

CS8952

CrystalLAN™ 100BASE-X and 10BASE-T Transceiver

Features

- **.** Single-Chip IEEE 802.3 Physical Interface IC for 100BASE-TX, 100BASE-FX and 10BASE-T
- **Adaptive Equalizer provides Extended** Length Operation (>160 m) with Superior Noise Immunity and NEXT Margin
- ! Extremely Low Transmit Jitter (<400 ps)
- **.** Low Common Mode Noise on TX Driver for Reduced EMI Problems
- ! Integrated RX and TX Filters for 10BASE-T
- ! Compensation for Back-to-Back "Killer Packets"
- Digital Interfaces Supported
	- Media Independent Interface (MII) for 100BASE-X and 10BASE-T
	- Repeater 5-bit code-group interface (100BASE-X)
	- 10BASE-T Serial Interface
- Register Set Compatible with DP83840A
- ! IEEE 802.3 Auto-Negotiation with Next Page Support
- **.** Six LED drivers (LNK, COL, FDX, TX, RX, and SPD)
- Low power (135 mA Typ) CMOS design operates on a single 5 V supply

Description

The CS8952 uses CMOS technology to deliver a highperformance, low-cost 100BASE-X/10BASE-T Physical Layer (PHY) line interface. It makes use of an adaptive equalizer optimized for noise and near end crosstalk (NEXT) immunity to extend receiver operation to cable lengths exceeding 160 m. In addition, the transmit circuitry has been designed to provide extremely low transmit jitter (<400 ps) for improved link partner performance. Transmit driver common mode noise has been minimized to reduce EMI for simplified FCC certification.

The CS8952 incorporates a standard Media Independent Interface (MII) for easy connection to a variety of 10 and 100 Mb/s Media Access Controllers (MACs). The CS8952 also includes a pseudo-ECL interface for use with 100Base-FX fiber interconnect modules.

ORDERING INFORMATION

100-pin TQFP **Evaluation Board**

Preliminary Product Information \vert This document contains information for a new product. Cirrus Logic reserves the right to modify this product without notice.

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1. SPECIFICATIONS AND CHARACTERISTICS

ABSOLUTE MAXIMUM RATINGS (AVSS, DVSS = 0 V, all voltages with respect to 0 V.)

Operation at or beyond these limits may result in permanent damage to the device. **WARNING:** Normal operation is not guaranteed at these extremes.

RECOMMENDED OPERATING CONDITIONS (AVSS, DVSS = 0 V, all voltages with respect

to $0 V.$)

QUARTZ CRYSTAL REQUIREMENTS (If a 25 MHz quartz crystal is used, it must meet the fol-

lowing specifications.)

DC CHARACTERISTICS (Over recommended operating conditions)

DC CHARACTERISTICS (CONTINUED) (Over recommended operating conditions)

Notes: 1. With digital outputs connected to CMOS loads.

10BASE-T CHARACTERISTICS

100BASE-X CHARACTERISTICS

RX/TX Signaling for 100Base-FX

100BASE-TX MII RECEIVE TIMING - 4B/5B ALIGNED MODES

100BASE-TX MII RECEIVE TIMING - 5B BYPASS ALIGN MODE

100BASE-TX MII TRANSMIT TIMING - 4B/5B ALIGN MODES

100BASE-TX MII TRANSMIT TIMING - 5B BYPASS ALIGN MODE

10BASE-T MII RECEIVE TIMING

10BASE-T MII TRANSMIT TIMING

10BASE-T Transmit Timing

10BASE-T SERIAL RECEIVE TIMING

10BASE-T SERIAL TRANSMIT TIMING

AUTO NEGOTIATION / FAST LINK PULSE TIMING

SERIAL MANAGEMENT INTERFACE TIMING

2. INTRODUCTION

The CS8952 is a complete physical-layer transceiver for 100BASE-TX and 10BASE-T applications. Additionally, the CS8952 can be used with an external optical module for 100BASE-FX.

2.1 **High Performance Analog**

The highly integrated mixed-signal design of the CS8952 eliminates the need for external analog circuitry such as external transmit or receive filters. The CS8952 builds upon Cirrus Logic's experience in pioneering the high-volume manufacturing of 10BASE-T integrated circuits with "true" internal filters. The CS8952, CS8920, CS8904, and CS8900 include fifth-order, continuous-time Butterworth 10BASE-T transmit and receive filters, allowing those products to meet 10BASE-T wave shape, emission, and frequency content requirements without external filters.

2.2 **Low Power Consumption**

The CS8952 is implemented in low power CMOS, consuming only 135 mA typically. Three low-power modes are provided to make the CS8952 ideal for power sensitive applications such as CardBus.

2.3 **Application Flexibility**

The CS8952's digital interface and operating modes can be tailored to efficiently support a wide variety of applications. For example, the Media Independent Interface (MII) supports 100BASE-TX, 100BASE-FX and 10BASE-T NIC cards, switch ports and router ports. Additionally, the low-latency "repeater" interface mode minimizes data delay through the CS8952, facilitating system compliance with overall network delay budgets. To support 10BASE-T applications, the CS8952 provides a 10BASE-T serial port (Seven-wire ENDEC interface).

2.4 **Typical Connection Diagram**

Figure 1 illustrates a typical MII to CS8952 application with twisted-pair and fiber interfaces. Refer to the Analog Design Considerations section for detailed information on power supply requirements and decoupling, crystal and magnetics requirements, and twisted-pair and fiber transceiver connections.

3. FUNCTIONAL DESCRIPTION

The CS8952 is a complete physical-layer transceiver for 100BASE-TX and 10BASE-T applications. It provides a Physical Coding Sub-layer for communication with an external MAC (Media Access Controller). The CS8952 also includes a complete Physical Medium Attachment layer and a 100BASE-TX and 10BASE-T Physical Medium Dependent layer. Additionally, the CS8952 provides a PECL interface to an external optical module for 100BASE-FX applications.

The primary digital interface to the CS8952 is an enhanced IEEE 802.3 Media Independent Interface (MII). The MII supports parallel data transfer, access to the CS8952 Control and Status registers, and several status and control pins. The CS8952's operating modes can be tailored to support a wide variety of applications, including low-latency 100BASE-TX repeaters, switches and MII-based network interface cards.

For 100BASE-TX applications, the digital data interface can be either 4-bit parallel (nibbles) or 5-bit parallel (code-groups). For 10BASE-T applications, the digital data format can be either 4-bit parallel (nibbles) or one-bit serial.

The CS8952 is controlled primarily by configuration registers via the MII Management Interface. Additionally, a number of the most fundamental register bits can be set at power-up and reset time by connecting pull-up or pull-down resistors to external pins.

The CS8952's MII interface is enhanced beyond IEEE requirements by register extensions and the addition of pins for \overline{MII} IRQ, RX EN, and ISO-DEF signals. The MII_IRQ pin provides an inter-

Figure 1. Typical Connection Diagram

rupt signal to the controller when a change of state has occurred in the CS8952, eliminating the need for the system to poll the CS8952 for state changes. The RX_EN signal allows the receiver outputs to be electrically isolated. The ISODEF pin controls the value of register bit ISOLATE in the Basic Mode Control Register (address 00h) which in turn electrically isolates the CS8952's MII data path.

