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Single-Supply, Rail-to-Rail Output, CMOS INSTRUMENTATION AMPLIFIER

FEATURES

- RAIL-TO-RAIL OUTPUT SWING: Within 10mV
- LOW OFFSET VOLTAGE: ±200μV
- LOW OFFSET DRIFT: ±5μV/°C
- INTERNAL FIXED GAIN = 10V/V OR 50V/V
- SPECIFIED TEMPERATURE RANGE: -55°C to +125°C
- LOW INPUT BIAS CURRENT: 0.2pA
- WIDE BANDWIDTH: 550kHz in G = 10
- HIGH SLEW RATE: 6.5V/µs
- LOW COST
- SO-8 AND TINY MSOP-8 PACKAGES

APPLICATIONS

- INDUSTRIAL SENSOR AMPLIFIERS Bridge, RTD, Thermocouple, Flow, Position
- MEDICAL EQUIPMENT
 ECG, EEG, EMG Amplifiers
- DRIVING A/D CONVERTERS
- PCMCIA CARDS
- AUDIO PROCESSING
- COMMUNICATIONS
- TEST EQUIPMENT
- LOW COST AUTOMOTIVE INSTRUMENTATION

DESCRIPTION

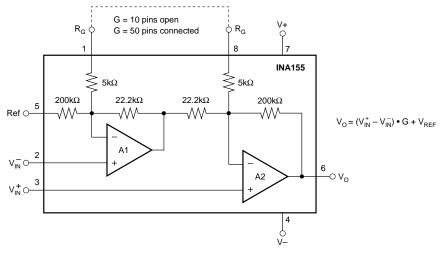
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The INA155 is a low-cost CMOS instrumentation amplifier with rail-to-rail output swing optimized for low voltage, single-supply operation.

Wide bandwidth (550kHz in G=10) and high slew rate (6.5V/µs) make the INA155 suitable for driving sampling A/D converters as well as general purpose and audio applications. Fast settling time allows use with higher speed sensors and transducers and rapid scanning data acquisition systems.

Gain can be set to 10V/V or 50V/V by pin strapping. Gains between these two values can be obtained with the addition of a single resistor. The INA155 is fully specified over the supply range of +2.7 to +5.5V.

The INA155 is available in MSOP-8 and SO-8 surface-mount packages. Both are specified for operation over the temperature range -55°C to 125°C.



International Airport Industrial Park • Mailing Address: PO Box 11400, Tucson, AZ 85734 • Street Address: 6730 S. Tucson Blvd., Tucson, AZ 85766 • Tel: (520) 746-1111

Twx: 910-952-1111 • Internet: http://www.burr-brown.com/ • Cable: BBRCORP • Telex: 066-6491 • FAX: (520) 889-1510 • Immediate Product Info: (800) 548-6132

SPECIFICATIONS: $V_S = +2.7V$ to +5.5V

Boldface limits apply over the specified temperature range, $T_A = -40^{\circ}C$ to $+85^{\circ}C$

At T_A = +25°C, R_L = 10k Ω connected to $V_S/2$. R_G pins open (G = 10), and Ref = $V_S/2$, unless otherwise noted.

