

## Overview

The LA9241M is an analog signal processing and servo control bipolar IC designed for use in compact disc players；a compact disc player can be configured by combining this IC with a CD－DSP such as the LC78622E and LC78620E，with a small number of additional components required．

## Functions

I／V amplifier，RF amplifier（with AGC），SLC，APC，FE，TE （with VCA and auto－balance function），focus servo amplifier （with offset cancellation function），tracking servo amplifier （with offset cancellation function），spindle servo amplifier （with gain switching function），sled servo amplifier（with off function），focus detection（DRF，FZD），track detection（HFL， TES），defect detection，and shock detection．

## Features

－The following automatic adjustment functions are built in．
－Focus offset auto cancel：FE（pin 20）
－Tracking offset auto cancel：TE（pin 7）
－EF balance auto adjustment
－RF level AGC function
－Tracking servo gain RF level following function
－Focus search smoothing setting pin：FSC（pin 46）
－EF balance adjustment variable range setting pin：（pin 47）
－Focus search mode switching pin：（pin 55）

## Package Dimensions

unit ：mm
3159－QFP64E


## Specifications

Maximum Ratings at $\mathbf{T a}=25^{\circ} \mathrm{C}$ ，Pins 22， $\mathbf{4 5}=\mathbf{G N D}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Maximum supply voltage | $\mathrm{V}_{\mathrm{CC}} \max$ | Pin 56,64 | 7 | V |
| Allowable power dissipation | Pd max |  | 350 | mW |
| Operating temperature | Topr |  | -25 to +75 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |

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Operating Conditions at Pins 22, $45=\mathbf{G N D}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Recommended supply voltage | $\mathrm{V}_{\mathrm{CC}}$ |  | 5 | V |
| Allowable operating supply voltage | $\mathrm{V}_{\mathrm{CC}}$ op |  | 3.2 to 5.5 | V |

## * Operating Supply Voltage at Limit of Operating Temperature at Pins 22, 45 = GND

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Operating temperature | Topr2 |  | -10 to +75 |  |
| Allowable operating supply voltage | $\mathrm{V}_{\mathrm{CC}}$ op2 |  | 3.0 to 5.5 |  |

Operating Characteristics at $\mathbf{T a}=\mathbf{2 5}^{\circ} \mathrm{C}$, Pins 22, $45=\mathbf{G N D}, \mathrm{V}_{\mathrm{CC}}($ pins 56, 64$)=5 \mathrm{~V}$

