

## STD95NH02L

## N-CHANNEL 24V - 0.0039Ω - 80A DPAK ULTRA LOW GATE CHARGE STripFET™ MOSFET

**Table 1: General Features** 

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	I <sub>D</sub>
STD95NH02L	24 V	< 0.005Ω	80(*) A

- TYPICAL  $R_{DS}(on) = 0.0039\Omega$  @ 10 V
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED

#### DESCRIPTION

The **STD95NH02L** is based on the latest generation of ST's proprietary STripFET™ technology. An innovative layout enables the device to also exhibit extremely low gate charge for the most demanding requirements in high-frequency DC-DC converters. It's therefore ideal for high-density converters in Telecom and Computer applications.

#### **APPLICATIONS**

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTERS

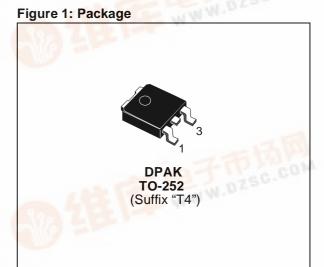
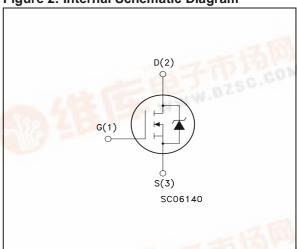


Figure 2: Internal Schematic Diagram



**Table 2: Order Codes** 

PART NUMBER	MARKING	PACKAGE	PACKAGING
STD95NH02LT4	STD95NH02LT4 D95NH02L		TAPE & REEL



#### STD95NH02L

**Table 3: Absolute Maximum ratings** 

Symbol	Parameter	Value	Unit	
V <sub>spike</sub> (1)	Drain-source Voltage Rating	30	V	
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	24	V	
$V_{DGR}$	Drain-gate Voltage ( $R_{GS} = 20 \text{ k}\Omega$ )	24	V	
V <sub>GS</sub>	Gate- source Voltage	± 20	V	
I <sub>D</sub> (*)	Drain Current (continuous) at T <sub>C</sub> = 25°C	80	А	
I <sub>D</sub>	Drain Current (continuous) at T <sub>C</sub> = 100°C	68	A	
I <sub>DM</sub> (2)	Drain Current (pulsed)	320	А	
P <sub>TOT</sub>	Total Dissipation at T <sub>C</sub> = 25°C	100	W	
	Derating Factor	0.67	W/°C	
E <sub>AS</sub> (3)	Single Pulse Avalanche Energy	600	mJ	
T <sub>stg</sub>	Storage Temperature	-55 to 175	°C	
Tj	Max. Operating Junction Temperature	-55 to 175 °C		

<sup>(1)</sup> Garanted when external  $R_g = 4.7~\Omega$  and  $t_f < t_f$  max. (2) Pulse width limited by safe operating area. (3) Starting  $T_j = 25^{\circ}\text{C}$ ,  $I_D = 40\text{A}$ ,  $V_{DD} = 22\text{V}$  (\*) Value limited by wires

**Table 4: Thermal Data** 

Rthj-case	Thermal Resistance Junction-case Max	1.5	°C/W
Rthj-amb	Thermal Resistance Junction-ambient Max	100	°C/W
Tı	Maximum Lead Temperature For Soldering Purpose	275	°C

#### **ELECTRICAL CHARACTERISTICS** (T<sub>CASE</sub> =25°C UNLESS OTHERWISE SPECIFIED) Table 5: On/Off

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	$I_D = 250 \mu\text{A},  V_{GS} = 0$	24			V
I <sub>DSS</sub>	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	$V_{DS}$ = Max Rating $V_{DS}$ = Max Rating, $T_{C}$ = 125 °C			1 10	μA μA
I <sub>GSS</sub>	Gate-body Leakage Current (V <sub>DS</sub> = 0)	V <sub>GS</sub> = ± 20V			±100	nA
V <sub>GS(th)</sub>	Gate Threshold Voltage	$V_{DS} = V_{GS}$ , $I_D = 250\mu A$	1			V
R <sub>DS(on)</sub>	Static Drain-source On Resistance	V <sub>GS</sub> = 10 V, I <sub>D</sub> = 40 A V <sub>GS</sub> = 5 V, I <sub>D</sub> = 40 A		0.0039 0.0055	0.005 0.009	Ω Ω

