





EL4393C

Triple 80 MHz Video Amplifier w/Disable

EL4393C

January 1996 Rev

Features

- 80 MHz -3 dB bandwidth for gains of 1 to 10
- 900 V/μs slew rate
- 10 MHz bandwidth flat to 0.1 dB
- Excellent differential gain and
- phase
- TTL/CMOS compatible
- Available in SOL-16

Applications

- RGB drivers
- RGB multiplexers
- RGB gain blocks
- Video gain blocks
- Coax cable driver
- ADC drivers/input multiplexer

Ordering Information

 Part No.
 Temp. Range
 Package
 Outline #

 EL4393CN
 -40°C to +85°C
 16-Lead P-DIP
 MDP0031

 EL4393CM
 -40°C to +85°C
 16-Lead SOL
 MDP0027

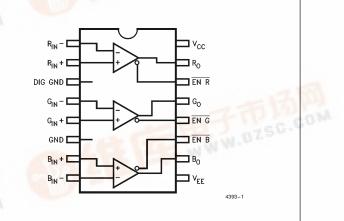
General Description

The EL4393C is three wideband current-feedback amplifiers optimized for video performance. Each amplifier can drive a load of 150Ω at video levels. Each amplifier has a disable capability, which is controlled by a TTL/CMOS compatible logic signal. The EL4393C operates on supplies as low as $\pm 4V$ up to $\pm 15V$.

Being a current-feedback design, the bandwidth stays relatively constant at approximately 80 MHz over the ± 1 to ± 10 gain range. The EL4393C has been optimized for use with 1300 Ω feedback resistors at a gain of 2.

When the outputs are disabled, the supply current consumption drops, by about 4 mA per channel that is disabled. This feature can be used to reduce power dissipation.

Connection Diagram



Note: All information contained in this data sheet has been carefully checked and is believed to be accurate as of the date of publication; however, this data sheet cannot be a "controlled document". Current revisions, if any, to these specifications are maintained at the factory and are available upon your request. We recommend checking the revision level before finalization of your design documentation.



Absolute Maximum Ratings (T_A = 25°C)

 $\begin{array}{l} \mbox{Voltage between } V_S{}^+ \mbox{ and } V_S{}^- \\ \mbox{Voltage at } V_S{}^+ \\ \mbox{Voltage at } V_S{}^- \\ \mbox{Voltage between } V_{IN}{}^+ \mbox{ and } V_{IN}{}^- \\ \mbox{Current into } V_{IN}{}^+ \mbox{ or } V_{IN}{}^- \end{array}$

Internal Power Dissipation Operating Ambient Temperature Range Operating Junction Temperature Storage Temperature Range See Curves -40°C to +85°C 150°C -65°C to +150°C

Important Note:

All parameters having Min/Max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality inspection. Elantec performs most electrical tests using modern high-speed automatic test equipment, specifically the LTX77 Series system. Unless otherwise noted, all tests are pulsed tests, therefore $T_J = T_C = T_A$.

+ 33V

+18V

-18V

 $\pm 6V$

5 mA

Test Level	Test Procedure
Ι	100% production tested and QA sample tested per QA test plan QCX0002.
Π	100% production tested at $T_{\rm A}=25^{\circ}{\rm C}$ and QA sample tested at $T_{\rm A}=25^{\circ}{\rm C}$,
	T_{MAX} and T_{MIN} per QA test plan QCX0002.
III	QA sample tested per QA test plan QCX0002.
IV	Parameter is guaranteed (but not tested) by Design and Characterization Data.
v	Parameter is typical value at $T_{\rm A}=25^{\circ}{\rm C}$ for information purposes only.

Open Loop DC Electrical Characteristics Supplies at $\pm 15V$, Load = $1 \text{ K}\Omega$

Parameter	Description	Temp	Min	Тур	Max	Test Level	Units
V _{OS}	Input Offset Voltage	+ 25°C		2	±15	I	mV
TCV _{OS}	Temperature Coefficient of V_{OS}	Full		50		v	μV/°C
I_B^+	${\rm I_{IN}}^+$ Input Bias Current	+ 25°C		0.2	5	I	μA
I_B^-	I_{IN}^- Input Bias Current	+ 25°C		10	65	I	μΑ
tci _b -	Temperature Coefficient of I_B^-	Full		25		v	nA/°C
CMRR	Common-Mode Rejection Ratio (Note 1)	+ 25°C	50	58		I	dB
-ICMR	I _{IN} ⁻ Input Common-Mode Current (Note 1)	+ 25°C		3	8	I	μA/V
PSRR	Power Supply Rejection Ratio (Note 2)	+ 25°C	50	58		I	dB
-IPSR	I _{IN} ⁻ Current Supply Rejection (Note 2)	+ 25°C		2	5	I	μA/V
R _{OL}	Transimpedance	+ 25°C	100	217		I	kΩ
R _{IN}	IN ⁺ Input Impedance	+ 25°C		2		I	MΩ
V _{IN}	IN ⁺ Input Range	+ 25°C	±13	±13.5		I	v
vo	Output Voltage Swing; $R_L = 1 k\Omega$	+ 25°C	±12	±13		I	v

