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National Semiconductor

LM2678 SIMPLE SWITCHER[®] High Efficiency 5A Step-Down Voltage Regulator

General Description

The LM2678 series of regulators are monolithic integrated circuits which provide all of the active functions for a step-down (buck) switching regulator capable of driving up to 5A loads with excellent line and load regulation characteristics. High efficiency (>90%) is obtained through the use of a low ON-resistance DMOS power switch. The series consists of fixed output voltages of 3.3V, 5V and 12V and an adjustable output version.

The SIMPLE SWITCHER concept provides for a complete design using a minimum number of external components. A high fixed frequency oscillator (260KHz) allows the use of physically smaller sized components. A family of standard inductors for use with the LM2678 are available from several manufacturers to greatly simplify the design process.

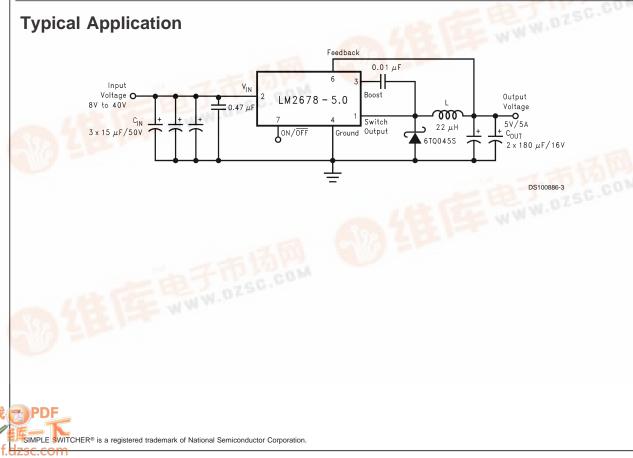
The LM2678 series also has built in thermal shutdown, current limiting and an ON/OFF control input that can power down the regulator to a low 50μ A quiescent current standby condition. The output voltage is guaranteed to a ±2% tolerance. The clock frequency is controlled to within a ±11% tolerance.

Features

- Efficiency up to 92%
- Simple and easy to design with (using off-the-shelf external components)
- 120 mΩ DMOS output switch
- 3.3V, 5V and 12V fixed output and adjustable (1.2V to 37V) versions
- 50µA standby current when switched OFF
- ±2%maximum output tolerance over full line and load conditions
- Wide input voltage range: 8V to 40V
- 260 KHz fixed frequency internal oscillator
- -40 to +125°C operating junction temperature range

Applications

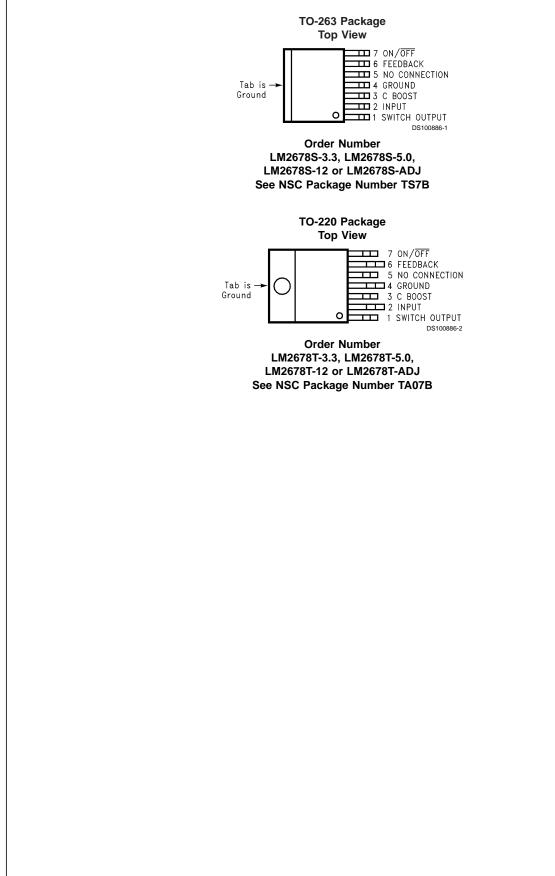
- Simple to design, high efficiency (>90%) step-down switching regulators
- Efficient system pre-regulator for linear voltage regulators
- Battery chargers



March 2000



Connection Diagram and Ordering Information



www.national.com

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Input Supply Voltage	45V
ON/OFF Pin Voltage	-0.1V to 6V
Switch Voltage to Ground	–1V to V _{IN}
Boost Pin Voltage	V _{SW} + 8V
Feedback Pin Voltage	–0.3V to 14V
Power Dissipation	Internally Limited
ESD (Note 2)	2 kV

Storage Temperature Range	-65°C to 150°C
Soldering Temperature	
Wave	4 sec, 260°C
Infrared	10 sec, 240°C
Vapor Phase	75 sec, 219°C

Operating Ratings

Supply Voltage Junction Temperature Range (T_J) 8V to 40V -40°C to 125°C

Electrical Characteristics Limits appearing in **bold type face** apply over the entire junction temperature range of operation, -40° C to 125° C. Specifications appearing in normal type apply for $T_A = T_J = 25^{\circ}$ C.

LM2678-3.3

Symbol	Parameter	Conditions	Typical	Min	Мах	Units
			(Note 3)	(Note 4)	(Note 4)	
V _{OUT}	Output Voltage	V_{IN} = 8V to 40V, 100mA $\leq I_{OUT} \leq 5A$	3.3	3.234/ 3.201	3.366/ 3.399	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 5A$	82			%

LM2678-5.0

Symbol	Parameter	Conditions	Typical	Min	Max	Units
			(Note 3)	(Note 4)	(Note 4)	
V _{OUT}	Output Voltage	V_{IN} = 8V to 40V, 100mA $\leq I_{OUT} \leq 5A$	5.0	4.900/ 4.850	5.100/ 5.150	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 5A$	84			%

LM2678-12

Symbol	Parameter	Conditions	Typical	Min	Max	Units
			(Note 3)	(Note 4)	(Note 4)	
V _{OUT}	Output Voltage	V_{IN} = 15V to 40V, 100mA $\leq I_{\text{OUT}} \leq 5\text{A}$	12	11.76/ 11.64	12.24/ 12.36	V
η	Efficiency	$V_{IN} = 24V, I_{LOAD} = 5A$	92			%

LM2678-ADJ

Symbol	Parameter	Conditions	Тур	Min	Max	Units
			(Note 3)	(Note 4)	(Note 4)	
V _{FB}	Feedback Voltage	V_{IN} = 8V to 40V, 100mA \leq $I_{\text{OUT}} \leq$ 5A V_{OUT} Programmed for 5V	1.21	1.186/ 1.174	1.234/ 1.246	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 5A$	84			%

3

All Output Voltage Versions Electrical Characteristics

Limits appearing in **bold type face** apply over the entire junction temperature range of operation, -40° C to 125°C. Specifications appearing in normal type apply for $T_A = T_J = 25^{\circ}$ C. Unless otherwise specified V_{IN}=12V for the 3.3V, 5V and Adjustable versions and V_{IN}=24V for the 12V version.

