

**FAIRCHILD**

SEMICONDUCTOR®

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## FGH50N3

### 300V, PT N-Channel IGBT

#### General Description

The FGH50N3 is a MOS gated high voltage switching device combining the best features of MOSFETs and bipolar transistors. These devices have the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C.

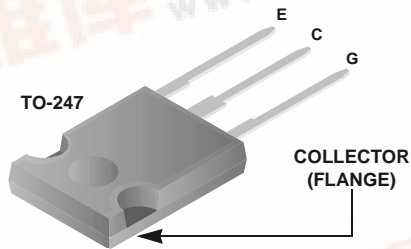
This IGBT is ideal for many high voltage switching applications operating at high frequencies where low conduction losses are essential. This device has been optimized for medium frequency switch mode power supplies.

Formerly Developmental Type TA49485

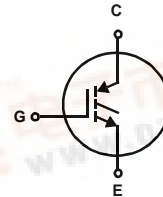
#### Features

- Low  $V_{CE(SAT)}$  ..... < 1.4V max
- Low  $E_{OFF}$  ..... < 200μJ
- SCWT (@  $T_J = 125^\circ\text{C}$ ) ..... > 8μs
- 300V Switching SOA Capability
- Positive  $V_{CE(SAT)}$  Temperature Coefficient above 50A

#### Package



#### Symbol



#### Device Maximum Ratings $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Ratings	Units
$BV_{CES}$	Collector to Emitter Breakdown Voltage	300	V
$I_{C25}$	Collector Current Continuous, $T_C = 25^\circ\text{C}$	75	A
$I_{C110}$	Collector Current Continuous, $T_C = 110^\circ\text{C}$	75	A
$I_{CM}$	Collector Current Pulsed (Note 1)	240	A
$V_{GES}$	Gate to Emitter Voltage Continuous	±20	V
$V_{GEM}$	Gate to Emitter Voltage Pulsed	±30	V
SSOA	Switching Safe Operating Area at $T_J = 150^\circ\text{C}$ , Figure 2	150A at 300V	
$E_{AS}$	Single Pulse Avalanche Energy, $I_{CE} = 30\text{A}$ , $L = 1.78\text{mH}$ , $V_{DD} = 50\text{V}$	800	mJ
$E_{ARV}$	Single Pulse Reverse Avalanche Energy, $I_{EC} = 30\text{A}$ , $L = 1.78\text{mH}$ , $V_{DD} = 50\text{V}$	800	mJ
$P_D$	Power Dissipation Total $T_C = 25^\circ\text{C}$	463	W
	Power Dissipation Derating $T_C > 25^\circ\text{C}$	3.7	W/°C
$T_J$	Operating Junction Temperature Range	-55 to 150	°C
$T_{STG}$	Storage Junction Temperature Range	-55 to 150	°C
$t_{SC}$	Short Circuit Withstand Time (Note 2)	8	μs

CAUTION: Stresses above those listed in "Device Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTE:

1. Pulse width limited by maximum junction temperature.
2.  $V_{CE(PK)} = 180\text{V}$ ,  $T_J = 125^\circ\text{C}$ ,  $V_{GE} = 12\text{Vdc}$ ,  $R_G = 5\Omega$



## Package Marking and Ordering Information

Device Marking	Device	Package	Tape Width	Quantity
FGH50N3	FGH50N3	TO-247	N/A	30

## Electrical Characteristics $T_J = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Typ	Max	Units
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### Off State Characteristics

$BV_{CES}$	Collector to Emitter Breakdown Voltage	$I_{CE} = 250\mu\text{A}, V_{GE} = 0\text{V}$	300V	-	-	V	
$BV_{ECS}$	Emitter to Collector Breakdown Voltage	$I_{EC} = 10\text{mA}, V_{GE} = 0\text{V}$	15V	-	-	V	
$I_{CES}$	Collector to Emitter Leakage Current	$V_{CE} = 300\text{V}$	$T_J = 25^\circ\text{C}$	-	-	250	$\mu\text{A}$
			$T_J = 125^\circ\text{C}$	-	-	2.0	mA
$I_{GES}$	Gate to Emitter Leakage Current	$V_{GE} = \pm 20\text{V}$	-	-	$\pm 250$	nA	

