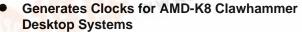
# 200-MHz CLOCK SYNTHESIZER/DRIVER

# WITH SPREAD SPECTRUM CAPABILITY AND DEVICE CONTROL INTERFACE

SCAS675 - APRIL 2002



- Uses a 14.318-MHz Crystal Input to **Generate Multiple Output Frequencies**
- Includes Spread Spectrum Clocking (SSC), 0.5% Downspread for Reduced EMI
- **Power Management Control Terminals**
- **SMBus Serial Interface Provides Output Enable and Control**
- Low-Output Skew and Low Jitter for Clock Distribution
- **Operates From Single 3.3-V Supply**
- **Generates the Following Clocks:** 
  - 2 CPU (3.3 V, 180° shifted pairs, 200/166/133/100 MHz)
  - 6 PCI (3.3 V, 33 MHz)
  - 1 PCI\_F (3.3 V, 33 MHz)
  - 3 REF (3.3 V, 14.318 MHz)
  - 1 USB (3.3 V, 48 MHz)
  - 1 FDC (3.3 V, 24 MHz or 48 MHz)
  - 3 PCI/LDT<sup>†</sup> (3.3 V, 33 MHz or 66 MHz)
- Packaged in 48-Pin SSOP Package

# description

The CDC960 is a clock synthesizer/driver and buffer that generates CPU, PCI, PCI/LDT, USB, FDC, and REF system clock signals to support PCs with an AMD-K8 Clawhammer-class system.

**DL PACKAGE** (TOP VIEW) FS0 & REF0 □ 48 TFS1 & REF1 47 T GND  $V_{DD}$ 2 XIN [3 46 VDD XOUT \ 4 45 | FS2 & REF2 44 N SPREAD GND 5 PCI/LDT SEL [ 43 VDDA PCI/LDT0 □ 42 GNDA PCI/LDT1 Γ 41 T CPU0 8 40 T CPU0  $\Lambda^{DD}$ 9 GND **1** 10 39 ND PCI/LDT2 [ 38 VDD 11 LDT\_Stop 37 CPU1 12 36 T CPU1 13 PCI1 [ 35 VDD 14 GND [ 34 GND 15 33 GNDF  $V_{DD}$ 16 PCI2 [ 17 32 VDDF PCI3 ∏ 18 31 NSB  $V_{DD}$  [ 30 NG GND 19 29 🛮 V<sub>DD</sub> GND [ 20 PCI4 28 24/48 SEL & FDC 21 PCI5 22 27 | GND PCI\_F [ SDATA 26 23 25 SCLK PCI\_Stop ∏ 24

All output frequencies are generated from a 14.318-MHz crystal input. A reference clock input can be provided at the XIN input instead of a crystal. It is recommended to use the bypass mode of the internal oscillator in this case. Two phase-locked loops (PLLs) are used to generate the host frequencies and 48-MHz clock frequencies. On-chip loop filters and internal feedback eliminate the need for external components.

The device provides a standard mode (100 kbps) SMBus 1.1 serial interface for device control. The implementation is as a slave with read and write capability. The device address is specified in the SMBus serial interface device address table. Both SMBus inputs (SDATA and SCLK) provide integrated pullup resistors (typically 150 k $\Omega$ ).

Seven 8-bit SMBus registers provide individual enable control for each of the outputs. The controllable outputs default to enabled at power up and can be placed in a disabled mode with a low-level output when a low-level control bit is written to the control register. The registers must be accessed in sequential order (i.e., random access of the registers not supported).

The CPU, PCI, PCI\_F, LDT, FDC (24/48-MHz), and USB (48-MHz) clock outputs provide low-skew/low-jitter clock signals for reliable clock operation. All outputs have 3-state capability, which can be selected via control inputs FS0, FS1, and FS2 at power-up preset condition.



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TIDT is equivalent to HT66 shown on AMD specification



# description (continued)

The CPU bus is a 3.3-V differential push-pull output type. All others are single-ended CMOS buffers.

The host frequencies are fixed and are controlled by the FS0, FS1 and FS2 signals at power-up. The CPU bus frequencies are 200, 166, 133 and 100 MHz.

Because the CDC960 is based on PLL circuitry, it requires a stabilization time to achieve phase-lock of the PLL. With use of external reference clock, this signal must be fixed-frequency and fixed-phase prior stabilization time starts.

#### **FUNCTION TABLES**

#### **DEVICE FREQUENCY SELECT FUNCTIONS**

SME CONTR			IN	PUTS						OUTPU	тѕ			
FS4 (Byte 0, Bit 5)	FS3 (Byte 0, Bit 4)	24/48_SEL <sup>‡</sup>	PCI/LDT_SEL <sup>‡</sup>	FS2	FS1	FS0	CPU	PCI_F	PCI/LDT	USBŤ	FDC <sup>†</sup>	REF <sup>†</sup>	Comment	FUNCTION
L	L	Н	Н	Н	Н	Н	200 MHz	33 MHz	33 MHz	48 MHz	24 MHz	14.31818 MHz		ins p
L	L	Н	L	Н	Н	Н	200 MHz	33 MHz	66 MHz	48 MHz	24 MHz	14.31818 MHz		-# pi
L	L	L	Н	Н	Н	Н	200 MHz	33 MHz	33 MHz	48 MHz	48 MHz	14.31818 MHz		FSx and 24/48_SEL# pins are latched at power up
L	L	L	L	Н	Н	Н	200 MHz	33 MHz	66 MHz	48 MHz	48 MHz	14.31818 MHz		/48_ d at
L	L	H/L	H/L	Н	Н	L	166 MHz	33 MHz	33/66 MHz	48 MHz	24/48 MHz	14.31818 MHz		d 24 che
L	L	H/L	H/L	Н	L	Н	133 MHz	33 MHz	33/66 MHz	48 MHz	24/48 MHz	14.31818 MHz		k and e lat
L	L	H/L	H/L	Н	L	L	100 MHz	33 MHz	33/66 MHz	48 MHz	24/48 MHz	14.31818 MHz		FS
L	L	Х	Н	L	L	Н	Xin	Xin/6	Xin/6	L	L	L	f(xin) = 0 to	Φ
L	L	Х	L	L	L	Н	Xin	Xin/6	Xin/3	L	L	L	$f_{(xin)} = 0 \text{ to}$ 200 MHz	Modes Test PLL by-pass mode
L	L	Н	Н	L	Н	Н	Xin	Xin/6	Xin/6	Xin/2	Xin/4	Xin		ss Te
L	L	Н	L	L	Н	Н	Xin	Xin/6	Xin/3	Xin/2	Xin/4	Xin	$f_{(xin)} = 0$ to	Modes by-pas
L	L	L	Н	L	Н	Н	Xin	Xin/6	Xin/6	Xin/2	Xin/2	Xin	$f_{(xin)} = 0 \text{ to}$ 16 MHz	
L	L	L	L	L	Н	Н	Xin	Xin/6	Xin/3	Xin/2	Xin/2	Xin		۵
L	L	Х	Χ	L	Н	L			Res	erved for fu	uture use			
L	L	Х	Χ	L	L	L	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z		
L	Н	H/L	H/L	Н	Н	Н	90 MHz	30 MHz	30/60 MHz	48 MHz	24/48 MHz	14.31818 MHz	-10%	
L	Н	H/L	H/L	Н	Н	L	119 MHz	30 MHz	30/60 MHz	48 MHz	24/48 MHz	14.31818 MHz	-10%	
L	Н	H/L	H/L	Н	L	Н	180 MHz	36.3 MHz	36.3/72.6 MHz	48 MHz	24/48 MHz	14.31818 MHz	-10%	
L	Н	H/L	H/L	Н	L	L	180 MHz	30 MHz	30/60 MHz	48 MHz	24/48 MHz	14.31818 MHz	-10%	
L	Н	H/L	H/L	L	Н	Н	111 MHz	36.9 MHz	36.9/73.9 MHz	48 MHz	24/48 MHz	14.31818 MHz	10%	
L	Н	H/L	H/L	L	Н	L	148 MHz	36.9 MHz	36.9/73.9 MHz	48 MHz	24/48 MHz	14.31818 MHz	10%	
L	Н	H/L	H/L	L	L	Н	222 MHz	44.4 MHz	44.4/88.8 MHz	48 MHz	24/48 MHz	14.31818 MHz	10%	
L	Н	H/L	H/L	L	L	L	222 MHz	36.9 MHz	36.9/73.9 MHz	48 MHz	24/48 MHz	14.31818 MHz	10%	
Н	Х	H/L	H/L	Χ	Χ	Χ			Not-	yet-defined	d settings			

<sup>†</sup> If the REF, USB, and FDC outputs are disabled in by pass mode, the Xin-input can be driven with an external clock signal from 0 MHz to 200 MHz. Otherwise the maximum input frequency is limited to 16 MHz.



<sup>‡24/48</sup>\_SEL and PCI/LDT\_SEL inputs operate independently from each other and the frequency of the corresponding bus, as shown in detail for the 200-MHz configuration.

# **FUNCTION TABLES (Continued)**

#### **SPREAD SPECTRUM**

INPU	Т	
0	0	Spread spectrum disabled
Spread	1	Spread spectrum enabled, –0.5% at CPU/CPU, PCI/LDT, PCI_F, PCI

### **DEVICE ENABLE FUNCTIONS**

SME CONTRO					II	NPUTS	3							OUT	PUTS					INTE	RNAL
FS4 (Byte 0, Bit 5)	FS3 (Byte 0, Bit 4)	24/48_SEL	PCI/LDT_SEL	LDT_Stop	PCI_Stop	FS2	FS1	FS0	SPREAD	NIX	CPU	± CPU ‡	PCI_F	PCI	PCI/LDT	USB	FDC	REF	Comment	CRYSTAL	VCOs
L	L	Х	Χ	Χ	Χ	L	L	L	Х	Xtal	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z	Hi-Z		Off	Off
L	X	Х	Н	Н	Н	Н	Χ	Χ	Χ	Xtal	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	Н	Н	L	Н	Χ	Χ	Χ	Xtal	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	L	L	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	Н	L	Н	Н	Χ	Χ	Χ	Xtal	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	Н	L	L	Н	Χ	Χ	Χ	Xtal	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	L	L	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	L	Н	Н	Н	Χ	Χ	Χ	Xtal	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	L	Н	L	Н	Χ	Χ	X	Xtal	$\uparrow\downarrow$	$\uparrow \downarrow$	$\uparrow\downarrow$	L	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow \downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	L	L	Н	Н	Χ	Χ	X	Xtal	$\uparrow\downarrow$	$\uparrow \downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	L	$\uparrow\downarrow$	$\uparrow \downarrow$	$\uparrow\downarrow$		$\uparrow\downarrow$	$\uparrow\downarrow$
L	X	Х	L	L	L	Н	Χ	Χ	Χ	Xtal	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$	L	L	$\uparrow\downarrow$	$\uparrow\downarrow$	$\uparrow\downarrow$		$\uparrow \downarrow$	$\uparrow\downarrow$
L	L	Х	Х	Х	Н	L	L	Н	Х	Н	Н	L	HL	HL	HL	L	L	L		Off	Off
L	L	Х	Н	Χ	Н	L	L	Н	Χ	L	L	Н	HL	HL	HL	L	L	L		Off	Off
L	L	Х	Н	Н	L	L	L	Н	Χ	L	L	Н	HL	L	L	L	L	L	Щ	Off	Off
L	L	Х	L	L	Н	L	L	Н	Χ	Н	Н	L	HL	HL	L	L	L	L	MODE	Off	Off
L	L	Х	L	Н	L	L	L	Н	Χ	L	L	Н	HL	L	HL	L	L	L		Off	Off
L	L	Х	Н	Н	L	L	Н	Χ	Χ	Н	Н	L	HL	L	L	HL	HL	HL	BYPASS	Off	Off
L	L	Х	Н	Н	L	L	Н	Χ	Χ	L	L	Н	HL	L	L	HL	HL	HL		Off	Off
L	L	Х	Н	L	Н	L	Н	Χ	Χ	Н	Н	L	HL	HL	HL	HL	HL	HL	PL	Off	Off
L	L	Х	Н	L	Н	L	Н	Χ	Χ	L	L	Н	HL	HL	HL	HL	HL	HL		Off	Off
L	L	Х	L	Н	L	L	Н	Χ	Χ	L/H	L/H	H/L	HL	L	HL	HL	HL	HL		Off	Off
L	L	Х	L	L	Н	L	Н	Χ	Χ	L/H	L/H	H/L	HL	HL	L	HL	HL	HL		Off	Off
L	L	Х	L	L	L	L	Н	Χ	Χ	L/H	L/H	H/L	HL	L	L	HL	HL	HL		Off	Off