Symbol	Parameter	Conditions	Тур	Min	Max	Units
DEVICE	PARAMETERS					
Ι _Q	Quiescent Current	V _{FEEDBACK} = 8V For 3.3V, 5.0V, and ADJ Versions	4.2		6	mA
		V _{FEEDBACK} = 15V For 12V Versions				
I _{stby}	Standby Quiescent Current	ON/OFF Pin = 0V	50		100/ 150	μA
I _{CL}	Current Limit		7	6.1/ 5.75	8.3/ 8.75	A
I _L	Output Leakage Current	$V_{IN} = 40V, ON/OFF Pin = 0V$ $V_{SWITCH} = 0V$ $V_{SWITCH} = -1V$	1 6		200 15	μA mA
R _{DS(ON)}	Switch On-Resistance	I _{SWITCH} = 5A	0.12		0.14/ 0.225	Ω
f _O	Oscillator Frequency	Measured at Switch Pin	260	225	280	kHz
D	Duty Cycle	Maximum Duty Cycle Minimum Duty Cycle	91 0			% %
I _{BIAS}	Feedback Bias Current	V _{FEEDBACK} = 1.3V ADJ Version Only	85			nA
V _{on/off}	ON/OFF Threshold Voltage		1.4	0.8	2.0	V
I _{ON/OFF}	ON/OFF Input Current	ON/OFF Input = 0V	20		45	μA
θ_{JA}	Thermal Resistance	T Package, Junction to Ambient (Note 5)	65			
θ_{JA}		T Package, Junction to Ambient (Note 6)	45			
θ_{JC}		T Package, Junction to Case	2			
θ _{JA}		S Package, Junction to Ambient (Note 7)	56			°C/W
θ_{JA}		S Package, Junction to Ambient (Note 8)	35			
θ_{JA}		(Note 9) S Package, Junction to Ambient (Note 9)	26			
θ_{JC}		S Package, Junction to Case	2			++

Note 1: Absolute Maximum Ratings are limits beyond which damage to the device may occur. Operating Ratings indicate conditions under which of the device is guaranteed. Operating Ratings do not imply guaranteed performance limits. For guaranteed performance limits and associated test condition, see the electrical Characteristics tables.

Note 2: ESD was applied using the human-body model, a 100pF capacitor discharged through a 1.5 k Ω resistor into each pin.

Note 3: Typical values are determined with $T_A = T_J = 25^{\circ}C$ and represent the most likely norm.

Note 4: All limits are guaranteed at room temperature (standard type face) and at **temperature extremes (bold type face)**. All room temperature limits are 100% tested during production with $T_A = T_J = 25$ °C. All limits at temperature extremes are guaranteed via correlation using standard standard Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

Note 5: Junction to ambient thermal resistance (no external heat sink) for the 7 lead TO-220 package mounted vertically, with ½ inch leads in a socket, or on a PC board with minimum copper area.

Note 6: Junction to ambient thermal resistance (no external heat sink) for the 7 lead TO-220 package mounted vertically, with ½ inch leads soldered to a PC board containing approximately 4 square inches of (1 oz.) copper area surrounding the leads.

(Continued)

Note 7: Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board area of 0.136 square inches (the same size as the TO-263 package) of 1 oz. (0.0014 in. thick) copper.

Note 8: Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board area of 0.4896 square inches (3.6 times the area of the TO-263 package) of 1 oz. (0.0014 in. thick) copper.

Note 9: Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board copper area of 1.0064 square inches (7.4 times the area of the TO-263 package) of 1 oz. (0.0014 in. thick) copper. Additional copper area will reduce thermal resistance further. See the thermal model in Switchers Made Simple® software.

Typical Performance Characteristics

V_{IN} = 20V

5

Normalized Output Voltage

Efficiency vs ILOAD

12

5١

3.3V

Standby Quiescent Current

1 1.5 2 2.5 3 3.5 4 4.5

= 125°C

= 25°C

= -40°C

25 50 75

JUNCTION TEMPERATURE (°C)

100 125

DS100886-15

-50-25 0

15 20 25

SUPPLY VOLTAGE (V) DS100886-40

T₁

10

5

Switching Frequency

LOAD CURRENT (A) DS100886-12

 $V_{ON/OFF} = 0V$

30 35 40

100

95

85

80

75

70

100

80

60

40

20

0

270

260

250

240

230

220

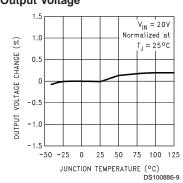
210

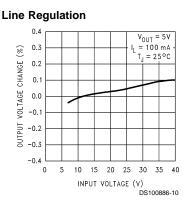
FREQUENCY (Hz)

CURRENT (µA)

⊗ 90

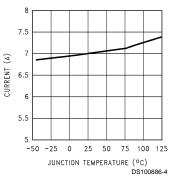
EFFICIENCY



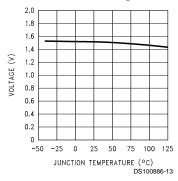




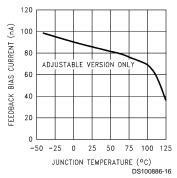
Switch Current Limit



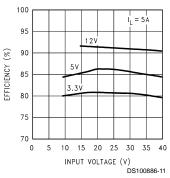
ON/OFF Threshold Voltage



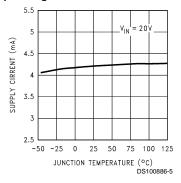
Feedback Pin Bias Current



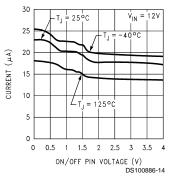
Efficiency vs Input Voltage



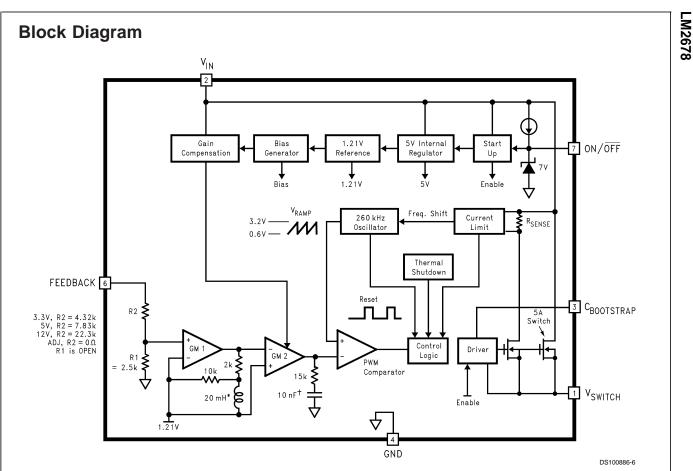
Operating Quiescent Current



ON/OFF Pin Current (Sourcing)



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* Active Inductor Patent Number 5,514,947

