

SEMICONDUCTOR®

FGH20N6S2 / FGP20N6S2 / FGB20N6S2

600V, SMPS II Series N-Channel IGBT

General Description

The FGH20N6S2, FGP20N6S2, FGB20N6S2, are Low Gate Charge, Low Plateau Voltage SMPS II IGBTs combining the fast switching speed of the SMPS IGBTs along with lower gate charge and plateau voltage and high avalanche capability (UIS). These LGC devices shorten delay times, and reduce the power requirement of the gate drive. These devices are ideally suited for high voltage switched mode power supply applications where low conduction loss, fast switching times and UIS capability are essential. SMPS II LGC devices have been specially designed for:

- Power Factor Correction (PFC) circuits
- Full bridge topologies
- Half bridge topologies
- Push-Pull circuits
- Uninterruptible power supplies
- Zero voltage and zero current switching circuits

Collector Current Continuous, T_C = 110°C

Switching Safe Operating Area at T_J = 150°C, Figure 2

Pulsed Avalanche Energy, I_{CE} = 7.0A, L = 4mH, V_{DD} = 50V

Pulsed Avalanche Energy, I_{CE} = 7.0A, L = 4mH, V_{DD} = 50V

Collector Current Pulsed (Note 1)

Gate to Emitter Voltage Pulsed

Gate to Emitter Voltage Continuous

Power Dissipation Total T_C = 25°C

Power Dissipation Derating T_C > 25°C

Operating Junction Temperature Range

Storage Junction Temperature Range

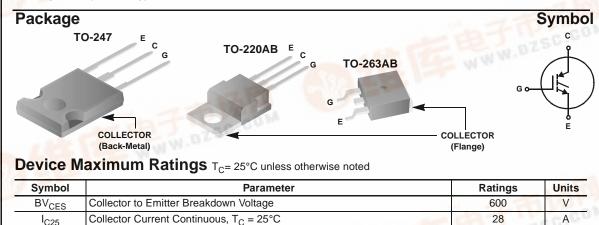
Formerly Developmental Type TA49330.

Features

- 100kHz Operation at 390V, 7A
- 200kHZ Operation at 390V, 5A
- 600V Switching SOA Capability

- Low Plateau Voltage6.5V Typical
- Low Conduction Loss
- Low E_{on}

August 2003



CAUTION: Stresses above those listed in "Device Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

PD	Pulse width limited by maximum junction temperature.	

