on whether or not the oscillator failure interrupt is enabled. If enabled (OSCFIF = 1), code execution will be clocked by the INTOSC multiplexor. An automatic transition back to the failed clock source will not occur.
If the interrupt is disabled, subsequent interrupts while in Idle mode will cause the CPU to begin executing instructions while being clocked by the INTOSC source.

### 20.5.4 POR OR WAKE-UP FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited Power-on Reset (POR) or low-power Sleep mode. When the primary device clock is either EC or INTRC modes, monitoring can begin immediately following these events.
For HS mode, the situation is somewhat different. Since the oscillator may require a start-up time considerably longer than the FSCM sample clock time, a false clock failure may be detected. To prevent this, the internal oscillator block is automatically configured as the device clock and functions until the primary clock is stable (the OST timer has timed out). This is identical to Two-Speed Start-up mode. Once the primary clock is stable, the INTRC returns to its role as the FSCM source.
Note: The same logic that prevents false oscillator failure interrupts on POR, or wake from Sleep, will also prevent the detection of the oscillator's failure to start at all following these events. This can be avoided by monitoring the OSTS bit and using a timing routine to determine if the oscillator is taking too long to start. Even so, no oscillator failure interrupt will be flagged.

As noted in Section 20.4.1 "Special Considerations for Using Two-Speed Start-up", it is also possible to select another clock configuration and enter an alternate power-managed mode while waiting for the primary clock to become stable. When the new power-managed mode is selected, the primary clock is disabled.

### 20.6 Program Verification and Code Protection

For all devices in the PIC18F45J10 family of devices, the on-chip program memory space is treated as a single block. Code protection for this block is controlled by one configuration bit, CPO. This bit inhibits external reads and writes to the program memory space. It has no direct effect in normal execution mode.

### 20.6.1 CONFIGURATION REGISTER PROTECTION

The Configuration registers are protected against untoward changes or reads in two ways. The primary protection is the write-once feature of the configuration bits which prevents reconfiguration once the bit has been programmed during a power cycle. To safeguard against unpredictable events, configuration bit changes resulting from individual cell-level disruptions (such as ESD events) will cause a parity error and trigger a device Reset.
The data for the Configuration registers is derived from the Flash Configuration Words in program memory. When the CPO bit is set, the source data for device configuration is also protected as a consequence.

### 20.7 In-Circuit Serial Programming

PIC18F45J10 family microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

### 20.8 In-Circuit Debugger

When the DEBUG configuration bit is programmed to a ' 0 ', the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB ${ }^{\circledR}$ IDE. When the microcontroller has this feature enabled, some resources are not available for general use. Table 20-3 shows which resources are required by the background debugger.

TABLE 20-3: DEBUGGER RESOURCES

| I/O pins: | RB6, RB7 |
| :--- | :--- |
| Stack: | 2 levels |
| Program Memory: | 512 bytes |
| Data Memory: | 32 bytes |

### 21.0 INSTRUCTION SET SUMMARY

PIC18F45J10 family devices incorporate the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions, for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

### 21.1 Standard Instruction Set

The standard PIC18 instruction set adds many enhancements to the previous PICmicro ${ }^{\circledR}$ instruction sets, while maintaining an easy migration from these PICmicro instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.
Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.
The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal operations
- Control operations

The PIC18 instruction set summary in Table 21-2 lists byte-oriented, bit-oriented, literal and control operations. Table 21-1 shows the opcode field descriptions.
Most byte-oriented instructions have three operands:

1. The file register (specified by ' $f$ ')
2. The destination of the result (specified by ' $d$ ')
3. The accessed memory (specified by ' $a$ ')

The file register designator ' $f$ ' specifies which file register is to be used by the instruction. The destination designator ' $d$ ' specifies where the result of the operation is to be placed. If ' $d$ ' is zero, the result is placed in the WREG register. If ' $d$ ' is one, the result is placed in the file register specified in the instruction.
All bit-oriented instructions have three operands:

1. The file register (specified by ' $f$ ')
2. The bit in the file register (specified by 'b')
3. The accessed memory (specified by 'a')

The bit field designator ' $b$ ' selects the number of the bit affected by the operation, while the file register designator ' $f$ ' represents the number of the file in which the bit is located.

The literal instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by ' $k$ ')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '-')
The control instructions may use some of the following operands:
- A program memory address (specified by ' $n$ ')
- The mode of the CALL or RETURN instructions (specified by ' $s$ ')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required
(specified by '-')
All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSbs are ' 1 's. If this second word is executed as an instruction (by itself), it will execute as a NOP.
All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the program counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles, with the additional instruction cycle(s) executed as a NOP.
The double-word instructions execute in two instruction cycles.
One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz , the normal instruction execution time is $1 \mu \mathrm{~s}$. If a conditional test is true, or the program counter is changed as a result of an instruction, the instruction execution time is $2 \mu \mathrm{~s}$. Two-word branch instructions (if true) would take $3 \mu \mathrm{~s}$.
Figure 21-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.
The Instruction Set Summary, shown in Table 21-2, lists the standard instructions recognized by the Microchip Assembler (MPASM ${ }^{\text {TM }}$ ).

Section 21.1.1 "Standard Instruction Set" provides a description of each instruction.

## PIC18F45J10 FAMILY

TABLE 21-1: OPCODE FIELD DESCRIPTIONS

| Field | Description |
| :---: | :---: |
| a | RAM access bit <br> $\mathrm{a}=0$ : RAM location in Access RAM (BSR register is ignored) <br> $a=1$ : RAM bank is specified by BSR register |
| b.bb | Bit address within an 8-bit file register (0 to 7). |
| BSR | Bank Select Register. Used to select the current RAM bank. |
| C, DC, Z, OV, N | ALU Status bits: Carry, Digit Carry, Zero, Overflow, Negative. |
| d | Destination select bit $d=0$ : store result in WREG $d=1$ : store result in file register $f$ |
| dest | Destination: either the WREG register or the specified register file location. |
| f | 8-bit Register file address (00h to FFh) or 2-bit FSR designator (0h to 3h). |
| $\mathrm{f}_{\text {S }}$ | 12-bit Register file address (000h to FFFh). This is the source address. |
| $\mathrm{f}_{\mathrm{d}}$ | 12-bit Register file address (000h to FFFh). This is the destination address. |
| GIE | Global Interrupt Enable bit. |
| k | Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value). |
| label | Label name. |
| mm <br> * <br> * <br> +* | The mode of the TBLPTR register for the table read and table write instructions. Only used with table read and table write instructions: <br> No change to register (such as TBLPTR with table reads and writes) Post-Increment register (such as TBLPTR with table reads and writes) Post-Decrement register (such as TBLPTR with table reads and writes) Pre-Increment register (such as TBLPTR with table reads and writes) |
| n | The relative address (2's complement number) for relative branch instructions or the direct address for Call/Branch and Return instructions. |
| PC | Program Counter. |
| PCL | Program Counter Low Byte. |
| PCH | Program Counter High Byte. |
| PCLATH | Program Counter High Byte Latch. |
| PCLATU | Program Counter Upper Byte Latch. |
| $\overline{\mathrm{PD}}$ | Power-down bit. |
| PRODH | Product of Multiply High Byte. |
| PRODL | Product of Multiply Low Byte. |
| S | Fast Call/Return mode select bit <br> $\mathrm{s}=0$ : do not update into/from shadow registers <br> $\mathrm{s}=1$ : certain registers loaded into/from shadow registers (Fast mode) |
| TBLPTR | 21-bit Table Pointer (points to a Program Memory location). |
| TABLAT | 8-bit Table Latch. |
| TO | Time-out bit. |
| TOS | Top-of-Stack. |
| u | Unused or unchanged. |
| WDT | Watchdog Timer. |
| WREG | Working register (accumulator). |
| x | Don't care (' 0 ' or ' 1 '). The assembler will generate code with $x=0$. It is the recommended form of use for compatibility with all Microchip software tools. |
| $\mathrm{z}_{\text {S }}$ | 7-bit offset value for indirect addressing of register files (source). |
| $z_{\text {d }}$ | 7-bit offset value for indirect addressing of register files (destination). |
| \{ \} | Optional argument. |
| [text] | Indicates an indexed address. |
| (text) | The contents of text. |
| [expr]<n> | Specifies bit n of the register indicated by the pointer expr. |
| $\rightarrow$ | Assigned to. |
| < > | Register bit field. |
| $\epsilon$ | In the set of. |
| italics | User defined term (font is Courier). |

FIGURE 21-1: GENERAL FORMAT FOR INSTRUCTIONS
Byte-oriented file register operations

d = 0 for result destination to be WREG register
d = 1 for result destination to be file register (f)
$\mathrm{a}=0$ to force Access Bank
$\mathrm{a}=1$ for BSR to select bank
$\mathrm{f}=8$-bit file register address
Byte to Byte move operations (2-word)

| $15 \quad 1211$ |  |
| :---: | :---: |
| OPCODE | f (Source FILE \#) |
| 1512 |  |
| 1111 | f (Destination FILE \#) |

$f=12$-bit file register address

## Bit-oriented file register operations

| 15 | 1211 | 98 | 0 |
| :--- | :--- | :--- | :--- | :--- |
| OPCODE | b (BIT \#) | a | f (FILE \#) |

$b=3$-bit position of bit in file register (f)
$\mathrm{a}=0$ to force Access Bank
a = 1 for BSR to select bank
$\mathrm{f}=8$-bit file register address
Literal operations

| 15 | 8 |
| :--- | :--- |
| OPCODE | k (literal) |

$\mathrm{k}=8$-bit immediate value

Control operations
CALL, GOTO and Branch operations

| 15 | 87 |  |  | 0 |  |  |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
|  | OPCODE |  |  |  | $\mathrm{n}<7: 0>$ (literal) |  |
| 15 | 12 |  |  |  | 11 | 0 |
| 1111 |  | $\mathrm{n}<19: 8>$ (literal) |  |  |  |  |



BRA MYFUNC

BC MYFUNC

TABLE 21-2: PIC18FXXXX INSTRUCTION SET

| Mnemonic, Operands |  | Description | Cycles | 16-Bit Instruction Word |  |  |  | Status <br> Affected | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MSb |  |  | LSb |  |  |
| BYTE-ORIENTED OPERATIONS |  |  |  |  |  |  |  |  |  |
| ADDWF | f, d, a |  | Add WREG and f | 1 | 0010 | 01da | ffff | ffff | C, DC, Z, OV, N | 1, 2 |
| ADDWFC | f, d, a | Add WREG and CARRY bit to f | 1 | 0010 | 00da | ffff | ffff | C, DC, Z, OV, N | 1, 2 |
| ANDWF | f, d, a | AND WREG with f | 1 | 0001 | 01da | ffff | ffff | Z, N | 1,2 |
| CLRF | f, a | Clear f | 1 | 0110 | 101a | ffff | ffff | Z | 2 |
| COMF | f, d, a | Complement f | 1 | 0001 | 11da | ffff | ffff | Z, N | 1, 2 |
| CPFSEQ | f, a | Compare f with WREG, skip = | 1 (2 or 3) | 0110 | 001 a | ffff | ffff | None | 4 |
| CPFSGT | f, a | Compare f with WREG, skip > | 1 (2 or 3) | 0110 | 010a | ffff | ffff | None | 4 |
| CPFSLT | f, a | Compare f with WREG, skip < | 1 (2 or 3) | 0110 | 000a | ffff | ffff | None | 1, 2 |
| DECF | f, d, a | Decrement f | 1 | 0000 | 01da | ffff | ffff | C, DC, Z, OV, N | 1, 2, 3, 4 |
| DECFSZ | f, d, a | Decrement f , Skip if 0 | 1 (2 or 3) | 0010 | 11da | ffff | ffff | None | 1, 2, 3, 4 |
| DCFSNZ | f, d, a | Decrement f, Skip if Not 0 | 1 (2 or 3) | 0100 | 11da | ffff | ffff | None | 1, 2 |
| INCF | f, d, a | Increment f | 1 | 0010 | 10da | ffff | ffff | C, DC, Z, OV, N | 1, 2, 3, 4 |
| INCFSZ | f, d, a | Increment f, Skip if 0 | 1 (2 or 3) | 0011 | 11da | ffff | ffff | None | 4 |
| INFSNZ | f, d, a | Increment f, Skip if Not 0 | 1 (2 or 3) | 0100 | 10da | ffff | ffff | None | 1, 2 |
| IORWF | f, d, a | Inclusive OR WREG with f | 1 | 0001 | 00da | ffff | ffff | Z, N | 1, 2 |
| MOVF | f, d, a | Move f | 1 | 0101 | 00da | ffff | ffff | Z, N | 1 |
| MOVFF | $f_{s}, f_{d}$ | Move $f_{s}$ (source) to 1 st word $\mathrm{f}_{\mathrm{d}}$ (destination) 2nd word | 2 | 1100 1111 | ffff | ffff | ffff | None |  |
| MOVWF | f, a | Move WREG to $f$ | 1 | 0110 | 111a | ffff | ffff | None |  |
| MULWF | f, a | Multiply WREG with f | 1 | 0000 | 001 a | ffff | ffff | None | 1, 2 |
| NEGF | f, a | Negate f | 1 | 0110 | 110a | ffff | ffff | C, DC, Z, OV, N |  |
| RLCF | f, d, a | Rotate Left f through Carry | 1 | 0011 | 01da | ffff | ffff | C, Z, N | 1, 2 |
| RLNCF | f, d, a | Rotate Left f (No Carry) | 1 | 0100 | 01da | ffff | ffff | Z, N |  |
| RRCF | f, d, a | Rotate Right f through Carry | 1 | 0011 | 00da | ffff | ffff | C, Z, N |  |
| RRNCF | f, d, a | Rotate Right f (No Carry) | 1 | 0100 | 00da | ffff | ffff | Z, N |  |
| SETF | f, a | Set f | 1 | 0110 | 100a | ffff | ffff | None | 1, 2 |
| SUBFWB | f, d, a | Subtract f from WREG with borrow | 1 | 0101 | 01da | ffff | ffff | C, DC, Z, OV, N |  |
| SUBWF | f, d, a | Subtract WREG from f | 1 | 0101 | 11da | ffff | ffff | C, DC, Z, OV, N | 1, 2 |
| SUBWFB | f, d, a | Subtract WREG from $f$ with borrow | 1 | 0101 | 10da | ffff | ffff | C, DC, Z, OV, N |  |
| SWAPF | f, d, a | Swap nibbles in f | 1 | 0011 | 10da | ffff | ffff | None |  |
| TSTFSZ | f, a | Test f, skip if 0 | 1 (2 or 3) | 0110 | 011a | ffff | ffff | None | 1, 2 |
| XORWF | f, d, a | Exclusive OR WREG with f | 1 | 0001 | 10da | ffff | ffff | Z, N |  |

Note 1: When a Port register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is ' 1 ' for a pin configured as input and is driven low by an external device, the data will be written back with a ' 0 '.
2: If this instruction is executed on the TMRO register (and where applicable, ' $d$ ' $=1$ ), the prescaler will be cleared if assigned.
3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.
4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

TABLE 21-2: PIC18FXXXX INSTRUCTION SET (CONTINUED)


Note 1: When a Port register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is ' 1 ' for a pin configured as input and is driven low by an external device, the data will be written back with a ' 0 '.
2: If this instruction is executed on the TMRO register (and where applicable, ' $d$ ' $=1$ ), the prescaler will be cleared if assigned.
3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.
4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

TABLE 21-2: PIC18FXXXX INSTRUCTION SET (CONTINUED)

| Mnemonic, Operands |  | Description | Cycles | 16-Bit Instruction Word |  |  |  | Status <br> Affected | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MSb |  |  |  | LSb |  |  |
| LITERAL OPERATIONS |  |  |  |  |  |  |  |  |  |
| ADDLW | k |  | Add literal and WREG | 1 | 0000 | 1111 | kkkk | kkkk | C, DC, Z, OV, N |  |
| ANDLW | k | AND literal with WREG | 1 | 0000 | 1011 | kkkk | kkkk | Z, N |  |
| IORLW | k | Inclusive OR literal with WREG | 1 | 0000 | 1001 | kkkk | kkkk | Z, N |  |
| LFSR | f, k | Move literal (12-bit) 2nd word | 2 | 1110 | 1110 | $00 f f$ | kkkk | None |  |
|  |  | to FSR(f) 1st word |  | 1111 | 0000 | kkkk | kkkk |  |  |
| MOVLB | k | Move literal to BSR<3:0> | 1 | 0000 | 0001 | 0000 | kkkk | None |  |
| MOVLW | k | Move literal to WREG | 1 | 0000 | 1110 | kkkk | kkkk | None |  |
| MULLW | k | Multiply literal with WREG | 1 | 0000 | 1101 | kkkk | kkkk | None |  |
| RETLW | k | Return with literal in WREG | 2 | 0000 | 1100 | kkkk | kkkk | None |  |
| SUBLW | k | Subtract WREG from literal | 1 | 0000 | 1000 | kkkk | kkkk | C, DC, Z, OV, N |  |
| XORLW | k | Exclusive OR literal with WREG | 1 | 0000 | 1010 | kkkk | kkkk | Z, N |  |
| DATA MEMORY $\leftrightarrow$ PROGRAM MEMORY OPERATIONS |  |  |  |  |  |  |  |  |  |
| TBLRD* <br> TBLRD* + <br> TBLRD*- <br> TBLRD+* <br> TBLWT* <br> TBLWT* + <br> TBLWT*- <br> TBLWT+ |  | Table Read | 2 | 0000 | 0000 | 0000 | 1000 | None |  |
|  |  | Table Read with post-increment |  | 0000 | 0000 | 0000 | 1001 | None |  |
|  |  | Table Read with post-decrement |  | 0000 | 0000 | 0000 | 1010 | None |  |
|  |  | Table Read with pre-increment |  | 0000 | 0000 | 0000 | 1011 | None |  |
|  |  | Table Write | 2 | 0000 | 0000 | 0000 | 1100 | None |  |
|  |  | Table Write with post-increment |  | 0000 | 0000 | 0000 | 1101 | None |  |
|  |  | Table Write with post-decrement |  | 0000 | 0000 | 0000 | 1110 | None |  |
|  |  | Table Write with pre-increment |  | 0000 | 0000 | 0000 | 1111 | None |  |

Note 1: When a Port register is modified as a function of itself (e.g., MOVF PORTB, 1, 0 ), the value used will be that value present on the pins themselves. For example, if the data latch is ' 1 ' for a pin configured as input and is driven low by an external device, the data will be written back with a ' 0 '.
2: If this instruction is executed on the TMRO register (and where applicable, ' $d$ ' $=1$ ), the prescaler will be cleared if assigned.
3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.
4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

### 21.1.1 STANDARD INSTRUCTION SET

| ADDLW | ADD Literal to W |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | ADDLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | $(\mathrm{W})+\mathrm{k} \rightarrow \mathrm{W}$ |  |  |  |
| Status Affected: | N, OV, C, DC, Z |  |  |  |
| Encoding: | 0000 | 1111 | kkkk | kkkk |
| Description: | The contents of W are added to the 8 -bit literal ' $k$ ' and the result is placed in W. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $k$ ' | Process Data |  | e to W |
| Example: | ADDLW | 15h |  |  |
| Before Instruction |  |  |  |  |
| $\mathrm{W}=$ |  |  |  |  |
| After Instruction |  |  |  |  |
| $\mathrm{W}=$ | 25h |  |  |  |

## PIC18F45J10 FAMILY

| ADDWFC | ADD W and Carry bit to f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | ADDWFC f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $(\mathrm{W})+(\mathrm{f})+(\mathrm{C}) \rightarrow$ dest |  |  |  |
| Status Affected: | N,OV, C, DC, Z |  |  |  |
| Encoding: | 0010 | 00da | ffff | ffff |
| Description: | Add W, the Carry flag and data memory location ' $f$ '. If ' $d$ ' is ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is placed in data memory location ' f '. If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ ( 5 Fh ). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' |  |  | Write to destination |
| Example: | ADDWFC | REG, 0, 1 |  |  |
| Before Instruction |  |  |  |  |
| Carry b REG | $\begin{aligned} & =1 \\ & =02 h \end{aligned}$ |  |  |  |
| W | $=4 \mathrm{Dh}$ |  |  |  |
| After Instruction Carry bit REG W | $\begin{array}{ll} = & 0 \\ = & 02 \mathrm{~h} \end{array}$ |  |  |  |


| ANDWF | AND W with f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | ANDWF f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | (W).AND. (f) $\rightarrow$ dest |  |  |  |
| Status Affected: | N, Z |  |  |  |
| Encoding: | 0001 | 01da | ffff | ffff |
| Description: | The contents of W are ANDed with register ' $f$ '. If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' $f$ ' (default). <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' f ' |  |  | Write to destination |
| Example: | ANDWF | REG, 0,0 |  |  |
| Before Instruction |  |  |  |  |
| W | $=17 \mathrm{~h}$ |  |  |  |
| REG | $=\mathrm{C} 2 \mathrm{~h}$ |  |  |  |
| After Instruction |  |  |  |  |
| W | $=02 \mathrm{~h}$ |  |  |  |
| REG | $=\mathrm{C} 2 \mathrm{~h}$ |  |  |  |


| BC | Branch if Carry |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | BC n |  |  |  |
| Operands: | $-128 \leq n \leq 127$ |  |  |  |
| Operation: | if Carry bit is ' 1 '$(P C)+2+2 n \rightarrow P C$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1110 | 0010 | nnnn | n nnnn |
| Description: | If the Carry bit is ' 1 ', then the program will branch. <br> The 2's complement number ' 2 n ' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\mathrm{PC}+2+2 \mathrm{n}$. This instruction is then a two-cycle instruction. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1(2) |  |  |  |
| Q Cycle Activity: If Jump: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $n$ ' | Process Data |  | Write to PC |
| No operation | No operation | No operation |  | No operation |
| If No Jump: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | $\begin{gathered} \text { Read literal } \\ \text { ' } n \text { ' } \\ \hline \end{gathered}$ | Process Data |  | No operation |
| Example: | HERE | BC | 5 |  |
| Before Instruction |  |  |  |  |
| PC <br> After Instruction | $=$ | address (HERE) |  |  |
| If Carry <br> If Carry PC | $\begin{array}{ll} = & 1 ; \\ = & a d \\ = & 0 ; \\ = & a d \end{array}$ | 1; address | address (HERE + 2) | + 12) <br> 2) |


| BCF | Bit Clear f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | BCF f, b $\{, \mathrm{a}\}$ |  |  |  |
| Operands: | $0 \leq f \leq 255$ $0 \leq b \leq 7$ $a \in[0,1]$ |  |  |  |
| Operation: | $0 \rightarrow \mathrm{f}<\mathrm{b}>$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1001 | bbba | ffff | ffff |
| Description: | Bit ' $b$ ' in register ' $f$ ' is cleared. If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' f ' | Process <br> Data |  | Write register 'f' |
| Example: | BCF | FLAG_REG, 7, |  | , 0 |
| Before Instruction |  |  |  |  |
| FLAG_REG = <br> After Instruction | $G=\quad C 7 h$ |  |  |  |
| FLAG_REG = |  | 47h |  |  |


| BN | Branch if Negative |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | BN n |  |  |  |
| Operands: | $-128 \leq n \leq 127$ |  |  |  |
| Operation: | if Negative bit is ' 1 '$(P C)+2+2 n \rightarrow P C$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1110 | 0110 |  | nnnn |
| Description: | If the Negative bit is ' 1 ', then the program will branch. <br> The 2's complement number ' $2 n$ ' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $P C+2+2 n$. This instruction is then a two-cycle instruction. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1(2) |  |  |  |
| Q Cycle Activity: If Jump: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' n ' | Process Data |  | e to PC |
| No operation | No operation | No operation |  | No ration |
| If No Jump: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' n ' | Process Data |  | No ration |

