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### Monolithic Linear IC

#### **LA4581MB**

## **Preamplifier + Power Amplifier** for 3V Headphone Stereos

Package Dimensions

[LA4581MB]

## **Overview**

The LA4581MB is an auto reverse-supported preamplifier + power amplifier IC that is intended for use in 3V headphone WWW.DZSC.COM stereos.

W.DZSC.CO

## **Features**

- · Preamplifier muting and preamplifier output on/off can be implemented with one pin. This IC can easily be used to construct a set with a radio.
- · The power amplifier needs no input/output coupling capacitor.
- · A high-frequency cut capacitor is connected to the preamplifier input pin and the power amplifier input pin. (Anti-buzz provision)
- Because  $V_{ref}$  AMP ( $r_0 = 10 \Omega$ ) is built in, the virtual grounding impedance is about 10  $\Omega$ . This eliminates the need for a large capacitor.
- 8  $\Omega$  speaker drivable.

# Specifications

#### Maximum Ratings at $Ta = 25 \circ C$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V <sub>CC</sub> max		4.5	V
Allowable power dissipation	Pd max		530	mW
Operating temperature	Topr		-20 to +75	°C
Storage temperature	Tstg		-40 to +125	°C

unit : mm

3112-MFP24S

#### **Operating Conditions at Ta = 25 °C**

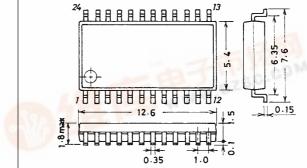
Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	V <sub>CC</sub>	64117	3.0	V
Operating supply voltage range	V <sub>CC</sub> op		1.8 to 3.6	V

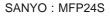
#### Operating Characteristics at Ta = $25 \circ C$ , $V_{CC} = 3.0 V$ , f = 1 kHz, 0.775 V = 0 dBm, $\mathbf{R}_{\mathrm{L}} = 10 \ \mathrm{k}\Omega$ (preamplifier), $\mathbf{R}_{\mathrm{L}} = 16 \ \Omega$ (power amplifier)

Parameter	Symbol	Output	min	typ	max	Unit
[Pre + Power]						
Quiescent current	Icco	$Rg = 2.2 \text{ k}\Omega$ (preamplifer) $V_{IN} = 0 \text{ V}$		17	27	mA
Voltage gain (Closed)	VGT	$V_{O} = -5 \text{ dBm}$	65	68	71	dB



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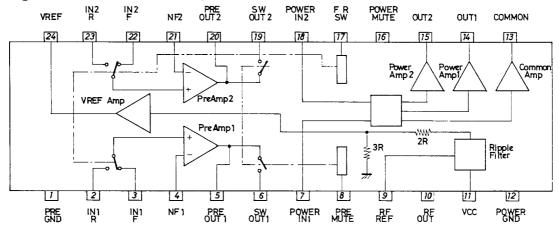
TOKYO OFFICE Tokyo Bldg., 1-10, 1 Chome, Ueno, Taito-ku, TOKYO, 110 JAPAN

