# Single／Dual，Ultra－Fast，Low－Power， Precision TTL Comparators 

$\qquad$
General Description
The MAX913 single and MAX912 dual high－speed， low－power comparators have differential inputs and complementary TTL outputs．Fast propagation delay （10ns typ），extremely low supply current，and a wide common－mode input range that includes the negative rail make the MAX912／MAX913 ideal for low－power， high－speed，single +5 V （or $\pm 5 \mathrm{~V}$ ）applications such as V／F converters or switching regulators．
The MAX912／MAX913 outputs remain stable through the linear region．This feature eliminates output instability common to high－speed comparators when driven with a slow－moving input signal．
The MAX912／MAX913 can be powered from a single +5 V supply or a $\pm 5 \mathrm{~V}$ split supply．The MAX913 is an improved plug－in replacement for the LT1016．It pro－ vides significantly wider input voltage range and equiva－ lent speed at a fraction of the power．The MAX912 dual comparator has equal performance to the MAX913 and includes independent latch controls．

## Applications

Zero－Crossing Detectors
Ethernet Line Receivers
Switching Regulators
High－Speed Sampling Circuits
High－Speed Triggers
Extended Range V／F Converters
Fast Pulse Width／Height Discriminators
Ultra Fast（10ns）
Single +5 V or Dual $\pm 5 \mathrm{~V}$ Supply Operation
－Input Range Extends Below Negative Supply
Low Power： $6 \mathrm{~mA}(+5 \mathrm{~V})$ Per Comparator
No Minimum Input Signal Slew－Rate Requirement
－No Power－Supply Current Spiking
Stable in the Linear Region
Lnputs Can Exceed Either Supply
Low Offset Voltage：0．8mV
Ordering Information

| PART | TEMP．RANGE | PIN－PACKAGE |
| :--- | :--- | :--- |
| MAX912CPE | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 16 Plastic DIP |
| MAX912CSE | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 16 Narrow SO |
| MAX912C／D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice ${ }^{*}$ |
| MAX912EPE | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 Plastic DIP |
| MAX912ESE | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 Narrow SO |
| MAX912MJE | $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 16 CERDIP |
| MAX913CPA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 Plastic DIP |
| MAX913CSA | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | 8 SO |
| MAX913C／D | $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | Dice ${ }^{*}$ |
| MAX913EPA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 Plastic DIP |
| MAX913ESA | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 SO |
| MAX913MJA | $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 8 CERDIP |

＊Dice are specified at $T_{A}=+25^{\circ} \mathrm{C}, D C$ parameters only．

Pin Configurations


MAXIA Maxim Integrated Products

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ABSOLUTE MAXIMUM RATINGS

Positive Supply Voltage............................................................ 7 V
Negative Supply Voltage .......................................................-7V
Differential Input Voltage ..................................................... $\pm 15 \mathrm{~V}$ Input Voltage (Referred to V-).......................................................... 0.3 V to 15 V Latch Pin Voltage $\qquad$
$\qquad$ Equal to Supplies Continuous Output Current. $\qquad$ $\pm 20 \mathrm{~mA}$ Continuous Power Dissipation ( $\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}$ )

8 -Pin Plastic DIP (derate $9.09 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) ... 727 mW
8 -Pin SO (derate $5.88 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) $\qquad$ . .471 mW
8 -Pin CERDIP (derate $8.00 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ).
 .640 mW


Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}+=+5 \mathrm{~V}, \mathrm{~V}-=-5 \mathrm{~V}, \mathrm{~V}_{\mathrm{Q}}=1.4 \mathrm{~V}, \mathrm{~V}_{\mathrm{LE}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{MIN}}\right.$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.)


# Single/Dual, Ultra-Fast, Low-Power, Precision TTL Comparators 

## ELECTRICAL CHARACTERISTICS (continued)

$\left(\mathrm{V}+=+5 \mathrm{~V}, \mathrm{~V}-=-5 \mathrm{~V}, \mathrm{~V}_{\mathrm{Q}}=1.4 \mathrm{~V}, \mathrm{~V}_{\mathrm{LE}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{MIN}}\right.$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted).


