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查询LA17000M供应商

## Ordering number: ENN6522

#### Monolithic Linear IC

LA17000M



## Tuner System IC with Built-in PLL for Car Audio Applications

## **Overview**

The LA17000M is an all-in-one car tuner IC that incorporates a PLL frequency synthesizer and all functions of an AM/FM tuner in a single chip. By combining two chips, a PLL (LC72144 equivalent) and an FM tuner IC (LA1781M equivalent) into a single chip (\*PLL + AM (up conversion) + FMFE + IF + NC + MCP + MRC), and as a result of optimal chip partitioning, the LA17000M improves the performance of car tuner systems, eliminates adjustments, and provides high reliability, all at a lower cost.

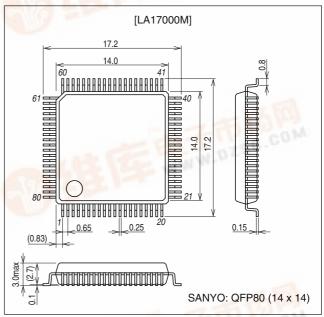
## **Features**

- PLL on chip
  - ADC (6 bits, 1 channel)
  - IF counter and I/O port on chip permit simplification of the interface.
  - Supports AM double conversion.
- Enhanced noise countermeasures
  - Excellent tri-signal characteristics
  - Improved medium and weak electric field NC characteristics
  - · Improved separation characteristics
  - Anti-birdie filter on chip (analog/digital output)
  - Multipath sensor output (analog/digital output)
- · Cost-saving features
  - AM double conversion (Up conversion method)
  - Enhanced FM-IF circuit (When there is interference from adjacent frequencies, the software handles switching of the CF between wide and narrow automatically.)
  - Because deviations in IF gain are only 1/3 that of earlier devices, adjustment is simplified when this IC is incorporated into a set; this IC also includes a shifter pin for VSM adjustment.
- · Suited for smaller devices
  - · Permits high-frequency signal line processing in a tuner pack.
  - · Easily conformes to FCC standards

## W.DZSC.COM Package Dimensions

## unit: mm

#### 3255-QFP80



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## Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
	V <sub>CC</sub> 1 max	Pins 6, 56, and 77	8.7	V
Maximum supply voltage	V <sub>CC</sub> 2 max	Pins 7, 61, 70, 75, and 76	12.0	V
	V <sub>DD</sub> max	Pin 19	6.0	V
Allowable power dissipation	Pd max	Ta $\leq$ 85°C, * With board	950	mW
Operating temperature	Topr		-40 to +85	°C
Storage temperature	Tstg		-40 to 150	°C

\* Specified board:  $114.3 \times 76.1 \times 1.6 \text{ mm}^3$ , glass epoxy

## Operating Conditions at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	Maa	Pins 6, 7, 56, 61, 70, 75, 76, and 77	8.0	V
	Vcc	Pin 19	5.0	V
Operating supply voltage range	V <sub>CC</sub> op		7.5 to 8.5	V
	V <sub>DD</sub> op		4.5 to 5.5	V

## Tuner Block Operating Characteristics at Ta = $25^{\circ}$ C, V<sub>CC</sub> = 8.0 V, V<sub>DD</sub> = 5.0 V, in the specified Test Circuit

De verse ete v	Cumhal	Conditions		Ratings		1.1
Parameter	Symbol	Symbol		typ	max	Unit
[FM characteristics] FM IF input	t					
Current drain	I <sub>CCO</sub> -FM	No input, I56 + I61 + I70 + I75 + I76 + I79	60	98	110	mA
Demodulated output		10.7 MHz, 100 dBµV, 1 kHz, 100%mod, pin 15 output	220	330	445	mVrms
Channel balance	СВ	10.7 MHz, 100 dBμV, 1 kHz, ratio of pin15 and pin 16	-1	0	+1	dB
Total harmonic distortion	THD-FMmono	10.7 MHz, 100 dBµV, 1 kHz, 100% mod, pin 15		0.4	1	%
Signal-to-noise ratio IF	S/N-FM IF	10.7 MHz, 100 dBµV, 1 kHz, 100% mod, pin 15	75	82		dB
AM suppression ratio IF	AMR IF	10.7 MHz, 100 dB $\mu$ , 1 kHz, fm = 1 kHz, pin 15 at 30% AM	55	68		dB
Muting attenuation	Att-1	10.7 MHz, 100 dBµV, 1 kHz, attenuation on pin 15 when V49 = 0 $\rightarrow$ 2 V	3	8	13	dB
	Att-2	10.7 MHz, 100 dBµV, 1 kHz, attenuation on pin 15 when V49 = 0 $\rightarrow$ 2 V *Note 1	13	18	23	dB
	Att-3	10.7 MHz, 100 dB $\mu$ V, 1 kHz, attenuation on pin 15 when V49 = 0 $\rightarrow$ 2 V *Note 2	26	31	36	dB
Separation	Separation	10.7 MHz, 100 dBμ, L + R = 90%, pilot = 10%, pin 15 output ratio	25	35		dB
Stereo ON level	ST-ON	Pilot modulation at which V17 < 0.5 V		4.1	6.6	%
Stereo OFF level	ST-OFF	Pilot modulation at which V17 > 3.5 V	1.2	3.1		%
Main total harmonic distortion	THD-Main L	10.7 MHz, 100 dBμV, L + R = 90%, pilot = 10%, pin 15		0.4	1.2	%
Pilot cancellation	PCAN	10.7 MHz, 100 dBμV, pilot = 10%, pin 15 signal/PILOT-LEVEL leak DIN AUDIO	12	22		dB
SNC output attenuation	AttSNC	10.7 MHz, 100 dB $\mu$ V, L – R = 90%, pilot = 10%, V44 = 3 V $\rightarrow$ 0.6 V, pin 15	1	5	9	dB
HCC output attenuation	AttHCC-1	10.7 MHz, 100 dB $\mu$ V, 10 kHz, L + R = 90%, pilot = 10%, V45 = 3 V $\rightarrow$ 0.6 V, pin 15	1	5	9	dB
	AttHCC-2	10.7 MHz, 100 dB $\mu$ V, 10 kHz, L + R = 90%, pilot = 10%, V45 = 3 V $\rightarrow$ 0.1 V, pin 15	6	10	14	dB

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Parameter	Symbol	Conditions	Ratings			Unit
			min typ m			Unit
Input limiting voltage	V <sub>IN</sub> -LIM	10.7 MHz, 100 dB $\mu$ V, 30% mod, IF input that decreases the input reference output by –3 dB	29	36		dBµV
Muting sensitivity	VIN-MUTE	IF input level non-mod when V49 = 2 V	19	27	35	dBµV
SD sensitivity	SD-sen1 FM	IF input non-mod (at least 100 mVrms) at which the IF count buffer output turns on	48	56	64	dBμV
	SD-sen2 FM		48	56	64	dBµV
IF counter buffer output	VIFBUFF-FM1	10.7 MHz, 100 dB $\mu$ V, non-mod, pin 38 output, during SEEK	145	245	330	mVrms
	VIFBUFF-FM2	10.7 MHz, 100 dB μV, non-mod, pin 38 output, during RDS mode	145	245	330	mVrms
Signal meter output	V <sub>SM</sub> FM-1	No input, pin 42 DC output non-mod	0.0	0.1	0.3	V
	V <sub>SM</sub> FM-2	50 dBµ, pin 42 DC output non-mod	0.65	1.6	2.4	V
	V <sub>SM</sub> FM-3	70 dBµ, pin 42 DC output non-mod	2.4	3.2	4.2	V
	V <sub>SM</sub> FM-4	100 dBµ, pin 42 DC output non-mod	4.9	5.8	6.5	V
Muting bandwidth	BW-MUTE	100 dB $\mu$ V, when V49 = 2 V Bandwidth non-mod	140	210	280	kHz
Muting drive output	V <sub>MUTE</sub> -100	100 dBµV, 0 dBµ, pin 49 DC output non-mod	0.00	0.1	0.3	V
[FM FE Block]	•					
N-AGC on input	V <sub>NAGC</sub>	83 MHz, non-mod, input at which pin 2 is 2.0 V or less	72	79	86	dBµV
W-AGC on input	Vwagc	83 MHz, non-mod, input at which pin 2 is 2.0 V or less (when KEYED-AGC is 4.0 V)	90	97	104	dBµV
Conversion gain	A. V1	83 MHz, 80 dBμ, non-mod, FECF output	9	13	17	dB
	A. V2	83 MHz, 80 dB $\mu$ , non-mod, 5 V applied to CF (pin 10), FECF output	13	17	21	dB
Oscillator buffer output	VOSCBUFFFM	No input, pin 5 output	51	67	102	mVrms
[NC Block] NC input (pin 30)						
Gate time	<sup>™</sup> GATE	f = 1 kHz, 1 μs, 100 mVp-o pulse input		15		μs
Noise sensitivity	SN	1 kHz, 1 μs pulse input that starts noise canceller operation. Measured at Pin 30.		18		mVp-o
[MRC Block]						
MRC output	VMRC	V42 = 5 V	2.1	2.25	2.4	V
MRC operating level	MRC-ON	Input level on pin 48 that is below pin 42 = 5 V and pin 43 = 2 V, f = 70 kHz	22	33	44	mVrms
MRC sensor output	VMRC-sensor1	V42 = 5 V, pin 34 output		1.5	1.9	V
	V <sub>MRC</sub> -sensor2	V42 = 5 V, pin 48 output, f = 70 kHz, 100 mVrms	2.1	2.9		V
[AM Characteristics] AM ANT	input					
Practical sensitivity	S/N-30	1 MHz, 30 dBµV, fm = 1 kHz, 30% mod, pin 15	15			dB
Detection output	V <sub>O</sub> -AM	1 MHz, 74 dBµV, fm = 1 kHz, 30% mod, pin 15	105	160	220	mVrms
AGC-F.O.M	VAGC-FOM	1 MHz, 74 dB $\mu$ V, output reference, input width at which output drops by 10 dB, pin 15	50	55	60	mVrms
Signal-to-noise ratio	S/N-AM	1 MHz, 74 dBµV, fm = 1 kHz, 30% mod	47	52		dB
Total harmonic distortion	THD-AM	1 MHz, 74 dBµV, fm = 1 kHz, 80% mod		0.5	1.2	%
Signal meter output	V <sub>SM</sub> AM-1	1 MHz, 30 dBµV, non - mod	0.6	1	1.4	V
	V <sub>SM</sub> AM-2	1 MHz, 120 dBµV, non - mod	3.4	4.5	5.9	V
Oscillator buffer output	VOSCBUFFAM-1	No input, pin 5 output	170	210		mVrms
Wideband AGC sensitivity	W-AGCsen1	1.4 MHz, input when V62 = 0.7 V	87	93	99	dBµV
,	W-AGCsen2	1.4 MHz, input when V62 = 0.7 V (during SEEK)	78	84	90	dBµV

#### Continued from preceding page.

Devenueter	Symbol Conditions					
Parameter	Parameter Symbol Conc		min	typ	max	Unit
SD sensitivity	SD-sen1AM	1 MHz, ANT input level at which IF count output turns on		33	39	dBμV
	SD-sen2AM	1 MHz, ANT input level at which SD pin turns on	27	33	39	dBµV
IF buffer output	VIFBUFF-AM	1 MHz, 74 dB $\mu$ V, non-mod, pin 38 output	150	220		mVrms

# PLL Block Allowable Operating Ranges at Ta = -40 to $+85^{\circ}$ C, $V_{DD}$ = 5 V, $V_{SS}$ = 0 V

			Ratings			
Parameter	Symbol	Conditions	min	typ	max	Unit
High-level input voltage	VIH1	CE, CL, DI, I/O-1, I/O-2	2.2		V <sub>DD</sub> + 0.3	V
Low-level Input voltage	VIL1	CE, CL, DI, I/O-1, I/O-2, SDSTSW	0		0.8	V
Output voltage	Vo1	DO	0		6.5	V
	V <sub>O</sub> 2	I/O-1, I/O-2	0		13	V
Input amplitude	fin1	X <sub>IN</sub> ; Sine wave, capacitor coupled	1		8	MHz
	f <sub>IN</sub> 2	PLL <sub>IN</sub> ; Sine wave, capacitor coupled	10		160	MHz
	f <sub>IN</sub> 3	HCTR; Sine wave, capacitor coupled	0.4		25	MHz
Guaranteed crystal oscillator ranges	X'tal	X <sub>IN</sub> , X <sub>OUT</sub> ; CI ≤ 70 Ω (X'tal: 10.25, 10.35 MHz); Note 1	10.1		10.5	MHz
Input amplitude	VIN1	X <sub>IN</sub>	200		1500	mVrms
	V <sub>IN</sub> 2-1	PLL <sub>IN</sub> ; 10 ≤ f < 130 MHz; Note 2	40		1500	mVrms
	V <sub>IN</sub> 2-2	PLL <sub>IN</sub> ; 130 ≤ f <160 MHz; Note 2	70		1500	mVrms
	V <sub>IN</sub> 3-1	HCTR; $0.4 \le f < 25$ MHz: Serial data; CTC = 0: Note 3	40		1500	mVrms
	V <sub>IN</sub> 3-2	HCTR; $8 \le f < 12$ MHz: Serial data; CTC = 1: Note 4	70		1500	mVrms
Data setup time	tsu	DI, CL: Note 5	0.45			μs
Data hold time	t <sub>HD</sub>	DI, CL: Note 5	0.45			μs
Clock low-level time	tCL	CL: Note 5	0.45			μs
Clock high-level time	tсн	CL: Note 5	0.45			μs
CE wait time	tEL	CE, CL: Note 5	0.45			μs
CE setup time	tES	CE, CL: Note 5	0.45			μs
CE hold time	tен	CE, CL: Note 5	0.45			μs
Data latch change time	tLC	Note 5			0.45	μs
Data output time	t <sub>DC</sub>	DO, CL; Dependent on pull-up resistance, board capacity: Note 5			0.2	μs
	t <sub>DH</sub>	DO, CL; Dependent on pull-up resistance, board capacity: Note 5			0.2	μs

Note 1: Recommended CI value for crystal oscillator CI  $\leq$  70  $\Omega$  (X'tal: 10.25, 10.35 MHz)

However, because the characteristics of the X'tal oscillation circuit depend on the board and circuit constants, we recommend requesting that the X'tal manufacturer perform the evaluation.