| Parameter | Symbol | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Current drain | ICCO | $\mathrm{V}_{\mathrm{CC}}{ }^{1}($ pin 64$)+\mathrm{V}_{\mathrm{CC}}{ }^{2}($ pin 56$)$ | 22 | 32 | 42 | mA |
| Reference voltage | $\mathrm{V}_{\text {REF }}$ | VR | 2.3 | 2.5 | 2.7 | V |
| [Interface] |  |  |  |  |  |  |
| CE-Vth | CEvth | CE |  | 0.8 |  | V |
| CL-Vth | CLvth | CL |  | 0.8 |  | V |
| DAT-Vth | DATvth | DAT |  | 0.8 |  | V |
| Maximum CL frequency | CLmax |  | 500 |  |  | kHz |
| [RF amplifier] |  |  |  |  |  |  |
| RFSM no signal voltage | RFSMo |  | 1.35 | 1.60 | 1.85 | V |
| Minimum gain | $\mathrm{RFSM}_{\mathrm{G}} \mathrm{min}$ | FIN1, FIN2 : $1 \mathrm{M} \Omega$-input, PH1 = 4 V freq $=200 \mathrm{kHz}$, RFSM | -14.0 | -12.5 | -11.0 | dB |
| [Focus amplifier] |  |  |  |  |  |  |
| FDO gain | $\mathrm{FD}_{\mathrm{G}}$ | FIN2 : $1 \mathrm{M} \Omega$-input, FDO | 3.5 | 5.0 | 6.5 | dB |
| FDO offset | FDost | Difference from reference voltage, servo on | -170 | 0 | +170 | mV |
| Off time offset | FDofost | Difference from reference voltage, servo off | -40 | 0 | +40 | mV |
| Offset adjustment step | FEstep | FE |  | 3 |  | mV |
| F search voltage H1 | FSmax1 | FDO, FSS = GND |  | 0.8 |  | V |
| F search voltage L1 | FSmin1 | FDO, FSS = GND |  | -0.8 |  | V |
| F search voltage H2 | FSmax2 | FDO, FSS $=\mathrm{V}_{\mathrm{CC}}$ |  | 0.8 |  | V |
| F search voltage L2 | FSmin2 | FDO, FSS $=\mathrm{V}_{\mathrm{CC}}$ |  | 0 |  | V |
| [Tracking amplifier] |  |  |  |  |  |  |
| TE gain MAX | TE ${ }_{\text {G }}$ max | $\mathrm{f}=10 \mathrm{kHz}, \mathrm{E}: 1 \mathrm{M} \Omega$-input, $\mathrm{PH} 1=4 \mathrm{~V}$ | 6.0 | 7.5 | 9.0 | dB |
| TE gain MIN | $\mathrm{TE}_{\mathrm{G}} \mathrm{min}$ | $\mathrm{f}=10 \mathrm{kHz}, \mathrm{E}: 1 \mathrm{M} \Omega$-input, $\mathrm{PH} 1=1 \mathrm{~V}$ | -0.5 | +1.8 | +4.0 | dB |
| TE-3 dB | TEfc | E: 1 M --input |  | 60 |  | kHz |
| TO gain | $\mathrm{TO}_{\mathrm{G}}$ | TH $\rightarrow$ TO gain, THLD mode | 4.0 | 6.0 | 8.0 | dB |
| TGL offset | TGLost | Servo on, TGL = H, TO | -250 | 0 | +250 | mV |
| TGH offset | TGHost | TGL = L, difference from TGL offset, TO | -50 | 0 | +50 | mV |
| THLD offset | THLDost | THLD mode, difference from TGL offset, TO | -50 | 0 | +50 | mV |
| Off 1 offset | OFF1ost | TOFF $=\mathrm{H}$ | -50 | 0 | +50 | mV |
| Off 2 offset | OFF2ost | TOFF2 off (IF) | -50 | 0 | +50 | mV |
| Offset adjustment step | TEstep | TE |  | 30 |  | mV |
| Balance range H | BAL-H | $\Delta$ Gain E/F input, TB $=5 \mathrm{~V}, \mathrm{TBC}=$ open |  | 3.5 |  | dB |
| Balance range L | BAL-L | $\Delta$ Gain E/F input, TB $=0 \mathrm{~V}, \mathrm{TBC}=$ open |  | -3.5 |  | dB |
| TOFF-VTH | TOFFvth |  | 1.0 | 2.5 | 3.0 | V |
| TGL-VTH | TGLvth |  | 1.0 | 2.5 | 3.0 | V |
| [PH] |  |  |  |  |  |  |
| No signal voltage | PHo | Difference from RFSM | -0.85 | -0.65 | -0.45 | V |
| [BH] |  |  |  |  |  |  |
| No signal voltage | BHo | Difference from RFSM | 0.45 | 0.65 | 0.85 | V |
| [DRF] |  |  |  |  |  |  |
| Detection voltage | DRFvth | Difference from VR at RFSM | -0.60 | -0.35 | -0.20 | V |
| Output voltage H | DRF-H |  | 4.5 | 4.9 |  | V |
| Output voltage L | DRF-L |  |  | 0 | +0.5 | V |

Continued from preceding page.