3.1 **Major Operating Modes**

The following sections describe the four major operating modes of the CS8952:

- 100BASE-X MII Modes (TX and FX)
- 100BASE-X Repeater Modes
- 10BASE-T MII Mode \overline{a}
- 10BASE-T Serial Mode

The choice of operating speed (10 Mb/s versus 100 Mb/s) is made using the auto-negotiation input pins (AN0, AN1) and/or the auto-negotiation MII registers. The auto-negotiation capability also is used to select a duplex mode (full or half duplex). Both speed and duplex modes can either be forced or negotiated with the far-end link partner.

The digital interface mode (MII, repeater, or 10BASE-T serial) is selected by input pins BPALIGN, BP4B5B and 10BT SER as shown in Table 1. Speed and duplex selection are made through the AN[1:0] pins as shown in Table 5.

Operating Mode BPALIGN BP4B5B 10BT SER		
100BASE-X MII		
10BASE-T MII		

Table 1.

Table 1.

$3.1.1$ **100BASE-X MII Application (TX)** and FX)

The CS8952 provides an IEEE 802.3-compliant MII interface. Data is transferred across the MII in four-bit parallel (nibble) mode. TX CLK and RX_CLK are nominally 25 MHz for 100BASE-X.

The 100BASE-X mode includes both the TX and FX modes, as determined by pin BPSCR (bypass scrambler), or the BPSCR bit (bit 13) in the Loopback, Bypass, and Receiver Error Mask Register (address 18h). In FX mode, an external optical module is connected to the CS8952 via pins TX_NRZ+, TX_NRZ-, RX_NRZ+, RX_NRZ-, SIGNAL+, and SIGNAL-. In FX mode, the MLT-3/NRZI conversion blocks and the scrambler/descrambler are bypassed.

$3.1.1.1$ **Symbol Encoding and Decoding**

In 100BASE-X modes, 4-bit nibble transmit data is encoded into 5-bit symbols for transmission onto the media as shown in Tables 2 and 3. The encoding is necessary to allow data and control symbols to be sent consecutively along the same media transparent to the MAC layer. This encoding causes the symbol rate transmitted across the wire (125) symbols/second) to be greater than the actual data rate of the system (100 symbols/second).

1. DATA code groups are indicated by $RX_DV = 1$

2. CONTROL code groups are inserted automatically during transmission in response to TX_EN. They are not generated through any combination of TXD[3:0] or TX_ER.

3. IDLE is indicated by $RX_DV = 0$.

1. CONTROL code groups become violations when found in the data portion of the frame.

2. Invalid code groups are mapped to 5h unless the Code Error Report select bit in the Loopback, Bypass, and Receiver Error Mask Register (address 18h) is set, in which case invalid code groups are mapped to 6h.

Table 3. 4B5B Code Violation Decoding

$3.1.1.2$ 100 Mb/s Loopback

One of two internal 100BASE-TX loopback modes can be selected. Local loopback redirects the $TXD[3:0]$ input data to $RXD[3:0]$ data outputs through the 4B5B coders and scramblers. Local loopback is selected by asserting pin LPBK, by setting the LPBK bit (bit 14) in the Basic Mode Control Register (address 00h) or by setting bits 8 and 11 in the Loopback, Bypass, and Receiver Error Mask Register (address 18h) as shown in Table 4.

Remote loopback redirects the analog line interface inputs to the analog line driver outputs. Remote loopback is selected by setting bit 9 in the Loopback, Bypass, and Receiver Error Mask Register (address 18h) as shown in Table 4.

Table 4.

When changing between local and non-loopback modes, the data on RXD[3:0] will be undefined for approximately 330 µs.

$3.1.2$ **100BASE-X Repeater Application**

The CS8952 provides two low latency modes for repeater applications. These are selected by asserting either pin BPALIGN or BP4B5B. Both pins have the effect of bypassing the 4B5B encoder and decoder. Bypassing the coders decreases latency, and uses a 5-bit wide parallel code group interface on pins RXD[4:0] and TXD[4:0] instead of the 4bit wide MII nibble interface on pins RXD[3:0] and TXD[3:0]. In repeater mode, pin RX_ER is redefined as the fifth receive data bit (RXD4), and pin TX ER is redefined as the fifth transmit data bit $(TXD4)$.

BPALIGN can also be selected by setting bit 12 in Loopback, Bypass, and Receiver Error Mask Register (address 18h). BP4B5B can be selected by setting bit 14 of the same register.

Pin BPALIGN causes more of the CS8952 to be bypassed than the BP4B5B pin. BPALIGN also bypasses the scrambler/descrambler, and the NRZI to NRZ converters (see Figure 1). Also, for repeater applications, pin REPEATER should be asserted to redefine the function of the CRS (carrier sense) pin. The REPEATER function may also be invoked by setting bit 12 in the PCS Sublayer Configuration Register (address 17h).

For repeater applications, the RX_EN pin can be used to gate the receive data pins $(RXD[4:0],$

RX_CLK, RX_DV, COL, and CRS) onto a shared, external repeater system bus.

$3.1.3$ **10BASE-T MII Application**

The digital interface used in this mode is the same as that used in the 100BASE-X MII mode except that TX_CLK and RX_CLK are nominally $2.5 MHz$

The CS8952 includes a full-featured 10BASE-T interface, as described in the following sections.

$3.1.3.1$ **Full and Half Duplex operation**

The 10BASE-T function supports full and half duplex operation as determined by pins $AN[1:0]$ and/or the corresponding MII register bits. (See Table 5).

3.1.3.2 **Collision Detection**

If half duplex operation is selected, the CS8952 detects a 10BASE-T collision whenever the receiver and transmitter are active simultaneously. When a collision is present, the collision is reported on pin COL. Collision detection is undefined for full-duplex operation.

$3.1.3.3$ **Jabber**

The jabber timer monitors the transmitter and disables the transmission if the transmitter is active for greater than approximately 105 ms. The transmitter stays disabled until approximately 406 ms after the internal transmit request is no longer enabled.

$3.1.3.4$ **Link Pulses**

To prevent disruption of network operation due to a faulty link segment, the CS8952 continually monitors the 10BASE-T receive pair (RXD+ and RXD-) for packets and link pulses. After each packet or link pulse is received, an internal Link-Loss timer is started. As long as a packet or link pulse is received before the Link-Loss timer finishes (between 50 and 100 ms), the CS8952 maintains normal operation. If no receive activity is detected, the CS8952 disables packet transmission to prevent "blind" transmissions onto the network (link pulses are still sent while packet transmission is disabled). To reactivate transmission, the receiver must detect a single packet (the packet itself is ignored), or two normal link pulses separated by more than 6 ms and no more than 50 ms.

The CS8952 automatically checks the polarity of the receive half of the twisted pair cable. To detect a reversed pair, the receiver examines received link pulses and the End-of-Frame (EOF) sequence of incoming packets. If it detects at least one reversed link pulse and at least four frames in a row with negative polarity after the EOF, the receive pair is considered reversed. If the polarity is reversed and bit 1 of the 10BASE-T Configuration Register (address 1Ch), is set, the CS8952 automatically corrects a reversal

In the absence of transmit packets, the transmitter generates link pulses in accordance with Section 14.2.1.1 of the Ethernet standard. Transmitted link pulses are positive pulses, one bit time wide, typically generated at a rate of one every 16 ms. The 16 ms timer also starts whenever the transmitter completes an End-of-Frame (EOF) sequence. Thus, a link pulse will be generated 16 ms after an EOF unless there is another transmitted packet.