Example: HERE BN Jump

Before Instruction

| PC | $=$ address (HERE) |
| ---: | :--- |
| After Instruction |  |
| If Negative | $=1 ;$ |
| PC | $=$ address (Jump) |
| If Negative | $=0 ;$ |
| PC | $=$ address (HERE +2 ) |



| BNN | Branch if Not Negative |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | BNN n |  |  |  |
| Operands: | $-128 \leq n \leq 127$ |  |  |  |
| Operation: | if Negative bit is ' 0 '$(P C)+2+2 n \rightarrow P C$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1110 | 0111 | nnnn | nnnn |
| Description: | If the Negative bit is ' 0 ', then the program will branch. <br> The 2's complement number ' $2 n$ ' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\mathrm{PC}+2+2 \mathrm{n}$. This instruction is then a two-cycle instruction. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1(2) |  |  |  |
| Q Cycle Activity: If Jump: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $n$ ' | Process Data |  | Write to PC |
| No operation | No operation | No operation |  | No operation |
| If No Jump: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | $\begin{gathered} \text { Read literal } \\ \text { ' } n \text { ' } \\ \hline \end{gathered}$ | Process Data |  | No operation |
| Example: | HERE | BNN | Jump |  |
| $\begin{aligned} & \text { Before Instruction } \\ & \text { PC }\end{aligned}=$ address (HERE) |  |  |  |  |
| After Instruction |  |  |  |  |
| If Negative | $\begin{array}{ll} = & 1 ; \\ = & a d \end{array}$ | ddress (Jump) |  |  |

## PIC18F45J10 FAMILY



| Example: | HERE | BNOV Jump |
| :---: | :--- | :--- | :--- |
| Before Instruction |  |  |
| PC | address (HERE) |  |
| After Instruction |  |  |
| If Overflow | $=$ | $0 ;$ |
| PC | $=$ | address (Jump) |
| If Overflow | $=1 ;$ |  |
| PC | $=$ | address (HERE +2 ) |


| BNZ | Branch if Not Zero |
| :---: | :---: |
| Syntax: | BNZ n |
| Operands: | $-128 \leq n \leq 127$ |
| Operation: | if Zero bit is ' 0 ' $(P C)+2+2 n \rightarrow P C$ |
| Status Affected: | None |
| Encoding: | 1110 0001 nnnn nnnn |
| Description: | If the Zero bit is ' 0 ', then the program will branch. <br> The 2's complement number ' $2 n$ ' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\mathrm{PC}+2+2 \mathrm{n}$. This instruction is then a two-cycle instruction. |
| Words: | 1 |
| Cycles: | 1(2) |

Q Cycle Activity:
If Jump:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read literal <br> 'n' | Process <br> Data | Write to PC |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

If No Jump:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read literal <br> ' $n$ ' | Process <br> Data | No <br> operation |

Example: HERE BNZ Jump
Before Instruction

| PC | $=$ address (HERE) |
| ---: | :--- |
| After Instruction |  |
| If Zero | $=0 ;$ |
| PC | $=$ address (Jump) |
| If Zero | $=1 ;$ |

If Zero $=1$;
PC = address (HERE + 2)



| BTFSS | Bit Test File, Skip if Set |
| :---: | :---: |
| Syntax: | BTFSS f, b $\{, \mathrm{a}\}$ |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & 0 \leq b<7 \\ & a \in[0,1] \end{aligned}$ |
| Operation: | skip if ( $\mathrm{f}<\mathrm{b}>$ ) = 1 |
| Status Affected: | None |
| Encoding: |  |
| Description: | If bit ' $b$ ' in register ' $f$ ' is ' 1 ', then the next instruction is skipped. If bit ' $b$ ' is ' 1 ', then the next instruction fetched during the current instruction execution is discarded and a NOP is executed instead, making this a two-cycle instruction. <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ (5Fh). <br> See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |
| Words: | 1 |
| Cycles: | 1(2) |
|  | 3 cycles if skip and followed by a 2 -word instruction. |

Q Cycle Activity:


If skip:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

If skip and followed by 2-word instruction:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |



| BTG | Bit Toggle f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | BTG f, b \{,a\} |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & 0 \leq b<7 \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $(\overline{f<b>}) \rightarrow f<b>$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0111 | bbba | ffff | ffff |
| Description: | Bit ' $b$ ' in data memory location ' $f$ ' is inverted. <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ ( 5 Fh ). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' |  |  | Write register ' $f$ ' |
| Example: | BTG Por | PORTC, 4, 0 |  |  |
| Before Instruction: |  |  |  |  |
| After Instruction: |  | - 01110101 [75h] |  |  |
| PORTC | $=01100101$ [65h] |  |  |  |



## PIC18F45J10 FAMILY



CALL

| Syntax: | CALL k $\{$ s\} |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Operands: | $\begin{aligned} & 0 \leq k \leq 1048575 \\ & s \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $\begin{aligned} & (\mathrm{PC})+4 \rightarrow \text { TOS, } \\ & \mathrm{k} \rightarrow \mathrm{PC}<20: 1>, \\ & \text { if } s=1 \\ & (\mathrm{~W}) \rightarrow \mathrm{WS}, \\ & \text { (STATUS) } \rightarrow \text { STATUSS, } \\ & (\text { BSR }) \rightarrow \text { BSRS } \end{aligned}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: <br> 1st word ( $k<7: 0>$ ) <br> 2nd word(k<19:8>) | 1110 | ( $\begin{gathered}110 s \\ k_{19} \mathrm{kkk}\end{gathered}$ | $\mathrm{k}_{7} \mathrm{kkk}$ kkkk | $\mathrm{kkkk}_{0}$ <br> kkkk |
| Description: | Subroutine call of entire 2-Mbyte memory range. First, return address (PC + 4) is pushed onto the return stack. If ' $s$ ' $=1$, the W, STATUS and BSR registers are also pushed into their respective shadow registers, WS, STATUSS and BSRS. If ' $s$ ' $=0$, no update occurs (default). Then, the 20 -bit value ' k ' is loaded into $\mathrm{PC}<20: 1>$. CALL is a two-cycle instruction. |  |  |  |

Words: 2
Cycles: 2
Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read literal <br> 'k'<7:0>, | PUSH PC to <br> stack | Read literal <br> ' $k$ '<19:8>, <br> Write to PC |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

Example: HERE CALL THERE, 1

| Before Instruction |  |  |
| :---: | :---: | :---: |
| PC | address | (HERE) |
| After Instruction |  |  |
| PC | address | (THERE) |
| TOS | address | (HERE + 4) |
| WS | W |  |
| BSRS | BSR |  |
| STATUSS = | STATUS |  |



| CLRWDT | Clear Watchdog Timer |  |  |
| :---: | :---: | :---: | :---: |
| Syntax: | CLRWDT |  |  |
| Operands: | None |  |  |
| Operation: | $\begin{aligned} & 000 \mathrm{~h} \rightarrow \text { WDT, } \\ & 000 \mathrm{~h} \rightarrow \text { WDT postscaler, } \\ & 1 \rightarrow \overline{\mathrm{TO},} \\ & 1 \rightarrow \overline{\mathrm{PD}} \end{aligned}$ |  |  |
| Status Affected:Encoding: | $\overline{\mathrm{TO}}, \overline{\mathrm{PD}}$ |  |  |
|  | 0000 | 0000 0000 | 0100 |
| Description: | CLRWDT instruction resets the Watchdog Timer. It also resets the postscaler of the WDT. Status bits, $\overline{T O}$ and $\overline{\mathrm{PD}}$, are set. |  |  |
| Words: | 1 |  |  |
| Cycles: | 1 |  |  |
| Q Cycle Activity: |  |  |  |
| Q1 | Q2 | Q3 | Q4 |
| Decode | No operation | Process Data | No operation |
| Example: | CLRWDT |  |  |
| Before Instruction |  |  |  |
| WDT Co | ter | ? |  |
| After Instruction |  |  |  |
| WDT Counter |  | 00h |  |
| WDT Postscaler |  | 0 |  |
| TO |  | 1 |  |
| $\overline{\mathrm{PD}}$ |  |  |  |


| COMF | Complement f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | COMF f $\{$, \{ $\{\mathrm{a}\}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | ( $\overline{\mathrm{f}}) \rightarrow$ dest |  |  |  |
| Status Affected: | N, Z |  |  |  |
| Encoding: | 0001 | 11da | ffff | ffff |
| Description: | The contents of register ' f ' are complemented. If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' $f$ ' (default). If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ ( 5 Fh ). See Section 21.2.3"Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' |  |  | Write to destination |
| Example: | COMF | REG, 0, 0 |  |  |
| Before Instruction |  |  |  |  |
| REG | $=13 \mathrm{~h}$ |  |  |  |
| After Instruction |  |  |  |  |
| REG | $\begin{array}{ll}= & 13 \mathrm{~h} \\ = & E C h\end{array}$ |  |  |  |
| W |  |  |  |  |



| CPFSLT | Compare f with W, Skip if $\mathbf{f}$ < W |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | CPFSLT f $\{, \mathrm{a}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $\begin{aligned} & \text { (f) }-(\mathrm{W}), \\ & \text { skip if }(\mathrm{f})<\text { (W) } \\ & \text { (unsigned comparison) } \end{aligned}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0110 | 000a | ffff | ffff |

Compares the contents of data memory location ' $f$ ' to the contents of $W$ by performing an unsigned subtraction. If the contents of ' $f$ ' are less than the contents of W, then the fetched instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.
If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default).
Words: $\quad 1$
Cycles: $\quad 1(2)$
Note: 3 cycles if skip and followed by a 2 -word instruction.

Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> register ' f ' | Process <br> Data | No <br> operation |

If skip:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

If skip and followed by 2-word instruction:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |


| Example: | HERE | CPFSLT REG, 1 |
| :--- | :--- | :--- |
|  | NLESS | $:$ |
|  | LESS | $:$ |

Before Instruction

| $\mathrm{PC}$ W | = | Address ? | (HERE) |
| :---: | :---: | :---: | :---: |
| After Instruction |  |  |  |
| If REG | < | W; |  |
| PC | $=$ | Address | (LESS) |
| If REG | $\geq$ | W; |  |
| PC | $=$ | Address | (NLESS) |

## PIC18F45J10 FAMILY

| DAW | Decimal Adjust W Register |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | DAW |  |  |  |
| Operands: | None |  |  |  |
| Operation: | If $[\mathrm{W}<3: 0>$ <br> ( $\mathrm{W}<3: 0>$ ) <br> else <br> ( $\mathrm{W}<3: 0>$ ) <br> If $[W<7: 4>$ <br> ( $\mathrm{W}<7: 4>$ ) <br> else <br> ( $\mathrm{W}<7: 4>$ ) | $\begin{aligned} & \text { If }[W<3: 0 \gg 9] \text { or }[D C=1] \text { then } \\ & (W<3: 0>)+6 \rightarrow W<3: 0> \\ & \text { else } \\ & (W<3: 0>) \rightarrow W<3: 0> \end{aligned}$ |  |  |
| Status Affected:Encoding: | C |  |  |  |
|  | 0000 | 0000 | 0000 | 0111 |
| Description: | DAW adjusts the eight-bit value in W , resulting from the earlier addition of two variables (each in packed BCD format) and produces a correct packed BCD result. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register W |  |  | Write W |
| Example 1: |  |  |  |  |
|  | DAW |  |  |  |
| Before Instruction |  |  |  |  |
| W | $=A 5 h$ |  |  |  |
| C | $=0$ |  |  |  |
| DC | $=0$ |  |  |  |
| After Instruction |  |  |  |  |
| W | $=05 \mathrm{~h}$ |  |  |  |
| C | $=1$ |  |  |  |
| Example 2: |  |  |  |  |
| Before Instruction |  |  |  |  |
| W | - CEh |  |  |  |
| C | 0 |  |  |  |
| DC | $=0$ |  |  |  |
| After Instruction |  |  |  |  |
| W | $=34 \mathrm{~h}$ |  |  |  |
|  | $\begin{aligned} & =1 \\ & = \\ & = \end{aligned}$ |  |  |  |


| DECF | Decrement f |
| :---: | :---: |
| Syntax: | DECF f $\{$, \{ $\{$ a \} $\}$ |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |
| Operation: | (f) $-1 \rightarrow$ dest |
| Status Affected: | C, DC, N, OV, Z |
| Encoding: |  |
| Description: | Decrement register ' $f$ '. If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' f ' (default). <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |
| Words: | 1 |
| Cycles: | 1 |

Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> register ' f ' | Process <br> Data | Write to <br> destination |

Example: $\quad$ DECF CNT, 1, 0
Before Instruction

| CNT | $=01 \mathrm{~h}$ |
| :--- | :--- |
| Z | $=0$ |

After Instruction

$$
\begin{aligned}
& \text { CNT }=00 h ~
\end{aligned}
$$



| DCFSNZ | Decrement f, Skip if not 0 |
| :---: | :---: |
| Syntax: | DCFSNZ f $\left\{, \mathrm{d}\left\{\right.\right.$ a $\left.{ }^{\text {¢ }}\right\}$ |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |
| Operation: | $\begin{aligned} & \text { (f) }-1 \rightarrow \text { dest, } \\ & \text { skip if result } \neq 0 \end{aligned}$ |
| Status Affected: | None |
| Encoding: | 0100 11daffff ffff |
| Description: | The contents of register ' $f$ ' are decremented. If ' $d$ ' is ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is placed back in register ' $f$ ' (default). If the result is not ' 0 ', the next instruction, which is already fetched, is discarded and a NOP is executed instead, making it a two-cycle instruction. <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |
| Words: | 1 |
| Cycles: | 1(2) |
|  | 3 cycles if skip and followed by a 2 -word instruction. |

Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> register ' $f$ ' | Process <br> Data | Write to <br> destination |

If skip:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

If skip and followed by 2-word instruction:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |


| Example: | HERE | DCFSNZ TEMP, 1, 0 |  |  |
| :--- | :--- | :--- | :--- | :--- |
|  | ZERO | $:$ |  |  |
|  | NZERO | $:$ |  |  |


| Before Instruction | $=?$ |
| ---: | :--- |
| TEMP | $=?$ |
| After Instruction | $=$ TEMP - 1, |
| TEMP | $=0 ;$ |
| If TEMP | $=$ Address (ZERO) |
| PC | $\neq 0 ;$ |
| If TEMP | $=$ Address (NZERO) |
| PC |  |

## PIC18F45J10 FAMILY




INFSNZ Increment f, Skip if not 0
Syntax:
Operands:

Operation:

Status Affected:
Encoding:
Description:

Words:
Cycles:
1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:


If skip:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

If skip and followed by 2-word instruction:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |
| No <br> operation | No <br> operation | No <br> operation | No <br> operation |

Example: |  | HERE |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
|  | ZERO |  |  |  |
|  | NZERO |  |  |  |
|  |  |  |  |  |

Before Instruction

$$
\text { PC } \quad=\text { Address (HERE) }
$$

After Instruction

| REG | $=$ | REG +1 |
| :--- | :--- | :--- |
| If REG | $\neq$ | $0 ;$ |
| PC | $=$ | Address (NZERO) |
| If REG | $=0 ;$ |  |
| PC | $=$ Address (ZERO) |  |

## PIC18F45J10 FAMILY

| IORLW | Inclusive OR Literal with W |  |  |  | IORWF | Inclusive OR W with f |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Syntax: | IORLW k |  |  |  | Syntax: | IORWF f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  | Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | (W) .OR. $\mathrm{k} \rightarrow \mathrm{W}$ |  |  |  |  |  |  |  |  |
| Status Affected: | N, Z |  |  |  | Operation: <br> Status Affected: <br> Encoding: <br> Description: |  |  |  |  |
| Encoding: | 0000 | 1001 | kkkk | kkkk |  |  |  |  |  |
| Description: | The contents of W are ORed with the eight-bit literal ' $k$ '. The result is placed in W. |  |  |  |  | 0001 | 0da | ffff | ffff |
|  |  |  |  |  | Inclusive | W with | gister | If ' d ' is |
| Words: | 1 |  |  |  |  |  | ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is placed back in register ' $f$ ' (default). |  |  |  |
| Cycles: | 1 |  |  |  |  |  |  |  |  |  |  |  |  |
| Q Cycle Activity: |  |  |  |  |  | If ' $a$ ' is ' 0 ', the Access Bank is selected. <br> If ' $a$ ' is ' 1 ', the BSR is used to select the |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |  | GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates |  |  |  |
| Decode | Read literal ' $k$ ' | $\begin{aligned} & \text { Proo } \\ & \mathrm{Da} \\ & \hline \end{aligned}$ |  | rite to W |  |  |  |  |  |  |  |  |  |
| Example: | IORLW | 35 h |  |  |  | mode whenever f $\leq 95$ (5Fh). See |  |  |  |
| Before Instruction |  |  |  |  |  | Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| W | 9Ah |  |  |  |  |  |  |  |  |  |  |  |  |
| After Instruction |  |  |  |  | Words: | 1 |  |  |  |
| W | = BFh |  |  |  | Cycles: | 1 |  |  |  |
|  |  |  |  |  | Q Cycle Activity: |  |  |  |  |
|  |  |  |  |  | Q1 | Q2 | Q |  | Q4 |
|  |  |  |  |  | Decode | $\begin{gathered} \text { Read } \\ \text { register 'f' } \end{gathered}$ | Proc |  | ite to ination |
|  |  |  |  |  | Example: | IORWF | ULT, |  |  |
|  |  |  |  |  | Before Instru |  |  |  |  |
|  |  |  |  |  | RESUL | $=13 \mathrm{~h}$ |  |  |  |
|  |  |  |  |  | W | 91h |  |  |  |
|  |  |  |  |  | After Instruc |  |  |  |  |
|  |  |  |  |  | RESUL | $=13 \mathrm{~h}$ |  |  |  |
|  |  |  |  |  | W | $=93 \mathrm{~h}$ |  |  |  |


| LFSR | Load FSR |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | LFSR f, k |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 2 \\ & 0 \leq k \leq 4095 \end{aligned}$ |  |  |  |
| Operation: | $\mathrm{k} \rightarrow$ FSRf |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1110 1111 | $\begin{aligned} & 1110 \\ & 0000 \end{aligned}$ | $\begin{gathered} 00 f f \\ \mathrm{k}_{7} \mathrm{kkk} \end{gathered}$ | $\begin{array}{c\|c} \mathrm{f} & \mathrm{k}_{11} \mathrm{kkk} \\ \mathrm{kk} & \mathrm{kkkk} \\ \hline \end{array}$ |
| Description: | The 12-bit literal ' $k$ ' is loaded into the File Select Register pointed to by ' $f$ '. |  |  |  |
| Words: | 2 |  |  |  |
| Cycles: | 2 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal 'k' MSB | $\begin{aligned} & \text { Proce } \\ & \text { Dat } \end{aligned}$ |  | Write <br> literal ' $k$ ' <br> MSB to <br> FSRfH |
| Decode | Read literal ' $k$ ' LSB | $\begin{array}{r} \hline \text { Proc } \\ \text { Dat } \end{array}$ |  | Write literal ' $k$ ' to FSRfL |

## Example: <br> LFSR 2, 3ABh

| After Instruction |  |  |
| :---: | :--- | :--- |
| FSR2H | $=03 \mathrm{~h}$ |  |
| FSR2L | $=$ ABh |  |


| MOVF | Move f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MOVF f $\{$, \{,a\}\} |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $\mathrm{f} \rightarrow$ dest |  |  |  |
| Status Affected: | N, Z |  |  |  |
| Encoding: | 0101 | 00da | ffff | ffff |
| Description: | The contents of register ' $f$ ' are moved to a destination dependent upon the status of ' $d$ '. If ' $d$ ' is ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is placed back in register ' $f$ ' (default). Location ' $f$ ' can be anywhere in the 256-byte bank. <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ ( 5 Fh ). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' | Process Data |  | Write W |
| Example: | MOVF REG, 0, 0 |  |  |  |
| Before Instruction |  |  |  |  |
| REG | $=22 \mathrm{~h}$ |  |  |  |
| W | $=\mathrm{FFh}$ |  |  |  |
| After Instruction |  |  |  |  |
| REG | $=22 \mathrm{~h}$ |  |  |  |
| W | $=22 \mathrm{~h}$ |  |  |  |