<sup>†</sup> SMBus bits set to their reset values

#### **OUTPUT BUFFER SPECIFICATIONS**

BUFFER NAME	V <sub>DD</sub> RANGE (V)	IMPEDANCE $(\Omega)$	LUMPED TEST LOAD		
CPU	3.135 – 3.465	40	10 pF		
PCI, PCI_F, LDT	3.135 – 3.465	25	30 pF		
REF, USB, FDC	3.135 - 3.465	35	20 pF		

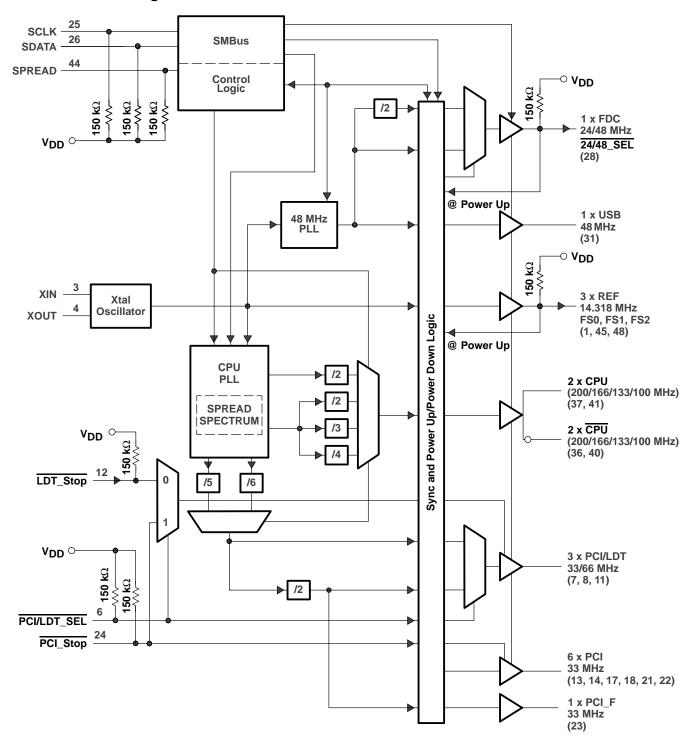


<sup>‡</sup> Hi-Z will have LOW state if external load circuit is applied, CPU and CPU are push-pull type outputs.

<sup>↑↓</sup> Outputs toggle at the selected frequency according to the Device Frequency Select FunctionS table above.

HL device output state is undefined, either L or H. It is L if Xin is held static at L or H before the bypass mode is selected.

# functional block diagram



# **Terminal Functions**

TERMINA	L			
NAME	NO.	1/0	DESCRIPTION	
CPU[0:1],CPU[0:1]	41, 37 40, 36	0	3.3-V, differential CPU clock outputs <u>CPU</u> Clock Outputs 0 and 1: CPU push-pull true clock outputs of the differential pair <u>CPU</u> Clock Outputs 0 and 1: CPU push-pull complementary clock outputs of the differential pair	
FS[0:2] & REF[0:2]	1, 48, 45	I/O	$3.3$ V, $14.318$ -MHz clock outputs Frequency Select inputs: Power–on strapping to set device operating frequency as described in the Device Frequency Select Functions table. These inputs have $150$ -k $\Omega$ internal pullup resistors. Low = 0, High = $1.3.3$ -V reference clock outputs: Fixed clock output at $14.318$ MHz	
GND	5, 10, 15, 20, 27, 30, 34, 39, 47	G	Power Connection: Connected to VSS. Used to ground digital portions of the chip	
GNDA	42	G	Analog GND: Connected to VSS through filter. Used to ground the main CPU-PLL on the chip	
GNDF	33	G	Analog GND for 48-MHz PLL: Connected to VSS through filter. Used to ground the 48-MHz PLL on the chip	
LDT_Stop	12	I	Control for 66-MHz PCI clocks: Active LOW control input to halt all 66-MHz PCI clocks except the free-running clock. This input has a 150-k $\Omega$ internal pullup resistor. Once this input has been asserted, PCI/LDT outputs if operating at 66-MHz must stop in the low state within 1 $\mu$ s. Low = stop, High = running	
PCI[0:5]	13, 14, 17, 18, 21, 22	0	3.3-V PCI clock outputs divided down from CPU-PLL 3.3-V PCI clock outputs: PCI clocks operate at 33 MHz.	
PCI_F	23	0	3.3-V, 33-MHz clocks divided down from CPU-PLL 3.3-V Free-Running PCI clock output: The free-running PCI clock pin operates at 33 MHz. The free-running PCI clock is not turned off when PCI_Stop# is activated LOW.	
PCI/LDT[0:2]	7, 8, 11	0	3.3-V PCI 33-MHz or LDT 66-MHz <u>outputs: This group</u> of outputs is selectable between 33 MHz and 66 MHz based upon the state of PCI/LDT_SEL. When running at 66 MHz these outputs are for use as reference clocks to LDT devices.	
PCI/LDT_SEL	6	I	PCI 33-MHz/LDT 66-MHz Select: This input selects the output frequency of PCI/LDT outputs to either 33 MHz or 66 MHz. This is a dedicated input pin to avoid corruption of the input state due to PCI add-in cards that may have termination resistors on the input clocks. This input has a 150-k $\Omega$ internal pullup resistor. Low = 66-MHz outputs, High = 33-MHz outputs	
PCI_Stop	24	I	3.3-V LVTTL-compatible input for $\overline{\text{PCI\_Stop}}$ active low Control for 33-MHz PCI clocks: Active LOW control input to halt all 33-MHz PCI clocks except the free-running clock. This input has a 150-k $\Omega$ internal pullup resistor. Once this input has been asserted, the PCI outputs and PCI/LDT outputs operating at 33 MHz must stop in the low state within 1 $\mu$ s. Low = stop, High = running	
SCLK	25	I	SMBus compatible SCLK. Clock pin for SMBus circuitry (SMBus revision 1.1). This input has an internal pull-up resistor of 150 k $\Omega$ . SCLK is a 3.6-V tolerant signal input. High impedance at power down is not supported.	
SDATA	26	I/O	SMBus compatible SDATA Data pin for SMBus circuitry (SMBus revision 1.1). This output is open drain and has an internal pullup resistor of 150 k $\Omega$ . SDATA is a 3.6V tolerant signal IO. High impedance at power down is not supported.	
SPREAD	44	I	Spread Spectrum Clocking Enable: Power-on strapping to set spread spectrum clocking as enabled or disabled. This input allows the default spread spectrum clocking mode to be enabled or disabled upon power up. This input has a 150-kΩ internal pullup resistor.  Low = disable, High = enable. Note that all Athlon and Hammer systems are recommended to use SSC; therefore, the default of this pin is enabled and should only be turned off for debug and test purposes.	
USB	31	0	3.3-V, fixed 48-MHz non-SSC clock output 3.3-V USB clock output: Fixed clock output at 48 MHz	
V <sub>DD</sub>	2, 9, 16, 19, 29, 35, 38, 46	9, 35,		



# **Terminal Functions (Continued)**

TERMINAL	_		DESCRIPTION			
NAME	NO.	1/0	DESCRIPTION			
V <sub>DDA</sub>	43	Р	Analog $V_{\mbox{\scriptsize DD}}$ : Connected to 3.3-V power supply through filter. Used to supply the main CPU-PLL on the chip			
V <sub>DDF</sub>	32	Р	Analog $\mbox{V}_{\mbox{DD}}$ for 48-MHz PLL: Connected to 3.3-V power supply through filter. Used to supple 48-MHz PLL on the chip			
XIN	3	I	Crystal input – 14.318 MHz Crystal Connection or External Reference: Reference crystal input or external reference clock input. This pin includes an internal 36-pF load capacitance to eliminate the need for an external load capacitor.			
XOUT	4	0	Crystal output – 14.318 MHz Crystal Connection: Reference crystal feedback. This output includes an internal 36-pF load capacitance to eliminate the need for an external load capacitor.			
24/48_SEL & FDC	28	I/O	3.3-V super I/O clock output: The super I/O clock can be strapped for 24 MHz or 48 MHz. This input has a 150-k $\Omega$ internal pullup resistor. Low = 48-MHz output, High = 24-MHz output			

# connecting SCLK and SDATA to 5-V SMBus signals

SCLK and SDATA of CDC960 have been designed to work within a 3.3-V supply voltage environment only. In order to connect SCLK and SDATA to a 5-V SMBus configuration, external circuitry is required. A simple and inexpensive solution is to use clamping diodes. Two approaches are recommended for this solution:

## 1. Using Zener diode to clamp to GND in reverse-biased direction

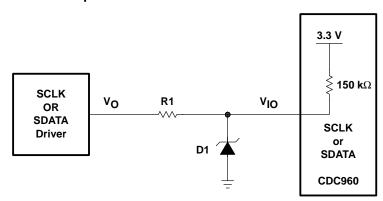


Figure 1. SCLK SDATA Connection to 5-V SMBus Using Zener Diode

Zener diode D1 in Figure 1 is chosen such that the Zener voltage ( $V_{ZK}$ ) cannot exceed 300 mV above  $V_{DD}$  of the CDC960. The minimum value of  $V_{ZK}$  must be greater than 2.1 V to meet minimum requirement for  $V_{IH}$  of the CDC960. The value of R1 is chosen to satisfy requirements both for  $I_{OH}$  of the driver of SCLK and SDATA and for  $V_{OL}$  and  $I_{OL}$  of SDATA of CDC960.

$$I_{OH} \le \frac{V_O - V_{IO}}{\left(R1 + R_S\right)}$$
 (For the driver of SCLK and SDATA.  $R_s$  is the source driver impedance.) (1)

$$2 \text{ mA} \le \frac{0.8 \text{ V}}{\text{R1} + \text{Z}_{\text{O}}} \le 6 \text{ mA}$$
 (For a SDATA of CDC960,  $25 \Omega < \text{Z}_{\text{O}} < 47 \Omega$ ) (2)

$$\left( \text{R1 + Z}_{\text{O}} \right) \times 1.75 \, \text{mA} < 0.4 \, \text{V} \quad \text{(For a SDATA of CDC960, } 25 \, \Omega < Z_{\text{O}} < 47 \, \Omega \text{)} \tag{3}$$



# connecting SCLK and SDATA to 5-V SMBus signals (continued)

There are many manufacturers making Zener diodes that can be used for this application. Panasonic MA8033 and Vishay BZX84C3V3 that have 3.1 V <  $V_Z$  < 3.5 V can be used for this application. In this case R1 is recommended as 150  $\Omega$ .

The worst  $I_{OH}$  in equation (1) is 16 mA when  $V_{OH}$  = 5.5 V,  $R_S$  = 0  $\Omega$  and  $V_{IO}$  = 3.1 V.

The current in equation (2), between 4 mA and 4.6 mA, satisfies the requirement.

Equation (3) is also satisfied with the selected R1 and diode.