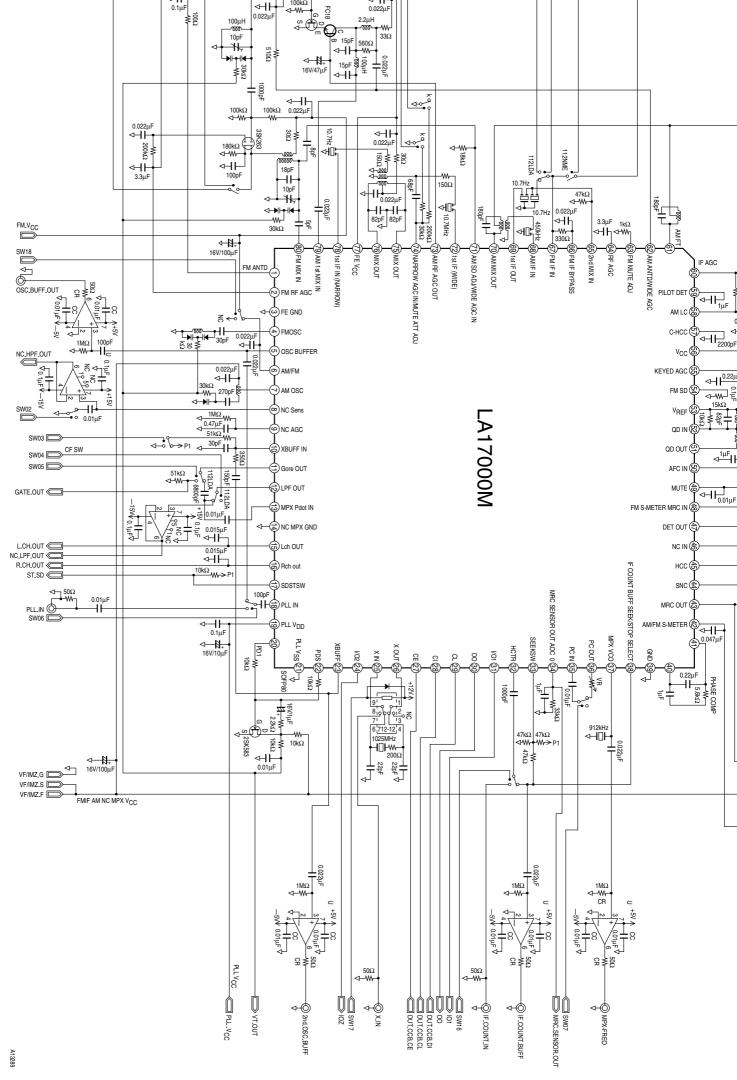
Note 2: Refer to the program divider configuration.

Note 3: Serial data: CTC = 0

- Note 4: Serial data: CTC = 1
- Note 5: Refer to the serial data timing.

# PLL Characteristics Electrical Characteristics at Ta = 25°C, V<sub>DD</sub> = 5 V, V<sub>SS</sub> = 0 V

Deremeter	Cumbal	Conditions		Ratings		11
Parameter	Symbol Conditions		min	typ	max	Unit
Built-in feedback resistors	Rf1	X <sub>IN</sub>		1		MΩ
	Rf2	PLL <sub>IN</sub>		500		kΩ
	Rf3	HCTR		250		kΩ
Hysterisis width	VHIS	CE, CL, DI		0.1V <sub>DD</sub>		V
High-level output voltage	V <sub>OH</sub> 1	PD1, PDS, SEEKSW; $I_0 = -1 \text{ mA}$	V <sub>DD</sub> -1.0			V
	V <sub>OH</sub> 2	XBUF; $I_0 = -0.5 \text{ mA}$	V <sub>DD</sub> -1.5			V
Low-level output voltage	V <sub>OL</sub> 1	PD1, PDS, SEEKSW; $I_0 = -1 \text{ mA}$			1	V
	Vol2	XBUFF; $I_0 = -0.5 \text{ mA}$			1.5	V
	V <sub>OL</sub> 3	I/O-1 to I/O-2; I <sub>O</sub> = 1.0 mA			0.2	V
		I/O-1 to I/O-2; I <sub>O</sub> = 2.5 mA			0.5	V
		I/O-1 to I/O-2; I <sub>O</sub> = 5.0 mA			1	V
		I/O-1 to I/O-2; I <sub>O</sub> = 9.0 mA			1.8	V
·	V <sub>OL</sub> 4	DO; I <sub>O</sub> = 5.0 mA			1	V
High-level input current	l <sub>IH</sub> 1	CE, CL, DI; V <sub>IN</sub> = 6.5 V			5	μA
	I <sub>IH</sub> 2	I/O-1 to I/O-2; V <sub>IN</sub> = 13 V			5	μA
	liH3	XIN; VIN = VDD	2		11	μA
·	lıH4	PLLIN; VIN = VDD	4		22	μA
Low-level input current	lı∟1	CE, CL, DI; V <sub>IN</sub> = 0 V			5	μA
-	l <sub>IL</sub> 2	I/O-1 to I/O-2; V <sub>IN</sub> = 0 V			5	μA
	lı∟3	$X_{IN}$ ; $V_{IN} = 0 V$	2		11	μA
	IIL4	PLL <sub>IN</sub> ; V <sub>IN</sub> = 0 V	4		22	μA
Output off leakage current	IOFF1	I/O-1 to I/O-2; V <sub>O</sub> = 13 V			5	μA
	IOFF2	DO; V <sub>O</sub> = 6.5 V			5	μA
High-level 3-state off leakage current	IOFFH	PD1, PDS; V <sub>IN</sub> = V <sub>DD</sub>		0.01	200	nA
Low-level 3-state off leakage current	IOFFL	PD1, PDS; V <sub>IN</sub> = 0 V		0.01	200	nA
Input capacitance	CIN			6		pF
A/D converter linearity error	Err	MRC SENSOR AUTO ADJ (MOS)	-0.5		+0.5	LSB
Pull-down transistor on resistance	Rpd1	PLLIN	80	200	600	kΩ
Supply current	I <sub>DD</sub> 1	$\label{eq:VDD} \begin{array}{l} V_{DD}; X'tal = 10.25 \mbox{ MHz}, \\ f_{IN}2 = 160 \mbox{ MHz}, \\ V_{IN}2 = 70 \mbox{ mVrms}, \\ f_{IN}3 = 25 \mbox{ MHz}, \\ V_{IN}3 = 40 \mbox{ mVrms} \end{array}$		10	15	mA
	I <sub>DD</sub> 2	V <sub>DD</sub> ; PLL block halt (PLL INHIBIT), X'tal OSC operation (10.25 MHz)		5	10	mA
	IDD3	V <sub>DD</sub> ; PLL block halt, X'tal OSC halt			3	mA



No. 6522-6/54

## [FM IF Selectivity Switching Circuit]

## Features

- 1) Comprises an FM/AM one-chip system.
- 2) Up conversion method is adopted for AM.
- 3) Uses an IF filter with a center frequency that is the same as the middle frequency of FM.
- 4) Uses a narrowband filter in AM mode.
- 5) Uses a narrowband filter in FM mode only during SEEK or when there is interference from adjacent frequencies.
- 6) Uses a wideband filter for normal reception in FM mode.
- 7) For an RDS AF search, switches to a narrowband filter and detects SD.
- 8) High sensitivity for detecting interference from adjacent frequencies.

## Advantages

- 1) This FM/AM one-chip tuner system (an IC that includes a microcontroller interface) allows for improved adjacent frequency interference characteristics without increased cost.
- 2) Prevents SD and IF count misdetection (station detection) during seek search, RDS AF search, and auto memory operations.
- 3) Permits adoption of an IC for certain functions without increasing the number of IC pins.
- 4) CF selectivity can be switched by the software in the microcontroller that controls the tuner, making it easy to achieve performance differentiation through the software.(The software can freely set the CF switching timing and conditions.)
- 5) Detects the radio wave status in the field through detection of SD, desired station field intensity, IF count output, and adjacent station field intensity. This IC offers improved adjacent frequency interference characteristics by switching the CF automatically when interference is being generated from an adjacent frequency.

## [IF Band Switching Circuit]

## Purpose

This AM/FM one-chip tuner IC automatically switches the FM selectivity, prevents misdetection during SEEK operations, and offers improved adjacent frequency interference characteristics without any increase in cost.

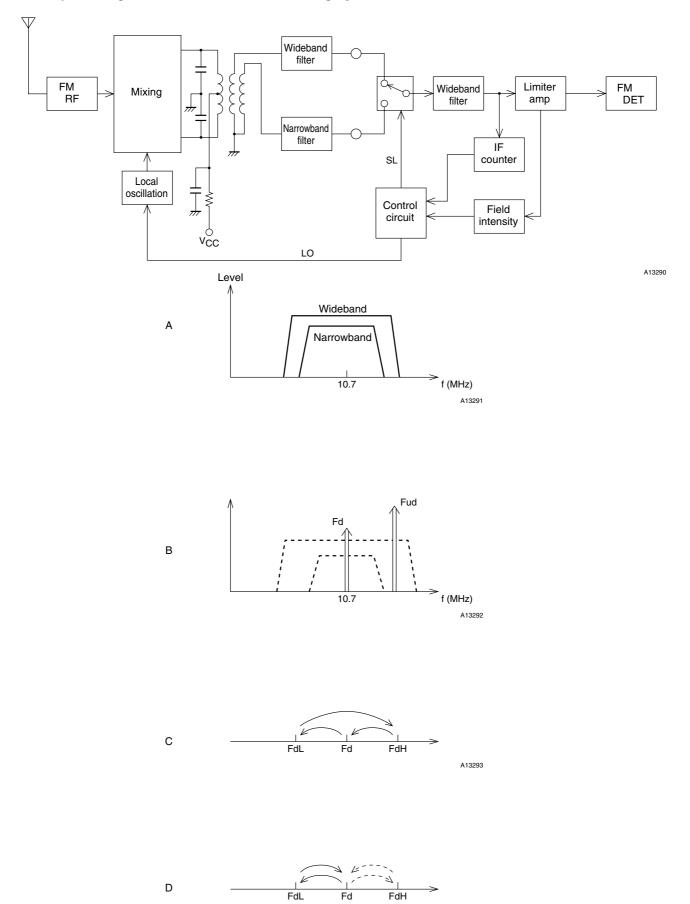
## **New Technological Features**

- 1. Comprises an AM/FM one-chip IC.
- 2. Because the narrowband CF that is used by the AM UP conversion system is also used for FM, additional external components required by earlier systems can be eliminated.
- 3. Uses a wideband CF during normal FM reception for high sound quality.
- 4. Uses a narrowband CF for AM reception, and if interference is being generated from adjacent frequencies during FM reception.
- 5. Uses a narrowband CF during SEEK and RDSAF search operations, preventing misdetection of SD and IF count due to adjacent stations.
- 6. CF switching is performed at the first IF amp input, and the amp gain is adjusted automatically to a suitable level according to the CF band form AM/FM or FM.
- 7. Switching of the CF input and the first IF amp gain is controlled by a microcontroller through the interface. The pins that are controlled are connected to the I/O ports of the microcontroller, and are controlled by the microcontroller's internal software.
- 8. Detection of adjacent frequency interference during FM reception is based on S-meter output, SD, and IF count output. The IF count buffer frequency fluctuates when interference is being generated from adjacent frequencies. This fluctuation is used to make the detection of interference from adjacent frequencies possible. (Related patents have been applied for.)

## **Conventional Technologies**

- 1. Comprised of a dedicated IC for IF band switching, or of multiple ICs.
- 2. None of the AM/FM all-in-one chip systems include the functions provided by the LA17000M.
- 3. Requires a narrowband CF especially for FM, resulting in increased costs. (Does not share the AM narrowband CF.)
- 4. Because CF switching control is handled by analog circuits or logic circuits, the switching timing can only be controlled through uniform conditions. Control by software is not possible.

