



DUAL, ISOLATED, BIDIRECTIONAL DIGITAL COUPLER

FEATURES

- Replaces High-Performance Optocouplers
- Data Rate: 80 M Baud, Typ
- Low Power Consumption: 25 mW Per Channel, Max
- Two Channels, Each Bidirectional, Programmable by User
- Partial Discharge Tested: 2400 Vrms
- Creepage Distance of 7,2 mm
- Low Cost Per Channel
- Available in SO Package
- UL 1577 Certified

APPLICATIONS

- Digital Isolation for A/D, D/A Conversion
- Isolated RS-485 Interface
- Multiplexed Data Transmission
- Isolated Parallel to Serial Interface
- Test Equipment
- Microprocessor System Interface
- Isolated Line Receiver
- Ground Loop Elimination

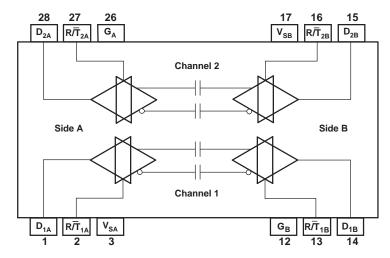
DESCRIPTION

The ISO150 is a 2-channel, galvanically-isolated data coupler capable of data rates of 80M Baud, typical. Each channel can be individually programmed to transmit data in either direction.

Data is transmitted across the isolation barrier by coupling complementary pulses through high voltage 0.4 pF capacitors. Receiver circuitry restores the pulses to standard logic levels. Differential signal transmission rejects isolation-mode voltage transients up to $1.6~\rm kV/\mu s$

The ISO150 avoids problems commonly associated with optocouplers. Optically isolated couplers require high current pulses and allowance must be made for LED aging. The ISO150's Bi-CMOS circuitry operates at 25 mW per channel.

The ISO150 is available in an SO-28 and is specified for operation from -40°C to 85°C.





Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.





This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ORDERING INFORMATION⁽¹⁾

PRODUCT	PACKAGE LEAD	PACKAGE DESIGNATOR	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER	TRANSPORT MEDIA, QUANTITY
ISO150AU	SO 20	DVB	−40°C to 85°C	ISO150AU	ISO150AU	Rails, 28
130 130AU	SO-28 DVB		-40 C 10 65 C	130 130AU	ISO150AU/1K	Tape and Reel, 1000

⁽¹⁾ For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) (1) (2)

		UNIT
	Storage temperature	-40°C to 125°C
Vs	Supply voltage	-0.5 V to 6 V
V_{I}	Transmitter input voltage	-0.5 V to V _S + 0.5 V
Vo	Receiver output voltage	-0.5 V to V _S + 0.5 V
	R/\overline{T}_{x} inputs	-0.5 V to V _S + 0.5 V
V_{ISO}	Isolation voltage dV/dt	500 kV/µs
D _x	Short to ground	Continuous
T_{J}	Junction temperature	175°C
	Lead temperature (soldering, 10s)	260°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute maximum rated conditions for extended periods may affect device reliability.

REGULATORY INFORMATION

UL
Recognized under 1577 Component Recognition Program ⁽¹⁾
File Number: E181974

(1) Production tested at 2400 VRMS for 1 second in accordance with UL 1577.

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⁽²⁾ This isolator is suitable for basic insulation applications within the safety limiting data. Maintenance of the safety data must be ensured by means of protective circuitry.



ELECTRICAL CHARACTERISTICS

At $T_A = 25^{\circ}C$ and $V_S = 5 \text{ V}$ (unless otherwise noted)