### On State Characteristics

$V_{CE(SAT)}$	Collector to Emitter Saturation Voltage	$I_{CE} = 30\text{A}$ $V_{GE} = 15\text{V}$	$T_J = 25^\circ\text{C}$	-	1.30	1.4	V
			$T_J = 125^\circ\text{C}$	-	1.25	1.4	V

### Dynamic Characteristics

$Q_{G(ON)}$	Gate Charge	$I_{CE} = 30\text{A}$ $V_{CE} = 150\text{V}$	$V_{GE} = 15\text{V}$	-	180	-	nC
			$V_{GE} = 20\text{V}$	-	228	-	nC
$V_{GE(TH)}$	Gate to Emitter Threshold Voltage	$I_{CE} = 250\mu\text{A}, V_{CE} = V_{GE}$	4.0	4.8	5.5	V	
$V_{GEP}$	Gate to Emitter Plateau Voltage	$I_{CE} = 30\text{A}, V_{CE} = 150\text{V}$	-	7.0	-	V	

### Switching Characteristics

SSOA	Switching SOA	$T_J = 150^\circ\text{C}, R_G = 5\Omega$ $V_{GE} = 15\text{V}, L = 25\mu\text{H}$ $V_{ce} = 300\text{V}$	150	-	-	A	
$t_{d(ON)I}$	Current Turn-On Delay Time	IGBT and Diode at $T_J = 25^\circ\text{C}$ , $I_{CE} = 30\text{A}$ , $V_{CE} = 180\text{V}$ , $V_{GE} = 15\text{V}$ , $R_G = 5\Omega$ , $L = 100\mu\text{H}$ , Test Circuit - Figure 20	-	20	-	ns	
$t_{rI}$	Current Rise Time		-	15	-	ns	
$t_{d(OFF)I}$	Current Turn-Off Delay Time		-	135	-	ns	
$t_{fI}$	Current Fall Time		-	12	-	ns	
$E_{ON2}$	Turn-On Energy (Note 1)		-	130	-	$\mu\text{J}$	
$E_{OFF}$	Turn-Off Energy (Note 2)		-	92	120	$\mu\text{J}$	
$t_{d(ON)I}$	Current Turn-On Delay Time		IGBT and Diode at $T_J = 125^\circ\text{C}$ , $I_{CE} = 30\text{A}$ , $V_{CE} = 180\text{V}$ , $V_{GE} = 15\text{V}$ , $R_G = 5\Omega$ , $L = 100\mu\text{H}$ , Test Circuit - Figure 20	-	19	-	ns
$t_{rI}$	Current Rise Time			-	13	-	ns
$t_{d(OFF)I}$	Current Turn-Off Delay Time			-	155	190	ns
$t_{fI}$	Current Fall Time			-	7	15	ns
$E_{ON2}$	Turn-On Energy (Note 1)			-	225	270	$\mu\text{J}$
$E_{OFF}$	Turn-Off Energy (Note 2)			-	135	200	$\mu\text{J}$

### Thermal Characteristics

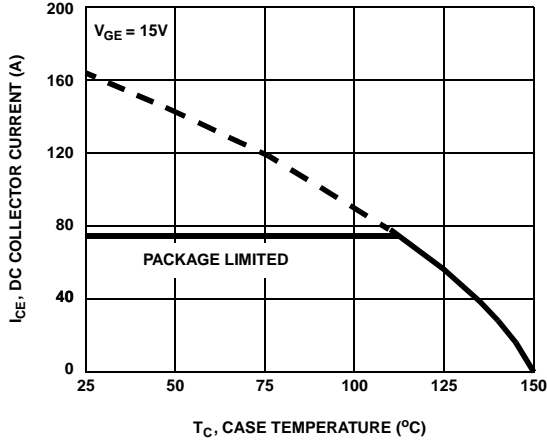
$R_{\theta JC}$	Thermal Resistance Junction-Case	TO-247	-	-	0.27	$^\circ\text{C/W}$
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NOTE:

