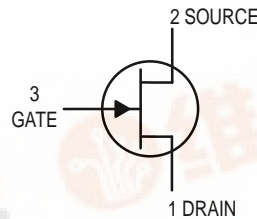
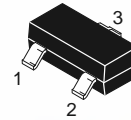


JFET VHF/UHF Amplifier Transistor

N-Channel



MMBFJ309LT1
MMBFJ310LT1



CASE 318-08, STYLE 10
SOT-23 (TO-236AB)

MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	V_{DS}	25	Vdc
Gate-Source Voltage	V_{GS}	25	Vdc
Gate Current	I_G	10	mAdc

THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Total Device Dissipation FR-5 Board ⁽¹⁾ $T_A = 25^\circ\text{C}$ Derate above 25°C	P_D	225 1.8	mW mW/ $^\circ\text{C}$
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	556	$^\circ\text{C}/\text{W}$
Junction and Storage Temperature	T_J, T_{stg}	-55 to +150	$^\circ\text{C}$

DEVICE MARKING

MMBFJ309LT1 = 6U; MMBFJ310LT1 = 6T

ELECTRICAL CHARACTERISTICS ($T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
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OFF CHARACTERISTICS

Gate-Source Breakdown Voltage ($I_G = -1.0 \mu\text{Adc}$, $V_{DS} = 0$)	$V_{(BR)GSS}$	-25	—	—	Vdc
Gate Reverse Current ($V_{GS} = -15 \text{ Vdc}$) ($V_{GS} = -15 \text{ Vdc}$, $T_A = 125^\circ\text{C}$)	I_{GSS}	— —	— —	-1.0 -1.0	nAdc μAdc
Gate Source Cutoff Voltage ($V_{DS} = 10 \text{ Vdc}$, $I_D = 1.0 \text{ nAdc}$)	MMBFJ309 MMBFJ310 $V_{GS(off)}$	-1.0 -2.0	— —	-4.0 -6.5	Vdc

ON CHARACTERISTICS

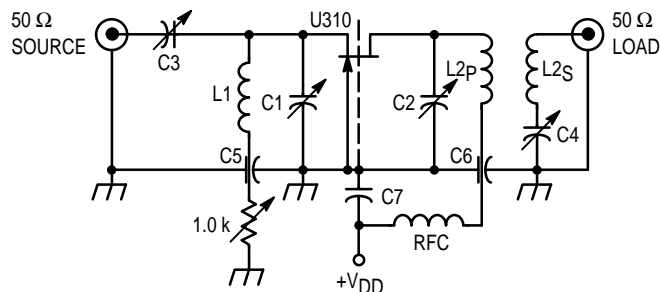
Zero-Gate-Voltage Drain Current ($V_{DS} = 10 \text{ Vdc}$, $V_{GS} = 0$)	MMBFJ309 MMBFJ310 I_{DSS}	12 24	— —	30 60	mAdc
Gate-Source Forward Voltage ($I_G = 1.0 \text{ mAdc}$, $V_{DS} = 0$)	$V_{GS(f)}$	—	—	1.0	Vdc

SMALL-SIGNAL CHARACTERISTICS

Forward Transfer Admittance ($V_{DS} = 10 \text{ Vdc}$, $I_D = 10 \text{ mAdc}$, $f = 1.0 \text{ kHz}$)	$ Y_{fs} $	8.0	—	18	mmhos
Output Admittance ($V_{DS} = 10 \text{ Vdc}$, $I_D = 10 \text{ mAdc}$, $f = 1.0 \text{ kHz}$)	$ Y_{os} $	—	—	250	μmhos
Input Capacitance ($V_{GS} = -10 \text{ Vdc}$, $V_{DS} = 0 \text{ Vdc}$, $f = 1.0 \text{ MHz}$)	C_{iss}	—	—	5.0	pF
Reverse Transfer Capacitance ($V_{GS} = -10 \text{ Vdc}$, $V_{DS} = 0 \text{ Vdc}$, $f = 1.0 \text{ MHz}$)	C_{rss}	—	—	2.5	pF
Equivalent Short-Circuit Input Noise Voltage ($V_{DS} = 10 \text{ Vdc}$, $I_D = 10 \text{ mAdc}$, $f = 100 \text{ Hz}$)	e_n	—	10	—	$\text{nV}/\sqrt{\text{Hz}}$

