



45MHz, 250V/µs Operational Amplifier

FEATURES

Gain-Bandwidth: 45MHz

■ Unity-Gain Stable■ Slew Rate: 250V/µs

■ C-LoadTM Op Amp Drives Capacitive Loads

Maximum Input Offset Voltage: 1mV

Maximum Input Bias Current: 300nA

Maximum Input Offset Current: 300nA

■ Minimum Output Swing Into 500Ω: ±12V

■ Minimum DC Gain: 20V/mV, $R_L = 500\Omega$

Settling Time to 0.1%: 75ns, 10V Step

Settling Time to 0.01%: 95ns, 10V Step

• Differential Gain: 0.1%, $A_V = 2$, $R_I = 150\Omega$

Differential Phase: 0.2°, A_V = 2, R_L = 150Ω

APPLICATIONS

- Wideband Amplifiers
- Buffers
- Active Filters
- Video and RF Amplification
- Cable Drivers
- 8-, 10-, 12-Bit Data Acquisition Systems

DESCRIPTION

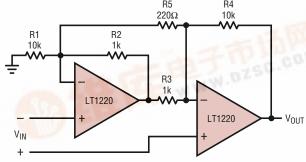
The LT®1220 is a high speed operational amplifier with superior DC performance. The LT1220 features reduced input offset voltage, lower input bias currents and higher DC gain than devices with comparable bandwidth and slew rate. The circuit is a single gain stage that includes proprietary DC gain enhancement circuitry to obtain precision with high speed. The high gain and fast settling time make the circuit an ideal choice for data acquisition systems. The circuit is also capable of driving large capacitive loads which makes it useful in buffer or cable driver applications.

The LT1220 is a member of a family of fast, high performance amplifiers that employ Linear Technology Corporation's advanced complementary bipolar processing. For applications with gains of 4 or greater the LT1221 can be used, and for gains of 10 or greater the LT1222 can be used for increased bandwidth.

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TYPICAL APPLICATION

Two Op Amp Instrumentation Amplifier



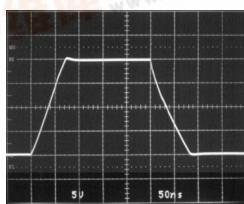
GAIN = [R4/R3][1 + (1/2)(R2/R1 + R3/R4) + (R2 + R3)/R5] = 102 TRIM R5 FOR GAIN

TRIM R1 FOR COMMON MODE REJECTION

dzsc.com

LT1220 • TA01

Inverter Pulse Response



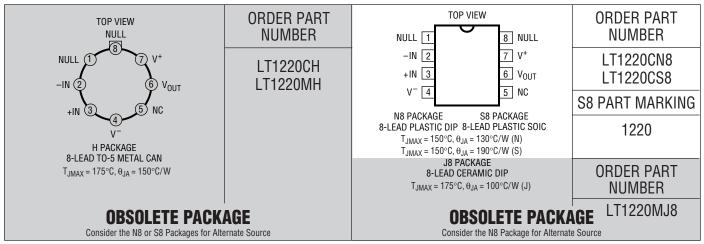
 $R_F = R_G = 1k$ $V_{IN} = 20V$ $V_S = \pm 15V$ f = 2MHz LT1220 • TA02

ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V+ to V-)
Differential Input Voltage ±6V
Input Voltage ±V _S
Output Short-Circuit Duration (Note 2) Indefinite
Specified Temperature Range
LT1220C (Note 3) 0°C to 70°C
LT1220M (OBSOLETE)55°C to 125°C