## **ELECTRICAL CHARACTERISTICS (CONTINUED)**

**Table 6: Dynamic** 

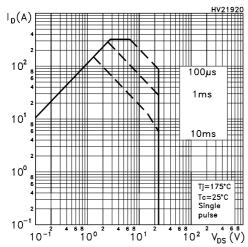
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
g <sub>fs</sub> (4)	Forward Transconductance	V <sub>DS</sub> = 10 V <sub>,</sub> I <sub>D</sub> = 10 A		30		S
C <sub>iss</sub> C <sub>oss</sub> C <sub>rss</sub>	Input Capacitance Output Capacitance Reverse Transfer Capacitance	$V_{DS} = 15V$ , $f = 1$ MHz, $V_{GS} = 0$		2070 990 90		pF pF pF
t <sub>d(on)</sub> t <sub>r</sub> t <sub>d(off)</sub> t <sub>f</sub>	Turn-on Delay Time Rise Time Turn-off Delay Time Fall Time	$V_{DD} = 12 \text{ V}, I_D = 40 \text{ A}, \\ R_{G} = 4.7 \Omega V_{GS} = 10 \text{ V} \\ \text{(see Figure 16)}$		20 110 47 20		ns ns ns ns
Q <sub>g</sub> Q <sub>gs</sub> Q <sub>gd</sub>	Total Gate Charge Gate-Source Charge Gate-Drain Charge	$V_{DD} = 12 \text{ V, } I_{D} = 80 \text{ A,}$ $V_{GS} = 5 \text{ V}$ (see Figure 19)		17 7.6 6.8		nC nC nC
Q <sub>oss</sub> (5)	Output Charge	V <sub>DS</sub> = 19 V, V <sub>GS</sub> = 0 V		22.6		nC
Q <sub>gls</sub> (6)	Third-Quadrant Gate Charge	V <sub>DS</sub> < 0 V, V <sub>GS</sub> = 5 V		15		nC
R <sub>G</sub>	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.8		Ω

#### **Table 7: Source Drain Diode**

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I <sub>SD</sub>	Source-drain Current				80	Α
I <sub>SDM</sub>	Source-drain Current (pulsed)				320	Α
V <sub>SD</sub> (4)	Forward On Voltage	I <sub>SD</sub> = 40A, V <sub>GS</sub> = 0			1.3	V
t <sub>rr</sub> Q <sub>rr</sub> I <sub>RRM</sub>	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 80A$ , di/dt = 100 A/ $\mu$ s, $V_{DD} = 20$ V, $T_j = 150$ °C (see Figure 16)		42 50.4 2.4		ns nC A

<sup>(4).</sup> Pulsed: Pulse duration = 300  $\mu$ s, duty cycle 1.5 %. (5).  $Q_{\rm OSS} = C_{\rm OSS}^* \Delta$   $V_{\rm in}$ .  $C_{\rm OSS} = C_{\rm gd} + C_{\rm ds}$ . See Appendix A. (6). Gate charge for Syncronous Operation.

Figure 3: Safe Operating Area



**Figure 4: Output Characteristics** 

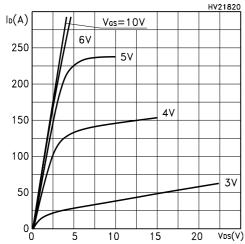


Figure 5: Transconductance

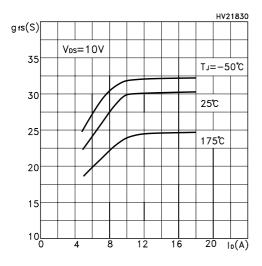
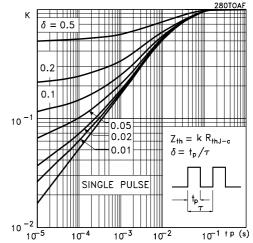