$3.1.3.5$ Receiver Squelch

The 10BASE-T squelch circuit determines when valid data is present on the RXD+/RXD- pair. Incoming signals passing through the receive filter are tested by the squelch circuit. Any signal with amplitude less than the squelch threshold (either positive or negative, depending on polarity) is reiected.

3.1.3.6 10BASE-T Loopback

When Loopback is selected, the TXD[3:0] pins are looped back into the RXD[3:0] pins through the

Manchester Encoder and Decoder. Selection is made via:

- setting bit 14 in the Basic Mode Control Register (address 00h) or
- setting bits 8 and 11 in the Loopback, Bypass, and Receiver Error Mask Register (address 18h) or
- asserting the LPBK pin.

3.1.3.7 **Carrier Detection**

The carrier detect circuit informs the MAC that valid receive data is present by asserting the Carrier Sense signal (CRS) as soon it detects a valid bit pattern (1010b or 0101b for 10BASE-T). During normal packet reception, CRS remains asserted while the frame is being received, and is de-asserted within 2.3 bit times after the last low-to-high transition of the End-of-Frame (EOF) sequence. Whenever the receiver is idle (no receive activity), CRS is de-asserted.

$3.1.4$ **10BASE-T Serial Application**

This mode is selected when pin 10BT_SERis asserted during power-up or reset, and operates similar to the 10BASE T MII mode except that data is transferred serially on pins RXD0 and TXD0 using a 10 MHz RX_CLK and TX_CLK. Receive data is framed by CRS rather than RX DV.

3.2 **Auto-Negotiation**

The CS8952 supports auto-negotiation, which is the mechanism that allows the two devices on either end of an Ethernet link segment to share information and automatically configure both devices for maximum performance. When configured for auto-negotiation, the CS8952 will detect and automatically operate full-duplex at 100 Mb/s if the device on the other end of the link segment also supports full-duplex, 100 Mb/s operation, and auto-negotiation. The CS8952 auto-negotiation capability is fully compliant with the relevant portions of section 28 of the IEEE 802.3u standard.

The CS8952 can auto-negotiate both operating speed (10 versus 100 Mb/s), duplex mode (half duplex versus full duplex), and flow control (pause frames), or alternatively can be set not to negotiate. At power-up and reset times, the auto-negotiation mode is selected via the auto-negotiation input pins $(AN[1:0])$. This selection can later be changed using the Auto-Negotiation Advertisement Register $(address 04h).$

Pins AN[1:0] are three level inputs, and have the function shown in Table 5.

Auto-Negotiation encapsulates information within a burst of closely spaced Link Integrity Test Pulses, referred to as a Fast Link Pulse (FLP) Burst. The FLP Burst consists of a series of Link Integrity Pulses which form an alternating clock / data sequence. Extraction of the data bits from the FLP Burst yields a Link Code Word which identifies the capability of the remote device.

In order to support legacy 10 and 100 Mb/s devices, the CS8952 also supports parallel detection. In parallel detection, the CS8952 monitors activity on the media to determine the capability of the link partner even without auto-negotiation having occurred.

3.3 **Reset Operation**

Reset occurs in response to six different conditions:

- 1) There is a chip-wide reset whenever the RE-SET pin is high for at least 200 ns. During a chip-wide reset, all circuitry and registers in the CS8952 are reset.
- 2) When power is applied, the CS8952 maintains reset until the voltage at the VDD supply pins reaches approximately 3.6 V. The CS8952 comes out of reset once VDD is greater than approximately 3.6 V and the crystal oscillator has stabilized.
- 3) There is a chip-wide reset whenever the RE-

SET bit (bit 15 of the Basic Mode Control Register (address 00h)) is set.

- 4) Digital circuitry is reset whenever bit 0 of the PCS Sub-Layer Configuration Register (address 17h) is set. Analog circuitry is unaffected.
- 5) Analog circuitry is reset and recalibrated whenever the CS8952 enters or exits the powerdown state, as requested by pin PWRDN.
- 6) Analog circuitry is reset and recalibrated whenever the CS8952 changes between 10 Mb/s and 100 Mb/s modes.

After a reset, the CS8952 latches the signals on various input pins in order to initialize key registers and goes through a self configuration. This includes calibrating on-chip analog circuitry. Time required for the reset calibration is typically 40 ms. External circuitry may access registers internal to the CS8952 during this time. Reset and calibration complete is indicated when bit 15 of the Basic Mode Control Register (address 00h) is clear.

3.4 **LED** Indicators

The LEDx, SPD100, and SPD10 output pins provide status information that can be used to drive LEDs or can be used as inputs to external control circuitry. Indication options include: receive activity, transmit activity, collision, carrier sense, polarity OK, descrambler synchronization status, autonegotiation status, speed (10 vs. 100), and duplex mode.

4. MEDIA INDEPENDENT INTERFACE (MII)

The Media Independent Interface (MII) provides a simple interconnect to an external Media Access Controller (MAC). This connection may be chip to chip, motherboard to daughterboard, or a connection between two assemblies attached by a limited length of shielded cable and an appropriate connector.

The MII interface uses the following pins:

STATUS Pins

- COL Collision indication, valid only for half duplex modes.
- **CRS** Carrier Sense indication

SERIAL MANAGEMENT Pins

- MDIO a bi-directional serial data path
- MDC clock for MDIO (16.7 MHz max)
- $\overline{MII _ IRQ}$ Interrupt indicating change in the Interrupt Status Register (address 11h)

RECEIVE DATA Pins

- RXD[3:0] Parallel data output path
- RX_CLK Recovered clock output
- RX DV Indicates when receive data is present and valid
- RX_ER Indicates presence of error in re- $\mathbf{L}^{\mathcal{A}}$ ceived data
- RX EN Can be used to tri-state receiver output pins

TRANSMIT DATA Pins

- TXD[3:0] Parallel data input path
- TX CLK Transmit clock \mathbf{r}
- TX EN Indicates when transmit data is present and valid
- TX_ER Request to transmit a 100BASE-T HALT symbol, ignored for 10BASE-T operation.

The interface uses TTL signal levels, which are compatible with devices operating at a nominal supply voltage of either 5.0 or 3.3 volts. It is capable of supporting either 10 Mb/s or 100 Mb/s data rates transparently. That is, all signaling remains identical at either data rate; only the nominal clock frequency is changed.

4.1 **MII Frame Structure**

Data frames transmitted through the MII have the following format:

Each frame is preceded by an inter-frame gap. The inter-frame gap is an unspecified time during which no data activity occurs on the media as indicated by the de-assertion of CRS for the receive path and TX_EN for the transmit path.

The Preamble consists of seven bytes of 10101010.

The Start of Frame Delimiter consists of a single byte of 10101011.

Data may be any number of bytes.

The End of Frame Delimiter is conveyed by the deassertion of RX DV and TX EN for receive and transmit paths, respectively.