Parameter	Description	Temp	Min	Тур	Max	Test Level	Units
I _{SC}	Short-Circuit Current (Note 3)	+ 25°C	40	70		I	mA
I _{O, DIS}	Output Current when Disabled	+ 25°C		5	150	I	μΑ
dis V _{IL}	Disable Voltage for Logic Low	+ 25°C			0.8	I	v
dis V _{IH}	Disable Voltage for Logic High	+ 25°C	2.2			I	v
DIS I _{IL}	Disable Logic Low Input Current	+ 25°C		3	25	I	μA
DIS I _{IH}	Disable Logic High Input Current	+ 25°C		0	5	I	μΑ
I _{CC (en)}	Positive Supply Current all Channels Enabled	+ 25°C	15	20	±29	I	mA
I _{CC (dis)}	Positive Supply Current all Channels Disabled	+ 25°C	6	11	16	I	mA
I _{EE (en)}	Negative Supply Current all Channels Enabled	+ 25°C	13	18	±28	I	mA
I _{EE (dis)}	Negative Supply Current all Channels Disabled	+ 25°C	4	9	14	I	mA

Note 1: $V_{CM} = \pm 10V$ for $V_S = \pm 15V$. Note 2: V_{OS} is measured at $V_S = \pm 4.5V$ and $V_S = \pm 16V$, both supplies are changed simultaneously. Note 3: Only one output short circuited. Pulse test or use heatsink.

Closed Loop AC Electrical Characteristics Supplies at \pm 15V, Load = 150 Ω and 15 pF, except where noted. Rf1 and Rf2 = 1500 Ω ; A_V = 2, T_A = 25°C. (See note 8 re: test fixture)

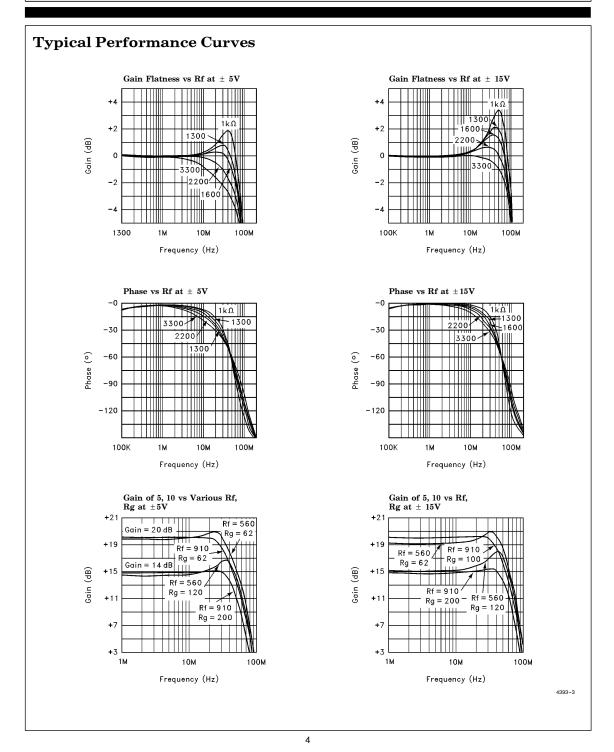
Parameter	Description	Min	Тур	Max	Test Level	Units
SR	Slew Rate (Note 4)		960		v	V/µs
SR	Slew Rate w/ \pm 5V Supplies (Note 5)		470		IV	V/µs
t _s	Settling Time to 1% 5V _{p-p} 5V Step (Note 6)		32		v	ns
BW	Bandwidth, -3 dB ±5V Supplies, -3 dB		80 60		IV IV	MHz MHz
BW	${f Bandwidth,-0.1~dB}\ \pm5V$ Supplies, $-0.1~dB$		16 21		IV IV	MHz MHz
Peaking	-3 dB BW Tests		0.6		IV	dB
dG	Differential Gain at 3.58 MHz at \pm 5V Supplies (Note 7)		0.03 0.30		v v	% %
dθ	Differential Phase at 3.58 MHz at \pm 5V Supplies (Note 7)		0.088 0.096		v v	(°) (°)