Figure 6: Thermal Impedance



**Figure 7: Transfer Characteristics** 

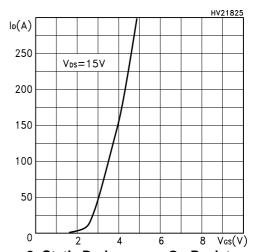


Figure 8: Static Drain-source On Resistance

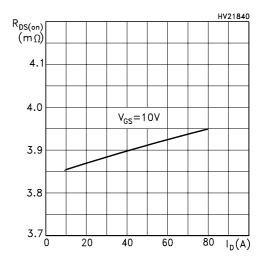


Figure 9: Gate Charge vs Gate-source Voltage

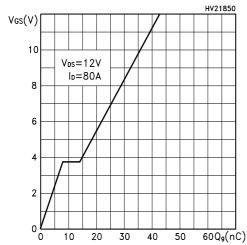


Figure 10: Normalized Gate Thereshold Voltage vs Temperature

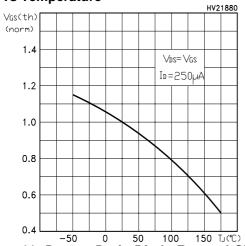


Figure 11: Dource-Drain Diode Forward Characteristics

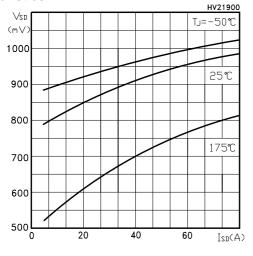


Figure 12: Capacitance Variations

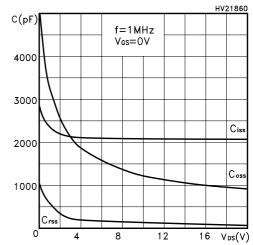


Figure 13: Normalized On Resistance vs Temperature

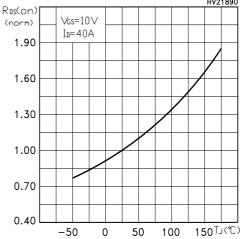


Figure 14: Normalized Breakdown Voltage vs Temperature

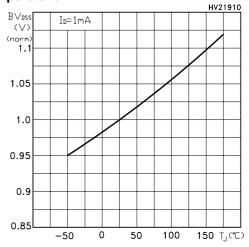


Figure 15: Unclamped Inductive Load Test Circuit

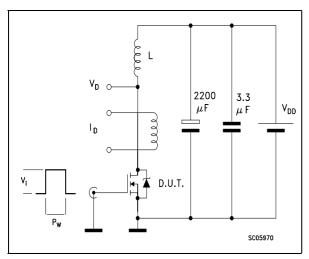


Figure 16: Switching Times Test Circuit For Resistive Load

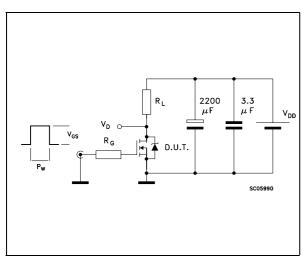


Figure 17: Test Circuit For Inductive Load Switching and Diode Recovery Times

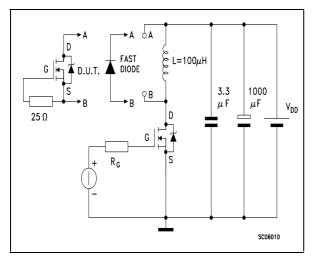


Figure 18: Unclamped Inductive Wafeform

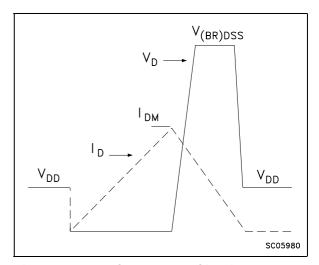
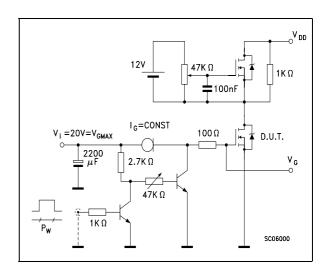
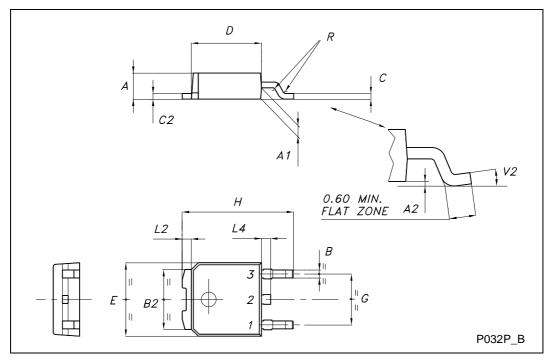


Figure 19: Gate Charge Test Circuit



## **TO-252 (DPAK) MECHANICAL DATA**

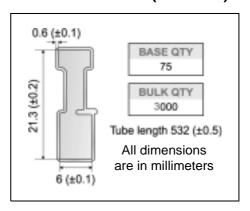
DIM.		mm		inch			
Dilvi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
А	2.20		2.40	0.087		0.094	
A1	0.90		1.10	0.035		0.043	
A2	0.03		0.23	0.001		0.009	
В	0.64		0.90	0.025		0.035	
B2	5.20		5.40	0.204		0.213	
С	0.45		0.60	0.018		0.024	
C2	0.48		0.60	0.019		0.024	
D	6.00		6.20	0.236		0.244	
E	6.40		6.60	0.252		0.260	
G	4.40		4.60	0.173		0.181	
Н	9.35		10.10	0.368		0.398	
L2		0.8			0.031		
L4	0.60		1.00	0.024		0.039	
V2	0°		8°	0°		0°	