Transmission and/or reception of each byte of data is done one nibble at a time in the following order:

4.2 **MII Receive Data**

The presence of recovered data on the RXD[3:0] bus is indicated by the assertion of RX DV. RX_DV will remain asserted from the beginning of the preamble (or Start of Frame Delimiter if preamble is not used) to the End of Frame Delimiter. Once RX DV is asserted, valid data will be driven

onto RXD[3:0] synchronously with respect to RX CLK.

Receive errors are indicated during frame reception by the assertion of RX ER. It indicates that an error was detected somewhere in the frame currently being transferred across the MII. RX_ER will transition synchronously with respect to the RX_CLK, and will be held high for one cycle for each error received. It is up to the MAC to ensure that a CRC error is detected in that frame by the Logical Link Control. Figure 2 illustrates reception without errors, and Figure 3 illustrates reception with errors.

4.3 **MII Transmit Data**

TX_EN is used by the MAC to signal to the CS8952 that valid nibbles of data are being presented across the MII via TXD[3:0]. TX_EN must be asserted synchronously with the first nibble of preamble, and must remain asserted as long as valid data is being presented to the MII.

TX EN must be de-asserted within one TX CLK cycle after the last nibble of data (CRC) has been presented to the CS8952. When TX_EN is not asserted, data on TXD[3:0] is ignored.

Transmit errors should be signaled by the MAC by asserting TX_ER for one or more TX_CLK cycles. TX ER must be synchronous with TX CLK. This will cause the CS8952 to replace the nibble with a HALT symbol in the frame being transmitted. This invalid data will be detected by the receiving PHY and flagged as a bad frame. Figure 4 illustrates transmission without errors, and Figure 5 illustrates transmission with errors.

MII Management Interface 4.4

The CS8952 provides an enhanced IEEE 802.3 MII Management Interface. The interface consists of three signals: a bi-directional serial data line (MDIO), a data clock (MDC), and an optional interrupt signal $(\overline{MII} \ \overline{IRQ})$. The Management Interface can be used to access status and control registers internal to the CS8952. The CS8952 implements an extended set of 16-bit MII registers. Eight of the registers are defined by the IEEE 802.3

Figure 3. Reception with errors

Figure 5. Transmission with errors

specification, while the remaining registers provide enhanced monitoring and control capabilities.

As many as 31 devices may share a single Management Interface. A unique five-bit PHY address is associated with each device, with all devices responding to PHY address 00000. The CS8952 determines its PHY address at power-up or reset through the PHYAD[4:0] pins.

4.5 **MII Management Frame Structure**

Frames transmitted through the MII Management Interface have the following format (Table 6):

When the management interface is idle, the MDIO signal will be tri-stated, and the MAC is required to keep MDIO pulled to a logic ONE.

At the beginning of each transaction, the MAC will typically send a sequence of 32 contiguous logic ONE bits on MDIO with 32 corresponding clock cycles on MDC to provide the CS8952 with a pattern that it can use to establish synchronization. Optionally, the CS8952 may be configured to operate without the preamble through bit 9 of the PCS Sub-Layer Configuration Register (address 17h).

Preamble	Start of	Opcode	PHY	Register	Turnaround	Data	Idle
(32 bits)	Frame	2 bits)	Address	Address	2 bits)	$'16 \text{ bits}$	
	(2 bits)		(5 bits)	(5 bits)			

Table 6. Format for Frame Transmitted through the MII Management Interface

The Start of Frame is indicated by a 01 bit pattern.

A read transaction is indicated by an Opcode of 10 and a write by 01.

The PHY Address is five bits, with the most significant bit sent first. If the PHY address included in the frame is not 00000 or does not match the PHY-AD field of the Self Status Register (address 19h), the rest of the frame is ignored.

The register address is five bits, with the most significant bit sent first, and indicates the CS8952 register to be written to/read from.

The Turnaround time is a two bit time spacing between when the MAC drives the last register address bit onto MDIO and the data field of a management frame in order to avoid contention during a read transaction. For a read transaction, the MAC should tri-state the MDIO pin beginning on the first bit time, and the CS8952 will begin driving the MDIO signal to a logic ZERO during the second bit time. During write transactions, since the MDIO direction does not need to be reversed, the MAC will drive the MDIO to a logic ONE for the first bit time and a logic ZERO for the second.

The data field is always 16 bits in length, with the most significant bit sent first.

5. CONFIGURATION

The CS8952 can be configured in a variety of ways. All control and status information can be accessed via the MII Serial Management Interface. Additionally, many configuration options can be set at power-up or reset times via individual control lines. Some configuration capabilities are available at any time via individual control lines.

5.1 **Configuration At Power-up/Reset Time**

At power-up and reset time, the following pins are

5.2 **Configuration Via Control Pins**

The following pins are for dedicated control signals and can be used at any time to configure the CS8952.

5.3 Configuration via the MII

The CS8952 supports configuration by software control through the use of 16-bit configuration and status registers accessed via the MDIO/MDC pins (MII Management Interface). The first seven registers are defined by the IEEE 802.3 specification. Additional registers extend the register set to provide enhanced monitoring and control capabilities.

6. CS8952 REGISTERS

The CS8952 register set is comprised of the 16-bit status and control registers described below. A detailed description each register follows.

Basic Mode Control Register - Address 00h 6.1

6.2 Basic Mode Status Register - Address 01h

PHY Identifier, Part 1 - Address 02h 6.3

PHY Identifier, Part 2 - Address 03h 6.4

6.5 Auto-Negotiation Advertisement Register - Address 04h

6.6 Auto-Negotiation Link Partner Ability Register - Address 05h

6.7 Auto-Negotiation Expansion Register - Address 06h

6.8 Auto-Negotiation Next-Page Transmit Register - Address 07h

Interrupt Mask Register - Address 10h 6.9

This register indicates which events will cause an interrupt event on the MII_IRQ pin. Each bit acts as an enable to the interrupt. Thus, when set, the event will cause the MILIRQ pin to be asserted. When clear, the event will not affect the MII_IRQ pin, but the status will still be reported via the Interrupt Status Register (address 11h).

Interrupt Status Register - Address 11h 6.10

This register indicates which event(s) caused an interrupt event on the MII_IRQ pin. All bits are selfclearing, and will thus be cleared upon readout.

Disconnect Count Register - Address 12h 6.11

6.12 False Carrier Count Register - Address 13h

6.13 Scrambler Key Initialization Register - Address 14h

Receive Error Count Register - Address 15h 6.14

6.15 Descrambler Key Initialization Register - Address 16h

6.16 PCS Sub-Layer Configuration Register - Address 17h

Self Status Register - Address 19h 6.18

6.19 10BASE-T Status Register - Address 1Bh

10BASE-T Configuration Register - Address 1Ch 6.20

7. DESIGN CONSIDERATIONS

The CS8952 is a mixed-signal device containing the high-speed digital and analog circuits required to implement Fast Ethernet communication. It is important the designer adhere to the following guidelines and recommendations for proper and reliable operation of the CS8952. These guidelines will also benefit the design with good EMC performance.

7.1 **Twisted Pair Interface**

The recommended connection of the twisted-pair interface is shown if Figure 6. The unused cable pairs are terminated to increase the common-mode performance. Common-mode performance is also improved by connecting the center taps of the RX and TX input circuits to the DC-isolated ground plane. The 0.01 µF capacitor C1 must provide 2KV $(1,500$ Vrms for 60 seconds) of isolation to meet 802.3 requirements. If a shielded RJ45 connector is used (recommended), the shield should be connected to chassis ground.