### 2. Clamping Diode to V<sub>DD</sub>

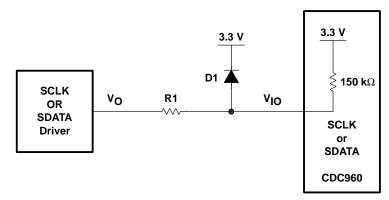


Figure 2. SCLK SDATA Connection to 5-V SMBus Using Clamping Diode to V<sub>DD</sub>

Diode D1 in Figure 2 should have a small forward voltage ( $V_F$ ). Ideally, we want  $V_F$  to be less than 300 mV to meet the input voltage requirement of the CDC960. International IOR Rectifier has a device (part number 10BQ015) with maximum  $V_F$  of 350 mV at 1.0 A. Using the 10BQ015 with R1 = 150  $\Omega$ , the worst-case  $I_{OH}$  is calculated using equation (4).

$$I_{OH} = \frac{5.5 \text{ V} - (3.0 - 0.35) \text{ V}}{150 \Omega} = 14 \text{ mA}$$
 (4)

The calculation for equations (2) and (3) is the same as in part 1.

When using the configuration in Figure 2, the power supply is required to have a capability of sinking current. The total amount of sinking current is dependent on the overall load connected to that power supply.

Using the above interface circuitry with a high-impedance source, the available high-level voltage on the SMBus is limited to about (Vzk) for the configuration in Figure 1 and ( $VDD_{(CDC960)} + V_{F(D1)}$ ) for the configuration in Figure 2. One has to choose which option best fits a given SMBus configuration.

Actually, the typical SMBus configuration is an open-drain configuration with pullup resistors to the corresponding power supply. It does not require a 5-V SMBus driver that has a low impedance to drive the CDC960 SMBus ports with its additional components as shown in Figure 1 and Figure 2. The external components are not needed if the pullup resistors of the SMBus are directly connected to a voltage equal to the supply voltage of the CDC960 (typically 3.3 V). This pullup resistor connection is strongly recommended.



# CDC960

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## power-up sequences

Sampled inputs are: FS0, FS1, FS2 and 24/48\_SEL.

State S1 is an analog controlled delay derived from internal reference voltages to ensure that a valid input state is captured. There is no specific delay in this state after power up.

Figure 3 shows the symbolic sequence of the CDC960 during power up. States S0–S4 are required to ensure proper configuration and operation of the device functions.

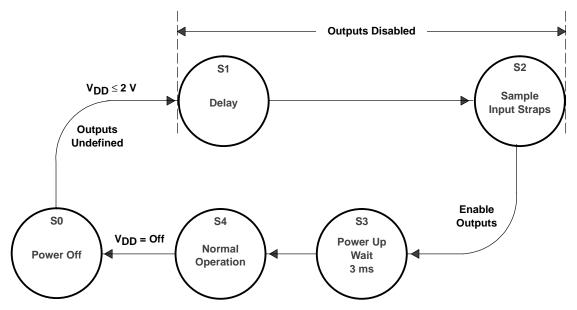


Figure 3. Power-Up State Transitions

### **SMBus serial interface**

The following section describes the SMBus interface programming.

In general the CDC960 SMBus protocol supports only block write and block read operations.

#### SMBus device address

A6	A5	A4	А3	A2	A1	A0	R/₩
1	1	0	1	0	0	1	0

0 = write to CDC960

1 = read from CDC960

## writing to the SMBus interface

- 1. Send the address D2<sub>(H)</sub> and validate the acknowledge from the slave.
- 2. Send the dummy byte as a command code and validate the acknowledge from the slave.
- 3. Send the number of data bytes to write and validate the acknowledge from the slave.
- 4. Write the desired data bytes to registers and validate the acknowledge from the slave for each data byte.

Clock Generator Addr (7 bits)		+8 bits dummy		+8 bits byte					
A(6:0) & R/W	ACK	command code	ACK	count	ACK	Data byte 0	ACK	Data byte N	ACK
D2(H)									



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# SMBus serial interface (continued)

# reading the SMBus interface, using address pre-phase

- 1. Send the address D2<sub>(H)</sub> and validate the acknowledge from the slave.
- 2. Send dummy byte as command code and validate the acknowledge from the slave.
- 3. Send repeated start condition followed by address D3<sub>(H)</sub> and validate the acknowledge from the slave.
- 4. The slave returns the number of bytes it is going to send (byte count) and validates the acknowledge from the master.
- 5. Read back the desired data bytes and validate the acknowledge sent by the master for each data byte.

Clock Generator Addr (7 bits)	ACK	+8 bits dummy	ACK	Repeated	Clock Generator Addr (7 bits)	ACK	+8 bit byte	ACK by	Data	ACK by	Data	ACK by
A(6: <u>0)</u> & R/W	ACK	command code	AOR	Start	A(6: <u>0)</u> & R/W	AOR	count	master	byte 0	master	byte N	master
D2(H)					D3(H)							

### reading the SMBus interface, using direct read

- 1. Send the address D3<sub>(H)</sub> and validate the acknowledge from the slave.
- 2. The slave returns the number of bytes it is going to send (byte count) and validates the acknowledge from the master.
- 3. Read back the desired data bytes and validate the acknowledge sent by the master for each data byte.

ĺ	Clock								ĺ
	Generator Addr (7 bits)	ACK	+8 bit byte	ACK by master	Data byte 0	ACK by master	Data byte N	ACK by master	
ĺ	A(6:0)& R/W		count	·		·	•	·	l
ſ	D3(H)								١

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# SMBus configuration command bitmap

Byte 0: Frequency and Spread Spectrum Control Register (see Note 1)

(H = Enable, L = Disable)

BIT	TYPE	PUD†			DESCRIPTION	N		PIN AFFECTED (WRITE OPERATION)	SOURCE PIN (READ OPERATION)
7	R/W	L	Ву		odification of all confi	bit after a 1 has been written guration registers until the d		1	Register value
						a software programmable co table for SSC is as follows:	n-		
				Spread (ext. Pin)	Byte0, Bit6	SSC Function			
6	R/W	L		L	L	Disabled		_	Register value
				L	Н				
				Н	L				
				Н	Н	Enabled			
5	R/W	L	FS	34 (corresponds to fre	equency selection ta	ble)			Register value
4	R/W	L	FS	3 (corresponds to fre	equency selection ta	ble)			Register value
3	R/W	Externally selected <sup>‡</sup>		62 (corresponds to fre n be set differently th		ole). If write is enabled, this Indition.	bit	1	45 at power up
2	R/W	Externally selected‡		S1 (corresponds to free n be set differently the		ole). If write is enabled, this Indition.	bit	_	48 at power up
1	R/W	Externally selected‡		60 (corresponds to fre n be set differently th		bit	_	1 at power up	
0	R/W	L	co ca ha til int	nfiguration registers a tion of all configurations s been written to Byte the device power cyc	and subsequent 0s we on registers except thing e0, Bit 7, all modification cles. Note also, that I. However, unless the	er up enables modification of rritten to this bit disable modi s single bit. Note that when a on is permanently disabled u block write transactions to the e interface has been previous	ifi- i 1 n- ne	_	Register value

<sup>†</sup> PUD = Power-up condition

NOTE 1: Byte0, Bit0 controls the write enable status for the device SMBus. If a 1 is written to Byte0, Bit0, the SMBus registers are write enabled. Once write has been enabled, a new block write protocol must be sent to the device to program the desired register values. Once after power up a 1 is written to Byte0, Bit0, the device functionality is according to the settings of the different registers. E.g., the device function table is according to setting of Bits[1...6] of Byte0 and other functions are according to corresponding SMBus register settings.

If a 0 is written to Byte0, Bit0, write is disabled and the device function is according to the previous settings of the last write cycle.

#### Byte 1: PCI Clock Control Register

(H = Enable, L = Disable)

ВІТ	TYPE	PUD†	DESCRIPTION	PIN AFFECTED (WRITE OPERATION)	SOURCE PIN (READ OPERATION)
7	R/W	Н	PCI/LDT1 enable	8	Register value
6	R/W	Н	PCI/LDT0 enable	7	Register value
5	R/W	Н	PCI5 enable	22	Register value
4	R/W	Н	PCI4 enable	21	Register value
3	R/W	Н	PCI3 enable	18	Register value
2	R/W	Н	PCI2 enable	17	Register value
1	R/W	Н	PCI1 enable	14	Register value
0	R/W	Н	PCI0 enable	13	Register value

<sup>†</sup> PUD = Power-up condition



<sup>&</sup>lt;sup>‡</sup> The value of this bit is according to level applied to corresponding device pin at power up.

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# SMBus configuration command bitmap (continued)

# Byte 2: PCI Clock USB FDC and REF Control Register

(H = Enable, L = Disable)

BIT	TYPE	PUD†	DESCRIPTION	PIN AFFECTED (WRITE OPERATION)	SOURCE PIN (READ OPERATION)
7	R/W	Н	CPU1 enable‡	36, 37	Register value
6	R/W	Н	CPU0 enable‡	40, 41	Register value
5	R/W	Н	REF2 enable	45	Register value
4	R/W	Н	REF1 enable	48	Register value
3	R/W	Н	REF0 enable	1	Register value
2	R/W	Н	FDC (24_48 MHz) enable	28	Register value
1	R/W	Н	USB enable	31	Register value
0	R/W	Н	PCI/LDT2 enable	11	Register value

<sup>†</sup> PUD = Power-up condition

#### Byte 3: PCI Clock Free Running Control Register (H = Free running, L = controlled by $\overline{PCI\_Stop}/\overline{LDT\_Stop}$ ))

BIT	TYPE	PUD <sup>†</sup>	DESCRIPTION	PIN AFFECTED (WRITE OPERATION)	SOURCE PIN (READ OPERATION)	
7	R/W	L	PCI/LDT1 free-running enable§	8	Register value	
6	R/W	L	PCI/LDT0 free-running enable§	7	Register value	
5	R/W	L	PCI5 free-running enable§	22	Register value	
4	R/W	L	PCI4 free-running enable§	21	Register value	
3	R/W	L	PCI3 free-running enable§	18	Register value	
2	R/W	L	PCI2 free-running enable§	17	Register value	
1	R/W	L	PCI1 free-running enable§	14	Register value	
0	R/W	L	PCI0 free-running enable§	13	Register value	

<sup>†</sup>PUD = Power-up condition



<sup>‡</sup> If a CPU clock is disabled by setting its control bit (bit 6 or bit 7) low, both the CPU and CPU outputs for the disabled clock are set low.

<sup>§</sup> The above individual free-running enable/disable controls are intended to allow individual clock outputs to be made free running. A clock output that has its free-running bit enabled (set to H) is not turned off with the assertion of either PCI\_Stop or LDT\_Stop. If a particular bit is disabled in Byte1, the Byte1 settings overwrite the Byte3 settings.

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# SMBus configuration command bitmap (continued)

# Byte 4: Pin Latched/Real Time State Control Register (see Note 2)

(H = Enable, L = Disable)

ВІТ	TYPE	PUD†	DESCRIPTION	DESCRIPTION PIN AFFECTED (WRITE OPERATION)	
7	R/W	Н	PCI_F enable	23	Register value
6	R	Externally selected‡§	SPREAD actual pin state	SPREAD actual pin state	
5	R	Externally selected‡	24/48_SEL pin power up latched state		28 at power up
4	R	Externally selected‡§	PCI/LDT_SEL actual pin state		6
3	R	Externally selected‡	FS2 power-up latched pin state		45 at power up
2	R	Externally selected‡	FS1 power-up latched pin state		48 at power up
1	R	Externally selected‡	FS0 power-up latched pin state		1 at power up
0	R/W	L	PCI/LDT2 free-running enable¶	11	Register value

<sup>†</sup> PUD = Power-up condition

NOTE 2: Byte4 holds the power-up information for pins latched at power up. In the case that an unintentional write has been made to these bits of Byte4, the SMBus write is ignored; the bits always return the power-up latched value during an SMBus read operation.

