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DISCRETE SEMICONDUCTORS

DATA SHEET

BFQ270 NPN 6 GHz wideband transistor

Product specification

File under Discrete Semiconductors, SC14

September 1995

NPN 6 GHz wideband transistor**BFQ270****FEATURES**

- High power gain
- Emitter-ballasting resistors for good thermal stability
- Gold metallization ensures excellent reliability.

PINNING

PIN	DESCRIPTION
1	collector
2	emitter
3	base
4	emitter

DESCRIPTION

Silicon NPN transistor mounted in a 4-lead dual-emitter SOT172A1. envelope with a ceramic cap. All leads are isolated from the mounting base.

It is primarily intended for use in MATV and CATV amplifiers.

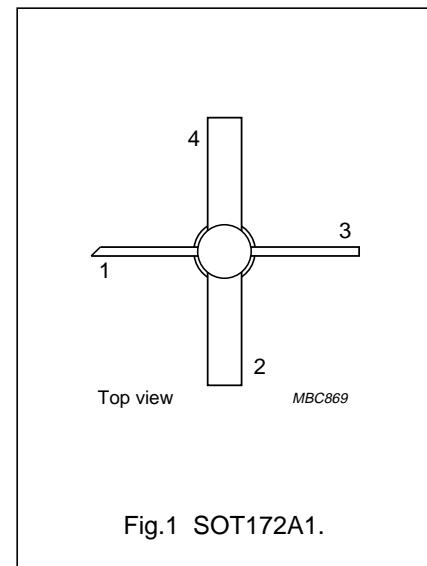


Fig.1 SOT172A1.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_{CBO}	collector-base voltage	open emitter	–	–	25	V
V_{CEO}	collector-emitter voltage	open base	–	–	19	V
I_C	DC collector current		–	–	500	mA
P_{tot}	total power dissipation	up to $T_c = 100^\circ\text{C}$	–	–	10	W
h_{FE}	DC current gain	$I_C = 240 \text{ mA}; V_{CE} = 18 \text{ V}; T_j = 25^\circ\text{C}$	60	–	–	
f_T	transition frequency	$I_C = 240 \text{ mA}; V_{CE} = 18 \text{ V}; f = 1 \text{ GHz}; T_{amb} = 25^\circ\text{C}$	–	6	–	GHz
G_{UM}	maximum unilateral power gain	$I_C = 240 \text{ mA}; V_{CE} = 18 \text{ V}; f = 800 \text{ MHz}; T_{amb} = 25^\circ\text{C}$	–	10	–	dB
V_O	output voltage	$d_{im} = -60 \text{ dB}; I_C = 240 \text{ mA}; V_{CE} = 18 \text{ V}; R_L = 75 \Omega; f_{(p+q-r)} = 793.25 \text{ MHz}$	–	1.6	–	V

WARNING**Product and environmental safety - toxic materials**

This product contains beryllium oxide. The product is entirely safe provided that the BeO discs are not damaged. All persons who handle, use or dispose of this product should be aware of its nature and of the necessary safety precautions. After use, dispose of as chemical or special waste according to the regulations applying at the location of the user. It must never be thrown out with the general or domestic waste.

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LIMITING VALUES

In accordance with the Absolute Maximum System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V_{CBO}	collector-base voltage	open emitter	–	25	V
V_{CEO}	collector-emitter voltage	open base	–	19	V
V_{EBO}	emitter-base voltage	open collector	–	2	V
I_C	DC collector current		–	500	mA
P_{tot}	total power dissipation	up to $T_c = 100^\circ\text{C}$	–	10	W
T_{stg}	storage temperature		–65	150	$^\circ\text{C}$
T_j	junction temperature		–	200	$^\circ\text{C}$

THERMAL RESISTANCE

SYMBOL	PARAMETER	THERMAL RESISTANCE
$R_{th\ j-c}$	thermal resistance from junction to case	10 K/W

CHARACTERISTICS $T_j = 25^\circ\text{C}$ unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I_{CBO}	collector cut-off current	$I_E = 0; V_{CB} = 18\text{ V}$	–	–	100	μA
h_{FE}	DC current gain	$I_C = 240\text{ mA}; V_{CE} = 18\text{ V}$	60	110	–	
C_c	collector capacitance	$I_E = i_e = 0; V_{CB} = 18\text{ V}; f = 1\text{ MHz}$	–	3.6	–	pF
C_e	emitter capacitance	$I_C = i_c = 0; V_{EB} = 0.5\text{ V}; f = 1\text{ MHz}$	–	11	–	pF
C_{re}	feedback capacitance	$I_C = 0; V_{CB} = 18\text{ V}; f = 1\text{ MHz}$	2	2.6	–	pF
C_{cs}	collector-stud capacitance		–	1.2	–	pF
f_T	transition frequency	$I_C = 240\text{ mA}; V_{CE} = 18\text{ V}; f = 1\text{ GHz}; T_{amb} = 25^\circ\text{C}$	4.5	6	–	GHz
G_{UM}	maximum unilateral power gain (note 1)	$I_C = 240\text{ mA}; V_{CE} = 18\text{ V}; f = 500\text{ MHz}; T_{amb} = 25^\circ\text{C}$	–	16	–	dB
		$I_C = 240\text{ mA}; V_{CE} = 18\text{ V}; f = 1\text{ GHz}; T_{amb} = 25^\circ\text{C}$	–	10	–	dB
V_O	output voltage	note 2	–	1.6	–	V
d_2	second order intermodulation distortion	note 3	–	–50	–	dB