PARAMETER					INA155E, U	ı	IN	NA155EA, U	IΔ	
NPUT Work Commendation Votal	DADAMETED		CONDITION			1			1	LINITE
Cliest Voltage, RTI			CONDITION	IVIIIN	ITP	IVIAA	IVIIIV	ITP	IVIAA	UNITS
Over Temperature Drift of V _{CO} /J _c vs Power Supply PSRR of V _S = 2.7V to +6V, V _{CM} = 0.2 • V _S ±15 to ±200 to ± ± ± ± ± ± ± ± ± ± ± ± ± ± ± ± ± ±			V .50V V V /0		100			.*.		\/
Drift Over Temperature O	o .	V _{OS}	$V_S = +5.0V, V_{CM} = V_S/2$		±0.2			*	1	1
Vs Form Supply PSRR Vs ± 2.7V to +6V, V _{CM} = 0.2 • Vs ± 50 ± 2200 ± ± ± µVV V V V V V V V V V	•	ط/ / (ط			15	±1.5			*	1
Over Temperature			\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\			1200		1		
Value Va		PSKK	$V_S = +2.7 \text{ V to } +6 \text{ V}, V_{CM} = 0.2 \cdot V_S$		±50			*		1 '
Sels Input Voltage Rangel	•				10.4	±250			*	
Safe Input Voltage Vo					±0.4			*		μν/πο
Common-Mode Rangel*1										
V ₂ = 2.7V 0.2 0.2 0.2 0.2 0.2 0.3 0.4 0.5	1 0			٠, ,			-		1	1
Common-Mode Rejection Ratio OMRR V ₃ = 5.5V, 0.6V < V _{CM} < 3.7V, 0 = 10 92 100 80 * dB 65 79 dB 65 66 dB 79 76 dB dB 76 dB	Common-Mode Range ⁽¹⁾	V_{CM}					-		1	
Vs 5.5V, 0.6V V _{CM} 3.7V, G 50 85 90 77 7 8 db						2.5(2)	-		*	
Over Temperature	•	CMRR	$V_S = 5.5V, 0.6V < V_{CM} < 3.7V, G = 10$		100			*		
Over Temperature	Over Temperature			85			_			
INPUT IMPEDANCE 1013 3 3			$V_S = 5.5V, 0.6V < V_{CM} < 3.7V, G = 50$	86	90		77	*		dB
Differential	Over Temperature			85			76			dB
Differential	INPUT IMPEDANCE									
CommonMode					10 ¹³ 3			*		ΩllpF
INPUT BIAS CURRENT Input B	Common-Mode							*		
Input Bias Current Input								-		
Offset Current I _{OS}		I.			+1	+10		*	*	nΔ
NOISE, RT Voltage Noise: f = 0.1Hz to 10Hz	•									
Voltage Noise 1 - 0.1 Hz Voltage Noise 1 - 0.1 Hz		ios	D 00 0 10 50		±1	110		*	*	PA
Voltage Noise Density, f = 10Hz	,		$R_S = 0.02, G = 10 \text{ or } 50$		4.5					\/\/n n
F = 100Hz	•									
Current Noise: f = 1kHz	•									1
Current Noise: f = 1kHz SAN FAV F										
Gain Equation Gain Emort										
Gain Equation Vs = 5.5V, Vo = 0.01V to 5.49V, G = 10					2			*		
Sain Error® V _S = 5.5V, V _O = 0.01V to 5.49V, G = 10							*		*	
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$				G = 1		Ų.				
vs Temperature V _S = 5.5V, V _O = 0.05V to 5.45V, G = 50 ±0.05 ±0.25 * * * ppm"C Nonlinearity V _S = 5.5V, G = 10 or 50 ±0.005 ±0.005 ±0.015 * * * ppm"C Our Temperature V _S = 5.5V, G = 10 or 50 ±0.005 ±0.005 ±0.015 * * % of FSR OUTPUT Voltage Output Swing from Rail R _L = 10kΩ, G _{ERR} < 0.1%			$V_S = 5.5V, V_O = 0.01V \text{ to } 5.49V, G = 10$			-				1
Stemperature Vs = 5.5V, G = 10 or 50	vs remperature		V 55V V 005V 545V 0 50		1	_				1 '''
Nonlinearity V _S = 5.5V, G = 10 or 50	-		$V_S = 5.5V, V_O = 0.05V \text{ to } 5.45V, G = 50$							1
Over Temperature ±0.015 * * % of FSF OUTPUT Voltage Output Swing from Rail R _L = 10kΩ, G _{ERR} < 0.1%	vs remperature				±15	±30				1 '''
Output Voltage Output Swing from Rail Over Temperature Short Circuit to Ground ±50 x mV	*		$V_S = 5.5V, G = 10 \text{ or } 50$		±0.005			*		% of FSR
Voltage Output Swing from Rail Over Temperature Short Circuit to Ground See Typical Curve See Typical Curve Short Circuit to Ground See Typical Curve See Typical	Over Temperature					±0.015			*	% of FSR
Short Circuit to Ground Short Circuit to Ground See Typical Curve Short Circuit to Ground See Typical Curve See T	OUTPUT									
Short-Circuit Current Capacitance Load (stable operation) Short Circuit to Ground See Typical Curve * MA			$R_L = 10k\Omega, G_{ERR} < 0.1\%$		5			*	*	mV
Capacitance Load (stable operation) See Typical Curve * FREQUENCY RESPONSE Bandwidth, -3dB BW G = 10 550 * kHz Slew Rate SR V _S = 5.5V, V _C = 100pF, G = 10 6.5 * V/µs Settling Time: 0.1% t _S V _S = 5.5V, V _O = 2V Step, C _L = 100pF, G = 50 11 * µs 0.01% V _S = 5.5V, V _O = 2V Step, C _L = 100pF, G = 50 11 * µs 0.01% V _S = 5.5V, V _O = 2V Step, C _L = 100pF, G = 50 11 * µs 0.01% V _S = 5.5V, V _O = 2V Step, C _L = 100pF, G = 50 15 * µs 0.01% V _S = 5.5V, V _O = 2V Step, C _L = 100pF, G = 50 15 * µs Overload Recovery See Typical Curve * µs Total Harmonic Distortion + Noise THD+N * * V Specified Voltage Range V _{IN} = 0, I _O = 0 1.7 2.1 * * V Quiescent Current V _{IN} = 0, I _O = 0 1.7 2.1	Over Temperature					10			*	mV
FREQUENCY RESPONSE Bandwidth, -3dB BW G = 10 550 110	Short-Circuit Current		Short Circuit to Ground					*		mA