This does not relate to the bits which hold the actual (current) pin state. Those bits can not be overwritten by software in order to get the hardware setting states back via software.

#### Byte 5: Vendor Identification Register

(H = Enable, L = Disable)

				,	<u> </u>
ВІТ	TYPE	PUD <sup>†</sup>	DESCRIPTION PIN AFFECTED (WRITE OPERATION) (		SOURCE PIN (READ OPERATION)
7	R	Н	Manufacturer ID (MSB)	-	Returns H
6	R	Н	Manufacturer ID	Manufacturer ID – Retu	
5	R	Н	Manufacturer ID, TI is shown for vendor ID = 111	_	Returns H
4	R	L	Device revision ID (MSB)	_	Returns L
3	R	L	Device revision ID	-	Returns L
2	R	L	Device revision ID	-	Returns L
1	R	L	Device revision ID	-	Returns L
0	R	Н	Device revision ID, device revision: 00001	_	Returns H

† PUD = Power-up condition



<sup>‡</sup> The value of this bit is determined by the level applied to the corresponding device pin at power up.

<sup>§</sup> If the SMBus is in read mode, and the byte-count byte is being sent, the device input pin is sampled again at the falling edge of SCLK at the same state as the acknowledge state for the byte count that is initiated by SCLK↓.

The above individual free running enable/disable controls are intended to allow individual clock outputs to be made free running. A clock output that has its free-running bit enabled (set to H) is not turned off with the assertion of either PCI\_Stop or LDT\_Stop. If a particular bit is disabled in Byte2, the Byte2 settings overwrite the Byte4 settings.

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# SMBus configuration command bitmap (continued)

# **Byte 6: Byte Count Control Register**

(H = Enable, L = Disable)

BIT	TYPE	PUD <sup>†</sup>	DESCRIPTION	DESCRIPTION PIN AFFECTED (WRITE OPERATION)		
7	R/W	L	Byte count bit, MSB	-	Register value	
6	R/W	L	Byte count bit	-	Register value	
5	R/W	L	Byte count bit	-	Register value	
4	R/W	L	Byte count bit	-	Register value	
3	R/W	L	Byte count bit	-	Register value	
2	R/W	Н	Byte count bit	-	Register value	
1	R/W	Н	Byte count bit	-	Register value	
0	R/W	Н	Byte count bit, LSB	-	Register value	

<sup>†</sup> PUD = Power-up condition

# Byte 7: Vendor Specific Register (reserved)

(H = Enable, L = Disable)

ВІТ	TYPE	PUD <sup>†</sup>	DESCRIPTION	PIN AFFECTED (WRITE OPERATION)	SOURCE PIN (READ OPERATION)
7	R/W	L	Must be set to L during the byte write	-	Register value
6	R/W	L	Must be set to L during the byte write	-	Register value
5	R/W	L	Must be set to L during the byte write	-	Register value
4	R/W	L	Must be set to L during the byte write	-	Register value
3	R/W	L	Must be set to L during the byte write	-	Register value
2	R/W	L	Must be set to L during the byte write	-	Register value
1	R	L	Must be set to L during the byte write	-	Register value
0	R	L	Must be set to L during the byte write	-	Register value

<sup>†</sup> PUD = Power-up condition



# CDC960 200-MHz CLOCK SYNTHESIZER/DRIVER

# WITH SPREAD SPECTRUM CAPABILITY AND DEVICE CONTROL INTERFACE

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# SMBus configuration command bitmap (continued)

# Byte 8: Vendor Specific Register (reserved)

(H = Enable, L = Disable)

BIT	TYPE	PUD†	DESCRIPTION PIN AFFECTED (WRITE OPERATION)		SOURCE PIN (READ OPERATION)
7	R/W	L	Trigger single pulse at the L-to-H transition of this bit after an SMBus write cycle completes. This bit must be written back to L in order to trigger a following pulse with a new L-to-H transition at the completion of a write protocol.	CPU, CPU	Register value
6	R/W	L	Single-pulse ARM bit H = enable, L = disable single-pulse feature	-	Register value
5	R/W	L	Must set to L during the byte write	-	Register value
4	R/W	L	Must set to L during the byte write	ı	Register value
3	R/W	L	Must set to L during the byte write	ı	Register value
2	R/W	Ĺ	Must set to L during the byte write	<ul> <li>Register value</li> </ul>	
1	R/W	L	Must set to L during the byte write		Register value
0	R/W	L	Must set to L during the byte write	-	Register value

<sup>†</sup> PUD = Power-up condition

## Single-pulse initialization

- 1. Device is in normal operating mode (frequencies selected by FS[4:0] as usual).
- 2. Put device into SMBus mode (set write enable bit according to specification).
- 3. Put device into required operating mode via the SMBus.
- 4. Set Byte8/Bit6 to H. Byte8 is a TI control byte, Bit6 is the ARM bit.
  - a. The device continues running as in the normal operating mode, but the CPUx/CPUx outputs are pulled to low/high, respectively; i.e., the clock is low.
  - b. All other clocks (PCI, LDT66, USB, 48-MHz, REFCLOCK) continue running as long as they are not disabled by the SMBus or other means.
- 5. Set Byte8/Bit7 to H. Byte8/Bit 7 is the SHOOT bit.
  - a. The device recognizes a rising edge on this bit and sends a single high pulse on CPUx. The CPUx output is complementary (low). The pulse duration depends on frequency settings for the CPU-BUS (half of the period).
  - b. CPU1 or CPU0 can still be enabled/disabled via the SMBus as usual.
- 6. Set Byte8/Bit7 back to L for the next shot.
  - a. Because the device only detects  $L \rightarrow H$  transitions, this bit must be reset to L.
- 7. Now the device is ready for the next pulse (write H to Byte8/Bit7).
- 8. When setting the ARM bit to L, the single-shot feature is disabled and the device runs as usual.



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# spread spectrum clock (SSC) implementation for CDC960

Simultaneously switching at a fixed frequency generates a significant power peak at the selected frequency, which in turn causes an EMI disturbance to the environment. The purpose of the internal frequency modulation of the CPU-PLL is to distribute the energy to many different frequencies, thus reducing the power peak.

A typical characteristic for a single-frequency spectrum and a modulated-frequency spectrum is shown in Figure 4.

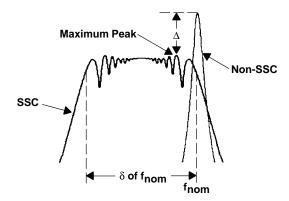


Figure 4. Frequency Power Spectrum With and Without the Use of SSC

The modulated spectrum has its distribution to the left side of the single-frequency spectrum, which indicates a *down-spread modulation*.

The peak reduction depends on the modulation scheme and modulation profile. System performance and timing requirements are the limiting factors for actual design implementations. The implementation is driven to keep the average clock frequency close to its upper specification limit. The modulation amount is set to -0.5%.

In order to allow a downstream PLL to follow the frequency modulated signal, the bandwidth of the modulation signal is limited in order to minimize SSC-induced tracking skew jitter.



# absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Supply voltage range, V <sub>DD</sub>	–0.5 V to 4.6 V
Input voltage range, V <sub>I</sub> (see Note 3)	$0.5 \text{ V to V}_{DD} + 0.5 \text{ V}$
Voltage range applied to any output in the high-impedance state or power-off state,	
V <sub>O</sub> (see Note 3)	$0.5 \text{ V to V}_{DD} + 0.5 \text{ V}$
Current into any output in the low state, IO	
Input clamp current: $I_{ K }(V_{ } < 0)$	18 mA
$I_{IK}(V_I > V_{DD})$	18 mA
Output clamp current: $I_{OK}(V_O < 0)$	–50 mA
$I_{OK}(V_O > V_{DD})$	50 mA
Package thermal impedance, θ <sub>JA</sub> (see Note 4)	95°C/W
Maximum power dissipation at $T_A = 55^{\circ}C$ (in still air) (see Note 5)	1.0 W
Operating free-air temperature range, T <sub>A</sub>	0°C to 70°C
Storage temperature range, T <sub>stq</sub>	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 3. The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.
  - 4. The package thermal impedance is calculated in accordance with EIA/JEDEC Std JESD51, except for the through-hole packages, which use a trace length of zero. The absolute maximum power dissipation allowed at T<sub>A</sub> = 55°C (in still air) is 1.0 W.
  - 5. The maximum package power dissipation is calculated using a junction temperature of 150°C and a board trace length of 750 mils. For more information, refer to the *Package Thermal Considerations* application note in the *ABT Advanced BiCMOS Technology Data Book*, literature number SCBD002.

#### **DISSIPATION RATING TABLE**

PACKAGE $T_A \le 25^{\circ}C$ POWER RATING		OPERATING FACTOR <sup>‡</sup> ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING
DL	1.3 W	10.7 mW/°C	0.85 W

<sup>&</sup>lt;sup>‡</sup> This is the inverse of the traditional junction-to-case thermal resistance ( $R_{\theta JA}$ ) and uses a board-mounted device at 95°C/W.

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# recommended operating conditions (see Notes 4 and 5)

		MIN	NOM†	MAX	UNIT	
Supply voltages, V <sub>DD</sub>	3.3 V	3.135		3.465	V	
	PCI_Stop, LDT_Stop, PCI/LDT_SEL, 24/48_SEL, FS0, FS1, FS2, SPREAD	2		V <sub>DD</sub> +0.3		
High-level input voltage, VIH	SDATA, SCLK (see Note 6)	2.0		V <sub>DD</sub> +0.3	V	
	XIN	2.0		V <sub>DD</sub> +0.3		
	PCI_Stop, LDT_Stop, PCI/LDT_SEL, 24/48_SEL, FS0, FS1, FS2, SPREAD	-0.3		0.8	.,	
Low-level input voltage, V <sub>IL</sub>	SDATA, SCLK (see Note 6)	-0.3		1.08	V	
	XIN	-0.3		0.5		
Input Voltage, V <sub>I</sub>	PCI_Stop, LDT_Stop, PCI/LDT_SEL, 24/48_SEL, FS0, FS1, FS2, SPREAD	-0.3		V <sub>DD</sub> +0.3	V	
	SDATA, SCLK (see Note 6)	-0.3		V <sub>DD</sub> +0.3		
	CPU			-18		
High-level output current, IOH	USB, FDC, REF			-12	mA	
	PCI, LDT			-12		
	CPU			18		
	USB, FDC, REF		9			
Low-level output current, IOL	PCI, LDT			9	mA	
	SDATA			4		
Input resistance to V <sub>DD</sub> , R <sub>I</sub>	PCI_Stop, LDT_Stop, PCI/LDT_SEL, 24/48_SEL, FS0, FS1, FS2, SPREAD	100		220	kΩ	
, 55, 1	SDATA, SCLK	100		220		
Reference frequency, f(XIN) <sup>‡</sup>	PLL BY-PASS MODE	0		200	MHz	
Crystal frequency, f(XTAL)§	NORMAL MODE	10	14.31818	16	MHz	
SCLK frequency, f(SCLK)¶				100	kHz	
Bus free time, t <sub>(BUS)</sub> ¶		4.7			μs	
START setup time, t <sub>SU(START)</sub>	T	4.7			μs	
START hold time, th(START)		4.0			μs	
SCLK low pulse duration, tw(SC	CLL)¶	4.7			μs	
SCLK high pulse duration, tw(S	CLH) <sup>¶</sup>	4.0			μs	
SDATA input rise time, tr(SDATA	n/1			1000	ns	
SDATA input fall time, tf(SDATA	Î ,			300	ns	
SDATA setup time, t <sub>SU(SDATA)</sub>		250			ns	
SDATA hold time, th(SDATA)		5			ns	
STOP setup time, t <sub>su(STOP)</sub> ¶		4			μs	
Operating free-air temperature,	TA	0		70	°C	

<sup>†</sup> All typical values are measured at their respective nominal VDD.