LA17000M



## Conceptual Diagram of the FM-IF Band Switching System

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#### I/O Port Assignment Table

I/O-0	OUTPUT PLL output port	L: Reception mode H: Seek mode
DI data	INPUT PLL input port	OPEN: RDS
I/O-1		Unused
I/O-2 DI data	OUTPUT PLL output port	H: Dx mode L: Lo mode
I/O-3 DO data	INPUT I/O-3 = 0 (input port) OUT3 = 1 (OPEN or high) PLL input port	When reception mode is set H: Monaural L: Stereo When seek mode is set H: SD ON
	Cannot be set as output port	L: SD OFF

The MRC sensor reads DO data from the PLL microcontroller's 6-bit A/D converter.

Currently, aside from the CCB data lines, only three lines are connected to the controller microcontroller: CF/SW, AUDIO mute, and AM/FM band switching port.

## Selectivity Switching Evaluation Software

## State-based Data Switching Table

I/O port state	Tuner processing	Seek	Manual preset	Receiving	Remarks
CF switching	WIDE		0	0	
Or switching	NARROW	0	0	0	
AUDIO mute output	ON	0	0		Switchable but fixed by software
	OFF			0	Switchable but fixed by software
Lo/Dx	Lo	0	0	0	Processing is performed according to the setting
20,07	Dx	0	0	0	Processing is performed according to the setting
	Seek mode	0		0	I/O-3 is SD output
Mode switching	Reception mode		0	0	I/O-3 is monaural/stereo output
	RDS mode			0	I/O-3 is SD output
IF count	Output ON	0		0	Seek mode RDS mode
	Output OFF		0		Reception mode

## Additional Settings (Added to the LC72144M)

#### Output (DI)

	Mode	Settings	When set
	Seek mode	DI data IN2 I/O-0 = 1 (output port) OUT0 = 1 (Hi)	For seek
Tuner mode switch	Reception mode	DI data IN2 I/O-0 = 1 (output port) OUT0 = 0 (Lo)	For seek-stop and for receiving
	RDS mode	DI data IN2 I/O-0 = 0 (input port) OUT0 = 1 (OPEN)	For AF search
	Lo mode	DI data IN2 I/O-2 = 0 (output port) OUT2 = 0 (Lo)	When setting Lo mode
Lo/Dx switch	Dx mode	DI data IN2 I/O-2 = 1 (output port) OUT2 = 1 (Hi)	When setting Dx mode
Hard mute *1	Mute ON	DI data IN2 I/O-0 = 1 (output port) OUT1 = 1 (Hi)	For tuning processing
	Mute OFF	DI data IN2 I/O-0 = 1 (output port) OUT1 = 1 (Lo)	When switching reception mode

Note: \*1. Depends on the I/O ports usage.

## Input (DO)

		DO data	Conditions
Sensor	Monaural/stereo	OUT data I3 = 1 (Hi) Monaural state OUT data I3 = 0 (Lo) Stereo state	When the tuner mode is set to reception mode
	SD	OUT data I3 = 1 (Hi) SD ON OUT data I3 = 0 (Lo) SD OFF	When the tuner mode is set to seek or RDS mode *2
MRC output		OUT data ADC0 AD00 to AD05 6 bit	Start AD conversion and then read after conversion is completed. 3.3 V at 6-bit resolution

Note: \*2. I/O-3 = 0 (input port) and OUT3 = 1 (Hi) must already be set in the DI data (IN2) settings.

Other settings

	In the LA17000	Setting	When set
CF switch	Pin 10	Hi: Wide (wideband setting) Lo: Narrow (narrowband setting)	For normal operation When there is interference from adjacent frequencies
Soft mute (AUDIO mute)	Pin 49	Hi: Forced mute Lo: Mute off	When setting mute When cancelling mute
AM/FM switch	Pin 6	Lo: AM Hi: FM	For AM reception For FM reception

LA1781 Pin No.	Pin Function	LA17000M Pin No.	Pin Function	LC72144N Pin No.
1	FN ANTD	1		
2	FM RF AGC	2		
3	FE GND	3		
4	FM OSC	4		
5	AM/FM OSC buff.	5		
6	FE V <sub>CC</sub>	6		
7	AM V <sub>CC</sub>	7		
8	Noise AGC-Sense	8		
9	Noise AGC-ADJ	9		
10	AM 2nd OSC	10		
11	Gate Out	11		
12	Memory circuit pin	12		
13	Pilot In	13		
14	NC, MPX GND	14		
15	MPX L-Out	15		
16	MPX R-Out	16		
26	$\begin{array}{l} Seek \to AM/FM \ SD \\ Stop \to FM \ ST \ IND \end{array}$	17	Both I/O-3 and SD/ST-IND	23
		18	FMIN	16
		19	VDD	17
		20	PD1	18
		21	V <sub>SS</sub>	19
		22	PDS	20
		23	XBUF	22
		24	I/O-2	8
		25	XIN	24
		26	XOUT	1
		27	CE	2
		28	DI	3
		29	CL	4
		30	DO	5
		31	I/O-1	9
		32	HCTR/I-6	11
		33	I/O-0	12
19	MRC sensor output	34	Both ADC0 and MRC sensor output	7
17	Pilot Can. ADJ	35		
18	Pilot Can. ADJ	36		
20	MPX VCO	37		
23	IF count buffer and seek/stop switch	38		
25	GND	39		
21	PHASE COMP.	40		
22	PHASE COMP.	41		
24	AM/FM S-meter	42		
27	MRC OUT	43		

## Correspondence of Pins Between the LA17000M, the LA1781M, and the LC72144M

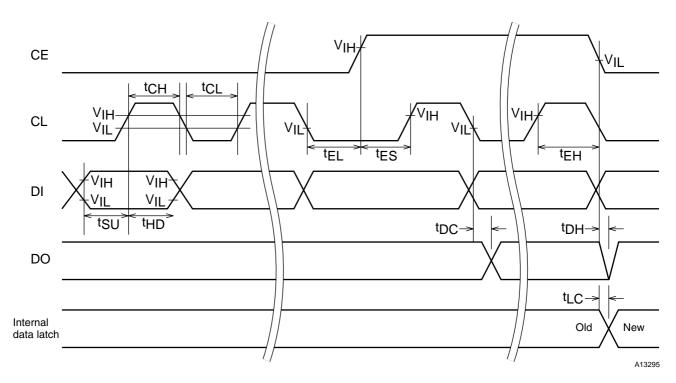
Continued from preceding page.

LA1781 Pin No.	Pin Function	LA17000M Pin No.	Pin Function	LC72144M Pin No.
28	SNC control input	44		
29	HCC control input	45		
30	Noise canceller IN	46		
31	AM/FM detector output	47		
32	FM S-meter output	48		
33	MUTE drive	49		
34	AFC IN	50		
35	QD OUT	51		
36	CD IN	52		
37	VREF	53		
38	FMSD	54		
39	GND Keyed AGC	55		
40	Vcc	56		
41	HCC capacitor	57		
42	AM L.C.	58		
43	Pilot detector	59		
44	IF AGC	60		
45	AM IFT (IF output)	61		
46	AM ANTD W-AGC IN	62		
47	FM Mute ON ADJ	63		
48	RF AGC	64		
49	AM 2nd MIX IN	65		
50	FM IF BYPASS	66		
51	FM IF IN	67		
52	AM IF IN	68		
53	1st IF amplifier output	69		
54	AM MIX OUT	70		
55	W-AGC IN AM SD ADJ	71		
56	1st IF IN	72		
57	AM RF AGC OUT	73		
58	N-AGC IN	74		
59	1st MIX OUT	75		
60	1st MIX OUT	76		
61	F.E.V <sub>CC</sub>	77		
64	FM MIX IN	78	1st IF narrow IN	
62	AM MIX IN	79		
63	FM MIX IN	80		

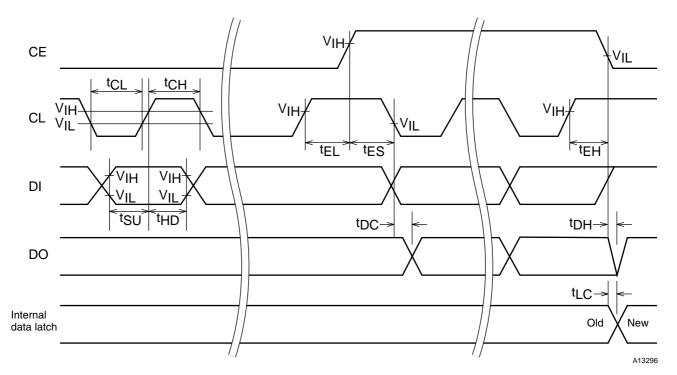
#### **PLL Block Functions**

- High-speed programmable divider
- FMIN : 10 to 160 MHz ..... Pulse swallower method
   General-purpose counter
- HCTR : 0.4 to 25.0 MHz ..... Frequency measurement
- Crystal oscillator : Two frequencies selectable: 10.35/10.25 MHz
- Reference frequencies : 12 frequencies selectable:
  - 50, <u>30</u>, 25, 12.5, 6.25, 3.125, 10, <u>9</u>, <u>3</u>, 5, and 1kHz \*1 \*1
  - \*1: Not available when using the 10.25 MHz crystal oscillator
- · Phase comparator
  - Dead zone can be controlled
  - Unlock detection circuit
  - Sub-charge pump for high-speed locking
  - Deadlock clear circuit on chip
- Serial data I/O
- Communications with controller possible in CCB format
- Power-on reset circuit
- · On-chip crystal oscillator output buffer
- 2nd IF injection signal for AM up conversion (10.35/10.25 MHz)
- I/O port ...... General-purpose I/O: four ports

## **Serial Data Timing**



When CL is Stopped at the low level



When CL is Stopped at the high level

LA17000M

## **PLL Block Pin Description**

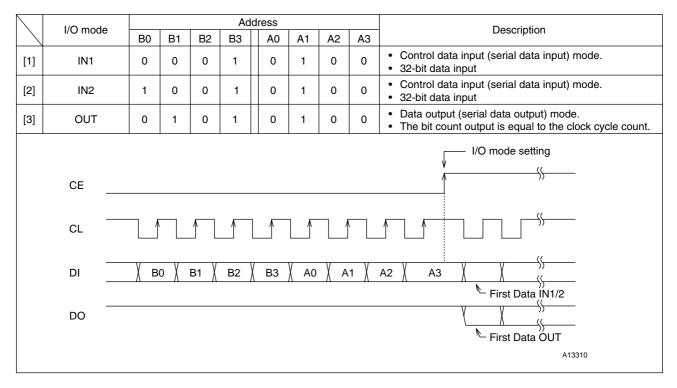
Symbol	Pin No.	Description	Function	Pin Circuit
XIN XOUT	25 26	X'tal OSC	<ul> <li>For connecting the crystal oscillator. (10.35, 10.25, 7.2 or 4.5 MHz)</li> </ul>	A13297
PLL IN	18	Local oscillator signal input	<ul> <li>FMIN is selected when DVS in the serial data input is set to 1.</li> <li>The input frequency range is from 10 to 160 MHz.</li> <li>The signal is transmitted to the swallow counter.</li> <li>The divisor can be set to a value in the range 272 to 65535.</li> </ul>	A13298
CE	27	Chip enable	• This pin is set high during serial data input to the PLL (DI) or during serial data output (DO).	A13299
CL	29	Clock	<ul> <li>This pin is the clock for data synchronization during serial data input to the PLL (DI) or during serial data output (DO).</li> </ul>	A13300
DI	28	Input data	• This is the input pin for serial data that is transferred from the controller to the PLL.	A13301
DO	30	Output data	• This is the output pin for serial data that is transferred from the controller to the PLL.	A13302
VDD	19	Power supply	<ul> <li>This is the PLL power supply pin. Supply 4.5 V to 5.5 V to this pin when the PLL is operating.</li> <li>When power is first applied to this pin, the power-on reset circuit operates.</li> </ul>	
Vss	21	Ground	This is the PLL ground pin.	
I/O-1 I/O-2 STSD SW	31 24 17	General- purpose I/O ports	<ul> <li>These are general-purpose I/O ports.</li> <li>The output circuits open-drain.</li> <li>During a power-on reset, I/O-1 and I/O-2 become input ports. STSD SW becomes an output port, and is fixed low.</li> <li>These ports can be switched between input and output according to the serial data that is transferred from the controller (I/O-1, I/O-2, STSD SW).</li> </ul>	A13303
SEEK SW	33	General- purpose I/O port	<ul> <li>This is a general-purpose I/O port.</li> <li>The output circuits are complementary circuits.</li> <li>During a power-on reset, this port becomes an input port.</li> <li>This port can be specified as an input or output port by the serial data that is transferred from the controller.</li> </ul>	A13304
ADC0	34	ADC input	<ul> <li>This is the A/D converter input pin. The converter is a 6-bit successive-approximation A/D converter. For details, refer to the page that describes the A/D converter configuration.</li> </ul>	