## PIC18F45J10 FAMILY

| MOVFF | Move f to f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MOVFF $\mathrm{f}_{\mathrm{s}}, \mathrm{f}_{\mathrm{d}}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f_{s} \leq 4095 \\ & 0 \leq f_{d} \leq 4095 \end{aligned}$ |  |  |  |
| Operation: | $\left(\mathrm{f}_{\mathrm{s}}\right) \rightarrow \mathrm{f}_{\mathrm{d}}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: 1st word (source) 2nd word (destin.) | 1100 | ffff ffff | ffff | ffff ffff d |
| Description: | The contents of source register ' 's ' are moved to destination register ' $f_{d}$ '. <br> Location of source ' $f$ ' can be anywhere in the 4096-byte data space ( 000 h to FFFh) and location of destination 'f ${ }_{d}$ ' can also be anywhere from 000h to FFFh. <br> Either source or destination can be W (a useful special situation). <br> MOVFF is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port). <br> The movff instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register. |  |  |  |
| Words: | 2 |  |  |  |
| Cycles: | 2 (3) |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' (src) | Process Data |  | No operation |
| Decode | No operation No dummy read | No operation |  | Write register ' f ' (dest) |
| Example: | MOVFF | REG1, REG2 |  |  |
| Before Instruction |  |  |  |  |
| REG1 REG2 | $\begin{aligned} & =33 \mathrm{~h} \\ & =\quad 11 \mathrm{~h} \end{aligned}$ |  |  |  |
| After Instruction |  |  |  |  |
| REG1 | $\begin{aligned} & =33 \mathrm{~h} \\ & =\quad 33 \mathrm{~h} \end{aligned}$ |  |  |  |


| MOVLB | Move Literal to Low Nibble in BSR |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MOVLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | $\mathrm{k} \rightarrow$ BSR |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 0001 | kkkk | kkkk |
| Description: | The eight-bit literal ' $k$ ' is loaded into the Bank Select Register (BSR). The value of $B S R<7: 4>$ always remains ' 0 ', regardless of the value of $k_{7}: k_{4}$. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $k$ ' | Process Data |  | Write literal ' $k$ ' to BSR |
| Example: | MOVLB |  | 5 |  |
| Before Instruction BSR Register $=02 \mathrm{~h}$ |  |  |  |  |


| MOVLW | Move Literal to W |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MOVLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | $\mathrm{k} \rightarrow \mathrm{W}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 1110 | kkkk | kkkk |
| Description: | The eight-bit literal ' k ' is loaded into W. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' k ' | Process Data |  | Write to W |
| Example: | MOVLW | 5Ah |  |  |
| After Instruction |  |  |  |  |
| W | $=5 \mathrm{Ah}$ |  |  |  |


| MOVWF | Move W to f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MOVWF f $\{, \mathrm{a}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | (W) $\rightarrow \mathrm{f}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0110 | 111a | ffff | ffff |
| Description: | Move data from W to register ' f '. <br> Location ' $f$ ' can be anywhere in the 256-byte bank. <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' f ' |  |  | Write ister ' f ' |

Example: MOVWF REG, 0
Before Instruction

| W | $=$ | 4 Fh |
| :---: | :---: | :---: |
| REG | $=$ | FFh |
| After Instruction |  |  |
| W | $=$ | 4 Fh |
| REG | $=$ | 4 Fh |

## PIC18F45J10 FAMILY

| MULLW | Multiply Literal with W |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MULLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | (W) $\mathrm{x} k \rightarrow$ PRODH:PRODL |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 1101 | kkkk | kkkk |
| Description: | An unsigned multiplication is carried out between the contents of W and the 8 -bit literal ' $k$ '. The 16 -bit result is placed in the PRODH:PRODL register pair. PRODH contains the high byte. $W$ is unchanged. None of the Status flags are affected. Note that neither overflow nor carry is possible in this operation. A zero result is possible but not detected. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $k$ ' | Process Data |  | Write registers PRODH: PRODL |
| Example: | MULLW OC4h |  |  |  |
| Before Instruction |  |  |  |  |
| $\begin{aligned} & \text { W } \\ & \text { PRODH } \\ & \text { PRODL } \end{aligned}$ | $\begin{aligned} & = \\ & = \\ & = \end{aligned}$ | $\begin{aligned} & \text { E2h } \\ & ? \\ & ? \end{aligned}$ |  |  |
| After Instruction |  |  |  |  |
| W PRODH PRODL | $\begin{aligned} & =\quad \text { E2h } \\ & =\quad A D h \\ & =08 h \end{aligned}$ |  |  |  |

MULWF Multiply W with f

| Syntax: | MULWF f $\{, \mathrm{a}\}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | (W) x (f) $\rightarrow$ PRODH:PRODL |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 001 a | ffff | ffff |

Description: An unsigned multiplication is carried out between the contents of W and the register file location ' $f$ '. The 16-bit result is stored in the PRODH:PRODL register pair. PRODH contains the high byte. Both W and ' f ' are unchanged.
None of the Status flags are affected. Note that neither overflow nor carry is possible in this operation. A zero result is possible but not detected. If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default).
If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever
$\mathrm{f} \leq 95$ (5Fh). See Section 21.2.3
"Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.
Words: 1
Cycles: 1
Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> register 'f' | Process <br> Data | Write <br> registers <br> PRODH: |

Example: MULWF REG, 1

Before Instruction

| $W$ | $=$ | $C 4 h$ |
| :--- | :--- | :--- |
| REG | $=$ | $B 5 h$ |
| PRODH | $=?$ |  |
| PRODL | $=$ | $?$ |
| After Instruction |  |  |
| $W$ | $=$ | $C 4 h$ |
| REG | $=$ | $B 5 h$ |
| PRODH | $=8 A h$ |  |
| PRODL | $=94 h$ |  |


| NEGF | Negate f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | NEGF f $\{, \mathrm{a}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $(\bar{f})+1 \rightarrow f$ |  |  |  |
| Status Affected: | N, OV, C, DC, Z |  |  |  |
| Encoding: | 0110 | 110a | ffff | ffff |
| Description: | Location ' f ' is negated using two's complement. The result is placed in the data memory location ' $f$ '. <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' f ' |  |  | Write register ' $f$ ' |
| Example: | NEGF | REG, 1 |  |  |
| Before Instruction |  |  |  |  |
| REG | 00111010 [3Ah] |  |  |  |
| After Instruction |  |  |  |
| REG |  |  |  |  | $=1100$ | 0110 |  |  |


| NOP | No Operation |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | NOP |  |  |  |
| Operands: | None |  |  |  |
| Operation: | No operation |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | $\begin{aligned} & 0000 \\ & 1111 \end{aligned}$ | 0000 0 <br> xxxx x | $\begin{aligned} & 0000 \\ & \mathrm{xxxx} \end{aligned}$ | $\begin{aligned} & 0000 \\ & x x x x \end{aligned}$ |
| Description: | No operation. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | $\begin{gathered} \text { No } \\ \text { operation } \end{gathered}$ | $\begin{gathered} \text { No } \\ \text { operation } \end{gathered}$ |  | No ration |

## Example:

## None.

## PIC18F45J10 FAMILY

| POP | Pop Top of Return Stack |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | POP |  |  |  |
| Operands: | None |  |  |  |
| Operation: | (TOS) $\rightarrow$ bit bucket |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 0000 0 | 0000 | 0110 |
| Description: | The TOS value is pulled off the return stack and is discarded. The TOS value then becomes the previous value that was pushed onto the return stack. This instruction is provided to enable the user to properly manage the return stack to incorporate a software stack. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | No operation | $\begin{aligned} & \text { POP TOS } \\ & \text { value } \\ & \hline \end{aligned}$ |  | No operation |
| Example: | POP |  |  |  |
|  | Gото | NEW |  |  |
| Before Instruction |  |  |  |  |
| TOS |  | 0031A2h0014332 h |  |  |
| Stack (1 level down) |  |  |  |  |
| After Instruction |  |  |  |  |
| TOS |  | $=014332 \mathrm{~h}$$=0 \mathrm{NEW}$ |  |  |



| RCALL | Relative Call |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | RCALL n |  |  |  |
| Operands: | $-1024 \leq n \leq 1023$ |  |  |  |
| Operation: | $\begin{aligned} & (\mathrm{PC})+2 \rightarrow \mathrm{TOS}, \\ & (\mathrm{PC})+2+2 \mathrm{n} \rightarrow \mathrm{PC} \end{aligned}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1101 | 1 nnn | nnnn | n nnnn |
| Description: | Subroutine call with a jump up to 1 K from the current location. First, return address $(P C+2)$ is pushed onto the stack. Then, add the 2's complement number ' $2 n$ ' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $\mathrm{PC}+2+2 \mathrm{n}$. This instruction is a two-cycle instruction. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 2 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $n$ ' PUSH PC to stack | Process Data |  | Write to PC |
| No operation | No operation | No operation |  | No operation |
| Example: | HERE | RCALL Jump |  |  |
| Before Instruction |  |  |  |  |
| $\mathrm{PC}=$ | Address (HERE) |  |  |  |
| After Instruction |  |  |  |  |
| $\mathrm{PC}=$ | Address (Jump) <br> Address (HERE + 2) |  |  |  |


| RESET | Reset |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | RESET |  |  |  |
| Operands: | None |  |  |  |
| Operation: | Reset all registers and flags that are affected by a MCLR Reset. |  |  |  |
| Status Affected: | All |  |  |  |
| Encoding: | 0000 | 0000 | 1111 | 1111 |
| Description: | This instruction provides a way to execute a $\overline{M C L R}$ Reset in software. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 |  | Q3 | Q4 |
| Decode | Start <br> Reset |  |  | No operation |
| Example: | RESET |  |  |  |
| After Instruction |  |  |  |  |
| Registers $=$ Flags* $=$ | Reset Value Reset Value |  |  |  |

## PIC18F45J10 FAMILY

| RETFIE | Return from Interrupt |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Syntax: | RETFIE \{s\} |  |  |  |  |
| Operands: | $s \in[0,1]$ |  |  |  |  |
| Operation: | $\begin{aligned} & (\mathrm{TOS}) \rightarrow \mathrm{F} \\ & 1 \rightarrow \mathrm{GIE} / \mathrm{G} \\ & \text { if } s=1 \\ & \text { (WS) } \rightarrow \mathrm{W} \\ & \text { (STATUSS } \\ & \text { (BSRS) } \rightarrow \\ & \text { PCLATU, } \end{aligned}$ | Hor $\rightarrow \text { STA }$ SR; LATH | US; | IEL, <br> chan |  |
| Status Affected: | GIE/GIEH, PEIE/GIEL. |  |  |  |  |
|  | 0000 | 0000 |  |  | 000s |
| Description: | Return from interrupt. Stack is popped and Top-of-Stack (TOS) is loaded into the PC. Interrupts are enabled by setting either the high or low priority global interrupt enable bit. If ' $s$ ' $=1$, the contents of the shadow registers, WS, STATUSS and BSRS, are loaded into their corresponding registers, W , STATUS and BSR. If ' $s$ ' $=0$, no update of these registers occurs (default). |  |  |  |  |
| Words: | 1 |  |  |  |  |
| Cycles: | 2 |  |  |  |  |
| Q Cycle Activity: |  |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |  |
| Decode | No operation | No operation |  |  | P PC <br> stack <br> GIEH or <br> GIEL |
| No operation | No operation |  |  |  | No ration |

## Example: RETFIE 1

| After Interrupt |  |
| :--- | :--- |
| PC | $=$ TOS |
| W | $=$ |
| BSR | $=$ |
| STATUS | $=$ |
| GIE/GIEH, PEIE/GIEL | $=1$ |


| RETLW | Return Literal to W |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | RETLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | $\begin{aligned} & \mathrm{k} \rightarrow \mathrm{~W}, \\ & \text { (TOS) } \rightarrow \mathrm{PC} \text {, } \\ & \text { PCLATU, PCLATH are unchanged } \end{aligned}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 1100 | kkkk | k kkkk |
| Description: | W is loaded with the eight-bit literal ' $k$ '. The program counter is loaded from the top of the stack (the return address). The high address latch (PCLATH) remains unchanged. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 2 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 |  |  | Q4 |
| Decode | Read literal ' $k$ ' |  |  | POP PC <br> from stack, Write to W |
| No operation | No operation |  |  | No operation |

## Example:

```
CALL TABLE ; W contains table
                                    ; offset value
                                    ; W now has
                                    ; table value
```

:
TABLE
ADDWF PCL ; W = offset
RETLW k0 ; Begin table
RETLW k1 ;
:
RETLW kn ; End of table
Before Instruction
$\mathrm{W}=07 \mathrm{~h}$
After Instruction
$\mathrm{W} \quad=\quad$ value of kn

| RETURN | Return from Subroutine |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Syntax: | RETURN \{s\} |  |  |  |  |
| Operands: | $s \in[0,1]$ |  |  |  |  |
| Operation: | $\begin{aligned} & (\text { TOS }) \rightarrow F \\ & \text { if } s=1 \\ & (W S) \rightarrow W \\ & \text { (STATUSS } \\ & \text { (BSRS) } \rightarrow \\ & \text { PCLATU, } \end{aligned}$ | $\rightarrow \text { STA }$ <br> BSR; | US <br> e | ha |  |
| Status Affected: | None |  |  |  |  |
| Encoding: | 0000 | 0000 |  |  | 001s |
| Description: | Return from subroutine. The stack is popped and the top of the stack (TOS) is loaded into the program counter. If ' $s$ ' $=1$, the contents of the shadow registers, WS, STATUSS and BSRS, are loaded into their corresponding registers, W, STATUS and BSR. If ' $s$ ' $=0$, no update of these registers occurs (default). |  |  |  |  |
| Words: | 1 |  |  |  |  |
| Cycles: | 2 |  |  |  |  |
| Q Cycle Activity: |  |  |  |  |  |
| Q1 | Q2 |  |  |  | Q4 |
| Decode | No operation |  |  |  | $\begin{aligned} & \text { OP PC } \\ & \mathrm{n} \text { stack } \end{aligned}$ |
| No operation | No operation |  |  |  | No eration |

## Example: <br> RETURN

After Instruction:
$\mathrm{PC}=\mathrm{TOS}$

| RLCF | Rotate Left f through Carry |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | RLCF f \{, d \{, a\}\} |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $\begin{aligned} & (\mathrm{f}<\mathrm{n}>) \rightarrow \text { dest }<\mathrm{n}+1>, \\ & (\mathrm{f}<7>) \rightarrow \mathrm{C}, \\ & (\mathrm{C}) \rightarrow \text { dest }<0> \end{aligned}$ |  |  |  |
| Status Affected: | C, N, Z |  |  |  |
| Encoding: | 0011 | 01da | ffff | ffff |

The contents of register ' $f$ ' are rotated one bit to the left through the Carry flag. If ' $d$ ' is ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' f ' (default).
If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever
$\mathrm{f} \leq 95$ (5Fh). See Section 21.2.3
"Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.


Words: $\quad 1$
Cycles: $\quad 1$
Q Cycle Activity:

| Q1 | Q2 |  | Q3 |
| :---: | :---: | :---: | :---: |
| Qecode | Read <br> register ' $f$ ' | Process <br> Data | Write to <br> destination |

Example: $\quad$ RLCF $\quad$ REG, 0, 0

Before Instruction
REG $=11100110$

| C | $=1$ |
| :--- | :--- |
| C | $=0$ |

After Instruction

| REG | $=11100110$ |
| :--- | :--- | :--- |
| W | $=11001100$ |
| C | $=1$ |


| RLNCF | Rotate Left f (No Carry) |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | RLNCF f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $\begin{aligned} & (f<n>) \rightarrow \text { dest }<n+1>, \\ & (f<7>) \rightarrow \text { dest }<0> \end{aligned}$ |  |  |  |
| Status Affected: | N, Z |  |  |  |
| Encoding: | 0100 | 01da | ffff | f ffff |
| Description: | The contents of register ' 'f' are rotated one bit to the left. If ' $d$ ' is ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' $f$ ' (default). If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
|  | 4 register f |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' f ' | Process Data |  | Write to destination |
| Example: | RLNCF | REG, 1, 0 |  |  |
| Before Instruction |  |  |  |  |
| REG | 10101011 |  |  |  |
| After Instruction |  |  |  |  |
| REG | - 01010111 |  |  |  |


| RRCF | Rotate Right f through Carry |
| :---: | :---: |
| Syntax: | RRCF f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |
| Operation: | $\begin{aligned} & (f<n>) \rightarrow \text { dest }<n-1>, \\ & (f<0>) \rightarrow C, \end{aligned}$ $\text { (C) } \rightarrow \text { dest }<7>$ |
| Status Affected: | C, N, Z |
|  |  |
| Description: | The contents of register ' $f$ ' are rotated one bit to the right through the Carry flag. If ' $d$ ' is ' 0 ', the result is placed in W. If ' $d$ ' is ' 1 ', the result is placed back in register ' $f$ ' (default). <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |



Words: $\quad 1$
Cycles: $\quad 1$
Q Cycle Activity:

| Q1 | Q2 |  | Q3 |  | Q4 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Decode | Read <br> register ' 'f' | Process <br> Data | Write to <br> destination |  |  |

Example: $\quad$ RRCF $\quad$ REG, 0, 0

Before Instruction

| REG | $=11100110$ |
| :--- | :--- | :--- |
| C | $=0$ |

After Instruction

| REG | $=$ | 1110 | 0110 |
| :--- | :--- | :--- | :--- |
| W | $=$ | 0111 | 0011 |
| C | $=$ | 0 |  |



Example 1: RRNCF REG, 1, 0
Before Instruction
REG $=11010111$
After Instruction
REG $=11101011$

Example 2: RRNCF REG, 0, 0
Before Instruction

$$
\mathrm{W} \quad=\quad ?
$$

$$
\begin{array}{lll}
\mathrm{W} & = & \\
\mathrm{REG} & = & 1101
\end{array} 0111
$$

After Instruction
$\begin{array}{ll}\mathrm{W} & =11101011 \\ \mathrm{REG} & =11010111\end{array}$

| SETF | Set f |
| :--- | :--- |
| Syntax: | SETF $\mathrm{f}\{, \mathrm{a}\}$ |
| Operands: | $0 \leq \mathrm{f} \leq 255$ |
|  | $\mathrm{a} \in[0,1]$ |
| Operation: | FFh $\rightarrow \mathrm{f}$ |
| Status Affected: | None |
| Encoding: | 0110 |
|  | $100 a$ |
|  | ffff |
|  |  |

