

The RF Sub-Micron MOSFET Line

RF Power Field Effect Transistors

N-Channel Enhancement-Mode Lateral MOSFETs

Designed for broadband commercial and industrial applications with frequencies up to 1.0 GHz. The high gain and broadband performance of these devices make them ideal for large-signal, common-source amplifier applications in 28 volt base station equipment.

- Typical Performance at 945 MHz, 28 Volts
 - Output Power — 45 Watts PEP
 - Power Gain — 19 dB
 - Efficiency — 41% (Two Tones)
 - IMD — -31 dBc
- Integrated ESD Protection
- Guaranteed Ruggedness @ Load VSWR = 5:1, @ 28 Vdc, 945 MHz, 45 Watts CW Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters
- Dual-Lead Boltdown Plastic Package Can Also Be Used As Surface Mount.
- TO-272 Available in Tape and Reel. R1 Suffix = 500 Units per 44 mm, 13 inch Reel.
- TO-270 Available in Tape and Reel. R1 Suffix = 500 Units per 24 mm, 13 inch Reel.

MRF9045MR1 **MRF9045MBR1**

945 MHz, 45 W, 28 V
LATERAL N-CHANNEL
BROADBAND
RF POWER MOSFETs

CASE 1265-08, STYLE 1
TO-270
PLASTIC
MRF9045MR1



CASE 1337-03, STYLE 1
TO-272 DUAL LEAD
PLASTIC
MRF9045MBR1

MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	V _{DSS}	65	Vdc
Gate-Source Voltage	V _{GS}	+15, -0.5	Vdc
Total Device Dissipation @ T _C = 25°C Derate above 25°C	P _D	177 1.18	Watts W/°C
Storage Temperature Range	T _{stg}	-65 to +150	°C
Operating Junction Temperature	T _J	175	°C

THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	R _{θJC}	0.85	°C/W

ESD PROTECTION CHARACTERISTICS

Test Conditions	Class
Human Body Model	1 (Minimum)
Machine Model	M2 (Minimum)

MOISTURE SENSITIVITY LEVEL

Test Methodology	Rating
Per JESD 22-A113	3

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

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ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS					
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 65 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	10	μAdc
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 28 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	1	μAdc
Gate-Source Leakage Current ($V_{GS} = 5 \text{ Vdc}$, $V_{DS} = 0 \text{ Vdc}$)	I_{GSS}	—	—	1	μAdc

ON CHARACTERISTICS

Gate Threshold Voltage ($V_{DS} = 10 \text{ Vdc}$, $I_D = 150 \mu\text{Adc}$)	$V_{GS(\text{th})}$	2	2.8	4	Vdc
Gate Quiescent Voltage ($V_{DS} = 28 \text{ Vdc}$, $I_D = 350 \text{ mA}$)	$V_{GS(Q)}$	3	3.7	5	Vdc
Drain-Source On-Voltage ($V_{GS} = 10 \text{ Vdc}$, $I_D = 1 \text{ Adc}$)	$V_{DS(\text{on})}$	—	0.22	0.4	Vdc
Forward Transconductance ($V_{DS} = 10 \text{ Vdc}$, $I_D = 3 \text{ Adc}$)	g_{fs}	—	4	—	S

