## Low－Power 100MHz Gain－of－2 Stable Operational Amplifier

The EL2045 is a high speed，low power，low cost monolithic operational amplifier built on Elantec＇s proprietary complementary bipolar process．The EL2045 is gain－of－2 stable and features a $275 \mathrm{~V} / \mu$ s slew rate and 100 MHz gain－ bandwidth at gain－of－2 while requiring only 5.2 mA of supply current．

The power supply operating range of the EL2045 is from $\pm 18 \mathrm{~V}$ down to as little as $\pm 2 \mathrm{~V}$ ．For single－supply operation， the EL2045 operates from 36 V down to as little as 2.5 V ．The excellent power supply operating range of the EL2045 makes it an obvious choice for applications on a single +5 V or +3 V supply．

The EL2045 also features an extremely wide output voltage swing of $\pm 13.6 \mathrm{~V}$ with $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ and $\mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega$ ．At $\pm 5 \mathrm{~V}$ ， output voltage swing is a wide $\pm 3.8 \mathrm{~V}$ with $R_{L}=500 \Omega$ and $\pm 3.2 \mathrm{~V}$ with $R_{L}=150 \Omega$ ．Furthermore，for single－supply operation at +5 V ，output voltage swing is an excellent 0.3 V to 3.8 V with $\mathrm{R}_{\mathrm{L}}=500 \Omega$ ．

At a gain of +2 ，the EL2045 has a $-3 d B$ bandwidth of 100 MHz with a phase margin of $50^{\circ}$ ．Because of its conventional voltage－feedback topology，the EL2045 allows the use of reactive or non－linear elements in its feedback network．This versatility combined with low cost and 75 mA of output－current drive makes the EL2045 an ideal choice for price－sensitive applications requiring low power and high speed．

## Ordering Information

| PART NUMBER | PACKAGE | TAPE \＆REEL | PKG．DWG．\＃ |
| :--- | :---: | :---: | :---: |
| EL2045CS | 8－Pin SO | - | MDP0027 |
| EL2045CS－T7 | 8－Pin SO | $7 \prime$ | MDP0027 |
| EL2045CS－T13 | 8－Pin SO | $13^{\prime \prime}$ | MDP0027 |
| EL2045CSZ <br> （See Note） | 8－Pin SO <br> （Pb－free） | - | MDP0027 |
| EL2045CSZ－T7 <br> （See Note） | 8－Pin SO <br> （Pb－free） | 7 ＂ | MDP0027 |
| EL2045CSZ－T13 <br> （See Note） | 8－Pin SO <br> （Pb－free） | $13^{\prime \prime}$ | MDP0027 |
| EL2045CN | 8－Pin PDIP | - | MDP0031 |

NOTE：Intersil Pb －free products employ special Pb －free material sets； molding compounds／die attach materials and 100\％matte tin plate termination finish，which are RoHS compliant and compatible with both SnPb and Pb －free soldering operations．Intersil Pb －free products are MSL classified at Pb－free peak reflow temperatures that meet or exceed the Pb－free requirements of IPC／JEDEC J STD－020．

## Features

－ 100 MHz gain－bandwidth at gain－of－2
－Gain－of－2 stable
－Low supply current -5.2 mA at $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$
－Wide supply range $- \pm 2 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ dual－supply and 2.5 V to 36 V single－supply
－High slew rate－ $275 \mathrm{~V} / \mu \mathrm{s}$
－Fast－settling－80ns to $0.1 \%$ for a 10 V step
－Low differential gain $-0.02 \%$ at $A_{V}=+2, R_{L}=150 \Omega$
－Low differential phase $-0.07^{\circ}$ at $A_{V}=+2, R_{L}=150 \Omega$
－Wide output voltage swing $- \pm 13.6 \mathrm{~V}$ with $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ ， $R_{L}=1 \mathrm{k} \Omega$ and $3.8 \mathrm{~V} / 0.3 \mathrm{~V}$ with $\mathrm{V}_{\mathrm{S}}=+5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=500 \Omega$
－Pb－Free available（RoHS compliant）

## Applications

－Video amplifiers
－Single－supply amplifiers
－Active filters／integrators
－High speed sample－and－hold
－High speed signal processing
－ADC／DAC buffers
－Pulse／RF amplifiers
－Pin diode receivers
－Log amplifiers
－Photo multiplier amplifiers
－Difference amplifiers

## Pinout

（8－PIN SO \＆8－PIN PDIP） TOP VIEW


Absolute Maximum Ratings $\left(\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right)$