1. FR-5 = $1.0 \times 0.75 \times 0.062 \text{ in.}$

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$C1 = C2 = 0.8 - 10$ pF, JFD #MVM010W.
 $C3 = C4 = 8.35$ pF Erie #539-002D.
 $C5 = C6 = 5000$ pF Erie (2443-000).
 $C7 = 1000$ pF, Allen Bradley #FA5C.
 $RFC = 0.33$ μ H Miller #9230-30.
 $L1 =$ One Turn #16 Cu, 1/4" I.D. (Air Core).
 $L2P =$ One Turn #16 Cu, 1/4" I.D. (Air Core).
 $L2S =$ One Turn #16 Cu, 1/4" I.D. (Air Core).

Figure 1. 450 MHz Common-Gate Amplifier Test Circuit

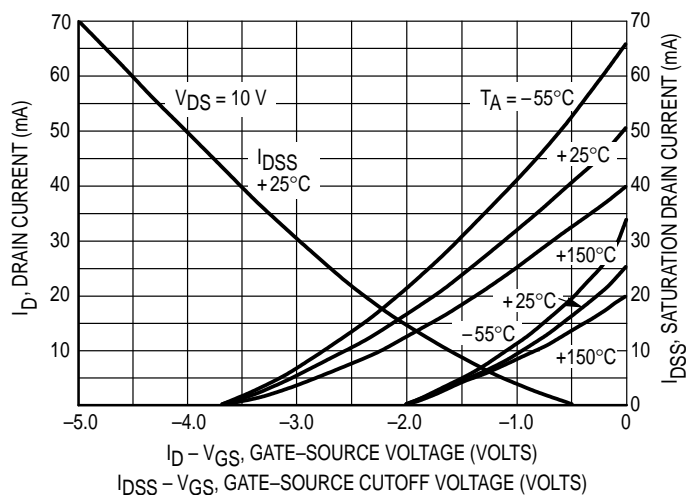


Figure 2. Drain Current and Transfer Characteristics versus Gate-Source Voltage

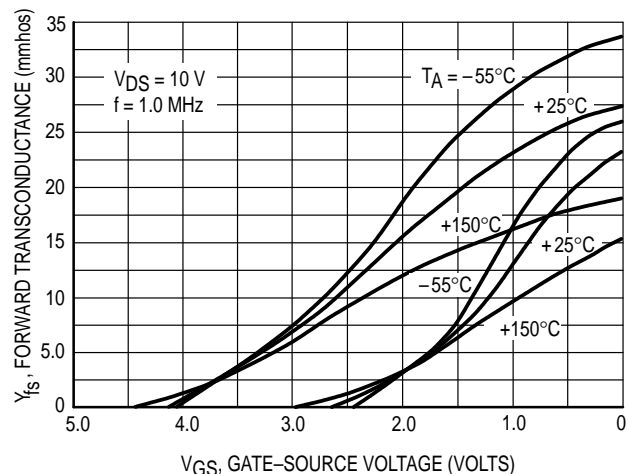


Figure 3. Forward Transconductance versus Gate-Source Voltage

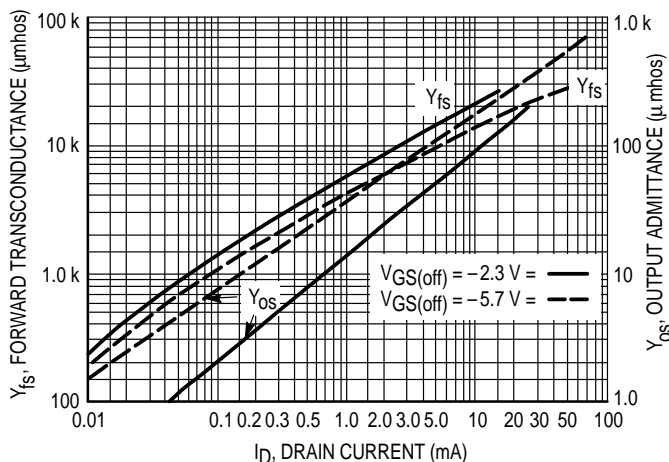


Figure 4. Common-Source Output Admittance and Forward Transconductance versus Drain Current