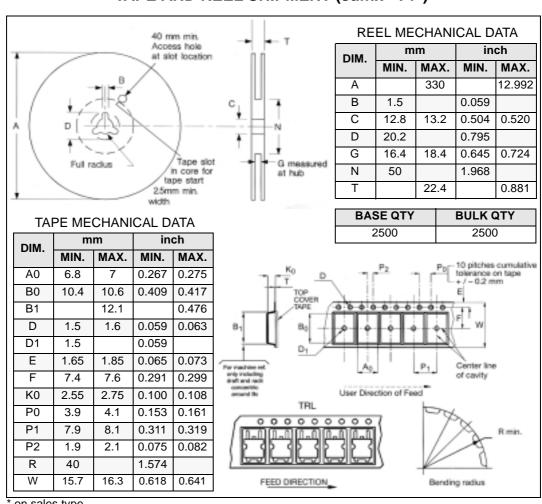
#### **DPAK FOOTPRINT**

# 3.0 2.3 6.7 2.3 1.6 All dimensions are in millimeters

#### **TUBE SHIPMENT (no suffix)\***



#### TAPE AND REEL SHIPMENT (suffix "T4")\*



on sales type

### **Appendix A: Buck Converter Power Losses Estimation**

#### **DESCRIPTION**

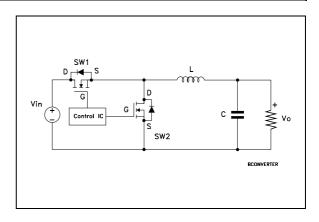
The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is removed to allow for a safer working junction temperature.

#### The low side (SW2) device requires:

- Very low RDS(on) to reduce conduction losses
- Small Q<sub>qls</sub> to reduce the gate charge losses
- Small  $C_{oss}^{\text{-}}$  to reduce losses due to output capaci tance
- Small Q<sub>rr</sub> to reduce losses on SW1 during its turn-on
- The C<sub>gd</sub>/C<sub>gs</sub> ratio lower than V<sub>th</sub>/V<sub>GG</sub> ratio especially with low drain to source voltage to avoid the cross conduction phenomenon

#### The high side (SW1) device requires:

- Small  $\rm R_g$  and  $\rm L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate
- Small  $\mathbf{Q}_{\mathbf{g}}$  to have a faster commutation and to reduce gate charge losses
- Low R<sub>DS(on)</sub> to reduce the conduction losses



		High Side Switch (SW1)	Low Side Switch (SW2)
Pconduction		$R_{ ext{DS(on)SW}}^{1} I_{ ext{L}}^{2} * \delta$	$R_{\rm DS(on)SW2}*I_{\rm L}^2*(1-\delta)$
Pswitchi	ng	$V_{\rm in} * (Q_{\rm gsth(SW1)} + Q_{\rm gd(SW1)}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P <sub>diode</sub>	Recovery	Not Applicable	<sup>1</sup> V <sub>in</sub> *Q <sub>rr(SW2)</sub> *f
	Conduction	Not Applicable	$V_{f(SW2)} * I_L * t_{deadtime} * f$
P <sub>gate(Q</sub>	)	$Q_{g(SWI)} * V_{gg} * f$	$Q_{\rm gls(SW2)}^{}^{}^{}^{}^{}_{}^{}^{}^{}_{}^{}^{}$
P <sub>Qoss</sub>		$\frac{V_{in} * Q_{oss(SW1)} * f}{2}$	$\frac{V_{in} * Q_{oss(SW2)} * f}{2}$

Parameter	Meaning		
δ	Duty-Cycle		
Q <sub>gsth</sub>	Post Threshold Gate Charge		
Q <sub>gls</sub>	Third Quadrant Gate Charge		
Pconduction	On State Losses		
Pswitching	On-off Transition Losses		
Pdiode	Conduction and Reverse Recovery Diode Losses		
Pdiode	Gate Drive Losses		
P <sub>Qoss</sub>	Output Capacitance Losses		



#### STD95NH02L

**Table 8: Revision History** 

Date	Revision	Description of Changes
27-Aug-2004	1	First Release.
10-Sep-2004	2	Values changed in table 7

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