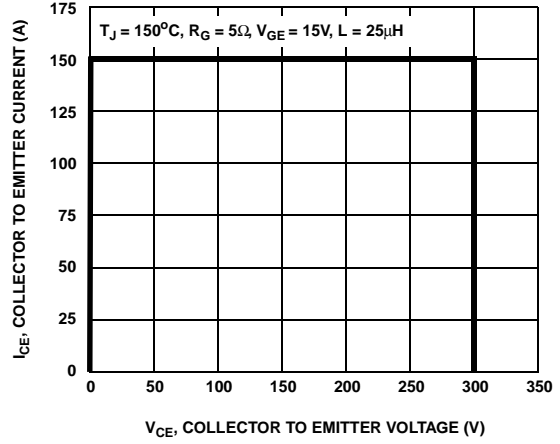
1.  $E_{ON2}$  is the turn-on loss when a typical diode is used in the test circuit and the diode is at the same  $T_J$  as the IGBT. The diode type is specified in figure 20.

2. Turn-Off Energy Loss ( $E_{OFF}$ ) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ( $I_{CE} = 0\text{A}$ ). All devices were tested per JEDEC Standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

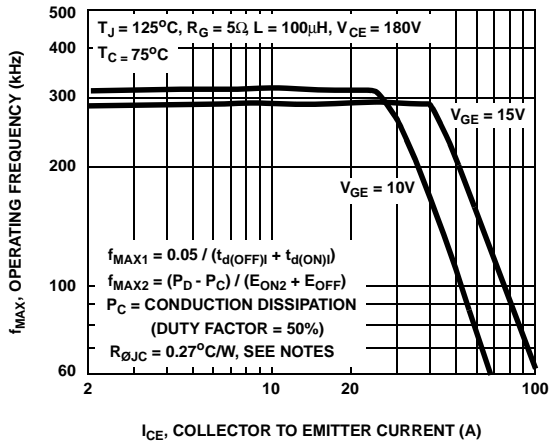
**Typical Performance Curves**  $T_J = 25^\circ\text{C}$  unless otherwise noted



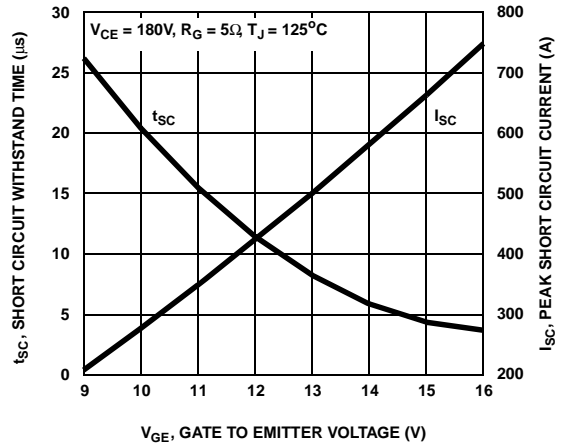
**Figure 1. DC Collector Current vs Case Temperature**



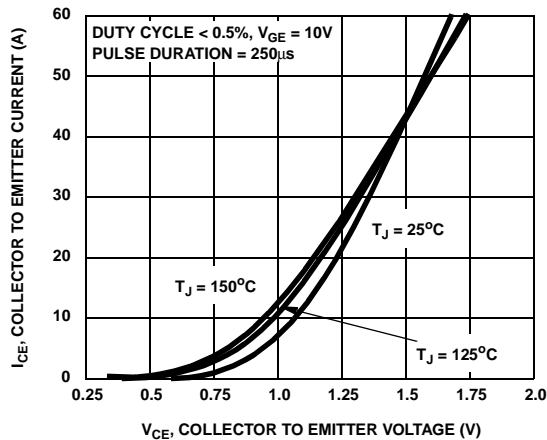
**Figure 2. Minimum Switching Safe Operating Area**