Bandwidth, -3dB BW G = 10 G = 50 110	Capacitance Load (stable operation)			Se	e Typical Cu	irve		*		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	FREQUENCY RESPONSE									
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Bandwidth, -3dB	BW	G = 10		550			*		kHz
Settling Time: 0.1%					110			*		kHz
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Slew Rate				6.5			*		V/μs
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Settling Time: 0.1%	ts	$V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100pF, G = 10$		5			*		μs
Overload Recovery $V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100 \text{pF, } G = 50 \\ 50\% \text{ Input Overload} \\ \hline \text{Total Harmonic Distortion + Noise} \\ \hline \text{THD+N} \\ \hline \text{POWER SUPPLY} \\ \hline \text{Specified Voltage Range} \\ \hline \text{Operating Voltage Range} \\ \hline \text{Over Temperature} \\ \hline \text{TEMPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{Over Storage Range} \\ \hline \text{Over Temperature} \\ \hline \text{TEMPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{Over Temperature} \\ \hline \text{Operating Range} \\ \hline \text{OPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{OPERATURE RANGE} \\ \hline \text{OPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{OPERATURE RANGE} \\ \hline$			$V_S = 5.5V, V_O = 2V \text{ Step}, C_L = 100pF, G = 50$		11			*		μs
Overload Recovery $V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100 \text{pF, } G = 50 \\ 50\% \text{ Input Overload} \\ \hline \text{Total Harmonic Distortion + Noise} \\ \hline \text{THD+N} \\ \hline \text{POWER SUPPLY} \\ \hline \text{Specified Voltage Range} \\ \hline \text{Operating Voltage Range} \\ \hline \text{Over Temperature} \\ \hline \text{TEMPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{Over Storage Range} \\ \hline \text{Over Temperature} \\ \hline \text{TEMPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{Over Temperature} \\ \hline \text{Operating Range} \\ \hline \text{OPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{OPERATURE RANGE} \\ \hline \text{OPERATURE RANGE} \\ \hline \text{Specified Range} \\ \hline \text{OPERATURE RANGE} \\ \hline$	0.01%		$V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100pF, G = 10$					*		μs
Overload Recovery			$V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100pF, G = 50$		15			*		μs
POWER SUPPLY Specified Voltage Range +2.7 +5.5 * * V Operating Voltage Range V _{IN} = 0, I _O = 0 +2.5 to +6 * * V Quiescent Current V _{IN} = 0, I _O = 0 1.7 2.1 * * mA Over Temperature V _{IN} = 0, I _O = 0 2.6 * * mA TEMPERATURE RANGE Specified Range -40 +85 * * °C Operating Range -65 +150 * * °C Storage Range -65 +150 * * °C Thermal Resistance θ _{JA} MSOP-8 Surface Mount * °C/W	Overload Recovery							*		μs
Specified Voltage Range	Total Harmonic Distortion + Noise	THD+N		Se	e Typical Cu	ırve		*		
Operating Voltage Range V _{IN} = 0, I _O = 0 +2.5 to +6 * * V Quiescent Current V _{IN} = 0, I _O = 0 1.7 2.1 * * mA Over Temperature V _{IN} = 0, I _O = 0 2.6 * mA mA TEMPERATURE RANGE Specified Range -40 +85 * * °C Operating Range -65 +150 * * °C Storage Range -65 +150 * * °C Thermal Resistance θ _{JA} MSOP-8 Surface Mount * °C/W	POWER SUPPLY									
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Specified Voltage Range			+2.7		+5.5	*		*	V
Quiescent Current Over Temperature $V_{\rm IN} = 0$, $I_{\rm O} = 0$ 1.7 2.1 * * mA mA TEMPERATURE RANGE Specified Range -40 +85 * * °C Operating Range -65 +150 * * °C Storage Range -65 +150 * * °C Thermal Resistance $\theta_{\rm JA}$ * °C * °C MSOP-8 Surface Mount 150 * * °C/W					+2.5 to +6			*		1
Over Temperature V _{IN} = 0, I _O = 0 2.6 * mA TEMPERATURE RANGE Specified Range -40 +85 * * °C Operating Range -65 +150 * * °C Storage Range -65 +150 * * °C Thermal Resistance θ _{JA} * °C/W			$V_{IN} = 0, I_{O} = 0$			2.1			*	mA
TEMPERATURE RANGE Specified Range -40 +85 * * °C Operating Range -65 +150 * * °C Storage Range -65 +150 * * °C Thermal Resistance θ _{JA} * °C/W										1
	•									
Operating Range $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$				-40		+85	*		*	°C
Storage Range $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$										1
Thermal Resistance θ_{JA} 150 * °C/W				1						1
MSOP-8 Surface Mount * °C/W		θ_{1A}							.	
		VJA			150			*		°C/W
OO-o outlace iviount I I I I I I I 本 I I I I I 本 I I I I I	SO-8 Surface Mount				150			*		°C/W