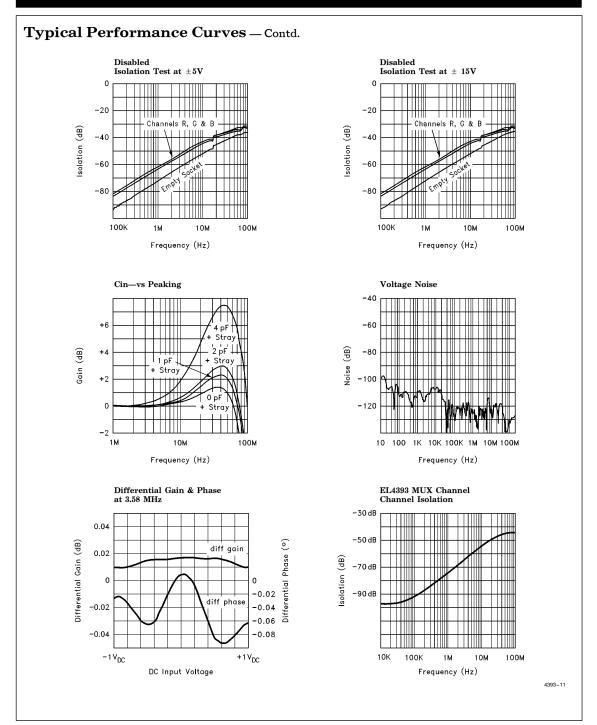
Note 4: $R_{\rm L}$ = 300 Ω , -5V to +5V swing, SR measured at 20% to 80%.

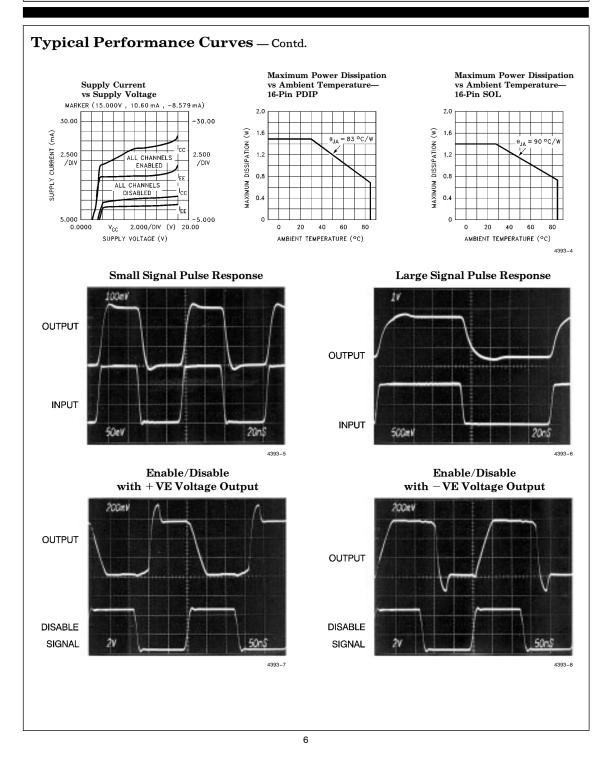
Note 5: -2V to +2V swing, SR measured at 20% to 80%.

Note 6: $R_L = 300\Omega$.

Note 7: DC offset from -0.7V through +0.7V AC amplitude is 286 mV_{P.P}, equivalent to 40 ire.
 Note 8: Test fixture was designed to minimize capacitance at the I_N⁺ input. A "good" fixture should have less than 2 pF of stray capacitance to ground at this very sensitive pin. See application notes for further details.







Typical Application for EL4393, and General Rules for PCB

Layout

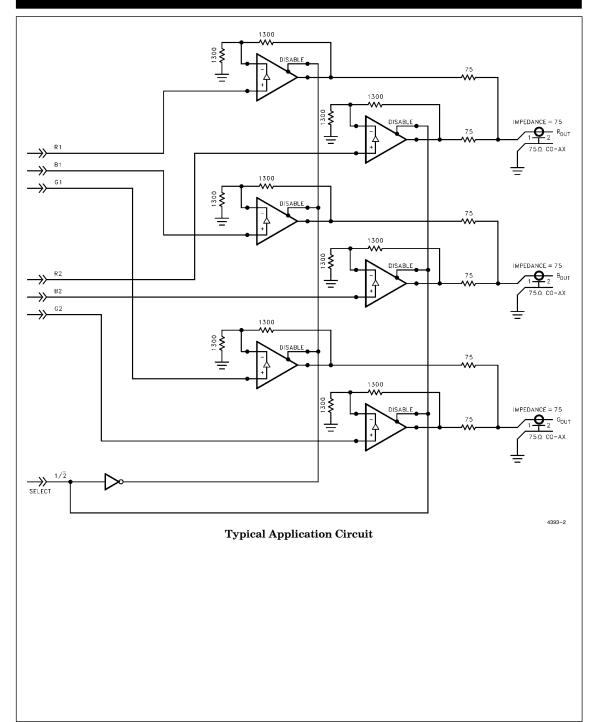
The figure shows two EL4393's configured as a 2:1 RGB multiplexer, and cable driver, driving 75 Ω , back terminated cables. Each channel of the EL4393 is configured to give a gain of two, to make up for the losses of the back terminating resistor.