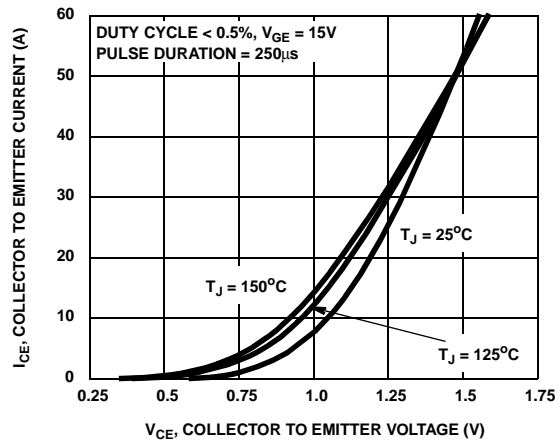
**Figure 3. Operating Frequency vs Collector to Emitter Current**



**Figure 4. Short Circuit Withstand Time**



**Figure 5. Collector to Emitter On-State Voltage**



**Figure 6. Collector to Emitter On-State Voltage**

Typical Performance Curves  $T_J = 25^\circ\text{C}$  unless otherwise noted (Continued)

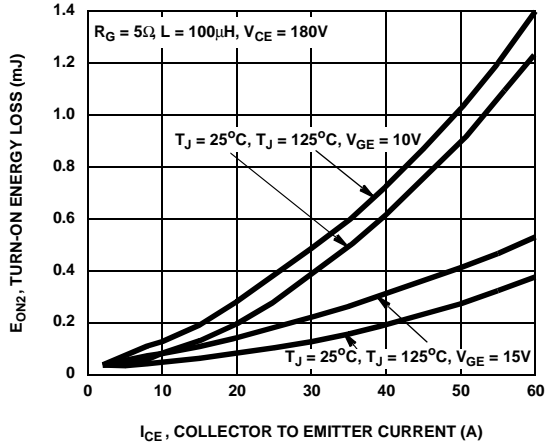


Figure 7. Turn-On Energy Loss vs Collector to Emitter Current

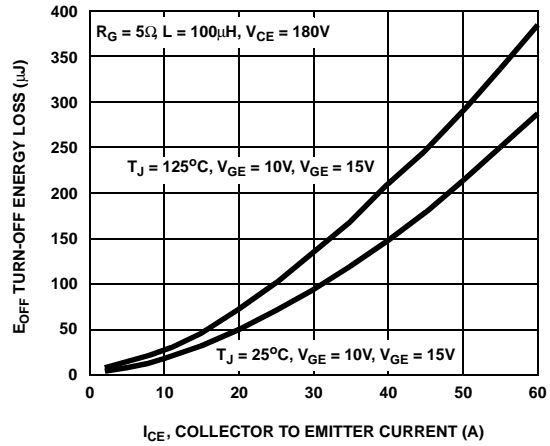


Figure 8. Turn-Off Energy Loss vs Collector to Emitter Current

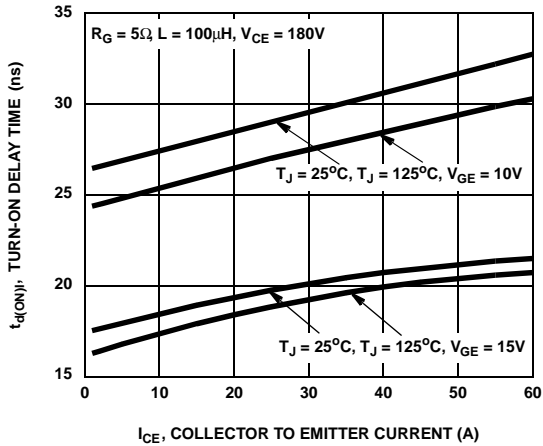