#### Continued from preceding page.

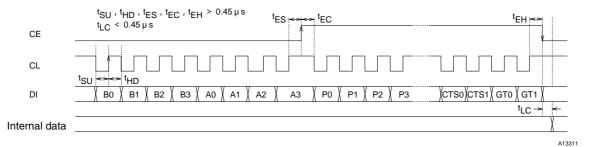
Symbol	Pin No.	Description	Function	Pin Circuit
PD1	PD1 20 Main 0 charge pump output		• This is the PLL charge pump output pin. When the frequency of the local oscillation signal frequency is divided by N is higher than the reference frequency, a high level signal is output from the PD1 pin. When the frequency is lower, a low level signal is output. If the frequencies match, the pin goes to high impedance.	A13306
PDS	22	Sub-charge pump output	<ul> <li>A high-speed lockup circuit can be formed by using this pin in combination with the main charge pump.</li> <li>For details, refer to page that describes the charge pump configuration.</li> </ul>	
HCTR	32	General- purpose counter	<ul> <li>Serial data: HCTR is selected if CTS1 = 1 is set.</li> <li>The input frequency is 0.4 to 25 MHz.</li> <li>The signal is passed through to the general-purpose counter internally, via the 1/2 frequency divider. An integrating count can also be kept.</li> <li>The count result is output from the MSB of the general-purpose counter through the output pin DO.</li> <li>For details, refer to page that describes the general-purpose counter configuration.</li> <li>Serial data: Prohibited when HCTR = 0.</li> </ul>	A13308
XBUF	23	X'tal oscillator buffer	<ul> <li>This is the output buffer for the crystal oscillator circuit.</li> <li>Serial data: When XB = 1 is set, the output buffer operates and the crystal oscillator signal (pulse) is output. When XB = 0, this pin outputs a low level. (When a power-on reset is executed, XB = 0 and the output buffer is fixed at the low level.)</li> </ul>	XOUT

#### Procedures for Input and Output of Serial Data

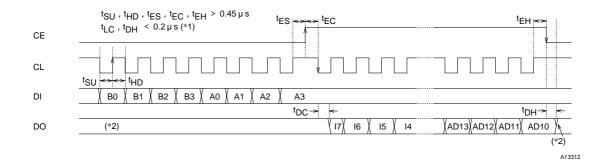
Data I/O is handled through the Computer Control Bus (CCB), SANYO's audio IC serial bus format. This IC uses CCB with 8-bit addressing.



#### i) Serial Data Input (IN1/IN2)



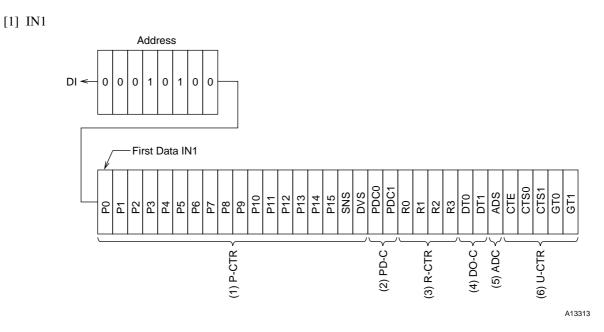
## ii) Serial data output (OUT)



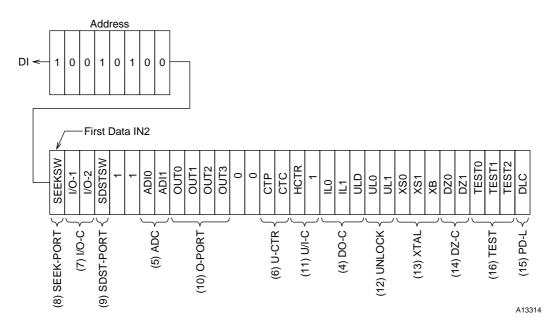
\*1: Because the DO pin is an N-channel open drain pin, the data transition time varies according to the pull-up resistance and the board capacitance.

\*2: The DO pin is normally open.

## DI Control Data (Serial Data Input) Configuration



## [2] IN2 Mode



## **Description of DI Control Data**

No.	Control block/Data	Description						Related data
	Programmable divider data						the programmable divider. P15 is a binary value that is The LSB changes depending on DVS and SNS.	
	P0 to P15		DVS					
			1	1		P0	272 to 65535	
(1)		•						
	DVS, SNS		DVS	SN	IS	Input pir	n Input pin frequency range	
			1	1		PLLIN	10 to 160 MHz	
			* For de	etails	, ref	er to "Pro	ogrammable Divider Configuration."	
	Sub-charge pump control data	•	This da	ta co	ntro	ls the su	b-charge pump.	
	control data		PDC	1	Ρ	PDC0	Subcharge pump status	UL0, UL1, DLC
(2)	PDC, PDC1				High impedance Charge pump on (when unlocked) Charge pump on (normal operation)			
	Reference divider data	•	combi For deta	natio ails, r	n w refe	r to the p	can be used to form a high-speed lockup circuit in and PD1 (main charge pump). bage on charge pump. quency (fref) selection data.	
		[	R3 F	32	R1	R0	Reference frequency	
			0	0	0	0	Prohibited	
			0	0	0	1	50	
			0	0	1	0	25	
			-	0	1	1	25	
			-	1	0	0	12.5	
				1	0 1	1	6.25	
			-	1	1	0	<u>3.125</u> <u>3.125</u>	
	R0 to R3		-	0	0	0	10	
(3)				0	0	1	9 *1	
. ,			1	0	1	0	5	
			1	0	1	1	1	
			1	1	0	0	9 *1	
			1	1	0	1	30 *1	
				1	1 1		2 PLL INHIBIT + X'tal OSC STOP 2 PLL INHIBIT	
			: Prohib : PLL IN The pr	ited v IHIBI rogra	whe T (b mm	n X'tal C backup m able divi	DSC = 10.25 MHz.	

## Continued from preceding page.

No.	Control block/Data		Description						
	DO, I/O-5 pin control data	• This dat							
		ULD	DT1	DT0	DO pin	]			
		0	0	0	Low when unlocked	-			
		0	0	1	end-AD		I/O-1		
	ULD DT0, DT1	0	1	0	end-UC		I/O-2		
	ILO, IL1	0	1	1	IN (*1)				
					sion by the A/D converter sion by the general-purp				
		DC							
(4)									
				Start	/1 1	End CE : Hi			
					(1-1	changes)			
						A13315			
		* 1							
		IL1	IL		IN				
		0	0	Open					
		0	1 0		n status) n status)				
			1		changes, DO goes low. (	Note)			
						,			
		* Howe		ne I/O-1	and I/O-2 pins are spec	ified as output ports, these pins			
		Note: C	annot b			STOP. (DO does not change.)			
		[\	When the	e referei	nce divider data: R3 = R	2 = R1 = 1 and $R0 = 0$ ]			
	A/D converter control data	• A/D cor	verter c	onversio	on start data.				
		ADS =			ion reset and start				
					ion reset				
(5)	100	ADI1	ADI0		AD input pin				
	ADS ADI0		1	Stopp ADC0					
		1	0 1	Not us					
		0	0	Not us					
							<u> </u>		

## Continued from preceding page.

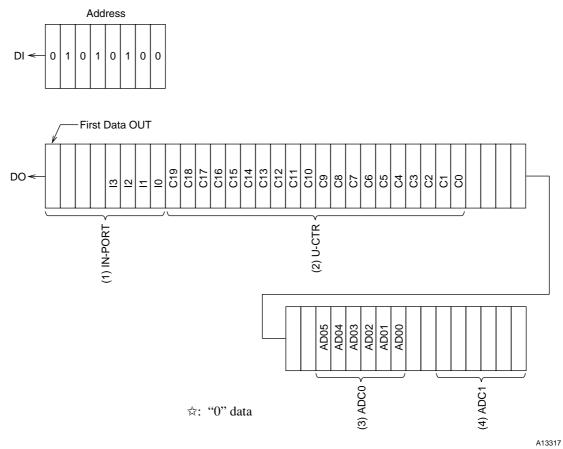
No.	Control block/Data					Description			Related data	
	General-purpose counter	•	This dat	ta sets t	he general-purpo	ose counter i	nput pin (HC	TR).		
	control data		CTS1 Measurement time Measurement mode							
			1     HCTR     Frequency       0     —     Not measured						HCTR	
	CTS0, CTS1 CTE GT0, GT1	$\Box = 1$ : Count start								
					Frequency	measuremer	nt mode	Period		
			GT1	GTO	Measurement		ne (ms)	measurement		
(6)					time (ms)	CTP = 0	CTP = 1	mode		
			0	0	4	3 to 4	1 to 2	1 period		
			0	1	8	3 to 4	1 to 2	1 period		
			1	0	32	7 to 8	1 to 2	2 periods		
			1	1	64	7 to 8	1 to 2	2 periods		
	CTP CTC		<ul> <li>CTP = 0: When a count reset is executed (CTE = 0), the general-purpose counter input is pulled down.</li> <li>= 1: When a count reset is executed (CTE = 0), the general-purpose counter input is not pulled down, and the wait time is reduced. However, immediately after CTP = 1 is set, the start of the count must wait until the general purpose counter input pin is biased.</li> <li>The input sensitivity is reduced when CTC = 1. (Sensitivity: 10 to 30 mVrms)</li> </ul>							
(7)	I/O port control data I/O-1 to I/0-2		"data"= =	0: Input 1: Outp	port ut port	·		r an output port.	OUT0 to OUT3 ULD	
		*			on reset, I/O-0 a		come input p	orts.		
(8)	SEEK SW	•	This da "data" = * This p "data" =	ta deterr 0: 2.5[\ in is ope 1: 0[V]	mines an output p mines the status /] output en and the midpo or 5[V] output / the OUT0 data	of the SEEK bint bias is ou	·	xternal circuit.	I/O-0 to I/O-3 ULD	
(9)	SDST SW	•	<ul> <li>AM/FM SD, FM-ST IND output dual-purpose pin</li> <li>"data" = 0: Fixed</li> <li>= 1: Prohibited</li> </ul>							
(10)	Output port data OUT0 to OUT2	•	<ul> <li>This data determines the output on output ports O-0 through O-3.</li> <li>"data" = 1: Open or Hi         <ul> <li>= 0: Low</li> <li>* Invalid if specified as an input port or unlocked output.</li> </ul> </li> </ul>							
(11)	General-purpose counter input control data HCTR	•	This dat HCTR =	ta conve = 0: Proh	erts the general-p	ourpose cour	nter pin to an	input port.	CTS1	

## Continued from preceding page.

No.	Control block/Data			Des	cription	Related data
	Unlock detection data	This data PLL lock If a phas generate the dete				
		UL1	DT0	øE detection width	Detection pin output	
(12)	UL1, UL0	0 0 1 1	0 1 0 1	Stop 0 ±0.5 μs ±1 μs	Open Direct output of øE Extend øE by 1 to 2 ms Extend øE by 1 to 2 ms	ULD DT0, DT1
		E DO				
		. This is A			A13316	
	Crystal oscillator circuit	• This is th XS1	XS0	al oscillator selection X'tal OSC		
		0	0	Prohibited		
		0	1	Prohibited		
(13)	XS0, XS1	1	0	10.25 MHz 10.35 MHz		R0 to R3
	ХВ	• Crystal o XB = 0: XB = 1:	bscillato Buffer o execute Buffer o	or buffer (XBUF) output output: OFF (This moded.) output ON	d, 10.25 MHz is selected. ut control data. de is selected when a power-on reset is in), XBUF output must be off.	
	Phase comparator control	• This dat	a contro	ols the phase compar	ator dead zone.	
(14)	data DZ0, DZ1	DZ1 0 0 1 1	DZ0 0 1 0 1	Dead zone mode DZA DZB DZC DZD		
	Charge nump control data			on reset is executed,	pump output to the low level (VSS level).	
(15)	Charge pump control data DLC	<ul> <li>This data</li> <li>DLC =</li> <li>=</li> <li>* If a deaconstruction</li> <li>oscillation</li> <li>charge p</li> <li>When a</li> </ul>				
(16)	IC test data TEST0 TEST1 TEST2	TE TE	t TEST( ST1 = 0 ST2 = 0	) = 0. )	d, all the test data is set to zero.	