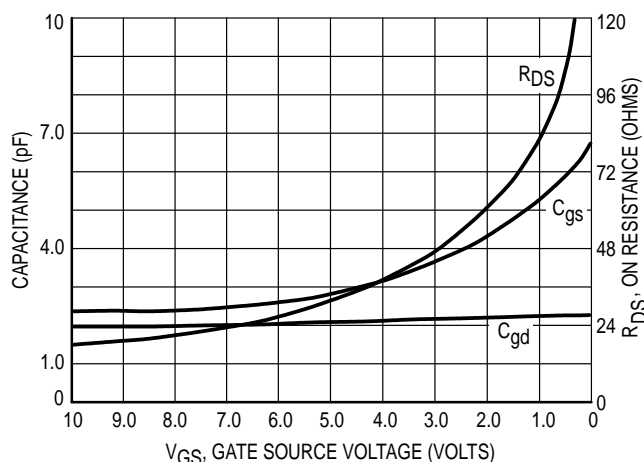


Figure 5. On Resistance and Junction Capacitance versus Gate-Source Voltage

MMBFJ309LT1 MMBFJ310LT1

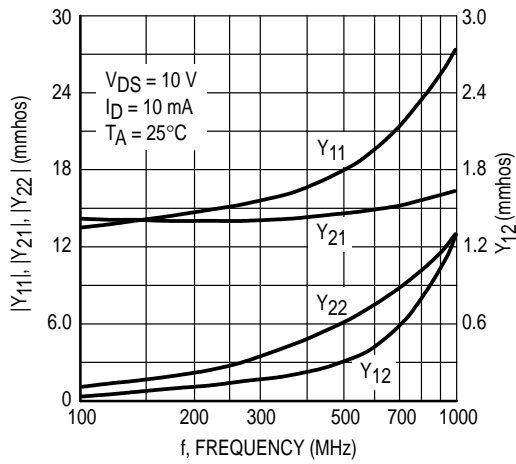


Figure 6. Common-Gate Y Parameter Magnitude versus Frequency

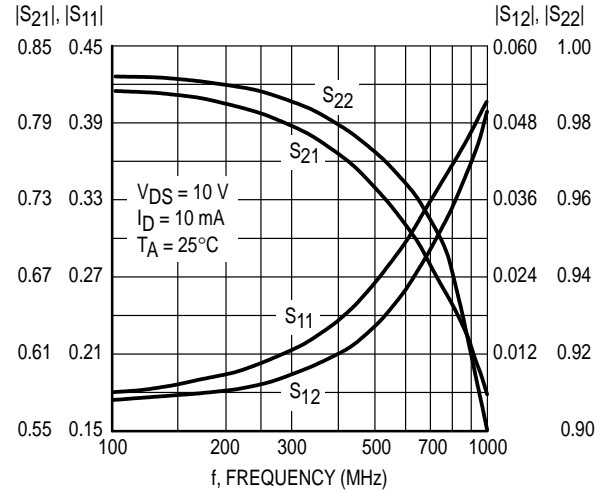


Figure 7. Common-Gate S Parameter Magnitude versus Frequency

