

# 捷多邦, 专业PCB打样エ厂, 24小时加急出货 HGTG15N120C3D

35A, 1200V, UFS Series N-Channel IGBT with Anti-Parallel Hyperfast Diode

May 1997

#### **Features**

- 35A, 1200V at T<sub>C</sub> = 25°C
- 1200V Switching SOA Capability
- Typical Fall Time at T<sub>J</sub> = 150°C ......350ns
- Short Circuit Rating
- Low Conduction Loss

# **Ordering Information**

PART NUMBER	PACKAGE	BRAND		
HGTG15N120C3D	TO-247	15N120C3D		

NOTE: When ordering, use the entire part number.

Formerly Developmental Type TA49133.

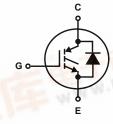
# Description

The HGTG15N120C3D is a MOS gated high voltage switching device combining the best features of MOSFETs and bipolar transistors. This device has the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as: AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.

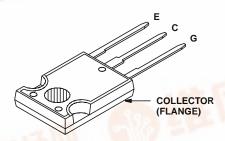
The diode used in anti-Parallel with the IGBT is the same as the RHRP15120. The IGBT was formerly development type TA49145.

# Symbol



#### **Packaging**

#### **JEDEC STYLE TO-247**



#### INTERSIL CORPRATION'S IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS:

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,567,641
4,587,713	4,598,461	4,605,948	4,618,872	4,620,211	4,631,564	4,639,754	4,639,762
4,641,162	4,644,637	4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690
4,794,432	4,801,986	4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606
4,860,080	4,883,767	4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951

<b>Absolute Maximum Ratings</b> $T_C = 25^{\circ}$ C, Unless Otherwise Specified	HGTG15N120C3D	UNITS
Collector to Emitter Voltage	1200	V
Collector Current Continuous		
At $T_C = 25^{\circ}C$	35	Α
At $T_C = 110^{\circ}C$ $I_{C110}$	15	Α
Collector Current Pulsed (Note 1)	120	Α
Gate to Emitter Voltage ContinuousV <sub>GES</sub>	±20	V
Gate to Emitter Voltage Pulsed	±30	V
Switching Safe Operating Area at T <sub>J</sub> = 150 <sup>o</sup> C, Figure 14SSOA	15A at 1200V	
Power Dissipation Total at T <sub>C</sub> = 25°C	164	W
Power Dissipation Derating T <sub>C</sub> > 25 <sup>o</sup> C	1.32	W/oC
Operating and Storage Junction Temperature Range	-55 to 150	οС
Maximum Lead Temperature for Soldering	260	οС
Short Circuit Withstand Time (Note 2) at V <sub>GE</sub> = 15V	6	μs
Short Circuit Withstand Time (Note 2) at V <sub>GE</sub> = 10V	25	μs

- 1. Repetitive Rating: Pulse width limited by maximum junction temperature.
- 2.  $V_{CE(PK)} = 360V$ ,  $T_J = 125^{\circ}C$ ,  $R_{GE} = 25\Omega$ .