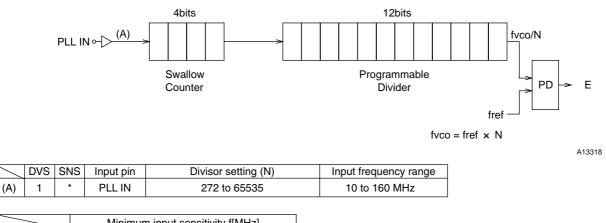
## DO Output Data (Serial Data Output) Configuration

[3] OUT mode



No.	Control block/Data	Description	Related data
(1)	I/O port data I3 to I0	<ul> <li>I0 to I3 is the latched data reflecting the status of the input ports: I/O-0 to I/O-3. The data is latched at the point that data output mode is set. The pin status is latched regardless of the input/output specification. Pin status = Hi: 1 Low: 0</li> </ul>	I/O-1 to I/O-2 SEEK SW HCTR
(2)	General-purpose counter binary data C19 to C0	<ul> <li>C19 to C0 is the latched data reflecting the contents of the general-purpose counter (a 20-bit binary counter).</li> <li>C19 ← MSB of binary counter</li> <li>C0 ← LSB of binary counter</li> </ul>	CTS0 CTS1 CTE
(3)	A/D converter ADC0 data AD05 to AD00	<ul> <li>AD05 to AD00 is the latched data reflecting the results when the ADC0 pin input signal undergoes AD conversion.</li> <li>AD05 ← MSB</li> <li>AD00 ← LSB</li> </ul>	ADI1 ADS

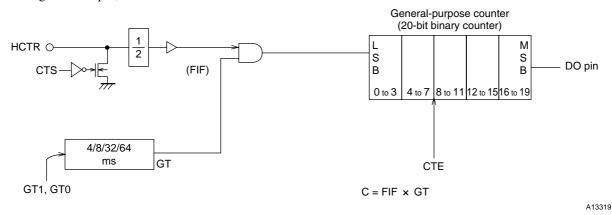
#### **Programmable Divider Configuration**



	Minimum input sensitivity f[MHz]						
	10 ≤ f < 130	130 ≤ f <160					
(A) PLL IN	40 mVrms	70 mVrms					

#### **General-purpose Counter Configuration**

In the LA17000M, the general-purpose counter consists of a 20-bit binary counter. The count results can be read through the DO pin, MSB first.



When using the general-purpose counter for cycle measurement, the measurement period can be selected from among 4, 8, 32, and 64 ms through the GT0 and GT1 data. The cycle of the signal that is input to the HCTR pin or the LCTR pin can then be measured by counting the number of pulses that are input to the general-purpose counter within this measurement period.

When using the general-purpose counter to measure a cycle, it is also possible to measure the cycle of a signal that is input to the LCTR pin according to the number of check signals (refer to the "Check Signal Frequency" table below) input to the general-purpose counter within one or two cycles of the signal that is input to the LCTR.

Check	Signal	Freq	uency
-------	--------	------	-------

X tol OSC 10.25 MHz			10.3	35 MI	Hz				
× tai 050	X tal OSC 10.25 MHz			fref = 30, 9, 3 kHz fref other than 30, 9, 3 kHz		κHz			
Check sigr	nal	10.2	5 kHz		1030 kHz		1150 kHz		
	C	CTS1	Inpu	t pin	in Measurement mod		Frequency range	Inp	out sensitivity
					_				

 S1
 1
 HCTR
 Frequency
 0.4 to
 25.0 MHz
 40 mVrms\*1

 \*1 CTC = 0: 40 mVrms; however, when CTC = 1, the frequency range is HCTR: 8 to 12 MHz

CTC = 1: 70 mVrms

CTC data: This is the input sensitivity switch data; when CTC = 1, the input sensitivity is degraded.

	HCTR: Minimum input sensitivity standard f [MHz]		
CTC	0.4 ≤ f < 8	8 ≤ f < 12	12 ≤ f < 25
0 (Normal mode)	40 mVrms	40 mVrms (1 to 10 mVrms)	40 mVrms
1 (Degraded mode)	_	70 mVrms (30 to 40 mVrms)	_

-: Not stipulated (operation not guaranteed)

(): Actual performance estimates (reference value)

CTP data: This is data that determines the status of the general-purpose counter input pin (HCTR/LCTR) when a general-purpose counter reset (CTE = 0) is executed.

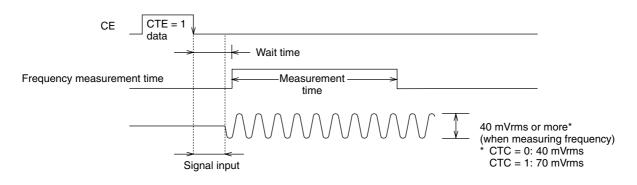
CTP = 0: Pulls down the general-purpose counter input pin.

= 1: Does not pull down the general-purpose counter input pin, reducing the wait time to 1 or 2 ms. When setting CTP = 1, do so at least 4 ms prior to starting the count (CTE = 1). If the counter is not to be used, set CTP = 0.

			Frequency measurement mode			Cycle
0	GT1 GT0		Measurement	Wait time		measurement
			time	CTP = 0	CTP = 1	mode
	0	0	4 ms	3 to 4 ms		1 0/010
	0	1	8	3 10 4 1115	1 to 2 ms	1 cycle
	1	0	32	7 to 8 ms	1 10 2 1115	
	1	1	64	7 10 8 1115		2 cycles

#### **IF Counter Operation**

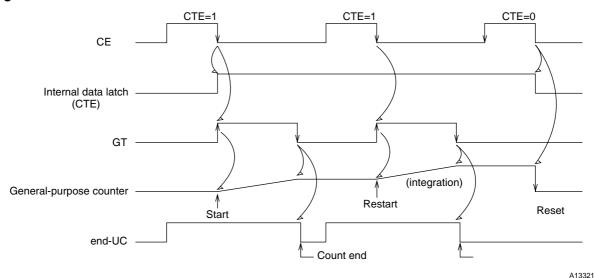
Before starting counting with the general-purpose counter, the general-purpose counter must first be reset by setting CTE = 0. The general-purpose counter is made to start counting by setting serial data CTE = 1. The serial data is finalized within the PLL by changing CE from high to low, but input to the HCTR pin must be started within the wait period after CE is sent low at the very latest. After measurement ends, the count results from the general-purpose counter must be read while CTE = 1. (Once CTE is set to zero, the general-purpose counter is reset.) Furthermore, the signal that was input to the HCTR pin is passed through to the general-purpose counter after having been divided by 1/2 internally. Therefore, the general-purpose count results are actually 1/2 the actual frequency of the signal that was input to the HCTR pin.



A13320



#### **Integrated Count**



\* CTE:  $0 \rightarrow$  • General-purpose counter reset

 $1 \rightarrow$  • General-purpose counter start

• Setting to "1" again causes a restart.

When using integrated counting, the count value is accumulated in the general-purpose counter.

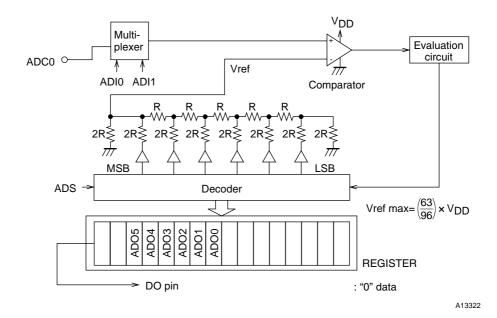
Be careful about counter overflows.

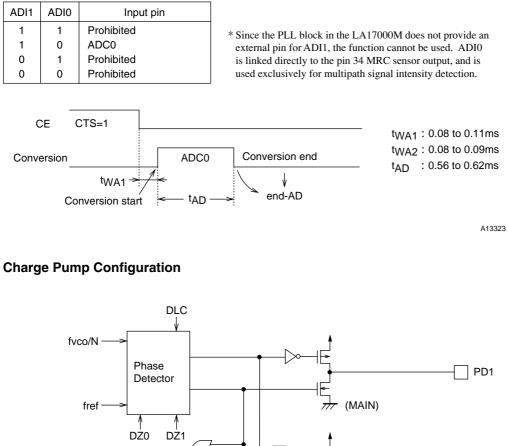
Count value: 0H to FFFFFH (1048575)

When using integrated counting, resending serial data (IN1) with CTE = 1 restarts measurement with the general-purpose counter, and the count results are added to the previous count results.

#### A/D Converter Configuration

This is a 6-bit successive-approximation converter with a conversion time of 0.56 ms. Full scale (when the data is 3FH) is (63/96) x VDD.

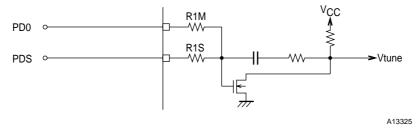




fref Detector fref DZ0 DZ1 Unlock Detector and Subcharge PL1 (MAIN) PDS PDS PDS A1324

PDC1	PDC0	PDS (sub-charge pump status)		DLC	PD1, PDS
0	*	High impedance		0	Normal operation
1	0	Charge pump on (when unlocked)		1	Forced low
1	1	Charge pump on (at all times)			1

If the unlocked state is detected during a channel change, the PDS (sub-charge pump) operates,  $R1 \leftarrow R1M/R1S$ , the low-pass filter time constant is reduced, and lockup is accelerated.



\* Unlock detection data: UL1 = 1 must be set. This sets the unlock detection width to " $\pm 0.5 \,\mu$ s" or " $\pm 1 \,\mu$ s" mode; if a phase difference that is greater than the value in question is detected, the signal is unlocked and the sub-charge pump operates. As the locked condition is approached and the phase difference falls to less than the unlocked detection width, the sub-charge pump stops operating (goes to high impedance).

#### **Other Items**

[1] Notes on the Phase Comparator Dead Zone

DZ1	DZ0	Dead zone mode	Charge pump	Dead zone
0	0	DZA	ON/OFF	<i>– –</i> 0 s
0	1	DZB	ON/ON	–0 s
1	0	DZC	OFF/OFF	+0 s
1	1	DZD	OFF/OFF	++0 s

Since correction pulses are output from the charge pump even if the PLL is locked when the charge pump is in the ON/ON state, the loop can easily become unstable. This point requires special care when designing application circuits.

The following problems may occur in the ON/ON state.

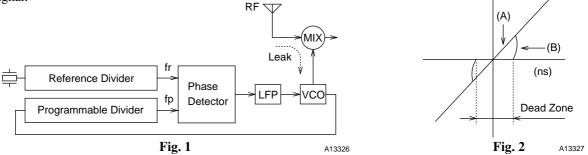
- Side band generation due to reference frequency leakage
- · Side band generation due to both the correction pulse envelope and low frequency leakage

Schemes in which a dead zone is present (OFF/ON) have good loop stability, but have the problem that acquiring a high C/N ratio can be difficult. On the other hand, although it is easy to acquire a high C/N ratio with schemes in which there is no dead zone, it is difficult to achieve high loop stability. Therefore, it can be effective to select DZA or DZB, which have no dead zone, in applications which require an FM S/N ratio in excess of 90 to 100 dB, or in which an increased AM stereo pilot margin is desired. On the other hand, we recommend selecting DZC or DZD, which provide a dead zone, for applications which do not require such a high FM signal-to-noise ratio and in which either AM stereo is not used or an adequate AM stereo pilot margin can be achieved.

#### Dead Zone

The phase comparator compares fp to a reference frequency (fr) as shown in Fig. 1. Although the characteristics of this circuit (see Fig. 2) are such that the output voltage is proportional to the phase difference  $\phi$  (line A), a region (the dead zone) in which it is not possible to compare small phase differences occurs in actual ICs due to internal circuit delays and other factors (line B). A dead zone as small as possible is desirable for products that must provide a high S/N ratio.

However, since a larger dead zone makes this circuit easier to use, a larger dead zone is appropriate for popularlypriced products. This is because it is possible for RF signals to leak from the mixer to the VCO and modulate the VCO in popularly-priced products in the presence of strong RF inputs. When the dead zone is narrow, the circuit outputs correction pulses and this output can further modulate the VCO and generate beat frequencies with the RF signal.



#### [2] Notes on the PLL IN and HCTR pins

Coupling capacitors must be placed extremely close to these pins. The capacitance should be about 100 pF. If a capacitor with a capacitance of 100 pF or less is not used with HCTR in particular, there will be a long wait until the bias level is reached, which may sometimes cause miscounting.

#### [3] Notes on IF counting

When using the general-purpose counter for IF counting, be certain to have the microcontroller determine whether the IF-IC SD (Station Detector) signal is present or not, and to turn on the IF count buffer output and conduct the count, but only if the SD signal is present. Conducting an auto search using only the IF count is not reliable, since there is a possibility of stopping even where there is no station due to leaked output from the IF count buffer.

[4] Using the DO pin

Aside from data output mode, the DO pin can also be used to check for the completion of counting by the generalpurpose counter, unlock detection output, and to check for changes in the input pins. It is also possible to input the status of the input pins (I/O-1, I/O-2) to the controller, unchanged, via the DO pin.

## [5] Cautions concerning the use of XBUF

When the XBUF output is on (AM up conversion is being used), the XBUF signal may leak to the adjacent pins (PD0, I/O-3), so do not use PD0 and I/O-3 for AM reception control. (Use the PD1 pin for the AM reception charge pump.) When using PD0 and I/O-3 for FM reception control, the XBUF output must be turned off (XB data = 0).