| Parameter | Symbol | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| [FZD] |  |  |  |  |  |  |
| Detection voltage 1 | FZD1 | FE, difference from VR | 0 | +0.2 |  | V |
| Detection voltage 2 | FZD2 | FE, difference from VR |  | 0 |  | V |
| [HFL] |  |  |  |  |  |  |
| Detection voltage | HFLvth | Difference from VR at RFSM | -0.55 | -0.4 | -0.25 | V |
| Output voltage H | HFL-H |  | 4.5 | 4.9 |  | V |
| Output voltage L | HFL-L |  |  | 0 | +0.5 | V |
| [TES] |  |  |  |  |  |  |
| Detection voltage LH | TES-LH | TESI, difference from VR | -0.15 | -0.10 | -0.05 | V |
| Detection voltage HL | TES-HL | TESI, difference from VR | 0.05 | 0.10 | 0.15 | V |
| Output voltage H | TES-H |  | 4.5 | 4.9 |  | V |
| Output voltage L | TES-L |  |  | 0 | +0.5 | V |
| [JP] |  |  |  |  |  |  |
| Output voltage H | JP-H | Difference from $\mathrm{JP}^{+}=0 \mathrm{~V}, \mathrm{JP}^{-}=0 \mathrm{~V}$ at $\mathrm{JP}^{+}=0 \mathrm{~V}$, $\mathrm{JP}^{-}=5 \mathrm{~V}, \mathrm{TO}$ | 0.35 | 0.5 | 0.65 | V |
| Output voltage L | JP-L | Difference from $\mathrm{JP}^{+}=0 \mathrm{~V}, \mathrm{JP}^{-}=0 \mathrm{~V}$ at $\mathrm{JP}^{+}=5 \mathrm{~V}$, $\mathrm{JP}^{-}=0 \mathrm{~V}, \mathrm{TO}$ | -0.65 | -0.5 | -0.35 | V |
| [Spindle amplifier] |  |  |  |  |  |  |
| Offset 12 | SPD12ost | Difference from VR at SPD, 12 cm mode | -40 | 0 | +40 | mV |
| Offset 8 | SPD8ost | Difference from VR at SPD, 8 cm mode | -40 | 0 | +40 | mV |
| Offset off | SPDof | Difference from VR at SPD, OFF mode | -30 | 0 | +30 | mV |
| Output voltage H12 | SPD-H12 | Difference from offset-12, 12 cm mode $\mathrm{CV}^{+}=5 \mathrm{~V}^{-\mathrm{CV}^{-}}=0 \mathrm{~V}$ | 0.75 | 1.0 | 1.25 | V |
| Output voltage L12 | SPD-L12 | Difference from offset-12, 12 cm mode $\mathrm{CV}^{+}=0 \mathrm{~V}^{-\mathrm{CV}^{-}}=5 \mathrm{~V}$ | -1.25 | -1.0 | -0.75 | V |
| Output voltage H8 | SPD-H8 | Difference from offset-8, 8 cm mode $\mathrm{CV}^{+}=5 \mathrm{~V}, \mathrm{CV}^{-}=0 \mathrm{~V}$ | 0.35 | 0.5 | 0.65 | V |
| [Sled amplifier] |  |  |  |  |  |  |
| SLEQ offset | SLEQost | Difference from TO at SLEQ | -30 | 0 | +30 | mV |
| Offset SLD | SLDost | SLEQ = VR, Difference from VR | -100 | 0 | +100 | mV |
| Offset off | SLDof | Off mode | -40 | 0 | +40 | mV |
| Off VTH | SLOFvth | SLOF | 1.0 | 1.4 | 2.0 | V |
| [SLC] |  |  |  |  |  |  |
| No signal voltage | SLCo | SLC | 2.25 | 2.5 | 2.75 | , |
| [Shock] |  |  |  |  |  |  |
| No signal voltage | SClo | SCI, difference from VR | -40 | 0 | +40 | mV |
| Detection voltage H | SClvthH | SCI, difference from VR | 60 | 100 | 140 | mV |
| Detection voltage L | SClvthL | SCI, difference from VR | -140 | -100 | -60 | mV |
| [DEF] |  |  |  |  |  |  |
| Detection voltage | DEFvth | Difference between LF2 voltage when RFSM = 3.5 V and DEF is detected and LF2 voltage when RFSM $=3.5 \mathrm{~V}$ | 0.20 | 0.35 | 0.50 | V |
| Output voltage H | DEF-H |  | 4.5 | 4.9 |  | V |
| Output voltage L | DEF-L |  |  | 0 | +0.5 | V |
| [APC] |  |  |  |  |  |  |
| Reference voltage | LDS | LDS voltage at which LDD $=3 \mathrm{~V}$ | 160 | 190 | 220 | mV |
| Off voltage | LDDof | LDD | 3.9 | 4.3 | 4.6 | V |