Figure 9. Turn-On Delay Time vs Collector to Emitter Current

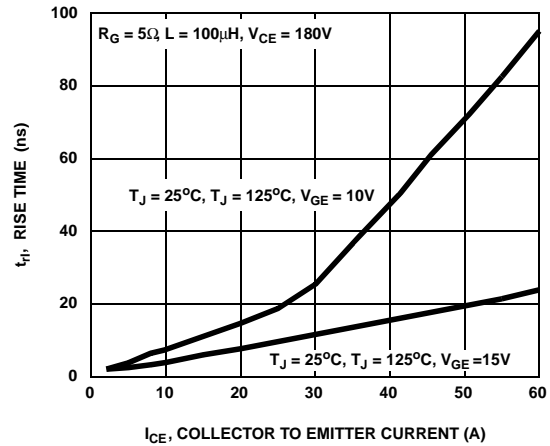


Figure 10. Turn-On Rise Time vs Collector to Emitter Current

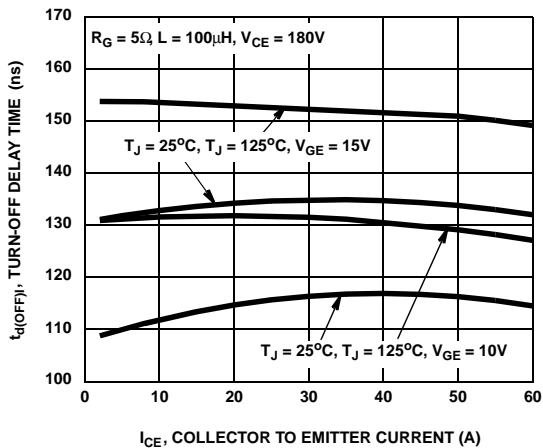


Figure 11. Turn-Off Delay Time vs Collector to Emitter Current

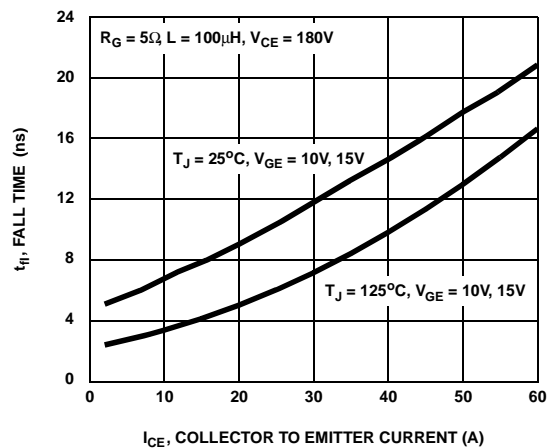
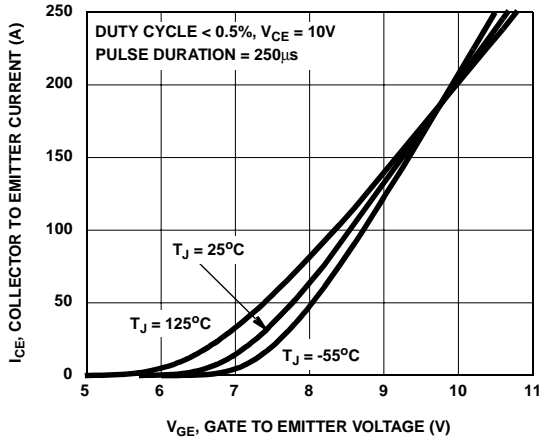
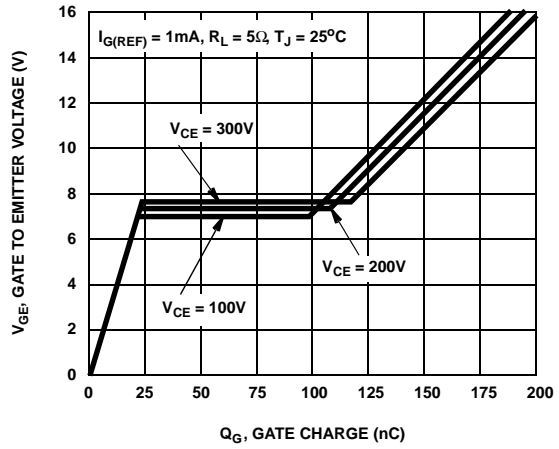


Figure 12. Fall Time vs Collector to Emitter Current

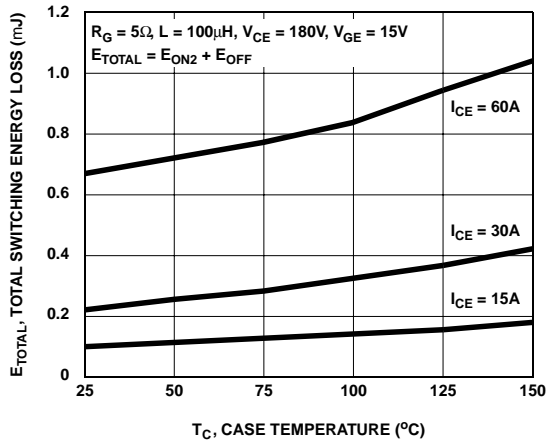
**Typical Performance Curves**  $T_J = 25^\circ\text{C}$  unless otherwise noted (Continued)