# **Electrical Specifications** $T_C = 25^{\circ}C$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS		MIN	TYP	MAX	UNITS
Collector to Emitter Breakdown Voltage	BV <sub>CES</sub>	$I_C = 250\mu A, V_{GE} = 0V$		1200	-	-	V
Collector to Emitter Leakage Current	I <sub>CES</sub>	V <sub>CE</sub> = BV <sub>CES</sub>	$T_{C} = 25^{\circ}C$	-	-	250	μА
		V <sub>CE</sub> = BV <sub>CES</sub>	$T_C = 150^{\circ}C$	-	-	3.0	mA
Collector to Emitter Saturation Voltage	V <sub>CE(SAT)</sub>	I <sub>C</sub> = I <sub>C110</sub> , V <sub>GE</sub> = 15V	$T_{C} = 25^{\circ}C$	-	2.3	3.5	V
			$T_C = 150^{\circ}C$	-	2.4	3.2	V
Gate to Emitter Threshold Voltage	V <sub>GE(TH)</sub>	I <sub>C</sub> = 250μA, V <sub>CE</sub>	= V <sub>GE</sub>	4.0	5.6	7.5	V
Gate to Emitter Leakage Current	I <sub>GES</sub>	V <sub>GE</sub> = ±20V		-	-	±100	nA
Switching SOA	SSOA	$T_{J} = 150^{\circ}C$	V <sub>CE(PK)</sub> = 960V	40	-	-	Α
		$R_G = 10\Omega$ $V_{GE} = 15V$ L = 1mH	V <sub>CE(PK)</sub> = 1200V	15	-	-	А
Gate to Emitter Plateau Voltage	$V_{GEP}$	$I_{C} = I_{C110}, V_{CE} = 0.5 \text{ BV}_{CES}$		-	8.8	-	V
On-State Gate Charge	-State Gate Charge $ \begin{array}{c c} Q_{g(ON)} & I_{C} = I_{C110}, & V_{GE} = 15V \\ V_{CE} = 0.5 \text{ BV}_{CES} & V_{GE} = 20V \\ \end{array} $	$I_C = I_{C110}$	V <sub>GE</sub> = 15V	-	75	100	nC
		V <sub>GE</sub> = 20V	-	100	130	nC	
Current Turn-On Delay Time	t <sub>d</sub> (ON)I	$T_J = 150^{\circ}C,$ $I_{CE} = I_{C110},$ $V_{CE(PK)} = 0.8 \text{ BV}_{CES},$ $V_{GE} = 15V,$		-	17	-	ns
Current Rise Time	t <sub>rl</sub>			-	25	-	ns
Current Turn-Off Delay Time	t <sub>d</sub> (OFF)I			-	470	550	ns
Current Fall Time	t <sub>fl</sub>	$R_G = 10\Omega$ , L = 1mH	-	350	400	ns	
Turn-On Energy (Note 3)	E <sub>ON</sub>		-	2100	-	μJ	
Turn-Off Energy (Note 3)	E <sub>OFF</sub>	1	-	4700	-	μJ	
Diode Forward Voltage	V <sub>EC</sub>	I <sub>EC</sub> = 15A		-	-	3.2	V
Diode Reverse Recovery Time	t <sub>rr</sub>	I <sub>EC</sub> = 1A, dI <sub>EC</sub> /dt = 200A/μs		-	-	65	ns
		$I_{EC} = 15A$ , $dI_{EC}/dt = 200A/\mu s$		-	-	75	ns
Thermal Resistance	$R_{ heta JC}$	IGBT		-	-	0.76	°C/W
		Diode		-	-	1.5	°C/W

#### NOTE:

<sup>3.</sup> Turn-Off Energy Loss (E<sub>OFF</sub>) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I<sub>CE</sub> = 0A). The HGTG15N120C3D was tested per JEDEC standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss. Turn-On Energy loss (E<sub>ON</sub>) includes losses due to the diode recovery.

# Typical Performance Curves Unless Otherwise Specified

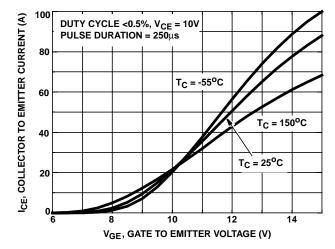
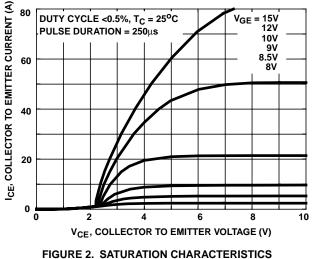


