

200MHz Current Feedback Amplifier

Features

- 200MHz -3dB bandwidth, $A_V = 2$
- Disable/enable
- 12ns settling to 0.05%
- $V_S = \pm 5V @ 15mA$
- Low distortion: HD2, HD3 @ -60dBc at 20MHz
- Differential gain 0.02% at NTSC, PAL
- Differential phase 0.01° at NTSC, PAL
- · Overload/short-circuit protected
- ±1 to ±8 closed-loop gain range
- · Low cost

Applications

- · Video gain block
- Video distribution
- · HDTV amplifier
- Analog multiplexing (using disable)
- Power-down mode (using disable)
- High-speed A/D conversion
- D/A I-V conversion
- Photodiode, CCD preamps
- IF processors
- High-speed communications

Ordering Information

Part No.	Temp. Range	Package	Outline #
EL2070CN	-40°C to +85°C	8-Pin P-DIP	MDP0031
EL2070CS	-40°C to +85°C	8-Lead SO	MDP0027

General Description

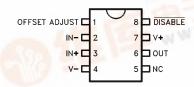
The EL2070C is a wide bandwidth, fast settling monolithic amplifier incorporating a disable/enable feature. Built using an advanced complementary bipolar process, this amplifier uses current-mode feedback to achieve more bandwidth at a given gain than conventional operational amplifiers. Designed for closed-loop gains of ± 1 to ± 8 , the EL2070C has a 200MHz -3dB bandwidth (Av = +2), and 12ns settling to 0.05% while consuming only 15mA of supply current. Furthermore, the fast disable/enable times of 200ns/100ns allow rapid analog multiplexing.

The EL2070C is an obvious high-performance solution for video distribution and line-driving applications, especially when its disable feature can be used for fast analog multiplexing. Furthermore, the low 15mA supply current, and the very low 5mA of supply current when disabled suggest use in systems where power is critical. With differential gain/phase of 0.02%/0.01°, guaranteed video specifications, and a minimum 50mA output drive, performance in these areas is assured.

The EL2070C's settling to 0.05% in 12ns, low distortion, and ability to drive capacitive loads make it an ideal flash A/D driver. The wide 200MHz bandwidth and extremely linear phase allow unmatched signal fidelity. D/A systems can also benefit from the EL2070C, especially if linearity and drive levels are important.

Connection Diagrams

DIP and SO Package - Top View



Manufactured under U.S. Patent No. 4,893,091

Note: All information contained in this data sheet has been carefully checked and is believed to be accurate as of the date of publication; however, this data sheet cannot be a "controlled document". Current revisions, if any, to these specifications are maintained at the factory and are available upon your request. We recommend checking the revision level before finalization of your design documentation.



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Absolute Maximum Ratings (T_A = 25°C)

Supply Voltage (V_S) $\pm 7V$ $\theta_{JA} = 175^{\circ}\text{C/W SO-8}$ Output Current Applied Output Voltage (Disabled) 70mA $\pm V_S$ (Output is short-circuit protected to ground, however, Power Dissipation See Curves maximum reliability is obtained if I_{OUT} does not exceed 70mA) Operating Temperature Common-Mode Input Voltage $\pm V_S$ EL2070C -40C to +85C Differential Input Voltage 5V Lead Temperature (Soldering, 5 Seconds) 300°C Disable Input Voltage +V_S, -1V 175°C Junction Temperature

Thermal Resistance Important Note:

All parameters having Min/Max specifications are guaranteed. Typ values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$.