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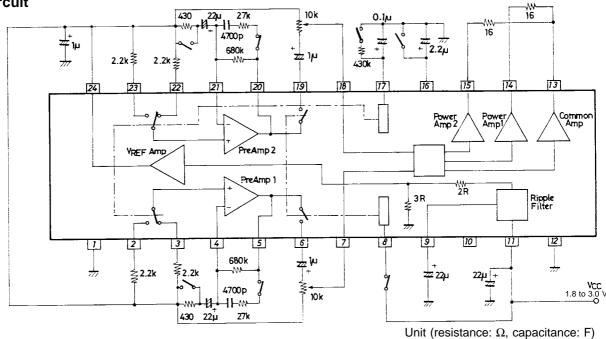
Parameter	Symbol	Output	min	typ	max	Unit
[Preamplifier]						-
Voltage gain (Open)	VGo	$V_{O} = -5 \text{ dBm}$	70	80		dB
Voltage gain (Closed)	VG1	$V_{O} = -5 \text{ dBm}$		40		dB
Maximum output voltage	V <sub>O</sub> max	THD = 1 %, V <sub>CC</sub> = 1.8 V	0.1	0.2		V
Total harmonic distortion	THD1	$V_{O} = 0.2 V, VG = 40 dB/NAB$		0.05	0.5	%
Equivalent input noise voltage	V <sub>N</sub> 1	Rg = 2.2 kΩ, B.P.F = 20 to 20 kHz		1.3	2.0	μV
Crosstalk	CT1	Rg = 2.2 kΩ, TUNE 1 kHz	60	80		dB
Ripple rejection ratio	R <sub>r</sub> 1	Rg = 2.2 kΩ, $V_{CC}$ = 1.8 V, Vr = -20 dBm, f = 100 Hz	40	50		dB
[Power Amplifer]						
Output voltage	Po	THD = 10%	23	32		mW
Voltage gain (Closed)	VG2	$V_{O} = -5 \text{ dBm}$	25	28	31	dB
Total harmonic distortion	THD2	P <sub>O</sub> = 1 mW		0.4	1.0	%
Interchannel crosstalk	CTT	$V_0 = -5 \text{ dBm}, R_v = 0 \Omega$	30	40		dB
Output noise voltage	V <sub>NO</sub>	Rg = 0, B.P.F = 20 to 20 kHz		24	40	μV
Ripple rejection ratio	R <sub>r</sub> 2	$Rg = 0, V_r = -20 \text{ dB}, f = 100 \text{ Hz}, V_{CC} = 1.8 \text{ V}$	45	60		dB
Input resistance	R <sub>IN</sub>		22	30	38	kΩ
DC offset voltage	V <sub>ODC</sub> off	Between 13-14 and 15	-90		+90	mV

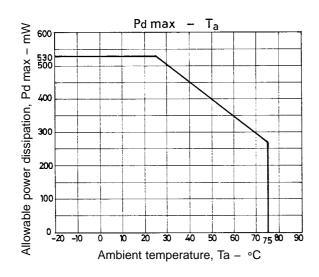
Note) Power amplifier voltage gain VG2 increases by about 1 dB for min/max respectively than specified above when  $R_L = 32 \Omega$ .