Note 1: Input Offset Voltage (VOS) is defined as the average of the two input offset voltages, measured by forcing first one output, then the other to 1.4 V . Input Offset Current (los) is defined the same way.
Note 2: Propagation Delay (tpD) and Differential Propagation Delay ( $\Delta \mathrm{tpD}$ ) cannot be measured in automatic handling equipment with low input overdrive values. The MAX912/MAX913 are sample tested to $0.1 \%$ AQL with a 1 V step and 500 mV overdrive at $+25^{\circ} \mathrm{C}$ only. Correlation tests show that tpD and $\Delta$ tPD can be guaranteed with this test, if additional DC tests are performed to guarantee that all internal bias conditions are correct. For low overdrive conditions, VOS is added to the overdrive. Differential Propagation Delay is defined as: $\triangle$ tPD $=$ tPD + tPD -
Note 3: Input latch setup time (tsu) is the interval in which the input signal must be stable prior to asserting the latch signal. The hold time ( $\mathrm{t}_{\mathrm{H}}$ ) is the interval after the latch is asserted in which the input signal must be stable. These parameters are guaranteed by design.
Note 4: Latch Propagation Delay (tLPD) is the delay time for the output to respond when the latch-enable pin is deasserted. See Timing Diagram.

Typical Operating Characteristics
$\left(\mathrm{V}_{+}=5 \mathrm{~V}, \mathrm{~V}-=-5 \mathrm{~V}, \mathrm{~V}_{\mathrm{LE}}=0 \mathrm{~V}, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, unless otherwise noted.)


## Single/Dual, Ultra-Fast, Low-Power, Precision TTL Comparators

MAX912/MAX913
Typical Operating Characteristics (continued)
$\left(\overline{\mathrm{V}+=5 \mathrm{~V}, \mathrm{~V}-=-5 \mathrm{~V}, \mathrm{~V}_{\mathrm{LE}}}=0 \mathrm{~V}, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}, \mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}\right.$, unless otherwise noted.)


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Typical Operating Characteristics (continued)


NEGATIVE-TO-POSITIVE PROPAGATION DELAY


MAX912/MAX913 RESPONSE TO 50 MHz ( $\pm 10 \mathrm{mV}$ p.p) SINE WAVE


MAX912/MAX913 RESPONSE TO SLOW-MOVING TRIANGLE WAVE

MAX912/MAX913 RESPONSE


INDUSTRY STANDARD 686 RESPONSE


## Single/Dual, Ultra-Fast, Low-Power, Precision TTL Comparators

| $\begin{gathered} \text { PIN } \\ \text { MAX912 } \end{gathered}$ | NAME | FUNCTION |
| :---: | :---: | :---: |
| 1 | QA | Comparator A TTL output |
| 2 | $\overline{\mathrm{Q}} \mathrm{A}$ | Comparator A complementary TTL output |
| 3, 14 | GND | Logic ground. Connect both GND pins to ground. |
| 4 | LEA | Comparator A latch enable. QA and $\overline{\mathrm{Q} A}$ are latched when LEA is high or floating. Comparator A latch is transparent when LEA is low. |
| 5, 12 | N.C. | Not internally connected |
| 6 | V- | Negative power supply: -5 V for dual supplies (bypass to GND with a 0.1 HF capacitor), or GND for a single supply |
| 7 | INA- | Comparator A inverting input |
| 8 | INA+ | Comparator A noninverting input |
| 9 | INB+ | Comparator B noninverting input |
| 10 | INB- | Comparator B inverting input |
| 11 | V+ | Positive power supply, +5 V . Bypass to GND with a $0.1 \mu \mathrm{~F}$ capacitor. |
| 13 | LEB | Comparator $B$ latch enable. QB and $\overline{\mathrm{Q} B}$ are latched when LEB is high or floating. Comparator B latch is transparent when LEB is low. |
| 15 | $\overline{\text { Q }}$ | Comparator B complementary TTL output |
| 16 | QB | Comparator B TTL output |
| $\begin{gathered} \text { PIN } \\ \text { MAX913 } \end{gathered}$ | NAME | FUNCTION |
| 1 | V+ | Positive power supply. Bypass to GND with a $0.1 \mu \mathrm{~F}$ capacitor. |
| 2 | $\mathrm{IN}_{+}$ | Noninverting input |
| 3 | IN- | Inverting input |
| 4 | V- | Negative power supply: -5 V for dual supplies (bypass to GND with a $0.1 \mu \mathrm{~F}$ capacitor), or GND for a single supply |
| 5 | LE | Latch enable. $Q$ and $\bar{Q}$ are latched when LE is $T \mathrm{LL}$ high or floating. The comparator latch is transparent when LE is low. |
| 6 | GND | Logic ground |
| 7 | Q | TL output |
| 8 | $\overline{\mathrm{Q}}$ | Complementary TTL output |