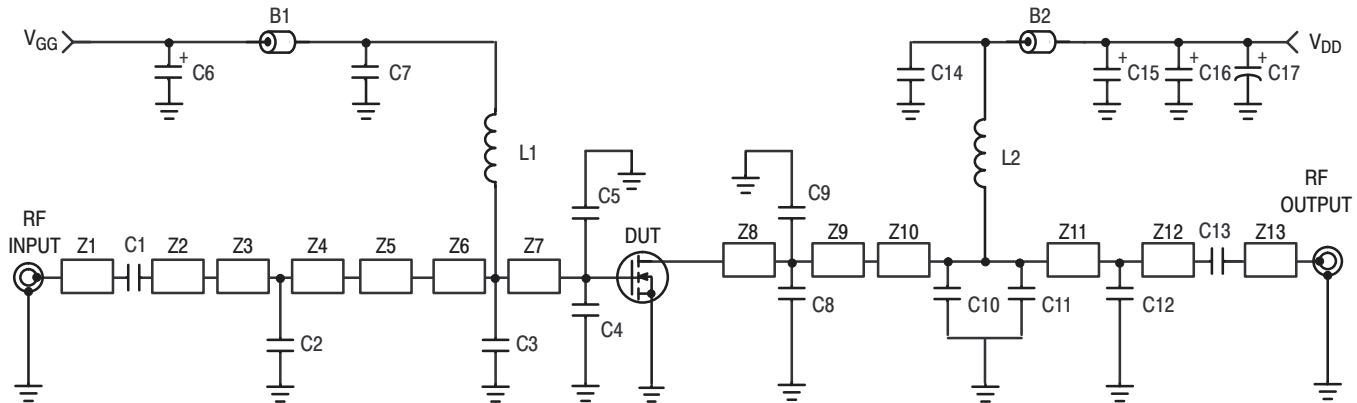
DYNAMIC CHARACTERISTICS

Input Capacitance ($V_{DS} = 28 \text{ Vdc} \pm 30 \text{ mV(rms)}$ ac @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$)	C_{iss}	—	70	—	pF
Output Capacitance ($V_{DS} = 28 \text{ Vdc} \pm 30 \text{ mV(rms)}$ ac @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$)	C_{oss}	—	38	—	pF
Reverse Transfer Capacitance ($V_{DS} = 28 \text{ Vdc} \pm 30 \text{ mV(rms)}$ ac @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$)	C_{rss}	—	1.7	—	pF

FUNCTIONAL TESTS (In Motorola Test Fixture, 50 ohm system)

Two-Tone Common-Source Amplifier Power Gain ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	G_{ps}	17	19	—	dB
Two-Tone Drain Efficiency ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	η	38	41	—	%
3rd Order Intermodulation Distortion ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	IMD	—	-31	-28	dBc
Input Return Loss ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	IRL	—	-14	-9	dB
Two-Tone Common-Source Amplifier Power Gain ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	G_{ps}	—	19	—	dB
Two-Tone Drain Efficiency ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	η	—	41	—	%
3rd Order Intermodulation Distortion ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	IMD	—	-31	—	dBc
Input Return Loss ($V_{DD} = 28 \text{ Vdc}$, $P_{out} = 45 \text{ W PEP}$, $I_{DQ} = 350 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	IRL	—	-13	—	dB

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B1, B2	Short Ferrite Beads, Surface Mount
C1, C7, C13, C14	47 pF Chip Capacitors, B Case
C2, C8	2.7 pF Chip Capacitors, B Case
C3	3.9 pF Chip Capacitor, B Case
C4, C5, C8, C9	10 pF Chip Capacitors, B Case
C6, C15, C16	10 μ F, 35 V Tantalum Surface Mount Capacitors
C10	2.2 pF Chip Capacitor, B Case
C11	4.7 pF Chip Capacitor, B Case
C12	1.2 pF Chip Capacitor, B Case
C17	220 μ F, 50 V Electrolytic Capacitor
L1, L2	12.5 nH Inductors
Z1	0.20" x 0.08" Microstrip
Z2	0.57" x 0.12" Microstrip

Z3	0.14" x 0.32" Microstrip
Z4	0.47" x 0.32" Microstrip
Z5	0.16" x 0.32" x 0.62" Taper
Z6	0.18" x 0.62" Microstrip
Z7	0.56" x 0.62" Microstrip
Z8	0.33" x 0.32" Microstrip
Z9	0.14" x 0.32" Microstrip
Z10	0.36" x 0.08" Microstrip
Z11	1.01" x 0.08" Microstrip
Z12	0.15" x 0.08" Microstrip
Z13	0.29" x 0.08" Microstrip