Notes

- G_{UM} is the maximum unilateral power gain, assuming S_{12} is zero and $G_{UM} = 10 \log \frac{|S_{21}|^2}{(1 - |S_{11}|^2)(1 - |S_{22}|^2)} \text{ dB}$
- $d_{im} = -60\text{ dB}$ (DIN 45004); $I_C = 240\text{ mA}; V_{CE} = 18\text{ V}; R_L = 75\text{ }\Omega$;
 $V_p = V_O; f_p = 795.25\text{ MHz};$
 $V_q = V_O - 6\text{ dB}; f_q = 803.25\text{ MHz};$
 $V_r = V_O - 6\text{ dB}; f_r = 805.25\text{ MHz};$
measured at $f_{(p+q-r)} = 793.25\text{ MHz}$.
- $I_C = 240\text{ mA}; V_{CE} = 18\text{ V}; R_L = 75\text{ }\Omega$;
 $V_p = V_q = V_O = 50.5\text{ dBmV} = 335\text{ mV};$
 $f_{(p+q)} = 810\text{ MHz}; f_p = 250\text{ MHz}; f_q = 560\text{ MHz}$.

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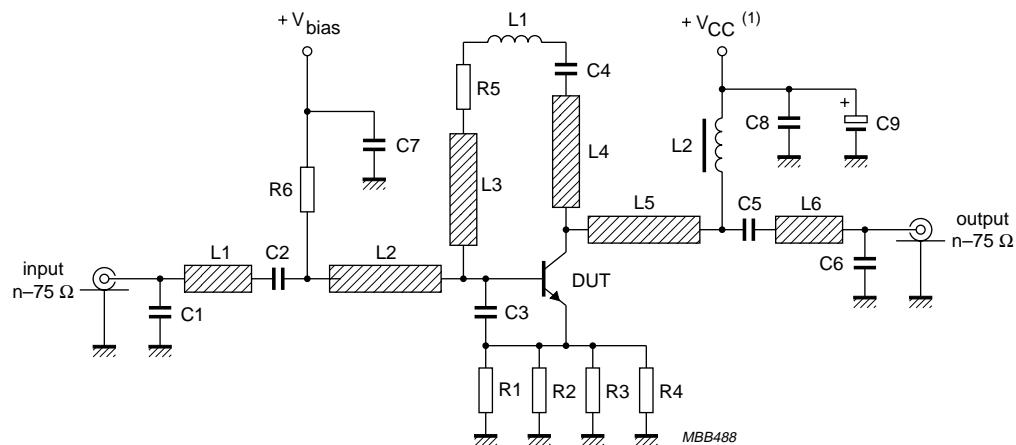


Fig.2 Intermodulation and second order intermodulation distortion test circuit.

List of components (see test circuit)

DESIGNATION	DESCRIPTION	VALUE	DIMENSIONS	CATALOGUE NO.
C1	miniature ceramic plate capacitor	0.82 pF		2222 680 03827
C2, C5, C7, C8	multilayer ceramic capacitor	10 nF		2222 852 47103
C3	multilayer ceramic chip capacitor	2.2 pF		2222 855 12228
C4 (note 1)	miniature ceramic plate capacitor	1 nF		2222 630 08102
C6	miniature ceramic plate capacitor	1.2 pF		2222 680 03128
C9	electrolytic capacitor	4.7 µF		2222 014 28478
L1 (note 1)	4.5 turns loosely wound 0.4 mm enamelled copper wire	=35 nH	internal coil diameter 2 mm	
L2	Ferroxcube choke	5 µH		3122 108 20153
ML1, ML6	microstripline	75 Ω	width 2.46 mm; length 9 mm	
ML2, ML5	microstripline	75 Ω	width 2.46 mm; length 22 mm	
ML3, ML4	microstripline	145 Ω	width 0.5 mm; length 12 mm	
R1, R2, R3, R4	metal film resistor	68 Ω	type MR25	2322 151 76819
R5 (note 1)	metal film resistor	240 Ω	type SFR16T	2322 180 73241
R6	metal film resistor	10 kΩ	type SFR16T	2322 180 73103

Note

- Components C4, L1, and R5 are mounted in a cavity in the brass ground plate.