^{*} Same as INA155E, U.

NOTES: (1) For further information, refer to typical performance curves on common-mode input range. (2) Operation above (V+) - 1.8V (max) results in reduced common-mode rejection. See discussion and Figure 6 in the text of this data sheet. (3) Does not include error and TCR of additional optional gain-setting resistor in series with R_g , if used.



SPECIFICATIONS: $V_S = +2.7V$ to +5.5V

Boldface limits apply over the specified temperature range, $T_A = -55^{\circ}C$ to $+125^{\circ}C$

At T_A = +25°C, R_L = 10k Ω connected to $V_S/2$. R_G pins open (G = 10), and Ref = $V_S/2$, unless otherwise noted.

				NA155E, L	J	II.] !		
PARAMETER		CONDITION	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
INPUT									1
Offset Voltage, RTI	Vos	$V_S = +5.0V, V_{CM} = V_S/2$		±0.2	±1		*	*	mV
Over Temperature	00	, om 6			±2			*	mV
Drift	dV_{OS}/d_{T}			±5			*		μV/°C
vs Power Supply	PSRR	$V_S = +2.7V \text{ to } +6V, V_{CM} = 0.2 \cdot V_S$		±50	±200		*	*	μV/V
Over Temperature	-	S S S S S S S S S S S S S S S S S S S			±250			*	μV/V
vs Time				±0.4			*		μV/mo
INPUT VOLTAGE RANGE									<u> </u>
Safe Input Voltage			(V-) - 0.5		(V+) + 0.5	*		*	V
Common-Mode Range ⁽¹⁾	V_{CM}	V _S = 5.5V	0.3		5.2 ⁽²⁾	*		*	V
Common-wode Range	v CM	$V_{S} = 3.5V$ $V_{S} = 2.7V$	0.3		2.5 ⁽²⁾	*		*	V
Common-Mode Rejection Ratio	CMDD	$V_S = 2.7 \text{ V}$ $V_S = 5.5 \text{V}, 0.6 \text{V} < V_{CM} < 3.7 \text{V}, G = 10$	92	100	2.3\	80	*	~	dB
Over Temperature	CIVILLIX	V _S = 3.3V, 0.0V < V _{CM} < 3.7V, G = 10	82 82	100		78	~		dB
Over remperature		$V_S = 5.5V$, $0.6V < V_{CM} < 3.7V$, $G = 50$	86	90		7 7	*		dB
• • • • • • • • • • • • • • • • • • •		$v_S = 5.5 \text{ V}, 0.6 \text{ V} < v_{CM} < 5.7 \text{ V}, G = 50$		90			*		1
Over Temperature			84			76			dB
INPUT IMPEDANCE				40			_		
Differential				10 ¹³ 3			*		Ω pF
Common-Mode				1013 3			*		Ω pF
INPUT BIAS CURRENT									
Input Bias Current	I _B			±1	±10		*	*	pA
Offset Current	Ios			±1	±10		*	*	pА
NOISE, RTI		$R_S = 0\Omega$, $G = 10 \text{ or } 50$							
Voltage Noise: f = 0.1Hz to 10Hz				4.5			*		μV/Vp-p
Voltage Noise Density: f = 10Hz				260			*		nV/√ Hz
f = 100Hz				99			*		nV/√ Hz
f = 1kHz				40			*		nV/√Hz
Current Noise: f = 1kHz				2			*		fA/√Hz
GAIN			10		50	*		*	V/V
Gain Equation			G = 10	0 + 400kΩ/(1	10kΩ + R _G)		*		V/V
Gain Error ⁽³⁾		$V_S = 5.5V$, $V_O = 0.01V$ to 5.49V, $G = 10$		±0.02	±0.1		*	*	%
vs Temperature				<u>+2</u>	±10		*	*	ppm/°C
·		$V_S = 5.5V, V_O = 0.05V \text{ to } 5.45V, G = 50$		±0.05	±0.25		*	*	%
vs Temperature				±15	±30		*	*	ppm/°C
Nonlinearity		V _S = 5.5V, G = 10 or 50		±0.005	±0.015		*	*	% of FSR
Over Temperature		3 ,			±0.015			*	% of FSR
OUTPUT									
Voltage Output Swing from Rail		$R_L = 10k\Omega, G_{ERR} < 0.1\%$		5	10		*	*	mV
Over Temperature		L - / - Link			10			*	mV
Short-Circuit Current		Short Circuit to Ground		±50			*	_	mA
Capacitance Load (stable operation))		Se	e Typical Cu	irve		*		
FREQUENCY RESPONSE									
Bandwidth, –3dB	BW	G = 10		550			*		kHz
, 	2	G = 50		110			*		kHz
Slew Rate	SR	$V_S = 5.5V, C_L = 100pF$		6.5			*		V/µs
Settling Time: 0.1%	t _s	$V_S = 5.5V$, $V_O = 2V$ Step, $C_L = 100$ pF, $G = 10$		5			*		μς
g	-3	$V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100pF, G = 50$		11			*		μs
0.01%		$V_S = 5.5V$, $V_O = 2V$ Step, $C_L = 100pF$, $G = 10$		8			*		μs
		$V_S = 5.5V, V_O = 2V \text{ Step, } C_L = 100pF, G = 50$		15			*		μs
Overload Recovery		50% Input Overload		0.2			*		μs
Total Harmonic Distortion + Noise	THD+N		Se	e Typical Cu	irve		*		
POWER SUPPLY									
Specified Voltage Range			+2.7		+5.5	*		*	V
Operating Voltage Range				+2.5 to +6			*		V
Quiescent Current		$V_{IN} = 0, I_{O} = 0$		1.7	2.1		*	*	mA
Over Temperature		$V_{IN} = 0$, $I_O = 0$			2.8			*	mA
TEMPERATURE RANGE							+		
Specified Range			-55		+125	*		*	∘c
Operating Range			-65		+150	*		*	.c
			-65		+150	*		*	-€
Storage Range					, I	-4"	1		
Storage Range Thermal Resistance	θ_{ij}								
Thermal Resistance	$ heta_{\sf JA}$			150			*		°C/W
	$ heta_{\sf JA}$			150 150			*		°C/W

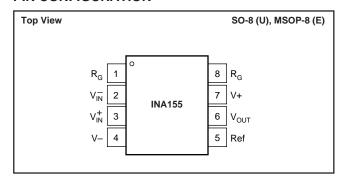
^{*} Same as INA155E, U.

