

## Low Offset, Fast Settling Video Operational Amplifier

HOS-060

FEATURES
<1mV V<sub>OS</sub>
Low Drift
80ns Settling to 0.1%; 200ns to 0.01%
100mA Output @ ±10V

APPLICATIONS
D/A Current Converter
Video Pulse Amplifier
CRT Deflection Amplifier
Wideband Current Booster

### GENERAL DESCRIPTION

The HOS-060 Operational Amplifier is an extension of the proven hybrid technology used in the HOS-050 series of op amps.

The FET input and high-performance characteristics, including wide bandwidth and fast settling, make it useful for a variety of applications in the processing of video signals.

Recent innovations in circuit design have been incorporated into the HOS-060 to make it extremely useful to the designer who needs outstanding performance in current boosting, voltage amplification, impedance matching, or a multiplicity of other high-frequency requirements.

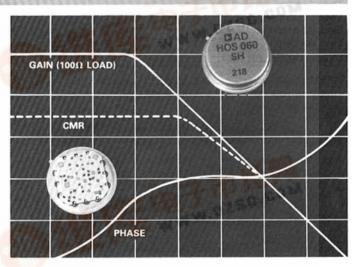
Voltage offset and its temperature coefficient have been dramatically improved in the HOS-060; offset is as low as on most monolithic op amps, despite being a thick-film hybrid.

These improvements, moreover, have been accomplished without any sacrifice in the other parameters which characterize its outstanding performance in video applications.

The HOS-060 op amp is pin-for-pin compatible with its forerunner HOS-050 and is useable in the same diversity of video requirements. The reader is strongly urged to refer to the six-page data sheet for the HOS-050 op amp to obtain additional insight and details on potential uses for the HOS-060.

The HOS-060 Operational Amplifier is housed in an industry standard TO-8 metal can and operates over a case temperature range of -55°C to +125°C; the model number for the standard unit is HOS-060SH.

For units processed to MIL-STD-883, Method 5008, specify model number HOS-060SH/883.



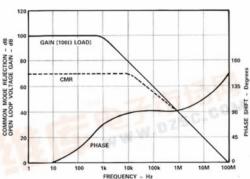


Figure 1. HOS-060 Frequency Response

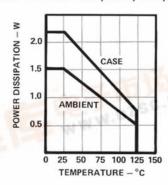


Figure 2. Power Derating

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West Coast Mid-W

Mid-West Texas

# **SPECIFICATIONS** (typical @ $+25^{\circ}$ C and $\pm 15$ V unless otherwise specified)

Model	A					HOS-060SH/883			
ABSOLUTE MAXIMUM RATINGS Supply Voltages (V <sub>S</sub> ) Power Dissipation Input Voltage Differential Input Voltage Operating Temperature Range (case) Junction Temperature Storage Temperature Range Lead Temperature (soldering, 10 sec.)		± 18V See Figure 2 ± V <sub>S</sub> ± V <sub>S</sub> − 55°C to + 125°C 175°C − 65°C to + 150°C 300°C				*  *  *  *  *  *  *  *  *  *  *  *  *			
DC ELECTRICAL CHARACTER	Conditions	Min	Тур	Max	Min	Тур	Max	Units	
Parameter Open Loop Gain Rated Output Current	$R_{L} = 100\Omega$	I I I	100	.vaax		*	.vaux	dB	
(not short circuit protected) Voltage	$\begin{array}{l} R_{L} = > 100\Omega \\ R_{L} = > 200\Omega \end{array}$	± 10	$\pm100$			*		mA V	
Input Offset Voltage Initial  - 25°C to + 125°C vs. Case Temperature  - 55°C to + 125°C	Adjustable to Zero @ + 25°C		± 0.5	±1 ±2			:	mV mV μV/°C	
vs. Power Supply Voltage			0.5			*		mV/V	
Input Bias Current Initial vs. Temperature	@ +25°C		l Double	2 es		*	*	nA /10°C	
Input Offset Current Initial	@ +25°C		± 100			*		pA	
Input Impedance Differential Common Mode	}In parallel with 5pF		10 <sup>10</sup> 10 <sup>10</sup>			*		$\Omega$	
Input Voltage Range Common Mode Differential Common Mode Rejection		± 10	70	± 18 ± 18	*	*	:	V V dB	
Input Noise dc to 100kHz dc to 2MHz	$R_{\mathrm{FF}}=100\Omega; R_{\mathrm{FB}}=1k\Omega$		5 7			*		μV rms μV rms	
AC ELECTRICAL CHARACTER	RISTICS1								
Parameter	Conditions	Min	Тур	Max	Min	Typ	Max	Units	
Slew Rate	$A = -1; R_{\mathrm{FF}} = R_{\mathrm{FB}} = 500\Omega; \\ Load = 100\Omega$		300			*		V/µs	
Noninverting Slew Rate Overload Recovery	$A=2; R_{FF}=R_{FB}=1000\Omega;$ $Load=100\Omega$ 50% Overdrive		320 400			*		$V/\mu s \\ ns$	
Gain Bandwidth Product Small Signal Bandwidth, -3dB	$\begin{split} R_{FF} &= R_{FB} = 500\Omega \\ A &= -1; R_{FF} = R_{FB} = 500\Omega \\ A &= -1; R_{FF} = R_{FB} = 1000\Omega \\ A &= -2; R_{FF} = 500\Omega; \end{split}$		100 45 35			*		MHz MHz MHz	
€a	$R_{FB} = 1000\Omega$ $A = -4; R_{FF} = 250\Omega;$ $R_{FB} = 1000\Omega$		35 30					MHz	
Output Impedance Noninverting Bandwidth, – 3dB	$100\Omega$ load; $10$ pF capacitance 5-volt p-p output 4-volt p-p output 2-volt p-p output A = 3; $R_{FF} = 500\Omega$ ; $R_{FB} = 1000\Omega$ ; load = $100\Omega$ , $1000\Omega$ , or $2000\Omega$ ;		25 30 55	<1		*	*	Ω MHz MHz MHz	
	capacitance = 10pF 10-volt p-p output 5-volt p-p output		17 25			*		MHz MHz	

