



# H3842S

## HIGH PERFORMANCE CURRENT MODE CONTROLERS

### Description

The H3842S is high performance fixed frequency current mode controllers. That is specifically designed for Off-Line and DC To DC converter applications offering the designer a costeffective solution with minimal external components. These integrated circuits feature a trimmed oscillator for precise duty cycle control. A temperature compensated reference, high gain error amplifier, current sensing compar-a-tor, and a high current totem pole output ideally suited for driving a power MOSFET. Also included are protective features consisting of input and reference undervoltage lockouts each with hysteresis, cycle-by-cycle current limiting, programmable output deadtime, and a latch for single pulse metering.

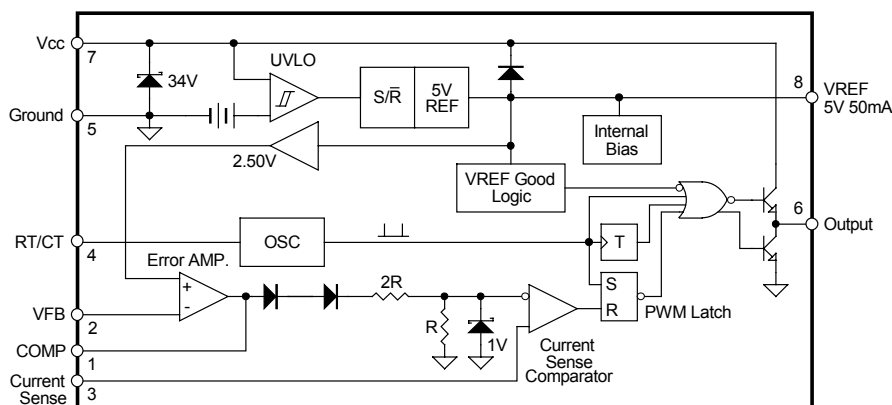
### Features

- Trimmed Oscillator for Precise Frequency Control
- Oscillator Frequency Guaranteed at 250 kHz
- Current Mode Operation to 500 kHz
- Automatic Feed Forward Compensation
- Latching PWM for Cycle-By-Cycle Current Limiting
- Internally Trimmed Reference with Undervoltage Lockout
- High Current Totem Pole Output
- Undervoltage Lockout with Hysteresis
- Low Startup and Operating Current

### Absolute Maximum Ratings

Symbol	Parameter	Value	Unit
$V_i$	Supply Voltage (low impedance source)	30	V
$V_i$	Supply Voltage ( $I_i < 30\text{mA}$ )	Self Limiting	
$I_o$	Output Current	$\pm 1$	A
$E_o$	Output Energy (capacitive load)	5	$\mu\text{J}$
	Analog Inputs (pin 2, 3)	-0.3 to 5.5	V
	Error Amplifier Output Sink Current	10	mA
$P_{tot}$	Power Dissipation at $T_{amb} \leq 25^\circ\text{C}$	800	mW
$T_{stg}$	Storage Temperature Range	-65 to 150	$^\circ\text{C}$
$T_J$	Junction Operating Temperature	-40 to 150	$^\circ\text{C}$
TL	Lead Temperature (soldering 10s)	300	$^\circ\text{C}$

### Block Diagram





### Pin Connection (Top View)

SOP-8P	Function	Description
	Pin1: Compensation	This pin is the Error Amplifier output and is made available for loop compensation.
	Pin2: Voltage Feedback	This is the inverting input of the Error Amplifier. It's normally connected to the switching power supply output through a resistor divider.
	Pin3: Current Sense	A voltage proportional to inductor current is connected to this input. The PWM uses this information to terminate the output switch conduction.
	Pin4: RT/CT	The oscillator frequency and maximum output duty cycle are programmed by connecting resistor RT to Vref and capacitor CT to ground. Operation to 500kHz is possible.
	Pin5: Ground	This pin is the combined control circuitry and power ground.
	Pin6: Output	This output directly drives the gate of a power MOSFET. Peak currents up to 1A are sourced and sunk by this pin.
	Pin7: Vcc	This pin is the positive supply of the control IC.
	Pin8: Vref	This is the reference output. It provides charging current for capacitor CT through resistor RT.

### Thermal Data

Symbol	Description	Max.	Units
Rth j-amb	Thermal Resistance Junction-ambient	100	°C/W

### Electrical Specifications

Parameter	Test Conditions	Min.	Typ.	Max.	Units
<i>Under-Voltage Lockout Section</i>					
Start Threshold		14.5	16	17.5	V
Min. Operating Voltage	After Turn On	8.5	10	11.5	V

### Electrical Characteristics

Unless otherwise stated, specifications apply for  $0 \leq T_a \leq 70^\circ\text{C}$   $V_{cc} = 15\text{V}$  (Note 1);  $R_T = 680\Omega$ ,  $C_T = 0.022\mu\text{F}$  for triangular mode,  $R_T = 10\text{k}$ ,  $C_T = 3.3\text{nf}$  for sawtooth mode