I_{C110}

I_{CM}

V_{GES} V_{GEM}

SSOA

E_{AS}

EARV

PD

T_J

T_{STG}

NOTE

13

40

±20

±30

35 at 600V

100

100

125

1.0

-55 to 150

-55 to 150

A

A

V

V

А

mJ

mJ

W

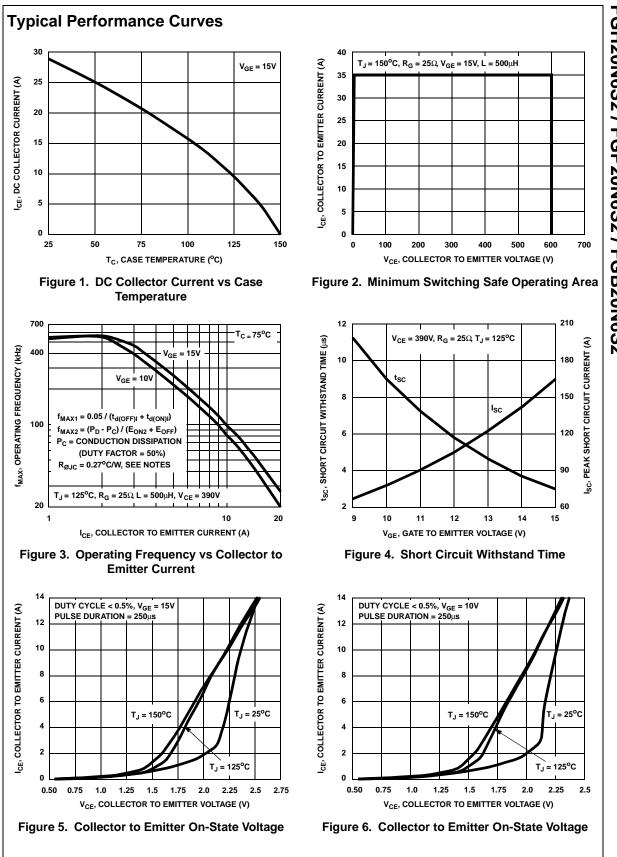
W/°C

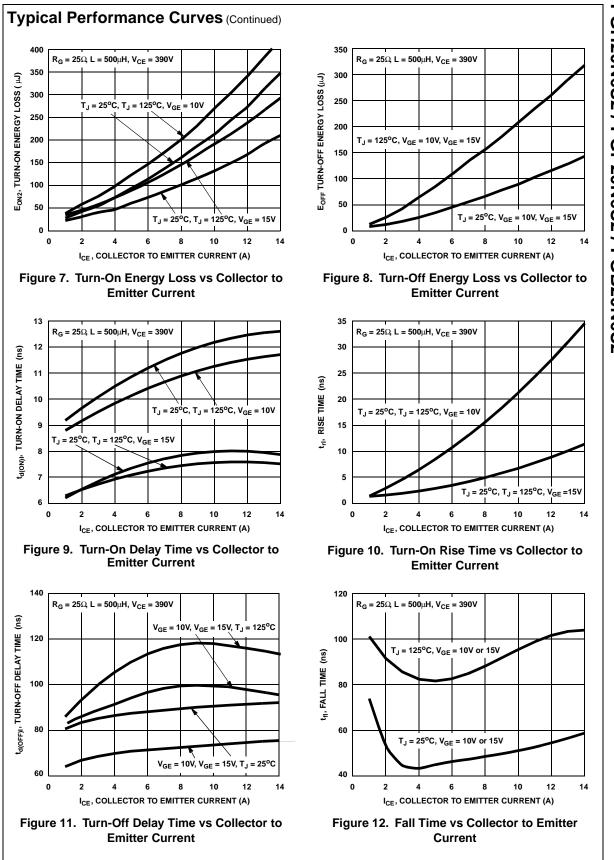
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Device N	larking	Device	Package	Reel Size	Таре	Width	Qua	ntity
20N6	S2	FGH20N6S2	TO-247	Tube	N	I/A	30 L	Jnits
20N6	S2	FGP20N6S2	TO-220AB	Tube	N/A		50 Units	
20N6S2 FGB20N6S2		TO-263AB	Tube	N	I/A	50 Units		
20N6	S2	FGB20N6S2T	TO-263AB	330mm	24mm		800 Units	
	al Chai	racteristics T _J = 25°C			Min	T	Maria	11
Symbol		Parameter	lest	Conditions	Min	Тур	Max	Units
Off State						•		
BV _{CES}		to Emitter Breakdown Volta		-	600	-	-	V
BV _{ECS}		o Collector Breakdown Volta			20	-	-	V
I _{CES}	Collector	to Emitter Leakage Current	$V_{CE} = 600V$	$T_J = 25^{\circ}C$	-	-	250	μA
				T _J = 125°C	-	-	2.0	mA
I _{GES}	Gate to E	Emitter Leakage Current	$V_{GE} = \pm 20V$		-	-	±250	nA
On State	Charact	eristics						
V _{CE(SAT)}	Collector	to Emitter Saturation Voltag	je I _C = 7.0A,	T _J = 25°C	-	2.2	2.7	V
OL(OAI)		·····3	$V_{GE} = 15V$	$T_{1} = 125^{\circ}C$	-	1.9	2.2	V
Dynamic (Charact	eristics				•	•	
Q _{G(ON)}	Gate Cha	arge	I _C = 7.0A,	V _{GE} = 15V	-	30	36	nC
-()			$V_{CE} = 300V$	$V_{GE} = 20V$	-	38	45	nC
V _{GE(TH)}	Gate to E	Emitter Threshold Voltage	I _C = 250μA, V	_{CE} = 600V	3.5	4.3	5.0	V
V _{GEP}	Gate to E	Emitter Plateau Voltage	I _C = 7.0A, V _{CI}		-	6.5	8.0	V
Switching	Charac	teristics				•	•	
SSOA	Switching	g SOA		$T_J = 150^{\circ}C, R_G = 25\Omega, V_{GE} = 15V, L = 0.5mH, Vce = 600V$		-	-	A
t _{d(ON)}	Current 7	Furn-On Delay Time	IGBT and Dic	IGBT and Diode at $T_J = 25^{\circ}C$,		7.7	-	ns
t _{rl}	Current F	Rise Time	I _{CE} = 7A,		-	4.5	-	ns
t _{d(OFF)} I	Current 7	Turn-Off Delay Time	$V_{CE} = 390V,$		-	87	-	ns
t _{fl}	Current F	Fall Time	— V _{GE} = 15V, — R _G = 25Ω		-	50	-	ns
E _{ON1}	Turn-On	Energy (Note 1)	L = 0.5 mH	÷		25	-	μJ
E _{ON2}	Turn-On	Energy (Note 1)	Test Circuit -	Figure 20	-	85	-	μJ
E _{OFF}	Turn-Off	Energy (Note 2)		1		58	75	μJ
t _{d(ON)I}	Current 7	Furn-On Delay Time	IGBT and Dic	ode at T _J = 125°C,	-	7	-	ns
t _{rl}		Rise Time	$I_{CE} = 7A,$	I _{CE} = 7A, V _{CE} = 390V,		4.5	-	ns
t _{d(OFF)} I	Current 7	Turn-Off Delay Time				120	145	ns
t _{fl}	Current F	Fall Time	— V _{GE} = 15V, R _G = 25Ω		-	85	105	ns
E _{ON1}	Turn-On	Energy (Note 1)	L = 0.5 mH			20	-	μJ
E _{ON2}	Turn-On	Energy (Note 1)				125	140	μJ
E _{OFF}	Turn-Off	Energy (Note 2)				135	180	μJ
Thermal C	haracte	eristics						
$R_{ extsf{ heta}JC}$		Resistance Junction-Case			-	-	1.0	°C/V