Supply Voltage ( $\mathrm{V}_{\mathrm{S}}$ )
) . . . . . . . . . . . . . . . . . . . . . . . . . . . . $\pm 18 \mathrm{~V}$ or 36 V
Input Voltage ( $\mathrm{V}_{\mathrm{IN}}$. . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . $\mathrm{IV}_{\mathrm{S}}$
Differential Input Voltage ( $\mathrm{dV} \mathrm{V}_{\mathrm{IN}}$ ) . . . . . . . . . . . . . . . . . . . . . . . . . $\pm 10 \mathrm{~V}$
Continuous Output Current . . . . . . . . . . . . . . . . . . . . . . . . . . . 60mA

Power Dissipation ( $\mathrm{P}_{\mathrm{D}}$ ) . . . . . . . . . . . . . . . . . . . . . . . . . . See Curves Operating Temperature Range $\left(\mathrm{T}_{\mathrm{A}}\right) \ldots \ldots . . . . . . . .40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ Operating Junction Temperature ( $\mathrm{T}_{\mathrm{J}}$ ) . . . . . . . . . . . . . . . . . . . $+150^{\circ} \mathrm{C}$ Storage Temperature (TST) . . . . . . . . . . . . . . . . . . . $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_{J}=T_{C}=T_{A}$

DC Electrical Specifications $\quad V_{S}= \pm 15 \mathrm{~V}, R_{L}=1 \mathrm{k} \Omega$, unless otherwise specified.

| PARAMETER | DESCRIPTION | CONDITION | TEMP | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Vos | Input Offset Voltage | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 0.5 | 7.0 | mV |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ |  |  | 9.0 | mV |
| TCV ${ }_{\text {OS }}$ | Average Offset Voltage Drift |  | All |  | 10.0 |  | $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ |
| $\mathrm{I}_{\mathrm{B}}$ | Input Bias Current | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 2.8 | 8.2 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ |  |  | 9.2 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 2.8 |  | $\mu \mathrm{A}$ |
| los | Input Offset Current | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 50 | 300 | nA |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ |  |  | 400 | nA |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 50 |  | nA |
| TClos | Average Offset Current Drift | (Note 1) | All |  | 0.3 |  | $n \mathrm{n} /{ }^{\circ} \mathrm{C}$ |
| AVOL | Open-loop Gain | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}= \pm 10 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega$ | $25^{\circ} \mathrm{C}$ | 1500 | 3000 |  | V/V |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ | 1500 |  |  | V/V |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}= \pm 2.5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=500 \Omega$ | $25^{\circ} \mathrm{C}$ |  | 2500 |  | V/V |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}= \pm 2.5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=150 \Omega$ | $25^{\circ} \mathrm{C}$ |  | 1750 |  | V/V |
| PSRR | Power Supply Rejection Ratio | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V} \text { to } \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ | 65 | 85 |  | dB |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ | 60 |  |  | dB |
| CMRR | Common-mode Rejection Ratio | $\mathrm{V}_{\mathrm{CM}}= \pm 12 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=0 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ | 70 | 95 |  | dB |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ | 70 |  |  | dB |
| CMIR | Common-mode Input Range | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | $\pm 14.0$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | $\pm 4.2$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{S}}=+5 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 4.2/0.1 |  | V |
| V OUT | Output Voltage Swing | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega$ | $25^{\circ} \mathrm{C}$ | $\pm 13.4$ | $\pm 13.6$ |  | V |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ | $\pm 13.1$ |  |  | V |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=500 \Omega$ | $25^{\circ} \mathrm{C}$ | $\pm 12.0$ | $\pm 13.4$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=500 \Omega$ | $25^{\circ} \mathrm{C}$ | $\pm 3.4$ | $\pm 3.8$ |  | V |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=150 \Omega$ | $25^{\circ} \mathrm{C}$ |  | $\pm 3.2$ |  | V |
|  |  | $V_{S}=+5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=500 \Omega$ | $25^{\circ} \mathrm{C}$ | 3.6/0.4 | 3.8/0.3 |  | V |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ | 3.5/0.5 |  |  | V |
| Isc | Output Short Circuit Current |  | $25^{\circ} \mathrm{C}$ | 40 | 75 |  | mA |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ | 35 |  |  | mA |