† Active Capacitor Patent Number 5,382,918



Typical Performance Characteristics Continuous Mode Switching Waveforms Discontinuous Mode Switching Waveforms V_{IN} = 20V, V_{OUT} = 5V, I_{LOAD} = 5A V_{IN} = 20V, V_{OUT} = 5V, I_{LOAD} = 500 mA L = 10 μ H, C_{OUT} = 400 μ F, C_{OUT}ESR = 13 m Ω L = 10 μ H, C_{OUT} = 400 μ F, C_{OUT}ESR = 13 m Ω 20V 20 A 10V A 10V ٥٧ 0٧ B 4A 2A 1A В 0A 0A C OV С 0٧ 1 $\mu sec/Div$ 1 μsec/Div DS100886-17 DS100886-18 A: VSW Pin Voltage, 10 V/div. A: V_{SW} Pin Voltage, 10 V/div. B: Inductor Current, 2 A/div B: Inductor Current, 1 A/div C: Output Ripple Voltage, 20 mV/div AC-Coupled C: Output Ripple Voltage, 20 mV/div AC-Coupled Horizontal Time Base: 1 µs/div Horizontal Time Base: 1 µs//iv Load Transient Response for Continuous Mode Load Transient Response for Discontinuous Mode $V_{IN} = 20V, V_{OUT} = 5V$ $V_{IN} = 20V, V_{OUT} = 5V,$ L = 10 μ H, C_{OUT} = 400 μ F, C_{OUT}ESR = 13 m Ω L = 10 μ H, C_{OUT} = 400 μ F, C_{OUT}ESR = 13 m Ω A OV A OV 3A 44 2A B 2A В 1A 0٧ ٥v 100 $\mu sec/Div$ DS100886-19 200 µsec/Div DS100886-20 A: Output Voltage, 100 mV//div, AC-Coupled. A: Output Voltage, 100 mV/div, AC-Coupled. B: Load Current: 500 mA to 5A Load Pulse B: Load Current: 200 mA to 3A Load Pulse Horizontal Time Base: 100 µs/div Horizontal Time Base: 200 µs/div

Application Hints

The LM2678 provides all of the active functions required for a step-down (buck) switching regulator. The internal power switch is a DMOS power MOSFET to provide power supply designs with high current capability, up to 5A, and highly efficient operation.

The LM2678 is part of the *SIMPLE SWITCHER* family of power converters. A complete design uses a minimum number of external components, which have been pre-determined from a variety of manufacturers. Using either this data sheet or a design software program called *LM267X Made Simple* (version 2.0) a complete switching power supply can be designed quickly. The software is provided free of charge and can be downloaded from National Semiconductor's Internet site located at http://www.national.com.

PIN 1 - Switch Output

This is the output of a power MOSFET switch connected directly to the input voltage. The switch provides energy to an inductor, an output capacitor and the load circuitry under control of an internal pulse-width-modulator (PWM). The PWM controller is internally clocked by a fixed 260KHz oscillator. In a standard step-down application the duty cycle (Time ON/Time OFF) of the power switch is proportional to the ratio of the power supply output voltage to the input voltage. The voltage on pin 1 switches between Vin (switch ON) and below ground by the voltage drop of the external Schottky diode (switch OFF).

PIN 2 - Input

The input voltage for the power supply is connected to pin 2. In addition to providing energy to the load the input voltage also provides bias for the internal circuitry of the LM2678. For guaranteed performance the input voltage must be in the range of 8V to 40V. For best performance of the power supply the input pin should always be bypassed with an input capacitor located close to pin 2.

DESIGN CONSIDERATIONS

PIN 3 - C Boost

A capacitor must be connected from pin 3 to the switch output, pin 1. This capacitor boosts the gate drive to the internal MOSFET above Vin to fully turn it ON. This minimizes conduction losses in the power switch to maintain high efficiency. The recommended value for C Boost is $0.01 \mu F$.

PIN 4 - Ground

This is the ground reference connection for all components in the power supply. In fast-switching, high-current applications such as those implemented with the LM2678, it is recommended that a broad ground plane be used to minimize signal coupling throughout the circuit

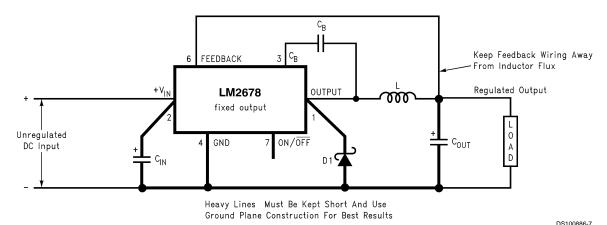
PIN 5 - No Connection

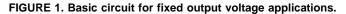
PIN 6 - Feedback

This is the input to a two-stage high gain amplifier, which drives the PWM controller. It is necessary to connect pin 6 to the actual output of the power supply to set the dc output voltage. For the fixed output devices (3.3V, 5V and 12V outputs), a direct wire connection to the output is all that is required as internal gain setting resistors are provided inside the LM2678. For the adjustable output version two external resistors are required to set the dc output voltage. For stable operation of the power supply it is important to prevent coupling of any inductor flux to the feedback input.

PIN 7 - ON/OFF

This input provides an electrical ON/OFF control of the power supply. Connecting this pin to ground or to any voltage less than 0.8V will completely turn OFF the regulator. The current drain from the input supply when OFF is only 50 μ A. Pin 7 has an internal pull-up current source of approximately 20 μ A and a protection clamp zener diode of 7V to ground. When electrically driving the ON/OFF pin the high voltage level for the ON condition should not exceed the 6V absolute maximum limit. When ON/OFF control is not required pin 7 should be left open circuited.





Application Hints (Continued)

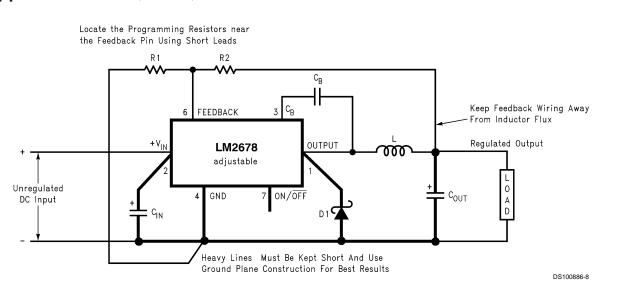


FIGURE 2. Basic circuit for adjustable output voltage applications

Power supply design using the LM2678 is greatly simplified by using recommended external components. A wide range of inductors, capacitors and Schottky diodes from several manufacturers have been evaluated for use in designs that cover the full range of capabilities (input voltage, output voltage and load current) of the LM2678. A simple design procedure using nomographs and component tables provided in this data sheet leads to a working design with very little effort. Alternatively, the design software, *LM267X Made Simple* (version 2.0), can also be used to provide instant component selection, circuit performance calculations for evaluation, a bill of materials component list and a circuit schematic.

The individual components from the various manufacturers called out for use are still just a small sample of the vast array of components available in the industry. While these components are recommended, they are not exclusively the only components for use in a design. After a close comparison of component specifications, equivalent devices from other manufacturers could be substituted for use in an application.

Important considerations for each external component and an explanation of how the nomographs and selection tables were developed follows.

INDUCTOR

The inductor is the key component in a switching regulator. For efficiency the inductor stores energy during the switch ON time and then transfers energy to the load while the switch is OFF.

Nomographs are used to select the inductance value required for a given set of operating conditions. The nomographs assume that the circuit is operating in continuous mode (the current flowing through the inductor never falls to zero). The magnitude of inductance is selected to maintain a maximum ripple current of 30% of the maximum load current. If the ripple current exceeds this 30% limit the next larger value is selected.

The inductors offered have been specifically manufactured to provide proper operation under all operating conditions of input and output voltage and load current. Several part types are offered for a given amount of inductance. Both surface mount and through-hole devices are available. The inductors from each of the three manufacturers have unique characteristics.