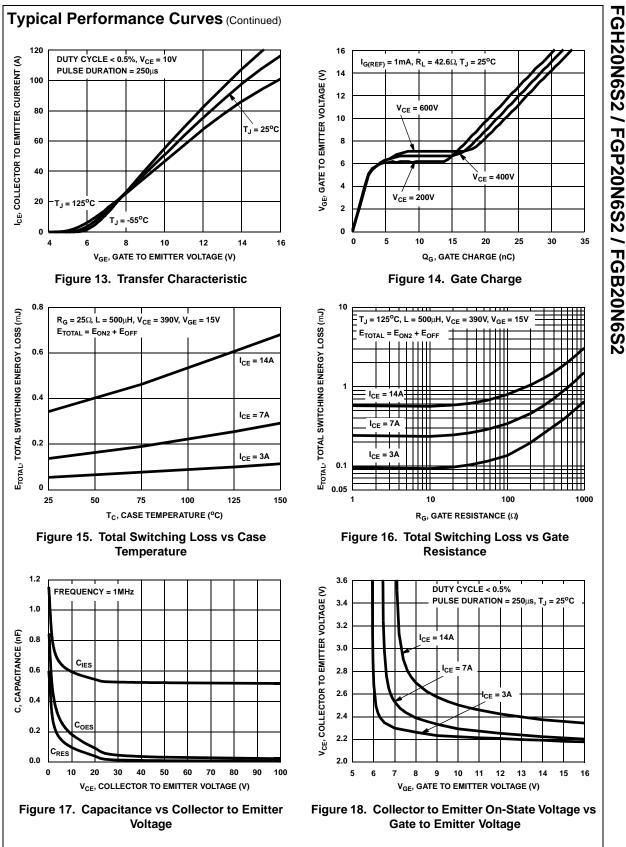
1. Values for two Turn-On loss conditions are shown for the convenience of the circuit designer. E_{ON1} is the turn-on loss of the IGBT only. E_{ON2} is the turn-on loss when a typical diode is used in the test circuit and the diode is at the same T_J as the IGBT. The diode type is specified in figure 20.