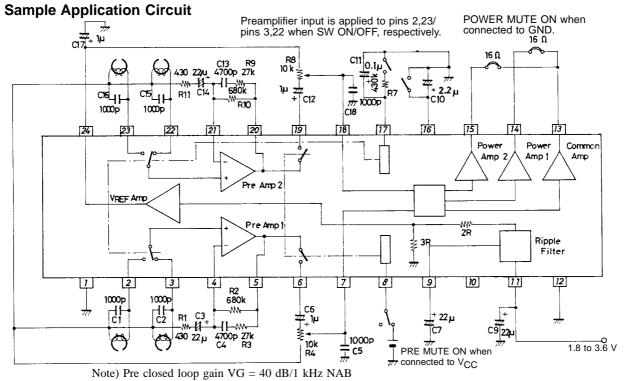
#### **Block Diagram**



## **Test Circuit**









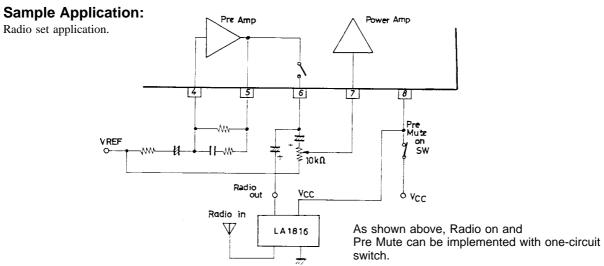


## Pin and external part functions (when the voltage is $V_{cc}$ = 3.0 V)

Pin No.	Pin Function
1	PRE GND
2	<ul> <li>PRE IN1R 1.8 V</li> <li>• Turns ON when pin 17 is grounded.</li> <li>• A bias resistor (2.2 kΩ) must be connected between pin 2 and pin 24 (V<sub>ref</sub>) when no head is in use.</li> </ul>
3	<ul> <li>PRE IN1F 1.8 V</li> <li>Turns ON when pin 17 is floating.</li> <li>A bias resistor (2.2 kΩ) must be connected between pin 3 and pin 24 (V<sub>ref</sub>) when no head is in use.</li> </ul>
4	PRE NF1 1.8 V
5	PRE OUT1 1.8 V • Like pin 6, 10 kΩ load drivable.
6	SW OUT1 1.8V • Provides PRE AMP1 output when pin 8 is floating (PRE MUTE OFF)(equivalent to pin 5). • Disconnects from PRE AMP1 and sets $R_{IN} \ge 500 \ k\Omega$ when pin 8 is at $V_{CC}$ (PRE MUTE ON).
7	POWER IN1 1.8V • Input resistance $R_{IN} = 30 \text{ k}\Omega$

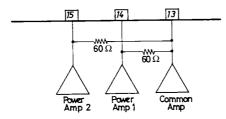
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Pin No.	Pin Function
8	PRE MUTE • When V <sub>CC</sub> is applied, PRE MUTE ON. • MUTE ON conditions: V <sub>8IN</sub> $\geq$ V <sub>CC</sub> - 0.2 V, inflow current I <sub>7</sub> $\doteqdot$ 60 µA (when V <sub>CC</sub> = 3 V)
9	Ripple Filter REF 2.7 V( $C_7 = 2.2 \ \mu$ F to 33 $\mu$ F) • Ripple Filter, V <sub>ref</sub> reference • The V <sub>ref</sub> ripple rejection ratio worsens when C <sub>7</sub> is made smaller. • R <sub>r</sub> is 55 dB for 22 $\mu$ F; 35 dB for 2.2 $\mu$ F.
10	Ripple Filter OUT 2.7 V • Ripple rejection ratio: $R_r$ is 38 dB when $C_7 = 22 \ \mu\text{F}$ ; 30 dB when $C_7 = 2.2 \ \mu\text{F}$ . • Outflow current $I_7$ max = 1 mA
11	V <sub>CC</sub> 3.0 V
12	POWER GND
13	COMMON 1.2 V
14	POWER OUT1 1.2 V • CH1 output.
15	POWER OUT2 1.2 V • CH2 output
16	POWER MUTE 0.7 V ( $C_{10}$ = 1.0 µF to 4.7 µF) • When connected to GND: POWER MUTE ON. • MUTE ON conditions: V <sub>16</sub> $\leq 0.3$ V, outflow current I <sub>16</sub> $\Rightarrow 2.5$ µA. • C <sub>10</sub> can be used to control MUTE TIME. • When C <sub>10</sub> = 2.2 µF, V <sub>CC</sub> = 3.0 V 0.7 sec.
17	FWD/REV SW (C11 $\leq 0.47 \ \mu$ F)• When connected to GND, PRE IN1R (pin 2) and IN2R (pin 23) turn on.• When floating, PRE IN1F (pin 3) and IN2F (pin 22) turn on.• C11 and R7 are intended for smoothing at the time of switching.• REV condition: V17 $\leq 0.2$ V.
18	POWER IN2 1.8 V • Input resistance $R_{IN} = 30 \text{ k}\Omega$
19	SW OUT2 1.8 V • Provides PRE AMP2 when pin 8 is floating (PRE MUTE OFF) (equivalent to pin 20). • Disconnects from PRE AMP2 and $R_{IN} \ge 500 \text{ k}\Omega$ when pin 8 is $V_{CC}$ (PRE MUTE ON).
20	PRE OUT2 1.8 V • Like pin 19, 10 kΩ load drivable.
21	PRE NF2 1.8V
22	<ul> <li>PRE IN2F 1.8 V</li> <li>• Turns on when pin 17 is floating.</li> <li>• A bias resistor (2.2 kΩ) must be connected between pin 22 and pin 24 (V<sub>ref</sub>) when no head is in use.</li> </ul>
23	<ul> <li>PRE IN2R 1.8 V</li> <li>• Turns on when pin 17 is connected to GND.</li> <li>• A bias resistor (2.2 kΩ) must be connected between pin 23 and pin 24(V<sub>ref</sub>) when no head is in use.</li> </ul>
24	$V_{ref}$ 1.8 V • The reference voltage is set to 3/5 × V <sub>CC</sub> . Because Vref AMP (r <sub>O</sub> $\Rightarrow$ 10 Ω) is built in, C <sub>17</sub> can be made smaller (1 µF) • Inflow/outflow current I <sub>24</sub> = ± 500 µA available.