Figure 1. MRF9045MR1 930–960 MHz Broadband Test Circuit Schematic

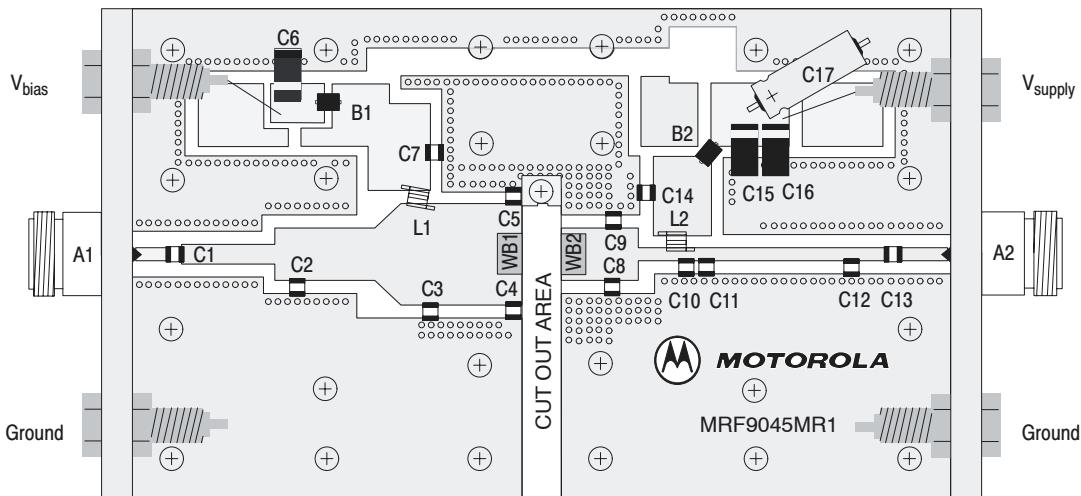
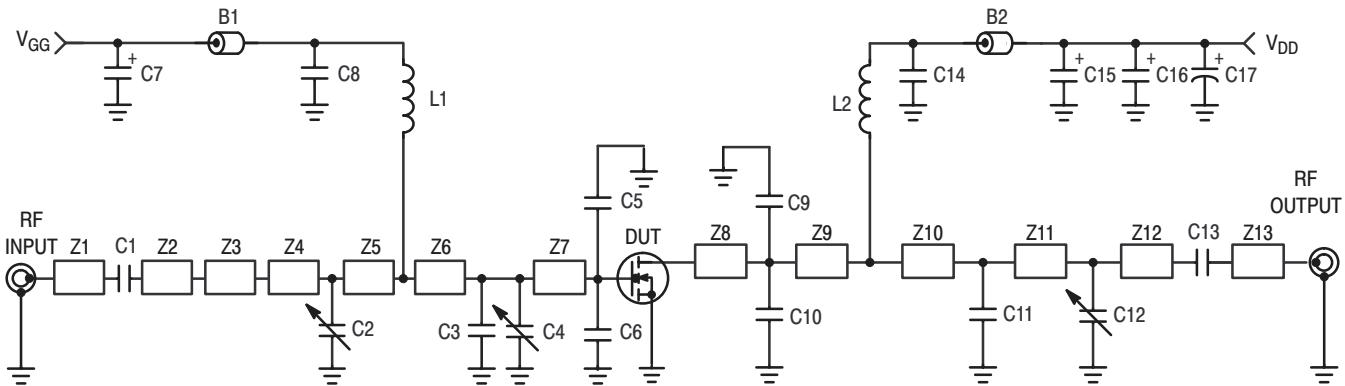


Figure 2. MRF9045MR1 930–960 MHz Broadband Test Circuit Component Layout

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B1	Short Ferrite Bead	Z1	0.260" x 0.060" Microstrip
B2	Long Ferrite Bead	Z2	0.240" x 0.060" Microstrip
C1, C8, C13, C14	47 pF Chip Capacitors, B Case	Z3	0.500" x 0.100" Microstrip
C2	0.4–2.5 pF Variable Capacitor, Johanson Gigatrim	Z4	0.215" x 0.270" Microstrip
C3	3.6 pF Chip Capacitor, B Case	Z5	0.315" x 0.270" Microstrip
C4	0.8–8.0 pF Variable Capacitor, Johanson Gigatrim	Z6	0.160" x 0.270" x 0.520" Taper
C5, C6, C9, C10	10 pF Chip Capacitors, B Case	Z7	0.285" x 0.520" Microstrip
C7, C15, C16	10 μ F, 35 V Tantalum Chip Capacitors	Z8	0.140" x 0.270" Microstrip
C11	7.5 pF Chip Capacitor, B Case	Z9	0.450" x 0.270" Microstrip
C12	0.6–4.5 pF Variable Capacitor, Johanson Gigatrim	Z10	0.250" x 0.060" Microstrip
C17	220 μ F Electrolytic Chip Capacitor	Z11	0.720" x 0.060" Microstrip
L1, L2	12.5 nH Surface Mount Inductors	Z12	0.490" x 0.060" Microstrip
WB1, WB2	10 mil Brass Wear Blocks	Z13	0.290" x 0.060" Microstrip
		Board	Taconic RF-35-0300, $\epsilon_r = 3.5$