7.2 **100BASE-FX Interface**

Figure 7 shows the recommended connection for a 100BASE-FX interface to a Hewlett-Packard HFBR-5103 fiber transceiver. Termination circuitry may need to be revised for other fiber transceivers. The FX Drive bit in the Loopback, Bypass, and Receiver Error Mask Register (address 18h) may be used to tailor the PECL interface for 50 Ω or 150 Ω loads.

Figure 6. Recommended Connection of Twisted-Pair Ports (Network Interface Card)

Figure 7. Recommended Connection of Fiber Port

TX_NRZ+/- termination components should be placed as close to the fiber transceiver as possible, while RX_NRZ+/- and SIGNAL+/- termination components should be placed close to the CS8952.

The CS8952 100BASE-FX interface IO pins (TX_NRZ+, TX_NRZ-, RX_NRZ+, RX_NRZ-, SIGNAL+, and SIGNAL-) may be left unconnected if a fiber interface is not used.

7.3 **Internal Voltage Reference**

A 4.99 k Ω biasing resistor must be connected between the CS8952 RES pin and ground. This resistor biases the internal analog circuits of the CS8952 and should be placed as close as possible to RES pin. Connect the other end of this resistor directly to the ground plane. Connect the adjacent CS8952 ground pins (pins 85 and 87) to the grounded end of the resistor forming a "shield" around the RES connection.

7.4 **Clocking Schemes**

The CS8952 may be clocked using one of three possible schemes: using a 25 MHz crystal and the internal oscillator, using an external oscillator sup-

Figure 8. Biasing Resistor Connection and Layout

plied through the XTAL I pin, or using an external clock source supplied through the TX_CLK pin.

When a 25 MHz crystal is used, it should be placed within one inch of the XTAL I and XTAL O pins of the CS8952. The crystal traces should be short, have no vias, and run on the component side. Table 7 lists examples of manufacturers of suitable crystals. The designer should evaluate their crystal selection for suitability in their specific design.

An external CMOS clock source may be connected to the XTAL_I pin, with the XTAL_O pin left open. The input capacitance of the XTAL I pin is larger than the other inputs (a maximum of 35pF), since it includes the additional load capacitance of the crystal oscillator. Care should be taken to assure any external clock source attached to XTAL_I is capable of driving higher capacitive loads. The clock signal should be 25 MHz $\pm 0.01\%$ with a duty cycle between 45% and 55%.

When the XTAL_I pin load is a problem, or only a TTL level clock source is available, the CS8952 can be clocked through the TX_CLK pin, providing the TX_CLK mode is set appropriately using the TCM pin. The clock frequency will be dependent on the operating mode.

7.5 **Recommended Magnetics**

The CS8952 requires an isolation transformer with a 1:1 turns ratio for both the transmit and receive signals. Table 7 lists examples of manufacturers

with transformers meeting these requirements. However, the designer should evaluate the magnetics for suitability in their specific design.

7.6 **Power Supply and Decoupling**

The CS8952 supports connection to either a 3.3 V or 5.0 V MII. When connected to a $+5.0$ V MII, all power pins should be provided $+5.0 V +/-5\%$, and all signal inputs should be referenced to $+5.0V$. When interfaced with a 3.3 V MII, VDD_MII power pins should be provided $+3.3$ V $+/-5\%$, VDD power pins should be provided $+5.0 V +/-5\%$, and all signal inputs should be referenced to $+3.3$ V.

Component	Manufacturer	Part Number
Crystal	Raltron Electronics Corp. 10651 NW 19th St. Miami, FL 33172 (305) 593-6033 www.raltron.com	AS-25.000-15-F- EXT-SMD-TR- CIR
Transformer	Halo Electronics, Inc. P.O. Box 5826 Redwood City, CA 94063 USA (650) 568-5800 www.haloelectronics.com	TG22-3506ND
	Bel Fuse, Inc. 198 Van Vorst Street Jersey City, NJ 07302 USA (201) 432-0463 www.belfuse.com	S5558-5999-46
	Pulse Engineering 12220 World Trade Drive San Diego, CA 92128 USA (619) 674-8100 www.pulseeng.com	PE-68515
Fiber Interface	Hewlett Packard Component Sales Response Center (408) 654-8675 www.hp.com/HP-COMP	HFBR-5103

Table 7. Support Component Manufactures

Each CS8952 power pin should be connected to a 0.1μ F bypass capacitor and then to the power plane. The bypass capacitors should be located as close to its corresponding power pin as possible. Connect ground pins directly to the ground plane.

7.7 **General Layout Recommendations**

The following PCB layout recommendations will help ensure reliable operation of the CS8952 and good EMC performance.

Use a multilayer Printed Circuit Board with at least one ground and one power plane. A typical +5V MII application would be as follows:

Layer 1: (top) Components and first choice signal routing Laver 2: Ground Layer 3: Power $(+5V)$ Layer 4: (bottom) Second choice signal routing, bypass components

- Place transformer TI as close to the RJ45 connector as possible with the secondary (network) side facing the RJ45 and the primary (chip) side facing the analog side (pins 76-100) of CS8952. Place the CS8952 in turn as close to T1 as possible.
- Use the bottom layer for signal routing as a sec- \bullet ond choice. You may place all components on the top layer. However, bypass capacitors are optimally placed as close to the chip as possible and may be best located underneath the CS8952 on the bottom layer. Termination components at the RJ-45 and fiber transceiver may also be optimally placed on the bottom layer.
- Connect a 0.1μ F bypass capacitor to each CS8952 VDD and VDD_MII pin. Place it as close to its corresponding power pin as possible and connect the other lead directly to the ground plane.
- The 4.99K reference resistor should be placed as close to the RES pin as possible. Connect the other end of this resistor to the ground plane using a via. Connect the adjacent VSS pins (pins 85 and 87) to the grounded end of the resistor forming a shield as illustrated in Figure 8.
- Controlled impedance is necessary for critical \bullet signals $TX+/-, RX+/-, TX_N RZ+/-$, and RX_NRZ+/-. These should be run as microstrip

transmission lines (100 Ω differential, 50 Ω single-ended). The MII signals should be 68 Ω microstrip transmission lines. (For short MII signal paths one may standardize on a given trace width for all traces without significant degradation in signal integrity.)

- Avoid routing traces other than the TX and RX signals under transformer T1 and the RJ45 connector. Signals may run on the bottom side underneath the CS8952 as long as they stay away from critical analog traces.
- Connect all CS8952 ground and power pins directly to the ground and power planes, respectively. Note: The VDD_MII power pins may need their own power plane or plane segment in $+3.3$ V MII applications.
- Depending on the orientation and location of the transformer, the CS8952, and the RJ-45, and on whether the application is for a NIC or a switch, the RX and TX pairs may need to cross. This should be done by changing layers on a pair by pair basis only, using the minimum number of vias, and making sure that each trace within a pair "sees" the same path as its peer.

Figure 6 shows the CS8952 in a NIC or adapter configuration. It may be configured for a hub or repeater application by changing the wiring to the RJ-45 as shown in Table 8 .