NOTES: (1) For further information, refer to typical performance curves on common-mode input range. (2) Operation above (V+) - 1.8V (max) results in reduced common-mode rejection. See discussion and Figure 6 in the text of this data sheet. (3) Does not include error and TCR of additional optional gain-setting resistor in series with R_G , if used.

3



PIN CONFIGURATION



ABSOLUTE MAXIMUM RATINGS(1)

Supply Voltage, V+ to VSignal Input Terminals, Voltage ⁽²⁾	
	(v–) – 0.3v to (v+) + 0.3v
Output Short-Circuit ⁽³⁾	
Operating Temperature	65°C to +150°C
Storage Temperature	65°C to +150°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C

NOTE: (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied. (2) Input terminals are diode-clamped to the power supply rails. Input signals that can swing more that 0.5V beyond the supply rails should be current limited to 10mA or less. (3) Short circuit to ground.

ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PACKAGE/ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER ⁽¹⁾	TRANSPORT MEDIA
INA155U	SO-8	182	-55°C to +125°C	INA155U	INA155U	Rails
"	n n	"	"	"	INA155U/2K5	Tape and Reel
INA155UA	SO-8	182	-55°C to +125°C	INA155UA	INA155UA	Rails
"	n n	"	"	"	INA155UA/2K5	Tape and Reel
INA155E	MSOP-8	337	-55°C to +125°C	A55	INA155E/250	Tape and Reel
"	п	"	"	"	INA155E/2K5	Tape and Reel
INA155EA	MSOP-8	337	-55°C to +125°C	A55	INA155EA/250	Tape and Reel
"	"	"	"	"	INA155EA/2K5	Tape and Reel

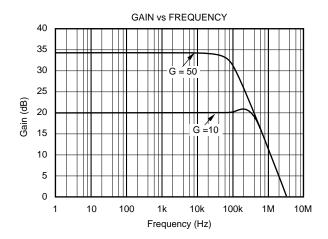
NOTES: (1) Models with a slash (/) are available only in Tape and Reel in the quantities indicated (e.g., /2K5 indicates 2500 devices per reel). Ordering 2500 pieces of "INA155UA/2K5" will get a single 2500-piece Tape and Reel.

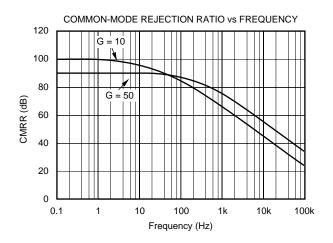
The information provided herein is believed to be reliable; however, BURR-BROWN assumes no responsibility for inaccuracies or omissions. BURR-BROWN assumes no responsibility for the use of this information, and all use of such information shall be entirely at the user's own risk. Prices and specifications are subject to change without notice. No patent rights or licenses to any of the circuits described herein are implied or granted to any third party. BURR-BROWN does not authorize or warrant any BURR-BROWN product for use in life support devices and/or systems.

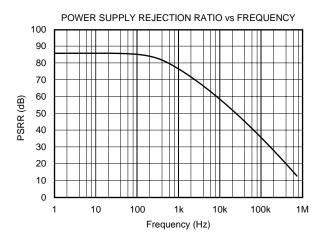


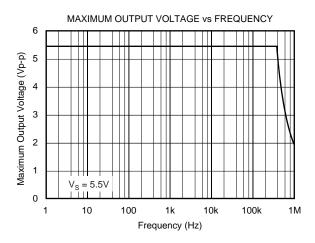
TYPICAL PERFORMANCE CURVES

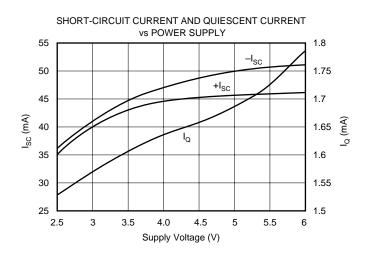
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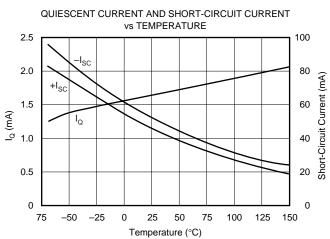






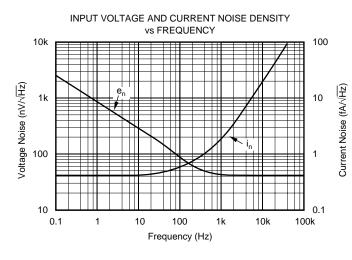


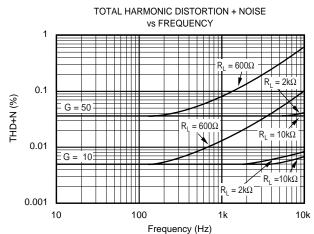


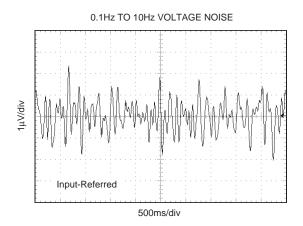


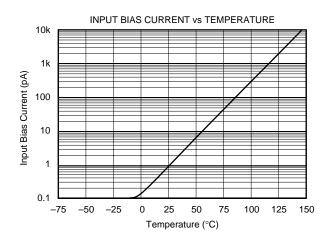
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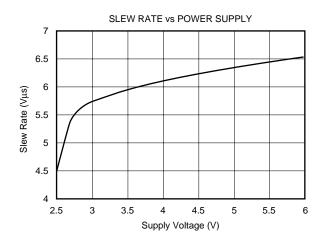
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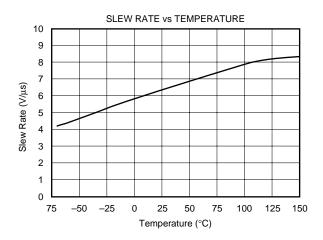