Operating Temperature Range	
LT1220C	−40°C TO 85°C
LT1220M (OBSOLETE)	. −55°C to 150°C
Maximum Junction Temperature (See	Below)
Diaglia Dagliana	
Plastic Package	150°C
Ceramic Package (OBSOLETE)	
	175°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS $T_A = 25^{\circ}C$, $V_S = \pm 15V$, $V_{CM} = 0V$, unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
$\overline{V_{0S}}$	Input Offset Voltage	(Note 4)		0.5	1	mV
I _{OS}	Input Offset Current			100	300	nA
I _B	Input Bias Current			100	300	nA
e _n	Input Noise Voltage	f = 10kHz		17		nV/√Hz
in	Input Noise Current	f = 10kHz		2		pA/√Hz
R _{IN}	Input Resistance	V _{CM} = ±12V Differential	20	45 150		MΩ kΩ
C _{IN}	Input Capacitance			2		pF
	Input Voltage Range (Positive) Input Voltage Range (Negative)		12	14 -13	-12	V
CMRR	Common Mode Rejection Ratio	V _{CM} = ±12V	92	114		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5V \text{ to } \pm 15V$	90	94		dB
A _{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10V$, $R_L = 500\Omega$	20	50		V/mV
V_{OUT}	Output Swing	$R_L = 500\Omega$	12	13		±V
I _{OUT}	Output Current	V _{OUT} = ±12V	24	26		mA
SR	Slew Rate	(Note 5)	200	250		V/µs
	Full Power Bandwidth	10V Peak (Note 6)		4		MHz
GBW	Gain-Bandwidth	f = 1MHz		45		MHz

ELECTRICAL CHARACTERISTICS $v_s = \pm 15 V$, $T_A = 25 ^{\circ}C$, $v_{CM} = 0 V$, unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN TYP MAX	UNITS
t _r , t _f	Rise Time, Fall Time	A _V = 1, 10% to 90%, 0.1V	2.5	ns
	Overshoot	A _V = 1, 0.1V	5	%
	Propagation Delay	A _V = 1, 50% V _{IN} to 50% V _{OUT} , 0.1V	4.9	ns
t _s	Settling Time	10V Step, 0.1% 10V Step, 0.01%	75 95	ns ns
	Differential Gain	$f = 3.58MHz$, $R_L = 150\Omega$ (Note 7) $f = 3.58MHz$, $R_L = 1k$ (Note 7)	0.10 0.02	% %
	Differential Phase	$f = 3.58MHz$, $R_L = 150\Omega$ (Note 7) $f = 3.58MHz$, $R_L = 1k$ (Note 7)	0.20 0.03	DEG DEG
$\overline{R_0}$	Output Resistance	A _V = 1, f = 1MHz	1	Ω
Is	Supply Current		8 10.5	mA

The ullet denotes the specifications which apply over the temperature range $0^{\circ}C \leq T_A \leq 70^{\circ}C$, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = \pm 15V$, $V_{CM} = 0V$, unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
$\overline{V_{0S}}$	Input Offset Voltage	(Note 4)	•		0.5	3.5	mV
	Input V _{OS} Drift		•		20		μV/°C
I _{OS}	Input Offset Current		•		100	400	nA
I _B	Input Bias Current		•		100	400	nA
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 12V$	•	92	114		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5V \text{ to } \pm 15V$	•	86	94		dB
A _{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10V$, $R_L = 500\Omega$	•	20	50		V/mV
V _{OUT}	Output Swing	$R_L = 500\Omega$	•	12	13		±V
I _{OUT}	Output Current	$V_{OUT} = \pm 12V$	•	24	26		mA
SR	Slew Rate	(Note 5)	•	180	250		V/µs
Is	Supply Current		•		8	11	mA

The ullet denotes the specifications which apply over the temperature range $-55^{\circ}C \leq T_A \leq 125^{\circ}C$, otherwise specifications are at $T_A = 25^{\circ}C$. $V_S = \pm 15V$, $V_{CM} = 0V$, unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage	(Note 4)	•		0.5	4	mV
	Input V _{OS} Drift		•		20		μV/°C
I _{OS}	Input Offset Current		•		100	800	nA
I _B	Input Bias Current		•		100	1000	nA
CMRR	Common Mode Rejection Ratio	V _{CM} = ±12V	•	92	114		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 5V \text{ to } \pm 15V$	•	82	94		dB
A _{VOL}	Large-Signal Voltage Gain	$V_{OUT} = \pm 10V$, $R_L = 500\Omega$	•	5	50		V/mV
V _{OUT}	Output Swing	$R_L = 500\Omega$ $R_L = 1k$	•	10 12	13 13		±V ±V
I _{OUT}	Output Current	$V_{OUT} = \pm 10V$ $V_{OUT} = \pm 12V$	•	20 12	26 13		mA mA
SR	Slew Rate	(Note 5)	•	130	250		V/µs
Is	Supply Current		•		8	11	mA