Figure 3. MRF9045MBR1 930–960 MHz Broadband Test Circuit Schematic

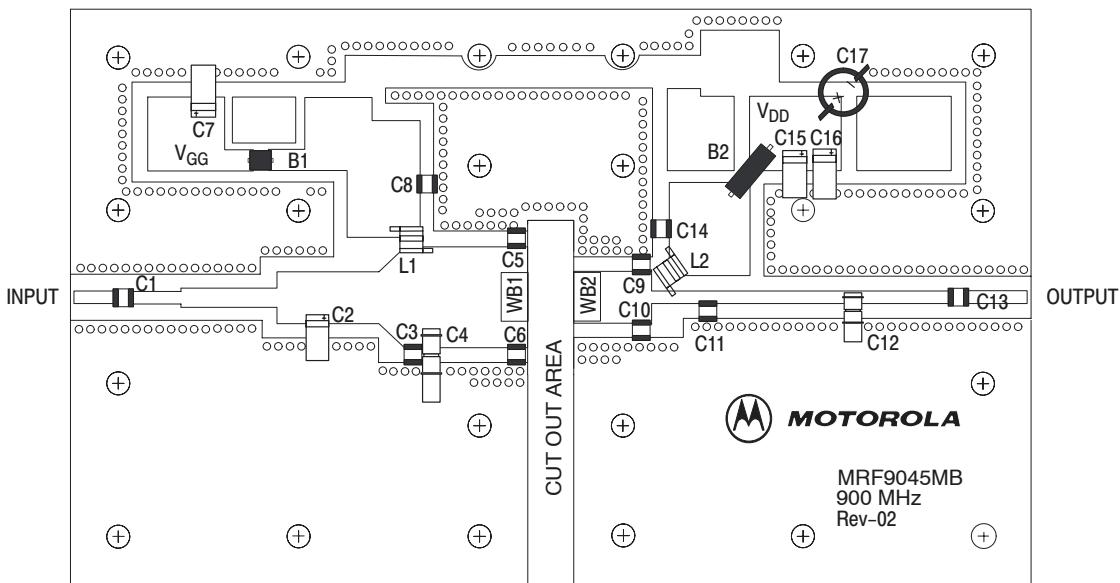


Figure 4. MRF9045MBR1 930–960 MHz Broadband Test Circuit Component Layout

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TYPICAL CHARACTERISTICS

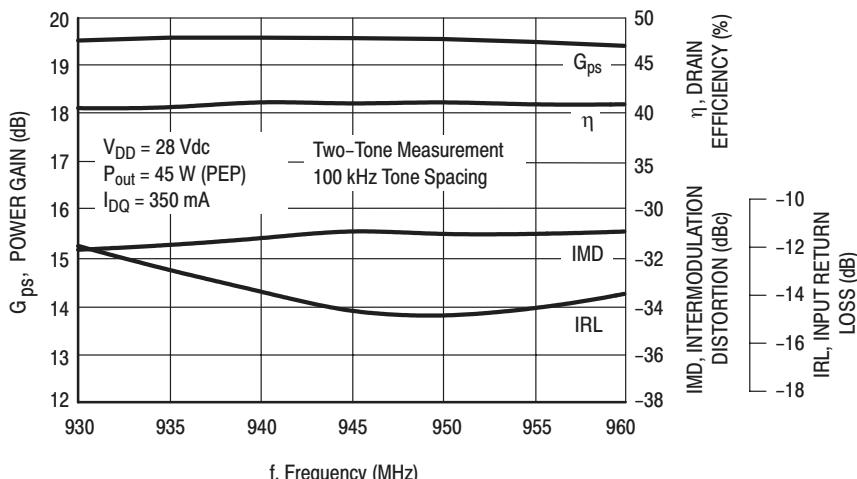


Figure 5. Class AB Broadband Circuit Performance

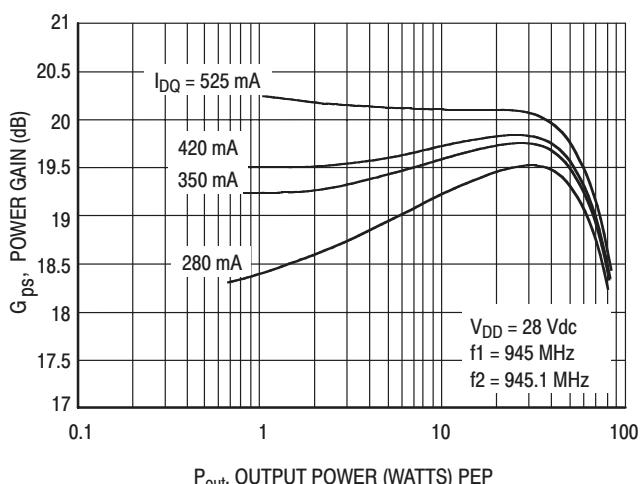


Figure 6. Power Gain versus Output Power

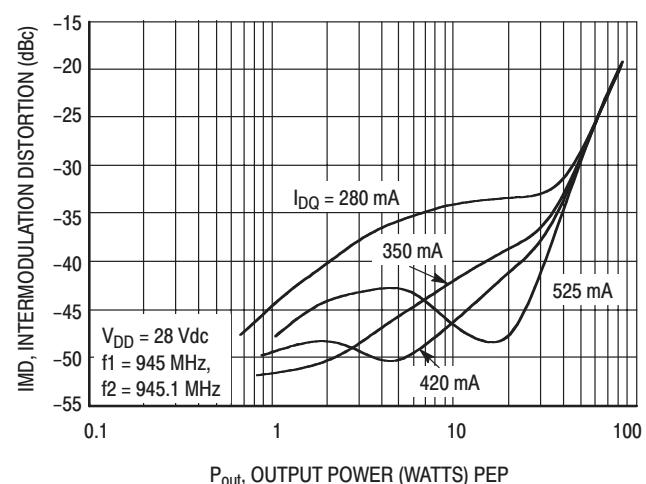