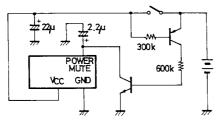


## **IC Usage Notes**

1. The power amplifier outputs and the common amplifier output are connected through resistors of about 60  $\Omega$ . The resistors are for common amplifier oscillation blocking.



- 2. The preamplifier muting function isolates the preamplifier outputs from SW OUT. The preamplifier is on even when the preamplifier muting is on.
- 3. If transient noise is noticeable when the power supply is turned off, add the external circuit described below. Transient noise when the power is turned off can be improved by rapidly applying the power amplifier muting.

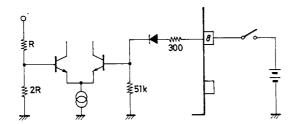


Unit (resistance:  $\Omega$ , capacitance: F)

Unit (resistance:  $\Omega$ )

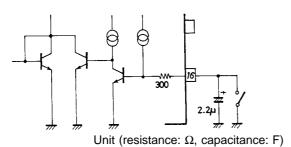
4. Internal equivalent circuit for each SW pin.

• Pre-mute



• Power mute

• F/R SW



Unit (resistance: Ω, capacitance: F)

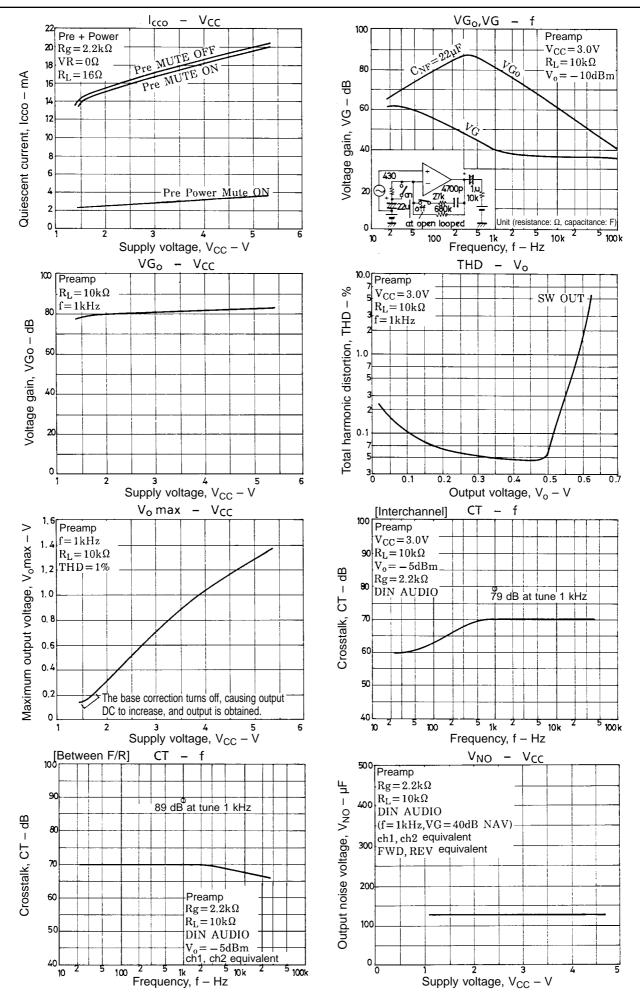
However, the standby current  $I_{ST}$  flows even when the power switch is off  $I_{ST} = (V_{CC} - V_{BE})/600 \ k\Omega$ When  $V_{CC} = 3.0 \ V$  $I_{ST} = (3.0 - 0.6)/600 \ k\Omega \doteqdot 4 \ \mu A$ 

MUTE ON condition :  $V_{8IN} \ge V_{CC} - 0.2 \text{ V}$ Inflow current :  $I_8 \doteq 60 \text{ }\mu\text{A}$  (when  $V_{CC} = 3.0 \text{ V}$ )