Description: The contents of the specified register are set to FFh.
If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default).
If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ (5Fh). See
Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.
Words: $\quad 1$
Cycles: 1
Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> register ' $f$ ' | Process <br> Data | Write <br> register ' $f$ ' |

```
\[
\text { Example: } \quad \text { SETF } \quad \text { REG, } 1
\]
```

| Before Instruction <br> REG | $=5 \mathrm{Ah}$ |
| :--- | :--- |
| After Instruction |  |
| REG | $=\mathrm{FFh}$ |



| SUBFWB | Subtract f from W with Borrow |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | SUBFWB f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | (W) - (f) - ( $\overline{\mathrm{C}}) \rightarrow$ dest |  |  |  |
| Status Affected: | N, OV, C, DC, Z |  |  |  |
| Encoding: | 0101 | 01da | ffff | ffff |
| Description: | Subtract register ' $f$ ' and Carry flag (borrow) from W (2's complement method). If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored in register ' 4 ' (default). <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $\mathrm{f} \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register 'f' |  |  | Write to destination |
| Example 1: | SUBFWB REG, 1, 0 |  |  |  |
| Before Instruction |  |  |  |  |
|  | 3 |  |  |  |
| W | $=2$ |  |  |  |
| C | - 1 |  |  |  |
| After Instruction |  |  |  |  |
| REG | = FF |  |  |  |
| W | 2 |  |  |  |
| C |  |  |  |  |
| Z | = 0 |  |  |  |
| N | 1 ; result is negative |  |  |  |
| Example 2: SUBFWB REG, 0, 0 |  |  |  |  |
| Before Instruction |  |  |  |  |
|  | 2 |  |  |  |
| W | - 1 |  |  |  |
| C |  |  |  |  |
| After Instruction |  |  |  |  |
| REG | 2 |  |  |  |
|  |  |  |  |  |
| C | - 1 |  |  |  |
| Z | $=\begin{aligned} & 0 \\ & 0\end{aligned}$; result is positive |  |  |  |
| N |  |  |  |  |
| Example 3: | SUBFWB REG, 1, 0 |  |  |  |
| Before Instruction |  |  |  |  |
| REG | 1 <br> 2 |  |  |  |
| W |  |  |  |  |
| After Instruction |  |  |  |  |
| REG | 0 |  |  |  |
| W | - 2 |  |  |  |
| C | $=1$. result is zero |  |  |  |
| $\mathrm{Z}_{\mathrm{N}}$ | 11 ; result is zero |  |  |  |


| SUBLW | Subtract W from Literal |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | SUBLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | $\mathrm{k}-\mathrm{W}) \rightarrow \mathrm{W}$ |  |  |  |
| Status Affected: | N, OV, C, DC, Z |  |  |  |
| Encoding: | 0000 | 1000 | kkkk |  |
| Description | W is subtracted from the eight-bit literal ' $k$ '. The result is placed in W. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' k ' |  |  | Write to W |
| Example 1: | SUBLW | 02h |  |  |
| Before Instruction |  |  |  |  |
| W | 01h |  |  |  |
|  | $=$ ? |  |  |  |
| After Instruction |  |  |  |  |
| W | 01h |  |  |  |
|  | - 1 | ; result is positive |  |  |
|  | $\begin{aligned} & 0 \\ & 0 \end{aligned}$ |  |  |  |
| Example 2: | SUBLW | 02h |  |  |
| Before Instruction |  |  |  |  |
| W | 02h |  |  |  |
| C |  |  |  |  |  |  |  |
| After Instruction |  |  |  |  |
| W | 00h |  |  |  |
|  | 1 1 ; result is zero | ; result is zero |  |  |
|  |  |  |  |  |
| Example 3: | SUBLW | 02h |  |  |
| Before Instruction |  |  |  |  |
| W | 03h |  |  |  |
| C |  |  |  |  |  |  |  |
| After Instruction |  |  |  |  |
| W |  |  |  |  |
| C | $=\quad$ FFh$=0$0 |  |  |  |
| Z | 0 0 1 |  |  |  |
| N | 1 |  |  |  |



| SUBWFB | Subtract W from f with Borrow |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | SUBWFB f \{, d \{,a\}\} |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | (f) - (W) - ( $\overline{\mathrm{C}}) \rightarrow$ dest |  |  |  |
| Status Affected: | N, OV, C, DC, Z |  |  |  |
| Encoding: | 0101 | 10da | ffff | ffff |
| Description: | Subtract W and the Carry flag (borrow) from register ' $f$ ' ( 2 's complement method). If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' $f$ ' (default). <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ ( 5 Fh ). See <br> Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 Q4 |  |  |
| Decode | $\begin{gathered} \text { Read } \\ \text { register ' } f \text { ' } \end{gathered}$ |  |  | Write to destination |
| Example 1: SUBWFB REG, 1, 0 |  |  |  |  |
| Before Instruction |  |  |  |  |
| REG | $=19 \mathrm{~h}$ | (0001 1001) |  |  |
| $\underset{C}{W}$ | $\begin{aligned} & =0 \mathrm{Dh} \\ & =1 \end{aligned}$ | (0000 1101) |  |  |
| After Instruction |  |  |  |  |
| REG | = 0Ch | (0000 1011) |  |  |
| W | $=0 \mathrm{Dh}$ | (0000 1101) |  |  |
| C | $=1$ $=0$ |  |  |  |
| N | $=0$ | ; result is positive |  |  |
| Example 2: | SUBWFB REG, 0, 0 |  |  |  |
| Before Instruction |  |  |  |  |
| REG | $=1 \mathrm{Bh}$ | (0001 1011) |  |  |
| W ${ }_{\text {W }}$ | $=1 \mathrm{Ah}$ $=0$ | (0001 1010) |  |  |
| After Instruction |  |  |  |  |
| REG | $=1 \mathrm{Bh}$ | (0001 1011) |  |  |
| W | $=00 \mathrm{~h}$ |  |  |  |
| C | $=1$ | ; result is zero |  |  |
| \% | $\begin{array}{ll} = & 1 \\ = & 0 \end{array}$ |  |  |  |
| Example 3: | SUBWFB | REG, 1, 0 |  |  |
| Before Instruction |  |  |  |  |
| REG | $=03 \mathrm{~h}$ | (0000 0011) |  |  |
| W | = 0Eh | (0000 1101) |  |  |
| C | $=1$ |  |  |  |
| After Instruction |  |  |  |  |
| REG | $=\mathrm{F} 5 \mathrm{~h}$ | $\begin{aligned} & \left(\begin{array}{ll} 11111 & 0100) \\ ; \\ ; \\ \text { [2's comp] } \end{array}\right. \end{aligned}$ |  |  |
| W | $=0$ Oh | (0000 1101) |  |  |
| C | $=0$ $=0$ |  |  |  |
| N |  | ; result is negative |  |  |




## PIC18F45J10 FAMILY

| TBLWT | Table Write |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | TBLWT ( *; *+; *-; +*) |  |  |  |
| Operands: | None |  |  |  |
| Operation: | if TBLWT *, <br> (TABLAT) $\rightarrow$ Holding Register; <br> TBLPTR - No Change <br> if TBLWT *+, <br> (TABLAT) $\rightarrow$ Holding Register; <br> (TBLPTR) + $1 \rightarrow$ TBLPTR <br> if TBLWT *-, <br> (TABLAT) $\rightarrow$ Holding Register; <br> (TBLPTR) - $1 \rightarrow$ TBLPTR <br> if TBLWT + ${ }^{*}$, <br> (TBLPTR) $+1 \rightarrow$ TBLPTR; <br> (TABLAT) $\rightarrow$ Holding Register |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0000 | 0000 | 0000 | $\begin{array}{rl}11 \mathrm{nn} \\ \mathrm{nn}=0 & \text { * } \\ =1 & *+ \\ =2 & *- \\ =3 & +*\end{array}$ |
| Description: | This instruction uses the 3 LSBs of TBLPTR to determine which of the 8 holding registers the TABLAT is written to. The holding registers are used to program the contents of Program Memory (P.M.). (Refer to Section 6.0 "Flash Program Memory" for additional details on programming Flash memory.) The TBLPTR (a 21-bit pointer) points to each byte in the program memory. <br> TBLPTR has a 2-MByte address range. The LSb of the TBLPTR selects which byte of the program memory location to access. <br> TBLPTR[0] $=0$ : Least Significant Byte of Program Memory Word <br> TBLPTR[0] = 1: Most Significant Byte of Program Memory Word <br> The TBLWT instruction can modify the value of TBLPTR as follows: <br> - no change <br> - post-increment <br> - post-decrement <br> - pre-increment |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 2 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
|  | Decode | No <br> operation | No <br> operation | No operation |
|  | No operation | No operation (Read TABLAT) | No $\begin{gathered}\text { No } \\ \text { operation }\end{gathered}$ | No operation (Write to Holding Register ) |



| XORLW | Exclusive OR Literal with W |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | XORLW k |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 255$ |  |  |  |
| Operation: | (W) . XOR. $\mathrm{k} \rightarrow \mathrm{W}$ |  |  |  |
| Status Affected: | N, Z |  |  |  |
| Encoding: | 0000 | 1010 | kkkk |  |
| Description: | The contents of W are XORed with the 8 -bit literal ' k '. The result is placed in $W$. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read literal ' $k$ ' | Process Data |  | Write to W |
| Example: | XORLW | OAFh |  |  |
| Before Instruction |  |  |  |  |
| W | B5h |  |  |  |
| After Instruction |  |  |  |  |
| W | 1Ah |  |  |  |

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| XORWF | Exclusive OR W with f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | XORWF f $\{, \mathrm{d}\{, \mathrm{a}\}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 255 \\ & d \in[0,1] \\ & a \in[0,1] \end{aligned}$ |  |  |  |
| Operation: <br> Status Affected: | (W). .XOR. (f) $\rightarrow$ dest |  |  |  |
| Encoding: | 0001 | 10da | ffff | ffff |
| Description: | Exclusive OR the contents of W with register ' $f$ '. If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored back in the register ' $f$ ' (default). <br> If ' $a$ ' is ' 0 ', the Access Bank is selected. If ' $a$ ' is ' 1 ', the BSR is used to select the GPR bank (default). <br> If ' $a$ ' is ' 0 ' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 21.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' |  |  | Write to destination |
| Example: | XORWF | REG, 1, 0 |  |  |
| Before Instruction |  |  |  |  |
| REG | AFh |  |  |  |
| W | B5h |  |  |  |
| After Instruction |  |  |  |  |
| REG | 1Ah |  |  |  |
| W | $=B 5 h$ |  |  |  |

### 21.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, PIC18F45J10 family devices also provide an optional extension to the core CPU functionality. The added features include eight additional instructions that augment indirect and indexed addressing operations and the implementation of Indexed Literal Offset Addressing mode for many of the standard PIC18 instructions.
The additional features of the extended instruction set are disabled by default. To enable them, users must set the XINST configuration bit.
The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for indexed addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULNK and SUBULNK) allow for automatic return after execution.
The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- function pointer invocation
- software stack pointer manipulation
- manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 21-3. Detailed descriptions are provided in Section 21.2.2 "Extended Instruction Set". The opcode field descriptions in Table 21-1 (page 242) apply to both the standard and extended PIC18 instruction sets.

Note: The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

### 21.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of indexed addressing, it is enclosed in square brackets ("[ ]"). This is done to indicate that the argument is used as an index or offset. MPASM ${ }^{\text {M }}$ Assembler will flag an error if it determines that an index or offset value is not bracketed.
When the extended instruction set is enabled, brackets are also used to indicate index arguments in byteoriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see Section 21.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands".

Note: In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("\{ \}").

TABLE 21-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

| Mnemonic, Operands |  | Description | Cycles | 16-Bit Instruction Word |  |  |  | Status <br> Affected |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MSb |  |  | LSb |  |
| ADDFSR | f, k |  | Add literal to FSR | 1 | 1110 | 1000 | ffkk | kkkk | None |
| ADDULNK | k | Add literal to FSR2 and return | 2 | 1110 | 1000 | 11 kk | kkkk | None |
| CALLW |  | Call subroutine using WREG | 2 | 0000 | 0000 | 0001 | 0100 | None |
| MOVSF | $z_{s}, f_{d}$ | Move $z_{s}$ (source) to 1 st word $\mathrm{f}_{\mathrm{d}}$ (destination) 2nd word | 2 | 1110 | $\begin{aligned} & 1011 \\ & \text { ffff } \end{aligned}$ | $\begin{aligned} & 0 \mathrm{zzz} \\ & \mathrm{ffff} \end{aligned}$ | $\begin{aligned} & z z z z \\ & \text { ffff } \end{aligned}$ | None |
| MOVSS | $\mathrm{z}_{\mathrm{s}}, \mathrm{z}_{\mathrm{d}}$ | Move $\mathrm{z}_{\mathrm{s}}$ (source) to 1 st word $\mathrm{z}_{\mathrm{d}}$ (destination) 2nd word | 2 | 1110 | 1011 xxxx | $\begin{aligned} & 1 z z z \\ & x z z z \end{aligned}$ | $\begin{aligned} & z z z z \\ & z z z z \end{aligned}$ | None |
| PUSHL | k | Store literal at FSR2, decrement FSR2 | 1 | 1110 | 1010 | kkkk | kkkk | None |
| SUBFSR | f, k | Subtract literal from FSR | 1 | 1110 | 1001 | ffkk | kkkk | None |
| SUBULNK | k | Subtract literal from FSR2 and return | 2 | 1110 | 1001 | 11kk | kkkk | None |

## PIC18F45J10 FAMILY

### 21.2.2 EXTENDED INSTRUCTION SET

| ADDFSR | Add Literal to FSR |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | ADDFSR f, k |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq k \leq 63 \\ & f \in[0,1,2] \end{aligned}$ |  |  |  |
| Operation: | $\mathrm{FSR}(\mathrm{f})+\mathrm{k} \rightarrow \mathrm{FSR}(\mathrm{f})$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1110 | 1000 | ffkk | kkkk |
| Description: | The 6-bit literal ' $k$ ' is added to the contents of the FSR specified by ' f '. |  |  |  |
| Words: | 1 - |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 |  | Q3 | Q4 |
| Decode | Read literal 'k' |  |  | Write to FSR |

Example: $\quad$ ADDFSR 2, 23h
Before Instruction FSR2 $=03 F F h$

After Instruction FSR2 $=0422 \mathrm{~h}$

| ADDULNK | Add Literal to FSR2 and Return |  |  |
| :--- | :--- | :---: | :---: |
| Syntax: | ADDULNK k |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 63$ |  |  |
| Operation: | FSR2 $+\mathrm{k} \rightarrow$ FSR2, |  |  |
|  | (TOS) $\rightarrow \mathrm{PC}$ |  |  |
| Status Affected: | None |  |  |
| Encoding: | 1110 |  |  |
|  | 1000 |  |  |
|  |  |  |  |

The 6-bit literal ' $k$ ' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS.
The instruction takes two cycles to execute; a NOP is performed during the second cycle. This may be thought of as a special case of the ADDFSR instruction, where $f=3$ (binary ' 11 '); it operates only on FSR2.
Words: 1
Cycles: 2

Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> literal 'k' | Process <br> Data | Write to <br> FSR |
| No <br> Operation | No <br> Operation | No <br> Operation | No <br> Operation |

Example: $\quad$ ADDULNK 23 h

| Before Instruction |  |  |
| :---: | :--- | :--- |
| FSR2 | $=03 F F h$ |  |
| PC | $=0100 \mathrm{~h}$ |  |
| After Instruction |  |  |
| FSR2 | $=0422 \mathrm{~h}$ |  |
| PC | $=$ | $(\mathrm{TOS})$ |

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction syntax then becomes: \{label\} instruction argument(s).


| MOVSF | Move Indexed to f |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | MOVSF [ $\mathrm{z}_{\mathrm{s}}$, $\mathrm{f}_{\mathrm{d}}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq \mathrm{z}_{\mathrm{s}} \leq 127 \\ & 0 \leq \mathrm{f}_{\mathrm{d}} \leq 4095 \end{aligned}$ |  |  |  |
| Operation: | $\left((\right.$ FSR2 $\left.)+\mathrm{z}_{\mathrm{s}}\right) \rightarrow \mathrm{f}_{\mathrm{d}}$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: |  |  |  |  |
| 2nd word (destin.) | 1111 | ffff | ffff | ffff |

Description:

Words: 2
Cycles: 2
Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Determine <br> source addr | Determine <br> source addr | Read <br> source reg |
| Decode | No <br> operation <br> No dummy <br> read | No <br> operation | Write <br> register 'f' <br> (dest) |

Example: MOVSF [05h], REG2
Before Instruction

| FSR2 <br> Contents <br> of 85 h | $=80 \mathrm{~h}$ |
| :--- | :--- |
| REG2 | $=33 \mathrm{~h}$ |
| After Instruction | $=11 \mathrm{~h}$ |
| FSR2 | $=80 \mathrm{~h}$ |
| Contents <br> of 85 h | $=33 \mathrm{~h}$ |
| REG2 | $=33 \mathrm{~h}$ |



PUSHL Store Literal at FSR2, Decrement FSR2


## Example: PUSHL 08h

Before Instruction

$$
\begin{array}{ll}
\text { FSR2H:FSR2L } & =01 E C h \\
\text { Memory (01ECh) } & =00 \mathrm{~h}
\end{array}
$$

After Instruction
FSR2H:FSR2L $=01 E B h$

| SUBFSR | Subtract Literal from FSR |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | SUBFSR f, k |  |  |  |
| Operands: | $0 \leq k \leq 63$ |  |  |  |
| Operation: | FSR(f) - $k \rightarrow$ FSRf |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1110 | 1001 | ffkk | kkkk |
| Description: | The 6-bit literal ' k ' is subtracted from the contents of the FSR specified by ' f '. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' |  |  | Write to destination |

Example:

SUBFSR 2, 23 h

Before Instruction
FSR2 $=03 F F h$
After Instruction
FSR2 $=03 D C h$

SUBULNK Subtract Literal from FSR2 and Return

| Syntax: | SUBULNK k |
| :--- | :--- |
| Operands: | $0 \leq \mathrm{k} \leq 63$ |
| Operation: | FSR2 $-\mathrm{k} \rightarrow$ FSR2 |
|  |  |
|  | (TOS) $\rightarrow$ PC |
| Status Affected: | None |
| Encoding: | 1110 |
|  | 1001 |
|  |  |

Description: The 6-bit literal ' $k$ ' is subtracted from the contents of the FSR2. A RETURN is then executed by loading the PC with the TOS. The instruction takes two cycles to execute; a NOP is performed during the second cycle.
This may be thought of as a special case of the SUBFSR instruction, where $f=3$ (binary ' 11 '); it operates only on FSR2.
Words: 1
Cycles: 2
Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read <br> register 'f' | Process <br> Data | Write to <br> destination |
| No <br> Operation | No <br> Operation | No <br> Operation | No <br> Operation |

Example: $\quad$ SUBULNK 23 h
Before Instruction
FSR2 $=03 F F h$
$\mathrm{PC}=0100 \mathrm{~h}$
After Instruction
FSR2 = 03DCh
$\mathrm{PC}=(\mathrm{TOS})$

### 21.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note: Enabling the PIC18 instruction set extension may cause legacy applications to behave erratically or fail entirely.

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing mode (Section 5.5.1 "Indexed Addressing with Literal Offset"). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank ('a' = 0), or in a GPR bank designated by the BSR ('a' = 1). When the extended instruction set is enabled and 'a' = 0 , however, a file register argument of 5 Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument - that is, all byte-oriented and bitoriented instructions, or almost half of the core PIC18 instructions - may behave differently when the extended instruction set is enabled.
When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between $C$ and assembly routines in order to preserve the stack pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 21.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands").
Although the Indexed Literal Offset Addressing mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5 Fh or less are used for Indexed Literal Offset Addressing.
Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset Addressing mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

### 21.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument, ' f ', in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value, ' $k$ '. As already noted, this occurs only when ' $f$ ' is less than or equal to 5 Fh . When an offset value is used, it must be indicated by square brackets ("[ ]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within brackets, will generate an error in the MPASM Assembler.
If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be ' 0 '. This is in contrast to standard operation (extended instruction set disabled) when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.
The destination argument, ' $d$ ', functions as before.
Refer to the MPLAB ${ }^{\circledR}$ IDE, MPASM ${ }^{\text {TM }}$ or MPLAB C18 documentation for information on enabling Extended Instruction set support

### 21.2.4 <br> CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.
Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F45J10 family, it is very important to consider the type of code. A large, re-entrant application that is written in ' $C$ ' and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

| ADDWF | ADD W to Indexed (Indexed Literal Offset mode) |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | ADDWF [k] $\{\mathrm{d}\}$ |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq k \leq 95 \\ & d \in[0,1] \end{aligned}$ |  |  |  |
| Operation: | $(\mathrm{W})+((\mathrm{FSR} 2)+\mathrm{k}) \rightarrow$ dest |  |  |  |
| Status Affected: | N, OV, C, DC, Z |  |  |  |
| Encoding: | 0010 | 01do | kkkk | k ${ }^{\text {k }}$ kkk |
| Description: | The contents of W are added to the contents of the register indicated by FSR2, offset by the value ' $k$ '. If ' $d$ ' is ' 0 ', the result is stored in W. If ' $d$ ' is ' 1 ', the result is stored back in register ' f ' (default). |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read 'k' | $\begin{gathered} \text { Process } \\ \text { Data } \\ \hline \end{gathered}$ |  | Write to destination |
| Example: | ADDWF [OFST], |  |  |  |
| Before Instruction |  |  |  |  |
| W | $=17 \mathrm{~h}$ |  |  |  |
| OFST FSR2 |  | $\begin{aligned} & \text { 2Ch } \\ & \text { OAOOh } \end{aligned}$ |  |  |
| $\begin{aligned} & \begin{array}{l} \text { Contents } \\ \text { of 0A2Ch } \end{array}=20 \mathrm{~h} \end{aligned}$ |  |  |  |  |
| After Instruction |  |  |  |  |
| w Contents of 0A2Ch |  | $=37 \mathrm{~h}$ |  |  |  |
|  | $=20 \mathrm{~h}$ |  |  |  |


| BSF | Bit Set Indexed (Indexed Literal Offset mode) |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | BSF [k], b |  |  |  |
| Operands: | $\begin{aligned} & 0 \leq f \leq 95 \\ & 0 \leq b \leq 7 \end{aligned}$ |  |  |  |
| Operation: | $1 \rightarrow(($ FSR2 $)+\mathrm{k})<\mathrm{b}>$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 1000 | bbbo | kkkk | kkkk |
| Description: | Bit 'b' of the register indicated by FSR2, offset by the value ' $k$ ', is set. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |
| Q Cycle Activity: |  |  |  |  |
| Q1 | Q2 | Q3 |  | Q4 |
| Decode | Read register ' $f$ ' |  |  | rite to tination |

Example: BSF [FLAG_OFST], 7
Before Instruction

| FLAG_OFST | $=0 A h$ |
| :--- | :--- |
| FSR2 | $=0 A 00 \mathrm{~h}$ |
| Contents <br> of OAOAh | $=55 \mathrm{~h}$ |

After Instruction
Contents
of OAOAh $=\mathrm{D} 5 \mathrm{~h}$

| SETF | Set Indexed (Indexed Literal Offset mode) |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Syntax: | SETF [k] |  |  |  |
| Operands: | $0 \leq \mathrm{k} \leq 95$ |  |  |  |
| Operation: | FFh $\rightarrow$ ((FSR2) +k$)$ |  |  |  |
| Status Affected: | None |  |  |  |
| Encoding: | 0110 | 1000 | kkkk | kkkk |
| Description: | The contents of the register indicated by FSR2, offset by ' $k$ ', are set to FFh. |  |  |  |
| Words: | 1 |  |  |  |
| Cycles: | 1 |  |  |  |

Q Cycle Activity:

| Q1 | Q2 | Q3 | Q4 |
| :---: | :---: | :---: | :---: |
| Decode | Read 'k' | Process <br> Data | Write <br> register |

Example: SETF [OFST]

Before Instruction

| OFST | $=2 \mathrm{Ch}$ |
| :--- | :--- |
| FSR2 | $=0 \mathrm{A00h}$ |
| Contents <br> of 0A2Ch | $=00 \mathrm{~h}$ |

of 0A2Ch $=00 \mathrm{~h}$
After Instruction
Contents
of OA2Ch $=$ FFh

## PIC18F45J10 FAMILY

### 21.2.5 SPECIAL CONSIDERATIONS WITH MICROCHIP MPLAB ${ }^{\circledR}$ IDE TOOLS

The latest versions of Microchip's software tools have been designed to fully support the extended instruction set of the PIC18F45J10 family of devices. This includes the MPLAB C18 C compiler, MPASM assembly language and MPLAB Integrated Development Environment (IDE).
When selecting a target device for software development, MPLAB IDE will automatically set default configuration bits for that device. The default setting for the XINST configuration bit is ' 0 ', disabling the extended instruction set and Indexed Literal Offset Addressing mode. For proper execution of applications developed to take advantage of the extended instruction set, XINST must be set during programming.