PARAM	IETED	TEST CONDITIONS	IS	ISO150AU			
FARAIV	IETEK	TEST CONDITIONS	MIN	TYP	TYP MAX		
ISOLATION PARAMETERS		·					
Rated Voltage, Continuous		60 Hz	1500			Vrms	
Partial Discharge, 100% Test ⁽¹⁾)	1s, 5pC	2400			Vrms	
Creepage distance (external)	SO-U Package			7.2		mm	
Internal isolation distance				0.10		mm	
Isolation voltage transient immu	unity ⁽²⁾			1.6		kV/μs	
Barrier impedance			:	>10 ¹⁴ 7		Ω pF	
Leakage current		240 Vrms, 50 Hz		0.6		μArms	
DC PARAMETERS							
Lania autout valtana	HIGH, V _{OH}	I _{OH} = 6 mA	V _S -1		Vs	V	
Logic output voltage	LOW, V _{OL}	I _{OL} = 6 mA	0		0.4		
Logic output short-circuit currer	nt	Source or sink		30		mA	
Lania innutualtana	HIGH ⁽³⁾		2		Vs	- V	
Logic input voltage	LOW ⁽³⁾		0		0.8		
Logic input capacitance				5		pF	
Logic input current				<1		nA	
Power-supply voltage range (3)			3	5	5.5	V	
	Tue :: :: :: :: :: :: :: :: :: :: :: :: ::	DC		0.001	100	μA	
Davies averali, average (4)	Transmit mode	50M Baud		14		mA	
Power-supply current ⁽⁴⁾	D	DC		7.2	10	mA	
	Receive mode	50M Baud		16			
AC PARAMETERS		·					
Maximum ⁽⁵⁾		C _L = 50 pF	50	80		M Baud	
Data rate	Minimum	DC					
Propagation time ⁽⁶⁾		C _L = 50 pF		27	40	ns	
Propagation delay skew ⁽⁷⁾		C _L = 50 pF		0.5	2	ns	
Pulse width distortion ⁽⁸⁾		C _L = 50 pF		1.5	6	ns	
Output rise-and-fall time, 10% t	o 90%	C _L = 50 pF		9	14	ns	
Mada suitab tima	Receive to Transmit			13		ns	
Mode switch time Transmit to receive (9)				75		ns	
TEMPERATURE RANGE	•		<u> </u>		1		
Operating range			-40		85	°C	
Storage			-40		125	°C	
Thermal resistance, θ _{JA}				75		°C/W	

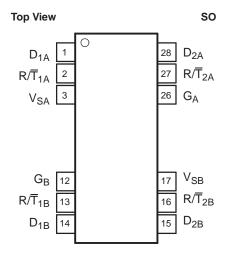
- (1) All devices receive a 1s test. Failure criterion is ≥ 5 PULSES OF ≥5 Pc.
- (2) The voltage rate-of-change across the isolation barrier that can be sustained without data errors.
- (3) Logic inputs are HCT-type and thresholds are a function of power-supply voltage with approximately 0.4 V hysteresis see text.
- (4) Supply current measured with both transceivers set for the indicated mode. Supply current varies with data rate see typical characteristics.
- (5) Calculated from the maximum pulse width distortion (PWD), where Data Rate = 0.3/PWD.
- (6) Propagation time measured from $V_{IN} = 1.5 \text{ V}$ to $V_{O} = 2.5 \text{ V}$.
- (7) The difference in propagation time of channel A and channel B in any combination of transmission directions.
- 8) The difference between propagation time of a rising edge and a falling edge.
- (9) When the device is powered up or direction is changed, the transceiver output is indeterminate (either high or low) and cannot be known until an input signal is applied. The output begins to track the input as soon as the input receives a change in logic state, either low to high or high to low.

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PIN CONFIGURATION



TERMINAL FUNCTIONS

TERM	IINAL	DESCRIPTION						
NAME	NO.	DESCRIPTION						
D _{1A}	1	Data in or data out for transceiver 1A, R/\overline{T}_{1A} held low makes D_{1A} and input pin.						
R/\overline{T}_{1A}	2	Receive/transmit switch controlling transceiver 1A.						
V_{SA}	3	+5V supply pin for side A, which powers transceivers 1A and 2A.						
G _B	12	Ground pin for transceivers 1B and 2B.						
R/\overline{T}_{1B}	13	Receive/transmit switch controlling transceiver 1B.						
D _{1B}	14	Data in or data out for transceiver 1B. R/T _{1B} held LOW makes D _{1B} an input pin.						
D_{2B}	15	Data in or data out for transceiver 2B. R/T _{2B} held LOW makes D _{2B} an input pin.						
R/\overline{T}_{2B}	16	Receive/transmit switch controlling D _{2B} .						
V_{SB}	17	+5V supply pin for side B, which powers transceivers 1B and 2B.						
G_A	26	Ground pin for transceivers 1A and 2A.						
R/\overline{T}_{2A}	27	Receive/transmit switch controlling transceiver 2A.						
D_{2A}	28	Data in or data out for transceiver 2A, R/T _{21A} held low makes D _{2A} an input pin.						

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TYPICAL CHARACTERISTICS

At $T_A = 25^{\circ}C$ and $V_S = 5$ V, unless otherwise noted.