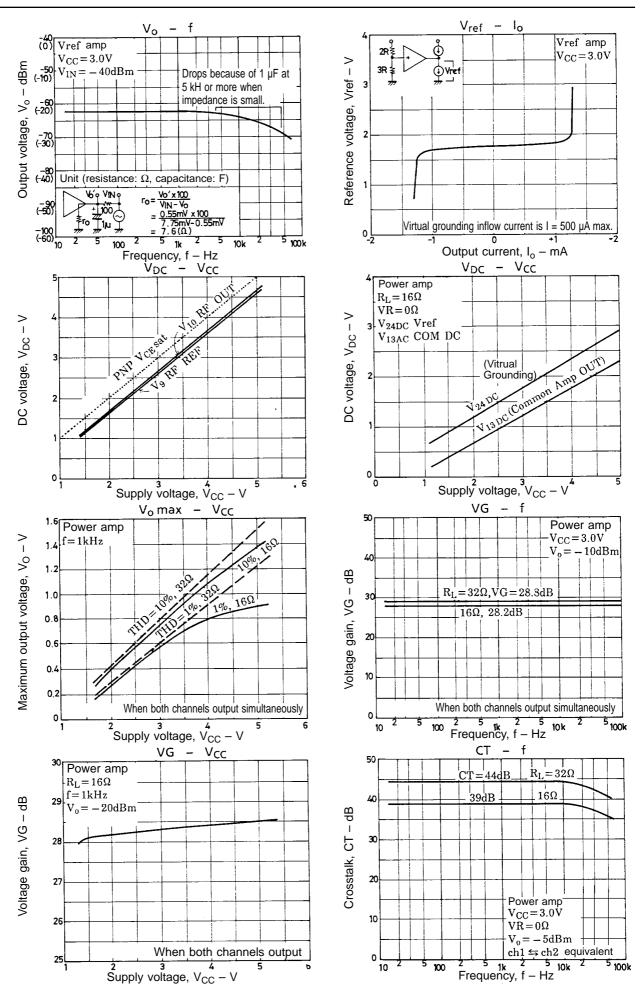
MUTE ON condition :  $V_{16} \leqq 0.3~V$  Outflow current :  $I_{16} \doteqdot 2.5~\mu A$ 

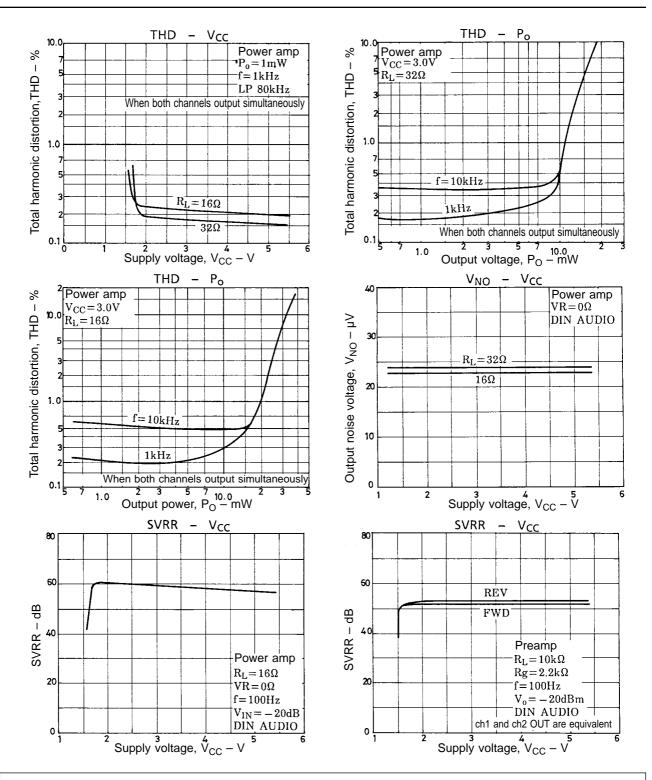
REV condition :  $V_{17} \leq 0.2 \text{ V}$ 

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