To develop software for the extended instruction set, the user must enable support for the instructions and the Indexed Addressing mode in their language tool(s). Depending on the environment being used, this may be done in several ways:

- A menu option, or dialog box within the environment, that allows the user to configure the language tool and its settings for the project
- A command line option
- A directive in the source code

These options vary between different compilers, assemblers and development environments. Users are encouraged to review the documentation accompanying their development systems for the appropriate information.

### 22.0 DEVELOPMENT SUPPORT

The PICmicro ${ }^{\circledR}$ microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
- MPLAB ${ }^{\circledR}$ IDE Software
- Assemblers/Compilers/Linkers
- MPASM ${ }^{\text {TM }}$ Assembler
- MPLAB C17 and MPLAB C18 C Compilers
- MPLINK ${ }^{\text {TM }}$ Object Linker/ MPLIB ${ }^{\text {M }}$ Object Librarian
- MPLAB C30 C Compiler
- MPLAB ASM30 Assembler/Linker/Library
- Simulators
- MPLAB SIM Software Simulator
- MPLAB dsPIC30 Software Simulator
- Emulators
- MPLAB ICE 2000 In-Circuit Emulator
- MPLAB ICE 4000 In-Circuit Emulator
- In-Circuit Debugger
- MPLAB ICD 2
- Device Programmers
- PRO MATE ${ }^{\circledR}$ II Universal Device Programmer
- PICSTART ${ }^{\circledR}$ Plus Development Programmer
- MPLAB PM3 Device Programmer
- Low-Cost Demonstration Boards
- PICDEM ${ }^{\text {TM }} 1$ Demonstration Board
- PICDEM.net ${ }^{\text {TM }}$ Demonstration Board
- PICDEM 2 Plus Demonstration Board
- PICDEM 3 Demonstration Board
- PICDEM 4 Demonstration Board
- PICDEM 17 Demonstration Board
- PICDEM 18R Demonstration Board
- PICDEM LIN Demonstration Board
- PICDEM USB Demonstration Board
- Evaluation Kits
- KEELOQ ${ }^{\circledR}$ Evaluation and Programming Tools
- PICDEM MSC
- microlD ${ }^{\circledR}$ Developer Kits
- CAN
- PowerSmart ${ }^{\circledR}$ Developer Kits
- Analog


### 22.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows ${ }^{\circledR}$ based application that contains:

- An interface to debugging tools
- simulator
- programmer (sold separately)
- emulator (sold separately)
- in-circuit debugger (sold separately)
- A full-featured editor with color coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High-level source code debugging
- Mouse over variable inspection
- Extensive on-line help

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PICmicro emulator and simulator tools (automatically updates all project information)
- Debug using:
- source files (assembly or C)
- mixed assembly and C
- machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increasing flexibility and power.

### 22.2 MPASM Assembler

The MPASM assembler is a full-featured, universal macro assembler for all PICmicro MCUs.

The MPASM assembler generates relocatable object files for the MPLINK object linker, Intel ${ }^{\circledR}$ standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.
The MPASM assembler features include:

- Integration into MPLAB IDE projects
- User defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process


### 22.3 MPLAB C17 and MPLAB C18 C Compilers

The MPLAB C17 and MPLAB C18 Code Development Systems are complete ANSI C compilers for Microchip's PIC17CXXX and PIC18CXXX family of microcontrollers. These compilers provide powerful integration capabilities, superior code optimization and ease of use not found with other compilers.
For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

### 22.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK object linker combines relocatable objects created by the MPASM assembler and the MPLAB C17 and MPLAB C18 C compilers. It can link relocatable objects from precompiled libraries, using directives from a linker script.
The MPLIB object librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction


### 22.5 MPLAB C30 C Compiler

The MPLAB C30 C compiler is a full-featured, ANSI compliant, optimizing compiler that translates standard ANSI C programs into dsPIC30F assembly language source. The compiler also supports many command line options and language extensions to take full advantage of the dsPIC30F device hardware capabilities and afford fine control of the compiler code generator.
MPLAB C30 is distributed with a complete ANSI C standard library. All library functions have been validated and conform to the ANSI C library standard. The library includes functions for string manipulation, dynamic memory allocation, data conversion, timekeeping and math functions (trigonometric, exponential and hyperbolic). The compiler provides symbolic information for high-level source debugging with the MPLAB IDE.

### 22.6 MPLAB ASM30 Assembler, Linker and Librarian

MPLAB ASM30 assembler produces relocatable machine code from symbolic assembly language for dsPIC30F devices. MPLAB C30 compiler uses the assembler to produce it's object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire dsPIC30F instruction set
- Support for fixed-point and floating-point data
- Command line interface
- Rich directive set
- Flexible macro language
- MPLAB IDE compatibility


### 22.7 MPLAB SIM Software Simulator

The MPLAB SIM software simulator allows code development in a PC hosted environment by simulating the PICmicro series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user defined key press, to any pin. The execution can be performed in Single-Step, Execute Until Break or Trace mode.
The MPLAB SIM simulator fully supports symbolic debugging using the MPLAB C17 and MPLAB C18 C Compilers, as well as the MPASM assembler. The software simulator offers the flexibility to develop and debug code outside of the laboratory environment, making it an excellent, economical software development tool.

### 22.8 MPLAB SIM30 Software Simulator

The MPLAB SIM30 software simulator allows code development in a PC hosted environment by simulating the dsPIC30F series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user defined key press, to any of the pins.
The MPLAB SIM30 simulator fully supports symbolic debugging using the MPLAB C30 C Compiler and MPLAB ASM30 assembler. The simulator runs in either a Command Line mode for automated tasks, or from MPLAB IDE. This high-speed simulator is designed to debug, analyze and optimize time intensive DSP routines.

### 22.9 MPLAB ICE 2000 High-Performance Universal In-Circuit Emulator

The MPLAB ICE 2000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PICmicro microcontrollers. Software control of the MPLAB ICE 2000 in-circuit emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.
The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the MPLAB ICE in-circuit emulator allows expansion to support new PICmicro microcontrollers.
The MPLAB ICE 2000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft ${ }^{\circledR}$ Windows 32 -bit operating system were chosen to best make these features available in a simple, unified application.

### 22.10 MPLAB ICE 4000 High-Performance Universal In-Circuit Emulator

The MPLAB ICE 4000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for highend PICmicro microcontrollers. Software control of the MPLAB ICE in-circuit emulator is provided by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICD 4000 is a premium emulator system, providing the features of MPLAB ICE 2000, but with increased emulation memory and high-speed performance for dsPIC30F and PIC18XXXX devices. Its advanced emulator features include complex triggering and timing, up to 2 Mb of emulation memory and the ability to view variables in real-time.
The MPLAB ICE 4000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

### 22.11 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low-cost, run-time development tool, connecting to the host PC via an RS-232 or high-speed USB interface. This tool is based on the Flash PICmicro MCUs and can be used to develop for these and other PICmicro microcontrollers. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the Flash devices. This feature, along with Microchip's In-Circuit Serial Programming ${ }^{\text {TM }}$ (ICSP ${ }^{\text {TM }}$ ) protocol, offers cost effective in-circuit Flash debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single-stepping and watching variables, CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real-time. MPLAB ICD 2 also serves as a development programmer for selected PICmicro devices.

### 22.12 PRO MATE II Universal Device Programmer

The PRO MATE II is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VdDMAX for maximum reliability. It features an LCD display for instructions and error messages and a modular detachable socket assembly to support various package types. In Stand-Alone mode, the PRO MATE II device programmer can read, verify and program PICmicro devices without a PC connection. It can also set code protection in this mode.

### 22.13 MPLAB PM3 Device Programmer

The MPLAB PM3 is a universal, CE compliant device programmer with programmable voltage verification at Vddmin and Vddmax for maximum reliability. It features a large LCD display ( $128 \times 64$ ) for menus and error messages and a modular detachable socket assembly to support various package types. The ICSP ${ }^{\text {TM }}$ cable assembly is included as a standard item. In StandAlone mode, the MPLAB PM3 device programmer can read, verify and program PICmicro devices without a PC connection. It can also set code protection in this mode. MPLAB PM3 connects to the host PC via an RS232 or USB cable. MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an SD/MMC card for file storage and secure data applications.

### 22.14 PICSTART Plus Development Programmer

The PICSTART Plus development programmer is an easy-to-use, low-cost, prototype programmer. It connects to the PC via a COM (RS-232) port. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. The PICSTART Plus development programmer supports most PICmicro devices up to 40 pins. Larger pin count devices, such as the PIC16C92X and PIC17C76X, may be supported with an adapter socket. The PICSTART Plus development programmer is CE compliant.

### 22.15 PICDEM 1 PICmicro Demonstration Board

The PICDEM 1 demonstration board demonstrates the capabilities of the PIC16C5X (PIC16C54 to PIC16C58A), PIC16C61, PIC16C62X, PIC16C71, PIC16C8X, PIC17C42, PIC17C43 and PIC17C44. All necessary hardware and software is included to run basic demo programs. The sample microcontrollers provided with the PICDEM 1 demonstration board can be programmed with a PRO MATE II device programmer or a PICSTART Plus development programmer. The PICDEM 1 demonstration board can be connected to the MPLAB ICE in-circuit emulator for testing. A prototype area extends the circuitry for additional application components. Features include an RS-232 interface, a potentiometer for simulated analog input, push button switches and eight LEDs.

### 22.16 PICDEM.net Internet/Ethernet Demonstration Board

The PICDEM.net demonstration board is an Internet/ Ethernet demonstration board using the PIC18F452 microcontroller and TCP/IP firmware. The board supports any 40-pin DIP device that conforms to the standard pinout used by the PIC16F877 or PIC18C452. This kit features a user friendly TCP/IP stack, web server with HTML, a 24L256 Serial EEPROM for Xmodem download to web pages into Serial EEPROM, ICSP/MPLAB ICD 2 interface connector, an Ethernet interface, RS-232 interface and a $16 \times 2$ LCD display. Also included is the book and CD-ROM "TCP/IP Lean, Web Servers for Embedded Systems," by Jeremy Bentham

### 22.17 PICDEM 2 Plus Demonstration Board

The PICDEM 2 Plus demonstration board supports many 18, 28 and 40-pin microcontrollers, including PIC16F87X and PIC18FXX2 devices. All the necessary hardware and software is included to run the demonstration programs. The sample microcontrollers provided with the PICDEM 2 demonstration board can be programmed with a PRO MATE II device programmer, PICSTART Plus development programmer, or MPLAB ICD 2 with a Universal Programmer Adapter. The MPLAB ICD 2 and MPLAB ICE in-circuit emulators may also be used with the PICDEM 2 demonstration board to test firmware. A prototype area extends the circuitry for additional application components. Some of the features include an RS-232 interface, a $2 \times 16$ LCD display, a piezo speaker, an on-board temperature sensor, four LEDs and sample PIC18F452 and PIC16F877 Flash microcontrollers.

### 22.18 PICDEM 3 PIC16C92X Demonstration Board

The PICDEM 3 demonstration board supports the PIC16C923 and PIC16C924 in the PLCC package. All the necessary hardware and software is included to run the demonstration programs.

### 22.19 PICDEM 4 8/14/18-Pin Demonstration Board

The PICDEM 4 can be used to demonstrate the capabilities of the 8, 14 and 18-pin PIC16XXXX and PIC18XXXX MCUs, including the PIC16F818/819, PIC16F87/88, PIC16F62XA and the PIC18F1320 family of microcontrollers. PICDEM 4 is intended to showcase the many features of these low pin count parts, including LIN and Motor Control using ECCP. Special provisions are made for low-power operation with the supercapacitor circuit and jumpers allow onboard hardware to be disabled to eliminate current draw in this mode. Included on the demo board are provisions for Crystal, RC or Canned Oscillator modes, a five volt regulator for use with a nine volt wall adapter or battery, DB-9 RS-232 interface, ICD connector for programming via ICSP and development with MPLAB ICD 2, $2 \times 16$ liquid crystal display, PCB footprints for H-Bridge motor driver, LIN transceiver and EEPROM. Also included are: header for expansion, eight LEDs, four potentiometers, three push buttons and a prototyping area. Included with the kit is a PIC16F627A and a PIC18F1320. Tutorial firmware is included along with the User's Guide.

### 22.20 PICDEM 17 Demonstration Board

The PICDEM 17 demonstration board is an evaluation board that demonstrates the capabilities of several Microchip microcontrollers, including PIC17C752, PIC17C756A, PIC17C762 and PIC17C766. A programmed sample is included. The PRO MATE II device programmer, or the PICSTART Plus development programmer, can be used to reprogram the device for user tailored application development. The PICDEM 17 demonstration board supports program download and execution from external on-board Flash memory. A generous prototype area is available for user hardware expansion.

### 22.21 PICDEM 18R PIC18C601/801 Demonstration Board

The PICDEM 18R demonstration board serves to assist development of the PIC18C601/801 family of Microchip microcontrollers. It provides hardware implementation of both 8-bit Multiplexed/Demultiplexed and 16-bit Memory modes. The board includes 2 Mb external Flash memory and 128 Kb SRAM memory, as well as serial EEPROM, allowing access to the wide range of memory types supported by the PIC18C601/801.

### 22.22 PICDEM LIN PIC16C43X Demonstration Board

The powerful LIN hardware and software kit includes a series of boards and three PICmicro microcontrollers. The small footprint PIC16C432 and PIC16C433 are used as slaves in the LIN communication and feature on-board LIN transceivers. A PIC16F874 Flash microcontroller serves as the master. All three microcontrollers are programmed with firmware to provide LIN bus communication.

### 22.23 PICkit $^{\text {TM }} 1$ Flash Starter Kit

A complete "development system in a box", the PICkit™ Flash Starter Kit includes a convenient multi-section board for programming, evaluation and development of 8/14-pin Flash PIC ${ }^{\circledR}$ microcontrollers. Powered via USB, the board operates under a simple Windows GUI. The PICkit 1 Starter Kit includes the User's Guide (on CD ROM), PICkit 1 tutorial software and code for various applications. Also included are MPLAB ${ }^{\circledR}$ IDE (Integrated Development Environment) software, software and hardware "Tips 'n Tricks for 8-pin Flash PIC ${ }^{\circledR}$ Microcontrollers" Handbook and a USB interface cable. Supports all current 8/14-pin Flash PIC microcontrollers, as well as many future planned devices.

### 22.24 PICDEM USB PIC16C7X5 Demonstration Board

The PICDEM USB Demonstration Board shows off the capabilities of the PIC16C745 and PIC16C765 USB microcontrollers. This board provides the basis for future USB products.

### 22.25 Evaluation and Programming Tools

In addition to the PICDEM series of circuits, Microchip has a line of evaluation kits and demonstration software for these products.

- KEELOQ evaluation and programming tools for Microchip's HCS Secure Data Products
- CAN developers kit for automotive network applications
- Analog design boards and filter design software
- PowerSmart battery charging evaluation/ calibration kits
- $\operatorname{IrDA}{ }^{\circledR}$ development kit
- microID development and rfLab ${ }^{\text {TM }}$ development software
- SEEVAL ${ }^{\circledR}$ designer kit for memory evaluation and endurance calculations
- PICDEM MSC demo boards for Switching mode power supply, high-power IR driver, delta sigma ADC and flow rate sensor
Check the Microchip web page and the latest Product Selector Guide for the complete list of demonstration and evaluation kits.


## PIC18F45J10 FAMILY

NOTES:

## PIC18F45J10 FAMILY

### 23.0 ELECTRICAL CHARACTERISTICS



Note 1: Power dissipation is calcula ted as tollows:
Pdis $=$ VDD $\times\left\{I D D-\sum(1 \varnothing H\}+\sum\{(\mathrm{VDD}-\mathrm{VOH}) \times \mathrm{IOH}\}+\sum(\mathrm{VOL} \times \mathrm{lOL})\right.$
$\dagger$ NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress ratingonly and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect oevice reliability.


## PIC18F45J10 FAMILY

FIGURE 23-1: PIC18LF45J10 FAMILY VOLTAGE-FREQUENCY GRAPH (INDUSTRIAL)


FIGURE 23-2: PIC18LF45 10 FANHNY VOLTAGE-FREQUENCY GRAPH (INDUSTRIAL)

23.1 DC Characteristics: Supply Voltage

PIC18F24J10/25J10/44J10/45J10 (Industrial)
PIC18LF24J10/25J10/44J10/45J10 (Industrial)

| PIC18F45J10 Family (Industrial) |  |  | Standard Operating Conditions (unless otherwise stated) Operating temperature $\quad-40^{\circ} \mathrm{C} \leq \mathrm{TA} \leq+85^{\circ} \mathrm{C}$ for industrial |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Param No. | Symbol | Characteristic | Min $\quad$ Typ | Max | Units | Conditions |
| D001 | VDD | Supply Voltage | VDDCORE ${ }^{(2)}$ | 3.6 | V |  |
| D001B | Vddcore | External Supply for Microcontroller Core | 2.0 | 2.75 |  | Valid only in parts designated "LF". See Section 20.3 "On-Chip Voltage-Regulator" for details. |
| D002 | VDR | RAM Data Retention Voltage ${ }^{(1)}$ | 1.5 |  | $2$ |  |
| D003 | VPOR | Vdd Start Voltage to ensure internal Power-on Reset signal |  | TB | $\sqrt{V}$ | See Section 4.3 "Power-on Reset (POR)" for details |
| D004 | SVDD | Vdd Rise Rate to ensure internal Power-on Reset signal |  | 1 | $\mathrm{V} / \mathrm{ms}$ | See Section 4.3 "Power-on Reset (POR)" for details |
| D005A | VBOR | Brown-out Reset Voltage |  | 2.6 | V | On-chip voltage regulator enabled. See Section 4.3 "Power-on Reset (POR)" for details. |

Legend: TBD = To Be Determined
Note 1: This is the limit to which VDD can be lowered in Sleep mode, or during a device Reset, without losing RAM data.
2: In parts designated with an"ت", XpQ $=2.7 \mathrm{~V}$.


### 23.2 DC Characteristics: Power-Down and Supply Current PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial)



OSC1 $=$ external square wave, from rail-to rail; all $M$ O pins tri-stated, pulled to VDD;
$\overline{M C L R}=$ VDD; WDT enabled/disabled as specifird.
3: For RC oscillator configurations, current through Rextis hot included. The current through the resistor can be estimated by the formula $\mathrm{Ir}=\mathrm{VDD} / 2 \operatorname{REXT}(\mathrm{~mA})$ with REXT in Ke.
4: Standard, low-cost 32 kHz crystals have an peerating temperature range of $-10^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Extended temperature crystals are available at a much higher cest


### 23.2 DC Characteristics: Power-Down and Supply Current PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)



### 23.2 DC Characteristics: Power-Down and Supply Current <br> PIC18F24J10/25J10/44J10/45J10 (Industrial) <br> PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)

| PIC18F45J10 Family (Industrial) |  | Standard Operating Conditions (unless otherwise stated) Operating temperature $\quad-40^{\circ} \mathrm{C} \leq \mathrm{TA} \leq+85^{\circ} \mathrm{C}$ for industrial |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Param No. | Device | Typ | Max | Units |  | Con |  |
|  | Supply Current (IDD) ${ }^{(2,3)}$ |  |  |  |  |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ | $\mathrm{VDD}=2.5 \mathrm{~V}$ | $\text { FOSC }=1 \mathrm{MHz}$ |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | mA | $-40^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | mA | $+25^{\circ} \mathrm{C}$ |  |  |  |
|  |  | TBD | TBD | mA | $+85^{\circ} \mathrm{C}$ |  |  |  |
|  | All devices | TBD | TBD | mA | $-40^{\circ} \mathrm{C}$ | N/V | Fosc $=4 \mathrm{MHz}$ (PRI_RUN mode, EC oscillator) |
|  |  | TBD | TBD | mA | $+25^{\circ} \mathrm{C}$ | $V B D=2.5 \mathrm{~V}$ |  |
|  |  | TBD | TBD | mA | $+85^{\circ} \mathrm{C}$ | $\square$ |  |
|  | All devices | TBD | TBD | mA | $-40^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | mA | $+25^{\circ} \mathrm{C}$ | $\mathrm{VD}=3.3 \mathrm{~V}$ |  |
|  |  | TBD | TBD | mA | $+85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | mA | $y^{-40^{\circ} \mathrm{C}}$ | $\mathrm{VDD}=2.5 \mathrm{~V}$ | $\text { Fosc }=40 \mathrm{MHz}$ <br> (PRI_RUN mode, EC oscillator) |
|  |  | TBD | TBD | mA | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | mA | $\lambda 85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | $m \mathrm{~A}$ | $-40^{\circ} \mathrm{C}$ | $\mathrm{VDD}=3.3 \mathrm{~V}$ |  |
|  |  | TBD | TBD | mA) | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | n $A$ | $+85^{\circ} \mathrm{C}$ |  |  |

Legend: TBD = To Be Determined
Note 1: The power-down current in Sleer hode dos not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with-atl//Dpins in high-impedance state and tied to VDD or Vss and all features that add delta current disabled (such as WDT, Timery oscillator, BOR, etc.).
2: The supply current is maink a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscinator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
$\widehat{M G \perp R}=V D D, W D T$ enabled/disabled as specified.
3: For RCOscikator Qonflgurations, current through REXT is not included. The current through the resistor can be estimated by the formula $l /=\operatorname{VDD/2REXT}(\mathrm{mA})$ with REXT in $\mathrm{k} \Omega$.
4: St\&ndard, O w- Cost 32 kHz crystals have an operating temperature range of $-10^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Extended temperature crystals are available at a much higher cost.