- Differential pair transmission lines should be \bullet routed close together (one trace width spacing edge-to-edge) and kept at least two trace widths away from other traces, components, etc. TX and RX pairs should be routed away from each other and may use opposite sides of the PCB as necessary, Each member of the differential pair should "see" the same PCB terrain as its peer.
- Unused spaces on the signal layers should be filled with ground fill (pour). Vias should connect the ground patches to the ground plane. This is especially recommended (symmetrical-

ly) on both sides of the $TX+/$ - traces.

Table 8. RJ-45 Wiring

- No signal current carrying planes, i.e. no ground or power plane, should be present underneath the region between the transformer secondary (network) side and the RJ-45. However, a chassis plane may be added in this region to pick up the metal tabs of a shielded RJ-45. This chassis plane should be separated from the ground and power planes by at least 50 mils. That is, all other ground and power planes should be "cookie cuttered" so they are voided in the area of the chassis plane. Generally speaking, parts should not cross the moat except for the transformer.
- Proper termination practices must be used with all transmission lines, especially if sending and receiving high speed signals on and off the board. Series terminations must be kept close to the source and load terminations close to the load. Thus the TX NRZ+/- termination components must be kept close to the fiber optic

transceiver, and the RX_NRZ+/- and SIG-NAL+/- termination components must be kept close to the CS8952.

- Locate the crystal as close to the CS8952 as possible, running short traces on the component side in order to reduce parasitic load capacitance.
- Add bulk capacitance at each connector where power may be supplied. For example, MII power may be provided at the MII connector and at a separate connector for test purposes. If so, and the two connectors are not adjacent, then the bulk capacitors should be duplicated in each locations.
- Use wide traces to connect the "Bob Smith" termination resistors at T1 and the RJ-45 to the 2 KV capacitor or capacitors in order to minimize their lead inductance.

8. PIN DESCRIPTIONS

Pin Diagram

MII Interface Pins

COL/PHYAD0 - Collision Detect/PHY Address 0. Input/Tri-State Output, Pin 48.

Asserted active-high to indicate a collision on the medium during half-duplex operation. In full-duplex operation, COL is undefined and should be ignored. When configured for 10 Mb/s operation, COL is also used to indicate a Signal Quality Error (SQE) condition.

At power-up or at reset, the logic value on this pin is latched into bit 0 of the PHY Address field of the Self Status Register (address 19h). This pin includes a weak internal pull-up (> 150 K Ω), or the value may be set by an external 4.7 K Ω pull-up or pull-down resistor.

CRS/PHYAD2 - Carrier Sense/PHY Address 2. Input/Tri-State Output, Pin 49.

The operation of CRS is controlled by the REPEATER pin as follows:

At power-up or at reset, the logic value of this pin is latched into bit 2 of the PHY Address Field of the Self Status Register (address 19h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $K\Omega$ pull-up or pull-down resistor.

MDC - Management Data Clock. Input, Pin 28.

Input clock used to transfer serial data on MDIO. The maximum clock rate is 16.67 MHz. This clock may be asynchronous to RX CLK and TX CLK.

MDIO - Management Data Input/Output. Bi-Directional, Pin 27.

Bi-directional signal used to transfer management data between the CS8952 and the Ethernet controller.

In order to conform with Annex 22B of the IEEE 802.3u specification, the MII DRV pin should be pulled high during power-up or reset, and the MDIO pin should have an external 1.5 KQ pull-up resistor. For systems not required to drive external connectors and cables as described in the IEEE802.3u specification, the external pull-up resistor may not be necessary.

MIL_IRQ - MII Interrupt. Open Drain Output, Pin 26.

Asserted low to indicate the status corresponding to one of the unmasked interrupt status bits in the Interrupt Status Register (address 11h) has changed. It will remain low until the ISR is read, clearing all status bits.

This open drain pin requires a 4.7 k Ω pull-up resistor.

RX_CLK - Receive Clock. Tri-State Output, Pin 36

Continuous clock output used as a reference clock for sampling RXD[3:0], RX_ER, and RX_DV. RX_CLK will have the following nominal frequency:

In order to conform with Annex 22B of the IEEE 802.3u specification, the MII DRV pin should be pulled high during power-up or reset, and the RX_CLK pin should have an external 33 Ω series resistor. For systems not required to drive external connectors and cables as described in the IEEE802.3u specification, the external series resistor may not be necessary.

RX DV/MII DRV - Receive Data Valid/MII Drive Strength. Input/Tri-State Output, Pin 33.

Asserted high to indicate valid data nibbles are present on RXD[3:0].

At power-up or at reset, this pin is used as an input to determine the drive strength of the MII output drivers. When the pin is low, all MII output drivers will be standard 4 mA CMOS drivers. When high, additional drive strength will be added to the MII output drivers. This pin includes a weak internal pulldown (> 20 K Ω), or the value may be set by an external 4.7 K Ω pull-up or pull-down resistor.

In order to conform with Annex 22B of the IEEE 802.3u specification, this pin should be pulled high during power-up or reset and should have an external 33 Ω series resistor. For systems not required to drive external connectors and cables as described in the IEEE802.3u specification, it may be possible to reduce overall power consumption by pulling the pin low at power-up or reset, and the external series resistor may not be necessary.

RX EN - Receive Enable. Input, Pin 14.

When high, signals RXD[3:0], RX_CLK, RX_DV, and RX_ER are enabled. When low, these signals are tri-stated. RX EN allows the received data signals of multiple PHY transceivers to share the same MII bus.

This pin includes a weak internal pull-up (> 150 K Ω), or the value may be set by an external 10 K Ω pullup or pull-down resistor.

RX ER/PHYAD4/RXD4 - Receive Error/PHY Address 4/Receive Data 4. Input/Tri-State Output, Pin 37.

During normal MII operation, this pin is defined as RX ER (Receive Error). When RX DV is high, RX ER asserted high indicates that an error has been detected in the current receive frame. When RX DV is low and \overline{R} XD[3:0] = "1110", RX ER high indicates a False Carrier condition.

If either BPALIGN or BP4B5B is asserted, then this pin is re-defined as RXD4 (Receive Data 4), the most-significant bit of the received five-bit code-group. If the 4B5B encoder is being bypassed, receive data is present when RX DV is asserted. If alignment is being bypassed, data reception is continuous.

At power-up or at reset, the logic value on this pin is latched into bit 4 of the PHY Address field of the Self Status Register (address 19h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $K\Omega$ pull-up or pull-down resistor.

In order to conform with Annex 22B of the IEEE 802.3u specification, the MII DRV pin should be pulled high during power-up or reset, and the RX ER pin should have an external 33 Ω series resistor. For systems not required to drive external connectors and cables as described in the IEEE802.3u specification, the external series resistor may not be necessary.

RXD3/PHYAD3 - Receive Data 3/PHY Address 3. Tri-State Output, Pin 29. RXD2 - Receive Data 2. Tri-State Output, Pin 30. RXD1/PHYAD1 - Receive Data 1/PHY Address 1. Tri-State Output, Pin 31. RXD0 - Receive Data 0. Tri-State Output, Pin 32.