## Pin Functions

| Pin No. | Symbol | Contents |
| :---: | :---: | :---: |
| 1 | FIN2 | Pickup photodiode connection pin. Added to FIN1 pin to generate the RF signal, subtracted from FIN1 pin to generate the FE signal. |
| 2 | FIN1 | Pickup photodiode connection pin. |
| 3 | E | Pickup photodiode connection pin. Subtracted from F pin to generate the TE signal. |
| 4 | F | Pickup photodiode connection pin. |
| 5 | TB | TE signal DC component input pin. |
| 6 | TE ${ }^{-}$ | Pin which connects the TE signal gain setting resistor between this pin and TE pin. |
| 7 | TE | TE signal output pin. |
| 8 | TESI | TES (Track Error Sense) comparator input pin. The TE signal is input through a bandpass filter. |
| 9 | SCI | Shock detection input pin. |
| 10 | TH | Tracking gain time constant setting pin. |
| 11 | TA | TA amplifier output pin. |
| 12 | TD ${ }^{-}$ | Pin for configuring the tracking phase compensation constant between the TD and VR pins. |
| 13 | TD | Tracking phase compensation setting pin. |
| 14 | JP | Tracking jump signal (kick pulse) amplitude setting pin. |
| 15 | TO | Tracking control signal output pin. |
| 16 | FD | Focusing control signal output pin. |
| 17 | FD ${ }^{-}$ | Pin for configuring the focusing phase compensation constant between the FD and FA pins. |
| 18 | FA | Pin for configuring the focusing phase compensation constant between the $\mathrm{FD}^{-}$and $\mathrm{FA}^{-}$pins. |
| 19 | $\mathrm{FA}^{-}$ | Pin for configuring the focusing phase compensation constant between the FA and FE pins. |
| 20 | FE | FE signal output pin. |
| 21 | $\mathrm{FE}^{-}$ | Pin which connects the FE signal gain setting resistor between this pin and FE pin. |
| 22 | AGND | Analog signal GND. |
| 23 | NC | No connection |
| 24 | SP | $\mathrm{CV}^{+}$and $\mathrm{CV}^{-}$pins input signal single-end output. |
| 25 | SPG | $12-\mathrm{cm}$ spindle mode gain setting resistor connection pin. |
| 26 | $\mathrm{SP}^{-}$ | Spindle phase compensation constant connection pin, along with the SPD pin. |
| 27 | SPD | Spindle control signal output pin. |
| 28 | SLEQ | Sled phase compensation constant connection pin. |
| 29 | SLD | Sled control signal output pin. |
| 30 | SL- | Input pin for sled movement signal from microprocessor. |
| 31 | SL+ | Input pin for sled movement signal from microprocessor. |
| 32 | $\mathrm{JP}^{-}$ | Input pin for tracking jump signal from DSP. |
| 33 | $\mathrm{JP}^{+}$ | Input pin for tracking jump signal from DSP. |
| 34 | TGL | Input pin for tracking gain control signal from DSP. Gain is low when TGL is high. |
| 35 | TOFF | Input pin for tracking off control signal from DSP. Tracking servo is off when TOFF is high. |
| 36 | TES | Output pin for TES signal to DSP. |
| 37 | HFL | The High Frequency Level is used to determine whether the main beam is positioned over a bit or over the mirrored surface. |
| 38 | SLOF | Sled servo off control input pin |
| 39 | $\mathrm{CV}^{-}$ | Input pin for CLV error signal from DSP. |
| 40 | $\mathrm{CV}^{+}$ | Input pin for CLV error signal from DSP. |
| 41 | RFSM | RF output pin. |
| 42 | RFS ${ }^{-}$ | RF gain setting and EFM signal 3T compensation constant setting pin, along with the RFSM pin. |
| 43 | SLC | Slice Level Control is an output pin that controls the data slice level used by the DSP for the RF waveform. |
| 44 | SLI | Input pin used by DSP for controlling the data slice level. |
| 45 | DGND | Digital system GND pin. |
| 46 | FSC | Focus search smoothing capacitor output pin. |
| 47 | TBC | Tracking Balance Control; EF balance adjustment variable range setting pin |
| 48 | NC | No connection |
| 49 | DEF | Disc defect detection output pin. |
| 50 | CLK | Reference clock input pin. 4.23 MHz signal from the DSP is input. |
| 51 | CL | Microprocessor command clock input pin. |

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Continued from preceding page.

| Pin <br> No. | Symbol |  |
| :---: | :--- | :--- |
| 52 | DAT | Microprocessor command data input pin. |
| 53 | CE | Microprocessor command chip enable input pin. |
| 54 | DRF | RF level detection output (Detect RF). |
| 55 | FSS | Focus Search Select; focus search mode ( $\pm$ search/+search vs. the reference voltage) switching pin |
| 56 | V $_{\text {CC } 2}$ | Servo system and digital system $\mathrm{V}_{\mathrm{CC}}$ pin. |
| 57 | REFI | By-pass capacitor connection pin for reference voltage. |
| 58 | VR | Reference voltage output pin. |
| 59 | LF2 | Disc defect detection time constant setting pin. |
| 60 | PH1 | RF signal peak hold capacitor connection pin. |
| 61 | BH1 | RF signal bottom hold capacitor connection pin. |
| 62 | LDD | APC circuit output pin. |
| 63 | LDS | APC circuit input pin. |
| 64 | $V_{\mathrm{CC}} 1$ | RF system $\mathrm{V}_{\mathrm{CC}}$ pin. |

## LA9241M

Equivalent Circuit Block Diagram


## Test Circuit



## Description of Operation

1. APC (auto laser power control)

This circuit controls the pickup laser power. The laser is turned on and off by commands from the microprocessor.
2. RF amplifier (eye pattern output)