Storage Temperature-60°C to +150°C

 $\theta_{JA} = 95^{\circ}C/W \text{ P-DIP}$

Open Loop DC Electrical Characteristics

 $V_S = \pm 5 V,\, R_L = 100 \Omega$ unless otherwise specified

Parameter	Description	Test Conditions	Temp	Min	Тур	Max	Unit
V _{OS}	Input Offset Voltage		25°C		2	5.5	mV
			T _{MIN}			8.2	mV
			T_{MAX}			9.0	mV
d(V _{OS})/dT	Average Offset	[1]	All		10.0	40.0	μV/°C
	Voltage Drift						
+I _{IN}	+Input Current		25°C, T _{MAX}		10	25.0	μΑ
			T _{MIN}			36.0	μΑ
d(+I _{IN})/dT	Average +Input	[1]	All		50.0	200.0	nA/°C
	Current Drift						
-I _{IN}	-Input Current		25.0°C		10	30	μΑ
			T _{MIN} , T _{MAX}			46	μΑ
d(-I _{IN})/dT	Average -Input	[1]	All		50.0	200.0	nA/°C
	Current Drift						
PSRR	Power Supply		All	45.0	50.0		dB
	Rejection Ratio						
CMRR	Common-Mode		All	40.0	50.0		dB
	Rejection Ratio						
I_S	Supply Current—Quiescent	No Load	All		16.0	20.0	mA
IS _{OFF}	Supply Current—Disabled	[2]	All		4.0	7.0	mA
$+R_{IN}$	+Input Resistance		25°C, T _{MAX}	100.0	200.0		kΩ
			T_{MIN}	50.0			kΩ
C _{IN}	Input Capacitance		All		0.5	2.0	pF
R _{OUT}	Output Impedance (DC)		All		0.1	0.2	Ω
R _{OUT} D	Output Resistance (DC)	Disabled	All	100.0	200.0		kΩ
C _{OUT} D	Output Capacitance (DC)	Disabled	All		0.5	2.0	pF
CMIR	Common-Mode	[3]	25°C, T _{MAX}	2.0	2.1		V
	Input Range		T _{MIN}	1.2			V
I _{OUT}	Output Current		25°C, T _{MAX}	50.0	70.0		mA
			T _{MIN}	35.0			mA
V _{OUT}	Output Voltage Swing	No Load	All	3.3	3.5		V
V _{OUT} L	Output Voltage Swing	100Ω	25°C	3.0	3.4		V

Open Loop DC Electrical Characteristics (Continued)

 $V_S = \pm 5V$, $R_L = 100\Omega$ unless otherwise specified

Parameter	Description	Test Conditions	Temp	Min	Тур	Max	Unit
-ICMR	Input Current Common		25°C		8.0	33.0	μA/V
	Mode Rejection						
+IPSR	+Input Current Power		25°C		1.0	3.6	μA/V
	Supply Rejection						
-IPSR	-Input Current Power		25°C		20	24	μA/V
	Supply Rejection						
R _{OL}	Transimpedance		25°C	30.0	125.0		V/mA
			T_{MIN}		80.0		V/mA
			T _{MAX}		140.0		V/mA
I _{LOGIC}	Pin 8 Current @ 0V		All		0.8	1.2	mA
V _{DIS}	Maximum Pin 8		All			0.5	V
	V to Disable						
V _{EN}	Minimum Pin 8		All	3.5			V
	V to Enable						
I_{DIS}	Minimum Pin 8		All	350.0			μΑ
	I to Disable						
I _{EN}	Maximum Pin 8		All			60.0	μΑ
	I to Enable						

- $1. \quad \text{Measured from T_{MIN} to T_{MAX}}.$
- 2. Supply current when disabled is measured at the negative supply.
- 3. Common-mode input range for rated performance.

Closed-Loop AC Electrical Characteristics

 $V_S=\pm 5V,\,R_F=250\Omega,\,A_V=+2,\,R_L=100\Omega$ unless otherwise specified

Parameter	Description	Test Conditions	Temp	Min	Тур	Max	Unit
FREQUENC'	Y RESPONSE	•	•	•	•		•
SSBW	-3dB Bandwidth		25°C	150.0	200.0		MHz
	$(V_{OUT} < 0.5V_{PP})$		T _{MIN}	150.0			MHz
			T_{MAX}	120.0			MHz
LSBW	-3dB Bandwidth	$A_V = +5$	All	35.0	50.0		MHz
	$(V_{OUT} < 5.0V_{PP})$						
GAIN FLAT	NESS				•		
GFPL	Peaking	<40MHz	25°C		0.0	0.3	dB
	$V_{OUT} < 0.5V_{PP}$		T _{MIN} , T _{MAX}			0.4	dB
GFPH	Peaking	>40MHz	25°C		0.0	0.5	dB
	$V_{OUT} < 0.5V_{PP}$		T _{MIN} , T _{MAX}			0.7	dB
GFR	$ \begin{array}{ccc} \text{GFR} & & \text{Rolloff} \\ & & V_{OUT} \leq 0.5 V_{PP} \end{array} $	<75MHz	25°C		0.6	1.0	dB
			T _{MIN}			1.0	dB
			T_{MAX}			1.3	dB
LPD	Linear Phase Deviation	<75MHz	25°C, T _{MIN}		0.2	1.0	0
	$V_{OUT} < 0.5V_{PP}$		T_{MAX}			1.2	0
TIME-DOM/	AIN RESPONSE	1	•	1	ı		
t_{r1}, t_{f1}	Rise Time, Fall Time	0.5V Step	All		1.6	2.4	ns
t_{r2}, t_{f2}	Rise Time, Fall Time	5.0V Step	All		6.5	10.0	ns