### 23.2 DC Characteristics: Power-Down and Supply Current PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)



Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc, $\rangle$.
2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and cirevit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements in active operation mode are:

> OSC1 $=$ external square puave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
> $\overline{M C L R}=$ VDD; WDT enabledddisabled as specified.

3: For RC oscillator configurations, \&urrentryrough REXT is not included. The current through the resistor can be estimated by the formula $\mathrm{Ir}=\mathrm{VDD} / 2 \mathrm{REXI}$ ( mA ) with REXT in $\mathrm{k} \Omega$.
4: Standard, low-cost 32 kHz erystałs have an operating temperature range of $-10^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Extended temperature crystals are availablegt a mach higher cost.

### 23.2 DC Characteristics: Power-Down and Supply Current PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)

| PIC18F45J10 Family (Industrial) |  | Standard Operating Conditions (unless otherwise stated) Operating temperature $\quad-40^{\circ} \mathrm{C} \leq \mathrm{TA} \leq+85^{\circ} \mathrm{C}$ for industrial |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Param No. | Device | Typ | Max | Units |  | Con | < |
|  | Supply Current (IDD) ${ }^{(2,3)}$ |  |  |  |  |  | ) |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ | $\mathrm{V} D \mathrm{D}=2.5 \mathrm{~V}$ | ( |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ |  | M |
|  | All devices | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ | $\angle D D=2.5 v$ | Fosc $=4 \mathrm{MHz}$ (PRI_IDLE mode, EC oscillator) |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | mA | $-40^{\circ} \mathrm{C}$ | $\bigcirc V D D=3.3 \mathrm{~V}$ |  |
|  |  | TBD | TBD | mA | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | mA | $+85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | mA | $-40^{\circ} \mathrm{C}$ | $\mathrm{VDD}=2.5 \mathrm{~V}$ | Fosc $=40 \mathrm{MHz}$ (PRI_IDLE mode, EC oscillator) |
|  |  | TBD | TBD | mA | + $25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | ma | $+85^{\circ} \mathrm{C}$ |  |  |
|  | All devices | TBD | TBD | $m \mathrm{~A}$ | $-40^{\circ} \mathrm{C}$ | $\mathrm{VDD}=3.3 \mathrm{~V}$ |  |
|  |  | TBD | TBQ | mA | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | ma | $+85^{\circ} \mathrm{C}$ |  |  |

Legend: TBD = To Be Determined
Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all IVQ pins in high-impedance state and tied to VDD or Vss and all features that add delta current disabled (such as $\forall$ DDT, Timerr oscillator, BOR, etc.).
2: The supply current is malnly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillatortype and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions forall
OSC1 = externat square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
$\overline{\mathrm{MCCBR}}=\mathrm{VBP}, \mathrm{WDT}$ enabled/disabled as specified.
3: For RC gscivaroy centigurations, current through REXT is not included. The current through the resistor can be estimated by the pomula $\operatorname{lr}=\mathrm{VDD} / 2$ REXT ( mA ) with REXT in $\mathrm{k} \Omega$.
4: Standard, lqw-cost 32 kHz crystals have an operating temperature range of $-10^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Extended temperature crystals ane available at a much higher cost.


### 23.2 DC Characteristics: Power-Down and Supply Current PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)



Legend: TBD = To Be Determined
Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or Vss and all features that add delta current disabled (such as WDT, Timer1 oscillato BQR, etc).
2: The supply current is mainly a function of operating yotage, frequency and mode. Other factors, such as $1 / O$ pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements intactive operation mode are:

> | $\mathrm{OSC1}=$ external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD; |
| :--- |
| $\mathrm{MCLR}=$ VDD; WDT enabledydisabled as specified. |

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be estimated by the formula $\mathrm{Ir}=\mathrm{VDD} / 2$ REXT $(\mathrm{mA})$ with REXT in $\mathrm{k} \Omega$.
4: Standard, low-cost 32 kHz erystals, have an operating temperature range of $-10^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Extended temperature crystals are available at a nuch higher cost.

### 23.2 DC Characteristics: Power-Down and Supply Current PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)

| PIC18F45J10 Family (Industrial) |  | Standard Operating Conditions (unless otherwise stated) Operating temperature $\quad-40^{\circ} \mathrm{C} \leq \mathrm{TA} \leq+85^{\circ} \mathrm{C}$ for industrial |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Param No. | Device | Typ | Max | Units |  | Cond | 5 |
| D022 ( $\Delta$ IWDT) | Module Differential Currents ( $\triangle$ IWDT, ${ }^{\text {a }}$ IOSCB, ${ }^{\text {IIAD }}$ ) |  |  |  |  |  |  |
|  | Watchdog Timer | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ | $\mathrm{VDD}=2.5 \mathrm{~V}$ |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ |  |  |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ | , |  |
| D025 ( $\Delta$ IOSCB) | Timer1 Oscillator | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ | 入 |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ | $V D D=2.54$ | 32 kHz on Timer1 ${ }^{(3)}$ |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ | $\cdots$ |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ | ) |  |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+25^{\circ} \mathrm{C}$ | $V \mathrm{VB}=3.3 \mathrm{~V}$ | 32 kHz on Timer1 ${ }^{(3)}$ |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $+85^{\circ} \mathrm{C}$ |  |  |
| $\begin{array}{\|l} \hline \text { D026 } \\ \text { ( } 1 \text { IAD) } \end{array}$ | A/D Converter | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{c} \times 10855 \mathrm{c}$ | $V \mathrm{VDD}=2.5 \mathrm{~V}$ | A/D on, not converting |
|  |  | TBD | TBD | $\mu \mathrm{A}$ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $\mathrm{VDD}=3.3 \mathrm{~V}$ |  |

Legend: TBD = To Be Determined
Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-îppedance state and tied to VDD or Vss and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).
2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and cireuiit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements inacetive operation mode are:
OSC1 $=$ external square waye, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
$\overline{M C L R}=$ VDD; WDT enabled/disabled as specified.
3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be estimated by the formula $\mathrm{Ir}=\mathrm{VDD} / 2$ REXT $(\mathrm{mA})$ with REXT in $\mathrm{k} \Omega$.
4: Standard, low-cost 32 kHz crysta) have an operating temperature range of $-10^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Extended temperature crystals are available at a nueh higher cost.


### 23.3 DC Characteristics: PIC18F24J10/25J10/44J10/45J10 (Industrial) PIC18LF24J10/25J10/44J10/45J10 (Industrial)



## Legend: $\triangle B D=$ To Be Determined

Note 1: AR scillator configuration, the OSC1/CLKI pin is a Schmitt Trigger input. It is not recommended that the $^{\text {n }}$
RIChicro ${ }^{\circledR}$ device be driven with an external clock while in RC mode.
2. The leakage current on the $\overline{M C L R}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.
3: Negative current is defined as current sourced by the pin.

### 23.3 DC Characteristics: PIC18F24J10/25J10/44J10/45J10 (Industrial)

PIC18LF24J10/25J10/44J10/45J10 (Industrial) (Continued)


Legend: TBD = To Be Determined
Note 1: In RC oscillator configuration, the $\varnothing$ SG1/CLKI pin is a Schmitt Trigger input. It is not recommended that the PICmicro ${ }^{\circledR}$ device be driven with an external clock while in RC mode.
2: The leakage current on the MGLR pin is strongly dependent on the applied voltage level. The specified levels represent normal eperating conditions. Higher leakage current may be measured at different input voltages.
3: Negative currest is defined as current sourced by the pin.

## TABLE 23-1: MEMORY PROGRAMMING REQUIREMENTS


$\dagger$ Data in "Typ" column is at $3.3 \mathrm{~V}, 25^{\circ} \mathrm{C}$ unless otherwise stated. Theselpakaneters are for design guidance only and are not tested.


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TABLE 23-2: COMPARATOR SPECIFICATIONS
Operating Conditions: $3.0 \mathrm{~V}<\mathrm{VDD}<3.6 \mathrm{~V},-40^{\circ} \mathrm{C}<\mathrm{TA}<+85^{\circ} \mathrm{C}$ (unless otherwise stated)

| Param No. | Sym | Characteristics | Min | Typ | Max | Units | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D300 | VIoff | Input Offset Voltage | - | $\pm 5.0$ | $\pm 10$ | mV |  |
| D301 | VICM | Input Common Mode Voltage* | 0 | - | Vdd - 1.5 | V | 5 |
| D302 | CMRR | Common Mode Rejection Ratio* | 55 | - | - | dB |  |
| 300 | TRESP | Response Time ${ }^{(1)^{*}}$ | - | 150 | 400 |  | ( ) |
| 301 | Tmc2ov | Comparator Mode Change to Output Valid* | - | - | 10 |  |  |

Note 1: Response time measured with one comparator input at (VDD-1.5)/2, while the other input transitions from Vss to VDD.

TABLE 23-3: VOLTAGE REFERENCE SPECIFICATIONS
Operating Conditions: $3.0 \mathrm{~V}<\mathrm{VDD}<3.6 \mathrm{~V},-40^{\circ} \mathrm{C}<\mathrm{TA}<+85^{\circ} \mathrm{C}$ (unless otherwise stated)

| Param <br> No. | Sym | Characteristics | Min | TyP | Max | Units | Comments |  |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| D310 | VRES | Resolution | VDD/24 |  | VDD/32 | LSb |  |  |
| D311 | VRAA | Absolute Accuracy |  | - | $1 / 2$ | LSb |  |  |
| D312 | VRUR | Unit Resistor Value (R) |  |  | 2 k | - | $\Omega$ |  |
| 310 | TSET | Settling Time ${ }^{(\mathbf{1 )}}$ |  |  | - | 10 | $\mu \mathrm{~S}$ |  |

Note 1: Settling time measured while CVRR =1 and CyR3:CVR0 transitions from ' 0000 ' to ' 1111 '.

TABLE 23-4: INTERNAL VOLTAGE REGYLATOR SPECIFICATIONS

| Operating Conditions: $-40^{\circ} \mathrm{C}<7 A<+85^{\circ} \mathrm{C}$ (unless otherwise stated) |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Param No. | Sym | Cnaracteristics | Min | Typ | Max | Units | Comments |
|  | VRgout | RegulatokOutput Voltage | - | 2.5 | - | V |  |
|  | Cefc | External Fifter Capacitor Varue | 1 | 10 | - | $\mu \mathrm{F}$ | Capacitor must be low ESR |

* These parameters are characterized but not tested. Parameter numbers not yet assigned for these specifications.


### 23.4 AC (Timing) Characteristics

### 23.4.1 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created following one of the following formats:

| 1. TppS2ppS <br> 2. TppS |
| :--- |
| T 3. Tcc:ST $\left(I^{2} \mathrm{C}\right.$ specifications only $)$  <br> F Frequency 4. Ts $\left(I^{2} \mathrm{C}\right.$ specifications only $)$ |

Lowercase letters (pp) and their meanings:

| pp |  |  |  |
| :--- | :--- | :--- | :--- |
| cc | CCP1 | osc | OSC1 |
| ck | CLKO | rd | $\overline{\mathrm{RD}}$ |
| cs | $\overline{\mathrm{CS}}$ | rw | $\overline{\mathrm{RD}}$ or $\overline{\mathrm{WR}}$ |
| di | SDI | sc | SCK |
| do | SDO | ss | $\overline{\mathrm{SS}}$ |
| dt | Data in | t0 | TOCKI |
| io | I/O port | t 1 | $\overline{\mathrm{~T} 1 \mathrm{CKI}}$ |
| mc | $\overline{\mathrm{MCLR}}$ | wr | $\overline{\mathrm{WR}}$ |


| S |  |  |
| :---: | :---: | :---: |
| F Fall | P | Period |
| H High | R | Rise |
| 1 Invalid (High-impedance) | V | Valid |
| L Low | Z | High-impedance |
| $I^{2} \mathrm{C}$ only |  |  |
| AA output access | High | High |
| BUF Bus free | Low | Low |
| TCC:ST ( ${ }^{2} \mathrm{C}$ specifications only) |  |  |
| CC |  |  |
| HD Hold | SU | Setup |
| ST |  |  |
| DAT DATA input hold | STO | Stop condition |
| STA Start condition |  |  |

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### 23.4.2 TIMING CONDITIONS

The temperature and voltages specified in Table 23-5 apply to all timing specifications unless otherwise noted. Figure 23-3 specifies the load conditions for the timing specifications.

TABLE 23-5: TEMPERATURE AND VOLTAGE SPECIFICATIONS - AC

|  | Standard Operating Conditions (unless otherwise stated) |
| :--- | :--- |
| AC CHARACTERISTICS | Operating temperature $-40^{\circ} \mathrm{C} \leq \mathrm{TA} \leq+85^{\circ} \mathrm{C}$ for industrial |
|  | Operating voltage VDD range as described in DC spec Section 23.1 and |
|  | Section 23.3. |

FIGURE 23-3: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS
Load Condition 1

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### 23.4.3 TIMING DIAGRAMS AND SPECIFICATIONS

FIGURE 23-4: EXTERNAL CLOCK TIMING (ALL MODES EXCEPT PLL)


TABLE 23-6: EXTERNAL CLOCK TIMING REQUIREMENTS

| Param. <br> No. | Symbol | Characteristic | Min | Conditions |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |

Note 1: Instruction cycle period (TCY) eqqals four times the input oscillator time base period for all configurations. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the "max." cycle time limit is "DC" (no clock) for all devices.

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TABLE 23-7: PLL CLOCK TIMING SPECIFICATIONS (VDD = 2.5V TO 3.6V)

| Param <br> No. | Sym | Characteristic | Min | Typt | Max | Units | Conditions |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: | :---: |
| F10 | FosC | Oscillator Frequency Range | 4 | - | 10 | MHz | HSmode mnly |
| F11 | FSYS | On-Chip VCO System Frequency | 16 | - | 40 | MHz | HS modde only |
| F13 | CLK | CLKO Stability (Jitter) | -2 | - | +2 | - |  |

$\dagger$ Data in "Typ" column is at $5 \mathrm{~V}, 25^{\circ} \mathrm{C}$, unless otherwise stated. These parameters 2 re for design guidance only and are not tested.

TABLE 23-8: AC CHARACTERISTICS: INTERNAL RC ACCURACY
PIC18F24J10/25J10/44J10745 y 10 (NDDUSTRIAL)

| Param <br> No. | Characteristic | Min | Typ | Max | Conditions |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | INTRC Accuracy @ Freq $=31 \mathrm{kHz}^{(1)}$ | 26.562 | - | 35.938 | kHz | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=2.0-3.3 \mathrm{~V}$ |

Note 1: INTRC frequency after calibration. Change of INTRC fequency as VDD changes.


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FIGURE 23-5: CLKO AND I/O TIMING


TABLE 23-9: CLKO AND I/O TIMING REQUIREMENTS

| Param No. | Symbol | Characteristic | Min | Typ | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 10 | TosH2ckL | OSC1 $\uparrow$ to CLKO $\downarrow>$ | - | TBD | TBD | ns |  |
| 11 | TosH2ckH | OSC1 $\uparrow$ to CLKO $\uparrow<$ | - | TBD | TBD | ns |  |
| 12 | TckR | CLKO Rise Time | - | TBD | TBD | ns |  |
| 13 | TCKF | CLKO Fall Time | - | TBD | TBD | ns |  |
| 14 | TскL2ıOV | CLKO $\downarrow$ to Rort Out Valid | - | - | 0.5 TCY + 20 | ns |  |
| 15 | TıoV2ckH | Port In Valid before CLKO $\uparrow$ | 0.25 TCY + 25 | - | - | ns |  |
| 16 | TckH2ıOI | Port In Hold a atter CLKO $\uparrow$ | 0 | - | - | ns |  |
| 17 | TosH2ıOV | O\$GT + (Q ¢ cycle) to Port Out Valid | - | 50 | 150 | ns |  |
| 18 | TosH2IOI QS@ 1 (Q2 cycle) to Port Input Invalid |  | 100 | - | - | ns |  |
| 18A |  |  | 200 | - | - | ns |  |
| 19 | Tiol2osh | Rery Input Valid to OSC1 $\uparrow$ <br> ( $/ / \mathrm{O}$ in setup time) | 0 | - | - | ns |  |
| 20 |  | Port Output Rise Time | - | TBD | TBD | ns |  |
| $20 A$ |  |  | - | - | TBD | ns |  |
| $\begin{aligned} & \hline 21 \\ & 21 \mathrm{~A} \end{aligned}$ |  | Port Output Fall Time | - | TBD | TBD | ns |  |
|  |  | - | - | TBD | ns |  |
| 22† | TINP |  | INT pin High or Low Time | TCY | - | - | ns |  |
| 23† | TRBP | RB7:RB4 Change INT High or Low Time | TCY | - | - | ns |  |

Legend: $\quad \mathrm{TBD}=\mathrm{To}$ Be Determined
$\dagger$ These parameters are asynchronous events not related to any internal clock edges.

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FIGURE 23-6: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND


FIGURE 23-7: BROWN-OUT RESET TIMING


TABLE 23-10: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET REQUIREMENTS

| Param <br> No. | Characteristic | Min | Typ | Max | Units | Conditions |  |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: | :---: |
| 30 | TMOL | MCLR Pulse Width (low) | 2 | - | - | $\mu \mathrm{s}$ |  |
| 31 | TVIDT | Watchdog Timer Time-out Period <br> (no postscaler) | TBD | TBD | TBD | ms |  |
| 32 | TOST | Oscillation Start-up Timer Period | 1024 Tosc | - | 1024 Tosc | - | TosC = OSC1 period |
| 33 | TPWRT | Power-up Timer Period | 55.6 | 65.5 | 75 | ms |  |
| 34 | TIOZ | I/O High-Impedance from $\overline{\text { MCLR }}$ <br> Low or Watchdog Timer Reset | - | 2 | - | $\mu \mathrm{s}$ |  |
| 38 | TCSD | CPU Start-up Time | - | 200 | - | $\mu \mathrm{s}$ |  |
| 39 | TIOBST | Time for INTRC to Stabilize | - | TBD | - | $\mu \mathrm{s}$ |  |

Legend: TBD = To Be Determined

FIGURE 23-8: TIMERO AND TIMER1 EXTERNAL CLOCK TIMINGS


TABLE 23-11: TIMERO AND TIMER1 EXTERNAL ELQCK REQUIREMENTS

| Param No. | Symbol | Characteristic |  |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 40 | TтOH | TOCKI High Pulse Width $\lambda$ No prescaler |  |  | 0.5 TcY + 20 | - | ns |  |
|  |  |  | - | With prescaler | 10 | - | ns |  |
| 41 | TTOL | TOCKI Low Pulse Vidth |  | No prescaler | 0.5 TcY + 20 | - | ns |  |
|  |  |  |  | With prescaler | 10 | - | ns |  |
| 42 | TтOP |  |  | No prescaler | TCY + 10 | - | ns |  |
|  |  |  |  | With prescaler | $\begin{aligned} & \text { Greater of: } \\ & 20 \mathrm{~ns} \text { or } \\ & (\mathrm{Tcy}+40) / \mathrm{N} \end{aligned}$ | - | ns | $\begin{aligned} & \mathrm{N}=\text { prescale } \\ & \text { value } \\ & (1,2,4, \ldots, 256) \end{aligned}$ |
| 45 | TT1H | $\begin{aligned} & \text { Tycki fligh } \\ & \text { tinge } \end{aligned}$ | Synchronous, no prescaler |  | 0.5 TCY + 20 | - | ns |  |
|  |  |  | Synchronous, with prescaler |  | 10 | - | ns |  |
|  |  |  | Asynchronous |  | 30 | - | ns |  |
| 46 |  | T1CKI Low Time | Synchronous, no | o prescaler | 0.5 TCY + 5 | - | ns |  |
|  |  |  | Synchronous, w | with prescaler | 10 | - | ns |  |
|  |  |  | Asynchronous |  | 30 | - | ns |  |
| 47 | $\sqrt{7 T 1 P}$ | T1CKI Input Period | Synchronous |  | Greater of: 20 ns or $(\mathrm{TCY}+40) / \mathrm{N}$ | - | ns | $\begin{array}{\|l} \mathrm{N}=\text { prescale } \\ \text { value } \\ (1,2,4,8) \\ \hline \end{array}$ |
|  |  |  | Asynchronous |  | 60 | - | ns |  |
|  | FT1 | T1CKI Oscillator Input Frequency Range |  |  | DC | 50 | kHz |  |
| 48 | TCKE2TMRI | Delay from External T1CKI Clock Edge to Timer Increment |  |  | 2 Tosc | 7 Tosc | - |  |

## PIC18F45J10 FAMILY

FIGURE 23-9: CAPTURE/COMPARE/PWM TIMINGS (INCLUDING ECCP MODULE)


TABLE 23-12: CAPTURE/COMPARE/PWM REQUIREMENTS (INQLUOING ECCP MODULE)

| Param No. | Symbol | Characteristic |  |  | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 50 | TccL | CCPx Input Low Time | No prescaler | 0.5 Fcy + 20 | - | ns |  |
|  |  |  | With prescaler | 10 | - | ns |  |
| 51 | Tcch | CCPx Input High Time | No prescaler | 0.5 Tcy + 20 | - | ns |  |
|  |  |  | With prescaler $\gg$ | 10 | - | ns |  |
| 52 | TccP | CCPx Input Period |  | $\frac{3 T C Y+40}{N}$ | - | ns | $\begin{array}{\|l\|} \hline N=\text { prescale } \\ \text { value (1, } 4 \text { or } 16) \\ \hline \end{array}$ |
| 53 | TccR | CCPx Output Fall Time |  | - | 25 | ns |  |
| 54 | TccF | CCPx Output Fall Time |  | - | 25 | ns |  |