<sup>6.</sup> The CMOS-level inputs fall within these limits:  $V_{IH}min = 0.7 \times V_{DD}$  and  $V_{IL}max = 0.3 \times V_{DD}$ .



<sup>‡</sup> Reference frequency is a test clock driven on the XIN input during the device test mode and normal mode. In test mode, XIN can be driven externally up to f(XIN) = 0 MHz to 200 MHz. If XIN is driven externally, XOUT is floating.

<sup>§</sup> This is a fundamental crystal with f<sub>O</sub> = 14.31818 MHz and 18 pF load in a parallel resonance application (Pierce-type oscillator)

<sup>¶</sup> This conforms to SMBus Specification, Version 1.1.

NOTES: 4. The package thermal impedance is calculated in accordance with EIA/JEDEC Std JESD51, except for the through-hole packages, which use a trace length of zero. The absolute maximum power dissipation allowed at T<sub>A</sub> = 55°C (in still air) is 1.0 W.

<sup>5.</sup> The maximum package power dissipation is calculated using a junction temperature of 150°C and a board trace length of 750 mils. For more information, refer to the *Package Thermal Considerations* application note in the *ABT Advanced BiCMOS Technology Data Book*, literature number SCBD002.

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER			TEST CO	ONDITIONS	MIN .	TYP†	MAX	UNIT
$V_{IK}$	Input clamp voltage	е	$V_{DD} = 3.135 \text{ V},$	I <sub>I</sub> = -18 mA		-0.7	-1.2	V
		XIN	$V_{DD} = 3.465 \text{ V},$	V <sub>I</sub> = 2.0		2.5		
ΙΗ	High-level input	PCI/LDT_SEL, PCI_Stop, LDT_Stop, SPREAD	V <sub>DD</sub> = 3.465 V,	$V_I = V_{DD}$			5	μΑ
	current	FS0, FS1, FS2, 24/48_SEL	$V_{DD} = 3.465 \text{ V},$	$V_I = V_{DD}$			5	
		SDATA, SCLK	$V_{DD} = 3.465 \text{ V},$	$V_I = V_{DD}$			5	
		XIN	$V_{DD} = 3.465 \text{ V},$	V <sub>I</sub> = GND		-1.5		mA
կլ‡	Low-level input current	PCI/LDT_SEL, PCI_Stop, LDT_Stop, SPREAD	V <sub>DD</sub> = 3.465 V,	V <sub>I</sub> = GND			-50	
	current	FS0, FS1, FS2, 24/48_SEL	$V_{DD} = 3.465 \text{ V},$	V <sub>I</sub> = GND			-50	μΑ
		SDATA, SCLK	$V_{DD} = 3.465 \text{ V},$	V <sub>I</sub> = GND			-50	
loz	High-impedance-state output current		$V_{DD} = 3.465 \text{ V},$	$V_O = V_{DD}$ or GND			±5	μΑ
loz	High-impedance-st	tate output current, SDATA	$V_{DD} = 3.465 \text{ V},$	$V_O = V_{DD}$			5	μΑ
I <sub>DD</sub>	Static supply current	All outputs open,	All outputs = low o TEST MODE,	r high, V <sub>DD</sub> = 3.465 V			4.5	mA
		SSC = ON/OFF, C <sub>L</sub> = MAX, LDT = 66 MHz, CPU outputs: TEST LOAD	CPU = 166 MHz,	V <sub>DD</sub> = 3.465 V			180	
I <sub>DD</sub>	Dynamic supply current	All others loaded with corresponding load capacitance only.	CPU = 200 MHz,	V <sub>DD</sub> = 3.465 V			185	mA
		All outputs disabled (LOW)	CPU =166 MHz/ 2 V <sub>DD</sub> = 3.465 V	00 MHz,		45	55	
I <sub>DD(Z)</sub>	High-impedance- state supply current	All outputs open, and outputs are in 3-state	CPU = 200 MHz,	V <sub>DD</sub> = 3.465 V		38	50	mA
Cl				$V_I = V_{DD}$ or GND		2.3	2.7	pF
XIN, XOUT	Input capacitance to GND		V <sub>DD</sub> = 3.3 V	V <sub>I</sub> = 1.5 V	27	29	31	pF
C <sub>XTAL</sub>	Cryatal terminal ca	pacitance (see Note 7)	$V_{DD} = 3.3 \text{ V},$	V <sub>I</sub> = 1.5 V		15		pF

<sup>&</sup>lt;sup>†</sup> All typical values are measured at their respective nominal V<sub>DD</sub>.

<sup>‡</sup>I<sub>IL</sub> is caused by internal pullup resistors.

NOTE 7: This is the corresponding electrical capacitive load for the crystal in this oscillator application (Pierce-type oscillator). Parasitic pin-to-pin capacitance = 2 pF.

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electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

# CPU (200/166/133/100 MHz)

	PARAMETI	ER	TEST CO	NDITIONS	MIN	TYP†	MAX	Unit
,,	LP ale la callacta de callact	_	$V_{DD} = MIN \text{ to } MAX,$	I <sub>OH</sub> = -1 mA	V <sub>DD</sub> – 0.1			.,
VOH	High-level output voltag	е	$V_{DD} = 3.135 V$ ,	I <sub>OH</sub> = -18 mA	2.3			V
,,			$V_{DD} = MIN \text{ to MAX},$	I <sub>OL</sub> = 1 mA			0.05	.,
VOL	Low-level output voltage		$V_{DD} = 3.135 V$ ,	I <sub>OL</sub> = 18 mA			0.6	V
			$V_{DD} = 3.465 V$ ,	V <sub>O</sub> = 2.0 V		-43		
lOH	High-level output current		$V_{DD} = 3.3 V$ ,	V <sub>O</sub> = 1.65 V	-27	-43	-56	mA
			$V_{DD} = 3.135 V$ ,	V <sub>O</sub> = 2.735 V		-14		
			$V_{DD} = 3.465 V$ ,	V <sub>O</sub> = 0.8 V		32		
IOL	Low-level output curren	t	$V_{DD} = 3.3 V$ ,	V <sub>O</sub> = 1.55 V	29	41	52	mA
			$V_{DD} = 3.135 V$ ,	V <sub>O</sub> = 0.4 V		17		
CO	Output capacitance		$V_{DD} = 3.3 V$ ,	$V_O = V_{DD}$ or GND		2.7	3.0	pF
_	0	High state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OH}$	25	40	55	
ZO	Output impedance	Low state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OL}$	25	40	55	Ω

<sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub> values.

# **REF (14.318 MHz)**

	PARAMET	ER	TEST CO	NDITIONS	MIN	TYP <sup>†</sup>	MAX	Unit
.,	18.1.1.1.1.1.1.1.1.1.1.1.1.1.1.1.1.1.1.		$V_{DD} = MIN \text{ to } MAX,$	I <sub>OH</sub> = -1 mA	V <sub>DD</sub> – 0.1			.,
VOH	High-level output volta	ge	$V_{DD} = 3.135 V$ ,	I <sub>OH</sub> = -12 mA	2.5			V
,,	Law law law to alter		$V_{DD} = MIN \text{ to } MAX,$	I <sub>OL</sub> = 1 mA			0.1	
VOL	Low-level output voltag	ge	$V_{DD} = 3.135 V$ ,	I <sub>OL</sub> = 9 mA			0.4	V
			$V_{DD} = 3.465 V,$	V <sub>O</sub> = 2.0 V		-46		
lOH	High-level output curre	ent	$V_{DD} = 3.3 V$ ,	V <sub>O</sub> = 1.65 V	-29	-47	-61	mA
			$V_{DD} = 3.135 V$ ,	V <sub>O</sub> = 2.735 V		-15		
			$V_{DD} = 3.465 V$ ,	V <sub>O</sub> = 0.8 V		33		
lOL	Low-level output curre	nt	$V_{DD} = 3.3 V$ ,	V <sub>O</sub> = 1.65 V	30	42	52	mA
			$V_{DD} = 3.135 V$ ,	V <sub>O</sub> = 0.4 V		17		
CO	Output capacitance		$V_{DD} = 3.3 V$ ,	$V_O = V_{DD}$ or GND		3.2	3.7	pF
_	O to Albania dana	High state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OH}$	22	35	52	
ZO	Output impedance	Low state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OL}$	22	35	52	Ω

<sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub>.

electrical characteristics over recommended operating free-air temperature range (unless otherwise noted) (continued)

# USB (48 MHz), FDC (24 MHz or 48 MHz)

	PARAMET	ER	TEST CO	NDITIONS	MIN	TYP†	MAX	Unit
V	Libertarial actions to called		$V_{DD} = MIN \text{ to MAX},$	I <sub>OH</sub> = -1 mA	V <sub>DD</sub> – 0.1			V
VOH	High-level output voltaç	je	$V_{DD} = 3.135 V$ ,	$I_{OH} = -16 \text{ mA}$	2.4			V
	Laurelaurelauren der eiter er	_	$V_{DD} = MIN \text{ to } MAX,$	I <sub>OL</sub> = 1 mA			0.1	.,
VOL	Low-level output voltag	е	$V_{DD} = 3.135 V$ ,	I <sub>OL</sub> = 9 mA			0.4	V
			$V_{DD} = 3.465 V,$	V <sub>O</sub> = 2.0 V		-46		
loH	OH High-level output current	ent	$V_{DD} = 3.3 V$ ,	V <sub>O</sub> = 1.65 V	-29	-47	-61	mA
			$V_{DD} = 3.135 V$ ,	V <sub>O</sub> = 2.735 V		-15		
			$V_{DD} = 3.465 V$ ,	V <sub>O</sub> = 0.8 V		33		
loL	Low-level output currer	nt	$V_{DD} = 3.3 V$ ,	V <sub>O</sub> = 1.65 V	30	42	52	mA
			$V_{DD} = 3.135 \text{ V},$	V <sub>O</sub> = 0.4 V		17		
CO	Output capacitance		$V_{DD} = 3.3 \text{ V},$	$V_O = V_{DD}$ or GND		3.2	3.7	pF
		High state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OH}$	22	35	52	
ZO	Output impedance	Low state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OL}$	22	35	52	Ω

 $<sup>^{\</sup>dagger}$  All typical values are measured at their nominal  $V_{DD}$ .

## PCI, PCI\_F (33 MHz) and LDT (33 MHz or 66 MHz)

	PARAMETE	R	TEST CO	NDITIONS	MIN	TYP <sup>†</sup>	MAX	Unit
	18.1.1.1.4.4.14		$V_{DD} = MIN \text{ to } MAX,$	I <sub>OH</sub> = -1 mA	V <sub>DD</sub> – 0.1			
VOH	High-level output voltag	е	$V_{DD} = 3.135 V$ ,	I <sub>OH</sub> = -12 mA	2.4			V
			$V_{DD} = MIN \text{ to MAX},$	I <sub>OL</sub> = 1 mA			0.1	
VOL	Low-level output voltage	9	$V_{DD} = 3.135 V,$	I <sub>OL</sub> = 9 mA			0.4	V
			$V_{DD} = 3.465 V,$	V <sub>O</sub> = 2.V		<b>-71</b>		
lOH	High-level output currer	t	$V_{DD} = 3.3 \text{ V},$	V <sub>O</sub> = 1.65 V	-40	-71	-97	mA
			$V_{DD} = 3.135 V,$	V <sub>O</sub> = 2.735 V		-23		
			V <sub>DD</sub> = 3.465 V,	V <sub>O</sub> = 0.8 V		38		
loL	Low-level output curren	t	$V_{DD} = 3.3 \text{ V},$	V <sub>O</sub> = 1.65 V	37	71	100	mA
			$V_{DD} = 3.135 V,$	V <sub>O</sub> = 0.4 V		19		
СО	Output capacitance		V <sub>DD</sub> = 3.3 V,	$V_O = V_{DD}$ or GND		3.2	3.7	pF
		High state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OH}$	12	25	37	_
ZO	Output impedance	Low state	$V_{O} = 0.5 V_{DD}$	$Z_{O} = V_{O}/I_{OL}$	22	25	37	Ω

<sup>†</sup> All typical values are measured at their nominal VDD.