The circuit is constructed on a printed circuit board and 10 mm thick brass ground plate, with a relative dielectric constant of ($\epsilon_r = 2.2$), thickness 1.57 mm; thickness of copper 0.017 mm (E.G. Rogers' RT/Duroid 5880).

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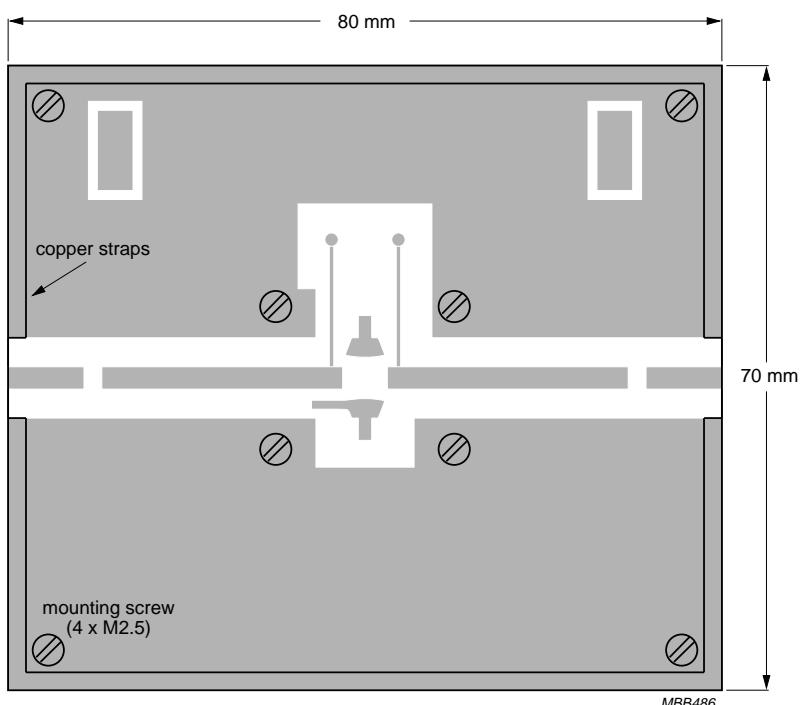
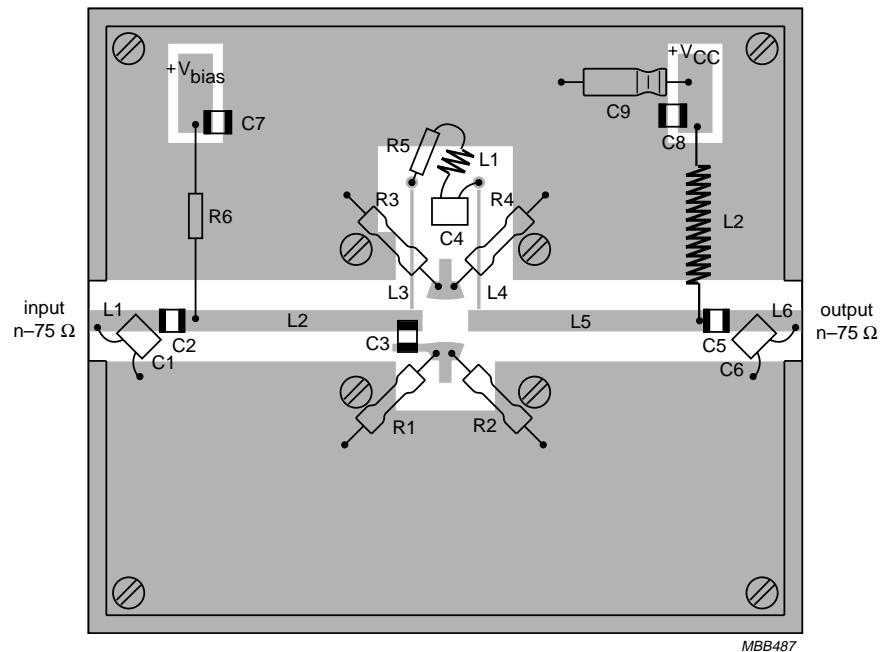


Fig.3 Intermodulation test circuit printed circuit board.

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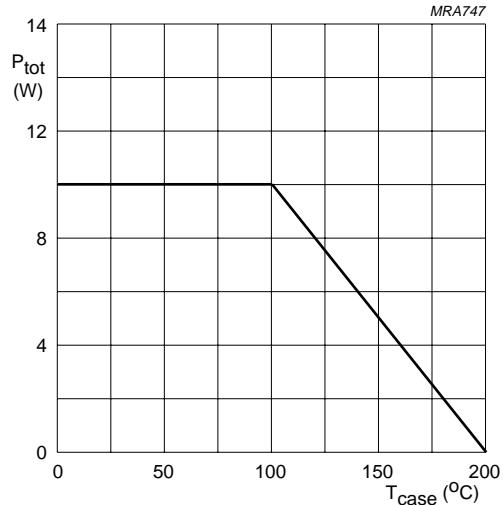


Fig.4 Power derating curve.