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DC Electrical Specifications $\quad \mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega$, unless otherwise specified. (Continued)

| PARAMETER | DESCRIPTION | CONDITION | TEMP | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Is | Supply Current | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$, no load | $25^{\circ} \mathrm{C}$ |  | 5.2 | 7 | mA |
|  |  |  | $\mathrm{T}_{\text {MIN }}, \mathrm{T}_{\text {MAX }}$ |  |  | 7.6 | mA |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$, no load | $25^{\circ} \mathrm{C}$ |  | 5.0 |  | mA |
| $\mathrm{R}_{\mathrm{IN}}$ | Input Resistance | Differential | $25^{\circ} \mathrm{C}$ |  | 150 |  | $\mathrm{k} \Omega$ |
|  |  | Common-mode | $25^{\circ} \mathrm{C}$ |  | 15 |  | $\mathrm{M} \Omega$ |
| $\mathrm{Cl}_{\mathrm{IN}}$ | Input Capacitance | $\mathrm{A}_{\mathrm{V}}=+2$ @10MHz | $25^{\circ} \mathrm{C}$ |  | 1.0 |  | pF |
| ROUT | Output Resistance | $A_{V}=+2$ | $25^{\circ} \mathrm{C}$ |  | 50 |  | $\mathrm{m} \Omega$ |
| PSOR | Power-Supply Operating Range | Dual-supply | $25^{\circ} \mathrm{C}$ | $\pm 2.0$ |  | $\pm 18.0$ | V |
|  |  | Single-supply | $25^{\circ} \mathrm{C}$ | 2.5 |  | 36.0 | V |

NOTE:

1. Measured from $\mathrm{T}_{\text {MIN }}$ To $\mathrm{T}_{\text {MAX }}$.

Closed-Loop AC Electrical Specifications $\quad V_{S}= \pm 15 \mathrm{~V}, A_{V}=+2, R_{F}=R_{G}=1 \mathrm{k} \Omega, C_{F}=3 p F, R_{L}=1 \mathrm{k} \Omega$ unless otherwise specified.

| PARAMETER | DESCRIPTION | CONDITION | TEMP | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BW | -3 dB Bandwidth ( $\mathrm{V}_{\mathrm{OUT}}=0.4 \mathrm{~V}_{\mathrm{PP}}$ ) | $\mathrm{V}_{S}= \pm 15 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=+2$ | $25^{\circ} \mathrm{C}$ |  | 100 |  | MHz |
|  |  | $\mathrm{V}_{S}= \pm 15 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=-1$ | $25^{\circ} \mathrm{C}$ |  | 75 |  | MHz |
|  |  | $\mathrm{V}_{S}= \pm 15 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=+5$ | $25^{\circ} \mathrm{C}$ |  | 20 |  | MHz |
|  |  | $\mathrm{V}_{S}= \pm 15 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=+10$ | $25^{\circ} \mathrm{C}$ |  | 10 |  | MHz |
|  |  | $\mathrm{V}_{S}= \pm 15 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=+20$ | $25^{\circ} \mathrm{C}$ |  | 5 |  | MHz |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~A}_{\mathrm{V}}=+2$ | $25^{\circ} \mathrm{C}$ |  | 75 |  | MHz |
| GBWP | Gain-bandwidth Product | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 200 |  | MHz |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 150 |  | MHz |
| PM | Phase Margin | $R_{L}=1 \mathrm{k} \Omega, C_{L}=10 \mathrm{pF}$ | $25^{\circ} \mathrm{C}$ |  | 50 |  | - |
| SR | Slew Rate (Note 1) | $\mathrm{V}_{S}= \pm 15 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega$ | $25^{\circ} \mathrm{C}$ | 200 | 275 |  | V/us |
|  |  | $\mathrm{V}_{S}= \pm 5 \mathrm{~V}, \mathrm{R}_{\mathrm{L}}=500 \Omega$ | $25^{\circ} \mathrm{C}$ |  | 200 |  | V/us |
| FPBW | Full-power Bandwidth (Note 2) | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ | 3.2 | 4.4 |  | MHz |
|  |  | $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}$ | $25^{\circ} \mathrm{C}$ |  | 12.7 |  | MHz |
| $t_{R}, t_{\text {F }}$ | Rise Time, Fall Time | 0.1V output step | $25^{\circ} \mathrm{C}$ |  | 3.0 |  | ns |
| OS | Overshoot | 0.1V output step | $25^{\circ} \mathrm{C}$ |  | 20 |  | \% |
| $t_{\text {PD }}$ | Propagation Delay |  | $25^{\circ} \mathrm{C}$ |  | 2.5 |  | ns |
| ts | Settling to $+0.1 \%\left(\mathrm{~A}_{V}=+2\right)$ | $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, 10 \mathrm{~V}$ step | $25^{\circ} \mathrm{C}$ |  | 80 |  | ns |
|  |  | $\mathrm{V}_{S}= \pm 5 \mathrm{~V}, 5 \mathrm{~V}$ step | $25^{\circ} \mathrm{C}$ |  | 60 |  | ns |
| dG | Differential Gain (Note 3) | NTSC/PAL | $25^{\circ} \mathrm{C}$ |  | 0.02 |  | \% |
| dP | Differential Phase (Note 3) | NTSC/PAL | $25^{\circ} \mathrm{C}$ |  | 0.07 |  | - |
| eN | Input Noise Voltage | 10 kHz | $25^{\circ} \mathrm{C}$ |  | 15.0 |  | $\mathrm{nV} / \sqrt{\mathrm{Hz}}$ |
| iN | Input Noise Current | 10 kHz | $25^{\circ} \mathrm{C}$ |  | 1.50 |  | $\mathrm{pA} / \sqrt{\mathrm{Hz}}$ |
| CI STAB | Load Capacitance Stability | $\mathrm{A}_{\mathrm{V}}=+2$ | $25^{\circ} \mathrm{C}$ |  | Infinite |  | pF |