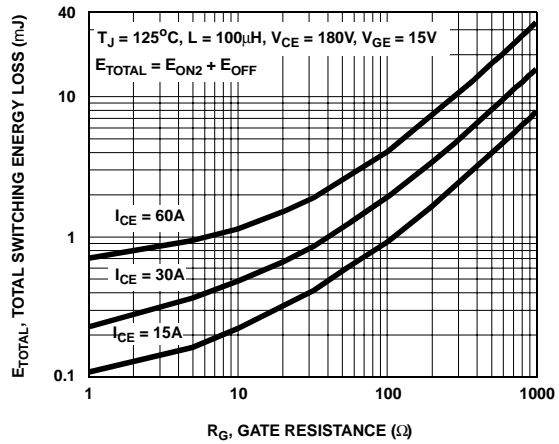
**Figure 13. Transfer Characteristic**



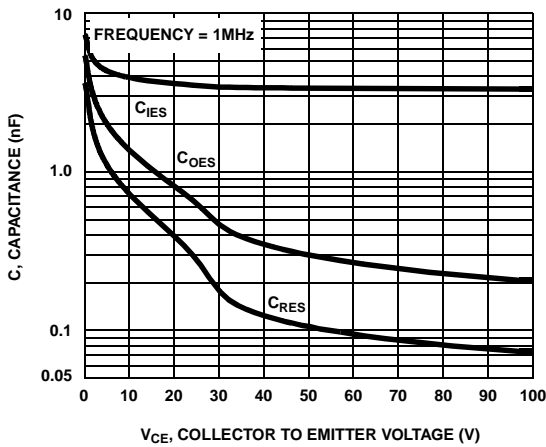
**Figure 14. Gate Charge**



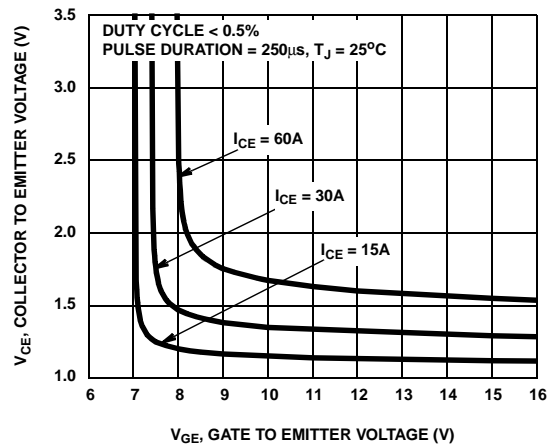
**Figure 15. Total Switching Loss vs Case Temperature**