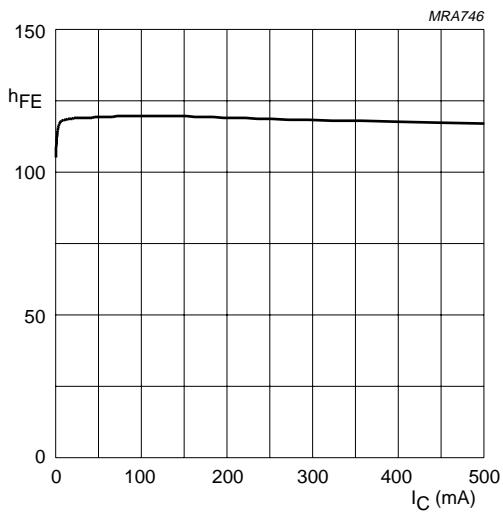
 $V_{CE} = 18 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$.

Fig.5 DC current gain as a function of collector current.

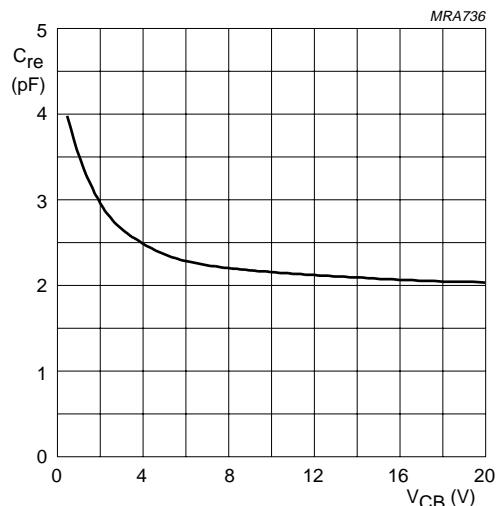
 $i_c = 0$; $f = 1 \text{ MHz}$.

Fig.6 Feedback capacitance as a function of collector-base voltage.

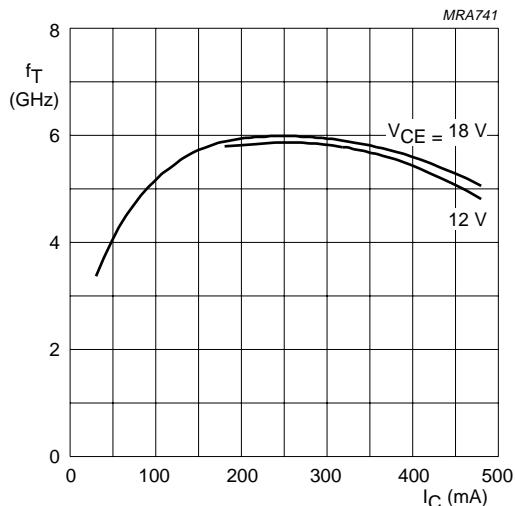
 $f = 1 \text{ GHz}$; $T_{amb} = 25 \text{ }^\circ\text{C}$.

Fig.7 Transition frequency as a function of collector current.

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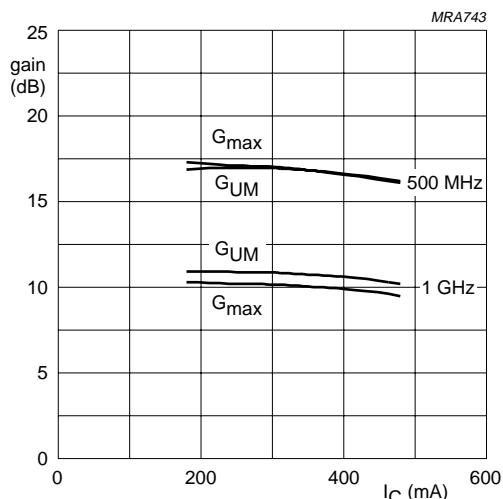
 $V_{CE} = 12$ V.

Fig.8 Gain as a function of collector current.

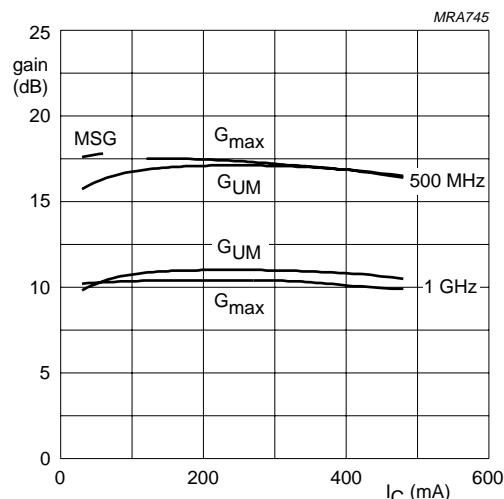
 $V_{CE} = 18$ V.