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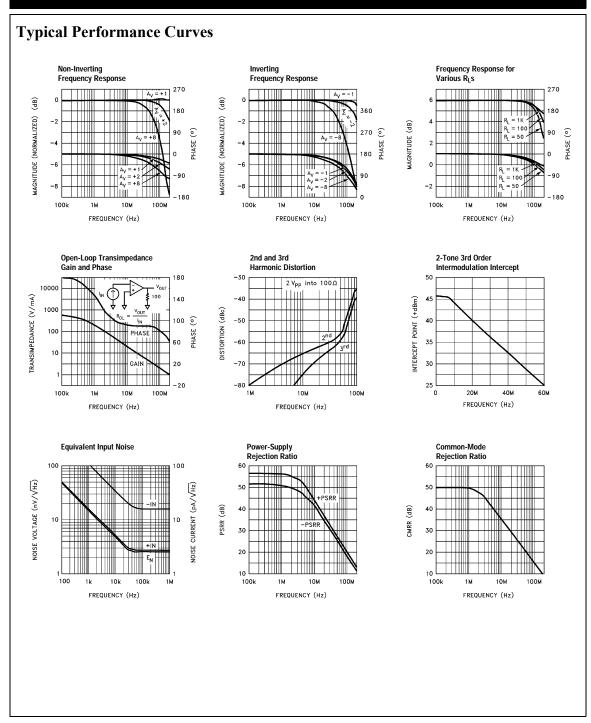
Closed-Loop AC Electrical Characteristics V_S = ±5V, R_F = 250 Ω , A_V = +2, R_L = 100 Ω unless otherwise specified

Parameter	Description	Test Conditions	Temp	Min	Тур	Max	Unit
t _{s1}	Settling Time to 0.1%	2.0V Step	All		10.0	13.0	ns
t _{s2}	Settling Time to 0.05%	2.0V Step	All		12.0	15.0	ns
OS	Overshoot	0.5V Step	25°C, T _{MAX}		0.0	10.0	%
			T _{MIN}			15.0	%
SR	Slew Rate	$A_V = +2$	All	430.0	700.0		V/µs
		A _V = - 2	All		1600.0		V/µs
DISTORTIO	N	•		•	•	•	•
HD2	2nd Harmonic Distortion	$2V_{PP}$	25°C		-60.0	-45.0	dBc
	at 20MHz		T _{MIN}			-40.0	dBc
			T _{MAX}			-45.0	dBc
HD3	3rd Harmonic Distortion	2V _{PP}	25°C		-60.0	-50.0	dBc
	at 20MHz		T _{MIN} , T _{MAX}			-50.0	dBc
EQUIVALEN	T INPUT NOISE	•	1				
NF	Noise Floor	[1]	25°C		-157.0	-154.0	dBm (1Hz)
	>100kHz		T _{MIN}			-154.0	dBm (1Hz)
			T _{MAX}			-153.0IV	dBm (1Hz)
INV	Integrated Noise	[1]	25°C		40.0	57.0	μV
	100kHz to 200MHz		T _{MIN}			57.0	μV
			T _{MAX}			63.0	μV
DISABLE/EN	NABLE PERFORMANCE	•			•	•	•
T _{OFF}	Disable Time to >50dB	10MHz	All		1000.0	IV	ns
Ton	Enable Time		All		200.0		ns
OFFIso	Off Isolation	10MHz	All	55.0	59.0		dB
VIDEO PERI	FORMANCE	•			•	•	•
d_G	Differential Gain [2]	NTSC/PAL	25°C		0.02	0.08	% pp
d _P	Differential Phase [2]	NTSC/PAL	25°C		0.01	0.08	° pp
d_G	Differential Gain [2]	30MHz	25°C		0.05	0.18	% pp
d _P	Differential Phase [2]	30MHz	25°C		0.05	0.18	° pp
VBW	-0.1dB Bandwidth [2]		25°C	30.0	60.0		MHz