SUPPLY CURRENT PER CHANNEL vs SUPPLY VOLTAGE

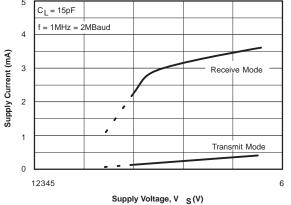


Figure 1.

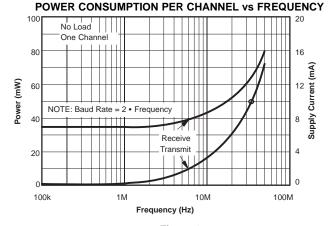


Figure 2.

SUPPLY CURRENT PER CHANNEL vs TEMPERATURE

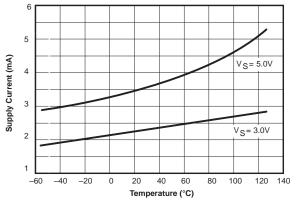


Figure 3.

TYPICAL RISE AND FALL TIMES VS CAPACITIVE LOAD VS SUPPLY VOLTAGE

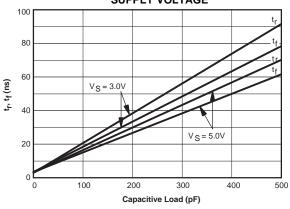


Figure 4.

NORMALIZED RISE-AND-FALL TIME vs TEMPERATURE

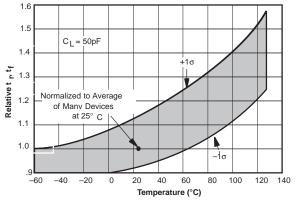
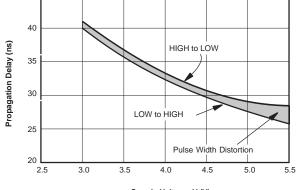


Figure 5.

PROPAGATION DELAY vs SUPPLY VOLTAGE



Supply Voltage, & (V)

Figure 6.



TYPICAL CHARACTERISTICS (continued)

At $T_A = 25$ °C and $V_S = 5$ V, unless otherwise noted.

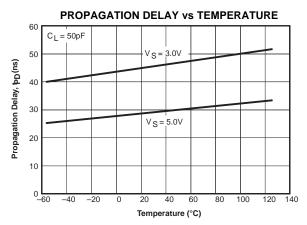


Figure 7.

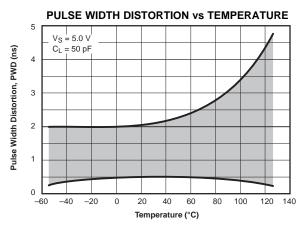
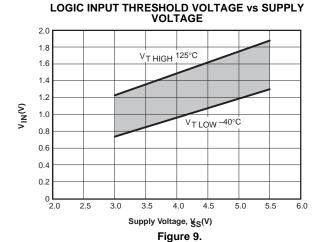
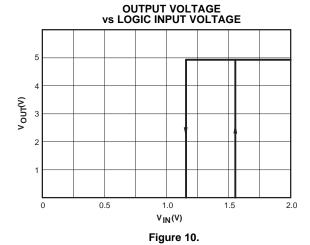


Figure 8.



ISOLATION LEAKAGE CURRENT



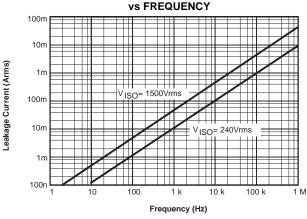


Figure 11.

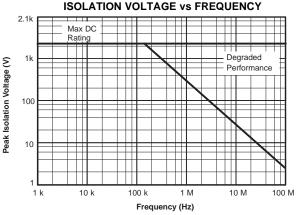


Figure 12.



TYPICAL CHARACTERISTICS (continued)

At $T_A = 25^{\circ}$ C and $V_S = 5$ V, unless otherwise noted.

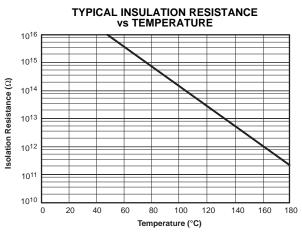


Figure 13.

ISOLATION BARRIER

Data is transmitted by coupling complementary logic pulses to the receiver through two 0.4 pF capacitors. These capacitors are built into the ISO150 package with Faraday shielding to guard against false triggering by external electrostatic fields.