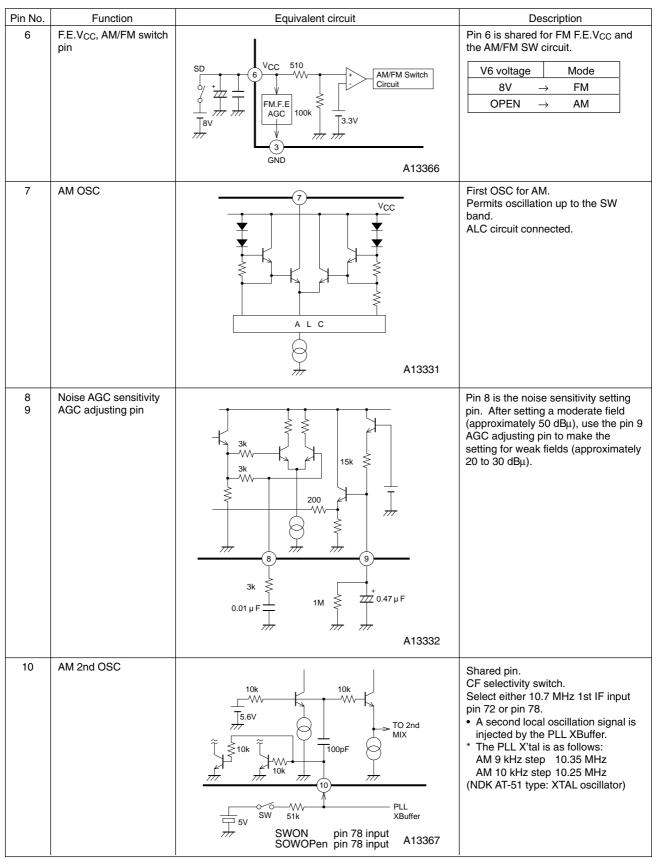
[6] Power supply pins

To filter out noise, insert a capacitor of at least 2000 pF between the power supply pins VDD and VSS. The capacitor must be located as close to the pins as possible.

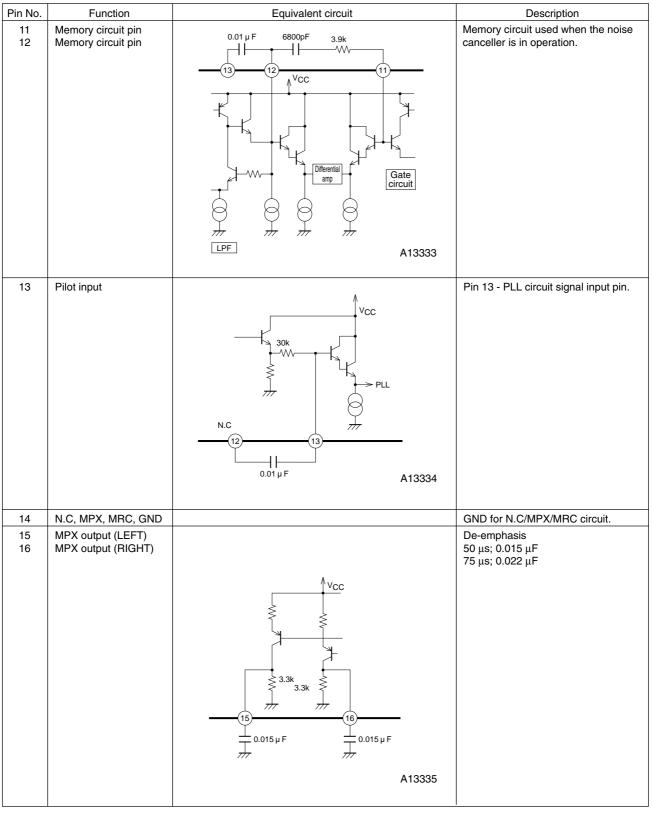
#### **Tuner Block Pin Description**

Pin No.	Function	Equivalent circuit	Description
1	Antenna damping drive pin.	ANT 1000pF 1000pF 1000pF 1000pF 1000pF 1000pF 1000pF AGC AGC AGC AGC AGC AGC AGC AGC	The antenna damping current flows to this pin when the pin 2: RF AGC voltage is $V_{CC}$ - $V_D$ .
2	RFAGC	FET 2ND GATE + + - - - - - - - - - - - - - - - - -	FET 2nd gate voltage control pin.
3	F.E.GND		
4	OSC	NT → TT →	OSC pin with built-in Tr. capacitor for oscillator circuit.

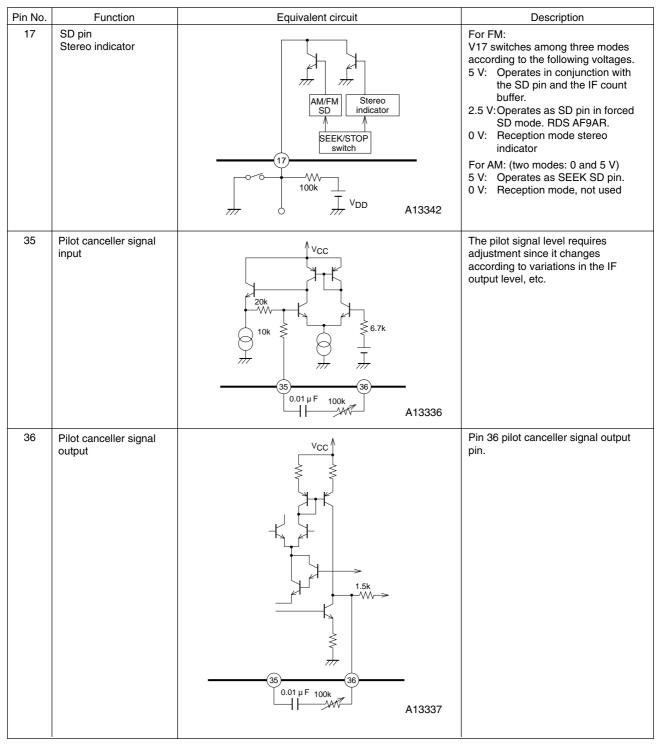
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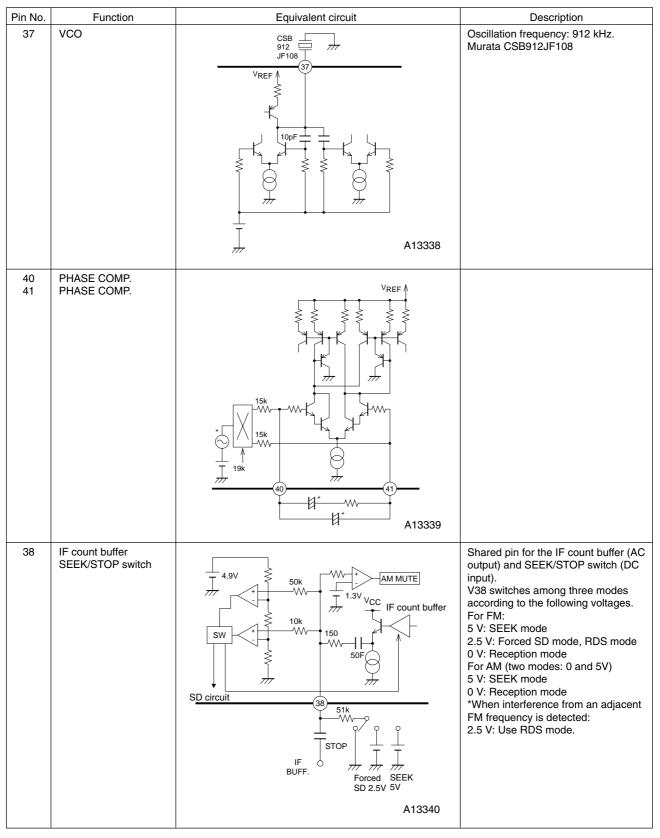


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LA17000M

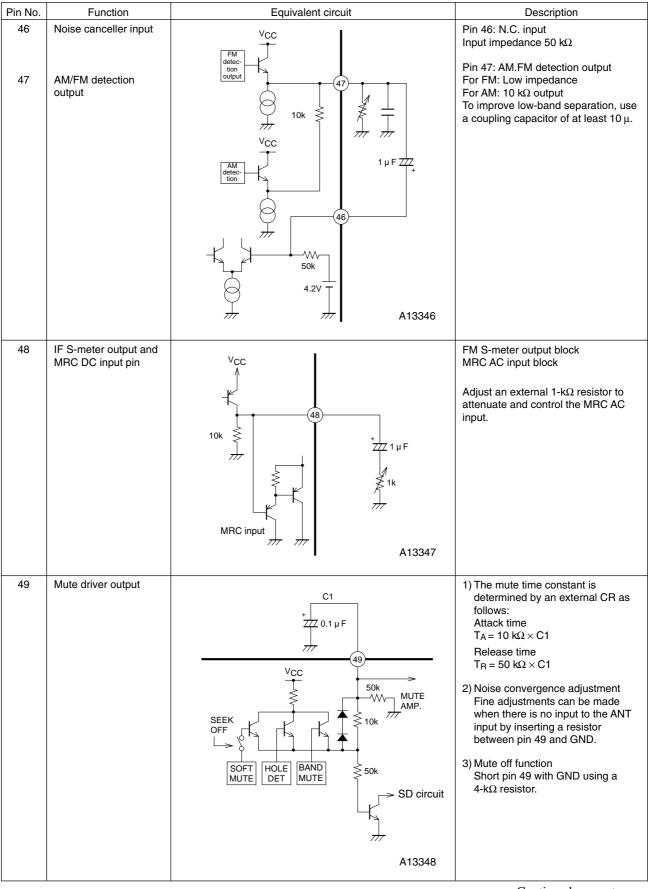
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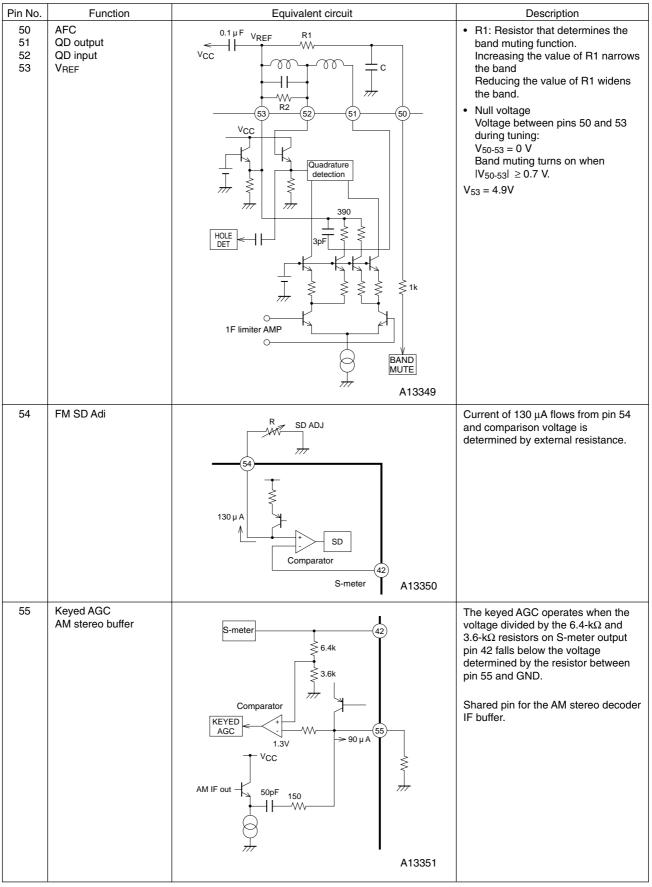
Pin No.	Function	Equivalent circuit	Description
42	AM/FM S-meter Dedicated FM S-meter	FM S-meter AM/FM Sweter 10k 10k MRC AM/FM W MRC A13341	Constant current drive-type S-meter output. When AM is set, pin 48 outputs a 1 mA current, which turns HCC OFF.
43	MRC control voltage time constant	VCC 7μA 500 5 5k to pin 44 A13343	The MRC detection time constant is determined 1 k $\Omega$ and C2 when discharging and by a constant current of 7 $\mu$ A and C2 when charging.
44	SNC control input pin	VCC + RA RA A13344	Controls sub-output with an input of 0 to 1 V. SNC voltage is determined by the RA and RB component voltage. This sets the separation blend curve. RB = 5 k $\Omega$ on chip RA is external
45	HCC control input pin	(48) (48) ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓	Controls high frequency output with an input of 0 to 1 V. Control through the MRC output is also possible. Use at least a 100 kΩ resistor when using pin 48 FM S-meter for control.