#### **SDATA**

	PARAMETER	TEST CO	NDITIONS	MIN	TYP <sup>†</sup>	MAX	Unit
.,		$V_{DD} = MIN \text{ to MAX},$	I <sub>OL</sub> = 4 mA	35 33 46		0.2	.,
VOL	Low-level output voltage, SDATA	$V_{DD} = 3.135 \text{ V},$	I <sub>OL</sub> = 6 mA			0.4	V
		$V_{DD} = 3.465 \text{ V},$	V <sub>O</sub> = 0.8 V		35		
lOL	Low-level output current, SDATA	$V_{DD} = 3.3 \text{ V},$	V <sub>O</sub> = 1.65 V	33	46	57	mA
		$V_{DD} = 3.135 \text{ V},$	V <sub>O</sub> = 0.4 V		19		
ZO	Output impedance, low state	0.5 V <sub>DD</sub> ,	$Z_{O} = V_{O}/I_{OL}$	25	36	47	Ω
C <sub>I/O</sub>	Input/output capacitance, SDATA	$V_{DD} = 3.3 \text{ V},$	$V_O = V_{DD}$ or GND		4.5	5.1	pF

<sup>&</sup>lt;sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub>.



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# switching characteristics, $\rm V_{DD}$ = MIN to MAX, $\rm T_A$ = 0°C to 70°C

	PARAMETER	TEST CONDITIONS	MIN	TYP <sup>†</sup>	MAX	UNIT
V(over)/V(under)	Overshoot/undershoot	All clocks			±0.7	
tsu(disable)	PCI_Stop↓ or LDT_Stop↓ to PCI_F↑	f(PCI/LDT) = 33/66 MHz to disable PCI/LDT in next cycle (PCI/LDT = low)	10			ns
<sup>t</sup> h(disable)	PCI_Stop↓ or LDT_Stop↓ to PCI_F↑	f(PCI/LDT) = 33/66 MHz to disable PCI/LDT in next cycle (PCI/LDT = low)	0			ns
<sup>t</sup> su(enable)	PCI_Stop↑ or LDT_Stop↑ to PCI_F↑	f(PCI/LDT) = 33/66 MHz to enable PCI/LDT in next cycle (PCI/LDT = high)	10			ns
<sup>t</sup> h(enable)	PCI_Stop↑ or LDT_Stop↑ to PCI_F↑	f(PCI/LDT) = 33/66 MHz to enable PCI/LDT in next cycle (PCI/LDT = high)	0			ns
SSC(midx)	SSC spread amount	f <sub>(CPU)</sub> = 100 MHz to 200 MHz		-0.5		%
f(mod)	SSC modulation frequency	f(CPU) = 100 MHz to 200 MHz		31.4	_	kHz
	Otal Parking Const	FS0, FS1, FS2 or SMBus update		0.03	3	
<sup>t</sup> stab	Stabilization time†	After power up		0.13	3	ms

The Stabilization time is the time required for the integrated PLL circuit to obtain phase lock of its feedback signal to its reference signal. In order for phase lock to be obtained, a fixed-frequency, fixed-phase reference signal must be present at XIN. Until phase lock is obtained, the specifications for propagation delay and skew parameters given in the switching characteristics tables are not applicable. Stabilization time is defined as the time from when VDD achieves its nominal operating level until the output frequency is stable and operating within specification.

switching characteristics,  $V_{DD}$  = 3.135 V to 3.465 V,  $T_A$  = 0°C to 70°C

# CPU, C<sub>L</sub> = 10 pF, R<sub>L</sub> = Test Load

PARAMI	ETER	FROM (INPUT)		TO OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	Unit
tpd1	Propagation delay time	XIN		CPUx	f <sub>(XIN)</sub> ≥ 1 MHz, TEST MODE	3.5		15	ns
t <sub>pd2</sub>	Propagation delay time	SCLK ↑	`	CPUx	Test mode		18		ns
					f(CPU) = 100 MHz	10.0		10.1	
	CPU clock period†				f(CPU) = 133 MHz f(CPU) = 166 MHz f(CPU) = 200 MHz f(CPU) = 100 to 200 MHz f(CPU) = 33.3 MHz	7.5		7.60	
t <sub>C</sub>	СРО сюск репоат				f(CPU) = 166 MHz	18 10.0 10 7.5 7. 6.0 6.0 5.0 1 47 -150 1 5 5 2 100 3			ns
					f(CPU) = 200 MHz	5.0		5.1	
tjit(cc)	Cycle to cycle jitter	Synthesize	er mode	)	f(CPU) = 100 to 200 MHz			160	ps
odc	Duty cycle				f(CPU) = 100 to 200 MHz	47		53	%
tjit(acc)	Accumulated jitter, SSC = ON,	see Note 8			f(CPU) = 100 to 200 MHz	-150		150	ps
tsk(b)	CPU bank skew ↑ edges	CPUx		CPUx	f <sub>(CPU)</sub> = 100 to 200 MHz			70	ps
	CPU x-point to ↑ edges Output skew window	↑ CPU	CPU	PCIx				500	
4-1	time independent (3.3 V)	200 MHz	CPUx	LDTx	f <sub>(LDT)</sub> = 66.7 MHz			500	
tsk(ow)	CPU x-point to ↑ edges	↑ CPU	CPU	PCIx	f <sub>(PCI)</sub> = 33.3 MHz			200	ps
	Output skew window time variant skew	200 MHz	CPU	LDTx	f <sub>(LDT)</sub> = 66.7 MHz			200	
t <sub>r</sub>	Rise time	Test load a	at the a	c coupling	V <sub>ref</sub> = 0V ±400 mV	100 2.5		300 8.0	ps V/ns
tf	Fall time	node inclu	ding CF	PU load.	differential measured	100 2.5		300 8.0	ps V/ns
v <sub>r</sub>	Edge rate rising edge (maintained during total transition)	Test load a			V <sub>ref</sub> = 0V ±400 mV differential measured	2.0		8.0	V/ns
Vf	Edge rate falling edge (maintained during total transition)	Test load a	at ac co	upling		2.0		8.0	V/ns

† All typical values are measured at their nominal V<sub>DD</sub> values.

NOTE 8: Accumulated jitter is the sum of individual consecutive cycle-to-cycle jitter reads added for a at least 32 μs (one SSC modulation period). The limit corresponds to the w/c cumulative shortest and longest jitter number found during evaluation time.

# CPU $C_L = 10$ pF, $R_L = Test Load$ (see Note 9) (continued)

	PARAMETER	TEST	CONDITIONS	MIN	TYP† MA	١X	Unit
VOD	Differential output voltage	Test load at ac coupling node including CPU load.	CPU to CPU	1.3	1	.7	<b>V</b>
ΔV <sub>OD_DC</sub>	Change in dc differential output voltage		CPU to CPU	-15		15	mV
Vосм	Common mode voltage		CPU to CPU	1.3	1	.4	V
ΔV <sub>OCM</sub>	Change in common mode voltage		CPU to CPU	-10		10	mV
V <sub>CM_AC</sub>	Common mode voltages (MIN/MAX)		CPU to CPU	1.0	1	.4	<b>V</b>
v <sub>cross</sub>	Absolute cross point voltages crosspoint (low and high)	Test load at ac coupling node including CPU load.	CPU and CPU	1.0	1	.2	V
$\Delta v_{cross}$	Variation of V <sub>Cross</sub> , rising edge		At ↑ CPU(xn), (max-min)			90	mV
T∆v <sub>cross</sub>	Total variation V <sub>cross</sub> , all edges		At $\uparrow$ or $\downarrow$ CPU(xn),(max-min)		1	40	mV

<sup>†</sup> The average over any 1-µs period of time is greater than the minimum specified period

NOTES: 9. This specification does not include variations caused by K8 input resistor network or K8 VDD voltage variations. The common mode voltage is calculated as: (VOH+VOL)/2. See the measurement information section for details.

<sup>10.</sup> This applies also to CPU outputs.

# switching characteristics, $V_{DD}$ = 3.135 V to 3.465 V, $T_A$ = 0°C to 70°C (continued)

USB, FDC (48 MHz) and FDC (24 MHz), CL = 20 pF, (USB) RL = 500  $\Omega$ 

	PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	түр†	MAX	UNIT
tpd1	Propagation delay time	XIN	USB/FDC	$f_{(XIN)} \ge 1$ MHz, TEST MODE	2		15	ns
t <sub>pd2</sub>	Propagation delay time	SCLK↑	USB/FDC	TEST MODE		18		ns
	USB/FDC (48 MHz) clock period‡			f(USB/FDC) = 48 MHz	20.8		20.84	
t <sub>C</sub>	FDC (24 MHz) clock period <sup>‡</sup>			f <sub>(FDC)</sub> = 24 MHz	41.6		41.68	ns
	Cycle to cycle jitter FDC (48 MHz)	or FDC (24 N	1Hz)	f(CPU) = 100 to 200 MHz			200	
<sup>t</sup> jit(cc)	Cycle to cycle jitter USB (48 MHz),	FDC=24 or 4	18 MHz	f(CPU) = 100 to 200 MHz			180	ps
tjit(acc)	Accumulated jitter USB (48 MHz), F	DC=24 or 4	8 MHz	f(CPU) = 100 to 200 MHz	-160		160	ps
	Duty cycle USB/FDC			f(USB/FDC) = 48 MHz	45		55	0.4
odc	Duty cycle FDC			f <sub>(FDC)</sub> = 24 MHz	45		55	%
<b>t</b>	USB to FDC skew ↑ edges time-independent and	USB	FDC	f(USB/FDC) = 48 MHz			500	ne
tsk(ow)	time-variant skew combined	ОЗВ	FDC	f(USB/FDC) = 48 / 24 MHz			500	ps
	USB/FDC pulse skew	USB/FDC	USB/FDC	f(USB/FDC) = 48 MHz	2		6.5	20
<sup>t</sup> sk(p)	FDC pulse skew	FDC	FDC	f <sub>(FDC)</sub> = 24 MHz	2		6.5	ns
	Dulas duration high			f(USB/FDC) = 48 MHz	7.5			20
t w(H)	Pulse duration, high			f(FDC) = 24  MHz	18			ns
	Pulse duration, low			f(USB/FDC) = 48 MHz	11.5			20
t w(L)	Pulse duration, low			f <sub>(FDC)</sub> = 24 MHz	22			ns
	Rise time		USB	\/ . 200/ to 200/ of \/ -	1.1		2.5	ns
t <sub>r</sub>	Rise time		FDC	$V_{ref}$ = 20% to 80% of $V_O$	2		0.7	V/ns
4.	Fall times		USB	V . 000/ to 000/ of V-	1.1		2.5	ns
t <sub>f</sub>	Fall time		FDC	$V_{ref}$ = 20% to 80% of $V_O$	2		0.7	V/ns
v <sub>r</sub>	Edge rate, rising edge (maintained	during total	USB	V <sub>ref</sub> = 20% to 60% of V <sub>DD</sub>	0.25		1.1	V/ns
	transition)		FDC	101 ==/				.,
v <sub>f</sub>	Edge rate falling edge (maintained transition)	during total	USB FDC	V <sub>ref</sub> = 20% to 60% of VDD	0.25		1.1	V/ns

<sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub> values.