The pickup photodiode output current $(\mathrm{A}+\mathrm{C})$ is input to FIN2 (pin 1), and $(\mathrm{B}+\mathrm{D})$ is input to FIN1 (pin 2). The current that is input is converted to the voltage, passes through the AGC circuit, and is then output from the RFSM amplifier output RFSM (pin 41). The internal AGC circuit has a variable range of $\pm 3 \mathrm{~dB}$, and the time constant can be changed through the external capacitor connected to PH1 (pin 60). In addition, this circuit also controls the bottom level of the EFM signal (RFSM output), and the response can be changed through the external capacitor connected to BH1 (pin 61). The center gain setting for the AGC variable range is set by the resistance between RFSM (pin 41) and $\operatorname{RFS}^{-}$(pin 42); if necessary, this resistance is also used for 3 T compensation for the EFM signal.
3. SLC (slice level control)

The SLC sets the duty ratio for the EFM signal that is input to the DSP to $50 \%$. The DC level is determined by integrating the EFMO signal output from the DSP to determine the duty factor.
4. Focus servo

The focus error signal is derived by detecting the difference between $(A+C)$ and $(B+D)$, which is $(B+D)-(A+C)$, and is then output from FE (pin 20). The focus error signal gain is set by the resistance between FE (pin 20) and $\mathrm{FE}^{-}$(pin 21). Offset cancellation is performed by the FE amplifier. "Offset cancellation" cancels the offset for the IC's internal IV amplifier, etc. Adjustment is initiated by the FOCUS-OFFSET ADJUST START command, and terminates after about 30 ms . The FOCUS-OFFSET ADJUST OFF command is used to return to the state before offset cancellation. The FA amplifier is the pickup phase compensation amplifier, and the equalizer curve is set by the external capacitor and resistance. Furthermore, this amplifier has a mute function which is applied when $\mathrm{V}_{\mathrm{CC}}$ is turned on, when the F-SERVO OFF command is sent, and during F-SEARCH. In order to turn the focus servo on, send either the LASER ON command or the F-SERVO ON command.
The FD amplifier has a phase compensation circuit, and a focus search signal composition function and is completed in about 560 ms . Focus search is initiated by the F-SEARCH command, and a ramp waveform is generated by the internal clock. This waveform is used for focus detection (focus zero cross) with the focus error signal and then turn the focus servo on. The ramp waveform amplitude is set by the resistance between FD (pin 16) and $\mathrm{FD}^{-}$(pin 17).
FSC ( $p$ in 46) is for smoothing the focus search ramp waveforms, and a capacitor is connected between FSC and REF. FSS (pin 55) is the focus search mode switching pin. If FSS is shorted with $\mathrm{V}_{\mathrm{CC}}$, the " + search" is set; if FSS is left open or is shorted with GND, the " $\pm$ search" is set.
5. Tracking servo

The pickup photodiode output current is input to $E$ ( $\operatorname{pin} 3$ ) and $F$ (pin 4). The current that is input is converted to the voltage, passes through the balance adjustment VCA circuit and then the VCA circuit that follows the gain in the RFAGC circuit, and is then output from TE (pin 7). The tracking error gain is set by the resistance between $\mathrm{TE}^{-}$(pin6) and TE (pin7). Offset cancellation is performed by the TE amplifier. Offset cancellation terminates after about 30 ms . The TRACK-OFFSET ADJUST OFF command is used to return to the state before the offset.
The TH amplifier alters the servo response characteristics according to the THLD signal, etc., generated internally after detection of the TGL signal from the DSP or the JP signal. When a defect is detected, the THLD mode goes into effect internally. To avoid this, short DEF (pin 49) to $\mathrm{L}=$ GND. By inserting an external bandpass filter to remove the shock component from the tracking error signal at SCI (pin 9), the gain is automatically boosted when a defect is detected. The TA output (pin 11) has a built-in resistance to allow configuration of a low-pass filter.
The TD amplifier performs servo loop phase compensation; the characteristics are set by external CR. Furthermore, this amplifier has a mute function, which is applied when $\mathrm{V}_{\mathrm{CC}}$ is turned on or the TRACK-SERVO OFF command is issued. The muting function is released by the TRACK-SERVO ON command.
The TOFF amplifier that is positioned immediately after TD (pin 13) functions to turn off the servo in response to the TOFF signal from the DSP.
The TO amplifier has a JP pulse composition function. The JP pulse is set by JP (pin 14). (THLD detection is performed internally.)
6. Sled servo