CELECTRICAL CHARACTERISTICS <sup>1</sup> (Continued)		н	OS-060SI	H .	HOS-060SH/883			
Parameter	Conditions	Min	Тур	Max	Min	Тур	Max	Units
Noninverting Bandwidth, -3dB (continued)	$\begin{split} A &= 5; R_{\mathrm{FF}} = 500\Omega; \\ R_{\mathrm{FB}} &= 2000\Omega; 100\Omega, 1000\Omega, \\ or 2000\Omega load/10pF \end{split}$							
	capacitance							
	5-volt p-p output		15			*		MHz
	4-volt p-p output		30			*		MHz
	2-volt p-p output		40					MHz MHz
	1-volt p-p output		40					MHZ
Full Power Bandwidth	Output = $\pm 5V$ ; A = $-1$ ;		20					MHz
(-3dB)	Load = $100\Omega$		20			100		MHZ
Settling Time to 0.1%	$A = -1; R_{FF} = R_{FB} = 500\Omega$		100					
Inverting	$V_{OUT} = \pm 5V$		100 80			÷		ns
	$V_{OUT} = \pm 2.5V$		80			-		ns
Noninverting	$A = 2$ ; $R_{FF} = R_{FB} = 500\Omega$ Max Load capacitance = 75pF	i		- 1				
	$V_{OUT} = \pm 5V$		200			*		ns
	$V_{OUT} = \pm 3V$ $V_{OUT} = \pm 2.5V$		135			*		ns
Harmonic Distortion	$A = -1$ ; Load = $1000\Omega$	1	133					110
(See Figure 5)	Signal = 4MHz; 2V output		-63			*		dB
Noninverting Harmonic	$A = 2$ ; $R_{FF} = R_{FB} = 1000\Omega$ ;							1000.
Distortion (See Figure 6)	Load = $1000\Omega$ ;							
Distortion (See Figure 0)	Signal = 4MHz; 2V output		- 59			*		dB
	organi marie, 2 v output							
Power Supply	D		± 15			*		V dc
Voltage	Rated performance Operating range	± 12	± 15	± 18	*		*	V dc
Voltage	Quiescent	± 12	± 20	± 25		*	*	mA
Current Power Consumption	Quiescent		0.6	123		*		W
Power Consumption Power Dissipation	Quiescent		0.0	1.25			*	w
CONTRACTOR DESIGNATION OF THE PROPERTY OF THE				1.20				
Temperature Range								90
Operating (Case)	(See Figure 2 for	- 55		+ 125	1		*	°C °C
Storage	Derating Information)	-65		+150	*			
Meantime Between Failures (MTBF)	MIL-HNBK 217; Ground; Fixed; Case = 70°C 883B Processing				2.78 × 10 <sup>6</sup>			Hours
Price:	1-4 100s	•						
HOS-060SH	\$150 \$105							
HOS-060SH/883		PPLICATI	ONS ASS	ISTANCE	CALL	(919) 292	-6427	

HOC MOCH

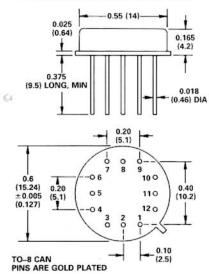
HOE MARTINES

Notes:

Individual socket assemblies (one per pin) are available from AMP as part number 6-330808-0. Specifications subject to change without notice.