Receive data output. Receive data is present when RX DV is asserted. RXD0 is the least-significant bit. For MII modes, nibble-wide data (synchronous to RX CLK) is transferred on pins RXD[3:0]. In 10 Mb/s serial mode, pin RXD0 is used as the serial output pin, and RXD[3:1] are ignored. When either BP4B5B or BPALIGN is selected, pin RXD4 contains the most-significant bit of the five-bit code-group.

At power-up or at reset, the value on RXD1/PHYAD1 is latched into bit 1 of the PHY Address field of the Self Status Register (address 19h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $K\Omega$ pull-up or pull-down resistor.

At power-up or at reset, the logic value on RXD3/PHYAD3 is latched into bit 3 of the PHY Address field of the Self Status Register (address 19h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $K\Omega$ pull-up or pull-down resistor.

In order to conform with Annex 22B of the IEEE 802.3u specification, the MII DRV pin should be pulled high during power-up or reset, and the RXD[3:0] pins should have external 33 Ω series resistors. For systems not required to drive external connectors and cables as described in the IEEE802.3u specification, the external series resistors may not be necessary.

TX_CLK - Transmit Clock. Input/Tri-State Output, Pin 42.

Continuous clock signal used by the CS8952 as a reference clock to sample TXD[3:0], TX ER, and TX_EN. TX_CLK can be referenced either internally (Output Mode) or externally (Input Mode) based upon the value of the TCM pin at power-up or at reset.

When the TCM pin is high on power-up or reset, the CLK25 pin may be used as a source for the TX CLK pin. When the TCM pin is floating on power-up or reset, TX CLK must be supplied externally. TX CLK should have the following nominal frequency:

TX EN - Transmit Enable. Input, Pin 43.

Asserted high to indicate valid data nibbles are present on TXD[3:0]. When BPALIGN is selected, TX_EN must be pulled up to VDD_MII.

TX ER/TXD4 - Transmit Error Encoding/Transmit Data 4. Input, Pin 38.

When high, TX ER indicates to the CS8952 that a transmit error has occurred. If TX ER is asserted simultaneously with TX_EN in 100 Mb/s mode, the CS8952 will ignore the data on the TXD[3:0] pins and transmit one or more 100 Mb/s HALT symbols in its place. In 10 Mb/s mode, TX_ER has no effect on the transmitted data.

If BP4B5B or BPALIGN are set, TX_ER/TXD4 is used to transmit the most-significant bit of the five-bit code group.

TXD[3:0] - Transmit Data. Input, Pins 47, 46, 45, and 44.

Transmit data input pins. For MII modes, nibble-wide data (synchronous to TX CLK) must be presented on pins TXD[3:0] when TX EN is asserted high. TXD0 is the least significant bit. In 10 Mb/s serial mode, pin TXD0 is used as the serial input pin, and TXD[3:1] are ignored.

When either BP4B5B or BPALIGN is selected, pin TXD4 contains the most significant bit of the five-bit code-group.

Control and Status Pins

10BT_SER - 10 Mb/s Serial Mode Select. Input, Pin 23.

When asserted high during power-up or reset and 10 Mb/s operation is selected, serial data will be transferred on pins RXD0 and TXD0. When low during power-up or reset and 10 Mb/s operation is selected, data is transferred a nibble at a time on RXD[3:0] and TXD[3:0]. This pin is ignored during 100 Mb/s operation.

10 Mb/s serial mode may also be entered under software control through bit 9 of the 10BASE-T Status Register (address 1Bh).

At power-up or at reset, the value on this pin is latched into bit 9 of the 10BASE-T Status Register (address 1Bh). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $K\Omega$ pull-up or pull-down resistor.

AN[1:0] - Auto-Negotiate Control. Input, Pins 58 and 57.

These three-level input pins are sampled during power-up or reset. They control the forced or advertised auto-negotiation operating modes. If one of these pins is left unconnected, internal logic pulls its signal to a mid-range value, 'M'.

Auto-Negotiation may also be enabled and the advertised capabilities modified under software control through bit 8 of the Basic Mode Control Register (address 00h), and bits 5, 6, 7, 8, and 10 of the Auto-Negotiation Advertisement Register (address 04h).

These pins are pulled to 'M' through weak internal resistors $($ > 150 K Ω). Other values may be set by tying them directly to VDD_MII or VSS, or through external 10 K Ω pull-up or pull-down resistors.

BP4B5B - Bypass 4B5B Coders. Input, Pin 56.

When driven high during power-up or reset, the transmit 4B5B encoder and receiver 5B4B decoder are bypassed. Five-bit code groups are output and input on pins RXD[4:0] and TXD[4:0].

The 4B5B Coders may also be bypassed under software control through bit 14 of the Loopback, Bypass, and Receiver Error Mask Register (address 18h).

At power-up or at reset, the value on this pin is latched into bit 14 of the Loopback, Bypass and Receiver Error Mask Register (address 18h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 K Ω pull-up or pull-down resistor.

BPALIGN - Bypass Symbol Alignment. Input, Pin 52.

When driven high during power-up or reset, the following blocks are bypassed: 4B5B encoder, 5B4B decoder, scrambler, descrambler, NRZI encoder, and NRZI decoder. Five-bit code groups are output and input on pins RXD[4:0] and TXD[4:0]. The receiver will output five-bit data with no attempt to identify code-group boundaries; therefore, the data in one RXD[4:0] word may contain data from two code groups.

Symbol alignment may also be bypassed under software control through bit 12 of the Loopback, Bypass, and Receiver Error Mask Register (address 18h).

At power-up or at reset, the value on this pin is latched into bit 12 of the Loopback, Bypass and Receiver Error Mask Register (address 18h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 K Ω pull-up or pull-down resistor.

BPSCR - Bypass Scrambler. Input, Pin 62.

When driven high during power-up or reset, the scrambler and descrambler is bypassed and NRZI FX mode is selected.

The 100BASE-FX mode may also be entered under software control through bit 13 of the Loopback, Bypass, and Receiver Error Mask Register (address 18h).

At power-up or at reset, the value on this pin is latched into bit 13 of the Loopback, Bypass and Receiver Error Mask Register (address 18h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 K Ω pull-up or pull-down resistor.

ISODEF - Isolate Default. Input, Pin 63.

When asserted high during power-up or reset, the MII will power-up electrically isolated except for the MDIO and MDC pins. When low, the part will exit reset fully electrically connected to the MII.

The MII may also be isolated under software control through bit 10 of the Basic Mode Control Register (address 00h).

At power-up or at reset, the value on this pin is latched into bit 10 of the Basic Mode Control Register (address 00h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $\widehat{K\Omega}$ pull-up or pull-down resistor.

LED1 - Transmit Active LED. Open Drain Output, Pin 69.

This active-low output indicates transmit activity. It contains a pulse stretcher to insure that the transmit events are visible when the pin is used to drive an LED. The definition of this pin may be modified to indicate Disconnect Detection (bit 5 of the Self Status Register (address 19h)) by setting bit 2 of the PCS Sub-layer Configuration Register (address 17h).

This pin can be simultaneously connected to an LED and to a TTL-compatible, CMOS input pin.

LED2 - Receive Activity LED. Open Drain Output, Pin 70.

This active-low output indicates receive activity. It contains a pulse stretcher to insure that the receive events are visible when the pin is used to drive an LED.