<sup>‡</sup> The average over any 1-µs period of time is greater than the minimum specified period

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# switching characteristics, $V_{DD}$ = 3.135 V to 3.465 V, $T_A$ = 0°C to 70°C (continued)

PCI, LDT (33 MHz), PCI\_F and LDT (66 MHz), CL = 30 pF, RL = 500  $\Omega$ 

	PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
tpd1	Propagation delay time	XIN	PCIx,LDT	$f(XIN) \ge 1$ MHz, Test mode	2.0		15	ns
t <sub>pd2</sub>	Propagation delay time	SCLK↑	PCIx, LDT	Test mode		18		ns
	DOI aloak maria d†			f <sub>(PCI)</sub> = 33.3 MHz	29.95		30.3	
t <sub>C</sub>	PCI clock period <sup>†</sup>			$f_{(LDT)} = 66.7MHz$	14.95		15.15	ns
4	Cycle-to-cycle jitter PCI/LDT (33 N	1Hz), LDT (33	MHz)	f			170	
<sup>t</sup> jit(cc)	Cycle-to-cycle jitter LDT (66 MHz)	PCI (33 MHz	<u>z</u> )	$f_{(CPU)} = 100 \text{ to } 200 \text{ MHz}$			290	ps
<sup>t</sup> jit(acc)	Accumulated jitter PCI/LDT (33 Mł	Hz), LDT (66 I	MHz)	f(CPU) = 100 to 200 MHz	-300		300	ps
odc	Duty cycle PCI (33 MHz)			f <sub>(PCI)</sub> = 33.3 MHz	45		55	%
t <sub>dc</sub>	Duty cycle LDT (66MHz)			f <sub>(LDT)</sub> = 66.7 MHz	45		55	%
	PCI bank skew ↑ edges time-independent (3.3 V)						500	
<sup>t</sup> sk(b)	PCI bank skew ↑ edges time-variant skew	PCIx	PCIx	f(PCI) = 33.3 MHz			200	ps
	↑ edges to CPU x-point time-independent (3.3 V)						500	
<sup>t</sup> sk(ow)	↑ edges to CPU x-point time-variant skew	PCIn	CPUx	f(PCI) = 33.3 MHz			200	ps
	LDT bank skew ↑ edges time-independent (3.3 V)	LDTx	LDTx				500	
<sup>t</sup> sk(b)	LDT bank skew ↑ edges time-variant skew			f <sub>(LDT)</sub> = 66.7 MHz			200	ps
	↑ edges to CPU x-point time-independent (3.3 V)	LDTx	CPUx				500	
<sup>t</sup> sk(ow)	↑ edges to CPU x-point tim- variant skew	LDTx	CPUx	f <sub>(LDT)</sub> = 66.7 MHz			15.15 170 290 300 55 55 500 200 200 500 200	ps

<sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub> values.

<sup>‡</sup> The average over any 1-µs period of time is greater than the minimum specified period

# PCI, LDT (33 MHz), PCI\_F and LDT (66 MHz), C<sub>L</sub> = 30 pF, R<sub>L</sub> = 500 $\Omega$ (continued)

	PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP <sup>†</sup>	MAX	Unit
	↑ edges to LDT time independent (3.3 V)	PCIx	LDTx				500	
	↑ edges to LDT time variant skew	PCIx	LDTx	f(PCI/LDT) = 33.3 MHz			200	
<sup>t</sup> sk(ow)	↑ edges to LDT time independent (3.3)	PCIx	LDTx				500	ps
†sk(s)	↑ edges to LDT time variant skew	PCIx	LDTx	f <sub>(PCI/LDT)</sub> = 33.3 MHz/66.7 MHz			200	
	PCI pulse skew	PCIn	PCIn	f(PCI) = 33.3 MHz	1.5		3.7	
tsk(p)	LDT pulse skew	LDTn	LDTn	f(LDT) = 66.7 MHz	1.4		3.6	ns
4	Pulse duration, high PCI (33 MF	lz)		f(PCI) = 33.3 MHz	13.6			
<sup>t</sup> w(H)	Pulse duration, high LDT (66 MI	Hz)		f(LDT) = 66.7 MHz	6.2			ns
	Pulse duration, low PCI (33 MH	z)		f(PCI) = 33.3 MHz	16.0			
tw(L)	Pulse duration, low LDT (66 MH	z)		$f_{(LDT)} = 66.7 \text{ MHz}$	8.4			ns
t <sub>r</sub>	Rise time PCI/LDT (33 MHz), LI	OT (66 MHz)		$V_{ref}$ = 20% to 80% of $V_{O}$	0.7 2.9		1.6 1.2	ns V/ns
tf	Fall time PCI/LDT (33 MHz), LD	T (66 MHz)		V <sub>ref</sub> = 20% to 80% of V <sub>O</sub>	0.6 3.5		1.6 1.2	ns V/ns
v <sub>r</sub>	Edge rate rising edge (maintain	ed during tota	l transition)	V <sub>ref</sub> = 20% to 60% of V <sub>DD</sub>	0.3		1.7	V/ns

<sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub> values.

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# switching characteristics, $V_{DD}$ = 3.135 V to 3.465 V, $T_A$ = 0°C to 70°C (continued)

# REF, $C_L$ = 20 pF, $R_L$ = 500 $\Omega$

	PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	түр†	MAX	UNIT
tpd1	Propagation delay time	XIN	REF	$f_{(XIN)} \ge 1$ MHz, TEST MODE	2		10.0	ns
t <sub>pd2</sub>	Propagation delay time	SCLK↑	REF	TEST MODE		18		ns
t <sub>C</sub>	REF clock period <sup>†</sup>			f <sub>(REF)</sub> = 14.318 MHz	69.8		69.84	ns
tjit(cc)	Cycle to cycle jitter			$f_{(CPU)} = 100 \text{ to } 200 \text{ MHz}$			250	ps
tjit(acc)	Accumulated jitter			f(CPU) = 100 to 200 MHz	-200		200	ps
tjit(∅)	Phase jitter			f(CPU) = 100 to 200 MHz			300	ps
odc	Duty cycle			f(REF) = 14.318 MHz	45		55	%
t <sub>sk(b)</sub>	REF bank skew ↑ edges	REFx	REFx	f(REF) = 14.318 MHz			500	ps
t <sub>sk(p)</sub>	REF pulse skew	REF	REF	f(REF) = 14.318 MHz	2		5.8	ps
t <sub>w(H)</sub>	Pulse duration width, high			f <sub>(REF)</sub> = 14.318 MHz	27			ns
t <sub>W(L)</sub>	Pulse duration width, low			f(REF) = 14.318 MHz	32			ns
t <sub>r</sub>	Rise time			V <sub>ref</sub> = 20% to 80% of Vo	1.1 2		2.7 0.7	ns V/ns
t <sub>f</sub>	Fall time			V <sub>ref</sub> = 20% to 80% of Vo	1.1 2		2.7 0.7	ns V/ns
v <sub>r</sub>	Edge rate rising edge (maintained d	uring total t	ransition)	V <sub>ref</sub> = 20% to 60% of VDD	0.25		1.1	V/ns
v <sub>f</sub>	Edge rate falling edge (maintained of	during total t	transition)		0.25		1.1	V/ns

<sup>†</sup> All typical values are measured at their nominal V<sub>DD</sub> values.

# SDATA, CL = 10 pF to 400 pF, RL = 1 k $\Omega$

	PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP <sup>†</sup>	MAX	UNIT
tPHL	Propagation delay time‡	SCLK↓	Data acknnowledge	See Figure 6	0.375		2	μs
tPLH	Propagation delay time‡	SCLK↓	Data valid	See Figure 6	0.375		2	μs
<sup>t</sup> PHL	Propagation delay time‡	SCLK↓	Data valid	See Figure 6	0.375		2	μs
	E 11.0			C <sub>L</sub> = 10 pF		86	250	
tf	Fall time			C <sub>L</sub> = 400 pF		115	250	ns

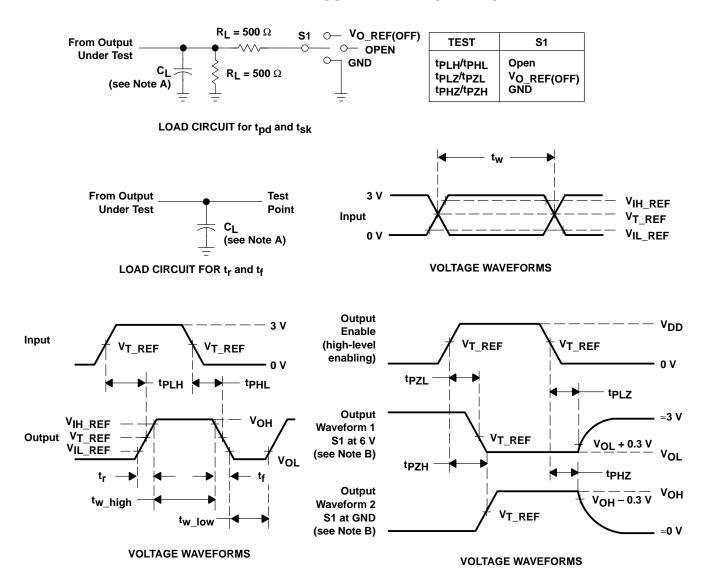
 $<sup>\</sup>dagger$  All typical values are measured at their nominal V<sub>DD</sub> values.



<sup>‡</sup> The average over any 1-µs period of time is greater than the minimum specified period

<sup>‡</sup> This is a digital controlled delay. It equals to 6 REF clock cycles plus the internal gate delay (20 ns).

#### PARAMETER MEASUREMENT INFORMATION



- NOTES: A. C<sub>L</sub> includes probe and jig capacitance. C<sub>L</sub> = 10 pF (CPU), C<sub>L</sub> = 20 pF (USB, FDC, REF), C<sub>L</sub> = 30 pF (PCI, LDC)
  - B. Waveform 1 is for an output with internal conditions such that the output is low except when disabled by the output control. Waveform 2 is for an output with internal conditions such that the output is high except when disabled by the output control.
  - C. All input pulses are supplied by generators having the following characteristics: PRR  $\leq$  14.318 MHz,  $Z_O = 50~\Omega$ ,  $t_f \leq 2.5~ns$ ,  $t_f \leq 2.5~ns$ .
  - D. The outputs are measured one at a time with one transition per measurement.

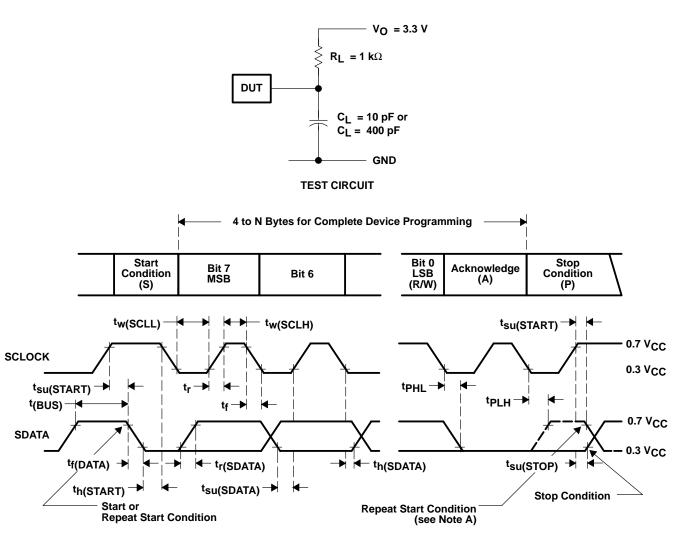
	PARAMETER	3.3-V INTERFACE	UNIT
VIH_REF	High-level reference voltage	2.4	V
VIL_REF	Low-level reference voltage	0.4	V
V <sub>T_REF</sub>	Input threshold reference voltage	1.5	V
VO REF	Off-state reference voltage	6	V

Figure 5. Load Circuit and Voltage Waveforms



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#### PARAMETER MEASUREMENT INFORMATION



#### **VOLTAGE WAVEFORMS**

NOTE A: The repeat start condition is supported, but not clock stretching.