Parameter	Test Conditions	Min.	Typ.	Max.	Units
<i>Reference Section</i>					
Reference Output Voltage	$T_j = 25^\circ\text{C}, I_o = 1\text{mA}$	4.90	5	5.1	V
Line Regulation	$12\text{V} \leq V_{in} \leq 25\text{V}$	-	2.0	20	mV
Load Regulation	$1\text{mA} \leq I_o \leq 20\text{mA}$	-	3.0	25	mV
Temperature Stability	(Note 2)	-	0.2	-	mV/°C
Total Output Variation	Line, Load, Temp (Note 2)	4.82	-	5.18	V
Output Noise Voltage	$10\text{Hz} \leq f \leq 10\text{kHz}, T_j = 25^\circ\text{C}$ (Note 2)	-	50	-	uV
Long Term Stability	$T_a = 125^\circ\text{C}, 1000\text{Hrs}$ (Note 2)	-	5	-	mV
Output Short current	$T_a = 25^\circ\text{C}$	-30	-100	-180	mA
<i>Oscillator Section</i>					
Frequency	$T_j = 25^\circ\text{C}$	47	52	57	KHz
	$T_{low} \leq T_a \leq T_{high}$	46	52	60	KHz
Frequency Chnge with Temperature	$12\text{V} \leq V_{cc} \leq 25\text{V}$	-	0.2	1	%
Frequency Change with Temperature	$T_{low} \leq T_a \leq T_{high}$ (Note 2)	-	5	-	%
	$T_j = 25^\circ\text{C}$	-	0.5	-	%
Oscillator Voltage Swing	Peak to Peak	-	1.7	-	V
Discharge Current	$T_j = 25^\circ\text{C}$	7.8	8.3	9.3	mA
	$T_{low} \leq T_a \leq T_{high}$	7.2	-	9.5	mA



Parameter	Test Conditions	Min.	Typ.	Max.	Units
<i>Oscillator Section</i>					
Frequency	Tj=25°C	47	52	57	KHz
	Tlow≤Ta≤Thigh	46	52	60	KHz
Frequency Chnge with Temperature	12V≤Vcc≤25V	-	0.2	1	%
Frequency Change with Temperature	Tlow≤Ta≤Thigh (Note 2)	-	5	-	%
	Tj=25°C	-	0.5	-	%
Oscillator Voltage Swing	Peak to Peak	-	1.7	-	V
Discharge Current	Tj=25°C	7.8	8.3	9.3	mA
	Tlow≤Ta≤Thigh	7.2	-	9.5	mA
<i>Error Amplifier Section</i>					
Voltage Feedback Input	Vo=2.5V	2.42	2.50	2.58	V
Input Bias Current	VFB=5.0V	-	-0.1	-2.0	uA
Open Loop Voltage Gain	2≤Vo≤4 V	65	90	-	dB
Unity Gain Bandwidth	Tj=25°C (Note 2)	0.7	1.0	-	MHz
Power Supply Rejection Ratio	12V≤Vcc≤25V	60	70	-	dB
Output Current	Sink (Vo=1.1V, VFB=2.7V)	2.0	12	-	mA
	Source (Vo=5.0V, VFB=2.3V)	-0.5	-1.0	-	mA
Output Voltage Swing High State	VFB=2.3V, RL=15K to ground	5	6	-	V
		Low State	VFB=2.7V, RL=15K to Vref	-	0.7
<i>Current Sense Section</i>					
Current Sense Input Voltage Gain	(Notes 3 & 4)	2.85	3.0	3.15	V/V
Maximum Current Sense Input Thresold	(Note 3)	0.9	1.0	1.1	V
Ppwer Supply Rejection Ratio	12V≤Vcc≤25V (Note 3)	-	70	-	dB
Input Bias Current		-	-2.0	-10	uA
Propagation Delay	Tj=25°C (Note 2)	-	150	300	nS
<i>Output Section</i>					
Output Voltage	Low State (Isink=20mA)	-	0.1	0.4	V
	Low State (Isink=200mA)	-	1.5	2.2	V
	High State (Isource=20mA)	13	13.5	-	V
	High State (Isource=200mA)	12	13.4	-	V
Output Voltage Rise Time	Tj=25°C, CL=1.0nF (Note 2)	-	50	150	nS
Output Voltage Fall Time	Tj=25°C, CL=1.0nF (Note 2)	-	50	150	nS
Output Voltage with UVLO Activated	Vcc=6V, Isink=1.0mA	-	-0.1	-1.1	V
<i>Total Device</i>					
Start-Up Current		-	0.1	0.5	mA
Operating Supply Current	Vpin2=Vpin3=0, RT=10K, CT=3.3nF	-	11	17	mA
Vcc Zener Voltage	Icc=25 mA	-	34	-	V

Note 1: Adjust Vcc above the start threshold before setting at 15V.