Fig.9 Gain as a function of collector current.

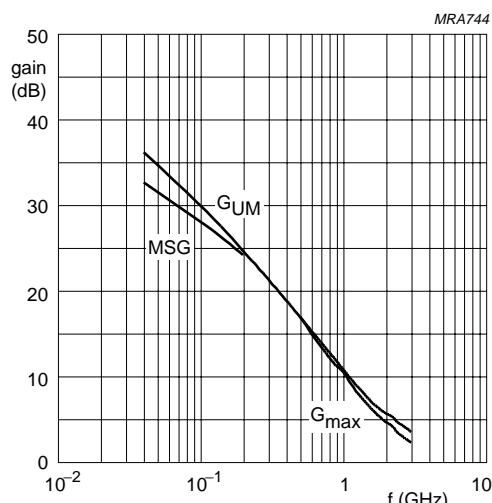
 $I_C = 240$ mA; $V_{CE} = 12$ V.

Fig.10 Gain as a function of frequency.

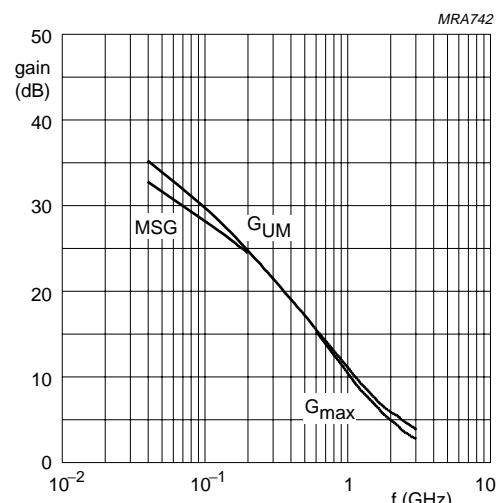
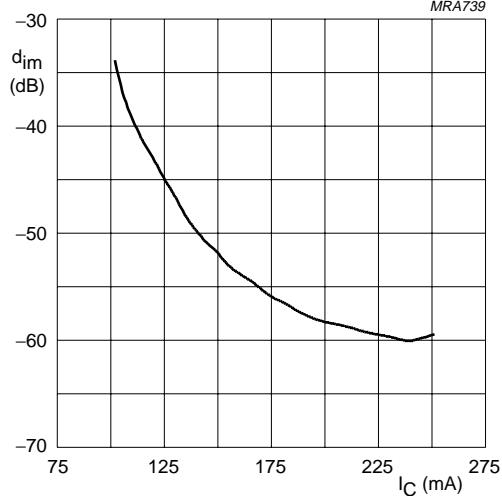
 $I_C = 240$ mA; $V_{CE} = 18$ V.

Fig.11 Gain as a function of frequency.

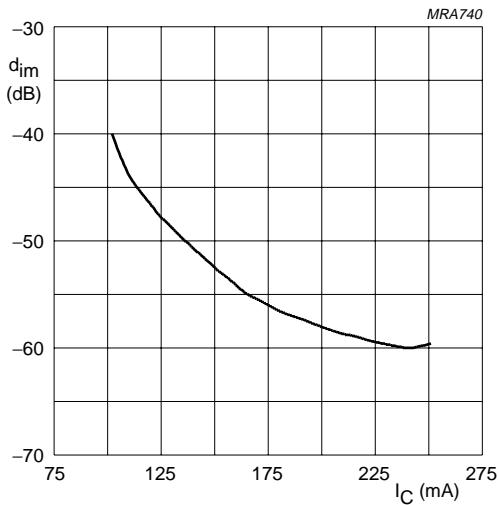
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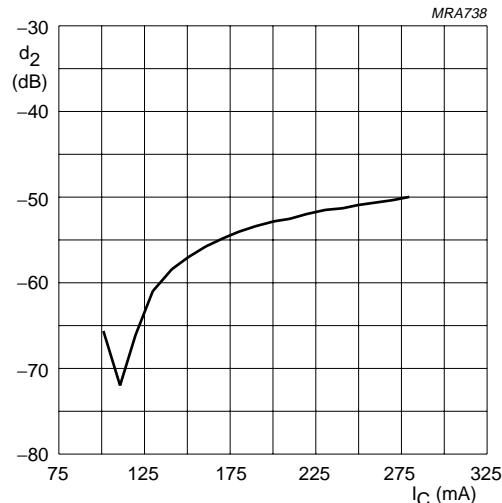
3-tone d_{im}; (V_{CE} = 18 V);
 $f_p = 445.25 \text{ MHz}; V_o = 65.11 \text{ dBmV (1.8 V)}$;
 $f_q = 453.25 \text{ MHz}; V_o = 59.11 \text{ dBmV}$;
 $f_r = 455.25 \text{ MHz}; V_o = 59.11 \text{ dBmV}$;
 $f_{(p+q+r)} = 443.25 \text{ MHz}$.