The integrity of the isolation barrier of the ISO150 is verified by partial discharge testing: 2400 Vrms, 60 Hz, is applied across the barrier for one second while measuring any tiny discharge currents that might flow through the barrier. These current pulses are

produced by localized ionization within the barrier; this is the most sensitive and reliable indicator of barrier integrity and longevity, and does not damage the barrier. A device fails the test if five or more current pulses of 5pC or greater are detected.

Conventional isolation barrier testing applies test voltage far in excess of the rated voltage to catastrophically break down a marginal device. A device that passes the test may be weakened, and lead to premature failure.



APPLICATION INFORMATION

Figure 14 shows the ISO150 connected for basic operation; Channel 1 is configured to transmit data from side B to A, whereas Channel 2 is set for transmission from side A to B. The R/\overline{T} pins for each of the four transceivers are shown connected to the required logic level for the transmission direction

shown. The transmission direction can be controlled by logic signals applied to the R/\overline{T} pins. Channel 1 and 2 can be independently controlled for the desired transmission direction. See Figure 15 and Figure 16 for application examples using the ISO150.

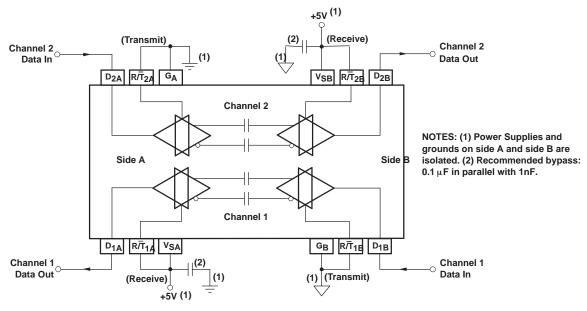


Figure 14. Basic Operation Diagram

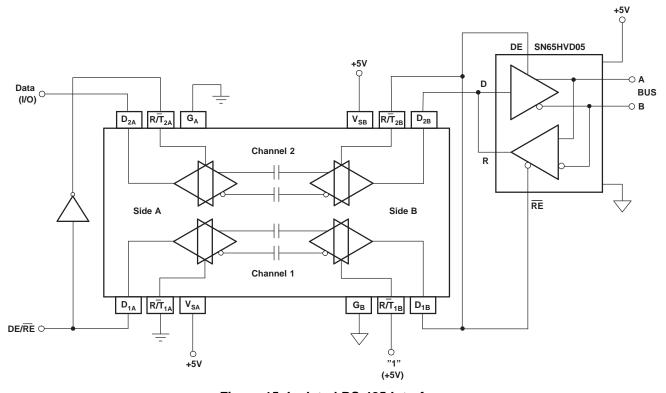


Figure 15. Isolated RS-485 Interface



LOGIC LEVELS

A single pin serves as a data input or output, depending on the mode selected. Logic inputs are CMOS with thresholds set for TTL compatibility. The logic threshold is approximately 1.3 V with 5 V supplies with approximately 400 mV of hysteresis. Input logic thresholds vary with the power-supply voltage. Drive the logic inputs with signals that swing the full logic voltage swing, note that the ISO150 will use somewhat greater quiescent current if logic inputs do not swing within 0.5 V of the power-supply rails.

In receive mode, the data output can drive 15 standard LS-TTL loads. It will also drive CMOS loads. The output drive circuits are CMOS.

POWER SUPPLY

Separate, isolated power supplies must be connected to side A and side B to provide galvanic isolation. Nominal rated supply voltage is 5 V. Operation extends from 3 V to 5.5 V. Power supplies should be bypassed close to the device pins on both sides of the isolation barrier.

The V_S pin for each side powers the transceivers for both channel 1 and 2. The specified supply current is the total of both transceivers on one side, both operating in the indicated mode. Supply current for one transceiver in transmit mode and one in receive mode can be estimated by averaging the specifications for transmit and receive operation. Supply current varies with the data transmission rate — see the typical characteristics.

POWER-UP STATE

When the device is powered up or direction is changed, the transceiver output is indeterminate (either high or low) and cannot be known until an input signal is applied. The output begins to track the input as soon as the input receives a change in logic state, either low to high or high to low.

SIGNAL LOSS

The ISO150's differential-mode signal transmission and careful receiver design make it highly immune to voltage across the isolation barrier (isolation-mode voltage). Rapidly changing isolation-mode voltage can cause data errors. As the rate of change of isolation voltage is increased, there is a very sudden increase in data errors. Approximately 50% of all ISO150s will begin to produce data errors with isolation-mode transients of 1.6kV/µs. This may occur as low as 500 V/µs in some devices. In comparison, a 1000 Vrms, 60 Hz isolation-mode voltage has a rate of change of approximately 0.5V/µs.