The response characteristics are set by SLEQ (pin 28). The amplifier positioned after SLEQ (pin 28) has a mute function that is applied either when SLOF (pin 38) goes high or the SLED OFF command is issued. The sled is moved by inputting current to $\mathrm{SL}^{-}$(pin 30) and $\mathrm{SL}^{+}$(pin 31); specifically, the pins are connected to the microprocessor output ports via resistors, and the movement gain is set by the resistance value of that resistor. It is important to note that if there is a deviation in the resistance values for $\mathrm{SL}^{-}$(pin 30) and $\mathrm{SL}^{+}$(pin 31), an offset will arise in the SLD output.
7. Spindle servo

This configures the servo circuit, which maintains the linear velocity of the disc at a constant speed, along with the DSP. This circuit accepts signals from the DSP through $\mathrm{CV}^{-}$(pin 39) and $\mathrm{CV}^{+}$(pin 40) and sets the equalizer characteristics through SP (pin 24), $\mathrm{SP}^{-}$(pin 26), and SPD (pin 27), which are output to SPD (pin 27). The $12-\mathrm{cm}$ mode amplifier gain is set by the resistor connected between SPG (pin 25) and the reference voltage. In 8-cm mode, this amplifier serves as an internal buffer, and SPG (pin 25) is ignored. Note that the gain setting is made for $8-\mathrm{cm}$ mode first, and then $12-\mathrm{cm}$ mode. If SPG (pin 25) is left open, the gain is forcibly set for $8-\mathrm{cm}$ mode, regardless of whether $8-\mathrm{cm}$ or $12-\mathrm{cm}$ mode is in effect.

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8. TES and HFL (traverse signals)

When moving the pickup from the outer track to the inner track, the EF output from the pickup must be connected so that the phase relationship of TES and HFL is as shown in the diagram below. For the TESI input, the TES comparator has negative polarity and hysteresis of approximately $\pm 100 \mathrm{mV}$. An external bandpass filter is needed in order to extract only the required signal from the TE signal.

9. DRF (luminous energy determination)

DRF goes high when the peak of the EFM signal (RFSM output) held by the PH1 (pin 60) capacitor exceeds approximately 2.1 V. The PH1 (pin 60) capacitor affects the DRF detection time constant and the RFAGC response bidirectional setting. The DRF output is driven by a constant current $(250 \mu \mathrm{~A})$.

10. Focus determination

Focus is assumed to be obtained when the focus error signal S curve reaching REF +0.2 V is detected, and the S curve subsequently returns to REF.


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## 11. DEFECT

The mirrored surface level is maintained by the capacitor for LF2 (pin 59); when a drop in the EFM signal (RFSM output) reaches 0.35 V or more, a high signal is output to DEF (pin 49). If DEF (pin 49) goes high, the tracking servo enters THLD mode. In order to prevent the tracking servo from entering THLD mode when a defect is detected, prevent DEFECT from being output by either shorting DEF (pin 49) to GND, or shorting LF2 (pin 59) to GND. The DEFECT output is driven by constant current (approximately $100 \mu \mathrm{~A}$ ).

12. Microprocessor interface

Because the Reset (Nothing) command initializes the LA9241M, it must be used carefully.
The LA9241M command acceptance (mode switching) timing is defined by the internal clock ( 4.23 MHz divided to 130 kHz ) after the falling edge of CE (RWC); therefore, when commands are sent consecutively, CE must go low for at least $10 \mu$ sec. The 4.23 MHz clock is required for that reason. 2BYTE-COMMAND DETECT and 2BYTE-COMMAND RESET are used only for the purpose of masking two-byte data.
All instructions can be input by setting CE high and sending commands synchronized with the CL clock from the microprocessor to DAT (pin 52) in LSB first format. Note that the command is executed at the falling edge of CE.

Timing

13. Reset circuit

The power-on reset is released when $\mathrm{V}_{\mathrm{CC}}$ exceeds approximately 2.6 V .
14. Pattern design notes

To prevent signal jump-in from $\mathrm{CV}^{+}$(pin 40) to RFSM (pin 41), a shielding line is necessary in between.
15. $\mathrm{V}_{\mathrm{CC}} / \mathrm{REF} / \mathrm{GND} / \mathrm{NC}$
$\mathrm{V}_{\mathrm{CC}} 1$ (pin 64) : RF system
$\mathrm{V}_{\mathrm{CC}}{ }^{2}$ (pin 56) : SERVO system, DIGITAL system
AGND (pin 22) : RF system, SERVO system
DGND (pin 45) : DIGITAL system
NC ( pin 23, 48)
: No connection
VR (Pin 58)
: Reference voltage