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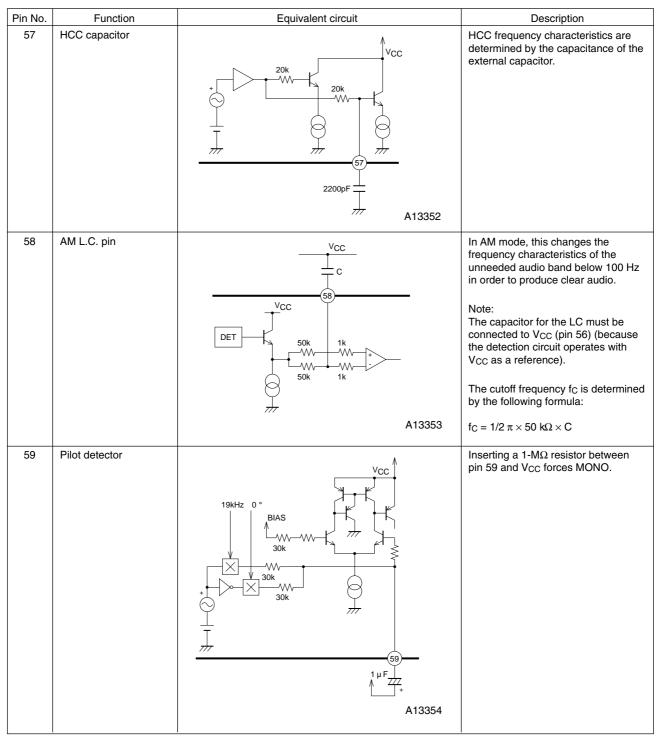


LA17000M

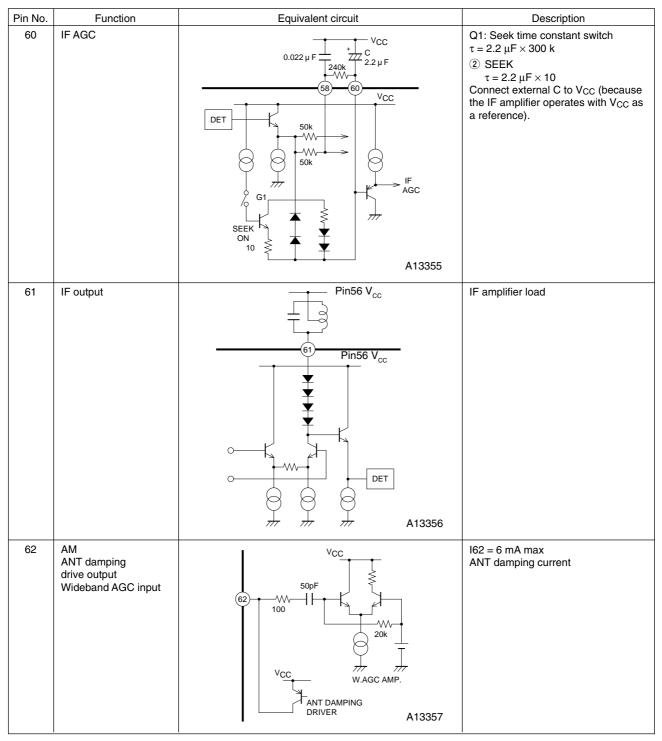
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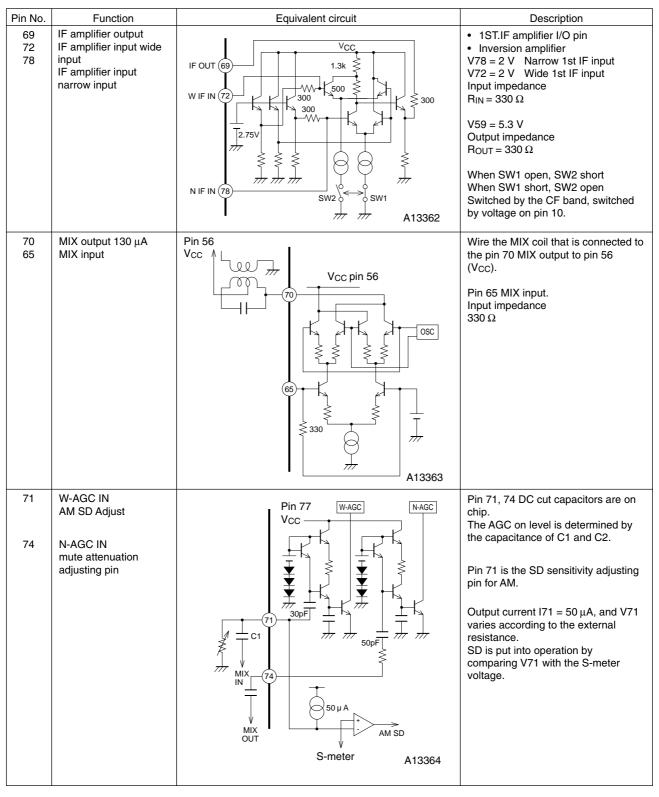
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Pin No.	Function	Equivalent circuit	Description
63	FM mute on Adjust	30k R m 24 μ/2 24 μ/2 R m (140 μ A) (140 μ	Vary the external resistor to adjust the mute on level.
64 73	RF AGC bypass RF AGC	For AGC $73$ $47 \mu F$ 777 $413359$	RF AGC rectification capacitor The distortion in low-frequency modulation is determined as follows. C64, C73 $\rightarrow$ Increase Distortion $\rightarrow$ Good Response $\rightarrow$ Slow C64, C73 $\rightarrow$ Decrease Distortion $\rightarrow$ Worsens Response $\rightarrow$ Fast
66	IF bypass		Be careful in regards to the GND
67	FM IF input	C1 0.022 µ F 330 IF in A13360	point for the limiter amplifier input C. Ground C1 at a point that does not increase AMR.
68	IF input		Input impedance 2 kΩ
		A13361	

#### Continued from preceding page.



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Pin No.	Function	Equivalent circuit	Description
75 76 80	MIX ouput MIX input	1ST.IF 0 S C 0 S C 0 VCC 0 VCCC 0 VCCC 0 VCCC 0 VCCC 0 VCCC 0 VCCC 0 VCCC 0 VCCC	Double-balance type mixer Pins 75 and 76, MIX output, 10.7 MHz output Pin 80, MIX input Emitter injection method and injection amount are determined by the values of C1 and C2. Note: The line for pin 80 must not approach pins 75 and 76.
79	1st MIX INPUT	to RF Amp. 79 10k 2.1V 77 10k 79 70 70 70 70 70 70 70 70 70 70 70 70 70	1st MIX input Input impedance: approximately 10 kΩ

### Methods for Using the LA17000M

# (1) About V<sub>CC</sub> and GND

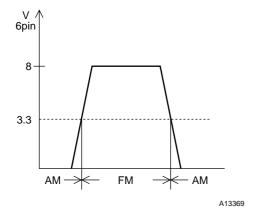
	-	
Pin 56	V <sub>CC</sub> for FM IF, AM, NC, MPX, and MRC	
Pin 39	GND for FM IF and AM	
Pin 14	GND for NC, MPX and MRC	
Pin 77	V <sub>CC</sub> for FM FE, AM 1st MIX, and 1st OSC	
*Pin 6	V <sub>CC</sub> for FM FE and AGC, and AM/FM switch	
Pin 3	GND for FM FE, AM 1st MIX, and 1st OSC	

#### (2) Notes on AM coil connection

 $V_{CC}$  for the 1st OSC coil that is connected to pin 7 should have the same electric potential as pin 77. Connect pin 61 IFT to pin 70 MIX coil.  $V_{CC}$  should have the same electric potential as pin 56.

(3) AM/FM switch

Pin 6 serves as FM, FE, and RFAGC V<sub>CC</sub>.



Pin 6 voltage	Mode
8	FM
OPEN	AM

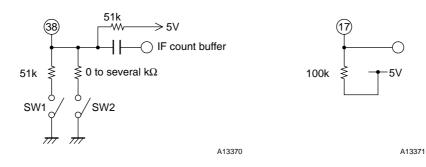
(4) Relationship between pin 38 and pin 17

4-1. For FM

Pin 17 STEREO indicator and SD dual-purpose pin

Pin 38 C DC input SEEK, STOP pin (control pin)

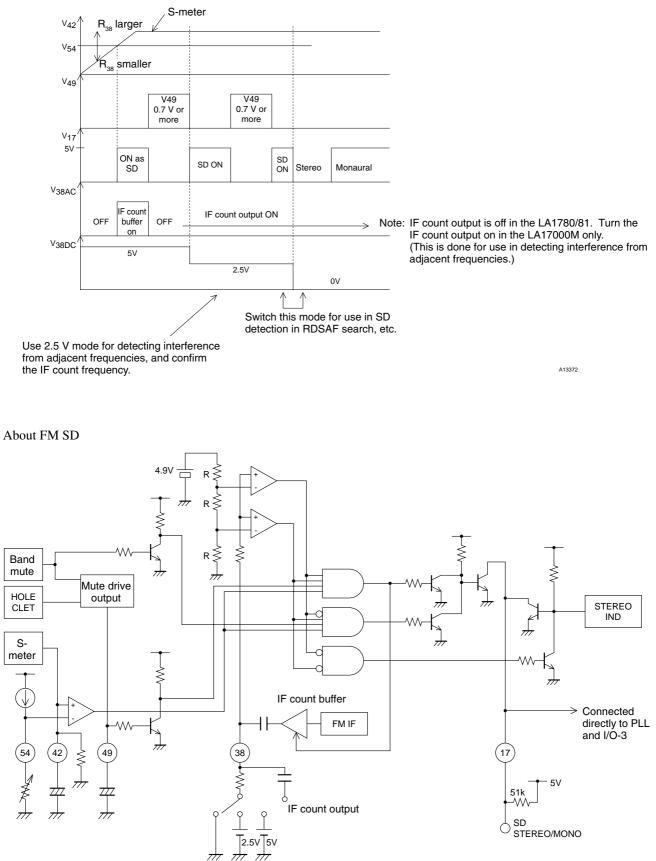
L AC output IF count buffer pin



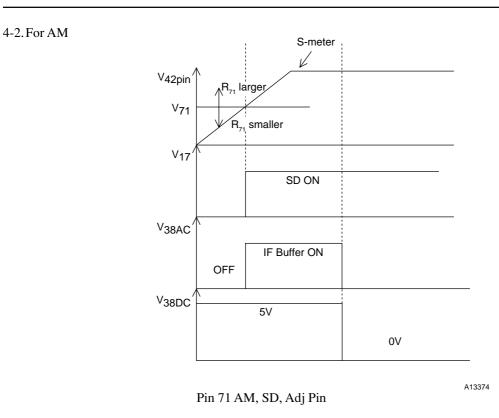
SW1	SW2	Pin 38 voltage	Pin 17	Pin 17
OPEN	OPEN	5 V	IF count buffer on	SD
ON	OPEN	2.5 V	IF count buffer on	High-speed SD
_	ON	0.7 V or less	OFF	Stereo indicator

Relationship Between Pin 38 Control Method and Output from Pins 38 and 17

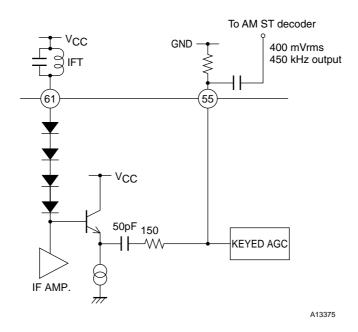




Relationship between FMSD, IF count buffer output, S-meter, and mute drive output



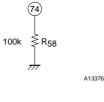
(5) AM STEREO support pin



• To attenuate the pin 55 AC level, add capacitance between GND and pin 55. For example, if pin 67 is added between GND and pin 55, the AM IF output decreases by about 6 dB.

### (6) About MUTE ATT

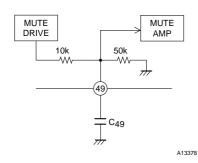
It is possible to switch to one of three levels (-20 dB, -30 dB, or -40 dB) by means of the resistor between pin 74 and GND. (This also has an effect on the total gain of the tuner.)



(49)



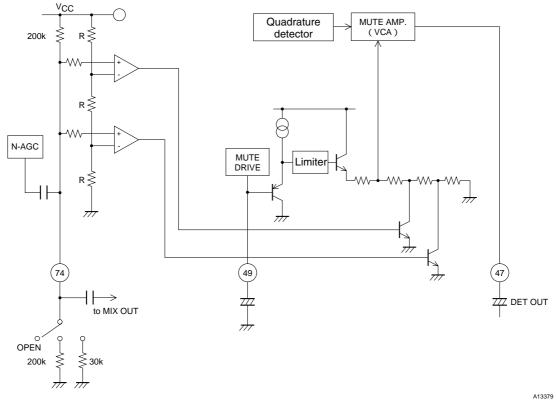
The attenuation can be reduced as shown in the table above by reducing R49.

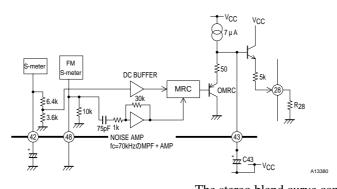


*≨* R<sub>49</sub>  $\overline{}$ 

A13377

MUTE time constant Attack  $10 \text{ k}\Omega \times \text{C49}$ Release 50 k $\Omega \times$  C49





The stereo blend curve can be adjusted through the R28 external resistor.

When there is no AC noise on pin 48
 V42 = V43-VBE
 ↑

QMRC

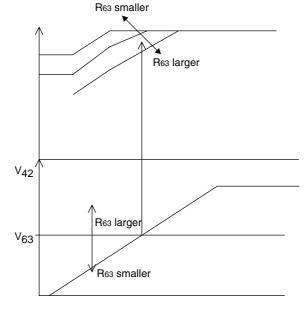
V43 is approximately 2.5 V when ANT input is 60 dB $\mu$  or higher.

2) Because the MRC noise amplifier gain is fixed, adjust MRC by reducing the AC input level.