Renco: ferrite stick core inductors; benefits are typically lowest cost and can withstand ripple and transient peak currents above the rated value. These inductors have an external magnetic field, which may generate EMI.

Pulse Engineering: powdered iron toroid core inductors; these also can withstand higher than rated currents and, being toroid inductors, will have low EMI.

Coilcraft: ferrite drum core inductors; these are the smallest physical size inductors and are available only as surface mount components. These inductors also generate EMI but less than stick inductors.

OUTPUT CAPACITOR

The output capacitor acts to smooth the dc output voltage and also provides energy storage. Selection of an output capacitor, with an associated equivalent series resistance (ESR), impacts both the amount of output ripple voltage and stability of the control loop.

The output ripple voltage of the power supply is the product of the capacitor ESR and the inductor ripple current. The capacitor types recommended in the tables were selected for having low ESR ratings.

In addition, both surface mount tantalum capacitors and through-hole aluminum electrolytic capacitors are offered as solutions.

Impacting frequency stability of the overall control loop, the output capacitance, in conjunction with the inductor, creates a double pole inside the feedback loop. In addition the capacitance and the ESR value create a zero. These frequency response effects together with the internal frequency compensation circuitry of the LM2678 modify the gain and phase shift of the closed loop system.

As a general rule for stable switching regulator circuits it is desired to have the unity gain bandwidth of the circuit to be limited to no more than one-sixth of the controller switching frequency. With the fixed 260KHz switching frequency of the LM2678, the output capacitor is selected to provide a unity gain bandwidth of 40KHz maximum. Each recommended capacitor value has been chosen to achieve this result.

In some cases multiple capacitors are required either to reduce the ESR of the output capacitor, to minimize output ripple (a ripple voltage of 1% of Vout or less is the assumed performance condition), or to increase the output capacitance to reduce the closed loop unity gain bandwidth (to less than 40KHz). When parallel combinations of capacitors are required it has been assumed that each capacitor is the exact same part type.

The RMS current and working voltage (WV) ratings of the output capacitor are also important considerations. In a typical step-down switching regulator, the inductor ripple current (set to be no more than 30% of the maximum load current by the inductor selection) is the current that flows through the output capacitor. The capacitor RMS current rating must be greater than this ripple current. The voltage rating of the output capacitor should be greater than 1.3 times the maximum output voltage of the power supply. If operation of the system at elevated temperatures is required, the capacitor voltage rating may be de-rated to less than the nominal room temperature rating. Careful inspection of the manufacturer's specification for de-rating of working voltage with temperature is important.

INPUT CAPACITOR

Fast changing currents in high current switching regulators place a significant dynamic load on the unregulated power source. An input capacitor helps to provide additional current to the power supply as well as smooth out input voltage variations.

Like the output capacitor, the key specifications for the input capacitor are RMS current rating and working voltage. The RMS current flowing through the input capacitor is equal to one-half of the maximum dc load current so the capacitor should be rated to handle this. Paralleling multiple capacitors proportionally increases the current rating of the total capacitance. The voltage rating should also be selected to be 1.3 times the maximum input voltage. Depending on the unregulated input power source, under light load conditions the maximum input voltage could be significantly higher than normal operation and should be considered when selecting an input capacitor.

The input capacitor should be placed very close to the input pin of the LM2678. Due to relative high current operation with fast transient changes, the series inductance of input connecting wires or PCB traces can create ringing signals at the input terminal which could possibly propagate to the output or other parts of the circuitry. It may be necessary in some designs to add a small valued $(0.1\mu F to 0.47\mu F)$ ceramic type capacitor in parallel with the input capacitor to prevent or minimize any ringing.

CATCH DIODE

When the power switch in the LM2678 turns OFF, the current through the inductor continues to flow. The path for this current is through the diode connected between the switch output and ground. This forward biased diode clamps the switch output to a voltage less than ground. This negative voltage must be greater than -1V so a low voltage drop (particularly at high current levels) Schottky diode is recommended. Total efficiency of the entire power supply is significantly impacted by the power lost in the output catch diode. The average current through the catch diode is dependent on the switch duty cycle (D) and is equal to the load current times (1-D). Use of a diode rated for much higher current than is required by the actual application helps to minimize the voltage drop and power loss in the diode.

During the switch ON time the diode will be reversed biased by the input voltage. The reverse voltage rating of the diode should be at least 1.3 times greater than the maximum input voltage.

BOOST CAPACITOR

The boost capacitor creates a voltage used to overdrive the gate of the internal power MOSFET. This improves efficiency by minimizing the on resistance of the switch and associated power loss. For all applications it is recommended to use a 0.01μ F/50V ceramic capacitor.

SIMPLE DESIGN PROCEDURE

Using the nomographs and tables in this data sheet (or use the available design software at http://www.national.com) a complete step-down regulator can be designed in a few simple steps.

Step 1: Define the power supply operating conditions:

Required output voltage

Maximum DC input voltage

Maximum output load current

Step 2: Set the output voltage by selecting a fixed output LM2678 (3.3V, 5V or 12V applications) or determine the required feedback resistors for use with the adjustable LM2678–ADJ

Step 3: Determine the inductor required by using one of the four nomographs, *Figure 3* through *Figure 6*. Table 1 provides a specific manufacturer and part number for the inductor.

Step 4: Using Table 3 (fixed output voltage) or Table 6 (adjustable output voltage), determine the output capacitance required for stable operation. Table 2 provides the specific capacitor type from the manufacturer of choice.

Step 5: Determine an input capacitor from Table 4 for fixed output voltage applications. Use Table 2 to find the specific capacitor type. For adjustable output circuits select a capacitor from Table 2 with a sufficient working voltage (WV) rating greater than Vin max, and an rms current rating greater than one-half the maximum load current (2 or more capacitors in parallel may be required).

Step 6: Select a diode from Table 5. The current rating of the diode must be greater than I load max and the Reverse Voltage rating must be greater than Vin max.

Step 7: Include a 0.01μ F/50V capacitor for Cboost in the design.

FIXED OUTPUT VOLTAGE DESIGN EXAMPLE

A system logic power supply bus of 3.3V is to be generated from a wall adapter which provides an unregulated DC voltage of 13V to 16V. The maximum load current is 4A. Through-hole components are preferred.

Step 1: Operating conditions are:

Vout = 3.3V

Vin max = 16V

lload max = 4A

Step 2: Select an LM2678T-3.3. The output voltage will have a tolerance of

 $\pm 2\%$ at room temperature and $\pm 3\%$ over the full operating temperature range.

Step 3: Use the nomograph for the 3.3V device ,*Figure 3*. The intersection of the 16V horizontal line (V_{in} max) and the 4A vertical line (I_{load} max) indicates that L46, a 15µH inductor, is required.

From Table 1, L46 in a through-hole component is available from Renco with part number RL-1283-15-43.