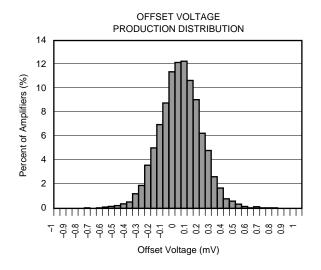


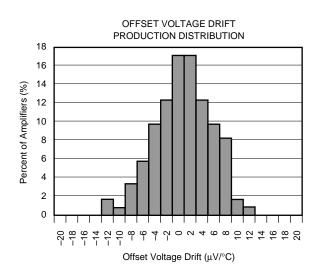


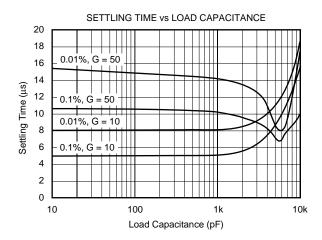


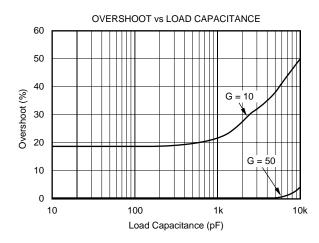
TYPICAL PERFORMANCE CURVES (Cont.)

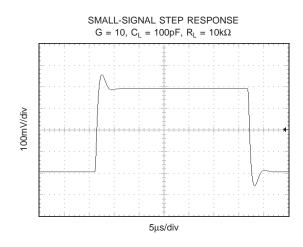
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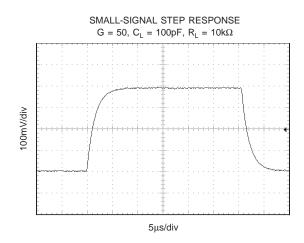






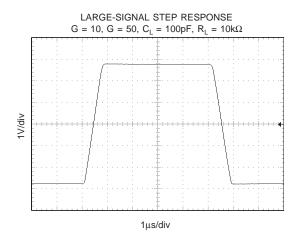


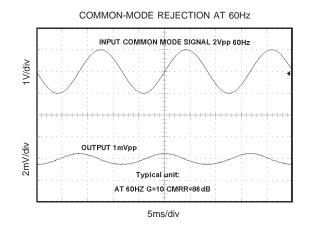


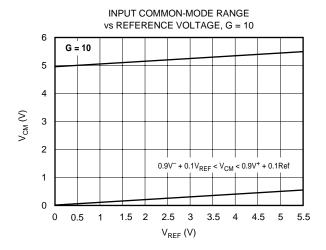


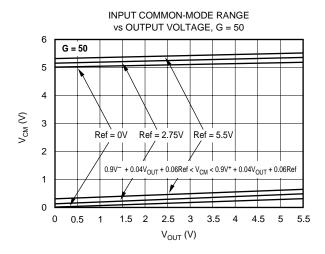
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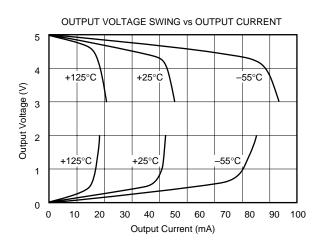
At T_A = +25°C, V_S = 5.5V, R_L = 10k Ω connected to $V_S/2$. R_G pins open (G = 10), and Ref = $V_S/2$, unless otherwise noted.











APPLICATIONS INFORMATION

Figure 1 shows the basic connections required for operation of the INA155. Applications with noisy or high impedance power supplies may require decoupling capacitors close to the device pins as shown.

The output is referred to the output reference terminal, Ref, which is normally set to $V_S/2$. This must be a low-impedance connection to ensure good common-mode rejection. A resistance of 200Ω in series with the Ref pin will cause a typical device to degrade to approximately 80dB CMRR.

In addition, for the G=50 configuration, the connection between pins 1 and 8 must be low-impedance. A connection impedance of 20Ω can cause a 0.2% shift in gain error.

OPERATING VOLTAGE

The INA155 is fully specified and guaranteed over the supply range +2.7V to +5.5V, with key parameters guaranteed over the temperature range of -55°C to +125°C. Parameters that vary significantly with operating voltages, load conditions or temperature are shown in the Typical Performance Curves.

The INA155 can be operated from either single or dual power supplies. By adjusting the voltage applied to the reference terminal, the input common-mode voltage range and the output range can be adjusted within the bounds shown in the Typical Performance Curves. Figure 2 shows a bridge amplifier circuit operated from a single +5V power supply. The bridge provides a relatively small differential voltage on top of an input common-mode voltage near 2.5V.