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: A heat sink may be required when the output is shorted indefinitely. **Note 3:** Commercial parts are designed to operate over -40° C to 85°C, but are not tested nor guaranteed beyond 0°C to 70°C. Industrial grade parts specified and tested over -40° C to 85°C are available on special request.

Consult factory.

Note 4: Input offset voltage is pulse tested and is exclusive of warm-up drift.

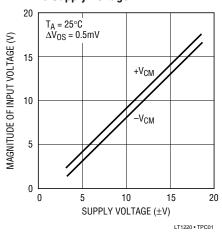
Note 5: Slew rate is measured between $\pm 10V$ on an output swing of $\pm 12V$.

Note 6: FPBW = $SR/2\pi V_P$.

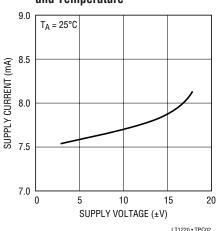
Note 7: Differential Gain and Phase are tested in $A_V = 2$ with five amps in series. Attenuators of 1/2 are used as loads $(75\Omega, 75\Omega)$ and 499Ω , 499Ω).

TYPICAL PERFORMANCE CHARACTERISTICS

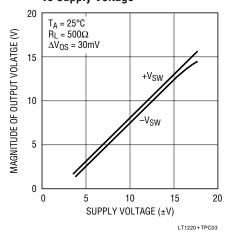
Input Common Mode Range vs Supply Voltage



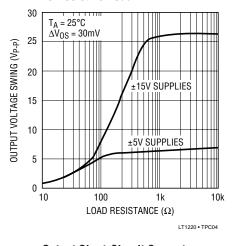
Supply Current vs Supply Voltage and Temperature



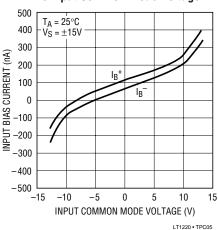
Output Voltage Swing vs Supply Voltage



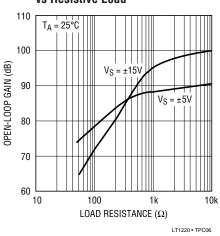
Output Voltage Swing vs Resistive Load



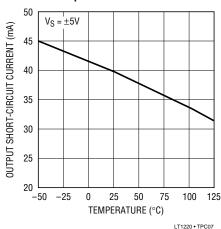
Input Bias Current vs Input Common Mode Voltage



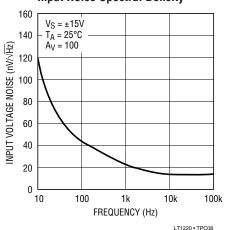
Open-Loop Gain vs Resistive Load



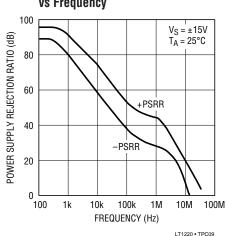
Output Short-Circuit Current vs Temperature



Input Noise Spectral Density

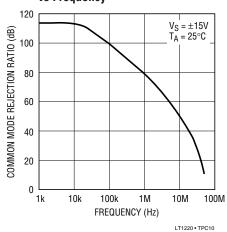


Power Supply Rejection Ratio vs Frequency

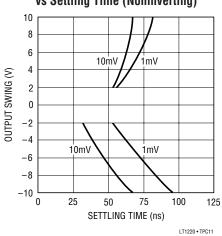


TYPICAL PERFORMANCE CHARACTERISTICS

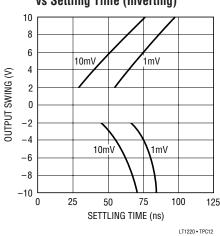
Common Mode Rejection Ratio vs Frequency