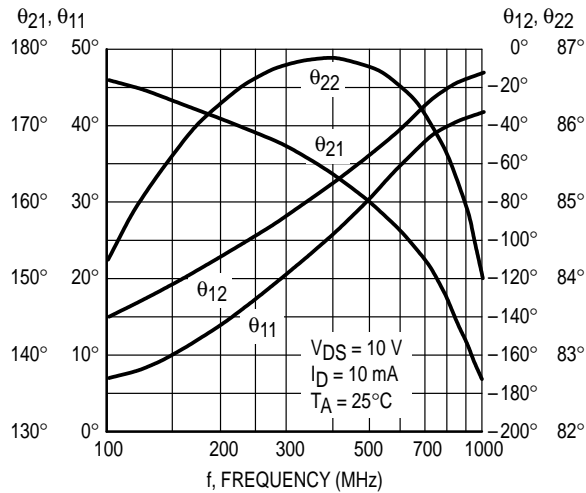


Figure 8. Common-Gate Y Parameter Phase-Angle versus Frequency

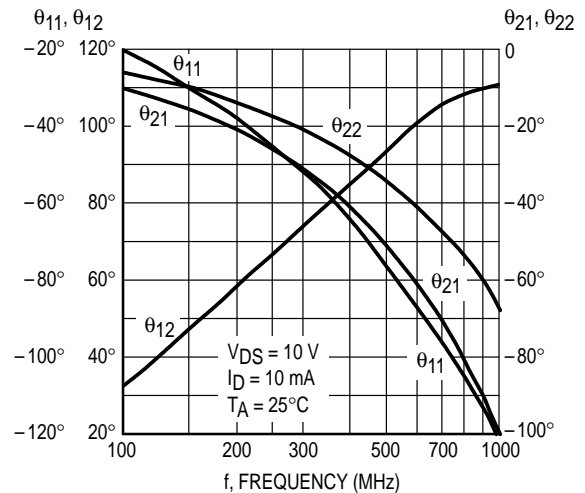


Figure 9. S Parameter Phase-Angle versus Frequency

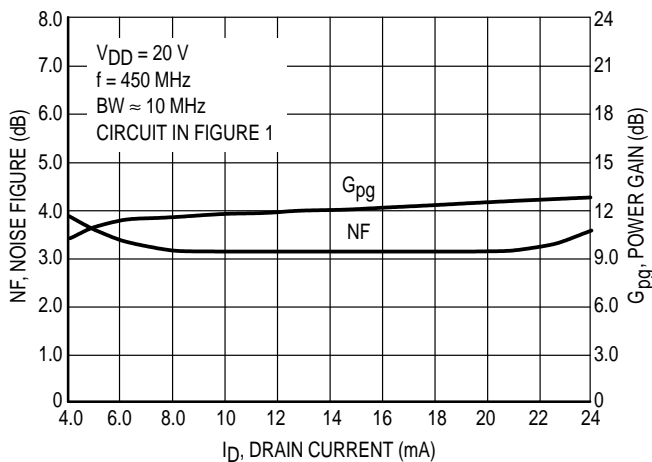


Figure 10. Noise Figure and Power Gain versus Drain Current

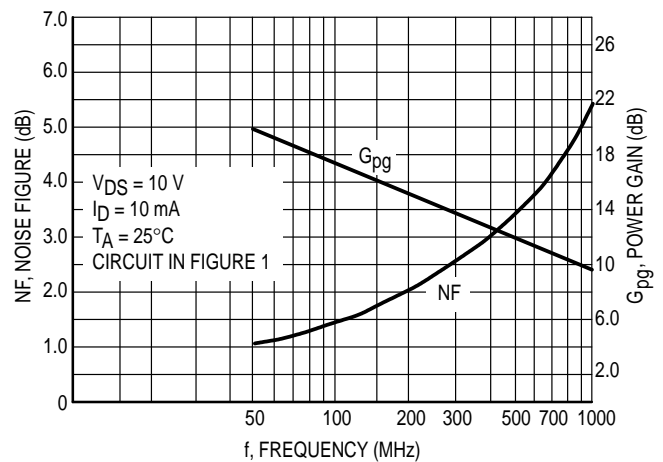
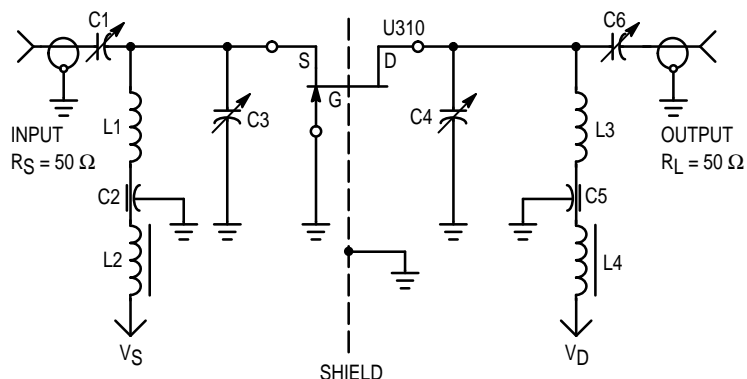