Still, some applications with large, noisy isolationmode voltage can produce data errors by causing the receiver output to change states. After a data error, subsequent changes in input data will produce correct output data.

PROPAGATION DELAY AND SKEW

Logic transitions are delayed approximately 27ns through the ISO150. Some applications are sensitive to data skew—the difference in propagation delay between channel 1 and channel 2. Skew is less than 2ns between channel 1 and channel 2. Applications using more than one ISO150 must allow for somewhat greater skew from device to device. As all devices are tested for delay times of 20ns min to 40ns max, 20ns is the largest device-to-device data skew.

MODE CHANGES

The transmission direction of a channel can be changed on the fly by reversing the logic levels at the channel's R/\overline{T} pin. Note that when channel direction is changed, the output state of the channel is indeterminate (either high or low) and cannot be known until an input signal is applied. The output begins to track the input as soon as the input receives a change in logic state, either low to high or high to low.

STANDBY MODE

Quiescent current of each transceiver circuit is very low in transmit mode when input data is not changing (1nA typical). To conserve power when data transmission is not required, program both side A and B transceivers for transmit mode. Input data applied to either transceiver is ignored by the other side. High-speed data applied to either transceiver will increase quiescent current.

CIRCUIT LAYOUT

The high speed of the ISO150 and its isolation barrier require careful circuit layout. Use good high speed logic layout techniques for the input and output data lines. Power supplies should be bypassed close to the device pins on both sides of the isolation barrier. Use low inductance connections. Ground planes are recommended.

Maintain spacing between side 1 and side 2 circuitry equal or greater than the spacing between the missing pins of the ISO150 (approximately 7mm).



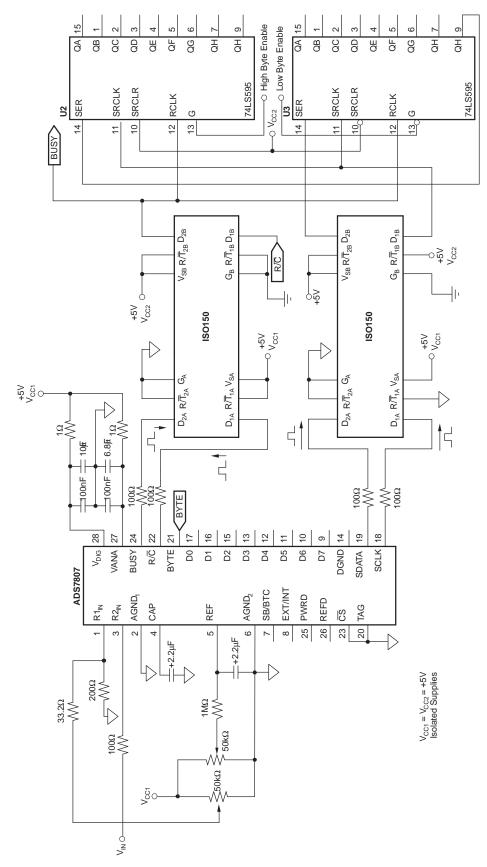


Figure 16. The ISO150 and the ADS7807 are Used to Reduce Circuit Noise in a Mixed-Signal Application





.com 19-Dec-2007

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
ISO150AP	OBSOLETE	PDIP	NVG	12		TBD	Call TI	Call TI
ISO150AU	ACTIVE	SOP	DVB	12	28	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
ISO150AU/1K	ACTIVE	SOP	DVB	12	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
ISO150AU/1KG4	ACTIVE	SOP	DVB	12	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
ISO150AUE4	ACTIVE	SOP	DVB	12	28	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
ISO150AUG4	ACTIVE	SOP	DVB	12	28	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





Α	0	Dimension designed to accommodate the component width
В	0	Dimension designed to accommodate the component length
		Dimension designed to accommodate the component thickness
٧	٧	Overall width of the carrier tape
ГР	1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device		Package Drawing			Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO150AU/1K	SOP	DVB	12	1000	330.0	24.4	10.9	18.3	3.2	12.0	24.0	Q1