## **OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).



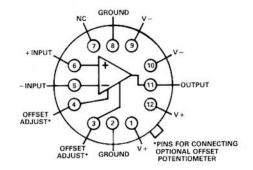
BOTTOM VIEW

## PIN DESIGNATIONS

PINS	FUNCTION
1	+ V
2	GROUND
3	OFFSET ADJ
4	OFFSET ADJ
5	-INPUT
6	+ INPUT
7	NC
8	GROUND
9	-v
10	-v
11	OUTPUT
12	+ V

\*PINS FOR CONNECTING OPTIONAL OFFSET POTENTIOMETER.

## HOS-060 OUTLINE AND PIN DESIGNATIONS



TO-8 PACKAGE BOTTOM VIEW

<sup>&</sup>lt;sup>1</sup>Specification for Inverting Mode unless otherwise noted.

<sup>\*</sup>Specifications same as HOS-060SH.

#### **VOLTAGE AMPLIFIERS/CURRENT BOOSTERS**

Video op amps such as the HOS-060 are generally characterized by high gain bandwidth products, fast settling times, and high output drive.

One of the most common uses of video op amps is for D/A converter output voltage amplification or current boosting. Figure 3 is one example of this type of application. In this circuit, the internal resistance of the D/A is the feed-forward resistor for the op amp.

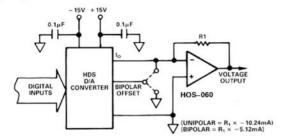


Figure 3. Inverting Unipolar or Bipolar Voltage Output

The circuit which is shown will provide a negative unipolar output with binary coding on the input, and bipolar offset grounded. It will provide a bipolar output with complementary offset binary coding on the input, and bipolar offset connected to I<sub>O</sub>.

### OFFSET AND GAIN ADJUSTMENT

The low value of offset may preclude the need for adjustment, but Figure 4 shows a method of adjusting both offset and gain.

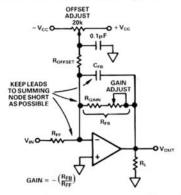


Figure 4. HOS-060 Offset and Gain Adjust

As shown, the gain of the circuit is established by the equation:

$$G = -\left(\frac{R_{FB}}{R_{FF}}\right)$$
 where  $R_{FB} = R_{GAIN} + Gain Adjust.$ 

Once the user has established the desired gain for the illustrated circuit, the value of  $R_{FB}$  can be used to determine the correct value of  $R_{OFFSET}$  with the equation:

$$R_{OFFSET} = \ - \ \left( \frac{V_{CC} \times R_{FB}}{\Delta E_O} \right) \label{eq:roffset}$$

where  $\Delta E_O$  is the desired amount of offset on the output.

Note: R<sub>FF</sub>, R<sub>GAIN</sub>, C<sub>FB</sub> and R<sub>OFFSET</sub> must be located as close to the summing node of the HOS-060 as physically possible. This helps prevent additional capacitance in the summing node and corresponding bad effects on frequency response and settling times.

Variable controls (such as Offset Adjust and Gain Adjust) should never be tied to the summing node of the op amp. Their correct electrical locations are those shown in Figure 4.

#### NONINVERTING OPERATION

The vast majority of video operational amplifiers display marked differences in settling times and bandwidths when operated in a noninverting mode instead of the inverting mode. There are a number of valid reasons for this characteristic.

Most high-speed op amps use feed-forward compensation for optimizing performance in the inverting mode. This is necessary to obtain wide gain-bandwidth products while maintaining dc performance in these types of devices. In effect, the op amp has a wideband ac channel which is not perfectly matched to the dc channel.

Feed-forward techniques enhance the performance of the op amp in the inverting mode by increasing the slew rate and smallsignal bandwidth. These techniques, however, also decrease the amplifier's tolerance to stray capacitances, so must be employed judiciously.

The Analog Devices HOS-060 has different performance characteristics when operating as a noninverting amplifier, but the care used in the design makes the differences less pronounced than they are in many competing units.

The HOS-060 can be considered a true differential video op amp. It requires little or no external compensation because its rolloff characteristics approach a 6dB/octave slope. This helps the user determine summing errors and loop response; and helps assure the stability of the system.

The performance parameters for both inverting and noninverting operation are shown elsewhere in this data sheet (see SPECIFI-CATIONS section and figures). A comparison of the characteristics will highlight the similarities in performance, with the exceptions noted above.

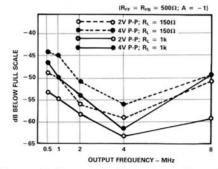


Figure 5. Harmonic Distortion - Inverting

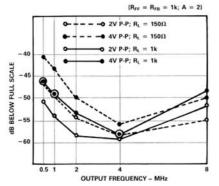


Figure 6. Harmonic Distortion - Noninverting

THE READER IS URGED TO CONSULT THE HOS-050 DATA SHEET FOR ADDITIONAL APPLICATIONS INFORMATION. THE HOS-060 IS PIN-FOR-PIN COMPATIBLE WITH THE HOS-050 SERIES AND CAN BE USED IN SIMILAR WAYS.