FIGURE 1. TRANSFER CHARACTERISTICS



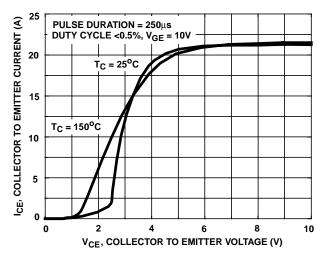


FIGURE 3. COLLECTOR TO EMITTER ON-STATE VOLTAGE

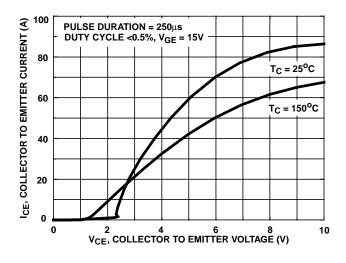
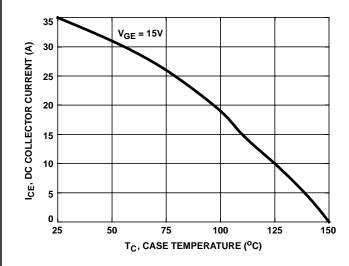


FIGURE 4. COLLECTOR TO EMITTER ON-STATE VOLTAGE

# Typical Performance Curves Unless Otherwise Specified (Continued)



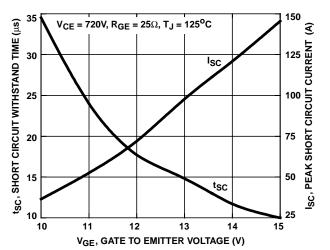
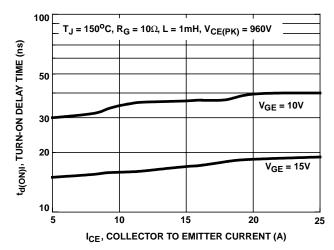


FIGURE 5. DC COLLECTOR CURRENT AS A FUNCTION OF CASE TEMPERATURE

FIGURE 6. SHORT CIRCUIT WITHSTAND TIME



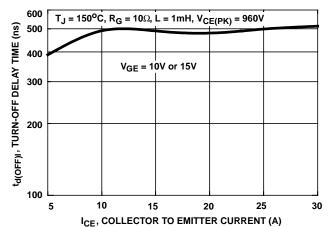
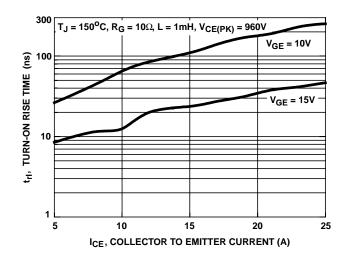


FIGURE 7. TURN-ON DELAY TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

FIGURE 8. TURN-OFF DELAY TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

# Typical Performance Curves Unless Otherwise Specified (Continued)



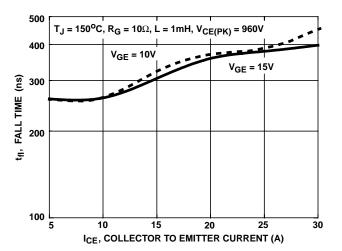
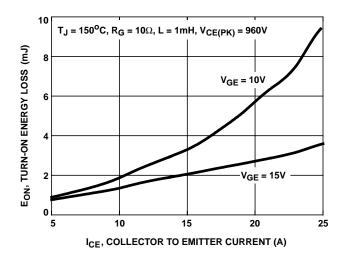


FIGURE 9. TURN-ON RISE TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

FIGURE 10. TURN-OFF FALL TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT



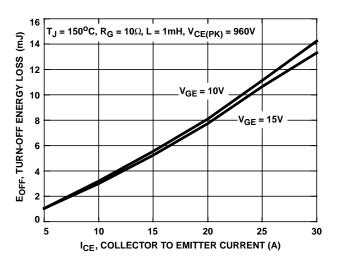
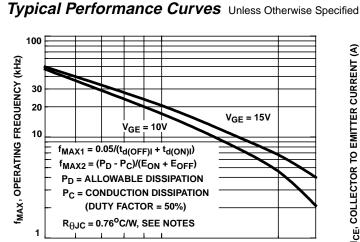