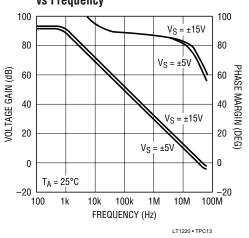
Output Swing and Error vs Settling Time (Noninverting)



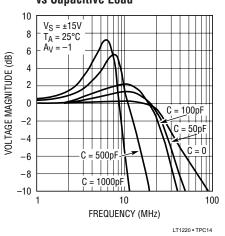
Output Swing and Error vs Settling Time (Inverting)



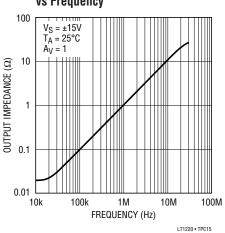
Voltage Gain and Phase vs Frequency



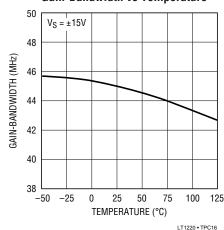
Frequency Response vs Capacitive Load



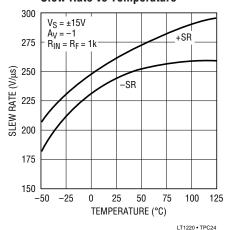
Closed-Loop Output Impedance vs Frequency



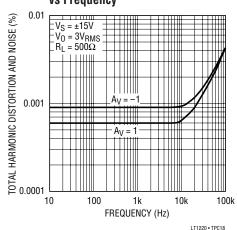
Gain-Bandwidth vs Temperature



Slew Rate vs Temperature

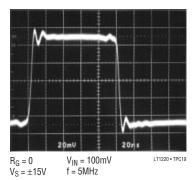


Total Harmonic Distortion vs Frequency

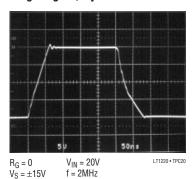


TYPICAL PERFORMANCE CHARACTERISTICS

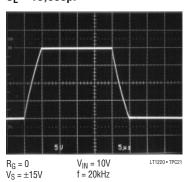
Small Signal, A_V = 1



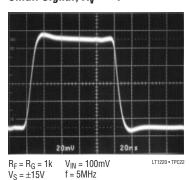
Large Signal, $A_V = 1$



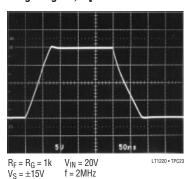
Large Signal, $A_V = 1$, $C_L = 10.000pF$



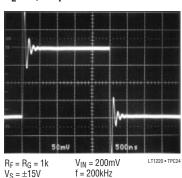
Small Signal, $A_V = -1$



Large Signal, $A_V = -1$



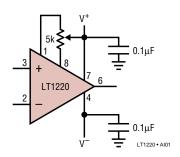
Small Signal, $A_V = -1$, $C_L = 1,000pF$



APPLICATIONS INFORMATION

The LT1220 may be inserted directly into HA2505/15/25, HA2541/2/4, AD817, AD847, EL2020, EL2044 and LM6361 applications, provided that the nulling circuitry is removed. The suggested nulling circuit for the LT1220 is shown in the following figure.

Offset Nulling



Layout and Passive Components

The LT1220 amplifier is easy to apply and tolerant of less than ideal layouts. For maximum performance (for example, fast settling time) use a ground plane, short lead lengths and RF-quality bypass capacitors (0.01 μ F to 0.1 μ F). For high driver current applications use low ESR bypass capacitors (1 μ F to 10 μ F tantalum). Sockets should be avoided when maximum frequency performance is required, although low profile sockets can provide reasonable performance up to 50MHz. For more details see Design Note 50. Feedback resistors greater than 5k are not recommended because a pole is formed with the input capacitance which can cause peaking or oscillations.