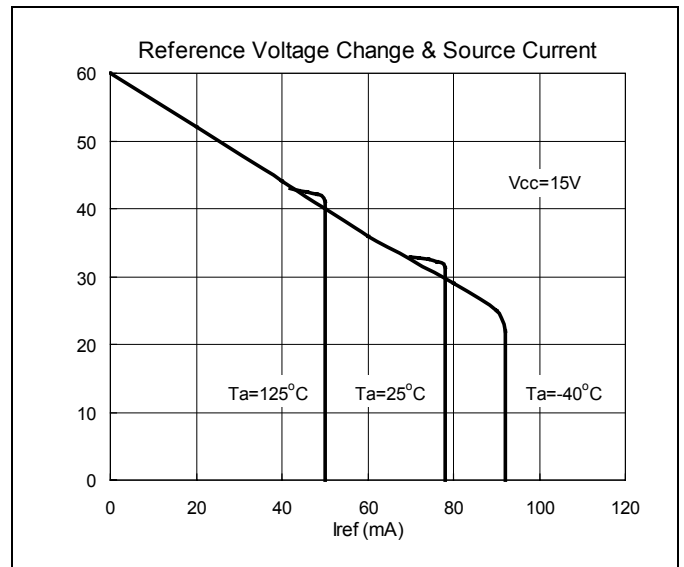
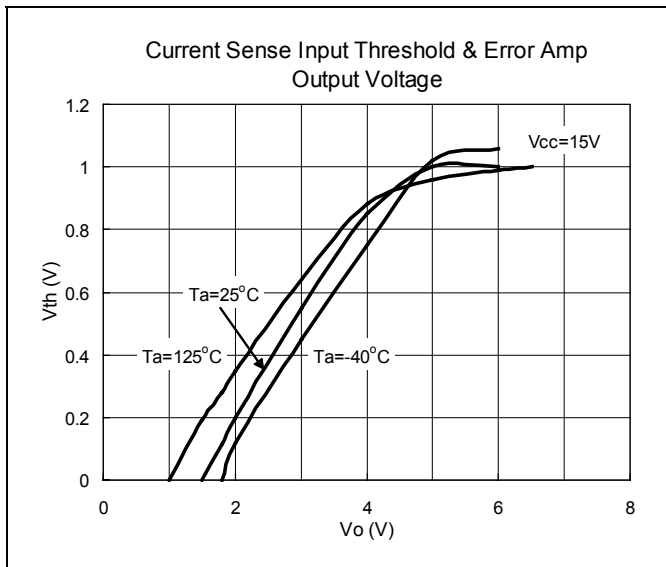
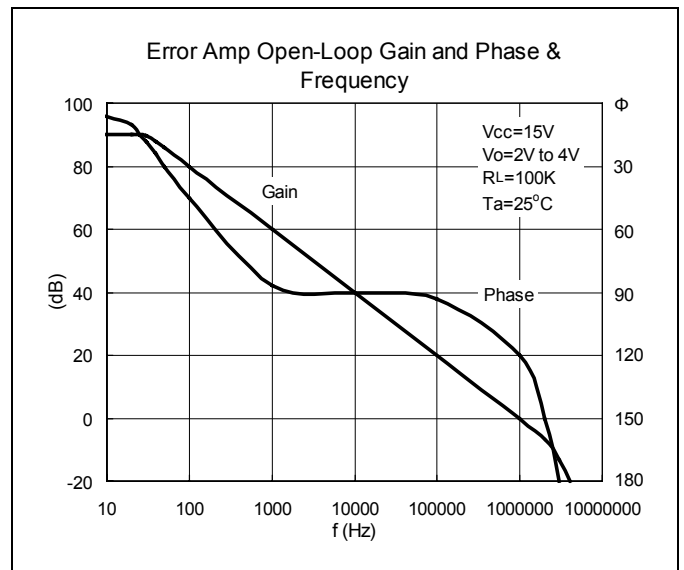
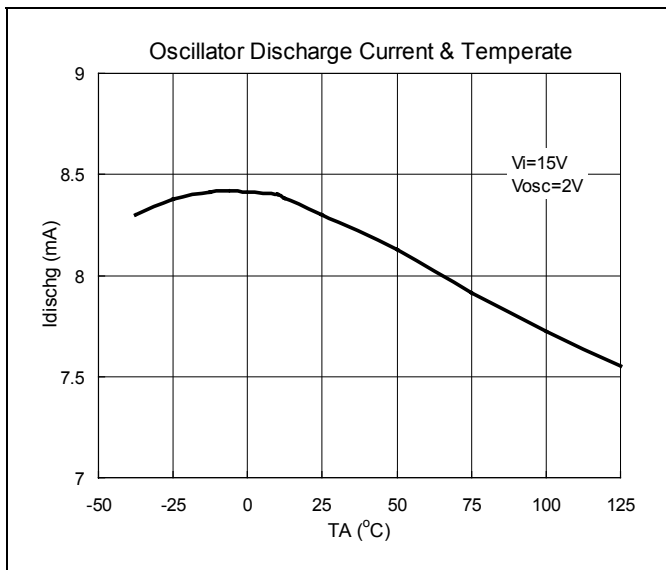
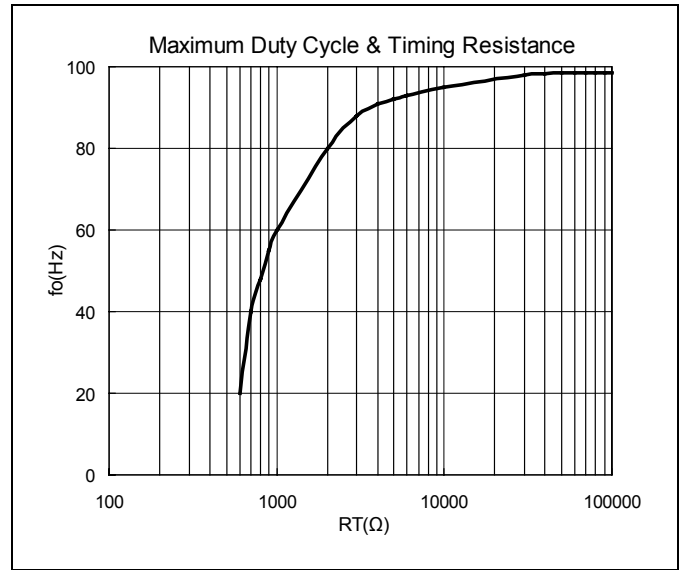
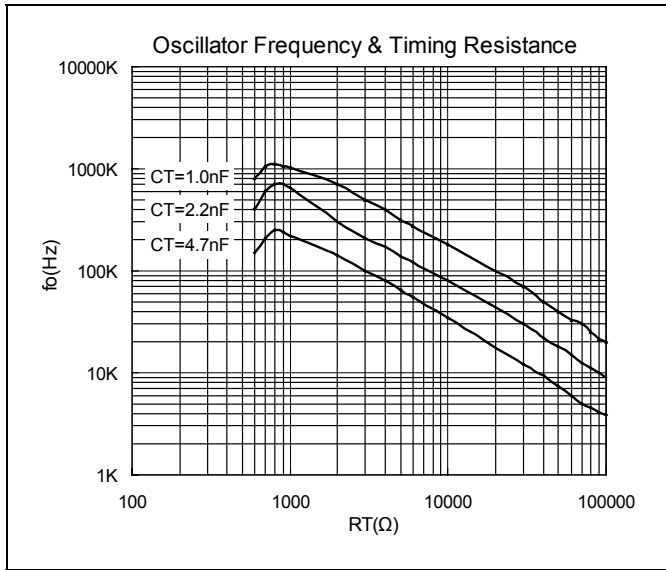
Note 2: These parameters, although guaranteed are not 100% tested in production.