NOTES:

1. Slew rate is measured on rising edge.
2. For $\mathrm{V}_{\mathrm{S}}= \pm 15 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=20 \mathrm{~V}_{\mathrm{PP}}$. For $\mathrm{V}_{\mathrm{S}}= \pm 5 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=5 \mathrm{~V}_{\text {PP }}$. Full-power bandwidth is based on slew rate measurement using: $\mathrm{FPBW}=\mathrm{SR} /(2 \pi$ * Vpeak$)$.
3. Video performance measured at $V_{S}= \pm 15 \mathrm{~V}, A_{V}=+2$ with 2 times normal video level across $R_{L}=150 \Omega$. This corresponds to standard video levels across a backterminated $75 \Omega$ load. For other values of $R_{L}$, see curves.

## EL2045 Test Circuit



## Typical Performance Curves



## Typical Performance Curves (Continued)



## Typical Performance Curves (Continued)



## Simplified Schematic



## Burn-In Circuit


all packages use the same schematic

## Applications Information

## Product Description

The EL2045 is a low-power wideband, gain-of-2 stable monolithic operational amplifier built on Elantec's proprietary high-speed complementary bipolar process. The EL2045 uses a classical voltage-feedback topology which allows it to be used in a variety of applications where current-feedback amplifiers are not appropriate because of restrictions placed upon the feedback element used with the amplifier. The conventional topology of the EL2045 allows, for example, a capacitor to be placed in the feedback path, making it an excellent choice for applications such as active filters, sample-and-holds, or integrators. Similarly, because of the
ability to use diodes in the feedback network, the EL2045 is an excellent choice for applications such as fast log amplifiers.

## Single-Supply Operation

The EL2045 has been designed to have a wide input and output voltage range. This design also makes the EL2045 an excellent choice for single-supply operation. Using a single positive supply, the lower input voltage range is within 100 mV of ground $\left(R_{L}=500 \Omega\right)$, and the lower output voltage range is within 300 mV of ground. Upper input voltage range reaches 4.2 V , and output voltage range reaches 3.8 V with a 5 V supply and $\mathrm{R}_{\mathrm{L}}=500 \Omega$. This results in a 3.5 V output swing on a single 5 V supply. This wide output voltage range also allows single-supply operation with a supply voltage as high as 36 V or as low as 2.5 V . On a single 2.5 V supply, the EL2045 still has 1V of output swing.

## Gain-Bandwidth Product and the -3dB Bandwidth

The EL2045 has a gain-bandwidth product of 100 MHz while using only 5.2 mA of supply current. For gains greater than 4 , its closed-loop -3dB bandwidth is approximately equal to the gain-bandwidth product divided by the noise gain of the circuit. For gains less than 4, higher-order poles in the amplifier's transfer function contribute to even higher closed loop bandwidths. For example, the EL2045 has a -3dB bandwidth of 100 MHz at a gain of +2 , dropping to 20 MHz at a gain of +5 . It is important to note that the EL2045 has been

## EL2045

designed so that this "extra" bandwidth in low-gain applications does not come at the expense of stability. As seen in the typical performance curves, the EL2045 in a gain of +2 only exhibits 1.0 dB of peaking with a $1 \mathrm{k} \Omega$ load.

## Video Performance

An industry-standard method of measuring the video distortion of a component such as the EL2045 is to measure the amount of differential gain (dG) and differential phase (dP) that it introduces. To make these measurements, a $0.286 \mathrm{~V}_{\mathrm{PP}}$ (40 IRE) signal is applied to the device with OV DC offset ( 0 IRE) at either 3.58 MHz for NTSC or 4.43 MHz for PAL. A second measurement is then made at 0.714 V DC offset (100 IRE). Differential gain is a measure of the change in amplitude of the sine wave, and is measured in percent. Differential phase is a measure of the change in phase, and is measured in degrees.