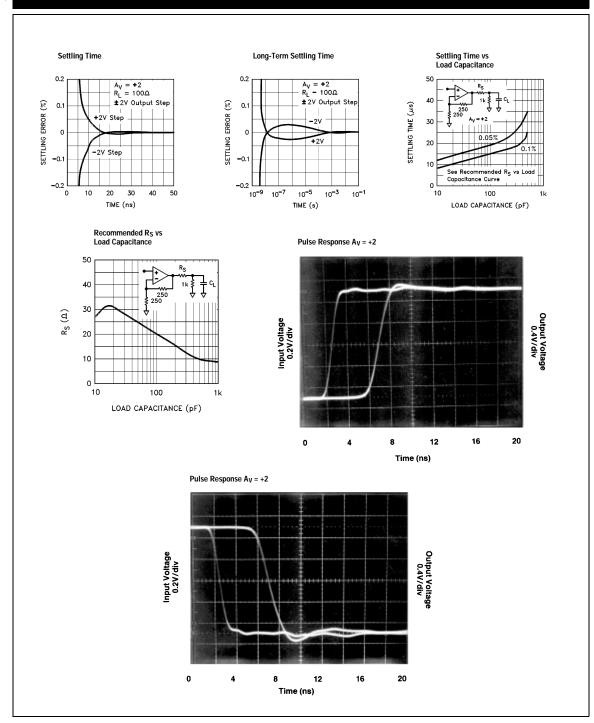
^{1.} Noise Tests are performed from 5MHz to 200MHz.

^{2.} Differential gain/phase tests are with R_L = 100 Ω . For other values of R_L , see curves.

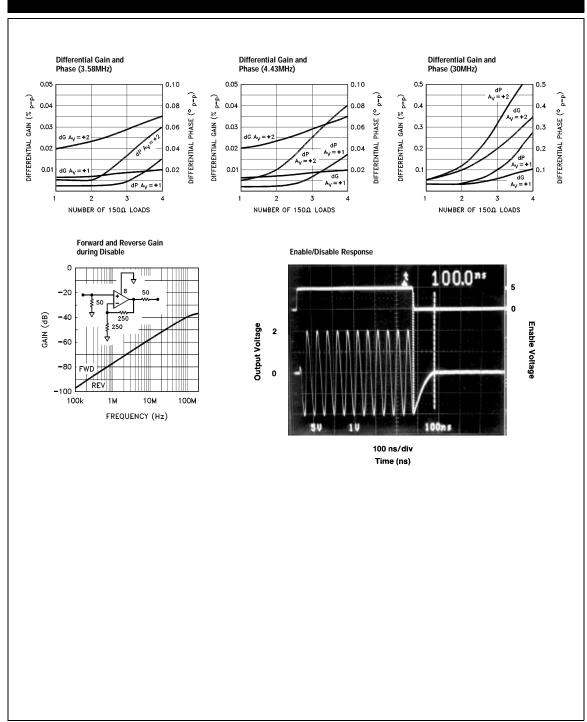
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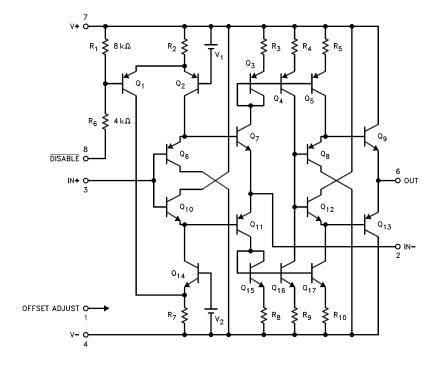
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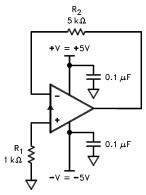
200MHz Current Feedback Amplifier



Equivalent Circuit



Burn-In Circuit



ALL PACKAGES USE THE SAME SCHEMATIC.