# Single/Dual, Ultra-Fast, Low-Power, Precision TTL Comparators 

## Detailed Description

The MAX913 (single) and MAX912 (dual) high-speed comparators have a unique design that prevents oscillation when the comparator is in its linear region. No minimum input slew rate is required.
Many high-speed comparators oscillate in the linear region, as shown in the Typical Operating Characteristics' industry-standard 686 response graph. One way to overcome this oscillation is to sample the output after it has passed through the unstable region. Another practical solution is to add hysteresis. Either solution results in a loss of resolution and bandwidth.
Because the MAX912/MAX913 do not need hysteresis, they offer high resolution to all signals-including lowfrequency signals.
The MAX912/MAX913 provide a TTL-compatible latch function that holds the comparator output state (Figure 1). As long as Latch Enable (LE) is high or floating, the input signal has no effect on the output state. With LE low, the outputs are controlled by the input differential voltage and the latch is transparent.

## Input Amplifier

A comparator can be thought of as having two sections: an input amplifier and a logic interface. The MAX912/MAX913's input amplifier is fully differential, with input offset voltage trimmed to below 2.0 mV at $+25^{\circ} \mathrm{C}$. Input common-mode range extends from 200 mV below the negative supply rail to 1.5 V below the positive power supply. The total common-mode range is 8.7 V when operating from $\pm 5 \mathrm{VDC}$ supplies.
The MAX912/MAX913's amplifier has no built-in hysteresis. For highest accuracy, do not add hysteresis. Figure 2 shows how hysteresis degrades resolution.

A comparator's ability to Resolution
A comparator's ability to resolve small signal differ-ences-its resolution-is affected by various factors. As with most amplifiers, the most significant factors are the input offset voltage (VOS) and the common-mode and power-supply rejection ratios (CMRR, PSRR). If source impedance is high, input offset current can be significant. If source impedance is unbalanced, the input bias current can introduce another error.
For high-speed comparators, an additional factor in resolution is the comparator's stability in its linear region. Many high-speed comparators are useless in their linear region because they oscillate. This makes the differential input voltage region around OV unusable, as does a high Vos. Hysteresis does not cure the problem, but acts to keep the input away from its linear range (Figure 2).
The MAX912/MAX913 do not oscillate in the linear region, which greatly enhances the comparator's resolution.

## Applications Information

Power Supplies and Bypassing
The MAX912/MAX913 are tested with $\pm 5 \mathrm{~V}$ power supplies that provide an input common-mode range (VCM) of $8.7 \mathrm{~V}(-5.2 \mathrm{~V}$ to $+3.5 \mathrm{~V})$. Operation from a single +5 V supply provides a common-mode input range of 3.7 V $(-0.2 \mathrm{~V}$ to $+3.5 \mathrm{~V})$. Connect V - to GND for single-supply operation. The MAX912/MAX913 will operate from a minimum single-supply voltage of +4.5 V .
The $\mathrm{V}_{+}$supply provides power to both the analog input stage and digital output circuits, whereas the V - supply only powers the analog section. Bypass $V+$ and $V$ - to ground with $0.1 \mu \mathrm{~F}$ to $1.0 \mu \mathrm{~F}$ ceramic capacitors in parallel with $10 \mu \mathrm{~F}$ or greater tantalum capacitors. Connect the ceramic capacitors very close to the MAX912/MAX913's


Figure 1. Timing Diagram

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supply pins, keeping leads short to minimize lead inductance. For particularly noisy applications, use ferrite beads on the power-supply lines.

## Board Layout

As with all high-speed components, careful attention to layout is essential for best performance.

1) Use a printed circuit board with an unbroken ground plane.
2) Pay close attention to the bandwidth of bypass components and keep leads short.
3) Avoid sockets; solder the comparator and other components directly to the board to minimize unwanted parasitic inductance and capacitance.

Input Slew Rate
The MAX912/MAX913 design eliminates the input slewrate requirement imposed on many standard comparators. As long as LE is high after the maximum propagation delay and the input is greater than the comparator's total DC error, the output will be valid without oscillations.

Maximum Clock (LE) and Signal Rate The maximum clock and signal rate is 70 MHz , based on the comparator's rise and fall time with a 5 mV overdrive at $+25^{\circ} \mathrm{C}$ (Figure 1). With a 20 mV overdrive, the maximum propagation delay is 12 ns and the clock and signal rate is 85 MHz .


Figure 2. Effect of Hysteresis on Input Resolution
Chip Topographies


TRANSISTOR COUNT: 197; SUBSTRATE CONNECTED TO V-.


TRANSISTOR COUNT: 100; SUBSTRATE CONNECTED TO V-.

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