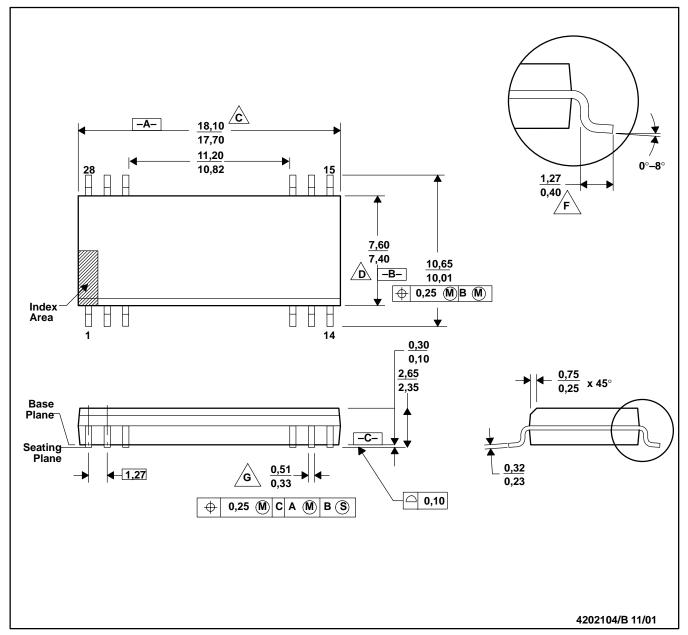


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO150AU/1K	SOP	DVB	12	1000	346.0	346.0	41.0

DVB(R-PDSO-G12/28)

PLASTIC SMALL-OUTLINE



NOTES: A. All linear dimensions are in millimeters.

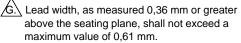
B. This drawing is subject to change without notice.

C.\ Body length dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, and gate burrs shall not exceed 0,15 mm per side.

D.\ Body width dimension does not include inter-lead flash or portrusions. Inter-lead flash and protrusions shall not exceed 0,25 mm per side.

E. The chamfer on the body is optional. If it is not present, a visual index feature must be located within the cross-hatched area.

F.\ Lead dimension is the length of terminal for soldering to a substrate.

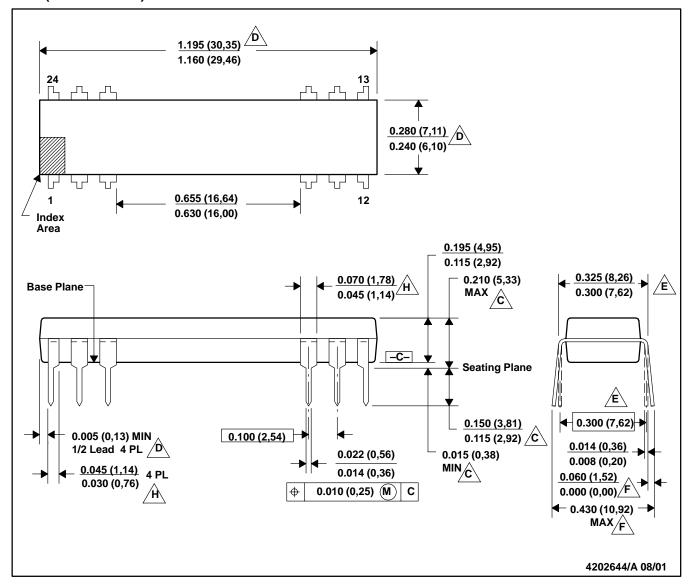


- H. Lead-to-lead coplanarity shall be less than 0,10 mm from seating plane.
- I. Falls within JEDEC MS-013-AE with the exception of the number of leads.



NVG (R-PDIP-T12/24)

PLASTIC DUAL-IN-LINE



- NOTES: A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - Dimensions are measured with the package seated in JEDEC seating plane gauge GS-3.
 - Dimensions do not include mold flash or protrusions.
 - Mold flash or protrusions shall not exceed 0.010 (0,25).

 Dimensions measured with the leads constrained to be perpendicular to Datum C.
 - Dimensions are measured at the lead tips with the leads unconstrained.
 - G. Pointed or rounded lead tips are preferred to ease $_{\mbox{\tiny A}}$ insertion.
 - Maximum dimensions do not include dambar protrusions. Dambar protrusions shall not exceed 0.010 (0,25).

- Distance between leads including dambar protrusions to be 0.005 (0,13) minimum.
- J. A visual index feature must be located within the cross–hatched area.
- K. For automatic insertion, any raised irregularity on the top surface (step, mesa, etc.) shall be symmetrical about the lateral and longitudinal package centerlines.
- L. Controlling dimension in inches.
- M. Falls within JEDEC-MS-001-BE with exception of lead count.



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