Step 4: Use Table 3 to determine an output capacitor. With a 3.3V output and a 15μ H inductor there are four through-hole output capacitor solutions with the number of same type capacitors to be paralleled and an identifying capacitor code given. Table 2 provides the actual capacitor characteristics. Any of the following choices will work in the circuit:

2 x 220µF/10V Sanyo OS-CON (code C5)

2 x 820µF/16V Sanyo MV-GX (code C5)

1 x 3900µF/10V Nichicon PL (code C7)

2 x 560µF/35V Panasonic HFQ (code C5)

Step 5: Use Table 4 to select an input capacitor. With 3.3V output and 15µH there are three through-hole solutions. These capacitors provide a sufficient voltage rating and an rms current rating greater than 2A (1/2 I_{load} max). Again using Table 2 for specific component characteristics the following choices are suitable:

2 x 680µF/63V Sanyo MV-GX (code C13)

1 x 1200µF/63V Nichicon PL (code C25)

1 x 1500µF/63V Panasonic HFQ (code C16)

Step 6: From Table 5 a 5A or more Schottky diode must be selected. For through-hole components only 40V rated diodes are indicated and 4 part types are suitable:

1N5825

MBR745

80SQ045

6TQ045

Step 7: A 0.01µF capacitor will be used for Cboost.

ADJUSTABLE OUTPUT DESIGN EXAMPLE

In this example it is desired to convert the voltage from a two battery automotive power supply (voltage range of 20V to 28V, typical in large truck applications) to the 14.8VDC alternator supply typically used to power electronic equipment from single battery 12V vehicle systems. The load current required is 3.5A maximum. It is also desired to implement the power supply with all surface mount components.

Step 1: Operating conditions are:

Vout = 14.8V

Vin max = 28V

lload max = 3.5A

Step 2: Select an LM2678S-ADJ. To set the output voltage to 14.9V two resistors need to be chosen (R1 and R2 in *Figure 2*). For the adjustable device the output voltage is set by the following relationship:

$$V_{OUT} = V_{FB} \left(1 + \frac{R_2}{R_1} \right)$$

Where V_{FB} is the feedback voltage of typically 1.21V.

A recommended value to use for R1 is 1K. In this example then R2 is determined to be:

$$R_{2} = R_{1} \left(\frac{V_{OUT}}{V_{FB}} - 1 \right) = 1 k\Omega \left(\frac{14.8V}{1.21V} - 1 \right)$$

R2 = 11.23KΩ

The closest standard 1% tolerance value to use is $11.3K\Omega$ This will set the nominal output voltage to 14.88V which is within 0.5% of the target value. **Step 3:** To use the nomograph for the adjustable device, *Figure 6*, requires a calculation of the inductor Volt•microsecond constant (E•T expressed in V• μ S) from the following formula:

$$E \cdot T = (V_{IN(MAX)} - V_{OUT} - V_{SAT}) \cdot \frac{V_{OUT} + V_{D}}{V_{IN(MAX)} - V_{SAT} + V_{D}} \cdot \frac{1000}{260} (V \cdot \mu s)$$

where V_{SAT} is the voltage drop across the internal power switch which is $R_{ds(ON)}$ times I_{load} . In this example this would be typically $0.12\Omega \times 3.5A$ or 0.42V and V_D is the voltage drop across the forward bisased Schottky diode, typically 0.5V. The switching frequency of 260KHz is the nominal value to use to estimate the ON time of the switch during which energy is stored in the inductor.

For this example E•T is found to be: E • T = $(28 - 14.8 - 0.42) \cdot \frac{14.8 + 0.5}{28 - 0.42 + 0.5} \cdot \frac{1000}{260} (V \cdot \mu s)$

$$E \cdot T = (12.78V) \cdot \frac{15.3V}{28.08V} \cdot 3.85 (V \cdot \mu s) = 26.8 (V \cdot \mu s)$$

Using *Figure 6*, the intersection of $27V \cdot \mu S$ horizontally and the 3.5A vertical line (I_{load} max) indicates that L48, a 47 μ H inductor, or L49, a 33 μ H inductor could be used. Either inductor will be suitable, but for this example selecting the larger inductance will result in lower ripple current.

From Table 1, L48 in a surface mount component is available from Pulse Engineering with part number P0848.

Step 4: Use Table 6 to determine an output capacitor. With a 14.8V output the 12.5 to 15V row is used and with a 47μ H inductor there are three surface mount output capacitor solutions. Table 2 provides the actual capacitor characteristics based on the C Code number. Any of the following choices can be used:

- 1 x 33µF/20V AVX TPS (code C6)
- 1 x 47µF/20V Sprague 594 (code C8)
- 1 x 47µF/20V Kemet T495 (code C8)

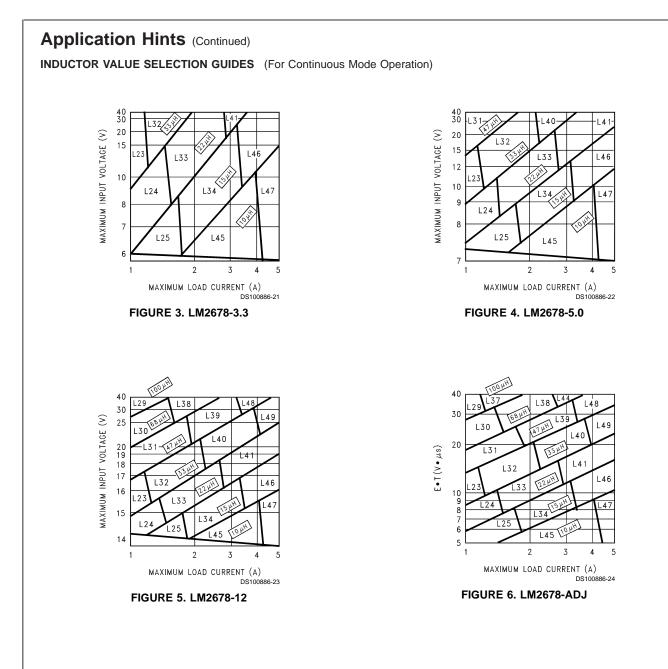
Important Note: When using the adjustable device in low voltage applications (less than 3V output), if the nomograph, Figure 6, selects an inductance of 22μ H or less, Table 6 does not provide an output capacitor solution. With these conditions the number of output capacitors required for stable operation becomes impractical. It is recommended to use either a 33μ H or 47μ H inductor and the output capacitors from Table 6.

Step 5: An input capacitor for this example will require at least a 35V WV rating with an rms current rating of 1.75A (1/2 lout max). From Table 2 it can be seen that C12, a 33μ F/35V capacitor from Sprague, has the highest voltage/current rating of the surface mount components and that two of these capacitor in parallel will be adquate.

Step 6: From Table 5 a 5A or more Schottky diode must be selected. For surface mount diodes with a margin of safety on the voltage rating one of two diodes can be used: MBRD1545CT

6TQ045S

Step 7: A 0.01µF capacitor will be used for Cboost.