APPLICATIONS INFORMATION

Input Considerations

Bias current cancellation circuitry is employed on the inputs of the LT1220 so the input bias current and input offset current have identical specifications. For this reason, matching the impedance on the inputs to reduce bias current errors is not necessary.

Capacitive Loading

The LT1220 is stable with capacitive loads. This is accomplished by sensing the load induced output pole and adding compensation at the amplifier gain node. As the capacitive load increases, both the bandwidth and phase margin decrease. There will be peaking in the frequency domain as shown in the curve of Frequency Response vs Capacitive Load. The small-signal transient response will have more overshoot as shown in the photo of the small-signal response with 1000pF load. The large-signal response with a 10,000pF load shows the output slew rate being limited to $4V/\mu s$ by the short-circuit current. The LT1220 can drive coaxial cable directly, but for best pulse fidelity a resistor of value equal to the characteristic impedance of the cable

(i.e., 75Ω) should be placed in series with the output. The other end of the cable should be terminated with the same value resistor to ground.

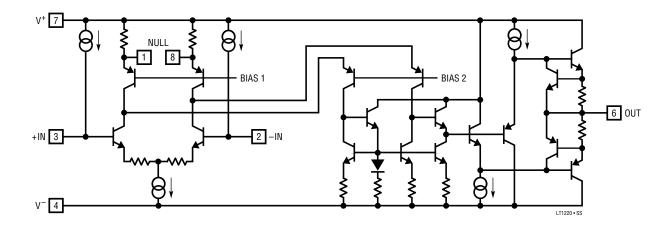
DAC Current-to-Voltage Amplifier

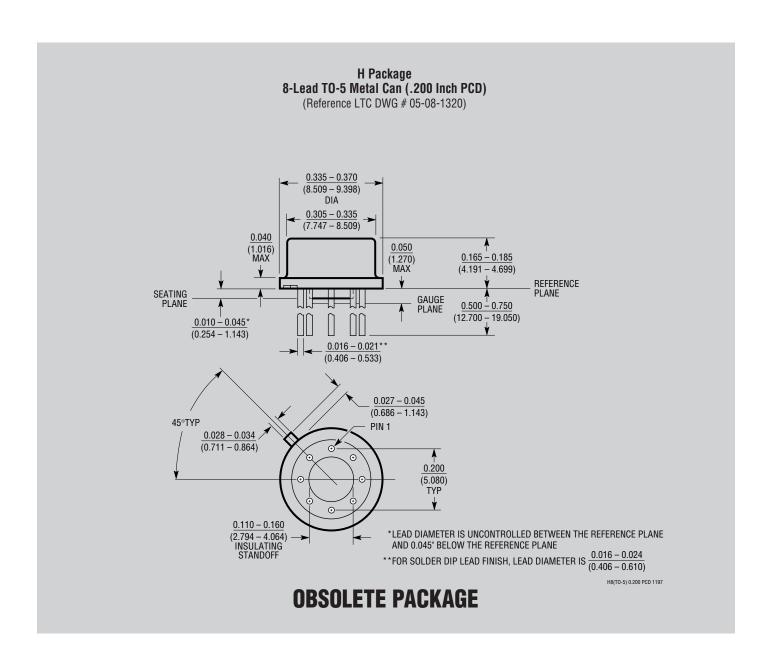
The high gain, low offset voltage, low input bias current, and fast settling of the LT1220 make it particularly useful as an I/V converter for current output DACs. A typical application is shown with a 565A type, 12-bit, 2mA full-scale output current DAC. The 5k resistor around the LT1220 is internal to the DAC and gives a 10V full-scale output voltage. A 5pF capacitor in parallel with the feedback resistor compensates for the DAC output capacitance and improves settling. The output of the LT1220 settles to 1/2LSB (1.2mV) in less than 300ns. The accuracy of this circuit is equal to:

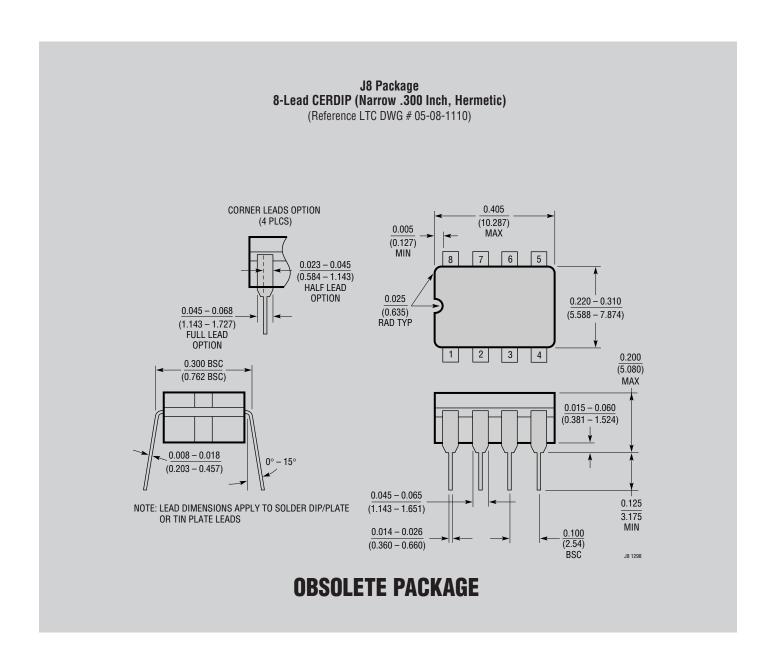
$$V_{ERROR} = V_{OS} + (I_{OS} \bullet 5k\Omega) + (V_{OUT}/A_{VOL})$$

At room temperature the worst-case error is 3mV (1.2LSB). Typically the error is 1.2mV (1/2LSB). Over the commercial temperature range the worse-case error is 6mV (2.5LSB).

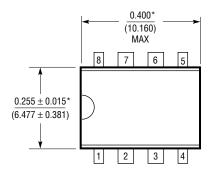
SIMPLIFIED SCHEMATIC

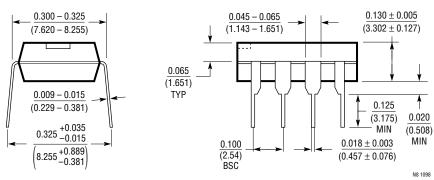






N8 Package 8-Lead PDIP (Narrow .300 Inch) (Reference LTC DWG # 05-08-1510)

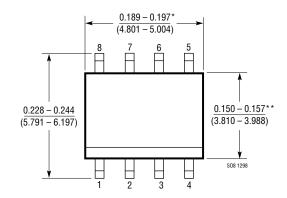


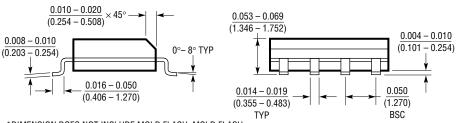


*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

\$8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch)

(Reference LTC DWG # 05-08-1610)



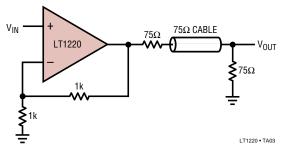


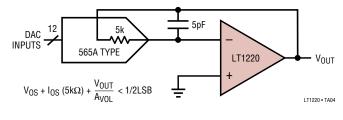
- *DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE
- **DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

TYPICAL APPLICATIONS

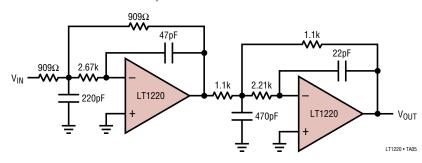
Cable Driver

DAC Current-to-Voltage Converter





1MHz, 4th Order Butterworth Filter



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1221	150MHz, 250V/µs Amplifier	$A_V \ge 4$ Version of the LT1220
LT1222	500MHz, 200V/μs Amplifier	$A_V \ge 10$ Version of the LT1220