Fig.12 Intermodulation distortion as a function of collector current.



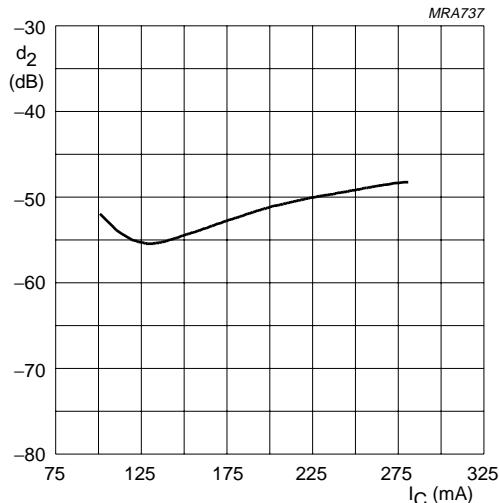
3-tone d_{im}; (V_{CE} = 18 V);
 $f_p = 795.25 \text{ MHz}; V_o = 64.08 \text{ dBmV (1.6 V)}$;
 $f_q = 803.25 \text{ MHz}; V_o = 58.08 \text{ dBmV}$;
 $f_r = 805.25 \text{ MHz}; V_o = 58.08 \text{ dBmV}$;
 $f_{(p+q+r)} = 793.25 \text{ MHz}$.

Fig.13 Intermodulation distortion as a function of collector current.



V_{CE} = 18 V;
 $f_p = 50 \text{ MHz}; V_o = 50.5 \text{ dBmV (0.335 V)}$;
 $f_q = 400 \text{ MHz}; V_o = 50.5 \text{ dBmV}$;
 $f_{(p+q)} = 450 \text{ MHz}$.

Fig.14 Second order intermodulation distortion as a function of collector current.

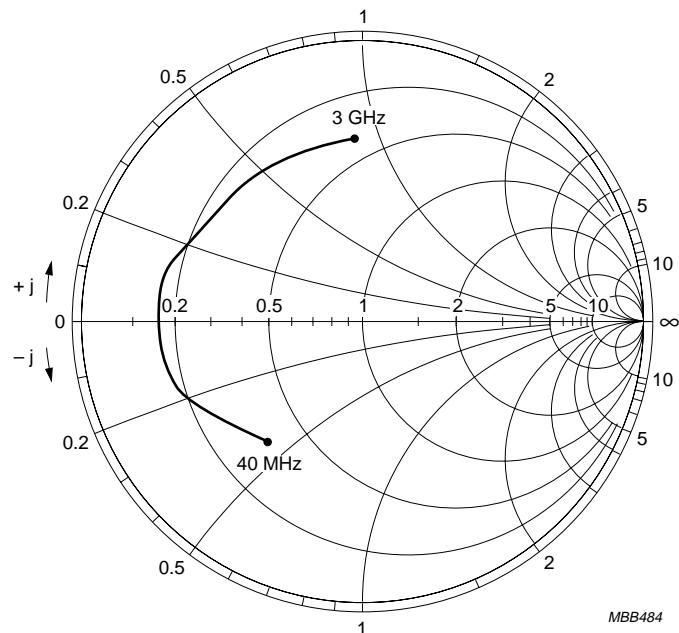


V_{CE} = 18 V;
 $f_p = 250 \text{ MHz}; V_o = 50.5 \text{ dBmV (0.335 V)}$;
 $f_q = 560 \text{ MHz}; V_o = 50.5 \text{ dBmV}$;
 $f_{(p+q)} = 810 \text{ MHz}$.