**Figure 16. Total Switching Loss vs Gate Resistance**



**Figure 17. Capacitance vs Collector to Emitter Voltage**



**Figure 18. Collector to Emitter On-State Voltage vs Gate to Emitter Voltage**

**Typical Performance Curves**  $T_J = 25^\circ\text{C}$  unless otherwise noted (Continued)

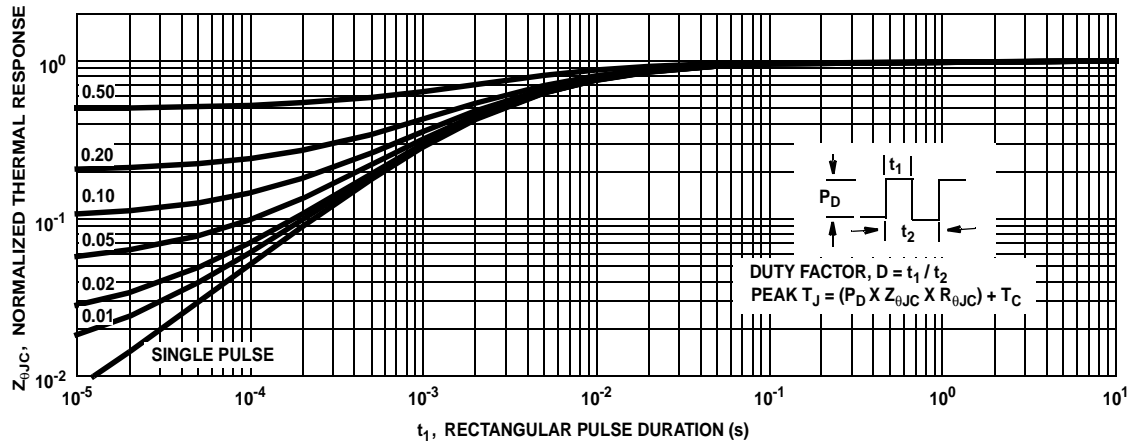


Figure 19. IGBT Normalized Transient Thermal Impedance, Junction to Case

**Test Circuit and Waveforms**

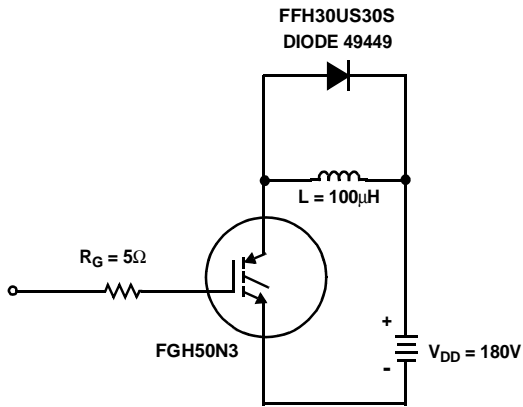


Figure 20. Inductive Switching Test Circuit

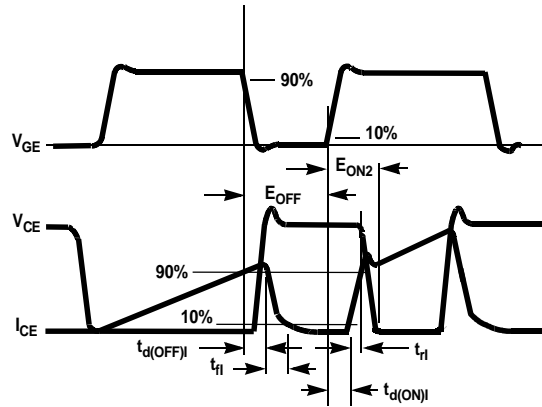


Figure 21. Switching Test Waveforms

## Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
3. Tips of soldering irons should be grounded.
4. Devices should never be inserted into or removed from circuits with power on.
5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
7. **Gate Protection** - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

## Operating Frequency Information

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 5, 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$ ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

$f_{MAX1}$  is defined by  $f_{MAX1} = 0.05 / (t_{d(OFF)1} + t_{d(ON)1})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)1}$  and  $t_{d(ON)1}$  are defined in Figure 21. Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{d(OFF)1}$  is important when controlling output ripple under a lightly loaded condition.

$f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON2})$ . The allowable dissipation ( $P_D$ ) is defined by  $P_D = (T_{JM} - T_C) / R_{\theta JC}$ . The sum of device switching and conduction losses must not exceed  $P_D$ . A 50% duty factor was used (Figure 3) and the conduction losses ( $P_C$ ) are approximated by  $P_C = (V_{CE} \times I_{CE}) / 2$ .

$E_{ON2}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 21.  $E_{ON2}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ )

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CROSSVOLT™	FRFET™	MicroPak™	QFET™	SuperSOT™-8
DOME™	GlobalOptoisolator™	MICROWIRE™	QS™	SyncFET™
EcoSPARK™	GTO™	MSX™	QT Optoelectronics™	TinyLogic™
E <sup>2</sup> CMOS™	HiSeC™	MSXPro™	Quiet Series™	TruTranslation™
EnSigna™	I <sup>2</sup> C™	OCX™	RapidConfigure™	UHC™
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Programmable Active Droop™		OPTOPLANAR™	SMART START™	

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