Figure 7. Intermodulation Distortion versus Output Power

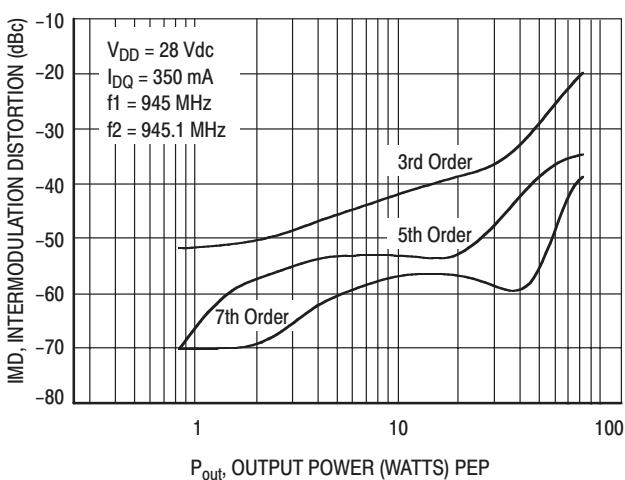


Figure 8. Intermodulation Distortion Products versus Output Power

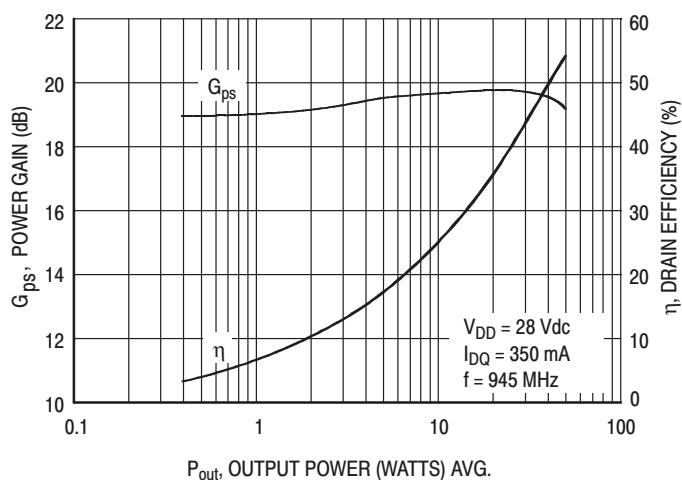


Figure 9. Power Gain and Efficiency versus Output Power

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TYPICAL CHARACTERISTICS

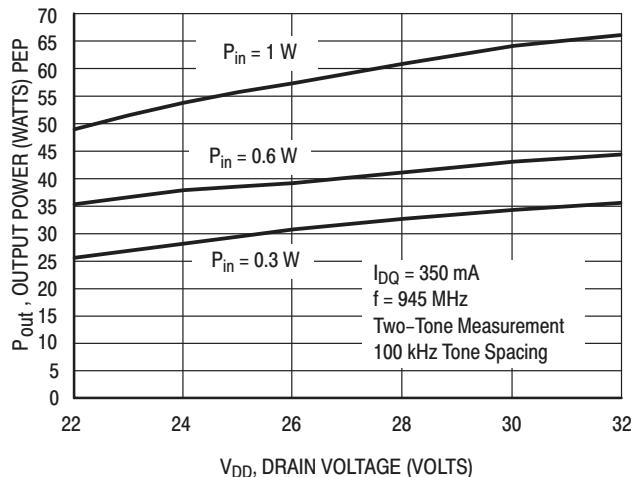


Figure 10. Output Voltage versus Supply Voltage (MRF9045MR1)