Applications Information

Theory of Operation

The EL2070C has a unity gain buffer from the non-inverting input to the inverting input. The error signal of the EL2070C is a current flowing into (or out of) the inverting input. A very small change in current flowing through the inverting input will cause a large change in the output voltage. This current amplification is called the transimpedance ($R_{\rm OL}$) of the EL2070C [$V_{\rm OUT}=(R_{\rm OL})*(-I_{\rm IN})$]. Since $R_{\rm OL}$ is very large, the current flowing into the inverting input in the steady-state (non-slewing) condition is very small.

Therefore we can still use op-amp assumptions as a firstorder approximation for circuit analysis, namely that:

- 1. The voltage across the inputs is approximately 0V.
- 2. The current into the inputs is approximately 0mA.

Resistor Value Selection and Optimization

The value of the feedback resistor (and an internal capacitor) sets the AC dynamics of the EL2070C. The nominal value for the feedback resistor is $250\Omega,$ which is the value used for production testing. This value guarantees stability. For a given closed-loop gain the bandwidth may be increased by decreasing the feedback resistor and, conversely, the bandwidth may be decreased by increasing the feedback resistor.

Reducing the feedback resistor too much will result in overshoot and ringing and eventually oscillations. Increasing the feedback resistor results in a lower -3dB frequency. Attenuation at high frequency is limited by a zero in the closed-loop transfer function which results from stray capacitance between the inverting input and ground. Consequently, it is very important to keep stray capacitance to a minimum at the inverting input.

Differential Gain/Phase

An industry-standard method of measuring the distortion of a video component is to measure the amount of differential gain and phase error it introduces. To measure these, a 40 IREpp reference signal is applied to the device with 0V DC offset (0 IRE) at 3.58MHz for NTSC, 4.43MHz for PAL, and 30MHz for HDTV. A

second measurement is then made with a 0.714V DC offset (100 IRE). Differential Gain is a measure of the change in amplitude of the sine wave, and is measured in percent. Differential Phase is a measure of the change in phase, and is measured in degrees. Typically, the maximum positive and negative deviations are summed to give peak values.

In general, a back terminated cable (75 Ω in series at the drive end and 75 Ω to ground at the receiving end) is preferred since the impedance match at both ends will absorb any reflections. However, when double-termination is used, the received signal is reduced by half; therefore a gain of 2 configuration is typically used to compensate for the attenuation. In a gain of 2 configuration, with output swing of 2Vpp, with each back-terminated load at 150 Ω . The EL2070C is capable of driving up to 4 back-terminated loads with excellent video performance. Please refer to the typical curves for more information on video performance with respect to frequency, gain, and loading.

Capacitive Feedback

The EL2070C relies on its feedback resistor for proper compensation. A reduction of the impedance of the feedback element results in less stability, eventually resulting in oscillation. Therefore, circuit implementations which have capacitive feedback should not be used because of the capacitor's impedance reduction with frequency. Similarly, oscillations can occur when using the technique of placing a capacitor in parallel with the feedback resistor to compensate for shunt capacitances from the inverting input to ground.

Offset Adjustment Pin

Output offset voltage of the EL2070C can be nulled by tying a 10k potentiometer between +V_S and -V_S with the slider attached to pin 1. A full-range variation of the voltage at pin 1 to $\pm 5V$ results in an offset voltage adjustment of at least $\pm 10 mV$. For best settling performance pin 1 should be bypassed to ground with a ceramic capacitor located near to the package, even if the offset voltage adjustment feature is not being used.

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Printed Circuit Layout

As with any high frequency device, good PCB layout is necessary for optimum performance. Ground plane construction is a requirement, as is good power-supply and Offset Adjust bypassing close to the package. The inverting input is sensitive to stray capacitance, therefore connections at the inverting input should be minimal, close to the package, and constructed with as little coupling to the ground plane as possible.