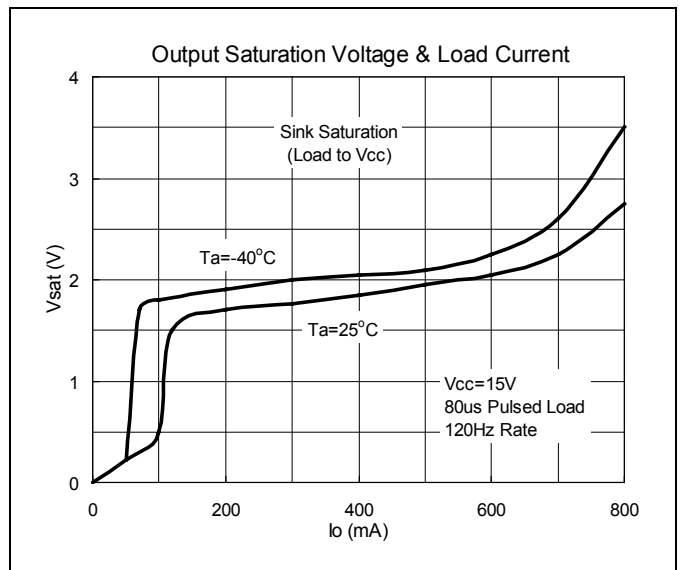
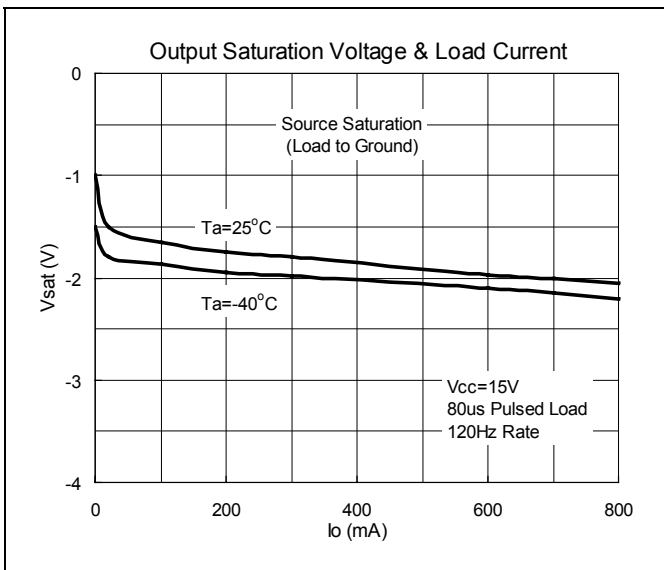
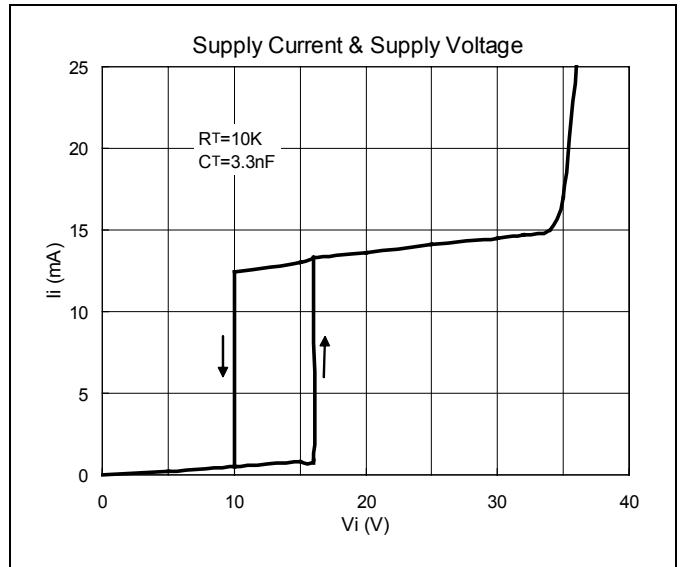
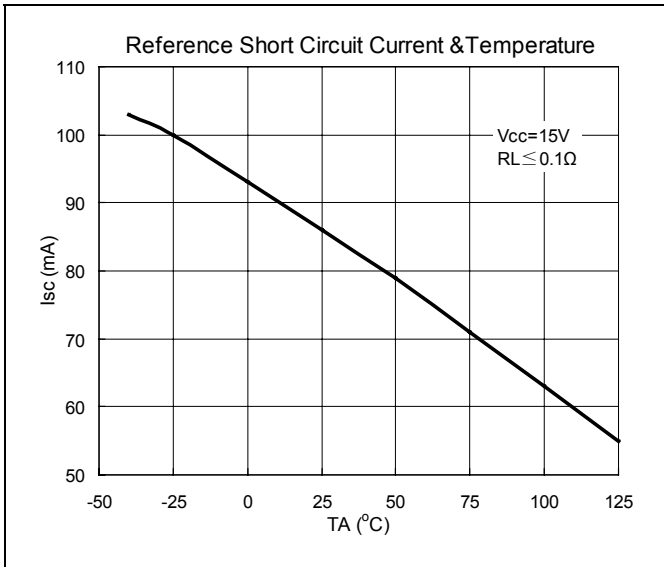
Note 3: Parameter measured at trip point of latch with Vpin2=0.