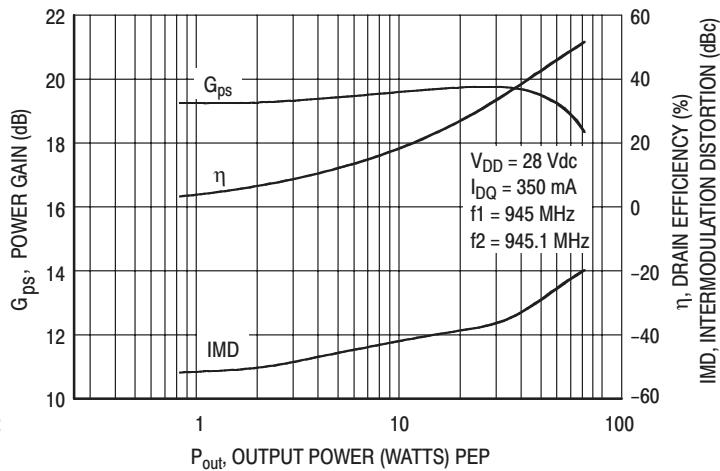
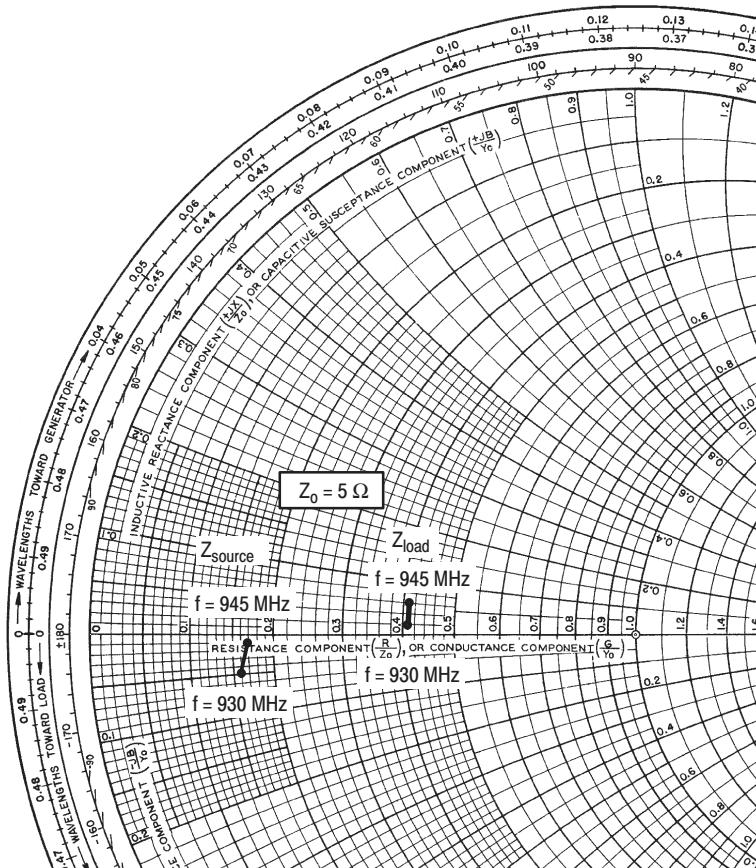


Figure 11. Power Gain, Efficiency and IMD versus Output Power (MRF9045MBR1)

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$V_{DD} = 28 \text{ V}, I_{DQ} = 350 \text{ mA}, P_{out} = 45 \text{ W (PEP)}$

f MHz	Z_{source} Ω	Z_{load} Ω
930	$0.81 - j0.25$	$2.03 + j0.09$
945	$0.85 - j0.05$	$2.03 + j0.28$

Z_{source} = Test circuit impedance as measured from gate to ground.

Z_{load} = Test circuit impedance as measured from drain to ground.

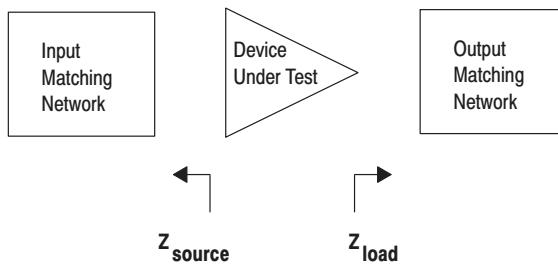
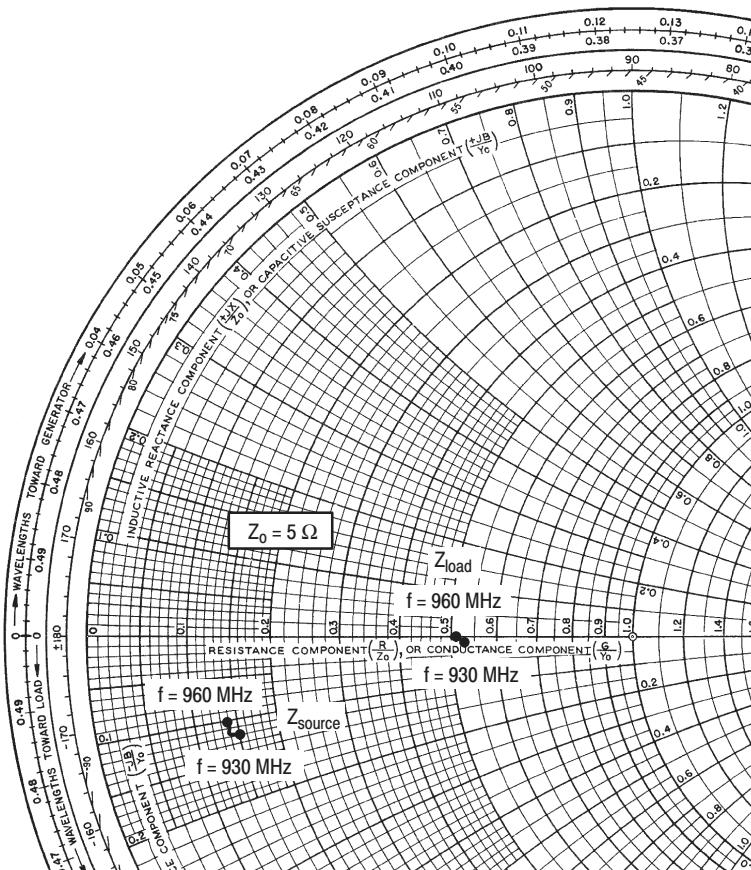