Application Hints (Continued)

Inductor	Inductance	Current	Ren	со	Pulse Er	gineering	Coilcraft
Reference	(µH)	(A)	Through Hole	Surface	Through	Surface	Surface Mount
Number	(μπ)	(~)		Mount	Hole	Mount	
L23	33	1.35	RL-5471-7	RL1500-33	PE-53823	PE-53823S	DO3316-333
L24	22	1.65	RL-1283-22-43	RL1500-22	PE-53824	PE-53824S	DO3316-223
L25	15	2.00	RL-1283-15-43	RL1500-15	PE-53825	PE-53825S	DO3316-153
L29	100	1.41	RL-5471-4	RL-6050-100	PE-53829	PE-53829S	DO5022P-104
L30	68	1.71	RL-5471-5	RL6050-68	PE-53830	PE-53830S	DO5022P-683
L31	47	2.06	RL-5471-6	RL6050-47	PE-53831	PE-53831S	DO5022P-473
L32	33	2.46	RL-5471-7	RL6050-33	PE-53932	PE-53932S	DO5022P-333
L33	22	3.02	RL-1283-22-43	RL6050-22	PE-53933	PE-53933S	DO5022P-223
L34	15	3.65	RL-1283-15-43	_	PE-53934	PE-53934S	DO5022P-153
L38	68	2.97	RL-5472-2	—	PE-54038	PE-54038S	—
L39	47	3.57	RL-5472-3	—	PE-54039	PE-54039S	—
L40	33	4.26	RL-1283-33-43	_	PE-54040	PE-54040S	—
L41	22	5.22	RL-1283-22-43		PE-54041	P0841	—
L44	68	3.45	RL-5473-3	_	PE-54044	_	—
L45	10	4.47	RL-1283-10-43		_	P0845	DO5022P-103HC
L46	15	5.60	RL-1283-15-43	_	_	P0846	DO5022P-153HC
L47	10	5.66	RL-1283-10-43	_	_	P0847	DO5022P-103HC
L48	47	5.61	RL-1282-47-43	_	_	P0848	—
L49	33	5.61	RL-1282-33-43		—	P0849	—

TABLE 1. Inductor Manufacturer Part Numbers

Inductor Manufacturer Contact Numbers

Coilcraft	Phone	(800) 322-2645
	FAX	(708) 639-1469
Coilcraft, Europe	Phone	+44 1236 730 595
	FAX	+44 1236 730 627
Pulse Engineering	Phone	(619) 674-8100
	FAX	(619) 674-8262
Pulse Engineering,	Phone	+353 93 24 107
Europe	FAX	+353 93 24 459
Renco Electronics	Phone	(800) 645-5828
	FAX	(516) 586-5562

Ormanitan				Su	rface Mo	unt				
Capacitor Reference	AVX	(TPS Se	ries	Sprag	ue 594D	Series	Keme	Kemet T495 Series		
Code	С	WV	Irms	С	WV	Irms	С	WV	Irms	
oouc	(µF)	(V)	(A)	(µF)	(V)	(A)	(µF)	(V)	(A)	
C1	330	6.3	1.15	120	6.3	1.1	100	6.3	0.82	
C2	100	10	1.1	220	6.3	1.4	220	6.3	1.1	
C3	220	10	1.15	68	10	1.05	330	6.3	1.1	
C4	47	16	0.89	150	10	1.35	100	10	1.1	
C5	100	16	1.15	47	16	1	150	10	1.1	
C6	33	20	0.77	100	16	1.3	220	10	1.1	
C7	68	20	0.94	180	16	1.95	33	20	0.78	
C8	22	25	0.77	47	20	1.15	47	20	0.94	
C9	10	35	0.63	33	25	1.05	68	20	0.94	
C10	22	35	0.66	68	25	1.6	10	35	0.63	
C11				15	35	0.75	22	35	0.63	
C12				33	35	1	4.7	50	0.66	
C13				15	50	0.9				

TABLE 2. Input and Output Capacitor Codes

Application Hints (Continued)

Input and Output Capacitor Codes (continued)

						Throug	h Hole					
Capacitor	Sanyo	OS-CO	N SA	Sanyo	Sanyo MV-GX Series Nichicon PL Series					Panasonic HFQ Series		
Reference	Series											
Code	C (µF)	WV	Irms	С	WV	Irms	С	WV	Irms	С	WV	Irms
		(V)	(A)	(µF)	(V)	(A)	(µF)	(V)	(A)	(µF)	(V)	(A)
C1	47	6.3	1	1000	6.3	0.8	680	10	0.8	82	35	0.4
C2	150	6.3	1.95	270	16	0.6	820	10	0.98	120	35	0.44
C3	330	6.3	2.45	470	16	0.75	1000	10	1.06	220	35	0.76
C4	100	10	1.87	560	16	0.95	1200	10	1.28	330	35	1.01
C5	220	10	2.36	820	16	1.25	2200	10	1.71	560	35	1.4
C6	33	16	0.96	1000	16	1.3	3300	10	2.18	820	35	1.62
C7	100	16	1.92	150	35	0.65	3900	10	2.36	1000	35	1.73
C8	150	16	2.28	470	35	1.3	6800	10	2.68	2200	35	2.8
C9	100	20	2.25	680	35	1.4	180	16	0.41	56	50	0.36
C10	47	25	2.09	1000	35	1.7	270	16	0.55	100	50	0.5
C11			1	220	63	0.76	470	16	0.77	220	50	0.92
C12				470	63	1.2	680	16	1.02	470	50	1.44
C13				680	63	1.5	820	16	1.22	560	50	1.68
C14				1000	63	1.75	1800	16	1.88	1200	50	2.22
C15							220	25	0.63	330	63	1.42
C16							220	35	0.79	1500	63	2.51
C17							560	35	1.43			
C18							2200	35	2.68			
C19							150	50	0.82			
C20							220	50	1.04			
C21							330	50	1.3			
C22							100	63	0.75			
C23							390	63	1.62			
C24							820	63	2.22			
C25							1200	63	2.51			

Capacitor Manufacturer Contact Numbers

Nichicon	Phone	(847) 843-7500
	FAX	(847) 843-2798
Panasonic	Phone	(714) 373-7857
	FAX	(714) 373-7102
AVX	Phone	(845) 448-9411
	FAX	(845) 448-1943
Sprague/Vishay	Phone	(207) 324-4140
	FAX	(207) 324-7223
Sanyo	Phone	(619) 661-6322
	FAX	(619) 661-1055
Kemet	Phone	(864) 963-6300
	FAX	(864) 963-6521

		Surface Mount								
Output Voltage	Inductance (µH)	AVX TPS Series			Sprague 594D Series		495 Series			
(V)		No.	C Code	No.	C Code	No.	C Code			
	10	5	C1	5	C1	5	C2			
	15	4	C1	4	C1	4	C3			
3.3	22	3	C2	2	C7	3	C4			
	33	1	C1	2	C7	3	C4			
	10	4	C2	4	C6	4	C4			
	15	3	C3	2	C7	3	C5			
5	22	3	C2	2	C7	3	C4			
	33	2	C2	2	C3	2	C4			
	47	2	C2	1	C7	2	C4			
	10	4	C5	3	C6	5	C9			
	15	3	C5	2	C7	4	C9			
	22	2	C5	2	C6	3	C8			
12	33	2	C5	1	C7	3	C8			
	47	2	C4	1	C6	2	C8			
	68	1	C5	1	C5	2	C7			
	100	1	C4	1	C5	1	C8			
			•		Throug	h Hole				
Output	Inductance	Sanvo O	S-CON SA	Sanyo MV-GX Series		Nichicon PL Series		Panasonic HFQ Series		
Voltage	(μH)	-	ries							
(V)		No.	C Code	No.	C Code	No.	C Code	No.	C Code	
	10	2	C5	2	C6	1	C8	2	C6	
	15	2	C5	2	C5	1	C7	2	C5	
3.3	22	1	C5	1	C10	1	C5	1	C7	
	33	1	C5	1	C10	1	C5	1	C7	
	10	2	C4	2	C5	1	C6	2	C5	
	15	1	C5	1	C10	1	C5	1	C7	
5	22	1	C5	1	C9	1	C5	1	C5	
	33	1	C4	1	C5	1	C4	1	C4	
	47	1	C4	1	C4	1	C2	2	C4	
	10	2	C7	1	C10	1	C14	2	C4	
	15	1	C8	1	C6	1	C17	1	C5	
	22	1	C7	1	C5	1	C13	1	C5	
12	33	1	C7	1	C4	1	C12	1	C4	
	47	1	C7	1	C3	1	C11	1	C3	
	4/	1	01		00				00	
	68	1	C6	1	C2	1	C10	1	C3	

Application Hints (Continued)

No. represents the number of identical capacitor types to be connected in parallel C Code indicates the Capacitor Reference number in Table 2 for identifying the specific component from the manufacturer.