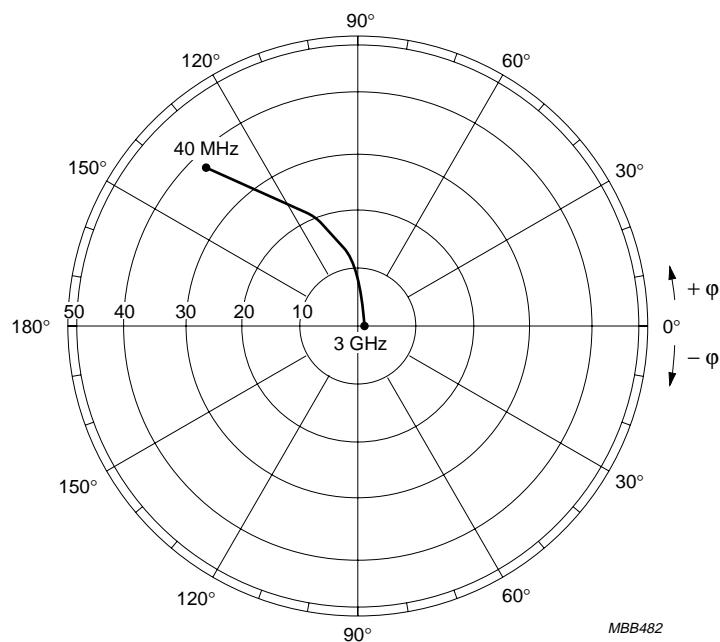
Fig.15 Second order intermodulation distortion as a function of collector current.

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$I_c = 240 \text{ mA}; V_{CE} = 18 \text{ V}.$
 $Z_0 = 50 \Omega.$

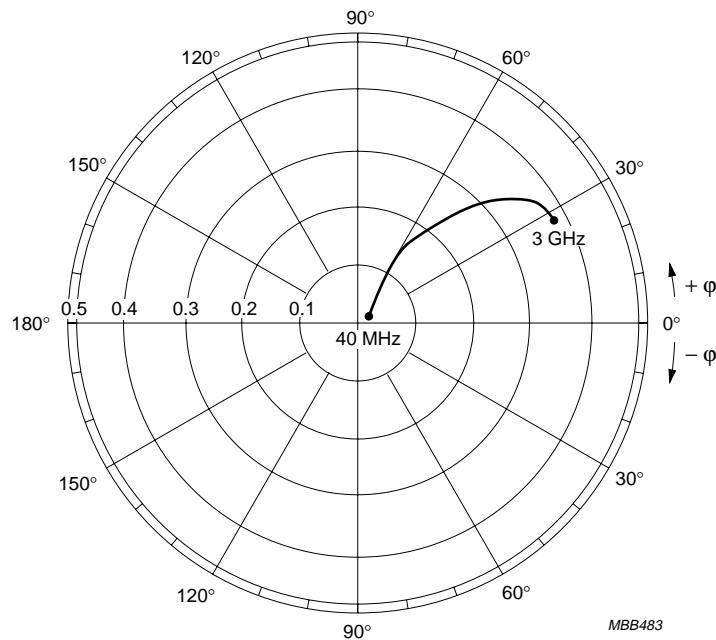
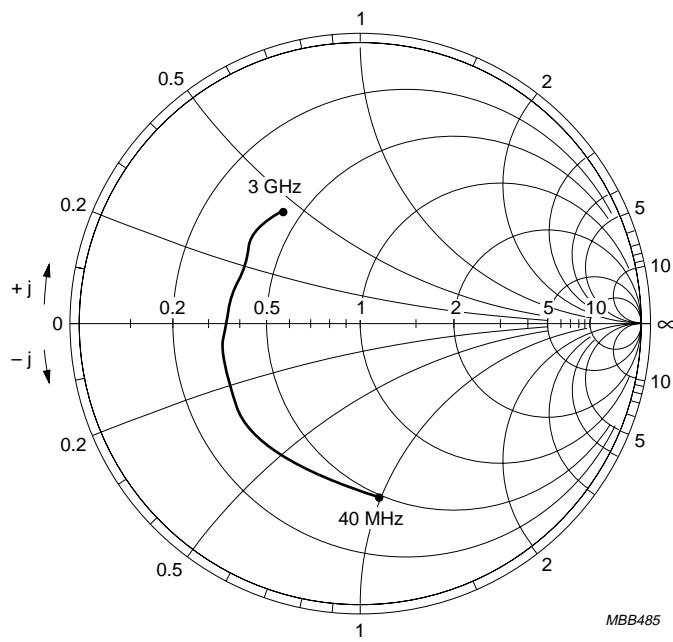
Fig.16 Common emitter input reflection coefficient (S_{11}).

$I_c = 240 \text{ mA}; V_{CE} = 18 \text{ V}.$

Fig.17 Common emitter forward transmission coefficient (S_{21}).

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 $I_c = 240 \text{ mA}$; $V_{CE} = 18 \text{ V}$.Fig.18 Common emitter reverse transmission coefficient (S_{12}). $I_c = 240 \text{ mA}$; $V_{CE} = 18 \text{ V}$. $Z_o = 50 \Omega$.Fig.19 Common emitter output reflection coefficient (S_{22}).