Figure 11. Noise Figure and Power Gain versus Frequency

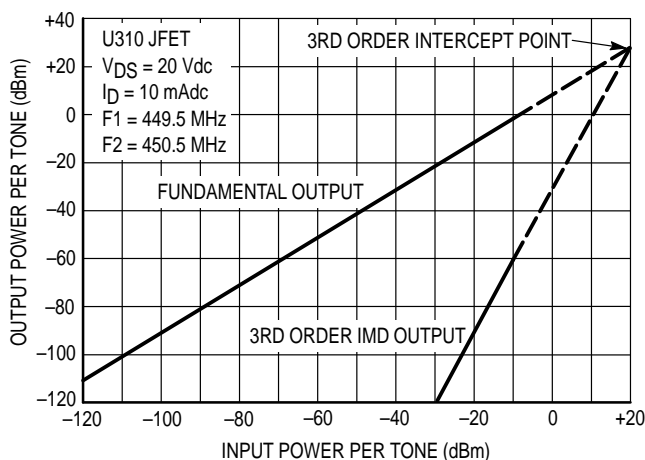
MMBFJ309LT1 MMBFJ310LT1



B_W (3 dB) – 36.5 MHz
 I_D – 10 mAdc
 V_{DS} – 20 Vdc
 Device case grounded
 IM test tones – $f_1 = 449.5$ MHz, $f_2 = 450.5$ MHz
 $C_1 = 1\text{--}10$ pF Johanson Air variable trimmer.
 $C_2, C_5 = 100$ pF feed thru button capacitor.
 $C_3, C_4, C_6 = 0.5\text{--}6$ pF Johanson Air variable trimmer.
 $L_1 = 1/8'' \times 1/32'' \times 1\text{--}5/8''$ copper bar.
 $L_2, L_4 =$ Ferroxcube Vx200 choke.
 $L_3 = 1/8'' \times 1/32'' \times 1\text{--}7/8''$ copper bar.

Figure 12. 450 MHz IMD Evaluation Amplifier

Amplifier power gain and IMD products are a function of the load impedance. For the amplifier design shown above with C_4 and C_6 adjusted to reflect a load to the drain resulting in a nominal power gain of 9 dB, the 3rd order intercept point (IP) value is 29 dBm. Adjusting C_4, C_6 to provide larger load values will result in higher gain, smaller bandwidth and lower IP values. For example, a nominal gain of 13 dB can be achieved with an intercept point of 19 dBm.



Example of intercept point plot use:

Assume two in-band signals of -20 dBm at the amplifier input. They will result in a 3rd order IMD signal at the output of -90 dBm. Also, each signal level at the output will be -11 dBm, showing an amplifier gain of 9.0 dB and an intermodulation ratio (IMR) capability of 79 dB. The gain and IMR values apply only for signal levels below comparison.

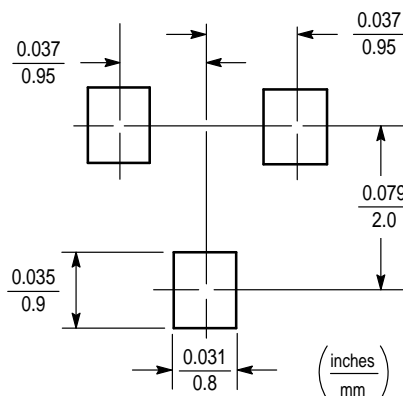
Figure 13. Two Tone 3rd Order Intercept Point

INFORMATION FOR USING THE SOT-23 SURFACE MOUNT PACKAGE

MINIMUM RECOMMENDED FOOTPRINT FOR SURFACE MOUNTED APPLICATIONS

Surface mount board layout is a critical portion of the total design. The footprint for the semiconductor packages must be the correct size to insure proper solder connection

interface between the board and the package. With the correct pad geometry, the packages will self align when subjected to a solder reflow process.