Figure 12. Series Equivalent Input and Output Impedance (MRF9045MR1)

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$V_{DD} = 28 \text{ V}, I_{DQ} = 350 \text{ mA}, P_{out} = 45 \text{ W (PEP)}$

f MHz	Z_{source} Ω	Z_{load} Ω
930	$0.75 - j0.6$	$2.65 - j0.05$
945	$0.72 - j0.6$	$2.60 - j0.05$
960	$0.70 - j0.5$	$2.55 - j0.02$

Z_{source} = Test circuit impedance as measured from gate to ground.

Z_{load} = Test circuit impedance as measured from drain to ground.

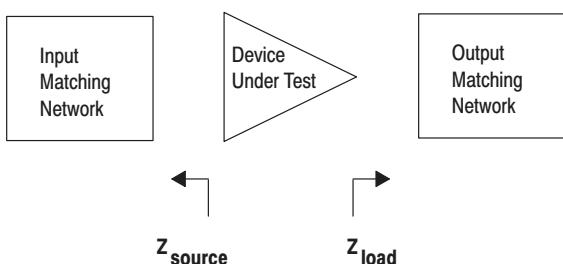
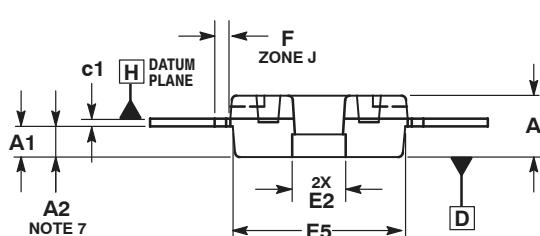
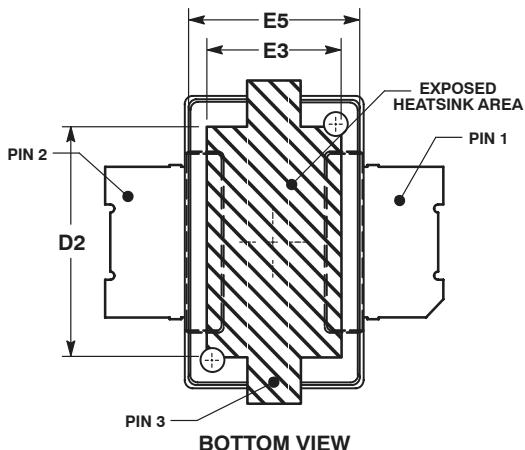
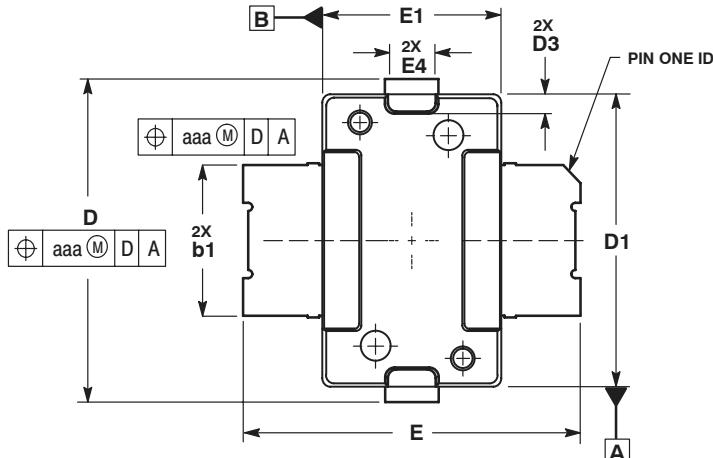


Figure 13. Series Equivalent Input and Output Impedance (MRF9045MBR1)

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NOTES

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PACKAGE DIMENSIONS



NOTES:

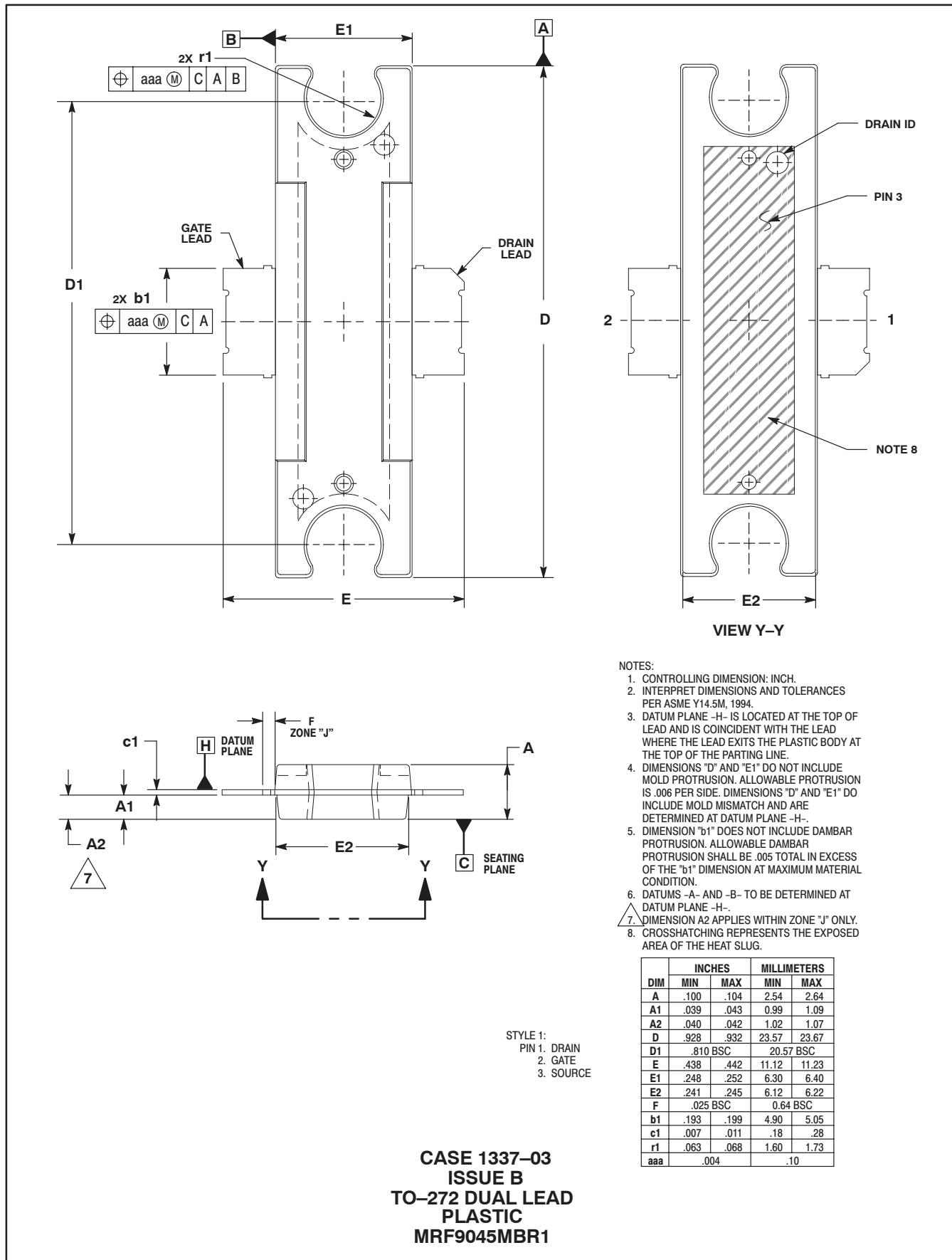
1. CONTROLLING DIMENSION: INCH.
2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M-1994.
3. DATUM PLANE -H- IS LOCATED AT TOP OF LEAD AND IS COINCIDENT WITH THE LEAD WHERE THE LEAD EXITS THE PLASTIC BODY AT THE TOP OF THE PARTING LINE.
4. DIMENSIONS "D1" AND "E1" DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS .006 PER SIDE. DIMENSIONS "D1" AND "E1" DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE -H-.
5. DIMENSION b1 DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE .005 TOTAL IN EXCESS OF THE b1 DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. DATUMS -A- AND -B- TO BE DETERMINED AT DATUM PLANE -H-.
7. DIMENSION A2 APPLIES WITHIN ZONE "J" ONLY.
8. DIMENSIONS "D" AND "E2" DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS .003 PER SIDE. DIMENSIONS "D" AND "E2" DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE -D-.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.078	.082	1.98	2.08
A1	.039	.043	0.99	1.09
A2	.040	.042	1.02	1.07
D	.416	.424	10.57	10.77
D1	.378	.382	9.60	9.70
D2	.290	.320	7.37	8.13
D3	.016	.024	0.41	0.61
E	.436	.444	11.07	11.28
E1	.238	.242	6.04	6.15
E2	.066	.074	1.68	1.88
E3	.150	.180	3.81	4.57
E4	.058	.066	1.47	1.68
E5	.231	.235	5.87	5.97
F	.025 BSC		0.64 BSC	
b1	.193	.199	4.90	5.06
c1	.007	.011	0.18	0.28
aaa	.004		0.10	

STYLE 1:
 1. PIN 1. DRAIN
 2. GATE
 3. SOURCE

CASE 1265-08
 ISSUE G
 TO-270
 PLASTIC
 MRF9045MR1

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