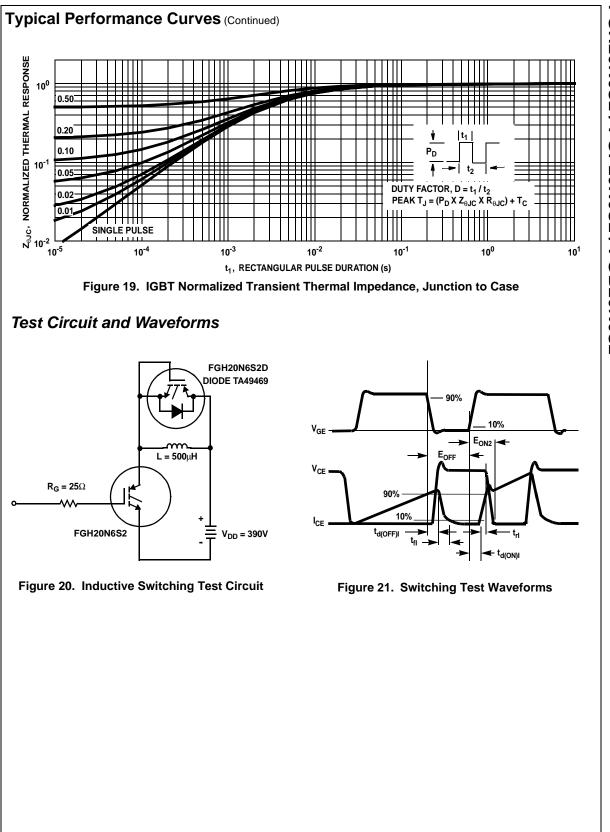
2. Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ($I_{CE} = 0A$). All devices were tested per JEDEC Standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.





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Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- 5. Gate Voltage Rating Never exceed the gatevoltage rating of V_{GEM} . Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region.
- 6. Gate Termination The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
- 7. **Gate Protection** These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

Operating Frequency Information

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information shown for a typical unit in Figures 5, 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows f_{MAX1} or f_{MAX2} ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 f_{MAX1} is defined by $f_{MAX1} = 0.05/(t_{d(OFF)I} + t_{d(ON)I})$. Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible. $t_{d(OFF)I}$ and $t_{d(ON)I}$ are defined in Figure 21. Device turn-off delay can establish an additional frequency limiting condition for an application other than T_{JM} . $t_{d(OFF)I}$ is important when controlling output ripple under a lightly loaded condition.

 $f_{MAX2} \text{ is defined by } f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON2}).$ The allowable dissipation (P_D) is defined by P_D = (T_{JM} - T_C)/R_{\theta JC}. The sum of device switching and conduction losses must not exceed P_D. A 50% duty factor was used (Figure 3) and the conduction losses (P_C) are approximated by P_C = (V_{CE} \times I_{CE})/2.

 E_{ON2} and E_{OFF} are defined in the switching waveforms shown in Figure 21. E_{ON2} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-on and E_{OFF} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-off. All tail losses are included in the calculation for E_{OFF} ; i.e., the collector current equals zero ($I_{CE} = 0$)

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The Power Fran		ise™ PACMAN™ Stealth™		
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