FIGURE 11. TURN-ON ENERGY LOSS AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

FIGURE 12. TURN-OFF ENERGY LOSS AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT



(Continued)

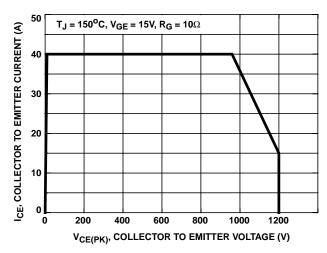
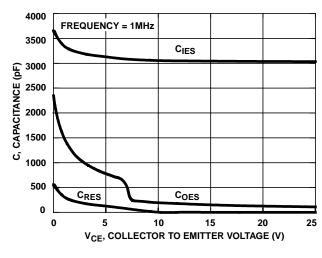


FIGURE 13. OPERATING FREQUENCY AS A FUNCTION OF **COLLECTOR TO EMITTER CURRENT** 

I<sub>CE</sub>, COLLECTOR TO EMITTER CURRENT (A)

FIGURE 14. MINIMUM SWITCHING SAFE OPERATING AREA



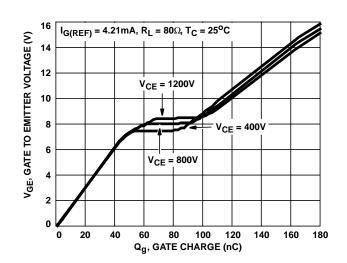


FIGURE 15. CAPACITANCE AS A FUNCTION OF COLLECTOR TO EMITTER VOLTAGE

FIGURE 16. GATE CHARGE WAVEFORM

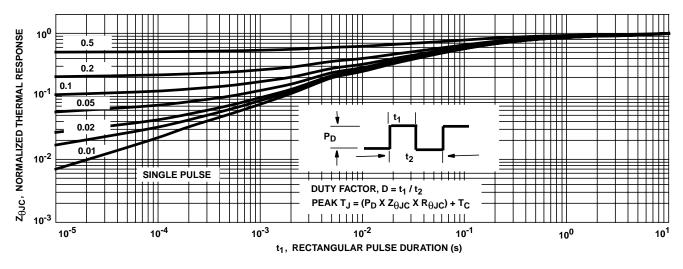
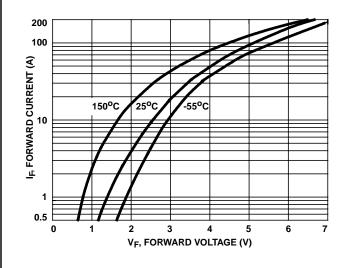


FIGURE 17. IGBT NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

# Typical Performance Curves Unless Otherwise Specified (Continued)



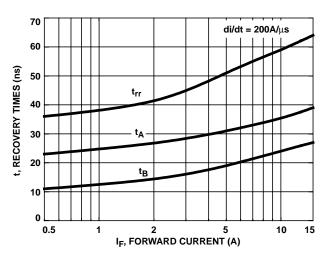


FIGURE 18. DIODE FORWARD CURRENT AS A FUNCTION OF FORWARD VOLTAGE DROP

FIGURE 19. RECOVERY TIMES AS A FUNCTION OF FORWARD CURRENT

## Test Circuit and Waveform

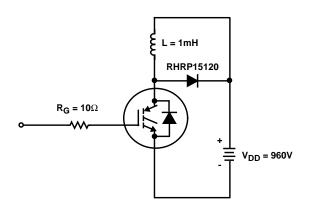


FIGURE 20. INDUCTIVE SWITCHING TEST CIRCUIT

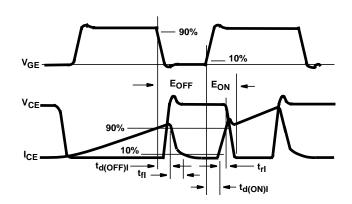


FIGURE 21. SWITCHING TEST WAVEFORMS

# Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- Gate Voltage Rating Never exceed the gate-voltage rating of V<sub>GEM</sub>. Exceeding the rated V<sub>GE</sub> can result in permanent damage to the oxide layer in the gate region.
- 6. Gate Termination The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
- Gate Protection These devices do not have an internal monolithic zener diode from gate to emitter. If gate protection is required an external zener is recommended.