For signal transmission and distribution, a back-terminated cable ( $75 \Omega$ in series at the drive end, and $75 \Omega$ to ground at the receiving end) is preferred since the impedance match at both ends will absorb any reflections. However, when double termination is used, the received signal is halved; therefore a gain of 2 configuration is typically used to compensate for the attenuation.

The EL2045 has been designed as an economical solution for applications requiring low video distortion. It has been thoroughly characterized for video performance in the topology described above, and the results have been included as typical dG and dP specifications and as typical performance curves. In a gain of +2 , driving $150 \Omega$, with standard video test levels at the input, the EL2045 exhibits dG and dP of only $0.02 \%$ and $0.07^{\circ}$ at NTSC and PAL. Because dG and dP can vary with different DC offsets, the video performance of the EL2045 has been characterized over the entire DC offset range from -0.714 V to +0.714 V . For more information, refer to the curves of dG and dP vs DC Input Offset.

The output drive capability of the EL2045 allows it to drive up to 2 back-terminated loads with good video performance. For more demanding applications such as greater output drive or better video distortion, a number of alternatives such as the EL2120, EL400, or EL2074 should be considered.

## Output Drive Capability

The EL2045 has been designed to drive low impedance loads. It can easily drive $6 \mathrm{~V}_{\mathrm{PP}}$ into a $150 \Omega$ load. This high output drive capability makes the EL2045 an ideal choice for RF, IF and video applications. Furthermore, the current drive of the EL2045 remains a minimum of 35 mA at low temperatures.

## Printed-Circuit Layout

The EL2045 is well behaved, and easy to apply in most applications. However, a few simple techniques will help assure rapid, high quality results. As with any high-frequency device, good PCB layout is necessary for optimum performance. Ground-plane construction is highly recommended, as is good power supply bypassing. A $0.1 \mu \mathrm{~F}$ ceramic capacitor is recommended for bypassing both supplies. Pin lengths should be as short as possible, and bypass capacitors should be as close to the device pins as possible. For good AC performance, parasitic capacitances should be kept to a minimum at both inputs and at the output. Resistor values should be kept under $5 \mathrm{k} \Omega$ because of the RC time constants associated with the parasitic capacitance. Metal-film and carbon resistors are both acceptable, use of wire-wound resistors is not recommended because of their parasitic inductance. Similarly, capacitors should be low-inductance for best performance.

## The EL2045 Macromodel

This macromodel has been developed to assist the user in simulating the EL2045 with surrounding circuitry. It has been developed for the PSPICE simulator (copywritten by the Microsim Corporation), and may need to be rearranged for other simulators. It approximates DC, AC, and transient response for resistive loads, but does not accurately model capacitive loading. This model is slightly more complicated than the models used for low-frequency op-amps, but it is much more accurate for AC analysis.

The model does not simulate these characteristics accurately:

- Noise
- Settling time
- Non-linearities
- Temperature effects
- Manufacturing variations
- CMRR
- PSRR


## EL2045

## EL2045 Macromodel

* Connections: +input
* | |-input
* | | +Vsupply
* | | | -Vsupply
* 

$\begin{array}{llllll}\text {.subckt M2045 } & 3 & 2 & 7 & 4 & 6\end{array}$
*

* Input stage
* 

ie 7370.9 mA
r6 3637400
r7 3837400
rc1 430850
rc2 439850
q1 30336 qp
q2 39238 qpa
ediff 33039301.0
rdiff 330 1Meg
*

* Compensation Section
* 

ga 034330 1m rh 3402 Meg ch 3401.5 pF rc 3440 1K cc 4001 pF
*

* Poles
* 

ep 4104001
rpa 4142200
cpa 420 2pF
rpb 4243200
cpb 430 2pF

* Output Stage
* 

ios1 7501.0 mA
ios2 5141.0 mA
q3 44350 qp
q4 74351 qn
q5 75052 qn
q6 45153 qp
ros1 52625
ros2 65325
*

* Power Supply Current
ips 742.7 mA
* 
* Models
* 

.model qn npn(is=800E-18 bf=200 tf=0.2nS)
.model qpa pnp(is=864E-18 bf=100 tf=0.2nS)
.model qp pnp(is=800E-18 bf=125 tf=0.2nS)
.ends

## EL2045 Macromodel (Continued)



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