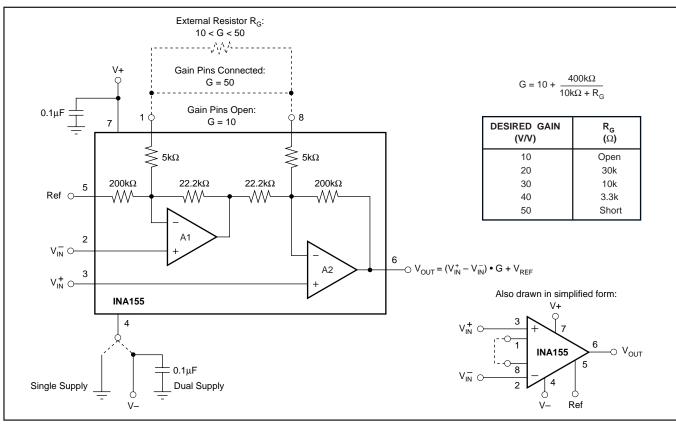
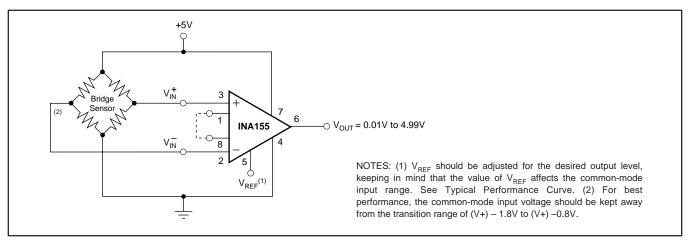


FIGURE 1. Basic Connections.



9

FIGURE 2. Single-Supply Bridge Amplifier.

SETTING THE GAIN

Gain of 10 is achieved simply by leaving the two gain pins (1 and 8) open. Gain of 50 is achieved by connecting the gain pins together directly. In the G=10 configuration, the gain error is less than 0.1%. In the G=50 configuration, the gain error is less than 0.25%.

Gain can be set to any value between 10 and 50 by connecting a resistor R_G between the gain pins according to the following equation:

$$10 + 400k\Omega/(10k\Omega + R_G) \tag{1}$$

This is demonstrated in Figure 1 and is shown with the commonly used gains and resistor R_G values. However, because the absolute value of internal resistors is not guaranteed, using the INA155 in this configuration will increase the gain error and gain error drift with temperature, as shown in Figure 3.

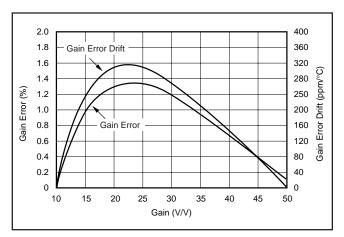


FIGURE 3. Typical Gain Error and Gain Error Drift with External Resistor.

OFFSET TRIMMING

The INA155 is laser trimmed for low offset voltage. In most applications, no external offset adjustment is required. However, if necessary, the offset can be adjusted by applying a correction voltage to the reference terminal. Figure 4 shows

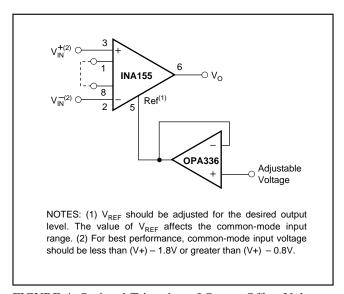


FIGURE 4. Optional Trimming of Output Offset Voltage.



an optional circuit for trimming the output offset voltage. The voltage applied to the Ref terminal is added to the output signal. An op amp buffer is used to provide low impedance at the Ref terminal to preserve good common-mode rejection.

INPUT BIAS CURRENT RETURN

The input impedance of the INA155 is extremely high—approximately $10^{13}\Omega$, making it ideal for use with high-impedance sources. However, a path must be provided for the input bias current of both inputs. This input bias current is less than 10pA and is virtually independent of the input voltage.

Input circuitry must provide a path for this input bias current for proper operation. Figure 5 shows various provisions for an input bias current path. Without a bias current path, the inputs will float to a potential that exceeds the commonmode range and the input amplifier will saturate.

If the differential source resistance is low, the bias current return path can be connected to one input (see the thermocouple in Figure 5). With higher source impedance, using two equal resistors provides a balanced input with advantages of lower input offset voltage due to bias current and better high-frequency common-mode rejection.

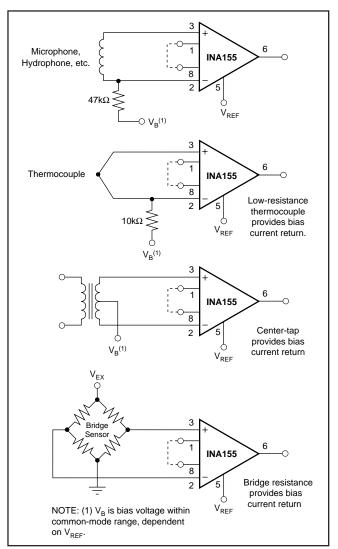


FIGURE 5. Providing an Input Common-Mode Current Path.