Application Hints (Continued)

Output Voltage (V)									
	Inductance (µH)	AVX TPS Series		Sprague 594D Series		Kemet T	495 Series		
(•)		No.	C Code	No.	C Code	No.	C Code		
	10	3	C7	2	C10	3	C9		
3.3	15	*	*	3	C13	4	C12		
3.3	22	*	*	2	C13	3	C12		
	33	*	*	2	C13	3	C12		
	10	3	C4	2	C6	3	C9		
	15	4	C9	3	C12	4	C10		
5	22	*	*	3	C13	4	C12		
	33	*	*	2	C13	3	C12		
	47	*	*	1	C13	2	C12		
	10	4	C9	2	C10	4	C10		
	15	4	C8	2	C10	4	C10		
	22	4	C9	3	C12	4	C10		
12	33	*	*	3	C13	4	C12		
	47	*	*	2	C13	3	C12		
	68	*	*	2	C13	2	C12		
	100	*	*	1	C13	2	C12		
					Throug	gh Hole			
Output	Inductance	Sanyo O	S-CON SA	Sanyo	MV-GX	Nichier		Panaso	nic HFQ
Voltage (V)	(µH)	Se	eries	Se	ries	Nichicon	PL Series	Series	
(•)		No.	C Code	No.	C Code	No.	C Code	No.	C Code
	10	2	C9	2	C8	1	C18	1	C8
	15	*	*	2	C13	1	C25	1	C16
22		*	*	1	C14	1	C24	1	C16
3.3	22	*		1	014		024	1	010
3.3	22 33	*	*	1	C14	1	C24	1	C10
3.3					-				

C Code C8 C16 C16 C16

C8

C16

C13

C11

C8

C8

C8

C14

C13

C15

C11

TABLE 4. Input Capacitors for Fixed Output Voltage Application

15 C8 5 22 * * 2 C13

*

*

*

C10

C10

*

*

*

*

*

2

1

1

2

2

2

2

1

1

1

C14

C12

C8

C8

C8

C12

C14

C13

C11

1

1

1

1

1

1

1

1

1

1

1

C25

C25

C23

C19

C18

C18

C18

C24

C23

C21

C22

1

1

1

1

1

1

1

1

1

1

1

*

*

*

2

2

*

*

*

*

*

33

47

10

15

22

33

47

68

100

* Check voltage rating of capacitors to be greater than application input voltage.

No. represents the number of identical capacitor types to be connected in parallel

C Code indicates the Capacitor Reference number in Table 2 for identifying the specific component from the manufacturer.

12

		Т	ABLE 5. Scho	ottky Diode		
Reverse	Surfa	ace Mount	Through Hole			
Voltage	3A	5A or More	3A	5A or		
(V)				More		
20V	SK32		1N5820			
			SR302			
30V	SK33	MBRD835L	1N5821			
	30WQ03F		31DQ03			
40V	SK34	MBRD1545CT	1N5822	1N5825		
	30BQ040	6TQ045S	MBR340	MBR745		
	30WQ04F		31DQ04	80SQ045		
	MBRS340		SR403	6TQ045		
	MBRD340					
50V or	SK35		MBR350			
More	30WQ05F		31DQ05			
			SR305			

Diode Manufacturer Contact Numbers

International Rectifier	Phone	(310) 322-3331
	FAX	(310) 322-3332
Motorola	Phone	(800) 521-6274
	FAX	(602) 244-6609
General	Phone	(516) 847-3000
Semiconductor		
	FAX	(516) 847-3236
Diodes, Inc.	Phone	(805) 446-4800
	FAX	(805) 446-4850

Application Hints (Continued)

			Surface Mount								
Output Voltage (V)	Inductance (µH)	AVX TPS Series		Spragu	ue 594D ries	Kemet T495 Series					
(-)	(17	No.	C Code	No.	C Code	No.	C Code				
	33*	7	C1	6	C2	7	C3				
1.21 to 2.50	47*	5	C1	4	C2	5	C3				
	33*	4	C1	3	C2	4	C3				
2.5 to 3.75	47*	3	C1	2	C2	3	C3				
	22	4	C1	3	C2	4	C3				
3.75 to 5	33	3	C1	2	C2	3	C3				
	47	2	C1	2	C2	2	C3				
	22	3	C2	3	C3	3	C4				
	33	2	C2	2	C3	2	C4				
5 to 6.25	47	2	C2	2	C3	2	C4				
	68	1	C2	1	C3	1	C4				
	22	3	C2	1	C4	3	C4				
	33	2	C2	1	C3	2	C4				
6.25 to 7.5	47	1	C3	1	C4	1	C6				
	68	1	C2	1	C3	1	C4				
	33	2	C5	1	C6	2	C8				
	47	1	C5	1	C6	2	C8				
7.5 to 10	68	1	C5	1	C6	1	C8				
	100	1	C4	1	C5	1	C8				
	33	1	C5	1	C6	2	C8				
	47	1	C5	1	C6	2	C8				
10 to 12.5	68	1	C5	1	C6	1	C8				
	100	1	C5	1	C6	1	C8				
	33	1	C6	1	C8	1	C8				
	47	1	C6	1	C8	1	C8				
12.5 to 15	68	1	C6	1	C8	1	C8				
	100	1	C6	1	C8	1	C8				
	33	1	C8	1	C10	2	C10				
	47	1	C8	1	C9	2	C10				
15 to 20	68	1	C8	1	C9	2	C10				
	100	1	C8	1	C9	1	C10				
	33	2	C9	2	C11	2	C11				
00.44.00	47	1	C10	1	C12	1	C11				
20 to 30	68	1	C9	1	C12	1	C11				
	100	1	C9	1	C12	1	C11				
	10			4	C13	8	C12				
	15			3	C13	5	C12				
20.4- 07	22	No Value	s Available	2	C13	4	C12				
30 to 37	33			1	C13	3	C12				
	47			1	C13	2	C12				
	68		-	1	C13	2	C12				

TABLE 6. Output Capacitors for Adjustable Output Voltage Applications

Output Capacitors for Adjustable Output Voltage Applications (continued)