This pin can be simultaneously connected to an LED and to a TTL-compatible, CMOS input pin.

LED3 - Link Good LED. Open Drain Output, Pin 71.

This active-low output indicates the CS8952 has detected a valid link.

This pin can be simultaneously connected to an LED and to a TTL-compatible, CMOS input pin.

LED4 - Polarity/Full Duplex LED. Open Drain Output, Pin 72.

This active-low output indicates:

1) for 100 Mb/s operation, the CS8952 is in full-duplex operation,

2) for 10 Mb/s operation, either good polarity exists or full duplex is selected (see bit 1 in the PCS Sublayer Configuration Register (address 17h)).

This pin can be simultaneously connected to an LED and to a TTL-compatible, CMOS input pin.

LED5 - Collision/Descrambler Lock LED. Open Drain Output, Pin 73.

This active-low output is asserted when either the CS8952 detects a collision (bit 11 of the PCS Sub-Layer Configuration Register (address 17h) is clear), or the 100BASE-TX descrambler is synchronized (bit 11 of the PCS Sub-Layer Configuration Register (address 17h) is set). It contains a pulse stretcher to insure that the collision events are visible when the pin is used to drive an LED.

This pin can be simultaneously connected to an LED and to a TTL-compatible, CMOS input pin.

LPBK - Loopback Enable. Input, Pin 51.

When this pin is asserted high and the CS8952 is operating in 100 Mb/s mode, the CS8952 will perform a local loopback inside the PMD block, routing the scrambled NRZI output to the NRZI input port on the descrambler. The loopback includes all CS8952 100 Mb/s functionality except the MLT-3 coders and the analog line interface blocks.

When asserted high and the CS8952 is operating in 10 Mb/s mode, the CS8952 will perform a local ENDEC loopback.

LPSTRT - Low Power Start. Input, Pin 50.

When this active-low input is asserted during power-up or reset, the CS8952 will exit reset in a low power configuration, where the only circuitry enabled is that necessary to maintain the media impedance. The CS8952 will remain in a low power state until RESET pin is asserted or the MDC pin toggles.

This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 K Ω pull-up or pull-down resistor.

PWRDN - Power Down. Input, Pin 64.

When this pin is asserted high, the CS8952 powers down all circuitry except that circuitry needed to maintain the network line impedance. This is the lowest power mode possible. The CS8952 will remain in low power mode until the PWRDN pin is deasserted.

A slightly higher power power-down mode may also be entered under software control through bit 11 of the Basic Mode Control Register (address 00h).

REPEATER - REPEATER Mode Select. Input, Pin 16.

This pin controls the operation of the CRS (Carrier Sense) pin as shown below:

At power-up or at reset, the value on this pin is latched into bit 12 of the PCS Sub-Layer Configuration Register (address 17h). This pin includes a weak internal pull-down (> 20 K Ω), or the value may be set by an external 4.7 $K\Omega$ pull-up or pull-down resistor.

SPD10 - 10 Mb/s Speed Indication. Output, Pin 68.

This pin is asserted high when the CS8952 is configured for 10 Mb/s operation. This pin can be used to drive a low-current LED to indicate 10 Mb/s operation.

SPD100 - 100 Mb/s Speed Indication. Output, Pin 67.

This pin is asserted high when the CS8952 is configured for 100 Mb/s operation. This pin can be used to drive a low-current LED to indicate 100 Mb/s operation.

TCM - Transmit Clock Mode Initialization. Input, Pin 59.

The logic value on this three-level pin during power-up or reset determines whether TX CLK is used as an input or an output, and whether an external 25 MHz clock reference is provided on the CLK25 output pin.

TEST[1:0] - Factory Test. Input, Pins 24 and 25.

These pins are for factory test only. They include weak internal pull-downs (> 20 K Ω), and should be tied directly to VSS for normal operation.

TXSLEW[1:0] - Transmit Slew Rate Control. Input, Pins 61 and 60.

These three-level pins allow adjustment to the rise and fall times of the 10BASE-TX transmitter output waveform. The rise and fall times are symmetric.

Media Interface Pins

RX+, RX- - 10/100 Receive. Differential Input Pair, Pins 91 and 92.

Differential input pair receives 10 or 100 Mb/s data from the receive port of the transformer primary.

TX+, TX- - 10/100 Transmit. Differential Output Pair, Pins 80 and 81.

Differential output pair drives 10 or 100 Mb/s data to the transmit port of the transformer primary.

RX NRZ+, RX NRZ- - FX Receive. Differential Input Pair, Pins 6 and 7.

PECL output pair receives 100 Mb/s NRZI-encoded data from an external optical module.

SIGNAL+, SIGNAL- - Signal Detect. Differential Input Pair, Pins 9 and 8.

PECL input pair receives signal detection indication from an external optical module.

TX NRZ+, TX NRZ- - FX Transmit. Differential Output Pair, Pins 5 and 4.

PECL output pair drives 100 Mb/s NRZI-encoded data to an external optical module.

General Pins

CLK25 - 25 MHz Clock. Output, Pin 17.

A 25 MHz Clock is output on this pin when the CS8952 is configured to use an external reference transmit clock in TX CLK IN MASTER mode. See the pin description for the Transmit Clock Mode Initialization pin (TCM) for more information on TX_CLK operating modes.

CLK25 may also be enabled regardless of the TCM pin state by clearing bit 7 of the PCS Sub-layer Configuration Register (address 17h).

RES - Reference Resistor. Input, Pin 86.

This input should be connected to ground with a 4.99 k Ω +/-1% series resistor. The resistor is needed for the biasing of internal analog circuits.

RESET - Reset. Input, Pin 15.

This active high input initializes the CS8952, and causes the CS8952 to latch the input signal on the following pins: COL/PHYAD0, CRS/PHYAD2, RX_ER/PHYAD4/RXD4, 10BT_SER, BP4B5B, BPALIGN, BPSCR, ISODEF, REPEATER, RXD[1]/PHYAD1, and RXD[3]/PHYAD3.

XTAL I - Crystal Input, Pin 96.

XTAL O - Crystal Output, Pin 97.

A 25 MHz crystal should be connected across pins XTAL_I and XTAL_O. If a crystal is not used, a 25 MHz CMOS level clock may be connected to XTAL I and XTAL O left open.

NOTE: The XTAL I pin capacitive load may be as high as 35pF. Any external clock source connected to this pin must be capable of driving larger capacitive loads.

RSVD - Reserved. Pins 74, 75, 76, 77, 84, 98, and 99.

These seven pins are reserved and should be tied to VSS.

VDD_MII - MII Power. Pins 21, 34, and 66.

These pins provide power to the CS8952 MII interface. Typically VDD_MII will be either +5V or +3.3V.

VDD - Core Power. Pins 2, 11, 19, 40, 54, 79, 82, 88, 89, 94, and 100.

These pins provide power to the CS8952 core. Typically, VDD should be +5V.

VSS - Ground. Pins 1, 3, 10, 12, 13, 18, 20, 22, 35, 39, 41, 53, 55, 65, 78, 83, 85, 87, 90, 93, and 95. These pins provide a ground reference for the CS8952.

9. PACKAGE DIMENSIONS

* Nominal pin pitch is 0.50 mm

Controlling dimension is mm. JEDEC Designation: MS026

· Notes •