## Microprocessor Command List

| MSB LSB | COMMAND | Reset mode Power-on mode | DSP |
| :---: | :---: | :---: | :---: |
| 00000000 | RESET |  | RESET(NOTHING) |
| 00001000 | FOCUS START |  | FOCUS START \#1 |
| $\begin{array}{lllllllll} \hline 1 & 1 & 1 & 1 & 0 & 0 & 0 & 0 \\ 1 & 1 & 1 & 1 & 1 & 0 & 0 & 0 \\ 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \end{array}$ | 2BYTE-COMMAND DETECT 2BYTE-COMMAND DETECT 2BYTE-COMMAND RESET |  | 2BYTE-COMMAND DETECT 2BYTE-COMMAND DETECT 2BYTE-COMMAND RESET |
| 10010000 | FOCUS-OFFSET ADJUST START |  | - |
| 10010001 | FOCUS-OFFSET ADJUST OFF | $\bigcirc$ | - |
| 10010010 | TRACK-OFFSET ADJUST START |  | - |
| 10010011 | TRACK-OFFSET ADJUST OFF | $\bigcirc$ | - |
| $\begin{array}{llllllll} \hline 1 & 0 & 0 & 1 & 0 & 1 & 0 & 0 \\ 1 & 0 & 0 & 1 & 0 & 1 & 0 & 1 \end{array}$ | $\begin{aligned} & \text { LASER ON } \\ & \text { LASER OFF : F-SERVO ON } \end{aligned}$ |  | - |
| 10010110 | LASER OFF : F-SERVO OFF | $\bigcirc$ | - |
| 10010111 | SPINDLE 8CM |  | - |
| 10011000 | SPINDLE 12 CM | $\bigcirc$ | - |
| 10011001 | SPINDLE OFF |  | - |
| 10011010 | SLED ON | $\bigcirc$ | - |
| 10011011 | SLED OFF |  | - |
| 10011100 | E/F BALANCE START | Non-adjusted | - |
| 10011101 | TRACK-SERVO OFF | $\bigcirc$ | - |
| 10011110 | TRACK-SERVO ON |  | - |

## Notes Concerning Microprocessor Program Creation

1. Commands

After sending the FOCUS START command and the E/F BALANCE START command, send 11111110 (FEH) in order to clear the internal registers of the IC.
Reason: Although the above commands are executed at point (1) in the timing chart below, the same commands will be executed again at point (2) if there is subsequent input to CE as shown below.

Timing


When sending a TRACK-OFFSET ADJUST START command or a FOCUS-OFFSET ADJUST START command after either $\mathrm{V}_{\mathrm{CC}}$ ON (POWER ON RESET), RESET command, or a corresponding OFFSET ADJUST OFF command, waiting time is necessary as listed below. (Only when a 4.2 MHz clock is input.)

TRACK-OFFSET ADJUST START: 4 ms or more
FOCUS-OFFSET ADJUST START: 4 ms or more
2. $\mathrm{E} / \mathrm{F}$ balance adjustment
$\mathrm{E} / \mathrm{F}$ balance adjustments should be made in a bit region of the disc, not a mirrored region.
Since there is no track-kick for LA9241M, measures must be taken during EF balance adjustment to obtain a stable TE signal. (By a sled movement signal from a microprocessor, for example.)

Pin Internal Equivalent Circuit
Pin No.

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Pin No.

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Pin No.