TABLE 23-13: PARALLELSLAVEPORT REQUIREMENTS

| $\begin{array}{\|c\|} \hline \text { Param. } \\ \text { No. } \end{array}$ | Symbol |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 62 | TdtV2wrH/ | PRta In Valid before $\overline{\mathrm{WR}} \uparrow$ or $\overline{\mathrm{CS}} \uparrow$ (setup time) | 20 | - | ns |  |
| 63 | TwrH2dtl | $\overline{\mathrm{WR}} \uparrow \uparrow$ or $\overline{\mathrm{CS}} \uparrow$ to Data-In Invalid (hold time) | 20 | - | ns |  |
| 64 | TrdL2dtK | $\stackrel{\rightharpoonup}{\text { PD }} \downarrow$ and $\overline{\mathrm{CS}} \downarrow$ to Data-Out Valid | - | 80 | ns |  |
| 65 | TrelH2dt | $\overline{\mathrm{RD}} \uparrow$ or $\overline{\mathrm{CS}} \downarrow$ to Data-Out Invalid | 10 | 30 | ns |  |
| $66$ | TinfinH | Inhibit of the IBF Flag bit being Cleared from $\overline{\mathrm{WR}} \uparrow$ or $\overline{\mathrm{CS}} \uparrow$ | - | 3 Tcy |  |  |

## PIC18F45J10 FAMILY

FIGURE 23-10: EXAMPLE SPITM MASTER MODE TIMING (CKE = 0)


TABLE 23-14: EXAMPLE SPITM MODE REQUHBEMENTS (MASTER MODE, CKE = 0)

| Param No. | Symbol | Characteristic |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 70 | $\begin{aligned} & \text { TssL2scH, } \\ & \text { TssL2scL } \end{aligned}$ | $\overline{\mathrm{SSx}} \downarrow$ to SCKx $\downarrow$ or SCKx $\uparrow$ inpyt |  | TCY | - | ns |  |
| 71 | Tsch | SCKx Input High Time (Slave mode) | Continuous | 1.25 TCY +30 | - | ns |  |
| 71A |  |  | Single Byte | 40 | - | ns | (Note 1) |
| 72 | TscL | SCKx Input Low Time (Slave Mode) | Continuous | 1.25 TCY + 30 | - | ns |  |
| 72A |  |  | Single Byte | 40 | - | ns | (Note 1) |
| 73 | ToIV2scH, TDIV2scL | Setup Timè of SDIx Data Input to SCKx Edge |  | 100 | - | ns |  |
| 73A | Tв2B | Last Clock Edge of Byte 1 to the 1st Clock Edge orbyte2 |  | 1.5 TCY + 40 | - | ns | (Note 2) |
| 74 | TscH2BLL Tscl2DIL | Held Time of SDIx Data Input to SCKx Edge |  | 100 | - | ns |  |
| 75 | TDOR | SDOx Data Output Rise Time |  | - | 25 | ns |  |
| 76 | TDQF | SDOx Data Output Fall Time |  | - | 25 | ns |  |
| 78 | Tsch | SCKx Output Rise Time (Master mode) |  | - | 25 | ns |  |
| 79 | TSCF | SCKx Output Fall Time (Master mode) |  | - | 25 | ns |  |
| 80 V | TscH2doV, TscL2doV | SDOx Data Output Valid after SCKx Edge |  | - | 50 | ns |  |

Note 1: Requires the use of Parameter \#73A.
2: Only if Parameter \#71A and \#72A are used.

## PIC18F45J10 FAMILY

FIGURE 23-11: EXAMPLE SPITM MASTER MODE TIMING (CKE = 1)
$\overline{s s x}$


Note: Refer to Figure 23-3 for load conditions.

TABLE 23-15: EXAMPLE SPITM MODE REQURREMENTS (MASTER MODE, CKE = 1)

| Param. No. | Symbol | Characteristic | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 71 | TscH | SCKx Input High Time continuous | 1.25 TCY + 30 | - | ns |  |
| 71A |  | (Slave mode) Single Byte | 40 | - | ns | (Note 1) |
| 72 | TscL | SCKx Input Lovy Time C Continuous | 1.25 TcY + 30 | - | ns |  |
| 72A |  | (Slave mode) Single Byte | 40 | - | ns | (Note 1) |
| 73 | ToIV2scH, TdIV2scL | Setup Time of SФix Data Input to SCKx Edge | 100 | - | ns |  |
| 73A | Тв2в | Læst Clock Edge of Byte 1 to the 1st Clock Edge of Bytec | 1.5 TCY + 40 | - | ns | (Note 2) |
| 74 | TscH2dIL, TscL2b爪 | Hota Tinme of SDIx Data Input to SCKx Edge | 100 | - | ns |  |
| 75 | TDLR | \$DOx Data Output Rise Time | - | 25 | ns |  |
| 76 | ToQ | SDOx Data Output Fall Time | - | 25 | ns |  |
| 78 < | TSCR | SCKx Output Rise Time (Master mode) | - | 25 | ns |  |
| 79 | TScF | SCKx Output Fall Time (Master mode) | - | 25 | ns |  |
|  | Tsch 2 DOV , TScL2doV | SDOx Data Output Valid after SCKx Edge | - | 50 | ns |  |
| 81 | TdoV2sch, TdoV2scL | SDOx Data Output Setup to SCKx Edge | TcY | - | ns |  |

Note 1: Requires the use of Parameter \#73A.
2: Only if Parameter \#71A and \#72A are used.

## PIC18F45J10 FAMILY

FIGURE 23-12: EXAMPLE SPITM SLAVE MODE TIMING (CKE = 0)


TABLE 23-16: EXAMPLE SPITM MODE REQUREMENTS (SLAVE MODE TIMING, CKE =0)

| Param No. | Symbol | Characteristic |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 70 | TssL2scH, TssL2scL | $\overline{\text { SSx }} \downarrow$ to SCKx $\downarrow$ or SCKX $\uparrow$ Input $\downarrow$ |  | TCY | - | ns |  |
| 71 | TscH | SCKx Input High Time (Slave mode) | Continuous | 1.25 TcY + 30 | - | ns |  |
| 71A |  |  | Single Byte | 40 | - | ns | (Note 1) |
| 72 | TscL | SCKx Input Low Time (Slave mode) | Continuous | 1.25 TcY + 30 | - | ns |  |
| 72A |  |  | Single Byte | 40 | - | ns | (Note 1) |
| 73 | TDIV2scH, TDIV2sCL | Setup tirse of SDIx Data Input to SCKx Edge |  | 100 | - | ns |  |
| 73A | Тв2в $\triangle$ | LasteloctyEdge of Byte 1 to the First Clock Edge of Byte 2 |  | 1.5 TCY + 40 | - | ns | (Note 2) |
| 74 | $\begin{aligned} & \text { TscH2DIL, } \\ & \text { TscL\& } 2 \mathrm{IIL} \end{aligned}$ | Hold Time of SDIx Data Input to SCKx Edge |  | 100 | - | ns |  |
| 75 | ToaR | SDOx Data Output Rise Time |  | - | 25 | ns |  |
| 76 | TROF) | SDOx Data Output Fall Time |  | - | 25 | ns |  |
| 77 | TSSHEDOZ | $\overline{\mathrm{SSx}} \uparrow$ to SDOx Output High-Impedance |  | 10 | 50 | ns |  |
| 78 | Tsch | SCKx Output Rise Time (Master mode) |  | - | 25 | ns |  |
| 79 | TscF | SCKx Output Fall Time (Master mode) |  | - | 25 | ns |  |
| 80 | TscH2doV, TscL2doV | SDOx Data Output Valid after SCKx Edge |  | - | 50 | ns |  |
| 83 | TscH2ssH, TscL2ssH | $\overline{\text { SSx }} \uparrow$ after SCKx Edge |  | 1.5 TCY + 40 | - | ns |  |

Note 1: Requires the use of Parameter \#73A.
2: Only if Parameter \#71A and \#72A are used.

## PIC18F45J10 FAMILY

FIGURE 23-13: EXAMPLE SPITM SLAVE MODE TIMING (CKE = 1)


TABLE 23-17: EXAMPLE SPITM SLAVE MODEREQUIREMENTS (CKE = 1)

| Param <br> No. | Symbol | Characteristic |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Note 1: Requires the use of Parameter \#73A.
2: Only if Parameter \#71A and \#72A are used.

## PIC18F45J10 FAMILY

FIGURE 23-14: $\quad I^{2} C^{\text {TM }}$ BUS START/STOP BITS TIMING


TABLE 23-18: $I^{2} C^{\text {TM }}$ BUS START/STOP BITS REQUIREMENTSISLAVE MODE)


FIGURE 23-15: $\quad I^{2} C^{\text {TM }}$ BUS DATA TIMING


TABLE 23-19: $\mathrm{I}^{2} \mathrm{C}^{\text {TM }}$ BUS DATA REQUIREMENTS (SLAVE MODE)

| Param. <br> No. | Symbol | Characteristic |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 100 | THIGH | Clock High Time | 100 kHz mode | 4.0 | - | $\mu \mathrm{s}$ | PIC18F45J10 must operate at a minimum of 1.5 MH |
|  |  |  | 400 kHz mode | 0.6 | - | $\mu \mathrm{s}$ | PIC18F45J10 must operate at a minimum of 10 MHz |
|  |  |  | MSSP Module | 1.5 TCY | - |  | $\rightarrow>$ |
| 101 | TLOW | Clock Low Time | 100 kHz mode | 4.7 | - | $\mu \mathrm{s}$ | PIC18F45 0 must pperate at a minimum or 1.5 MHz |
|  |  |  | 400 kHz mode | 1.3 | - | $\mu \mathrm{s}$ | PIC 18 F 45 S 10 must operate at a minimumbo 10 MHz |
|  |  |  | MSSP Module | 1.5 TCY | - |  | - |
| 102 | TR | SDAx and SCLx Rise Time | 100 kHz mode | - | 1000 |  | , |
|  |  |  | 400 kHz mode | $20+0.1$ CB |  |  | CB is specified to be from 10 to 400 pF |
| 103 | TF | SDAx and SCLx Fall Time | 100 kHz mode | - | 300 | ns |  |
|  |  |  | 400 kHz mode | $20+0.1 \mathrm{CB}$ | $2^{300}$ | $\lambda^{0 \mathrm{~S}}$ | CB is specified to be from 10 to 400 pF |
| 90 | TSU:STA | Start Condition Setup Time | 100 kHz mode | 4.7 | $\square$ | $\mu \mathrm{s}$ | Only relevant for Repeated |
|  |  |  | 400 kHz mode | 0.6 | $\Sigma$ | $\mu \mathrm{s}$ | Start condition |
| 91 | THD:STA | Start Condition Hold Time | 100 kHz mode | 4.0 | - | $\mu \mathrm{s}$ | After this period, the first clock pulse is generated |
|  |  |  | 400 kHz mode | Q6 | - | $\mu \mathrm{s}$ |  |
| 106 | THD:DAT | Data Input Hold Time | 100 kHz mode | ) 0 | - | ns |  |
|  |  |  | 400 kHz mode | $>0$ | 0.9 | $\mu \mathrm{s}$ |  |
| 107 | TSU:DAT | Data Input Setup Time | 100 kHz mode | 250 | - | ns | (Note 2) |
|  |  |  | 400 kAz mode | 100 | - | ns |  |
| 92 | Tsu:sto | Stop Condition Setup Time | 100 kHz mpde | 4.7 | - | $\mu \mathrm{s}$ |  |
|  |  |  | 400 kHz mode | 0.6 | - | $\mu \mathrm{s}$ |  |
| 109 | TAA | Output Valid from Clock 100 kHz mode |  | - | 3500 | ns | (Note 1) |
|  |  |  |  | - | - | ns |  |
| 110 | TBuF | Bus Free Time | 100 kHz mode | 4.7 | - | $\mu \mathrm{s}$ | Time the bus must be free before a new transmission can start |
|  |  |  | 400 kHz mode | 1.3 | - | $\mu \mathrm{s}$ |  |
| D102 | Cв | Bus Capacitive Loading |  | - | 400 | pF |  |

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns ) of the falling edge ofsCLx to avoid unintended generation of Start or Stop conditions.
2: A Fast mqde $1^{2}$ CTM bus device can be used in a Standard mode $I^{2} \mathrm{C}$ bus system, but the requirement, TSU:DAT $\geq 250 \mathrm{~ns}$, must then pe met. This will automatically be the case if the device does not stretch the LOW period of the SCLx signal. If such adewice does stretch the LOW period of the SCLx signal, it must output the next data bit to the SDAx line, Trmax. + TsU:DAT $=1000+250=1250 \mathrm{~ns}$ (according to the Standard mode $\mathrm{I}^{2} \mathrm{C}$ bus specification), before the SCLx line is yeleased.

## PIC18F45J10 FAMILY

FIGURE 23-16: MASTER SSP ${ }^{2}{ }^{2} \mathbf{C}^{\text {TM }}$ BUS START/STOP BITS TIMING WAVEFORMS


TABLE 23-20: MASTER SSP $I^{2} C^{\text {TM }}$ BUS START/STOP BITS REQUIREMENTS

| Param. No. | Symbol | Characteristic |  |  | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 90 | Tsu:STA | Start Condition Setup Time | 100 kHz mode | $2(T o s c)(B R C)+1)$ | - | ns | Only relevant for Repeated Start condition |
|  |  |  | 400 kHz mode | 2(tosc)(BRG + 1) | - |  |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | $2(T \sigma s c)($ BRG + 1) | - |  |  |
| 91 | THD:STA | Start Condition Hold Time | 100 kHz mode | 2(TOsC)(BRG + 1) | - | ns | After this period, the first clock pulse is generated |
|  |  |  | 400 kHz mode | $2($ (\%SC) (BRG + 1) | - |  |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | $2(\mathrm{Tosc})(\mathrm{BRG}+1)$ | - |  |  |
| 92 | Tsu:sto | Stop Condition Setup Time | 100 kHz mode | 2(Tosc)(BRG + 1) | - | ns |  |
|  |  |  | 400 kitzmode | 2(Tosc)(BRG + 1) | - |  |  |
|  |  |  | $141+2 z$ phode ${ }^{(1)}$ | 2(Tosc)(BRG + 1) | - |  |  |
| 93 | THD:Sto | Stop ConditionHold Time400 kHz mode400 MHz mode ${ }^{(\mathbf{1 )}}$ |  | 2(Tosc)(BRG + 1) | - | ns |  |
|  |  |  |  | 2(Tosc)(BRG + 1) | - |  |  |
|  |  |  |  | 2(Tosc)(BRG + 1) | - |  |  |

Note 1: Maximum pin capacifance $=10 \mathrm{pF}$ for all $\mathrm{I}^{2} \mathrm{C}^{\text {TM }}$ pins.

FIGURE 23-17:
MASTER SSP $I^{2} C^{\text {TM }}$ BUS DATA TIMING


Note: Refer to Figure 23-3 for load conditions.

## PIC18F45J10 FAMILY

TABLE 23-21: MASTER SSP I ${ }^{2} C^{\text {TM }}$ BUS DATA REQUIREMENTS

| Param. No. | Symbol | Characteristic |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 100 | THIGH | Clock High Time | 100 kHz mode | 2(Tosc)(BRG + 1) | - | ms |  |
|  |  |  | 400 kHz mode | 2(Tosc)(BRG + 1) | - | ms |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | 2(Tosc)(BRG + 1) | - | ms |  |
| 101 | TLow | Clock Low Time | 100 kHz mode | 2(Tosc)(BRG + 1) | - | ms |  |
|  |  |  | 400 kHz mode | 2(Tosc)(BRG + 1) | - | ms | ) |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | 2(Tosc)(BRG + 1) | - | ms |  |
| 102 | TR | SDAx and SCLx Rise Time | 100 kHz mode | - | 1000 |  | CB is specified to be from te to 400 pF |
|  |  |  | 400 kHz mode | $20+0.1 \mathrm{CB}$ | 300 |  |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | - | 300 | ns |  |
| 103 | TF | SDAx and SCLx Fall Time | 100 kHz mode | - | 300 | $n 8$ | $C_{B}$ is specified to be from 10 to 400 pF |
|  |  |  | 400 kHz mode | $20+0.1$ Cв | 309 | ns) |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | - | 10 | Hs |  |
| 90 | TSU:STA | Start Condition Setup Time | 100 kHz mode | 2(Tosc)(BRG +1$)$ | $\rightarrow$ | ms | Only relevant for Repeated Start condition |
|  |  |  | 400 kHz mode | 2(Tosc)(BRG +1 ) | 7 | ms |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | 2(Tosc)(BRG + 1) | - | ms |  |
| 91 | THD:STA | Start Condition Hold Time | 100 kHz mode | 2(Tos¢) (BRG +1 1) | - | ms | After this period, the first clock pulse is generated |
|  |  |  | 400 kHz mode | 2(XQsc)(BRG+1) | - | ms |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | $2(\operatorname{tas})(\mathrm{BRG}+1)$ | - | ms |  |
| 106 | THD:DAT | Data Input Hold Time | 100 kHz mode | $\gg$ | - | ns |  |
|  |  |  | 400 kHz mode | $\Rightarrow 0$ | 0.9 | ms |  |
|  |  |  | 1 MHz mpde $(1)$ | TBD | - | ns |  |
| 107 | TSu:dat | Data Input Setup Time | 100 kHz mode $>$ | 250 | - | ns | (Note 2) |
|  |  |  | 400 kklz mode | 100 | - | ns |  |
|  |  |  | MM1 +2 mode ${ }^{(1)}$ | TBD | - | ns |  |
| 92 | Tsu:Sto | Stop Condition Setup Time | 100 kHz mode | 2(Tosc)(BRG + 1) | - | ms |  |
|  |  |  | 400 kHz mode | 2(Tosc)(BRG + 1) | - | ms |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | 2(Tosc)(BRG + 1) | - | ms |  |
| 109 | TAA | Outpux Latid from Clock $\nabla$ | 100 kHz mode | - | 3500 | ns |  |
|  |  |  | 400 kHz mode | - | 1000 | ns |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | - | - | ns |  |
| 110 |  | Bus Free Time | 100 kHz mode | 4.7 | - | ms | Time the bus must be free before a new transmission can start |
|  |  |  | 400 kHz mode | 1.3 | - | ms |  |
|  |  |  | 1 MHz mode ${ }^{(1)}$ | TBD | - | ms |  |
|  |  | Bus Capacitive Loading |  | - | 400 | pF |  |

Kegend: TBD = To Be Determined
Note 1: Maximum pin capacitance $=10 \mathrm{pF}$ for all $\mathrm{I}^{2} \mathrm{C}^{\mathrm{TM}}$ pins.
A Fast mode $I^{2} \mathrm{C}$ bus device can be used in a Standard mode $\mathrm{I}^{2} \mathrm{C}$ bus system, but parameter \#107 $\geq 250 \mathrm{~ns}$ must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCLx signal. If such a device does stretch the LOW period of the SCLx signal, it must output the next data bit to the SDAx line, parameter \#102 + parameter \#107 = $1000+250=1250 \mathrm{~ns}$ (for 100 kHz mode), before the SCLx line is released.

FIGURE 23-18: EUSART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING


TABLE 23-22: EUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

| Param <br> No. | Symbol | Characteristic |  | Units | Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 120 | TCKH2DTV | SYNC XMIT (MASTER and SLAVE) |  |  |  |
| Clock High to Data Out Valid |  |  |  |  |  |

FIGURE 23-19: EUSART SYNCHRONOUS RECEWE (MASTER/SLAVE) TIMING


TABLE 23-23: EUSARTSYNCHRONOUS RECEIVE REQUIREMENTS


## PIC18F45J10 FAMILY

TABLE 23-24: A/D CONVERTER CHARACTERISTICS: PIC18F24J10/25J10/44J10/45J10 (INDUSTRIAL)

| Param <br> No. | Symbol | Characteristic | Min | Typ | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A01 | NR | Resolution | - | - | 10 | bit | $\Delta \mathrm{VREF} \geq 3.0 \mathrm{~K}$ |
| A03 | EIL | Integral Linearity Error | - | - | < $\pm 1$ | LSb | $\triangle \mathrm{VREF} \geq 3.0 \mathrm{~V}$ |
| A04 | EdL | Differential Linearity Error | - | - | $< \pm 1$ | LSb | $\Delta \mathrm{VREF} \geq 3.0 \mathrm{X}$ |
| A06 | EOFF | Offset Error | - | - | < $\pm 1.5$ | LSb | $\Delta V_{R} E F \geq 3.0 \mathrm{~V}$ |
| A07 | EGN | Gain Error | - | - | < $\pm 1$ | LSb | $\triangle V R E F \geq 3.00$ |
| A10 | - | Monotonicity | Guaranteed ${ }^{(1)}$ |  |  |  | VSs $\leq$ Vain $\leq$ Vref |
| A20 | $\Delta$ VREF | Reference Voltage Range (VRefh - Vrefl) | $\begin{gathered} 1.8 \\ 3 \end{gathered}$ | - | - | $\langle v$ | $\begin{aligned} & Y D D<3.0 \mathrm{~V} \\ & Y D D \geq 3.0 \mathrm{~V} \end{aligned}$ |
| A21 | Vrefe | Reference Voltage High | Vss | - | VREFH | $\xrightarrow{N}$ |  |
| A22 | VREFL | Reference Voltage Low | Vss -0.3V | - | $\operatorname{VDD}+3$ |  |  |
| A25 | VAIN | Analog Input Voltage | Vrefl | - | VREFH |  |  |
| A30 | ZAIN | Recommended Impedance of Analog Voltage Source | - |  | $22^{2.5}$ | $\mathrm{k} \Omega$ |  |
| A50 | IREF | VREF Input Current ${ }^{(2)}$ | - |  | $\sqrt{5}$ | $\mu \mathrm{A}$ $\mu \mathrm{A}$ | During VAIN acquisition. During A/D conversion cycle. |

Note 1: The A/D conversion result never decreases with ay increase in the input voltage and has no missing codes.
2: VREFH current is from RA3/AN3/VREF+ pin or $\langle$ DQ, Shichever is selected as the VREFH source.
VREFL current is from RA2/AN2/VREF- pin or VsS, whichever is selected as the VREFL source.