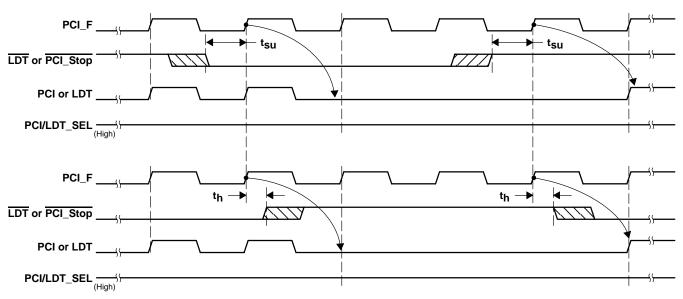
BYTE	DESCRIPTION		
1	SMBus address		
2	Command (dummy value, ignored)		
3	Byte count		
4	SMBus data byte 0		
5 – N	SMBus data byte 1 – N		

Figure 6. Propagation Delay Times,  $t_r$  and  $t_f$ 



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### PARAMETER MEASUREMENT INFORMATION



NOTE: Assertion and deassertion of PCI\_STOP or LDT\_STOP maintain signals duty cycle. 
tsu(disable) is the time at which no pulse exists in following period.
tsu(enable) is the time at which a pulse exists in following period.

Figure 7. PCI\_Stop or LDT\_Stop ↓↑ to PCI (LDT)

#### PARAMETER MEASUREMENT INFORMATION

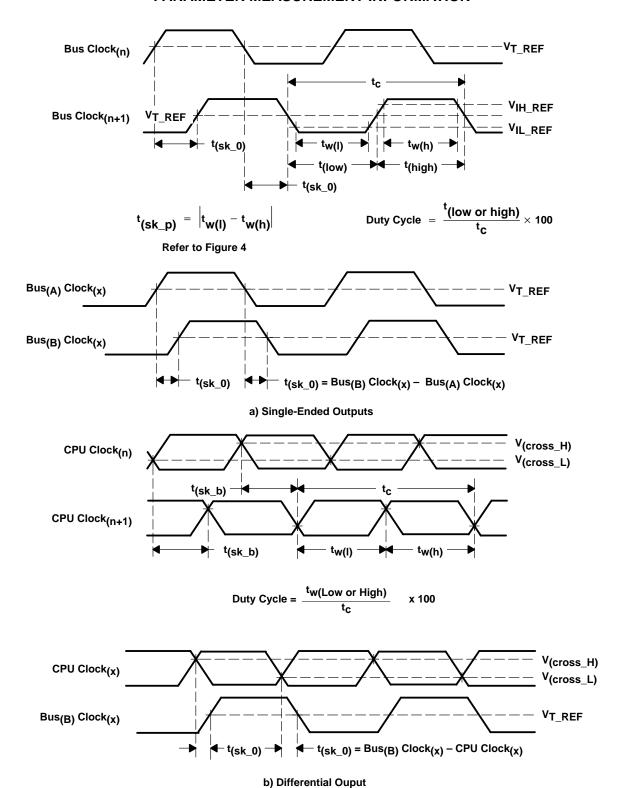
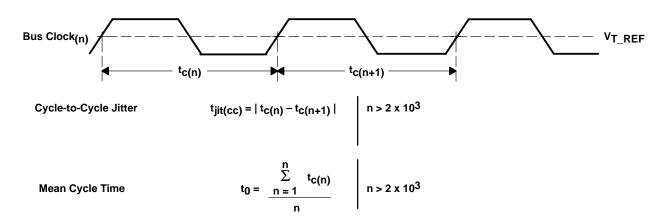
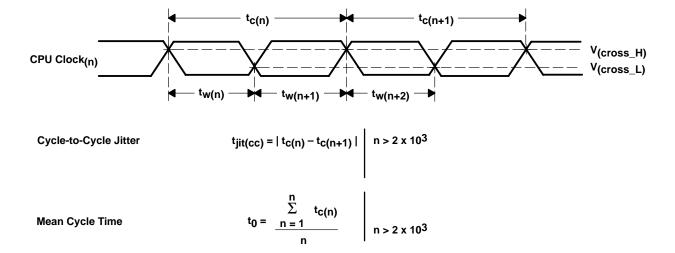


Figure 8. Waveforms for Calculation of Skew and Offset

#### PARAMETER MEASUREMENT INFORMATION



a) Single-Ended Output



b) Differential Output

Figure 9. Waveforms for Calculation of Jitter

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#### PARAMETER MEASUREMENT INFORMATION

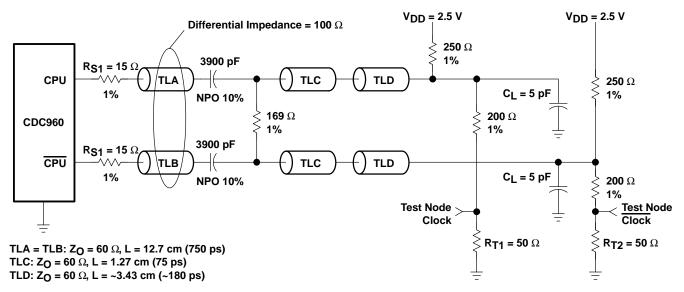


Figure 10. Load Circuit and Voltage Waveforms for CPU Bus

### correction for measurements at 50 $\Omega$ nodes

Voltage levels and readings are scaled for the voltage divider 200  $\Omega$  to 50  $\Omega$  versus all reads and reference levels must be multiplied/divided by the fixed scale of five.

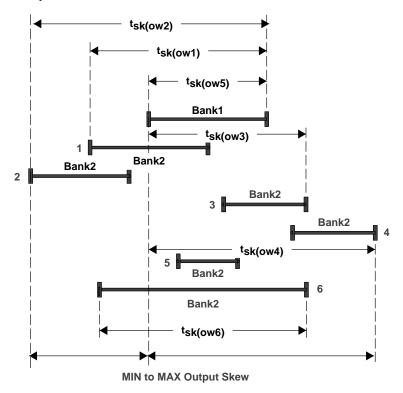


Figure 11. Bank and Output Skew;  $t_{sk(owx)}$ : Output Skew Window and MIN-to-MAX Phase

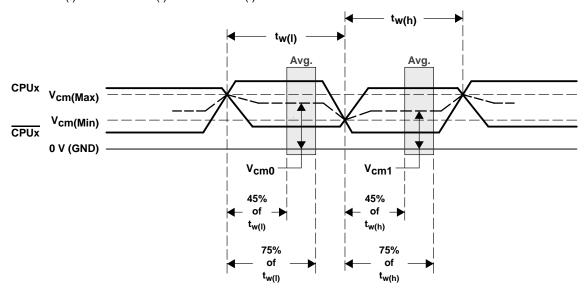


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#### PARAMETER MEASUREMENT INFORMATION

The common mode voltage is measured single-ended and is the result of the following calculation:

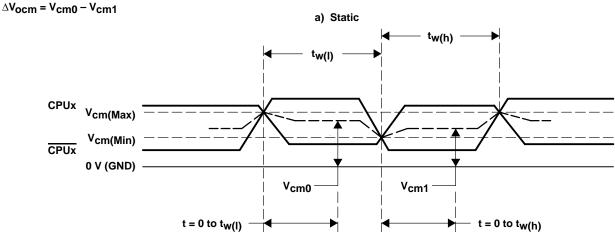
$$Vocm_{(t)} = [Vo(CPUT)_{(t)} + Vo(CPUC)_{(t)}]/2$$



 $V_{\text{cm0}}$  is calculated from the average of Vocm within 45%–75% (region without switching noise) of the pulse while CPUx is in the LOW state.

 $V_{\text{cm1}}$  is calculated from the average of Vocm within 45%–75% (region without switching noise) of the pulse while CPUx is in the HIGH state.

$$V_{\text{ocm}} = (V_{\text{cm0}} + V_{\text{cm1}})/2$$



 $\Delta V_{ocm(t)} = \text{MAX} \ (V_{cm0(t)}) - \text{MIN} \ (V_{cm1(t)}) \ \text{and} \\ \Delta V_{ocm(t)} = \text{MIN} \ (V_{cm0(t)}) - \text{MAX} \ (V_{cm1(t)})$ 

b) Dynamic

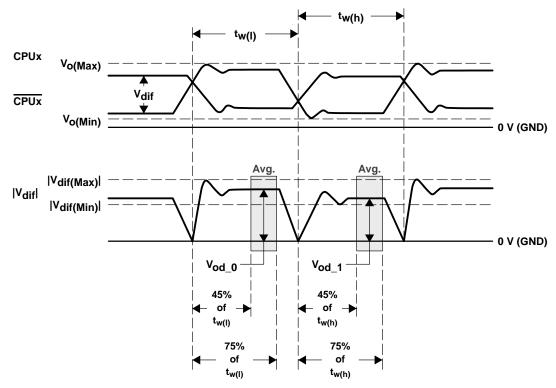
Figure 12. Common Mode Voltage



#### PARAMETER MEASUREMENT INFORMATION

The differential voltage is measured single-ended and is the result of the following calculation:

$$Vod_{(t)} = Vo(CPUx)_{(t)} - Vo(\overline{CPUx})_{(t)}$$



Vod\_0 is calculated from the average of Vod within 45%–75% (region without switching noise) of the pulse while CPUx is in the LOW state.

Vod\_1 is calculated from the average of Vod within 45%–75% (region without switching noise) of the pulse while CPUx is in the HIGH state.

Vod = (Vod\_0 + Vod\_1/2 ∆Vod\_DC = Vod\_0 − Vod\_1 ∆Vod\_AC = |Vdif(max0| − |Vdif(min)|

Figure 13. Differential Output Voltage



# CDC960

# 200-MHz CLOCK SYNTHESIZER/DRIVER WITH SPREAD SPECTRUM CAPABILITY AND DEVICE CONTROL INTERFACE

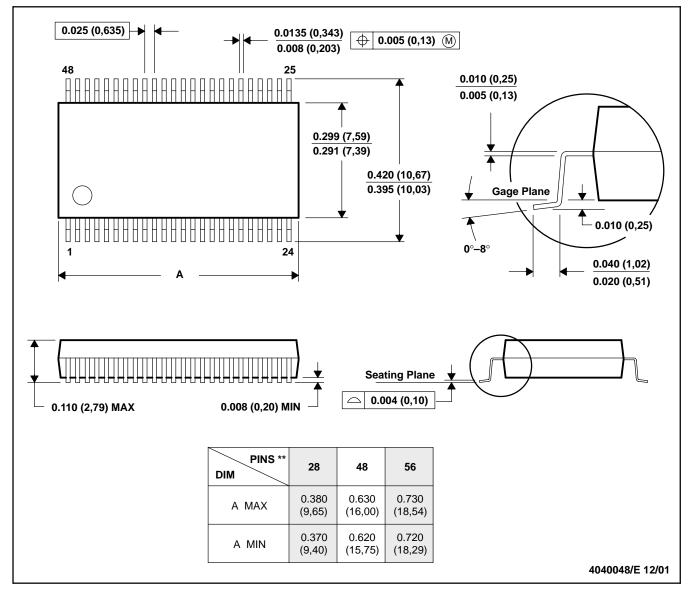
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#### **MECHANICAL DATA**

# DL (R-PDSO-G\*\*)

#### **48 PINS SHOWN**

## PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
- D. Falls within JEDEC MO-118





# PACKAGE OPTION ADDENDUM

4-Mar-2005

## **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins F	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
CDC960DL	ACTIVE	SSOP	DL	48	25	None	Call TI	Level-1-220C-UNLIM
CDC960DLR	ACTIVE	SSOP	DL	48	1000	None	Call TI	Level-1-220C-UNLIM

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - May not be currently available - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

None: Not yet available Lead (Pb-Free).

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Green (RoHS & no Sb/Br): TI defines "Green" to mean "Pb-Free" and in addition, uses package materials that do not contain halogens, including bromine (Br) or antimony (Sb) above 0.1% of total product weight.

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDECindustry standard classifications, and peak solder temperature.

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