Note 4: Gain defined as:  $A = V_{pin1}/V_{pin3}$ ,  $0 \leq V_{pin3} \leq 0.8V$



### Characteristics Curve

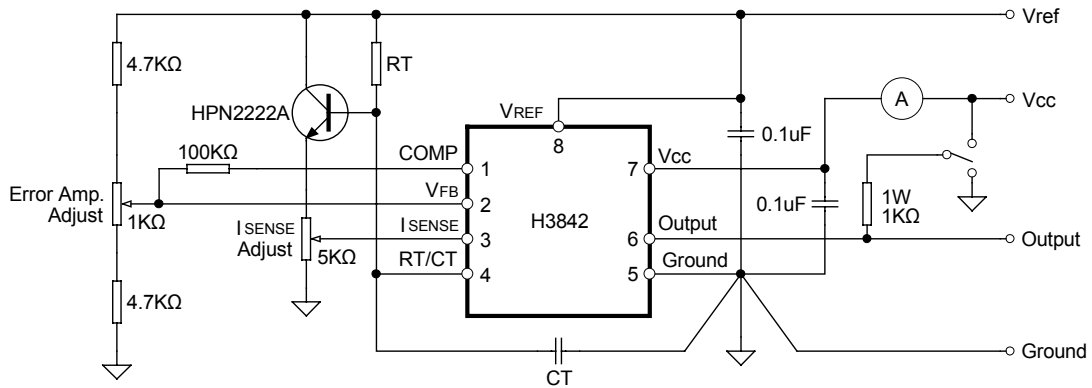






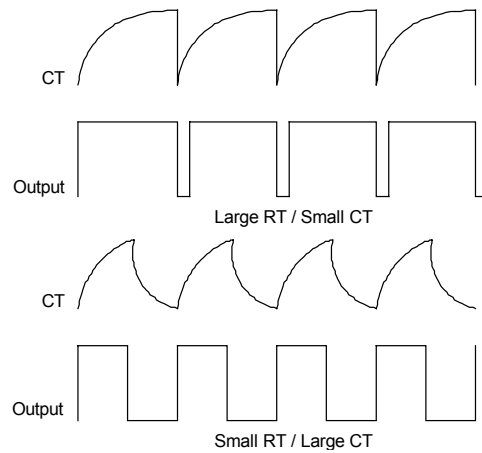
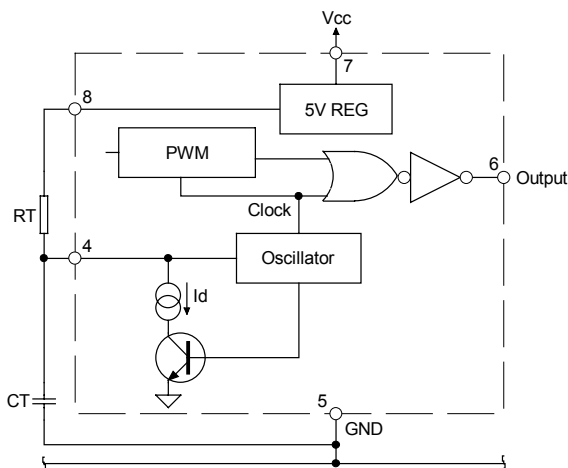
### Application Information

#### Open Loop Test Circuit

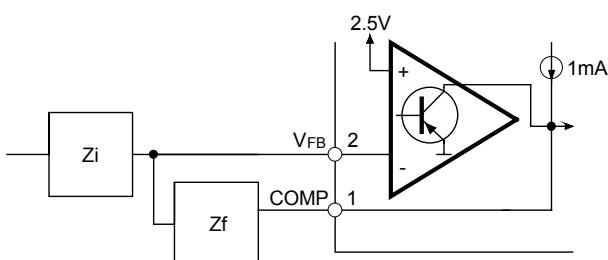


High peak currents associated with capacitive loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected close to pin5 in a single point ground. The transistor and 5KΩ potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to pin3.