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Pin No. Pin name

## Sample Application Circuit



Function Comparisons for CD-ASP

| Type | LA9220M | LA9230M | LA9231M | LA9233M | LA9240M | LA9241M |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Package | QFP-64E | QFP-64E | QFP-64E | QFP-64E | QFP-64E | QFP-64E |
| Allowable operating supply voltage <br> $V_{\text {CCOP }}$ max <br> $V_{\text {CCOP }}$ min1 <br> $\mathrm{V}_{\mathrm{CC}}$ op min2 | $\begin{aligned} & \text { 5.5 } \mathrm{V} \\ & 3.6 \mathrm{~V}: \mathrm{t}=-25 \text { to }+75^{\circ} \mathrm{C} \end{aligned}$ | $\begin{aligned} & 5.5 \mathrm{~V} \\ & 3.6 \mathrm{~V}: \mathrm{t}=-25 \text { to }+75^{\circ} \mathrm{C} \end{aligned}$ | $\begin{aligned} & 5.5 \mathrm{~V} \\ & 3.6 \mathrm{~V}: \mathrm{t}=-25 \text { to }+75^{\circ} \mathrm{C} \end{aligned}$ | $\begin{aligned} & \text { 5.5 } \mathrm{V} \\ & 3.6 \mathrm{~V}: \mathrm{t}=-25 \text { to }+75^{\circ} \mathrm{C} \end{aligned}$ | $\begin{aligned} & \text { 5.5 } \mathrm{V} \\ & 3.6 \mathrm{~V}: \mathrm{t}=-25 \text { to }+75^{\circ} \mathrm{C} \\ & 3.4 \mathrm{~V}: \mathrm{t}=-5 \text { to }+75^{\circ} \mathrm{C} \end{aligned}$ | $\begin{aligned} & 5.5 \mathrm{~V} \\ & 3.2 \mathrm{~V}: \mathrm{t}=-25 \text { to }+75^{\circ} \mathrm{C} \\ & 3.0 \mathrm{~V}: \mathrm{t}=-10 \text { to }+75^{\circ} \mathrm{C} \end{aligned}$ |
| Current drain (When $\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}$ ) | 40 mA | 32 mA | 32 mA | 32 mA | 32 mA | 32 mA |
| VCO circuit built in | Yes | No | No | No | No | No |
| Auto adjustment function |  |  |  |  |  |  |
| - Focus offset adjsutment Maximum adjsutment time | Adjustment position: FD 270 ms | Adjustment position: FD 270 ms | Adjustment position: FD 270 ms | Adjustment position: FD 270 ms | Adjustment position: FE 30 ms | Adjustment position: FE 30 ms |
| Tracking offset adjustment Maximum adjustment time | Adjustment position: TO 30 ms | Adjustment position: TO 30 ms | Adjustment position: TO 30 ms | Adjustment position: TO 30 ms | Adjustment position: TE 30 ms | Adjustment position: TE 30 ms |
| - E/F balance auto adjustment | Yes | Yes | Yes | Yes | Yes | Yes |
| - RF level AGC function RF waveform amplitude when $\mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}$ <br> RF waveform amplitude when $V_{C C}$ min | $\begin{aligned} & \text { Yes } \\ & 1.8 \mathrm{Vp}-\mathrm{p} \\ & 1.3 \mathrm{Vp}-\mathrm{p} \end{aligned}$ | $\begin{aligned} & \text { Yes } \\ & 1.8 \mathrm{Vp}-\mathrm{p} \\ & 1.3 \mathrm{Vp}-\mathrm{p} \end{aligned}$ | $\begin{aligned} & \text { Yes } \\ & \text { 1.8 Vp-p } \\ & 1.3 \mathrm{Vp}-\mathrm{p} \end{aligned}$ | $\begin{aligned} & \text { Yes } \\ & \text { 1.8 Vp-p } \\ & \text { 1.3 Vp-p } \end{aligned}$ | $\begin{aligned} & \text { Yes } \\ & 1.8 \mathrm{Vp}-\mathrm{p} \\ & 1.2 \mathrm{Vp}-\mathrm{p}: \mathrm{V}_{\mathrm{CC}}=3.4 \mathrm{~V} \end{aligned}$ | $\begin{aligned} & \text { Yes } \\ & 1.5 \mathrm{Vp}-\mathrm{p} \\ & 0.9 \mathrm{Vp}-\mathrm{p}: \mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V} \end{aligned}$ |
| - Tracking servo gain RF level following function | Yes | Yes | Yes | Yes | Yes | Yes |
| Focus search time | Approx. 280 msec | Approx. 280 msec | Approx. 1.1 sec | Approx. 1.1 sec | Approx. 560 msec | Approx. 560 msec |
| Regeneration speed | $\times 2$ | $\times 2$ | $\times 2$ | $\times 4$ | $\times 4$ | $\times 4$ |
| Track-kick signal output (Track-kick during E/F balance adjsutment) | Built in | Built in | No output | No output | No output | No output |
| Focus search smoothing capacitor pin: FSC | No | No | Yes | Yes | Yes | Yes |
| E/F balance variable range setting pin: TBC | No | No | No | No | Yes | Yes |
| Focus search mode switch pin: FSS | No | No | No | No | Yes | Yes |
| Vth for HFL detection | 2.3 V | 2.3 V | 2.3 V | 2.3 V | 2.1 V | 2.1 V |
| DRF current capacity | Approx. $100 \mu \mathrm{~A}$ | Approx. $100 \mu \mathrm{~A}$ | Approx. $100 \mu \mathrm{~A}$ | Approx. $100 \mu \mathrm{~A}$ | Approx. $250 \mu \mathrm{~A}$ | Approx. $250 \mu \mathrm{~A}$ |
| APC reference voltage LCD voltage where LDD $=3 \mathrm{~V}$ | 180 mV : typ | 180 mV : typ | 180 mV : typ | 180 mV : typ | 180 mV : typ | 190 mV: typ |
| No connection |  | Pin 46, 47, 48, 55 | Pin 47, 48, 55 | Pin 47, 48, 55 | Pin 48 | Pin 23,48 |

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