FIGURE 23-20: A/D CONVERSION TIMNNG


TABLE 23-25: A/D CONVERSION REQUIREMENTS

| Param No. | Symbol | Characteristic | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 130 | TAD | A/D Clock Period | 0.7 | $25.0^{(1)}$ | $\mu \mathrm{s}$ | Tosc based, VREF $\geq 3.0 \mathrm{~V}$ |
|  |  |  | TBD | 1 | $\mu \mathrm{s}$ | A/D RC mode |
| 131 | TcNv | Conversion Time (not including acquisition time) (Note 2) | 11 | 12 | TAD | $\bar{S}$ |
| 132 | TACQ | Acquisition Time (Note 3) | $\begin{gathered} 1.4 \\ \text { TBD } \end{gathered}$ | - | $\begin{aligned} & \mu \mathrm{s} \\ & \mu \mathrm{~s} \end{aligned}$ | $\begin{aligned} & -40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} \\ & 0^{\circ} \mathrm{C} \text { sto }=+85^{\circ} \mathrm{C} \end{aligned}$ |
| 135 | Tswc | Switching Time from Convert $\rightarrow$ Sample | - | (Note 4) |  | $\rightarrow>$ |

Legend: TBD = To Be Determined
Note 1: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.
2: ADRES registers may be read on the following Tcy cycle.
3: The time for the holding capacitor to acquire the "New" input voltage when the roltage changes full scale after the conversion (VDD to Vss or Vss to VDD). The source impedance(RS) on the input channels is $50 \Omega$.
4: On the following cycle of the device clock.

## PIC18F45J10 FAMILY

NOTES:

### 24.0 DC AND AC <br> CHARACTERISTICS GRAPHS <br> AND TABLES

Graphs and tables are not available at this time.

## PIC18F45J10 FAMILY

NOTES:

### 25.0 PACKAGING INFORMATION

### 25.1 Package Marking Information

28-Lead SPDIP


## 28-Lead SOIC



28-Lead SSOP


## 28-Lead QFN



Example


## Example



## Example



## Example



```
Legend: XX...X Customer-specific information
Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')
NNN Alphanumeric traceability code
(e3) Pb-free JEDEC designator for Matte Tin (Sn)
* This package is Pb-free. The Pb-free JEDEC designator (e3)
    can be found on the outer packaging for this package.
Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.
```


## PIC18F45J10 FAMILY

## Package Marking Information (Continued)

40-Lead PDIP


44-Lead QFN


44-Lead TQFP


## Example



Example


Example


### 25.2 Package Details

The following sections give the technical details of the packages.

## 28-Lead Skinny Plastic Dual In-line (SP) - 300 mil Body (PDIP)



| Units |  | INCHES* |  |  | MILLIMETERS |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dimension Limits |  | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Pins | n | 28 |  |  | 28 |  |  |
| Pitch | p |  | . 100 |  |  | 2.54 |  |
| Top to Seating Plane | A | . 140 | . 150 | . 160 | 3.56 | 3.81 | 4.06 |
| Molded Package Thickness | A2 | . 125 | . 130 | . 135 | 3.18 | 3.30 | 3.43 |
| Base to Seating Plane | A1 | . 015 |  |  | 0.38 |  |  |
| Shoulder to Shoulder Width | E | . 300 | . 310 | . 325 | 7.62 | 7.87 | 8.26 |
| Molded Package Width | E1 | . 275 | . 285 | . 295 | 6.99 | 7.24 | 7.49 |
| Overall Length | D | 1.345 | 1.365 | 1.385 | 34.16 | 34.67 | 35.18 |
| Tip to Seating Plane | L | . 125 | . 130 | . 135 | 3.18 | 3.30 | 3.43 |
| Lead Thickness | c | . 008 | . 012 | . 015 | 0.20 | 0.29 | 0.38 |
| Upper Lead Width | B1 | . 040 | . 053 | . 065 | 1.02 | 1.33 | 1.65 |
| Lower Lead Width | B | . 016 | . 019 | . 022 | 0.41 | 0.48 | 0.56 |
| Overall Row Spacing § | eB | . 320 | . 350 | . 430 | 8.13 | 8.89 | 10.92 |
| Mold Draft Angle Top | $\alpha$ | 5 | 10 | 15 | 5 | 10 | 15 |
| Mold Draft Angle Bottom | $\beta$ | 5 | 10 | 15 | 5 | 10 | 15 |

* Controlling Parameter
§ Significant Characteristic
Notes:
Dimension D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed
$010^{\prime \prime}(0.254 \mathrm{~mm})$ per side.
JEDEC Equivalent: MO-095
Drawing No. C04-070


## PIC18F45J10 FAMILY

## 28-Lead Plastic Small Outline (SO) - Wide, 300 mil Body (SOIC)



| Units |  | INCHES* |  |  | MILLIMETERS |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dim | imits | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Pins | n | 28 |  |  | 28 |  |  |
| Pitch | p |  | . 050 |  |  | 1.27 |  |
| Overall Height | A | . 093 | . 099 | . 104 | 2.36 | 2.50 | 2.64 |
| Molded Package Thickness | A2 | . 088 | . 091 | . 094 | 2.24 | 2.31 | 2.39 |
| Standoff § | A1 | . 004 | . 008 | . 012 | 0.10 | 0.20 | 0.30 |
| Overall Width | E | . 394 | . 407 | . 420 | 10.01 | 10.34 | 10.67 |
| Molded Package Width | E1 | . 288 | . 295 | . 299 | 7.32 | 7.49 | 7.59 |
| Overall Length | D | . 695 | . 704 | . 712 | 17.65 | 17.87 | 18.08 |
| Chamfer Distance | h | . 010 | . 020 | . 029 | 0.25 | 0.50 | 0.74 |
| Foot Length | L | . 016 | . 033 | . 050 | 0.41 | 0.84 | 1.27 |
| Foot Angle Top | $\phi$ | 0 | 4 | 8 | 0 | 4 | 8 |
| Lead Thickness | c | . 009 | . 011 | . 013 | 0.23 | 0.28 | 0.33 |
| Lead Width | B | . 014 | . 017 | . 020 | 0.36 | 0.42 | 0.51 |
| Mold Draft Angle Top | $\alpha$ | 0 | 12 | 15 | 0 | 12 | 15 |
| Mold Draft Angle Bottom | $\beta$ | 0 | 12 | 15 | 0 | 12 | 15 |

* Controlling Parameter
§ Significant Characteristic


## Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed
$010 "$ ( 0.254 mm ) per side.
JEDEC Equivalent: MS-013
Drawing No. C04-052

## PIC18F45J10 FAMILY

## 28-Lead Plastic Shrink Small Outline (SS) - 209 mil Body, 5.30 mm (SSOP)



| Units |  | INCHES |  |  | MILLIMETERS* |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dime |  | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Pins | n | 28 |  |  | 28 |  |  |
| Pitch | p |  | . 026 |  |  | 0.65 |  |
| Overall Height | A | - | - | . 079 | - | - | 2.0 |
| Molded Package Thickness | A2 | . 065 | . 069 | . 073 | 1.65 | 1.75 | 1.85 |
| Standoff | A1 | . 002 | - | - | 0.05 | - | - |
| Overall Width | E | . 295 | . 307 | . 323 | 7.49 | 7.80 | 8.20 |
| Molded Package Width | E1 | . 009 | . 209 | . 220 | 5.00 | 5.30 | 5.60 |
| Overall Length | D | . 390 | . 402 | . 413 | 9.90 | 10.20 | 10.50 |
| Foot Length | L | . 022 | . 030 | . 037 | 0.55 | 0.75 | 0.95 |
| Lead Thickness | c | . 004 | - | . 010 | 0.09 | - | 0.25 |
| Foot Angle | f | $0^{\circ}$ | $4^{\circ}$ | $8^{\circ}$ | $0^{\circ}$ | $4^{\circ}$ | $8^{\circ}$ |
| Lead Width | B | . 009 | - | . 015 | 0.22 | - | 0.38 |

*Controlling Parameter
Notes:
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010 " ( 0.254 mm ) per side.

JEDEC Equivalent: MO-150
Drawing No. C04-073

## PIC18F45J10 FAMILY

## 28-Lead Plastic Quad Flat No Lead Package (ML) $6 x 6$ mm Body (QFN) With 0.55 mm Contact Length (Saw Singulated)



| Units |  | INCHES |  |  | MILLIMETERS* |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dimension Limits |  | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Pins | n | 28 |  |  | 28 |  |  |
| Pitch | e | . 026 BSC |  |  | 0.65 BSC |  |  |
| Overall Height | A | . 031 | . 035 | . 039 | 0.80 | 0.90 | 1.00 |
| Standoff | A1 | . 000 | . 001 | . 002 | 0.00 | 0.02 | 0.05 |
| Contact Thickness | A3 | . 008 REF |  |  | 0.20 REF |  |  |
| Overall Width | E | . 232 | . 236 | . 240 | 5.90 | 6.00 | 6.10 |
| Exposed Pad Width | E2 | . 140 | . 146 | . 152 | 3.55 | 3.70 | 3.85 |
| Overall Length | D | . 232 | . 236 | . 240 | 5.90 | 6.00 | 6.10 |
| Exposed Pad Length | D2 | . 140 | . 146 | . 152 | 3.55 | 3.70 | 3.85 |
| Contact Width | b | . 009 | . 011 | . 013 | 0.23 | 0.28 | 0.33 |
| Contact Length | L | . 018 | . 022 | . 024 | 0.45 | 0.55 | 0.65 |

*Controlling Parameter
Notes:
JEDEC equivalent: MO-220
Drawing No. C04-105

## 40-Lead Plastic Dual In-line (P) - $\mathbf{6 0 0}$ mil Body (PDIP)



| Units |  | INCHES* |  |  | MILLIMETERS |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dimension Limits |  | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Pins | n | 40 |  |  | 40 |  |  |
| Pitch | p |  | . 100 |  |  | 2.54 |  |
| Top to Seating Plane | A | . 160 | . 175 | . 190 | 4.06 | 4.45 | 4.83 |
| Molded Package Thickness | A2 | . 140 | . 150 | . 160 | 3.56 | 3.81 | 4.06 |
| Base to Seating Plane | A1 | . 015 |  |  | 0.38 |  |  |
| Shoulder to Shoulder Width | E | . 595 | . 600 | . 625 | 15.11 | 15.24 | 15.88 |
| Molded Package Width | E1 | . 530 | . 545 | . 560 | 13.46 | 13.84 | 14.22 |
| Overall Length | D | 2.045 | 2.058 | 2.065 | 51.94 | 52.26 | 52.45 |
| Tip to Seating Plane | L | . 120 | . 130 | . 135 | 3.05 | 3.30 | 3.43 |
| Lead Thickness | c | . 008 | . 012 | . 015 | 0.20 | 0.29 | 0.38 |
| Upper Lead Width | B1 | . 030 | . 050 | . 070 | 0.76 | 1.27 | 1.78 |
| Lower Lead Width | B | . 014 | . 018 | . 022 | 0.36 | 0.46 | 0.56 |
| Overall Row Spacing § | eB | . 620 | . 650 | . 680 | 15.75 | 16.51 | 17.27 |
| Mold Draft Angle Top | $\alpha$ | 5 | 10 | 15 | 5 | 10 | 15 |
| Mold Draft Angle Bottom | $\beta$ | 5 | 10 | 15 | 5 | 10 | 15 |

* Controlling Parameter
§ Significant Characteristic
Notes:
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed $.010^{\prime \prime}$ ( 0.254 mm ) per side.
JEDEC Equivalent: MO-011
Drawing No. C04-016


## PIC18F45J10 FAMILY

## 44-Lead Plastic Quad Flat No Lead Package (ML) 8x8 mm Body (QFN)



| Units |  | INCHES |  |  | MILLIMETERS* |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dimension Limits |  | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Contacts | n | 44 |  |  | 44 |  |  |
| Pitch | p | . 026 BSC |  |  | 0.65 BSC |  |  |
| Overall Height | A | . 031 | . 035 | . 039 | 0.80 | 0.90 | 1.00 |
| Standoff | A1 | . 000 | . 001 | . 002 | 0 | 0.02 | 0.05 |
| Base Thickness | (A3) | . 010 REF |  |  | 0.25 REF |  |  |
| Overall Width | E | . 309 | . 315 | . 321 | 7.85 | 8.00 | 8.15 |
| Exposed Pad Width | E2 | . 246 | . 268 | . 274 | 6.25 | 6.80 | 6.95 |
| Overall Length | D | . 309 | . 315 | . 321 | 7.85 | 8.00 | 8.15 |
| Exposed Pad Length | D2 | . 246 | . 268 | . 274 | 6.25 | 6.80 | 6.95 |
| Contact Width | B | . 008 | . 013 | . 013 | 0.20 | 0.33 | 0.35 |
| Contact Length | L | . 014 | . 016 | . 019 | 0.35 | 0.40 | 0.48 |

*Controlling Parameter
Notes:

1. BSC: Basic Dimension. Theoretically exact value shown without tolerances. See ASME Y14.5M
2. REF: Reference Dimension, usually without tolerance, for information purposes only. See ASME Y14.5M
3. Contact profiles may vary.

JEDEC equivalent: M0-220
Drawing No. C04-103


| Units |  | INCHES |  |  | MILLIMETERS* |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dimension Limits |  | MIN | NOM | MAX | MIN | NOM | MAX |
| Number of Pins | n |  | 44 |  |  | 44 |  |
| Pitch | p |  | . 031 |  |  | 0.80 |  |
| Pins per Side | n1 |  | 11 |  |  | 11 |  |
| Overall Height | A | . 039 | . 043 | . 047 | 1.00 | 1.10 | 1.20 |
| Molded Package Thickness | A2 | . 037 | . 039 | . 041 | 0.95 | 1.00 | 1.05 |
| Standoff § | A1 | . 002 | . 004 | . 006 | 0.05 | 0.10 | 0.15 |
| Foot Length | L | . 018 | . 024 | . 030 | 0.45 | 0.60 | 0.75 |
| Footprint (Reference) | (F) |  | . 039 |  | 1.00 |  |  |
| Foot Angle | $\phi$ | 0 | 3.5 | 7 | 0 | 3.5 | 7 |
| Overall Width | E | . 463 | . 472 | . 482 | 11.75 | 12.00 | 12.25 |
| Overall Length | D | . 463 | . 472 | . 482 | 11.75 | 12.00 | 12.25 |
| Molded Package Width | E1 | . 390 | . 394 | . 398 | 9.90 | 10.00 | 10.10 |
| Molded Package Length | D1 | . 390 | . 394 | . 398 | 9.90 | 10.00 | 10.10 |
| Lead Thickness | c | . 004 | . 006 | . 008 | 0.09 | 0.15 | 0.20 |
| Lead Width | B | . 012 | . 015 | . 017 | 0.30 | 0.38 | 0.44 |
| Pin 1 Corner Chamfer | CH | . 025 | . 035 | . 045 | 0.64 | 0.89 | 1.14 |
| Mold Draft Angle Top | $\alpha$ | 5 | 10 | 15 | 5 | 10 | 15 |
| Mold Draft Angle Bottom | $\beta$ | 5 | 10 | 15 | 5 | 10 | 15 |

* Controlling Parameter
§ Significant Characteristic
Notes:
Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed
.010" ( 0.254 mm ) per side.
JEDEC Equivalent: MS-026
Drawing No. C04-076


## PIC18F45J10 FAMILY

NOTES:

## APPENDIX A: REVISION HISTORY

Revision A (March 2005)
Original data sheet for PIC18F45J10 family devices.

## APPENDIX B: MIGRATION <br> BETWEEN HIGH-END DEVICE FAMILIES

Devices in the PIC18F45J10 family and PIC18F4520 families are very similar in their functions and feature sets. However, there are some potentially important differences which should be considered when migrating an application across device families to achieve a new design goal. These are summarized in Table B-1. The areas of difference which could be a major impact on migration are discussed in greater detail later in this section.

## TABLE B-1: NOTABLE DIFFERENCES BETWEEN PIC18F4520 AND PIC18FXXXX FAMILIES

| Characteristic | PIC18FXXXX Family | PIC18F4520 Family |
| :---: | :---: | :---: |
| Operating Frequency | 40 MHz @ 2.15 V | 40 MHz @ 4.2V |
| Supply Voltage | 2.0V-3.6V | 2.0V-5.5V |
| Operating Current | Low | Lower |
| Program Memory Endurance | 1,000 write/erase cycles (typical) | 100,000 write/erase cycles (typical) |
| I/O Sink/Source at 25 mA | PORTB and PORTC only | All ports |
| Input Voltage Tolerance on I/O pins | 5.5 V on digital only pins | VdD on all I/O pins |
| I/O | 32 | 36 |
| Pull-ups | PORTB | PORTB |
| Oscillator Options | Limited options (EC, HS, fixed 32 kHz INTRC) | More options (EC, HS, XT, LP, RC, PLL, flexible INTRC) |
| Program Memory Retention | 10 years (minimum) | 40 years (minimum) |
| Programming Time (Normalized) | 156 \%s/byte (10 ms/64-byte block) | 15.6 ¢ $/$ byte ( $1 \mathrm{~ms} / 64$-byte block) |
| Programming Entry | Low Voltage, Key Sequence | VPP and LVP |
| Code Protection | Single block, all or nothing | Multiple code protection blocks |
| Configuration Words | Stored in last 4 words of Program Memory space | Stored in Configuration Space, starting at 300000h |
| Start-up Time from Sleep | $200 \mu$ s (typical) | $10 \mu$ (typical) |
| Power-up Timer | Always on | Configurable |
| Data EEPROM | Not available | Available |
| BOR | Simple BOR ${ }^{(1)}$ | Programmable BOR |
| LVD | Not available | Available |
| A/D Calibration | Required | Not required |
| In-Circuit Emulation | Not available | Available |
| TMR3 | Not available | Available |
| Second MSSP | Available ${ }^{(2)}$ | Not available |

Note 1: BOR is not available on PIC18LFXXJ10 devices.
2: Available on 40/44-pin devices only.

## B. 1 Power Requirement Differences

The most significant difference between the PIC18F45J10 family and PIC18F4520 device families is the power requirements. PIC18F45J10 family devices are designed on a smaller process; this results in lower maximum voltage and higher leakage current.
The operating voltage range for PIC18F45J10 family devices is 2.0 V to 3.6 V . One of the VDD pins is separated for the core logic supply (VddCore). This pin has specific voltage and capacitor requirements as described in Section 23.0 "Electrical Characteristics".
The current specifications for PIC18F45J10 family devices are yet to be determined.

## B. 2 Pin Differences

There are several differences in the pinouts between the PIC18F45J10 family and the PIC18F4520 families:

- Input voltage tolerance
- Output current capabilities
- Available I/O

Pins on the PIC18F45J10 family that have digital only input capability will tolerate voltages up to 5.5 V and are thus tolerant to voltages above VDD. Table 9-1 in Section 9.0 "I/O Ports" contains the complete list.
In addition to input differences, there are output differences as well. Not all I/O pins can source or sink equal levels of current. Only PORTB and PORTC support the 25 mA source/sink capability that is supported by all output pins on the PIC18F4520. Table 9-2 in Section 9.0 "I/O Ports" contains the complete list of output capabilities.
There are additional differences in how some pin functions are implemented on PIC18F45J10 family devices. First, the OSC1/OSC2 oscillator pins are strictly dedicated to the external oscillator function; there is no option to re-allocate these pins to I/O (RA6 or RA7) as on PIC18F4520 devices. Second, the $\overline{\mathrm{MCLR}}$ pin is dedicated only to $\overline{\mathrm{MCLR}}$ and cannot be configured as an input (RE3). Finally, RA4 does not exist on PIC18F45J10 family devices.
All of these pin differences (including power pin differences) should be accounted for when making a conversion between PIC18F4520 and PIC18F45J10 family devices.

## B. 3 Oscillator Differences

PIC18F4520 family devices have a greater range of oscillator options than PIC18F45J10 family devices. The latter family is limited primarily to operating modes that support HS and EC oscillators.

In addition, the PIC18F45J10 family has an internal RC oscillator with only a fixed 32 kHz output. The higher frequency RC modes of the PIC18F4520 family are not available.

## B. 4 Peripherals

The PIC18F45J10 family is able to operate at 40 MHz down to 2.15 volts unlike the PIC18F4520 family where 40 MHz operation is limited to $4.2+\mathrm{V}$ applications.

Peripherals must also be considered when making a conversion between the PIC18F45J10 family and the PIC18F4520 families:

- Data EEPROM: PIC18F45J10 family devices do not have this module.
- BOR: PIC18F45J10 family devices do not have a programmable BOR. Simple brown-out capability is provided through the use of the internal voltage regulator (not available in PIC18LFXXJ10 devices).
- LVD: PIC18F45J10 family devices do not have this module.
- Timer3 (TMR3) has been removed from the PIC18F45J10 family.
- The TOCKI/C1OUT pins have been moved from RA4 to RB5.
- The 40/44-pin devices in the PIC18F45J10 family have a second MSSP module available on pins RD0:RD3


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[^0]:    Microchip received ISO/TS-16949:2002 quality system certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona and Mountain View, California in October 2003. The Company's quality system processes and procedures are for its PICmicro ${ }^{\circledR}$-bit MCUs, KEELOQ ${ }^{\circledR}$ code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.

[^1]:    Legend: - = unimplemented, read as ' 0 ’. Shaded cells are not used by the Parallel Slave Port.

[^2]:    Note: Upon Reset, Timer0 is enabled in 8-bit mode with clock input from TOCKI max. prescale.