		Through Hole								
Output Voltage (V)	Inductance (µH)	Sanyo OS-CON SA Series		Sanyo MV-GX Series		Nichicon	PL Series	Panasonic HFQ Series		
		No.	C Code	No.	C Code	No.	C Code	No.	C Code	
1.21 to 2.50	33*	2	C3	5	C1	5	C3	3	С	
1.21 to 2.50	47*	2	C2	4	C1	3	C3	2	C5	
0 E to 0 7E	33*	1	C3	3	C1	3	C1	2	C5	
2.5 to 3.75	47*	1	C2	2	C1	2	C3	1	C5	
	22	1	C3	3	C1	3	C1	2	C5	
3.75 to 5	33	1	C2	2	C1	2	C1	1	C5	
	47	1	C2	2	C1	1	C3	1	C5	
	22	1	C5	2	C6	2	C3	2	C5	
E (= 0.0E	33	1	C4	1	C6	2	C1	1	C5	
5 to 6.25	47	1	C4	1	C6	1	C3	1	C5	
	68	1	C4	1	C6	1	C1	1	C5	
	22	1	C5	1	C6	2	C1	1	C5	
	33	1	C4	1	C6	1	C3	1	C5	
6.25 to 7.5	47	1	C4	1	C6	1	C1	1	C5	
	68	1	C4	1	C2	1	C1	1	C5	
	33	1	C7	1	C6	1	C14	1	C5	
	47	1	C7	1	C6	1	C14	1	C5	
7.5 to 10	68	1	C7	1	C2	1	C14	1	C2	
	100	1	C7	1	C2	1	C14	1	C2	
	33	1	C7	1	C6	1	C14	1	C5	
	47	1	C7	1	C2	1	C14	1	C5	
10 to 12.5	68	1	C7	1	C2	1	C9	1	C2	
	100	1	C7	1	C2	1	C9	1	C2	
	33	1	C9	1	C10	1	C15	1	C2	
	47	1	C9	1	C10	1	C15	1	C2	
12.5 to 15	68	1	C9	1	C10	1	C15	1	C2	
	100	1	C9	1	C10	1	C15	1	C2	
	33	1	C10	1	C7	1	C15	1	C2	
	47	1	C10	1	C7	1	C15	1	C2	
15 to 20	68	1	C10	1	C7	1	C15	1	C2	
	100	1	C10	1	C7	1	C15	1	C2	
	33	L	1	1	C7	1	C16	1	C2	
	47	No V	/alues	1	C7	1	C16	1	C2	
20 to 30	68		ilable	1	C7	1	C16	1	C2	
	100			1	C7	1	C16	1	C2	
	10			1	C12	1	C20	1	C10	
	15			1	C11	1	C20	1	C11	
	22	No V	/alues	1	C11	1	C20	1	C10	
30 to 37	33		ilable	1	C11	1	C20	1	C10	
	47	7.00		1	C11	1	C20	1	C10	
	68			1	C11	1	C20	1	C10	

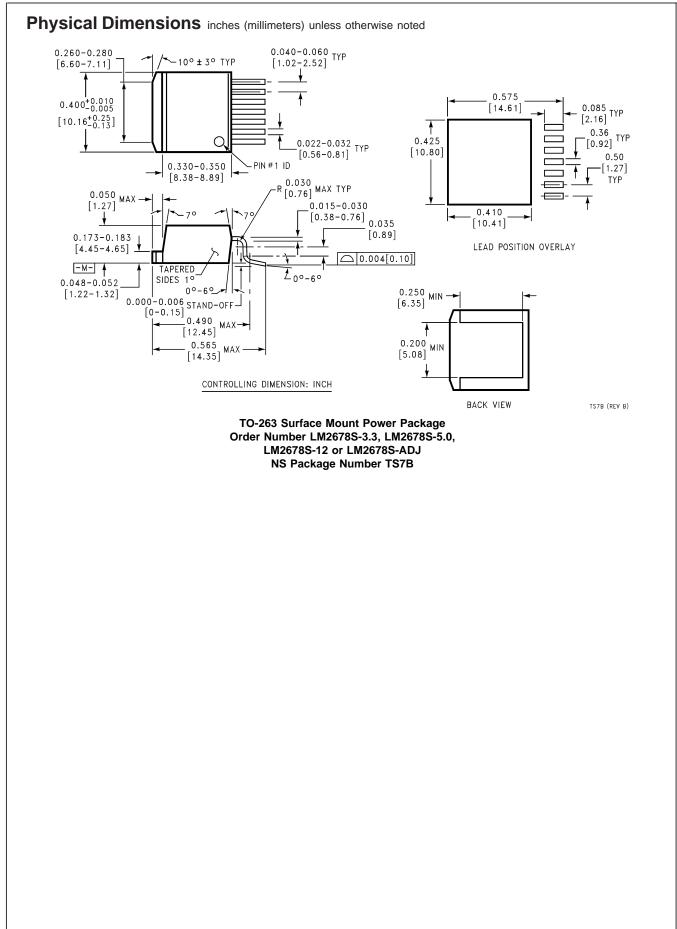
 * Set to a higher value for a practical design solution. See Applications Hints section

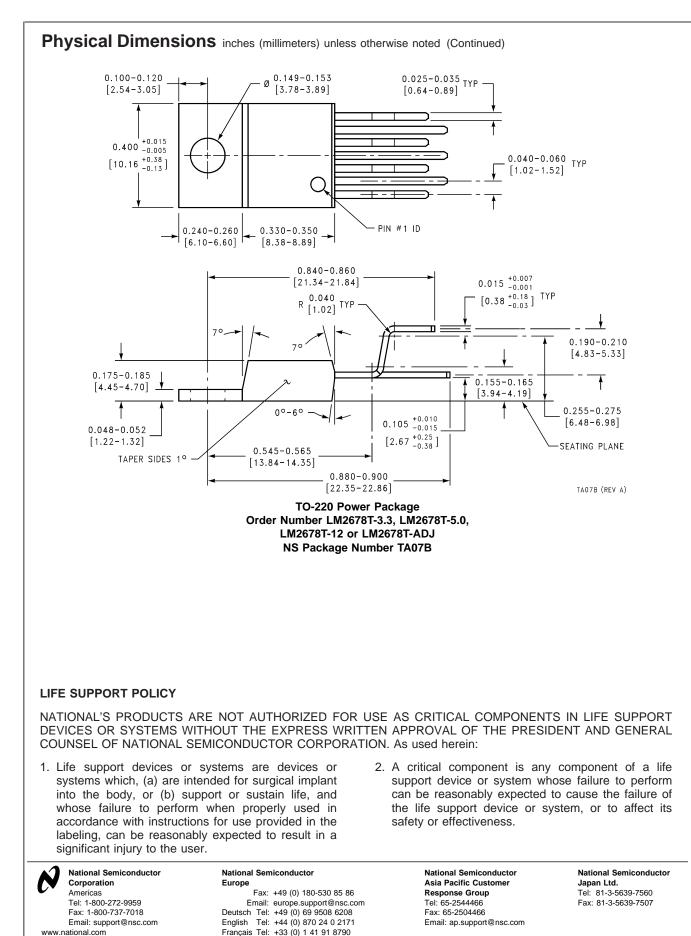
No. represents the number of identical capacitor types to be connected in parallel

C Code indicates the Capacitor Reference number in Table 2 for identifying the specific component from the manufacturer.

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