INPUT COMMON-MODE RANGE

The input common-mode range of the INA155 for various operating conditions is shown the in Typical Performance Curves. The common-mode input range is limited by the output voltage swing of A1, an internal circuit node. For the G = 10 configuration, output voltage of A1 can be expressed as:

$$V_{OUTA1} = -\frac{1}{9}V_{REF} + (1 + \frac{1}{9})V_{IN}$$
 (2)

Using this equation given that the output of A1 can swing to within 10mV of either rail, the input common-mode voltage range can be calculated. When the input common-mode range is exceeded (A1's output is saturated), A2 can still be in linear operation and respond to changes in the non-inverting input voltage. However, the output voltage will be invalid.

The common-mode range for the G=50 configuration is included in the Typical Performance Curve, "Input Common-Mode Range vs Output Voltage."

INPUT RANGE FOR BEST ACCURACY

The internal amplifiers have rail-to-rail input stages, achieved by using complementary n- and p-channel input pairs. The common-mode input voltage determines whether the p-channel or the n-channel input stage is operating. The transition between the input stages is gradual and occurs between (V+) - 1.8V to (V+) - 0.8V. Due to these characteristics operating the INA155 with input voltages within the transition region of (V+) - 1.8V to (V+) - 0.8V results in a shift in input offset voltage and reduced common-mode and power supply rejection performance. Typical patterns of the offset voltage change throughout the input common-mode range are illustrated in Figure 6. The INA155 can be operated below or above the transition region with excellent results. Figure 7 demonstrates the use of the INA155 in a single-supply, high-side current monitor. In this application, the INA155 is operated above the transition region.

RAIL-TO-RAIL OUTPUT

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. For resistive loads greater than $10k\Omega$, the output voltage can swing to within a few millivolts of the supply rail while maintaining low gain error. For heavier loads and over temperature, see the typical performance curve "Output Voltage Swing vs Output Current." The INA155's low output impedance at high frequencies makes it suitable for directly driving Capacitive Digital-to-Analog (CDAC) input A/D converters, as shown in Figure 9.

INPUT PROTECTION

Device inputs are protected by ESD diodes that will conduct if the input voltages exceed the power supplies by more than 500mV. Momentary voltages greater than 500mV beyond the power supply can be tolerated if the current on the input pins is limited to 10mA. This is easily accomplished with input resistors R_{LIM} as shown in Figure 8. Many input signals are inherently current-limited to less than 10mA, therefore, a limiting resistor is not required.

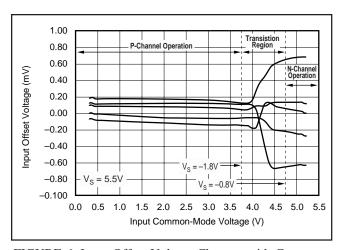


FIGURE 6. Input Offset Voltage Changes with Common-Mode Voltage.

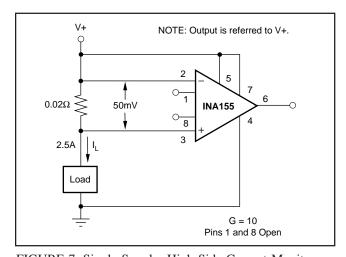


FIGURE 7. Single-Supply, High-Side Current Monitor.

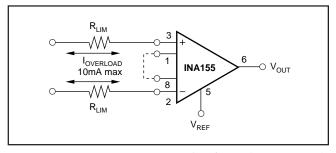


FIGURE 8. Input Current Protection for Voltages Exceeding the Supply Voltage.

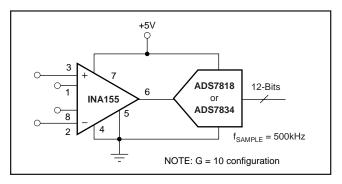


FIGURE 9. INA155 Directly Drives Capacitive-Input, High-Speed A/D Converter.







ti.com 18-Jul-2006

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Packag Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
INA155E/250	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155E/250G4	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155E/2K5	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155E/2K5G4	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155EA/250	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155EA/250G4	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155EA/2K5	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155EA/2K5G4	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155U	ACTIVE	SOIC	D	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155U/2K5	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155U/2K5G4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155UA	ACTIVE	SOIC	D	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155UA/2K5	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155UA/2K5G4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155UAG4	ACTIVE	SOIC	D	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
INA155UG4	ACTIVE	SOIC	D	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.



PACKAGE OPTION ADDENDUM

18-Jul-2006

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





Α	0	Dimension designed to accommodate the component width
В	0	Dimension designed to accommodate the component length
		Dimension designed to accommodate the component thickness
٧	٧	Overall width of the carrier tape
ГР	1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
INA155E/250	MSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
INA155E/2K5	MSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
INA155EA/250	MSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
INA155EA/2K5	MSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
INA155U/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
INA155UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1





*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
INA155E/250	MSOP	DGK	8	250	184.0	184.0	50.0
INA155E/2K5	MSOP	DGK	8	2500	346.0	346.0	29.0
INA155EA/250	MSOP	DGK	8	250	184.0	184.0	50.0
INA155EA/2K5	MSOP	DGK	8	2500	346.0	346.0	29.0
INA155U/2K5	SOIC	D	8	2500	346.0	346.0	29.0
INA155UA/2K5	SOIC	D	8	2500	346.0	346.0	29.0

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