- 3) The MRC attack and release are determined by C43 on pin 43. Attack 7  $\mu$ A C27 Release 500  $\Omega$  C27
- (8) FM soft mute

(7) MRC circuit

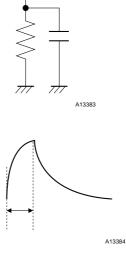


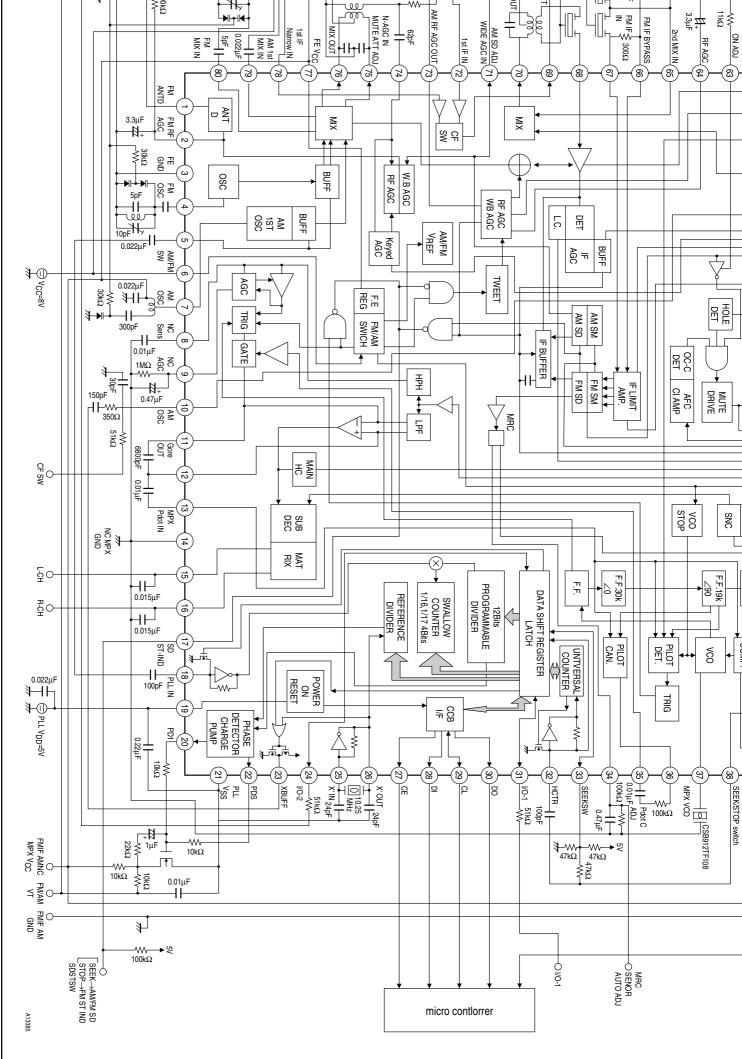
A13382

#### (9) About the noise canceller

The noise canceller improves the characteristics by implementing the circuits that determine the gate time with a logic circuit.

Because a conventional noise canceller determines the time constant according to CR as shown in Fig. 5, the rise time is dependent on the CR, as shown in Fig. 6. This caused a delay in the rise, which resulted in a deterioration of noise filtering performance when the rise was delayed too much. In the LA17000, the circuits that determine the gate time have been configured with logic, resulting in a faster rise and making more reliable noise filtering possible.





No. 6522-48/54

# **Recommended External Components**

Component name	Manufacturer	Component number	Component model number
AM loading coil	Toko Sumida Electronics Co., Ltd.	L1	7TL-269ANS-0720Z SA-1062
AM ANT-IN	Toko Sumida Electronics Co., Ltd.	L2	7PSU-385BNS-027Z SA-1048
AM RF LPF	Toko Sumida Electronics Co., Ltd.	L3	5VUS-A286LBIS-15327 SA-1051
AM choke coil	Toko Sumida Electronics Co., Ltd.	L4	8RB-187LY-222J RC875-222J
AM 2nd MIX coil	Toko Sumida Electronics Co., Ltd.	L7	5PG-5PGLC-5310N SA-264
AM IF coil	Toko Sumida Electronics Co., Ltd.	L8	7PSGTC-50002Y=S SA-1063/SA-1112
AM OSC1 coil	Toko Sumida Electronics Co., Ltd.	L9	7KSS-V666SNS-213BY SA-359
AM/FM MIX coil with selectivity switch	Toko	L10	7PSG-8261N-5202D=S
AM/FM MIX coil without selectivity switch	Sumida Electronics Co., Ltd. Toko	L10	SA-266 371DH-1108FYH
FM detection coil	Sumida Electronics Co., Ltd. Toko	L14	SA-208 DM600DEAS-8407GLF
FM OSC coil	Sumida Electronics Co., Ltd. Toko	L11	SA-125 (JP), SA-278 (US) T-666NF-251APZ (JP), T-666SNF-2471B (US)
FM RF coil	Sumida Electronics Co., Ltd. Toko	L12	SA-143 (JP), SA-250 (US) T-666NF-269X (JP), T-666SNF-246JA (US)
FM ANT coil	Sumida Electronics Co., Ltd. Toko	L13	SA-144 (JP), SA-231 (US) T-666NF-268Z (JP), T-666SNF-244X (US)
MPX ceramic oscillator	Murata Manufacturing Co., Ltd. Kyocera	VCO1	CSB912JF108 (912 kHz) KRB-912F108 (912kHz)
PLL X'tal oscillator	Nihon Dempa kogyo	VCO2	LN-P-0001 (10.25, 10.35 MHz)
FM ceramic filter	Murata Manufacturing Co., Ltd.	CF1	SFE 10.7MS3A50K-A
FM/AM narrow band ceramic filter	Murata Manufacturing Co., Ltd.	CF2	SFE 10.7 MTE
AM ceramic filter	Toko Murata Manufacturing Co., Ltd.	CF3	LFCM450H SFPS450H
AM pin diode	SANYO Electric Co., Ltd.	PIN1	1SV234/267
AMRF FET+TR	SANYO Electric Co., Ltd.	FET1	FC18
AM OSC1 varactor	SANYO Electric Co., Ltd.	VD2	SVC252/253
FM pin diode	SANYO Electric Co., Ltd.	PIN2	1SV234
FM RF amplifier FET	SANYO Electric Co., Ltd.	FET2	3SK263/264
FM RF/ANT/OSC varactor	SANYO Electric Co., Ltd.	VD3	SVC231/208

### **Crystal oscillator**

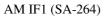
Nihon Dempa Kogyo Co., Ltd. Frequency: 10.25 MHz 10.35 MHz CL: 16pF 16pF Model name: LN-P-0001 LN-P-0001

# **Coil specifications**

Sumida Electronics Co., Ltd. [AM block]

AM FILTER (SA-1051)



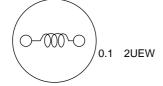




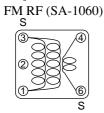
AM loading (SA-1062)



For AM RF amplifier (RC875-222J)



[FM block]



FM OSC (SA-1052)



FM DET (SA-208)



AM OSC (SA-359)



AM IF2 (SA-1063)



AM ANT IN (SA-1048)



# FM ANT (SA-1061)



(without selectivity switch)

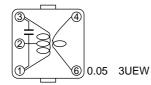
FM MIX (SA-266)



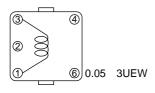
TOKO Co., Ltd. [AM block] AM FILTER (A286LBIS-15327)



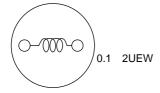
AM IF1 (7PSGTC-5001A=S)



AM loading (269ANS-0720Z)

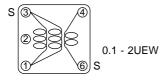


For AM RF amplifier (187LY-222)

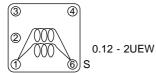


# [FM block]

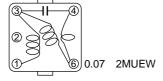
FM RF (V666SNS-208AQ)



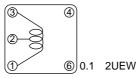
FM OSC (V666SNS-205APZ)



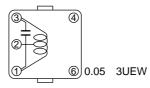
FM DET (DM600DEAS-8407GLF)



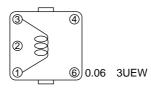
AM OSC (V666SNS-213BY)



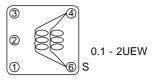
AM IF2 (7PSGTC-5002Y=S)



AM ANT IN (385BNS-027Z)



FM ANT (V666SNS-209BS)



### FM MIX (371DH-1108FYH)



(without selectivity switch) 0.07 - 2UEW

FM MIX (826IN-5202D=S)

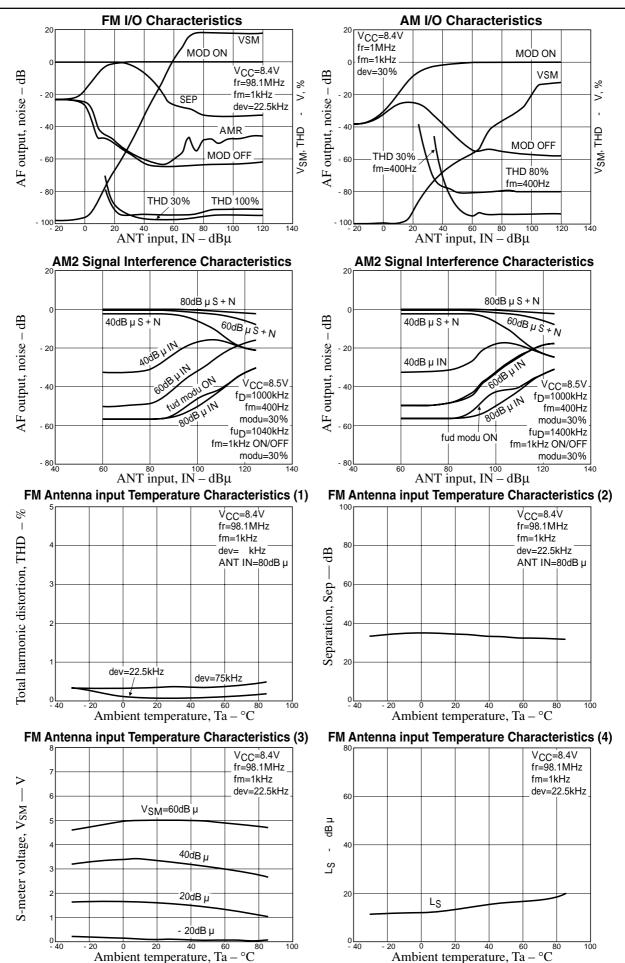
s

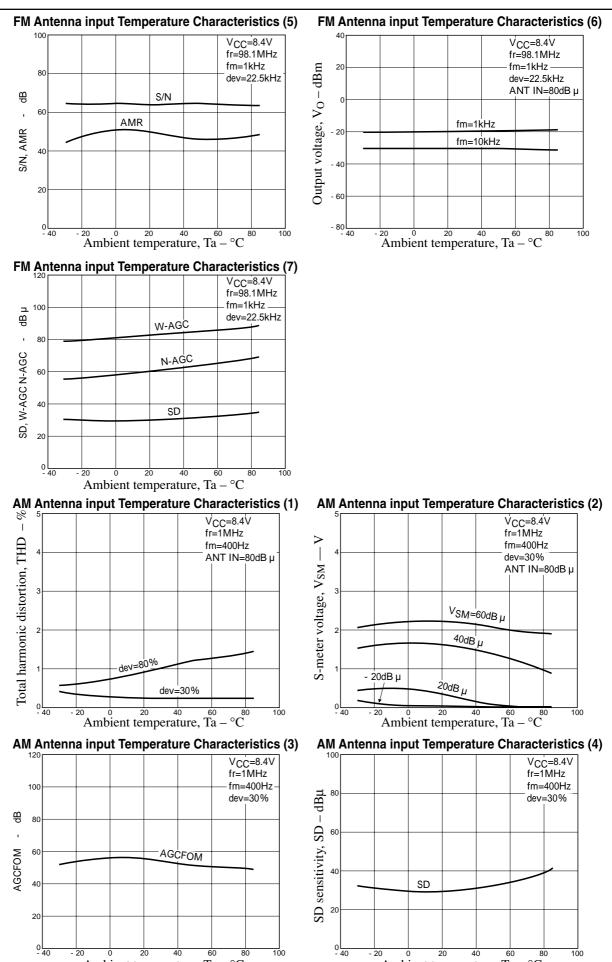


<sup>5</sup> (with selectivity switch) External 82P2 in parallel (1-G, 3-G)

0.062UEW

LA17000M

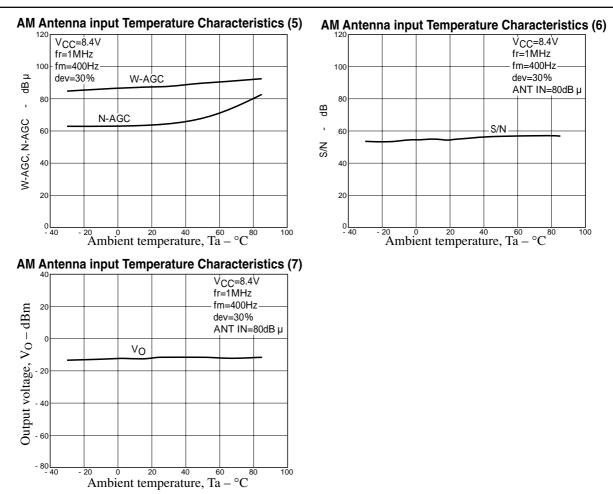




Ambient temperature, Ta - °C

Ambient temperature, Ta - °C





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