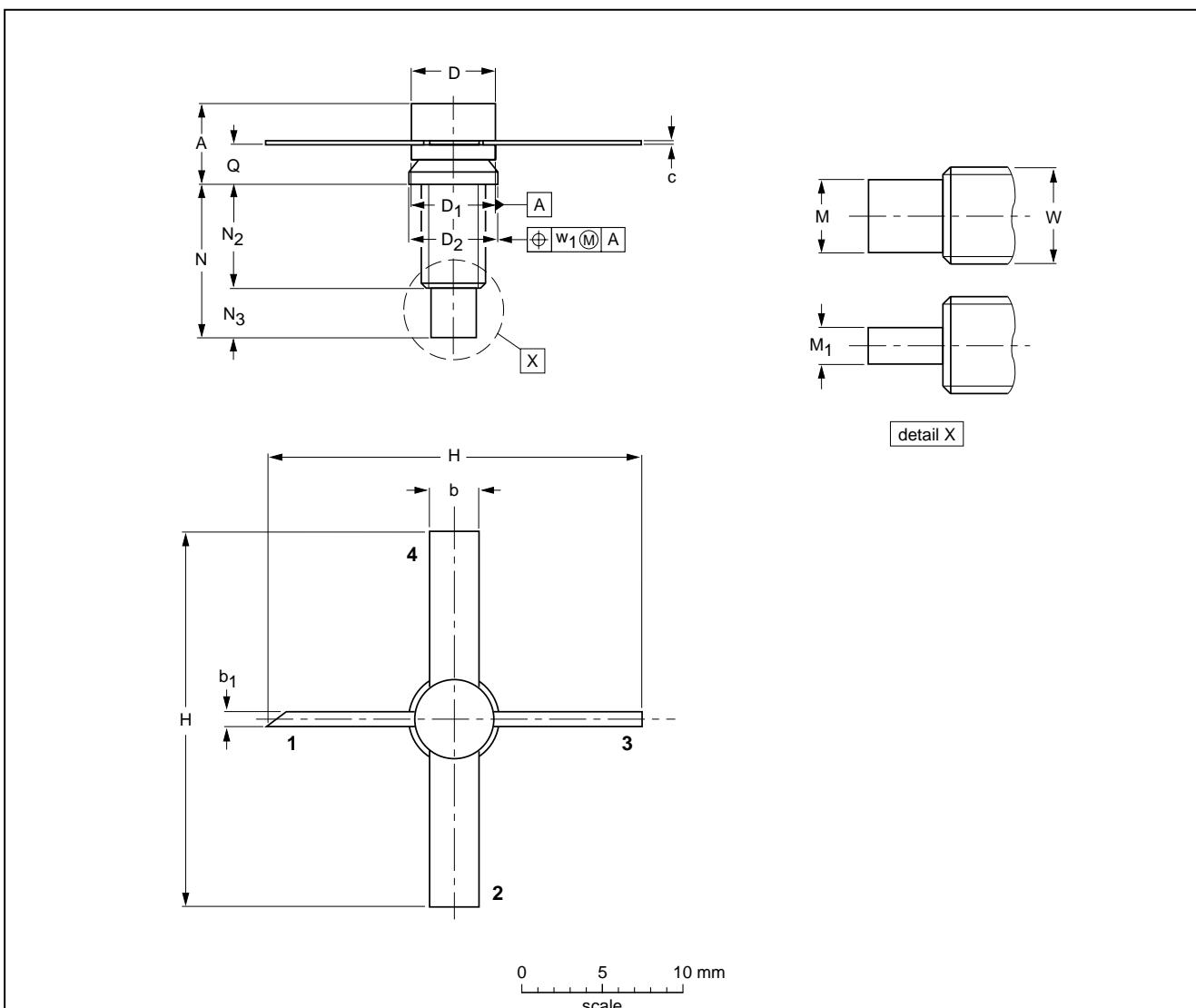
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PACKAGE OUTLINE

Studded ceramic package; 4 leads

SOT172A1



DIMENSIONS (millimetre dimensions are derived from the original inch dimensions)

UNIT	A	b	b ₁	c	D	D ₁	D ₂	H	M	M ₁	N	N ₂	N ₃	Q	W	w ₁
mm	5.31 4.34	3.31 3.04	0.89 0.63	0.16 0.10	5.20 4.95	5.33 5.08	5.33 5.08	26.17 24.63	3.05 2.79	1.66 1.39	11.82 10.89	8.89 6.90	3.69 2.92	2.90 2.31	8-32 UNC	0.38 0.015
inches	0.209 0.171	0.130 0.120	0.035 0.025	0.006 0.004	0.205 0.195	0.210 0.200	0.210 0.200	1.03 0.97	0.12 0.11	0.065 0.055	0.465 0.429	0.350 0.272	0.145 0.115	0.114 0.091		

OUTLINE VERSION	REFERENCES					EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ				
SOT172A1							97-06-28

NPN 6 GHz wideband transistor**BFQ270****DEFINITIONS**

Data Sheet Status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
Application information	
Where application information is given, it is advisory and does not form part of the specification.	

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