In this example, the Disable pins of each RGB section are driven by a complementary TTL "select" signal. Larger multiplexers can be assembled, with a 1-of-n TTL decoder selecting each RGB triplet.

The circuit gives channel isolations of typically better than -50 dB at 10 MHz, and with a 20 dB/decade slope, extending down to better than -90 dB at frequencies below 100 kHz.

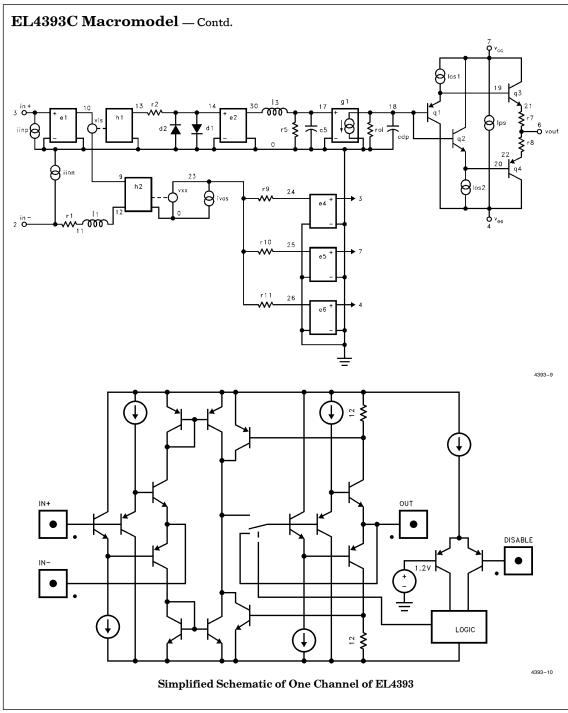
The schematic does not show things like power supply decoupling, or pcb layout, grounding and signal returns, but these will all affect the overall performance of the circuit, and care should be taken with these aspects. It is recommended that the V_{CC} and V_{EE} pins each be decoupled by a 0.1 μ F NPO or X7R dielectric ceramic capacitors to ground within 0.1 inch of the part, and in parallel with the 0.1 μ F, A 47 μ F tantalum capacitor, also to ground. The 47 μ F capacitors should be within 0.25 inch of their power pins. The ground plane should be underneath the package, but cut away from the In - inputs. Care should be taken with the center channel feedback-it must be kept away from any of the in + or in - pins, if it has to go under the package. Route the G-out line between the pin 3 ground and the pin 4 In - if going under the package is essential. Otherwise, loop the Gout trace around all the other circuitry, to its Rf resistor. The Rf and if used, Rg resistors should be on the input side of the package, to minimize trace length on the In - pins.

The digital input disables are on the output side of the package, so that a good ground plane down the center of the board underneath the package will isolate any fast edges from the sensitive inputs.



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EL4393C Macromodel * Revision A, July 1993 * Output Stage * Enhancements include PSRR, CMRR, and Slew Rate Limiting * Connections: + input q1 4 18 19 qp * q2 7 18 20 qn -Input* T +Vsupply q3 7 19 21 qn q4 4 20 22 qp * -Vsupplyr7 21 6 4 * Putput r8 22 6 4 * ios1 7 19 2.5 mA * subckt EL4393/EL 3 2 7 4 6 ios2 20 4 2.5 mA * Input Stage * Error Terms e1 10 0 3 0 1.0 ivos 0 23 2 mA vis 10 9 0V vxx 23 0 0V h2 9 12 vxx 1.0 e4 24 0 3 0 1.0 r1 2 11 50 e5 25 0 7 0 1.0 11111229 nH e6 26 0 4 0 1.0 iinp 3 0 0.2 μA r9 24 23 1K iinm 2 0 10 μ A r10 25 23 1K r11 26 33 1K * Slew Rate Limiting * Models h1 13 0 vis 600 r2 13 14 1K .model qn npn (is = 5e – 15 bf = 100 tf = 0.2nS) d1 14 0 dclamp .model qp pnp (is = 5e-15 bf= 100 tf= 0.2nS) d2 0 14 dclamp .model dclamp d (is = 1e - 30 ibv = 0.266 bv = 1.5 n = 4) .ends * High Frequency Pole e2 30 0 14 0 0.00166666666 15 30 17 1.2 μH $c5\ 17\ 0\ 1\ pF$ r5 17 0 500 * Transimpedance Stage g1 0 18 17 0 1.0 rol 18 0 250k $cdp\;18\;0\;2.2\;p\mathbf{F}$



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General Disclaimer

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