Capacitance at the output node will reduce stability, eventually resulting in peaking, and finally oscillation if the capacitance is large enough. The design of the EL2070C allows a larger capacitive load than comparable products, yet there are occasions when a series resistor before the capacitance may be needed. Please refer to the graphs to determine the proper resistor value needed.

Disable/Enable Operation

The EL2070C has a disable/enable control input at pin 8. The device is enabled and operates normally when pin 8 is left open or tied to pin 7. When more than $350\mu A$ is pulled from pin 8, the EL2070C is disabled. The output becomes a high impedance, the inverting input is no longer driven to the positive input voltage, and the supply current is reduced by 2/3. To make it easy to use this feature, there is an internal resistor to limit the current to a safe level (0.8mA) if pin 8 is grounded.

To draw current out of pin 8 an open-collector TTL output, a 5V CMOS output, or an NPN transistor can be used.

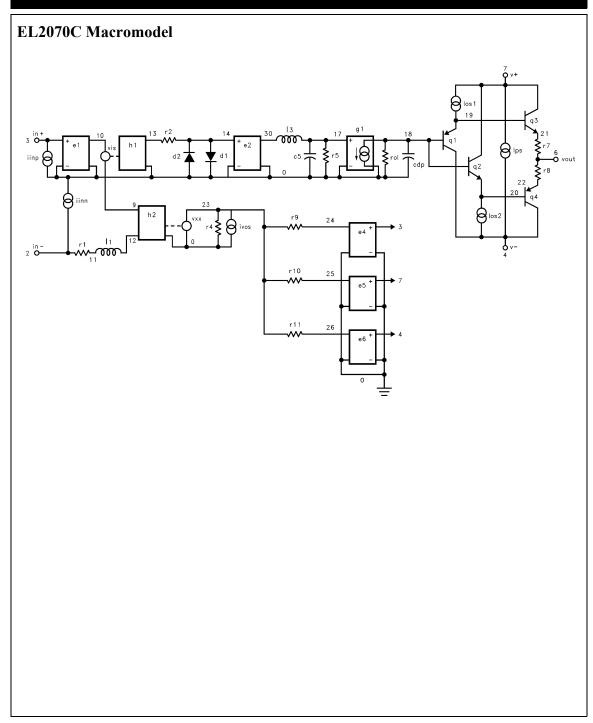
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EL2070C Macromodel
* Revision A. March 1992
* Enhancements include PSRR, CMRR, and Slew Rate Limiting
* Connections: +input

* | -input
              -input
                  +Vsupply
                  | -Vsupply
                     output
subckt M2070C 3 2 7 4 6
* Input Stage
e1 10 0 3 0 1.0
vis 10 9 0V
h2 9 12 vxx 1.0
r1 2 11 50
11 11 12 48nH
iinp 3 0 8µA
iinm 2 0 8μA
* Slew Rate Limiting
h1 13 0 vis 600
r2 13 14 1K
d1 14 0 dclamp
d2 0 14 dclamp
* High Frequency Pole
e2 30 0 14 0 0.00166666666
13\ 30\ 17\ 0.1 \mu H
c5 17 0 0.1pF
r5 17 0 500
* Transimpedance Stage
g1 0 18 17 0 1.0
rol 18 0 150K
cdp 18 0 2.8pF
* Output Stage
q1 4 18 19 qp
q2 7 18 20 qn
q3 7 19 21 qn
q4 4 20 22 qp
r7 21 6 2
r8 22 6 2
ios1 7 19 2.5mA
ios2 20 4 2.5mA
* Supply Current
ips 7 4 9mA
* Error Terms
ivos 0 23 5mA
vxx 23 0 0V
e4 24 0 3 0 1.0
e5 25 0 7 0 1.0
```

EL2070C
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e6 26 0 4 0 1.0 r9 24 23 3K r10 25 23 1K r11 26 23 1K
* Models * Models .model qn npn (is=5e-15 bf=200 tf=0.05nS) .model qp pnp (is=5e-15 bf=200 tf=0.05nS) .model dclamp d(is=1e-30 ibv=0.266 bv=1.3 n=4)
.model dclamp d(is=1e-30 ibv=0.266 bv=1.3 n=4) .ends

EL2070C 200MHz Current Feedback Amplifier



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14



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September 26, 2001