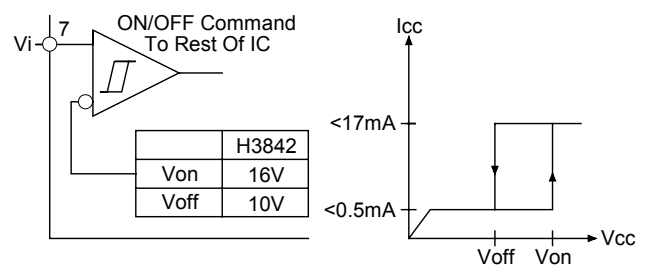
#### Oscillator and Output Waveforms



#### Error Amp Configuration

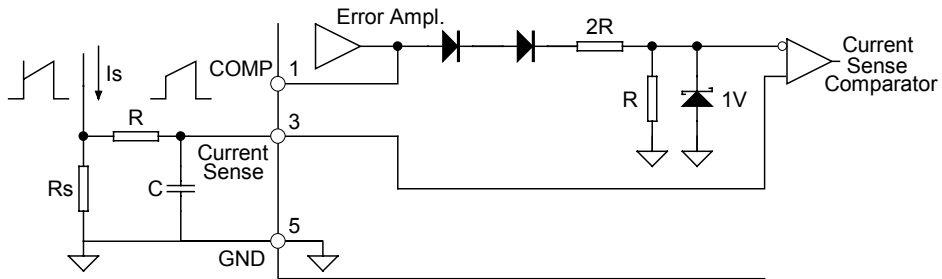


#### Under Voltage Lockout



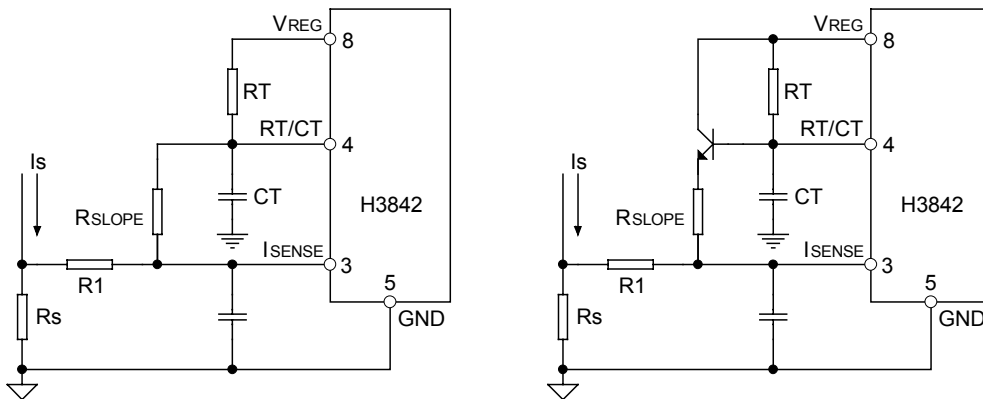


### Current Sense Circuit

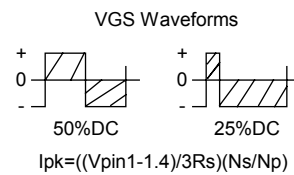
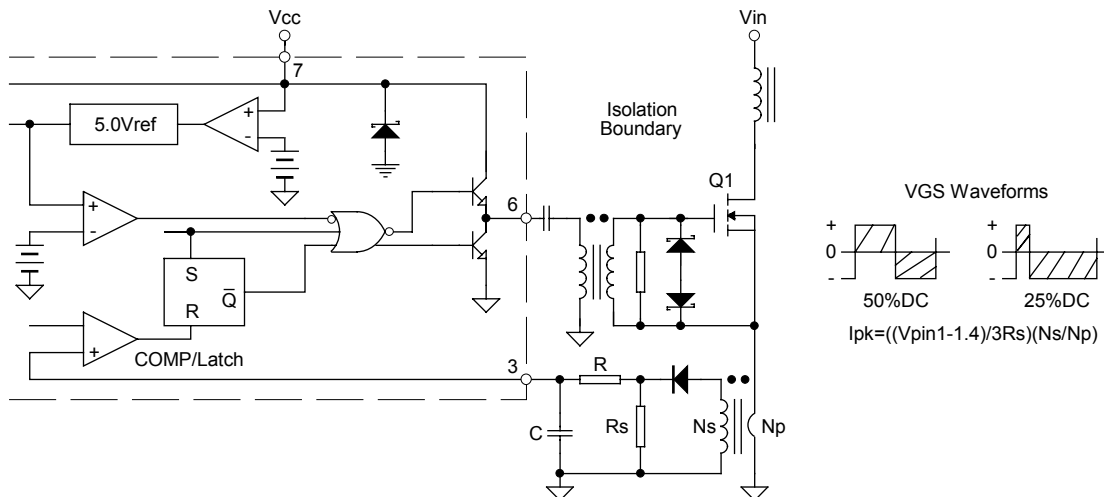


Peak current ( $I_s$ ) is determined by the formula  
 $I_s(\max.) \approx 1V/R_s$   
 A small RC filter may be required to suppress switch transients.

### Slope Compensation Techniques

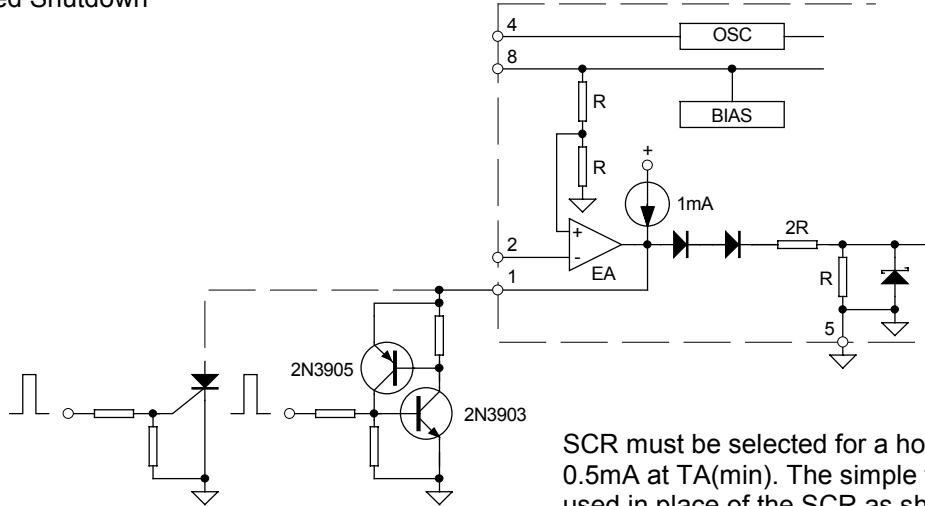


### Isolated MOSFET Drive and Current Transformer Sensing



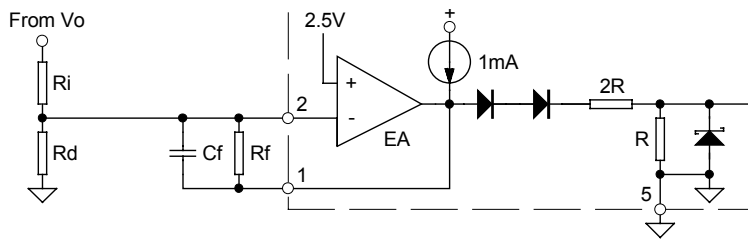


### Latched Shutdown

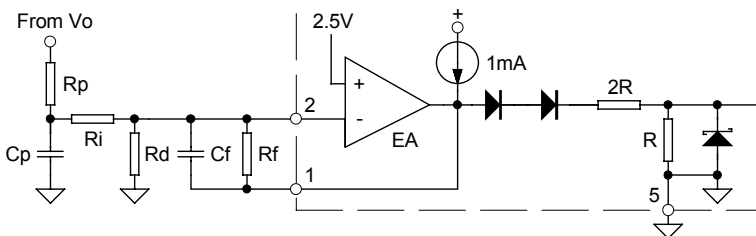


SCR must be selected for a holding current of less than 0.5mA at TA(min). The simple two transistor circuit can be used in place of the SCR as shown. All resistors are 10K.