ECCOSORBD™ is a Trademark of Emerson and Cumming, Inc.

### **Operating Frequency Information**

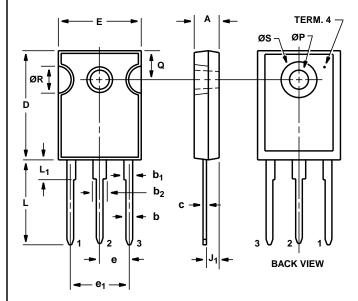
Operating frequency information for a typical device (Figure 13) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 4, 7, 8, 11 and 12. The operating frequency plot (Figure 13) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$  whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 $f_{MAX1}$  is defined by  $f_{MAX1}=0.05/(t_{d(OFF)I}+t_{d(ON)I}).$  Deadtime (the denominator) has been arbitrarily held to 10% of the on- state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 21. Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JMAX}.$   $t_{d(OFF)I}$  is important when controlling output ripple under a lightly loaded condition.

 $f_{MAX2}$  is defined by  $f_{MAX2}=(P_D-P_C)/(E_{OFF}+E_{ON}).$  The allowable dissipation  $(P_D)$  is defined by  $P_D=(T_{JMAX}-T_C)/R_{\theta JC}.$  The sum of device switching and conduction losses must not exceed  $P_D.$  A 50% duty factor was used (Figure 13) and the conduction losses  $(P_C)$  are approximated by  $P_C=(V_{CE}\times I_{CE})/2.$ 

 $E_{ON}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 21.  $E_{ON}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e. the collector current equals zero ( $I_{CE} = 0$ ).

**TO-247**3 LEAD JEDEC STYLE TO-247 PLASTIC PACKAGE



LEAD 1 - GATE

LEAD 2 - COLLECTOR LEAD 3 - EMITTER TERM. 4 - COLLECTOR

	INCHES		MILLIM		
SYMBOL	MIN	MAX	MIN	MAX	NOTES
Α	0.180	0.190	4.58	4.82	-
b	0.046	0.051	1.17	1.29	2, 3
b <sub>1</sub>	0.060	0.070	1.53	1.77	1, 2
b <sub>2</sub>	0.095	0.105	2.42	2.66	1, 2
С	0.020	0.026	0.51	0.66	1, 2, 3
D	0.800	0.820	20.32	20.82	-
Е	0.605	0.625	15.37	15.87	-
е	0.219 TYP		5.56 TYP		4
e <sub>1</sub>	0.438 BSC		11.12 BSC		4
J <sub>1</sub>	0.090	0.105	2.29 2.66		5
L	0.620	0.640	15.75	16.25	-
L <sub>1</sub>	0.145	0.155	3.69	3.93	1
ØP	0.138	0.144	3.51	3.65	-
Q	0.210	0.220	5.34	5.58	-
ØR	0.195	0.205	4.96	5.20	-
ØS	0.260	0.270	6.61	6.85	-

#### NOTES:

- 1. Lead dimension and finish uncontrolled in L<sub>1</sub>.
- 2. Lead dimension (without solder).
- 3. Add typically 0.002 inches (0.05mm) for solder coating.
- 4. Position of lead to be measured 0.250 inches (6.35mm) from bottom of dimension D.
- Position of lead to be measured 0.100 inches (2.54mm) from bottom of dimension D.
- 6. Controlling dimension: Inch.
- 7. Revision 1 dated 1-93.

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