SOT-23

SOT-23 POWER DISSIPATION

The power dissipation of the SOT-23 is a function of the pad size. This can vary from the minimum pad size for soldering to a pad size given for maximum power dissipation. Power dissipation for a surface mount device is determined by $T_{J(max)}$, the maximum rated junction temperature of the die, $R_{\theta JA}$, the thermal resistance from the device junction to ambient, and the operating temperature, T_A . Using the values provided on the data sheet for the SOT-23 package, P_D can be calculated as follows:

$$P_D = \frac{T_{J(max)} - T_A}{R_{\theta JA}}$$

The values for the equation are found in the maximum ratings table on the data sheet. Substituting these values into the equation for an ambient temperature T_A of 25°C, one can calculate the power dissipation of the device which in this case is 225 milliwatts.

$$P_D = \frac{150^\circ\text{C} - 25^\circ\text{C}}{556^\circ\text{C/W}} = 225 \text{ milliwatts}$$

The 556°C/W for the SOT-23 package assumes the use of the recommended footprint on a glass epoxy printed circuit board to achieve a power dissipation of 225 milliwatts. There are other alternatives to achieving higher power dissipation from the SOT-23 package. Another alternative would be to use a ceramic substrate or an aluminum core board such as Thermal Clad™. Using a board material such as Thermal Clad, an aluminum core board, the power dissipation can be doubled using the same footprint.

SOLDERING PRECAUTIONS

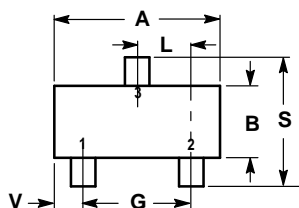
The melting temperature of solder is higher than the rated temperature of the device. When the entire device is heated to a high temperature, failure to complete soldering within a short time could result in device failure. Therefore, the following items should always be observed in order to minimize the thermal stress to which the devices are subjected.

- Always preheat the device.
- The delta temperature between the preheat and soldering should be 100°C or less.*
- When preheating and soldering, the temperature of the leads and the case must not exceed the maximum temperature ratings as shown on the data sheet. When using infrared heating with the reflow soldering method, the difference shall be a maximum of 10°C.
- The soldering temperature and time shall not exceed 260°C for more than 10 seconds.
- When shifting from preheating to soldering, the maximum temperature gradient shall be 5°C or less.
- After soldering has been completed, the device should be allowed to cool naturally for at least three minutes. Gradual cooling should be used as the use of forced cooling will increase the temperature gradient and result in latent failure due to mechanical stress.
- Mechanical stress or shock should not be applied during cooling.

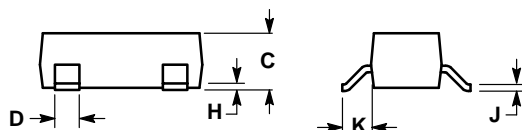
* Soldering a device without preheating can cause excessive thermal shock and stress which can result in damage to the device.

MMBFJ309LT1 MMBFJ310LT1

PACKAGE DIMENSIONS



STYLE 10:
PIN 1. DRAIN
2. SOURCE
3. GATE




NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH THICKNESS. MINIMUM LEAD THICKNESS IS THE MINIMUM THICKNESS OF BASE MATERIAL.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.1102	0.1197	2.80	3.04
B	0.0472	0.0551	1.20	1.40
C	0.0350	0.0440	0.89	1.11
D	0.0150	0.0200	0.37	0.50
G	0.0701	0.0807	1.78	2.04
H	0.0005	0.0040	0.013	0.100
J	0.0034	0.0070	0.085	0.177
K	0.0180	0.0236	0.45	0.60
L	0.0350	0.0401	0.89	1.02
S	0.0830	0.0984	2.10	2.50
V	0.0177	0.0236	0.45	0.60

**CASE 318-08
ISSUE AE
SOT-23 (TO-236AB)**

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