### Error Amplifier Compensation

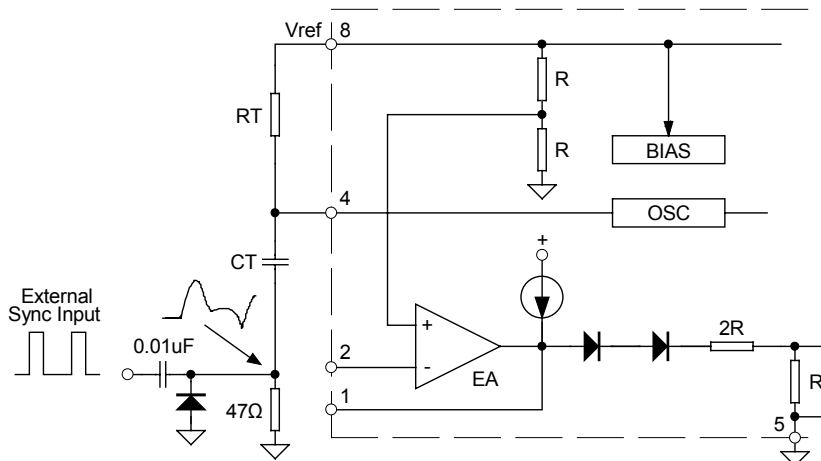


Error Amp compensation circuit for stabilizing any current-mode topology except for boost and flyback converters operating with continuous inductor current



Error Amp compensation circuit for stabilizing current-mode boost and flyback topologies operating with continuous inductor current.

### External Clock Synchronization

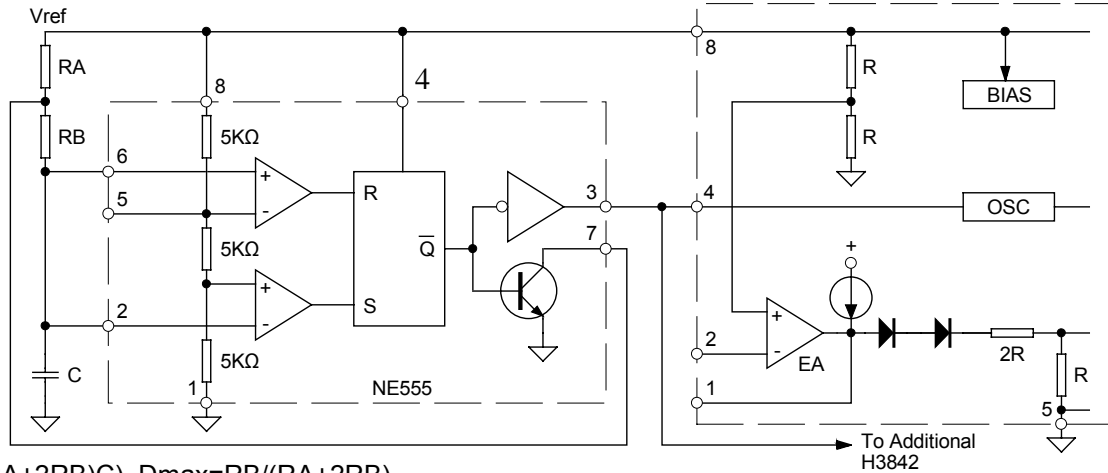


The diode clamp is required if the Sync amplitude is large enough to cause the bottom side of CT to go more than 300mV below ground



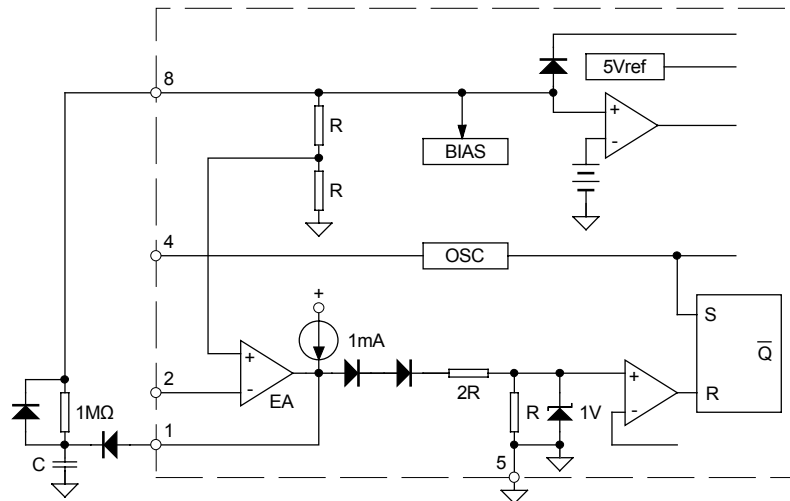


### External Duty Cycle Clamp and Multi Unit Synchronization

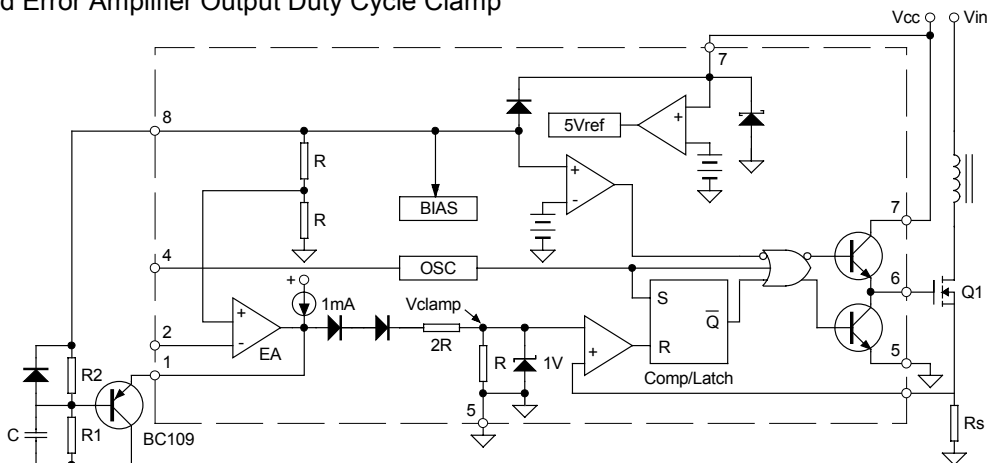


$$f = 1.44 / ((RA + 2RB)C), D_{max} = RB / (RA + 2RB)$$

### Soft-Start Circuit



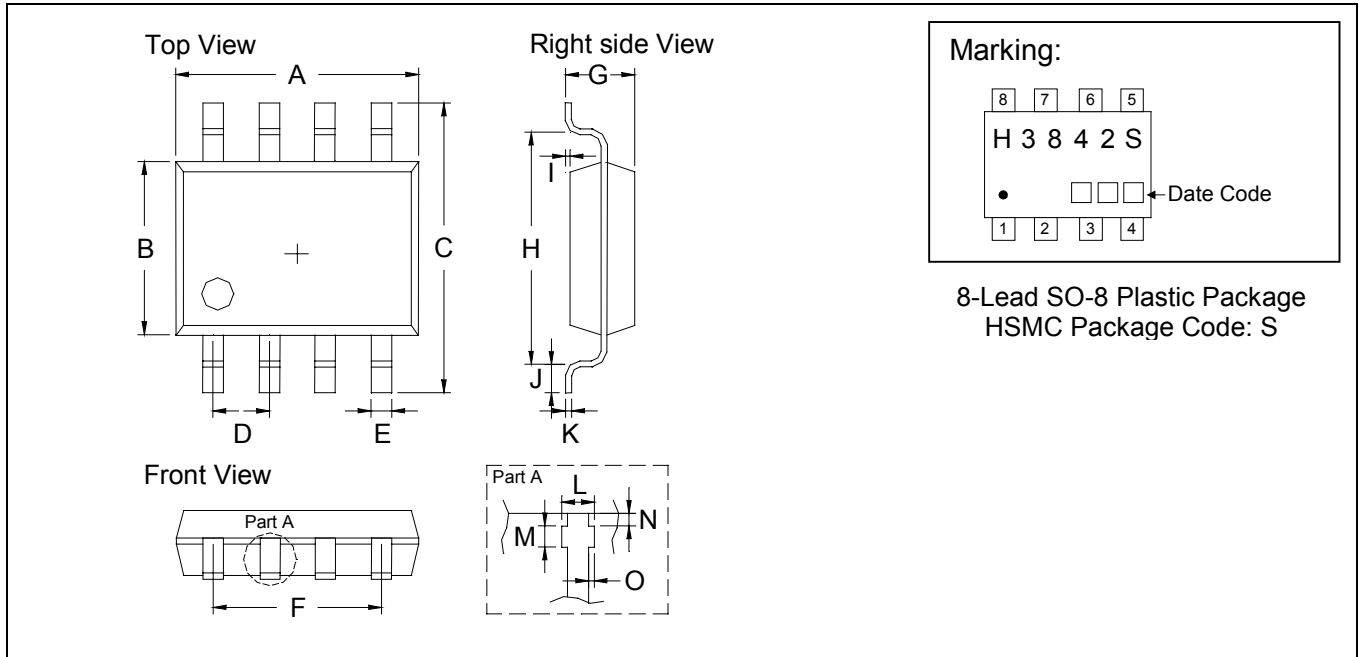
### Soft-Start and Error Amplifier Output Duty Cycle Clamp



$$V_{clamp} = R1 / (R1 + R2) \text{ where } 0 < V_{clamp} < 1V, I_{pk(max)} = V_{clamp} / R_S$$



### SO-8 Dimension



\*: Typical

DIM	Inches		Millimeters		DIM	Inches		Millimeters	
	Min.	Max.	Min.	Max.		Min.	Max.	Min.	Max.
A	0.1909	0.2007	4.85	5.10	I	0.0019	0.0078	0.05	0.20
B	0.1515	0.1555	3.85	3.95	J	0.0118	0.0275	0.30	0.70
C	0.2283	0.2441	5.80	6.20	K	0.0074	0.0098	0.19	0.25
D	0.0480	0.0519	1.22	1.32	L	0.0145	0.0204	0.37	0.52
E	0.0145	0.0185	0.37	0.47	M	0.0118	0.0197	0.30	0.50
F	0.1472	0.1527	3.74	3.88	N	0.0031	0.0051	0.08	0.13
G	0.0570	0.0649	1.45	1.65	O	0.0000	0.0059	0.00	0.15
H	0.1889	0.2007	4.80	5.10					

- Notes:**
- 1.Dimension and tolerance based on our Spec. dated Aug. 01,1999.
  - 2.Controlling dimension: millimeters.
  - 3.Maximum lead thickness includes lead finish thickness, and minimum lead thickness is the minimum thickness of base material.
  - 4.If there is any question with packing specification or packing method, please contact your local HSMC sales office.

**Material:**

- Lead: 42 Alloy; solder plating
- Mold Compound: Epoxy resin family, flammability solid burning class: UL94V-0

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